ANALOG DEVICES

Low Cost, Voltage Output, High-Side, Current-Sense Amplifier
ADM4073

FEATURES

Low cost, compact, current-sense solution
Three available gain versions
  20 V/V (ADM4073T)
  50 V/V (ADM4073F)
  100 V/V (ADM4073H)
Typical ±1.0% full-scale accuracy
500 µA supply current
Wide 1.8 MHz bandwidth
3 V to 28 V operating supply
Wide 2 V to 28 V common-mode range
  Independent of supply voltage
Operates from −40°C to +125°C
Available in a 6-lead SOT-23 package
Pin-to-pin compatibility with the MAX4073

APPLICATIONS

Cell phones, PDAs
Notebook computers
Portable/battery-powered systems
Smart battery packs/chargers
Automotive
Power management systems
PA bias control
General system/board-level current monitoring
Precision current sources

FUNCTIONAL BLOCK DIAGRAM

Figure 1.

APPLICATION DIAGRAM

Figure 2.

GENERAL DESCRIPTION

The ADM4073 is a low cost, high-side, current-sense amplifier ideal for small portable applications such as cell phones, notebook computers, PDAs, and other systems where current monitoring is required. The device is available in three different gain models, eliminating the need for gain-setting resistors. Because the ground path is not interrupted, the ADM4073 is particularly useful in rechargeable battery-powered systems, while its wide 1.8 MHz bandwidth makes it suitable for use inside battery-charger control loops. The input common-mode range of 2 V to 28 V is independent of the supply voltage.
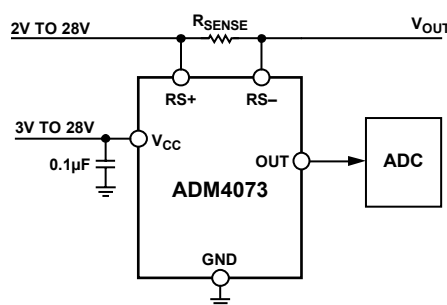
The voltage on the output pin is determined by the current flowing through the selectable external sense resistor and the gain of the version selected. The operating range is 3 V to 28 V with a typical supply current of 500 µA. The ADM4073 is available in a 6-lead SOT-23 package and is specified over the automotive operating temperature range (−40°C to +125°C).

ADM4073

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REVISION HISTORY

7/06—Revision 0: Initial Version

SPECIFICATIONS

V_{RS+} = 2 V to 28 V, V_{SENSE} = ($V_{RS+} - V_{RS-}$) = 0, V_{CC} = 3 V to 28 V, T_A = −40°C to +125°C, unless otherwise noted. Typical values are at T_A = 25°C.[1]

Table 1.

Parameter	Min	Typ	Max	Unit	Conditions
POWER SUPPLY					
Operating Voltage Range, V_{CC}	3		28	V	Inferred from PSRR test
Common-Mode Input Range, V_{CMR}	2		28	V	Inferred OUT voltage error test
Common-Mode Input Rejection, CMR		90		dB	V_{SENSE} = 100 mV, V_{CC} = 12 V
Supply Current, I_{CC}		0.5	1.2	mA	V_{CC} = 28 V
Leakage Current, I_{RS+}/I_{RS-}		0.05	2	µA	V_{CC} = 0 V, V_{RS+} = 28 V, T_A = 85°C
Input Bias Current, I_{RS+}		20	60	µA	
Input Bias Current, I_{RS-}		40	120	µA	
Full Scale Sense Voltage, V_{SENSE}		150		mV	V_{SENSE} = ($V_{RS+} - V_{RS-}$)
Total Out Voltage Error[2]		±1		%	V_{SENSE} = 100 mV, V_{CC} = 12 V, V_{RS+} = 2 V
		±1.0	±5.0	%	V_{SENSE} = 100 mV, V_{CC} = 12 V, V_{RS+} = 12 V, T_A = +25°C
			±5.0	%	V_{SENSE} = 100 mV, V_{CC} = 12 V, V_{RS} = 12V, T_A = −40°C to +125°C
		±1.0	±5.0	%	V_{SENSE} = 100 mV, V_{CC} = 28 V, V_{RS} = 28 V, T_A = +25°C
			±5.0	%	V_{SENSE} = 100 mV, V_{CC} = 28 V, V_{RS} = 28 V, T_A = −40°C to +125°C
		±7.5		%	V_{SENSE} = 6.25 mV,[3] V_{CC} = 12 V, V_{RS} = 12 V
Extrapolated Input Offset Voltage, V_{OS}		1.0		mV	V_{CC} = V_{RS+} = 12 V, V_{SENSE} > 10 mV
Out High Voltage ($V_{CC} - V_{OH}$)		0.8	1.2	V	V_{CC} = 3 V, V_{SENSE} = 150 mV (ADM4073T)
		0.8	1.2	V	V_{CC} = 7.5 V, V_{SENSE} = 150 mV (ADM4073F)
		0.8	1.2	V	V_{CC} = 15 V, V_{SENSE} = 150 mV (ADM4073H), T_A = 25°C
DYNAMIC CHARACTERISTICS					
Bandwidth, BW		1.8		MHz	V_{SENSE} = 100 mV, V_{CC} = 12 V, V_{RS+} = 12 V, C_{LOAD} = 5 pF (ADM4073T)
		1.7		MHz	V_{SENSE} = 100 mV, V_{CC} = 12 V, V_{RS+} = 12 V, C_{LOAD} = 5 pF (ADM4073F)
		1.6		MHz	V_{SENSE} = 100 mV, V_{CC} = 12 V, V_{RS+} = 12 V, C_{LOAD} = 5 pF (ADM4073H)
		600		kHz	V_{SENSE} = 6.25 mV,[3] V_{CC} = 12 V, V_{RS+} = 12 V, C_{LOAD} = 5 pF (ADM4073T/F/H)
Gain, A_V		20		V/V	ADM4073T
		50		V/V	ADM4073F
		100		V/V	ADM4073H
Gain Accuracy, dA_V		±1.0	±2.0	%	V_{SENSE} = 10 mV to 150 mV, V_{CC} = 12 V, V_{RS+} = 12 V, T_A = +25°C (ADM4073T/F)
			±2.0	%	V_{SENSE} = 10 mV to 150 mV, V_{CC} = 12 V, V_{RS+} = 12 V, T_A = −40°C to +125°C (ADM4073T/F)
		±1.0	±1.5	%	V_{SENSE} = 10 mV to 100 mV, V_{CC} = 12 V, V_{RS+} = 12 V, T_A = +25°C (ADM4073H)
			±3.0	%	V_{SENSE} = 10 mV to 100 mV, V_{CC} = 12 V, V_{RS+} = 12 V, T_A = −40°C to +125°C (ADM4073H)
OUT Settling Time to 1% of Final Value		400		ns	V_{SENSE} = 6.25 mV to 100 mV, V_{CC} = 12 V, V_{RS+} = 12 V, C_{LOAD} = 5 pF
		800		ns	V_{SENSE} = 100 mV to 6.25 mV, V_{CC} = 12 V, V_{RS+} = 12 V, C_{LOAD} = 5 pF
Output Resistance, R_{OUT}		12		kΩ	
Power Supply Rejection Ratio, PSRR		78		dB	V_{SENSE} = 60 mV, V_{CC} = 3 V to 28 V (ADM4073T)
		85		dB	V_{SENSE} = 24 mV, V_{CC} = 3 V to 28 V (ADM4073F)
		90		dB	V_{SENSE} = 12 mV, V_{CC} = 3 V to 28 V (ADM4073H)
Power-Up Time[4]		5		µs	C_{LOAD} = 5 pF, V_{SENSE} = 100 mV
Saturation Recovery Time[5]		5		µs	C_{LOAD} = 5 pF, V_{CC} = 12 V, V_{RS+} = 12 V

[1] 100% production tested at T_A = 25°C. Specifications over temperature limit are guaranteed by design.
[2] Total out voltage error is the sum of the gain and offset errors.
[3] 6.25 mV = 1/16th of 100 mV full-scale sense voltage.
[4] Output settles to within 1% of final value.
[5] The device does not experience phase reversal when overdriven.

ADM4073

ABSOLUTE MAXIMUM RATINGS

Table 2.

Parameter	Rating
V_{CC} to GND	−0.3 V to +30 V
RS± to GND	−0.3 V to +30 V
OUT to GND	−0.3 V to (V_{CC} + 0.3 V)
OUT Short-Circuit to GND	Continuous
Differential Input Voltage (V_{RS+} − $V_{RS−}$)	±5 V
Current into any Pin	±20 mA
Storage Temperature Range	−65°C to +125°C
Operating Temperature Range	−40°C to +125°C
Lead Temperature, Soldering (10 sec)	300°C
Junction Temperature	150°C

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

THERMAL CHARACTERISTICS

θ_{JA} is specified for the worst-case conditions, that is, a device soldered in a circuit board for surface-mount packages.

Table 3. Thermal Resistance

Package Type	θ_{JA}	Unit
6-Lead SOT-23	169.5	°C/W

ESD CAUTION

ESD (electrostatic discharge) sensitive device. Electrostatic charges as high as 4000 V readily accumulate on the human body and test equipment and can discharge without detection. Although this product features proprietary ESD protection circuitry, permanent damage may occur on devices subjected to high energy electrostatic discharges. Therefore, proper ESD precautions are recommended to avoid performance degradation or loss of functionality.

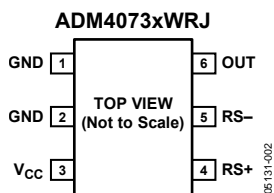# PIN CONFIGURATION AND FUNCTION DESCRIPTIONS

ADM4073xWRJ

GND	1		6	OUT
GND	2	TOP VIEW (Not to Scale)	5	RS–
V_{CC}	3		4	RS+

05131-002

Figure 3. Pin Configuration

Table 4. Pin Function Descriptions

Pin No.	Mnemonic	Description
1, 2	GND	Chip Ground Pin.
3	V_{CC}	Chip Power Supply. Requires a 0.1µF capacitor to ground.
4	RS+	Power-Side Connection to the External Sense Resistor.
5	RS–	Load-Side Connection to the External Sense Resistor.
6	OUT	Voltage Output. V_{OUT} is proportional to V_{SENSE}. Output impedence is approximately 12 kΩ.

TYPICAL PERFORMANCE CHARACTERISTICS
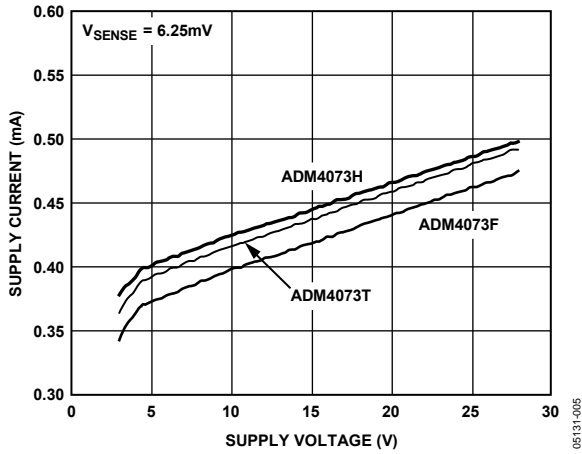
Figure 4. Supply Current vs. Supply Voltage (V_{SENSE} = 6.25mV)
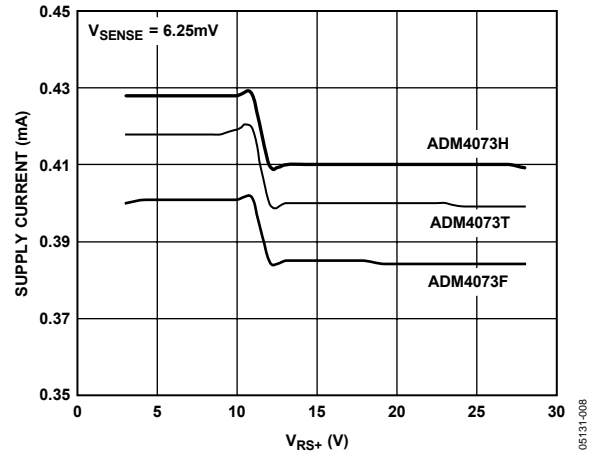
Figure 7. Supply Current vs. RS+ Voltage (V_{SENSE} = 6.25 mV)
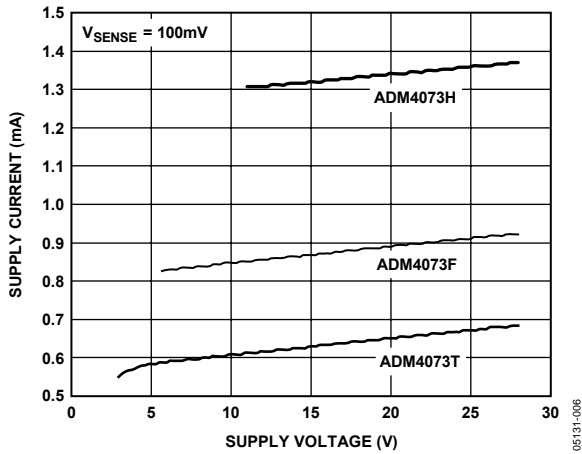
Figure 5. Supply Current vs. Supply Voltage (V_{SENSE} = 100 mV)
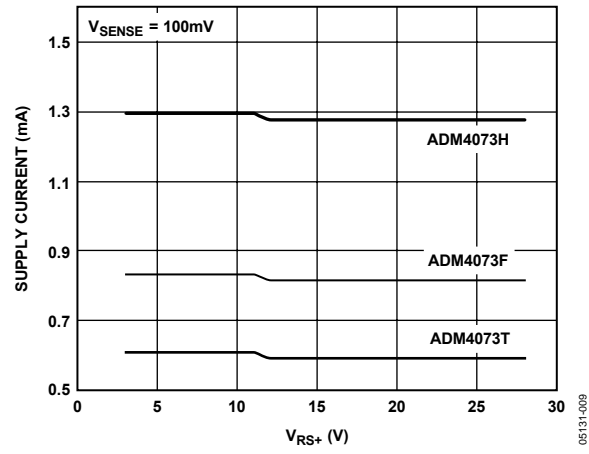
Figure 8. Supply Current vs. RS+ Voltage (V_{SENSE} = 100 mV)
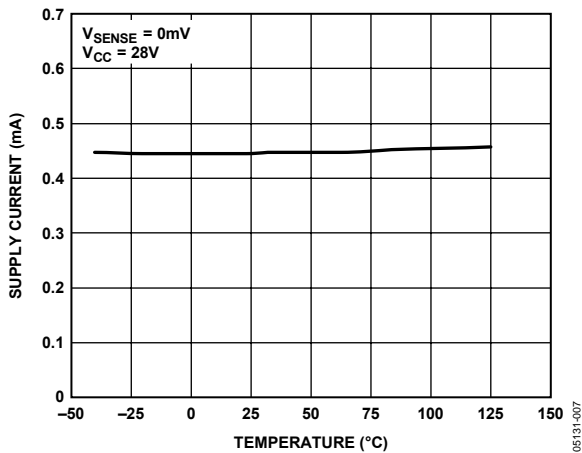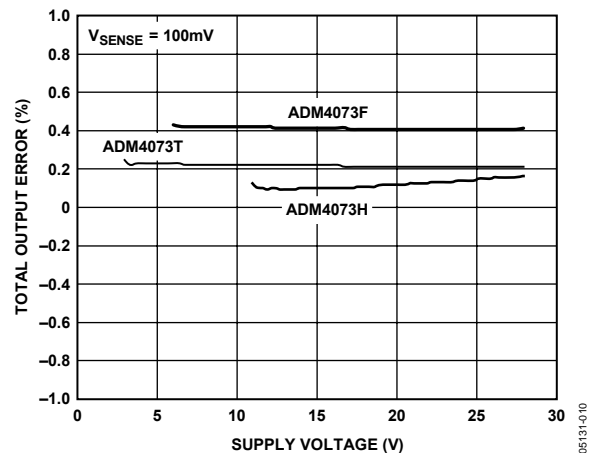

Figure 6. Supply Current vs. Temperature

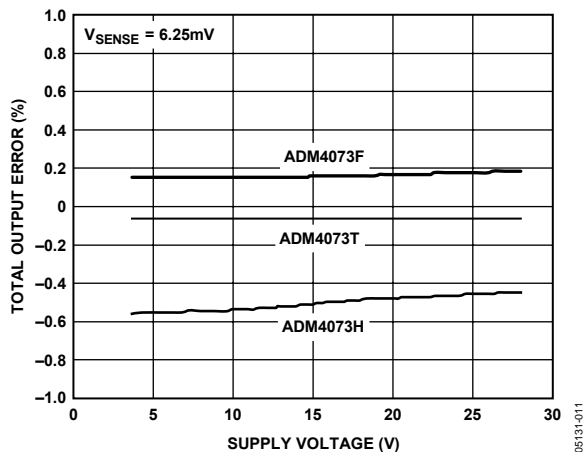Figure 9. Total Output Error vs. Supply Voltage (V_{SENSE} = 100 mV)

Figure 10. Total Output Error vs. Supply Voltage (V$_{SENSE}$ = 6.25 mV)

Figure 11. Total Output Error vs. Common-Mode Voltage

Figure 12. Total Output Error vs. Temperature

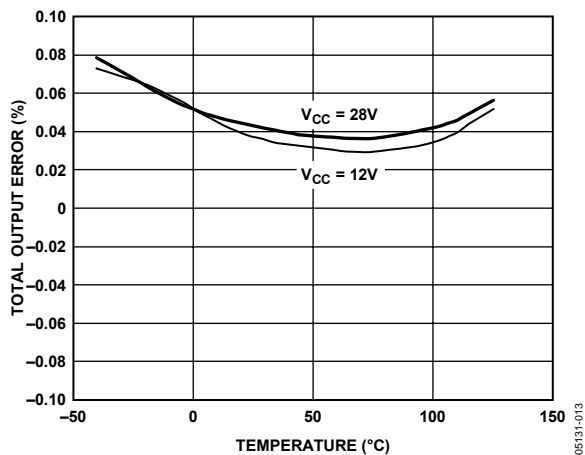*Figure 13. Gain Accuracy vs. Temperature*

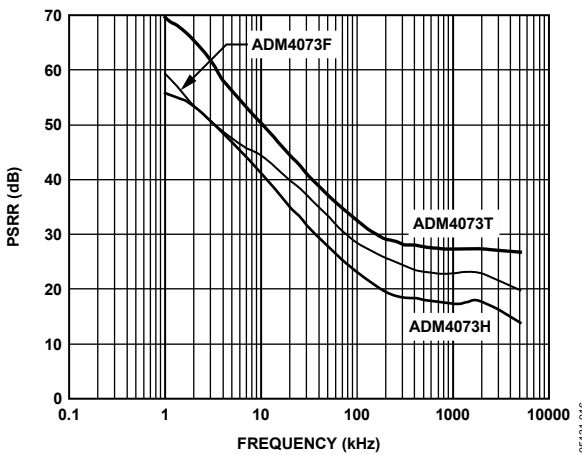*Figure 14. O/P High Voltage (V$_{CC}$ − V$_{OH}$) vs. Temperature*

Figure 15. PSRR vs. Frequency
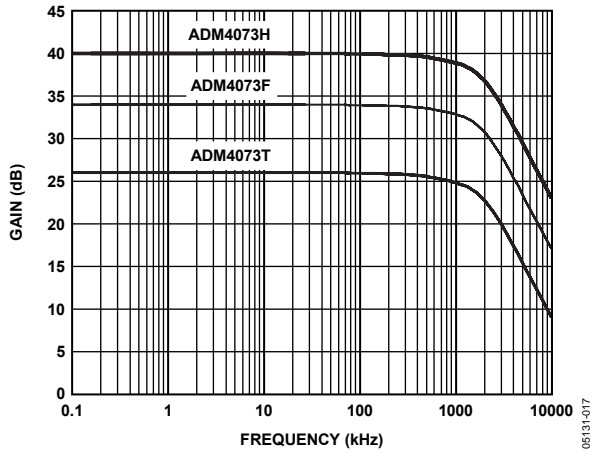
ADM4073

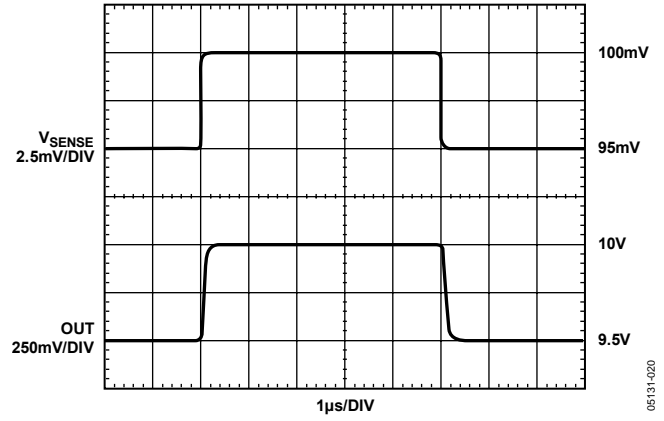Figure 16. Small Signal Gain vs. Frequency

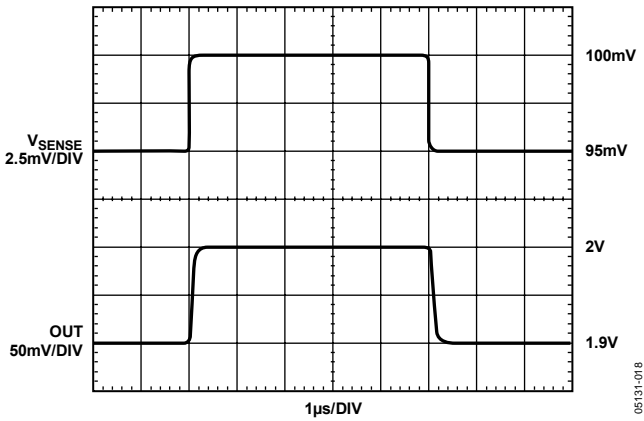Figure 19. ADM4073H Small Signal Transient Response

Figure 17. ADM4073T Small Signal Transient Response
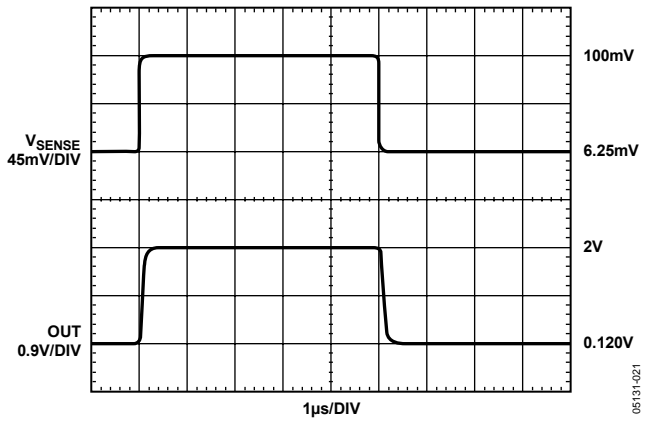

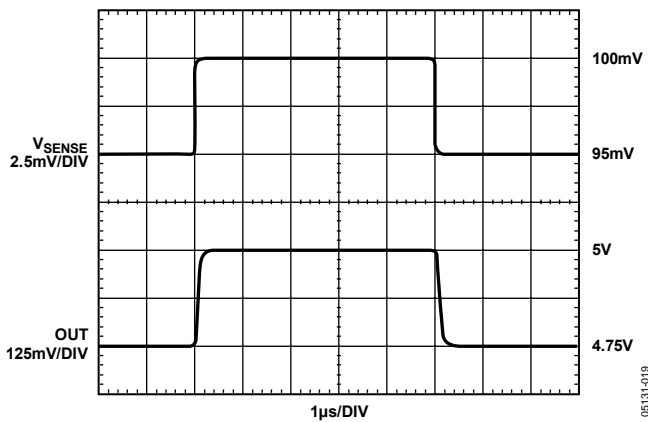Figure 20. ADM4073T Large Signal Transient Response

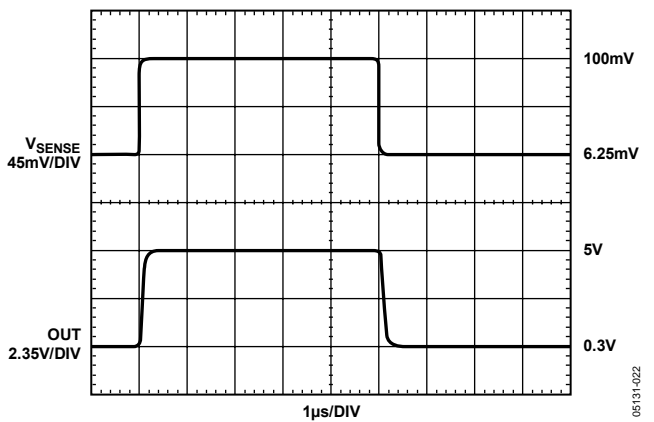Figure 18. ADM4073F Small Signal Transient Response
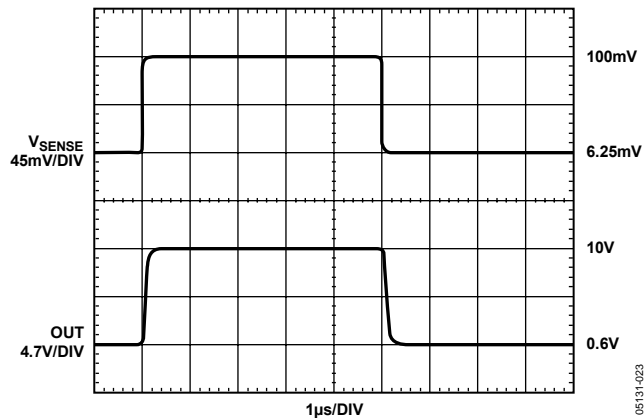

Figure 21. ADM4073F Large Signal Transient Response

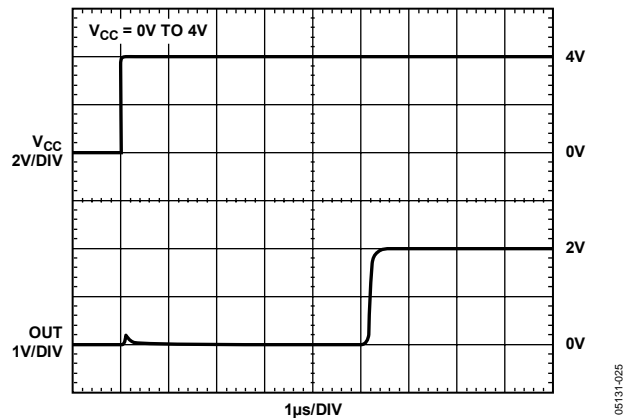Figure 22. ADM4073H Large Signal Transient Response

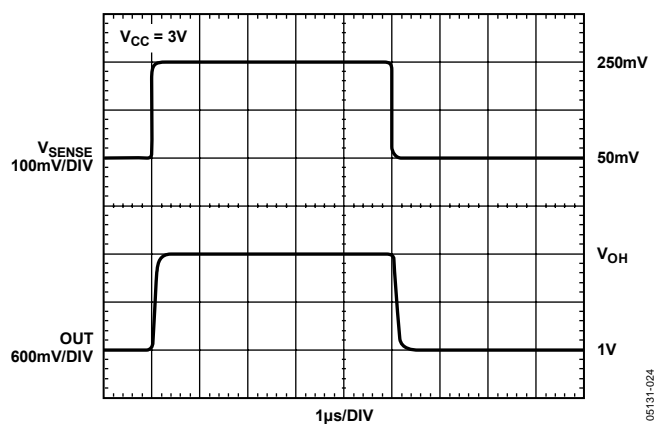Figure 24. ADM4073T Start-Up Delay

Figure 23. ADM4073T Overdrive Response

THEORY OF OPERATION

The current from the source flows through R_{SENSE}, which generates a voltage drop, V_{SENSE}, across the RS+ and RS– terminals of the sense amplifier. The Input Stage Amplifier A1 regulates its inputs to be equal, thereby shunting a current proportional to V_{SENSE}/R_{G1} to the output current mirror. This current is then multiplied by a gain factor b in the output stage current mirror and flows through R_{GD} to generate V_{OUT}. Therefore, V_{OUT} is related to V_{SENSE} by the ratio of R_{G1} to R_{GD} and the current gain b.

$$V_{OUT} = A_V \times V_{SENSE}$$

where:

$$A_V = R_{GD}/R_{G1} \times b$$

where:

A_V = 20 V/V (for ADM4073T).
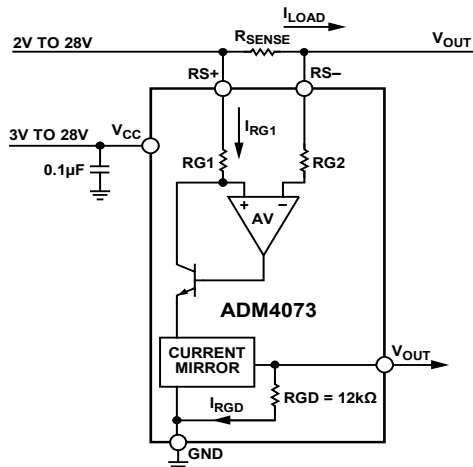A_V = 50 V/V (for ADM4073H).
A_V = 100 V/V (for ADM4073F).

Figure 25. Functional Block Diagram

R_SENSE

The ADM4073 has the ability to sense a wide variety of currents by selecting a particular sense resistor. Select a suitable output voltage for full-scale current, for example, 10 V for 10 A. Then select a gain model that gives the most efficient use of the sense voltage range (150 mV max).

In the example above, using the ADM4073H (gain of 100) give an output voltage of 10 V when the sense voltage is 100 mV. Use the following equation to determine what value of sense resistor gives 100 mV with 10 A flowing through it:

$$R_{SENSE} = 100 \text{ mV}/10 \text{ A}$$

$$R_{SENSE} = 10 \text{ m}\Omega$$

$$V_{OUT} = (I_{LOAD} \times R_{SENSE}) \times A_V$$

To measure lower currents accurately, as large a sense resistor as possible should be used to utilize the higher end of the sense voltage range. This reduces the effects of the offset voltage errors in the internal amplifier.

When currents are very large, it is important to take the I^2R power losses across the sense resistor into account. If the sense resistor's rated power dissipation is not sufficient, its value can drift, giving an inaccurate output voltage, or it could fail altogether. This, in turn, causes the voltage across the RS+ and RS– pins to exceed the absolute maximum ratings.

If the monitored supply rail has a large amplitude high frequency component, a sense resistor with low inductance should be chosen.
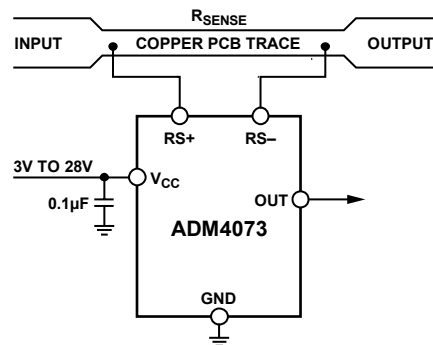

Figure 26. Using PCB Trace for Current Sensing

OUTPUT (OUT)

The output stage of the ADM4073 is a current source driving a pull-down resistance. To ensure optimum accuracy, care must be taken not to load this output externally. To minimize output errors, ensure OUT is connected to a high impedance input stage. If this is not possible, output buffering is recommended.

The percent error introduced by output loading is determined with the following formula:
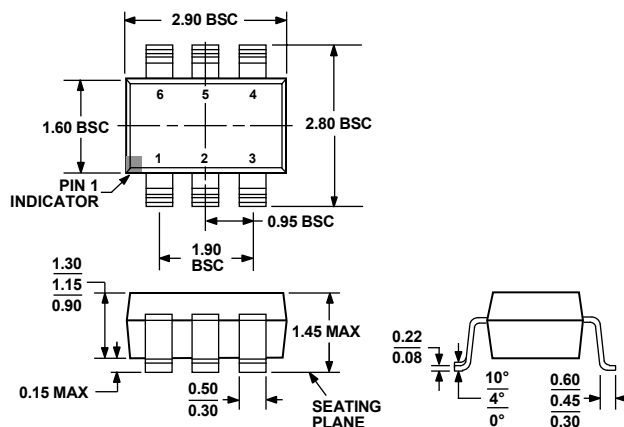
$$\% \, Error = 100 \left(\left(R_{LOAD} / R_{OUT\_INT} + R_{SENSE} \right) - 1 \right)$$

where:

R_{LOAD} is the external load applied to OUT.
R_{OUT_INT} is the internal output resistance (12 kΩ).

OUTLINE DIMENSIONS

COMPLIANT TO JEDEC STANDARDS MO-178-AB

Figure 27. 6-Lead Small Outline Transistor Package [SOT-23]
(RJ-6)
Dimensions shown in millimeters

ORDERING GUIDE

Model	Gain	Temperature Range	Package Description	Package Option	Branding
ADM4073TWRJZ-REEL7[1]	20	−40°C to +125°C	6-Lead SOT-23	RJ-6	M2E
ADM4073FWRJZ-REEL7[1]	50	−40°C to +125°C	6-Lead SOT-23	RJ-6	M2C
ADM4073HWRJZ-REEL7[1]	100	−40°C to +125°C	6-Lead SOT-23	RJ-6	M2D

[1] Z = Pb-free part.

NOTES

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