

1X/1.5X/2X Charge Pump White LED Driver for Main and Sub-Displays

#### **Features**

- ±1.5% LED Current Matching
- · Powers Main and Sub-Display LEDs
- High Efficiency Up to 90% Over Li-ion Battery
   Discharge
- Output Current Up to 30mA per LED
- 2.7V to 5.5V Operating Voltage
- · 1x/1.5x/2x Charge Pump Modes
- Low Shutdown Current: 2mA Maximum
- · Low Input Ripple and EMI
- · Internal Soft-Start Limits Inrush Current
- Short Circuit Current Limit
- Thermal Shutdown Protection
- Output Over-Voltage Protection
- 16-pin QFN Package
- Lead Free and Green Devices Available (RoHS Compliant)

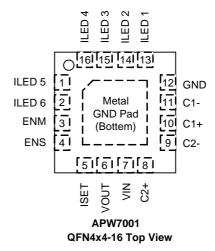
## **Applications**

- · Cellular Phone White LED Back Light
- Portable Device
- PDA, Handheld Computer
- · DSC

## **General Description**

The APW7001 is a high efficiency charge pump white LED driver; the device drives up to four white LEDs in the main display and up to two white LEDs in the sub-display with regulated constant current for uniform intensity. The supply voltage ranges from 2.7V to 5.5V and it is optimized for a Li-ion battery application. The APW7001 operates in 1x, 1.5x, and 2x charge pump modes and automatically switches the charge pump modes depend on the input voltage to maintain the required power for high power efficiency. The APW7001 provides up to 30mA per LED and allows several methods such as the PWM signals on the ENM pin for main-display dimming and on the ENS pin for sub-display dimming. The two control logic pins, ENM and ENS, allow disabling or enabling the main and sub-displays. The supply current is only 2mA in 2x mode, and the ENM and ENS are kept low for 20ms will allow the device to enter shutdown mode with 2µA quiescent current. The APW7001 features current limit and short circuit protection. The APW7001 switches at 1MHz frequency and only requires four 1µF ceramic capacitors and one resistor, and ensures low input current ripple and EMI. The APW7001 is available in a 16-pin QFN package.

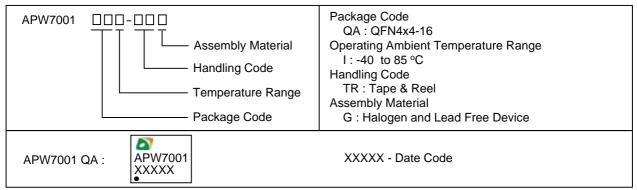
## **Pin Configuration**



ANPEC reserves the right to make changes to improve reliability or manufacturability without notice, and advise customers to obtain the latest version of relevant information to verify before placing orders.



## Ordering and Marking Information



Note: ANPEC lead-free products contain molding compounds/die attach materials and 100% matte tin plate termination finish; which are fully compliant with RoHS. ANPEC lead-free products meet or exceed the lead-free requirements of IPC/JEDEC J-STD-020D for MSL classification at lead-free peak reflow temperature. ANPEC defines "Green" to mean lead-free (RoHS compliant) and halogen free (Br or Cl does not exceed 900ppm by weight in homogeneous material and total of Br and Cl does not exceed 1500ppm by weight).

## Absolute Maximum Ratings (Note 1)

Symbol	Parameter	Rating	Unit
$V_{OUT}$	VOUT to GND	-0.3 to +6	V
V <sub>IN</sub>	VIN to GND	-0.3 to +6	V
V <sub>C1+</sub> , V <sub>C1-</sub> , V <sub>C2+</sub> , V <sub>C2-</sub>	C1+, C1-, C2+, C2- to GND	-0.3 to +6	V
V <sub>ILED1-6</sub>	ILED1-6 to GND	-0.3 to +6	V
$V_{\text{ENM}}, V_{\text{ENS}}$	ENM, ENS to GND	-0.3 to +6	V
V <sub>ISET</sub>	ISET to GND	-0.3 to 2	V
TJ	Maximum Junction Temperature	+150	°C
T <sub>STG</sub>	Storage Temperature	-65 ~ 150	°C
$T_{SDR}$	Maximum Lead Soldering Temperature, 10 Seconds	260	°C

Note 1: Absolute Maximum Ratings are those values beyond which the life of a device may be impaired. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

### Thermal Characteristics (Note 2)

Symbol	Parameter	Typical Value	Unit
$\theta_{JA}$	Thermal Resistance -Junction to Ambient  QFN4x4-16	40	°C/W

Note 2:  $\theta_{JA}$  is measured with the component mounted on a high effective thermal conductivity test board in free air.

## **Recommended Operating Conditions**

Symbol	Parameter	Rating	Unit
$V_{IN}$	Input Voltage	2.8 to 4.5	V
V <sub>OUT</sub>	Output Voltage	3 to 4	V
I <sub>LED</sub>	LED Current	5 to 30	mA



# **Recommended Operating Conditions (Cont.)**

Symbol	Parameter	Rating	Unit
	Output Current, V <sub>IN</sub> >3.5V, V <sub>F</sub> =3.1V, 1x mode	180	
I <sub>OUT</sub>	Output Current, 3.5V <v<sub>IN&gt;3.1V, V<sub>F</sub>=3.1V, 1.5x mode</v<sub>	120	mA
	Output Current, 3.1V <v<sub>IN&gt;2.8V, V<sub>F</sub>=3.1V, 2X mode</v<sub>	90	
T <sub>A</sub>	Ambient Temperature	-40 to 85	°C

### **Electrical Characteristics**

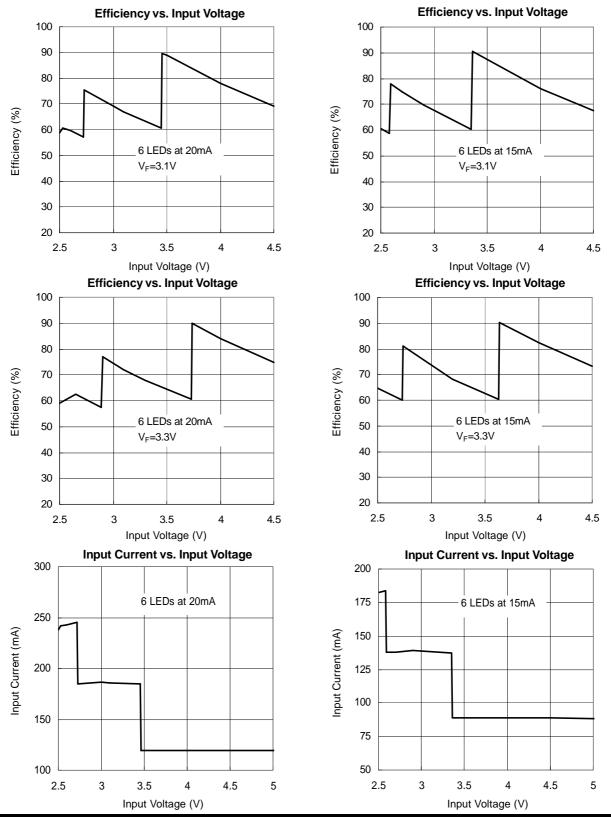
 $V_{_{IN}}=2.85~to~5.5V,~C_{_{IN}}=C_{_{OUT}}=C1=C2=1\mu F~(ESR=0.03\Omega),~I_{_{LED}}=20mA,~T_{_{A}}=-40^{\circ}C~to~+85^{\circ}C,~unless~otherwise~noted.~Typical~values~are~at~T_{_{A}}=+25^{\circ}C.$ 

Council of	Baramatar	Took Comditions		APW7001		Unit	
Symbol	Parameter	Test Conditions	Min.	Тур.	Max.		
V <sub>IN</sub>	Input Voltage		2.7	-	5.5	V	
$V_{\text{UVLO}}$	Under-voltage Lockout Threshold	V <sub>IN</sub> falling	2.2	2.4	2.6	V	
	Under-voltage Lockout Hysteresis		-	50	-	mV	
		in 1.5x/2x mode	-	2	4	mA	
IQ	Quiescent Current	No switching in 1x mode	-	0.5	1	mA	
		EN=0	-	0.1	2	μΑ	
I <sub>LED-ERR</sub>	LED Current Accuracy	5mA <i<sub>LED&lt;30mA <sup>(Note 3)</sup></i<sub>	-	±2	±8	%	
	Current Matching	5mA <i<sub>LED&lt;30mA <sup>(Note 4)</sup></i<sub>	-	±1.5	±5	%	
I <sub>ISET</sub>	ISET Current		5	-	1000	μΑ	
	ISET to LED Current Ratio	$I_{LED}$ / (1.2V/R <sub>SET</sub> ) 5mA< $I_{LED}$ <30mA, T <sub>A</sub> = +25°C	370	400	420		
$V_{ILED-TH}$	ILED Threshold Voltage	V <sub>ILED</sub> falling	-	100	-	mV	
	1.5x mode to 1x mode Transition Hysteresis	V <sub>IN</sub> rising, V <sub>IN</sub> -V <sub>OUT</sub>	-	300	-	mV	
	2x mode to 1.5x mode Transition Hysteresis	V <sub>IN</sub> rising, V <sub>OUT</sub> -V <sub>IN</sub>	-	300	-	mV	
Fosc	Switching Frequency		0.8	1	1.2	MHz	
		1x mode (V <sub>IN</sub> -V <sub>OUT</sub> ) / I <sub>OUT</sub>	-	1.6	3		
$R_{\text{OUT}}$	Open Loop VOUT Resistance	1.5x mode (1.5xV <sub>IN</sub> -V <sub>OUT</sub> ) / I <sub>OUT</sub>	-	7	12	Ω	
		2x mode (2xV <sub>IN</sub> -V <sub>OUT</sub> ) / I <sub>OUT</sub>	-	16	28		
I <sub>SHORT</sub>	Short Circuit Current Limit	V <sub>OUT</sub> < 1V	-	40	-	mA	
$V_{OVP}$	Output Over Voltage Protection		5	5.5	6	V	
$V_{IH}$	Logic Pins High Threshold		1.3	0.7	-	V	
V <sub>IL</sub>	Logic Pins Low Threshold		-	0.6	0.3	V	
I <sub>IH</sub>	Logic Pins High Current	V <sub>IH</sub> =VIN	-	-	1	μΑ	
I <sub>IL</sub>	Logic Pins Low Current	V <sub>IL</sub> =GND	-	-	1	μΑ	
	Thermal Shutdown		-	150	-	°C	
	Thermal Shutdown Hysteresis		-	20	-	°C	

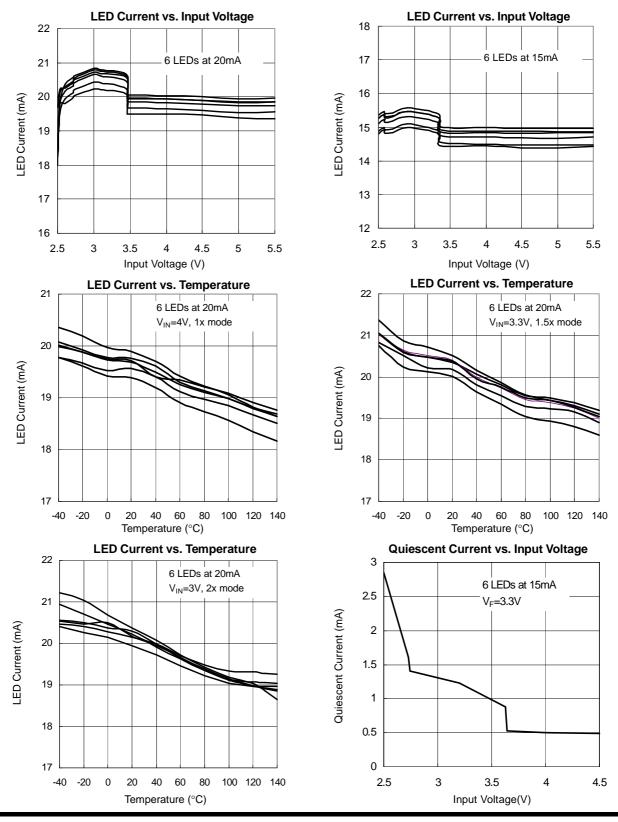
Note 3: LED current accuracy is defined as:  $\pm (I_{\text{LED-MEASURED}} - I_{\text{LED-SET}}) / I_{\text{LED-SET}}$ Note 4: LED current matching is defined as:  $\pm (I_{\text{LED-MAX}} - I_{\text{LED-MIN}}) / (I_{\text{LED-MAX}} + I_{\text{LED-MIN}})$ 



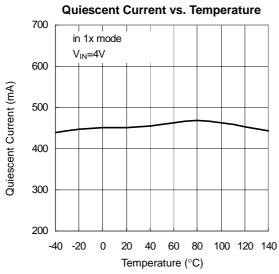
# **Typical Operating Characteristics**

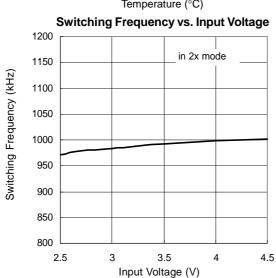


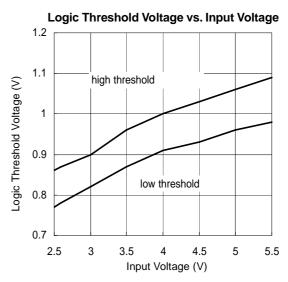


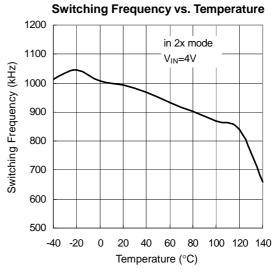






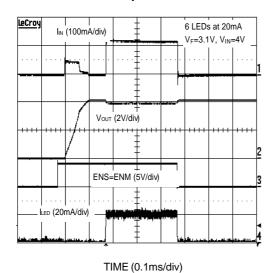




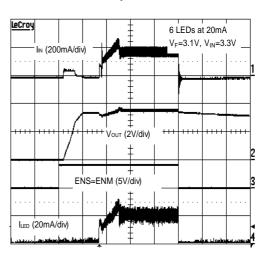




### Start Up in 1x Mode

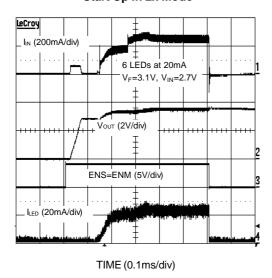


Start Up in 1.5x Mode

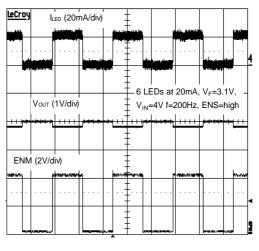


TIME (0.1ms/div)

### Start Up in 2x Mode



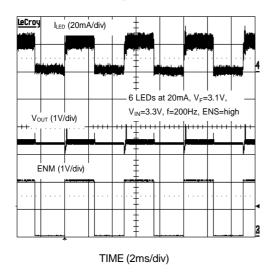
### **Dimming in 1x Mode**



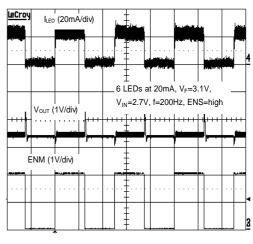
TIME (2ms/div)



Dimming in 1.5x Mode

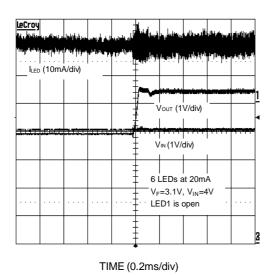


### Dimming in 2x Mode

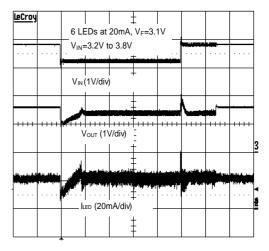


TIME (2ms/div)

### **OVP Even with LED Open Circuit**



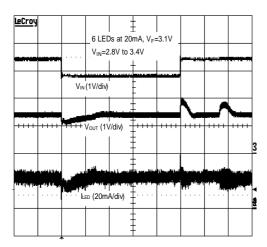
Line Transient Response in 1x to 1.5x Mode



TIME (0.1ms/div)



### Line Transient Response in 1.5x to 2x Mode



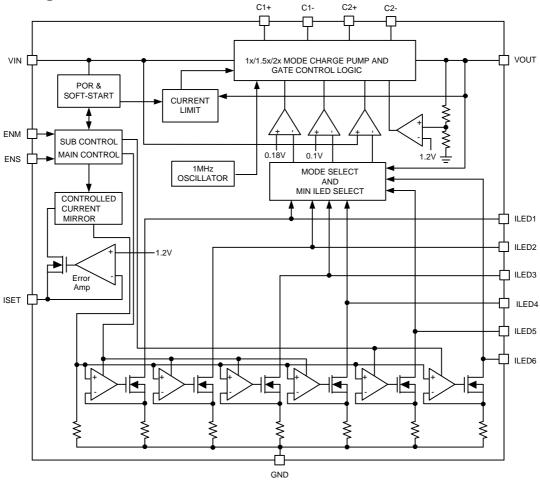
TIME (0.1ms/div)

# **Pin Description**

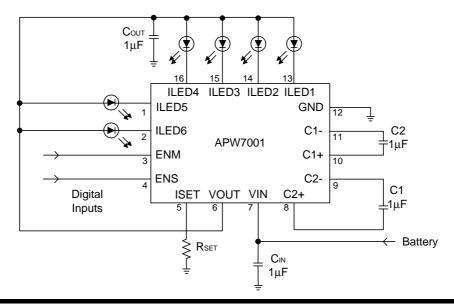
Р	IN	FUNCTION				
NO.	NAME	FUNCTION				
1	ILED5	Sub-Display LEDs Cathode Connection. The LED current flows from VOUT through LED into ILED_pin. The				
2	ILED6	charge pump regulates the lowest V <sub>ILED</sub> to 180mV. Connect ILED_ pin to VOUT if the LED is not used.				
3	ENM	On/Off and Dimming Control for LED1-4 (Main-Display).				
4	ENS	On/Off and Dimming Control for LED5-6 (Sub-Display).				
5	ISET	LED Current Set Input. Connect a resistor from ISET to GND to set the LED current. VISET is typically 1.2V.				
6	VOUT	Output Voltage Pin. Connect VOUT to the LED anodes. Connect a 1µF capacitor from VOUT to GND.				
7	VIN	Supply Voltage Input Pin. Connect a 1μF capacitor from VIN to GND.				
8	C2+	Bucket Capacitor1 Positive Terminal. Connect a 1μF capacitor from C2+ to C2				
9	C2-	Bucket Capacitor1 Negative Terminal. Connect a 1μF capacitor from C2+ to C2				
10	C1+	Bucket Capacitor1 Positive Terminal. Connect a 1μF capacitor from C1+ to C1				
11	C1-	Bucket Capacitor1 Negative Terminal. Connect a 1μF capacitor from C1+ to C1				
12	GND	Device Ground Pin.				
13	ILED1					
14	ILED2	Main-Display LEDs Cathode Connection. The LED current flows from VOUT through LED into ILED_ pin.				
15	ILED3	The charge pump regulates the lowest V <sub>ILED</sub> to 180mV. Connect ILED_ pin to VOUT if the LED is not use				
16	ILED4					



## **Block Diagram**



# **Typical Application Circuit**





## **Function Description**

#### Soft-Start

The APW7001 provides the soft-start function to limit the inrush current during startup. When the input voltage is supplied to the device and exceeds the UVLO voltage, the output capacitor is charged directly from input with a limited current source. Approximate 100µs after the output voltage approaches the input voltage, the device starts to provide the programmed LED current and determines which of 1x, and 1.5x, or 2x mode is required. When the programmed LED current can be reached with 1x mode, the soft-start is completed and the device operates in 1x mode. When the programmed LED current cannot be reached, the charge pump goes into 1.5x mode. If the 1.5x mode charge pump cannot suffice for the LED current need, the charge pump will switch to 2x mode.

#### **Mode Transition**

The APW7001 operates in 1x, 1.5x, and 2x charge pump modes and automatically switches the charge pump modes depend on the input voltage to maintain the required power for high power efficiency. If the APW7001 operates in 1x mode, the VOUT is pulled up to V<sub>IN</sub>. When  $V_{IN}$  decreases, the  $V_{ILED}$  will decease to maintain the regulated LED current. Until  $V_{\text{ILED}}$  is below 100mV, the device will switch to 1.5x mode. In 1.5x mode, the VILED is regulated to 0.18V, and the output voltage is VF+0.18V. If  $V_{_{\rm IN}}$ continues to decrease until  $V_{ILED}$  is below 100mV again, the device will switch to 2x mode. When the  $V_{IN}$  rises and reaches by approximately VOUT-300mV, the APW7001 switches back to 1.5x mode. If the  $V_{IN}$  continues to rise and reaches by approximately VOUT+300mV, the APW7001 switches back to 1x mode. The 2x charge pump is enough to suffice the White LED for a Li-ion battery application. The APW7001 ensures that in the 1x mode for as long as possible to increase the efficiency and extend the operating range by using the 2x mode. The transition voltages from 1x to 1.5x, and 1.5x to 2x are given by:

$$V_{TRANS1X} = V_F + 0.1V + (I_{OUT} \times R_{OUT1X})$$
$$V_{TRANS1.5X} = [V_F + 0.1V + (I_{OUT} \times R_{OUT1.5X})] / 1.5$$

where

 $V_{\scriptscriptstyle F}$  is the forward voltage of LED

I is the output current

 $R_{OUT1X}$  is the output impedance in 1x mode = 1.6 $\Omega$ 

 $R_{\text{OUT1.5X}}$  is the output impedance in 1.5x mode =  $7\Omega$ 

#### **ENM/ENS Control Logic Pins**

The APW7001 provides two logic input pins to enable or disable the main-display and sub-display. When the ENM or ENS is high, the VOUT is supplied and the respective LEDs are enabled. When ENM or ENS is low, the respective LEDs are disabled. If both logic pins are kept low for more than 20ms, the APW7001 enters shutdown mode. In shutdown mode, all internal control circuits are turned off and the quiescent current is below  $2\mu A$ . When the device exits shutdown mode, the output has the soft-start function as the input voltage startup.

ENM	ENS	LED Status
0 (for more than 20ms)	0 (for more than 20ms)	IC enters shutdown
1	0	LED1, 2, 3, 4 are turned on / LED5, 6 are turned off
0	1	LED1, 2, 3, 4 are turned off / LED5, 6 are turned on
1	1	LED1, 2, 3, 4 are turned on / LED5, 6 are turned on

Table 1. The Truth Table of ENM and ENS

#### **LED Current Setting**

Connect a resistor from ISET pin to GND to set the LED current. The ISET voltage is 1.2V, and the LED current is typically 400 times the current through the ISET resistor. The LED current is given by:

$$R_{SET} = \frac{400 \times 1.2 V}{I_{LED}}$$

The APW7001 provides up to 30mA of LED current per LED and the device has a max current matching of  $\pm 5\%$  between any two LED currents and a max current accuracy of  $\pm 8\%$ . If high accuracy is required, using a 1% precision surface mount resistor for the need.

ILED (mA)	RSET (kW)
5	92
10	47
15	32
20	24
30	16.5

Table 2.  $\rm R_{\rm SET}$  Value Selection



# **Function Description (Cont.)**

### **LED Current Setting (Cont.)**

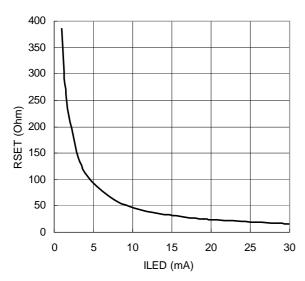


Figure 1.  $R_{\text{SET}}$  Value vs. LED Current

### **Over-Voltage Protection**

If any of LEDs is failed or unused, LED channel is not connected to VOUT, the charge pump mode will go into 2x mode and the output voltage will be pumped to 2 times the input voltage. If the output voltage is over 5.5V, the over-voltage protection circuit will limit the output voltage to approximately 5.5V.



## Application Information

### **Capacitor Selection**

For lower input and output voltage ripples, both input and output capacitors should be larger values and lower ESR capacitors. However, the larger output capacitor values will increase the soft-start time. The lower charge pump flying capacitors values and ESR improve the efficiency, but lower capacitor values may limit the LED's currents at low input voltage.

It is recommended that the low ESR and low variation over temperature, such as the ceramic capacitors with X7R or X5R and the value is  $1\mu F$  for the input capacitor, output capacitor, and the charge pump flying capacitors.

#### **Brightness Control**

#### 1. PWM dimming using ENM, or/and EMS

The first method for dimming the LEDs is to apply a PWM signal into the ENM or/and ENS pins. Figure 2 shows the application circuit. The average LED current is proportional to the PWM signal duty cycle. Note that the frequency of PWM signal will affect the minimum dimming duty. Figure 3 shows the LED current vs. dimming frequency and dimming duty, the recommend dimming frequency is below 10kHz. The PWM signal can either be applied to ENM or ENS, or both inputs can be tied together and the PWM signal can be applied to both pins. Table 3 shows the truth table of ENM and ENS dimming control. The average LED current is calculated by the following equation:

$$I_{LED(avg)} = \frac{ton \times I_{LED(max)}}{ton + toff}$$

#### Where:

 $I_{\text{LED(max)}}$  is programmed LED current by  $I_{\text{SET}}$  pin toff is the off time of the PWM signal ton is the on time of the PWM signal

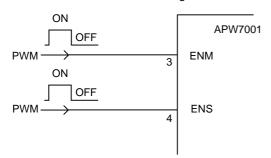


Figure 2. PWM Dimming Application Circuit

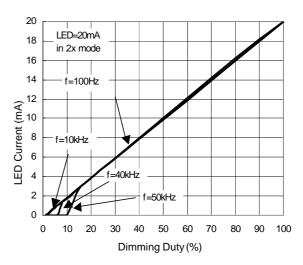


Figure 3. PWM Dimming Frequency vs. LED Current

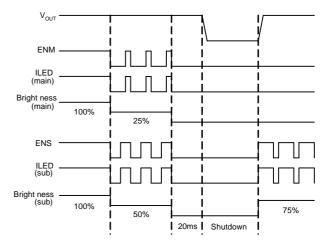


Figure 4. ENM/ENS Timing

ENM	ENS	LED Status
0	PWM	LED1, 2, 3, 4 are turned off LED5, 6 are PWM dimming
PWM	0	LED1, 2, 3, 4 are PWM dimming LED5, 6 are turned off
1	PWM	LED1, 2, 3, 4 are turned on LED5, 6 are PWM dimming
PWM	1	LED1, 2, 3, 4 are PWM dimming LED5, 6 are turned on
PWM	PWM	LED1, 2, 3, 4 are PWM dimming LED5, 6 are PWM dimming

Table 3. The Truth Table of ENM and ENS Dimming Control



## **Application Information (Cont.)**

### **Brightness Control (Cont.)**

#### 2. Analog Dimming With Analog Voltage

The second method for dimming the LEDs is to apply a voltage through a resistor into the ISET pin. The variation of LED current is proportional to the variation of the analog voltage. If the resistor values are chosen correctly, the analog control voltage varies the output current from 0mA to full LED current. Figure 5 shows the application circuit, and the LED current is calculated by the following equation:

$$I_{LED} = 400 \times \left(\frac{1.2V - V_{ADJ}}{R2} + \frac{1.2V}{R1}\right)$$

$$APW7001$$

$$ISET$$

$$V_{ADJ}$$

$$R2$$

$$R1$$

Figure 5. Analog Voltage Dimming Application Circuit

3. Digital dimming with external NMOS transistors
The third method for dimming the LEDs is to change the
equivalent resistance for RSET with the external NMOS
transistors. The equivalent resistance is the parallel combinations of the R1, R2, R3, and R4. R4 is always connected and selected for the minimum LED current. Figure 6 shows the application circuit.

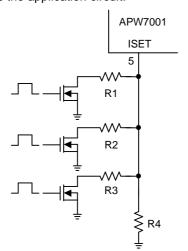


Figure 6. Digital Dimming Application Circuit

#### 4. PWM dimming with EN pin

Another method for dimming the LEDs is to apply a PWM signal into the EN pin. The average LED current is proportional to the PWM signal duty cycle. Note that the frequency of PWM signal will affect the minimum dimming duty. The recommend dimming frequency is between 100Hz and 1kHz. The average LED current is calculated by the following equation:

$$I_{LED(avg)} = \frac{ton \times I_{LED(max)}}{ton + toff}$$

#### Where

 $I_{\text{LED}(\text{max})}$  is programmed LED current by ISET pin toff is the off time of the PWM signal ton is the on time of the PWM signal

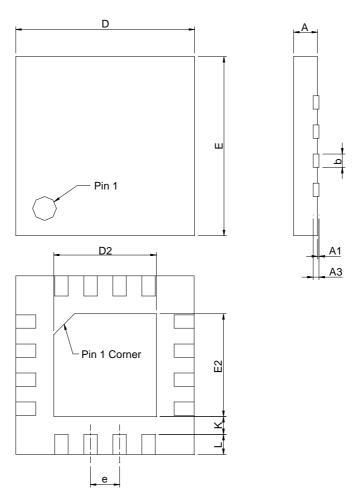
#### **Layout Consideration**

The APW7001 is a high frequency charge pump for white LED driver and requires some care when laying out the printed circuit board. The metal GND pad of the bottom of the package must be soldered to the PCB and connected to the GND plane on the backside through several thermal vias. Place the CIN, COUT, C1, and C2 as close to IC as possible for reducing the switching noise.



# Package Information

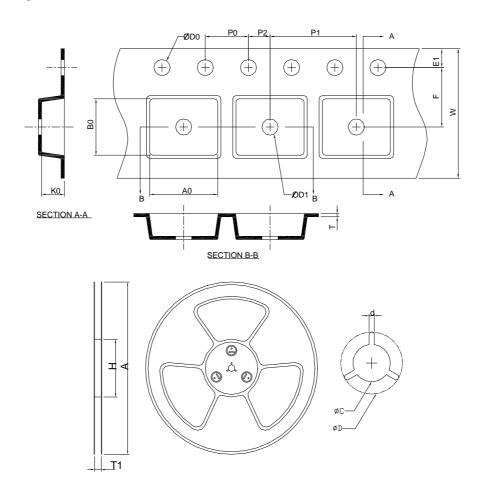
### QFN4x4-16



Ş		x4-16					
SYMBOL	MILLIM	MILLIMETERS		MILLIMETERS		NCHES	
P	MIN.	MAX.	MIN.	MAX.			
Α	0.80	1.00	0.031	0.039			
A1	0.00	0.05	0.000	0.002			
А3	0.20	REF	0.00	8 REF			
b	0.25	0.35	0.010	0.014			
D	3.90	4.10	0.154	0.161			
D2	2.50	2.80	0.098	0.110			
Е	3.90	4.10	0.154	0.161			
E2	2.50	2.80	0.098	0.110			
е	0.65 BSC		0.020	6 BSC			
L	0.30	0.50	0.012	0.020			
K	0.20		0.008				



# **Carrier Tape & Reel Dimensions**



Application	Α	Н	T1	С	d	D	W	E1	F
	330.0 €.00	50 MIN.	12.4+2.00 -0.00	13.0+0.50 -0.20	1.5 MIN.	20.2 MIN.	12.0 ±0.30	1.75 ±0.10	5.5 <b>±</b> 0.05
QFN4x4-16	P0	P1	P2	D0	D1	Т	A0	В0	K0
	4.0 <b>±</b> 0.10	8.0 <b>±</b> 0.10	2.0 ±0.05	1.5+0.10 -0.00	1.5 MIN.	0.6+0.00 -0.40	4.30 ±0.20	4.30 <b>±</b> 0.20	1.30 ±0.20

(mm)

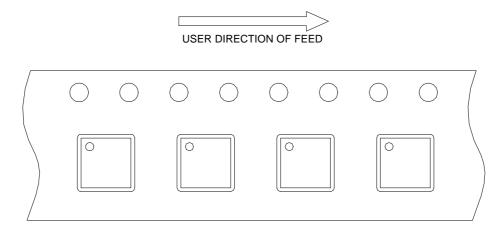
## **Devices Per Unit**

Package Type	Unit	Quantity
QFN4x4-16	Tape & Reel	3000

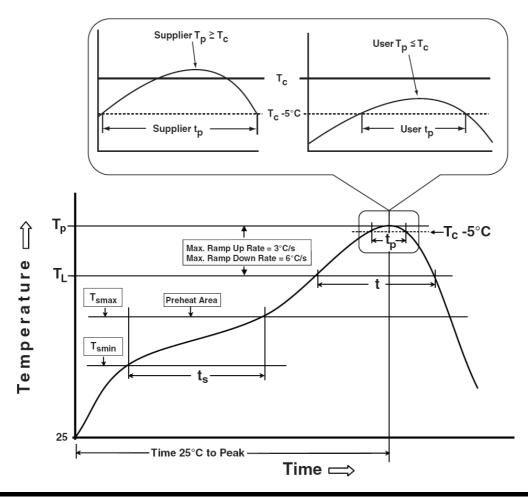


# **Taping Direction Information**

QFN4x4-16



## **Classification Profile**





## **Classification Reflow Profiles**

Profile Feature	Sn-Pb Eutectic Assembly	Pb-Free Assembly	
Preheat & Soak Temperature min (T <sub>smin</sub> ) Temperature max (T <sub>smax</sub> ) Time (T <sub>smin</sub> to T <sub>smax</sub> ) (t <sub>s</sub> )	100 °C 150 °C 60-120 seconds	150 °C 200 °C 60-120 seconds	
Average ramp-up rate (T <sub>smax</sub> to T <sub>P</sub> )	3 °C/second max.	3°C/second max.	
Liquidous temperature (T <sub>L</sub> ) Time at liquidous (t <sub>L</sub> )	183 °C 60-150 seconds	217 °C 60-150 seconds	
Peak package body Temperature (T <sub>p</sub> )*	See Classification Temp in table 1	See Classification Temp in table 2	
Time (t <sub>P</sub> )** within 5°C of the specified classification temperature (T <sub>c</sub> )	20** seconds	30** seconds	
Average ramp-down rate (T <sub>p</sub> to T <sub>smax</sub> )	6 °C/second max.	6 °C/second max.	
Time 25°C to peak temperature	6 minutes max.	8 minutes max.	
* Tolorance for peak profile Tomporat	ure (T ) is defined as a supplier minimu	m and a usor maximum	

<sup>\*</sup> Tolerance for peak profile Temperature (Tp) is defined as a supplier minimum and a user maximum.

Table 1. SnPb Eutectic Process – Classification Temperatures (Tc)

Package	Volume mm <sup>3</sup>	Volume mm <sup>3</sup>
Thickness	<350	<sup>3</sup> 350
<2.5 mm	235 °C	220 °C
≥2.5 mm	220 °C	220 °C

Table 2. Pb-free Process – Classification Temperatures (Tc)

Package Thickness	Volume mm <sup>3</sup> <350	Volume mm <sup>3</sup> 350-2000	Volume mm <sup>3</sup> >2000
<1.6 mm	260 °C	260 °C	260 °C
1.6 mm – 2.5 mm	260 °C	250 °C	245 °C
≥2.5 mm	250 °C	245 °C	245 °C

## **Reliability Test Program**

Test item	Method	Description
SOLDERABILITY	JESD-22, B102	5 Sec, 245°C
HOLT	JESD-22, A108	1000 Hrs, Bias @ 125°C
PCT	JESD-22, A102	168 Hrs, 100%RH, 2atm, 121°C
ТСТ	JESD-22, A104	500 Cycles, -65°C~150°C
НВМ	MIL-STD-883-3015.7	VHBM 2KV
MM	JESD-22, A115	VMM 200V
Latch-Up	JESD 78	10ms, 1 <sub>tr</sub> 100mA

<sup>\*\*</sup> Tolerance for time at peak profile temperature  $(t_p)$  is defined as a supplier minimum and a user maximum.



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