

2N7638-GA

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600 V

20 A

110

170 mΩ

Normally – OFF Silicon Carbide Junction Transistor

Features

- 225°C maximum operating temperature
- Electrically Isolated Base Plate
- Gate Oxide Free SiC Switch
- Exceptional Safe Operating Area
- Excellent Gain Linearity
- Temperature Independent Switching Performance
- Low Output Capacitance
- Positive Temperature Coefficient of RDS,ON
- Suitable for Connecting an Anti-parallel Diode

Advantages

- Compatible with Si MOSFET/IGBT Gate Drive ICs
- > 20 µs Short-Circuit Withstand Capability
- Lowest-in-class Conduction Losses
- High Circuit Efficiency
- Minimal Input Signal Distortion
- High Amplifier Bandwidth

Package

RoHS Compliant



V_{DS}

R_{DS(ON)}

I_{D (Tc = 25°C)}

h_{FE (Tc = 25°C)}

SMD0.5 / TO - 276 (Hermetic Package)

Applications

- Down Hole Oil Drilling
- Geothermal Instrumentation
- Solenoid Actuators
- General Purpose High-Temperature Switching
- Amplifiers
- Solar Inverters
- Switched-Mode Power Supply (SMPS)
- Power Factor Correction (PFC)

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Section I: Absolute Maximum Ratings

Parameter	Symbol	Conditions	Values	Unit
Drain – Source Voltage	V _{DS}	$V_{GS} = 0 V$	600	V
Continuous Drain Current	ID	$T_J = 225^{\circ}C, T_C = 25^{\circ}C$	20	А
Continuous Gate Current	I _{GM}		1.25	А
Turn-Off Safe Operating Area	RBSOA	T _{VJ} = 225°C, I _G = 1.25 A, Clamped Inductive Load	$I_{D,max} = 20$ @ $V_{DS} \le V_{DSmax}$	А
Short Circuit Safe Operating Area	SCSOA	T_{VJ} = 225°C, I_G = 1.25 A, V_{DS} = 400 V, Non Repetitive	>20	μs
Reverse Gate – Source Voltage	V _{GS}		30	V
Reverse Drain – Source Voltage	V _{DS}		40	V
Power Dissipation	P _{tot}	$T_J = 225^{\circ}C, T_C = 25^{\circ}C$	200	W
Operating and Storage Temperature	T _j , T _{stg}		-55 to 225	°C



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Section II: Static Electrical Characteristics

Parameter	Symbol Conditions	Values		11		
	Symbol	Conditions	min.	typ.	max.	Unit
A: On State						
		I _D = 7 A, T _j = 25 °C		170		
Drain – Source On Resistance	R _{DS(ON)}	I _D = 7 A, T _j = 175 °C		320		mΩ
		I _D = 7 A, T _j = 220 °C		440		
Gate – Source Saturation Voltage	V _{GS,SAT}	$I_D = 10 \text{ A}, I_D/I_G = 40, T_j = 25 \text{ °C}$		3.50		V
Gale – Source Saluration Voltage	V GS,SAT	$I_D = 10 \text{ A}, I_D/I_G = 30, T_j = 175 \text{ °C}$		3.27		v
DC Current Gain	h _{FE}	$V_{DS} = 5 \text{ V}, \text{ I}_{D} = 10 \text{ A}, \text{ T}_{j} = 25 ^{\circ}\text{C}$	80	110		
	IIFE	$V_{DS} = 5 \text{ V}, \text{ I}_{D} = 10 \text{ A}, \text{ T}_{j} = 220 ^{\circ}\text{C}$	50	80		
B: Off State						
		$V_R = 600 \text{ V}, \text{ V}_{GS} = 0 \text{ V}, \text{ T}_j = 25 \text{ °C}$		10	100	
Drain Leakage Current	IDSS	$V_R = 600 \text{ V}, V_{GS} = 0 \text{ V}, T_j = 175 \text{ °C}$		40	400	μA
		$V_R = 600 \text{ V}, V_{GS} = 0 \text{ V}, T_j = 220 \text{ °C}$		100	600	-
C: Thermal						
Thermal resistance, junction - case	R _{thJC}			1.0		°C/W

Section III: Dynamic Electrical Characteristics

Baramatar	Symbol	Conditions	١	Values		Unit	
Parameter	Symbol	Conditions	min.	typ.	max.	Unit	

A: Capacitance and Gate Charge

Input Capacitance	C _{iss}	V _{GS} = 0 V, V _D = 500 V, <i>f</i> = 1 MHz	685	pF
Reverse Transfer/Output Capacitance	C _{rss} /C _{oss}	$V_{\rm D} = 500 \text{ V}, f = 1 \text{ MHz}$	24	pF
Output Capacitance Stored Energy	Eoss	V _{GS} = 0 V, V _D = 500 V, <i>f</i> = 1 MHz	3.1	μJ
Effective Output Capacitance, time related	$C_{oss,tr}$	I_D = constant, V_{GS} = 0 V, V_{DS} = 0400 V	50	pF
Effective Output Capacitance, energy related	$C_{\text{oss,er}}$	$V_{GS} = 0 \text{ V}, \text{ V}_{DS} = 0400 \text{ V}$	37	pF
Gate-Source Charge	Q_{GS}	V _{GS} = -53 V	11	nC
Gate-Drain Charge	Q_{GD}	V _{GS} = 0 V, V _{DS} = 0400 V	20	nC
Gate Charge - Total	Q_{G}		31	nC

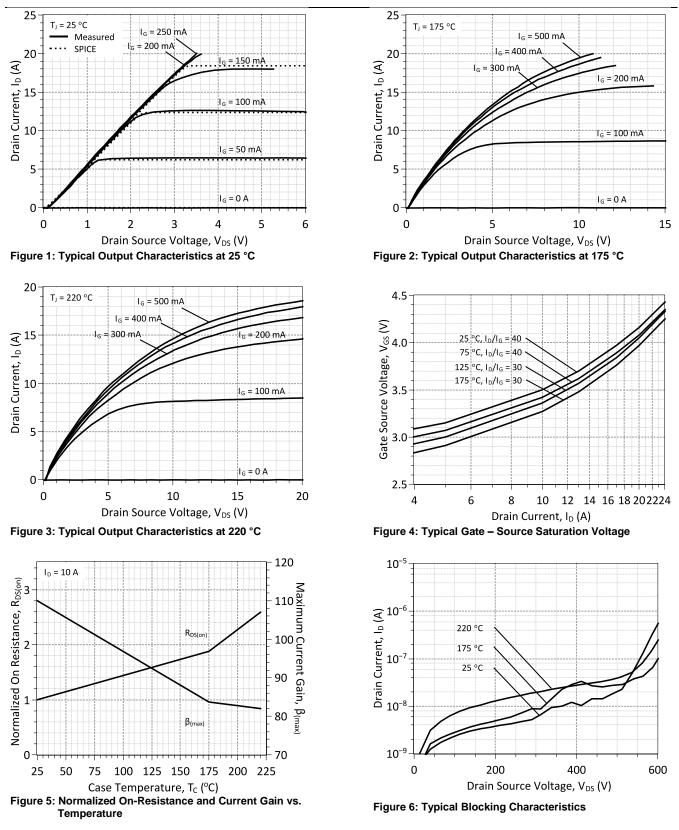
B: Switching

Turn On Delay Time	t _{d(on)}		10	ns
Rise Time	tr	T _j = 175 °C, V _{DS} = 400 V,	30	ns
Turn Off Delay Time	t _{d(off)}		75	ns
Fall Time	t _f	I _D = 7 A, Inductive Load Refer to Section V for additional	40	ns
Turn-On Energy Per Pulse	Eon	driving information.	35	μJ
Turn-Off Energy Per Pulse	E _{off}		65	μJ
Total Switching Energy	Ets		100	μJ
Turn On Delay Time	t _{d(on)}		10	ns
Rise Time	t _r		30	ns
Turn Off Delay Time	t _{d(off)}	$T_j = 225 \text{ °C}, V_{DS} = 400 \text{ V},$	75	ns
Fall Time	t _f	I _D = 7 A, Inductive Load Refer to Section V for additional driving information.	60	ns
Turn-On Energy Per Pulse	Eon		45	μJ
Turn-Off Energy Per Pulse	E _{off}		80	μJ
Total Switching Energy	E _{ts}		125	μJ

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Section IV: Figures





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B: Dynamic Characteristics

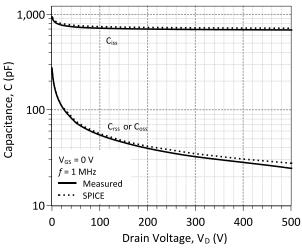
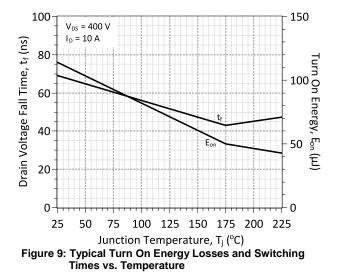
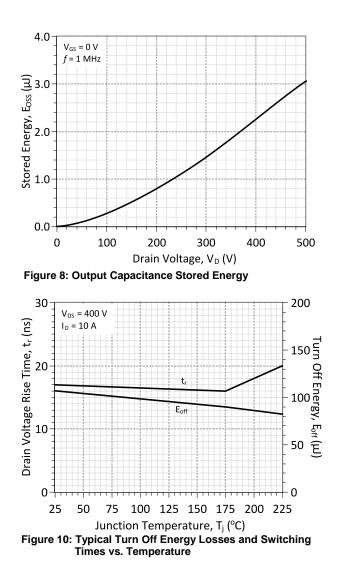


Figure 7: Capacitance Characteristics





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Section V: Driving the 2N7638-GA

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The 2N7638-GA is a current controlled SiC transistor which requires a positive gate current for turn-on and to remain in on-state. It may be driven by different drive topologies depending on the intended application.

Drive Topology	Gate Drive Power Consumption	Switching Frequency
Simple TTL	High	Low
Constant Current	Medium	Medium
High Speed – Boost Capacitor	Medium	High
High Speed – Boost Inductor	Low	High
Proportional	Lowest	Medium
Pulsed Power	Medium	N/A

Table 1: Estimated Power Consur	nption and switching frequencies	for various Gate Drive topologies.

A: Simple TTL Drive

The 2N7638-GA may be driven by 5 V TTL logic using a simple current amplification stage. The current amplifier output current must meet or exceed the steady state gate current, $I_{G, steady}$, required to operate the 2N7638-GA. An external gate resistor R_G , shown in the Figure 11 topology, sets $I_{G, steady}$ to the required level which is dependent on the SJT drain current I_D and DC current gain h_{FE} , R_G may be calculated from the equation below. The value of $V_{EC, sat}$ can be taken from the PNP datasheet, a partial list of high-temperature PNP and NPN transistors options is given below. High-temperature MOSFETs may also be used in the topology.

$$R_{G,max} = \frac{\left(5.0 \, V - V_{EC,sat}(PNP) - V_{GS,sat}(SJT)\right) * h_{FE}(T, I_D)}{I_D * 1.5}$$

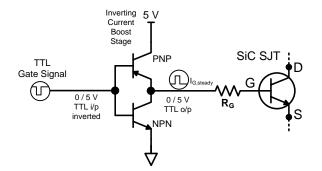


Figure 11: Simple TTL Gate Drive Topology

BJT Part Number	Туре	Т _{ј,max} (°С)
PHPT60603PY	PNP	175
PHPT60603NY	NPN	175
2N2222	NPN	200
2N6730	PNP	200
2N2905	PNP	200
2N5883	PNP	200
2N5885	NPN	200

Table 2: Partial List of High-Temperature BJTs for TTL Gate Driving

B: High Speed Driving

For ultra high speed 2N7638-GA switching (t_n , t_r < 20 ns) while maintaining low gate drive losses the supplied gate current should include a positive current peak during turn-on, a negative voltage peak during turn-off, and continuous gate current I_G to remain on.

An SJT is rapidly switched from its blocking state to on-state, when the necessary gate charge for turn-on, Q_G , is supplied by a burst of high gate current until the gate-source capacitance, C_{GS} , and gate-drain capacitance, C_{GD} , are fully charged. Ideally, the burst should terminate when the drain voltage has fallen to its on-state value in order to avoid unnecessary drive losses. A negative voltage peak is recommended for the turn-off transition in order to ensure that the gate current is not being supplied under high dV/dt due to the Miller effect. While satisfactory turn off can be achieved with $V_{GS} = 0 V$, a negative V_{GS} value may be used in order to speed up the turn-off transition.

B:1: High Speed, Low Loss Drive with Boost Capacitor

The 2N7638-GA may be driven using a High Speed, Low Loss Drive with Boost Capacitor topology in which multiple voltage levels, a gate resistor, and a gate capacitor are used to provide current peaks at turn-on and turn-off for fast switching and a continuous gate current while in on-state. As shown in Figure 12, in this topology two gate driver ICs are utilized. An external gate resistor R_G is driven by a low voltage driver to supply the continuous gate current throughout on-state. and a gate capacitor C_G is driven at a higher voltage level to supply a high current peak at turn-on and turn-off. A 3 kV isolated evaluation gate drive board (GA03IDDJT30-FR4) from GeneSiC Semiconductor utilizing this topology is commercially available for high and low-side driving, its datasheet provides additional details about this drive topology.

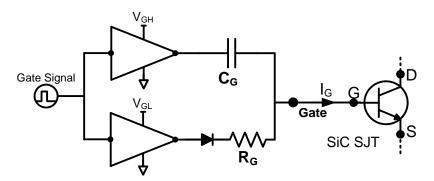


Figure 12: High Speed, Low Loss Drive with Boost Capacitor Topology

B:2: High Speed, Low Loss Drive with Boost Inductor

A High Speed, Low-Loss Driver with Boost Inductor is also capable of driving the 2N7638-GA at high-speed. It utilizes a gate drive inductor instead of a capacitor to provide the high-current gate current pulses $I_{G,on}$ and $I_{G,off}$. During operation, inductor L is charged to a specified $I_{G,on}$ current value then made to discharge I_L into the SJT gate pin using logic control of S₁, S₂, S₃, and S₄, as shown in Figure 13. After turn on, while the device remains on the necessary steady state gate current $I_{G,steady}$ is supplied from source V_{CC} through R_G. Please refer to the article "A current-source concept for fast and efficient driving of silicon carbide transistors" by Dr. Jacek Rąbkowski for additional information on this driving topology.³

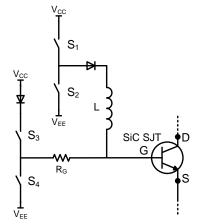


Figure 13: High Speed, Low-Loss Driver with Boost Inductor Topology

³ – Archives of Electrical Engineering. Volume 62, Issue 2, Pages 333–343, ISSN (Print) 0004-0746, DOI: 10.2478/aee-2013-0026, June 2013



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C: Proportional Gate Current Driving

A proportional gate drive topology may be beneficial for applications in which the 2N7638-GA will operate over a wide range of drain current conditions to lower the gate drive power consumption. A proportional gate driver relies on instantaneous drain current I_D feedback to vary the steady state gate current $I_{G,steady}$ supplied to the 2N7638-GA.

C:1: Voltage Controlled Proportional Driver

A voltage controlled proportional driver relies on a gate drive integrated circuit to detect the 2N7638-GA drain-source voltage V_{DS} during onstate to sense I_D . The integrated circuit will then increase or decrease I_G in response to I_D . This allows I_G and gate drive power consumption to reduce while I_D is low or for I_G to increase when I_D increases. A high voltage diode connected between the drain and sense protects the integrated circuit from high-voltage when blocking. A simplified version of this topology is shown in Figure 14. Additional information will be available in the future at http://www.genesicsemi.com/references/product-notes/.

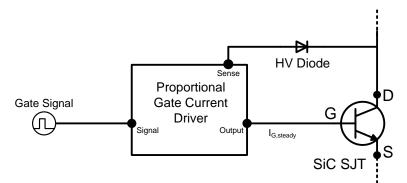


Figure 14: Simplified Voltage Controlled Proportional Driver

C:2: Current Controlled Proportional Driver

The current controlled proportional driver relies on a low-loss transformer in the drain or source path to provide feedback of the 2N7638-GA drain current during on-state to supply $I_{G,steady}$ into the gate. $I_{G,steady}$ will increase or decrease in response to I_D at a fixed forced current gain which is set be the turns ratio of the transformer, $h_{force} = I_D / I_G = N_2 / N_1$. 2N7638-GA is initially tuned-on using a gate current pulse supplied into an RC drive circuit to allow I_D current to begin flowing. This topology allows $I_{G,steady}$ and the gate drive power consumption to reduce while I_D is relatively low or for $I_{G,steady}$ to increase when I_D increases. A simplified version of this topology is shown in Figure 15. Additional information will be available in the future at http://www.genesicsemi.com/references/product-notes/.

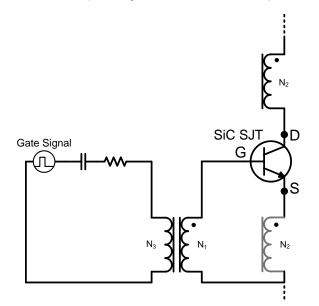
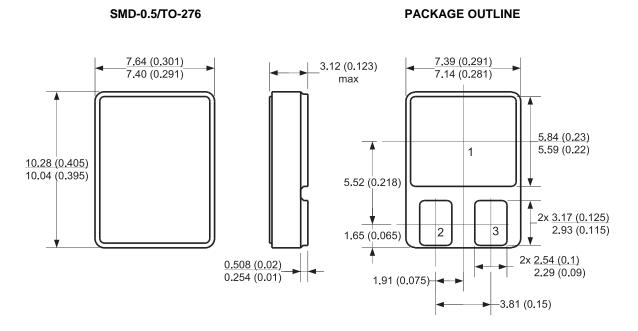


Figure 15: Simplified Current Controlled Proportional Driver



2N7638-GA

Section VI: Package Dimensions:



NOTE

1. CONTROLLED DIMENSION IS INCH. DIMENSION IN BRACKET IS MILLIMETER.

2. DIMENSIONS DO NOT INCLUDE END FLASH, MOLD FLASH, MATERIAL PROTRUSIONS

	Revision History				
Date	Revision	Comments	Supersedes		
2014/12/12	6	Updated Electrical Characteristics			
2014/08/23	5	Updated Electrical Characteristics			
2014/03/20	4	Updated Gate Drive Section			
2014/02/11	3	Updated Electrical Characteristics			
2013/12/19	2	Updated Gate Drive Section			
2013/11/18	1	Updated Electrical Characteristics			
2012/08/24	0	Initial release			

Published by GeneSiC Semiconductor, Inc. 43670 Trade Center Place Suite 155 Dulles, VA 20166

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GeneSiC SEMICONDUCTOR

Section VII: SPICE Model Parameters

This is a secure document. Please copy this code from the SPICE model PDF file on our website (http://www.genesicsemi.com/images/hit_sic/sjt/2N7638-GA_SPICE.pdf) into LTSPICE (version 4) software for simulation of the 2N7638-GA.

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       MODEL OF GeneSiC Semiconductor Inc.
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       $Revision: 1.3
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       $Date: 12-DEC-2014
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* These models are provided "AS IS, WHERE IS, AND WITH NO WARRANTY
* OF ANY KIND EITHER EXPRESSED OR IMPLIED, INCLUDING BUT NOT LIMITED
* TO ANY IMPLIED WARRANTIES OF MERCHANTABILITY AND FITNESS FOR A
* PARTICULAR PURPOSE."
* Models accurate up to 2 times rated drain current.
.model 2N7638 NPN
              9.8338E-48
+ IS
              1.0733E-26
+ ISE
+ EG
              3.23
+ BF
              130
+ BR
              0.55
              200
+ IKF
+ NF
               1
+ NE
              2.
              7.2
+ RB
+ IRB
              0.002
              0.2
+ RBM
              0.1039
+ RE
+ RC
              0.06188
+ CJC
              2.73E-10
+ VJC
              3.04
+ MJC
              0.448
+ CJE
              6.86E-10
+ VJE
              2.89
+ MJE
              0.466
+ XTI
              3
+ XTB
              -0.35
+ TRC1 1.90E-2
+ VCEO 600
+ ICRATING
               20
             GeneSiC_Semiconductor
+ MFG
* End of 2N7638-GA SPICE Model
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