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AK4645A

Stereo CODEC with MIC/HP-AMP

GENERAL DESCRIPTION

The AK4645A is a stereo CODEC with a built-in Microphone-Amplifier and Headphone-Amplifier. The AK4645A features analog mixing circuit that allows easy interfacing in mobile phone and portable A/V player designs. The AK4645A is available in a 32pin QFN (4mm x 4mm), utilizing less board space than competitive offerings.

FEATURES

1. Recording Function
 - 4 Stereo Input Selector
 - Stereo Mic Input (Full-differential or Single-ended)
 - Stereo Line Input
 - MIC Amplifier (+32dB/+26dB/+20dB or 0dB)
 - Digital ALC (Automatic Level Control)
(+36dB ~ -54dB, 0.375dB Step, Mute)
 - ADC Performance: S/(N+D): 83dB, DR, S/N: 86dB (MIC-Amp=+20dB)
S/(N+D): 88dB, DR, S/N: 95dB (MIC-Amp=0dB)
 - Wind-noise Reduction Filter
 - Stereo Separation Emphasis
 - Programmable EQ
2. Playback Function
 - Digital De-emphasis Filter (tc=50/15(s, fs=32kHz, 44.1kHz, 48kHz)
 - Bass Boost
 - Soft Mute
 - Digital Volume (+12dB (115.0dB, 0.5dB Step, Mute)
 - Digital ALC (Automatic Level Control)
(+36dB ~ -54dB, 0.375dB Step, Mute)
 - Stereo Separation Emphasis
 - Programmable EQ
 - Stereo Line Output
 - Performance: S/(N+D): 88dB, S/N: 92dB
 - Stereo Headphone-Amp
 - S/(N+D): 65dB@7.5mW, S/N: 90dB
 - Output Power: 70mW@16Ω (HVDD=5V), 62mW@16Ω (HVDD=3.3V)
 - Pop Noise Free at Power ON/OFF
 - Analog Mixing: 4 Stereo Input
3. Power Management
4. Master Clock:
 - External Clock Mode
 - Frequencies: 256fs, 384fs, 512fs or 1024fs (MCKI pin)
5. Sampling Rate:
 - EXT Master/Slave Mode:
 - 7.35kHz ~ 48kHz (256fs), 7.35kHz ~ 48kHz (384fs),
7.35kHz ~ 26kHz (512fs), 7.35kHz ~ 13kHz (1024fs)
6. μ P I/F: 3-wire serial, I²C Bus (Ver 1.0, 400kHz Fast-mode)
7. Master/Slave mode
8. Audio Interface Format: MSB First, 2's complement
 - ADC: 16bit MSB justified, I²S, DSP Mode
 - DAC: 16bit MSB justified, 16bit LSB justified, 16-24bit I²S, DSP Mode

9. $T_a = -30 \sim 85^\circ\text{C}$
10. Power Supply:
 - AVDD, DVDD: 2.6 ~ 3.5V (typ. 3.3V)
 - HVDD: 2.6 ~ 5.25V (typ. 3.3V/5.0V)
 - TVDD (Digital I/O): 1.6 ~ 3.5V (typ. 3.3V)
11. Package: 32pin QFN (4mm x 4mm, 0.4mm pitch)
12. Register Compatible with AK4644

■ Block Diagram

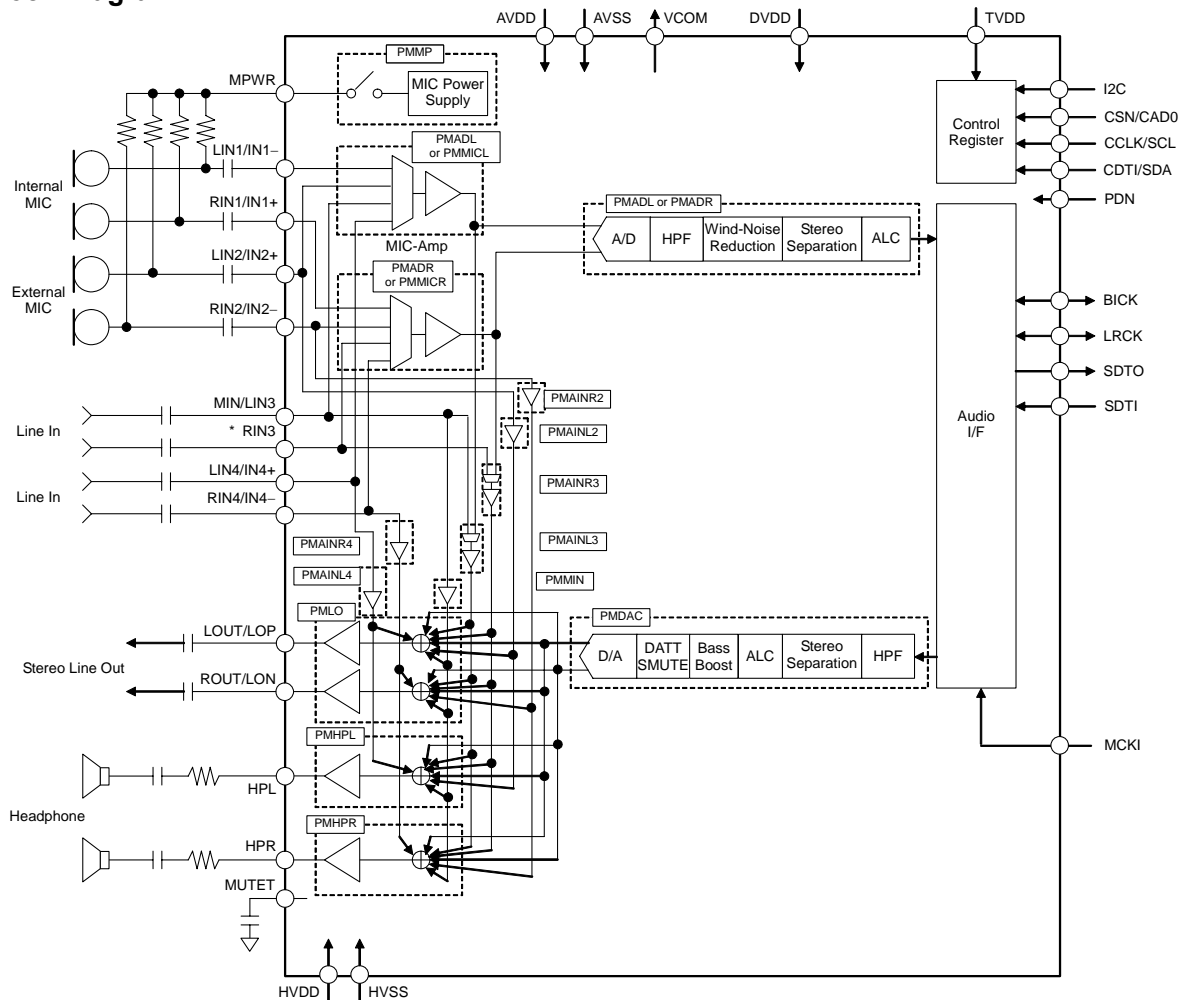


Figure 1. Block Diagram

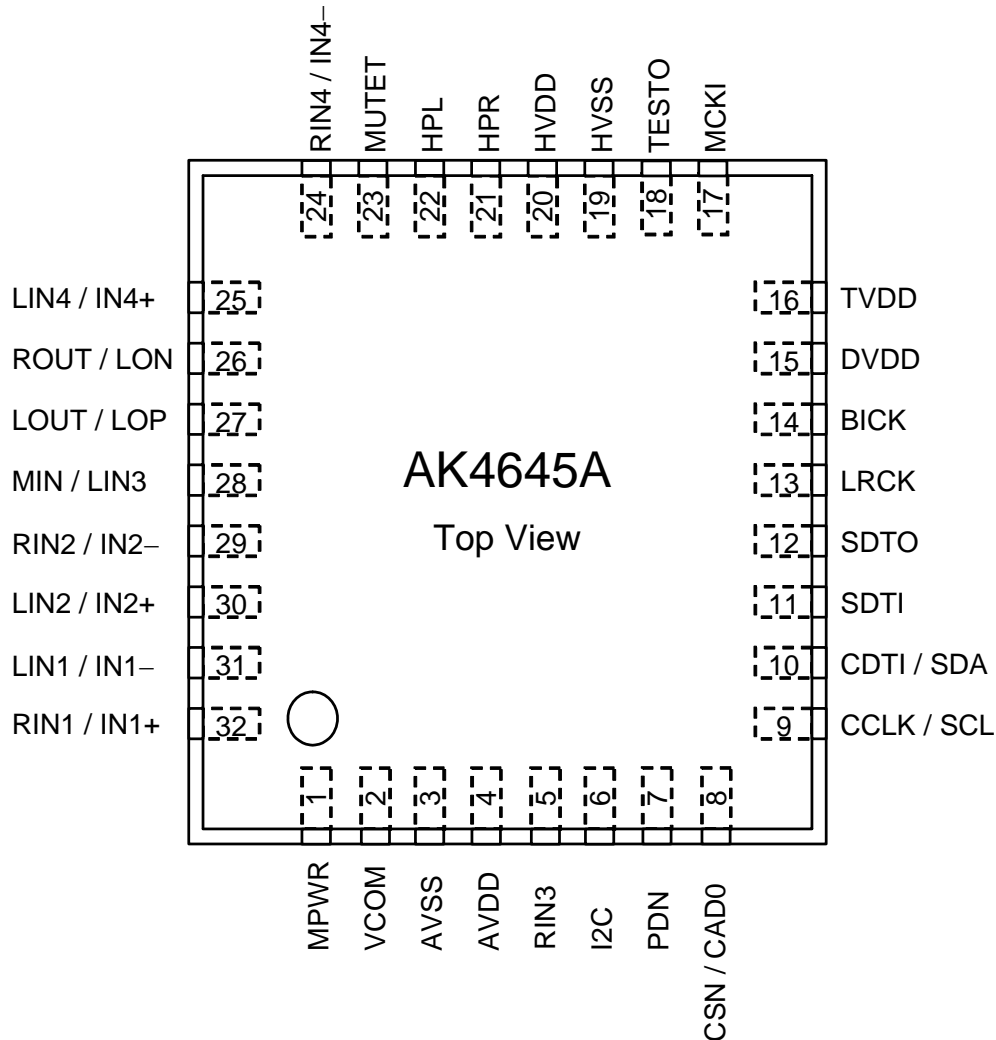
■ Ordering Guide

 AK4645AEZ
 AKD4645A

 -30 ~ +85°C
 Evaluation board for AK4645A

32pin QFN (0.4mm pitch)

■ Pin Layout



■ Compatibility with AK4643/44

1. Function

Function	AK4643	AK4644	AK4645A
Digital I/O of μ P I/F	2.6 to 3.6V	←	1.6 to 3.5V
Analog Mixing for Playback	3 Stereo	←	4 Stereo
Input Selector for Recording	3 Stereo	←	4 Stereo
HP-Amp Hi-Z Setting for wired OR	No	←	Yes
PLL	11.2896/12/12.288/ 13.5/24/27MHz	←	No
Speaker-Amp	Yes	No	←
Receiver-Amp	Yes	←	No
Package	32QFN (5mm x 5mm, 0.5mm pitch)	←	32QFN (4mm x 4mm, 0.4mm pitch)

2. Pin

Pin#	AK4643	AK4644	AK4645A
16	DVSS	←	TVDD
18	MCKO	MCKO	TESTO
19	SPN	TEST1	HVSS
20	SPP	TEST2	HVDD
21	HVDD	←	HPR
22	HVSS	←	HPL
23	HPR	←	MUTET
24	HPL	←	RIN4 / IN4-
25	MUTET	←	LIN4 / IN4+
26	ROUT/RCN	←	ROUT/LON
27	LOUT/RCP	←	LOUT/LOP

3. Register (difference from AK4644)

Addr	Register Name	D7	D6	D5	D4	D3	D2	D1	D0
00H	Power Management 1	0	PMVCM	PMMIN	0	PMLO	PMDAC	0	PMADL
01H	Power Management 2	HPZ	HPMTN	PMHPL	PMHPR	M/S	0	0	0
02H	Signal Select 1	0	0	0	DACL	0	PMMP	0	MGAIN0
03H	Signal Select 2	LOVL	LOPS	MGAIN1	0	0	MINL	0	0
04H	Mode Control 1	0	0	0	0	BCKO	0	DIF1	DIF0
05H	Mode Control 2	0	0	FS3	MSBS	BCKP	FS2	FS1	FS0
06H	Timer Select	DVTM	WTM2	ZTM1	ZTM0	WTM1	WTM0	RFST1	RFST0
07H	ALC Mode Control 1	0	0	ALC	ZELMN	LMAT1	LMAT0	RGAIN0	LMTH0
08H	ALC Mode Control 2	REF7	REF6	REF5	REF4	REF3	REF2	REF1	REF0
09H	Lch Input Volume Control	IVL7	IVL6	IVL5	IVL4	IVL3	IVL2	IVL1	IVL0
0AH	Lch Digital Volume Control	DVL7	DVL6	DVL5	DVL4	DVL3	DVL2	DVL1	DVL0
0BH	ALC Mode Control 3	RGAIN1	LMTH1	0	0	0	0	VBAT	0
0CH	Rch Input Volume Control	IVR7	IVR6	IVR5	IVR4	IVR3	IVR2	IVR1	IVR0
0DH	Rch Digital Volume Control	DVR7	DVR6	DVR5	DVR4	DVR3	DVR2	DVR1	DVR0
0EH	Mode Control 3	0	LOOP	SMUTE	DVOLC	BST1	BST0	DEM1	DEM0
0FH	Mode Control 4	0	0	0	0	IVOLC	HPM	MINH	DACH
10H	Power Management 3	INR1	INL1	HPG	MDIF2	MDIF1	INR0	INL0	PMADR
11H	Digital Filter Select	GN1	GN0	0	FIL1	EQ	FIL3	0	0
12H	FIL3 Co-efficient 0	F3A7	F3A6	F3A5	F3A4	F3A3	F3A2	F3A1	F3A0
13H	FIL3 Co-efficient 1	F3AS	0	F3A13	F3A12	F3A11	F3A10	F3A9	F3A8
14H	FIL3 Co-efficient 2	F3B7	F3B6	F3B5	F3B4	F3B3	F3B2	F3B1	F3B0
15H	FIL3 Co-efficient 3	0	0	F3B13	F3B12	F3B11	F3B10	F3B9	F3B8
16H	EQ Co-efficient 0	EQA7	EQA6	EQA5	EQA4	EQA3	EQA2	EQA1	EQA0
17H	EQ Co-efficient 1	EQA15	EQA14	EQA13	EQA12	EQA11	EQA10	EQA9	EQA8
18H	EQ Co-efficient 2	EQB7	EQB6	EQB5	EQB4	EQB3	EQB2	EQB1	EQB0
19H	EQ Co-efficient 3	0	0	EQB13	EQB12	EQB11	EQB10	EQB9	EQB8
1AH	EQ Co-efficient 4	EQC7	EQC6	EQC5	EQC4	EQC3	EQC2	EQC1	EQC0
1BH	EQ Co-efficient 5	EQC15	EQC14	EQC13	EQC12	EQC11	EQC10	EQC9	EQC8
1CH	FIL1 Co-efficient 0	F1A7	F1A6	F1A5	F1A4	F1A3	F1A2	F1A1	F1A0
1DH	FIL1 Co-efficient 1	F1AS	0	F1A13	F1A12	F1A11	F1A10	F1A9	F1A8
1EH	FIL1 Co-efficient 2	F1B7	F1B6	F1B5	F1B4	F1B3	F1B2	F1B1	F1B0
1FH	FIL1 Co-efficient 3	0	0	F1B13	F1B12	F1B11	F1B10	F1B9	F1B8
20H	Power Management 4	PMMAINR4	PMMAINL4	PMMAINR3	PMMAINL3	PMMAINR2	PMMAINL2	PMMICR	PMMICL
21H	Mode Control 5	0	0	MICR3	MICL3	L4DIF	MIX	AIN3	LODIF
22H	Lineout Mixing Select	LOM	LOM3	RINR4	LINL4	RINR3	LINL3	RINR2	LINL2
23H	HP Mixing Select	0	HPM3	RINH4	LINH4	RINH3	LINH3	RINH2	LINH2
24H	Reserved	0	0	0	0	0	0	0	0

These bits are added in the AK4645A.

PIN/FUNCTION

No.	Pin Name	I/O	Function
1	MPWR	O	MIC Power Supply Pin
2	VCOM	O	Common Voltage Output Pin, 0.45 x AVDD Bias voltage of ADC inputs and DAC outputs.
3	AVSS	-	Analog Ground Pin
4	AVDD	-	Analog Power Supply Pin, 2.6 ~ 3.5V
5	RIN3	I	Rch Analog Input 3 Pin (AIN3 bit = "1")
6	I2C	I	Control Mode Select Pin "H": I ² C Bus, "L": 3-wire Serial
7	PDN	I	Power-Down Mode Pin "H": Power-up, "L": Power-down, reset and initializes the control register.
8	CSN	I	Chip Select Pin (I2C pin = "L": 3-wire Serial Mode)
	CAD0	I	Chip Address 1 Select Pin (I2C pin = "H": I ² C Bus Mode)
9	CCLK	I	Control Data Clock Pin (I2C pin = "L": 3-wire Serial Mode)
	SCL	I	Control Data Clock Pin (I2C pin = "H": I ² C Bus Mode)
10	CDTI	I	Control Data Input Pin (I2C pin = "L": 3-wire Serial Mode)
	SDA	I/O	Control Data Input Pin (I2C pin = "H": I ² C Bus Mode)
11	SDTI	I	Audio Serial Data Input Pin
12	SDTO	O	Audio Serial Data Output Pin
13	LRCK	I/O	Input / Output Channel Clock Pin
14	BICK	I/O	Audio Serial Data Clock Pin
15	DVDD	-	Digital Power Supply Pin, 2.6 ~ 3.5V
16	TVDD	-	Digital I/O Power Supply Pin, 1.6 ~ 3.5V
17	MCKI	I	External Master Clock Input Pin
18	TESTO	O	Test Pin This pin must be open.
19	HVSS	-	Headphone Amp Ground Pin
20	HVDD	-	Headphone Amp Power Supply Pin
21	HPR	O	Rch Headphone-Amp Output Pin
22	HPL	O	Lch Headphone-Amp Output Pin
23	MUTET	O	Mute Time Constant Control Pin Connected to HVSS pin with a capacitor for mute time constant.
24	RIN4	I	Rch Analog Input 4 Pin (L4DIF bit = "0": Single-ended Input)
	IN4-	I	Negative Line Input 4 Pin (L4DIF bit = "1": Full-differential Input)
25	LIN4	I	Lch Analog Input 4 Pin (L4DIF bit = "0": Single-ended Input)
	IN4+	I	Positive Line Input 4 Pin (L4DIF bit = "1": Full-differential Input)
26	ROUT	O	Rch Stereo Line Output Pin (LODIF bit = "0": Single-ended Stereo Output)
	LON	O	Negative Line Output Pin (LODIF bit = "1": Full-differential Mono Output)
27	LOUT	O	Lch Stereo Line Output Pin (LODIF bit = "0": Single-ended Stereo Output)
	LOP	O	Positive Line Output Pin (LODIF bit = "1": Full-differential Mono Output)
28	MIN	I	Mono Signal Input Pin
	LIN3	I	Lch Analog Input 3 Pin
29	RIN2	I	Rch Analog Input 2 Pin (MDIF2 bit = "0": Single-ended Input)
	IN2-	I	Microphone Negative Input 2 Pin (MDIF2 bit = "1": Full-differential Input)
30	LIN2	I	Lch Analog Input 2 Pin (MDIF2 bit = "0": Single-ended Input)
	IN2+	I	Microphone Positive Input 2 Pin (MDIF2 bit = "1": Full-differential Input)
31	LIN1	I	Lch Analog Input 1 Pin (MDIF1 bit = "0": Single-ended Input)
	IN1-	I	Microphone Negative Input 1 Pin (MDIF1 bit = "1": Full-differential Input)
32	RIN1	I	Rch Analog Input 1 Pin (MDIF1 bit = "0": Single-ended Input)
	IN1+	I	Microphone Positive Input 1 Pin (MDIF1 bit = "1": Full-differential Input)

Note 1. All input pins except analog input pins (MIN/LIN3, LIN1, RIN1, LIN2, RIN2, RIN3, RIN4, LIN4) must not be left floating.

Note 2. AVDD or AVSS voltage must be input to I2C pin.

■ Handling of Unused Pin

The unused I/O pins must be processed appropriately as below.

Classification	Pin Name	Setting
Analog	MPWR, RIN3, HPR, HPL, MUTET, RIN4/IN4-, LIN4/IN4+, ROU/LOP, LOU/LON, MIN/LIN3, RIN2/IN2-, LIN2/IN2+, LIN1/IN1-, RIN1/IN1+	These pins must be open.
Digital	TESTO	This pin must be open.
	MCKI	This pin must be connected to HVSS.

ABSOLUTE MAXIMUM RATINGS

(AVSS=HVSS=0V; Note 3, Note 4)

Parameter	Symbol	min	max	Units
Power Supplies:				
Analog	AVDD	-0.3	6.0	V
Digital	DVDD	-0.3	6.0	V
Digital I/O	TVDD	-0.3	6.0	V
Headphone-Amp	HVDD	-0.3	6.0	V
Input Current, Any Pin Except Supplies	IIN	-	±10	mA
Analog Input Voltage (Note 5)	VINA	-0.3	AVDD+0.3	V
Digital Input Voltage (Note 6)	VIND	-0.3	TVDD+0.3	V
Ambient Temperature (powered applied)	Ta	-30	85	°C
Storage Temperature	Tstg	-65	150	°C

Note 3. All voltages with respect to ground.

Note 4. AVSS and HVSS must be connected to the same analog ground plane.

Note 5. I2C, RIN4/IN4-, LIN4/IN4+, MIN/LIN3, RIN3, RIN2/IN2-, LIN2/IN2+, LIN1/IN1-, RIN1/IN1+ pins

Note 6. PDN, CSN/CAD0, CCLK/SCL, CDTI/SDA, SDTI, LRCK, BICK, MCKI pins

Pull-up resistors at SDA and SCL pins should be connected to (TVDD+0.3)V or less voltage.

WARNING: Operation at or beyond these limits may result in permanent damage to the device.

Normal operation is not guaranteed at these extremes.

RECOMMENDED OPERATING CONDITIONS

(AVSS=HVSS=0V; Note 3)

Parameter	Symbol	min	typ	max	Units
Power Supplies					
Analog	AVDD	2.6	3.3	3.5	V
Digital	DVDD	2.6	3.3	3.5	V
Digital I/O	TVDD	1.6	3.3	3.5	V
HP-Amp	HVDD	2.6	3.3 / 5.0	5.25	V
Difference	AVDD-DVDD	-0.3	0	+0.3	V
Difference	TVDD-DVDD	-	0	+0.3	V

Note 3. All voltages with respect to ground.

Note 7. The power-up sequence between AVDD, DVDD, TVDD and HVDD is not critical. PDN pin must be held to "L" upon power-up. PDN pin must be set to "H" after all power supplies are powered-up.

The AK4645A must be operated by the recommended power-up/down sequence shown in "System Design (Grounding and Power Supply Decoupling)" to avoid the pop noise at line output and headphone output.

When one of power supplies is partially powered OFF, the power supply current at power-down mode may be increased. All the power supplies should be powered OFF when the power supply is powered OFF.

* AKEMD assumes no responsibility for the usage beyond the conditions in this datasheet.

ANALOG CHARACTERISTICS

(Ta=25°C; AVDD=DVDD=TVDD=HVDD=3.3V; AVSS=HVSS=0V; fs=44.1kHz, BICK=64fs;
Signal Frequency=1kHz; 16bit Data; Measurement frequency=20Hz ~ 20kHz; unless otherwise specified)

Parameter		min	typ	max	Units
MIC Amplifier: LIN1/RIN1/LIN2/RIN2/LIN4/RIN4 pins & LIN3/RIN3 pins (AIN3 bit = "1"); MDIF1=MDIF2 bits = "0" (Single-ended inputs)					
Input Resistance	MGAIN1-0 bits = "00"	40	60	80	kΩ
	MGAIN1-0 bits = "01", "10" or "11"	20	30	40	kΩ
Gain	MGAIN1-0 bits = "00"	-	0	-	dB
	MGAIN1-0 bits = "01"	-	+20	-	dB
	MGAIN1-0 bits = "10"	-	+26	-	dB
	MGAIN1-0 bits = "11"	-	+32	-	dB
MIC Amplifier: IN1+/IN1-/IN2+/IN2- pins; MDIF1 = MDIF2 bits = "1" (Full-differential input)					
Maximum Input Voltage (Note 8)					
	MGAIN1-0 bits = "01"	-	-	0.228	Vpp
	MGAIN1-0 bits = "10"	-	-	0.114	Vpp
	MGAIN1-0 bits = "11"	-	-	0.057	Vpp
MIC Power Supply: MPWR pin					
Output Voltage (Note 9)		2.22	2.47	2.72	V
Load Resistance		0.5	-	-	kΩ
Load Capacitance		-	-	30	pF
ADC Analog Input Characteristics: LIN1/RIN1/LIN2/RIN2/LIN4/RIN4 pins & LIN3/RIN3 pins (AIN3 bit = "1") → ADC → IVOL, IVOL=0dB, ALC=OFF					
Resolution		-	-	16	Bits
Input Voltage (Note 10)	(Note 11)	0.168	0.198	0.228	Vpp
	(Note 12)	1.68	1.98	2.28	Vpp
S/(N+D) (-1dBFS)	(Note 11, LIN1/RIN1/LIN2/RIN2)	71	83	-	dBFS
	(Note 11, LIN3/RIN3/LIN4/RIN4)	-	83	-	dBFS
	(Note 12, except for LIN3/RIN3)	-	88	-	dBFS
	(Note 12, LIN3/RIN3)	-	72	-	dBFS
D-Range (-60dBFS, A-weighted)	(Note 11)	76	86	-	dB
	(Note 12)	-	95	-	dB
S/N (A-weighted)	(Note 11)	76	86	-	dB
	(Note 12)	-	95	-	dB
Interchannel Isolation	(Note 11)	75	90	-	dB
	(Note 12)	-	100	-	dB
Interchannel Gain Mismatch	(Note 11)	-	0.1	0.8	dB
	(Note 12)	-	0.1	0.8	dB

Note 8. The voltage difference between IN1/2+ and IN1/2- pins. AC coupling capacitor should be inserted in series at each input pin. Full-differential mic input is not available at MGAIN1-0 bits = "00". Maximum input voltage of IN1+, IN1-, IN2+ and IN2- pins are proportional to AVDD voltage, respectively.

Vin = 0.069 x AVDD (max)@MGAIN1-0 bits = "01", 0.035 x AVDD (max)@MGAIN1-0 bits = "10", 0.017 x AVDD (max)@MGAIN1-0 bits = "11".

When the signal larger than above value is input to IN1+, IN1-, IN2+ or IN2- pin, ADC does not operate normally.

Note 9. Output voltage is proportional to AVDD voltage. Vout = 0.75 x AVDD (typ)

Note 10. Input voltage is proportional to AVDD voltage. Vin = 0.06 x AVDD (typ)@MGAIN1-0 bits = "01" (+20dB),
Vin = 0.6 x AVDD(typ)@MGAIN1-0 bits = "00" (0dB)

Note 11. MGAIN1-0 bits = "01" (+20dB)

Note 12. MGAIN1-0 bits = "00" (0dB)

Parameter		min	typ	max	Units
DAC Characteristics:					
Resolution		-	-	16	Bits
Stereo Line Output Characteristics: DAC → LOUT/ROUT pins, ALC=OFF, IVOL=0dB, DVOL=0dB, LOVL bit = "0", LODIF bit = "0", $R_L=10k\Omega$ (Single-ended); unless otherwise specified.					
Output Voltage (Note 13)	LOVL bit = "0"	1.78	1.98	2.18	V_{pp}
	LOVL bit = "1"	2.25	2.50	2.75	V_{pp}
S/(N+D) (-3dBFS)		78	88	-	dBFS
S/N (A-weighted)		82	92	-	dB
Interchannel Isolation		80	100	-	dB
Interchannel Gain Mismatch		-	0.1	0.5	dB
Load Resistance		10	-	-	$k\Omega$
Load Capacitance		-	-	30	pF
Mono Line Output Characteristics: DAC → LOP/LON pins, ALC=OFF, IVOL=0dB, DVOL=0dB, LOVL bit = "0", LODIF bit = "1", $R_L=10k\Omega$ for each pin (Full-differential)					
Output Voltage (Note 14)	LOVL bit = "0"	3.52	3.96	4.36	V_{pp}
	LOVL bit = "1"	-	5.00	-	V_{pp}
S/(N+D) (-3dBFS)		78	88	-	dBFS
S/N (A-weighted)		85	95	-	dB
Load Resistance (LOP/LON pins, respectively)		10	-	-	$k\Omega$
Load Capacitance (LOP/LON pins, respectively)		-	-	30	pF

Note 13. Output voltage is proportional to AVDD voltage. $V_{out} = 0.6 \times AVDD$ (typ)@LOVL bit = "0".

Note 14. Output voltage is proportional to AVDD voltage. $V_{out} = (LOP) - (LON) = 1.2 \times AVDD$ (typ)@LOVL bit = "0".

Parameter		min	typ	max	Units
Headphone-Amp Characteristics: DAC → HPL/HPR pins, ALC=OFF, IVOL=0dB, DVOL=0dB, VBAT bit = "0"; unless otherwise specified.					
Output Voltage (Note 15)					
HPG bit = "0", 0dBFS, HVDD=3.3V, $R_L=22.8\Omega$		1.58	1.98	2.38	V _{pp}
HPG bit = "1", 0dBFS, HVDD=5V, $R_L=100\Omega$		2.40	3.00	3.60	V _{pp}
HPG bit = "1", 0dBFS, HVDD=3.3V, $R_L=16\Omega$ (Po=62mW)		-	1.0	-	V _{rms}
HPG bit = "1", 0dBFS, HVDD=5V, $R_L=16\Omega$ (Po=70mW)		-	1.06	-	V _{rms}
S/(N+D)					
HPG bit = "0", -3dBFS, HVDD=3.3V, $R_L=22.8\Omega$		57	67	-	dBFS
HPG bit = "1", -3dBFS, HVDD=5V, $R_L=100\Omega$		-	75	-	dBFS
HPG bit = "1", 0dBFS, HVDD=3.3V, $R_L=16\Omega$ (Po=62mW)		-	20	-	dBFS
HPG bit = "1", 0dBFS, HVDD=5V, $R_L=16\Omega$ (Po=70mW)		-	65	-	dBFS
S/N (A-weighted)	(Note 16)	80	90	-	dB
	(Note 17)	-	90	-	dB
Interchannel Isolation	(Note 16)	65	75	-	dB
	(Note 17)	-	80	-	dB
Interchannel Gain Mismatch	(Note 16)	-	0.1	0.8	dB
	(Note 17)	-	0.1	0.8	dB
Load Resistance		16	-	-	Ω
Load Capacitance	C1 in Figure 2	-	-	30	pF
	C2 in Figure 2	-	-	300	pF

Note 15. Output voltage is proportional to AVDD voltage.

$$V_{out} = 0.6 \times AVDD(\text{typ}) @ \text{HPG bit} = "0", 0.91 \times AVDD(\text{typ}) @ \text{HPG bit} = "1".$$

Note 16. HPG bit = "0", HVDD=3.3V, $R_L=22.8\Omega$.

Note 17. HPG bit = "1", HVDD=5V, $R_L=100\Omega$.

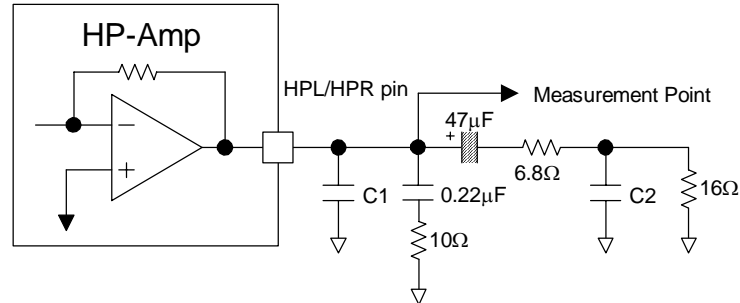


Figure 2. Headphone-Amp output circuit

Parameter	min	typ	max	Units	
Mono Input: MIN pin (AIN3 bit = "0"; External Input Resistance=20kΩ)					
Maximum Input Voltage (Note 18)	-	1.98	-	V _{pp}	
Gain (Note 19)					
MIN → LOU/ROUT	LOVL bit = "0"	-4.5	0	+4.5	dB
	LOVL bit = "1"	-	+2	-	dB
MIN → HPL/HPR	HPG bit = "0"	-24.5	-20	-15.5	dB
	HPG bit = "1"	-	-16.4	-	dB
Stereo Input: LIN2/RIN2/LIN4/RIN4 pins; LIN3/RIN3 pins (AIN3 bit = "1")					
Maximum Input Voltage (Note 20)	-	1.98	-	V _{pp}	
Gain					
LIN/RIN → LOU/ROUT	LOVL bit = "0"	-4.5	0	+4.5	dB
	LOVL bit = "1"	-	+2	-	dB
LIN/RIN → HPL/HPR	HPG bit = "0"	-4.5	0	+4.5	dB
	HPG bit = "1"	-	+3.6	-	dB
Full-differential Mono Input: IN4+/- pins (L4DIF bit = "1")					
Maximum Input Voltage (Note 21)	-	3.96	-	V _{pp}	
Gain					
IN4+/- → LOU/ROUT (LODIF bit = "0")	LOVL bit = "0"	-10.5	-6	-1.5	dB
	LOVL bit = "1"	-	-4	-	dB
IN4+/- → LOP/LON (LODIF bit = "1", Note 22)	LOVL bit = "0"	-4.5	0	+4.5	dB
	LOVL bit = "1"	-	+2	-	dB
IN4+/- → HPL/HPR	HPG bit = "0"	-10.5	-6	-1.5	dB
	HPG bit = "1"	-	-2.4	-	dB
Power Supplies:					
Power-Up (PDN pin = "H")					
All Circuit Power-up:					
AVDD+DVDD+TVDD (Note 23)	-	12	18	mA	
HVDD: HP-Amp Normal Operation No Output (Note 24)	-	3	4.5	mA	
Power-Down (PDN pin = "L") (Note 25)					
AVDD+DVDD+TVDD+HVDD	-	1	100	μA	

Note 18. Maximum voltage is in proportion to both AVDD and external input resistance (R_{in}). $V_{in} = 0.6 \times AVDD \times R_{in} / 20k\Omega$ (typ).

Note 19. The gain is in inverse proportion to external input resistance.

Note 20. Maximum Input voltage is proportional to AVDD voltage. $V_{out} = 0.6 \times AVDD$ (typ).

Note 21. Maximum Input voltage is proportional to AVDD voltage. $V_{out} = (IN4+) - (IN4-) = 1.2 \times AVDD$ (typ). The signals with same amplitude and inverted phase should be input to IN4+ and IN4- pins, respectively.

Note 22. $V_{out} = (LOP) - (LON)$ at LODIF bit = "1".

Note 23. When EXT Slave Mode (M/S = bit = "0", MCKI = 12.288MHz), PMVCM = PMADL = PMADR = PMDAC = PMLO = PMHPL = PMHPR = PMMIN = PMMP = "1":

AVDD=9mA (typ), DVDD=3mA (typ), TVDD=0.03mA (typ).

Note 24. When PMADL = PMADR = PMDAC = PMLO = PMHPL = PMHPR = PMVCM = PMMIN bits = "1"

Note 25. All digital input pins are fixed to TVDD or HVSS.

■ Power Consumption for Each Operation Mode

Conditions: Ta=25°C; AVDD=DVDD=TVDD=HVDD=3.3V; AVSS=HVSS=0V; fs=44.1kHz, External Slave Mode, BICK=64fs; 1kHz, 0dBFS input; Headphone = No output.

Mode	Power Management Bit															AVDD [mA]	DVDD [mA]	TVDD [mA]	HVDD [mA]	Total Power [mW]											
	00H					01H		10H																							
	PMVCM	PMMIN	PML0	PMDAC	PMADL	PMHPL	PMHPR	PMADR	PMMICL	PMMICR	PMAINL2	PMAINR2	PMAINL3	PMAINR3	PMAINL4						PMAINR4										
All Power-down	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0					
DAC → Lineout	1	0	1	1	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	3.2	1.8	0.03	0.2	17.2
DAC → HP	1	0	0	1	0	1	1	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	2.6	1.8	0.03	3.0	24.5
LIN2/RIN2 → HP	1	0	0	0	0	1	1	0	0	0	0	1	1	0	0	0	0	0	0	0	0	0	0	0	0	0	1.9	0	0	3.0	16.2
LIN2/RIN2 → ADC	1	0	0	0	1	0	0	1	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	5.5	1.6	0.03	0.2	24.2
LIN1 (Mono) → ADC	1	0	0	0	1	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	3.5	1.5	0.03	0.2	17.3
LIN2/RIN2 → ADC & DAC → HP	1	0	0	1	1	1	1	1	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	7.1	2.7	0.03	3.0	42.3

Table 1. Power Consumption for each operation mode (typ)

FILTER CHARACTERISTICS

(Ta=25°C; AVDD=DVDD=2.6 ~ 3.5V; TVDD=1.6 ~ 3.5V; HVDD=2.6 ~ 5.25V; fs=44.1kHz; DEM=OFF; FIL1=FIL3=EQ=OFF)

Parameter		Symbol	min	typ	max	Units	
ADC Digital Filter (Decimation LPF):							
Passband (Note 26)	±0.16dB	PB	0	-	17.3	kHz	
	-0.66dB		-	19.4	-	kHz	
	-1.1dB		-	19.9	-	kHz	
	-6.9dB		-	22.1	-	kHz	
Stopband		SB	26.1	-	-	kHz	
Passband Ripple		PR	-	-	±0.1	dB	
Stopband Attenuation		SA	73	-	-	dB	
Group Delay (Note 27)		GD	-	19	-	1/fs	
Group Delay Distortion		ΔGD	-	0	-	μs	
ADC Digital Filter (HPF): (Note 28)							
Frequency Response (Note 26)	-3.0dB	FR	-	0.9	-	Hz	
	-0.5dB		-	2.7	-	Hz	
	-0.1dB		-	6.0	-	Hz	
DAC Digital Filter (LPF):							
Passband (Note 26)	±0.1dB	PB	0	-	19.6	kHz	
	-0.7dB		-	20.0	-	kHz	
	-6.0dB		-	22.05	-	kHz	
Stopband		SB	25.2	-	-	kHz	
Passband Ripple		PR	-	-	±0.01	dB	
Stopband Attenuation		SA	59	-	-	dB	
Group Delay (Note 27)		GD	-	25	-	1/fs	
DAC Digital Filter (LPF) + SCF:							
Frequency Response: 0 ~ 20.0kHz		FR	-	±1.0	-	dB	
DAC Digital Filter (HPF): (Note 28)							
Frequency Response (Note 26)	-3.0dB	FR	-	0.9	-	Hz	
	-0.5dB		-	2.7	-	Hz	
	-0.1dB		-	6.0	-	Hz	
BOOST Filter: (Note 29)							
Frequency Response	MIN	20Hz	FR	-	5.76	-	dB
		100Hz		-	2.92	-	dB
		1kHz		-	0.02	-	dB
	MID	20Hz	FR	-	10.80	-	dB
		100Hz		-	6.84	-	dB
		1kHz		-	0.13	-	dB
	MAX	20Hz	FR	-	16.06	-	dB
		100Hz		-	10.54	-	dB
		1kHz		-	0.37	-	dB

Note 26. The passband and stopband frequencies scale with fs (system sampling rate).

For example, DAC is PB=0.454*fs (@-0.7dB). Each response refers to that of 1kHz.

Note 27. The calculated delay time caused by digital filtering. This time is from the input of analog signal to setting of the 16-bit data of both channels from the input register to the output register of the ADC. This time includes the group delay of the HPF. For the DAC, this time is from setting the 16-bit data of both channels from the input register to the output of analog signal. Group delay of DAC part is 25/fs(typ) at PMADL=PMADR bits = "0".

Note 28. When PMADL bit = "1" or PMADR bit = "1", the HPF of ADC is enabled but the HPF of DAC is disabled. When PMADL=PMADR bits = "0", PMDAC bit = "1", the HPF of DAC is enabled but the HPF of ADC is disabled.

Note 29. These frequency responses scale with fs. If a high-level signal is input, the analog output clips to the full-scale at low frequency.

DC CHARACTERISTICS

(Ta=25°C; AVDD=DVDD=2.6 ~ 3.5V; TVDD=1.6 ~ 3.5V; HVDD=2.6 ~ 5.25V)

Parameter		Symbol	min	typ	max	Units
High-Level Input Voltage	2.2V ≤ TVDD ≤ 3.5V	VIH	70%TVDD	-	-	V
	1.6V ≤ TVDD < 2.2V	VIH	75%TVDD	-	-	V
Low-Level Input Voltage	2.2V ≤ TVDD ≤ 3.5V	VIL	-	-	30%TVDD	V
	1.6V ≤ TVDD < 2.2V	VIL	-	-	25%TVDD	V
High-Level Output Voltage (Iout = -200μA)		VOH	TVDD - 0.2	-	-	V
Low-Level Output Voltage (Except SDA pin: Iout = 200μA) (SDA pin: Iout = 3mA)		VOL	-	-	0.2	V
		VOL	-	-	0.4	V
Input Leakage Current		Iin	-	-	±10	μA

SWITCHING CHARACTERISTICS

 (Ta=25°C; AVDD=DVDD=2.6 ~ 3.5V; or 3.3V TVDD=1.6 ~ 3.5V; HVDD=2.6 ~ 5.25V or AVdd=DVdd=3.3V;
 CL=20pF; unless otherwise specified)

Parameter		Symbol	min	typ	max	Units
External Slave Mode						
MCKI Input Timing						
Frequency	256fs	fCLK	1.8816	-	12.288	MHz
	384fs	fCLK	2.8224	-	18.432	MHz
	512fs	fCLK	3.7632	-	13.312	MHz
	1024fs	fCLK	7.5264	-	13.312	MHz
Pulse Width Low		tCLKL	0.4/fCLK	-	-	ns
Pulse Width High		tCLKH	0.4/fCLK	-	-	ns
LRCK Input Timing						
Frequency	256fs	fs	7.35	-	48	kHz
	384fs	fs	7.35	-	48	kHz
	512fs	fs	7.35	-	26	kHz
	1024fs	fs	7.35	-	13	kHz
DSP Mode: Pulse Width High		tLRCKH	tBCK - 60	-	1/fs - tBCK	ns
Except DSP Mode: Duty Cycle		Duty	45	-	55	%
BICK Input Timing						
Period		tBCK	312.5	-	-	ns
Pulse Width Low		tBCKL	130	-	-	ns
Pulse Width High		tBCKH	130	-	-	ns
External Master Mode						
MCKI Input Timing						
Frequency	256fs	fCLK	1.8816	-	12.288	MHz
	384fs	fCLK	2.8224	-	18.432	MHz
	512fs	fCLK	3.7632	-	13.312	MHz
	1024fs	fCLK	7.5264	-	13.312	MHz
Pulse Width Low		tCLKL	0.4/fCLK	-	-	ns
Pulse Width High		tCLKH	0.4/fCLK	-	-	ns
LRCK Output Timing						
Frequency		fs	7.35	-	48	kHz
DSP Mode: Pulse Width High		tLRCKH	-	tBCK	-	ns
Except DSP Mode: Duty Cycle		Duty	-	50	-	%
BICK Output Timing						
Period	BCKO bit = "0"	tBCK	-	1/(32fs)	-	ns
	BCKO bit = "1"	tBCK	-	1/(64fs)	-	ns
Duty Cycle		dBCK	-	50	-	%

Parameter	Symbol	min	typ	max	Units
Audio Interface Timing (DSP Mode)					
Master Mode					
LRCK “↑” to BICK “↑” (Note 30)	tDBF	0.5 x tBCK – 40	0.5 x tBCK	0.5 x tBCK + 40	ns
LRCK “↑” to BICK “↓” (Note 31)	tDBF	0.5 x tBCK – 40	0.5 x tBCK	0.5 x tBCK + 40	ns
BICK “↑” to SDTO (BCKP bit = “0”)	tBSD	–70	-	70	ns
BICK “↓” to SDTO (BCKP bit = “1”)	tBSD	–70	-	70	ns
SDTI Hold Time	tSDH	50	-	-	ns
SDTI Setup Time	tSDS	50	-	-	ns
Slave Mode					
LRCK “↑” to BICK “↑” (Note 30)	tLRB	0.4 x tBCK	-	-	ns
LRCK “↑” to BICK “↓” (Note 31)	tLRB	0.4 x tBCK	-	-	ns
BICK “↑” to LRCK “↑” (Note 30)	tBLR	0.4 x tBCK	-	-	ns
BICK “↓” to LRCK “↑” (Note 31)	tBLR	0.4 x tBCK	-	-	ns
BICK “↑” to SDTO (BCKP bit = “0”)	tBSD	-	-	80	ns
BICK “↓” to SDTO (BCKP bit = “1”)	tBSD	-	-	80	ns
SDTI Hold Time	tSDH	50	-	-	ns
SDTI Setup Time	tSDS	50	-	-	ns
Audio Interface Timing (Right/Left justified & I²S)					
Master Mode					
BICK “↓” to LRCK Edge (Note 32)	tMBLR	–40	-	40	ns
LRCK Edge to SDTO (MSB) (Except I ² S mode)	tLRD	–70	-	70	ns
BICK “↓” to SDTO	tBSD	–70	-	70	ns
SDTI Hold Time	tSDH	50	-	-	ns
SDTI Setup Time	tSDS	50	-	-	ns
Slave Mode					
LRCK Edge to BICK “↑” (Note 32)	tLRB	50	-	-	ns
BICK “↑” to LRCK Edge (Note 32)	tBLR	50	-	-	ns
LRCK Edge to SDTO (MSB) (Except I ² S mode)	tLRD	-	-	80	ns
BICK “↓” to SDTO	tBSD	-	-	80	ns
SDTI Hold Time	tSDH	50	-	-	ns
SDTI Setup Time	tSDS	50	-	-	ns

Note 30. MSBS, BCKP bits = “00” or “11”.

Note 31. MSBS, BCKP bits = “01” or “10”.

Note 32. BICK rising edge must not occur at the same time as LRCK edge.

Parameter	Symbol	min	typ	max	Units
Control Interface Timing (3-wire Serial mode)					
CCLK Period	tCCK	200	-	-	ns
CCLK Pulse Width Low	tCCKL	80	-	-	ns
Pulse Width High	tCCKH	80	-	-	ns
CDTI Setup Time	tCDS	40	-	-	ns
CDTI Hold Time	tCDH	40	-	-	ns
CSN "H" Time	tCSW	150	-	-	ns
CSN Edge to CCLK "↑" (Note 34)	tCSS	50	-	-	ns
CCLK "↑" to CSN Edge (Note 34)	tCSH	50	-	-	ns
Control Interface Timing (I²C Bus mode): (Note 33)					
SCL Clock Frequency	fSCL	-	-	400	kHz
Bus Free Time Between Transmissions	tBUF	1.3	-	-	μs
Start Condition Hold Time (prior to first clock pulse)	tHD:STA	0.6	-	-	μs
Clock Low Time	tLOW	1.3	-	-	μs
Clock High Time	tHIGH	0.6	-	-	μs
Setup Time for Repeated Start Condition	tSU:STA	0.6	-	-	μs
SDA Hold Time from SCL Falling (Note 35)	tHD:DAT	0	-	-	μs
SDA Setup Time from SCL Rising	tSU:DAT	0.1	-	-	μs
Rise Time of Both SDA and SCL Lines	tR	-	-	0.3	μs
Fall Time of Both SDA and SCL Lines	tF	-	-	0.3	μs
Setup Time for Stop Condition	tSU:STO	0.6	-	-	μs
Capacitive Load on Bus	Cb	-	-	400	pF
Pulse Width of Spike Noise Suppressed by Input Filter	tSP	0	-	50	ns
Power-down & Reset Timing					
PDN Pulse Width (Note 36)	tPD	150	-	-	ns
PMADL or PMADR "↑" to SDTO valid (Note 37)	tPDV	-	1059	-	1/fs

Note 33. I²C is a registered trademark of Philips Semiconductors.

Note 34. CCLK rising edge must not occur at the same time as CSN edge.

Note 35. Data must be held long enough to bridge the 300ns-transition time of SCL.

Note 36. The AK4645A can be reset by the PDN pin = "L".

Note 37. This is the count of LRCK "↑" from the PMADL or PMADR bit = "1".

■ Timing Diagram

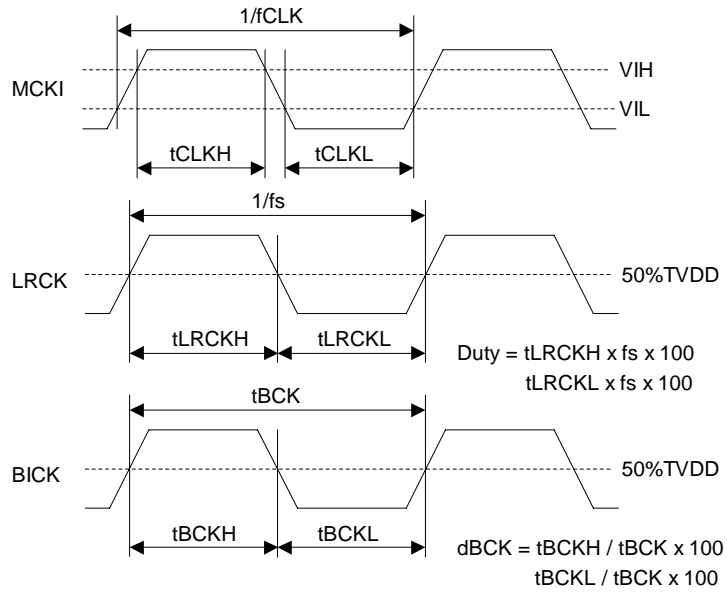


Figure 3. Clock Timing (EXT Master mode)

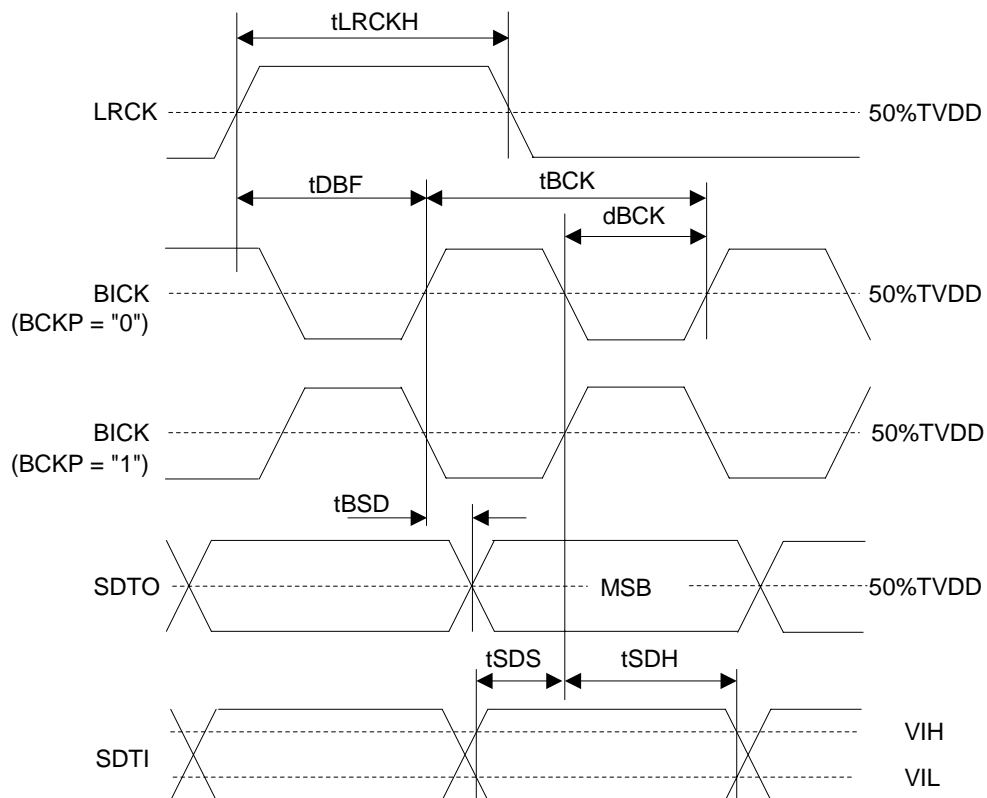


Figure 4. Audio Interface Timing (EXT Master mode, DSP mode, MSBS = "0")

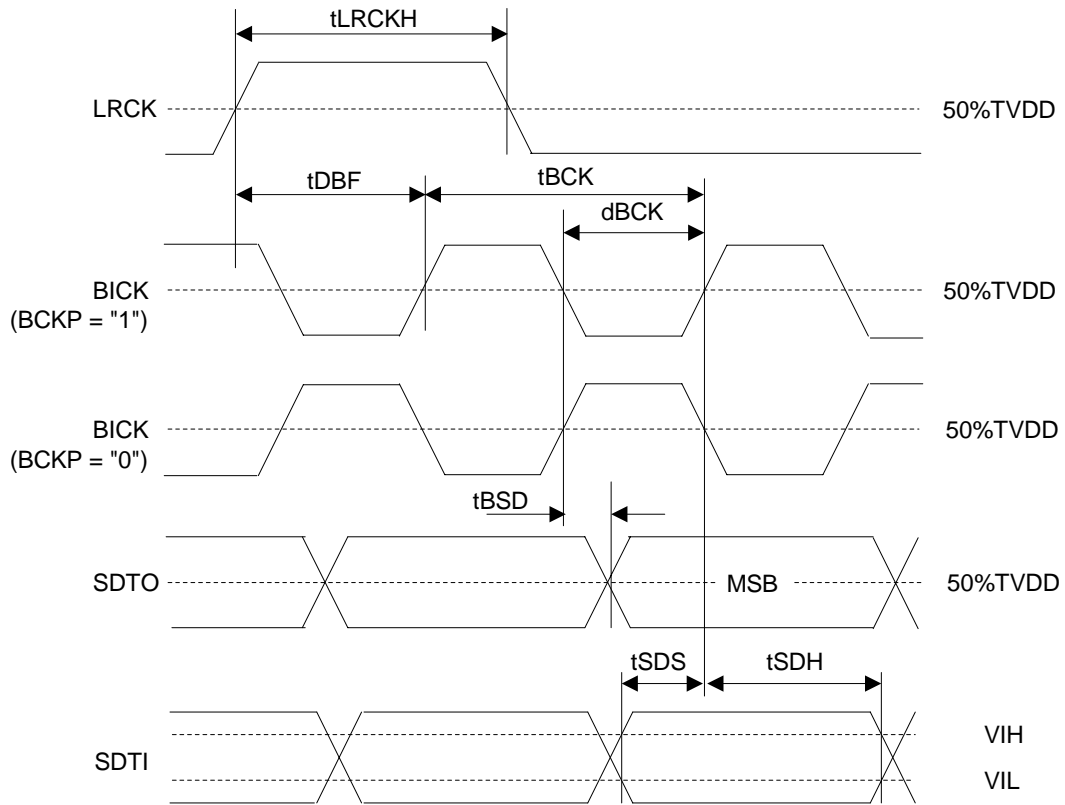


Figure 5. Audio Interface Timing (EXT Master mode, DSP mode, MSBS = "1")

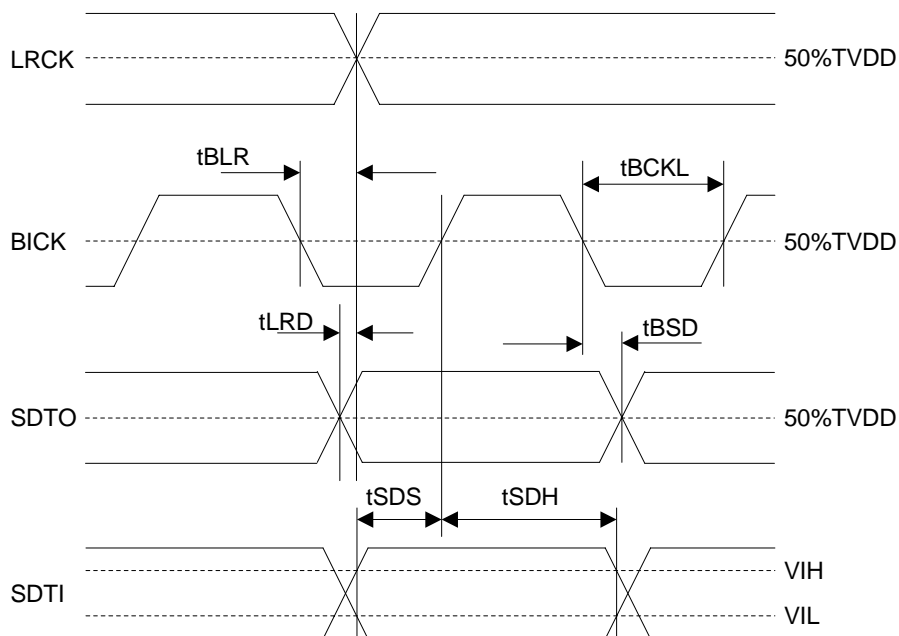


Figure 6. Audio Interface Timing (EXT Master mode, Except DSP mode)

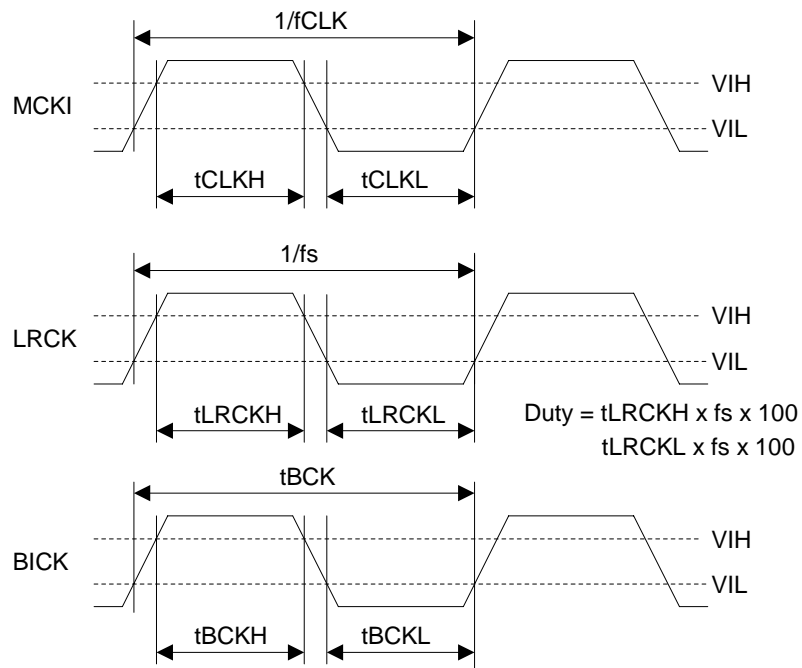


Figure 7. Clock Timing (EXT Slave mode)

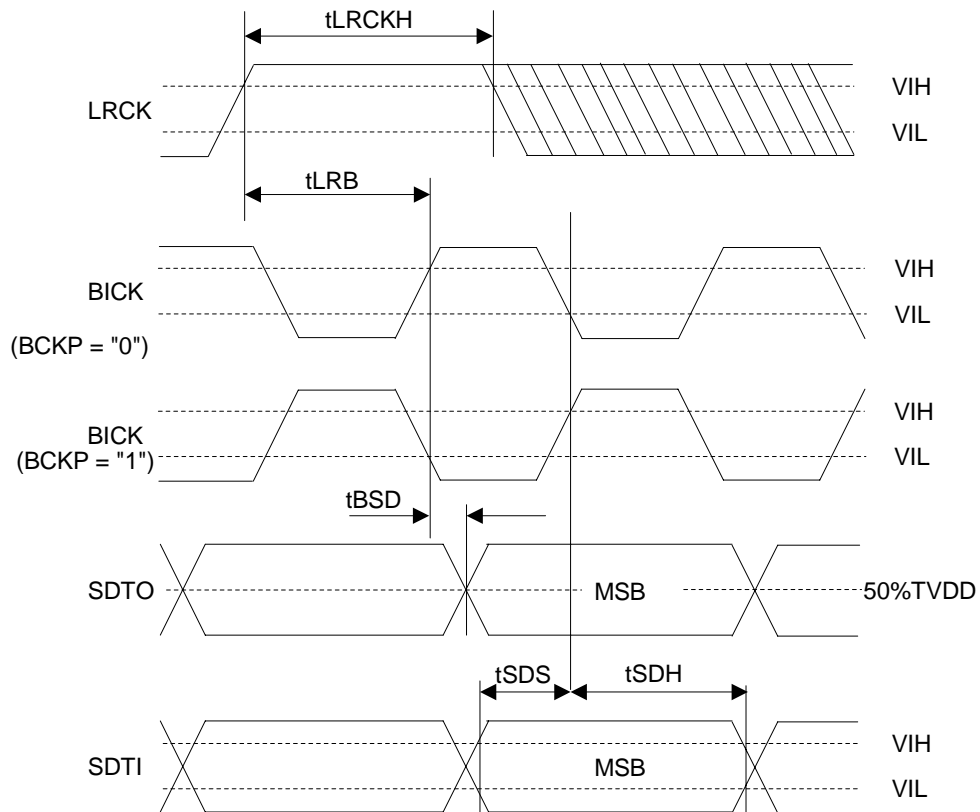


Figure 8. Audio Interface Timing (EXT Slave mode, DSP mode; MSBS = "0")

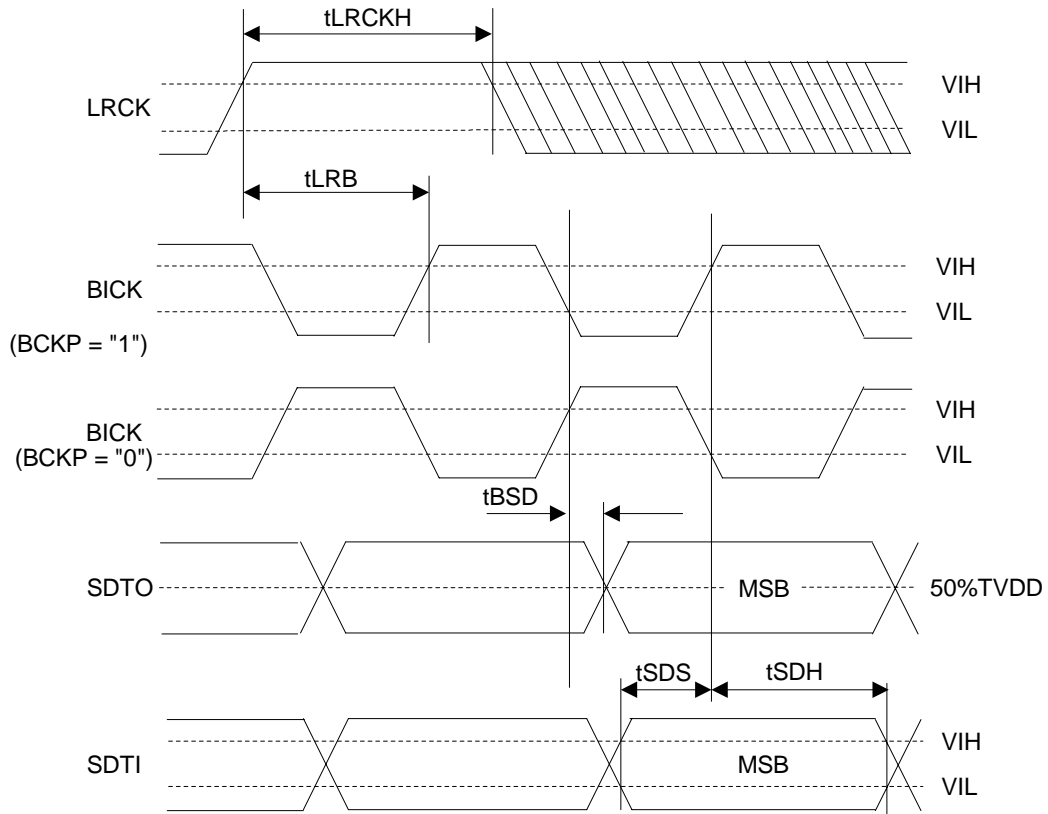


Figure 9. Audio Interface Timing (EXT Slave mode, DSP mode, MSBS = "1")

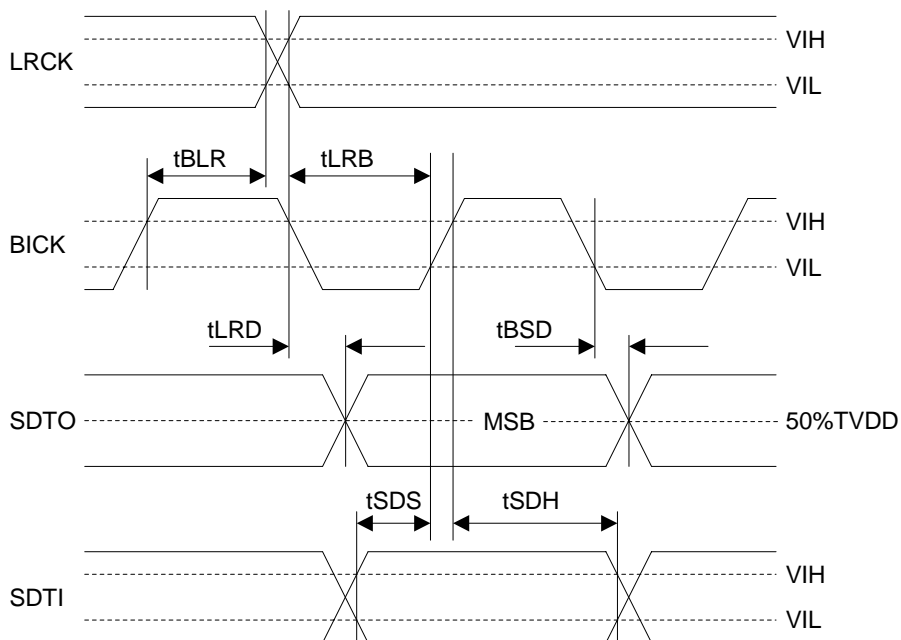


Figure 10. Audio Interface Timing (EXT Slave mode, Except DSP mode)

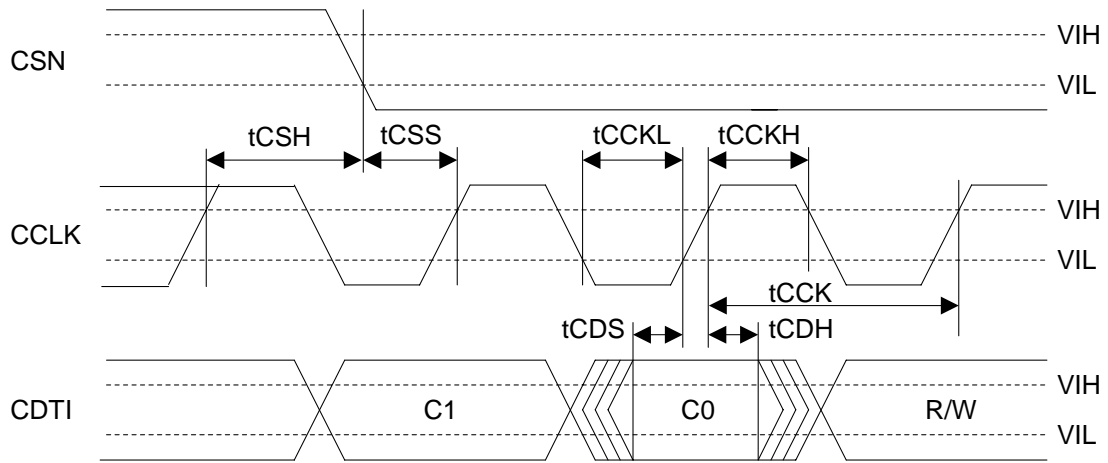


Figure 11. WRITE Command Input Timing

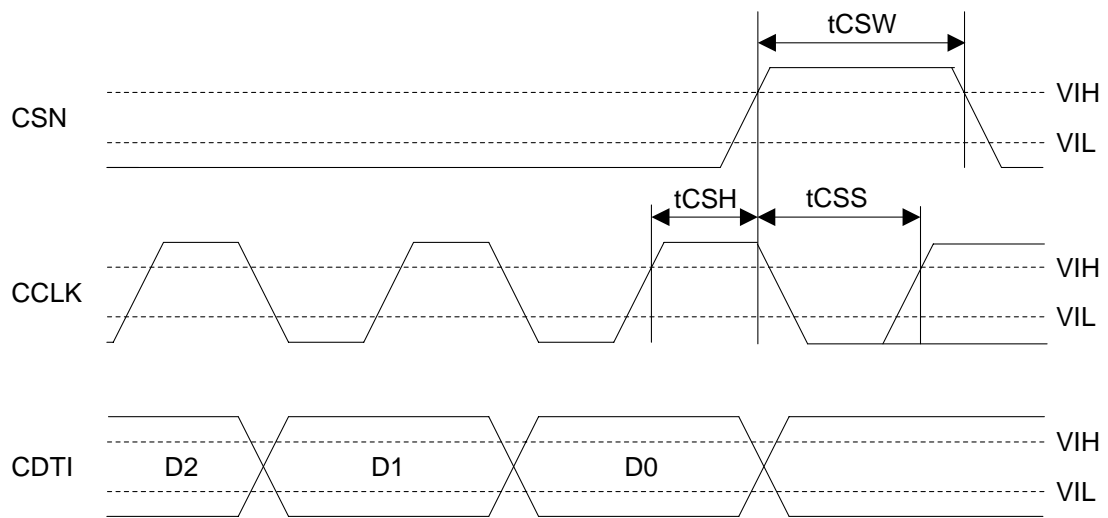


Figure 12. WRITE Data Input Timing

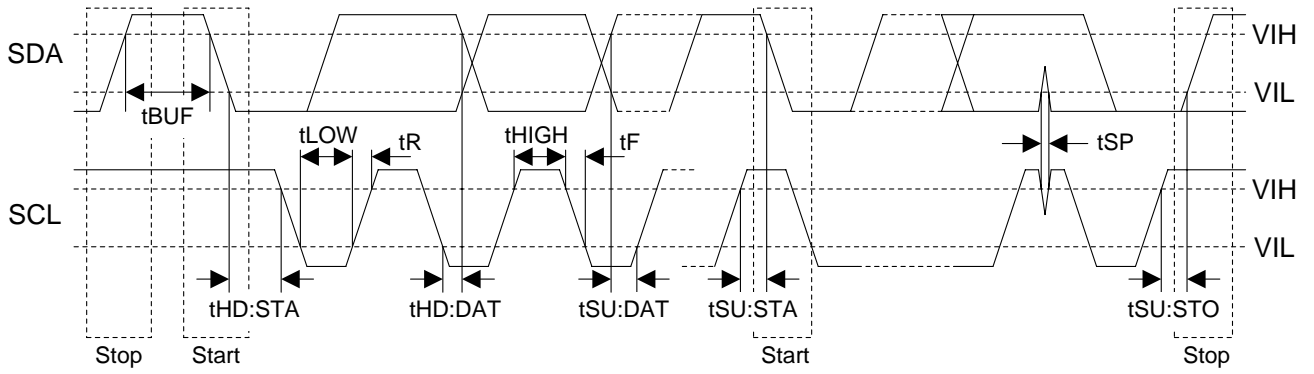


Figure 13. I²C Bus Mode Timing

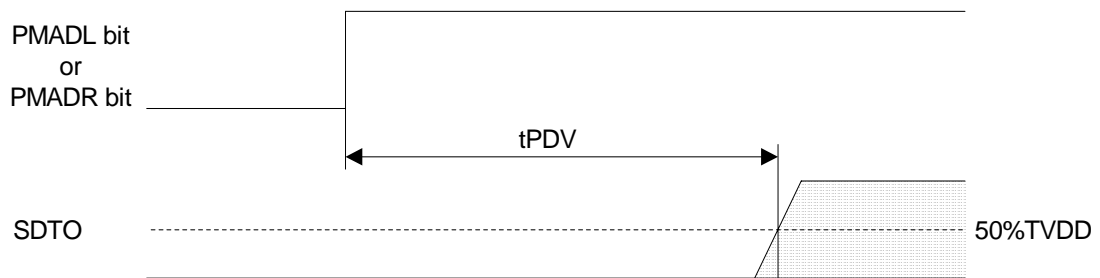


Figure 14. Power Down & Reset Timing 1

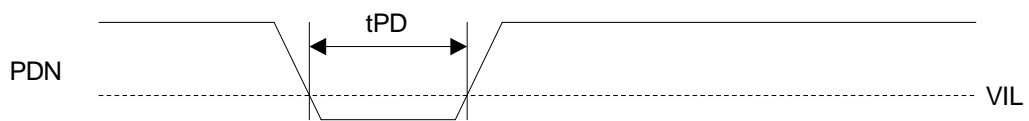


Figure 15. Power Down & Reset Timing 2

OPERATION OVERVIEW

■ System Clock

There are the following two clock modes to interface with external devices (Table 2 and Table 3).

Mode	M/S bit	Figure
EXT Slave Mode	0	Figure 16
EXT Master Mode	1	Figure 17

Table 2. Clock Mode Setting (x: Don't care)

Mode	MCKI pin	BICK pin	LRCK pin
EXT Slave Mode	Selected by FS1-0 bits	Input ($\geq 32\text{fs}$)	Input (1fs)
EXT Master Mode	Selected by FS1-0 bits	Output (Selected by BCKO bit)	Output (1fs)

Table 3. Clock pins state in Clock Mode

■ Master Mode/Slave Mode

The M/S bit selects either master or slave mode. M/S bit = "1" selects master mode and "0" selects slave mode. When the AK4645A is powered-down (PDN pin = "L") and exits reset state, the AK4645A is in slave mode. After exiting reset state, the AK4645A goes to master mode by changing M/S bit = "1".

When the AK4645A is in master mode, LRCK and BICK pins are floating state until M/S bit becomes "1". LRCK and BICK pins of the AK4645A should be pulled-down or pulled-up by the resistor (about 100k Ω) externally to avoid the floating state.

M/S bit	Mode
0	Slave Mode
1	Master Mode

(default)

Table 4. Select Master/Slave Mode

■ EXT Slave Mode (M/S bit = "0")

The Master clock is input from the MCKI pin for ADC and DAC. This mode is compatible with I/F of the normal audio CODEC. The clocks required to operate are MCKI (256fs, 384fs, 512fs or 1024fs), LRCK (fs) and BICK (≥ 32 fs). The master clock (MCKI) should be synchronized with LRCK. The phase between these clocks does not matter. The input frequency of MCKI is selected by FS1-0 bits (Table 5).

Mode	FS3-2 bits	FS1 bit	FS0 bit	MCKI Input Frequency	Sampling Frequency Range	(default)
0	Don't care	0	0	256fs	7.35kHz ~ 48kHz	
1	Don't care	0	1	1024fs	7.35kHz ~ 13kHz	
2	Don't care	1	0	384fs	7.35kHz ~ 48kHz	
3	Don't care	1	1	512fs	7.35kHz ~ 26kHz	

Table 5. MCKI Frequency at EXT Slave Mode (M/S bit = "0")

The S/N of the DAC at low sampling frequencies is worse than at high sampling frequencies due to out-of-band noise. The out-of-band noise can be improved by using higher frequency of the master clock. The S/N of the DAC output through LOUT/ROUT pins at fs=8kHz is shown in Table 6.

MCKI	S/N (fs=8kHz, 20kHzLPF + A-weighted)
256fs	83dB
384fs	83dB
512fs	93dB
1024fs	93dB

Table 6. Relationship between MCKI and S/N of LOUT/ROUT pins

The external clocks (MCKI, BICK and LRCK) should always be present whenever the ADC or DAC is in operation (PMADL bit = "1", PMADR bit = "1" or PMDAC bit = "1"). If these clocks are not provided, the AK4645A may draw excess current and it is not possible to operate properly because utilizes dynamic refreshed logic internally. If the external clocks are not present, the ADC and DAC should be in the power-down mode (PMADL=PMADR=PMDAC bits = "0").

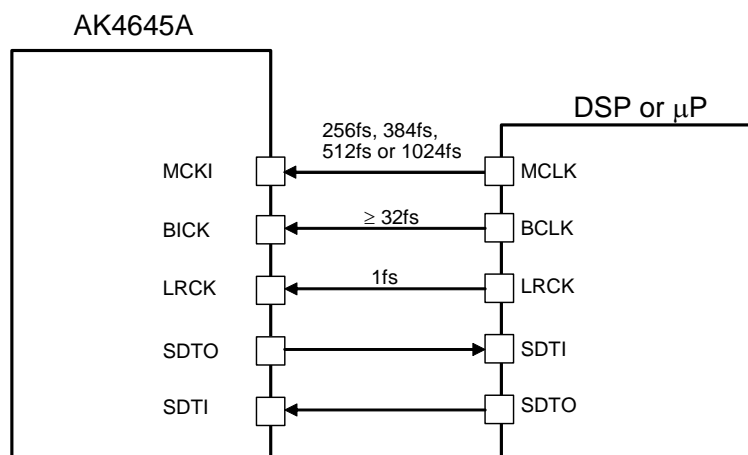


Figure 16. EXT Slave Mode

■ EXT Master Mode (M/S bit = “1”)

The AK4645A becomes EXT Master Mode by setting M/S bit = “1”. Master clock is input from the MCKI pin for ADC and DAC. The clock required to operate is MCKI (256fs, 384fs, 512fs or 1024fs). The input frequency of MCKI is selected by FS1-0 bits (Table 7).

Mode	FS3-2 bits	FS1 bit	FS0 bit	MCKI Input Frequency	Sampling Frequency Range	(default)
0	Don't care	0	0	256fs	7.35kHz ~ 48kHz	
1	Don't care	0	1	1024fs	7.35kHz ~ 13kHz	
2	Don't care	1	0	384fs	7.35kHz ~ 48kHz	
3	Don't care	1	1	512fs	7.35kHz ~ 26kHz	

Table 7. MCKI Frequency at EXT Master Mode (M/S bit = “1”)

The S/N of the DAC at low sampling frequencies is worse than at high sampling frequencies due to out-of-band noise. The out-of-band noise can be improved by using higher frequency of the master clock. The S/N of the DAC output through LOUT/ROUT pins at fs=8kHz is shown in Table 8.

MCKI	S/N (fs=8kHz, 20kHzLPF + A-weighted)
256fs	83dB
384fs	83dB
512fs	93dB
1024fs	93dB

Table 8. Relationship between MCKI and S/N of LOUT/ROUT pins

MCKI should always be present whenever the ADC or DAC is in operation (PMADL bit = “1”, PMADR bit = “1” or PMDAC bit = “1”). If MCKI is not provided, the AK4645A may draw excess current and it is not possible to operate properly because utilizes dynamic refreshed logic internally. If MCKI is not present, the ADC and DAC should be in power-down mode (PMADL=PMADR=PMDAC bits = “0”).

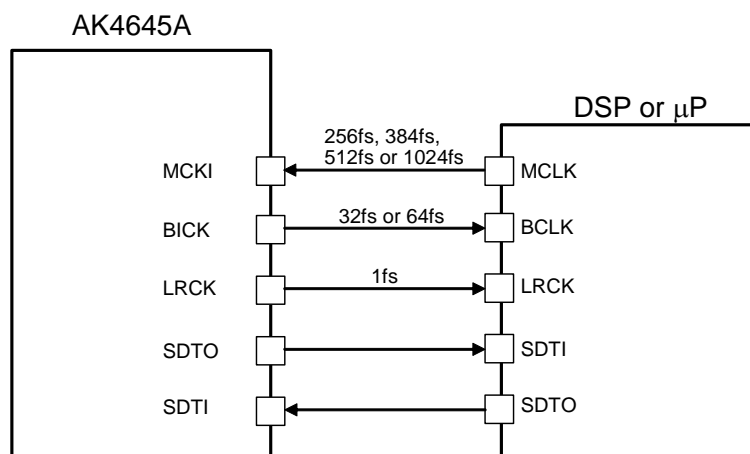


Figure 17. EXT Master Mode

BCKO bit	BICK Output Frequency	(default)
0	32fs	
1	64fs	

Table 9. BICK Output Frequency at Master Mode

■ System Reset

Upon power-up, the AK4645A must be reset by bringing the PDN pin = “L”. This ensures that all internal registers reset to their initial values.

The ADC enters an initialization cycle when the PMADL or PMADR bit is changed from “0” to “1” at PMDAC bits is “0”. The initialization cycle time is $1059/fs=24ms@fs=44.1kHz$. During the initialization cycle, the ADC digital data of both channels is forced to output a 2’s compliment, “0” data. The ADC output reflects the analog input signal after the initialization cycle is completed. When PMDAC bit is “1”, the ADC does not require an initialization cycle.

The DAC enters an initialization cycle when the PMDAC bit is changed from “0” to “1” at PMADL and PMADR bits are “0”. The initialization cycle time is $1059/fs=24ms@fs=44.1kHz$. During the initialization cycle, the DAC input digital data of both channels are internally forced to a 2’s compliment, “0”. The DAC output reflects the digital input data after the initialization cycle is completed. When PMADL or PMADR bit is “1”, the DAC does not require an initialization cycle.

■ Audio Interface Format

Four types of data formats are available and are selected by setting the DIF1-0 bits (Table 10). In all modes, the serial data is MSB first, 2’s complement format. Audio interface formats can be used in both master and slave modes. LRCK and BICK are output from the AK4645A in master mode, but must be input to the AK4645A in slave mode.

Mode	DIF1 bit	DIF0 bit	SDTO (ADC)	SDTI (DAC)	BICK	Figure
0	0	0	DSP Mode	DSP Mode	$\geq 32fs$	Table 11
1	0	1	MSB justified	LSB justified	$\geq 32fs$	Figure 22
2	1	0	MSB justified	MSB justified	$\geq 32fs$	Figure 23
3	1	1	I ² S compatible	I ² S compatible	$\geq 32fs$	Figure 24

(default)

Table 10. Audio Interface Format

In modes 1, 2 and 3, the SDTO is clocked out on the falling edge (“↓”) of BICK and the SDTI is latched on the rising edge (“↑”). In Modes 0 (DSP mode), the audio I/F timing is changed by BCKP and MSBS bits (Table 11).

DIF1	DIF0	MSBS	BCKP	Audio Interface Format	Figure
0	0	0	0	MSB of SDTO is output by the rising edge (“↑”) of the first BICK after the rising edge (“↑”) of LRCK. MSB of SDTI is latched by the falling edge (“↓”) of the BICK just after the output timing of SDTO’s MSB.	Figure 18
		0	1	MSB of SDTO is output by the falling edge (“↓”) of the first BICK after the rising edge (“↑”) of LRCK. MSB of SDTI is latched by the rising edge (“↑”) of the BICK just after the output timing of SDTO’s MSB.	Figure 19
		1	0	MSB of SDTO is output by next rising edge (“↑”) of the falling edge (“↓”) of the first BICK after the rising edge (“↑”) of LRCK. MSB of SDTI is latched by the falling edge (“↓”) of the BICK just after the output timing of SDTO’s MSB.	Figure 20
		1	1	MSB of SDTO is output by next falling edge (“↓”) of the rising edge (“↑”) of the first BICK after the rising edge (“↑”) of LRCK. MSB of SDTI is latched by the rising edge (“↑”) of the BICK just after the output timing of SDTO’s MSB.	Figure 21

(default)

Table 11. Audio Interface Format in Mode 0

If 16-bit data that ADC outputs is converted to 8-bit data by removing LSB 8-bit, “-1” at 16bit data is converted to “-1” at 8-bit data. And when the DAC playbacks this 8-bit data, “-1” at 8-bit data will be converted to “-256” at 16-bit data and this is a large offset. This offset can be removed by adding the offset of “128” to 16-bit data before converting to 8-bit data.

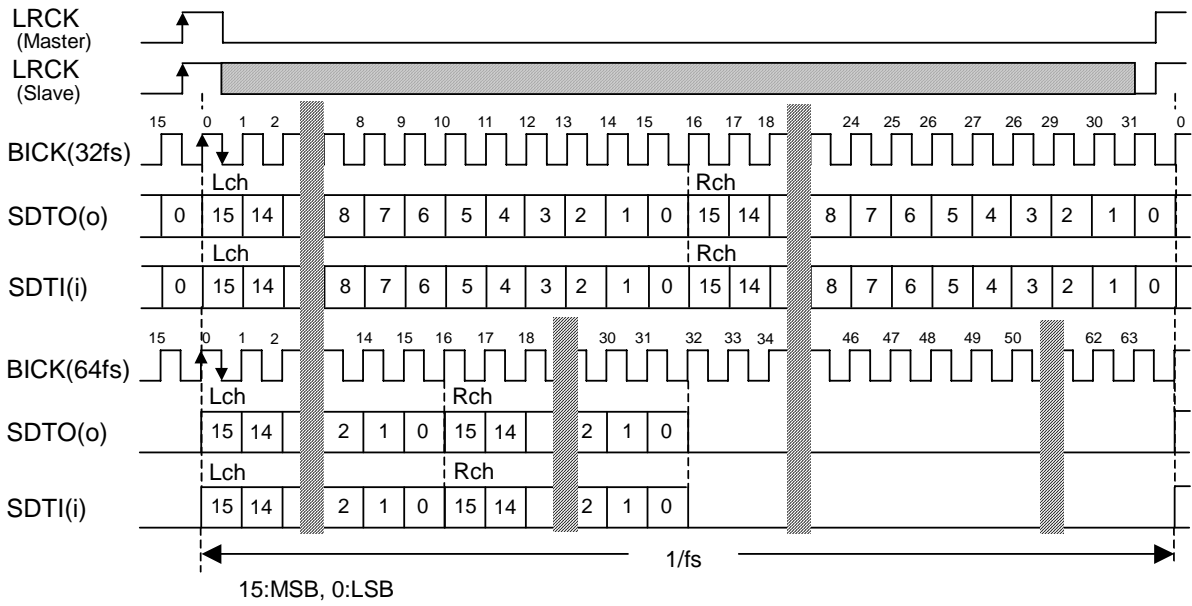


Figure 18. Mode 0 Timing (BCKP = "0", MSBS = "0")

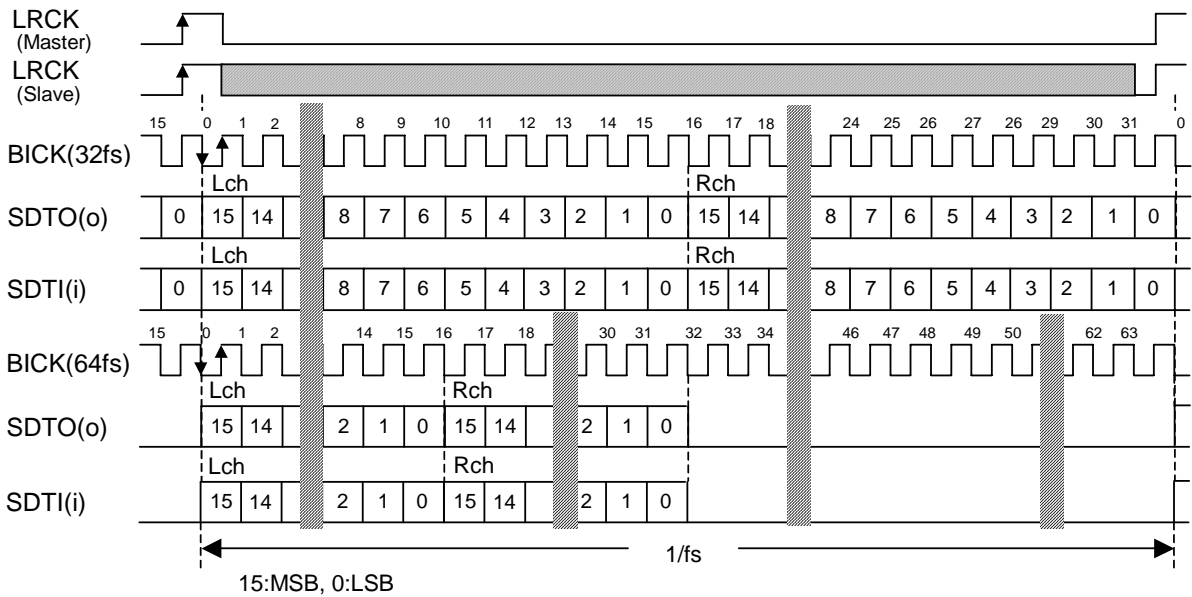


Figure 19. Mode 0 Timing (BCKP = "1", MSBS = "0")

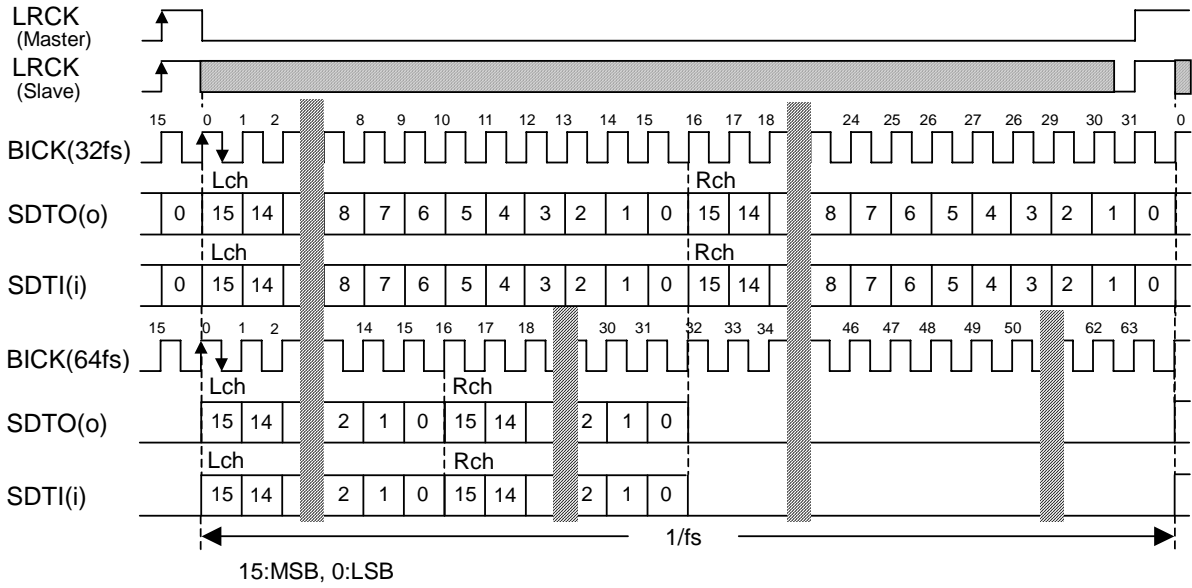


Figure 20. Mode 0 Timing (BCKP = "0", MSBS = "1")

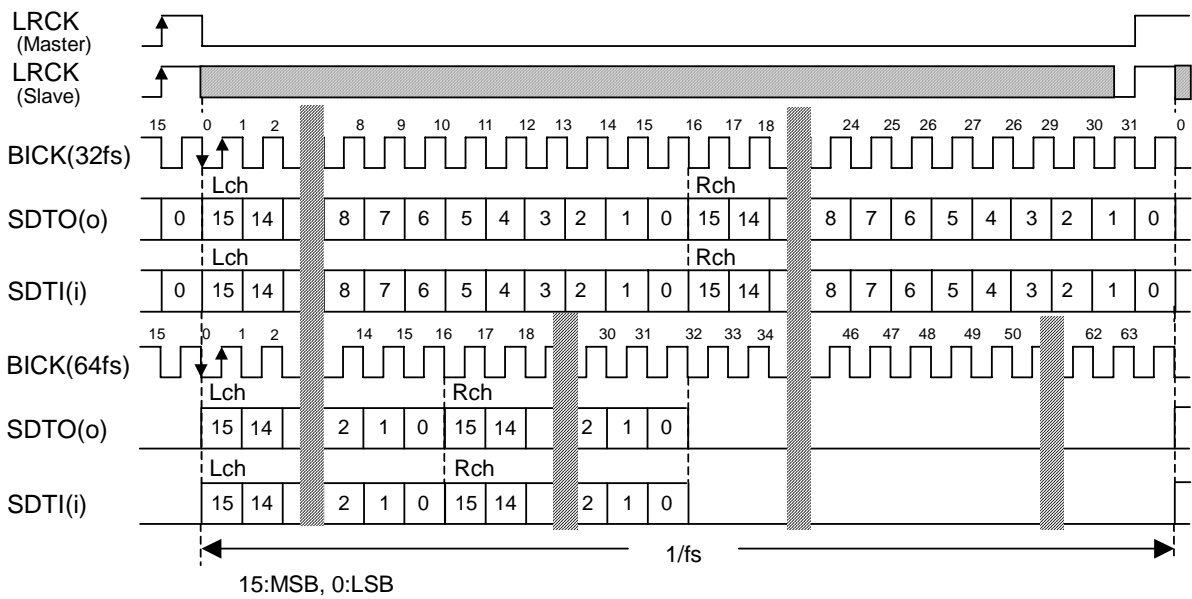


Figure 21. Mode 0 Timing (BCKP = "1", MSBS = "1")

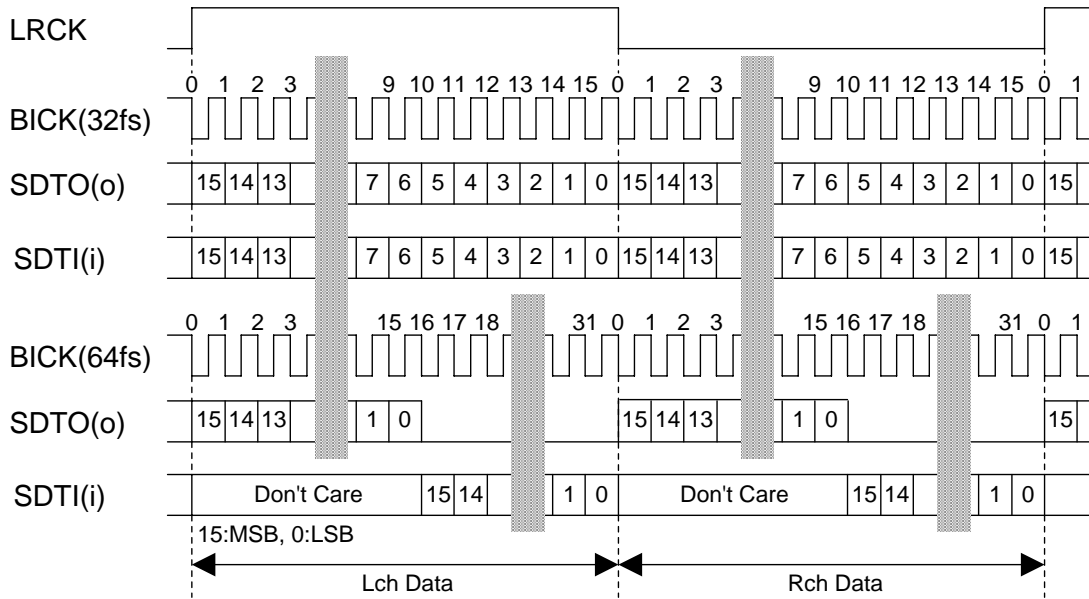


Figure 22. Mode 1 Timing

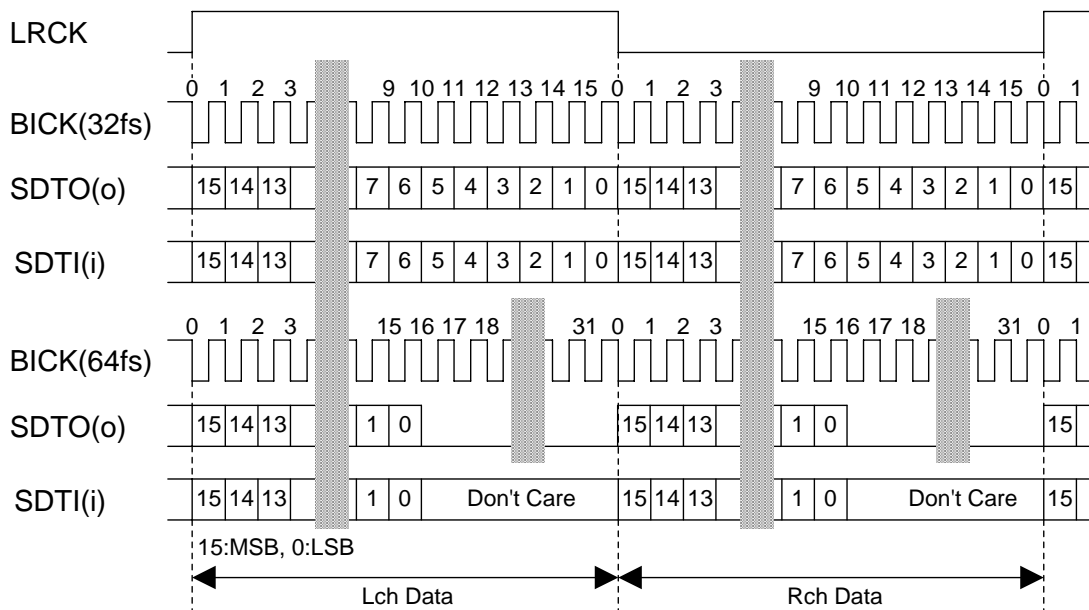


Figure 23. Mode 2 Timing

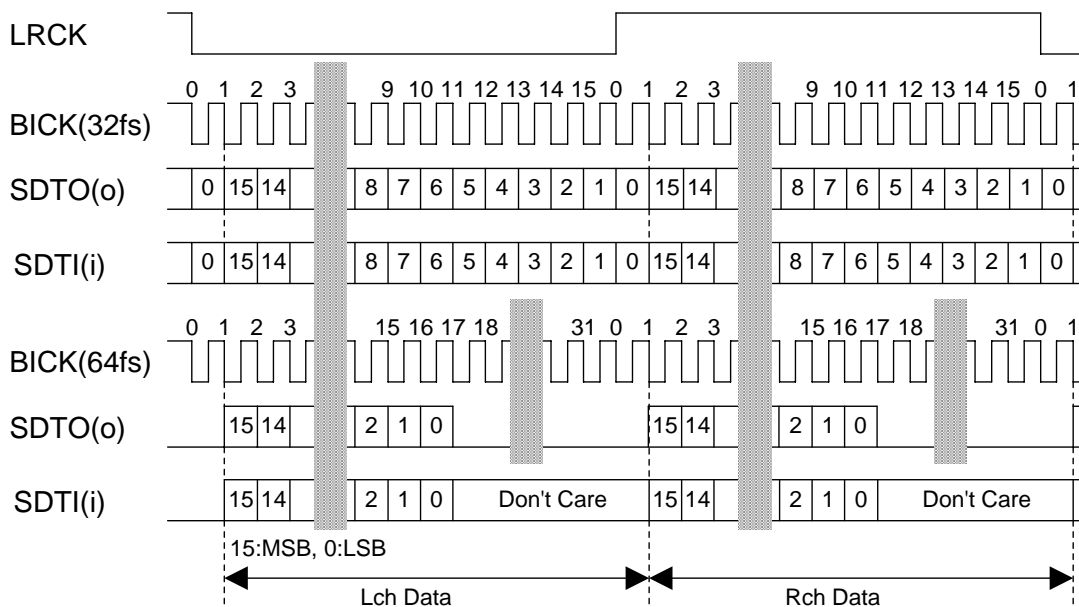


Figure 24. Mode 3 Timing

■ Mono/Stereo Mode

PMADL, PMADR and MIX bits set mono/stereo ADC operation. When MIX bit = "1", EQ and FIL3 bits should be set to "0". ALC operation (ALC bit = "1") or digital volume operation (ALC bit = "0") is applied to the data in [Table 12](#).

PMADL bit	PMADR bit	MIX bit	ADC Lch data	ADC Rch data	
0	0	x	All "0"	All "0"	(default)
0	1	x	Rch Input Signal	Rch Input Signal	
1	0	x	Lch Input Signal	Lch Input Signal	
1	1	0	Lch Input Signal	Rch Input Signal	
		1	(L+R)/2	(L+R)/2	

Table 12. Mono/Stereo ADC operation (x: Don't care)

■ Digital High Pass Filter

The AK4645A has a digital high pass filter for DC offset cancellation. The cut-off frequency of the HPF is 0.9Hz (@fs=44.1kHz) and scales with sampling rate (fs). When PMADL bit = "1" or PMADR bit = "1", the HPF of ADC is enabled but the HPF of DAC is disabled. When PMADL=PMADR bits = "0", PMDAC bit = "1", the HPF of DAC is enabled but the HPF of ADC is disabled.

■ MIC/LINE Input Selector

The AK4645A has input selector for MIC-Amp. When MDIF1 and MDIF2 bits are “0”, INL1-0 and INR1-0 bits select LIN1/LIN2/LIN3/LIN4 and RIN1/RIN2/RIN3/RIN4, respectively. When MDIF1 and MDIF2 bits are “1”, LIN1, RIN1, LIN2 and RIN2 pins become IN1-, IN1+, IN2+ and IN2- pins respectively. In this case, full-differential input is available (Figure 26). When full-differential input is used, the signal must not be input to the pins marked by “X” in Table 14.

MDIF1 bit	MDIF2 bit	INL1 bit	INL0 bit	INR1 bit	INR0 bit	Lch	Rch	
0	0	0	0	0	0	LIN1	RIN1	(default)
0	0	0	0	0	1	LIN1	RIN2	
0	0	0	0	1	0	LIN1	RIN3	
0	0	0	0	1	1	LIN1	RIN4	
0	0	0	1	0	0	LIN2	RIN1	
0	0	0	1	0	1	LIN2	RIN2	
0	0	0	1	1	0	LIN2	RIN3	
0	0	0	1	1	1	LIN2	RIN4	
0	0	1	0	0	0	LIN3	RIN1	
0	0	1	0	0	1	LIN3	RIN2	
0	0	1	0	1	0	LIN3	RIN3	
0	0	1	0	1	1	LIN3	RIN4	
0	0	1	1	0	0	LIN4	RIN1	
0	0	1	1	0	1	LIN4	RIN2	
0	0	1	1	1	0	LIN4	RIN3	
0	0	1	1	1	1	LIN4	RIN4	
0	1	0	0	0	0	LIN1	IN2+/-	
0	1	1	0	0	0	LIN3	IN2+/-	
0	1	1	1	0	0	LIN4	IN2+/-	
1	0	0	0	0	1	IN1+/-	RIN2	
1	0	0	0	1	0	IN1+/-	RIN3	
1	0	0	0	1	1	IN1+/-	RIN4	
1	1	0	0	0	0	IN1+/-	IN2+/-	
Others						N/A	N/A	

Table 13. MIC/Line In Path Select

Register			Pin							
AIN3 bit	MDIF1 bit	MDIF2 bit	LIN1 IN1-	RIN1 IN1+	LIN2 IN2+	RIN2 IN2-	MIN LIN3	RIN3	LIN4 IN4+	RIN4 IN4-
0	0	0	O	O	O	O	O	-	O	O
0	0	1	O	X	O	O	O	-	O	X
0	1	0	O	O	X	O	O	-	X	O
0	1	1	O	O	O	O	O	-	X	X
1	0	0	O	O	O	O	O	O	O	O
1	0	1	O	X	O	O	O	X	O	X
1	1	0	O	O	X	O	X	O	X	O
1	1	1	O	O	O	O	X	X	X	X

Table 14. Handling of MIC/Line Input Pins (“-”: N/A; “X”: Signal should not be input.)

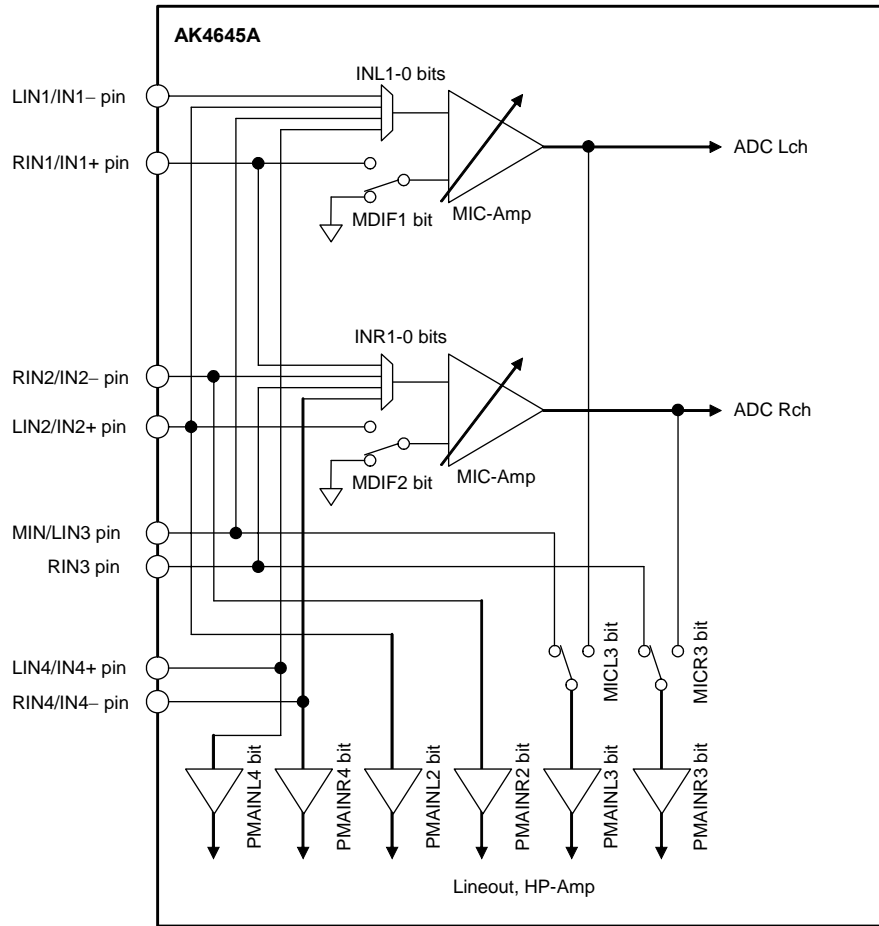


Figure 25. Mic/Line Input Selector

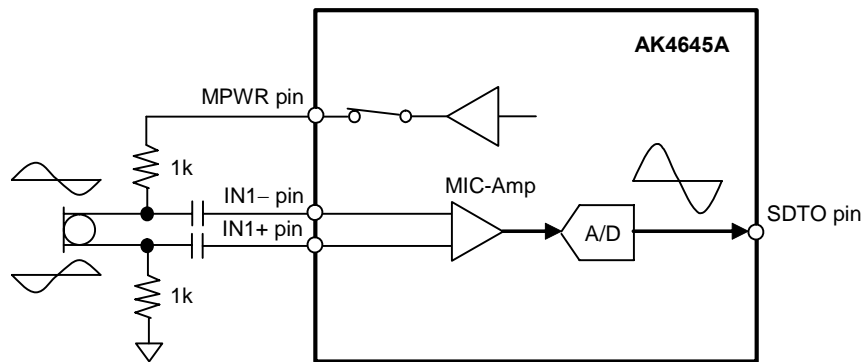


Figure 26. Connection Example for Full-differential Mic Input (MDIF1/2 bits = "1")

<Input Selector Setting Example>

In case that IN1+/- pins are used as full-differential mic input and LIN2/RIN2 pins are used as stereo line input, it is recommended that the following two modes are set by register setting according to each case.

MDIF1 bit	MDIF2 bit	INL1 bit	INL0 bit	INR1 bit	INR0 bit	Lch	Rch
1	0	0	0	0	1	IN1+/-	RIN2
0	0	0	1	0	1	LIN2	RIN2

Table 15. MIC/Line In Path Select Example

■ MIC Gain Amplifier

The AK4645A has a gain amplifier for microphone input. The gain of MIC-Amp is selected by the MGAIN1-0 bits (Table 16). The typical input impedance is $60\text{k}\Omega$ (typ)@MGAIN1-0 bits = “00” or $30\text{k}\Omega$ (typ)@MGAIN1-0 bits = “01”, “10” or “11”.

MGAIN1 bit	MGAIN0 bit	Input Gain
0	0	0dB
0	1	+20dB
1	0	+26dB
1	1	+32dB

(default)

Table 16. Mic Input Gain

■ MIC Power

When PMMP bit = “1”, the MPWR pin supplies power for the microphone. This output voltage is typically $0.75 \times AVDD$ and the load resistance is minimum $0.5\text{k}\Omega$. In case of using two sets of stereo mic, the load resistance is minimum $2\text{k}\Omega$ for each channel. Capacitor must not be connected directly to MPWR pin (Figure 27).

PMMP bit	MPWR pin
0	Hi-Z
1	Output

(default)

Table 17. MIC Power

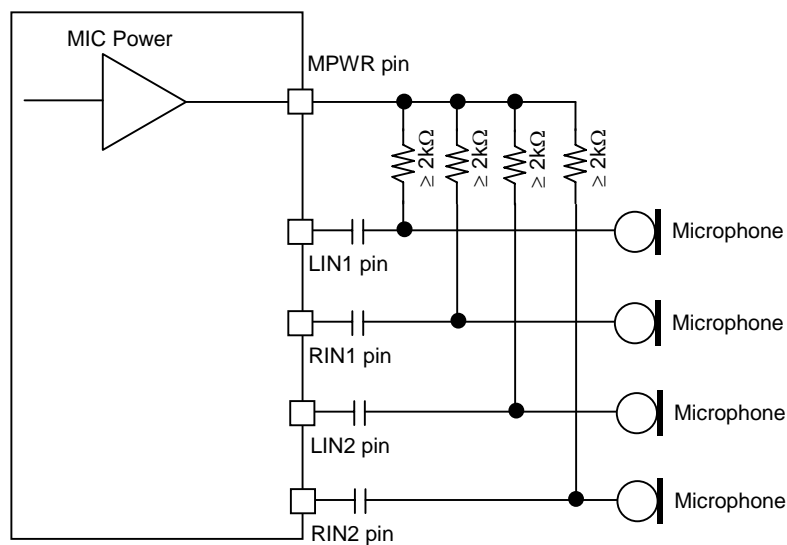


Figure 27. MIC Block Circuit

■ Digital EQ/HPF/LPF

The AK4645A performs wind-noise reduction filter, stereo separation emphasis, gain compensation and ALC (Automatic Level Control) by digital domain for A/D converted data (Figure 28). FIL1, FIL3 and EQ blocks are IIR filters of 1st order. The filter coefficient of FIL3, EQ and FIL1 blocks can be set to any value. Refer to the section of “ALC operation” about ALC.

When only DAC is powered-up, digital EQ/HPF/LPF circuit operates at playback path. When only ADC is powered-up or both ADC and DAC are powered-up, digital EQ/HPF/LPF circuit operates at recording path. Even if the path is switched from recording to playback, the register setting of filter coefficient at recording remains. Therefore, FIL3, EQ, FIL1 and GN1-0 bits should be set to “0” if digital EQ/HPF/LPF is not used for playback path.

PMADL bit, PMADR bit	PMDAC bit	LOOP bit	Status	Digital EQ/HPF/LPF	(default)
“00”	0	x	Power-down	Power-down)
	1	x	Playback	Playback path	
“01”, “10” or “11”	0	x	Recording	Recording path	
	1	0	Recording & Playback	Recording path	
		1	Recording Monitor Playback	Recording path	

Note 38. Stereo separation emphasis circuit is effective only at stereo operation.

Table 18. Digital EQ/HPF/LPF Circuit Setting (x: Don't care)

FIL3 coefficient also sets the attenuation of the stereo separation emphasis.

The combination of GN1-0 bit (Table 19) and EQ coefficient set the compensation gain.

FIL1 and FIL3 blocks become HPF when F1AS and F3AS bits are “0” and become LPF when F1AS and F3AS bits are “1”, respectively.

When EQ and FIL1 bits are “0”, EQ and FIL1 blocks become “through” (0dB). When FIL3 bit is “0”, FIL3 block become “MUTE”. When each filter coefficient is changed, each filter should be set to “through” (“MUTE” in case of FIL3).

When MIX bit = “1”, only FIL1 is available. In this case, EQ and FIL3 bits should be set to “0”.

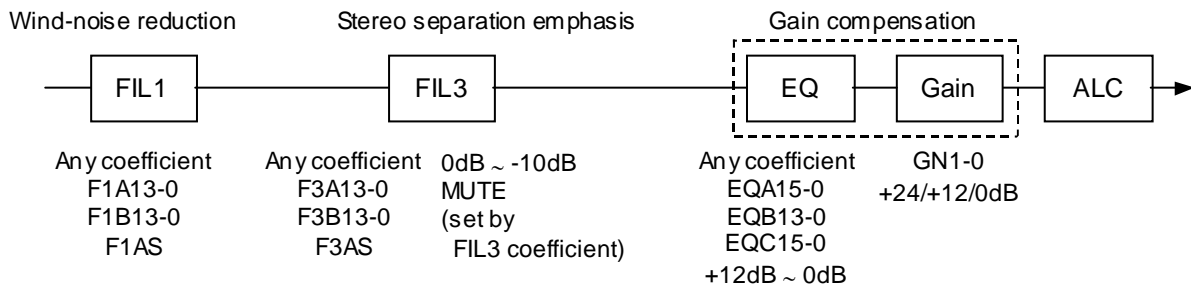


Figure 28. Digital EQ/HPF/LPF

GN1	GN0	Gain	(default)
0	0	0dB	
0	1	+12dB	
1	x	+24dB	

Table 19. Gain select of gain block (x: Don't care)

[Filter Coefficient Setting]

1) When FIL1 and FIL3 are set to “HPF”

fs: Sampling frequency
 fc: Cut-off frequency
 f: Input signal frequency
 K: Filter gain [dB] (Filter gain of should be set to 0dB.)

Register setting

FIL1: F1AS bit = “0”, F1A[13:0] bits =A, F1B[13:0] bits =B
 FIL3: F3AS bit = “0”, F3A[13:0] bits =A, F3B[13:0] bits =B
 (MSB=F1A13, F1B13, F3A13, F3B13; LSB=F1A0, F1B0, F3A0, F3B0)

$$A = 10^{K/20} \times \frac{1 / \tan (\pi f c / f s)}{1 + 1 / \tan (\pi f c / f s)}, \quad B = \frac{1 - 1 / \tan (\pi f c / f s)}{1 + 1 / \tan (\pi f c / f s)}$$

Transfer function	Amplitude	Phase
$H(z) = A \frac{1 - z^{-1}}{1 + Bz^{-1}}$	$M(f) = A \sqrt{\frac{2 - 2\cos (2\pi f / f s)}{1 + B^2 + 2B\cos (2\pi f / f s)}}$	$\theta(f) = \tan^{-1} \frac{(B+1)\sin (2\pi f / f s)}{1 - B + (B-1)\cos (2\pi f / f s)}$

2) When FIL1 and FIL3 are set to “LPF”

fs: Sampling frequency
 fc: Cut-off frequency
 f: Input signal frequency
 K: Filter gain [dB] (Filter gain of FIL1 should be set to 0dB.)

Register setting

FIL1: F1AS bit = “1”, F1A[13:0] bits =A, F1B[13:0] bits =B
 FIL3: F3AS bit = “1”, F3A[13:0] bits =A, F3B[13:0] bits =B
 (MSB=F1A13, F1B13, F3A13, F3B13; LSB=F1A0, F1B0, F3A0, F3B0)

$$A = 10^{K/20} \times \frac{1}{1 + 1 / \tan (\pi f c / f s)}, \quad B = \frac{1 - 1 / \tan (\pi f c / f s)}{1 + 1 / \tan (\pi f c / f s)}$$

Transfer function	Amplitude	Phase
$H(z) = A \frac{1 + z^{-1}}{1 + Bz^{-1}}$	$M(f) = A \sqrt{\frac{2 + 2\cos (2\pi f / f s)}{1 + B^2 + 2B\cos (2\pi f / f s)}}$	$\theta(f) = \tan^{-1} \frac{(B-1)\sin (2\pi f / f s)}{1 + B + (B+1)\cos (2\pi f / f s)}$

3) EQ

fs: Sampling frequency
 fc₁: Pole frequency
 fc₂: Zero-point frequency
 f: Input signal frequency
 K: Filter gain [dB] (Maximum +12dB)

Register setting

EQA[15:0] bits =A, EQB[13:0] bits =B, EQC[15:0] bits =C
 (MSB=EQA15, EQB13, EQC15; LSB=EQA0, EQB0, EQC0)

$$A = 10^{K/20} \times \frac{1 + 1 / \tan(\pi fc_2 / fs)}{1 + 1 / \tan(\pi fc_1 / fs)}, \quad B = \frac{1 - 1 / \tan(\pi fc_1 / fs)}{1 + 1 / \tan(\pi fc_1 / fs)}, \quad C = 10^{K/20} \times \frac{1 - 1 / \tan(\pi fc_2 / fs)}{1 + 1 / \tan(\pi fc_1 / fs)}$$

Transfer function	Amplitude	Phase
$H(z) = \frac{A + Cz^{-1}}{1 + Bz^{-1}}$	$M(f) = \sqrt{\frac{A^2 + C^2 + 2AC \cos(2\pi f / fs)}{1 + B^2 + 2B \cos(2\pi f / fs)}}$	$\theta(f) = \tan^{-1} \frac{(AB - C) \sin(2\pi f / fs)}{A + BC + (AB + C) \cos(2\pi f / fs)}$

[Translation the filter coefficient calculated by the equations above from real number to binary code (2's complement)]

$$X = (\text{Real number of filter coefficient calculated by the equations above}) \times 2^{13}$$

X should be rounded to integer, and then should be translated to binary code (2's complement).

MSB of each filter coefficient setting register is sine bit.

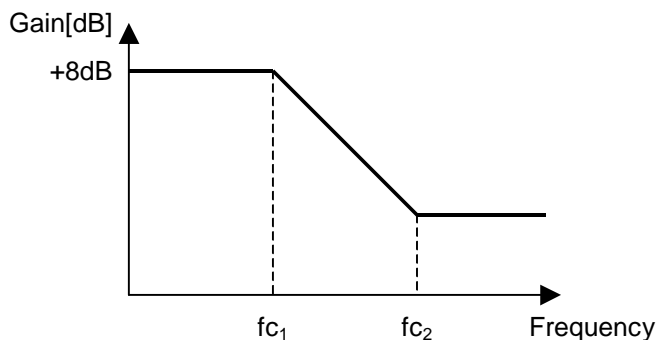
[Filter Coefficient Setting Example]

1) FIL1 block

Example: HPF, fs=44.1kHz, fc=100Hz
 F1AS bit = "0"
 F1A[13:0] bits = 01 1111 1100 0110
 F1B[13:0] bits = 10 0000 0111 0100

2) EQ block

Example: fs=44.1kHz, fc₁=300Hz, fc₂=3000Hz, Gain=+8dB



EQA[15:0] bits = 0000 1001 0110 1110
 EQB[13:0] bits = 10 0001 0101 1001
 EQC[15:0] bits = 1111 1001 1110 1111

■ ALC Operation

The ALC (Automatic Level Control) is operated by ALC block when ALC bit is “1”. When only DAC is powered-up, ALC circuit operates at playback path. When only ADC is powered-up or both ADC and DAC are powered-up, ALC circuit operates at recording path.

PMADL bit, PMADR bit	PMDAC bit	LOOP bit	Status	Digital EQ/HPF/LPF	(default)
“00”	0	x	Power-down	Power-down	
	1	x	Playback	Playback path	
“01”, “10” or “11”	0	x	Recording	Recording path	
	1	0	Recording & Playback	Recording path	
		1	Recording Monitor Playback	Recording path	

Table 20. ALC Setting (x: Don’t care)

1. ALC Limiter Operation

During the ALC limiter operation, when either Lch or Rch exceeds the ALC limiter detection level (Table 21), the IVL and IVR values (same value) are attenuated automatically by the amount defined by the ALC limiter ATT step (Table 22). The IVL and IVR are then set to the same value for both channels.

When ZELMN bit = “0” (zero cross detection is enabled), the IVL and IVR values are changed by ALC limiter operation at the individual zero crossing points of Lch and Rch or at the zero crossing timeout. ZTM1-0 bits set the zero crossing timeout period of both ALC limiter and recovery operation (Table 23).

When ZELMN bit = “1” (zero cross detection is disabled), IVL and IVR values are immediately (period: 1/fs) changed by ALC limiter operation. Attenuation step is fixed to 1 step regardless of the setting of LMAT1-0 bits.

The attenuation operation is executed continuously until the input signal level becomes ALC limiter detection level (Table 21) or less. After completing the attenuate operation, unless ALC bit is changed to “0”, the operation repeats when the input signal level exceeds LMTH1-0 bits.

LMTH1	LMTH0	ALC Limier Detection Level	ALC Recovery Waiting Counter Reset Level	(default)
0	0	ALC Output \geq -2.5dBFS	-2.5dBFS > ALC Output \geq -4.1dBFS	
0	1	ALC Output \geq -4.1dBFS	-4.1dBFS > ALC Output \geq -6.0dBFS	
1	0	ALC Output \geq -6.0dBFS	-6.0dBFS > ALC Output \geq -8.5dBFS	
1	1	ALC Output \geq -8.5dBFS	-8.5dBFS > ALC Output \geq -12dBFS	

Table 21. ALC Limiter Detection Level / Recovery Counter Reset Level

ZELMN	LMAT1	LMAT0	ALC Limiter ATT Step		(default)
0	0	0	1 step	0.375dB	
	0	1	2 step	0.750dB	
	1	0	4 step	1.500dB	
	1	1	8 step	3.000dB	
1	x	x	1step	0.375dB	

Table 22. ALC Limiter ATT Step (x: Don’t care)

ZTM1	ZTM0	Zero Crossing Timeout Period				(default)
			8kHz	16kHz	44.1kHz	
0	0	128/fs	16ms	8ms	2.9ms	
0	1	256/fs	32ms	16ms	5.8ms	
1	0	512/fs	64ms	32ms	11.6ms	
1	1	1024/fs	128ms	64ms	23.2ms	

Table 23. ALC Zero Crossing Timeout Period

2. ALC Recovery Operation

The ALC recovery operation waits for the WTM2-0 bits (Table 24) to be set after completing the ALC limiter operation. If the input signal does not exceed “ALC recovery waiting counter reset level” (Table 21) during the wait time, the ALC recovery operation is executed. The IVL and IVR values are automatically incremented by RGAIN1-0 bits (Table 25) up to the set reference level (Table 26) with zero crossing detection which timeout period is set by ZTM1-0 bits (Table 23). Then the IVL and IVR are set to the same value for both channels. The ALC recovery operation is executed at a period set by WTM2-0 bits. When zero cross is detected at both channels during the wait period set by WTM2-0 bits, the ALC recovery operation waits until WTM2-0 period and the next recovery operation is executed. If ZTM1-0 is longer than WTM2-0 and no zero crossing occurs, the ALC recovery operation is executed at a period set by ZTM1-0 bits.

For example, when the current IVOL value is 30H and RGAIN1-0 bits are set to “01”, IVOL is changed to 32H by the auto limiter operation and then the input signal level is gained by 0.75dB (=0.375dB x 2). When the IVOL value exceeds the reference level (REF7-0), the IVOL values are not increased.

When

“ALC recovery waiting counter reset level (LMTH1-0) ≤ Output Signal < ALC limiter detection level (LMTH1-0)” during the ALC recovery operation, the waiting timer of ALC recovery operation is reset. When

“ALC recovery waiting counter reset level (LMTH1-0) > Output Signal”, the waiting timer of ALC recovery operation starts.

The ALC operation corresponds to the impulse noise. When the impulse noise is input, the ALC recovery operation becomes faster than a normal recovery operation (Fast Recovery Operation). When large noise is input to microphone instantaneously, the quality of small level in the large noise can be improved by this fast recovery operation. The speed of fast recovery operation is set by RFST1-0 bits (Table 27).

WTM2	WTM1	WTM0	ALC Recovery Operation Waiting Period			(default)
				8kHz	16kHz	
0	0	0	128/fs	16ms	8ms	2.9ms
0	0	1	256/fs	32ms	16ms	5.8ms
0	1	0	512/fs	64ms	32ms	11.6ms
0	1	1	1024/fs	128ms	64ms	23.2ms
1	0	0	2048/fs	256ms	128ms	46.4ms
1	0	1	4096/fs	512ms	256ms	92.9ms
1	1	0	8192/fs	1024ms	512ms	185.8ms
1	1	1	16384/fs	2048ms	1024ms	371.5ms

Table 24. ALC Recovery Operation Waiting Period

RGAIN1	RGAIN0	GAIN STEP		(default)
0	0	1 step	0.375dB	
0	1	2 step	0.750dB	
1	0	3 step	1.125dB	
1	1	4 step	1.500dB	

Table 25. ALC Recovery GAIN Step

REF7-0	GAIN(dB)	Step
F1H	+36.0	0.375dB (default)
F0H	+35.625	
EFH	+35.25	
:	:	
E2H	+30.375	
E1H	+30.0	
E0H	+29.625	
:	:	
03H	-53.25	
02H	-53.625	
01H	-54.0	
00H	MUTE	

Table 26. Reference Level at ALC Recovery operation

RFST1 bit	RFST0 bit	Recovery Speed
0	0	4 times (default)
0	1	8 times
1	0	16times
1	1	N/A

Table 27. Fast Recovery Speed Setting

3. Example of ALC Operation

Table 28 shows the examples of the ALC setting for mic recording.

Register Name	Comment	fs=8kHz		fs=44.1kHz	
		Data	Operation	Data	Operation
LMTH1-0	Limiter detection Level	01	-4.1dBFS	01	-4.1dBFS
ZELMN	Limiter zero crossing detection	0	Enable	0	Enable
ZTM1-0	Zero crossing timeout period	01	32ms	11	23.2ms
WTM2-0	Recovery waiting period *WTM2-0 bits should be the same or longer data as ZTM1-0 bits.	001	32ms	011	23.2ms
REF7-0	Maximum gain at recovery operation	E1H	+30dB	E1H	+30dB
IVL7-0, IVR7-0	Gain of IVOL	E1H	+30dB	E1H	+30dB
LMAT1-0	Limiter ATT step	00	1 step	00	1 step
RGAIN1-0	Recovery GAIN step	00	1 step	00	1 step
RFST1-0	Fast Recovery Speed	00	4 times	00	4 times
ALC	ALC enable	1	Enable	1	Enable

Table 28. Example of the ALC setting

The following registers should not be changed during the ALC operation. These bits should be changed after the ALC operation is finished by ALC bit = "0" or PMADL=PMADR bits = "0".

- LMTH1-0, LMAT1-0, WTM2-0, ZTM1-0, RGAIN1-0, REF7-0, ZELMN, RFST1-0

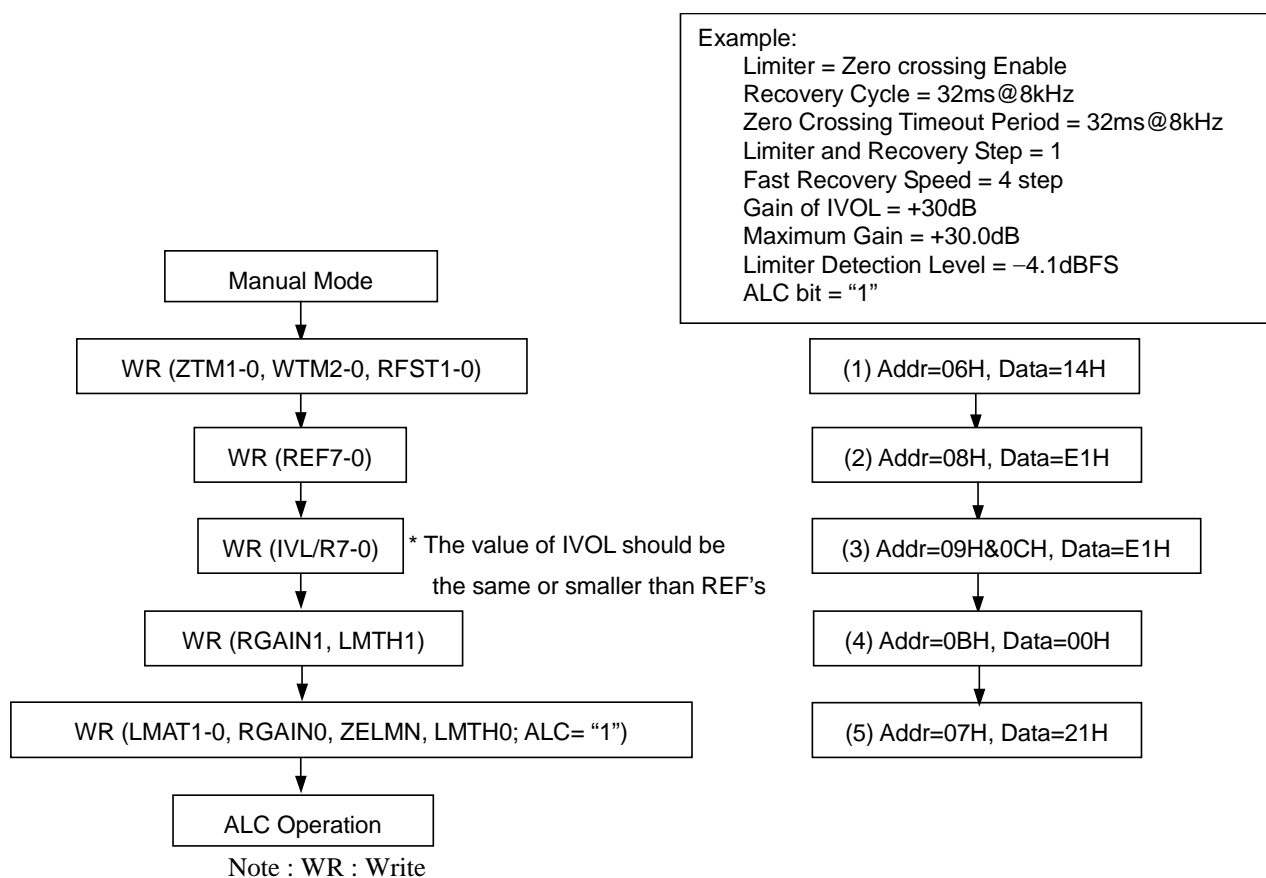


Figure 29. Registers set-up sequence at ALC operation

■ Input Digital Volume (Manual Mode)

The input digital volume becomes manual mode when ALC bit is “0”. This mode is used in the case shown below.

1. After exiting reset state, set-up the registers for the ALC operation (ZTM1-0, LMTH1-0 and etc)
2. When the registers for the ALC operation (Limiter period, Recovery period and etc) are changed.
For example; when the change of the sampling frequency.
3. When IVOL is used as a manual volume.

IVL7-0 and IVR7-0 bits set the gain of the volume control (Table 29). The IVOL value is changed at zero crossing or timeout. Zero crossing timeout period is set by ZTM1-0 bits. If IVL7-0 or IVR7-0 bits are written during PMADL=PMADR bits = “0”, IVOL operation starts with the written values at the end of the ADC initialization cycle after PMADL or PMADR bit is changed to “1”.

Even if the path is switched from recording to playback, the register setting of IVOL remains. Therefore, IVL7-0 and IVR7-0 bits should be set to “91H” (0dB).

IVL7-0 IVR7-0	GAIN (dB)	Step
F1H	+36.0	0.375dB (default)
F0H	+35.625	
EFH	+35.25	
:	:	
E2H	+30.375	
E1H	+30.0	
E0H	+29.625	
:	:	
03H	-53.25	
02H	-53.625	
01H	-54	
00H	MUTE	

Table 29. Input Digital Volume Setting

When writing to the IVL7-0 and IVR7-0 bits continuously, the control register should be written by an interval more than zero crossing timeout. If not, IVL and IVR are not changed since zero crossing counter is reset at every write operation. If the same register value as the previous write operation is written to IVL and IVR, zero crossing counter is not reset. Therefore, IVL and IVR can be written by an interval less than zero crossing timeout.


ALC bit			
ALC Status	Disable	Enable	Disable
IVL7-0 bits	E1H(+30dB)		
IVR7-0 bits	C6H(+20dB)		
Internal IVL	E1H(+30dB)	E1(+30dB) --> F1(+36dB)	E1(+30dB)
Internal IVR	C6H(+20dB)	E1(+30dB) --> F1(+36dB)	C6H(+20dB)

Figure 30. IVOL value during ALC operation

- (1) The IVL value becomes the start value if the IVL and IVR are different when the ALC starts. The wait time from ALC bit = "1" to ALC operation start by IVL7-0 bits is at most recovery time (WTM2-0 bits) plus zerocross timeout period (ZTM1-0 bits).
- (2) Writing to IVL and IVR registers (09H and 0CH) is ignored during ALC operation. After ALC is disabled, the IVOL changes to the last written data by zero crossing or timeout. When ALC is enabled again, ALC bit should be set to "1" by an interval more than zero crossing timeout period after ALC bit = "0".

■ De-emphasis Filter

The AK4645A includes the digital de-emphasis filter ($t_c = 50/15\mu s$) by IIR filter. Setting the DEM1-0 bits enables the de-emphasis filter (Table 30).

DEM1	DEM0	Mode
0	0	44.1kHz
0	1	OFF
1	0	48kHz
1	1	32kHz

(default)

Table 30. De-emphasis Control

■ Bass Boost Function

The BST1-0 bits control the amount of low frequency boost applied to the DAC output signal (Table 31). If the BST1-0 bits are set to “01” (MIN Level), use a $47\mu F$ capacitor for AC-coupling. If the boosted signal exceeds full scale, the analog output clips to the full scale. Figure 31 shows the boost frequency response at $-20dB$ signal input.

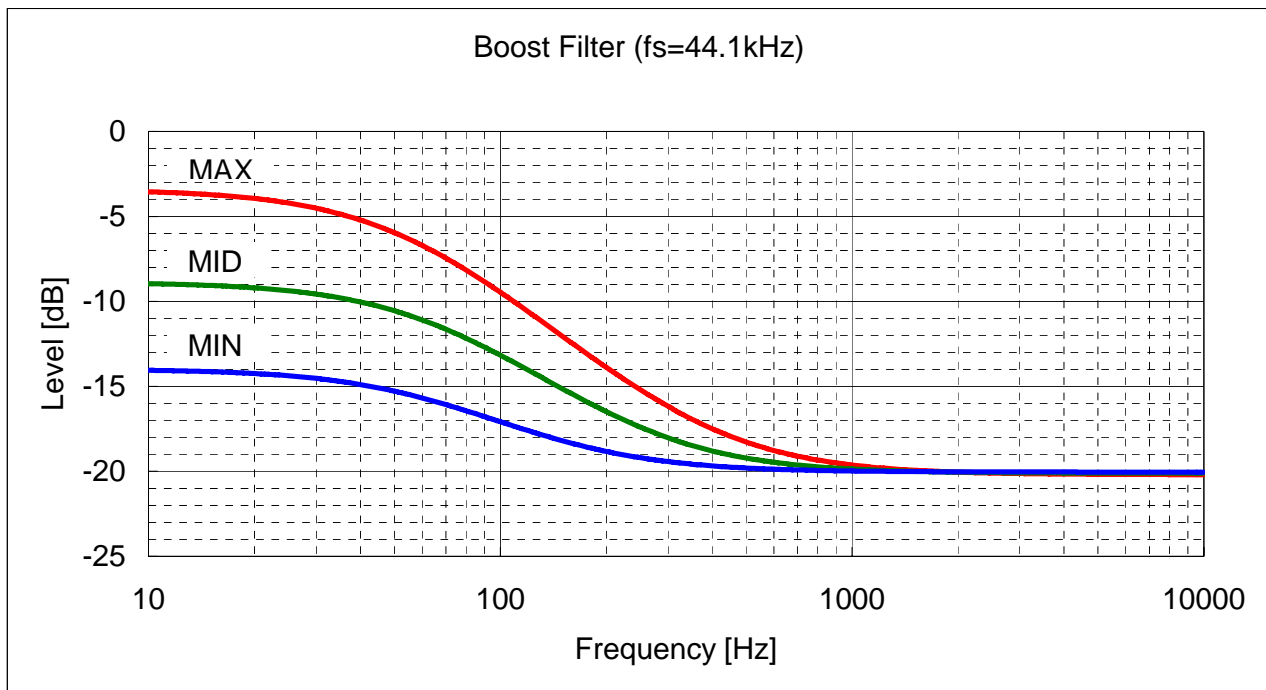


Figure 31. Bass Boost Frequency Response ($f_s=44.1kHz$)

BST1	BST0	Mode
0	0	OFF
0	1	MIN
1	0	MID
1	1	MAX

(default)

Table 31. Bass Boost Control

■ Digital Output Volume

The AK4645A has a digital output volume (256 levels, 0.5dB step, Mute). The volume can be set by the DVL7-0 and DVR7-0 bits. The volume is included in front of the DAC block. The input data of DAC is changed from +12 to -115dB or MUTE. When the DVOLC bit = "1", the DVL7-0 bits control both Lch and Rch attenuation levels. When the DVOLC bit = "0", the DVL7-0 bits control Lch level and DVR7-0 bits control Rch level. This volume has a soft transition function. The DVTM bit sets the transition time between set values of DVL/R7-0 bits as either 1061/fs or 256/fs (Table 33). When DVTM bit = "0", a soft transition between the set values occurs (1062 levels). It takes 1061/fs (=24ms@fs=44.1kHz) from 00H (+12dB) to FFH (MUTE).

DVL/R7-0	Gain	Step
00H	+12.0dB	0.5dB (default)
01H	+11.5dB	
02H	+11.0dB	
:	:	
18H	0dB	
:	:	
FDH	-114.5dB	
FEH	-115.0dB	
FFH	MUTE ($-\infty$)	

Table 32. Digital Volume Code Table

DVTM bit	Transition time between DVL/R7-0 bits = 00H and FFH		
	Setting	fs=8kHz	fs=44.1kHz
0	1061/fs	133ms	24ms
1	256/fs	32ms	6ms

Table 33. Transition Time Setting of Digital Output Volume

■ Soft Mute

Soft mute operation is performed in the digital domain. When the SMUTE bit goes to “1”, the output signal is attenuated by $-\infty$ (“0”) during the cycle set by the DVTM bit. When the SMUTE bit is returned to “0”, the mute is cancelled and the output attenuation gradually changes to the value set by the DVL/R7-0 bits during the cycle set of the DVTM bit. If the soft mute is cancelled within the cycle set by the DVTM bit after starting the operation, the attenuation is discontinued and returned to the value set by the DVL/R7-0 bits. The soft mute is effective for changing the signal source without stopping the signal transmission (Figure 32).

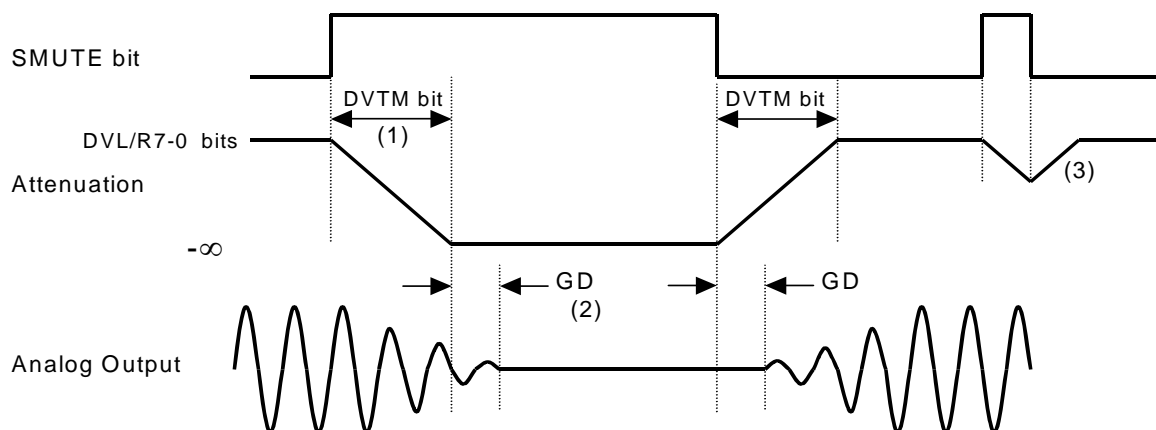


Figure 32. Soft Mute Function

- (1) The output signal is attenuated until $-\infty$ (“0”) by the cycle set by the DVTM bit.
- (2) Analog output corresponding to digital input has group delay (GD).
- (3) If the soft mute is cancelled within the cycle set by the DVTM bit, the attenuation is discontinued and returned to the value set by the DVL/R7-0 bits.

■ Analog Mixing: Stereo Input (LIN2/RIN2/LIN4/RIN4, AIN3 bit = “1”: LIN3/RIN3 pins)

When PMAINL2=PMAINR2 bits = “1”, LIN2 and RIN2 pins can be used as stereo line input for analog mixing. When LINH2 and RINH2 bits are set to “1”, the input signal from the LIN2/RIN2 pins is output to Headphone-Amp. When LINL2/RINR2 bits are set to “1”, the input signal from the LIN2/RIN2 pins is output to the stereo line output amplifier.

When PMAINL4=PMAINR4 bits = “1”, LIN4 and RIN4 pins can be used as stereo line input for analog mixing. When LINH4 and RINH4 bits are set to “1”, the input signal from the LIN4/RIN4 pins is output to Headphone-Amp. When LINL4/RINR4 bits are set to “1”, the input signal from the LIN4/RIN4 pins is output to the stereo line output amplifier.

When the analog mixing is used, A/D converter is also available if PMADL or PMADR bit is “1”. In this case, the input resistance of LIN2/RIN2/LIN4/RIN4 pins becomes 30kΩ (typ) at MGAIN1-0 bits = “00” and 20kΩ (typ) at MGAIN1-0 bits = “01”, “10” or “11”, respectively.

When AIN3 bit = “1”, MIN and VCOC pins becomes LIN3 and RIN3 pins, respectively. When PMAINL3=PMAINR3 bits = “1”, LIN3 and RIN3 pins can be used as stereo line input for analog mixing. When PMMICL=PMMICR=MICL3=MICR3 bits = “1”, analog mixing source is changed from LIN3/RIN3 input to MIC-Amp output signal. When the LINH3 and RINH3 bits are set to “1”, the input signal from the LIN3/RIN3 pins is output to Headphone-Amp. When the LINL3/RINR3 bits are set to “1”, the input signal from the LIN3/RIN3 pins is output to the stereo line output amplifier.

When the analog mixing is used, A/D converter is also available if PMADL or PMADR bit is “1”. When the analog mixing is used at MICL3=MICR3 bits = “0”, the input resistance of LIN3/RIN3 pins becomes 30kΩ (typ) at MGAIN1-0 bits = “00” and 20kΩ (typ) at MGAIN1-0 bits = “01”, “10” or “11”, respectively. When the analog mixing is used at MICL3=MICR3 bits = “1”, the input resistance of LIN3/RIN3 pins becomes 60kΩ (typ) at MGAIN1-0 bits = “00” and 30kΩ (typ) at MGAIN1-0 bits = “01”, “10” or “11”, respectively.

Table 34, Table 35 and Table 36 show the typical gain.

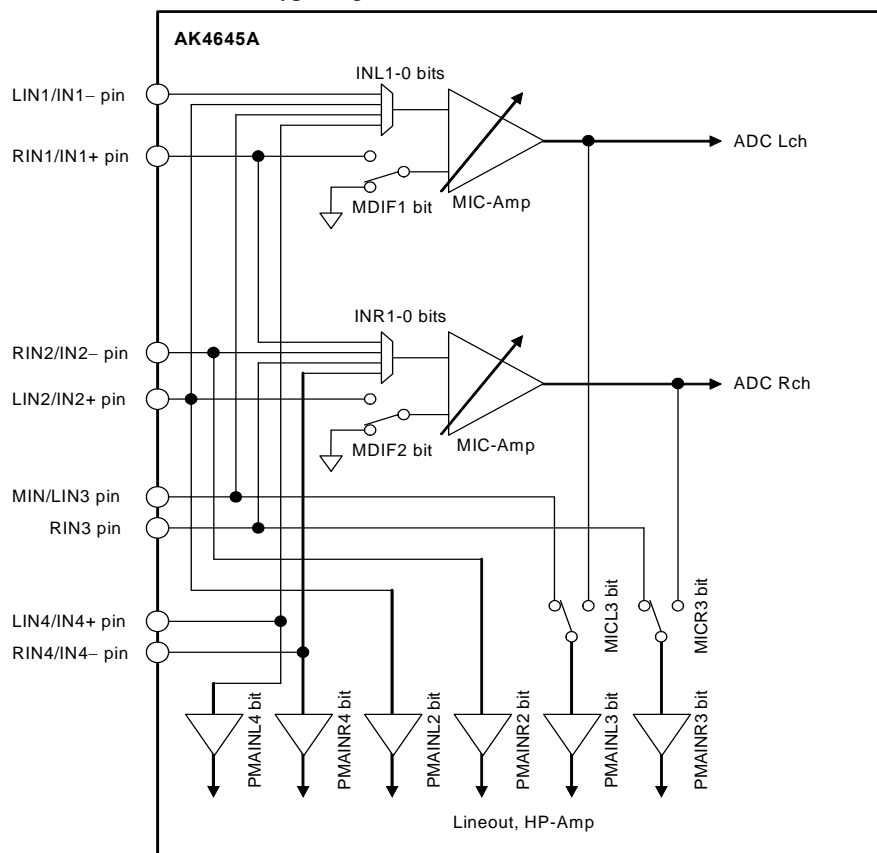


Figure 33. Analog Mixing Circuit (Stereo Input)

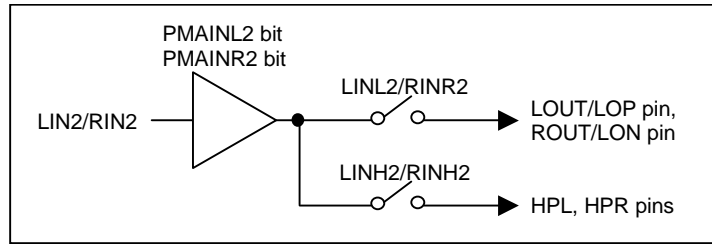


Figure 34. Analog Mixing Circuit (LIN2/RIN2)

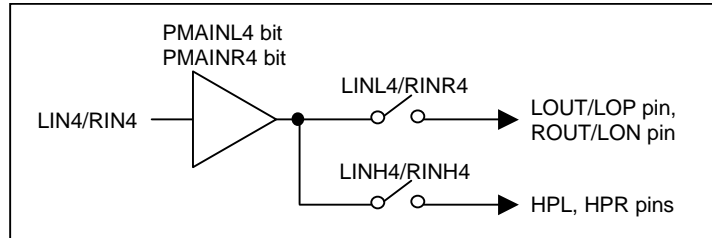


Figure 35. Analog Mixing Circuit (LIN4/RIN4)

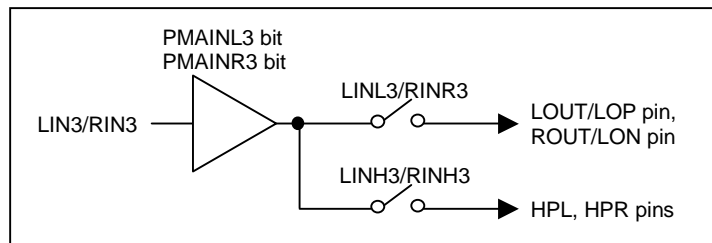


Figure 36. Analog Mixing Circuit (LIN3/RIN3)

LOVL bit	LIN2/RIN2/LIN3/RIN3/LIN4/RIN4 → LOUT/ROUT	(default)
0	0dB	
1	+2dB	

Table 34. LIN2/RIN2/LIN3/RIN3/LIN4/RIN4 Input → LOUT/ROUT Output Gain (typ)

LOVL bit	LIN2/RIN2/LIN3/RIN3/LIN4/RIN4 → LOP/LON	(default)
0	0dB	
1	+2dB	

Table 35. LIN2/RIN2/LIN3/RIN3/LIN4/RIN4 Input → LOP/LON Output Gain (typ)

HPG bit	LIN2/RIN2/LIN3/RIN3/LIN4/RIN4 → HPL/HPR	(default)
0	0dB	
1	+3.6dB	

Table 36. LIN2/RIN2/LIN3/RIN3/LIN4/RIN4 Input → Headphone-Amp Output Gain (typ)

■ Analog Mixing: Full-differential Mono Input (L4DIF bit = “1”: IN4+/IN4– pins)

When L4DIF bit = “1”, LIN4 and RIN4 pins becomes IN4+ and IN4– pins, respectively.

When PMAINL4 bit = “1”, IN4+ and IN4– pins can be used as full-differential mono line input for analog mixing. When the LINH4 and RINH4 bits are set to “1”, the input signal from the IN4+/IN4– pins is output to Headphone-Amp. When the LINL4/RINR4 bits are set to “1”, the input signal from the IN4+/IN4– pins is output to the stereo line output amplifier.

Table 37, Table 38 and Table 39 show the typical gain. Input signal amplitude is defined as (IN4+) – (IN4–).

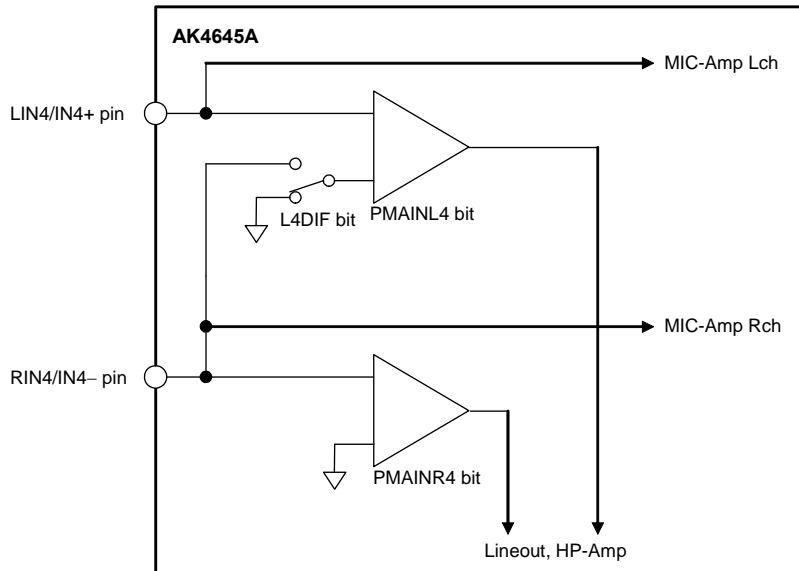


Figure 37. Full-differential Mono Analog Mixing Circuit

LOVL bit	IN4+/IN4– → LOUT/ROUT	(default)
0	–6dB	
1	–4dB	

Table 37. IN4+/IN4– Input → LOUT/ROUT Output Gain (typ)

LOVL bit	IN4+/IN4– → LOP/LON	(default)
0	0dB	
1	+2dB	

Table 38. IN4+/IN4– Input → LOP/LON Output Gain (typ)

HPG bit	IN4+/IN4– → HPL/HPR	(default)
0	–6dB	
1	–2.4dB	

Table 39. IN4+/IN4– Input → Headphone-Amp Output Gain (typ)

■ Analog Mixing: Mono Input (AIN3 bit = "0": MIN pin)

When AIN3 bit = "0", the MIN pin is used as mono input for analog mixing. When the PMMIN bit is set to "1", the mono input is powered-up. When the MINH bit is set to "1", the input signal from the MIN pin is output to Headphone-Amp. When the MINL bit is set to "1", the input signal from the MIN pin is output to the stereo line output amplifier. The external resistor R_i adjusts the signal level of MIN input. Table 40, Table 41 and Table 42 show the typical gain example at $R_i = 20k\Omega$. This gain is in inverse proportion to R_i .

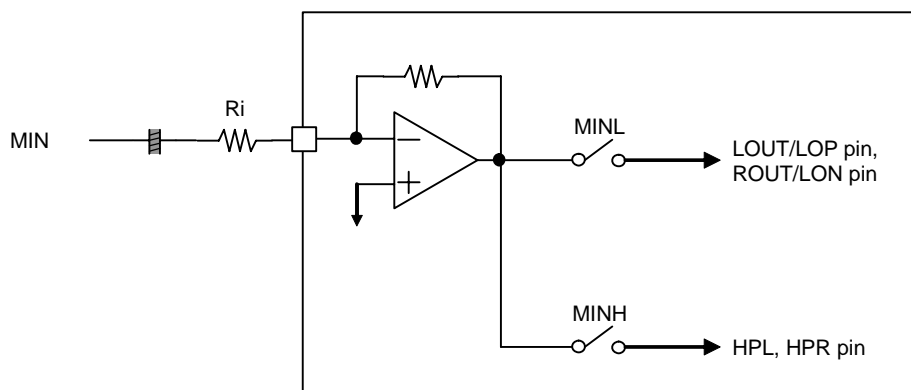


Figure 38. Block Diagram of MIN pin

LOVL bit	MIN → LOUT/ROUT	(default)
0	0dB	
1	+2dB	

Table 40. MIN Input → LOUT/ROUT Output Gain (typ) at $R_i = 20k\Omega$

LOVL bit	MIN → LOP/LON	(default)
0	+6dB	
1	+8dB	

Table 41. MIN Input → LOP/LON Output Gain (typ) at $R_i = 20k\Omega$

HPG bit	MIN → HPL/HPR	(default)
0	-20dB	
1	-16.4dB	

Table 42. MIN Input → Headphone-Amp Output Gain (typ) at $R_i = 20k\Omega$

■ Stereo Line Output (LOUT/ROUT pins)

When DACL bit is “1”, Lch/Rch signal of DAC is output from the LOUT/ROUT pins which is single-ended. When DACL bit is “0”, output signal is muted and LOUT/ROUT pins output VCOM voltage. The load impedance is 10kΩ (min.). When the PMLO=LOPS bits = “0”, the stereo line output enters power-down mode and the output is pulled-down to AVSS by 100kΩ(typ). When the LOPS bit is “1”, stereo line output enters power-save mode. Pop noise at power-up/down can be reduced by changing PMLO bit at LOPS bit = “1”. In this case, output signal line should be pulled-down to AVSS by 20kΩ after AC coupled as [Figure 40](#). Rise/Fall time is 300ms(max) at C=1μF and AVDD=3.3V. When PMLO bit = “1” and LOPS bit = “0”, stereo line output is in normal operation.

LOVL bit set the gain of stereo line output.

When LOM bit = “1”, DAC output signal is output to LOUT and ROUT pins as (L+R)/2 mono signal.

When LOM3 bit = “1”, the signal selected by MICL3 and MICR3 bits (LIN3/RIN3 inputs or MIC-Amp outputs) to LOUT and ROUT pins as (L+R)/2 mono signal.

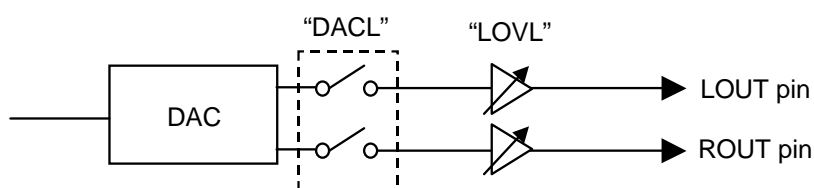


Figure 39. Stereo Line Output

LOPS	PMLO	Mode	LOUT/ROUT pin
0	0	Power-down	Pull-down to AVSS
	1	Normal Operation	Normal Operation
1	0	Power-save	Fall down to AVSS
	1	Power-save	Rise up to VCOM

(default)

Table 43. Stereo Line Output Mode Select (x: Don't care)

LOVL	Gain	Output Voltage (typ)
0	0dB	0.6 x AVDD
1	+2dB	0.757 x AVDD

(default)

Table 44. Stereo Line Output Volume Setting

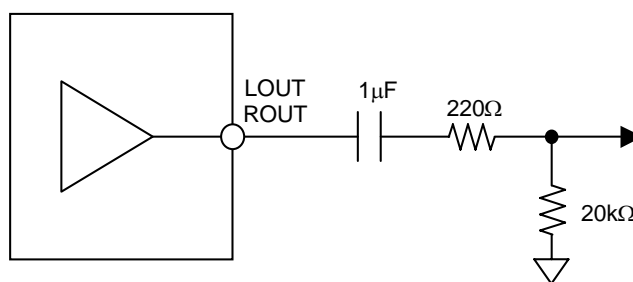


Figure 40. External Circuit for Stereo Line Output (in case of using Pop Reduction Circuit)

<Stereo Line Output Control Sequence (in case of using Pop Reduction Circuit)>

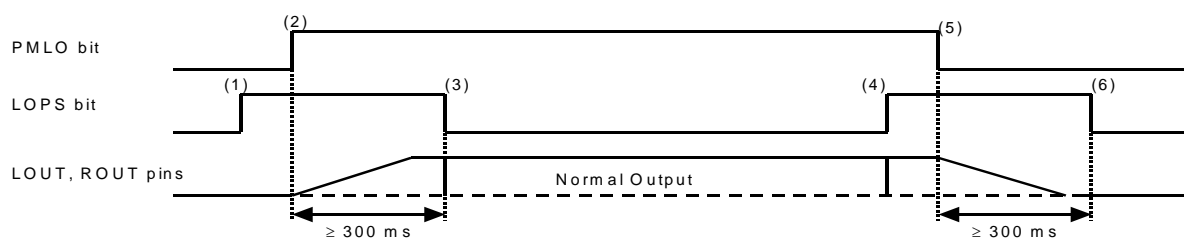


Figure 41. Stereo Line Output Control Sequence (in case of using Pop Reduction Circuit)

- (1) Set LOPS bit = "1". Stereo line output enters the power-save mode.
- (2) Set PMLO bit = "1". Stereo line output exits the power-down mode.
LOUT and ROUT pins rise up to VCOM voltage. Rise time is 200ms (max 300ms) at $C=1\mu\text{F}$ and $\text{AVDD}=3.3\text{V}$.
- (3) Set LOPS bit = "0" after LOUT and ROUT pins rise up. Stereo line output exits the power-save mode.
Stereo line output is enabled.
- (4) Set LOPS bit = "1". Stereo line output enters power-save mode.
- (5) Set PMLO bit = "0". Stereo line output enters power-down mode.
LOUT and ROUT pins fall down to AVSS. Fall time is 200ms (max 300ms) at $C=1\mu\text{F}$ and $\text{AVDD}=3.3\text{V}$.
- (6) Set LOPS bit = "0" after LOUT and ROUT pins fall down. Stereo line output exits the power-save mode.

<Analog Mixing Circuit for Stereo Line Output>

When AIN3 bit = "0", DACL, MINL, LINL2, RINR2, LINL4 and RINR4 bits controls each path switch. MIN path mixing gain is 0dB(typ)@LOVL bit = "0" when the external input resistance is 20kΩ. LIN2, RIN2, LIN4, RIN4 and DAC pathe's mixing gains are 0dB(typ)@LOVL bit = "0".

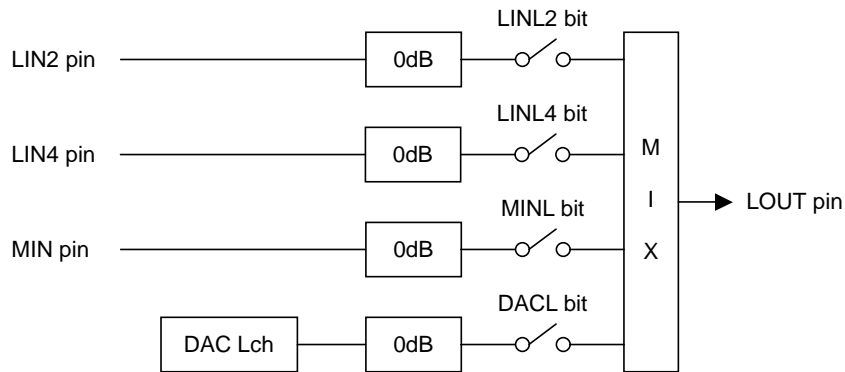


Figure 42. LOUT Mixing Circuit (AIN3 bit = "0", LOVL bit = "0")

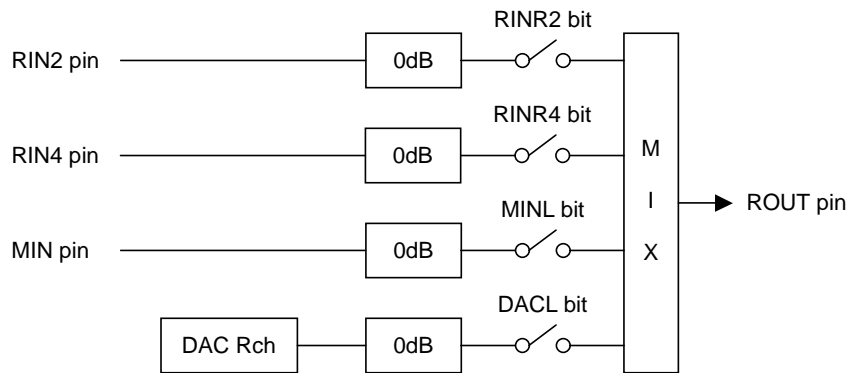


Figure 43. ROUT Mixing Circuit (AIN3 bit = "0", LOVL bit = "0")

When AIN3 bit = "1", DACL, LINL2, RINR2, LINL3, RINR3, LINL4, RINR4, MICL3 and MICR3 bits controls each path switch. All paths' mixing gains are 0dB(typ)@LOVL bit = "0".

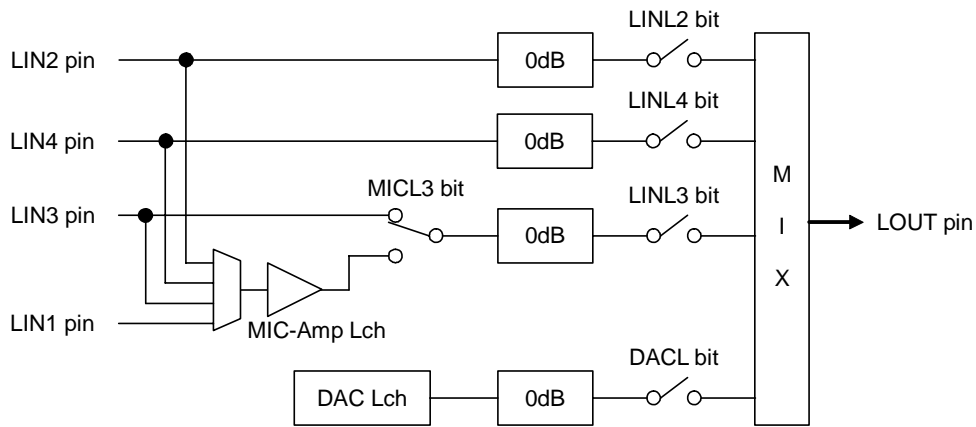


Figure 44. LOUT Mixing Circuit (AIN3 bit = "1", LOVL bit = "0")

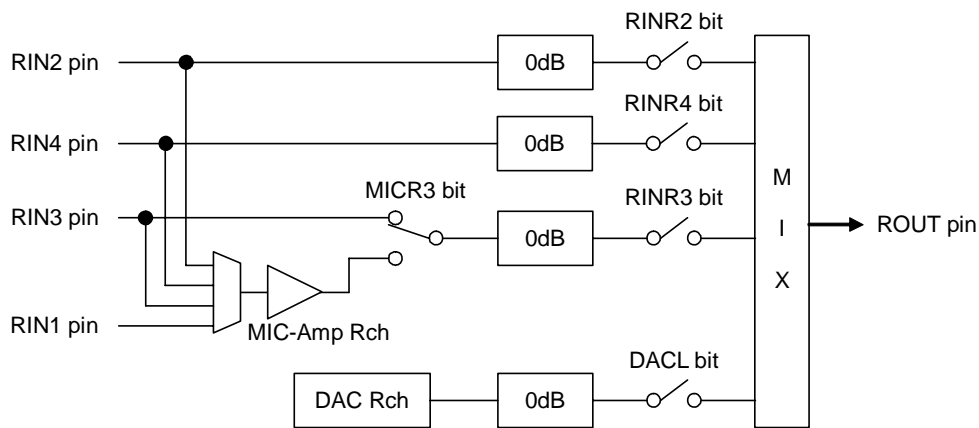


Figure 45. ROUT Mixing Circuit (AIN3 bit = "1", LOVL bit = "0")

■ Full-differential Mono Line Output (LOP/LON pins)

When LODIF bit = “1”, LOUT/ROUT pins become LOP/LON pins, respectively. Lch/Rch signal of DAC or LIN2/RIN2/LIN3/RIN3/LIN4/RIN4 is output from the LOP/LON pins which is full-differential as $(L+R)/2$ signal. The load impedance is 10k Ω (min) for LOP and LON pins, respectively. When the PMLO bit = “0”, the mono line output enters power-down mode and the output is Hi-Z. When the PMLO bit is “1” and LOPS bit is “1”, mono line output enters power-save mode. Pop noise at power-up/down can be reduced by changing PMLO bit at LOPS bit = “0”. When PMLO bit = “1” and LOPS bit = “0”, mono line output enters in normal operation. LOVL bit set the gain of mono line output.

When L4DIF=LODIF bits = “1”, full-differential output signal is as follows: $(LOP) - (LON) = (IN4+) - (IN4-)$.

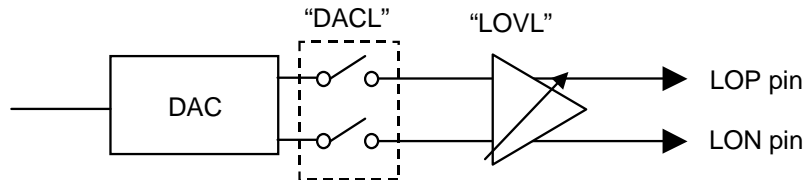


Figure 46. Mono Line Output

PMLO	LOPS	Mode	LOP	LON
0	x	Power-down	Hi-Z	Hi-Z
1	1	Power-save	Hi-Z	VCOM/2
	0	Normal Operation	Normal Operation	Normal Operation

(default)

Table 45. Mono Line Output Mode Setting (x: Don't care)

LOVL	Gain	Output Voltage (typ)
0	+6dB	1.2 x AVDD
1	+8dB	1.5 x AVDD

(default)

Table 46. Mono Line Output Volume Setting

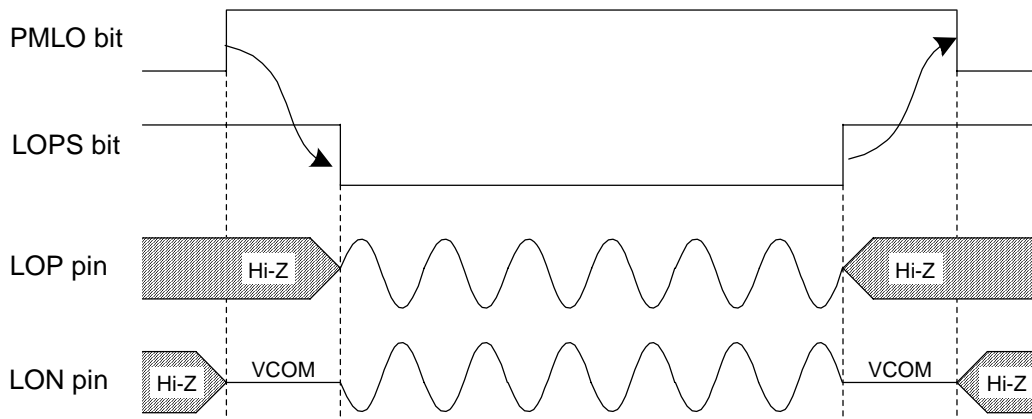


Figure 47. Power-up/Power-down Timing for Mono Line Output

<Analog Mixing Circuit for Mono Line Output>

When AIN3 bit = "0", DACL, MINL, LINL2, RINR2, LINL4 and RINR4 bits controls each path switch. MIN path mixing gain is +6dB(typ)@LOVL bit = "0" when the external input resistance is 20kΩ. LIN2, RIN2, LIN4, RIN4 and DAC pathes mixing gain is 0dB(typ)@LOVL bit = "0".

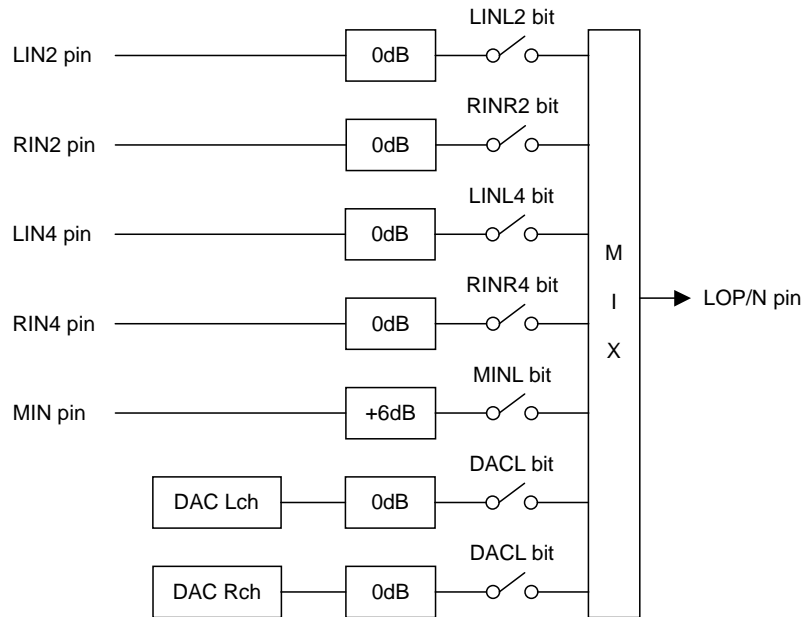


Figure 48. Mono Line Output Mixing Circuit (AIN3 bit = "0", LOVL bit = "0")

When AIN3 bit = "1", DACL, LINL2, RINR2, LINL3, RINR3, LINL4, RINR4, MICL3 and MICR3 bits controls each path switch. All pathes's mixing gains are 0dB(typ)@LOVL bit = "0".

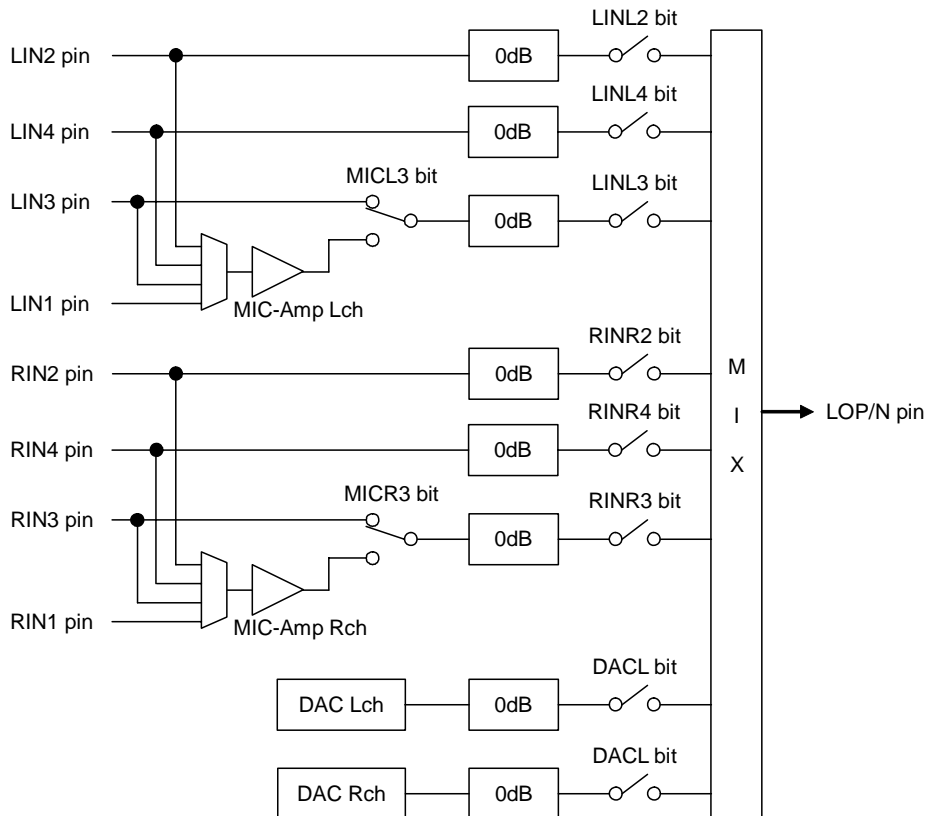


Figure 49. Mono Line Output Mixing Circuit (AIN3 bit = "1", LOVL bit = "0")

■ Headphone Output

Power supply voltage for the Headphone-Amp is supplied from the HVDD pin and centered on the HVDD/2 voltage at VBAT bit = "0". The load resistance is 16Ω (min). HPG bit selects the output voltage (Table 47).

When HPM bit = "1", DAC output signal is output to HPL and HPR pins as (L+R)/2 mono signal.

When HPM3 bit = "1", the signal selected by MICL3 and MICR3 bits (LIN3/RIN3 inputs or MIC-Amp outputs) to HPL and HPR pins as (L+R)/2 mono signal.

HPG bit	0	1
Output Voltage [Vpp]	0.6 x AVDD	0.91 x AVDD

Table 47. Headphone-Amp Output Voltage

When the HPMTN bit is "0", the common voltage of Headphone-Amp falls and the outputs (HPL and HPR pins) go to "L" (HVSS). When the HPMTN bit is "1", the common voltage rises to HVDD/2 at VBAT bit = "0". A capacitor between the MUTET pin and ground reduces pop noise at power-up. Rise/Fall time constant is in proportional to HVDD voltage and the capacitor at MUTET pin.

[Example]: A capacitor between the MUTET pin and ground = 1.0μF, HVDD=3.3V:

Rise/fall time constant: $\tau = 100\text{ms}(\text{typ}), 250\text{ms}(\text{max})$

Time until the common goes to HVSS when HPMTN bit = "1" → "0": 500ms(max)

When PMHPL and PMHPR bits are "0", the Headphone-Amp is powered-down, and the outputs (HPL and HPR pins) go to "L" (HVSS).

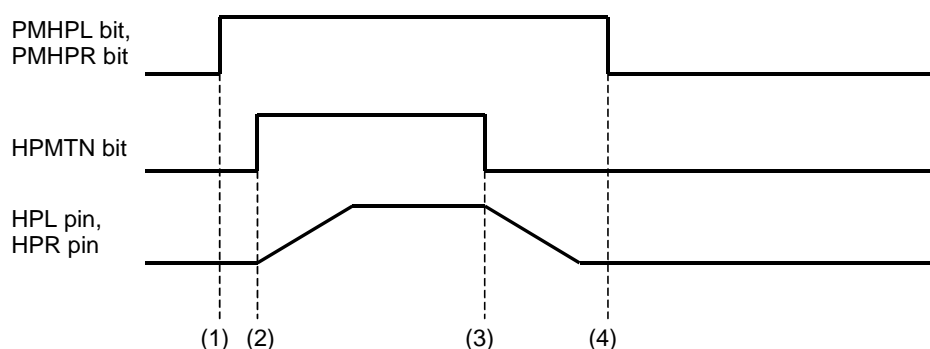


Figure 50. Power-up/Power-down Timing for Headphone-Amp

- (1) Headphone-Amp power-up (PMHPL, PMHPR bit = "1"). The outputs are still HVSS.
- (2) Headphone-Amp common voltage rises up (HPMTN bit = "1"). Common voltage of Headphone-Amp is rising.
- (3) Headphone-Amp common voltage falls down (HPMTN bit = "0"). Common voltage of Headphone-Amp is falling.
- (4) Headphone-Amp power-down (PMHPL, PMHPR bit = "0"). The outputs are HVSS. If the power supply is switched off or Headphone-Amp is powered-down before the common voltage goes to HVSS, some pop noise occurs.

<External Circuit of Headphone-Amp >

When BOOST=OFF, the cut-off frequency (f_c) of Headphone-Amp depends on the external resistor and capacitor. This f_c can be shifted to lower frequency by using bass boost function. Table 48 shows the cut off frequency and the output power for various resistor/capacitor combinations. The headphone impedance R_L is 16Ω. Output powers are shown at HVDD = 3.0, 3.3 and 5.0V. The output voltage of headphone is 0.6 x AVDD (Vpp) @HPG bit = "0" and 0.91 x AVDD (Vpp) @HPG bit = "1".

When an external resistor R is smaller than 12Ω, put an oscillation prevention circuit (0.22μF±20% capacitor and 10Ω±20% resistor) because it has the possibility that Headphone-Amp oscillates.

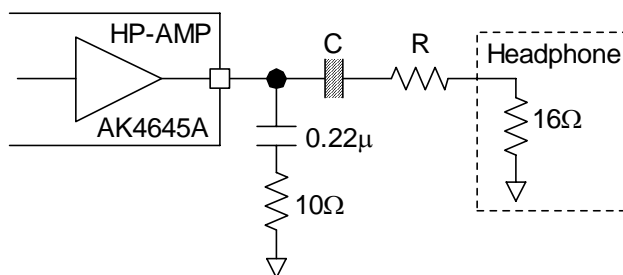


Figure 51. External Circuit Example of Headphone

HPG bit	R [Ω]	C [μF]	fc [Hz] BOOST =OFF	fc [Hz] BOOST =MIN @fs=44.1kHz	Output Power [mW]@0dBFS		
					HVDD=3.0V AVDD=3.0V	HVDD=3.3V AVDD=3.3V	HVDD=5V AVDD=3.3V
0	0	220	45	17	25.3	30.6	30.6
		100	100	43			
	6.8	100	70	28	12.5	15.1	15.1
		47	149	78			
	16	100	50	19	6.3	7.7	7.7
		47	106	47			
1	0	220	45	17	51	62	70
		100	100	43	(Note 40)	(Note 40)	
	100	22	62	25	1.1	1.3	1.3
		10	137	69			

Note 39. Output power at 16Ω load.

Note 40. Output signal is clipped.

Table 48. External Circuit Example

<Headphone-Amp PSRR>

When HVDD is directly supplied from the battery in a mobile phone system, RF noise may influence headphone output performance. When VBAT bit is set to “1”, HP-Amp PSRR for the noise applied to HVDD is reduced. In this case, HP-Amp common voltage is 0.64 x AVDD (typ). When AVDD is 3.3V, common voltage is 2.1V. Therefore, when HVDD voltage becomes lower than 4.2V, the output signal will be clipped easily.

VBAT bit	0	1
Common Voltage [V]	0.5 x HVDD	0.64 x AVDD

Table 49. HP-Amp Common Voltage

<Wired OR with External Headphone-Amp>

When PMVCM=PMHPL=PMHPR bits = “0” and HPZ bit = “1”, HP-Amp is powered-down and HPL/R pins are pulled-down to HVSS by 200kΩ (typ). In this setting, it is available to connect HP-Amp of AK4645A and external single supply HP-Amp by “wired OR”. In this mode, power supply current is 20μA(typ).

PMVCM	PMHPL/R	HPMTN	HPZ	Mode	HPL/R pins	
x	0	x	0	Power-down & Mute	HVSS	(default)
0	0	x	1	Power-down	Pull-down by 200kΩ	
1	1	0	x	Mute	HVSS	
1	1	1	x	Normal Operation	Normal Operation	

Table 50. HP-Amp Mode Setting (x: Don't care)

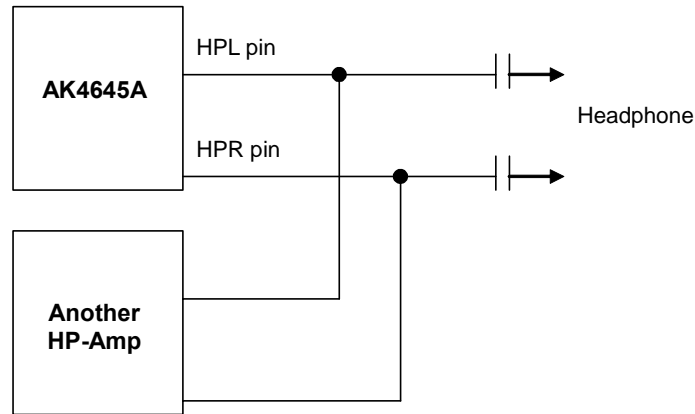


Figure 52. Wired OR with External HP-Amp

<Analog Mixing Circuit for Headphone Output>

When AIN3 bit = "0", DACH, MINH, LINH2, RINH2, LINH4 and RINH4 bits controls each path switch. MIN path mixing gain is -20dB(typ)@HPG bit = "0" when the external input resistance is 20kΩ. LIN2, RIN2, LIN4, RIN4 and DAC pathes mixing gain is 0dB(typ)@HPG bit = "0".

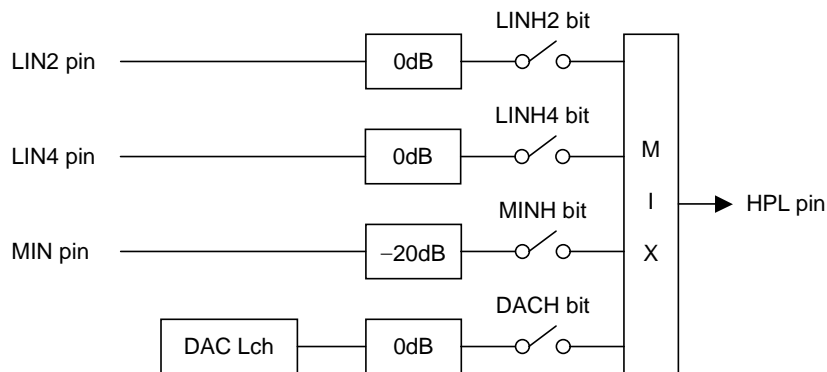


Figure 53. HPL Mixing Circuit (AIN3 bit = "0", HPG bit = "0")

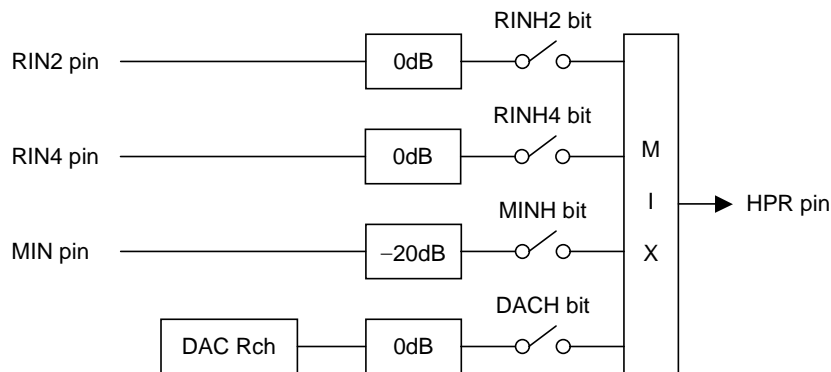


Figure 54. HPR Mixing Circuit (AIN3 bit = "0", HPG bit = "0")

When AIN3 bit = "1", DACH, LINH2, RINH2, LINH3, RINH3, LINH4, RINH4, MICL3 and MICR3 bits controls each path switch. All paths' mixing gains are 0dB(typ)@HPG bit = "0".

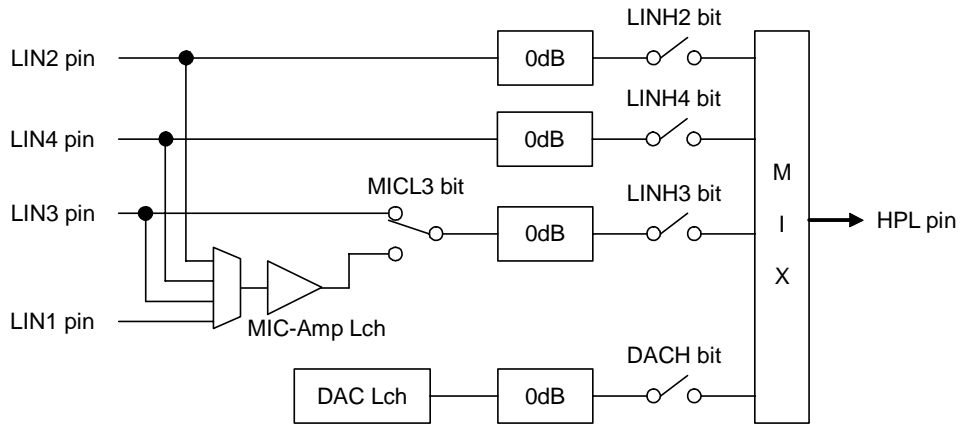


Figure 55. HPL Mixing Circuit (AIN3 bit = "1", HPG bit = "0")

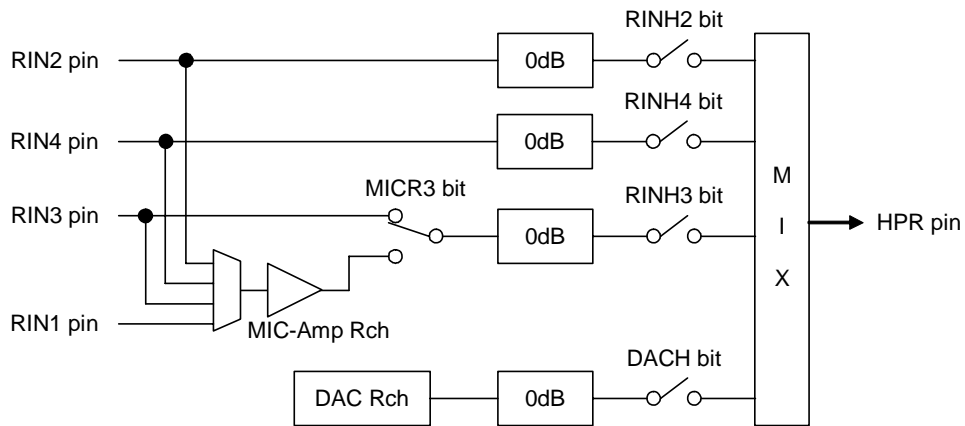
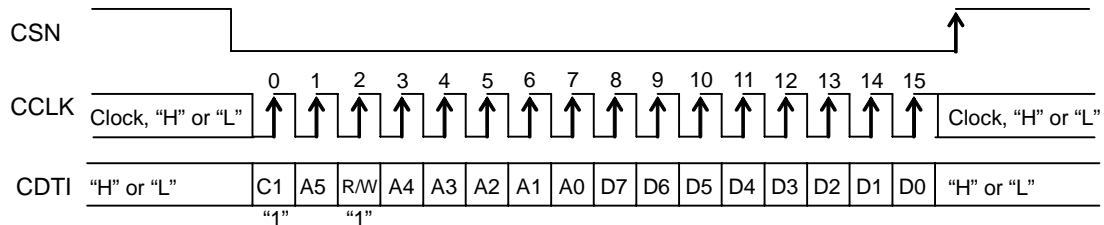


Figure 56. HPR Mixing Circuit (AIN3 bit = "1", HPG bit = "0")

■ Serial Control Interface

(1) 3-wire Serial Control Mode (I2C pin = "L")

Internal registers may be written by using the 3-wire μ P interface pins (CSN, CCLK and CDTI). The data on this interface consists of a 1-bit Chip address (Fixed to "1"), Read/Write (Fixed to "1"), Register address (MSB first, 6bits) and Control data (MSB first, 8bits). Each bit is clocked in on the rising edge ("↑") of CCLK. Address and data are latched on the 16th CCLK rising edge ("↑") after CSN falling edge("↓"). CSN should be set to "H" once after 16 CCLKs for each address. Clock speed of CCLK is 5MHz (max). The value of internal registers are initialized by PDN pin = "L".



- C1: Chip Address; Fixed to "1"
- R/W: READ/WRITE ("1": WRITE, "0": READ); Fixed to "1"
- A5-A0: Register Address
- D7-D0: Control data

Figure 57. Serial Control I/F Timing

(2) I²C-bus Control Mode (I2C pin = "H")

The AK4645A supports the fast-mode I²C-bus (max: 400kHz). Pull-up resistors at SDA and SCL pins should be connected to (TVDD+0.3)V or less voltage.

(2)-1. WRITE Operations

Figure 58 shows the data transfer sequence for the I²C-bus mode. All commands are preceded by START condition. A HIGH to LOW transition on the SDA line while SCL is HIGH indicates START condition (Figure 64). After the START condition, a slave address is sent. This address is 7 bits long followed by the eighth bit that is a data direction bit (R/W). The most significant six bits of the slave address are fixed as "001001". The next bit is CAD0 (device address bit). This bit identifies the specific device on the bus. The hard-wired input pin (CAD0 pin) sets these device address bits (Figure 59). If the slave address matches that of the AK4645A, the AK4645A generates an acknowledge and the operation is executed. The master must generate the acknowledge-related clock pulse and release the SDA line (HIGH) during the acknowledge clock pulse (Figure 65). R/W bit value of "1" indicates that the read operation is to be executed. "0" indicate that the write operation is to be executed.

The second byte consists of the control register address of the AK4645A. The format is MSB first, and those most significant 2-bits are fixed to zeros (Figure 60). The data after the second byte contains control data. The format is MSB first, 8bits (Figure 61). The AK4645A generates an acknowledge after each byte has been received. A data transfer is always terminated by STOP condition generated by the master. A LOW to HIGH transition on the SDA line while SCL is HIGH defines STOP condition (Figure 64).

The AK4645A can perform more than one byte write operation per sequence. After receiving the third byte the AK4645A generates an acknowledge and awaits the next data. The master can transmit more than one byte instead of terminating the write cycle after the first data byte is transferred. After receiving each data packet the internal 6-bit address counter is incremented by one, and the next data is automatically taken into the next address. If the address exceeds 24H prior to generating stop condition, the address counter will "roll over" to 00H and the previous data will be overwritten.

The data on the SDA line must remain stable during the HIGH period of the clock. HIGH or LOW state of the data line can only change when the clock signal on the SCL line is LOW (Figure 66) except for the START and STOP conditions.

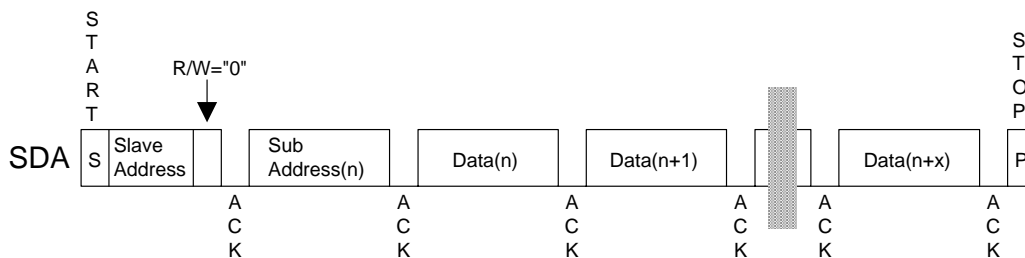
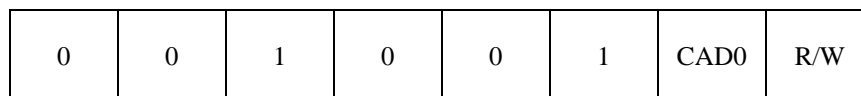


Figure 58. Data Transfer Sequence at the I²C-Bus Mode



(Those CAD1/0 should match with CAD1/0 pins)

Figure 59. The First Byte

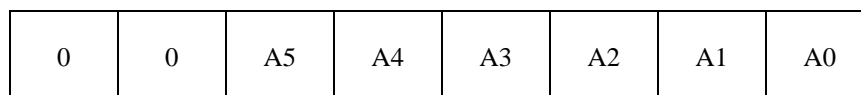


Figure 60. The Second Byte

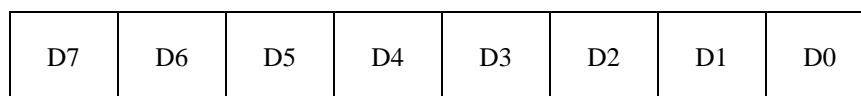


Figure 61. Byte Structure after the second byte

(2)-2. READ Operations

Set the R/W bit = "1" for the READ operation of the AK4645A. After transmission of data, the master can read the next address's data by generating an acknowledge instead of terminating the write cycle. After receiving each data packet the internal 6-bit address counter is incremented by one, and the next data is automatically taken into the next address. If the address exceeds 24H prior to generating a stop condition, the address counter will "roll over" to 00H and the data of 00H will be read out.

The AK4645A supports two basic read operations: CURRENT ADDRESS READ and RANDOM ADDRESS READ.

(2)-2-1. CURRENT ADDRESS READ

The AK4645A contains an internal address counter that maintains the address of the last word accessed, incremented by one. Therefore, if the last access (either a read or write) were to address "n", the next CURRENT READ operation would access data from the address "n+1". After receiving the slave address with R/W bit set to "1", the AK4645A generates an acknowledge, transmits 1-byte of data to the address set by the internal address counter and increments the internal address counter by 1. If the master does not generate an acknowledge but instead generates stop condition, the AK4645A ceases transmission.

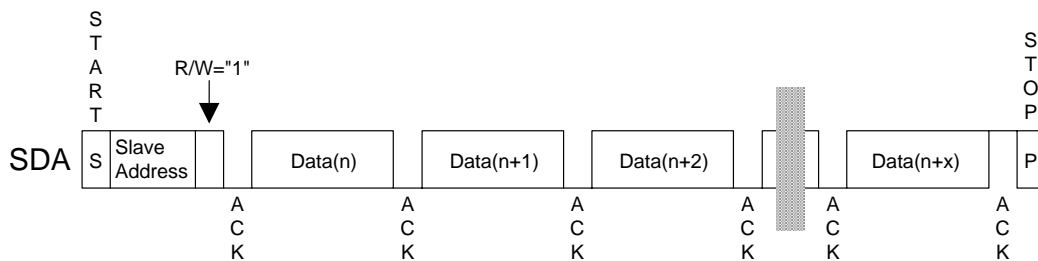


Figure 62. CURRENT ADDRESS READ

(2)-2-2. RANDOM ADDRESS READ

The random read operation allows the master to access any memory location at random. Prior to issuing the slave address with the R/W bit set to "1", the master must first perform a "dummy" write operation. The master issues a start request, a slave address (R/W bit = "0") and then the register address to read. After the register address is acknowledged, the master immediately reissues the start request and the slave address with the R/W bit set to "1". The AK4645A then generates an acknowledge, 1 byte of data and increments the internal address counter by 1. If the master does not generate an acknowledge but instead generates stop condition, the AK4645A ceases transmission.

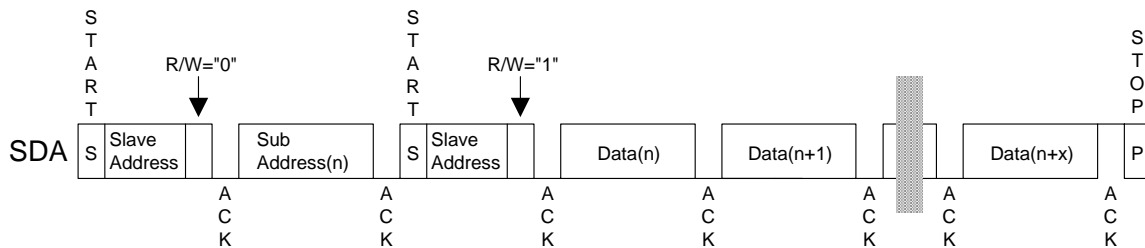


Figure 63. RANDOM ADDRESS READ

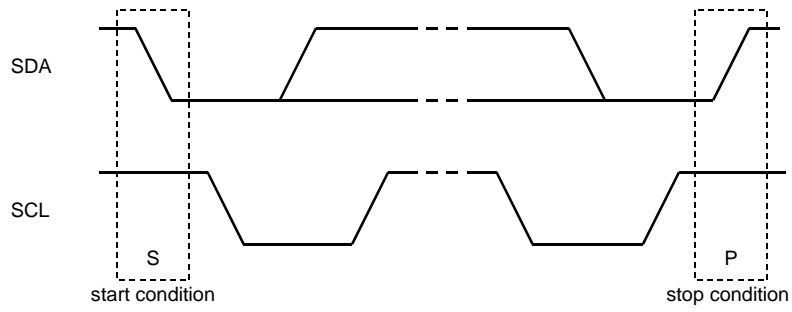


Figure 64. START and STOP Conditions

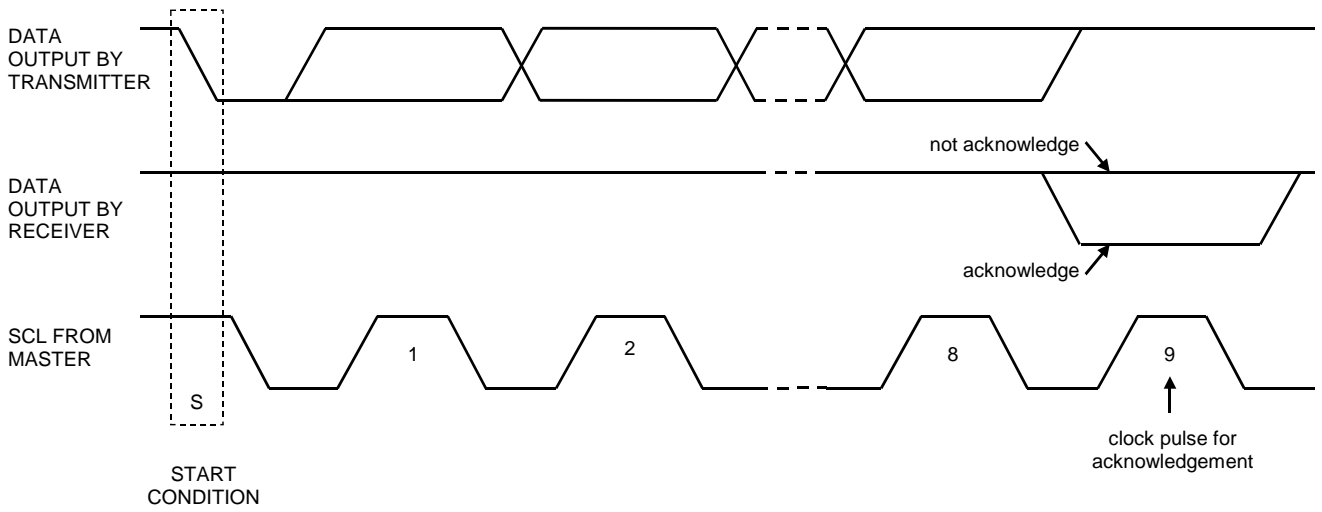


Figure 65. Acknowledge on the I²C-Bus

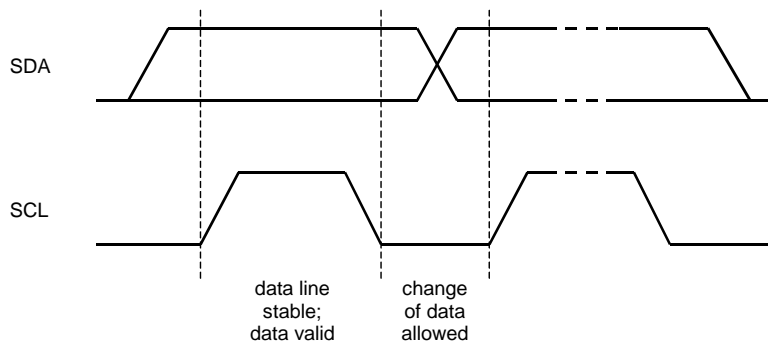


Figure 66. Bit Transfer on the I²C-Bus

■ Register Map

Addr	Register Name	D7	D6	D5	D4	D3	D2	D1	D0
00H	Power Management 1	0	PMVCM	PMMIN	0	PMLO	PMDAC	0	PMADL
01H	Power Management 2	HPZ	HPMTN	PMHPL	PMHPR	M/S	0	0	0
02H	Signal Select 1	0	0	0	DACL	0	PMMP	0	MGAIN0
03H	Signal Select 2	LOVL	LOPS	MGAIN1	0	0	MINL	0	0
04H	Mode Control 1	0	0	0	0	BCKO	0	DIF1	DIF0
05H	Mode Control 2	0	0	FS3	MSBS	BCKP	FS2	FS1	FS0
06H	Timer Select	DVTM	WTM2	ZTM1	ZTM0	WTM1	WTM0	RFST1	RFST0
07H	ALC Mode Control 1	0	0	ALC	ZELMN	LMAT1	LMAT0	RGAIN0	LMTH0
08H	ALC Mode Control 2	REF7	REF6	REF5	REF4	REF3	REF2	REF1	REF0
09H	Lch Input Volume Control	IVL7	IVL6	IVL5	IVL4	IVL3	IVL2	IVL1	IVL0
0AH	Lch Digital Volume Control	DVL7	DVL6	DVL5	DVL4	DVL3	DVL2	DVL1	DVL0
0BH	ALC Mode Control 3	RGAIN1	LMTH1	0	0	0	0	VBAT	0
0CH	Rch Input Volume Control	IVR7	IVR6	IVR5	IVR4	IVR3	IVR2	IVR1	IVR0
0DH	Rch Digital Volume Control	DVR7	DVR6	DVR5	DVR4	DVR3	DVR2	DVR1	DVR0
0EH	Mode Control 3	0	LOOP	SMUTE	DVOLC	BST1	BST0	DEM1	DEM0
0FH	Mode Control 4	0	0	0	0	IVOLC	HPM	MINH	DACH
10H	Power Management 3	INR1	INL1	HPG	MDIF2	MDIF1	INR0	INL0	PMADR
11H	Digital Filter Select	GN1	GN0	0	FIL1	EQ	FIL3	0	0
12H	FIL3 Co-efficient 0	F3A7	F3A6	F3A5	F3A4	F3A3	F3A2	F3A1	F3A0
13H	FIL3 Co-efficient 1	F3AS	0	F3A13	F3A12	F3A11	F3A10	F3A9	F3A8
14H	FIL3 Co-efficient 2	F3B7	F3B6	F3B5	F3B4	F3B3	F3B2	F3B1	F3B0
15H	FIL3 Co-efficient 3	0	0	F3B13	F3B12	F3B11	F3B10	F3B9	F3B8
16H	EQ Co-efficient 0	EQA7	EQA6	EQA5	EQA4	EQA3	EQA2	EQA1	EQA0
17H	EQ Co-efficient 1	EQA15	EQA14	EQA13	EQA12	EQA11	EQA10	EQA9	EQA8
18H	EQ Co-efficient 2	EQB7	EQB6	EQB5	EQB4	EQB3	EQB2	EQB1	EQB0
19H	EQ Co-efficient 3	0	0	EQB13	EQB12	EQB11	EQB10	EQB9	EQB8
1AH	EQ Co-efficient 4	EQC7	EQC6	EQC5	EQC4	EQC3	EQC2	EQC1	EQC0
1BH	EQ Co-efficient 5	EQC15	EQC14	EQC13	EQC12	EQC11	EQC10	EQC9	EQC8
1CH	FIL1 Co-efficient 0	F1A7	F1A6	F1A5	F1A4	F1A3	F1A2	F1A1	F1A0
1DH	FIL1 Co-efficient 1	F1AS	0	F1A13	F1A12	F1A11	F1A10	F1A9	F1A8
1EH	FIL1 Co-efficient 2	F1B7	F1B6	F1B5	F1B4	F1B3	F1B2	F1B1	F1B0
1FH	FIL1 Co-efficient 3	0	0	F1B13	F1B12	F1B11	F1B10	F1B9	F1B8
20H	Power Management 4	PMAINR4	PMAINL4	PMAINR3	PMAINL3	PMAINR2	PMAINL2	PMMICR	PMMICL
21H	Mode Control 5	0	0	MICR3	MICL3	L4DIF	MIX	AIN3	LODIF
22H	Lineout Mixing Select	LOM	LOM3	RINR4	LINL4	RINR3	LINL3	RINR2	LINL2
23H	HP Mixing Select	0	HPM3	RINH4	LINH4	RINH3	LINH3	RINH2	LINH2
24H	Reserved	0	0	0	0	0	0	0	0

Note 41. PDN pin = "L" resets the registers to their default values.

Note 42. Unused bits must contain a "0" value.

■ Register Definitions

Addr	Register Name	D7	D6	D5	D4	D3	D2	D1	D0
00H	Power Management 1	0	PMVCM	PMMIN	0	PMLO	PMDAC	0	PMADL
	Default	0	0	0	0	0	0	0	0

PMADL: MIC-Amp Lch and ADC Lch Power Management

0: Power-down (default)

1: Power-up

When the PMADL or PMADR bit is changed from “0” to “1”, the initialization cycle ($1059/f_s=24\text{ms}$ @44.1kHz) starts. After initializing, digital data of the ADC is output.

PMDAC: DAC Power Management

0: Power-down (default)

1: Power-up

PMLO: Stereo Line Out Power Management

0: Power-down (default)

1: Power-up

PMMIN: MIN Input Power Management

0: Power-down (default)

1: Power-up

PMMIN or PMAINL3 bit should be set to “1” for playback.

PMVCM: VCOM Power Management

0: Power-down (default)

1: Power-up

When any blocks are powered-up, the PMVCM bit must be set to “1”. PMVCM bit can be set to “0” only when all power management bits of 00H, 01H, 02H, 10H, 20H and MCKO bits are “0”.

Each block can be powered-down respectively by writing “0” in each bit of this address. When the PDN pin is “L”, all blocks are powered-down regardless of setting of this address. In this case, register is initialized to the default value.

When all power management bits are “0” in the 00H, 01H, 02H, 10H and 20H addresses and MCKO bit is “0”, all blocks are powered-down. The register values remain unchanged. Power supply current is $20\mu\text{A}$ (typ) in this case. For fully shut down (typ. $1\mu\text{A}$), the PDN pin should be “L”.

When neither ADC nor DAC are used, external clocks may not be present. When ADC or DAC is used, external clocks must always be present.

Addr	Register Name	D7	D6	D5	D4	D3	D2	D1	D0
01H	Power Management 2	HPZ	HPMTN	PMHPL	PMHPR	M/S	0	0	0
	Default	0	0	0	0	0	0	0	0

M/S: Master / Slave Mode Select

0: Slave Mode (default)

1: Master Mode

PMHPR: Headphone-Amp Rch Power Management

0: Power-down (default)

1: Power-up

PMHPL: Headphone-Amp Lch Power Management

0: Power-down (default)

1: Power-up

HPMTN: Headphone-Amp Mute Control

0: Mute (default)

1: Normal operation

HPZ: Headphone-Amp Pull-down Control

0: Shorted to GND (default)

1: Pulled-down by 200k Ω (typ)

Addr	Register Name	D7	D6	D5	D4	D3	D2	D1	D0
02H	Signal Select 1	0	0	0	DACL	0	PMMP	0	MGAIN0
	Default	0	0	0	0	0	0	0	1

MGAIN1-0: MIC-Amp Gain Control ([Table 16](#))

MGAIN1 bit is D5 bit of 03H.

PMMP: MPWR pin Power Management

0: Power-down: Hi-Z (default)

1: Power-up

DACL: Switch Control from DAC to Line Output

0: OFF (default)

1: ON

When PMLO bit is “1”, DAACL bit is enabled. When PMLO bit is “0”, the LOUT/ROUT pins go to AVSS.

Addr	Register Name	D7	D6	D5	D4	D3	D2	D1	D0
03H	Signal Select 2	LOVL	LOPS	MGAIN1	0	0	MINL	0	0
	Default	0	0	0	0	0	0	0	0

MINL: Switch Control from MIN pin to Stereo Line Output

0: OFF (default)

1: ON

When PMLO bit is “1”, MINL bit is enabled. When PMLO bit is “0”, the LOUT/ROUT pins go to AVSS.

MGAIN1: MIC-Amp Gain Control ([Table 16](#))

LOPS: Stereo Line Output Power-Save Mode

0: Normal Operation (default)

1: Power-Save Mode

LOVL: Stereo Line Output Gain Select ([Table 44](#) and [Table 46](#))

0: 0dB/+6dB (default)

1: +2dB/+8dB

Addr	Register Name	D7	D6	D5	D4	D3	D2	D1	D0
04H	Mode Control 1	0	0	0	0	BCKO	0	DIF1	DIF0
	Default	0	0	0	0	0	0	1	0

DIF1-0: Audio Interface Format ([Table 10](#))

Default: “10” (Left justified)

BCKO: BICK Output Frequency Select at Master Mode ([Table 9](#))

Addr	Register Name	D7	D6	D5	D4	D3	D2	D1	D0
05H	Mode Control 2	0	0	FS3	MSBS	BCKP	FS2	FS1	FS0
	Default	0	0	0	0	0	0	0	0

FS3-0: MCKI Frequency Select ([Table 5](#))

FS3-0 bits select MCKI frequency.

BCKP: BICK Polarity at DSP Mode ([Table 11](#))

“0”: SDTO is output by the rising edge (“↑”) of BICK and SDTI is latched by the falling edge (“↓”). (default)

“1”: SDTO is output by the falling edge (“↓”) of BICK and SDTI is latched by the rising edge (“↑”).

MSBS: LRCK Polarity at DSP Mode ([Table 11](#))

“0”: The rising edge (“↑”) of LRCK is half clock of BICK before the channel change. (default)

“1”: The rising edge (“↑”) of LRCK is one clock of BICK before the channel change.

Addr	Register Name	D7	D6	D5	D4	D3	D2	D1	D0
06H	Timer Select	DVTM	WTM2	ZTM1	ZTM0	WTM1	WTM0	RFST1	RFST0
	Default	0	0	0	0	0	0	0	0

RFST1-0: ALC First recovery Speed ([Table 27](#))

Default: "00"(4times)

WTM2-0: ALC Recovery Waiting Period ([Table 24](#))

Default: "000" (128/fs)

ZTM1-0: ALC Limiter/Recovery Operation Zero Crossing Timeout Period ([Table 23](#))

Default: "00" (128/fs)

DVTM: Digital Volume Transition Time Setting ([Table 33](#))

0: 1061/fs (default)

1: 256/fs

This is the transition time between DVL/R7-0 bits = 00H and FFH.

Addr	Register Name	D7	D6	D5	D4	D3	D2	D1	D0
07H	ALC Mode Control 1	0	0	ALC	ZELMN	LMAT1	LMAT0	RGAIN0	LMTH0
	Default	0	0	0	0	0	0	0	0

LMTH1-0: ALC Limiter Detection Level / Recovery Counter Reset Level ([Table 21](#))

Default: "00"

LMTH1 bit is D6 bit of 0BH.

RGAIN1-0: ALC Recovery GAIN Step ([Table 25](#))

Default: "00"

RGAIN1 bit is D7 bit of 0BH.

LMAT1-0: ALC Limiter ATT Step ([Table 22](#))

Default: "00"

ZELMN: Zero Crossing Detection Enable at ALC Limiter Operation

0: Enable (default)

1: Disable

ALC: ALC Enable

0: ALC Disable (default)

1: ALC Enable

Addr	Register Name	D7	D6	D5	D4	D3	D2	D1	D0
08H	ALC Mode Control 2	REF7	REF6	REF5	REF4	REF3	REF2	REF1	REF0
	Default	1	1	1	0	0	0	0	1

REF7-0: Reference Value at ALC Recovery Operation. 0.375dB step, 242 Level ([Table 26](#))

Default: "E1H" (+30.0dB)

Addr	Register Name	D7	D6	D5	D4	D3	D2	D1	D0
09H	Lch Input Volume Control	IVL7	IVL6	IVL5	IVL4	IVL3	IVL2	IVL1	IVL0
0CH	Rch Input Volume Control	IVR7	IVR6	IVR5	IVR4	IVR3	IVR2	IVR1	IVR0
Default		1	1	1	0	0	0	0	1

IVL7-0, IVR7-0: Input Digital Volume; 0.375dB step, 242 Level ([Table 29](#))
 Default: "E1H" (+30.0dB)

Addr	Register Name	D7	D6	D5	D4	D3	D2	D1	D0
0AH	Lch Digital Volume Control	DVL7	DVL6	DVL5	DVL4	DVL3	DVL2	DVL1	DVL0
0DH	Rch Digital Volume Control	DVR7	DVR6	DVR5	DVR4	DVR3	DVR2	DVR1	DVR0
Default		0	0	0	1	1	0	0	0

DVL7-0, DVR7-0: Output Digital Volume ([Table 32](#))
 Default: "18H" (0dB)

Addr	Register Name	D7	D6	D5	D4	D3	D2	D1	D0
0BH	ALC Mode Control 3	RGAIN1	LMTH1	0	0	0	0	VBAT	0
Default		0	0	0	0	0	0	0	0

VBAT: HP-Amp Common Voltage ([Table 49](#))
 0: 0.5 x HVDD (default)
 1: 0.64 x AVDD

LMTH1: ALC Limiter Detection Level / Recovery Counter Reset Level ([Table 21](#))

RGAIN1: ALC Recovery GAIN Step ([Table 25](#))

Addr	Register Name	D7	D6	D5	D4	D3	D2	D1	D0
0EH	Mode Control 3	0	LOOP	SMUTE	DVOLC	BST1	BST0	DEM1	DEM0
Default		0	0	0	1	0	0	0	1

DEM1-0: De-emphasis Frequency Select ([Table 30](#))
 Default: "01" (OFF)

BST1-0: Bass Boost Function Select ([Table 31](#))
 Default: "00" (OFF)

DVOLC: Output Digital Volume Control Mode Select

0: Independent

1: Dependent (default)

When DVOLC bit = "1", DVL7-0 bits control both Lch and Rch volume level, while register values of DVL7-0 bits are not written to DVR7-0 bits. When DVOLC bit = "0", DVL7-0 bits control Lch level and DVR7-0 bits control Rch level, respectively.

SMUTE: Soft Mute Control

0: Normal Operation (default)

1: DAC outputs soft-muted

LOOP: Digital Loopback Mode

0: SDTI → DAC (default)

1: SDTO → DAC

Addr	Register Name	D7	D6	D5	D4	D3	D2	D1	D0
0FH	Mode Control 4	0	0	0	0	IVOLC	HPM	MINH	DACH
	Default	0	0	0	0	1	0	0	0

DACH: Switch Control from DAC to Headphone-Amp

- 0: OFF (default)
- 1: ON

MINH: Switch Control from MIN pin to Headphone-Amp

- 0: OFF (default)
- 1: ON

HPM: Headphone-Amp Mono Output Select

- 0: Stereo (default)
- 1: Mono

When the HPM bit = "1", DAC output signal is output to Lch and Rch of the Headphone-Amp as (L+R)/2.

IVOLC: Input Digital Volume Control Mode Select

- 0: Independent
- 1: Dependent (default)

When IVOLC bit = "1", IVL7-0 bits control both Lch and Rch volume level, while register values of IVL7-0 bits are not written to IVR7-0 bits. When IVOLC bit = "0", IVL7-0 bits control Lch level and IVR7-0 bits control Rch level, respectively.

Addr	Register Name	D7	D6	D5	D4	D3	D2	D1	D0
10H	Power Management 3	INR1	INL1	HPG	MDIF2	MDIF1	INR0	INL0	PMADR
	Default	0	0	0	0	0	0	0	0

PMADR: MIC-Amp Lch and ADC Rch Power Management

- 0: Power-down (default)
- 1: Power-up

INL1-0: ADC Lch Input Source Select ([Table 13](#))

Default: 00 (LIN1 pin)

INR1-0: ADC Rch Input Source Select ([Table 13](#))

Default: 00 (RIN1 pin)

MDIF1: Single-ended / Full-differential Input Select 1

- 0: Single-ended input (LIN1/RIN1 pins: Default)
 - 1: Full-differential input (IN1+/IN1- pins)
- MDIF1 bit selects the input type of pins #32 and #31.

MDIF2: Single-ended / Full-differential Input Select 2

- 0: Single-ended input (LIN2/RIN2 pins: Default)
 - 1: Full-differential input (IN2+/IN2- pins)
- MDIF2 bit selects the input type of pins #30 and #29.

HPG: Headphone-Amp Gain Select ([Table 47](#))

- 0: 0dB (default)
- 1: +3.6dB

Addr	Register Name	D7	D6	D5	D4	D3	D2	D1	D0
11H	Digital Filter Select	GN1	GN0	0	FIL1	EQ	FIL3	0	0
	Default	0	0	0	0	0	0	0	0

GN1-0: Gain Select at GAIN block (Table 19)

Default: "00"

FIL3: FIL3 (Stereo Separation Emphasis Filter) Coefficient Setting Enable

0: Disable (default)

1: Enable

When FIL3 bit is "1", the settings of F3A13-0 and F3B13-0 bits are enabled. When FIL3 bit is "0", FIL3 block is OFF (MUTE).

EQ: EQ (Gain Compensation Filter) Coefficient Setting Enable

0: Disable (default)

1: Enable

When EQ bit is "1", the settings of EQA15-0, EQB13-0 and EQC15-0 bits are enabled. When EQ bit is "0", EQ block is through (0dB).

FIL1: FIL1 (Wind-noise Reduction Filter) Coefficient Setting Enable

0: Disable (default)

1: Enable

When FIL1 bit is "1", the settings of F1A13-0 and F1B13-0 bits are enabled. When FIL1 bit is "0", FIL1 block is through (0dB).

Addr	Register Name	D7	D6	D5	D4	D3	D2	D1	D0
12H	FIL3 Co-efficient 0	F3A7	F3A6	F3A5	F3A4	F3A3	F3A2	F3A1	F3A0
13H	FIL3 Co-efficient 1	F3AS	0	F3A13	F3A12	F3A11	F3A10	F3A9	F3A8
14H	FIL3 Co-efficient 2	F3B7	F3B6	F3B5	F3B4	F3B3	F3B2	F3B1	F3B0
15H	FIL3 Co-efficient 3	0	0	F3B13	F3B12	F3B11	F3B10	F3B9	F3B8
16H	EQ Co-efficient 0	EQA7	EQA6	EQA5	EQA4	EQA3	EQA2	EQA1	EQA0
17H	EQ Co-efficient 1	EQA15	EQA14	EQA13	EQA12	EQA11	EQA10	EQA9	EQA8
18H	EQ Co-efficient 2	EQB7	EQB6	EQB5	EQB4	EQB3	EQB2	EQB1	EQB0
19H	EQ Co-efficient 3	0	0	EQB13	EQB12	EQB11	EQB10	EQB9	EQB8
1AH	EQ Co-efficient 4	EQC7	EQC6	EQC5	EQC4	EQC3	EQC2	EQC1	EQC0
1BH	EQ Co-efficient 5	EQC15	EQC14	EQC13	EQC12	EQC11	EQC10	EQC9	EQC8
1CH	FIL1 Co-efficient 0	F1A7	F1A6	F1A5	F1A4	F1A3	F1A2	F1A1	F1A0
1DH	FIL1 Co-efficient 1	F1AS	0	F1A13	F1A12	F1A11	F1A10	F1A9	F1A8
1EH	FIL1 Co-efficient 2	F1B7	F1B6	F1B5	F1B4	F1B3	F1B2	F1B1	F1B0
1FH	FIL1 Co-efficient 3	0	0	F1B13	F1B12	F1B11	F1B10	F1B9	F1B8
	Default	0	0	0	0	0	0	0	0

F3A13-0, F3B13-0: FIL3 (Stereo Separation Emphasis Filter) Coefficient (14bit x 2)

Default: "0000H"

F3AS: FIL3 (Stereo Separation Emphasis Filter) Select

0: HPF (default)

1: LPF

EQA15-0, EQB13-0, EQC15-0: EQ (Gain Compensation Filter) Coefficient (14bit x 2 + 16bit x 1)

Default: "0000H"

F1A13-0, F1B13-0: FIL1 (Wind-noise Reduction Filter) Coefficient (14bit x 2)

Default: "0000H"

F1AS: FIL1 (Wind-noise Reduction Filter) Select

0: HPF (default)

1: LPF

Addr	Register Name	D7	D6	D5	D4	D3	D2	D1	D0
20H	Power Management 4	PMAINR4	PMAINL4	PMAINR3	PMAINL3	PMAINR2	PMAINL2	PMMICR	PMMICL
	Default	0	0	0	0	0	0	0	0

PMMICL: MIC-Amp Lch Power Management

0: Power down (default)

1: Power up

PMMICR: MIC-Amp Rch Power Management

0: Power down (default)

1: Power up

PMAINL2: LIN2 Mixing Circuit Power Management

0: Power down (default)

1: Power up

PMAINR2: RIN2 Mixing Circuit Power Management

0: Power down (default)

1: Power up

PMAINL3: LIN3 Mixing Circuit Power Management

0: Power down (default)

1: Power up

PMAINR3: RIN3 Mixing Circuit Power Management

0: Power down (default)

1: Power up

PMAINL4: LIN4 Mixing Circuit Power Management

0: Power down (default)

1: Power up

PMAINR4: RIN4 Mixing Circuit Power Management

0: Power down (default)

1: Power up

Addr	Register Name	D7	D6	D5	D4	D3	D2	D1	D0
21H	Mode Control 5	0	0	MICR3	MICL3	L4DIF	MIX	AIN3	LODIF
	Default	0	0	0	0	0	0	0	0

LODIF: Lineout Select

- 0: Single-ended Stereo Line Output (LOUT/ROUT pins) (default)
- 1: Full-differential Mono Line Output (LOP/LON pins)

AIN3: Analog Mixing Select

- 0: Mono Input (MIN pin) (default)
- 1: Stereo Input (LIN3/RIN3 pins)

MIX: Mono Recording

- 0: Stereo (default)
- 1: Mono: (L+R)/2

L4DIF: Line Input Type Select

- 0: Stereo Single-ended Input: LIN4/RIN4 pins (default)
- 1: Mono Full-differential Input: IN4+/- pins

MICL3: Switch Control from MIC-Amp Lch to Analog Output

- 0: LIN3 input signal is selected. (default)
- 1: MIC-Amp Lch output signal is selected.

MICR3: Switch Control from MIC-Amp Rch to Analog Output

- 0: RIN3 input signal is selected. (default)
- 1: MIC-Amp Rch output signal is selected.

Addr	Register Name	D7	D6	D5	D4	D3	D2	D1	D0
22H	Lineout Mixing Select	LOM	LOM3	RINR4	LINL4	RINR3	LINL3	RINR2	LINL2
	Default	0	0	0	0	0	0	0	0

LINL2: Switch Control from LIN2 pin to Stereo Line Output (without MIC-Amp)

0: OFF (default)

1: ON

RINR2: Switch Control from RIN2 pin to Stereo Line Output (without MIC-Amp)

0: OFF (default)

1: ON

LINL3: Switch Control from LIN3 pin (or MIC-Amp Lch) to Stereo Line Output

0: OFF (default)

1: ON

RINR3: Switch Control from RIN3 pin (or MIC-Amp Lch) to Stereo Line Output

0: OFF (default)

1: ON

LINL4: Switch Control from LIN4 pin to Stereo Line Output (without MIC-Amp)

0: OFF (default)

1: ON

RINR4: Switch Control from RIN4 pin to Stereo Line Output (without MIC-Amp)

0: OFF (default)

1: ON

LOM3: Mono Mixing from MIC-Amp (or LIN3/RIN3) to Stereo Line Output

0: Stereo Mixing (default)

1: Mono Mixing

LOM: Mono Mixing from DAC to Stereo Line Output

0: Stereo Mixing (default)

1: Mono Mixing

Addr	Register Name	D7	D6	D5	D4	D3	D2	D1	D0
23H	HP Mixing Select	0	HPM3	RINH4	LINH4	RINH3	LINH3	RINH2	LINH2
	Default	0	0	0	0	0	0	0	0

LINH2: Switch Control from LIN2 pin to Headphone Output (without MIC-Amp)

0: OFF (default)

1: ON

RINH2: Switch Control from RIN2 pin to Headphone Output (without MIC-Amp)

0: OFF (default)

1: ON

LINH3: Switch Control from LIN3 pin (or MIC-Amp Lch) to Headphone Output

0: OFF (default)

1: ON

RINH3: Switch Control from RIN3 pin (or MIC-Amp Lch) to Headphone Output

0: OFF (default)

1: ON

LINH4: Switch Control from LIN4 pin to Headphone Output (without MIC-Amp)

0: OFF (default)

1: ON

RINH4: Switch Control from RIN4 pin to Headphone Output (without MIC-Amp)

0: OFF (default)

1: ON

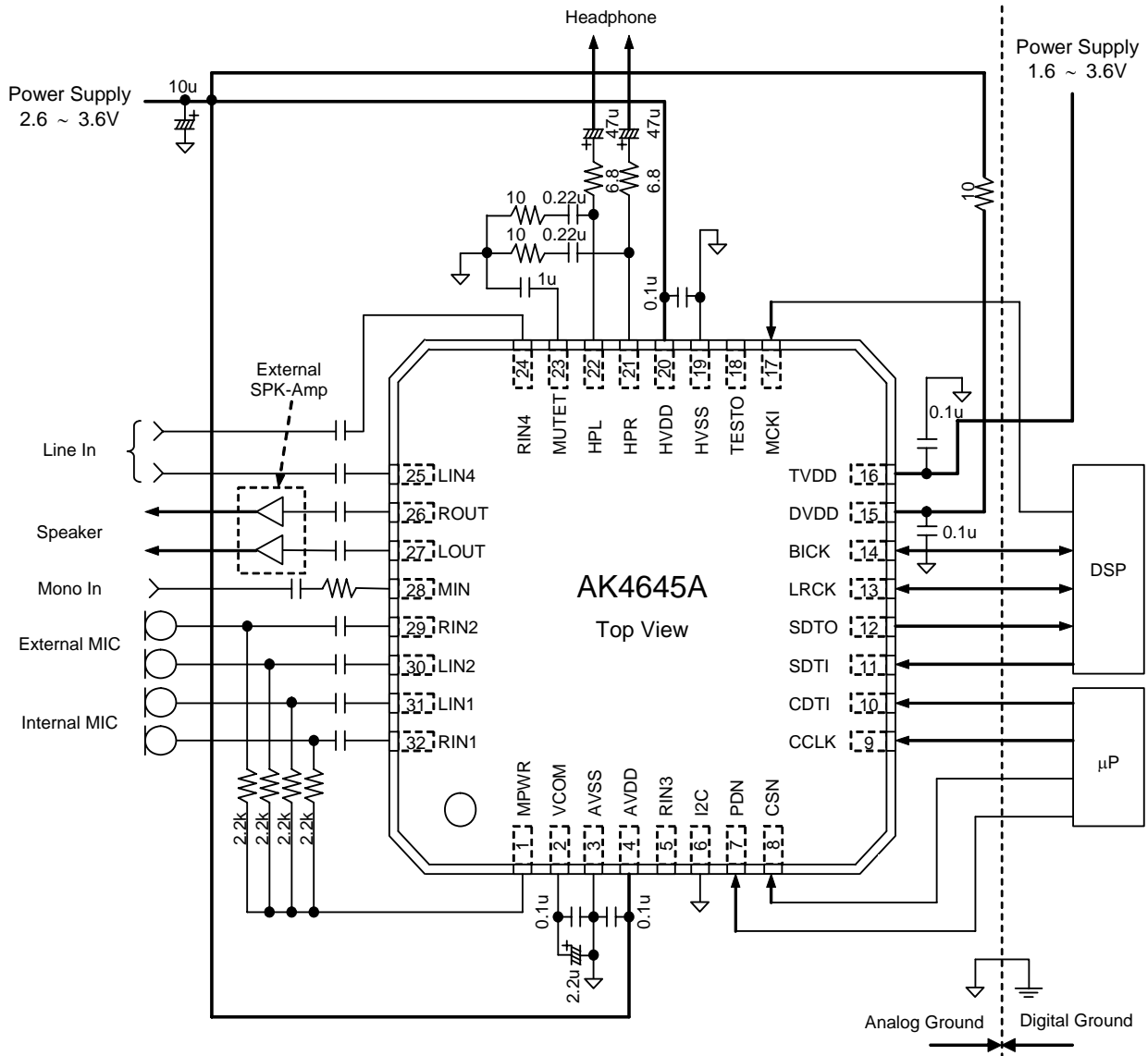
HPM3: Mono Mixing from MIC-Amp (or LIN3/RIN3) to Headphone Output

0: Stereo Mixing (default)

1: Mono Mixing

SYSTEM DESIGN

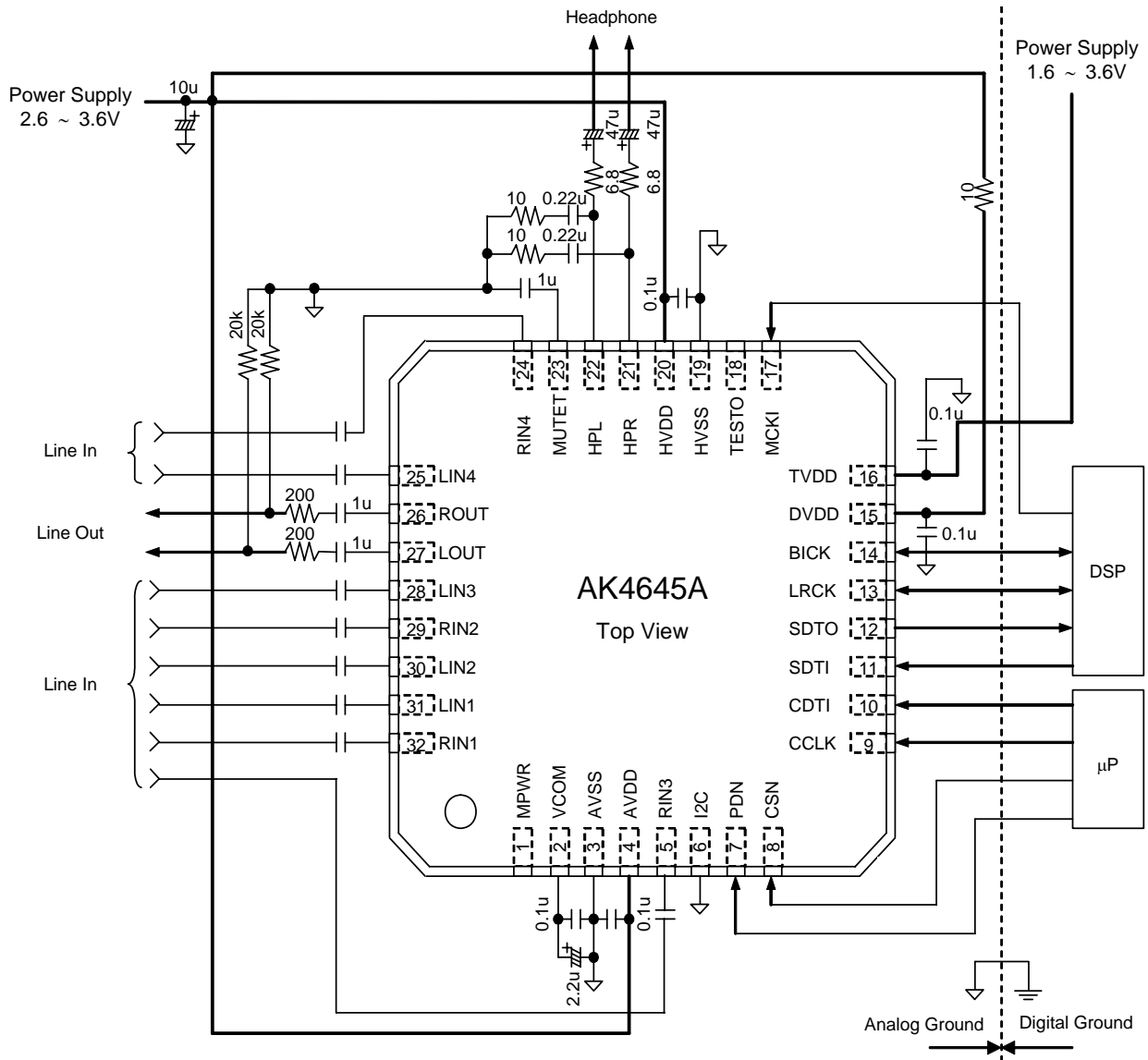
Figure 67 and Figure 68 shows the system connection diagram for the AK4645A. An evaluation board [AKD4645] is available which demonstrates the optimum layout, power supply arrangements and measurement results.



Notes:

- AVSS and HVSS of the AK4645A must be distributed separately from the ground of external controllers.
- All digital input pins must not be left floating.
- When the AK4645A is used at master mode, LRCK and BICK pins are floating before M/S bit is changed to "1". Therefore, 100kΩ around pull-up resistor must be connected to LRCK and BICK pins of the AK4645A.
- 0.1μF ceramic capacitor must be attached to each supply pins. The type of other capacitors is not critical.
- When DVDD is supplied from AVDD via 10Ω series resistor, the capacitor larger than 0.1μF must not be connected between DVDD and the ground.

Figure 67. Typical Connection Diagram (AIN3 bit = "0", MIC Input)


Notes:

- AVSS and HVSS of the AK4645A must be distributed separately from the ground of external controllers.
- All digital input pins must not be left floating.
- When the AK4645A is used at master mode, LRCK and BICK pins are floating before M/S bit is changed to "1". Therefore, 100kΩ around pull-up resistor must be connected to LRCK and BICK pins of the AK4645A.
- 0.1µF ceramic capacitor should be attached to each supply pins. The type of other capacitors is not critical.
- When DVDD is supplied from AVDD via 10Ω series resistor, the capacitor larger than 0.1µF must not be connected between DVDD and the ground.

Figure 68. Typical Connection Diagram (AIN3 bit = "1", Line Input)

1. Grounding and Power Supply Decoupling

The AK4645A requires careful attention to power supply and grounding arrangements. AVDD, DVDD, TVDD and HVDD are usually supplied from the system's analog supply. If AVDD, DVDD, TVDD and HVDD are supplied separately, the power-up sequence is not critical. The PDN pin must be held to "L" upon power-up. The PDN pin must be set to "H" after all power supplies are powered-up.

In case that pop noise have to be avoided at line output and headphone output, the AK4645A should be operated by the following recommended power-up/down sequence.

- 1) Power-up
 - The PDN pin must be held to "L" upon power-up. The AK4645A must be reset by bringing PDN pin "L" for 150ns or more.
 - In case that the power supplies are separated in two or more groups, the power supply including TVDD should be powered ON at first. Regarding the relationship between DVDD and HVDD, the power supply including DVDD should be powered ON prior to the power supply including HVDD.
- 2) Power-down
 - Each power supplies must be powered OFF after the PDN pin is set to "L".
 - In case that the power supplies are separated in two or more groups, the power supply including TVDD should be powered OFF at last. Regarding the relationship between DVDD and HVDD, the power supply including HVDD should be powered OFF prior to the power supply including DVDD.

AVSS and HVSS of the AK4645A must be connected to the analog ground plane. System analog ground and digital ground must be connected together near to where the supplies are brought onto the printed circuit board. Decoupling capacitors should be as near to the AK4645A as possible, with the small value ceramic capacitor being the nearest.

2. Voltage Reference

VCOM is a signal ground of this chip. A 2.2 μ F electrolytic capacitor in parallel with a 0.1 μ F ceramic capacitor attached to the VCOM pin eliminates the effects of high frequency noise. No load current may be drawn from the VCOM pin. All signals, especially clocks, should be kept away from the VCOM pin in order to avoid unwanted coupling into the AK4645A.

3. Analog Inputs

The Mic, Line and MIN inputs are single-ended. The input signal range scales with nominally at 0.06 x AVDD Vpp(typ) @MGAIN1-0 bits = "01", 0.03 x AVDD Vpp(typ) @MGAIN1-0 bits = "10", 0.015 x AVDD Vpp(typ) @MGAIN1-0 bits = "11" or 0.6 x AVDD Vpp(typ) @MGAIN1-0 bits = "00" for the Mic/Line input and 0.6 x AVDD Vpp (typ) for the MIN input, centered around the internal common voltage (0.45 x AVDD). Usually the input signal is AC coupled using a capacitor. The cut-off frequency is $f_c = (1/2\pi RC)$. The AK4645A can accept input voltages from AVSS to AVDD.

4. Analog Outputs

The input data format for the DAC is 2's complement. The output voltage is a positive full scale for 7FFFH(@16bit) and a negative full scale for 8000H(@16bit). The ideal output is VCOM voltage for 0000H(@16bit). Stereo Line Output is centered at 0.45 x AVDD. The Headphone-Amp output is centered at HVDD/2.

CONTROL SEQUENCE

■ Clock Set up

When ADC or DAC is powered-up, the clocks must be supplied.

1. 1. EXT Slave Mode

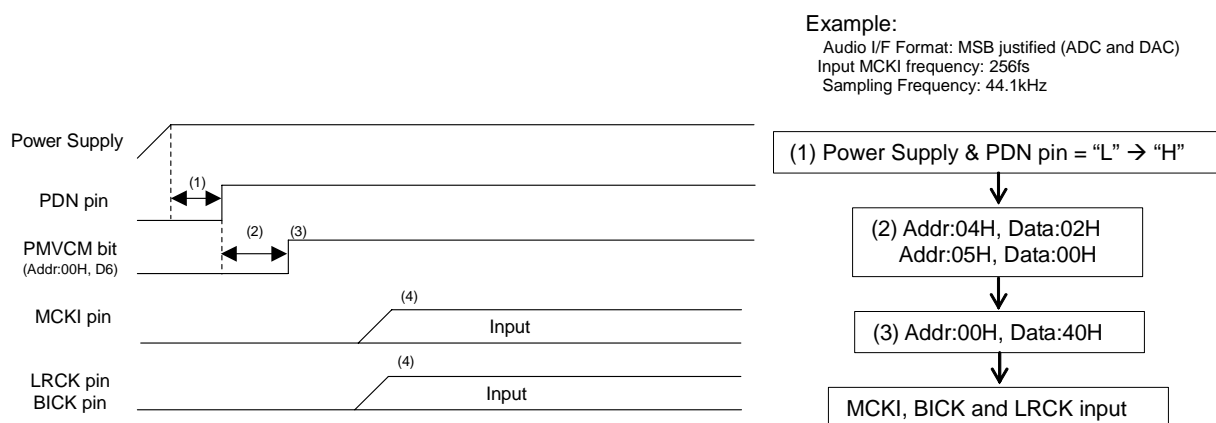


Figure 69. Clock Set Up Sequence (4)

<Example>

- (1) After Power Up, PDN pin = "L" → "H". "L" time of 150ns or more is needed to reset the AK4645A. The AK4645A should be operated by the recommended power-up/down sequence shown in "System Design (Grounding and Power Supply Decoupling)" to avoid pop noise at line output and headphone output.
- (2) DIF1-0 and FS1-0 bits must be set during this period.
- (3) Power Up VCOM: PMVCM bit = "0" → "1"
VCOM should first be powered up before the other block operates.
- (4) Normal operation starts after the MCKI, LRCK and BICK are supplied.

2. EXT Master Mode

Example:

Audio I/F Format: MSB justified (ADC and DAC)
 Input MCKI frequency: 256fs
 Sampling Frequency: 44.1kHz

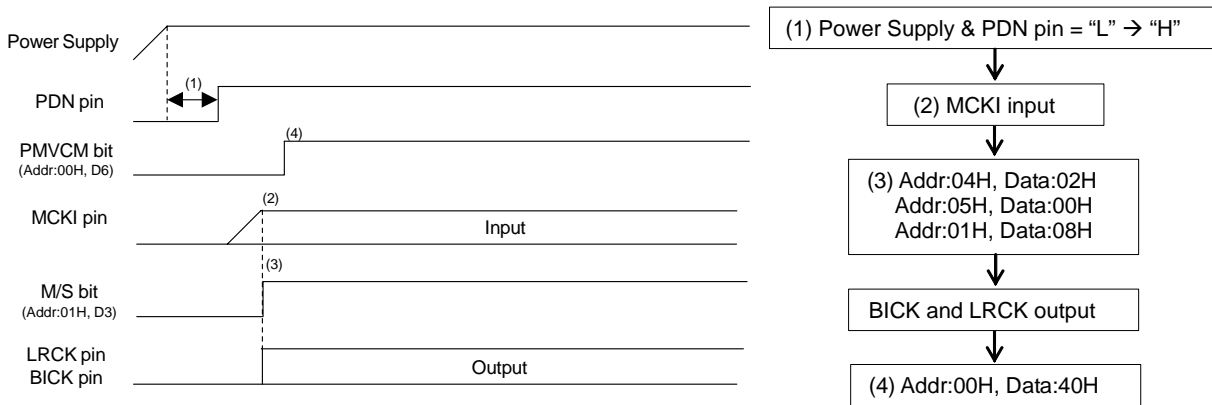


Figure 70. Clock Set Up Sequence (5)

<Example>

- (1) After Power Up, PDN pin = "L" → "H". "L" time of 150ns or more is needed to reset the AK4645A. The AK4645A must be operated by the recommended power-up/down sequence shown in "System Design (Grounding and Power Supply Decoupling)" to avoid the pop noise at line output and headphone output.
- (2) MCKI should be input.
- (3) After DIF1-0 and FS1-0 bits are set, M/S bit must be set to "1". Then LRCK and BICK are output.
- (4) Power Up VCOM: PMVCM bit = "0" → "1"
VCOM should first be powered up before the other block operates.

■ MIC Input Recording (Stereo)

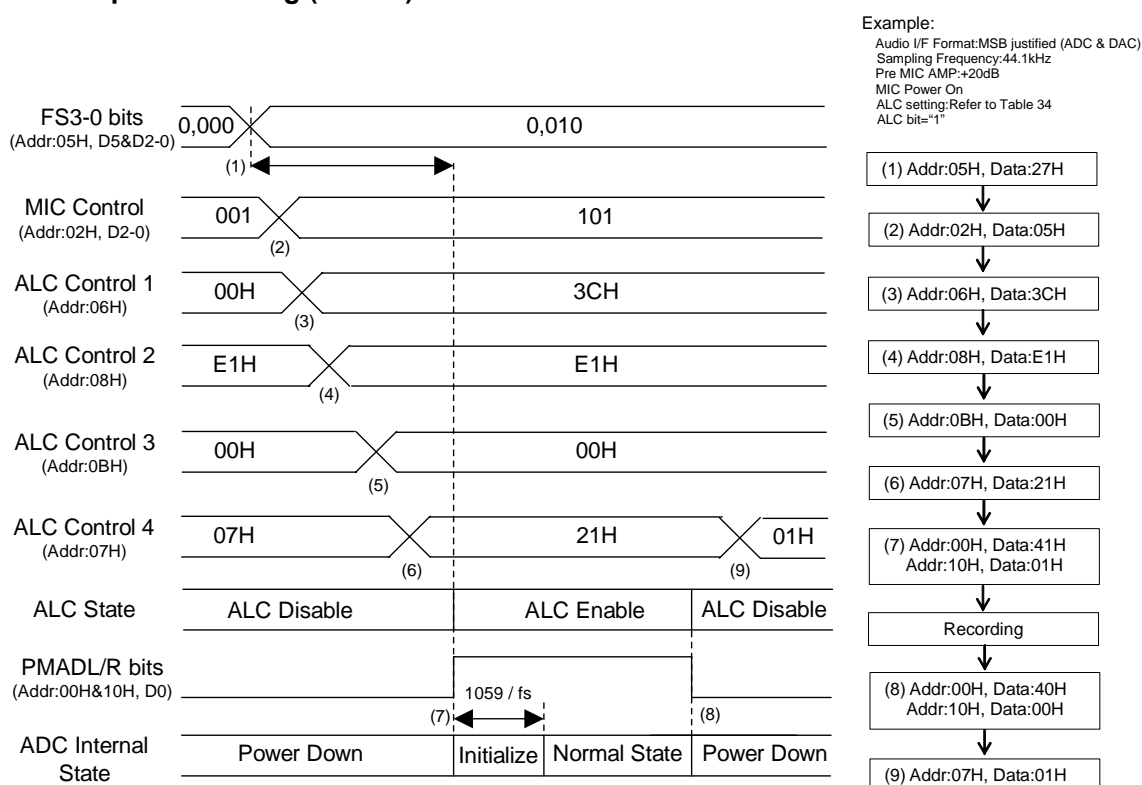


Figure 71. MIC Input Recording Sequence

<Example>

This sequence is an example of ALC setting at $fs=44.1\text{kHz}$. If the parameter of the ALC is changed, please refer to “Figure 29. Registers set-up sequence at ALC operation”

At first, clocks must be supplied according to “Clock Set Up” sequence.

- (1) Set up a sampling frequency (FS3-0 bit).
- (2) Set up MIC input (Addr: 02H)
- (3) Set up Timer Select for ALC (Addr: 06H)
- (4) Set up REF value for ALC (Addr: 08H)
- (5) Set up LMTH1 and RGAIN1 bits (Addr: 0BH)
- (6) Set up LMTH0, RGAIN0, LMAT1-0 and ALC bits (Addr: 07H)
- (7) Power Up MIC and ADC: PMADL = PMADR bits = “0” → “1”

The initialization cycle time of ADC is $1059/fs=24\text{ms}@fs=44.1\text{kHz}$.

After the ALC bit is set to “1” and MIC&ADC block is powered-up, the ALC operation starts from IVOL default value (+30dB).

The time of offset voltage going to “0” after the ADC initialization cycle depends on both the time of analog input pin going to the common voltage and the time constant of the offset cancel digital HPF. This time can be shorter by using the following sequence:

At first, PMVCM and PMMP bits must set to “1”. Then, the ADC should be powered-up. The wait time to power-up the ADC should be longer than 4 times of the time constant that is determined by the AC coupling capacitor at analog input pin and the internal input resistance 60k(typ).

- (8) Power Down MIC and ADC: PMADL = PMADR bits = “1” → “0”

When the registers for the ALC operation are not changed, ALC bit may be keeping “1”. The ALC operation is disabled because the MIC&ADC block is powered-down. If the registers for the ALC operation are also changed when the sampling frequency is changed, it should be done after the AK4645A goes to the manual mode (ALC bit = “0”) or MIC&ADC block is powered-down (PMADL=PMADR bits = “0”). IVOL gain is not reset when PMADL=PMADR bits = “0”, and then IVOL operation starts from the setting value when PMADL or PMADR bit is changed to “1”.

- (9) ALC Disable: ALC bit = “1” → “0”

Headphone-amp Output

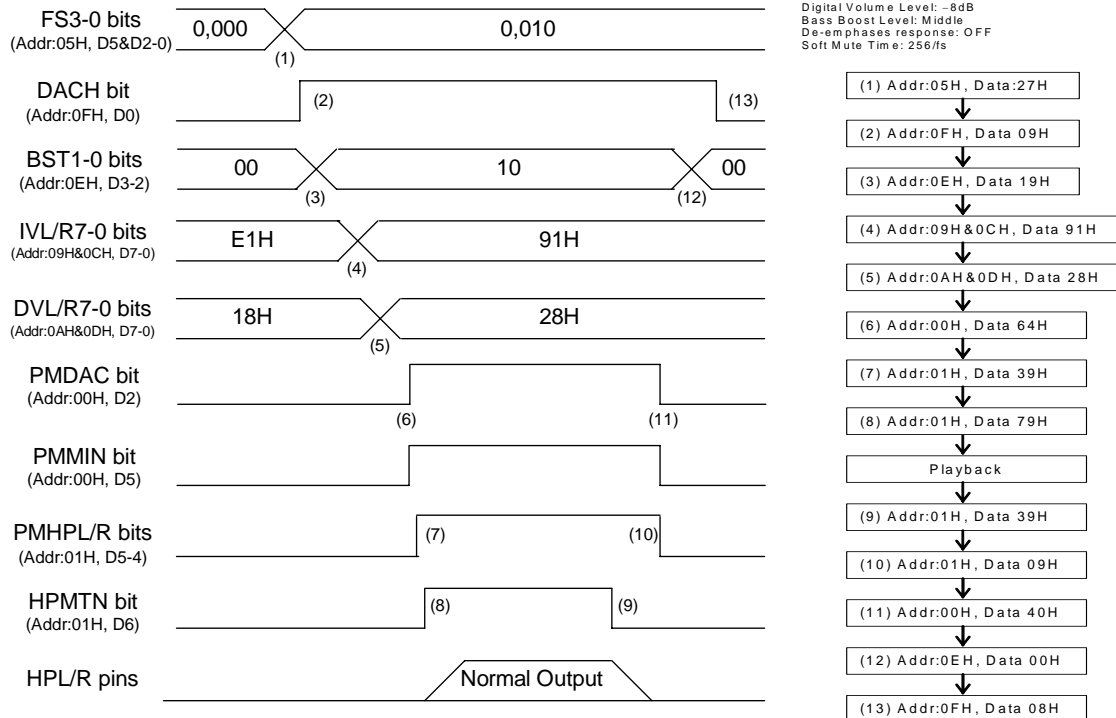


Figure 72. Headphone-Amp Output Sequence

<Example>

At first, clocks must be supplied according to “Clock Set Up” sequence.

- (1) Set up a sampling frequency (FS3-0 bits).
- (2) Set up the path of “DAC → HP-Amp”: DACH bit = “0” → “1”
- (3) Set up the low frequency boost level (BST1-0 bits)
- (4) Set up the input digital volume (Addr: 09H and 0CH)
When PMADL = PMADR bits = “0”, IVL7-0 and IVR7-0 bits must be set to “91H”(0dB).
- (5) Set up the output digital volume (Addr: 0AH and 0DH)
When DVOLC bit is “1” (default), DVL7-0 bits set the volume of both channels. After DAC is powered-up, the digital volume changes from default value (0dB) to the register setting value by the soft transition.
- (6) Power up DAC and MIN-Amp: PMDAC = PMMIN bits = “0” → “1”
The DAC enters an initialization cycle that starts when the PMDAC bit is changed from “0” to “1” at PMADL and PMADR bits are “0”. The initialization cycle time is $1059/fs=24ms@fs=44.1kHz$. During the initialization cycle, the DAC input digital data of both channels are internally forced to a 2's complement, “0”. The DAC output reflects the digital input data after the initialization cycle is complete. When PMADL or PMADR bit is “1”, the DAC does not require an initialization cycle. When ALC bit is “1”, ALC is disable (ALC gain is set by IVL/R7-0 bits) during an initialization cycle ($1059/fs=24ms@fs=44.1kHz$). After the initialization cycle, ALC operation starts from the gain set by IVL/R7-0 bits.
- (7) Power up headphone-amp: PMHPL = PMHPR bits = “0” → “1”
Output voltage of headphone-amp is still HVSS.
- (8) Rise up the common voltage of headphone-amp: HPMTN bit = “0” → “1”
The rise time depends on HVDD and the capacitor value connected with the MUTET pin. When HVDD=3.3V and the capacitor value is 1.0 μ F, the time constant is $\tau_r = 100ms$ (typ), 250ms(max).
- (9) Fall down the common voltage of headphone-amp: HPMTN bit = “1” → “0”
The fall time depends on HVDD and the capacitor value connected with the MUTET pin. When HVDD=3.3V and the capacitor value is 1.0 μ F, the time constant is $\tau_f = 100ms$ (typ), 250ms(max).
If the power supply is powered-off or headphone-Amp is powered-down before the common voltage goes to GND, pop noise occurs. It takes twice of τ_f that the common voltage goes to GND.
- (10) Power down headphone-amp: PMHPL = PMHPR bits = “1” → “0”
- (11) Power down DAC and MIN-Amp: PMDAC = PMMIN bits = “1” → “0”
- (12) Off the bass boost: BST1-0 bits = “00”
- (13) Disable the path of “DAC → HP-Amp”: DACH bit = “1” → “0”

■ Stereo Line Output

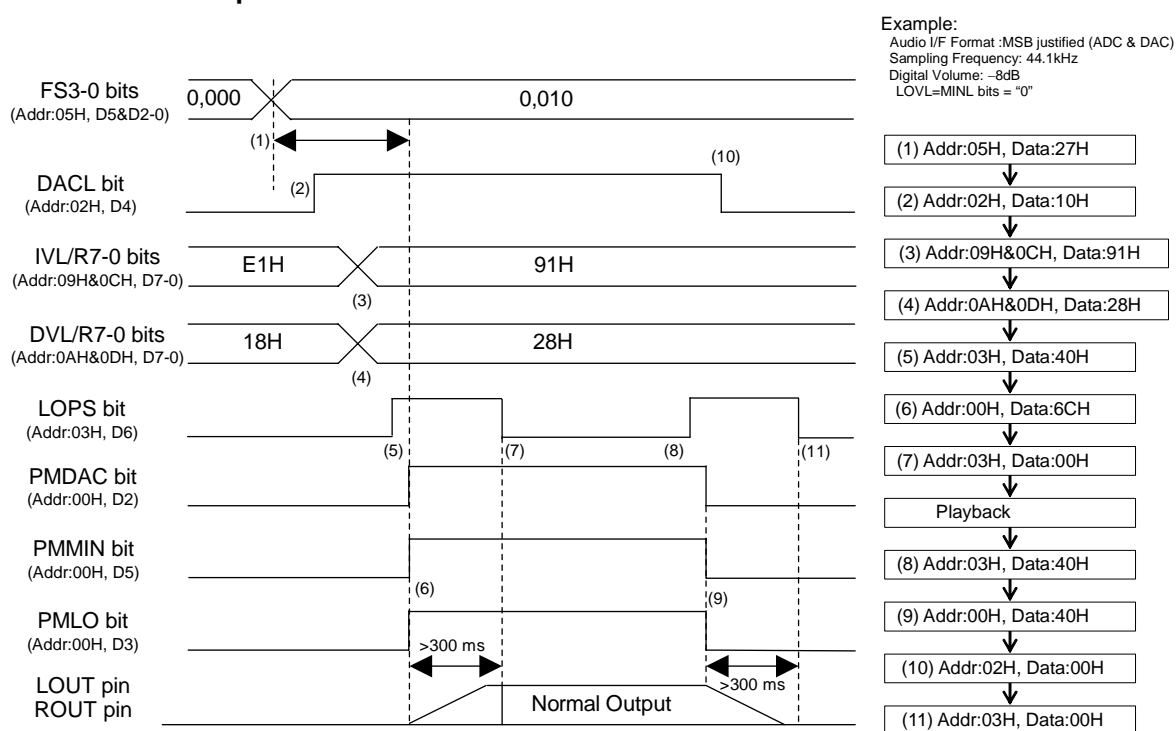


Figure 73. Stereo Lineout Sequence

<Example>

At first, clocks must be supplied according to “Clock Set Up” sequence.

- (1) Set up the sampling frequency (FS3-0 bits).
- (2) Set up the path of “DAC → Stereo Line Amp”: DACL bit = “0” → “1”
- (3) Set up the input digital volume (Addr: 09H and 0CH)
When PMADL = PMADR bits = “0”, IVL7-0 and IVR7-0 bits must be set to “91H”(0dB).
- (4) Set up the output digital volume (Addr: 0AH and 0DH)
When DVOLC bit is “1” (default), DVL7-0 bits set the volume of both channels. After DAC is powered-up, the digital volume changes from default value (0dB) to the register setting value by the soft transition.
- (5) Enter power-save mode of Stereo Line Amp: LOPS bit = “0” → “1”
- (6) Power-up DAC, MIN-Amp and Stereo Line-Amp: PMDAC = PMMIN = PMLO bits = “0” → “1”
The DAC enters an initialization cycle that starts when the PMDAC bit is changed from “0” to “1” at PMADL and PMADR bits are “0”. The initialization cycle time is $1059/fs=24ms@fs=44.1kHz$. During the initialization cycle, the DAC input digital data of both channels are internally forced to a 2's complement, “0”. The DAC output reflects the digital input data after the initialization cycle is completed. When PMADL or PMADR bit is “1”, the DAC does not require an initialization cycle. When ALC bit is “1”, ALC is disable (ALC gain is set by IVL/R7-0 bits) during an initialization cycle ($1059/fs=24ms@fs=44.1kHz$). After the initialization cycle, ALC operation starts from the gain set by IVL/R7-0 bits.
LOUT and ROUT pins rise up to VCOM voltage after PMLO bit is changed to “1”. Rise time is 300ms(max) at $C=1\mu F$ and $AVDD=3.3V$.
- (7) Exit power-save mode of Stereo Line-Amp: LOPS bit = “1” → “0”
LOPS bit should be set to “0” after LOUT and ROUT pins rise up. Stereo Line-Amp goes to normal operation by setting LOPS bit to “0”.
- (8) Enter power-save mode of Stereo Line-Amp: LOPS bit: “0” → “1”
- (9) Power-down DAC, MIN-Amp and Stereo Line-Amp: PMDAC = PMMIN = PMLO bits = “1” → “0”
LOUT and ROUT pins fall down to AVSS. Fall time is 300ms(max) at $C=1\mu F$ and $AVDD=3.3V$.
- (10) Disable the path of “DAC → Stereo Line-Amp”: DACL bit = “1” → “0”
- (11) Exit power-save mode of Stereo Line-Amp: LOPS bit = “1” → “0”
LOPS bit should be set to “0” after LOUT and ROUT pins fall down.

■ Stop of Clock

Master clock can be stopped when ADC and DAC are not in operation.

1. EXT Slave Mode

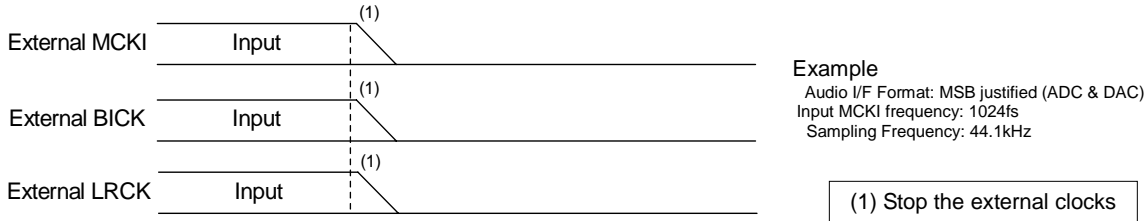


Figure 74. Clock Stopping Sequence (4)

<Example>

(1) Stop the external MCKI, BICK and LRCK clocks.

2. EXT Master Mode

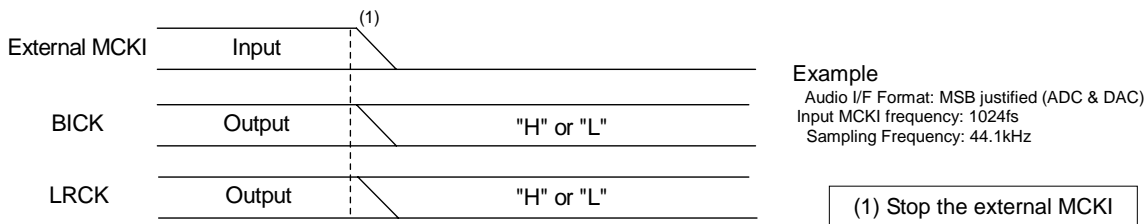


Figure 75. Clock Stopping Sequence (5)

<Example>

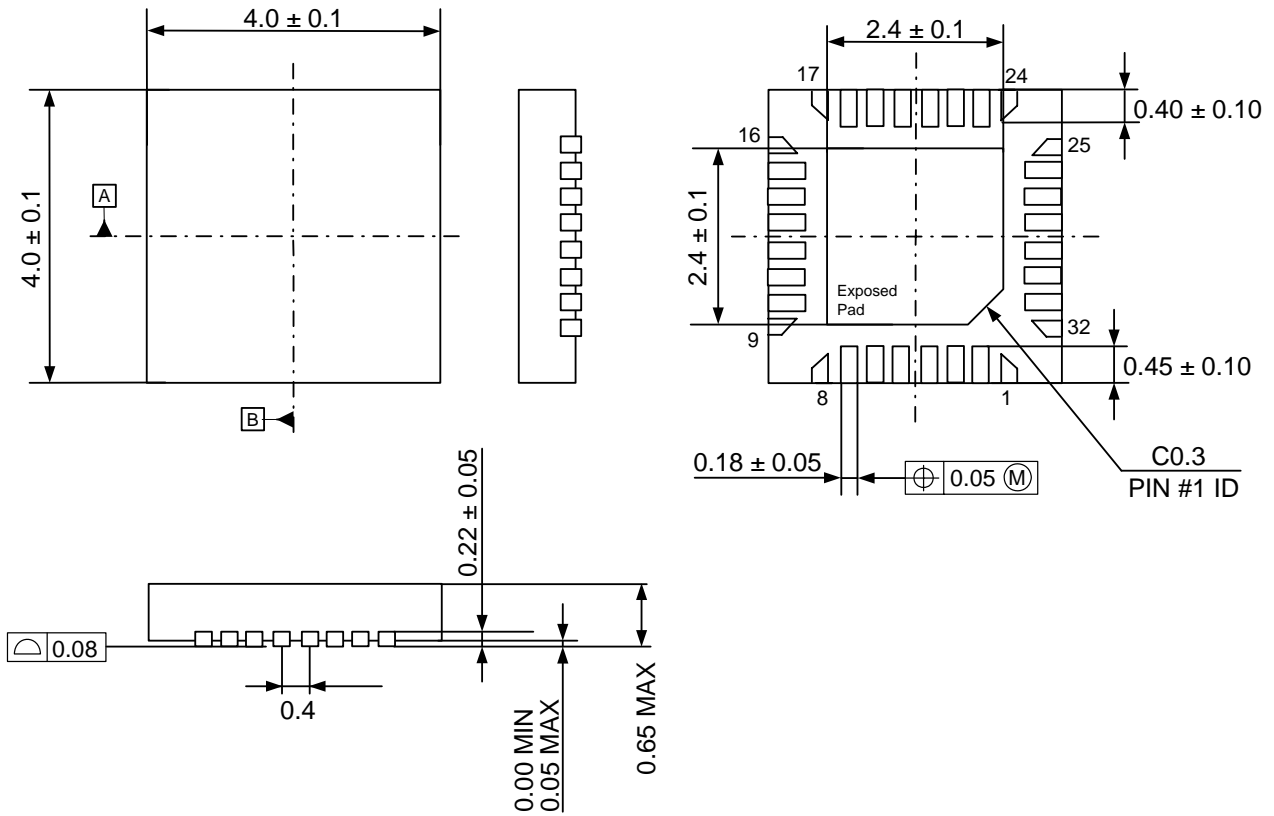
(1) Stop MCKI clock. BICK and LRCK are fixed to "H" or "L".

■ Power down

Power supply current can be shut down (typ. 20 μ A) by stopping clocks and setting PMVCM bit = "0" after all blocks except for VCOM are powered-down. Power supply current can also be shut down (typ. 1 μ A) by stopping clocks and setting the PDN pin = "L". When the PDN pin = "L", the registers are initialized.

PACKAGE

● 32pin QFN (Unit: mm)

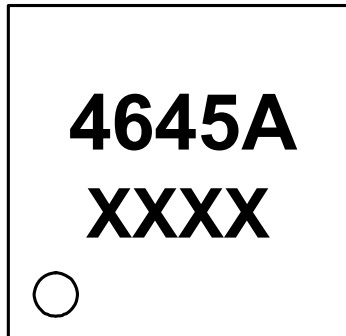


Note) The exposed pad on the bottom surface of the package must be open or connected to the ground.

■ Material & Lead finish

Package molding compound:	Epoxy
Lead frame material:	Cu
Lead frame surface treatment:	Solder (Pb free) plate

MARKING



1

XXXX: Date code (4 digit)
Pin #1 indication

REVISION HISTORY

Date (YY/MM/DD)	Revision	Reason	Page	Contents
08/07/31	00	First Edition		

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