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CMOS 8-Bit Microcontroller

# TMP87CH38N/F, TMP87CK38N/F

The 87CH38/K38 is the high speed and high performance 8-bit single chip microcomputer. This MCU contains CPU core, ROM, RAM, input/output ports, six multi-function timer / counter, serial bus interface, on-screen display, PWM, 8-bit A/D converter and remote control signal preprocessor on a chip.

| Part No. | ROM | RAM | Package | OTP MCU |
|---|---|---|---|---|
| TMP87CH38N/F | 16 Kbytes | 512 bytes | SDIP42-P-600-1.78 / QFP44-P-1414-0.80D | TMP87PS38N/F |
| TMP87CK38N/F | 24 Kbytes | | | |

## Features

◆8-bit single chip microcomputer TLCS-870 Series
◆Instruction execution time : 0.5 $\mu$s (at 8 MHz)
◆412 basic instructions
- Multiplication and Division (8 bits × 8 bits , 16 bits ÷ 8 bits)
- Bit manipulations(Set/Clear/Complement/Move/Test/Exclusive Or)
- 16-bit data operations
- 1-byte jump / subroutine-call (Short relative jump / Vector call)

◆14 interrupt sources (External : 5, Internal : 9)
- All sources have independent latches each, and nested interrupt control is available.
- 3 edge-selectable external interrupts with noise reject
- High-speed task switching by register bank changeover

◆ROM Corrective Function
◆6 Input / Output ports (33 pins)
- High current output : 4 pins (typ. 20 mA)

◆Two 16-bit Timer / Counters
- Timer, Event counter, Pulse width measurement, External trigger timer, window modes

◆Two 8-bit Timer / Counters
- Timer, Event counter, Capture (Pulse width / duty measurement) modes

◆Time Base Timer (Interrupt frequency : 1 Hz to 16 kHz)
◆Watchdog Timer
- Interrupt source / reset output (programmable)

◆Serial bus Interface
- I²C-bus, 8-bit SIO modes

980910EBP1


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◆On-screen display circuit
- Character patterns : 256 characters
- Characters displayed : 24 columns × 8 lines
- Composition : 14 × 18 dots
- Size of character : 3 kinds (line by line)
- Color of character : 8 kinds (character by character)
- Variable display position : Horizontal 128 steps, Vertical 256 steps
- Fringing, Smoothing function

◆D/A conversion (Pulse Width Modulation) outputs
- 14-bit resolution (1 channel)
- 7-bit resolution (9 channels)

◆8-bit successive approximate type A/D converter with sample and hold

◆Remote control signal preprocessor

◆Two Power saving operating modes
- STOP mode : Oscillation stops. Battery / Capacitor back-up. Port output hold / high-impedance.
- IDLE mode : CPU stops, and Peripherals operate. Release by interrupts.

◆Jitter Elimination

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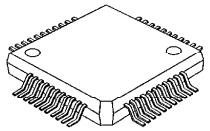
Package

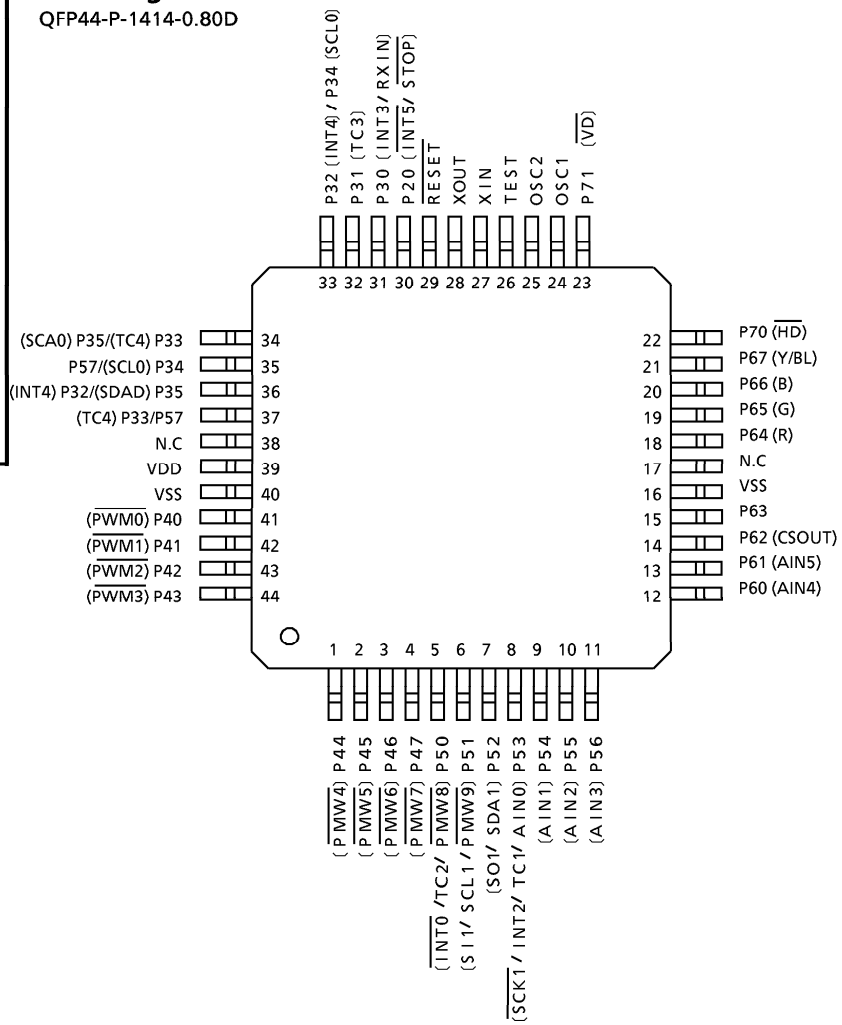
QFP44-P-1414-0.80D



TMP87CH38F
TMP87CK38F
TMP87PS38F

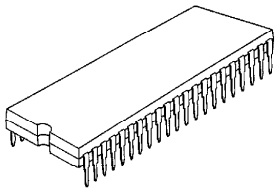
## Pin Assignments

QFP44-P-1414-0.80D




Package

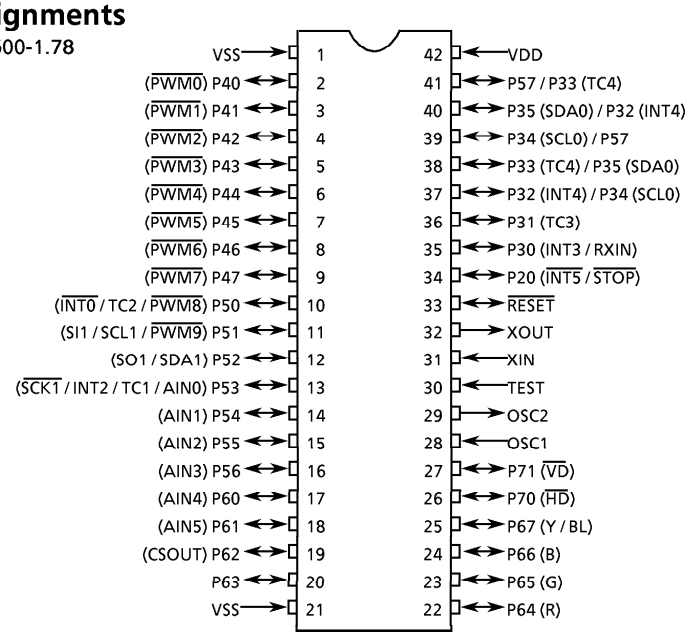
SDIP42-P-600-1.78



TMP87CH38N
TMP87CK38N
TMP87PS38N

## Pin Assignments

SDIP42-P-600-1.78

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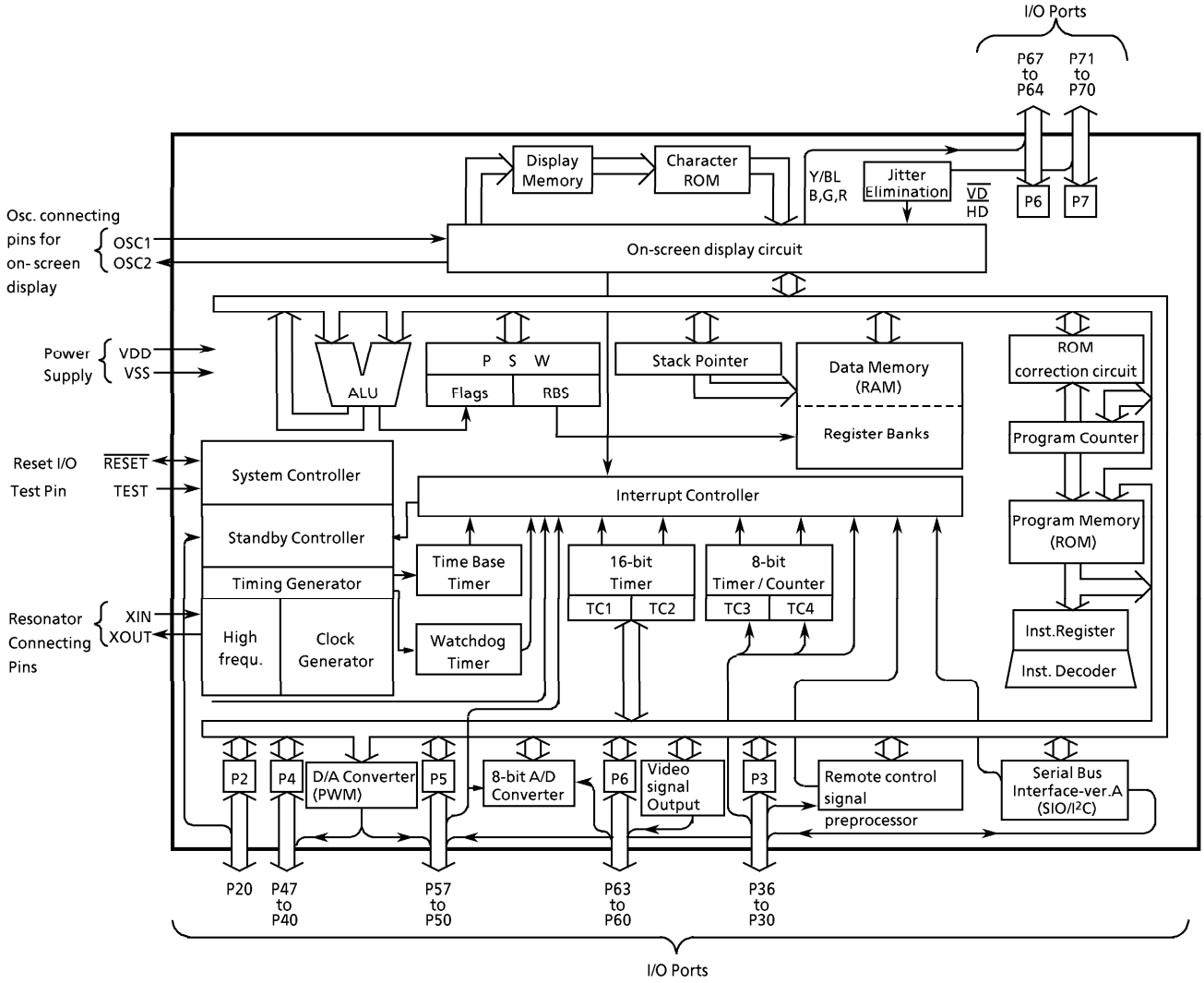
## Block Diagram

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## Pin Function

| Pin Name | Input / Output | Function | | |
|---|---|---|---|---|
| P20 ($\overline{INT5}$/$\overline{STOP}$) | I/O (Input) | 1-bit input / output port with latch. When used as an input port, the latch must be set to "1". | External interrupt input 5 or STOP mode release signal input | |
| P35 (SDA0) | I/O (I/O) | 6-bit input/output port with latch. When used as an input port, a serial bus interface input/output, a timer/counter input, a remote control signal preprocessor input, or an external interrupt input, the latch must be set to "1". | I²Cbus serial data input/output | |
| P34 (SCL0) | I/O (I/O) | | I²Cbus serial clock input/output | |
| P33 (TC4) | I/O (Input) | | Timer / Counter 4 input | |
| P32 (INT4) | | | External interrupt input 4 | |
| P31 (TC3) | | | Timer / Counter 3 input | |
| P30 (INT3/RXIN) | I/O (Input/Input) | | External interrupt input 3 or remote control signal preprocessor input | |
| P47 ($\overline{PWM7}$) to P41 ($\overline{PWM1}$) | I/O (Output) | 8-bit programmable input/output port (tri-state). Each bit of this port can be individually configured as an input or an output under software control. During reset, all bits are configured as inputs. When used as a PWM output, the latch must be set to "1". | 7-bit D/A conversion (PWM) outputs | |
| P40 ($\overline{PWM0}$) | | | 14-bit D/A conversion (PWM) output | |
| P57 | I/O (Input) | 8-bit input/output port with latch. When used as an input port, a analog input, a PWM output, or a pulse output, the latch must be set to "1". | | |
| P56 (AIN3) to P53 (AIN0) | | | 8-bit A/D conversion inputs | |
| P53 (AIN0 / TC1/ INT2 / $\overline{SCK1}$) | I/O | | A/D conversion inputs or Timer / Counter 1 inputs or interrupt input 2 or SIO serial clock input/output | |
| P52 (SDA1/SO1) | I/O (Input/Output) | | I²Cbus serial data input/output or SIO Serial data output | |
| P51 ($\overline{PWM9}$ /SCL1/SI1) | | | 7-bit D/A conversion (PWM) outputs | I²Cbus serial data input / output or SIO Serial data input |
| P50 ($\overline{PWM8}$/TC2/ $\overline{INT0}$) | | | | Timer / Counter 2 input / External interrupt input 0 |
| P67 (Y/BL) | I/O(Output) | 8-bit programmable input/output port (P67 to P64 : tri-state, P63 to P60 : High current output). Each bit of this port can be individually configured as an input or an output under software control. During reset, all bits are configured as inputs. When used as the R, G, B, Y / BL outputs of on-screen display circuit, each bit of the P6 port data selection register (bits 7 to 4 in address $0F91_H$) must be set to "1". | Focus signal output or Background blanking control signal output | |
| P66 (B) | | | RGB output | |
| P65 (G) | | | | |
| P64 (R) | | | | |
| P63 | I/O | | High current output. | |
| P62 (CSOUT) | | | | Test video signal output |
| P61 (AIN5) | | | | A/D conversion inputs |
| P60 (AIN4) | | | | |
| P71 ($\overline{VD}$) | I/O (Input) | 2-bit input/output port with latch. When used as an input ports, or a vertical synchronous signal input and horizontal synchronous signal input, the latch must be set to "1". | Vertical synchronous signal input | |
| P70 ($\overline{HD}$) | | | Horizontal synchronous signal input | |
| OSC1, OSC2 | Input, Output | Resonator connecting pins for on-screen display circuitry. | | |
| XIN, XOUT | | Resonator connecting pins. For inputting external clock, XIN is used and XOUT is opened. | | |
| $\overline{RESET}$ | I/O | Reset signal input or watchdog timer output/address-trap- reset output/system-clock-reset output. | | |
| TEST | Input | Test pin for out-going test. Be tied to low. | | |
| VDD, VSS | Power Supply | + 5 V, 0 V (GND) | | |

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## Operational Description

### 1. CPU Core Functions

The CPU core consists of a CPU, a system clock controller, an interrupt controller, and a watchdog timer. This section provides a description of the CPU core, the program memory (ROM), the data memory (RAM), and the reset circuit.

### 1.1 Memory Address Map

The TLCS-870 Series is capable of addressing 64 K bytes of memory. Figure 1-1 shows the memory address maps of the 87CH38/K38. In the TLCS-870 Series, the memory is organized 4 address spaces (ROM, RAM, SFR, and DBR). It uses a memory mapped I/O system, and all I/O registers are mapped in the SFR / DBR address spaces. There are 16 banks of general-purpose registers. The register banks are also assigned to the first 128 bytes of the RAM address space.

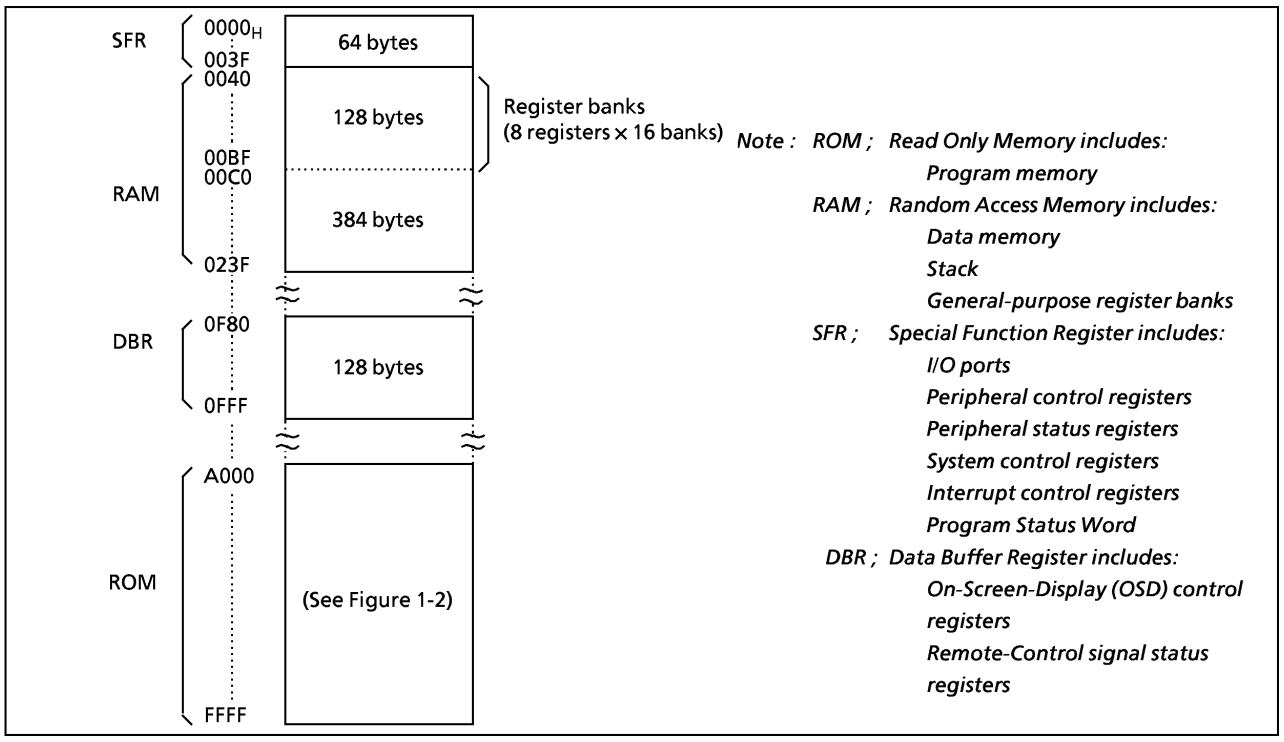


Figure 1-1. Memory address map

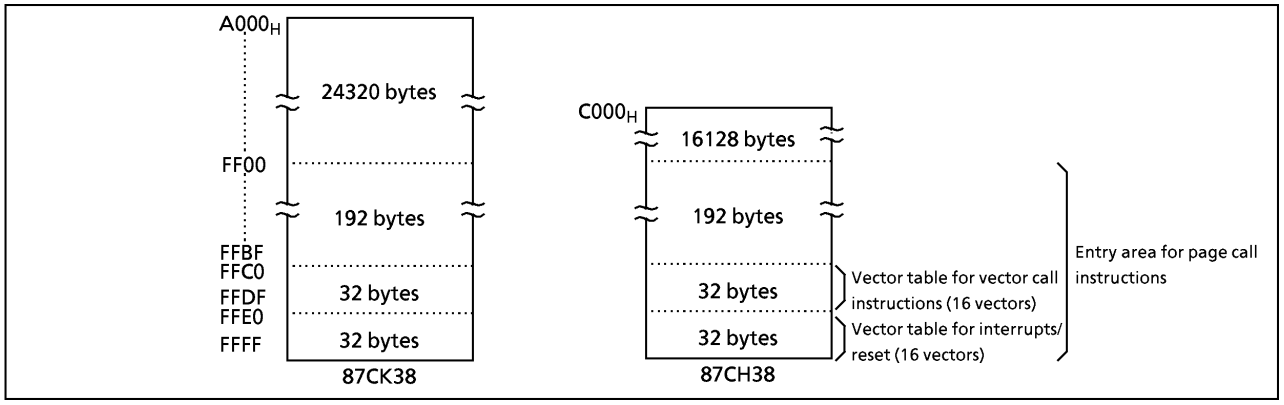


Figure 1-2. ROM address maps

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## 1.2 Program Memory (ROM)

The 87CH38 has a 16K bytes (addresses $C000_H$ to $FFFF_H$), and the 87CK38 has a 24K bytes (addresses $A000_H$ to $FFFF_H$) of program memory (mask programmed ROM). Addresses $FF00_H$-$FFFF_H$ in the program memory can also be used for special purposes. Figure 1-2 shows the ROM address maps of the 87CH38/K38.

(1) **Interrupt / Reset** vector table (addresses $FFE0_H$ to $FFFF_H$)
This table consists of a reset vector and 15 interrupt vectors (2 bytes / vector). These vectors store a reset start address and 15 interrupt service routine entry addresses.

(2) Vector table for **vector call** instructions (addresses $FFC0_H$ to $FFDF_H$)
This table stores call vectors (subroutine entry address, 2 bytes/vector) for the vector call instructions [CALLV n]. There are 16 vectors. The CALLV instruction increases memory efficiency when utilized for frequently used subroutine calls (called from 3 or more locations).

(3) Entry area (addresses $FF00_H$ to $FFFF_H$) for **page call** instructions
This is the subroutine entry address area for the page call instructions [CALLP n]. Addresses $FF00_H$ to $FFBF_H$ are normally used because address $FFC0_H$ to $FFFF_H$ are used for the vector tables.

Programs and fixed data are stored in the program memory. The instruction to be executed next is read from the address indicated by the current contents of the program counter (PC). There are relative jump and absolute jump instructions. The concepts of page or bank boundaries are not used in the program memory concerning any jump instruction.

Example : The relationship between the jump instructions and the PC.

① 5-bit PC-relative jump [JRS cc, $ + 2 + d]
E8C4H: JRS T, $ + 2 + 08H
When JF = 1, the jump is made to $E8CE_H$, which is $08_H$ added to the contents of the PC. (The PC contains the address of the instruction being executed + 2; therefore, in this case, the PC contents are $E8C4_H + 2 = E8C6_H$.)

② 8-bit PC-relative jump [JR cc, $ + 2 + d]
E8C4H : JR Z, $ + 2 + 80H
When ZF = 1, the jump is made to $E846_H$, which is $FF80_H$ ( – 128) added to the current contents of the PC.

③ 16-bit absolute jump [JP a]
E8C4H : JP 0C235H
An unconditional jump is made to address $C235_H$. The absolute jump instruction can jump anywhere within the entire 64K-byte space.

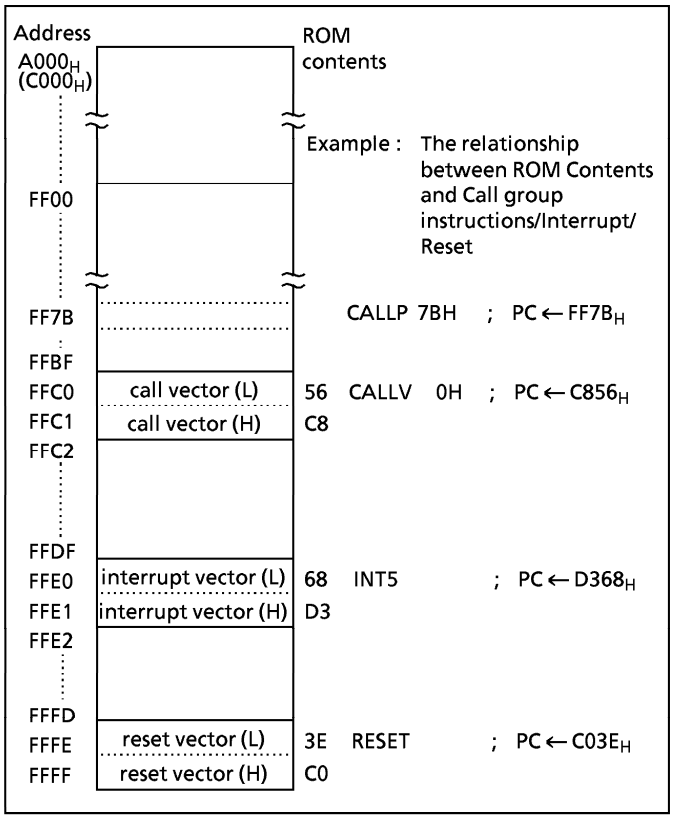


Figure 1-3. Program memory map

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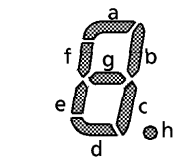
In the TLCS-870 Series, the same instruction used to access the data memory (e.g. [LD A, (HL)] ) is also used to read out fixed data (ROM data) stored in the program memory. The register-offset PC-relative addressing (PC + A) instructions can also be used, and the code conversion, table look-up and n-way multiple jump processing can easily be programmed.

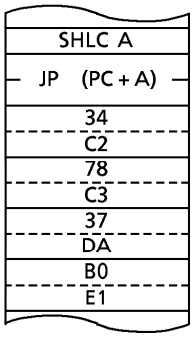
Example 1 : Loads the ROM contents at the address specified by the HL register pair contents into the accumulator (HL≧$A000_H$ for 87CK38):

```
LD  A, (HL) ; A←ROM (HL)
```

Example 2 : Converts BCD to 7-segment code (common anode LED). When A = $05_H$, $92_H$ is output to port P5 after executing the following program:

```
        ADD  A,  TABLE-$-4 ; P5←ROM (TABLE + A)
        LD   (P5), (PC+A)
        JRS  T,  SNEXT
TABLE : DB   0C0H, 0F9H, 0A4H, 0B0H, 99H, 92H, 82H,
             0D8H, 80H, 98H
SNEXT :
```




*Notes : "$" is a header address of ADD instruction.*
*DB is a byte data difinition instruction.*

Example 3 : N-way multiple jump in accordance with the contents of accumulator (0≦A≦3):

```
SHLC  A       ; if A = 00H then PC←C234H
JP    (PC+A)    if A = 01H then PC←C378H
                if A = 02H then PC←DA37H
                if A = 03H then PC←E1B0H
DW    0C234H, 0C378H, 0DA37H, 0E1B0H
```

*Note : DW is a word data definition instruction.*

## 1.3 Program Counter (PC)

The program counter (PC) is a 16-bit register which indicates the program memory address where the instruction to be executed next is stored. After reset, the user defined reset vector stored in the vector table (addresses $FFFF_H$ and $FFFE_H$) is loaded into the PC ; therefore, program execution is possible from any desired address. For example, when $C0_H$ and $3E_H$ are stored at addresses $FFFF_H$ and $FFFE_H$, respectively, the execution starts from address $C03E_H$ after reset.

The TLCS-870 Series utilizes pipelined processing (instruction pre-fetch); therefore, the PC always indicates 2 addresses in advance. For example, while a 1-byte instruction stored at address $C123_H$ is being executed, the PC contains $C125_H$.

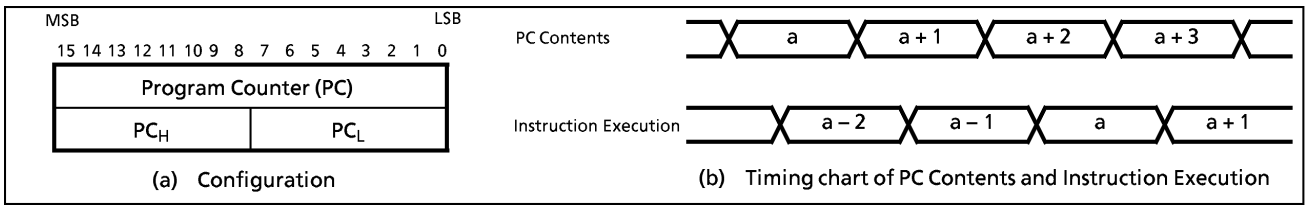



Figure 1-4. Program counter

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## 1.4 Data Memory (RAM)

The 87CH38/K38 have a 512 bytes (addresses $0040_H$ to $023F_H$) of data memory (static RAM). Figure 1-5 shows the data memory map.

Addresses $0000_H$ to $00FF_H$ are used as a direct addressing area to enhance instructions which utilize this addressing mode; therefore, addresses $0040_H$ to $00FF_H$ in the data memory can also be used for user flags or user counters. General-purpose register banks (8 registers × 16 banks) are also assigned to the 128 bytes of addresses $0040_H$ to $00BF_H$. Access as data memory is still possible even when being used for registers. For example, when the contents of the data memory at address $0040_H$ is read out, the contents of the accumulator in the bank 0 are also read out. The stack can be located anywhere within the data memory except the register bank area. The stack depth is limited only by the free data memory size. For more details on the stack, see section "1.7 Stack and Stack Pointer".

The TLCS-870 Series cannot execute programs placed in the data memory. When the program counter indicates a data memory address, a bus error occurs and an address-trap-reset applies. The $\overline{RESET}$ pin goes low during the address-trap-reset.

Example 1 : If bit 2 at data memory address $00C0_H$ is "1", $00_H$ is written to data memory at address $00E3_H$; otherwise, $FF_H$ is written to the data memory at address $00E3_H$:

```
        TEST  (00C0H).2   ; if (00C0H) 2 = 0 then jump
        JRS   T,SZERO
        CLR   (00E3H)     ; (00E3H) ← 00H
        JRS   T,SNEXT
SZERO : LD (00E3H), 0FFH  ; (00E3H) ← FFH
SNEXT :
```

Example 2 : Increments the contents of data memory at address $00F5_H$, and clears to $00_H$ when $10_H$ is exceeded:

```
 INC  (00F5H)       ; (00F5H) ← (00F5H) + 1
 AND  (00F5H), 0FH  ; (00F5H) ← (00F5H)∧0FH
```

The data memory contents become unstable when the power supply is turned on; therefore, the data memory should be initialized by an initialization routine. Note that the general-purpose registers are mapped in the RAM ; therefore, *do not clear RAM at the current bank addresses.*

Example : Clears RAM to "$00_H$" except the bank 0:

```
        LD  HL,  0048H ; Sets start address to HL register pair
        LD  A,   H     ; Sets initial data (00H) to A register
        LD  BC,  03F7H ; Sets number of byte to BC register pair
SRAMCLR : LD  (HL+),  A
        DEC BC
        JRS F,  SRAMCLR
```

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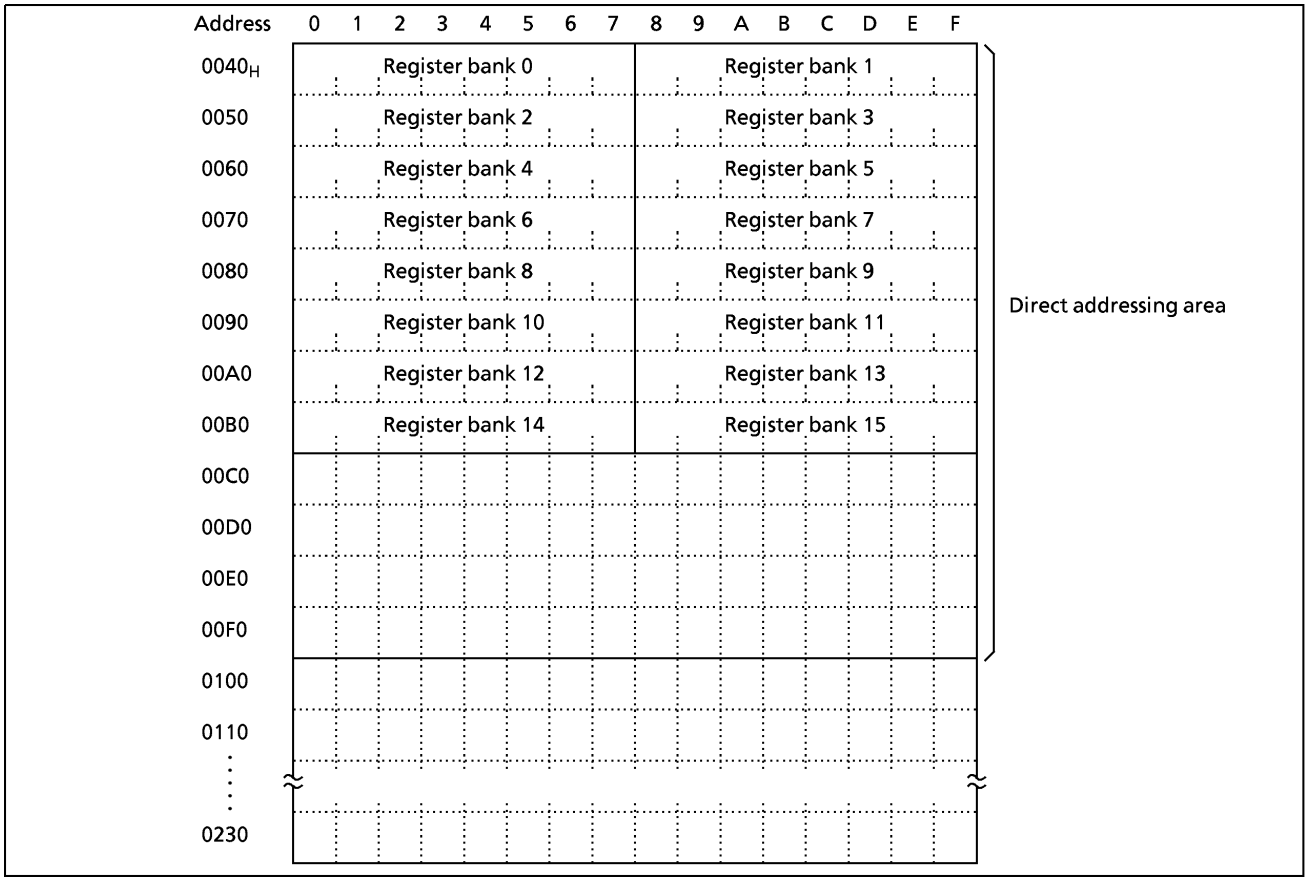


Figure 1-5. Data memory map

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## 1.5 General-purpose Register Banks

General-purpose registers are mapped into addresses $0040_H$ to $00BF_H$ in the data memory as shown in Figure 1-5. There are 16 register banks, and each bank contains eight 8-bit registers W, A, B, C, D, E, H, and L. Figure 1-6 shows the general-purpose register bank configuration.

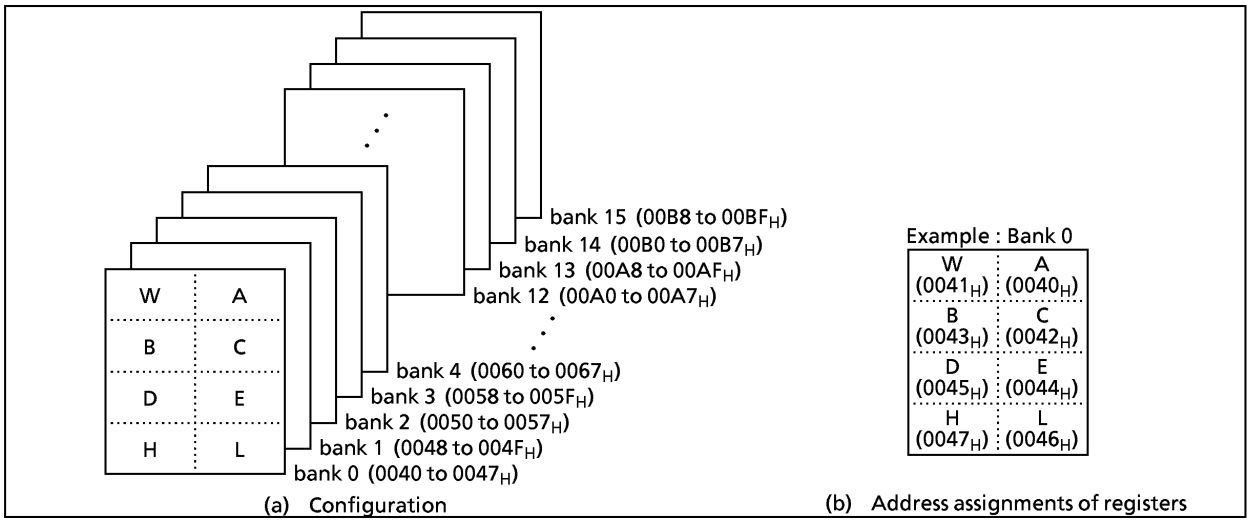


Figure 1-6. General-purpose register banks

In addition to access in 8-bit units, the registers can also be accessed in 16-bit units as the register pairs WA, BC, DE, and HL. Besides its function as a general-purpose register, the register also has the following functions:

(1) **A, WA**

The A register functions as an 8-bit accumulator and WA the register pair functions as a 16-bit accumulator (W is high byte and A is low byte). Registers other than A can also be used as accumulators for 8-bit operations.

```
Examples :  ① ADD  A, B       ;  Adds B contents to A contents and stores the result into A.
            ② SUB  WA, 1234H  ;  Subtracts 1234H from WA contents and stores the result into
                                 WA.
            ③ SUB  E, A       ;  Subtracts A contents from E contents, and stores the result
                                 into E.
```

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(2) **HL, DE**

The HL and DE specify a memory address. The HL register pair functions as data pointer (HL) / index register (HL + d) / base register (HL + C), and the DE register pair function as a data pointer (DE). The HL also has an auto-post- increment and auto-pre-decrement functions. This function simplifies multiple digit data processing, software LIFO (last-in first-out) processing, etc.

Example 1 :

| | | |
|---|---|---|
| ① LD A, (HL) | ; | Loads the memory contents at the address specified by HL into A. |
| ② LD A, (HL + 52H) | ; | Loads the memory contents at the address specified by the value obtained by adding $52_H$ to HL contents into A. |
| ③ LD A, (HL + C) | ; | Loads the memory contents at the address specified by the value obtained by adding the register C contents to HL contents into A. |
| ④ LD A, (HL + ) | ; | Loads the memory contents at the address specified by HL into A. Then increments HL. |
| ⑤ LD A, ( – HL) | ; | Decrements HL. Then loads the memory contents at the address specified by new HL into A. |

The TLCS-870 Series can transfer data directly memory to memory, and operate directly between memory data and memory data. This facilitates the programming of block processing.

Example 2 : Block transfer

```
        LD   B,   n-1      ; Sets (number of bytes to transfer) – 1 to B
        LD   HL,  DSTA     ; Sets destination address to HL
        LD   DE,  SRCA     ; Sets source address to DE
SLOOP : LD   (HL), (DE)    ; (HL) ← (DE)
        INC  HL            ; HL ← HL + 1
        INC  DE            ; DE ← DE + 1
        DEC  B             ; B ← B – 1
        JRS  F,  SLOOP     ; if B ≧ 0 then loop
```

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(3) **B, C, BC**

Registers B and C can be used as 8-bit buffers or counters, and the BC register pair can be used as a 16-bit buffer or counter. The C register functions as an offset register for register-offset index addressing (refer to example 1 ③ above) and as a divisor register for the division instruction [DIV gg, C].

Example 1 : Repeat processing

```
        LD B,  n          ; Sets n as the number of repetitions to B
SREPEAT : processing       (n + 1 times processing)
        DEC B
        JRS F,  SREPEAT
```

Example 2 : Unsigned integer division (16-bit ÷ 8-bit)

```
        DIV WA,  C        ; Divides the WA contents by the C contents, places the quotient in
                            Aand the remainder in W.
```

The general-purpose register banks are selected by the 4-bit register bank selector (RBS). During reset, the RBS is initialized to "0". The bank selected by the RBS is called the current bank.
Together with the flag, the RBS is assigned to address $003F_H$ in the SFR as the program status word (PSW). There are 3 instructions [LD RBS, n], [PUSH PSW] and [POP PSW] to access the PSW. The PSW can be also operated by the memory access instruction.

Example 1 : Incrementing the RBS

```
        INC  (003FH)      ; RBS ← RBS + 1
```

Example 2 : Reading the RBS

```
        LD   A, (003FH)   ; A ← PSW (A3-0 ← RBS, A7-4 ← Flags)
```

Highly efficient programming and high-speed task switching are possible by using bank changeover to save registers during interrupt and to transfer parameters during subroutine processing.
During interrupt, the PSW is automatically saved onto the stack. The bank used before the interrupt was accepted is restored automatically by executing an interrupt return instruction [RETI]/[RETN] ; therefore, there is no need for the RBS save / restore software processing.
The TLCS-870 Series supports a maximum of 15 interrupt sources. One bank is assigned to the main program, and one bank can be assigned to each source. Also, to increase the efficiency of data memory usage, assign the same bank to interrupt sources which are not nested.

Example : Saving /restoring registers during interrupt task using bank changeover.

```
PINT1 : LD RBS,  n ; RBS ← n (Bank changeover)
        Interrupt processing
        RETI       ; Maskable interrupt return (Bank restoring)
```

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## 1.6 Program Status Word (PSW)

The program status word (PSW) consists of a register bank selector (RBS) and four flags, and the PSW is assigned to address $003F_H$ in the SFR.

The RBS can be read and written using the memory access instruction (e. g. [LD A, (003FH)], [LD (003FH), A], however the flags can only be read. When writing to the PSW, the change specified by the instruction

is made without writing data to the flags. For example, when the instruction [LD (003FH), 05H] is executed, "5" is written to the RBS and the JF is set to "1", but the other flags are not affected.

[PUSH PSW] and [POP PSW] are PSW access instructions.

### 1.6.1 Register bank selector (RBS)

The register bank selector (RBS) is a 4-bit register used to select general-purpose register banks. For example, when RBS = 2, bank 2 is currently selected. During reset, the RBS is initialized to "0".

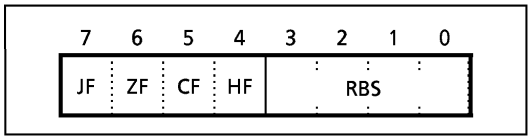


Figure 1-7. PSW (Flags, RBS) configuration

### 1.6.2 Flags

The flags are configured with the upper 4 bits : a zero flag, a carry flag, a half carry flag and a jump status flag. The flags are set or cleared under conditions specified by the instruction. These flags except the half carry flag are used as jump condition "cc" for conditional jump instructions [JR cc, $ + 2 + d]/[JRS cc, $ + 2 + d]. After reset, the jump status flag is initialized to "1", other flags are not affected.

(1) **Zero flag (ZF)**

The ZF is set to "1" if the operation result or the transfer data is $00_H$ (for 8-bit operations and data transfers) / $0000_H$ (for 16-bit operations); otherwise the ZF is cleared to "0".

During the bit manipulation instructions [SET, CLR, and CPL], the ZF is set to "1" if the contents of the specified bit is "0"; otherwise the ZF is cleared to "0".

This flag is set to "1" when the upper 8 bits of the product are $00_H$ during the multiplication instruction [MUL], and when $00_H$ for the remainder during the division instruction [DIV]; otherwise it is cleared to "0".

(2) **Carry flag (CF)**

The CF is set to "1" when a carry out of the MSB (most significant bit) of the result occurred during addition or when a borrow into the MSB of the result occurred during subtraction; otherwise the CF is cleared to "0". During division, this flag is set to "1" when the divisor is $00_H$ (divided by zero error), or when the quotient is $100_H$ or higher (quotient overflow error); otherwise it is cleared. The CF is also affected during the shift / rotate instructions [SHLC, SHRC, ROLC, and RORC]. The data shifted out from a register is set to the CF.

This flag is also a 1-bit register (a boolean accumulator) for the bit manipulation instructions. Set/clear/complement are possible with the CF manipulation instructions.

Example1 : Bit manipulation

```
LD  CF,  (0007H) . 5 ; (0001H)2 ← (0007H)5 ∨ (009AH)0
XOR CF,  (009AH) . 0
LD (0001H) . 2,  CF
```

Example2 : Arithmetic right shift

```
LD CF,  A . 7        ; A ← A / 2
RORC A
```

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(3) **Half carry flag (HF)**

The HF is set to "1" when a carry occurred between bits 3 and 4 of the operation result during an 8-bit addition, or when a borrow occurred from bit 4 into bit 3 of the result during an 8-bit subtraction; otherwise the HF is cleared to "0". This flag is useful in the decimal adjustment for BCD operations (adjustments using the [DAA r], or [DAS r] instructions).

Example : BCD operation

(The A becomes $47_H$ after executing the following program when A = $19_H$, B = $28_H$)

```
ADD  A,  B  ; A ← 41H, HF ← 1
DAA  A      ; A ← 41H + 06H = 47H (decimal-adjust)
```

(4) **Jump status flag (JF)**

Zero or carry information is set to the JF after operation (e. g. INC, ADD, CMP, TEST).

The JF provides the jump condition for conditional jump instructions [JRS T/F, $ + 2 + d], [JR T/F, $ + 2 + d] (T or F is a condition code). Jump is performed if the JF is "1" for a true condition (T), or the JF is "0" for a false condition (F).

The JF is set to "1" after executing the load/exchange/swap/nibble rotate/jump instruction, so that [JRS T, $ + 2 + d] and [JR T, $ + 2 + d] can be regarded as an unconditional jump instruction.

Example : Jump status flag and conditional jump instruction

```
INC  A
JRS  T,  SLABLE1 ; Jump when a carry is caused by the immediately
 ⋮                 preceding operation instruction.
LD   A, (HL)
JRS  T,  SLABLE2 ; JF is set to "1" by the immediately preceding
 ⋮                 instruction, making it an unconditional jump instruction.
```

Example : The accumulator and flags become as shown below after executing the following instructions when the WA register pair, the HL register pair, the data memory at address $00C5_H$, the carry flag and the half carry flag contents being "$219A_H$", "$00C5_H$", "$D7_H$", "1" and "0", respectively.

| Instruction | Acc. after execution | Flag after execution | | | |
|---|---|---|---|---|---|
| | | JF | ZF | CF | HF |
| ADDC A, (HL) | 72 | 1 | 0 | 1 | 1 |
| SUBB A, (HL) | C2 | 1 | 0 | 1 | 0 |
| CMP A, (HL) | 9A | 0 | 0 | 1 | 0 |
| AND A, (HL) | 92 | 0 | 0 | 1 | 0 |
| LD A, (HL) | D7 | 1 | 0 | 1 | 0 |
| ADD A, 66H | 00 | 1 | 1 | 1 | 1 |

| Instruction | Acc. after execution | Flag after execution | | | |
|---|---|---|---|---|---|
| | | JF | ZF | CF | HF |
| INC A | 9B | 0 | 0 | 1 | 0 |
| ROLC A | 35 | 1 | 0 | 1 | 0 |
| RORC A | CD | 0 | 0 | 0 | 0 |
| ADD WA, 0F508H | 16A2 | 1 | 0 | 1 | 0 |
| MUL W, A | 13DA | 0 | 0 | 1 | 0 |
| SET A.5 | BA | 1 | 1 | 1 | 0 |

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## 1.7 Stack and Stack Pointer

### 1.7.1 Stack

The stack provides the area in which the return address or status, etc. are saved before a jump is performed to the processing routine during the execution of a subroutine call instruction or the acceptance of an interrupt. On a subroutine call instruction [CALL a] / [CALLP n] / [CALLV n], the contents of the PC (the return address) is saved; on an interrupt acceptance, the contents of the PC and the PSW are saved (the PSW is pushed first, followed by $PC_H$ and $PC_L$). Therefore, a subroutine call occupies two bytes on the stack; an interrupt occupies three bytes.

When returning from the processing routine, executing a subroutine return instruction [RET] restores the contents to the PC from the stack; executing an interrupt return instruction [RETI] / [RETN] restores the contents to the PC and the PSW (the $PC_L$ is popped first, followed by $PC_H$ and PSW).

The stack can be located anywhere within the data memory space except the register bank area, therefore the stack depth is limited only by the free data memory size.

### 1.7.2 Stack pointer (SP)

The stack pointer (SP) is a 16-bit register containing the address of the next free locations on the stack.

The SP is post-decremented when a subroutine call or a push instruction is executed, or when an interrupt is accepted; and the SP is pre-incremented when a return or a pop instruction is executed. Figure 1-9 shows the stacking order.

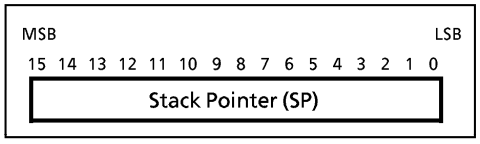


Figure 1-8. Stack pointer

The SP is not initialized hardware-wise but requires initialization by an initialize routine (sets the highest stack address). [LD SP, mn], [LD SP, gg] and [LD gg, SP] are the SP access instructions (mn ; 16-bit immediate data, gg ; register pair).

Example 1 : To initialize the SP

```
LD  SP,  043FH  ; SP←043F_H
```

Example 2 : To read the SP

```
LD  HL,  SP     ; HL←SP
```

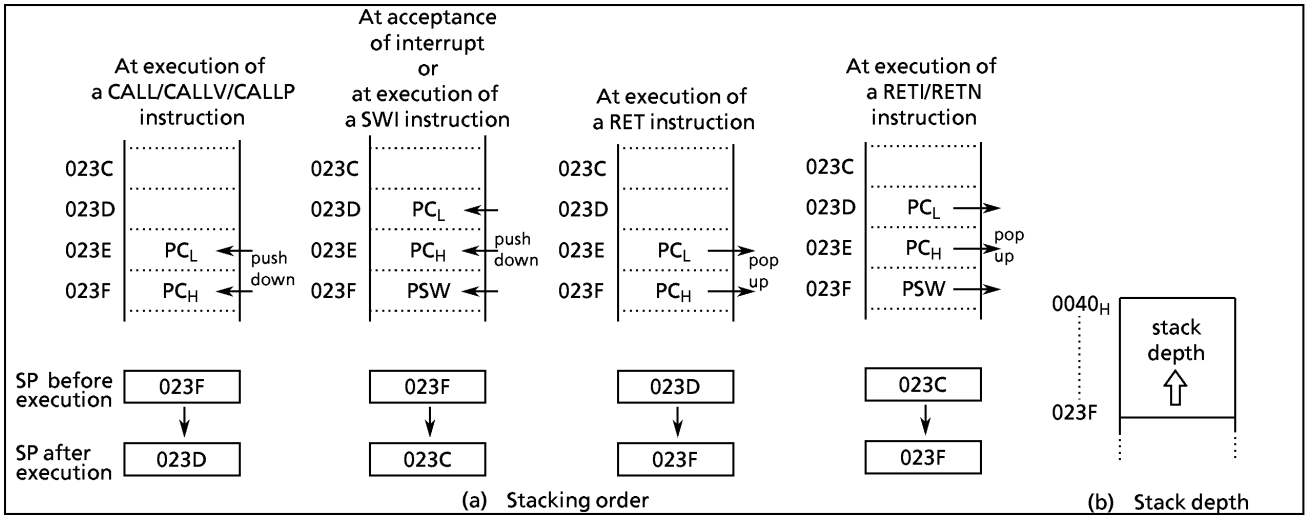


Figure 1-9. Stack

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## 1.8 System Clock Controller

The system clock controller consists of a clock generator, a timing generator, and a stand-by controller.

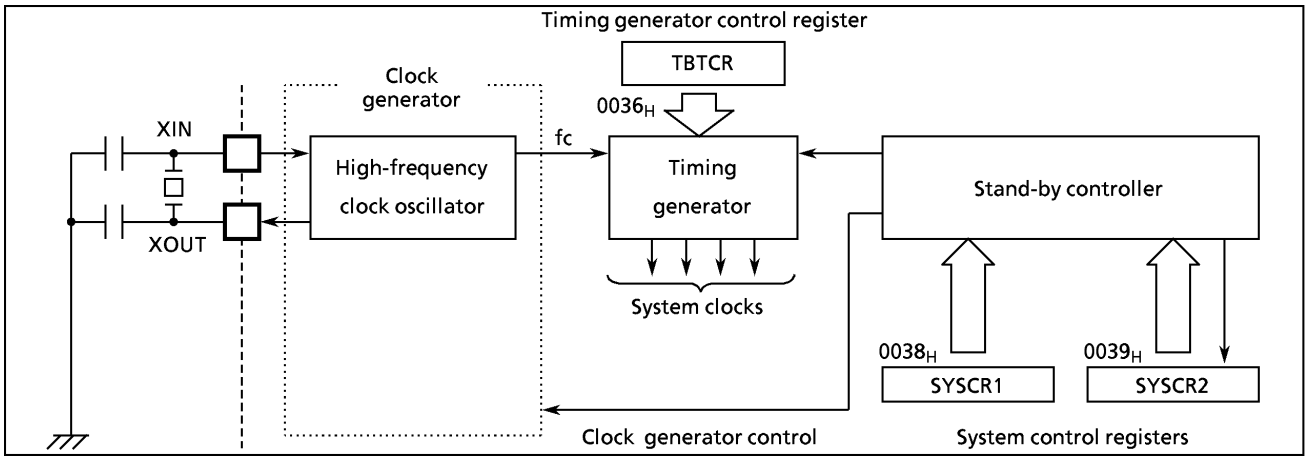


Figure 1-10. System clock controller

### 1.8.1 Clock generator

The clock generator generates the basic clock which provides the system clocks supplied to the CPU core and peripheral hardware. It contains a oscillation circuit for the high-frequency clock.

The high-frequency (fc) clock can be easily obtained by connecting a resonator between the XIN / XOUT pins, respectively. Clock input from an external oscillator is also possible. In this case, external clock is applied to the XIN pin with the XOUT pin not connected. The 87CH38/K38 is not provided an RC oscillation.

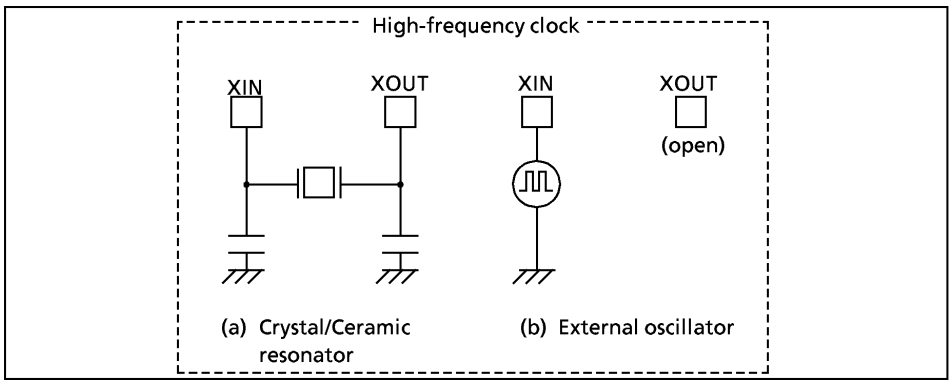


Figure 1-11. Examples of Resonator Connection

*Note : Accurate Adjustment of the Oscillation Frequency:*
*Although hardware to externally and directly monitor the basic clock pulse is not provided, the oscillation frequency can be adjusted by providing a program to output fixed frequency pulses to the port while disabling all interrupts and monitoring this pulse. With a system requiring adjustment of the oscillation frequency, the adjusting program must be created beforehand.*

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### 1.8.2 Timing generator

The timing generator generates from the basic clock the various system clocks supplied to the CPU core and peripheral hardware. The timing generator provides the following functions :

① Generation of main system clock
② Generation of source clocks for time base timer
③ Generation of source clocks for watchdog timer
④ Generation of internal source clocks for timer / counters TC1 to TC4
⑤ Generation of warm-up clocks for releasing STOP mode
⑥ Generation of a clock for releasing reset output

(1) Configuration of timing generator

The timing generator consists of a 21-stage divider with a divided-by-4 prescaler, a main system clock generator, and machine cycle counters, shown in Figure 1-12 as follows. During reset and upon releasing STOP mode, the divider is cleared to "0", however, the prescaler is not cleared.

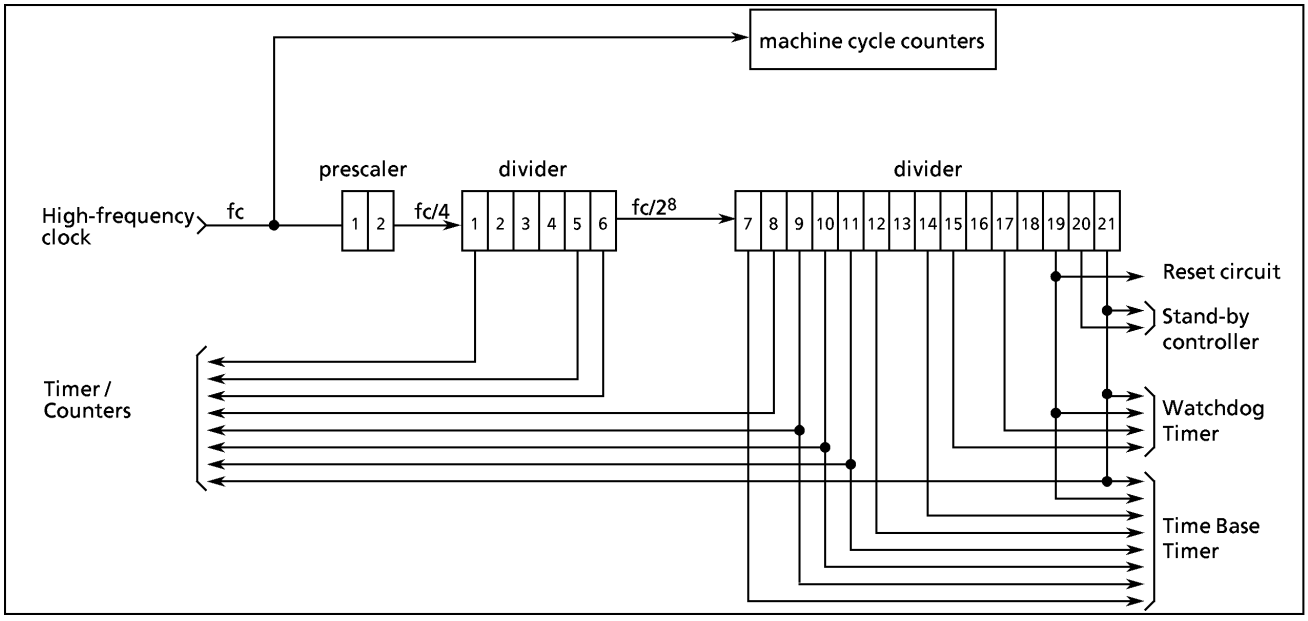


Figure 1-12. Configuration of timing generator

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(2) Machine cycle

Instruction execution and peripherals hardware operation are synchronized with the main system clock. The minimum instruction execution unit is called an "machine cycle". There are a total of 10 different types of instructions for the TLCS-870 Series: ranging from 1-cycle instructions which require one machine cycle for execution to 10-cycle instructions which require 10 machine cycles for execution.

A machine cycle consists of 4 states (S0 to S3), and each state consists of one main system clock.

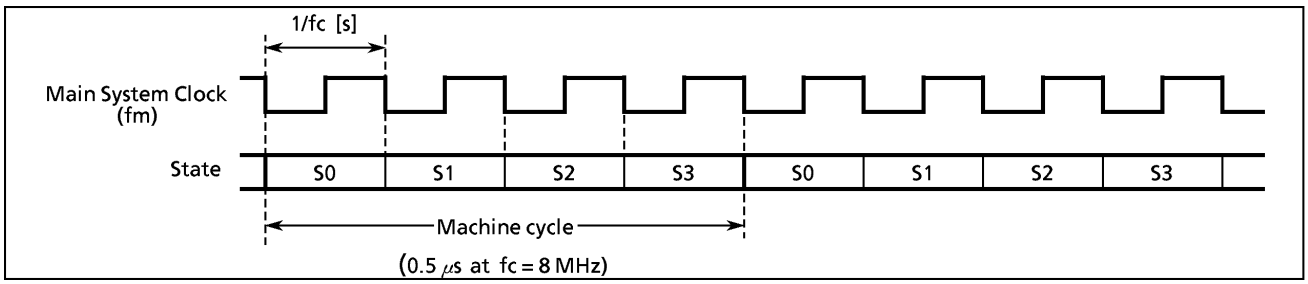


Figure 1-13. Machine cycle

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### 1.8.3 Stand-by controller

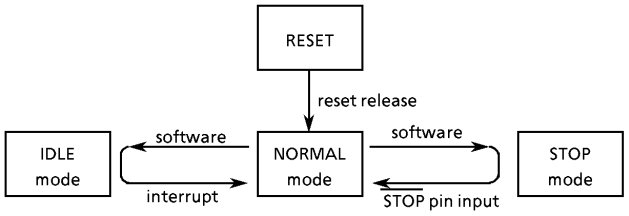
The stand-by controller starts and stops the oscillation circuit for the high-frequency clock. Operating modes are controlled by the system control registers (SYSCR1, SYSCR2).
Figure 1-14 shows the operating mode transition diagram and Figure 1-15 shows the system control registers.

(1) **Operating mode**

① NORMAL mode
In this mode, both the CPU core and on-chip peripherals operate.

② IDLE mode
In this mode, the internal oscillation circuit remains active. The CPU and the watchdog timer are halted; however, on-chip peripherals remain active. IDLE mode is started by setting IDLE bit in the system control register 2 (SYSCR2), and IDLE mode is released to NORMAL mode by an interrupt request from on-chip peripherals or external interrupt inputs. When IMF (interrupt master enable flag) is "1" (interrupt enable), the execution will resume upon acceptance of the interrupt, and the operation will return to normal after the interrupt service is completed. When IMF is "0" (interrupt disable), the execution will resume with the next instruction which follows IDLE mode start instruction.

③ STOP mode
In this mode, the internal oscillation circuit is turned off, causing all system operations to be halted. The internal status immediately prior to the halt is held with the lowest power consumption during this mode. The output status of all output ports can be set to either output hold or high-impedance under software control.
STOP mode is started by setting STOP bit in the system control register 1 (SYSCR1), and STOP mode is released by an input (either level-sensitive or edge-sensitive can be programmably selected) to the $\overline{\text{STOP}}$ pin. After the warming-up period is completed, the execution resumes with the next instruction which follows the STOP mode start instruction.




| Operating mode | Frequency: High-frequency | CPU core | On-chip Peripherals | Machine cycle time |
|---|---|---|---|---|
| RESET | turning on oscillation | reset | reset | 4/fc [s] |
| NORMAL | | operate | operate | |
| IDLE | | halt | | |
| STOP | turning off oscillation | | halt | — |

Figure 1-14. Operating mode transition diagram

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System Control Register 1

**SYSCR1** ($0038_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| STOP | RELM | RETM | OUTEN | WUT | | | |

(Initial value: 0000 00**)

| | | | |
|---|---|---|---|
| STOP | STOP mode start | 0 : CPU core and peripherals remain active<br>1 : CPU core and peripherals are halted (start STOP mode) | R/W |
| RELM | Release method for STOP mode | 0 : Edge-sensitive release<br>1 : Level-sensitive release | |
| RETM | Operating mode after STOP mode | 0 : Return to NORMAL mode<br>1 : Reserved | |
| OUTEN | Port output control during STOP mode | 0 : High-impedance<br>1 : Remain unchanged | |
| WUT | Warming-up time at releasing STOP mode | 00 : $3 \times 2^{19}$ / fc [s]<br>01 : $2^{19}$ / fc<br>1* : Reserved | |

*Note 1 : Always set RETM to "0" when transiting from NORMAL mode to STOP mode.*

*Note 2 : If 87CH38/K38 are moved to STOP mode while OUTEN = "0", internal inputs fix "0". Then there is a possibility to set interrupt of falling edge.*

*Note 3 : Bits 1 and 0 in SYSCR1 are read in as undefined data when a read instruction is executed.*

*Note 4 : fc ; high-frequency clock [Hz]*
** ; don't care*

*Note 5 : 87CH38/K38 returns to NORMAL mode without value of RETM, when STOP mode is retuned by input of $\overline{\text{RESET}}$ pin.*

System Control Register 2

**SYSCR2** ($0039_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| "1" | "0" | "0" | IDLE | | | | |

(Initial value: 1000 ****)

| | | | |
|---|---|---|---|
| IDLE | IDLE mode start | 0 : CPU and watchdog timer remain active<br>1 : CPU and watchdog timer are stopped (start IDLE mode) | R/W |

*Note 1 : A reset is applied ($\overline{\text{RESET}}$ pin output goes low) if bit 7 in SYSCR2 are cleared to "0".*

*Note 2 : Do not clear bit 7 in SYSCR2 to "0" , and do not set bits 6-5 in SYSCR2 to "1" .*

*Note 3 : * ; don't care*

*Note 4 : Bits 3 to 0 in SYSCR2 are always read in as "1" when a read instruction is executed.*

Figure 1-15. System control registers

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### 1.8.4 Operating mode control

(1) **STOP** mode

STOP mode is controlled by the system control register 1 (SYSCR1) and the $\overline{STOP}$ pin input. The $\overline{STOP}$ pin is also used both as a port P20 and an $\overline{INT5}$ (external interrupt input 5) pin. STOP mode is started by setting STOP (bit 7 in SYSCR1 ) to "1". During STOP mode, the following status is maintained.

① Oscillation is turned off, and all internal operations are halted.

② The data memory, registers and port output latches are all held in the status in effect before STOP mode was entered. The port output can be select either output hold or high-impedance by setting OUTEN ( bit 4 in SYSCR1).

③ The divider of the timing generator is cleared to "0".

④ The program counter holds the address of the instruction following the instruction which started STOP mode.

STOP mode includes a level-sensitive release mode and an edge-sensitive release mode, either of which can be selected with RELM (bit 6 in SYSCR1).

a. Level-sensitive release mode (RELM = 1)

In this mode, STOP mode is released by setting the $\overline{STOP}$ pin high. This mode is used for capacitor back-up when the main power supply is cut off and for long term battery back-up.

When the $\overline{STOP}$ pin input is high, executing an instruction which starts the STOP mode will not place in STOP mode but instead will immediately start the release sequence (warm-up). Thus, to start STOP mode in the level-sensitive release mode, it is necessary for the program to first confirm that the $\overline{STOP}$ pin input is low. The following method can be used for confirmation:

- Using an external interrupt input INT5 ($\overline{INT5}$ is a falling edge-sensitive input).

Example : Starting STOP mode with an INT5 interrupt.

```
PINT5 : TEST  (P2) . 0                ; To reject noise, STOP mode does not start
        JRS   F,  SINT5                 if port P20 is at high
        LD    (SYSCR1), 01000000B     ; Sets up the level-sensitive release mode.

        SET   (SYSCR1) . 7            ; Starts STOP mode
        LDW   (IL), 1110011111111111B ; IL12, 11 ← 0
                                        (clears interrupt latches)
SINT5 : RETI
```

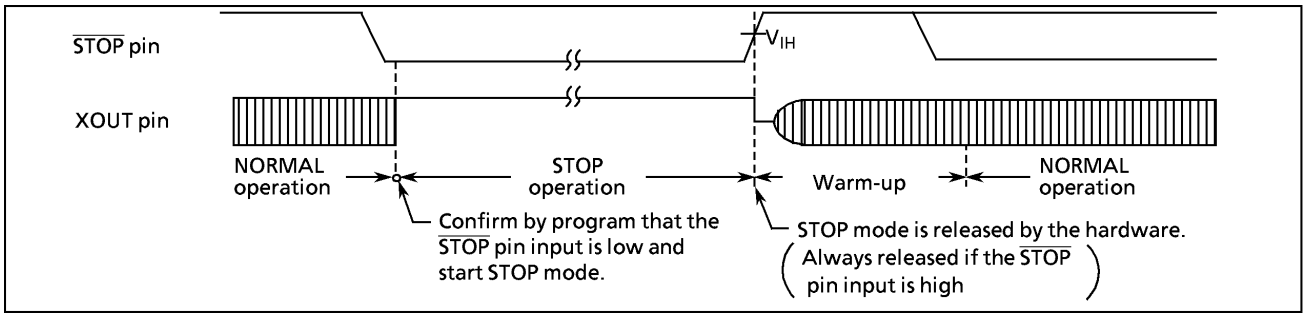


Figure 1-16. Level-sensitive release mode

*Note1 : After warming up is started, when $\overline{STOP}$ pin input is changed "L" level, STOP mode is not placed.*

*Note2 : When changing to the level-sensitive release mode from the edge-sensitive release mode, the release mode is not switched until a rising edge of the $\overline{STOP}$ pin input is detected.*

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b. Edge-sensitive release mode (RELM = 0)

In this mode, STOP mode is released by a rising edge of the $\overline{STOP}$ pin input. This is used in applications where a relatively short program is executed repeatedly at periodic intervals. This periodic signal (for example, a clock from a low-power consumption oscillator) is input to the $\overline{STOP}$ pin.

In the edge-sensitive release mode, STOP mode is started even when the $\overline{STOP}$ pin input is high.

Example : Starting STOP mode operation in the edge-sensitive release mode

```
LD   (SYSCR1),  00000000B     ; OUTEN ← 0 (specifies high-impedance)
DI                            ; IMF ← 0 (disables interrupt service)
SET  (SYSCR1). STOP           ; STOP ← 1 (activates stop mode)
LDW  (IL), 1110011111111111B ; IL12, 11 ← 0
                              (clears interrupt latches)
EI                            ; IMF ← 1 (enables interrupt service)
```

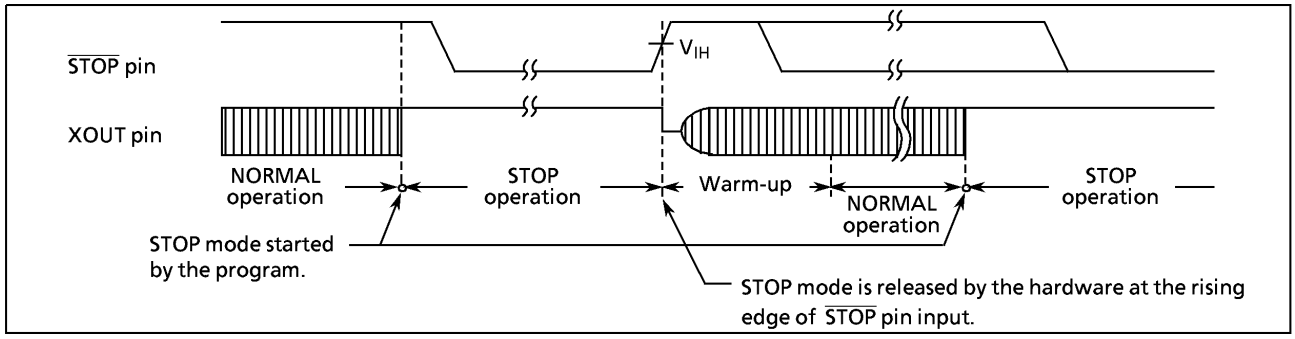


Figure 1-17. Edge-sensitive release mode

STOP mode is released by the following sequence:

① The high-frequency clock oscillator is turned on.

② A warming-up period is inserted to allow oscillation time to stabilize. During warm-up, all internal operations remain halted. Two different warming-up times can be selected with WUT ( bits 2 and 3 in SYSCR1) as determined by the resonator characteristics.

③ When the warming-up time has elapsed, normal operation resumes with the instruction following the STOP mode start instruction (e.g. [SET (SYSCR1). 7]). The start is made after the divider of the timing generator is cleared to "0".

Table 1-1. Warming-up time example

| WUT | At fc = 4.194304 MHz | At fc = 8 MHz |
|---|---|---|
| $3 \times 2^{19}$ / fc [s] | 375 [ms] | 196.6 [ms] |
| $2^{19}$ / fc | 125 | 65.5 |

*Note : The warming-up time is obtained by dividing the basic clock by the divider: therefore, the warming-up time may include a certain amount of error if there is any fluctuation of the oscillation frequency when STOP mode is released. Thus, the warming-up time must be considered an approximate value.*

STOP mode can also be released by setting the $\overline{RESET}$ pin low, which immediately performs the normal reset operation.

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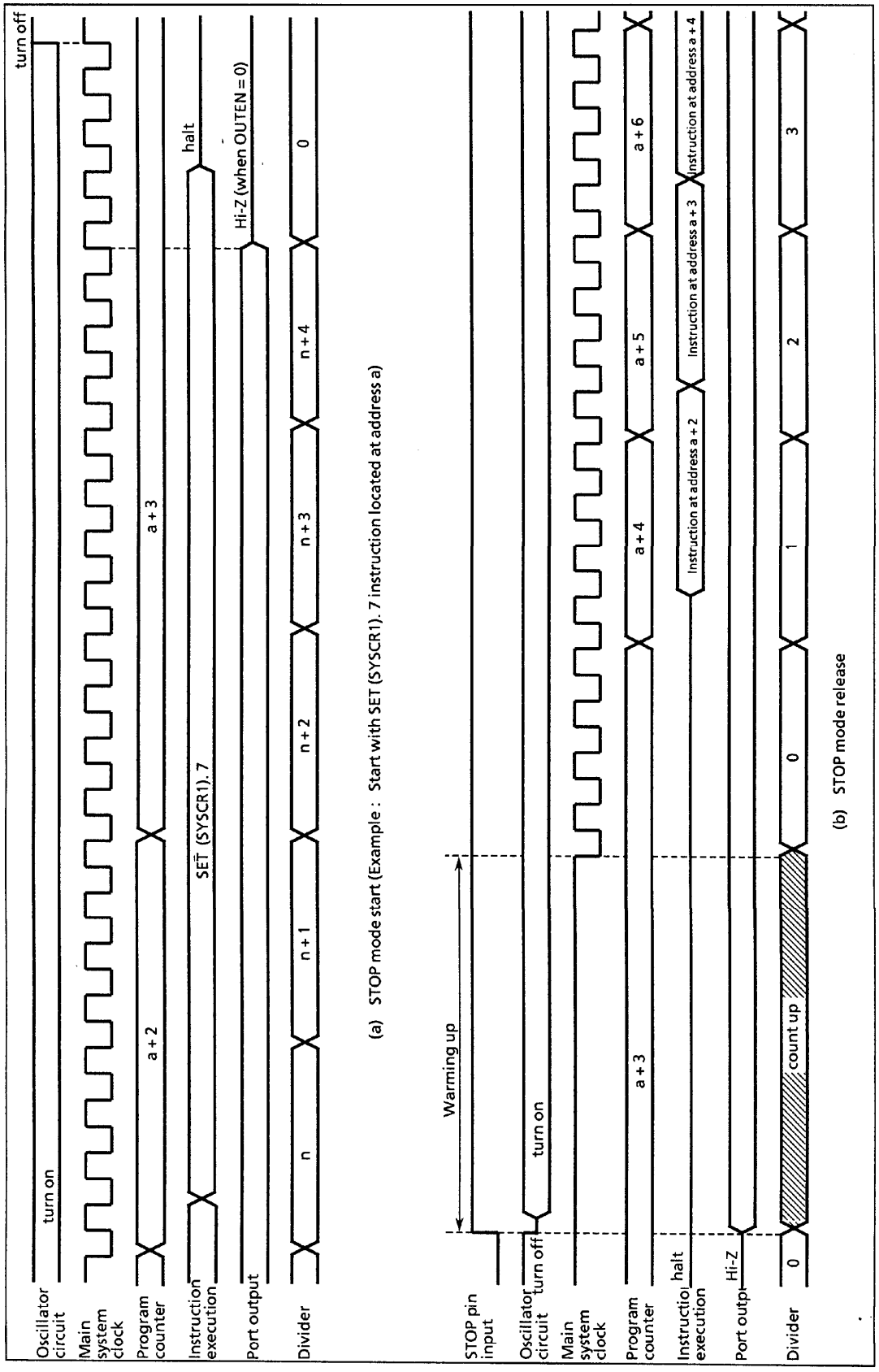


Figure 1-18. STOP mode start / release

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*Note : When STOP mode is released with a low hold voltage, the following cautions must be observed.*
*The power supply voltage must be at the operating voltage level before releasing the STOP mode. The $\overline{\text{RESET}}$ pin input must also be high, rising together with the power supply voltage. In this case, if an external time constant circuit has been connected, the $\overline{\text{RESET}}$ pin input voltage will increase at a slower rate than the power supply voltage. At this time, there is a danger that a reset may occur if input voltage level of the $\overline{\text{RESET}}$ pin drops below the non-inverting high-level input voltage (hysteresis input).*

(2) **IDLE** mode

IDLE mode is controlled by the system control register 2 and maskable interrupts. The following status is maintained during IDLE mode.

① Operation of the CPU and watchdog timer is halted. On-chip peripherals continue to operate.
② The data memory, CPU registers and port output latches are all held in the status in effect before IDLE mode was entered.
③ The program counter holds the address of the instruction following the instruction which started IDLE mode.

Example : Starting IDLE mode.

```
SET (SYSCR2) . 4 ; IDLE←1
```

IDLE mode includes a normal release mode and an interrupt release mode. Selection is made with the interrupt master enable flag (IMF). Releasing the IDLE mode returns to NORMAL mode.

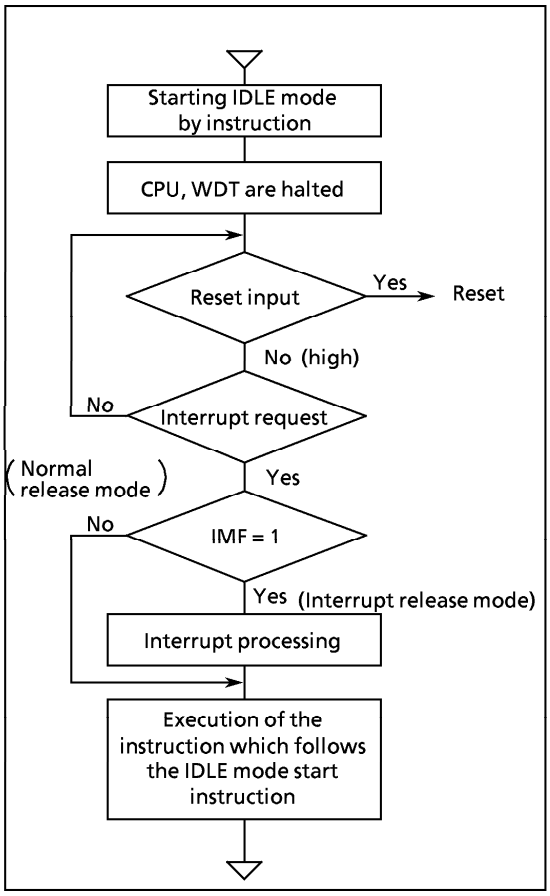


Figure 1-19. IDLE mode

a. Normal release mode (IMF = "0")

IDLE mode is released by any interrupt source enabled by the individual interrupt enable flag (EF). Execution resumes with the instruction following the IDLE mode start instruction (e.g. [SET (SYSCR2).4]). Normally, IL (Interrupt Latch) of interrupt source to release IDLE mode must be cleared by load instructions.

b. Interrupt release mode (IMF = "1")

IDLE mode is released and interrupt processing is started by any interrupt source enabled with the individual interrupt enable flag (EF). After the interrupt is processed, the execution resumes from the instruction following the instruction which started IDLE mode.

IDLE mode can also be released by setting the $\overline{\text{RESET}}$ pin low, which immediately performs the reset operation. After reset, the 87CM38/P38/S38 are placed in NORMAL mode.

*Note : When a watchdog timer interrupt is generated immediately before IDLE mode is started, the watchdog timer interrupt will be processed but IDLE mode will not be started.*

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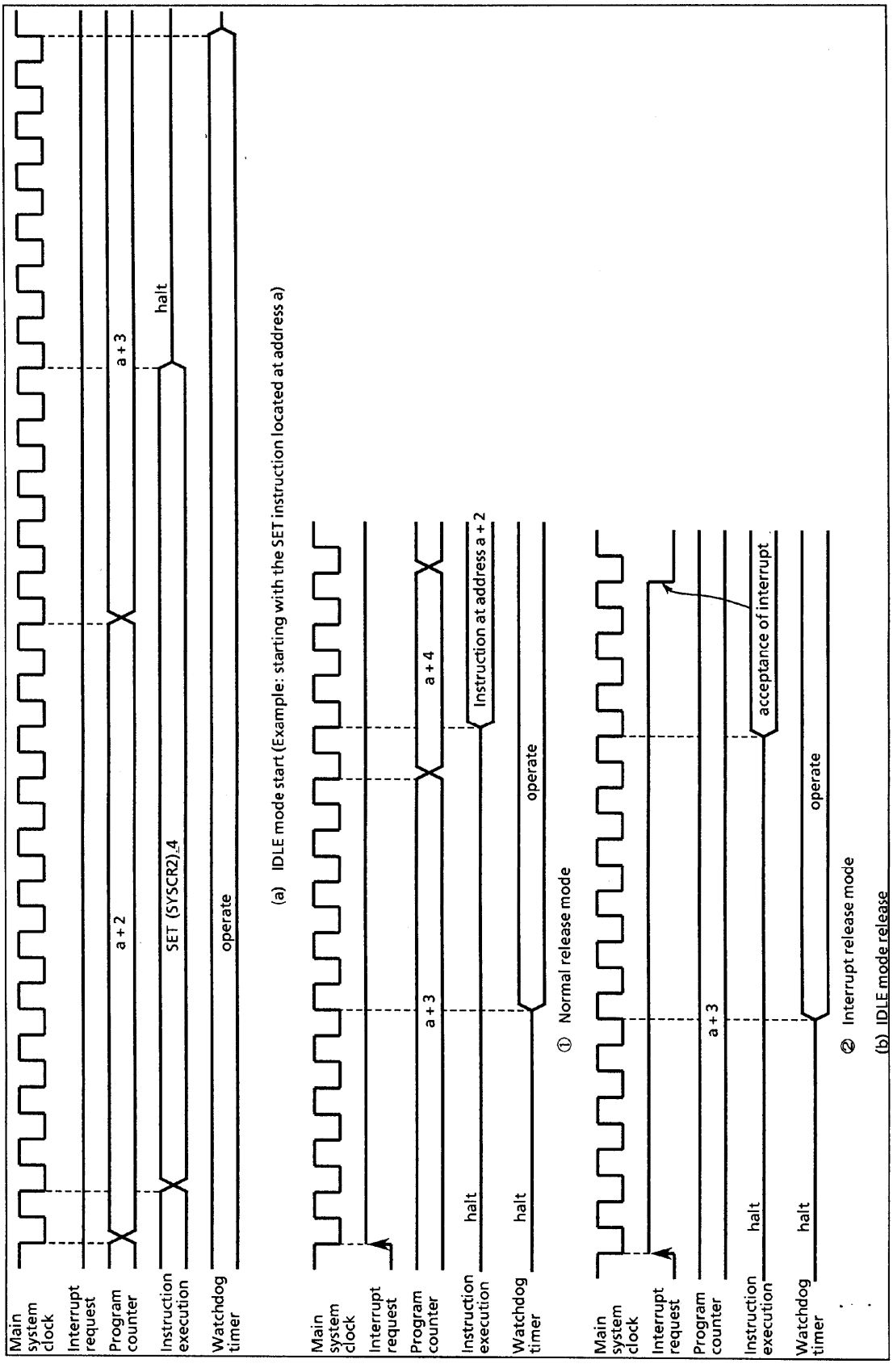


Figure 1-20. IDLE mode start/release

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## 1.9 Interrupt Controller

The 87CH38/K38 has a total of 14 interrupt sources: 5 externals and 9 internals. Nested interrupt control with priorities is also possible. Two of the internal sources are pseudo non-maskable interrupts; the remainder are all maskable interrupts.

Interrupt latches (IL) that hold the interrupt requests are provided for interrupt sources. Each interrupt vector is independent.

The interrupt latch is set to "1" when an interrupt request is generated and requests the CPU to accept the interrupt. The acceptance of maskable interrupts can be selectively enabled and disabled by the program using the interrupt master enable flag (IMF) and the individual interrupt enable flags (EF). When two or more interrupts are generated simultaneously, the interrupt is accepted in the highest priority order as determined by the hardware. Figure 1-21 shows the interrupt controller.

Table 1-2. Interrupt sources

| Interrupt Source | | Enable Condition | Interrupt Latch | Vector Table Address | Priority |
|---|---|---|---|---|---|
| Internal/ External | (Reset) | Non-Maskable | — | $FFFE_H$ | High 0 |
| Internal | INTSW (Software interrupt) | Pseudo non-maskable | — | $FFFC_H$ | 1 |
| Internal | INTWDT (Watchdog Timer interrupt) | | $IL_2$ | $FFFA_H$ | 2 |
| External | INT0 (External interrupt 0) | IMF = 1, INT0EN = 1 | $IL_3$ | $FFF8_H$ | 3 |
| Internal | INTTC1 (16-bit TC1 interrupt) | IMF · $EF_4$ = 1 | $IL_4$ | $FFF6_H$ | 4 |
| | reserved | IMF · $EF_5$ = 1 | $IL_5$ | $FFF4_H$ | 5 |
| Internal | INTTBT (Time Base Timer interrupt) | IMF · $EF_6$ = 1 | $IL_6$ | $FFF2_H$ | 6 |
| External | INT2 (External interrupt 2) | IMF · $EF_7$ = 1 | $IL_7$ | $FFF0_H$ | 7 |
| Internal | INTTC3 (8-bit TC3 interrupt) | IMF · $EF_8$ = 1 | $IL_8$ | $FFEE_H$ | 8 |
| Internal | INTSBI (Serial bus Interface interrupt) | IMF · $EF_9$ = 1 | $IL_9$ | $FFEC_H$ | 9 |
| Internal | INTTC4 (8-bit TC4 interrupt) | IMF · $EF_{10}$ = 1 | $IL_{10}$ | $FFEA_H$ | 10 |
| External | INT3 (External interrupt 3, Remote control receive interrupt) | IMF · $EF_{11}$ = 1 | $IL_{11}$ | $FFE8_H$ | 11 |
| External | INT4 (External interrupt 4) | IMF · $EF_{12}$ = 1 | $IL_{12}$ | $FFE6_H$ | 12 |
| Internal | INTOSD (OSD interrupt) | IMF · $EF_{13}$ = 1 | $IL_{13}$ | $FFE4_H$ | 13 |
| Internal | INTTC2 (16-bit TC2 interrupt) | IMF · $EF_{14}$ = 1 | $IL_{14}$ | $FFE2_H$ | 14 |
| External | INT5 (External interrupt 5) | IMF · $EF_{15}$ = 1 | $IL_{15}$ | $FFE0_H$ | Low 15 |

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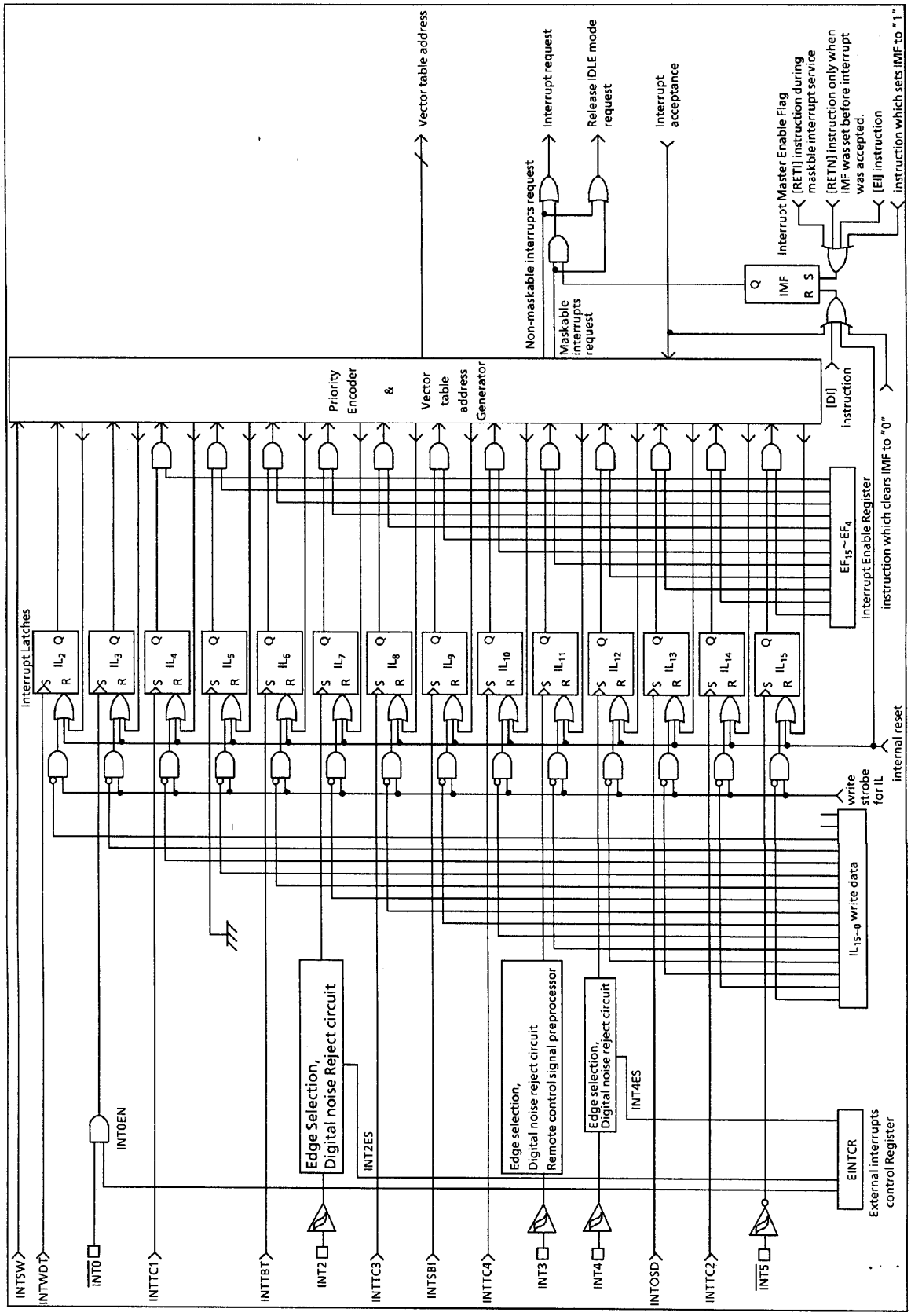


Figure 1-21. Interrupt controller block diagram

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(1) **Interrupt latches** ($IL_{15\ to\ 2}$)

Interrupt latches are provided for each source, except for a software interrupt. The latch is set to "1" when an interrupt request is generated, and requests the CPU to accept the interrupt. The latch is cleared to "0" just after the interrupt is accepted. All interrupt latches except INT3 are initialized to "0" during reset. The interrupt latch of INT3 is unstable during reset.

The interrupt latches are assigned to addresses $003C_H$ and $003D_H$ in the SFR. Each latch can be cleared to "0" individually by an instruction; however, the read-modify-write instruction such as bit manipulation or operation instructions cannot be used (Do not clear the $IL_2$ for a watchdog timer interrupt to "0"). Thus, interrupt requests can be cancelled and initialized by the program. Note that interrupt latches cannot be set to "1" by any instruction.

The contents of interrupt latches can be read out by an instruction. Therefore, testing interrupt requests by software is possible.

Example 1 : Clears interrupt latches

```
LDW  (IL),  1110101010111111B  ; IL12, IL10, IL8, IL6←0
```

Example 2 : Reads interrupt latches

```
LD   WA,  (IL)                 ; W←ILH, A←ILL
```

Example 3 : Tests an interrupt latch

```
TEST (ILH).4                   ;  if IL12 = 1 then jump
JR   F, SSET
```

(2) **Interrupt enable register** (EIR)

The interrupt enable registers (EIR) enable and disable the acceptance of interrupts, except for the pseudo non-maskable interrupts (software and watchdog timer interrupts). Pseudo non-maskable interrupts are accepted regardless of the contents of the EIR; however, the pseudo non-maskable interrupts cannot be nested more than once at the same time. For example, the watchdog timer interrupt is not accepted during the software interrupt service.

The EIR consists of an interrupt master enable flag (IMF) and individual interrupt enable flags (EF). These registers are assigned to addresses $003A_H$ and $003B_H$ in the SFR, and can be read and written by an instruction (including read-modify-write instructions such as bit manipulation instructions).

① **Interrupt master enable flag** (IMF)

The interrupt master enable flag (IMF) enables and disables the acceptance of all interrupts, except for pseudo non-maskable interrupts. Clearing this flag to "0" disables the acceptance of all maskable interrupts. Setting to "1" enables the acceptance of interrupts.

When an interrupt is accepted, this flag is cleared to "0" to temporarily disable the acceptance of maskable interrupts. After execution of the interrupt service program, this flag is set to "1" by the maskable interrupt return instruction [RETI] to again enable the acceptance of interrupts. If an interrupt request has already been occurred, interrupt service starts immediately after execution of the [RETI] instruction.

Pseudo non-maskable interrupts are returned by the [RETN] instruction. In this case, the IMF is set to "1" only when pseudo non-maskable interrupt service is started with interrupt acceptance enabled (IMF = 1). Note that the IMF remains "0" when cleared by the interrupt service program.

The IMF is assigned to bit 0 at address $003A_H$ in the SFR, and can be read and written by an instruction. The IMF is normally set and cleared by the [EI] and [DI] instructions, and the IMF is initialized to "0" during reset.

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② **Individual interrupt enable flags** ($EF_{15}$ to $EF_4$)

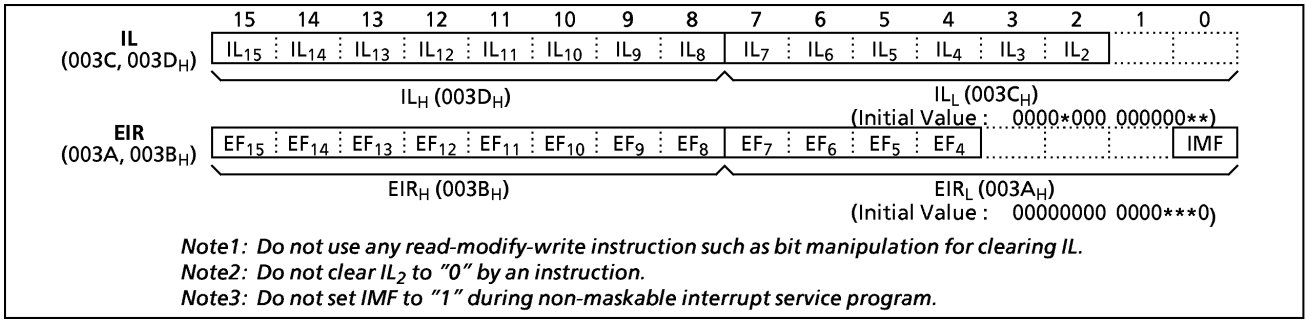
These flags enable and disable the acceptance of individual maskable interrupts. Setting the corresponding bit of an individual interrupt enable flag to "1" enables acceptance of an interrupt, setting the bit to "0" disables acceptance.

Example 1 : Sets EF for individual interrupt enable, and sets IMF to "1".

```
LDW (EIR), 1110100000000001B ; EF15 to EF13, EF11, IMF←1
```

Example 2 : Sets an individual interrupt enable flag to "1".

```
SET (EIRH).4                 ; EF12←1
```




*Note1: Do not use any read-modify-write instruction such as bit manipulation for clearing IL.*
*Note2: Do not clear $IL_2$ to "0" by an instruction.*
*Note3: Do not set IMF to "1" during non-maskable interrupt service program.*

Figure 1-22. Interrupt latch (IL) and interrupt enable register (EIR)

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### 1.9.1 Interrupt sequence

An interrupt request is held until the interrupt is accepted or the interrupt latch is cleared to "0" by a reset or an instruction. Interrupt acceptance sequence requires 8 machine cycles (4 $\mu$s at fc = 8 MHz in NORMAL mode) after the completion of the current instruction execution. The interrupt service task terminates upon execution of an interrupt return instruction [RETI] (for maskable interrupts) or [RETN] (for pseudo non-maskable interrupts).

(1) Interrupt acceptance processing is as follows:

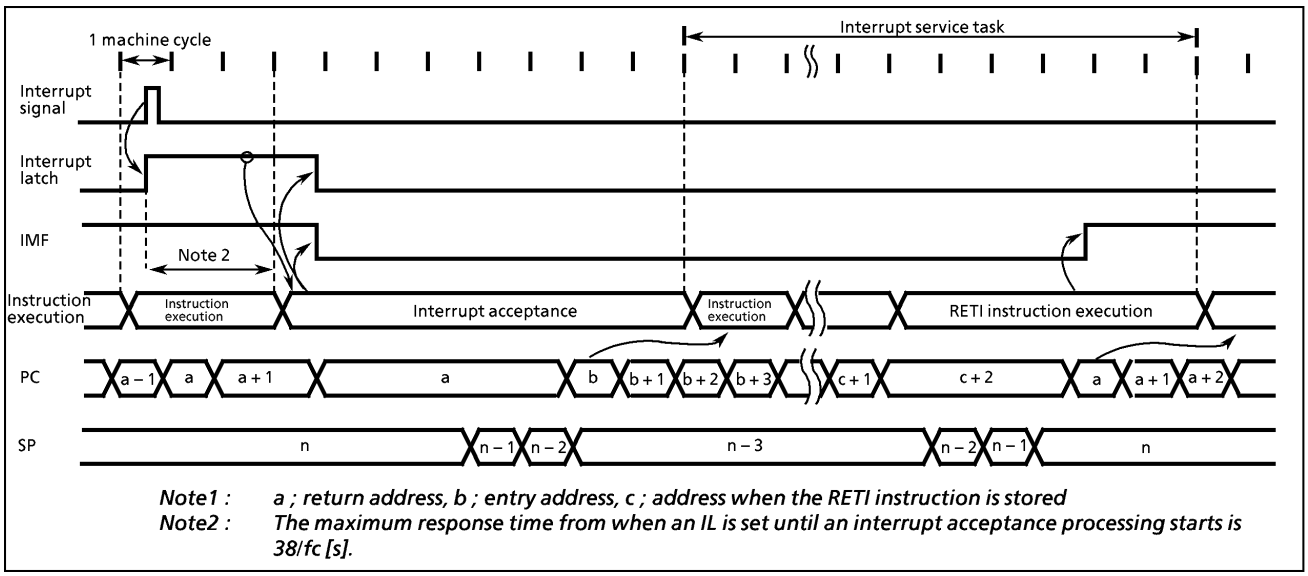
① The interrupt master enable flag (IMF) is cleared to "0" to temporarily disable the acceptance of any following maskable interrupts. When a non-maskable interrupt is accepted, the acceptance of any following interrupts is temporarily disabled.

② The interrupt latch (IL) for the interrupt source accepted is cleared to "0".

③ The contents of the program counter (return address) and the program status word are saved (pushed) onto the stack. The stack pointer is decremented 3 times.

④ The entry address of the interrupt service program is read from the vector table address, and the entry address is loaded to the program counter.

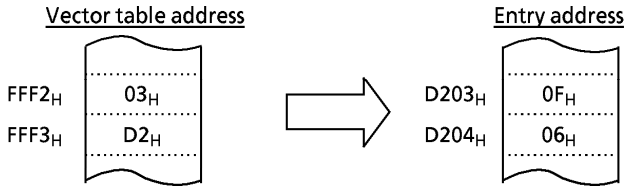
⑤ The instruction stored at the entry address of the interrupt service program is executed.




*Note1 : a ; return address, b ; entry address, c ; address when the RETI instruction is stored*
*Note2 : The maximum response time from when an IL is set until an interrupt acceptance processing starts is 38/fc [s].*

Figure 1-23. Timing chart of interrupt acceptance and interrupt return instruction

Example : Correspondence between vector table address for INTTBT and the entry address of the interrupt service program.




A maskable interrupt is not accepted until the IMF is set to "1" even if a maskable interrupt of higher priority than that of the current interrupt being serviced.
When nested interrupt service is necessary, the IMF is set to "1" in the interrupt service program. In this case, acceptable interrupt sources are selectively enabled by the individual interrupt enable flags.

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(2) Saving / restoring general-purpose register

During interrupt acceptance processing, the program counter and the program status word are automatically saved on the stack, but not the accumulator and other registers. These registers are saved by the program if necessary. Also, when nesting multiple interrupt services, it is necessary to avoid using the same data memory area for saving registers.

The following method is used to save/restore the general-purpose registers:

① General-purpose register save/restore by register bank changeover:

General-purpose registers can be saved at high-speed by switching to a register bank that is not in use. Normally, bank 0 is used for the main task and banks 1 to 15 are assigned to interrupt service tasks. To increase the efficiency of data memory utilization, the same bank is assigned for interrupt sources which are not nested.

The switched bank is automatically restored by executing an interrupt return instruction [RETI] or [RETN]. Therefore, it is not necessary for a program to save the RBS.

Example : Register Bank Changeover

```
PINTxx : LD   RBS,   n  ; Switches to bank n (1 μs at 8 MHz)
         Interrupt processing
         RETI           ; Restores bank and Returns
```

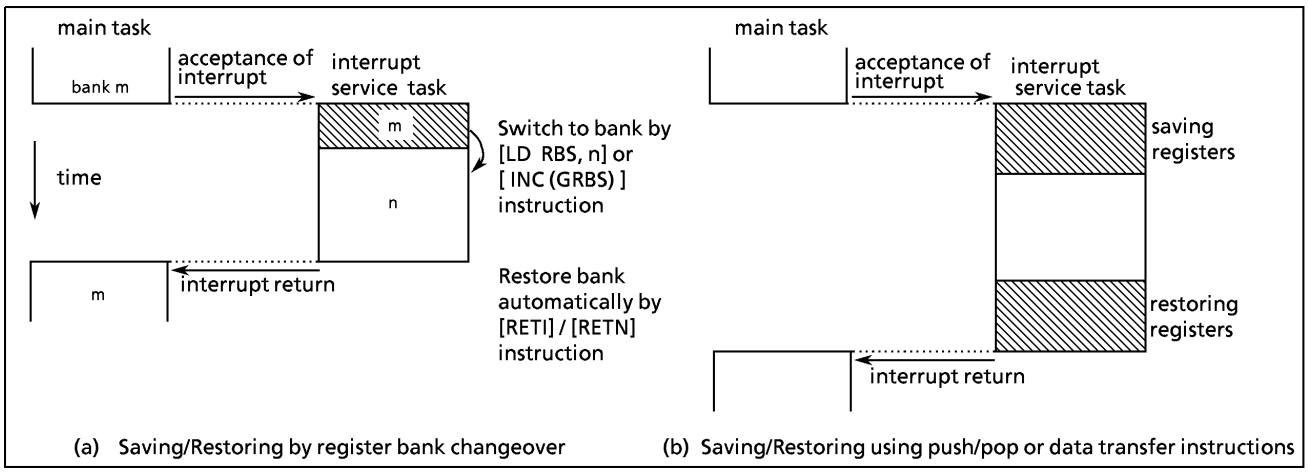


Figure 1-24. Saving/restoring general-purpose registers

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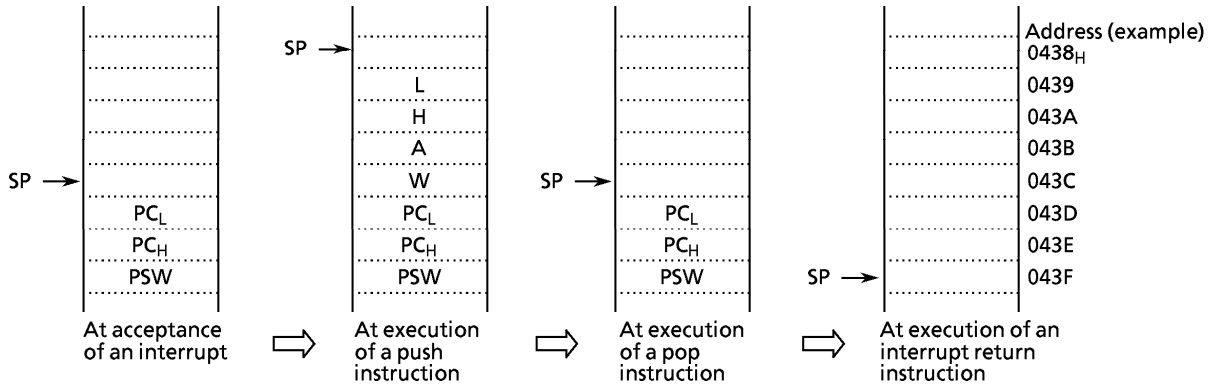
② General-purpose register save / restore using push and pop instructions:
To save only a specific register, and when the same interrupt source occurs more than once, the general-purpose registers can be saved/restored using push/pop instructions.
Example : Register save using push and pop instructions

```
PINTxx : PUSH  WA ; Save WA register pair
         PUSH  HL ; Save HL register pair
           interrupt processing
         POP   HL ; Restore HL register pair
         POP   WA ; Restore WA register pair
         RETI     ; Return
```




③ General-purpose registers save/restore using data transfer instructions:
Data transfer instructions can be used to save only a specific general-purpose register during processing of a single interrupt.

Example : Saving/restoring a register using data transfer instructions

```
PINTxx : LD  (GSAVA), A ; Save A register
           interrupt processing
         LD  A, (GSAVA) ; Restore A register
         RETI           ; Return
```

The interrupt return instructions [RETI] / [RETN] perform the following operations.

| [RETI] Maskable interrupt return | [RETN] Non-maskable interrupt return |
|---|---|
| ① The contents of the program counter and the program status word are restored from the stack. | ① The contents of the program counter and program status word are restored from the stack. |
| ② The stack pointer is incremented 3 times. | ② The stack pointer is incremented 3 times. |
| ③ The interrupt master enable flag is set to "1". | ③ The interrupt master enable flag is set to "1" only when a non-maskable interrupt is accepted in interrupt enable status. However, the interrupt master enable flag remains at "0" when so clear by an interrupt service program. |

Interrupt requests are sampled during the final cycle of the instruction being executed. Thus, the next interrupt can be accepted immediately after the interrupt return instruction is executed.

*Note : When the interrupt processing time is longer than the interrupt request generation time, the interrupt service task is performed but not the main task.*

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### 1.9.2 Software interrupt (INTSW)

Executing the [SWI] instruction generates a software interrupt and immediately starts interrupt processing (INTSW is highest prioritized interrupt). However, if processing of a non-maskable interrupt is already underway, executing the SWI instruction will not generate a software interrupt but will result in the same operation as the [NOP] instruction. Thus, the [SWI] instruction behaves like the [NOP] instruction.

*Note : At the development tool, if processing of a non-maskable interrupt is already underway, executing the SWI instruction will generate a software interrupt as a software brake.*

Use the [SWI] instruction only for detection of the address error or for debugging.

① Address error detection

$FF_H$ is read if for some cause such as noise the CPU attempts to fetch an instruction from a non-existent memory address. Code $FF_H$ is the SWI instruction, so a software interrupt is generated and an address error is detected. The address error detection range can be further expanded by writing $FF_H$ to unused areas of the program memory. the address trap reset is generated in case that an instruction is fetched from RAM or SFR areas.

*Note : The fetch data from addresses $1080_H$ to $10FF_H$ (test ROM area) is not "$FF_H$".*

② Debugging

Debugging efficiency can be increased by placing the SWI instruction at the software break point setting address.

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### 1.9.3 External interrupts

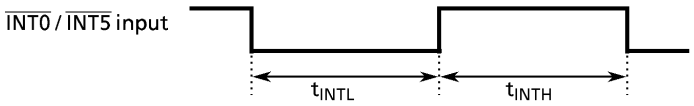
The 87CH38/K38 have three external interrupt inputs ($\overline{INT0}$, INT2, INT3, INT4, and $\overline{INT5}$). Three of these are equipped with digital noise rejection circuits (pulse inputs of less than a certain time are eliminated as noise). Edge selection is also possible with INT2, INT3 and INT4.

The $\overline{INT0}$ / P50 pin can be configured as either an external interrupt input pin or an input / output port, and is configured as an input port during reset.

Edge selection of INT2, INT4 pin input is performed by the external interrupt control register (EINTCR). Edge selection and noise rejection control for INT3 pin input are performed by the Remote-control signal processor control registers. (refer to the selection of the Remote-control signal processor.) When INT0EN = 0, the $IL_3$ will not be set even if the falling edge of $\overline{INT0}$ pin input is detected.

| Source | Pin | Secondary function pin | Enable conditions | Edge | Digital noise reject |
|---|---|---|---|---|---|
| INT0 | $\overline{INT0}$ | P50/TC2/$\overline{PWM8}$ | IMF = 1, INT0EN = 1 | falling edge | — (hysteresis input) |
| INT2 | INT2 | P53/TC1 | IMF · $EF_7$ = 1 | falling edge or rising edge | Pulses of less than 7/fc [s] are eliminated as noise. Pulses equal to or more than 24/fc [s] are regarded as signals. |
| INT3 | INT3 | P30/RXIN | IMF · $EF_{11}$ = 1 | falling edge, rising edge or falling / rising edge | Refer to the selection of the Remote control signal preprocessor. |
| INT4 | INT4 | P32 | IMF · $EF_{12}$ = 1 | falling edge or rising edge | Pulses of less than 7/fc [s] are eliminated as noise. Pulses equal to or more than 24/fc [s] are regarded as signals. |
| INT5 | $\overline{INT5}$ | P20/$\overline{STOP}$ | IMF · $EF_{15}$ = 1 | falling edge | — (hysteresis input) |

*Note 1 : The pulse width (both "H" and "L" level) for input to the $\overline{INT0}$ and $\overline{INT5}$ pin must be over 1 machine cycle.*




$t_{INTL}$, $t_{INTH}$ > tcyc (*Note : tcyc = 4/fc [s]*)

*Note 2 : The noise reject function is also affected for timer / counter input (TC1 pin).*

*Note 3 : If a noiseless signal is input to the external interrupt pin in the NORMAL 1 or IDLE 1 mode, the maximum time from the edge of input signal until the IL is set is as follows :*

① INT2,INT4 pins 25/fc [s]

② INT3 pin Refer to the section of the Remote control preprocessor.

*Note 4 : When high-impedance is specified for port output in stop mode, port input is forcibly fixed to low level internally. Thus, interrupt latches of external interrupt inputs except P20 ($\overline{INT5}$/$\overline{STOP}$) which are also used as ports may be set to "1". To specify high-impedance for port output in stop mode, first disable interrupt service (IMF = 0) , activate stop mode. After releasing stop mode, clear interrupt latches using load instruction, then, enable interrupt service.*

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**EINTCR** ($0037_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| "0" | INT0EN | (TC4ES) | INT4 ES | (TC3ES) | INT2 ES | "0" | "0" |

(Initial value : *000 000*)

| | | | |
|---|---|---|---|
| INT0EN | P50 / $\overline{INT0}$ pin configuration | 0 : P50 input / output<br>1 : $\overline{INT0}$ pin (The output latch should be set to "1".) | write only |
| INT4 ES<br>INT2 ES | INT4, INT2 edge select | 0 : Rising edge<br>1 : Falling edge | |

*Note 1 :* * ; *don't care*

*Note 2 :* *Do not change EINTCR when IMF = 1. After changing EINTCR, interrupt latches of external interrupt inputs must be cleared to "0" using load instruction.*

*Note 3 :* *In order to change of external interrupt input by rewriting the contents of INT2ES and INT4ES during NORMAL mode, clear interrupt latches of external interrupt inputs (INT2 and INT4) after 8 machine cycles from the time of rewriting.*

*Note 4 :* *In order to change an edge of timer counter input by rewriting the contents of INT2ES and INT4ES during NORMAL mode, rewrite the contents after timer counter is stopped (TC*s = 0), that is, interrupt disable state. Then, clear interrupt laches of external interrupt inputs (INT2 and INT4) after 8 machine cycles from the time of rewriting to change to interrupt enable state. Finally, start timer counter.*

Example : When changing TC1 pin inputs edge in external trigger timer mode from rising edge to falling edge.

```
                LD (TC1CR),01001000B    ; TC1S ← 00 (stops TC1)
                DI                      ; IMF ← 0 (disables interrupt service)
                LD (EINTCR),00000100B   ; INT2ES ← 1 (change edge selection)
                NOP
  8 machine     ~
     cycles     NOP
                LD (ILL),01111111B      ; IL7 ← 0 (clears interrupt latch)
                EI                      ; IMF ← 1 (enables interrupt service)
                LD (TC1CR),01111000B    ; TC1S ← 11 (starts TC1)
```

*Note 5 :* *Always write "0" to bit7, 6, 1, 0 in EINTCR.*

*Note 6 :* *When high-impedance is specified for port output in stop mode, port input is forcibly fixed to low level internally. Thus, interrupt latches of external interrupt inputs except P20 ($\overline{INT5}$/$\overline{STOP}$) which are also used as ports may be set to "1". To specify high-impedance for port output in stop mode, first disable interrupt service (IMF = 0) , activate stop mode. After releasing stop mode, clear interrupt latches using load instruction, then, enable interrupt service.*

*Example : Activating stop mode (CM38/P38/S38):*

```
LD    (SYSCR1),01000000B      ; OUTEN ← 0 (specifies high-impedance)
DI                            ; IMF ← 0 (disables interrupt service)
SET   (SYSCR1).STOP           ; STOP ← 1 (activates stop mode)
LDW   (IL),1110011101010111B  ; IL12,11,7,5,3 ← 0 (clears interrupt latches)
EI                            ; IMF ← 1 (enables interrupt service)
```

Figure 1-25. External interrupt control register

## 1.10 Watchdog Timer (WDT)

The watchdog timer rapidly detects the CPU malfunction such as endless looping caused by noise or the like, and resumes the CPU to the normal state.

The watchdog timer signal for detecting malfunction can be selected either as a reset output or a non-maskable interrupt request. However, selection is possible only once after reset. At first, the reset output is selected.

When the watchdog timer is not being used for malfunction detection, it can be used as a timer to generate an interrupt at fixed intervals.

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### 1.10.1 Watchdog timer configuration

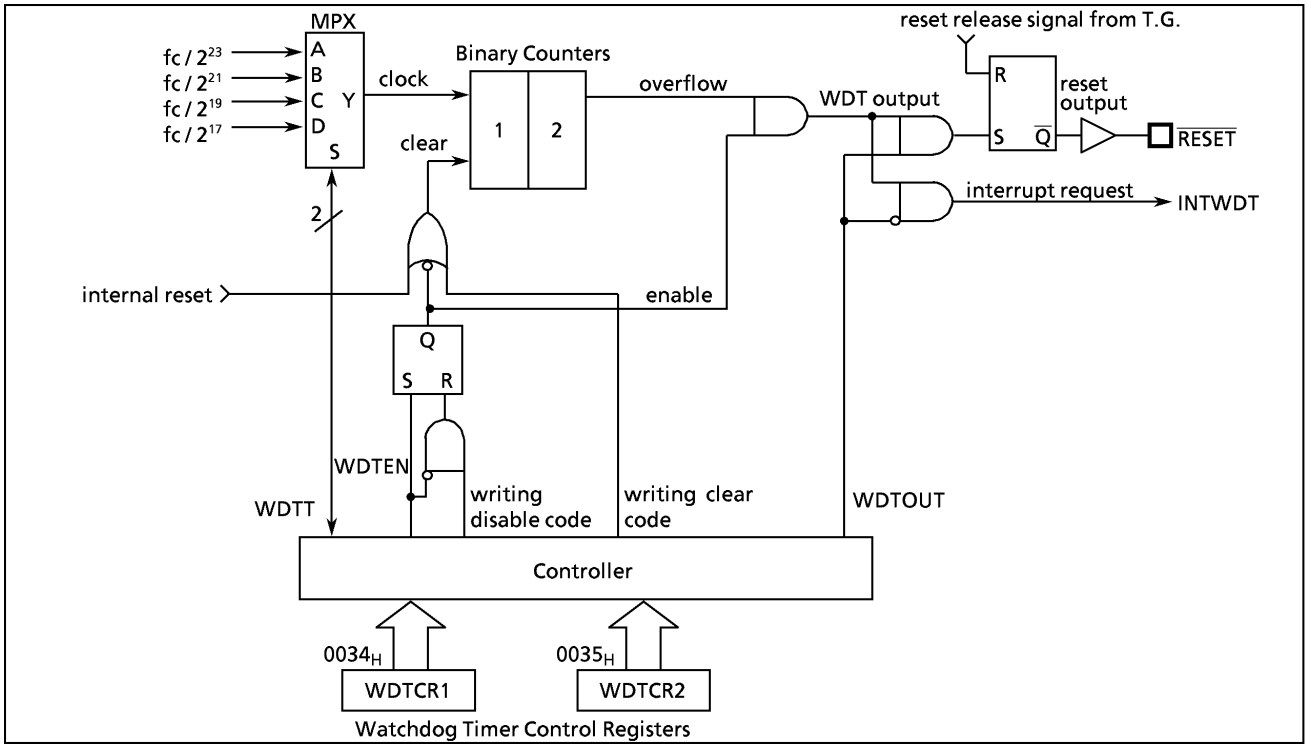


Figure 1-26. Watchdog timer configuration

### 1.10.2 Watchdog timer control

Figure 1-26 shows the watchdog timer control registers (WDTCR1, WDTCR2). The watchdog timer is automatically enabled after reset.

(1) Malfunction detection methods using the watchdog timer
The CPU malfunction is detected as follows:
① Setting the detection time, selecting output, and clearing the binary counter.
② Repeatedly clearing the binary counter within the setting detection time.

If a CPU malfunction occurs for any cause, the watchdog timer output will become active on the rise of an overflow from the binary counters unless the binary counters are cleared. At this time, when WDTOUT = 1 a reset is generated, which drives the $\overline{\text{RESET}}$ pin low to reset the internal hardware and the external circuits. When WDTOUT = 0, a watchdog timer interrupt (INTWDT) is generated.
The watchdog timer temporarily stops counting in STOP mode (including warm-up) or IDLE mode, and automatically restarts (continues counting) when STOP/IDLE mode is released.

Example : Sets the watchdog timer detection time to $2^{21}$/fc [s] and resets the CPU malfunction.

```
        LD   (WDTCR2),  4EH         ; Clears the binary counters
        LD   (WDTCR1),  00001101B   ; WDTT←10, WDTOUT←1
Within WDT
detection time
        LD   (WDTCR2),  4EH         ; Clears the binary counters
        :                             (always clear immediately after changing WDTT)
        LD   (WDTCR2),  4EH         ; Clears the binary counters
Within WDT
detection time
        :
        LD   (WDTCR2),  4EH         ; Clears the binary counters
```

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Watchdog Timer Control Register 1

**WDTCR1** ($0034_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| | | | | WDT EN | WDTT | | WDT OUT |

(Initial value : **** 1001)

| | | | |
|---|---|---|---|
| WDTEN | Watchdog timer enable/disable | 0 : Disable (It is necessary to write the disable code to WDTCR2)<br>1 : Enable | write only |
| WDTT | Watchdog timer detection time | 00 : $2^{25}$ / fc [s] (4.194 s at fc = 8 MHz)<br>01 : $2^{23}$ / fc (1.048 s at fc = 8 MHz)<br>10 : $2^{21}$ / fc (262.1 ms at fc = 8 MHz)<br>11 : $2^{19}$ / fc (65.5 ms at fc = 8 MHz) | |
| WDTOUT | Watchdog timer output select | 0 : Interrupt request<br>1 : Reset output | |

*Note 1 : WDTOUT cannot be set to "1" by program after clearing WDTOUT to "0".*
*Note 2 : fc ; High-frequency clock [Hz] * ; don't care*
*Note 3 : WDTCR1 is a write-onry-register and must not be used with any of read-modify-write instructions*
*Note 4 : Disable the watchdog timer or clear the counter just before switching to STOP mode.*
*When the counter is cleared just before switching to STOP mode, clear the counter again subsequently to releasing STOP mode.*

Watchdog Timer Control Register 2

**WDTCR2** ($0035_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| | | | | | | | |

(Initial value : **** ****)

| | | | |
|---|---|---|---|
| WDTCR2 | Watchdog timer control code write register | $4E_H$ : Watchdog timer binary counter clear (clear code)<br>$B1_H$ : Watchdog timer disable (disable code)<br>others : Invalid | write only |

*Note 1 : The disable code is invalid unless written when WDTEN = 0.*
*Note 2 : * ; don't care*

Figure 1-27. Watchdog timer control registers

(2) Watchdog timer enable

The watchdog timer is enabled by setting WDTEN (bit 3 in WDTCR1) to "1". WDTEN is initialized to "1" during reset, so the watchdog timer operates immediately after reset is released.

Example : Enables watchdog timer

```
LD (WDTCR1), 00001000B ; WDTEN←1
```

(3) Watchdog timer disable

The watchdog timer is disabled by writing the disable code ($B1_H$) to WDTCR2 after clearing WDTEN (bit 3 in WDTCR1) to "0". The watchdog timer is not disabled if this procedure is reversed and the disable code is written to WDTCR2 before WDTEN is cleared to "0". The watchdog timer is halted temporarily in STOP mode (including warm-up) and IDLE mode, and restarts automaticallyafter STOP or IDLE mode is released.

During disabling the watchdog timer, the binary counters are cleared to "0".

Example : Disables watchdog timer

```
LDW (WDTCR1) , 0B101H ; WDTEN←0, WDTCR2←disable code
```

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### 1.10.3 Watchdog timer interrupt (INTWDT)

This is a pseudo non-maskable interrupt which can be accepted regardless of the contents of the EIR. If a watchdog timer interrupt or a software interrupt is already accepted, however, the new watchdog timer interrupt waits until the previous non-maskable interrupt processing is completed (the end of the [RETN] instruction execution).

The stack pointer (SP) should be initialized before using the watchdog timer output as an interrupt source with WDTOUT.

Example : Watchdog timer interrupt setting up.

```
LD   SP,   043FH              ; Sets the stack pointer
LD   (WDTCR1) ,   00001000B ; WDTOUT←0
```

### 1.10.4 Watchdog timer reset

If the watchdog timer output becomes active, a reset is generated, which drives the $\overline{\text{RESET}}$ pin (sink open drain output) low to reset the internal hardware and the external circuits. The reset output time is $2^{20}$/fc [s] (131 ms at fc = 8 MHz).

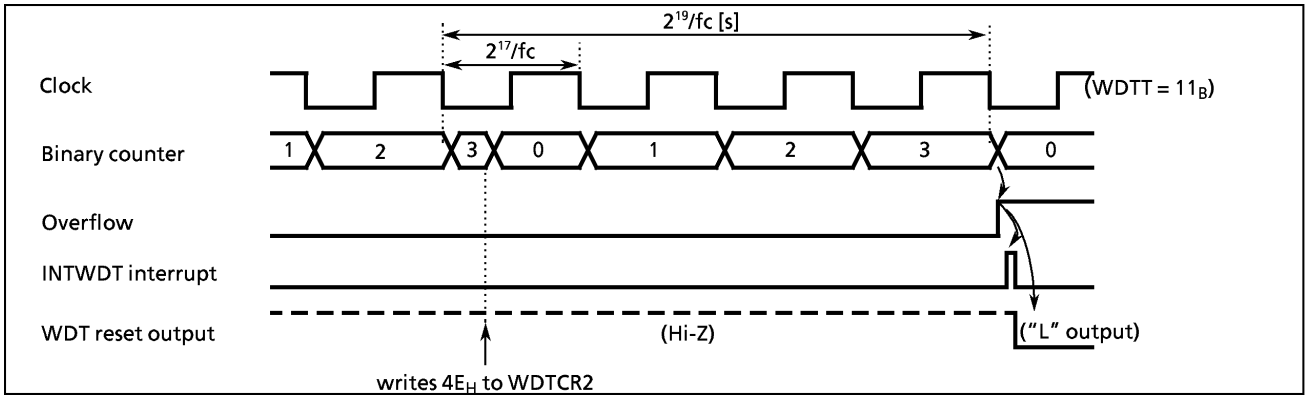


Figure 1-28. Watchdog timer interrupt / reset

## 1.11 Reset Circuit

The TLCS-870 Series has four types of reset generation procedures: an external reset input, an address-trap-reset, a watchdog timer reset and a system-clock-reset. Table 1-4 shows on-chip hardware initialization by reset action. The internal source reset circuit (watchdog timer reset, address trap reset, and system clock reset) is not initialized when power is turned on. Thus, output from the $\overline{\text{RESET}}$ pin may go low ($2^{20}$/fc [s.] 131 ms at 8 MHz) when power is turned on.

Table 1-4. Initializing internal status by reset action

| On-chip Hardware | | Initial Value | On-chip Hardware | Initial Value |
|---|---|---|---|---|
| Program counter | (PC) | (FFFF$_H$) · (FFFE$_H$) | Divider of Timing generator | 0 |
| Register bank selector | (RBS) | 0 | Watchdog timer | Enable |
| Jump status flag | (JF) | 1 | Output latches of I/O ports | Refer to I/O port circuitry |
| Interrupt master enable flag | (IMF) | 0 | Control registers | Refer to each of control register |
| Interrupt individual enable flags | (EF) | 0 | | |
| Interrupt latches | (IL) | 0 | | |

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### 1.11.1 External reset input

When the $\overline{\text{RESET}}$ pin is held at low for at least 3 machine cycles (12/fc [s]) with the power supply voltage within the operating voltage range and oscillation stable, a reset is applied and the internal state is initialized.

When the $\overline{\text{RESET}}$ pin input goes high, the reset operation is released and the program execution starts at the vector address stored at addresses $FFFE_H$ to $FFFF_H$.

The $\overline{\text{RESET}}$ pin contains a Schmitt trigger (hysteresis) with an internal pull-up resistor. A simple power-on-reset can be applied by connecting an external

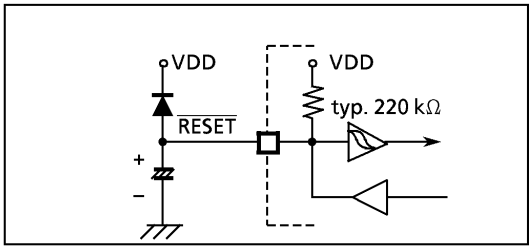
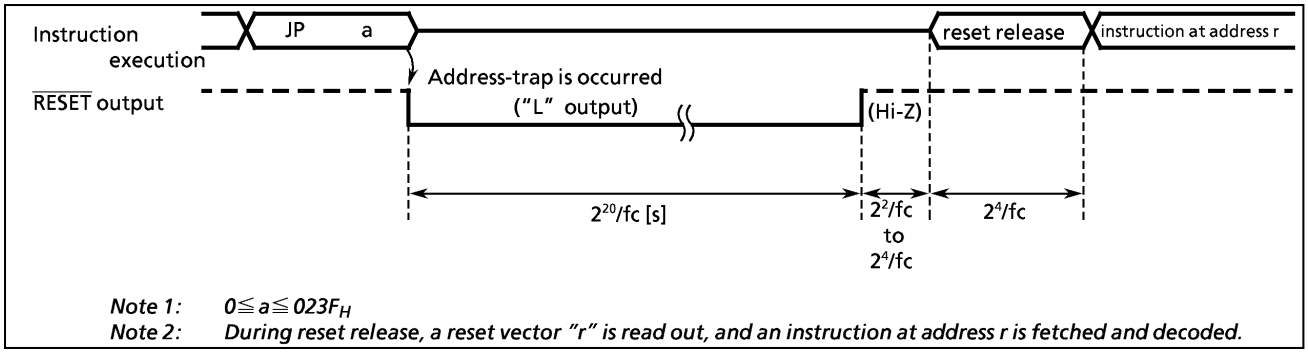



Figure 1-29. Simple power-on-reset circuitry

### 1.11.2 Address-trap-reset

If a CPU malfunction occurs and an attempt is made to fetch an instruction from the RAM or the SFR area (addresses $0000_H$ to $023F_H$ [87CH38/K38]), an address-trap-reset will be generated. Then, the $\overline{\text{RESET}}$ pin output will go low. The reset time is $2^{20}$/fc [s] (131 ms at fc = 8 MHz).




*Note 1:* $0 \leqq a \leqq 023F_H$

*Note 2:* *During reset release, a reset vector "r" is read out, and an instruction at address r is fetched and decoded.*

Figure 1-30. Address-trap-reset

### 1.11.3 Watchdog timer reset

Refer to Section "1.10 Watchdog Timer".

### 1.11.4 System-clock-reset

Clearing both bits 7 and 6 in SYSCR2 to "0" stops high-frequency oscillation, and causes the MCU to deadlock. This can be prevented by automatically generating a reset signal whenever (bit 7 in SYSCR2) = (bit 6 in SYSCR2) = 0 is detected to continue the oscillation. Then, the $\overline{\text{RESET}}$ pin output goes low from high-impedance. The reset time is $2^{20}$/fc [s] (131 ms at fc = 8 MHz).

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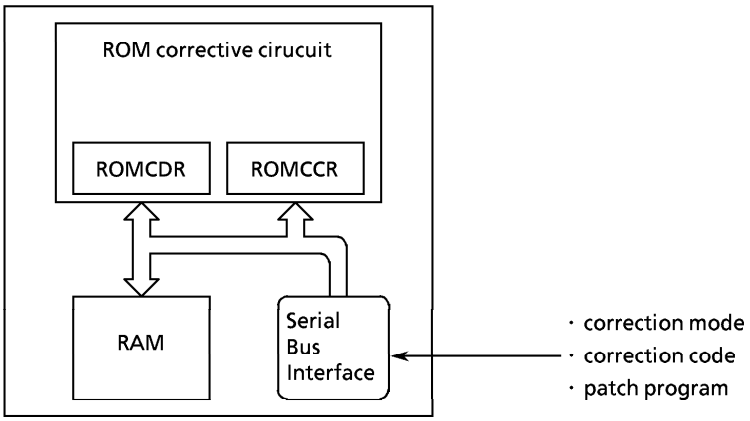
## 1.12 ROM corrective function

The ROM corrective function can patch the part (s) of on-chip ROM with some bugs. The patched data should be loaded from an external memory at the initialized routine beforehand. The figure below shows one example of configurations loading the patched data via I²C-bus. The ROM corrective function have two modes. One is to replace the instruction on a certain address in the ROM with the jump instruction to branch into the RAM area where the patched codes and/or data are loaded (Program Jump Mode) . The other is to replace a byte or a word (2 byte) length data in the ROM with the patched data (Data Replacement Mode) . When the ROM corrective function is enabled, the address-trap-reset is automatically disabled on the RAM area from $0140_H$ where the patched program is running.

*Note1 : When use ROM correction circuit, it is necessary to contain a program which operates to load patch data or program code from external memory to internal date RAM in an initial routine.*
*Note2 : BM87CS38N0A does not support the ROM corrective function.*
*Note3 : RAM area which is used for ROM correction circuit in 87CH38/K38 can use address from 0140H to 023FH, but RAM area which is used for ROM correction circuit in OTP (87PS38) can use address from 0240H to 083FH. Therefore, when using ROM correction circuit in 87CH38/K38, load address for patch program codes and jump vector must be changed after debugging a program by OTP.*

Example :

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### 1.12.1 Configuration

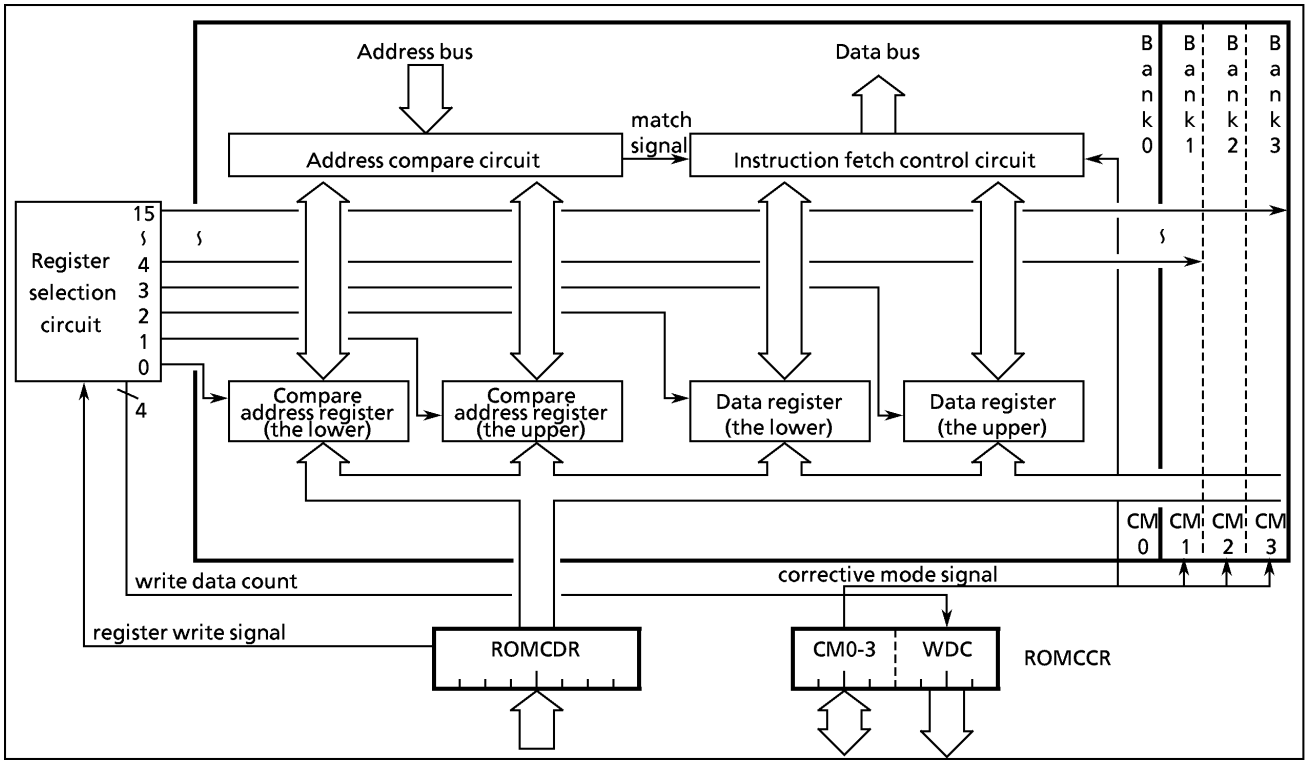


Figure 1-31. ROM correction circuit

### 1.12.2 Control

The ROM corrective function is controlled by ROM corrective control register (ROMCCR) and ROM corrective data register (ROMCDR).

ROMCCR ($0FC0_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| CM3 | CM2 | CM1 | CM0 | WDC | | | |

(Initial value 0000 0000)

| | | | |
|---|---|---|---|
| CM3 | corrective mode setting (BANK3) | 0 : program jump mode<br>1 : data replacement mode | Read/ Write |
| CM2 | corrective mode setting (BANK2) | | |
| CM1 | corrective mode setting (BANK1) | | |
| CM0 | corrective mode setting (BANK0) | | |
| WDC | write data counter | Counting the number of the byte written in ROMCDR | read only |

ROMCDR ($0FC1_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| | | | | | | | |

(Initial value 0000 0000) write only

Figure 1-32. ROM corrective control register and ROM corrective data register

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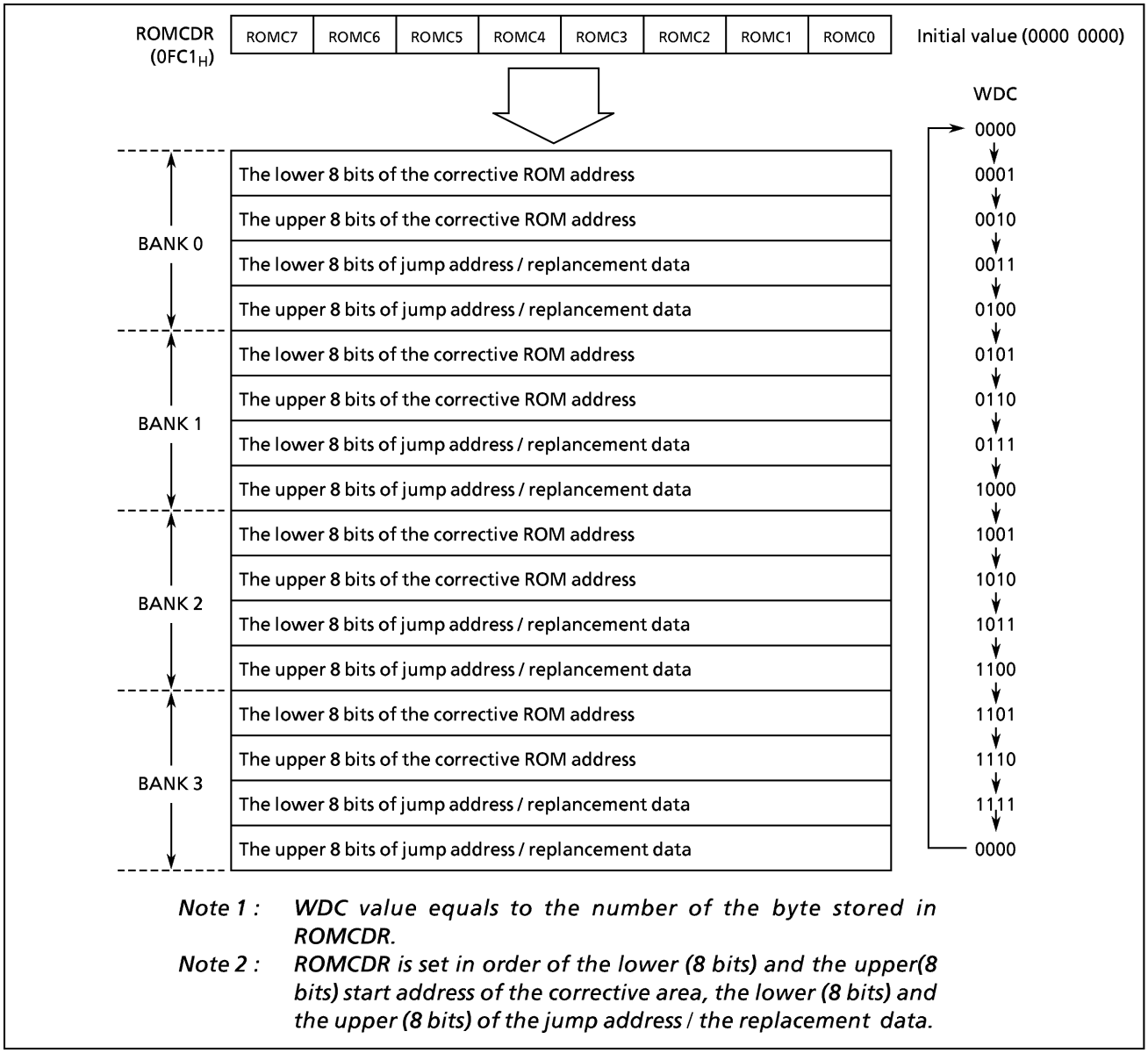
(1) The ROM corrective data register writing
The ROM corrective data register has four banks corresponding to four independent locations to patch. The write data counter (WDC) points each bank to set. (Figure 1-33.)




*Note 1 : WDC value equals to the number of the byte stored in ROMCDR.*

*Note 2 : ROMCDR is set in order of the lower (8 bits) and the upper(8 bits) start address of the corrective area, the lower (8 bits) and the upper (8 bits) of the jump address / the replacement data.*

Figure 1-33. Banks and WDC value of the program corrective data register

Whenever ROMCDR is written, WDC is incremented to indicate what data is writen via ROMCDR. During reset, WDC is intialized to "0".

(1) The lower start address of the corrective area (8 bits)
(2) The upper start address of the corrective area (8 bits)
(3) The lower jump address / replacement data (8 bits)
(4) The upper jump address / replacement data (8 bits)

*Note : Corrective addresses must have over five address each other.*

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### 1.12.3 Functions

The ROM corrective function can correct maximum four ROM areas with their corresponding four banks of ROM corrective registers. Either program jump mode or data replacement mode is selected for each bank by CM0 - CM3 respectively.

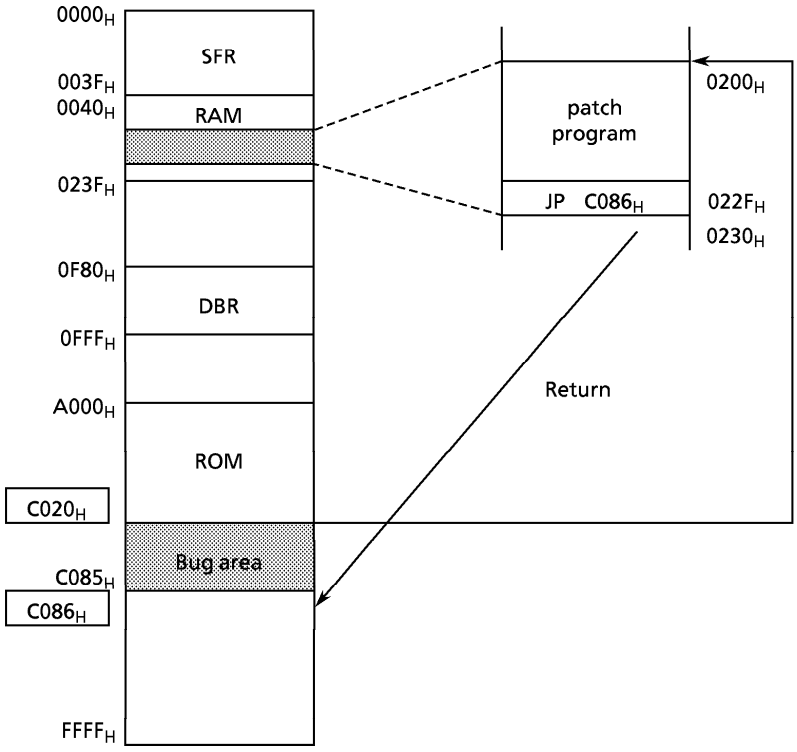
(1) Program jump mode

The program jump mode is to execute the program in the RAM area to correct the bug (s) in the ROM. The start address of ROM that should be patched and the jump vector pointing the RAM area are specified by ROMCDR. When the program is about to run on the code at this start address, the jump instruction is issued, the program branches into the RAM at the jump vector, and the subsequent program codes primarily loaded into this RAM area are excuted. After this patch program execution, the program must be returned to the ROM area by any of the jump instructions at the end of this RAM area. By doing these, the correction of the bug is completed. The program jump mode can be selected at CMn = 0 (n = 0 - 3 for each bank) . The start address must point the 1 st byte of the instruction codes (Op-Code) .

| | |
|---|---|
| *Note :* | *Corrective address must be assigned to 1st byte of instruction codes on the program jump mode.* |

Example : There is bugs on the locations from $C020_H$ to $C085_H$

The corrective address, the jump vector, the program patch codes and other information to patch the ROM with the bugs must be read out from any of memory storage that holds them during initial program routine. CMn = 0 specifies the program jump mode. Subsequently, the patch program codes are loaded into RAM ($0200_H$ to $022F_H$) . The start address ($C020_H$) of the ROM necessary to patch is written to the corrective ROM address registers, and the start address ($0200_H$) of the RAM area to patch is loaded onto the jump address registers. When the instruction at $C020_H$ is fetched, the instruction to jump into $0200_H$ is unconditionally executed instead of the instruction at $C020_H$, and the subsequent patch program codes are executed. The jump instruction at the end of the patch program codes returns to the ROM at $C086_H$.

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(2) Data replacement mode

The data replacement mode is to directly replace a single byte or word (2 byte) length data with the replacement data which are written via ROMCDR.

The program jump mode can work as the equivalent data replacement mode. However, when many instructions refer a certain data in the ROM which must be patched, the program jump mode consumes the same number of banks as that of the instructions referring this (these) data. ROM data replace mode reduces this kind of bank consumption.

*Note : The instruction that accesses to an only byte is replaced to an only start byte.*

By setting CMn to "1", the data replacement mode is selected. The start address of ROM data is set to the corrective ROM address, and two bytes replacement data is set to the patch data register via ROMCDR.

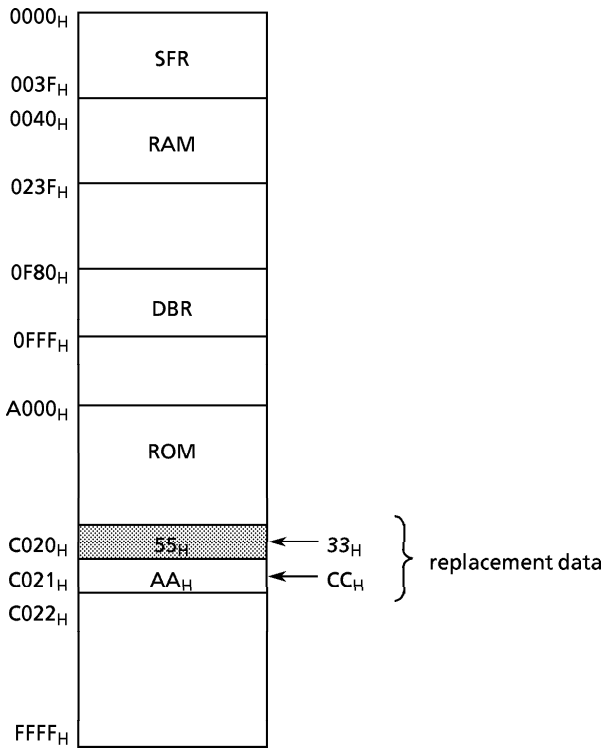
The corrective address must point the constant data in the data replacement mode. It is impossible to replace opecode and operand in the data replacement mode.

*Note1 : Corrective address must be assigned to constant data area on the data replacement mode. (Ope-code and operand can not be replaced by ROM correction circuit)*

*Note2 : Instructions which incluedes "(HL+)" or "(–HL)" operation can not be replaced by ROM correction circuit on the data replacement mode.*

Example :

The start address is set to $C020_H$ as the location of the replaced data. Two bytes of the patch data are set $33_H$ for $C020_H$, $CC_H$ for $C021_H$.




1. At HL = $C020_H$, Executing LD A, (HL) loads $33_H$ in A. (Data replacement)
2. At HL = $C021_H$, Executing LD A, (HL) loads $AA_H$ in A. (No data replacement)
3. At HL = $C020_H$, Executing LD WA, (HL) loads $CC33_H$ in WA. (Data replacement)

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# 2. On-Chip Peripherals Functions

## 2.1 Special Function Registers (SFR) and Data Buffer Registers (DBR)

The TLCS-870 Series uses the memory mapped I/O system and all peripherals control and data transfers are performed through the special function registers (SFR) and data buffer registers (DBR).
The SFR are mapped to addresses $0000_H$ to $003F_H$, and the DBR to addresses $0F80_H$ to $0FFF_H$.
Figure 2-1 shows the list of the 87CH38/K38 SFRs and DBRs.

| Address | Read | Write |
|---|---|---|
| $0000_H$ | reserved | |
| 01 | reserved | |
| 02 | P2 Port | |
| 03 | P3 Port | |
| 04 | P4 Port | |
| 05 | P5 Port | |
| 06 | P6 Port | |
| 07 | P7 Port | |
| 08 | reserved | |
| 09 | reserved | |
| 0A | reserved | |
| 0B | reserved | |
| 0C | – | P4CR (P4 I/O control) |
| 0D | – | P6CR (P6 I/O control) |
| 0E | ADCCR (A/D Converter Control) | |
| 0F | ADCDR (A/D CONV. Result) | |
| 10 | – | $TREG1A_L$ (Timer register 1A) |
| 11 | – | $TREG1A_H$ |
| 12 | $TREG1B_L$ (Timer register 1B) | |
| 13 | $TREG1B_H$ | |
| 14 | – | TC1CR (TC1 control) |
| 15 | – | TC2CR (TC2 control) |
| 16 | – | $TREG2_L$ (Timer register 2) |
| 17 | – | $TREG2_H$ |
| 18 | TREG3A (Timer register 3A) | |
| 19 | TREG3B (Timer register 3B) | – |
| 1A | – | TC3CR (TC3 control) |
| 1B | – | TREG4 (Timer register 4) |
| 1C | – | TC4CR (TC4 control) |
| 1D | reserved | |
| 1E | reserved | |
| 1F | reserved | |

| Address | Read | Write |
|---|---|---|
| $0020_H$ | SBICR1 (ACK/CHS) | SBICR1 (SBI control 1) |
| 21 | SBIDBR (SBI data buffer) | |
| 22 | – | I2CAR ($I^2C$-bus address) |
| 23 | SBISR (SBI status) | SBICR2 (SBI control 2) |
| 24 | reserved | |
| 25 | PWMSR (PWM status) | PWMCR (PWM control) |
| 26 | – | PWMDBR (PWM data buffer) |
| 27 | PMPXCR (Port control) | |
| 28 | reserved | |
| 29 | reserved | |
| 2A | reserved | |
| 2B | reserved | |
| 2C | reserved | |
| 2D | reserved | |
| 2E | reserved | |
| 2F | reserved | |
| 30 | reserved | |
| 31 | reserved | |
| 32 | reserved | |
| 33 | reserved | |
| 34 | – | WDTCR1 (WDT control) |
| 35 | – | WDTCR2 |
| 36 | TBTCR (TBT control) | |
| 37 | EINTCR (Exter. interrupt control) | |
| 38 | SYSCR1 (System control) | |
| 39 | SYSCR2 | |
| 3A | $EIR_L$ (Interrupt enable register) | |
| 3B | $EIR_H$ | |
| 3C | $IL_L$ (Interrupt latch) | |
| 3D | $IL_H$ | |
| 3E | reserved | |
| 3F | PSW (Program status word) | RBS (Register bank selector) |

(a) Special Function Registers

Figure 2-1-1. SFR & DBR

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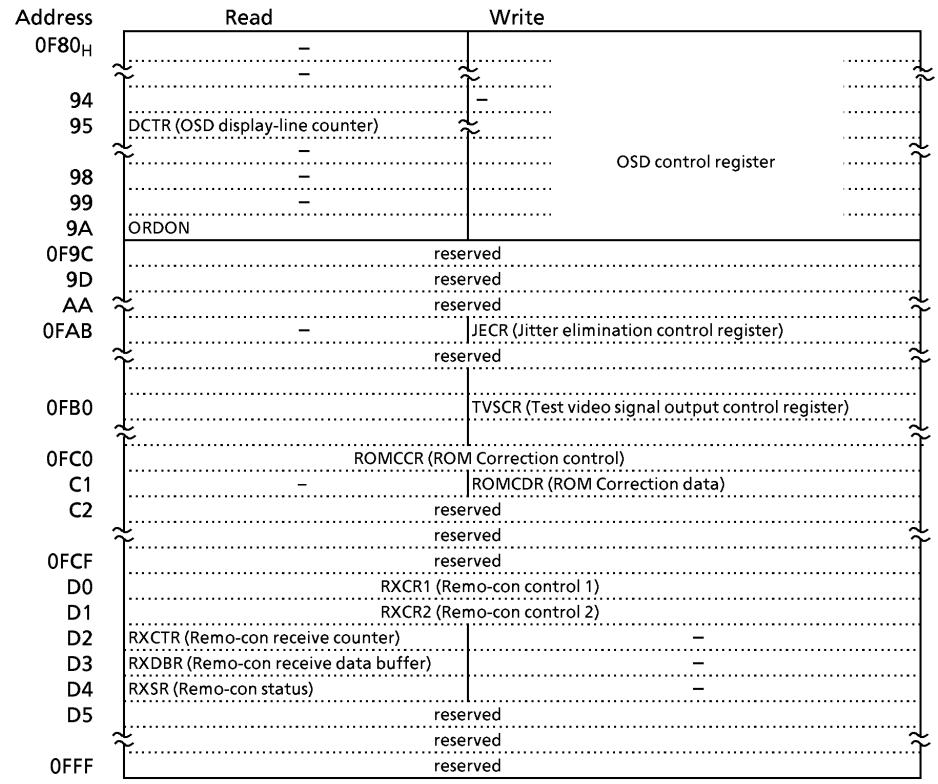


(b) Data Buffer Registers

*Note 1 : Do not access reserved areas by program.*
*Note 2 : – : Cannot be accessed.*
*Note 3 : When defining address 003F$_H$ with assembler symbols, use GPSW and GRBS.*
*Note 4 : Write-only registers and interrupt latches cannot use the read-modify-write instructions (bit manipulation instructions such as SET, CLR, etc. and logical operation instructions such as AND, OR, etc.)*
*Note 5 : SBI : Serial Bus Interface*
*PWM : Pulse Width Madulation*
*OSD : On screen display*
*Remo-con : Remote control signal preprocessor*

Figure 2-1-2. DBR

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## 2.2 I/O Ports

The 87CH38/K38 has 6 parallel input/output ports (33pins) as follows:

| | Primary Function | Secondary Functions |
|---|---|---|
| Port P2 | 1-bit I/O port | external interrupt input, and STOP mode release signal input |
| Port P3 | 6-bit I/O port | external interrupt input, remote control signal input, timer / counter input, and serial bus interface input/output |
| Port P4 | 8-bit I/O port | pulse width modulation output |
| Port P5 | 8-bit I/O port | pulse width modulation output, external interrupt input, serial bus interface input / output and A/D converter input |
| Port P6 | 8-bit I/O port | R, G, B and Y/BL output from OSD circuitry, A/D converter input, test video signal output |
| Port P7 | 2-bit I/O port | horizontal synchronous pulse input and vertical synchronous pulse input to OSD circuitry |

Each output port contains a latch, which holds the output data. All input ports do not have latches, so the external input data should either be held externally until read or reading should be performed several times before processing. Figure 2-2 shows input/output timing examples.
External data is read from an I/O port in the S1 state of the read cycle during execution of the read instruction. This timing can not be recognized from outside, so that transient input such as chattering must be processed by the program.
Output data changes in the S2 state of the write cycle during execution of the instruction which writes to an I/O port.

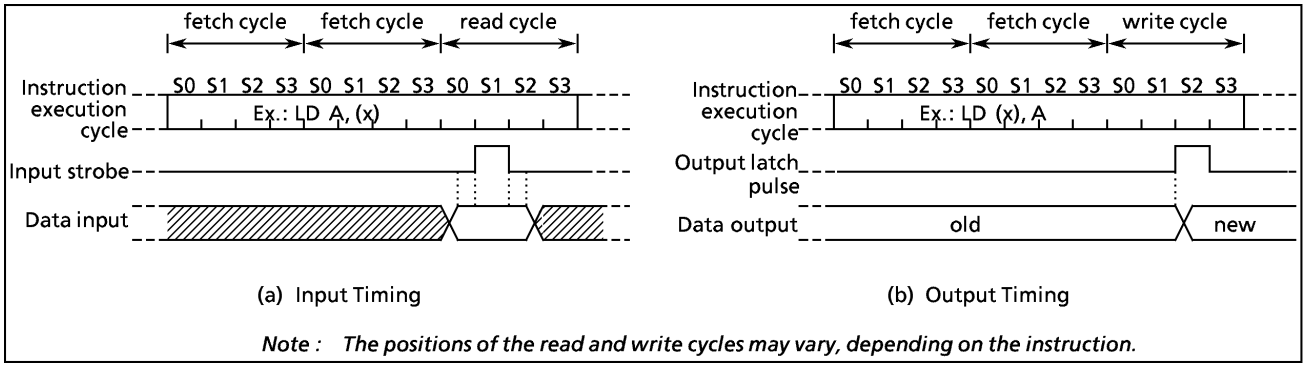


Figure 2-2. Input/output timing (Example)

When reading an I/O port except programmable I/O ports, whether the pin input data or the output latch contents are read depends on the instructions, as shown below:

(1) Instructions that read the output latch contents
- ① XCH r, (src)
- ② CLR/SET/CPL (src).b
- ③ CLR/SET/CPL (pp).g
- ④ LD (src).b, CF
- ⑤ LD (pp) . b, CF
- ⑥ ADD/ADDC/SUB/SUBB/AND/OR/XOR (src), n
- ⑦ (src) side of ADD/ADDC/SUB/SUBB/AND/OR/XOR (src), (HL)

(2) Instructions that read the pin input data
- ① Instructions other than the above (1)
- ② (HL) side of ADD/ADDC/SUB/SUBB/AND/OR/XOR (src), (HL)

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### 2.2.1 Port P2 (P20)

Port P2 is a 1-bit input/output port. It is also used as an external interrupt input, and a STOP mode release signal input. When used as an input port, or a secondary function pin, the output latch should be set to "1". During reset , the output latch is initialized to "1".

It is recommended that pin P20 should be used as an external interrupt input, a STOP mode release signal input, or an input port. If used as an output port, the interrupt latch is set on the falling edge of the P20 output pulse.

When a read instruction for port P2 is executed, bits 7 to 1 in P2 are read in as undefined data.

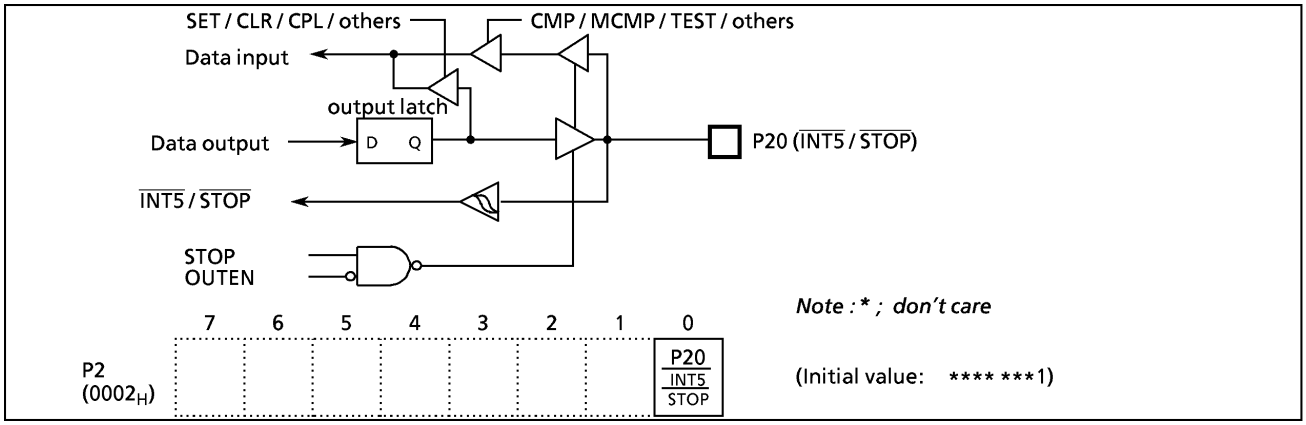


Figure 2-3. Port P2

### 2.2.2 Port P3 (P35 to P30)

Port P3 is a 6-bit input / output port, and is also used as serial bus interface input/output, an exrernal interrupt input, a timer / counter input, and Remote-control signal input. When used as an input port or a secondary function pin, the output latch should be set to "1". The output latches are initialized to "1" during reset.

Example 1 : Outputs an immediate data $5A_H$ to port P3.

```
LD  (P3), 5AH        ; P3←5AH
```

Example 2 : Inverts the output of the lower 4bits (P33 to P30) in port P3.

```
XOR (P3), 00001111B ; P33 to P30←P33 to P30
```

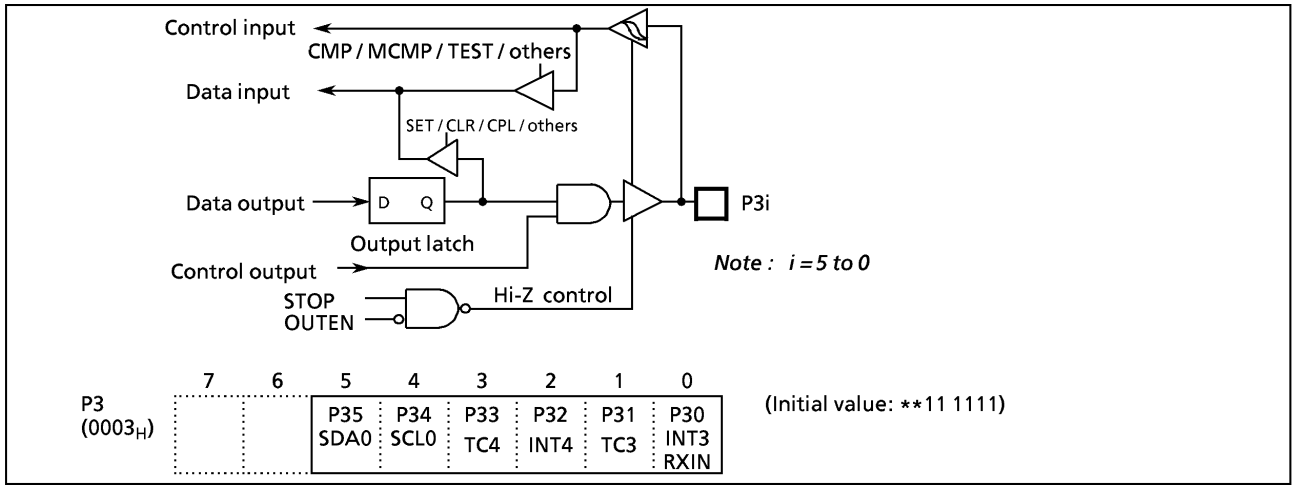


Figure 2-4. Port P3

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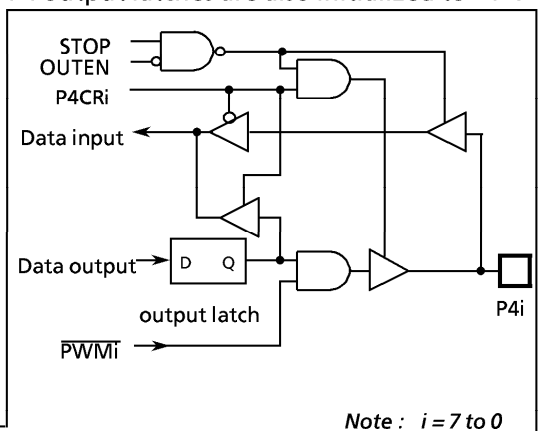
### 2.2.3 Port P4 (P47 to P40)

Port P4 is an 8-bit input / output port which can be configured as an input or an output in one-bit unit under software control. Input / output mode is specified by the corresponding bit in the port P4 input/output control register (P4CR). Port P4 is configured as an input if its corresponding P4CR bit is cleared to "0", and as an output if its corresponding P4CR bit is set to "1". During reset, P4CR is initialized to "0", which configures port P4 as an input . The P4 output latches are also initialized to "1". Data is written into the output latch regardless of the P4CR contents. Therfore initial output data should be written into the output latch before setting P4CR. Port P4 is also used as a pulse width modulation (PWM) output. When used as a PWM output pin, the output pins should be set to the output mode and beforehand the output latch should be set to "1".

*Note : Input mode port is read the state of input pin. When input / output mode is used to mixed, the contents of output latch setting input mode may be changed by executing bit manipulation instructions.*




P4 ($0004_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| P47 | P46 | P45 | P44 | P43 | P42 | P41 | P40 |
| PWM7 | PWM6 | PWM5 | PWM4 | PWM3 | PWM2 | PWM1 | PWM0 |

(Initial value : 1111 1111)

P4CR ($000C_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| | | | | | | | |

(Initial value : 0000 0000)

| | | | |
|---|---|---|---|
| P4CR | I/O control for port P4 | 0 : input mode<br>1 : output mode | write only |

Figure 2-5. Ports P4 and P4CR

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### 2.2.4 Port P5 (P57 to P50)

P5 port is an 8-bit input/output port. And is also used as analog input for A/D converter, a pulse width modulation(PWM) output and serial bus interface input/output. During reset, the P5 output latches are initialized to "1". And AINDS (bit4 in the ADCCR) and SAIN (bit3 to 0 in the ADCCR) are also initialized to "0", which configure port P53 as an analog input.

When used as an input port or a secondary function pin, the output latch should be set to "1".

When used as an analog input, AINDS must be cleared to "0". The bit used as an analog input must be selected by SAIN. Unused pin as analog input can be used as input/output port. But it is recommendable that the contents of output pins in P5 port should not be changed during A/D conversion, because an accuracy of A/D conversion is changed for the worse.

Pin P50 (INT0) can be configured as either an I/O port or an external interrupt input with INT0EN (bit6 in EINTCR). During reset, pin P50 (INT0) is configured as an input port P50.

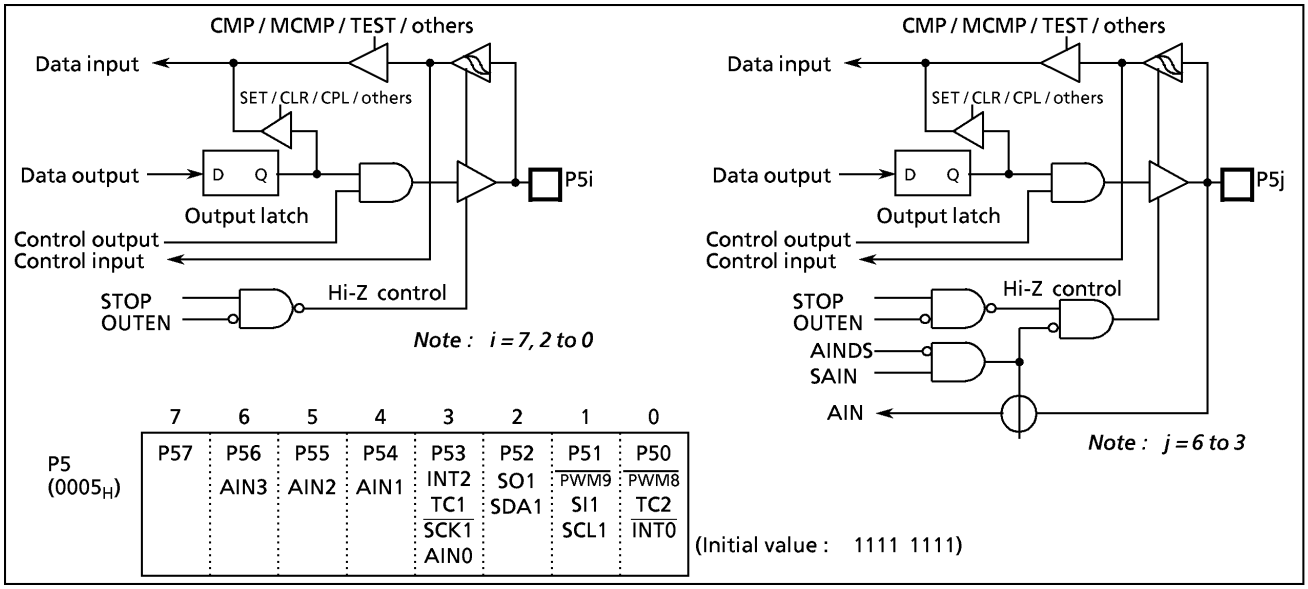


Figure 2-6. Ports P5

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### 2.2.5 Port P6 (P67 to P60)

Port P6 is an 8-bit input / output port which can be configured as an input or an output in one-bit unit under software control. Input or output mode is selected by the corresponding bit in the input/output control register (P6CR). For example, port P6 is configured as an input if its corresponding P6CR bit is cleared to "0", and as an output if its corresponding bit is set to "1". During reset, P6CR is initialized to "0", which configures port P6 as an input. The P6 output latches are also initialized to "1".

Data is written into the output latch regardless of the P6CR contents. Therefore initial output data should be written into the output latch before setting P6CR. Pins P63 to P60 are available high current output, so LEDs can be driven directly.

Port P6 is also used an on screen display (OSD) output (R,G,B and Y/BL signal), an analog input for A/D converter and test video signal output.

When used as an OSD output pin, the OSD output pins should be set to the output mode and beforehand the port P6 data selection register (P67DS to P64DS) should be set to "1".

When used as an analog input, AINDS (bit4 in the ADCCR) must be cleared to "0". The bit used as an analog input must be selected by SAIN and be specified as input port by P6CR. Unused pin as analog input can be used as input/output port. But it is recommendable that the contents of output pins in P6 port should not be changed during A/D conversion, because an accuracy of A/D conversion is changed for the worse. When used as a test video signal output, the output pins should be set to the output mode and beforehand the output latch should be set to "1".

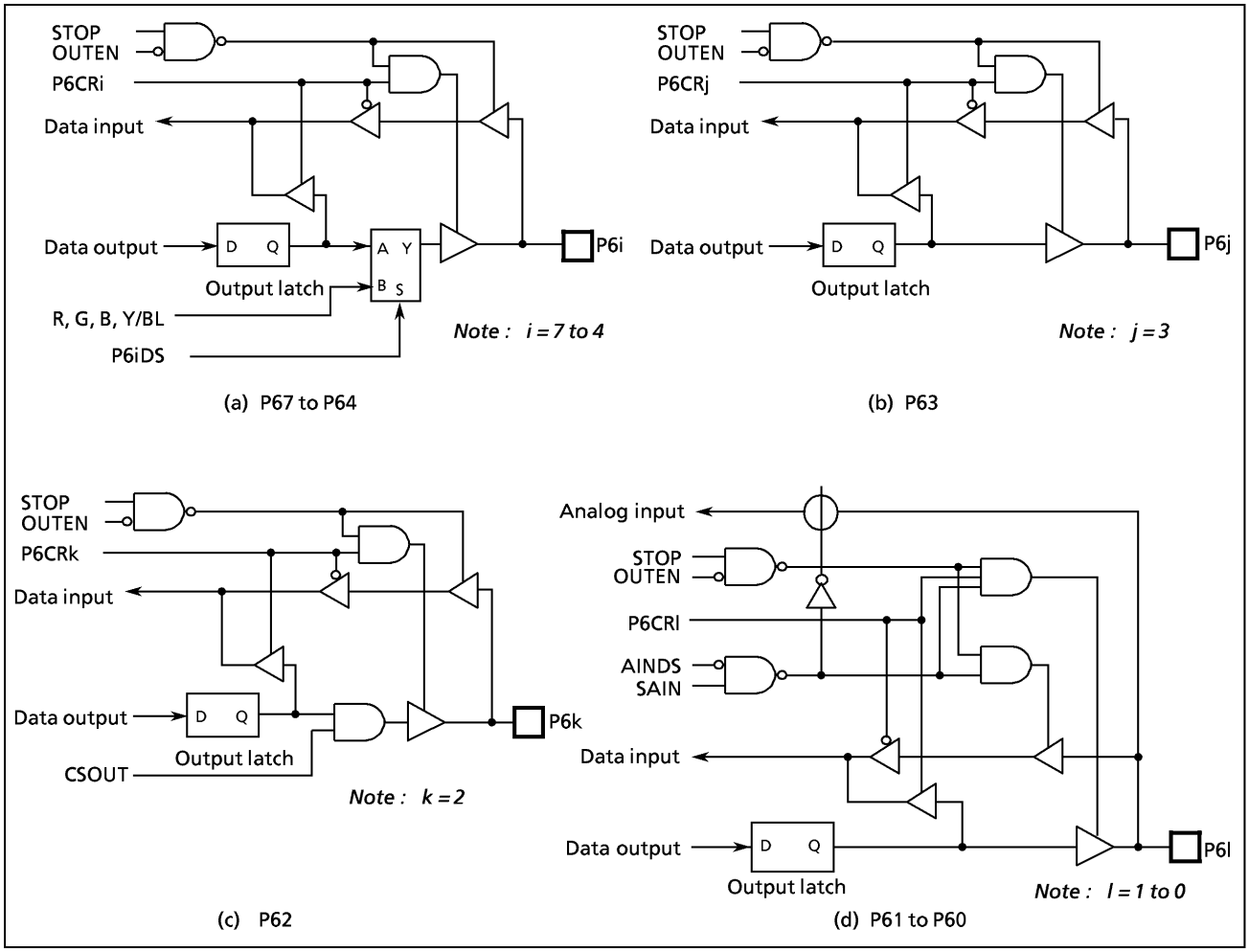


Figure 2-7-1. Ports P6, P6CR, and P67DS to P64DS

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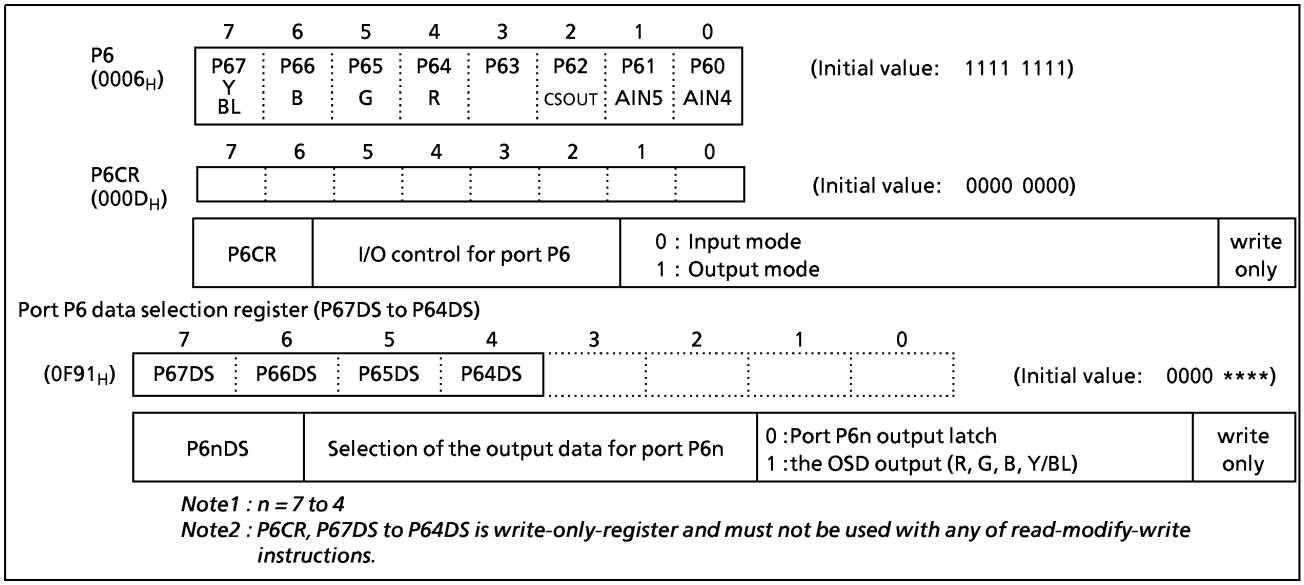


*Note1 : n = 7 to 4*
*Note2 : P6CR, P67DS to P64DS is write-only-register and must not be used with any of read-modify-write instructions.*

Figure 2-7-2. Ports P6, P6CR, and P67DS to P64DS

*Note : Input mode port is read the state of input pin. When input/output mode is used to mixed, the contents of output latch setting input mode may be changed by executing bit manipulation instructions.*

Example : Set the lower 4 bit in port P6 (P63 to 60) to the output port and Set the other to the input port.

```
LD (P6CR), 0FH ; P6CR ← 0000 1111B
```

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### 2.2.6 Port P7 (P71 to P70)

Port P7 is a 2-bit input /output port, and is also used as a vertical synchronous signal ($\overline{VD}$) input and a horizontal synchronous signal ($\overline{HD}$) input for the on screen display (OSD) circuitry.

The output latches are initialized to "1" during reset. When used as an input port or a secondary function pin, the output latch should be set to "1".

When a read instruction for port P7 is executed, bits 7 to 2 in P7 are read in as undefined data.

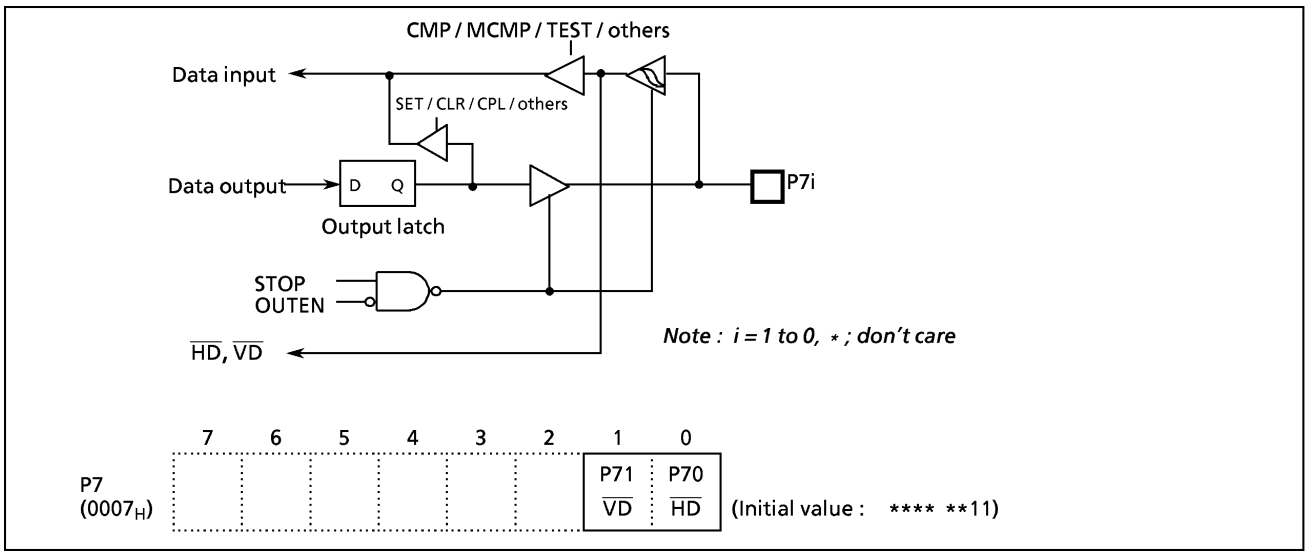


Figure 2-8. Ports P7

### 2.2.7 Port Switching Control

There are two types for pin assignment of pin 37 to pin 41. The pin assignment is selected with PMPX(PMPXCR bit 7). PMPX is initialized to "0" at reset.

| Mode \ Pin No. | 37 | 38 | 39 | 40 | 41 |
|---|---|---|---|---|---|
| MODE0 (PMPX = 0) | P32 (INT4) | P33 (TC4) | P34 (SCL0) | P35 (SDA0) | P57 |
| MODE1 (PMPX = 1) | P34 (SCL0) | P35 (SDA0) | P57 | P32 (INT4) | P33 (TC4) |

Port switching mode and pin assignment

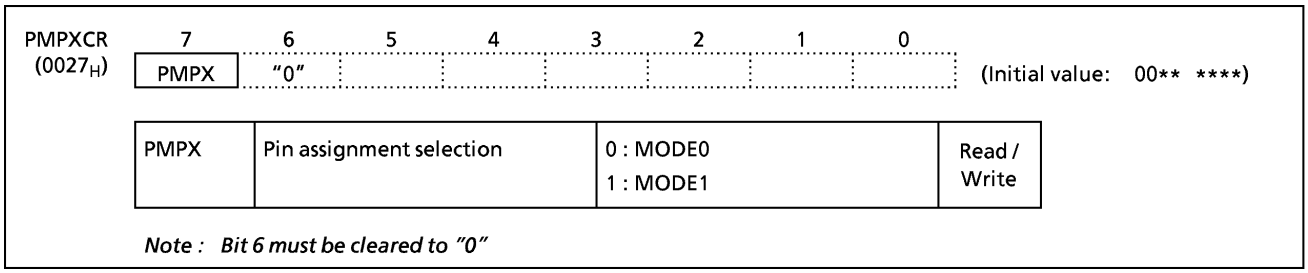


Figure 2-9. Port switching register

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## 2.3 Time Base Timer (TBT)

The time base timer generates time base for key scanning, dynamic displaying, etc. It also provides a time base timer interrupt (INTTBT). The time base timer is controlled by a control register (TBTCR) shown in Figure 2-10.

An INTTBT is generated on the first rising edge of source clock (the divider output of the timing generator) after the time base timer has been enabled. The divider is not cleared by the program; therefore, only the first interrupt may be generated ahead of the set interrupt period.

*The interrupt frequency (TBTCK) must be selected with the time base timer disabled* (When the time base timer is changed from enabling to disabling, the interrupt frequency can't be changed.) both frequency selection and enabling can be performed simultaneously.


Example : Sets the time base timer frequency to fc/$2^{16}$ [Hz] and enables an INTTBT interrupt.

```
LD      (TBTCR) , 00001010B
SET     (EIRL) . 6
```

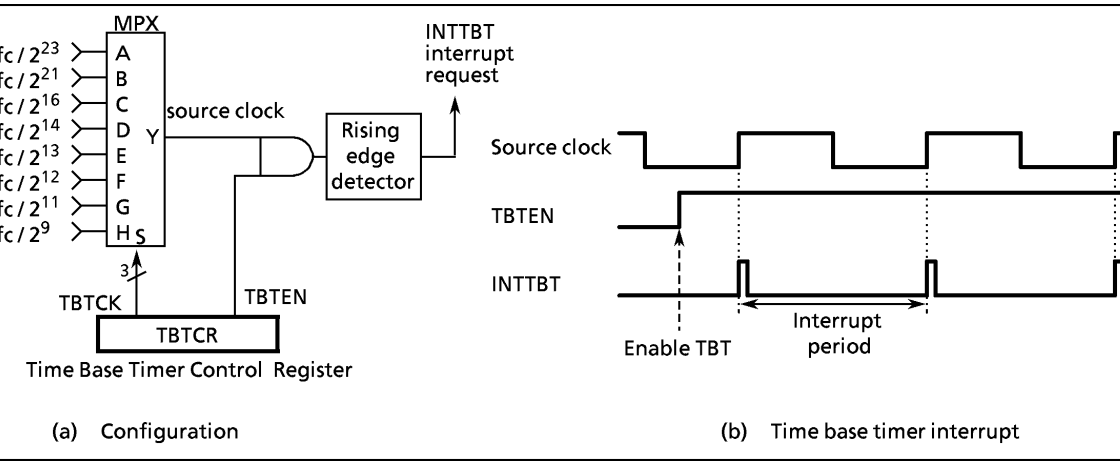
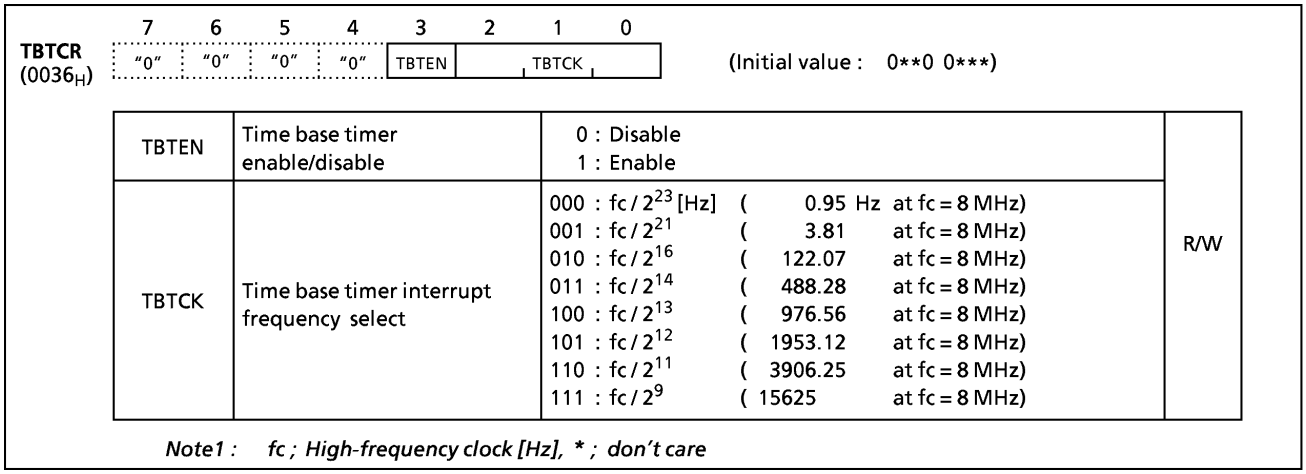



Figure 2-10. Time base timer



TBTCR ($0036_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| "0" | "0" | "0" | "0" | TBTEN | | TBTCK | |

(Initial value : 0**0 0***)

| | | | |
|---|---|---|---|
| TBTEN | Time base timer enable/disable | 0 : Disable<br>1 : Enable | R/W |
| TBTCK | Time base timer interrupt frequency select | 000 : fc / $2^{23}$ [Hz] ( 0.95 Hz at fc = 8 MHz)<br>001 : fc / $2^{21}$ ( 3.81 at fc = 8 MHz)<br>010 : fc / $2^{16}$ ( 122.07 at fc = 8 MHz)<br>011 : fc / $2^{14}$ ( 488.28 at fc = 8 MHz)<br>100 : fc / $2^{13}$ ( 976.56 at fc = 8 MHz)<br>101 : fc / $2^{12}$ ( 1953.12 at fc = 8 MHz)<br>110 : fc / $2^{11}$ ( 3906.25 at fc = 8 MHz)<br>111 : fc / $2^{9}$ ( 15625 at fc = 8 MHz) | |

*Note1 : fc ; High-frequency clock [Hz], * ; don't care*

Figure 2-11. Time base timer control register

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## 2.4 16-bit Timer / Counter 1 (TC1)

### 2.4.1 Configuration

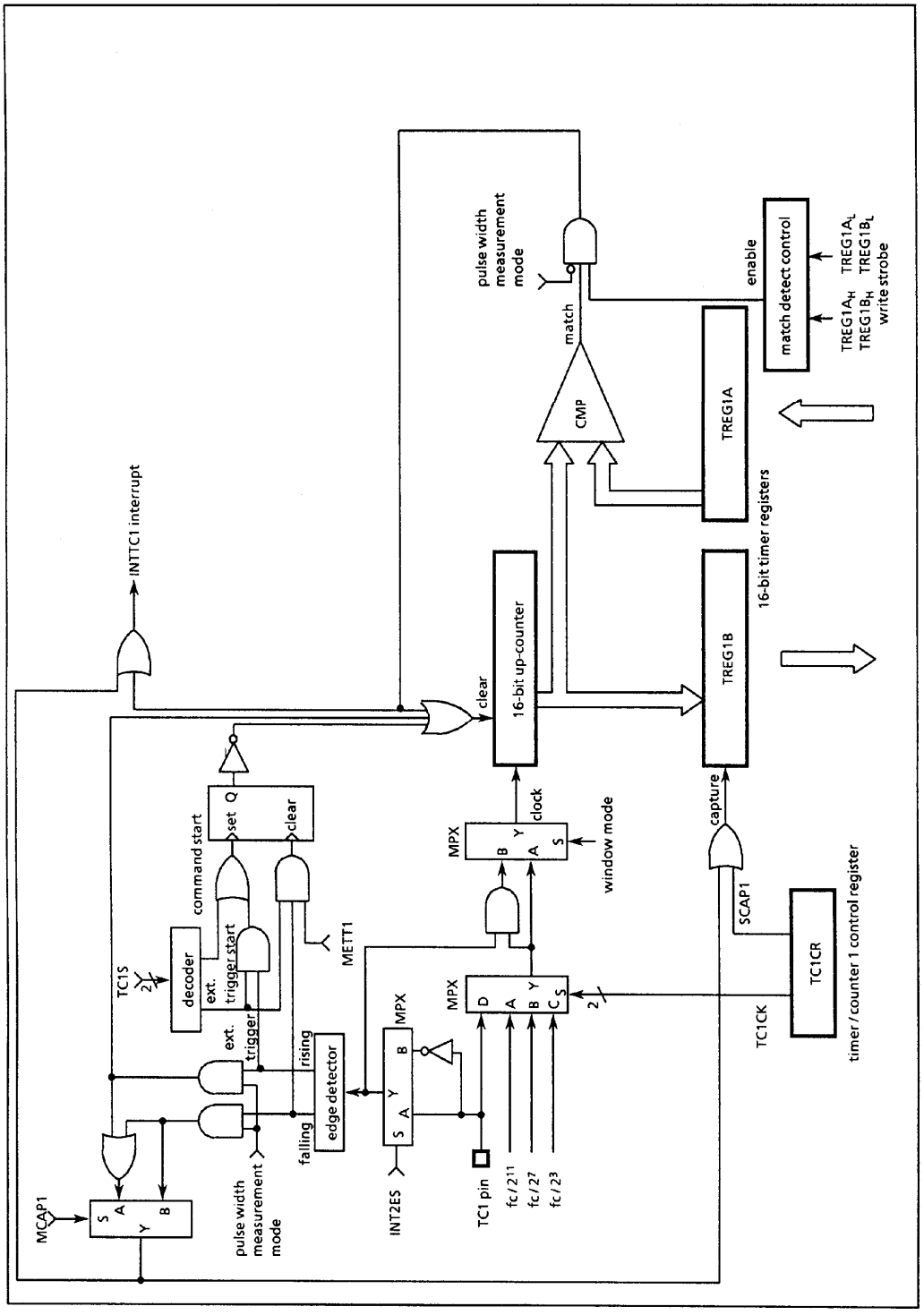


Figure 2-12. Timer / counter 1

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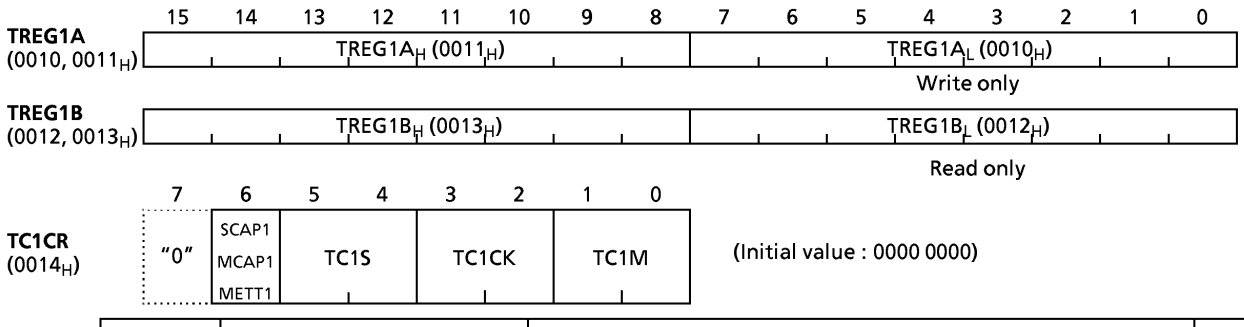
## 2.4.2 Control

The timer / counter 1 is controlled by a timer/counter 1 control register (TC1CR) and two 16-bit timer registers (TREG1A and TREG1B). Reset does not affect the TREG1A and TREG1B.




| | | | |
|---|---|---|---|
| TC1M | TC1 mode select | 00 : timer / external trigger timer / event counter mode<br>01 : window mode<br>10 : pulse width measurement mode<br>11 : reserved | write only |
| TC1CK | TC1 source clock select | 00 : internal clock fc/$2^{11}$ [Hz]<br>01 : internal clock fc/$2^{7}$<br>10 : internal clock fc/$2^{3}$<br>11 : external clock (TC1 pin input) | |
| TC1S | TC1 start control | 00 : stop & counter clear<br>01 : command start<br>10 : reserved<br>11 : external trigger start | |
| SCAP1 | software capture control | 0 : – 1 : software capture trigger (Note 3) | |
| MCAP1 | pulse width measurement control | 0 : double edge capture 1 : single edge capture | |
| METT1 | external trigger timer control | 0 : trigger srart 1 : trigger start & stop | |

*Note 1 : fc ; High-frequency clock [Hz]*

*Note 2 : Writing to the low-byte of the timer registers (TREG1A$_L$, TREG1B$_L$), the comparison is inhibited until the high-byte (TREG1A$_H$, TREG1B$_H$) is written.*

*Note 3 : Set the mode, source clock, edge, PPG control and timer F/F control when TC1 stops (TC1S = 00).*

*Note 4 : Software capture can be used in only timer and event counter modes.*

*Note 5 : Values to be loaded to timer registers must satisfy the following condition.*

*TREG1A > 0*

*Note6 : TC1CR is write-only register and must not be used with any of the read-modify-write instructions such as SET, CLR, etc. TC1CR must be used with load instructions.*

figure 2-13. Timer registers and TC1 control register

## 2.4.3 Function

Timer/counter 1 has five operating modes: timer, external trigger timer, event counter, window, pulse width measurement mode.

(1) **Timer** mode

In this mode, counting up is performed using the internal clock. The contents of the timer register 1A (TREG1A) are compared with the contents of the up-counter. If a match is found, an INTTC1 interrupt is generated, and the counter is cleared to"0". Counting up resumes after the counter is cleared. The current contents of the up-counter can be transferred to the timer register 1B (TREG1B) by setting SCAP1 (bit 6 in TC1CR) to "1" (software capture function). SCAP1 is automatically cleared to "0" after capturing.

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Table 2-1. Timer 1 source clock (Internal clock)

| Source clock | Resolution ( At fc = 8 MHz) | Maximum time setting (At fc = 8 MHz) |
|---|---|---|
| fc / $2^3$ [Hz] | 1 $\mu$s | 65.535 ms |
| fc / $2^7$ | 16 $\mu$s | 1.04856 s |
| fc / $2^{11}$ | 256 $\mu$s | 16.77696 s |

Example 1 : Sets the source clock to fc/$2^7$[Hz] and generates an interrupt 1 [s] later (at fc = 8 MHz).

```
LD      (TC1CR), 00000100B      ; Sets the TC1 source clock
LDW     (TREG1A), 0F424H        ; Sets the timer register (1 s ÷ fc / 2^7 = F424H)
SET     (EIRL). EF4             ; Enables INTTC1 interrupt
EI
LD      (TC1CR), 00010100B      ; Starts TC1
```

Example 2 : Software capture

```
LD      (TC1CR), 01010100B      ; SCAP1←1 (Captures)
LD      WA, (TREG1B)            ; Reads captured value
```

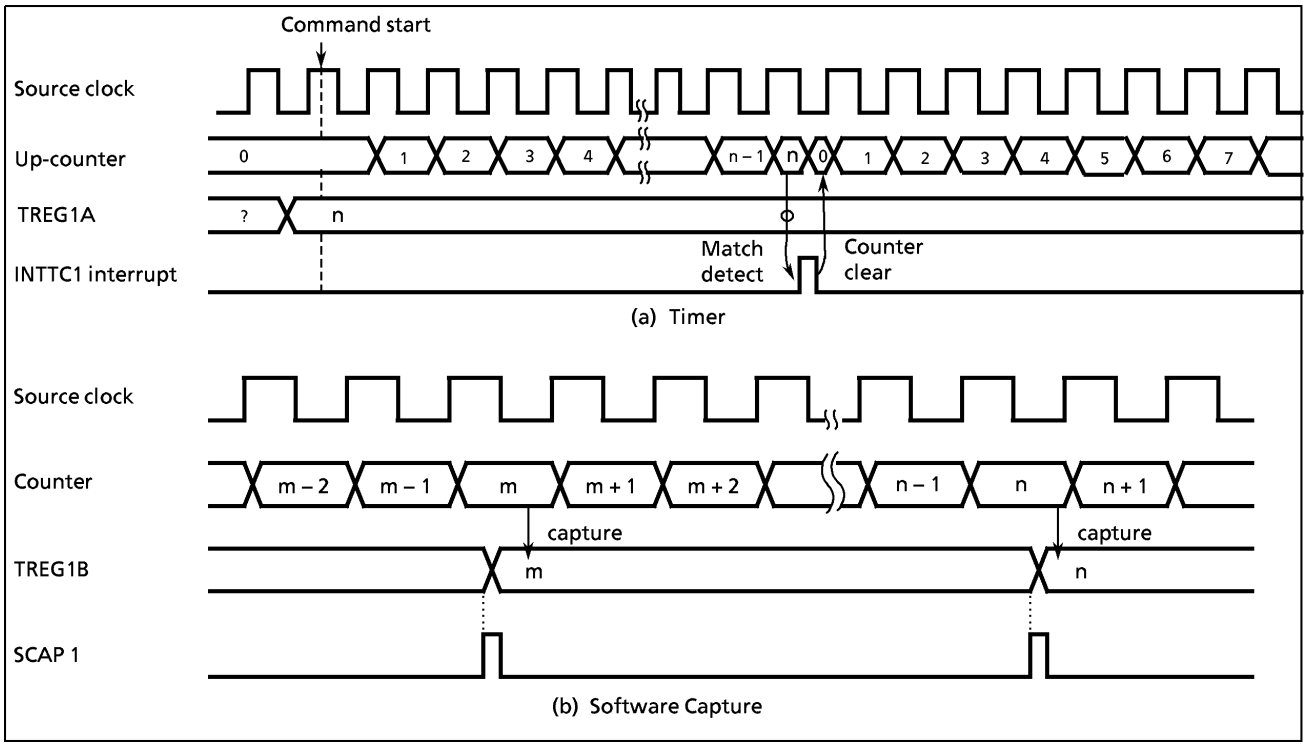


Figure 2-14. Timer mode timing chart

(2) **External trigger timer** mode

This is the timer mode to start counting up by the external trigger. The trigger is the edge of the TC1 pin input. Either rising or falling edge can be selected with INT2ES. Edge selection is the same as for the external interrupt input INT2 pin. Source clock is used an internal clock selected with the TC1CK. The contents of the TREG1A is compared with the contents of the up-counter. If a match is found, an INTTC1 interrupt is generated, and the counter is cleared to"0" and halted. The counter is restarted by the selected edge of the TC1 pin input.

The TC1 pin input has the same noise rejection as the INT2 pin; therefore, pulses of 7/fc [s] or less are eliminated as noise. A pulse width of 24/fc [s] or more is required for edge detection in the NORMAL or IDEL mode.

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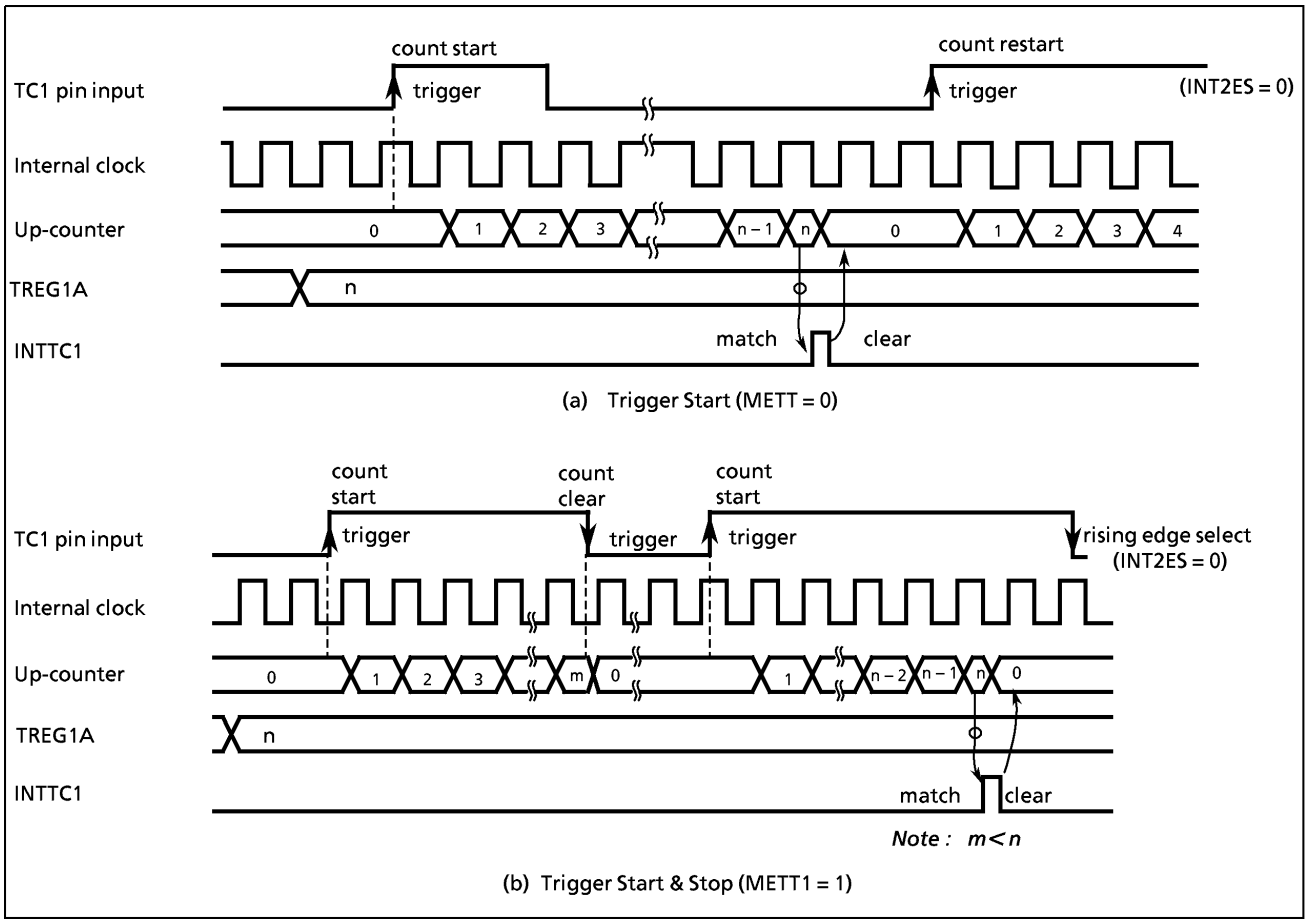


Figure 2-15. External trigger timer mode timing chart

(3) **Event counter** mode

In this mode, events are counted at the edge of the TC1 pin input. Either rising or falling edge can be selected with INT2ES in EINTCR. The contents of the TREG1A are compared with the contents of the up-counter. If a match is found, an INTTC1 interrupt is generated, and the counter is cleared. The maximum applied frequency is fc/$2^4$ [Hz] in the NORMAL or IDLE mode.

Setting SCAP1 to "1" transferres the current contents of the up-counter to the TREG1B (software capture function).

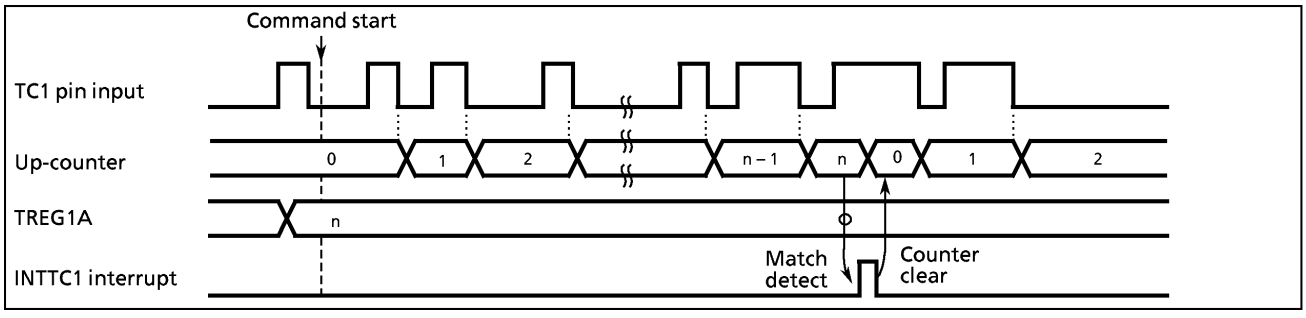


Figure 2-16. Event counter mode timing chart (INT2ES = 1)

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(4) **Window** mode

Counting up is performed at the rising edge of the pulse that is the logical AND-ed product of the TC1 pin input (window pulse) and an internal clock. The contents of the TREG1A are compared with the contents of the up-counter. If a match is found, an INTTC1 interrupt is generated, and the counter is cleared. Positive or negative logic for the TC1 pin input can be selected with INT2ES. Setting SCAP1 to "1" transferres the current contents of the up-counter to the TREG1B. It is necessary that the maximum applied frequency (TC1 input) be such that the counter value can be analyzed by the program. That is, the frequency must be considerably slower than the selected internal clock.

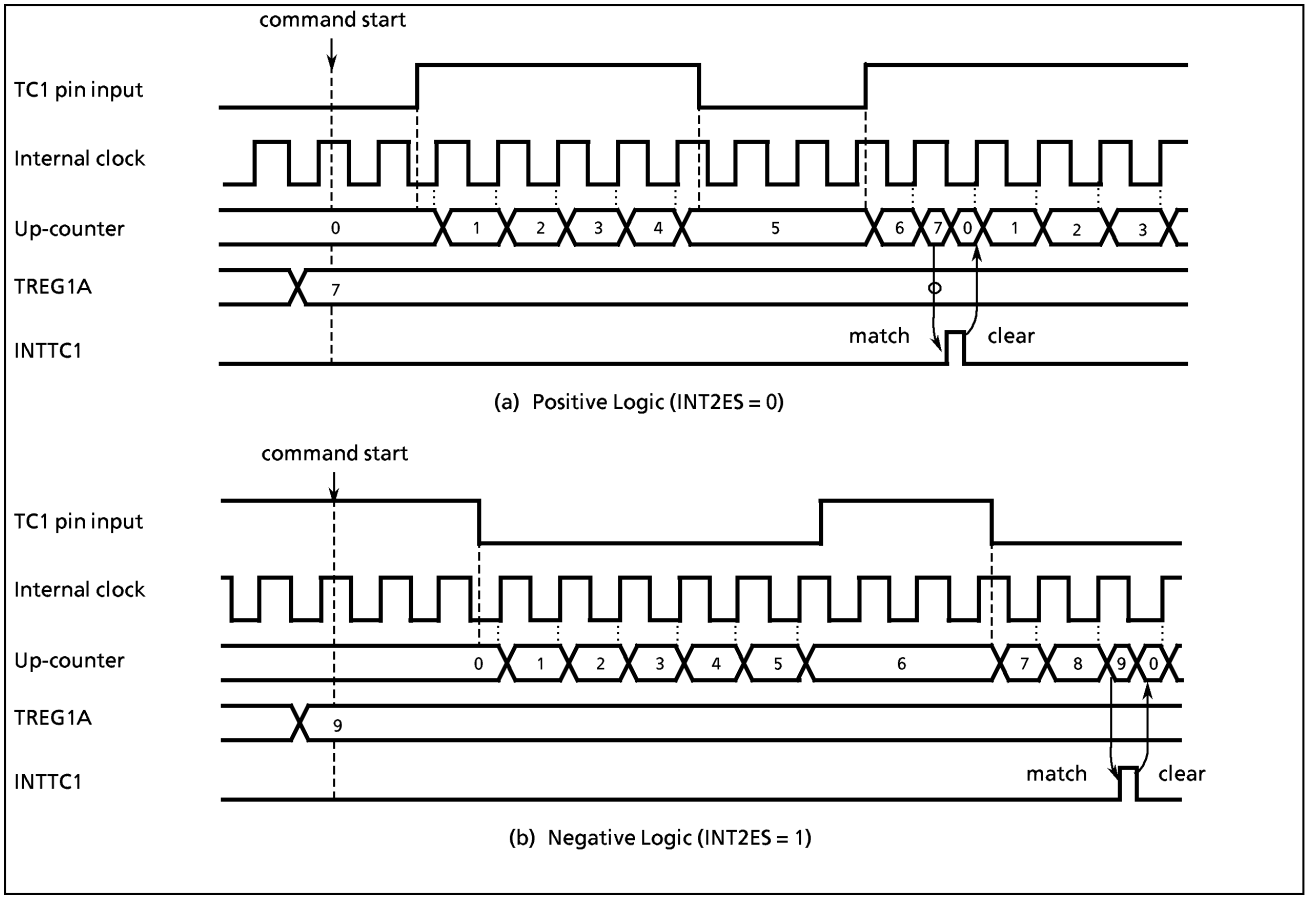


Figure 2-17. Window mode timing chart

(5) **Pulse width measurement** mode

Counting is started by the external trigger (set to external trigger start by TC1S). The trigger is selected either rising or falling edge of the TC1 pin input. The source clock is used an internal clock. At the next falling (rising) edge, the counter contents are transferred to the TREG1B and an INTTC1 interrupt is generated. The counter is cleared when single edge capture mode is set. When double edge capture is set, the counter continues and, at the next rising (falling) edge, the counter contents are again transferred to the TREG1B. If a falling (rising) edge capture value is required, it is necessary to read out the TREG1B contents until a rising (falling) edge is detected. Falling or rising edge is selected with INT2ES, and single edge or double edge is selected with MCAP1 ( bit 6 in TC1CR).

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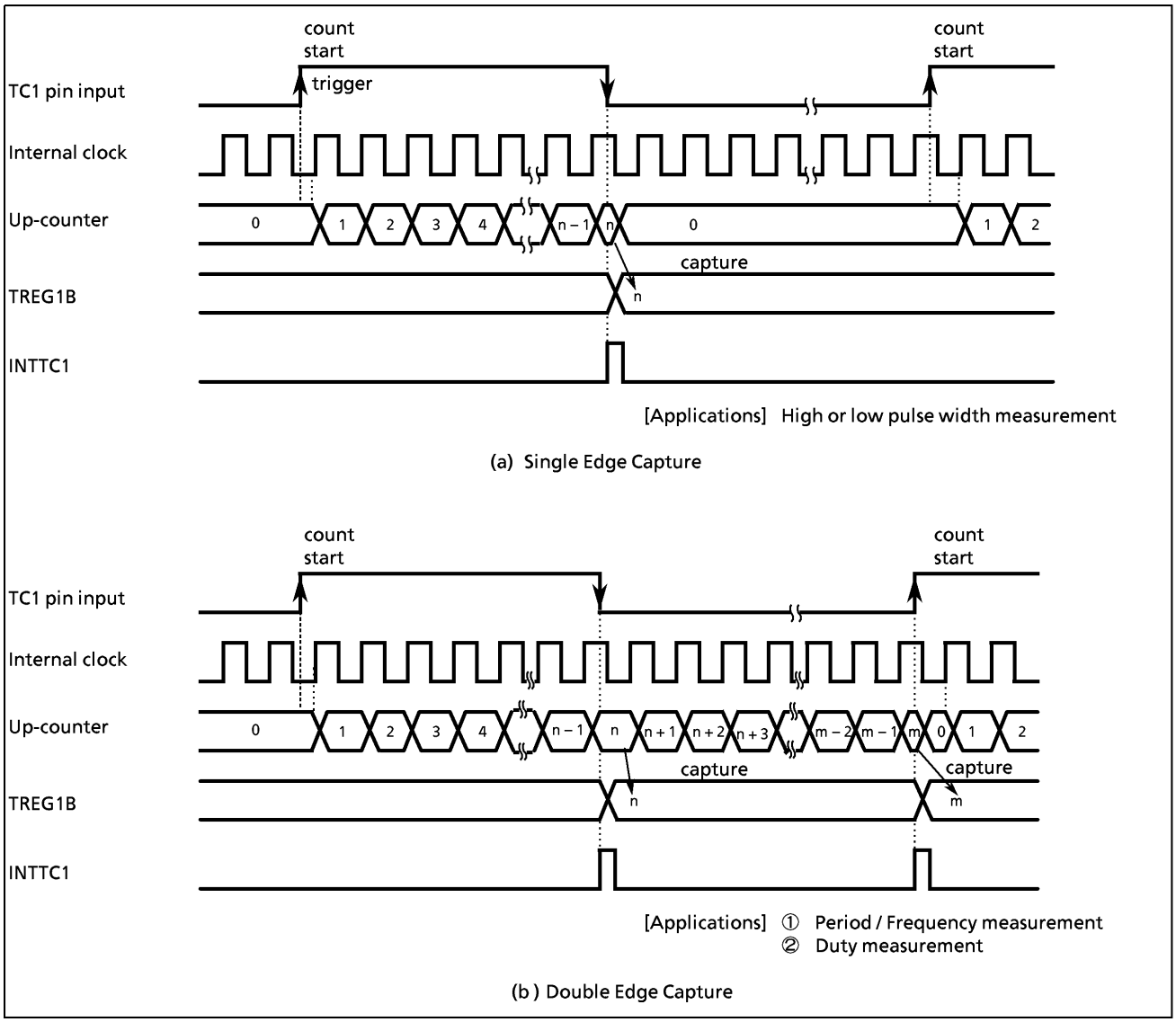


Figure 2-18. Pulse width measurement mode timing chart

Example : Duty measurement (Resolution $fc/2^7$ [Hz])

```
                CLR     (INTTC1C). 0            ;   INTTC1 service switch initial setting
                LD      (EINTCR), 00000000B     ;   Sets the rise edge at the INT2 edge
                LD      (TC1CR), 00000110B      ;   Sets the TC1 mode and source clock
                SET     (EIRL). 4               ;   Enables INTTC1
                LD      (TC1CR), 00110110B      ;   Starts TC1 with an external trigger
                 :
PINTTC1 :       CPL     (INTTC1C). 0            ;   Complements INTTC1 service switch
                JRS     F, SINTTC1
                LD      (HPULSE), (TREG1BL)     ;   Reads TREG1B
                LD      (HPULSE + 1), (TREG1BH)
                RETI
SINTTC1 :       LD      (WIDTH), (TREG1BL)      ;   Reads TREG1B (Period)
                LD      (WIDTH + 1), (TREG1BH)
                 :
```

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## 2.5 16-bit Timer / Counter 2 (TC2)

### 2.5.1 Configuration

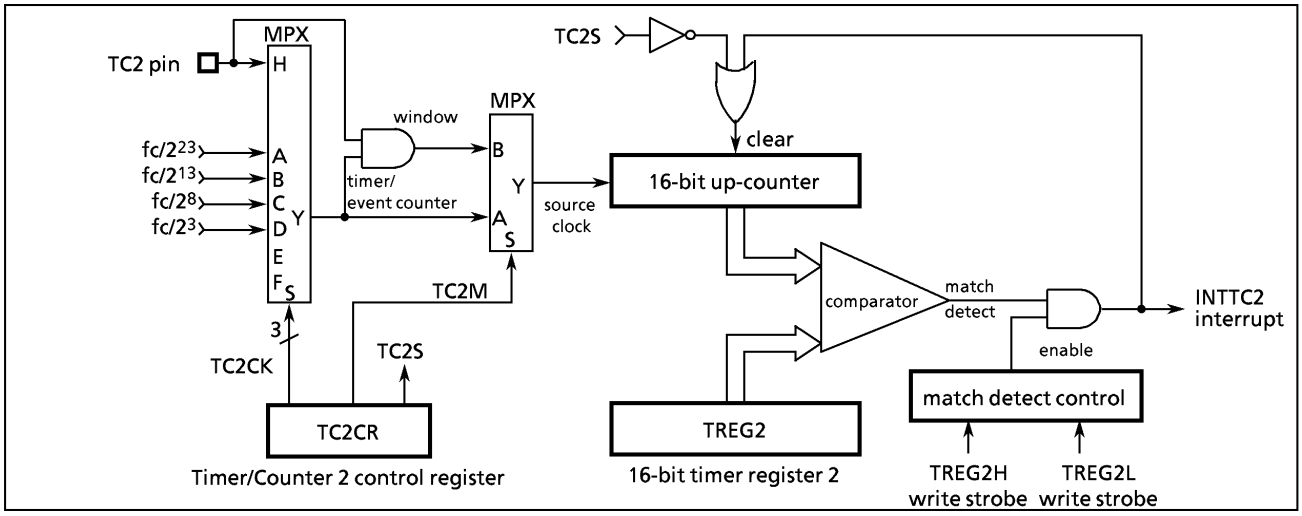


Figure 2-19. Timer/counter 2 (TC2)

### 2.5.2 Control

The timer / counter 2 is controlled by a timer/counter 2 control register (TC2CR) and a 16-bit timer register 2 (TREG2). Reset does not affect the TREG2.

**TREG2** (0016, 0017$_H$)

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|
| TREG2$_H$ (0017$_H$) | | | | | | | | TREG2$_L$ (0016$_H$) | | | | | | | |

write only

**TC2CR** (0015$_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| | | TC2S | TC2CK | | | | TC2M |

(Initial value : **00 00*0)

| | | | |
|---|---|---|---|
| TC2M | Timer/counter 2 operating mode select | 0 : Timer/Event counter mode<br>1 : Window mode | write only |
| TC2CK | Timer/counter 2 source clock select | 000 : Internal clock fc / $2^{23}$ [Hz]<br>001 : 〃 fc / $2^{13}$<br>010 : 〃 fc / $2^{8}$<br>011 : 〃 fc / $2^{3}$<br>100 : Reserved<br>101 : Reserved<br>110 : Reserved<br>111 : External clock (TC2 pin input) | |
| TC2S | Timer/counter 2 start control | 0 : Stop and counter clear<br>1 : Start | |

*Note 1 : fc; High-frequency clock [Hz], *; don't care*

*Note 2 : When writing to the low-byte of the timer register 2 (TREG2$_L$), the comparison is inhibited until the high-byte (TREG2$_H$) is written.*
*After writing to the high-byte, any match during 1 machine cycle (instruction execution cycle) is ignored.*

*Note 3 : Set the mode and source clock when timer/counter stop (TC2S = 0).*

*Note 4 : Values to be loaded to the timer register must satisfy the following condition.*
*TREG2 > 0 (TREG2$_{15\ to\ 11}$ > 0 when warm-up).*

*Note 5 : The TC2CR and the TREG2 are write-only registers and must not be used with any of the read-modify-write instructions such as SET, CLR, etc.*

Figure 2-20. Timer register 2 and TC2 control register

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## 2.5.3 Function

The timer/counter 2 has three operating modes: timer, event counter and window modes.

(1) **Timer** mode

In this mode, the internal clock is used for counting up. The contents of the timer register 2 (TREG2) are compared with the contents of the up-counter. If a match is found, a timer/counter 2 interrupt (INTTC2) is generated, and the counter is cleared. Counting up is resumed after the counter is cleared.

Table 2-2. Source clock (Internal clock) for timer 2

| Source clock | Resolution (At fc = 8 MHz) | Maximum time setting (At fc = 8 MHz) |
|---|---|---|
| $fc/2^{23}$ [Hz] | 1.048576 s | 19 hour 5 min 18.4 s |
| $fc/2^{13}$ | 1.024 ms | 1 min 7.1 s |
| $fc/2^{8}$ | 32 $\mu$s | 2.09712 s |
| $fc/2^{3}$ | 1 $\mu$s | 65.535 ms |

Example : Sets the source clock $fc/2^3$ [Hz] and generates an interrupt every 25ms (at fc = 8 MHz).

```
LD      (TC2CR), 00001100B      ; Sets the source clock
LDW     (TREG2), 61A8H          ; Sets TREG2 (25 ms ÷ 2^3/fc = 61A8H)
SET     (EIRH). EF14            ; Enables INTTC2 interrupt
EI
LD      (TC2CR), 00101100B      ; Starts TC2
```

(2) **Event counter** mode

In this mode, events are counted at the rising edge of the TC2 pin input. The contents of TREG2 are compared with the contents of the up-counter. If a match is found, an INTTC2 interrupt is generated, and the counter is cleared. The maximum frequency applied to the TC2 pin is $fc/2^4$ [Hz] in the NORMAL and IDLE mode.

Example : Sets the event counter mode and generates an INTTC2 interrupt 640 counts later.

```
LD      (TC2CR), 00011100B      ; Sets the TC2 mode
LDW     (TREG2), 640            ; Sets TREG2
LD      (TC2CR), 00111100B      ; Starts TC2
```

(3) **Window** mode

In this mode, counting up is performed at rising edge of the pulse that is the logical AND-ed product of the TC2 pin input (window pulse) and an internal clock. The internal clock is selected with the TC2CK. The contents of the TREG2 are compared with the contents of the up-counter. If a match is found, an INTTC2 interrupt is generated, and the up-counter is cleared to "0". It is necessary that the maximum applied frequency (TC2 input) be such that the counter value can be analyzed by the program. That is, the frequency must be considerably slower than the selected internal clock.

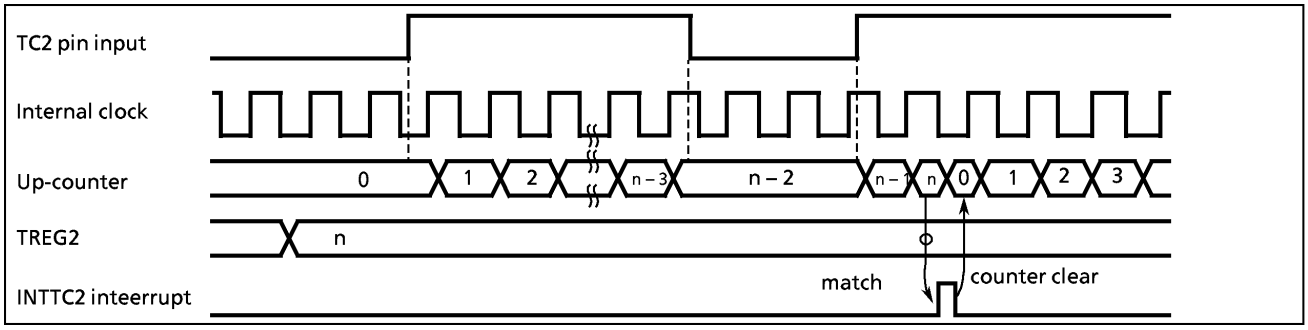


Figure 2-21. Window mode timing chart

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## 2.6 8-Bit Timer / Counter 3 (TC3)

### 2.6.1 Configuration

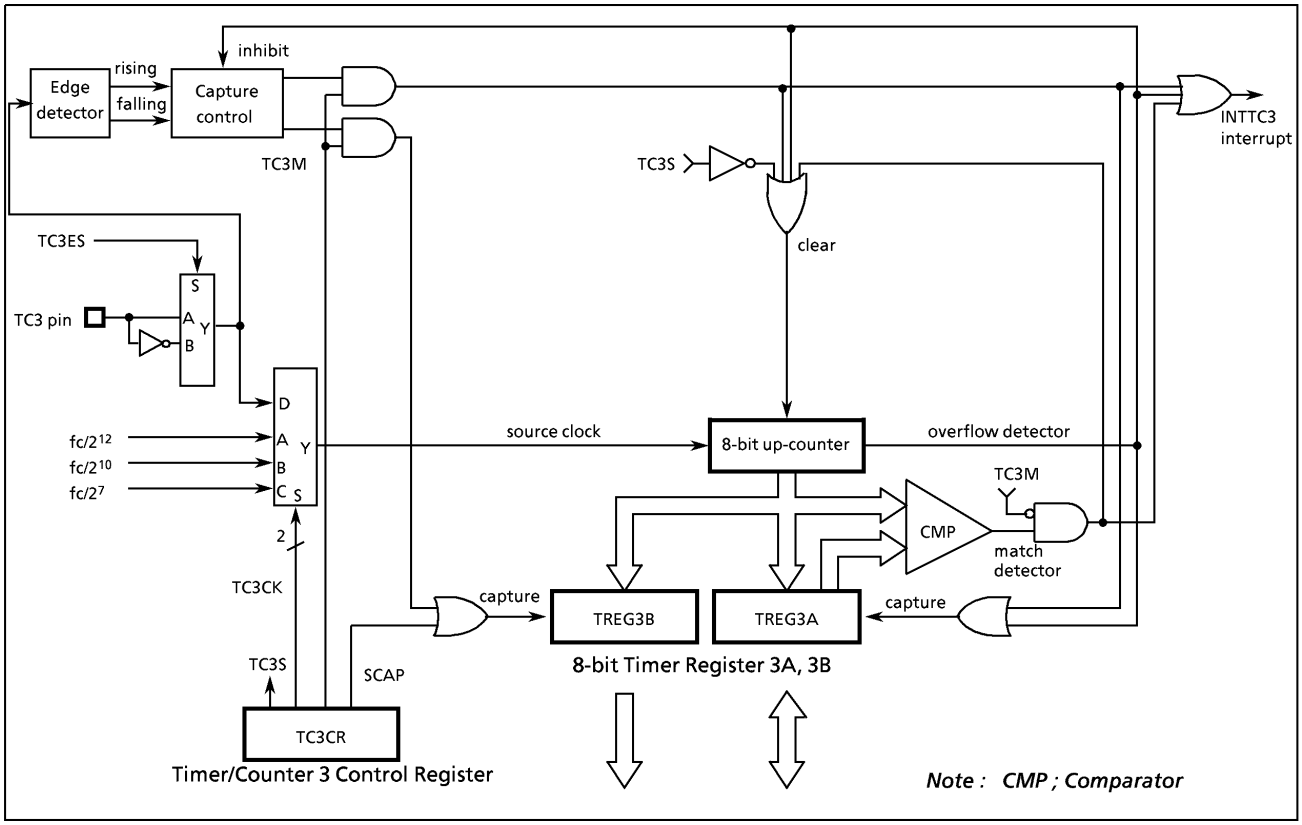


Figure 2-22. Timer/counter (TC3)

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## 2.6.2 Control

The timer / counter 3 is controlled by a timer / counter 3 control register (TC3CR) and two 8-bit timer registers (TREG3A and TREG3B). Reset does not affect these timer registers.

TREG3A ($0018_H$) — 7 6 5 4 3 2 1 0 — Read / Write

TREG3B ($0019_H$) — Read only

TC3CR ($001A_H$) — 7 6 5 4 3 2 1 0 — SCAP | TC3S | TC3CK | TC3M — (Initial value : *0*0 00*0)

| | | | |
|---|---|---|---|
| TC3M | Timer/counter 3 operation mode select | 0 : Timer / event counter mode<br>1 : Capture mode | write only |
| TC3CK | Timer/counter 3 source clock select | 00 : Internal clock fc / $2^{12}$ [Hz]<br>01 : Internal clock fc / $2^{10}$<br>10 : Internal clock fc / $2^{7}$<br>11 : External clock (TC3 pin input) | |
| TC3S | Timer/counter 3 start control | 0 : Stop and counter clear<br>1 : Start | |
| SCAP | Software capture control | 0 : –<br>1 : Software capture | |

EINTCR ($0037_H$) — 7 6 5 4 3 2 1 0 — "0" | (INT0EN) | (TC4ES) | (INT4ES) | TC3ES | (INT2ES) | "0" | "0" — (Initial value : *000 000*)

| | | | |
|---|---|---|---|
| TC3ES | TC3 edge select | 0 : Rising edge<br>1 : Falling edge | R/W |

*Note 1 : fc ; High-frequency clock [Hz], * ; don't care*
*Note 2 : Set the mode, the source clock and the edge selection (TC3ES) when the TC3 stops (TC3S = 0).*
*Note 3 : Values to be loaded to timer register 3A must satisfy the following condition.*
*TREG3A > 0 (in the timer/event counter mode)*
*Note 4 : TC3CR and EINTCR is a write-only-register and must not be used with any of read-modify-write instructions.*
*Note 5 : Always write "0" to bit 7, 6, 2, 1 in EINTCR.*

Figure 2-23. Timer register 3 and TC3 control registers

## 2.6.3 Function

The timer / counter 3 has three operating modes : timer, event counter, and capture mode.

(1) **Timer** mode

In this mode, the internal clock shown in Table 2-3 is used for counting up. The contents of TREG3A are compared with the contents of the up-counter. If a match is found, a timer / counter 3 interrupt (INTTC3) is generated, and the up-counter is cleared. Counting up resumes after the up-counter is cleared. The current contents of up-counter are loaded into TREG3B by setting SCAP (bit 6 in TC3CR) to "1". SCAP is automatically cleared after capturing.

Table 2-3. Source clock (internal clock) for timer / counter 3

| Source clock | Resolution (AT fc = 8 MHz) | Maximum setting time (AT fc = 8 MHz) |
|---|---|---|
| fc / $2^{12}$ | 512 μs | 130.56 ms |
| fc / $2^{10}$ | 128 μs | 32.64 ms |
| fc / $2^{7}$ | 16 μs | 4.08 ms |

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(2) **Event counter** mode

In this mode, the TC3 pin input pulse are used for counting up. Either the rising or falling edge can be selected with TC3ES (bit 3 in EINTCR). The contents of TREG3A are compared with the contents of the up-counter. If a match is found, an INTTC3 interrupt is generated and the counter is cleared. The maximum applied frequency is fc/$2^4$ [Hz]. Two or more machine cycles are required for both the high and low levels of the pulse width.

The current contents of up-counter are loaded into TREG3B by setting SCAP (bit 6 in TC3CR) to "1". SCAP is automatically cleared after capturing.

Example : Generetes an interrupt every 0.5 s, inputting 50Hz pulses to the TC3 pin.

```
LD      (TC3CR), 00001100B      ; Sets TC3 mode and source clock
LD      (TREG3A), 19H           ; 0.5 [s] ÷ 1 / 50 = 25 = 19H
SET     (EIRH).EF8              ; Enables INTTC3 itnerrupt
EI
LD      (TC3CR), 00011100B      ; Starts TC3
```

(3) **Capture** mode

The pulse width, period and duty of the TC3 pin input are measured in this mode, which can be used in decoding the remote control signal, etc. The counter is running free by the internal clock. On the rising (falling) edge of the TC3 pin input, the current contents of counter is loaded into TREG3A, then the up-counter is cleared to "0" and an INTTC3 interrupt is generated. On the falling (rising) edge of the TC3 pin input, the current contents of the counter is loaded into TREG3B. In this case, counting continued. On the next rising (falling) edge of the TC3 pin input, the current contents of counter are loaded into TREG3A, then the counter is cleared again and an interrupt is generated. If the counter overflows before the edge is detected, $FF_H$ is set to the TREG3A and an overflow interrupt (INTTC3) is generated. During interrupt processing, it can determine whether or not there is an overflow by checking whether or not the TREG3A value is $FF_H$. Also, after an interrupt (capture to TREG3A, or overflow detection) is generated, capture and overflow detection are halted until TREG3A has been read out; however, the counter continues. After TREG3A has been read out, capture and overflow detection start again. Therefore, TREG3B must be read out earlier than TREG3A.

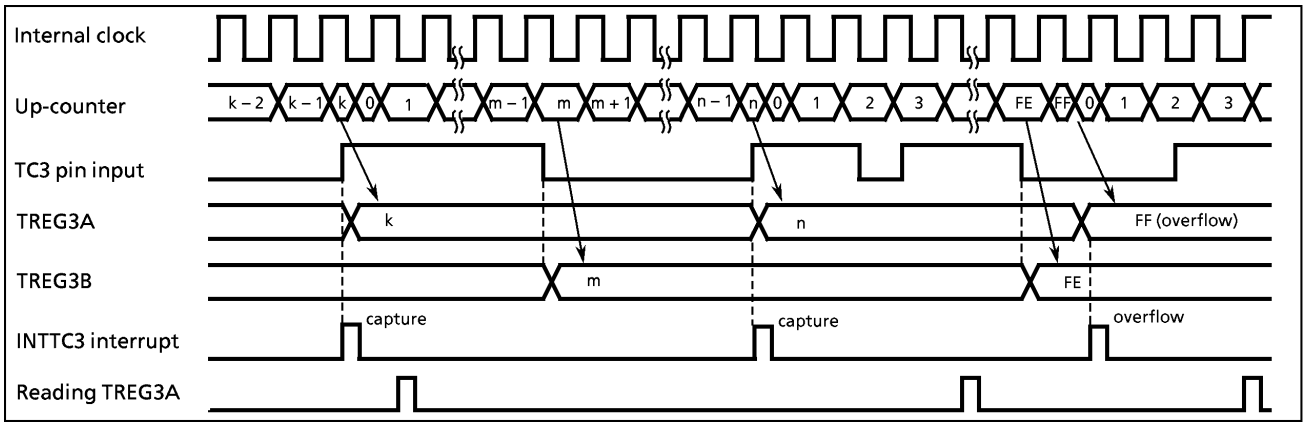


Figure 2-24. Timing chart for capture mode (TC3ES = 0)

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## 2.7 8-bit Timer / Counter (TC4)

### 2.7.1 Configuration

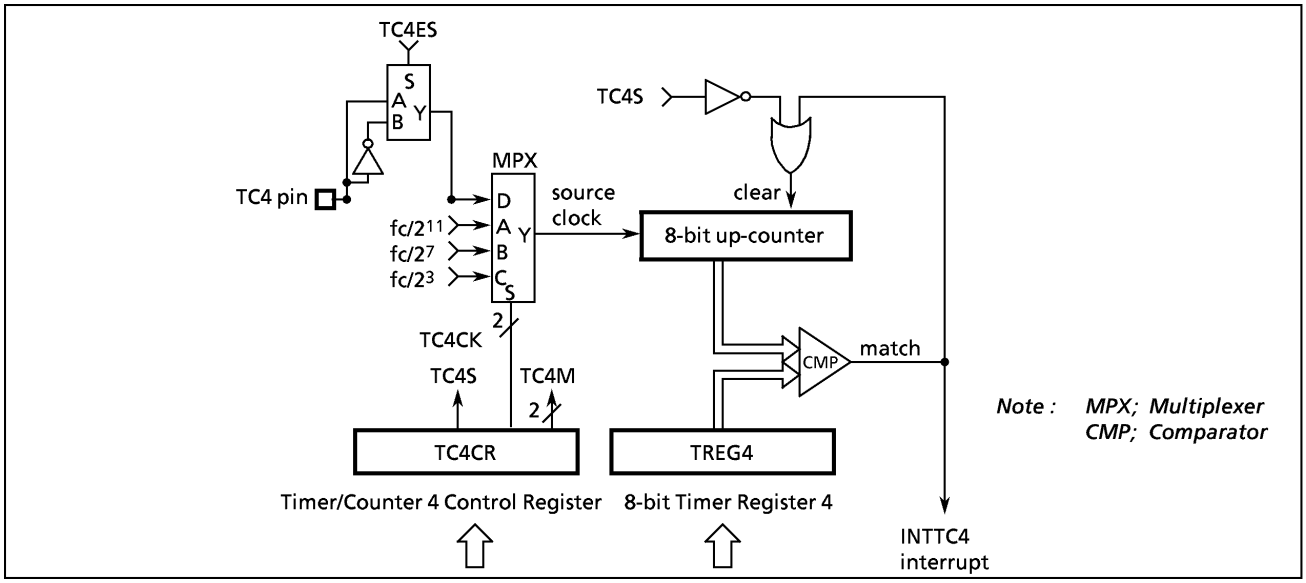


Figure 2-25. Timer / counter 4

### 2.7.2 Control

The timer / counter 4 is controlled by a timer / counter 4 control register (TC4CR) and an 8-bit timer register 4 (TREG4). Reset does not affect the TREG4.

**TREG4** ($001B_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| | | | | | | | |

Write only

**TC4CR** ($001C_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| "1" | "1" | "0" | TC4S | TC4CK | | TC4M | |

(Initial value : 00*0 0000)

| | | | |
|---|---|---|---|
| TC4M | TC4 operating mode select | 00 : Timer / event counter mode<br>01 : Reserved<br>1* : Reserved | write only |
| TC4CK | TC4 source clock select | 00 : Internal clock $fc/2^{11}$ [Hz]<br>01 : Internal clock $fc/2^7$<br>10 : Internal clock $fc/2^3$<br>11 : External clock (TC4 pin input) | |
| TC4S | TC4 start control | 0 : Stop and counter clear<br>1 : Start | |

**EINTCR** ($0037_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| "0" | (INT0EN) | TC4ES | (INT4ES) | (TC3ES) | (INT2ES) | "0" | "0" |

(Initial value : *000 000*)

| | | | |
|---|---|---|---|
| TC4ES | TC4 edge select | 0 : Rising edge<br>1 : Falling edge | R/W |

*Note 1 : fc; High-frequency clock [Hz], *; don't care*

*Note 2 : Set the operating mode, the source clock selection and the edge selection (TC4ES) when the TC4 stops (TC4S = 0).*

*Note 3 : Always write "1" to bit 7 and bit 6 in TC4CR.*

*Note 4 : Values to be loaded to the timer register must satisfy the following condition.*
*TREG4 > 0*

*Note 5 : TC4CR and the TREG4 are write-only registers and must not be used with any of the read-modify-write instructions such as SET, CLR, etc.*

*Note 6 : Always write "0" to bit5 in TC4CR.*

Figure 2-26. Timer register 4 and TC4 control registers

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### 2.7.3 Function

The timer / counter 4 has two operating modes : timer and event counter mode.

(1) **Timer** mode

In this mode, the internal clock is used for counting up. The contents of TREG4 are compared with the contents of the up-counter. If a match is found, a timer/counter 4 interrupt (INTTC4) is generated and the counter is cleared. Counting up resumes after the counter is cleared.

Table 2-4. Source clock (internal clock) for timer/counter 4

| Source clock | Resolution (At fc = 8 MHz) | Maximum setting time (At fc = 8 MHz) |
|---|---|---|
| $fc / 2^{11}$ [Hz] | 256 $\mu$s | 65.28 ms |
| $fc / 2^{7}$ | 16 $\mu$s | 4.08 ms |
| $fc / 2^{3}$ | 1 $\mu$s | 255 $\mu$s |

(2) **Event counter** mode

In this mode, the TC4 pin input (external clock) pulse is used for counting up. The contents of TREG4 are compared with the contents of the up-counter. If a match is found, an INTTC4 interrupt is generated and the counter is cleared. Counting up resumes after the counter is cleared. The maximum applied frequency is $fc/2^{4}$ [Hz]. Two or more machine cycles are required for both the high and low levels of the pulse width.

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## 2.8 Serial Bus Interface (SBI-ver.A)

The 87CH38/K38 has a 1-channel serial bus interface which employs a clocked-synchronous 8-bit serial bus interface and an I²C bus.

The serial bus interface is connected to an external device through P35 (SDA0) / P52 (SDA1) and P34 (SCL0) / P51 (SCL1) in the I²C bus mode; and through P53 ($\overline{SCK1}$), P52 (SO1), and P51 (SI1) in the clocked-synchronous 8-bit SIO mode.

The serial bus interface pins are also used as the P3 / P5 port. When used as serial bus interface pins, set the P3 / P5 output latches of these pins to "1". When not used as serial bus interface pins, the P3 / P5 port is used as a normal I/O port.

I²C bus has no an arbitration function which is necessary when two or more master devices scramble for the bus control. In master mode, other devices which are connected on the same bus need be slave devices. (single master)

*Note : When a multi master I²C bus system operates in I²C bus mode of this serial bus interface circuit, there is a possibility that the following problems raise. I²C bus mode of this serial bus interface circuit should be used by a single master I²C bus system.*

1. *The SCL line is fixed to low level and transferring stops by the serial bus interface circuit. The other devices can not run on the SCL line. Thus the bus locks.*
2. *The SCL pin is pulled down to low level regardless of the state of the SCL line by the serial bus interface circuit. A period of high-level SCL clock pulse which other devices output is shortened. The minimum value of which the SCL clock holds high level is not satisfied, which is specified with the I2C bus standard.*

### 2.8.1 Configuration

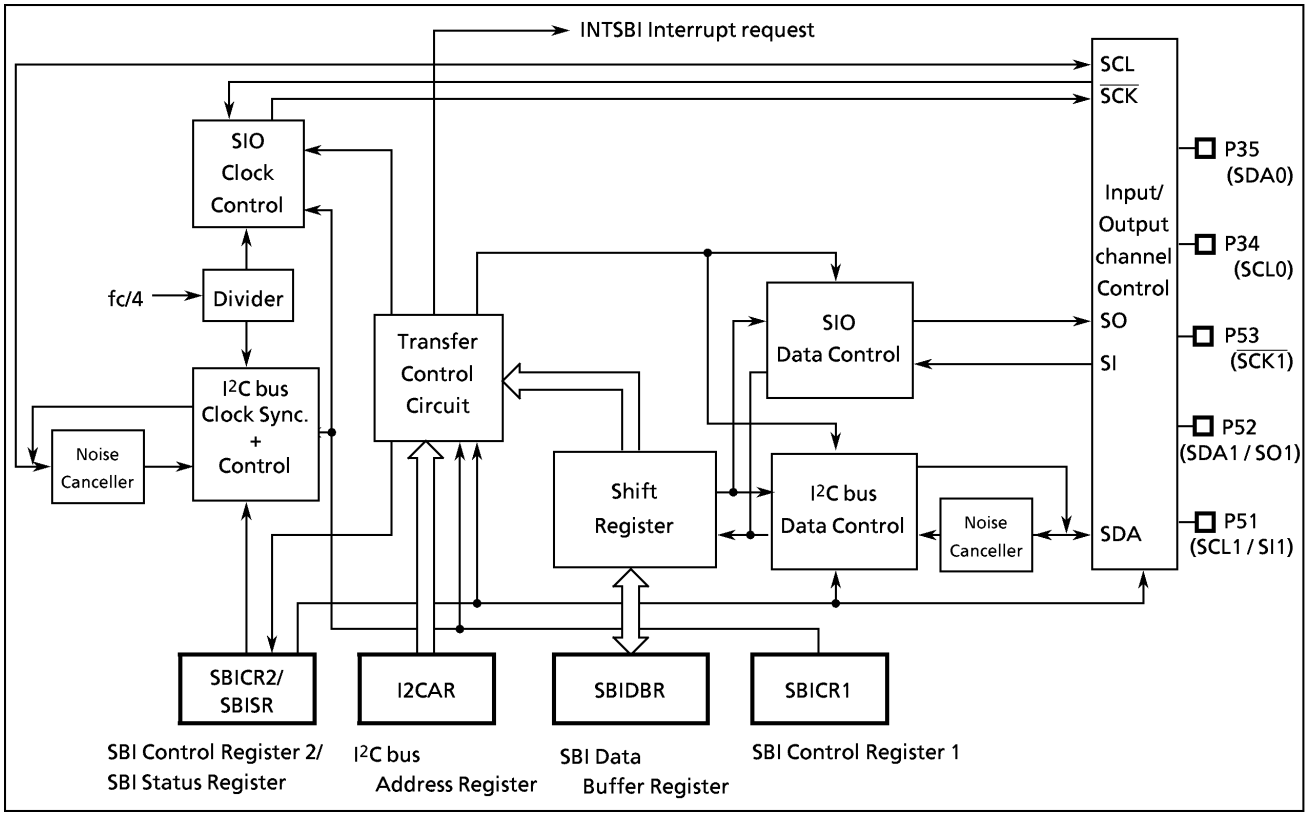


Figure 2-27. Serial bus interface (SBI-ver.A)

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### 2.8.2 Serial bus interface (SBI-ver.A) control

The following reginsters are used for control and operation status monitoring when using the serial bus interface (SBI-ver.A).

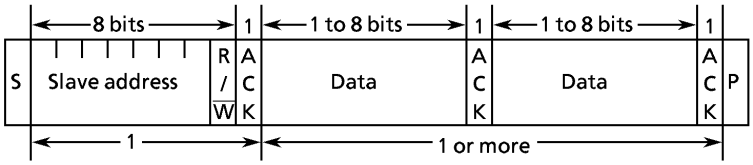
- Serial bus interface control register 1 (SBICR1)
- Serial bus interface control register 2 (SBICR2)
- Serial bus interface data buffer register (SBIDBR)
- $I^2C$ bus address register (I2CAR)
- Serial bus interface status register (SBISR)

The above registers differ depending on a mode to be used.
Refer to Section "2.8.4 $I^2C$ bus Mode Control" and "2.8.6 Clocked-synchronous 8-bit SIO Mode Control".

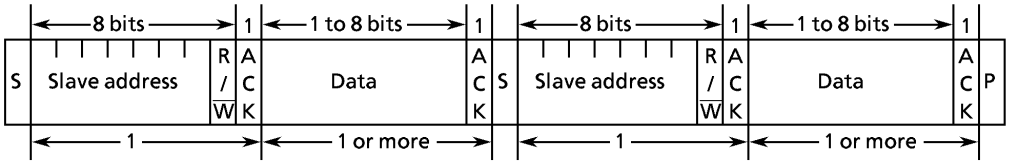
### 2.8.3 The data formats in the $I^2C$ bus mode

The data formats when using the serial bus interface circuit in the $I^2C$ bus mode are shown below.

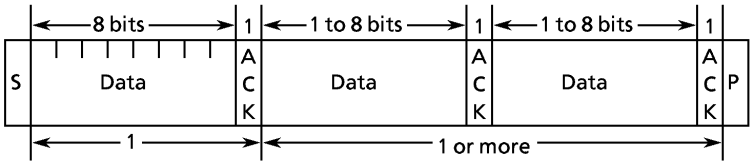
(a) Addressing format




(b) Addressing format (with restart)




(c) Free data format




*Notes* *S : Start condition*
*R/$\overline{W}$ : Direction bit*
*ACK : Acknowledge bit*
*P : Stop condition*

Figure 2-28. Data format

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## 2.8.4 I²C bus mode control

The following registers are used for control and operation status monitoring when using the serial bus interface (SBI-ver.A) in the I²C bus mode.

### Serial Bus Interface Control Register 1

SBICR1 ($0020_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| BC | | | ACK | CHS | SCK | | |

(Initial value: 0000 0000)

| | | BC | ACK = 0 | | ACK = 1 | | |
|---|---|---|---|---|---|---|---|
| | | | Number of Clock | Bits | Number of Clock | Bits | |
| BC | Number of transferred bits | 000 | 8 | 8 | 9 | 8 | Write only |
| | | 001 | 1 | 1 | 2 | 1 | |
| | | 010 | 2 | 2 | 3 | 2 | |
| | | 011 | 3 | 3 | 4 | 3 | |
| | | 100 | 4 | 4 | 5 | 4 | |
| | | 101 | 5 | 5 | 6 | 5 | |
| | | 110 | 6 | 6 | 7 | 6 | |
| | | 111 | 7 | 7 | 8 | 7 | |
| ACK | Acknowledge mode specification | 0 : Does not generate clock pulse for acknowledgment. (master mode) / Does not count clock pulse for acknowledgment. (slave mode)<br>1 : Generates clock pulse for acknowledgment. (master mode) / Counts clock pulse for acknowledgment. (slave mode) | | | | | Read/ Write |
| CHS | Input / Output channel selection | 0 : Channel0 (SCL0, SDA0 )<br>1 : Channel1 (SCL1, SDA1 ) | | | | | Read/ Write |
| SCK | Serial clock selection | 000 : 181.8 kHz<br>001 : 105.3 kHz<br>010 : 57.1 kHz<br>011 : 29.9 kHz<br>100 : 15.3 kHz<br>101 : 7.72 kHz<br>110 : 3.88 kHz<br>111 : reserved | at fc = 8 MHz (Output on SCL pin) | | | | Write only |

*Note 1 : fc ; high-frequency clock [Hz], * ; don't care*
*Note 2 : Set the BC to "000" before switching to a clock-synchronous 8-bit SIO mode.*
*Note 3 : SBICR1 has write-only register bits, which cannot access any of in read-modify-write instructions such as bit operate, etc.*
*Note 4 : Can not use SIO function when setting the CHS to "0".*

### Serial Bus Interface Data Buffer Register

SBIDBR ($0021_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| | | | | | | | |

(Initial value: **** ****) Read / Write

*Note 1 : When writing transmitted data, start from the MSB.*
*Note 2 : Cannot read the data which was written into SBIDBR, since a write data buffer and a read data buffer are independent in SBIDBR. Therefore, cannot access it any of in read-modify-write instructions such as bit operate, etc.*
*Note3 : A value written to SBIDBR is cleared to "0" by INTSBI interrupt request signal.*
*Note4 : * ; don't care*

### I²C bus Address Register

I2CAR ($0022_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| Slave address | | | | | | | ALS |
| SA6 | SA5 | SA4 | SA3 | SA2 | SA1 | SA0 | |

(Initial value: 0000 0000)

| | | | |
|---|---|---|---|
| SA | 87CH38/K38 slave address selection | | Write only |
| ALS | Address recognition mode specification | 0 : Slave address recognition<br>1 : Non slave address recognition | |

*Note : I2CAR is a write-only register, which cannot access any of in read-modify-write instructions such as bit operate, etc.*

Figure 2-29. Serial bus interface control register 1 / serial bus interface data buffer register/ I²C bus address register in the I²C bus mode

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Serial Bus Interface Control Register 2

SBICR2 ($0023_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| MST | TRX | BB | PIN | SBIM | | "0" | "0" |

(Initial value: 0001 00**)

| | | | |
|---|---|---|---|
| MST | Master / slave selection | 0 : Slave<br>1 : Master | Write only |
| TRX | Transmitter / receiver selection | 0 : Receiver<br>1 : Transmitter | |
| BB | Start / stop generation | 0 : Generate the stop condition when the MST, TRX, and PIN are "1".<br>1 : Generate the start condition when the MST, TRX, and PIN are "1". | |
| PIN | Cancel interrupt service request | 0 : –<br>1 : Cancel interrupt service request | |
| SBIM | Serial bus interface operating mode selection | 00 : Port mode (serial bus interface output disable)<br>01 : SIO mode<br>10 : $I^2C$ bus mode<br>11 : Reserved | |

*Note 1 : * ; don't care*
*Note 2 : Switch a mode to port mode after confirming that the bus is free.*
*Note 3 : Swich a mode to I2Cbus mode after confirming that input signals via port are high level.*
*Note 4 : SBICR2 has write-only register bits, which can not access any of in read-modify-write instructions such as bit operate, etc.*
*Note 5 : Clear bits 1 and 0 in SBICR2 to "0".*

Serial Bus Interface Status Register

SBISR ($0023_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| MST | TRX | BB | PIN | AL | AAS | AD0 | LRB |

(Initial value: 0001 0000)

| | | | |
|---|---|---|---|
| MST | Master / Slave selection status monitor | 0 : Slave<br>1 : Master | Read only |
| TRX | Transmitter / Receiver selection status monitor | 0 : Receiver<br>1 : Transmitter | |
| BB | Bus status monitor | 0 : Bus free<br>1 : Bus busy | |
| PIN | Interrupt service request status monitor | 0 : Requesting interrupt service<br>1 : Releasing interrupt service request | |
| AL | Noise detection monitor | 0 : Does not detect noise<br>1 : Detects noise | |
| AAS | Slave address match detection monitor | 0 : Does not detect slave address match or "GENERAL CALL"<br>1 : Detects slave address match or "GENERAL CALL" | |
| AD0 | "GENERAL CALL" detection monitor | 0 : Does not detect "GENERAL CALL"<br>1 : Detects "GENERAL CALL" | |
| LRB | Last received bit monitor | 0 : Last received bit is "0"<br>1 : Last received bit is "1" | |

Figure 2-30. Serial bus interface control register 2 / serial bus interface status register in the $I^2C$bus mode

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(1) **Acknowledgment mode specification**

Set the ACK (bit 4 in SBICR1) to "1" for operation in acknowledgment mode. When the serial bus interface circuit is the master mode, an additional clock pulse is generated for an acknowledge signal. In the transmitter mode during this additional clock pulse cycle, the SDA pin is released in order to receive the acknowledge signal from the receiver. In the receiver mode during this additional clock pulse cycle, the SDA pin is set to low level generating the acknowledge signal.

Clear the ACK to "0" for operation in the non-acknowledgment mode. When the serial bus interface circuit is the master mode, a clock pulse for the acknowledge signal is not generated.

In the acknowledgment mode, when the serial bus interface circuit is the slave mode, clocks are counted for the acknowledge signal. During the clock for the acknowledge signal, when a received slave address matches to a slave address set to the I2CAR or a "GENERAL CALL" is received, the SDA pin is set to low level generating an acknowledge signal.

After a received slave address matches to a slave address set to the I2CAR and a "GENERAL CALL" is received, in the transmitter mode during the clock for the acknowledge signal, the SDA pin is released in order to receive the acknowledge signal from the receiver. In the receiver mode, the SDA pin is set to low level generating an acknowledge signal.

In the non-acknowledgment mode, when the serial bus interface circuit is the slave mode, clocks for the acknowledge signal are not counted.

(2) **Number of transfer bits**

The BC (bits 7 to 5 in the SBICR1) is used to select a number of bits for next transmitting and receiving data.

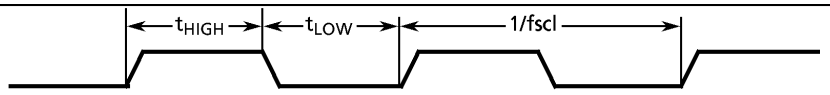
Since the BC is cleared to "000" by a start condition, a slave address and direction bit transmissions are executed in 8 bits. Other than these, the BC retains a specified value.

(3) **Serial clock**

a. Clock source

The SCK (bits 2 to 0 in the SBICR1) is used to select a maximum transfer frequency outputed on the SCL pin in the master mode.

Four or more machine cycles are required for both the high and low levels of the pulse width of a clock which is input externally in both the master and slave mode.




$t_{LOW} = 2^n/fc$

$t_{HIGH} = 2^n/fc + 12/fc$

$fscl = 1/(t_{Low} + t_{HIGH})$

*Note : fc ; high-frequency clock*

| SCK (bits2 to 0 in the SBICR1) | n |
|---|---|
| 000 | 4 |
| 001 | 5 |
| 010 | 6 |
| 011 | 7 |
| 100 | 8 |
| 101 | 9 |
| 110 | 10 |

Figure 2-31. Clock source

b. Clock synchronization

The $I^2C$ bus has a clock synchronization function to meet the transfer speed to a slow processing device when a transfer is performed between devices which have different process speed.

The clock synchronization functions when the SCL pin is high level and the SCL line of the bus is low level in the serial bus interface circuit. The serial bus interface circuit waits counting a clock pulse in high level until the SCL line of the bus is high level. When the SCL line of the bus is high level, the serial bus interface circuit starts counting during high level. The clock synchronization function holds clocks which are output from the serial interface circuit to be high level.

The slave device can stop the clock output of the master device on one word or one bit basis. Additionally, the transfer speed by the master device matches to the process speed of the slave device.

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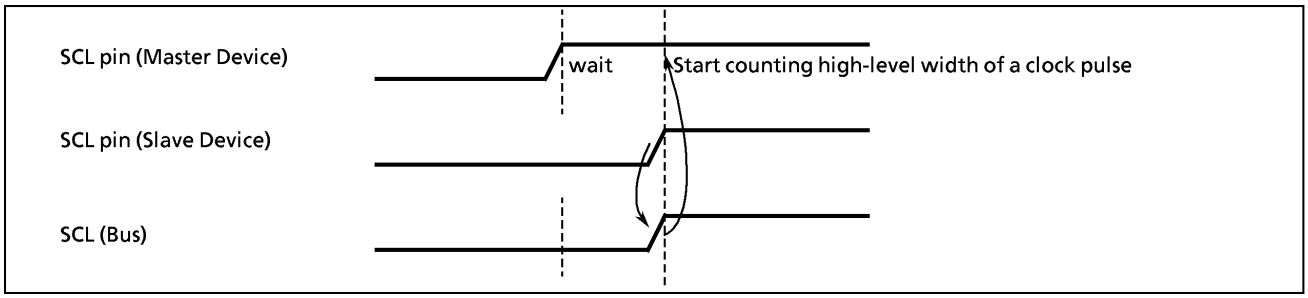


Figure 2-32. Clock synchronization

(4) **Slave address and address recognition mode specification**
To operate the serial bus interface circuit in the addressing format which recognizes the slave address, clear the ALS (bit 0 in I2CAR) to "0" and set the slave address to the SA (bits 7 to 1 in I2CAR).
To operate the serial bus interface circuit in the free data format which does not recognize the slave address, set the ALS to "1". When the serial bus interface circuit is used in the free data format, the slave address and the direction bit are not recognized. They are handled as data just after generation of start conditions.

(5) **Master/slave selection**
Set the MST (bit 7 in the SBICR2) to "1" for operating the serial bus interface as a master device. Clear the MST to "0" for operation as a slave device. The MST is cleared to "0" by the hardware after a stop condition on a bus is detected or the noise is detected.

(6) **Transmitter / receiver selection**
Set the TRX (bit 6 in the SBICR2) to "1" for operating the serial bus interface circuit as a transmitter. Clear the TRX to "0" for operation as a receiver. When data with an addressing format is transferred in the slave mode, the TRX is set to "1" by the hardware if the direction bit (R/W) sent from the master device is "1", and is cleared to "0" by the hardware if the bit is "0". In the master mode, after an acknowledge signal is returned from the slave device, the TRX is cleared to "0" by the hardware if a transmitted direction bit is "1", and is set to "1" by the hardware if it is "0". When an acknowledge signal is not returned, the current condition is maintained.
The TRX is cleared to "0" by the hardware after a stop condition on the bus is detected or the noise is detected.

The following shows TRX change conditions in each mode and TRX after changing.

| Mode | Direction bit | Change condition | TRX after changing |
|---|---|---|---|
| Slave mode | 0 | A received slave address is the same as a value set to I2CAR. | 0 |
| | 1 | | 1 |
| Master mode | 0 | ACK signal is returned. | 1 |
| | 1 | | 0 |

When the serial bus interface circuit operates in the free data format, the slave address and the direction bit are not recognized. They are handled as data just after generating a start condition. The TRX was not changed by the hardware.

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(7) **Start/stop condition generation**

When the BB (bit 5 in the SBICR2) is "0", the slave address and the direction bit which are set to the SBIDBR are output on a bus after generating a start condition by writing "1" to the MST, TRX, BB, and PIN. It is necessary to set transmitted data to the data buffer register (SBIDBR) and set "1" to ACK beforehand.

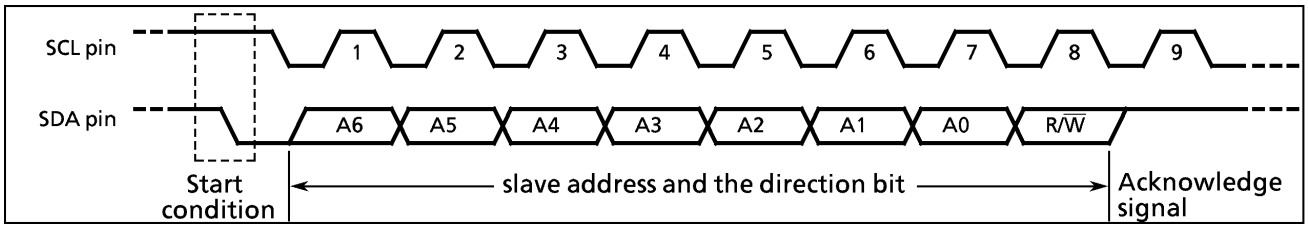


Figure 2-33. Start condition generation and slave address generation

When the BB is "1", a sequence of generating a stop condition is started by writing "1" to the MST, TRX, and PIN, and "0" to the BB. Do not modify the contents of MST, TRX, BB and PIN until a stop condition is generated on a bus.

When a stop condition is generated and the SCL line on the bus is set to low level by another device, a stop condition is generated after releasing the SCL line.

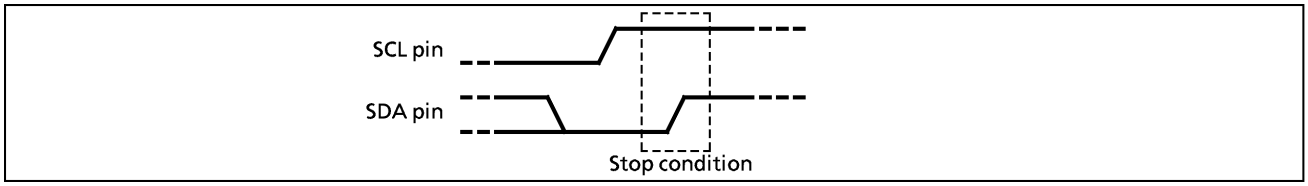


Figure 2-34. Stop condition generation

The bus condition can be indicated by reading the contents of the BB (bit 5 in the SBISR). The BB is set to "1" when a start condition on a bus is detected,and is cleared to "0" when a stop condition is detected on a bus.

(8) **Interrupt service request and cancel**

When the serial bus interface circuit is the master mode and transferring a number of clocks set by the BC and the ACK is complete, a serial bus interface interrupt request (INTSBI) is generated.

In the slave mode, the INTSBI is generated when the received slave address is the same as the value set to the I2CAR and an acknowledge signal is output, when a "GENERAL CALL" is received and an acknowledge signal is output, or when transferring / receiving data is complete after the received slave address is the same as the value set to the I2CAR and a "GENERAL CALL" is received.

When the serial bus interface interrupt request occurs, the PIN (bit 4 in the SBISR) is cleared to "0". During the time that the PIN is "0", the SCL pin is set to low level.

Either writing or reading data to or from the SBIDBR sets the PIN to "1".

The time from the PIN being set to "1" until the SCL pin is released takes $t_{LOW}$.

Although the PIN (bit 4 in the SBICR2) can be set to "1" by the program, the PIN is not cleared to "0" when it is written "0".

(9) **Serial bus interface operating mode selection**

The SBIM (bits 3 and 2 in the SBICR2) is used to specify the serial bus interface operation mode.

Set the SBIM to "10" when used in the $I^2C$ bus mode after confirming that the serial bus interface pin is high level. Switch a mode to port after confirming that the bus is free.

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(10) **Noise detection monitor**

The $I^2C$ bus is easy to be affected by noise, because the bus is driven by the open drain and the pull-up resistor.

With the serial bus interface circuit, the SDA pin output and the SDA line level are compared at a rise of the SCL line on the bus, and whether data are output correctly on the bus is detected only in the master transmitter mode.

When the SDA pin output differs from the SDA line level, the AL (bit 3 in the SBISR) is set to "1". When the AL is set to "1", the SDA pin is released and the MST and the TRX are cleared to "0" by the hardware. The serial bus interface circuit changes to the slave receiver mode, and the serial bus interface circuit continues outputting clocks until transferring data when the AL was set to "1" is completed.

Either writing or reading data to or from the SBIDBR, or writing data to the SBICR2 clears to the AL to "0".

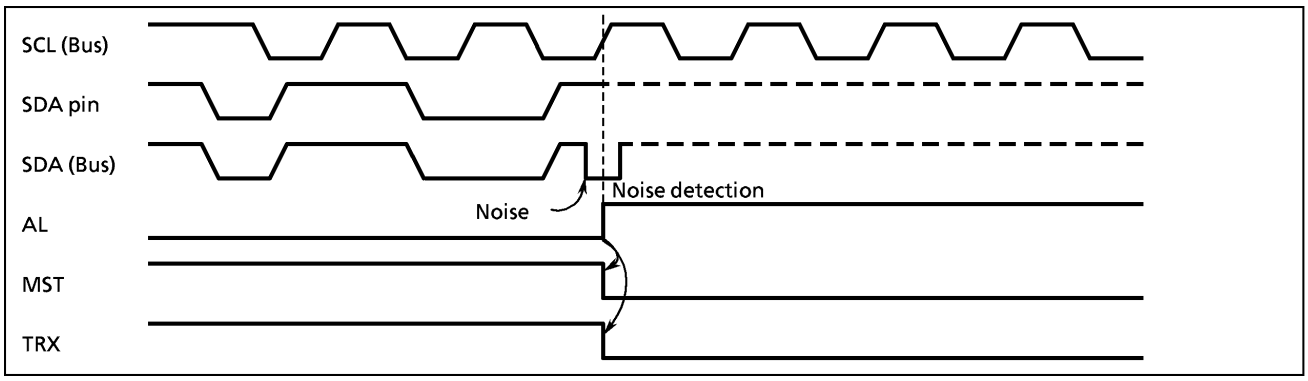


Figure 2-35. Noise detection monitor

(11) **Slave address match detection monitor**

The AAS (bit 2 in the SBISR) is set to "1" in the slave mode, in the address recognition mode (ALS = 0), when receiving "GENERAL CALL" or a slave address with the same value that is set to the I2CAR. When the ALS is "1", the AAS is set to "1" after receiving the first 1-word of data. The AAS is cleared to "0" by writing / reading data to / from a data buffer register.

(12) **GENERAL CALL detection monitor**

The AD0 (bit 1 in the SBISR) is set to "1" in the slave mode, when all 8-bit received data is "0", after a start condition (GENERAL CALL). The AD0 is cleared to "0" when a start or stop condition is detected on a bus.

(13) **Last received bit monitor**

The SDA value stored at the rising edge of the SCL is set to the LRB (bit 0 in the SBISR). In the acknowledge mode, immediately after an INTSBI interrupt request is generated, an acknowledge signal is read by reading the contents of the LSB.

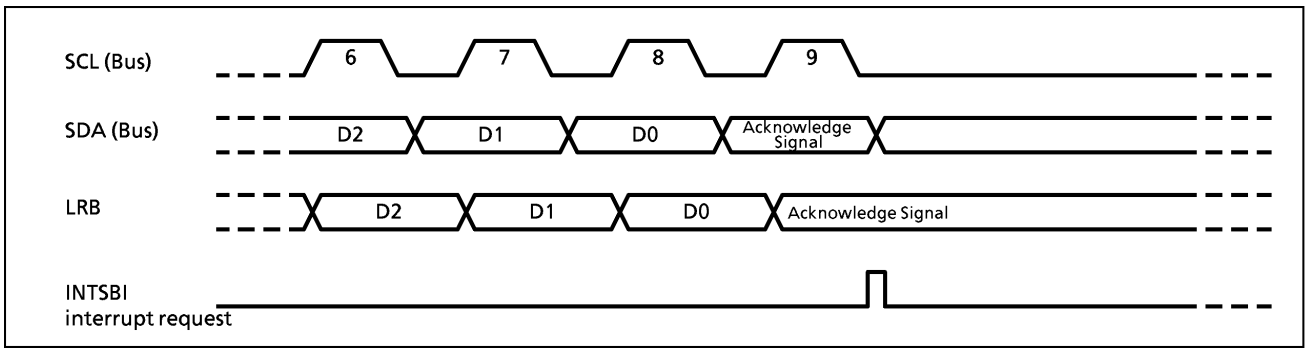


Figure 2-36. Last received bit monitor

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### 2.8.5 Data transfer in I²C bus mode

(1) **Device Initialization**

Set the ACK in the SBICR1 to "1", and the BC to 000. Specify the data length to 8 bits to count clocks for acknowledge. Set a transfer frequency to the SCK and a serial bus interface pin to the CHS.
Subsequently, set a slave address to the SA in the I2CAR and clear the ALS to "0" to set an addressing format.
After confirming that the serial bus interface pin is high-level, for specifying the default setting to a slave receiver mode, clear "0" to the MST, TRX, and BB in the SBICR2, set "1" to the PIN, "10" to the SBIM, and "0" to bits 1 and 0,

*Note : The initialization of the serial bus interface circuit must be complete within the time from all devices which are connected to the bus have initialized to any device does not generate a start condition. If not, there is a possibility that another device starts transferring before an end of the initialization of the serial bus interface circuit. Data can not be received correctly.*

(2) **Start condition and slave address generation**

Confirm a bus free status (when BB = 0).
Set the ACK to "1" and specify a slave address and a direction bit to be transmitted to the SBIDBR.
When the BB is "0", the start condition are generated and the slave address and the direction bit which are set to the SBIDBR are output on a bus by writing "1" to the MST, TRX, BB and PIN. An INTSBI interrupt request occurs at the 9th falling edge of the SCL clock cycle, and the PIN is cleared to "0". The SCL pin is pulled down to the low-level while the PIN is "0". When an interrupt request occurs, the TRX changes by the hardware according to the direction bit only when an acknowledge signal is returned from the slave device.

*Note 1 : Do not write a slave address to be output to the SBIDBR while data are transferred. If data is written to the SBIDBR, data to been outputting may be destroyed.*
*Note 2 : Do not start transferring due to another master from writing a slave address to be output to the SBIDBR to writing a start condition generation command to the SBICR2. The serial bus interface circuit malfunctions because it has not an arbitration function.*

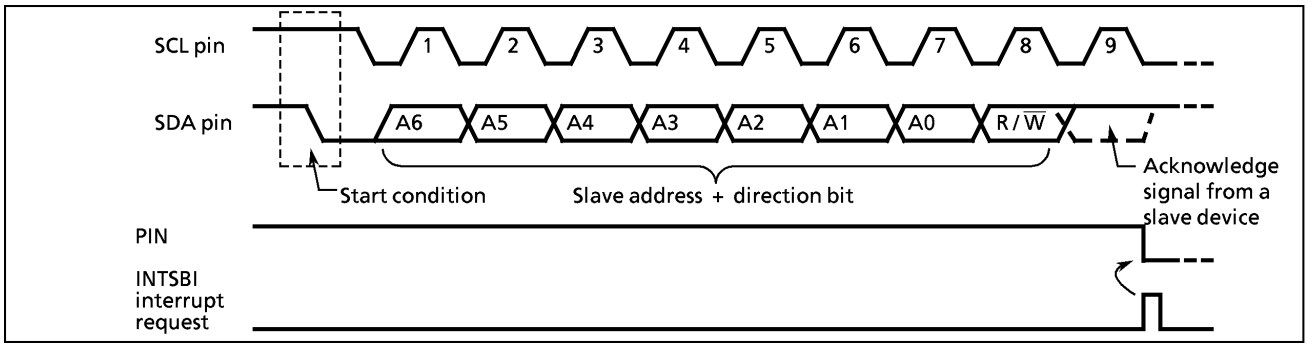


Figure 2-37. Start condition generation and slave address transfer

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(3) **1-word data transfer**

Check the MST by the INTSBI interrupt process after an 1-word data transfer is completed, and determine whether the mode is a master or slave.

a. When the MST is "1" (Master mode)

Check the TRX and determine whether the mode is a transmitter or receiver.

① When the TRX is "1" (Master mode)

Test the LRB. When the LRB is "1", a receiver does not request data. Implement the process to generate a stop condition (described later) and terminate data transfer.

When the LRB is "0", the receiver requests new data. When the next transmitted data is other than 8 bits, set the BC, set the ACK to "1", and write the transmitted data to the SBIDBR. After writing the data, the PIN becomes "1", a serial clock pulse is generated for transferring a new 1-word of data from the SCL pin, and then the 1-word data is transmitted. After the data is transmitted, and an INTSBI interrupt request occurs. The PIN becomes "0" and the SCL pin is set to low level. If the data to be transferred is more than one word in length, repeat the procedure from the LRB test above.

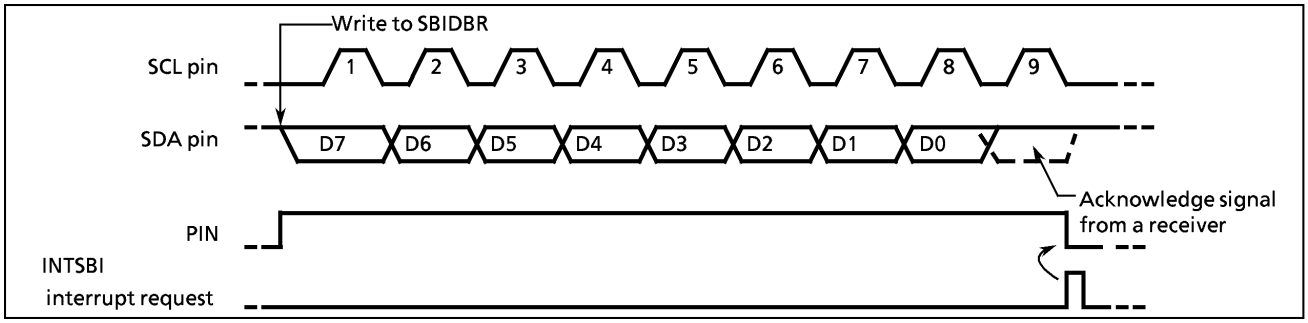


Figure 2-38. Example when BC = "000", ACK = "1" in transmitter mode

② When the TRX is "0" (Receiver mode)

When the next transmitted data is other than 8 bits, set the BC again. Set the ACK to "1" and read the received data from the SBIDBR (data which is read immediately after a slave address is sent is undefined). After the data is read, the PIN becomes "1". The serial bus interface circuit outputs a serial clock pulse to the SCL to transfer new 1-word of data and sets the SDA pin to "0" at the acknowledge signal timing.

An INTSBI interrupt request occurs and the PIN becomes "0". Then the serial bus interface circuit pulls down the SCL pin to the low level. The serial bus interface circuit outputs a clock pulse for 1-word of data transfer and the acknowledge signal each time that received data is read from the SBIDBR.

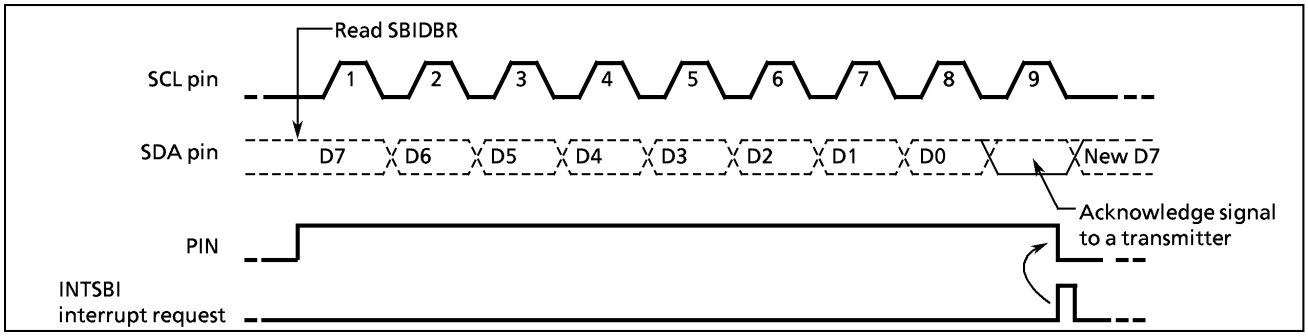


Figure 2-39. Example when BC = "000", ACK = "1" in receiver mode

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In order to terminate transmitting data to a transmitter, clear the ACK to "0" before reading data which is 1 word before the last data to be received. The last data does not generate a clock pulse for the acknowledge signal. After the data is transmitted and an interrupt request has occurred, set the BC to "001" and read the data. The serial bus interface circuit generates a clock pulse for a 1-bit data transfer. Since the master device is a receiver, the SDA line on a bus keeps the high level. The transmitter receives the high-level signal as an ACK signal. The receiver indicates to the transmitter that data transfer is complete.
After 1-bit data is received and an interrupt request has occurred, the serial bus interface circuit generates a stop condition (Refer to 2.8.5. (4) ) and terminates data transfer.

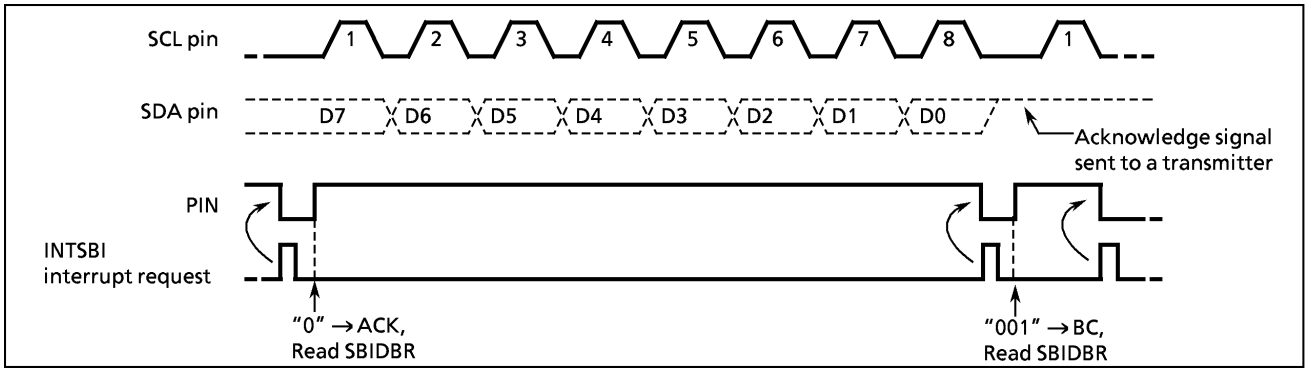


Figure 2-40. Termination of data transfer in master receiver mode

b. When the MST is "0" (Slave mode)

In the slave mode, the serial bus interface circuit operates either in normal slave mode or in recovery process after a noise detection.
In the slave mode, an INTSBI interrupt request occurs when the serial bus interface circuit receives a slave address or a "GENERAL CALL" from the master device, or when a "GENERAL CALL" is received and data transfer is complete after matching a received slave address. In the master mode, the serial bus interface circuit operates in a slave mode if a noise is detected. An INTSBI interrupt request occurs when word data transfer terminates after a noise detection. When an INTSBI interrupt request occurs, the PIN (bit 4 in the SBICR2) is reset, and the SCL pin is set to low level. Either reading or writing from or to the SBIDBR or setting the PIN to "1" releases the SCL pin after taking $t_{LOW}$ time.
The serial bus interface circuit tests the AL (bit 3 in the SBISR), the TRX (bit 6 in the SBISR), the AAS (bit 2 in the SBISR), and the AD0 (bit 1 in the SBISR) and implements processes according to conditions listed in the next table.

Table 2-5. Operation in the slave mode

| TRX | AL | AAS | AD0 | Conditions | Process |
|---|---|---|---|---|---|
| 1 | 0 | 1 | 0 | In the slave receiver mode, the serial bus interface circuit receives a slave address of which the value of the direction bit sent from the master is "1". | Set the number of bits in 1-word to the BC and write transmitted data to the SBIDBR. |
| | | 0 | 0 | In the slave transmitter mode, 1-word data is transmitted. | Check the LRB. If the LRB is set to "1", set the PIN to "1" since the receiver does not request next data. Then, clear the TRX to "0" release the bus. If the LRB is cleared to "0", set the number of bits in a word to the BC and write transmitted data to the SBIDBR since the receiver requests next data. |
| 0 | 1 | 0 | 0 | The serial bus interface circuit detects the noise when transmitting a slave address or data and terminates transferring word data. | There is a possibility that a serial bus interface circuit does not receive data normally. The recovery process such as a data re-transfer, etc. is needed. |
| | 0 | 1 | 1/0 | In the slave receiver mode, the serial bus interface circuit receives a slave address or GENERAL CALL of which the value of the direction bit sent from the master is "0". | Read the SBIDBR for setting the PIN to "1" (reading dummy data) or set the PIN to "1". |
| | | 0 | 1/0 | In the slave receiver mode, the serial bus interface circuit terminates receiving of 1-word data. | Set the number of bits in a word to the BC and read received data from the SBIDBR. |

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(4) **Stop condition generation**

When the BB is "1", a sequence of generating a stop condition is started by setting "1" to the MST, TRX and PIN, and "0" to the BB. Do not modify the contents of the MST, TRX, BB, PIN until a stop condition is generated on a bus. When a SCL line of bus is pulled down by other devices, the serial bus interface circuit generates a stop condition after they release a SCL line.

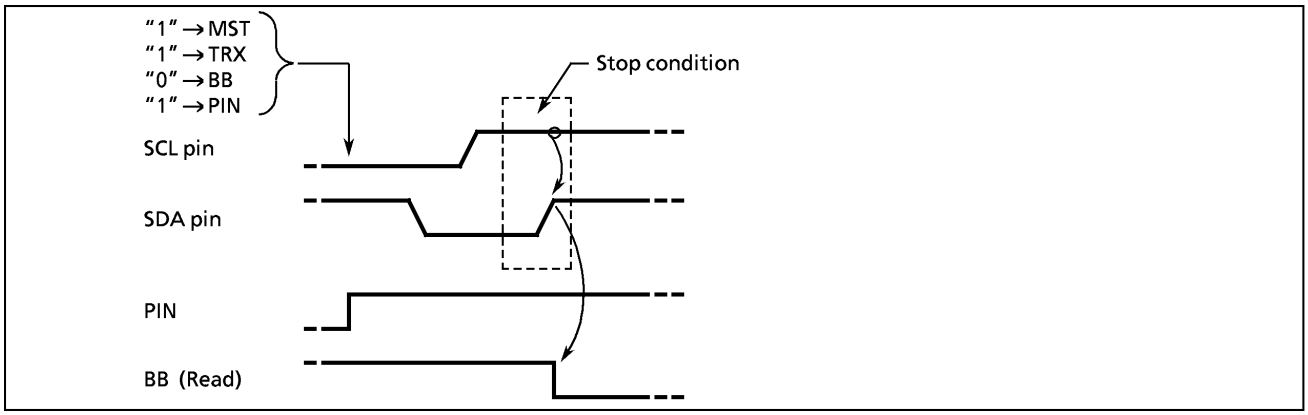


Figure 2-41. Stop condition generation

(5) **Restart**

Restart is used to change the direction of data transfer between a master device and a slave device during transferring data. The following explains how to restart the serial bus interface circuit.

Clear "0" to the MST, TRX, and BB and set "1" to the PIN. The SDA pin retains the high level and the SCL pin is released. Since a stop condition is not generated on the bus, the bus is assumed to be in a busy state from other devices. Test the BB until it becomes "0" to check that the SCL pin of the serial bus interface circuit is released. Test the LRB until it becomes "1" to check that the SCL line of the bus is not set to low level by other devices. After confirming that the bus stays in a free state, generate a start condition with procedure (2).

In order to meet setup time when restarting, take at least 4.7 $\mu$s of waiting time by software from the time of restarting to confirm that the bus is free until the time to generate the start condition.

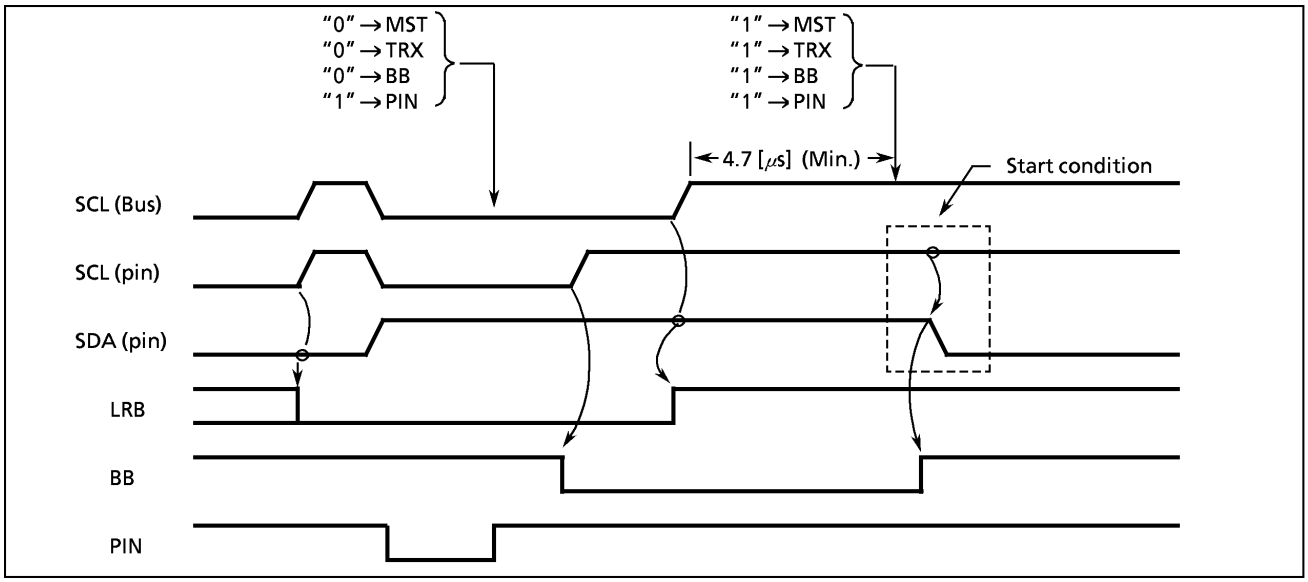


Figure 2-42. Timing diagram when restarting

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## 2.8.6 Clocked-synchronous 8-bit SIO mode control

The following registers are used for control and operation status monitoring when using the serial bus interface (SBI-ver.A) in the clocked-synchronous 8-bit SIO mode.

**Serial Bus Interface Control Register 1**

SBICR1 ($0020_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| SIOS | SIO INH | SIOM | | "1" | | SCK | |

(Initial value 0000 0000)

| | | | |
|---|---|---|---|
| SIOS | Indicate transfer start/stop | 0 : Stop<br>1 : Start | Write only |
| SIOINH | Continue/abort transfer | 0 : Continue transfer<br>1 : Abort transfer (automatically cleared after abort) | |
| SIOM | Transfer mode select | 00 : 8-bit transmit mode<br>01 : reserved<br>10 : 8-bit transmit/receive mode<br>11 : 8-bit receive mode | |
| SCK | Serial clock select | 000 : fc/$2^5$ ( 250 kHz)<br>001 : fc/$2^6$ ( 125 kHz)<br>010 : fc/$2^7$ ( 62.5 kHz)<br>011 : fc/$2^8$ (31.25 kHz)<br>100 : fc/$2^9$ (15.62 kHz)<br>101 : fc/$2^{10}$ ( 7.81 kHz)<br>110 : fc/$2^{11}$ ( 3.90 kHz)<br>111 : External clock (input from $\overline{SCK}$ pin)<br>} at fc = 8 MHz ($\overline{SCK}$ pin: Output on) | |

*Note 1 : fc ; high-frequency clock [Hz]*
*Note 2 : Clear the SISO to "0" and set the SIOINH to "1" when setting the transfer mode and serial clock.*
*Note 3 : SBICR1 is a write-only register, which cannot access any of in read-modify-write instructions such as bit operate, etc.*
*Note 4 : Set bit 3 to "1" in SBICR1 when SBI is used as SIO mode.*

**Serial Bus Interface Data Buffer Register**

SBIDBR ($0021_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| | | | | | | | |

(Initial value **** ****) Read / Write

*Note 1 : Cannot read the data which was written into SBIDBR, since a write data buffer and a read data buffer are independent in SBIDBR. Therefore, cannot access it any of in read-modify-write instructions such as bit operate, etc.*
*Note 2 : don't care*

**Serial Bus Interface Control Register 2**

SBICR2 ($0023_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| "0" | "0" | "0" | "1" | SBIM | | "0" | "0" |

(Initial value **** 00**)

| | | | |
|---|---|---|---|
| SBIM | Serial bus interface operation mode selection | 00 : Port mode (serial bus interface output disable)<br>01 : SIO mode<br>10 : $I^2C$ bus mode<br>11 : reserved | Write only |

*Note 1 : * ; don't care*
*Note 2 : Switch a mode to port after data transfer is complete.*
*Note 3 : Switch a mode to SIO mode after confirming that input signals via port are high level.*
*Note 4 : SBICR2 is a write-only regiter, which cannot access any of in read-modify-write instructions such as bit operate, etc.*
*Note 5 : Clear bits 7 to 5 in the SBICR2 to "0", and set bit 4 to "1".*

Figure 2-43-1. Serial bus interface control register 1 / serial bus interface data buffer register / serial bus interface control register 2 in SIO mode

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Serial Bus Interface Status Register

SBISR ($0023_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| "1" | "1" | "1" | "1" | SIOF | SEF | "1" | "1" |

| | | | |
|---|---|---|---|
| SIOF | Serial transfer operating status monitor | 0 : Transfer terminated<br>1 : Transfer in process | Read only |
| SEF | Shift operating status monitor | 0 : Shift operation terminated<br>1 : Shift operation in process | |

Figure 2-43-2. Serial bus interface status register in SIO mode

(1) Serial clock

a. Clock source

The SCK (bit 2 to 0 in the SBICR1) is used to select the following functions.

① Internal clock

In an internal clock mode, any of seven frequencies can be selected. The serial clock is output to the outside on the $\overline{SCK}$ pin. The $\overline{SCK}$ pin becomes a high-level when data transfer starts. When writing (in the transmit mode) or reading (in the receive mode) data cannot follow the serial clock rate, an automatic-wait function is executed to stop the serial clock automatically and hold the next shift operation until reading or writing is complete.

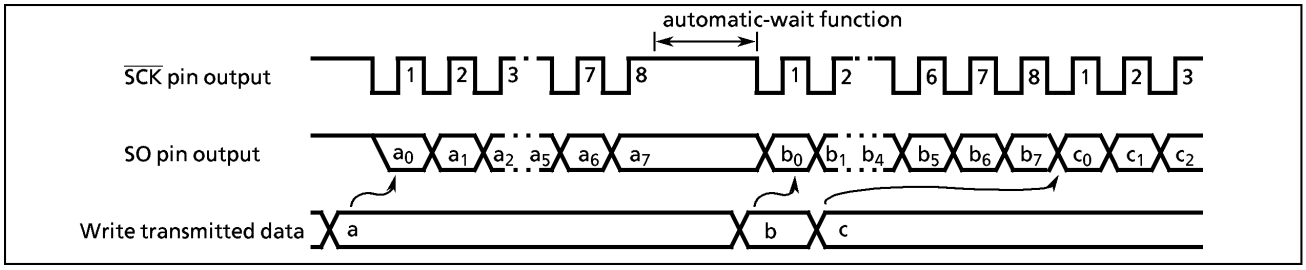


Figure 2-44. Automatic-wait function

② External clock (SCK = "111")

An external clock supplied to the $\overline{SCK}$ pin is used as the serial clock. In order to ensure shift operation, a pulse width of at least 4 machine cycles is required for both high-level and low-level in the serial clock. The maximum data transfer frequency is 250 kHz (when fc = 8 MHz).

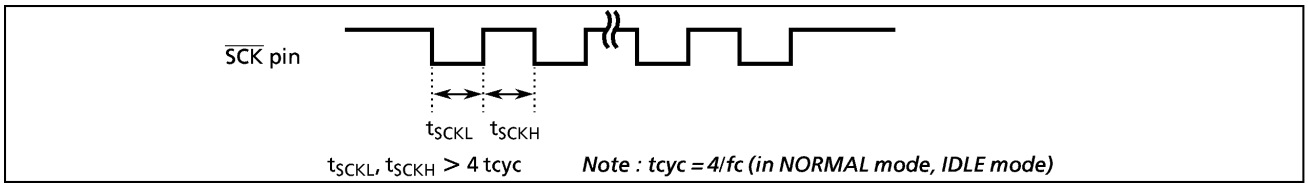


Figure 2-45. Maximum data transfer frequency when external clock input

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b. Shift edge

The leading edge is used to transmit data, and the trailing edge is used to receive data.

① Leading edge shift
Data is shifted on the leading edge of the serial clock (at a falling edge of the $\overline{SCK}$ pin input/output).

② Trailing edge shift
Data is shifted on the trailing edge of the serial clock (at a rising edge of the $\overline{SCK}$ pin input/output).

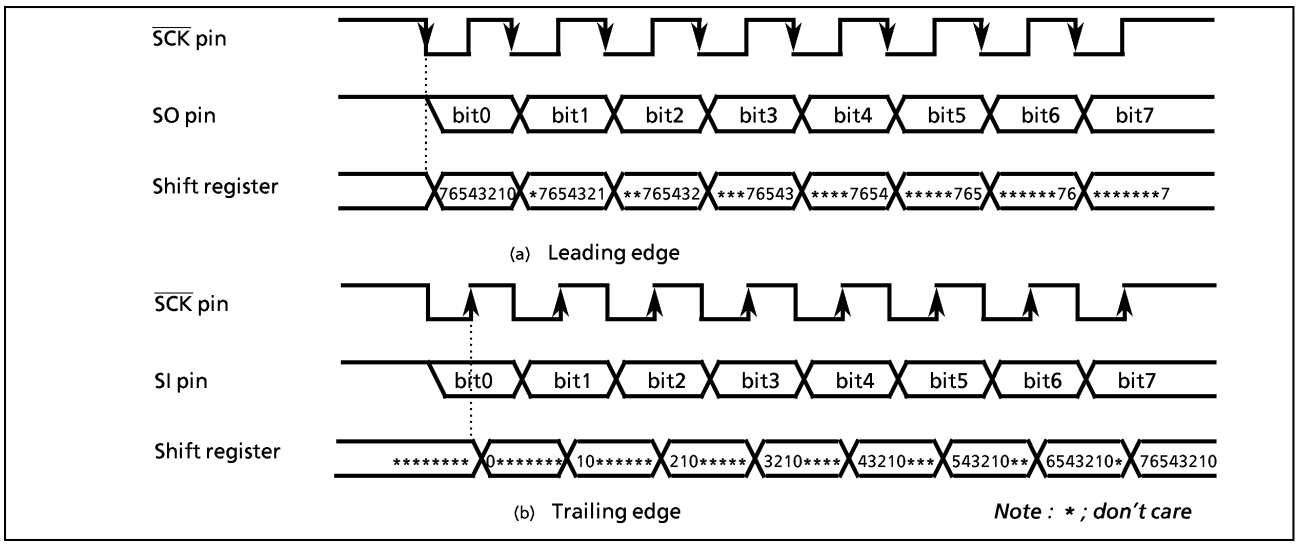


Figure 2-46. Shift edge

(2) Transfer mode

The SIOM (bit 5 and 4 in the SBICR1) is used to select a transmit, receive, or transmit/receive mode.

a. 8-bit transmit mode

Set a control register to a transmit mode and write transmit data to the SBIDBR.

After the transmit data is written, set the SIOS to "1" to start data transfer. The transmitted data is transferred from the SBIDBR to the shift register and output to the SO pin in synchronous with the serial clock, starting from the least significant bit (LSB). When the transmit data is transferred to the shift register, the SBIDBR becomes empty. The INTSBI (buffer empty) interrupt request is generated to request new data.

When the internal clock is used, the serial clock will stop and automatic-wait function will be initiated if new data is not loaded to the data buffer register after the specified 8-bit data is transmitted. When new transmit data is written, automatic-wait function is canceled.

When the external clock is used, data should be written to the SBIDBR before new data is shifted. The transfer speed is determined by the maximum delay time between the time when an interrupt request is generated and the time when data is written to the SBIDBR by the interrupt service program.

When the transmit is started, after the SIOF goes "1" output from the SO pin holds final bit of the last data until falling edge of the $\overline{SCK}$.

Transmitting data is ended by clearing the SIOS to "0" by the buffer empty interrupt service program or setting the SIOINH to "1". When the SIOS is cleared, the transmitted mode ends when all data is output. In order to confirm if data is surely transmitted by the program, set the SIOF (bit 3 in the SBISR) to be sensed. The SIOF is cleared to "0" when transmitting is complete. When the SIOINH is set, transmitting data stops. The SIOF turns "0".

When the external clock is used, it is also necessary to clear the SIOS to "0" before new data is shifted; otherwise, dummy data is transmitted and operation ends.

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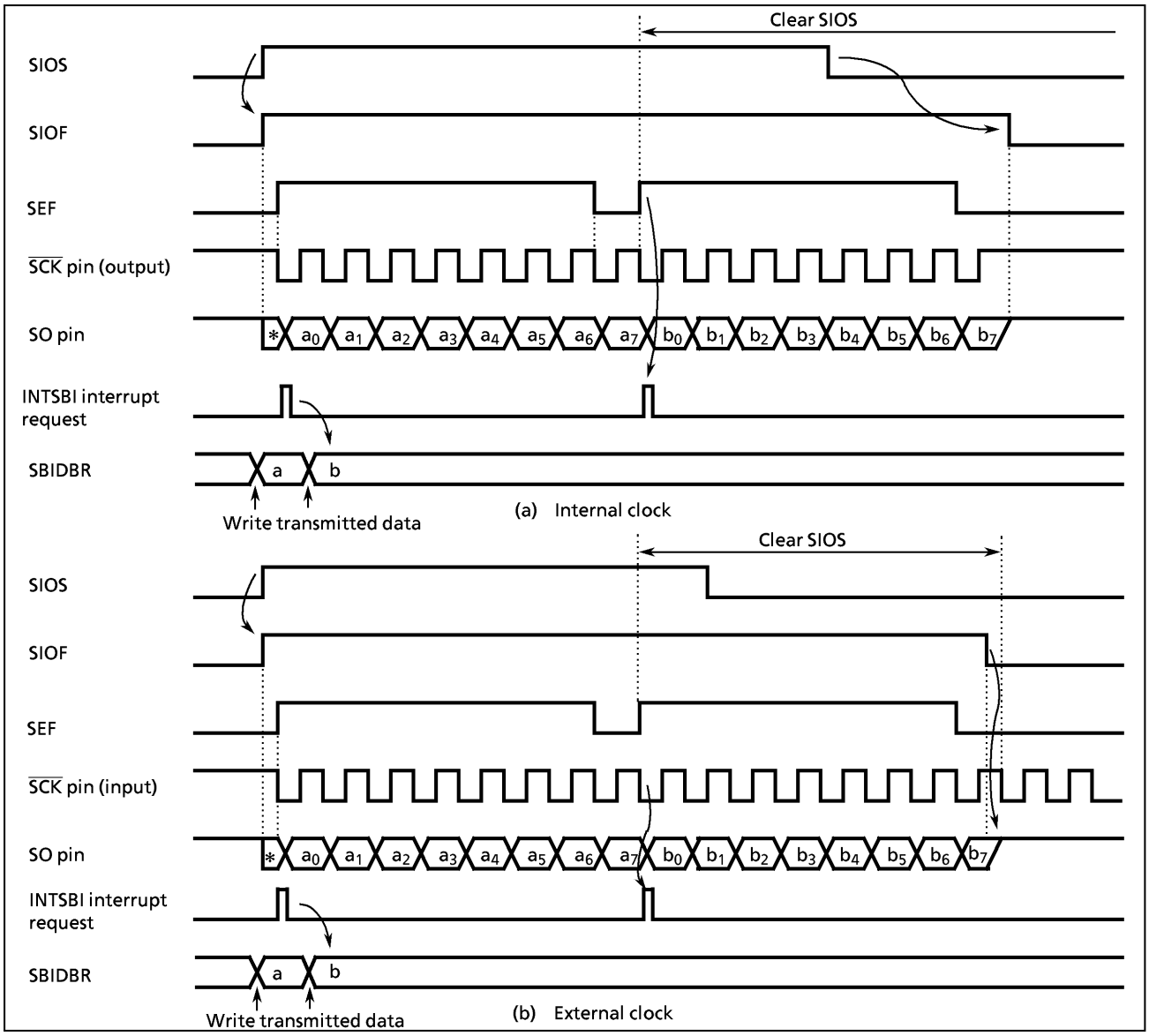


Figure 2-47. Transfer mode

```
Example   :  Program to stop transmitting data (when external clock is used)
 STEST1 : TEST  (SBISR) . SEF          ; If SEF = 1 then loop
          JRS   F , STEST1
 STEST2 : TEST  (P5) . 3               ; If SCK = 0 then loop
          JRS   T , STEST2
          LD    (SBICR1) , 00000111B ; SIOS ← 0
```

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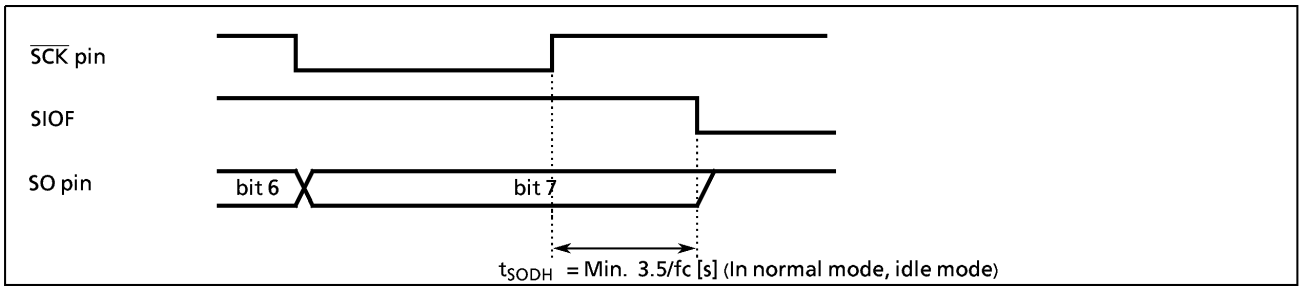


Figure 2-48. Transmitted data hold time at end of transmit

b. 8-bit receive mode

Set the control register to receive mode and the SIOS to "1" for switching to receive mode. Data is received from the SI pin to the shift register in synchronous with the serial clock, starting from the least significant bit (LSB). When the 8-bit data is received, the data is transferred from the shift register to the SBIDBR. The INTSBI (buffer full) interrupt request is generated to request of reading the received data. The data is then read from the SBIDBR by the interrupt service program.

When the internal clock is used, the serial clock will stop and automatic-wait function will be initiated until the received data is read from the SBIDBR.

When the external clock is used, since shift operation is synchronized with the clock pulse provided externally, the received data should be read from the SBIDBR before next serial clock is input. If the received data is not read, further data to be received is canceled. The maximum transfer speed when the external clock is used is determined by the delay time between the time when an interrupt request is generated and the time when received data is read.

Receiving data is ended by clearing the SIOS to "0" by the buffer full interrupt service program or setting the SIOINH to "1". When the SIOS is cleared, received data is transferred to the SBIDBR in complete blocks. The received mode ends when the transfer is complete. In order to confirm if data is surely received by the program, set the SIOF (bit 3 in the SBIDBR) to be sensed. The SIOF is cleared to "0" when receiving is complete. After confirming that receiving has ended, the last data is read. When the SIOINH is set, receiving data stops. The SIOF turns "0" (the received data becomes invalid, therefore no need to read it).

*Note : When the transfer mode is switched, the SBIDBR contents are lost. In case that the mode needs to be switched, receiving data is concluded by clearing the SIOS to "0", read the last data, and then switch the mode.*

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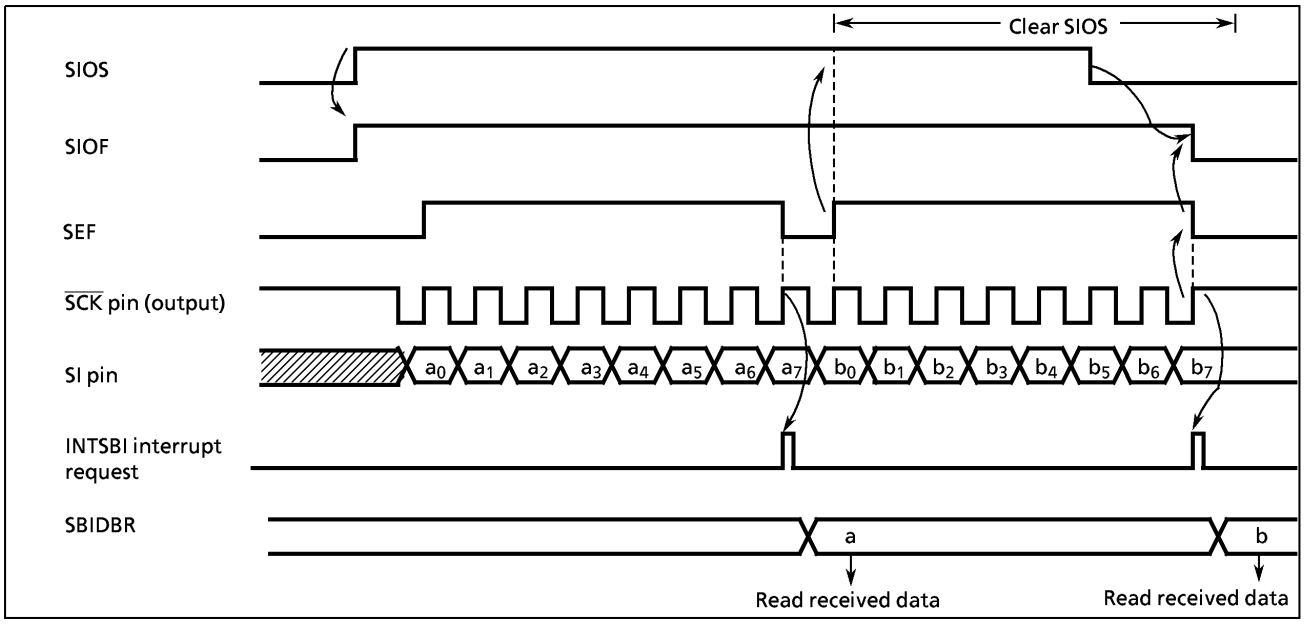


Figure 2-49. Receive mode (Example : Internal clock)

c. 8-bit transmit / receive mode

Set a control register to a transmit / receive mode and write data to the SBIDBR. After the data is written, set the SIOS to "1" to start transmitting / receiving. When transmitting, the data is output from the SO pin on the leading edges in synchronous with the serial clock, starting from the least significant bit (LSB). When receiving, the data is input to the SI pin on the trailing edges of the serial clock. 8-bit data is transferred from the shift register to the SBIDBR, and the INTSBI interrupt request occurs. The interrupt service program reads the received data from the data buffer register and writes data to be transmitted. The SBIDBR is used for both transmitting and receiving. Transmitted data should always be written after received data is read.

When the internal clock is used, automatic-wait function is initiated until received data is read and next data is written.

When the external clock is used, since the shift operation is synchronized with the external clock, received data is read and transmitted data is written before new shift operation is executed. The maximum transfer speed when the external clock is used is determined by the delay time between the time when an interrupt request is generated and the time when received data is read and transmitted data is written.

When the transmit is started, after the SIOF goes "1" output from the SO pin holds final bit of the last data until falling edge of the $\overline{SCK}$.

Transmitting / receiving data is ended by clearing the SIOS to "0" by the INTSBI interrupt service program or setting the SIOINH to "1". When the SIOS is cleared, received data is transferred to the SBIDBR in complete blocks. The transmit / receive mode ends when the transfer is complete. In order to confirm if data is surely transmitted / received by the program, set the SIOF (bit3 in the SBISR) to be sensed. The SIOF becomes "0" after transmitting / receiving is complete. When the SIOINH is set, transmitting / receiving data stops. The SIOF turns "0".

*Note : When the transfer mode is switched, the SBIDBR contents are lost. In case that the mode needs to be switched, conclude transmitting / receiving data by clearing the SIOS to "0", read the last data, and then switch the transfer mode.*

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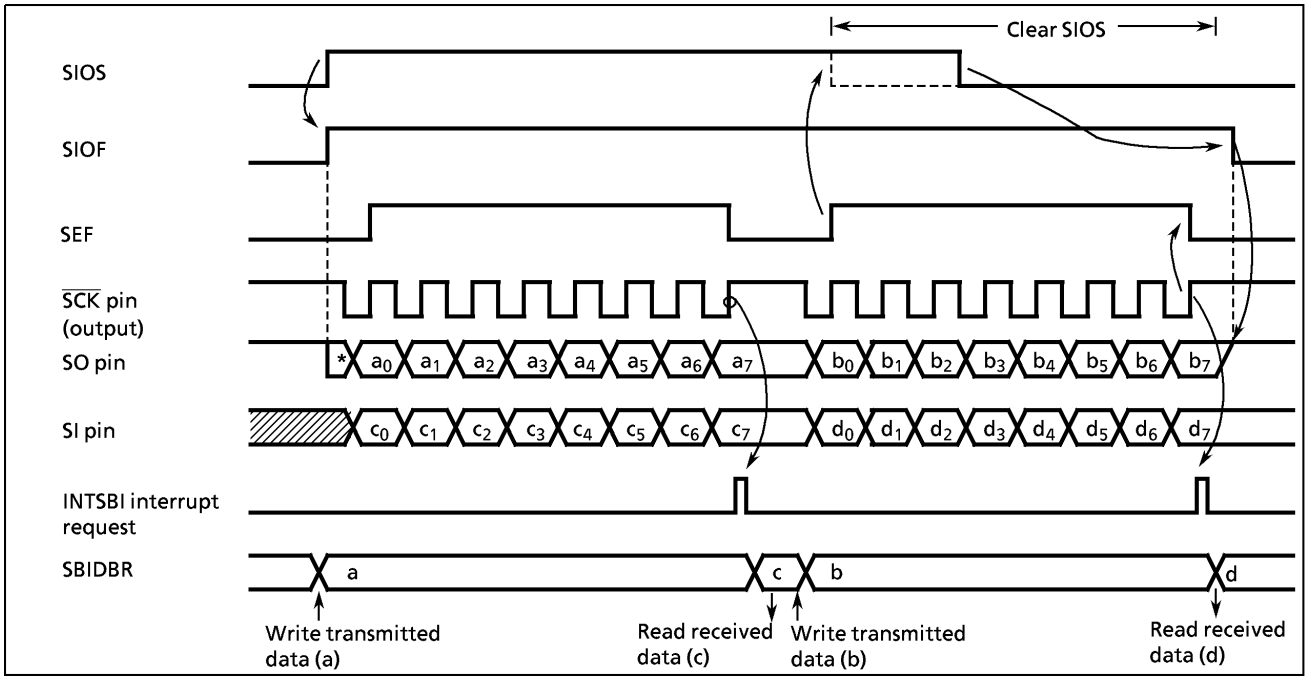


Figure 2-50. Transmit / receive mode (Example : Internal clock)

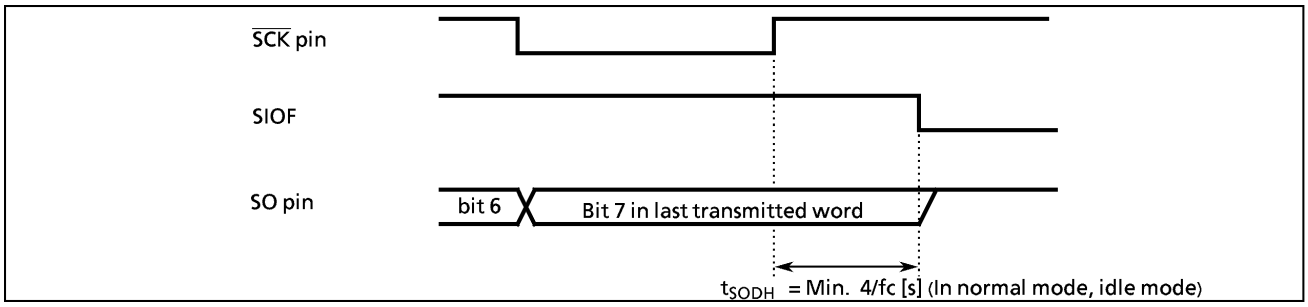


Figure 2-51. Transmitted data hold time at end of transmit / receive

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## 2.9 Remote Control Signal Preprocessor / External Interrupt 3 Input Pin

The remote control signal waveform can be determined by inputting the remote control signal waveform from which the carrier wave was eliminated by the receive circuit to P30 (INT3 / RXIN) pin. When the remote control signal preprocessor / external interrupt 3 pin is also used as the P30 port, set the P30 port output latch to "1". When it is not used as the remote control signal preprocessor/external interrupt 3 input pin, it can be used for normal port.

### 2.9.1 Configuration

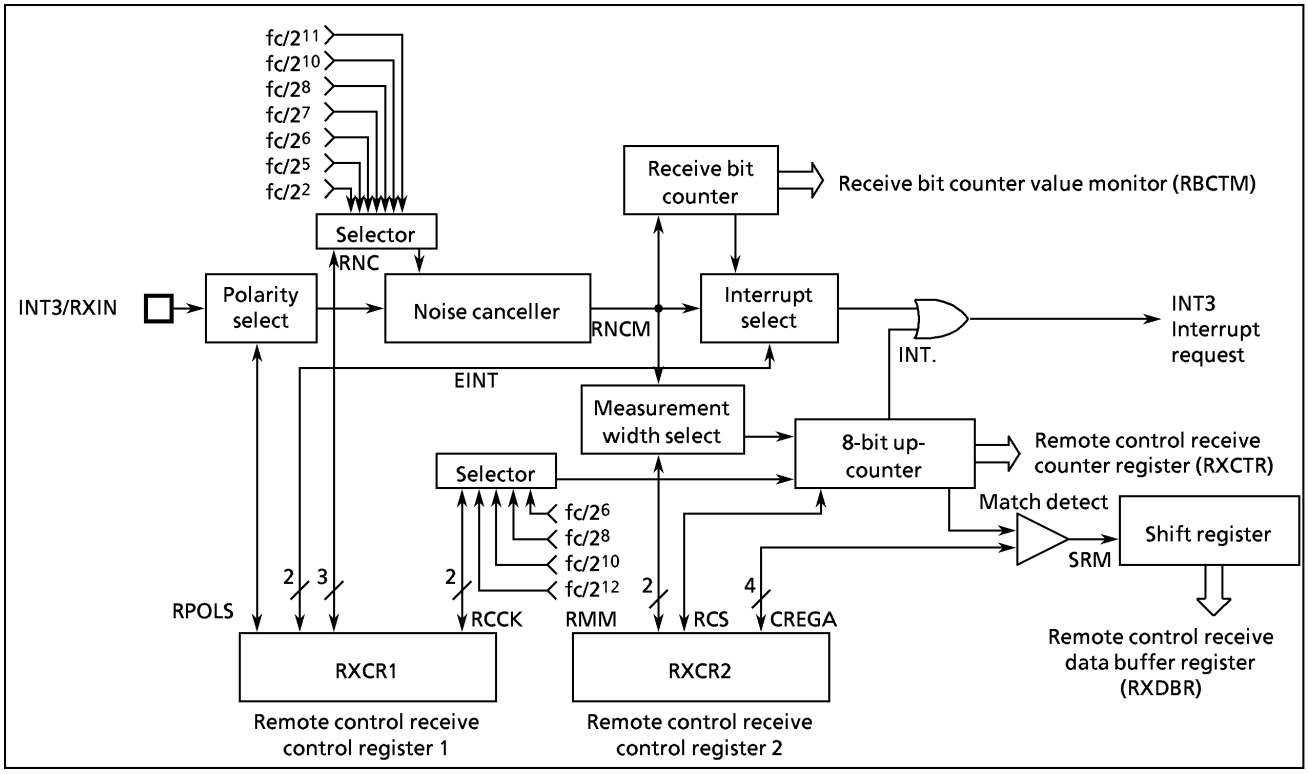


Figure 2-52. Remote control signal preprocessor

### 2.9.2 Remote control signal preprocessor control

When the remote control signal preprocessor is used, operating states are controlled and monitored by the following registers. Interrupt requests also use the remote control signal preprocessor / external interrupt 3 input pin.

- Remote control receive control register 1 (RXCR1)
- Remote control receive control register 2 (RXCR2)
- Remote control receive counter register (RXCTR)
- Remote control receive data buffer register (RXDBR)
- Remote control receive status register (RXSR)

When this pin is used for the external interrupt 3 input, set EINT in RXCR1 to other than"11".

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Remote control receive control register 1

RXCR1 ($0FD0_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| RCCK | | RPOLS | EINT | | | RNC | |

(Initial value : 0000 0000)

| | | | |
|---|---|---|---|
| RCCK | 8-bit up-counter source clock select | 00 : $fc/2^6$ [Hz]<br>01 : $fc/2^8$<br>10 : $fc/2^{10}$<br>11 : $fc/2^{12}$ | Read/ Write |
| RPOLS | Remote control signal polarity select | 0 : Positive<br>1 : Negative | |
| EINT | Interrupt source select | 00 : Rising edge<br>01 : Falling edge ) when RPOLS = 0<br>10 : Rising / Falling edge<br>11 : 8-bit receive end | |
| RNC | Noise canceler noise eliminating time select | 000 : Noise canceler disable<br>001 : $2^2/fc \times 7 - 1/fc$ [s]<br>010 : $2^5/fc \times 7 - 1/fc$<br>011 : $2^6/fc \times 7 - 1/fc$<br>100 : $2^7/fc \times 7 - 1/fc$<br>101 : $2^8/fc \times 7 - 1/fc$<br>110 : $2^{10}/fc \times 7 - 1/fc$<br>111 : $2^{11}/fc \times 7 - 1/fc$ | |

*Note1 : fc ; High-frequency clock [Hz]*

*Note2 : After reset, RPOLS do not change the set value in the receiving remote control signal. For setting interrupt edge and measurement data, use EINT and RMM.*

Remote control receive control register 2

RXCR2 ($0FD1_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| CREGA | | | | RCS | | RMM | |

(Initial value : 0000 0*00)

| | | | |
|---|---|---|---|
| CREGA | Setting of detect time for match with 8-bit up-counter upper 4 bits | Match detect time (Tth) = 16 × CREGA / RCCK [s]<br>CREGA = $0_H$ to $F_H$<br>Example : CREGA = $2_H$, RCCK = $fc/2^6$ [Hz] ,at fc = 8 MHz<br>Tth = 256 [$\mu$s] | Read/ Write |
| RCS | 8-bit up-counter start control | 0 : Stop and counter clear<br>1 : Start | |
| RMM | Measurement mode select (invalid when EINT = "10") | 00 :<br>01 :<br>10 : Refer to table 2-6<br>11 : | |

*Note1 : fc ; High-frequency clock [Hz], * ;don't care*

*Note2 : When an interrupt source is set for rising / falling edge, low and high widths are forcibly measured separately.*

*Note3 : Set CREGA ($0_H$ to $F_H$) before EINT sets to 8-bit receive end.*

Figure 2-53. Remote control receive control register 1, 2

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Remote control receive counter register

RXCTR ($0FD2_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| | | | | | | | |

Read Only
(Initial value : 0000 0000)

Remote control receive data buffer register

RXDBR ($0FD3_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| | | | | | | | |

Read Only
(Initial value : 0000 0000)

Remote control receive status register

RXSR ($0FD4_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| RBCTM | | | | | OVFF | SRM | RNCM |

(Initial value : 0000 * 000)

| | | | |
|---|---|---|---|
| RBCTM | Receive bit counter value monitor | | Read Only |
| OVFF | 8-bit up-counter overflow flag | 0 : No overflow<br>1 : Overflow | |
| SRM | Data buffer register input monitor | 0 : Upper 4bits of 8-bit up-counter < CREGA<br>1 : Upper 4bits of 8-bit up-counter ≧ CREGA | |
| RNCM | Remote control signal monitor after passing through noise canceler | | |

*Note : * ; don't care*

Figure 2-54. Remote control receive counter register, data buffer register, status register

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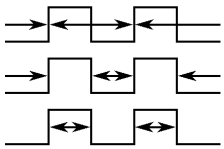
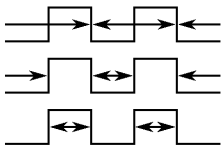
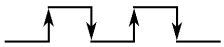
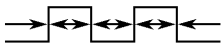
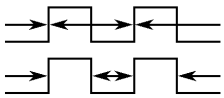
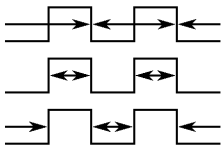
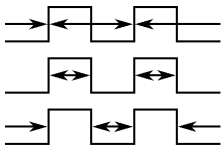
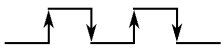
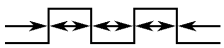
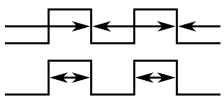
Table 2-6. Combination of interrupt source and measurement mode

| RPOLS | EINT | RMM | Interrupt source | Measurement mode |
|---|---|---|---|---|
| 0 | 00 | 00 |  |  |
| | | 10 | | |
| | | 11 | | |
| | 01 | 01 |  |  |
| | | 10 | | |
| | | 11 | | |
| | 10 | — |  |  |
| | 11 | 00 | Receive end |  |
| | | 10 | | |
| 1 | 00 | 00 |  |  |
| | | 10 | | |
| | | 11 | | |
| | 01 | 01 |  |  |
| | | 10 | | |
| | | 11 | | |
| | 10 | — |  |  |
| | 11 | 00 | Receive end |  |
| | | 10 | | |

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### 2.9.3 Noise elimination time setting

The remote control receive circuit has a noise canceler. By setting RNC in RXCR1, input signals shorter than the fixed time can be eliminated as noise.

Table 2-7. Noise elimination time setting

| RNC | Minimum signal pulse width | at fc = 8 MHz | Maximum noise width to be eliminated | at fc = 8 MHz |
|---|---|---|---|---|
| 000 | ——— | —— | ——— | —— |
| 001 | $(2^5 + 5) / fc$ [s] | 4.63 [$\mu$s] | $(2^2 \times 7 - 1) / fc$ [s] | 3.38 [$\mu$s] |
| 010 | $(2^8 + 5) / fc$ | 32.63 | $(2^5 \times 7 - 1) / fc$ | 27.88 |
| 011 | $(2^9 + 5) / fc$ | 64.63 | $(2^6 \times 7 - 1) / fc$ | 55.88 |
| 100 | $(2^{10} + 5) / fc$ | 128.63 | $(2^7 \times 7 - 1) / fc$ | 111.88 |
| 101 | $(2^{11} + 5) / fc$ | 256.63 | $(2^8 \times 7 - 1) / fc$ | 223.88 |
| 110 | $(2^{13} + 5) / fc$ | 1.025 [ms] | $(2^{10} \times 7 - 1) / fc$ | 895.88 |
| 111 | $(2^{14} + 5) / fc$ | 2.049 | $(2^{11} \times 7 - 1) / fc$ | 1.792 [ms] |

### 2.9.4 Operation

(1) interrupts at rising, falling, or rising / falling edge, and measurement modes

First set EINT and RMM. Next, set RCS to "1" ; the 8-bit up-counter is counted up by the internal clock. After measurement, the 8-bit up-counter value is saved in RXCTR. Then, the 8-bit up-counter is cleared, an INT3 request is generated, and the 8-bit up-counter resumes counting.

If the 8-bit up-counter overflows ($FF_H$) before measurement is completed, an INT3 request is generated and the overflow flag (OVFF) is set to "1". Then, the 8-bit up-counter is cleared. An overflow can be detected by reading OVFF by the interrupt processing. To restart the 8-bit up-counter, set RCS to "1".

Setting RCS to "1" zero-clears OVFF.

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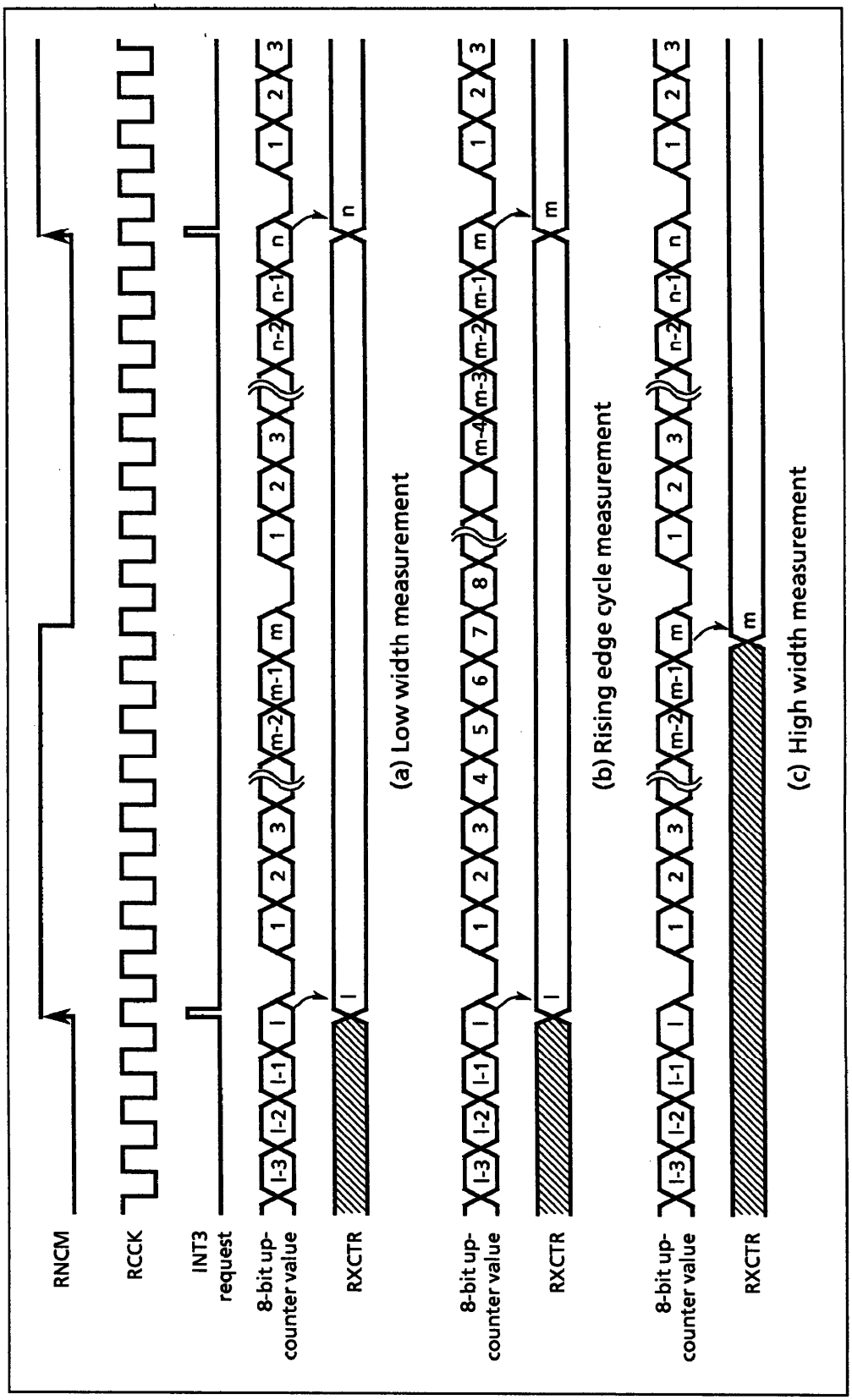


Figure 2-55. Rising edge interrupt timing chart (RPOLS = 0)

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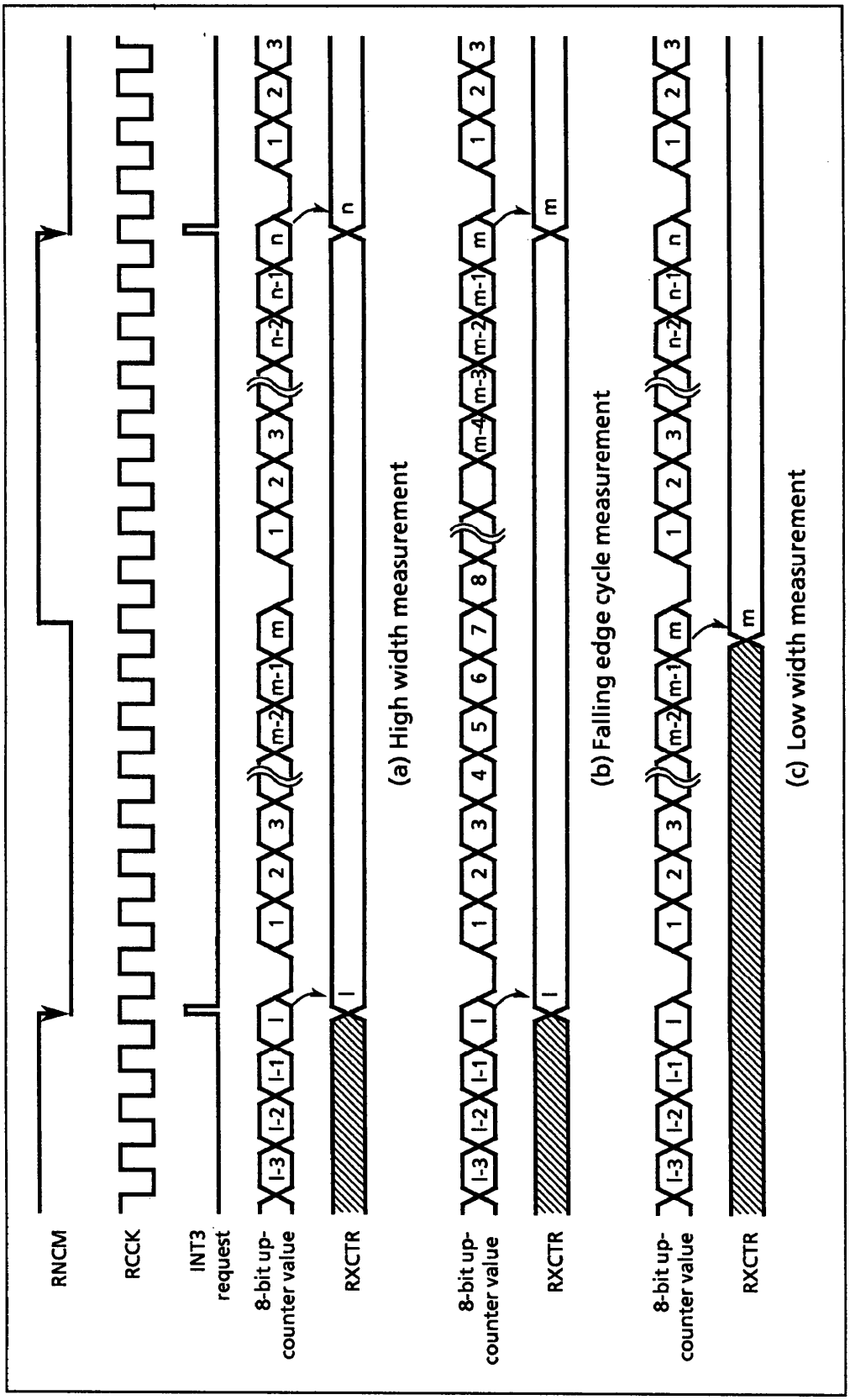


Figure 2-56. Falling edge interrupt timing chart (RPOLS = 0)

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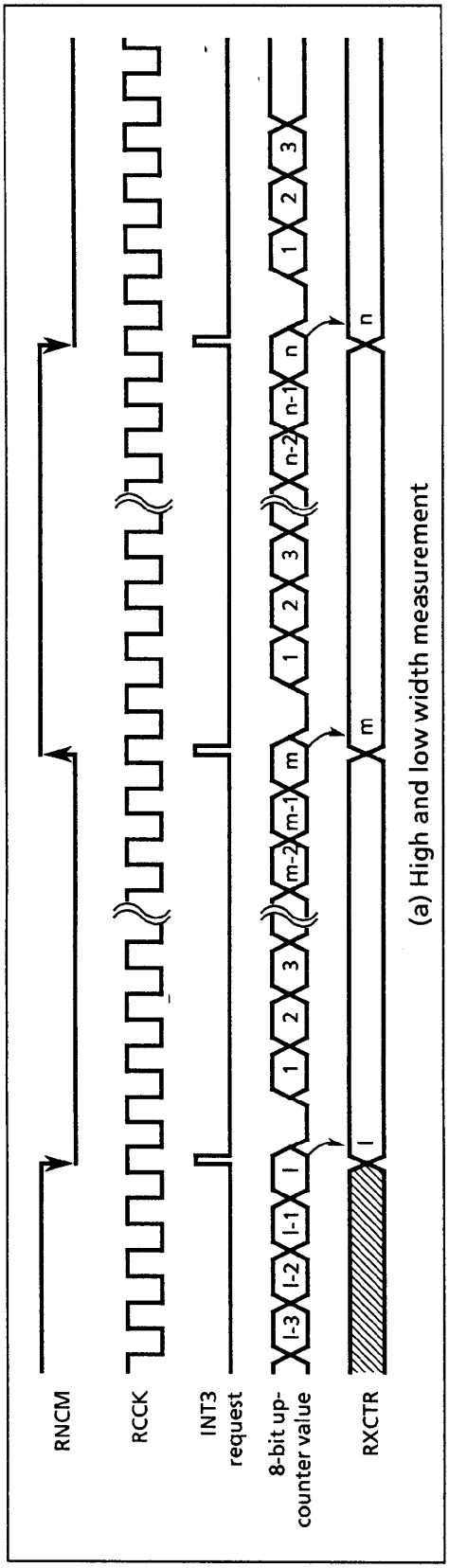


Figure 2-57. Rising / falling edge interrupt timing chart

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(2) 8-bit receive end interrupts and measurement modes

By determining one-cycle remote control signal as one-bit data set to "0" or one-pulse width remote control signal as one-bit data set to "1", an INT3 request is generated after 8-bit data is received. When "0" is determined, this means the upper four bits in the 8-bit up-counter have not reached the CREGA value. When "1" is determined, this means the upper four bits in the 8-bit up-counter have reached or exceeded the CREGA value. The 8-bit up-counter value is saved in RXCTR after one bit is determined. The determined data is saved, bit by bit, in RXDBR at the rising edge of the remote control signal (when RPOLS = 1, falling edge). The number of bits saved in RXDBR is counted by the receive bit counter and saved in RBCTM. RBCTM is set to "0001B" at the rising edge of the input (when RPOLS = 1, falling edge) after the INT3 request is generated.

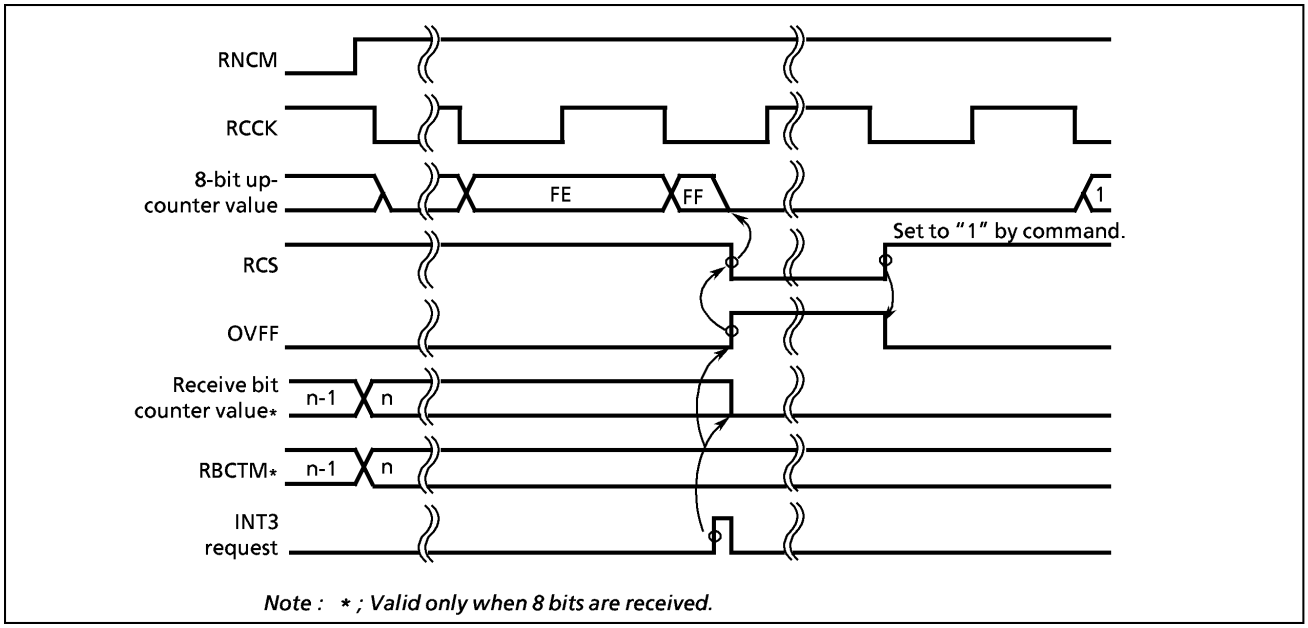


Figure 2-58. Overflow interrupt timing chart

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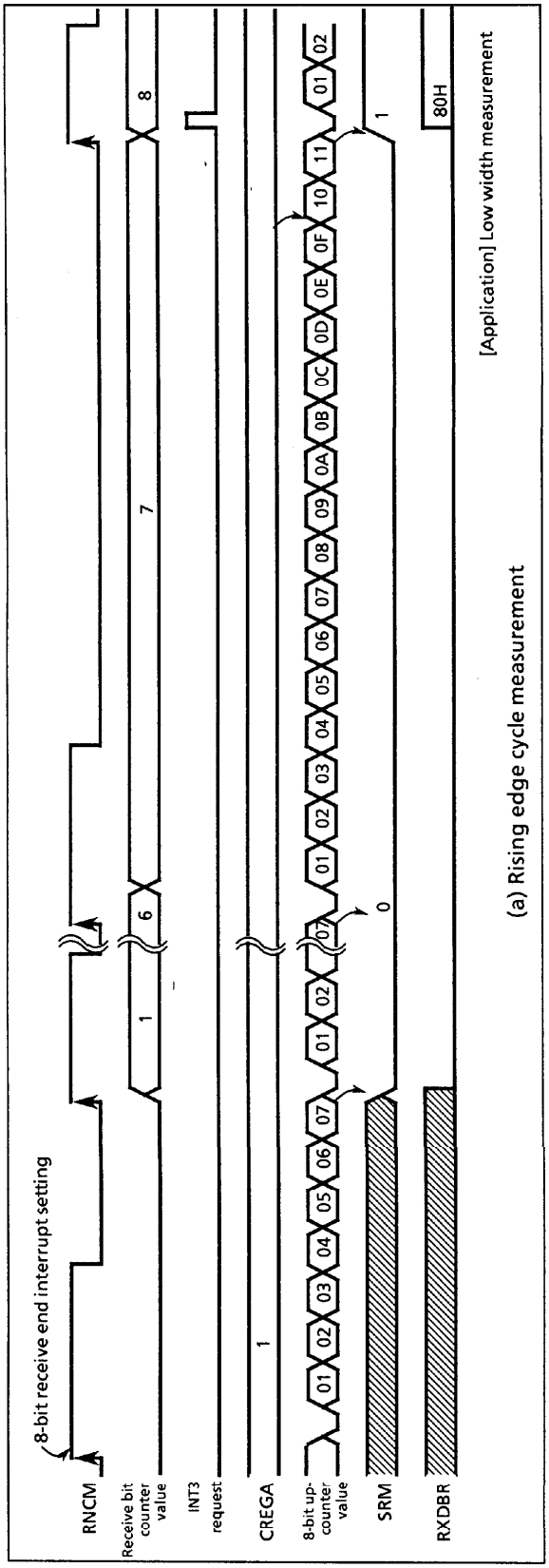


Figure 2-59. 8-bit receive end interrupt timing chart (RPOLS = 0)

Table 2-8. Count clock for remote control preprocessor circuit

| Count clock (RCCK) | at fc = 8 MHz | | | | |
|---|---|---|---|---|---|
| | Resolution | | | Maximum setting time | |
| $fc/2^6$ [Hz] | 8 | μs | | 2.048 | ms |
| $fc/2^8$ | 32 | μs | | 8.192 | ms |
| $fc/2^{10}$ | 128 | μs | | 32.768 | ms |
| $fc/2^{12}$ | 512 | μs | | 131.072 | ms |

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## 2.10 8-bit A/D Converter (ADC)

The 87CH38/K38 each have an 6-channel multiplexed-input 8-bit successive approximate type A/D converter with sample and hold.

### 2.10.1 Configuration

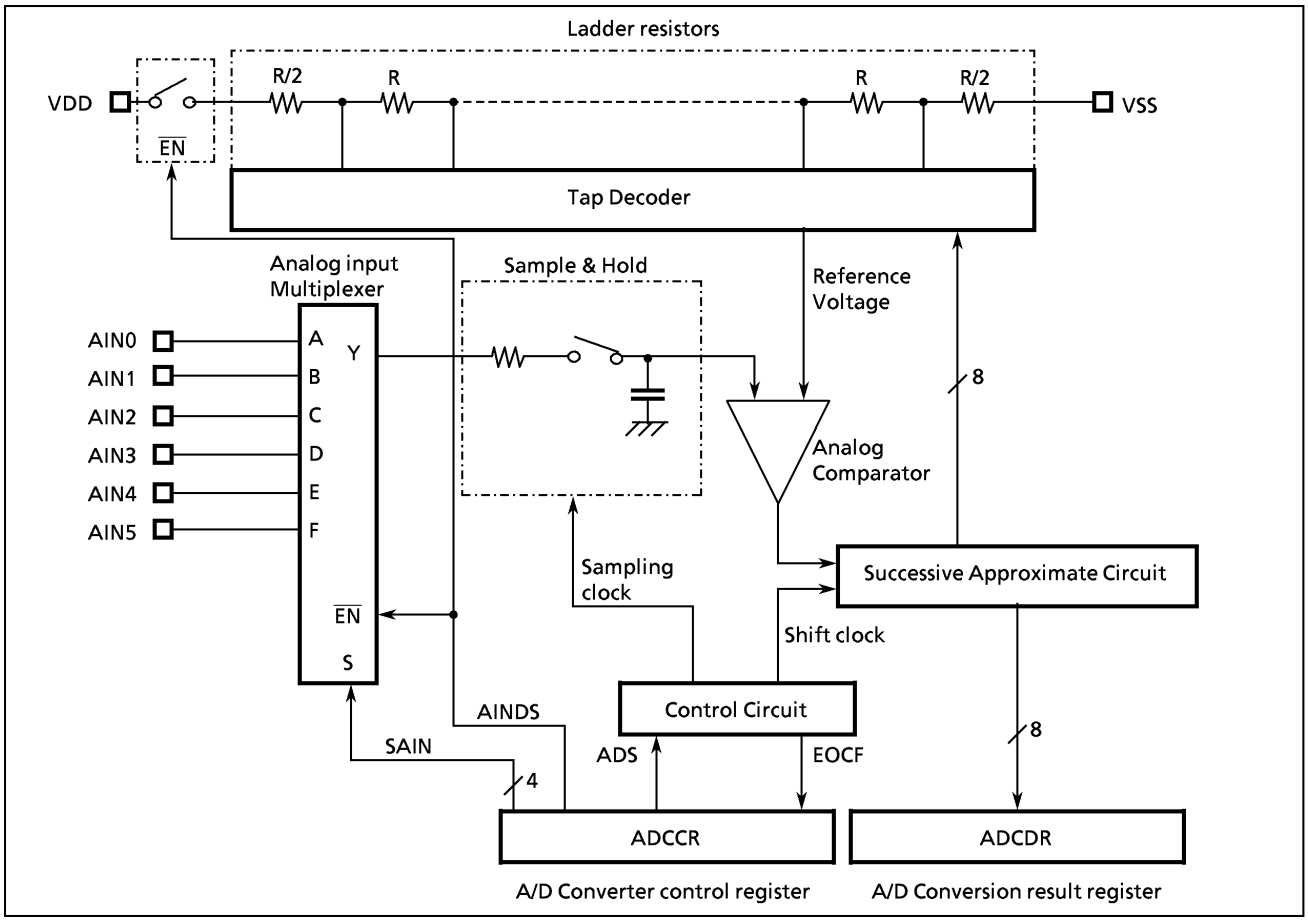


Figure 2-60. A/D converter

### 2.10.2 Control

The A/D converter is controlled by an A/D converter control register (ADCCR) .

A/D Conversion Result Register

| | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | |
|---|---|---|---|---|---|---|---|---|---|
| **ADCDR** ($000F_H$) | | | | | | | | | Read only |

Figure 2-61. A/D conversion result register

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A/D Converter Control Register

ADCCR ($000E_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| EOCF | ADS | "0" | AINDS | SAIN | | | |

(Initial value : 00*0 0000)

| | | | |
|---|---|---|---|
| SAIN | Analog input selection | 0000 : AIN0<br>0001 : AIN1<br>0010 : AIN2<br>0011 : AIN3<br>0100 : AIN4<br>0101 : AIN5<br>011* : reserved<br>1*** : reserved | R/W |
| AINDS | Analog input control | 0 : Enable<br>1 : Disable | |
| ADS | A/D conversion start | 0 : –<br>1 : A/D conversion start | |
| EOCF | End of A/D conversion flag | 0 : Under conversion or Before conversion<br>1 : End of conversion | R |

*Note 1 : * ; don't care*

*Note 2 : Select analog input when A/D converter stops.*

*Note 3 : The ADS is automatically cleared to "0" after starting conversion.*

*Note 4 : The EOCF is cleared to "0" when reading the ADCDR.*

*Note 5 : The EOCF is read-only.*

Figure 2-62. A/D converter control register

## 2.10.3 Operation

(1) Start of A/D conversion

First, clear the corressponding P6CR bit to "0" for analog input. Clear the AINDS (bit 4 in ADCCR) to "0" and select one of six analog inputs AIN5-AIN0 with the SAIN (bit 3 to 0 in ADCCR).

*Note : The pin that is not used as an analog input can be used as regular input/output pins. During conversion, do not perform output instruction to maintain a precision for all of the pins.*

A/D conversion is started by setting the ADS (bit 6 in ADCCR) to "1".
Conversion is accomplished in 46 machine cycles (184/fc [s] ).
The EOCF (bit 7 in ADCCR) is set to "1" at end of conversion.
When setting the ADS to "1" under A/D conversion, the A/D converter circuit is initialized and the A/D conversion try again from start.
The sampling of the analog input voltage is excuted at 4 machine cycles after setting the ADS to "1".

*Note : The circuit of sample and hold is included in a condenser (12 pF (typ.)) through a register (5 kΩ (typ.)).*
*Therefore, until 4 machine cycles is over, this condenser must be charged.*

(2) Reading of A/D conversion result

After the end of conversion, read the conversion result from the ADCDR.
The EOCF is automatically cleared to "0" when reading the ADCDR.

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(3) A/D conversion in STOP mode

When the MCU places in the STOP mode during the A/D conversion, the conversion is terminated and the ADCDR contents become indefinite.

However, if the STOP mode is started after the end of conversion (EOCF = 1), the ADCDR contents are held.

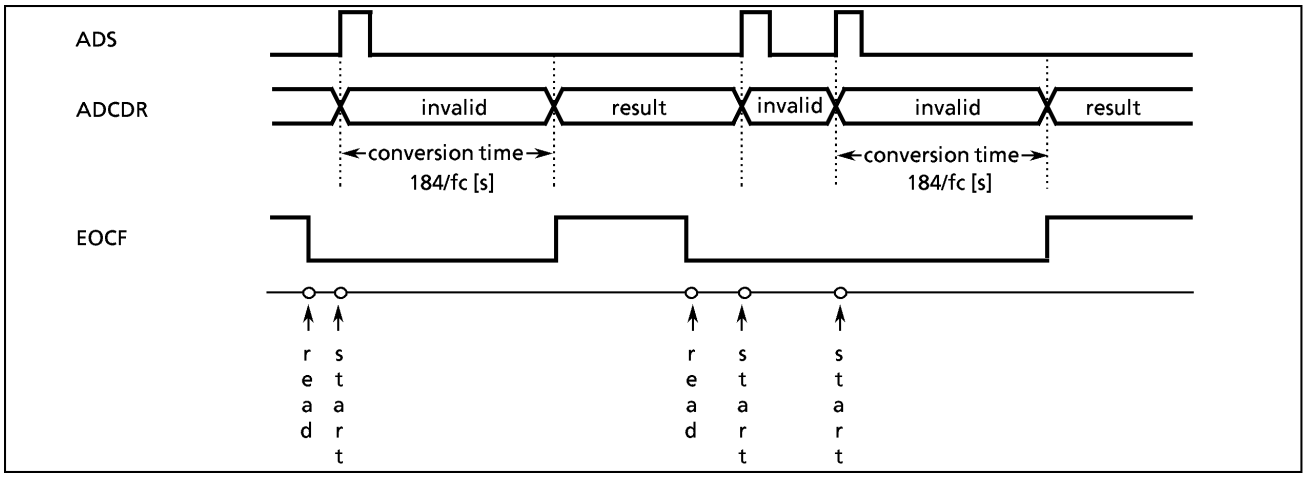


Figure 2-63. A/D conversion timing chart

Example:

```
                ; AIN SELECT
                LD          (ADCCR) , 00000100B     ;   selects AIN4
                ; A/D CONVERT START
                SET         (ADCCR) . 6             ;   ADS = 1
SLOOP    :      TEST        (ADCCR) . 7             ;   EOCF = 1 ?
                JRS         T, SLOOP
                ; RESULT DATA READ
                LD          (9EH), (ADCDR)
```

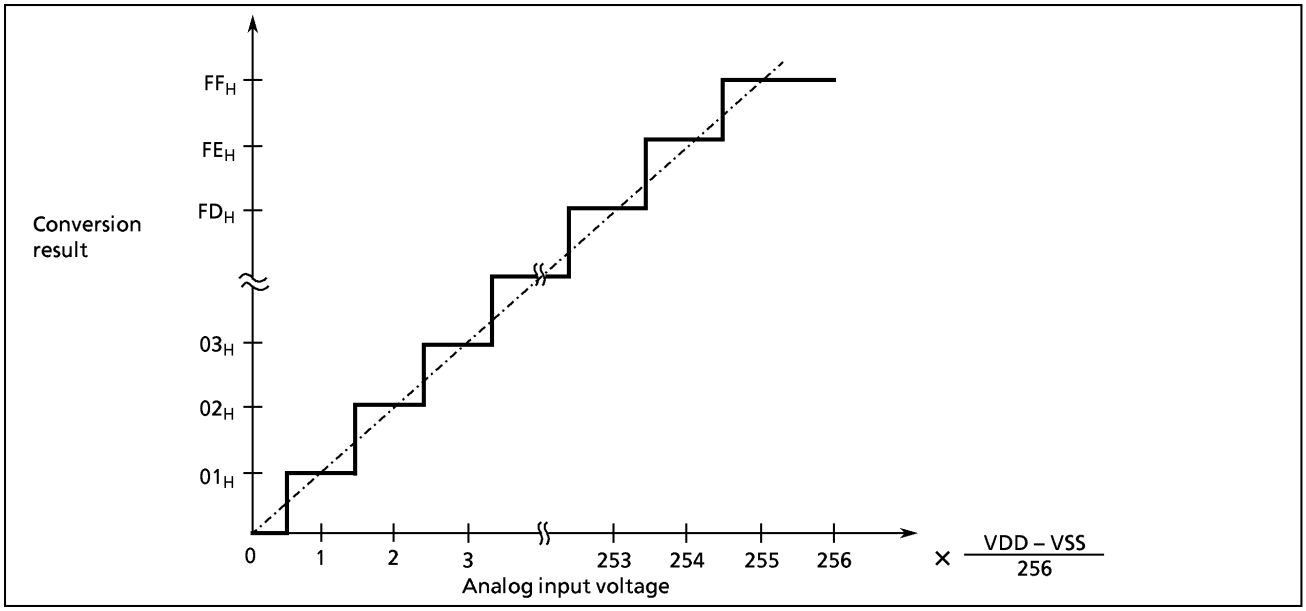


Figure 2-64. Analog input voltage vs A/D conversion result (typ.)

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## 2.11 Pulse Width Modulation Circuit Output

87CH38/K38 have a 14-bit resolution pulse with modulation (PWM) channel and 9 7-bit resolution PWM channels. D/A converter output can easily be obtained by connecting an external low-pass filter. PWM outputs are multiplexed with general purpose I/O ports as; P40 ($\overline{PWM0}$) to P47 ($\overline{PWM7}$), P50 ($\overline{PWM8}$), P51 ($\overline{PWM9}$). When these ports are used PWM outputs,the corresponding bits of P4, P5 output latches and input/output control latches should be set to "1".

### 2.11.1 Configuration

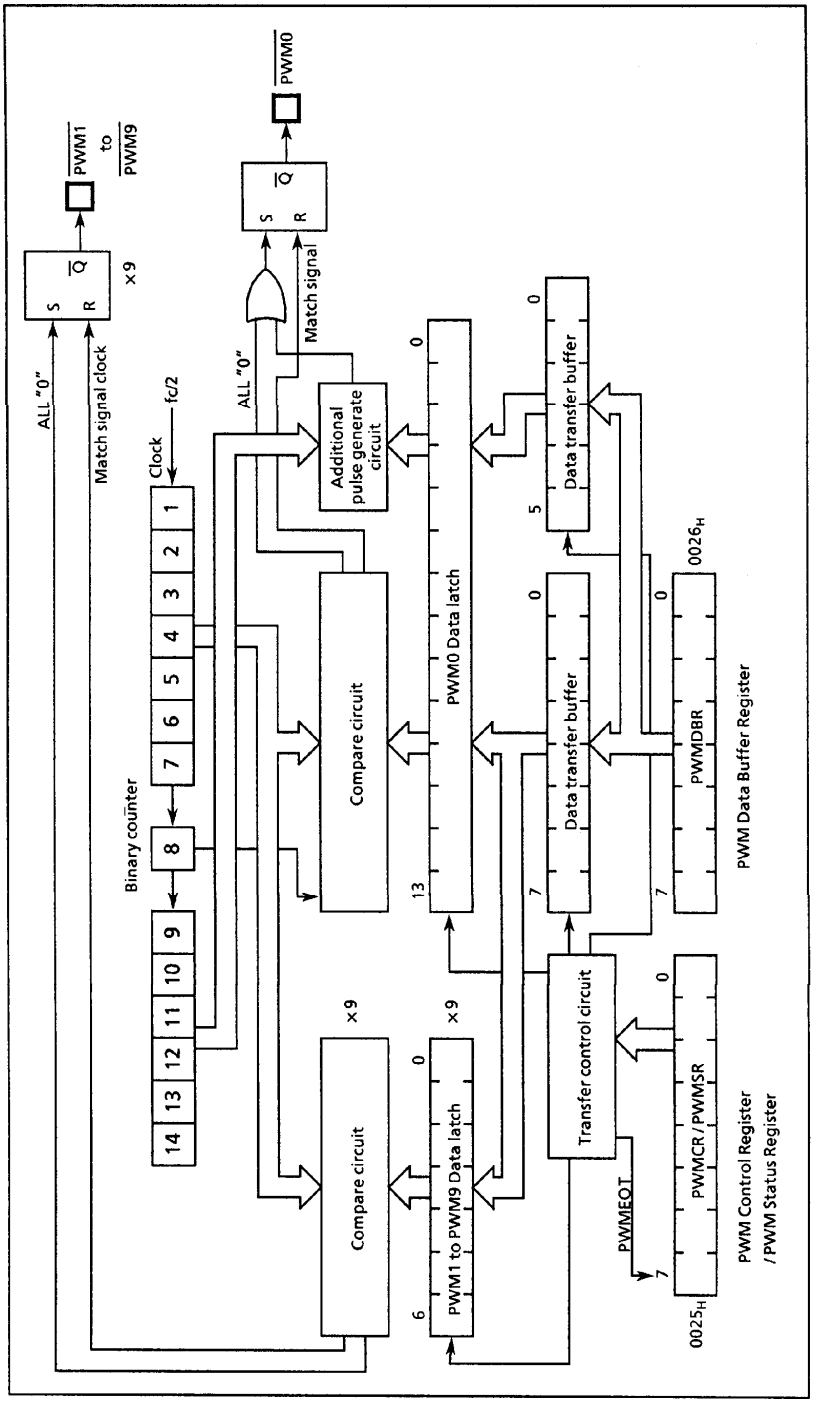


Figure 2-65. Pulse width modulation circuit

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### 2.11.2 PWM output wave form

(1) $\overline{PWM0}$ output

This is 14-bit resolution PWM output and one period is $T_M = 2^{15}/fc$ [s].

The 8 high-order bits of the PWM data latch control the pulse width of the pulse output with a period of $T_S$ ($T_S = T_M/64$), which is the sub-period of the $\overline{PWM0}$. When the 8-bit data are decimal $n$ ($0 \leqq n \leqq 255$), this pulse width becomes $n \times t_0$, where $t_0 = 2/fc$.

The lower 6-bit of 14 bit data are used to control the generation of additional to wide pulse in each $T_S$ period. When the 6-bit data are decimal $m$ ($0 \leqq m \leqq 63$), the additional pulse is generated in each of $m$ periods out of 64 periods contained in a $T_M$ period. The relationship between the 6 bits data and the position of $T_S$ period where the additional pulse is generated is shown in Table 2-10.

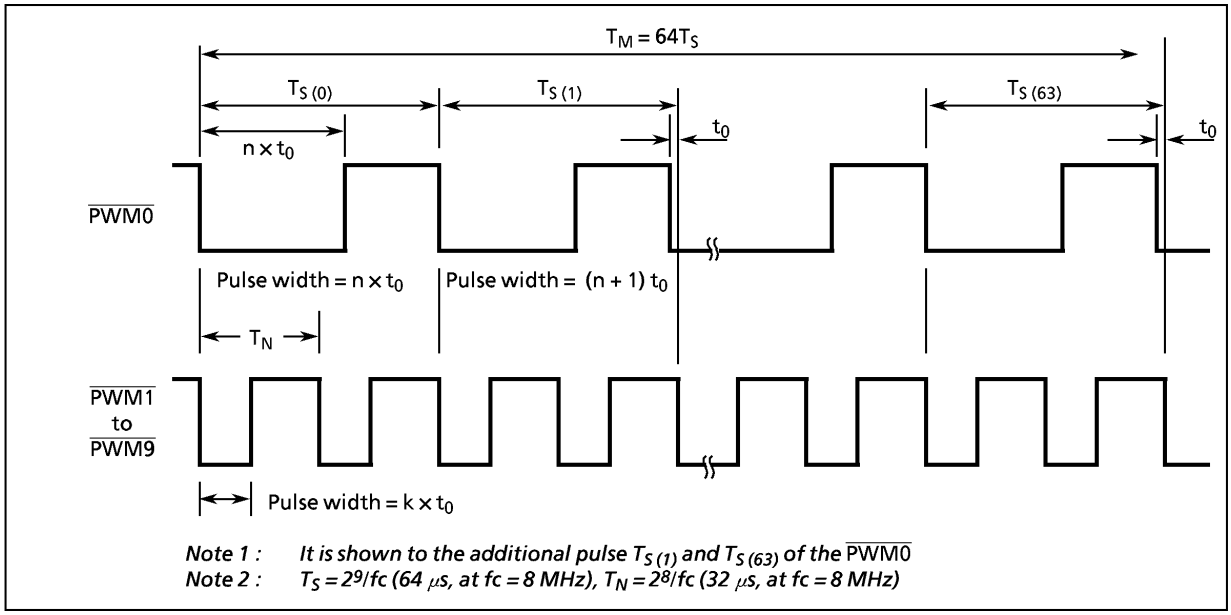
Table 2-10. Correspondence between 6 Bits data and the additional pulse generated $T_S$ period

| Bit position of 6 bits data | Relative position of Ts where the output pulse is generated. (Number i of $T_{S(i)}$ is listed) |
|---|---|
| Bit 0 | 32 |
| Bit 1 | 16, 48 |
| Bit 2 | 8, 24, 40, 56 |
| Bit 3 | 4, 12, 20, 28, 36, 44, 52, 60 |
| Bit 4 | 2, 6, 10, 14, 18, 22, 26, 30, ·····, 58, 62 |
| Bit 5 | 1, 3, 5, 7, 9, 11, 13, 15, 17, ·····, 59, 61, 63 |

*Note : When the corresponding bit is "1", it is output.*

(2) $\overline{PWM1}$ to $\overline{PWM9}$ outputs

These are 7-bit resolution PWM outputs and one period is $T_N = 2^8/fc$ [s]. When the 7-bit data are decimal $k$ ($0 \leqq k \leqq 127$), the pulse width becomes $k \times t_0$. The wave form is illustrated in Figure 2-66.




*Note 1 : It is shown to the additional pulse $T_{S(1)}$ and $T_{S(63)}$ of the $\overline{PWM0}$*
*Note 2 : $T_S = 2^9/fc$ (64 μs, at fc = 8 MHz), $T_N = 2^8/fc$ (32 μs, at fc = 8 MHz)*

Figure 2-66. PWM output wave form

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### 2.11.3 Control

PWM output is controlled by PWM Control Register (PWMCR) and PWM Data Buffer Register (PWMDBR). The status of transfer PWM data from PWMDBR to PWM data latch is read by PWMEOT of PWM status register (PWMSR).

PWM Control Register

| | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | |
|---|---|---|---|---|---|---|---|---|---|
| PWMCR (0025H) | | | | | PWMDLS | | | | (Initial value **** 0000) |

| | | | |
|---|---|---|---|
| PWMDLS | Selection of PWM data latch and request of data transfer | 0000 : Lower 6-bit of PWM0<br>0001 : 8 High-order bits of PWM0<br>0010 : PWM1<br>0011 : PWM2<br>0100 : PWM3<br>0101 : PWM4<br>0110 : PWM5<br>0111 : PWM6<br>1000 : PWM7<br>1001 : PWM8<br>1010 : PWM9<br>1011 : reserved<br>1100 : PWM Data Transfer Request<br>1101 : reserved<br>111* : reserved | write only |

*Note : * ; don't care*

PWM Status Register

| | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | |
|---|---|---|---|---|---|---|---|---|---|
| PWMSR (0025H) | PWMEOT | "1" | "1" | "1" | "1" | "1" | "1" | "1" | (Initial value 0111 1111) |

| | | | |
|---|---|---|---|
| PWMEOT | PWM data<br>End of PWM data transfer flag | 0 : End of Transfer<br>1 : Under Transfer | read only |

PWM Data Buffer Register

| | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | |
|---|---|---|---|---|---|---|---|---|---|
| PWMDBR (0026H) | | | | | | | | | write only |

Figure 2-67. PWM control register / PWM status register / PWM data buffer register

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(1) Programing of PWM data

PWM output is controlled by writing the output data to data latches.
The sequence of writing the output data to data latch is shown as follows;

1. Write the channel number of PWM data latch to the PWMDLS.
2. Write PWM output data to the PWMDBR.
3. Write "$0C_H$" to the PWMCR.
   When transferring of the output data is completed, the PWMEOT becomes "0", indicating that the next data can be written. Do not write PWM data when the PWMEOT is "1" because write errors can occur in this case.

| *Note : When writing the output data to PWM0 data latch, write "$0C_H$" to the PWMCR after writing of the 14-bit output data is completed.* |
|---|

While the output data are being written to the data latch, the previously written data are being output. The maximum time from the point at which "$0C_H$" is written to the data latch until PWM output is switched is $2^{15}$/fc [s] (4.096 ms, at fc = 8 MHz) for PWM0 output and $2^9$/fc [s] (64 $\mu$s, at fc = 8 MHz) for PWM1 to PWM9 output.

Example : $\overline{\text{PWM0}}$ pin outputs a PWM wave form with a low-level of 32 $\mu$s width and no additional pulse.
$\overline{\text{PWM1}}$ pin outputs a PWM wave form with a low-level of 16 $\mu$s width.
$\overline{\text{PWM2}}$ pin outputs a PWM wave form with a low-level of 8 $\mu$s width.

| *Note : at fc = 8 MHz* |
|---|

```
        LD      (PWMCR), 00H    ;  Select lower 6-bit of PWM0
        LD      (PWMDBR), 00H   ;  No additional pulse
        LD      (PWMCR), 01H    ;  Select 8 high-order bits of PWM0
        LD      (PWMDBR), 80H   ;  32 μs ÷ 2/fc = 80 H
        LD      (PWMCR), 0CH    ;  Request PWM Data Transfer
WAIT0 : TEST    (PWMSR). 7      ;  PWMEOT = 0?
        JRS     F, WAIT0
        LD      (PWMCR), 02H    ;  Select PWM1
        LD      (PWMDBR), 40H   ;  16 μs ÷ 2/fc = 40 H
        LD      (PWMCR), 0CH    ;  Request PWM Data Transfer
WAIT1 : TEST    (PWMSR). 7      ;  PWMEOT = 0?
        JRS     F, WAIT1
        LD      (PWMCR), 03H    ;  Select PWM2
        LD      (PWMDBR), 20H   ;  8 μs ÷ 2/fc = 20 H
        LD      (PWMCR), 0CH    ;  Request PWM Data Transfer
WAIT2 : TEST    (PWMSR). 7      ;  PWMEOT = 0?
        JRS     F, WAIT2
```

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## 2.12 Test Video Signal Output for Adjusting TV Screen

TMP87CH38/K38 have a built-in video signal output circuit to output necessary signal for TV screen adjustment.

Mode : NTSC (at fc = 8.056 MHz)
PAL (at fc = 8.000 MHz)
Picture pattern : Total eight types, Monochromatic inversion possible
Output format : Three states (H, L, Hi-Z) output

| | |
|---|---|
| Comp.Sync duration time | L output |
| Black level / Pedestal duration time | Hi-Z output |
| White level duration time | H output |

### 2.12.1 Configuration

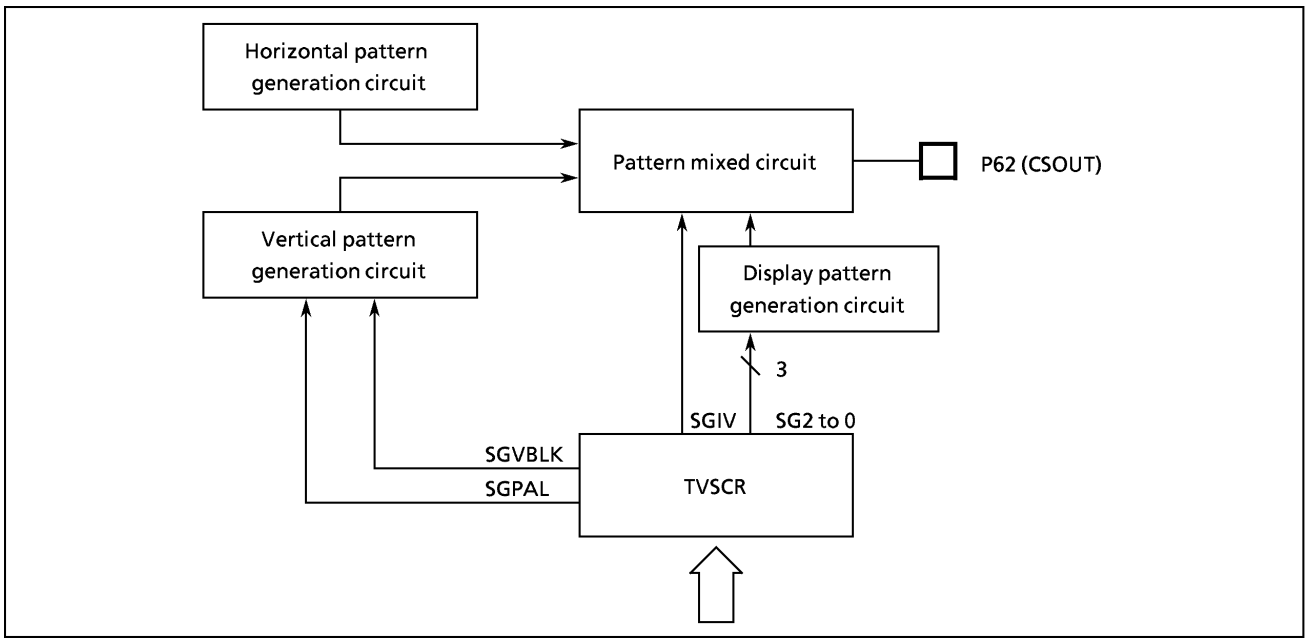


Figure 2-68. Test video signal output circuit

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## 2.12.2 Control

The test video signal output circuit can be controlled with the test video signal control register.

TVSCR ($0FB0_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| SGEN | SGVBLK | SGPAL | SGIV | SGCHS | SGPAT | | |

(Initial Value 0000 0000)

| | | | |
|---|---|---|---|
| SGEN | SG function selection | 0 : disable<br>1 : enable | Write only |
| SGVBLK | Picture signal for VBLK duration time | 0 : Output<br>1 : No output | |
| SGPAL | PAL/NTSC selection | 0 : NTSC<br>1 : PAL | |
| SGIV | Pattern monochromatic inversion | 0 : No inversion<br>1 : Inversion | |
| SGCHS | OSD synchronous signal selection | 0 : Port<br>1 : Pseudo signal circuit | |
| SGPAT | Display pattern | 000 : Black on the whole screen<br>001 : White on the whole screen<br>010 : Cross hatch<br>011 : Cross dot pattern<br>100 : Cross bar<br>101 : White on the upper side / Black on the lower side<br>110 : H signal pattern<br>111 : H resolution pattern | |

Figure 2-69. Test video signal control register

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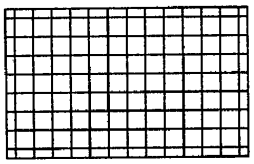
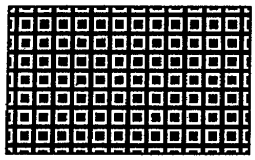
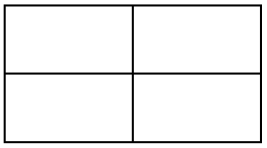
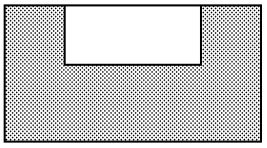
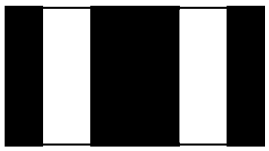
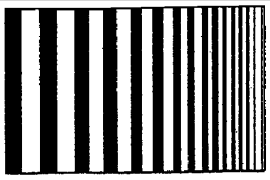
### 2.12.3 Functions

Video signal output is to generate monochromatic picture signal output to take easily the necessary tests such as TV screen white adjustment and screen distortion amplitude adjustment implemented on the final manufacturing process of a TV receiver set.

| Display pattern | TV screen |
|---|---|
| 000<br>(Black on the whole surface) |  |
| 001<br>(White on the whole surface) |  |
| 010<br>(Cross hatch) |  |
| 011<br>(Cross dot) |  |
| 100<br>(Cross bar) |  |
| 101<br>(White on the upper side /<br>Black on the lower side) |  |
| 110<br>(H signal pattern) |  |
| 111<br>(H resolution pattern) |  |

Display pattern and TV screen

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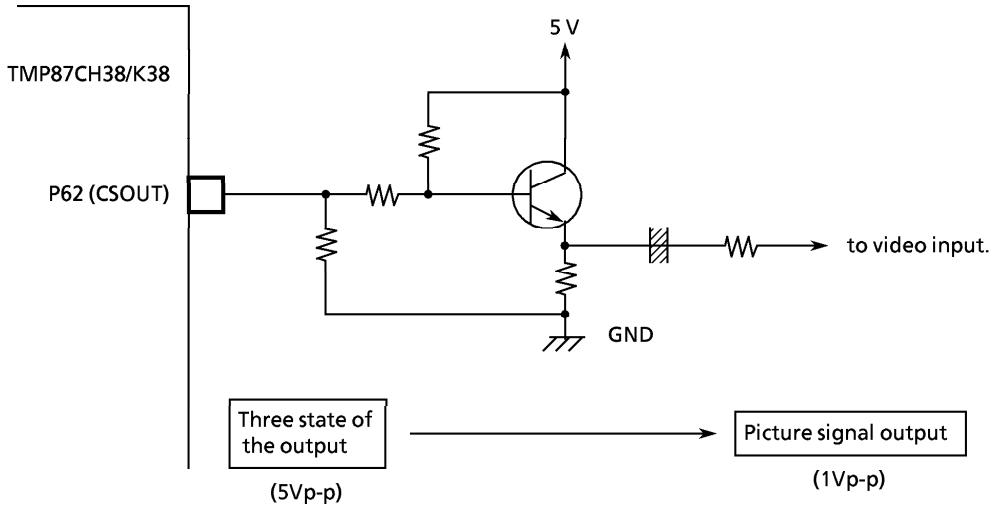
There are three states of the output to generate picture signal with the external circuit of the resistance divided voltage.

Example of picture output generation)

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## 2.13 On-Screen Display (OSD) Circuit

The TMP87CH38/K38 features a built-in on-screen display circuit used to display characters and symbols on the TV screen. 192 characters in any of 256 character fonts can be displayed in 24 characters x 8 rows.

OSD circuit functions are as follows:

① Number of character fonts 256 (including blank character)
② Number of display characters 192 (24 characters x 8 rows)
③ Composition of a character 14 x 18 dots
④ Character sizes 3 (selectable line by line)
⑤ Display colors Character colors : 8 (selectable character by character)
Fringe color : 8 (selectable page by page)
Background color : 8 (selectable page by page)
⑥ Fringing function (for large, middle, and small characters)
⑦ Smoothing function (for large and middle characters)
⑧ Display position horizontal : 128 steps ; vertical : 256 steps
⑨ Full-raster blanking function
⑩ Blinking function
⑪ Reverse function
⑫ Reverse Blinking function
⑬ Window function

### 2.13.1 Configuration

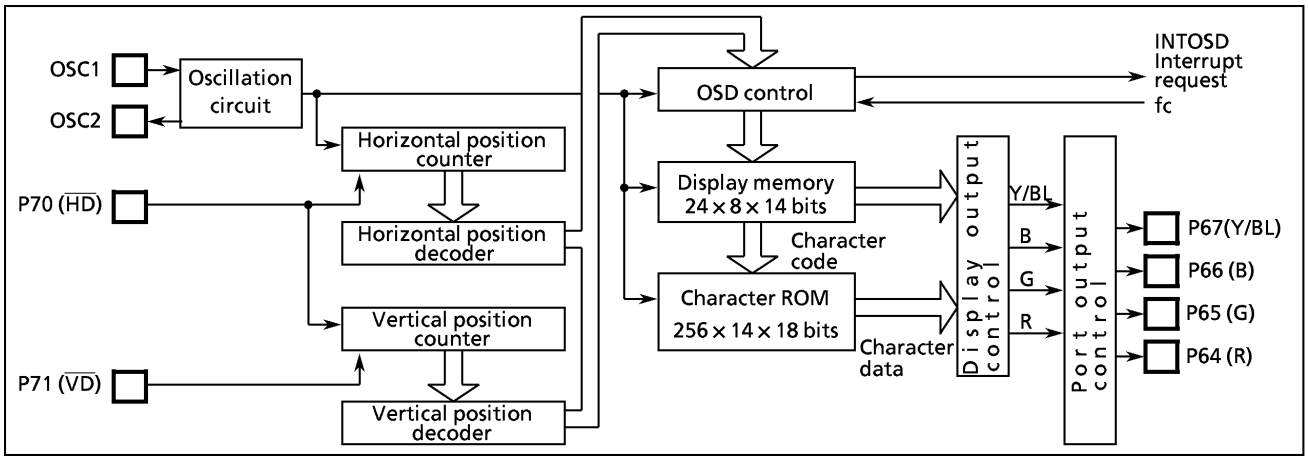


Figure 2-70. OSD circuit

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## 2.13.2 Character ROM and display memory

(1) Character ROM

The character ROM contains 256 character fonts. The user can set fonts as desired. The character ROM consists of 256 characters in 14 x 18 dots (character codes $00_H$ to $FF_H$). Each dot corresponds to one bit in the character ROM. When a bit in the character ROM is set to "1", the corresponding dot is displayed; if set to "0", the dot is not displayed. The start address in the character ROM corresponding to a character code is determined by the following expression:

$$\text{Start address in character ROM} = CRA \times 40_H + 4000_H$$

Since character code $00_H$ is used as blank character, the character font for this character code cannot be changed. Write "0" in the data of character code $00_H$.

Set all unused bits (bit 7 with $0_H$ to $8_H$ in the lower 4-bit of an address) to "1" and write the data "$FF_H$" to all unused address (the lower 4-bit of an address are $9_H$ to $F_H$) in character ROM.

Figure 2-71. (a) shows an example of the character font configuration for the character code $00_H$ and $01_H$, together with the ROM addresses and data.

Figure 2-71. (b) shows the character ROM dump list for these 2 character fonts .

*Note 1 : CRA ; Character code ($00_H$ to $FF_H$).*
*Note 2 : A data can not be read fran character ROM by software.*
*Note 3 : When ordering a mask, load the data to character ROM at addresses $4000_H$ to $7FFF_H$.*

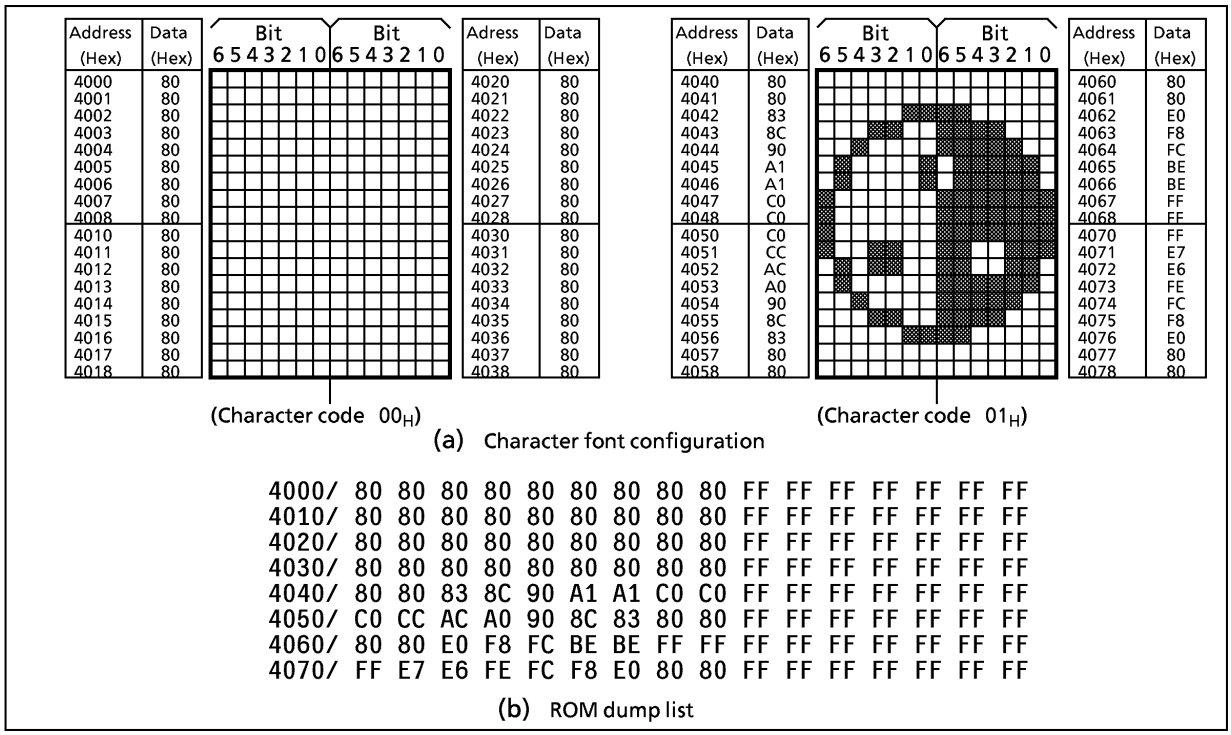


Figure 2-71. Character font configuration and ROM dump list

(2) Display memory

Each character out of the 192 characters displayed in 24 characters x 8 rows consists of 14 bits in the display memory. Five data items are written to the display memory: character code, color data, blinking specification, reverse specification, and reverse blinking specification. The display memory contents become unstable after the reset operation is released.

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There are two modes for writing display data to the display memory. One mode is for writing all display data (character code, color data, blinking specification, reverse specification, and reverse blinking specification) simultaneously. The other mode is for changing either character code or character ornamentation data (color data, blinking specification, reverse specification, and reverse blinking specification). How the display data is written to the display memory is described in section 2.13.3 (18).

Display memory configuration

- Character code specification register (8 bits) ··· CRA7 to 0
- Color data specification register (3 bits) ··· RDT / GDT / BDT
- Blinking specification register (1 bit) ··· BLF
- Reverse specification register (1 bit) ··· RVF
- Reverse blinking specification register (1 bit) ··· RBF

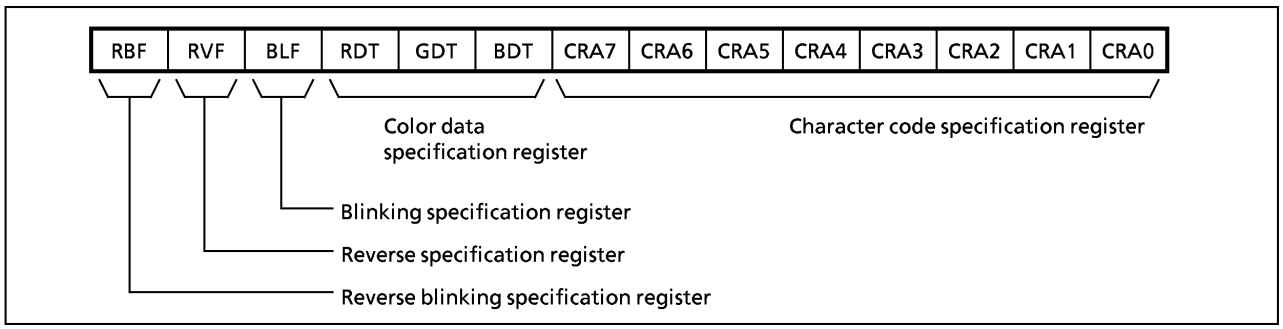


Figure 2-72. Display memory bit configuration

| Character / Row | 1 | 2 | 3 | 4 | 5 | 6 | 7 | 8 | 9 | 10 | 11 | 12 | 13 | 14 | 15 | 16 | 17 | 18 | 19 | 20 | 21 | 22 | 23 | 24 |
|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|
| 1 | 000 | 001 | 002 | 003 | 004 | 005 | 006 | 007 | 008 | 009 | 00A | 00B | 00C | 00D | 00E | 00F | 010 | 011 | 012 | 013 | 014 | 015 | 016 | 017 |
| 2 | 020 | 021 | 022 | 023 | 024 | 025 | 026 | 027 | 028 | 029 | 02A | 02B | 02C | 02D | 02E | 02F | 030 | 031 | 032 | 033 | 034 | 035 | 036 | 037 |
| 3 | 040 | 041 | | | | | | | | | | | | | | | | | | | | | | 057 |
| 4 | 060 | 061 | | | | | | | | | | | | | | | | | | | | | | 077 |
| 5 | 080 | 081 | | | | | | | | | | | | | | | | | | | | | | 097 |
| 6 | 0A0 | 0A1 | | | | | | | | | | | | | | | | | | | | | | 0B7 |
| 7 | 0C0 | 0C1 | | | | | | | | | | | | | | | | | | | | | | 0D7 |
| 8 | 0E0 | 0E1 | | | | | | | | | | | | | | | | | | | | | | 0F7 |

*Note: Numerals in the table indicate (hexadecimal) addresses in the display memory.*

Figure 2-73. Display memory address configuration

## 2.13.3 OSD circuit control

The OSD circuit is controled by using the OSD control registers assigned to addresses $0F80_H$ to $0F9A_H$ in the data buffer register (DBR). For write to or read from the OSD control registers, see section 2.13.3 (19). The OSD control registers are used to set display start position, display character ornamentations (that is, fringing, smoothing, color data, character size, and etc.), display memory addresses, character codes, and etc.

After all settings are complete, setting the display on-off control bit, EDISP (bit 0 in ORDON) to "1" enables display (starts display). Setting EDISP to "0" disables display (halts display).

*Note : The contents of OSD control registers are not initialized in STOP mode.*

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(1) Display position

The horizontal display start position can be set in 128 steps. The vertical display start positions can be specified for each line using 256 steps. The horizontal display start position is set with OSD control registers HS16 to HS10 (bit 6 to 0 in ORHS1). The vertical display start position of the line 1 is set with VS17 to VS10 (in ORVS1). The vertical display start position of the line 2 to 8 are determined by setting VS27 to VS20 ... VS87 to VS80 (ORVS2 to ORVS8) in the same way.

Horizontal display start position

Specification unit : Display page

Specification steps : 128

Specification horizontal display start position : Line 1 to 8 : HS16 to HS10

When FORS is "0" (Normal mode)

$HS1 = (HS16 \text{ to } HS10)_H \times 2T_{OSC} + 11T_{OSC}$ (Line1 to 8)

When FORS is "1" (Double frequency mode)

$HS1 = (HS16 \text{ to } HS10)_H \times T_{OSC} + 6.5T_{OSC}$ (Line1 to 8)

*Note : $T_{OSC}$ ; One cycle of OSC oscillation*

Vertical display start position

Specification unit : Line

Specification steps : 256

Specification vertical display start position : Line1 : VS17 to VS10
Line2 : VS27 to VS20
.
Line8 : VS87 to VS80

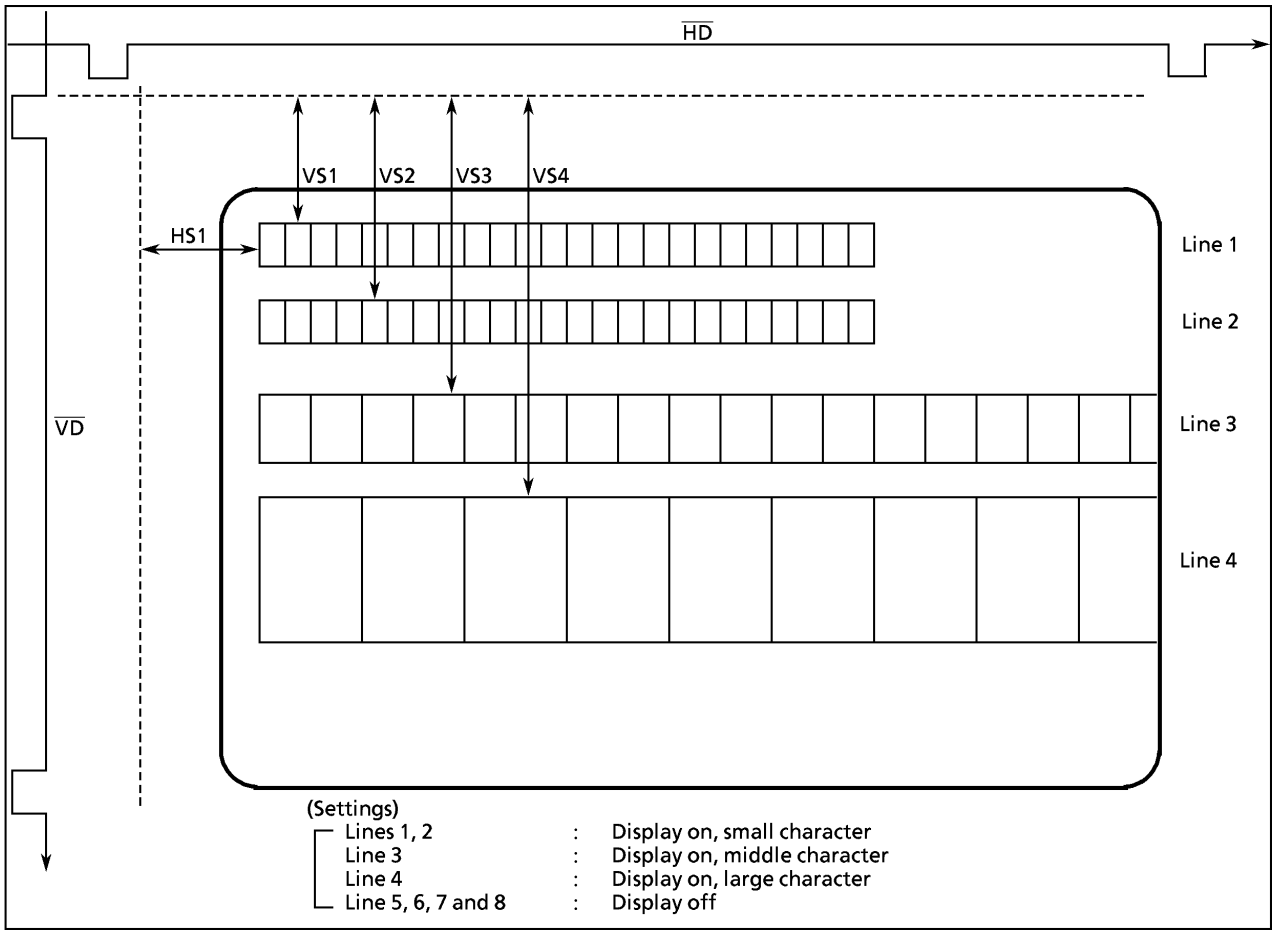


Figure 2-74. TV screen image

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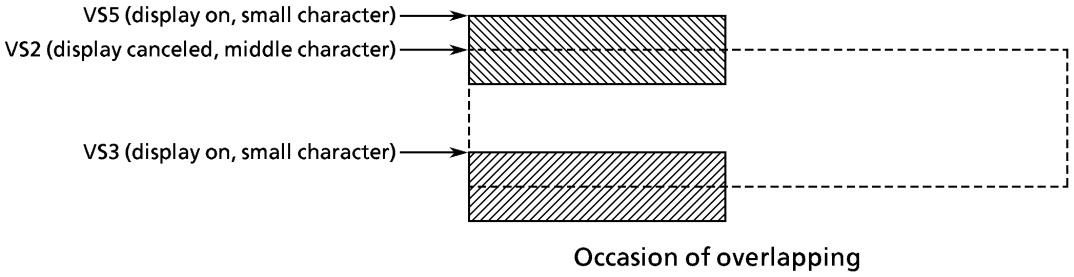
When VDSMD is "0" (Normal mode)
Line n : VSn = (VSn7 to VSn0) $_H \times 2T_{HD\ (n\ ;\ 1\ to\ 8)}$

When VDSMD is "1" (Double scan mode)
Line n : VSn = (VSn7 to VSn0) $_H \times 4T_{HD\ (n\ ;\ 1\ to\ 8)}$

*Note1 : $T_{HD}$ ; One cycle of $\overline{HD}$ signal*
*Note2 : If display lines are overlapped each other, previous display line is enabled and next line is disabled. Set the vertical display start position not to overlap display lines.*



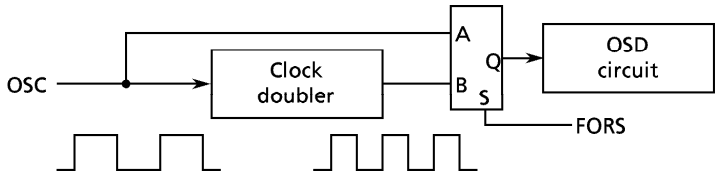

Occasion of overlapping

*Note3 : The line which is displayed off is managed as a small size character line. It is recommendable that its vertical display start position should be set out of TV screen.*
*Note4 : Transfer the contents of vertical display start position registers into OSD circuit before the position of the scanning line coincides with their own vertical display start position.*

(2) Double scan mode
The double scan mode is used to handle non-interlaced scanning TV. When double scan mode is enabled, the vertical display counter increases every 4 scan lines and a vertical size of a dot is double. This function is enabled by setting VDSMD (bit 3 in ORETC) in the OSD control register to "1".

Scan mode select register (1 bit) ··· VDSMD (bit 3 in ORETC)
"0" ··· Normal mode
"1" ··· Double scan mode

(3) Double OSC frequency mode
The double OSC frequency mode is used to display OSD by the OSC frequency doubled. When this function is enabled, the clock which is doubled by a clock doubler is inputted into an OSD circuit.




This function is ebabled by setting FORS (bit 4 in ORDON) in the OSD control register to "1".

OSC frequency select register (1 bit) ··· FORS (bit 4 in ORDON)
"0" ··· Normal mode
"1" ··· Double frequency mode

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(4) Character sizes and display on / off

Character size can be selected line by line from 3 sizes. And display on / off also can be set line by line. Small, middle and large character size and display on / off can be set with OSD control registers CS11, CS10...CS81, CS80 (ORCS4 and ORCS8) in the OSD control registers.

Character sizes : 3 sizes (Small, middle and large)
Character size and display on / off specification unit : Line
Character size select/display on / off register (2 bits x 8)
Line 1: CS11 and CS10
Line 2: CS21 and CS20
: :
Line 8: CS81 and CS80

Table 2-11. Character size and display on / off specifications (n;1 to 12)

| CSn1 | CSn0 | Character size | Display on/off |
|---|---|---|---|
| 1 | 1 | Small | On |
| 1 | 0 | Middle | On |
| 0 | 1 | Large | On |
| 0 | 0 | – | Off |

*Note : The line which is displayed off is managed as a small character sizeline by the overlap of vertical display start position, the display line counter function, and etc.*

Table 2-12. Dot and character sizes

| | | VDSMD = 0 (Normal mode) | | VDSMD = 1 (Double scan mode) | |
|---|---|---|---|---|---|
| | | Dot size | Character size | Dot size | Character size |
| FORS = 0 (normal mode) | Small | 1 $T_{OSC}$ × 1 $T_{HD}$ | 14 $T_{OSC}$ × 18 $T_{HD}$ | 1 $T_{OSC}$ × 2 $T_{HD}$ | 14 $T_{OSC}$ × 36 $T_{HD}$ |
| | Middle | 2 $T_{OSC}$ × 2 $T_{HD}$ | 28 $T_{OSC}$ × 36 $T_{HD}$ | 2 $T_{OSC}$ × 4 $T_{HD}$ | 28 $T_{OSC}$ × 72 $T_{HD}$ |
| | Large | 4 $T_{OSC}$ × 4 $T_{HD}$ | 56 $T_{OSC}$ × 72 $T_{HD}$ | 4 $T_{OSC}$ × 8 $T_{HD}$ | 56 $T_{OSC}$ × 144 $T_{HD}$ |
| FORS = 1 (double frequency mode) | Small | 0.5 $T_{OSC}$ × 1 $T_{HD}$ | 7 $T_{OSC}$ × 18 $T_{HD}$ | 0.5 $T_{OSC}$ × 2 $T_{HD}$ | 7 $T_{OSC}$ × 36 $T_{HD}$ |
| | Middle | 1 $T_{OSC}$ × 2 $T_{HD}$ | 14 $T_{OSC}$ × 36 $T_{HD}$ | 1 $T_{OSC}$ × 4 $T_{HD}$ | 14 $T_{OSC}$ × 72 $T_{HD}$ |
| | Large | 2 $T_{OSC}$ × 4 $T_{HD}$ | 28 $T_{OSC}$ × 72 $T_{HD}$ | 2 $T_{OSC}$ × 8 $T_{HD}$ | 28 $T_{OSC}$ × 144 $T_{HD}$ |

*Note : $T_{OSC}$; One cycle of OSC oscillation $T_{HD}$; One cycle of $\overline{HD}$ signal*

(5) Smoothing function

The smoothing function is used to make characters look smooth. Enabling smoothing displays 1/4 dot between two dots connecting corner to corner within a character. Small size character can not be enabled smoothing. Smoothing is enabled by setting ESMZ (bit 4 in ORETC) in the OSD control register to "1".

Smoothing specification unit: Display page
Smoothing specification register (1 bit) ··· ESMZ (bit 4 in ORETC)
"0" ··· Disable smoothing
"1" ··· Enable smoothing

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(6) Fringing function

The fringing function is used to display a character with a fringe width is 1/2 dot in a different color from that of the character. For small characters, fringe width is 1 dot. When a character is displayed with the maximum of 14 vertical dots and 18 horizontal dots, the fringe exceeds right and left, top, and bottom of the character display area. The exceeded fringe can be displayed; however, display characters have higher priority to fringe horizontally.

Fringing is enabled for each line by setting EFR1 to EFR8 (OREFR) in the OSD control register to "1".

A color for fringe is specified common to all lines using OSD control registers, RFDT, GFDT, and BFDT (bit 2 to 0 in ORBK).

Fringing specification unit: Line

Fringing enable register (1 bit x 8) ··· EFRn (n ; 1 to 8) (OREFR)

"0" ··· Disable fringing

"1" ··· Enable fringing

Fringe color specification unit: Display page

Fringe color register (3 bits) ··· RFDT, GFDT, BFDT (bit 2 to 0 in ORBK)

*Note : When a display line is enable fringing function, its vertical size is increased by one dot (by two dots when its character size is small) independent of its character font.*
*Therefore, when a vertical display start position is specified to no space between the lines, the display line which is overlapped with increasing dot (s) is canceled.*

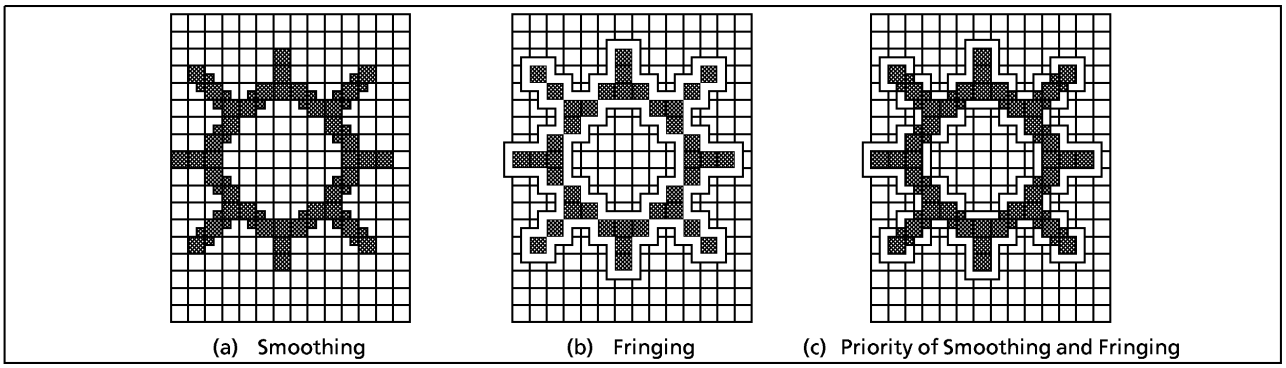


Figure 2-75. Smoothing / fringing / priority of smoothing and fringing

(7) Background color function

Background color function is used to color the entire background for the character area (14 × 18 dots). Except the character area whose character code is $00_H$

This function is specified for each display page by setting EBKGD (bit 7 in ORBK) in the OSD control register to "1".

A background color is specified for each display page by setting RBDT, GBDT, and BBDT (bit 5 to 3 in ORBK) in the OSD control registers. A Color specification is same as them for full-roster blanking.

Background specification unit: Display page

Background enable register (1 bit) ··· EBKGD (bit 7 in ORBK)

"0" ··· Disable background

"1" ··· Enable background

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Background color specification unit: Display page
Background color specification registers (3 bits) ··· RBDT, GBDT, BBDT (bit 5 to 3 in ORBK)

*Note: When the background color function is used, the blank character (Code $00_H$) can not be used as the first character on the fringing line.*

(8) Full-raster blanking function

Full-raster blanking function is used to color the entire background for the display area (TV screen). When using the full-raster blanking function, set YBLCS (bit7 in ORETC) to "1", output BL signal from Y/BL pin, because Y signal cannot delete whole display page from video signal.
This function is specified for each display page by setting EXBL (bit 6 in ORBK) in the OSD register to "1". Color specification is same as them for background color.

Full-raster blanking specification unit: Display page
Full-raster blanking enable register (1 bit) ··· EXBL (bit 6 in ORBK)
"0" ··· Disable full-raster blanking
"1" ··· Enable full-raster blanking

Full-raster blanking color specification registers (3 bits) ··· RBDT, GBDT, BBDT (bit 5 to 3 in ORBK)

(9) Reverse function

This function is used to reverse the background and character colors. However, when fringing is specified, the fringe color does not change.
Reverse function is enabled by setting RVF (bit 4 in ORDSN) in the OSD control register to "1".

Reverse specification: Character
Reverse enable register (1 bit) ··· RVF (bit 4 in ORDSN)
"0" ··· Disable reverse
"1" ··· Enable reverse

(10) Reverse blinking function

Reverse blinking function is used to reverse the background and character colors.
When RBMF is "1", characters specified for blinking by RBF are reversed the background and character colors. However, when fringing is specified, the fringe color does not change.

Reverse blinking specification unit: Character
Reverse blinking specification register (1 bit) ··· RBF (bit 5 in ORDSN)
"0" ··· No reverse blinking
"1" ··· Reverse blinking
Reverse blinking master specification register (1 bit) ··· RBMF (bit 5 in ORETC)
"0" ··· Disable reverse blinking
"1" ··· Enable reverse blinking
(Characters whose RBF is set to "1" are reversed the background and character colors.)

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Table 2-13. Display mode

| RBF | RVF | RBMF | Display |
|---|---|---|---|
| 0 | 0 | * | Normal |
| 0 | 1 | * | Reverse |
| 1 | 0 | 0 | Normal |
| | | 1 | Reverse |
| 1 | 1 | * | reserved |

* ; don't care

(11) Blinking function

Blinking function is used to blink display characters.

When BKMF is "1", characters specified for blinking by BLF are not displayed. (If the background color function is used, the background color is not disappeared.)

Blinking specification unit : Character

Blinking specification register (1 bit) ··· BLF (bit 3 in ORDSN)

"0" ··· No blinking

"1" ··· Blinking

Blinking master specification register (1 bit) ··· BKMF (bit 6 in ORETC)

"0" ··· Disable blinking

"1" ··· Enable blinking (Characters whose BLF are set to "1" are not displayed.)

(12) Character

Characters: 256 (including blank character)

Character specification register (8 bits) ··· CRA7 to CRA0 (bit 7 to 0 in ORCRA)

Character code "$00_H$" ··· Blank character

Character code "$01_H$" to "$FF_H$" ··· User programmable by character ROM

(13) Character color

Character colors: 8

Character color specification unit: Character

Character color specification register (3 bits) ··· RDT / GDT / BDT (bit 2 to 0 in ORDSN)

Table 2-14. Character color

| RDT | GDT | BDT | Character Color |
|---|---|---|---|
| 0 | 0 | 0 | Black |
| 0 | 0 | 1 | Blue |
| 0 | 1 | 0 | Green |
| 0 | 1 | 1 | Cyan |
| 1 | 0 | 0 | Red |
| 1 | 0 | 1 | Magenta |
| 1 | 1 | 0 | Yellow |
| 1 | 1 | 1 | White |

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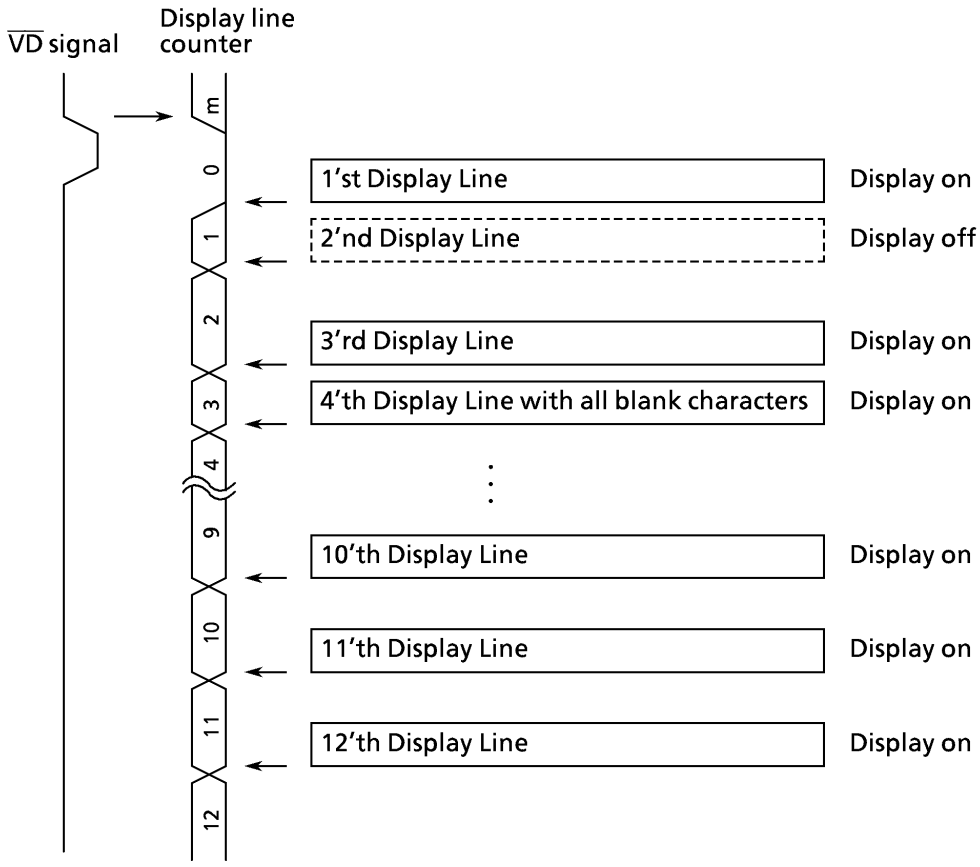
(14) OSD interrupt

1) Display line counter

The display line counter indicates number of display line(s) by OSD circuit on the TV screen. The display line counter is a 4-bit counter which is initialized to "0" by the falling edge of the $\overline{VD}$ signal and which increments when last scanning of each display line is completed (falling edge of the $\overline{HD}$ signal). It is necessary to be read out display line counter several times, because it does not synchronize CPU clock.

Display line counter register (4 bits) ··· DCTR (bit 3 to 0 in ORIRC)

"0000" ··· No display line is completed.
"0001" ··· 1'st display line is completed.
"0010" ··· 2'nd display line is completed.
to
"1111" ··· 15'th display line is completed.




*Note : The display line counter also increments when a line with all blank characters or a line with display off is specified.*

Figure 2-76. Display line counter

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2) Interrupt generator circuit

An interrupt request is generated when a falling edge of $\overline{VD}$ signal or when line counter (DCTR) is counted to the certain value specified by ISDC.

Interrupt souce select register (1 bit) ··· SVD (bit4 in ORIRC)

"0" ··· Interrupt request generated when the display line counter (DCTR) is counted to the certain value which is specified by ISDC.

"1" ··· Interrupt request is generated when a falling edge of $\overline{VD}$ signal.

Interrupt generation line specification register (4 bits) ··· ISDC (bit 3 to 0 in ORIRC)

"0000" ··· Interrupt request generated when the display line counter is cleared.

"0001" ··· Interrupt request generated at end points of the last scanning line of the first displey line

"0010" ··· Interrupt request generated at end points of the last scanning line of the 2'nd display line

to

"1111" ··· Interrupt request generated at end points of the last scanning line of the 15'th display line

(15) P6 port output select function

This function is used to select whether the contents of port P67 to P64 will be output or R, G, B, Y/BL signals of the OSD circuit will be output on pins P67 to P64.

P6 port output select registers (4 bits) ··· P67DS to P64DS (bit 7 to 4 in ORP6DS)

"1" ··· R, G, B, Y/BL signal output

"0" ··· Port contents output

(16) OSD pin output polarity control function

This function is used to select the polarity of the OSD outputs for RGB and Y/BL.

Output polarity control register (3 bits) ··· BLIV, YIV, RGBIV (bit 7 to 5 in ORIRC)

Table 2-15. Control of OSD Output Polarity

| Symbol | Output port | Data "0" | Data "1" |
|---|---|---|---|
| BLIV | BL | Active High | Active Low |
| YIV | Y | Active High | Active Low |
| RGBIV | RGB | Active High | Active Low |

(17) Y/BL signal select function

This function is used to select either Y or BL signal output from the Y/BL pin.

Y/BL signal select register (1 bit) ··· YBLCS (bit 7 in ORETC)

"0" ··· Y signal output

"1" ··· BL signal output

Y signal ··· Logical OR for R, G, B Character data, and Fringing data.

BL signal ··· When EXBL is "0":

Output in all display character areas

(except for character code $00_H$: blank character)

When EXBL is "1":

Output in the whole page

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(18) Writing display data to the display memory

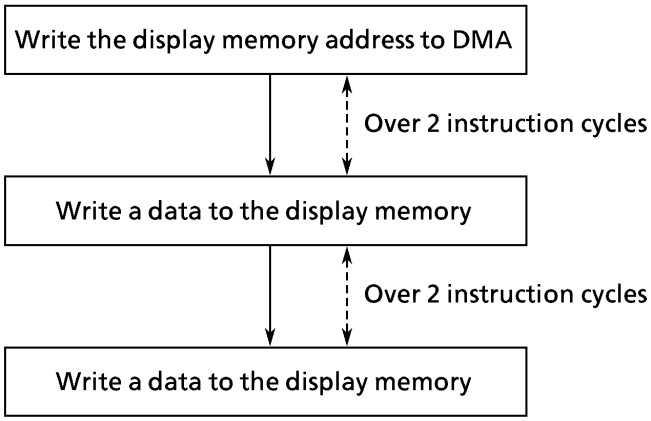
Data are written to the display memory using DMA8 to DMA0, CRA7 to CRA0, RDT, GDT, BDT, BLF, RBF, RVF, and MBK registers.

Display memory address specification register (9 bits) ···
DMA8 to DMA0 (bit0 in ORETC and ORDMA)

Display memory bank switching register (1 bit) ··· MBK (bit1 in ORETC)
"0" ··· For changing either character code or character ornamentation
"1" ··· For changing both character code and character ornamentation

*Note 1: Don't use the 2 bytes transfer operation such as [LDW (HL), mn] when accessing to display memory.*

*Note 2: When writing a data to the display memory immediately after setting the display memory address to DMA, or when continuously writing a data to the display memory, insert over 2 instruction cycles between the instrunction for writing a data.*




Example: Setting a character code ($20_H$) to the display memory (Address : $020_H$ to $021_H$)

```
LD      HL, ORCRA           ; Set ORCRA address to HL reg.
LD      A, 20H              ; Load character code to A reg.

LD      DE, ORDMA           ; Set lower 8-bit addresses to DMA7 to 0
LD      (DE), 20H
LD      DE, ORETC           ; Set the most upper address to DMA8 and set MBK to "0"
LD      (DE), 00000001B

LD      DE, ORDMA           ; Set lower 8-bit addresses to DMA7 to 0
LD      (DE), 20H
LD      DE, ORETC           ; Set the most upper address to DMA8 and set MBK to "0"
LD      (DE), 00000000B

NOP                         ; Insert 2 instruction cycles
NOP

LD      (HL),A              ; Write a character code to display memory (Address : 120H)

NOP                         ; Insert 2 instruction cycles
NOP

LD      (HL),A              ; Write a character code to display memory (Address : 121H)
```

*Note 3: Transfer the contents of display memory which affect displaying characters into OSD circuit, before the position of scanning line coincides with their own vertical display start position.*

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a. Display memory write sequence when writing both character code and character ornamentation.

① Write lower 8-bit addresses of display memory to DMA7 to DMA0 (in ORDMA).

② Write the most upper addresses of display memory to DMA8 (bit 0 in ORETC) and set MBK (bit 1 in ORETC) to 1.

> *Note: It is necessary to write all bits of display memory address, writting DMA8 after DMA7 to DMA0, when writing display address, and repeat this sequence.*

③ Write character ornamentation data (blinking, reverse, reverse blinking, and color data) to RDT, GDT, BDT, BLF, RVF, and RBF.
At this time, the character ornamentation data is transferred to the display memory.

④ Write character code to CRA7 to CRA0.
At this time, character code is transferred to the display memory together with the character ornamentation data which is written in ③ and DMA8 to DMA0 are automatically incremented.

b. Display memory write sequence when writing either character code or character ornamentation

① Write lower 8-bit addresses of display memory to DMA7 to DMA0 (in ORDMA).
② Write the most upper address of display memory to DMA8 (bit 0 in ORETC) and clear MBK (bit 1 in ORETC) to 0.

> *Note: It is necessary to write all bits of display memory address, writting DMA8 after DMA7 to DMA0, when writing display address, and repeat this sequence.*

③ Write character ornamentation data (blinking, reverse, reverse blinking, and color data) to RDT, GDT, BDT, BLF, RVF, and RBF or write character code to CRA7 to CRA0.
At this time, written data are transferred to the display memory and DMA8 to DMA0 are automatically incremented.

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(19) OSD control register write / read

The address of the OSD control registers are assigned to the DBR area.

To write or to read from the OSD control registers, the method is the same as for accessing ordinary DBR registers.

The written data are transferred to the OSD circuit at the end point of the scanning line without display by setting RGWR register to "1" and become valid. And, be able to write value of OSD control register after RGWR flag is cleared to "0".

*Note 1 : Do not write the contents of OSD control registers during RGWR flag is "1". If contents of OSD control registers are written during RGWR flag is "1", the written data are broken.*

*Note 2 : Do not clear RGWR register to "0". If RGWR register is cleared to "0", the contents of OSD control registers may be transferred to OSD circuit at unexpected timing.*

*Note 3 : Insert over 3 instruction cycles between the instruction which sets RGWR register to "1" and the instruction which checks RGWR flag.*

*Note 4 : Transfer the contents of all OSD control registers which affect displaying characters into OSD circuit before the position of scanning line coincides with their own vertical display start position.*

Example 1 : In the case of writing the data into OSD registers, setting RGWR register to "1" and checking RGWR flag

Writing the data into OSD registers

↓

```
        LD     A, (TEMP_ORDON)      ; Set bit 2 of work-area to "1"
        SET    A.2

        LD     HL, ORDON            ; Set RGWR register to "1" (Request data transfer)
        LD     (HL), A

        NOP                         ; Insert 3 instruction cycles
        NOP
        NOP
CHECK_RGWR_FLAG:
        TEST   (HL).2               ; Check RGWR flag until RGWR flag is "0"
        JR     F,CHECK_RGWR_FLAG
```

Example 2 : In the case of checking RGWR flag, writing the data into OSD registers, and writing RGWR register to "1"

```
        LD     HL, ORDON
CHECK_RGWR_FLAG:
        TEST   (HL).2               ; Checking RGWR flag until RGWR flag is "0"
        JR     F,CHECK_RGWR_FLAG
```

↓

Writing the data into OSD registers

↓

```
        LD     A, (TEMP_ORDON)      ; Set bit 2 of work-area to "1"
        SET    A.2

        LD     HL, ORDON            ; Set RGWR register to "1" (Request data transfer)
        LD     (HL), A

        NOP                         ; Insert 3 instruction cycles
        NOP
        NOP
```

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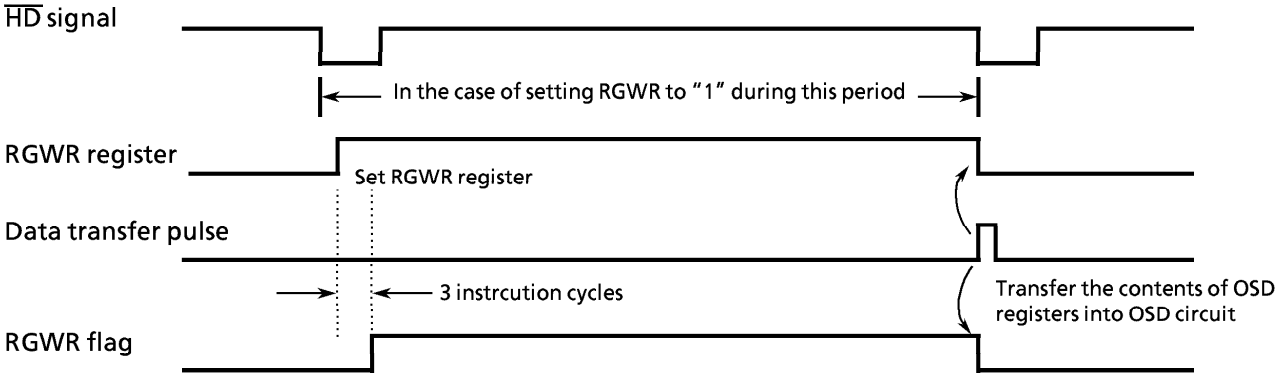
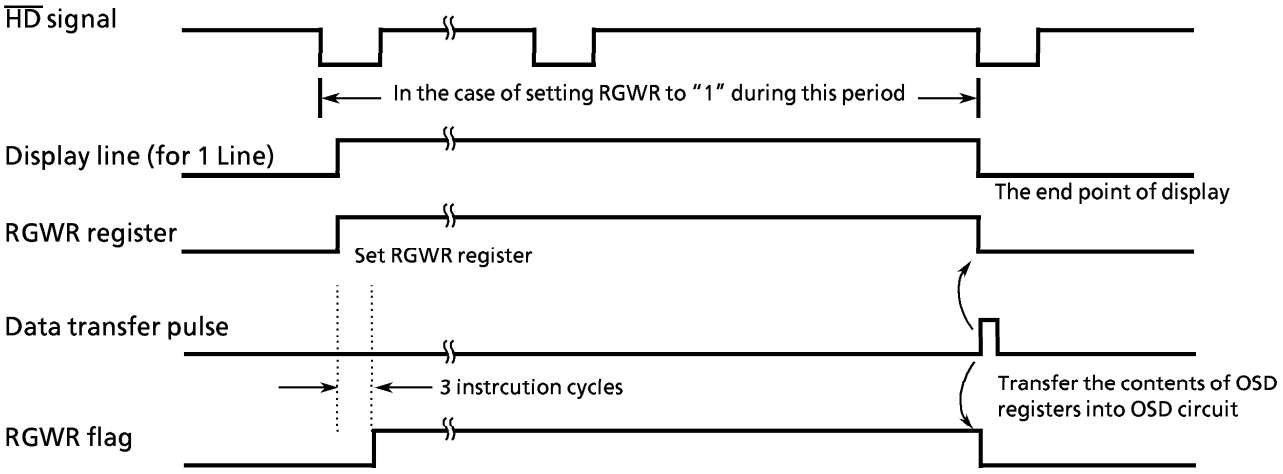
The timing chart of transferring the contents of OSD registers into OSD circuit is shown as follows;

1. In the case of setting RGWR register to "1" during the position of the scanning line is in no display area (except any lines specified as display off by CSn).
   The contents of OSD registers are transferred into OSD circuit when the position of the scanning line is at the falling edge of $\overline{HD}$ signal.




2. In the case of setting RGWR register to "1" during the position of the scanning line is in display area (including any lines specified as display off by CSn).
   The contents of OSD registers are transferred into OSD circuit when the position of the scanning line is at the falling edge of $\overline{HD}$ signal of finishing the display line.




For registers (DMA8 to DMA0, CRA7 to CAR0, RDT, GDT, BDT, BLF, RBF, RVF, MBK) used for updating the display memory, P67DS to P64DS, YBLCS, BKMF, RBMF, ESMZ, VDSMD, FORS and RGWR, the data become valid as soon as they are written.

Written data transfer register (1 bit) ··· RGWR (bit 2 in ORDON)
- "0" ··· Initial state
- "1" ··· Transfer written data to OSD circuit. (After transfer, RGWR register and RGWR flag are automatically cleared to "0".)

Written data transfer monitor flag (1 bit) ··· RGWR (bit 2 in ORDON)
- "0" ··· Transfer completed.
- "1" ··· During transfer

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(20) Display on / off

This function is used to display characters specified for on / off display.

Display on / off specification unit : Display page

Display on / off specification register (1 bit) ··· EDISP (bit 0 in ORDON)

"0" ··· Disable display

"1" ··· Enable display

*Note: Do not start STOP mode during display is enable.*

(21) Window function

This function is used to set upper and lower limit of display page. Window upper limit is specified by WVSH (ORWVSH). Window lower limit is specified by WVSL (ORWVSL). This function is enabled by setting EWDW (bit 1 in ORDON ) in the OSD control register to 1.

Window specification unit: Display page

Window function enable specification register (1 bit) ··· EWDW (bit 1 in ORDON)

"0" ··· Disable window function

"1" ··· Enable window function

Window upper limit specification register (8 bits) ··· WVSH7 to 0 (ORWVSH)

Window lower limit specification register (8 bits) ··· WVSL7 to 0 (ORWVSL)

Window upper and lower limit position ···

When VDSMD is "0" (Normal mode) :

$WVSH = (WVSH7 \text{ to } WVSH0)_H \times 2T_{HD}$

$WVSL = (WVSL7 \text{ to } WVSL0)_H \times 2T_{HD}$

When VDSMD is "1" ( Double scan mode) :

$WVSH = (WVSH7 \text{ to } WVSH0)_H \times 4T_{HD}$

$WVSL = (WVSL7 \text{ to } WVSL0)_H \times 4T_{HD}$

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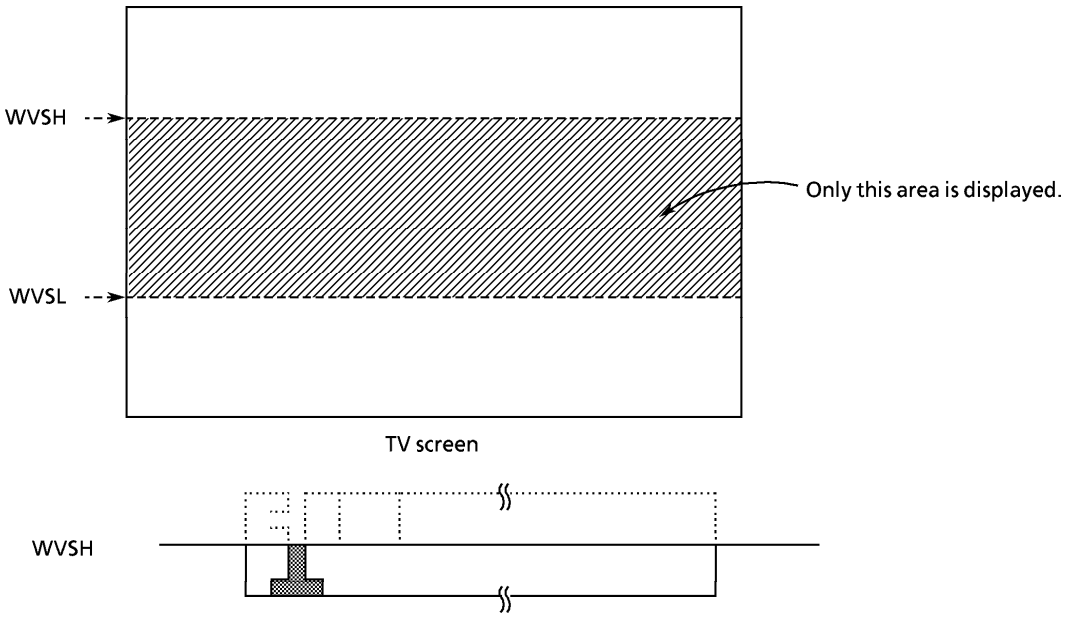
*Note 1 :* $T_{HD}$ *; One cycle of* $\overline{HD}$ *signal*

*Note 2 :* $WVSL > WVSH \geqq$ "1"

*Note 3 : Modify the value of window upper and lower limit register as follows:*

1. *When* $WVSH_{NEW} \leqq WVSH_{OLD}$
   *Finish to transfer the new value, during* $\overline{VD}$ *signal is low or before the position of the scanning line coincides with* $WVSH_{NEW}$.
2. *When* $WVSL > WVSH_{NEW} > WVSH_{OLD}$
   *Finish to transfer the new value, during* $\overline{VD}$ *signal is low or before the position of the scanning line coincides with* $WVSH_{OLD}$.
3. *When* $WVSL_{NEW} \leqq WVSL_{OLD}$
   *Finish to transfer the new value, during* $\overline{VD}$ *signal is low or before the position of the scanning line coincides with* $WVSL_{NEW}$.
4. *When* $WVSL_{NEW} > WVSL_{OLD}$
   *Finish to transfer the new value, during* $\overline{VD}$ *signal is low or before the position of the scanning line coincides with* $WVSL_{OLD}$.

*Note 4 : It is recommendable that the window function is always enabled (EWDW = "1") and set WVSH to "*$01_H$*", WVSL to "*$FE_H$*". When the window function should be set to disable, clear EWDW to "0" independent of the value which this register has been set from detecting the rising edge of* $\overline{HD}$ *signal by software until the falling edge of* $\overline{HD}$ *signal.*

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(22) OSD control registers

Can not access all OSD control registers in any of read-modify-write instructions such as bit operation, etc.

**ORHS1** ($0F80_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | |
|---|---|---|---|---|---|---|---|---|
| "0" | HS16 | HS15 | HS14 | HS13 | HS12 | HS11 | HS10 | (Initial value *000 0000) |

| HS16 to 10 | Horizontal display start position | Write only |
|---|---|---|

**ORVS1** ($0F81_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | |
|---|---|---|---|---|---|---|---|---|
| VS17 | VS16 | VS15 | VS14 | VS13 | VS12 | VS11 | VS10 | (Initial value 0000 0000) |

**ORVS8** ($0F88_H$)

| VS87 | VS86 | VS85 | VS84 | VS83 | VS82 | VS81 | VS80 | (Initial value 0000 0000) |
|---|---|---|---|---|---|---|---|---|

| VSn7 to 0 | Vertical display start position for line n | Write only |
|---|---|---|

(n = 1 to 8)

*Note 1: If display lines are overlapped each other, previous display line is enabled and next line is disabled. Set the vertical display start position not to overlap display lines.*

*Note 2: Transfer the contents of vertical display start position registers into OSD circuit before a position of the scanning line coincides with their own vertical display start position.*

**ORCS4** ($0F8D_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | |
|---|---|---|---|---|---|---|---|---|
| CS4 | | CS3 | | CS2 | | CS1 | | (Initial value 0000 0000) |

**ORCS8** ($0F8E_H$)

| CS8 | | CS7 | | CS6 | | CS5 | | (Initial value 0000 0000) |
|---|---|---|---|---|---|---|---|---|

| CSn | Character size and display on/off for line n | 00: Display off<br>01 : Large size<br>10 : Middle size<br>11 : Small size | Write only |
|---|---|---|---|

(n = 1 to 8)

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**OREFR** ($0F90_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| EFR8 | EFR7 | EFR6 | EFR5 | EFR4 | EFR3 | EFR2 | EFR1 |

(Initial value **** 0000)

| | | | |
|---|---|---|---|
| EFRn | Fringing enable specification register | 0 : Disable fringing<br>1 : Enable fringing | Write only |

(n = 1 to 8)

*Note: When a display line is enabled fringing function, its vertical size is increased by one dot (by two dots when its character size is small) independent of its character font. Therefore, when a vertical display start position is specified to no space between the lines, the display line which is overlapped with increasing dot(s) is canceled.*

**ORP6DS** ($0F91_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| P67DS | P66DS | P65DS | P64DS | | | | |

(Initial value 0000 ****)

| | | | |
|---|---|---|---|
| P67DS to P64DS | P6 port output select | 0 : Port contents output<br>1 : R, G, B, Y/BL signal output | Write only |

**ORWVSH** ($0F92_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| WVSH7 | WVSH6 | WVSH5 | WVSH4 | WVSH3 | WVSH2 | WVSH1 | WVSH0 |

(Initial value 0000 0000)

| | | |
|---|---|---|
| WVSH7 to 0 | Window upper limit position (WVSL > WVSH ≧ 1) | Write only |

**ORWVSL** ($0F93_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| WVSL7 | WVSL6 | WVSL5 | WVSL4 | WVSL3 | WVSL2 | WVSL1 | WVSL0 |

(Initial value 0000 0000)

| | | |
|---|---|---|
| WVSL7 to 0 | Window lower limit position (WVSL > WVSH ≧ 1) | Write only |

**ORBK** ($0F94_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| EBKGD | EXBL | RBDT | GBDT | BBDT | RFDT | GFDT | BFDT |

(Initial value 0000 0000)

| | | | |
|---|---|---|---|
| EBKGD | Background function enable specification register | 0 : Disable background<br>1 : Enable background | Write only |
| EXBL | Full-raster blanking enable specification register | 0 : Disable full-raster blanking<br>1 : Enable full-raster blanking | |
| RBDT / GBDT / BBDT | Background color select | 000 : Black<br>001 : Blue<br>010 : Green<br>011 : Cyan<br>100 : Red<br>101 : Magenta<br>110 : Yellow<br>111 : White | |
| RFDT / GFDT / BFDT | Fringing color select | 000 : Black<br>001 : Blue<br>010 : Green<br>011 : Cyan<br>100 : Red<br>101 : Magenta<br>110 : Yellow<br>111 : White | |

*Note: When the background color function is used, the blank character (code $00_H$) can not be used as the first character on the fringing line.*

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**ORIRC** ($0F95_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| BLIV | YIV | RGBIV | SVD | ISDC | | | |

(Initial value 0000 0000)

| | | | |
|---|---|---|---|
| BLIV | BL output polarity select | 0 : Active high<br>1 : Active low | Write only |
| YIV | Y output polarity select | 0 : Active high<br>1 : Active low | |
| RGBIV | R, G, B output polarity select | 0 : Active high<br>1 : Active low | |
| SVD | Interrupt source select | 0: Interrpt request by ISDC value<br>1 : Interrupt request at falling edge of $\overline{VD}$ signal | |
| ISDC | Interrupt generation line select | | |

**ORIRC** ($0F95_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| | | | | DCTR | | | |

(Initial value **** 0000)

| | | |
|---|---|---|
| DCTR | Display line counter | Read only |

*Note: The display line counter also increments when a line with all blank data or a line with display off is specified.*

**ORETC** ($0F96_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| YBLCS | BKMF | RBMF | ESMZ | VDSMD | "0" | MBK | DMA8 |

(Initial value 0000 0000)

| | | | |
|---|---|---|---|
| YBLCS | Y/BL signal select | 0 : Y signal output<br>1 : BL signal output | Write only |
| BKMF | Blinking master enable specification register | 0 : Disable blinking<br>1 : Enable blinking | |
| RBMF | Reverse blinking master enable specification register | 0 : Disable reverse blinking<br>1 : Enable reverse blinking | |
| ESMZ | Smoothing enable specification register | 0 : Disable smoothing<br>1 : Enable Smoothing | |
| VDSMD | Double scan mode select | 0 : Normal mode<br>1 : Double scan mode | |
| MBK | Display memory bank switching | 0 : Access to either character code or character display options<br>1 : Access to Both character code and character display options | |
| DMA8 | Display memory address (bit 8) | | |

*Note1: Clear "0" to bit 2 in ORETC.*
*Note2: It is necessary to write all bits of display memory address, writing DMA8 after DMA7 to DMA0, when writing display address, and repeat this sequence.*

**ORDMA** ($0F97_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| DMA7 | DMA6 | DMA5 | DMA4 | DMA3 | DMA2 | DMA1 | DMA0 |

(Initial value 0000 0000)

| | | |
|---|---|---|
| DMA7 to 0 | Display memory address | Write only |

*Note: It is necessary to write all bits of display memory address, writing DMA8 after DMA7 to DMA0, when writing display address, and repeat this sequence.*

**ORDSN** ($0F98_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| | | RBF | RVF | BLF | RDT | GDT | BDT |

(Initial value **** ****)

| | | | |
|---|---|---|---|
| RBF | Reverse blinking enable specification register | 0 : Disable reverse blinking<br>1 : Enable reverse blinking | Write only |
| RVF | Reverse enable specification register | 0 : Disable reverse<br>1 : Enable reverse | |
| BLF | Blinking enable specification register | 0 : Disable blinking<br>1 : Enable blinking | |
| RDT/<br>GDT/<br>BDT | Character color select | 000 : Black<br>001 : Blue<br>010 : Green<br>011 : Cyan<br>100 : Red<br>101 : Magenta<br>110 : Yellow<br>111 : White | |

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**ORCRA** ($0F99_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| CRA7 | CRA6 | CRA5 | CRA4 | CRA3 | CRA2 | CRA1 | CRA0 |

(Initial value **** ****)

| | | |
|---|---|---|
| CRA7 to 0 | Character code | Write only |

**ORDON** ($0F9A_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| "0" | "0" | "0" | FORS | "1" | RGWR | EWDW | EDISP |

(Initial value ***0 0000)

| | | | |
|---|---|---|---|
| FORS | fosc frequency select | 0 : Normal frequency mode<br>1 : Double frequency mode | Write only |
| RGWR | Written data transfer request register | 0 : (Initial state)<br>1 : Transfer written data to OSD circuit. (After transfer, RGWR is cleared to "0".) | |
| EWDW | Window function enable specification register | 0 : Disable window function<br>1 : Enable window function | |
| EDISP | Display on/off specification register | 0 : Disable Display<br>1 : Enable Display | |

*Note1 : * ; don't care*

*Note2 : The data written to OSD control registers except the register followed table is transmitted to OSD circuit by setting RGWR (bit2 in ORDON) to "1". RGWR is cleared to "0" automatically after the transfer is completed.*

| | |
|---|---|
| P67DS, P66DS, P65DS, P64DS | bits 7 to 4 in ORP6DS |
| YBLCS, BKMF, RBMF, ESMZ, VDSMD, MBK, DMA8 | bits 7 to 3, 1 to 0 in ORETC |
| DMA7, DMA6, DMA5, DMA4, DMA3, DMA2, DMA1, DMA0 | bits 7 to 0 in ORDMA |
| RBF, RVF, BLF, RDT, GDT, BDT | bits 5 to 0 in ORDSN |
| CRA7, CRA6, CRA5, CRA4, CRA3, CRA2, CRA1, CRA0 | bits 7 to 0 in ORCRA |
| FORS, RGWR | bits 4 and 2 in ORDON |

*Note3 : When EWDW is cleared to "0", clear EWDW to "0" independent of the value which this register has been set from detecting the rising edge of $\overline{HD}$ signal by software until the falling edge of $\overline{HD}$ signal.*

*Note4 : Write "1" to bit 3 of ORDON when writing to ORDON.*

*Note5 : Do not clear RGWR register to "0". If RGWR register is cleared to "0", the contents of OSD control registers may be transferred to OSD circuit at unexpected timing.*

**ORDON** ($0F9A_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| | | | FORS | | RGWR | EWDW | EDISP |

(Initial value ***0 *000)

| | | | |
|---|---|---|---|
| FORS | fosc frequency select status | 0 : Normal frequency mode<br>1 : Double frequency mode | Read only |
| RGWR | Written data transfer monitor flag | 0 : Transfer completed<br>1 : During transfer | |
| EWDW | Window function enable specification status | 0 : Disable window function<br>1 : Enable window function | |
| EDISP | Display on/off specification status | 0 : Disable display<br>1 : Enable display | |

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## 2.14 Jitter Elimination Circuit

The 87CH38/K38 have a jitter elimination circuit which can display stably without moving up and down in an interlace TV, even if a vertical synchronous signal input at the on-screen display is disarranged.

### 2.14.1 Configuration

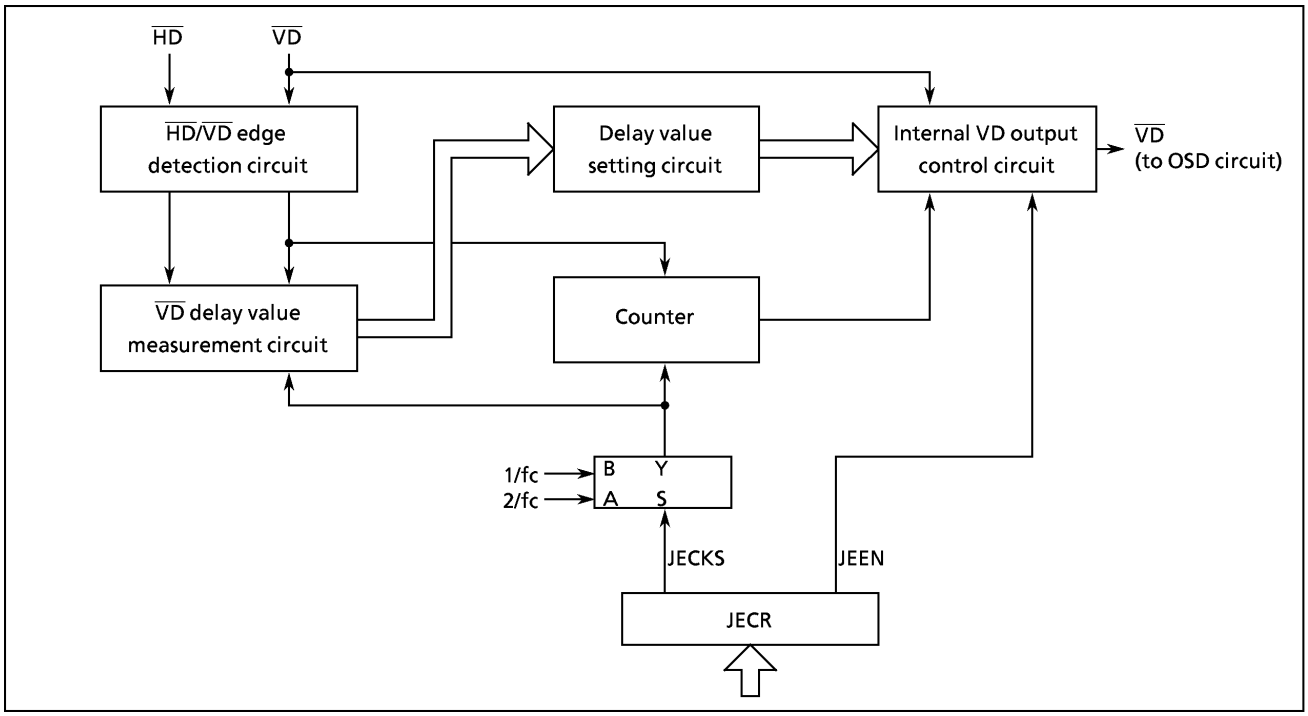


Figure 2-77. Jitter elimination circuit

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## 2.14.2 Control

Jitter elimination circuit is controlled by the jitter elimination control register (JECR).

Jitter elimination control register

JECR ($0FAB_H$)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|---|---|---|---|---|---|---|---|
| | | | | "0" | JEEN | "0" | JECKS |

(initial value: **** *0*0)

| | | | |
|---|---|---|---|
| JEEN | Jitter elimination enable specification | 0 : disable<br>1 : enable | write only |
| JECKS | Source clock select | 0 : 2/fc (Select this, when fc = 8 MHz)<br>1 : 1/fc (Select this, when fc = 4 MHz) | |

*Note 1: When using jitter elimination function, set the JECKS using the CPU clock of 4 MHz or 8 MHz. Using different frequency clock may cause malfunctions.*

*Note 2: The bit 1 and bit 3 of JECR are always cleared to "0".*

*Note 3: In a non-interlace TV, the jitter elimination circuit must be disabled. If not disabled, OSD display is disarranged.*

Figure 2-77. Jitter elimination control register

## 2.14.3 Functions

Jitter elimination circuit detects the different phase between the falling edge of the external $\overline{VD}$ signal and $\overline{HD}$ signal. When $\overline{VD}$ signal is falling within $\overline{HD}$ signal falling +/– 1/4 $\overline{HD}$, the jitter is automatically eliminated and internal $\overline{VD}$ signal is set to the stable location.

This function is enabled by setting JEEN (bit2 in JECR) in the jitter elimination control register to "1". CPU clock must be used at 8 MHz or 4 MHz at using the jitter elimination. JECKS (bit0 in JECR) must be set to "0" at 8 MHz, and "1" at 4 MHz.

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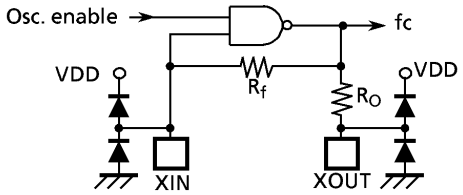
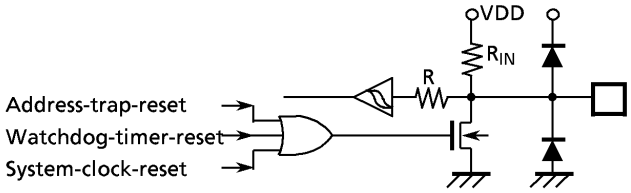
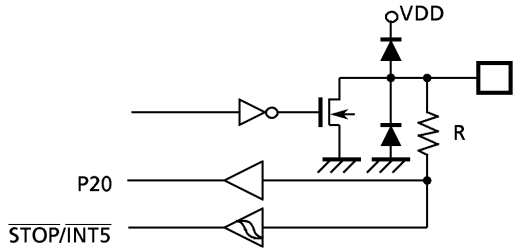
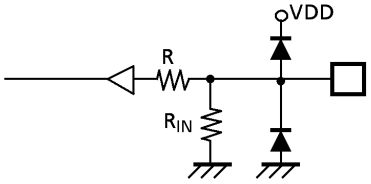
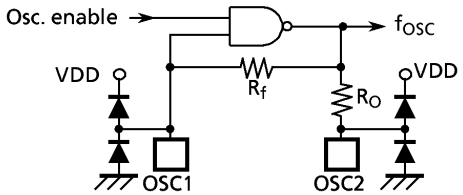
## Input / Output Circuitry

(1) Control pins

The input/output circuitries of the 87CH38/K38 control pins are shown below.

| Control Pin | I/O | Input/Output Circuitry | Remarks |
|---|---|---|---|
| XIN<br>XOUT | Input<br>Output |  | Resonator connecting pins (high-frequency)<br>$R_f$ = 1.2 MΩ (typ.)<br>$R_O$ = 1.5 kΩ (typ.) |
| $\overline{\text{RESET}}$ | I/O |  | Sink open drain output<br>Hysteresis input<br>Pull-up resistor<br>$R_{IN}$ = 220 kΩ (typ.)<br>R = 1 kΩ (typ.) |
| $\overline{\text{STOP}}$ / $\overline{\text{INT5}}$ | Input |  | Hysteresis input<br>R = 1 kΩ (typ.) |
| TEST | Input |  | Pull-down resistor<br>$R_{IN}$ = 70 kΩ (typ.)<br>R = 1 kΩ (typ.) |
| OSC1<br>OSC2 | Input<br>Output |  | Osc. connecting pin for on-screen display<br>$R_f$ = 1.2 MΩ (typ.)<br>$R_O$ = 1.5 kΩ (typ.) |

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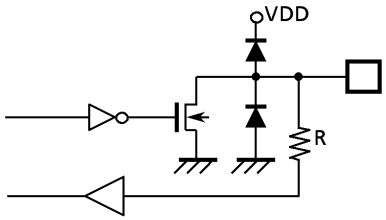
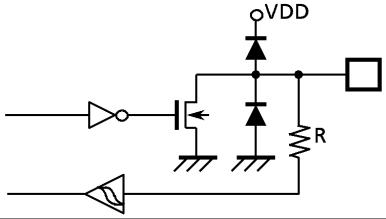
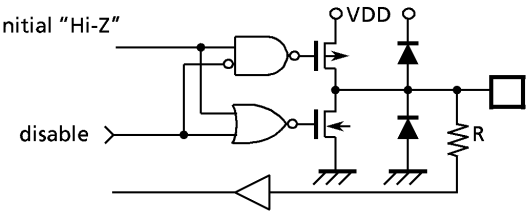
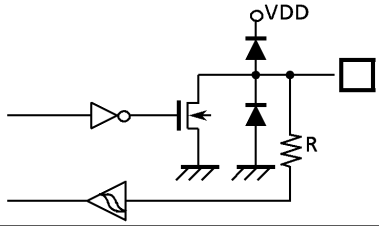
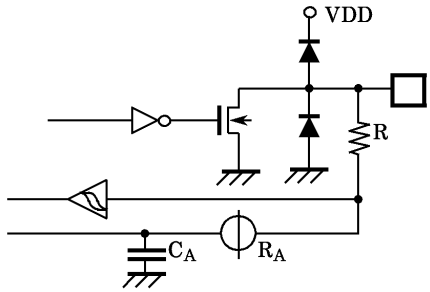
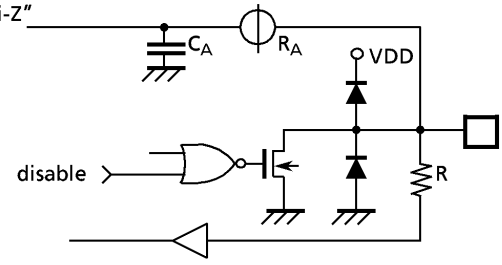
(2) Input / output ports

The input / output circuitries of the 87CH38/K38 I/O ports are shown below.

| Port | I/O | Input / Output Circuitry | Remarks |
|---|---|---|---|
| P20 | I/O | initial "Hi-Z"   | Sink open drain output<br>R = 1 kΩ (typ.) |
| P3 | I/O | initial "Hi-Z"   | Sink open drain output<br>Hysteresis input<br>R = 1 kΩ (typ.) |
| P4<br>P64<br>to<br>P67 | I/O | initial "Hi-Z"   | Tri-state I/O<br>R = 1 kΩ (typ.) |
| P50<br>to<br>P52<br>P57 | I/O | initial "Hi-Z"   | Sink open drain output<br>Hysteresis input<br>R = 1 kΩ (typ.) |
| P53<br>to<br>P56 | I/O | initial "Hi-Z"   | Sink open drain output<br>Hysteresis input<br>R = 1 kΩ (typ.)<br>$R_A$ = 5 kΩ (typ.)<br>$C_A$ = 12 pF (typ.) |
| P60<br>to<br>P61 | I/O | initial "Hi-Z"   | Sink open drain output<br>Higi current output<br>$I_{OL}$ = 20mA (typ.)<br>R = 1 kΩ (typ.)<br>$R_A$ = 5 kΩ (typ.)<br>$C_A$ = 12 pF (typ.) |

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| Port | I/O | Input / Output Circuitry | Remarks |
|---|---|---|---|
| P62 to P63 | I/O | initial "Hi-Z"<br>VDD<br>disable<br>R | Sink open drain output<br>High current output<br>$I_{OL}$ = 20 mA(typ.)<br>R = 1 kΩ (typ.) |
| P70 P71 | I/O | initial "Hi-Z"<br>VDD<br>R | Sink open drain output<br>Hysteresis input<br>R = 1 kΩ (typ.) |

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## Electrical Characteristics

### Absolute maximum ratings

($V_{SS}$ = 0 V)

| Parameter | Symbol | Pins | Ratings | Unit |
|---|---|---|---|---|
| Supply Voltage | $V_{DD}$ | | – 0.3 to 6.5 | V |
| Input Voltage | $V_{IN}$ | | – 0.3 to $V_{DD}$ + 0.3 | V |
| Output Voltage | $V_{OUT1}$ | | – 0.3 to $V_{DD}$ + 0.3 | V |
| Output Current (Per 1 pin) | $I_{OUT1}$ | Ports P2, P3, P4, P5, P64 to P67, P7 | 3.2 | mA |
| | $I_{OUT2}$ | Ports P60 to P63 | 30 | |
| Output Current (Total) | $\Sigma I_{OUT1}$ | Ports P2, P3, P4, P5, P64 to P67, P7 | 120 | mA |
| | $\Sigma I_{OUT2}$ | Ports P60 to P63 | 120 | |
| Power Dissipation [Topr = 70°C] | PD | | 600 | mW |
| Soldering Temperature (time) | Tsld | | 260 (10 s) | °C |
| Storage Temperature | Tstg | | – 55 to 125 | °C |
| Operating Temperature | Topr | | – 30 to 70 | °C |

*Note: The absolute maximum ratings are rated values which must not be exceeded during operation, even for an instant. Any one of the ratings must not be exceeded. If any absolute maximum rating is exceeded, a device may break down or its performance may be degraded, causing it to catch fire or explode resulting in injury to the user. Thus, when designing products which include this device, ensure that no absolute maximum rating value will ever be exceeded.*

### Recommended operating conditions

($V_{SS}$ = 0 V, Topr = – 30 to 70°C)

| Parameter | Symbol | Pins | Conditions | Min | Max | Unit |
|---|---|---|---|---|---|---|
| Supply Voltage | $V_{DD}$ | | NORMAL mode | 4.5 | 5.5 | V |
| | | | IDLE mode | | | |
| | | | STOP mode | 2.0 | | |
| Input High Voltage | $V_{IH1}$ | Except hysteresis input | | $V_{DD} \times 0.70$ | $V_{DD}$ | V |
| | $V_{IH2}$ | Hysteresis input | | $V_{DD} \times 0.75$ | | |
| Input Low Voltage | $V_{IL1}$ | Except hysteresis input | | 0 | $V_{DD} \times 0.30$ | V |
| | $V_{IL2}$ | Hysteresis input | | | $V_{DD} \times 0.25$ | |
| Clock Frequency | fc | XIN, XOUT | | 4.0 | 8.0 | MHz |
| | $f_{OSC}$ | OSC1, OSC2 | Normal frequency mode (FORS = 0, $V_{DD}$ = 4.5 to 5.5 V) | 4.0 | $f_{OSC} \leqq$ fc × 1.2 ≦ 8.0 | |
| | | | Double frequency mode (FORS = 1, $V_{DD}$ = 4.5 to 5.5 V) | 2.0 | $f_{OSC} \leqq$ fc × 0.6 ≦ 4.0 | |

*Note 1: The recommended operating conditions for a device are operating conditions under which it can be guaranteed that the device will operate as specified. If the device is used under operating conditions other than the recommended operating conditions (supply voltage, operating temperature range, specified AC/DC values etc.), malfunction may occur. Thus, when designing products which include this device, ensure that the recommended operating conditions for the device are always adhered to.*

*Note 2: Clock Frequency fc ; The condition of supply voltage range is the value in NORMAL and IDLE modes.*

*Note 3: When using test video signal circuit, high frequency must be 8 MHz.*

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## D.C. characteristics

($V_{SS}$ = 0 V, $T_{opr}$ = – 30 to 70°C)

| Parameter | Symbol | Pins | Conditions | Min | Typ. | Max | Unit |
|---|---|---|---|---|---|---|---|
| Hysteresis Voltage | $V_{HS}$ | Hysteresis inputs | | – | 0.9 | – | V |
| Input Current | $I_{IN1}$ | TEST | $V_{DD}$ = 5.5 V, $V_{IN}$ = 5.5 V / 0 V | – | – | ± 2 | μA |
| | $I_{IN2}$ | Open drain ports | $V_{DD}$ = 5.5 V, $V_{IN}$ = 5.5 V / 0 V | – | – | ± 2 | |
| | $I_{IN3}$ | Tri-state ports | $V_{DD}$ = 5.5 V, $V_{IN}$ = 5.5 V / 0 V | – | – | ± 2 | |
| | $I_{IN4}$ | $\overline{RESET}$, $\overline{STOP}$ | $V_{DD}$ = 5.5 V, $V_{IN}$ = 5.5 V / 0 V | – | – | ± 2 | |
| Input Resistance | $R_{IN2}$ | $\overline{RESET}$ | | 100 | 220 | 450 | kΩ |
| Output Leakage Current | $I_{LO1}$ | Sink open drain ports | $V_{DD}$ = 5.5 V, $V_{OUT}$ = 5.5 V | – | – | 2 | μA |
| | $I_{LO2}$ | Tri-state ports | $V_{DD}$ = 5.5 V, $V_{OUT}$ = 5.5 V / 0 V | – | – | ± 2 | |
| Output High Voltage | $V_{OH2}$ | Tri- state port | $V_{DD}$ = 4.5 V, $I_{OH}$ = – 0.7 mA | 4.1 | – | – | V |
| Output Low Voltage | $V_{OL}$ | Except XOUT, OSC2 and ports P60 to P63 | $V_{DD}$ = 4.5 V, $I_{OL}$ = 1.6 mA | – | – | 0.4 | V |
| Output Low Current | $I_{OL3}$ | Ports P60 to P63 | $V_{DD}$ = 4.5 V, $V_{OL}$ = 1.0 V | – | 20 | – | mA |
| Supply Current in NORMAL mode | | | $V_{DD}$ = 5.5 V<br>fc = 8 MHz<br>$V_{IN}$ = 5.3 V / 0.2 V | – | 14 | 17 | mA |
| Supply Current in IDLE mode | | | | – | 7 | 10 | mA |
| Supply Current in STOP mode | | | $V_{DD}$ = 5.5 V<br>$V_{IN}$ = 5.3 V / 0.2 V | – | 0.5 | 10 | μA |

*Note 1: Typical values show those at $T_{opr}$ = 25°C, $V_{DD}$ = 5 V.*

*Note 2: Input Current $I_{IN1}$, $I_{IN4}$ ; The current through pull-up or pull-down resistor is not included.*

*Note3: Supply Current $I_{DD}$ ; The current (Typ. 0.5 mA) through ladder resistors of ADC is included in NORMAL mode and IDEL mode.*

## A/D conversion characteristics

($V_{SS}$ = 0 V, $V_{DD}$ = 4.5 Vto 5.5V Topr = – 30 to 70°C)

| Parameter | Symbol | Conditions | Min | Typ. | Max | Unit |
|---|---|---|---|---|---|---|
| Analog Reference Voltage | $V_{AREF}$ | supplied from $V_{DD}$ pin. | – | $V_{DD}$ | – | V |
| | $V_{ASS}$ | supplied from $V_{SS}$ pin. | – | 0 | – | |
| Analog Reference Voltage Range | $\triangle V_{AREF}$ | = $V_{DD}$ – $V_{SS}$ | – | $V_{DD}$ | – | |
| Analog Input Voltage | $V_{AIN}$ | | $V_{SS}$ | – | $V_{DD}$ | |
| Nonlinearity Error | | $V_{DD}$ = 4.5 V to 5.5 V | – | – | ± 1 | LSB |
| zero Point Error | | | – | – | ± 2 | |
| Full Scale Error | | | – | – | ± 2 | |
| Total Error | | | – | – | ± 3 | |

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## A.C. characteristics

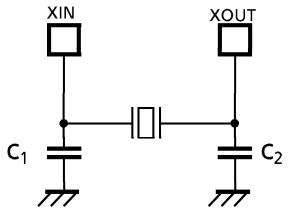
($V_{SS}$ = 0 V, $V_{DD}$ = 4.5 to 5.5 V, $T_{opr}$ = – 30 to 70°C)

| Parameter | Symbol | Conditions | Min | Typ. | Max | Unit |
|---|---|---|---|---|---|---|
| Machine Cycle Time | tcy | In NORMAL mode | 0.5 | – | 1.0 | μs |
| | | In IDLE mode | | | | |
| High-Level Clock Pulse Width | $t_{WCH}$ | For external clock operation (XIN input) , fc = 8 MHz | 62.5 | – | – | ns |
| Low-Level Clock Pulse Width | $t_{WCL}$ | | | | | |

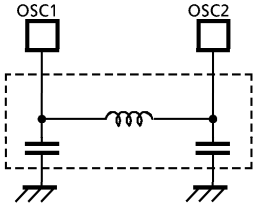
## Recommended oscillating condition

($V_{SS}$ = 0 V, $V_{DD}$ = 4.5 to 5.5 V, $T_{opr}$ = – 30 to 70°C)

| Parameter | Oscillator | Frequency | Recommended Oscillator | Recommended Conditions $C_1$ | Recommended Conditions $C_2$ |
|---|---|---|---|---|---|
| High-frequency Oscillation | Ceramic Resonator | 8 MHz | KYOCERA KBR8.0M | 30 pF | 30 pF |
| | | 4 MHz | KYOCERA KBR4.0MS | | |
| | | | MURATA CSA4.00MG | | |
| | Crystal Oscillator | 8 MHz | TOYOCOM 210B 8.0000 | 20 pF | 20 pF |
| | | 4 MHz | TOYOCOM 204B 4.0000 | | |
| OSD | LC Resonator | 8 MHz | TOKO A285TNIS-11695 (5 mm) | – | – |
| | | 7 MHz | TOKO TBEKSES-30375FBY | | |




(1) High-frequency




(2) LC Resonator for OSD

*Note : On our OSD circuit, the horizontal display start position is determined by counting the clock from LC oscillator. So, the unstable start of oscillation after the rising edge of Horizontal Sync. Signal will be cause the OSD distortion. Generally, smaller C and larger L make clearer wave form at the beginning of oscillation. We recommend that the value of LC oscillator should be equal and bigger than 33μH.*

*Note : To keep reliable operation, shield the device electrically with the metal plate on its package mold surface against the high electric field, for example, by CRT (Cathode Ray Tube) .*

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## Notice when developing a program of TMP87CH38/K38

When developing a program of 87CH38/K38 by using an OTP (87PS38) and an emulator (BM87CS38N0A), it is necessary to take notice as follows for emulating the operation of 87CH38/K38 with them.

### On-Screen-Display Circuit

(1) Setting ORVFn (Vertical Display Start Position) of line 9 to line 12
Set "$FF_H$" into ORVS9 to ORVF12. If these registers are set other value or have an initial value, cannot emulate the operation of 87CH38/K38 with an OTP and an emulator.

| | | | | | | | | | OTP/Emulator Initial value |
|---|---|---|---|---|---|---|---|---|---|
| **ORVS9** ($0F89_H$) | VS97 | VS96 | VS95 | VS94 | VS93 | VS92 | VS91 | VS90 | (0000 0000) |
| **ORVS12** ($0F8C_H$) | VS127 | VS126 | VS125 | VS124 | VS123 | VS122 | VS121 | VS120 | (0000 0000) |

87CH38/K38 does not have ORVF9 to ORVF12. But the operation of OSD interrupt is similar with setting "$FF_H$" into ORVF9 to ORVF12.

(2) Setting CSn (Character size and display on/off) of line 9 to line 12
Set "$00_H$" (Initial value) into ORCS12. If these registers are set other value, cannot emulate the operation of 87CH38/K38 with an OTP and an emulator.

| | | | | | OTP/Emulator Initial value |
|---|---|---|---|---|---|
| **ORCS12** ($0F8F_H$) | CS12 | CS11 | CS10 | CS9 | (0000 0000) |

(3) Setting EFRn (Fringing Specification) of line 9 to line 12
Set "0" (Initial value) into ORP6DS (EFR12 to EFR9). If these registers are set other value, cannot emulate the operation of 87CH38/K38 with an OTP and an emulator.

| | | | | | ↓ | ↓ | ↓ | ↓ | OTP/Emulator Initial value |
|---|---|---|---|---|---|---|---|---|---|
| **ORP6DS** ($0F91_H$) | (P67DS) | (P66DS) | (P65DS) | (P64DS) | EFR12 | EFR11 | EFR10 | EFR9 | (0000 0000) |

### ROM Correction Circuit

RAM area which is used for ROM correction circuit in 87CH38/K38 can use address from 0140H to 023FH, but RAM area which is used for ROM correction circuit in OTP (87PS38) can use address from 0240H to 083FH. Therefore, when using ROM correction circuit in 87CH38/K38, load address for patch program codes and jump vector must be changed after debugging a program by OTP.

*Note : Development tool does not have a ROM correction circuit.*

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