
ARM[®]-based 32-bit Cortex[®]-M4 MCU 64 KB Flash, sLib, 10 timers, 1 ADC, 1 CMP, 2 OP, 7 communication interfaces

Features

- **Core: ARM[®] 32-bit Cortex[®]-M4 CPU**
 - 120 MHz maximum frequency, with a memory protection unit (MPU), single-cycle multiplication and hardware division
 - DSP instructions
- **Memories**
 - 64 KB of internal Flash memory
 - 4 Kbytes of boot memory as a Bootloader
 - sLib: configurable part of main Flash as a library area with code executable but secured, non-readable
 - 16 Kbytes of SRAM
- **Power control (PWC)**
 - 2.4 to 3.6 V power supply
 - Power-on reset (POR), low voltage reset (LVR) and power voltage monitoring (PVM)
 - Low power modes: Sleep, DeepSleep and Standby modes, 4 WKUP pins waking up Standby mode
 - Supports 5 x 32-bit battery powered registers (BPR)
- **Clock and reset management (CRM)**
 - 4 to 25 MHz crystal oscillator (HEXT)
 - Internal 48 MHz factory-trimmed HICK ($\pm 1\%$ at TA=25 °C, $\pm 2\%$ at TA= -40 to +105 °C)
 - Configurable PLL 31 to 500 multiplication factors and 1 to 15 division factors
 - 32 kHz crystal (LEXT)
 - Low speed internal clock (LICK)
- **Analog**
 - 1 x 12-bit 2 MSPS A/D converter, up to 13 external input channels and 5 input channels
 - Temperature sensor (V_{TS})
 - Internal reference voltage (V_{INTRV})
 - 1 x CMP with 4 external input channels
 - 2 x Operational Amplifiers (OP)
- **DMA: 1x 5-channel DMA controller**
- **Up to 39 fast GPIOs**
 - All mappable on 16 external interrupts (EXINT)
 - Almost all 5 V-tolerant
- **Up to 10 timers (TMR)**
 - 1 x 16-bit 7-channel advanced timer, including 3 pairs of complementary PWM outputs with dead-time generator and emergency break
 - Up to 5 x 16-bit timers, each with 4 IC/OC /PWM or pulse counter and quadrature (incremental) encoder input
 - 1 x 16-bit basic timer
 - 2 x watchdog timers (general WDT and windowed WWDT)
 - SysTick timer: a 24-bit downcounter
- **ERTC: enhanced RTC, with alarm, subsecond accuracy, hardware calendar and calibration**
- **Up to 7 communication interfaces**
 - 2 x I²C interfaces (SMBus/PMBus)
 - 2 x USARTs supporting master synchronous SPI and modem control, with ISO7816 interface, LIN, IrDA capability and swappable TX/RX
 - 2 x SPIs (36 Mbit/s), all with multiplexed half-duplex I²S
 - Infrared transmitter (IRTMR)
- **CRC calculation unit**
- **96-bit unique ID (UID)**
- **Serial wire debug (SWD) interface**
- **Operating temperature: -40 to +105 °C**
- **Packages**
 - LQFP48 7 x 7 mm

Table 1. Device summary

Internal Flash	Part number
64 Kbytes	AT32F4212C8T7

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1 Descriptions

The AT32F4212 series is based on the high-performance ARM®Cortex®-M4 32-bit RISC core operating up to 120 MHz. The Cortex®-M4 core features a full set of DSP instructions and a memory protection unit (MPU).

The AT32F4212 series incorporates high-speed embedded memories (up to 64 Kbytes of internal Flash memory and 16 Kbytes of SRAM), a wide range of enhanced GPIO ports and peripherals connected to two APB buses. Any block of the embedded Flash memory can be protected by the “sLib”, functioning as a security area with code-executable only.

The AT32F4212 series offers one 12-bit ADC, one analog comparator (CMP), two operational amplifiers, five general-purpose 16-bit timers, one advanced timer and one low-power ERTC. It also features standard and advanced communication interfaces: up to two I²C interfaces, two SPIs (all multiplexed as I²S), two USARTs and one infrared transmitter.

The AT32F4212 device operates in the -40 to +105 °C temperature range, from a 2.4 to 3.6 V power supply. A comprehensive set of power-saving mode allows the design of low-power application.

Table 2. AT32F4212 features and peripheral counts

Part number		AT32F4212C8T7
CPU frequency (MHz)		120
Flash (Kbytes)		64
SRAM (Kbytes)		16
Timers	Advanced	1
	16-bit general-purpose	5
	Basic	1
	SysTick	1
	WDT	1
	WWDT	1
	ERTC	1
Communication	I ² C	2
	SPI ⁽¹⁾	2
	I ² S (half-duplex) ⁽¹⁾	2
	USART+UART	2+0
	IR	1
Analog	12-bit ADC numbers/ external channels	1 13
	Comparator	1
	Operational amplifier	2
	GPIO	39
Operating temperature		-40 °C to +105 °C
Package		LQFP48 7 x 7 mm

(1) Half-duplex I²S shares the same pin with SPI.

2 Functional overview

2.1 ARM®Cortex®-M4

The ARM®Cortex®-M4 processor is the latest generation of ARM® core processors for embedded systems. It is a 32-bit high-performance RISC processor featuring exceptional code efficiency, outstanding computing power and advanced response to interrupts. The processor supports a set of DSP instructions allowing for efficient signal processing and complex algorithm execution.

2.2 Memory

2.2.1 Internal Flash memory

Up to 64 Kbytes of embedded Flash is available for storing programs and data. Any part of the embedded Flash memory can be protected by the sLib, a security area with code-executable only but non-readable. The “sLib” is a mechanism designed to protect the intelligence of solution providers and facilitates the second-level development at user level.

The AT32F4212 series embeds 4 Kbytes of boot memory, in which the Bootloader is resided.

A User System Data block is included for hardware configurations such as read/erase/write protection and watchdog self-enable. User System Data allows users to set erase/write and read protection individually. There are two levels of access protection: low level and high level.

2.2.2 Memory protection unit (MPU)

The memory protection unit (MPU) is used to manage the CPU accesses to memory to prevent one task to accidentally corrupt the memory or resources used by any other active task. This memory area consists of up to 8 protected areas that can in turn be divided up into 8 subareas. The protection area sizes are between 32 bytes and the whole 4 gigabytes of addressable memory. The MPU is especially suited to the applications where some critical or certified code has to be protected against the misbehavior of other tasks. It is usually managed by an RTOS (real-time operating system).

2.2.3 Embedded SRAM

Up to 16 Kbytes of embedded SRAM accessed (read/write) at CPU clock speed with 0 wait states.

2.3 Interrupts

2.3.1 Nested vectored interrupt controller (NVIC)

The AT32F4212 series embeds a nested vectored interrupt controller capable of managing 16 priority levels and handling maskable interrupt channels plus the 16 interrupt lines of the Cortex®-M4. This hardware block provides flexible interrupt management features with minimal interrupt latency.

2.3.2 External interrupts (EXINT)

The external interrupt (EXINT), which is connected directly to NVIC, consists of 20 edge detector lines used to generate interrupt requests. Each line can be independently configured to select the trigger event (rising edge, falling edge, both) and can be masked independently. A pending register maintains the status of the interrupt requests. The external interrupt lines connects up to 16 GPIOs.

2.4 Power control (PWC)

2.4.1 Power supply schemes

- $V_{DD} = 2.4 \sim 3.6$ V: power supply for GPIOs, ERTC, external 32 kHz crystal (LEXT), battery powered register (BPR) and the internal block such as regulator (LDO), provided via V_{DD} pins.
- $V_{DDA} = 2.4 \sim 3.6$ V: power supply for ADC and CMP via V_{DDA} pin. V_{DDA} and V_{SSA} must be the same voltage potential as V_{DD} and V_{SS} , respectively.

2.4.2 Reset and power voltage monitoring (POR / LVR / PVM)

The device has an integrated power-on reset (POR)/low voltage reset (PDR) circuitry. It is always active, and allows proper operation starting from/down to 2.4 V. The device remains in reset mode when V_{DD} is below a specified threshold (V_{LVR}) without the need for an external reset circuit.

The device embeds a power voltage monitor (PVM) that monitors the V_{DD} power supply and compares it to the V_{PVM} threshold. An interrupt is generated when V_{DD} drops below the V_{PVM} threshold and/or when V_{DD} rises above the V_{PVM} threshold. The PVM is enabled by software.

2.4.3 Voltage regulator (LDO)

The LDO has three operation modes: normal, low-power and extra low-power, power down.

- Normal mode: used in Run/Sleep mode and in Deepsleep mode.
- Low-power and extra low-power mode: used in Deepsleep mode.
- Power down mode: used in Standby mode: The LDO output is in high impedance and the kernel circuitry is powered down but the content of the registers and SRAM are lost.

The LDO operates always in its normal mode after reset.

2.4.4 Low-power modes

The AT32F4212 series supports three low-power modes:

- **Sleep mode**
In Sleep mode, only the CPU is stopped. All peripherals continue to operate and can wake up the CPU when an interrupt/event occurs.
- **Deepsleep mode**
Deepsleep mode achieves low power consumption while retaining the content of SRAM and registers. All clocks in the LDO domain are stopped, disabling the PLL, the HICK clock, and the HEXT crystal. The LDO can also be put in normal or low-power mode.
The device can be woken up from Deepsleep mode by any of the EXINT line. The EXINT line source can be one of the 16 external lines, the PVM output, the ERTC alarm/tamper detection/time stamp, or the CMP wakeup.
- **Standby mode**
The Standby mode is used to achieve the lowest power consumption. The internal LDO is switched off so that the entire LDO domain is powered off. The PLL, the HICK clock and the HEXT crystal are also switched off. After entering Standby mode, SRAM and register content are lost except for registers in the BPR domain and ERTC domain.
The device exits Standby mode when an external reset (NRST pin), a WDT reset, a rising edge on the WKUPx pin, or an ERTC alarm /tamper detection/time stamp occurs.

Note: The corresponding clock sources of ERTC and WDT are not stopped by entering Deepsleep or Standby mode.

2.5 Boot modes

At startup, BOOT0 and nBOOT1 (User System Data area) are used to select one of three boot options:

- Boot from user Flash memory;
- Boot from boot memory;
- Boot from embedded SRAM.

The bootloader is stored in boot memory. It is used to reprogram the Flash memory through USART1 or USART2. [Table 3](#) shows the pin configurations for Bootloader in the AT32F4212.

Table 3. Bootloader pin configurations

Peripherals	Pin
USART1	PA9: USART1_TX PA10: USART1_RX
USART2	PA2: USART2_TX PA3: USART2_RX

2.6 Clocks

The internal 48 MHz clock (HICK) divided by 6 (that is 8 MHz) is selected as the default CPU clock after any reset. An external 4 to 25 MHz clock (HEXT) can be selected, in which case it is monitored for failure. If failure is detected, HEXT will be switched off and the system automatically switches back to the internal HICK. A software interrupt is generated. Similarly, the system take the same action once HEXT fails when it is used as the source of PLL.

Several prescalers are available to configure AHB and APB (APB1 and APB2) frequency. The maximum frequency of both AHB and APB domains is 120 MHz.

2.7 General-purpose inputs/outputs (GPIO)

Each of the GPIO pins can be configured by software as output (push-pull or open-drain), as input (with or without pull-up or pull-down), or as alternate functions. Most of the GPIO pins are shared with digital or analog multiple functions. All GPIOs are high current-capable.

The GPIO's configuration can be locked, if needed, in order to avoid spurious writing to the GPIO's registers by following a specific sequence.

2.8 Direct Memory Access Controller (DMA)

The AT32F4212 series features a 5-channel general-purpose DMA which is able to manage memory-to-memory, peripheral-to-memory, and memory-to-peripheral transfers.

The DMA controller supports circular buffer management, without the need of the intervention of user code when the controller reaches the end of the buffer.

Each channel is connected to dedicated hardware DMA requests, with support for software trigger on each channel. Configuration is made by software and the number of transfers between source and destination are independent.

The DMA can be used with the main peripherals: SPI, I²S, I²C, USART, all timers TMRx (except TMR14) and ADC.

2.9 Timers (TMR)

The AT32F4212 device includes one advanced timer, up to five general-purpose timers, one basic timer and a SysTick timer.

The table below compares the features of the advanced, general-purpose, and basic timers.

Table 4. Timer feature comparison

Timer type	Timer	Counter resolution	Counter type	Prescaler factor	DMA request generation	Capture/compare channels	Complementary outputs
Advanced	TMR1	16-bit	Up, down, up/down	Any integer between 1 and 65536	Yes	4	3
General-purpose	TMR3	16-bit	Up, down, up/down	Any integer between 1 and 65536	Yes	4	No
	TMR14	16-bit	Up, down, up/down	Any integer between 1 and 65536	No	1	No
	TMR15	16-bit	Up	Any integer between 1 and 65536	Yes	2	1
	TMR16 TMR17	16-bit	Up	Any integer between 1 and 65536	Yes	1	1
Basic	TMR6	16-bit	Up	Any integer between 1 and 65536	Yes	No	No

2.9.1 Advanced timer (TMR1)

The advanced timer (TMR1) can be seen as a three-phase PWM generator assigned to 6 channels. It has complementary PWM outputs with programmable dead-time insertion. It can also be used as a complete general-purpose timer. These four independent channels can be used for:

- Input capture
- Output compare
- PWM generation (edge or center-aligned modes)
- One-cycle mode output

If configured as a standard 16-bit timer, it has the same features as the TMRx timer. If configured as the 16-bit PWM generator, it has full modulation capability (0-100%).

In debug mode, the advanced timer counter can be frozen and the PWM outputs disabled to turn off any power switch driven by these outputs.

Many features are shared with those of the general-purpose TMR which have the same architecture. The advanced timer can therefore work together with the TMR timers via the link feature for synchronization or event chaining.

2.9.2 General-purpose timers (TMR3, TMR14, TMR15, TMR16 and TMR17)

There are five synchronizable general-purpose timers embedded in the AT32F4212.

- **TMR3**

The TMR3 timer is based on a 16-bit auto-reload upcounter/downcounter and a 16-bit prescaler. They can offer four independent channels on the largest package. Each channel can be used for input capture/output compare, PWM or one-cycle mode output.

It can work together with the advanced timers via the link feature for synchronization or event chaining. In debug mode, the counter can be frozen. TMR3 can be used to generate PWM outputs. It has an individual DMA request mechanism.

This timer TMR3 is also capable of handling incremental encoder signals and the digital outputs coming from 1 to 3 hall-effect sensors.

- **TMR14**

The TMR14 timer is based on a 16-bit auto-reload upcounter, a 16-bit prescaler, and one independent channels for input capture/output compare, PWM, or one-cycle mode output. It can not only be synchronized with full-featured general-purpose timers, and but also can be used as simple time bases. In debug mode, counter can be frozen.

- **TMR15, TMR16 and TMR17**

These three timers are based on a 16-bit auto-reload upcounter, a 16-bit prescaler. TMR15 features two channels and one complementary channel. TMR16 and TMR17 have one channel and one complementary channel, respectively. All channels can be used for input capture/output compare, PWM, or one-pulse mode output.

They can work together via the Timer link feature for synchronization or event chaining.

In debug mode, counter can be frozen. These timers have their independent DMA request generation mechanism.

2.9.3 Basic timer (TMR6)

This timer is used as a generic 16-bit time base.

2.9.4 SysTick timer

This timer is dedicated to real-time operating systems, but could also be used as a standard downcounter. Its features include:

- A 24-bit downcounter
- Auto reload capability
- Maskable system interrupt generation when the counter reaches 0
- Programmable clock source (HICK or HICK/8)

2.10 Watchdog (WDT)

The watchdog consists of a 12-bit downcounter and 8-bit prescaler. It is clocked by an internal LICK clock. As it operates independently from the main clock, it can operate in DeepSleep and Standby modes. It can be used either as a watchdog to reset the device when an error occurs, or as a free running timer for application timeout management. It is self-enabling or not through the User System Data configuration. The counter can be frozen in debug mode.

2.11 Window watchdog (WWDT)

The window watchdog embeds a 7-bit downcounter that can be set as free running. It can be used as a watchdog to reset the device when a problem occurs. It is clocked from the main clock. It has an early warning interrupt capability and the counter can be frozen in debug mode.

2.12 Enhanced real-time clock (ERTC) and battery powered registers (BPR)

The battery powered domain includes:

- Enhanced real-time clock (ERTC)
- Five 32-bit battery powered registers

The enhanced real-time clock (ERTC) is an independent BCD timer/counter. It supports the following features:

- Calendar with second, minute, hour (12 or 24 format), week day, date, month, year, in BCD (binary-coded decimal) format.
- Support sub-seconds value in binary format.
- Automatic correction for 28, 29 (leap year), 30, and 31 days of the month.
- Programmable alarms with wakeup capability from DeepSleep or Standby mode
- Digital calibration with 512 Hz external output to compensate quartz crystal inaccuracy.

The alarm registers are used to generate an alarm at a specific time and calendar fields can be independently masked for alarm comparison. Other 32-bit registers also contain the programmable alarm subseconds, seconds, minutes, hours, day, and date.

A prescaler is used as a time base clock. It is by default configured to generate a time base of 1 second from a clock at 32.768 kHz.

The battery powered registers are 32-bit registers used to store 20 bytes of user application data. Battery powered registers are not reset by a system, or when the device wakes up from the Standby mode.

2.13 Communication interfaces

2.13.1 Serial peripheral interface (SPI)

Two SPIs can communicate at up to 36 Mbits/s in slave and master modes in full-duplex and half-duplex communication modes. A prescaler is used to generate multiple master mode frequencies. The frame is configurable to 8 bits or 16 bits. The hardware CRC generation/verification supports basic SD Card/MMC/SDHC modes. Both SPIs can be served by the DMA controller.

2.13.2 Inter-integrated sound interface (I²S)

Two standard I²S interfaces (multiplexed with SPI) are available, that can be operated in master or slave mode. These interfaces can be configured to operate with 16/24/32 bit resolution, as input or output channels. Audio sampling frequencies from 8 kHz up to 192 kHz are supported. When any of the I²S interfaces is/are configured in master mode, the master clock can be output at 256 times the sampling frequency. Both I²S can be served by the DMA controller.

2.13.3 Universal synchronous / asynchronous receiver transmitters (USART)

The AT32F4212 series embeds two universal synchronous/asynchronous receivers/transmitters (USART1 and USART2).

These two USART interfaces provide asynchronous communication, IrDA SIR ENDEC support, multiprocessor communication mode, master synchronous communication, single-wire half-duplex communication mode, and have LIN Master/Slave capability. These two USART interfaces also provide hardware management of the CTS and RTS signals, Smart Card mode (ISO 7816 compliant) and SPI-like communication capability. These two USART interfaces can be served by the DMA controller. TX/RX pins are swappable configuration.

Two USARTs are able to communicate at speeds of up to 7.5 Mbit/s.

2.13.4 Inter-integrated-circuit interface (I²C)

Two I²C bus interfaces can operate in multi-master and slave modes. They can support standard (up to 100 Kbit/s) and fast modes (up to 400 Kbit/s). The I²C bus frequency can be increased up to 1 MHz. For more details, please contact your nearest Artery sales office for technical support.

They support 7/10-bit addressing mode and 7-bit dual addressing mode (as slave). A hardware CRC generation/verification is included.

They can be served by DMA and they support SMBus 2.0/PMBus.

2.13.5 Infrared transmitter (IRTMR)

The AT32F4212 device offers an infrared transmitter solution. The solution is based on the internal connection between TMR16, USART1 or USART2 and TMR17. TMR17 is used to provide the carrier frequency, and TMR16, USART1 or USART2 provides the main signal to be sent. The infrared output signal is available on PB9 or PA13.

To generate infrared remote control signals, TMR16 channel 1 and TMR17 channel 1 must be correctly configured to generate correct waveforms. All standard IR pulse modulation modes can be obtained by programming two timer output compare channels.

2.14 Cyclic redundancy check (CRC) calculation unit

The CRC (cyclic redundancy check) calculation unit is used to get a CRC code from a 32-bit data word and a fixed generator polynomial. Among other applications, CRC-based techniques are used to verify data transmission or storage integrity.

2.15 Analog-to-digital converter (ADC)

One 12-bit analog-to-digital converter is embedded into AT32F4212 device and it has up to 13 external channels and five internal channels. The five internal channels are internally connected to two operational amplifier outputs (OPx_OUT), temperature sensor (V_{TS}), internal reference voltage (V_{INTRV}), and V_{SSA} , respectively. In sequence mode, automatic conversion is performed on a selected group of analog inputs.

The ADC can be served by the DMA controller.

An analog watchdog feature allows very precise monitoring of the converted voltage of one, some or all selected channels. An interrupt is generated when the converted voltage is outside the programmed thresholds.

The events generated by general-purpose timers (TMRx) and advanced timer (TMR1) can be internally connected to ADC start trigger and injection trigger, respectively, to synchronize ADC conversion with timers.

2.15.1 Operational amplifier output (OP_OUT)

The OP_OUT, which is internally connected to ADC inputs, can be used as an ADC internal input channel. The OP1_OUT and OP2_OUT are linked to ADC1_IN2 and ADC1_IN6 input channels, respectively, as shown in [Figure 1](#). Thus both ADC1_IN2 and ADC1_IN6 can no longer be used as external input channels.

2.15.2 Temperature sensor (V_{TS})

The temperature sensor has to generate a voltage V_{TS} that varies linearly with temperature. The temperature sensor is internally connected to the ADC1_IN16 input channel which is used to convert the sensor output voltage into a digital value.

The offset of this line varies from chip to chip due to process variation. The internal temperature sensor is more suited to applications that detect temperature variations instead of absolute temperatures. If accurate temperature readings are needed, an external temperature sensor part should be used.

2.15.3 Internal reference voltage (V_{INTRV})

The internal reference voltage (V_{INTRV}) provides a stable voltage source for ADC and CMP. The V_{INTRV} is internally connected to the ADC1_IN17 input channel.

2.16 Comparator (CMP)

The AT32F4212 series embeds one rail-to-rail comparator with programmable reference voltage (internal or external), hysteresis, speed, selectable output polarity, output blanking and noise filter.

The reference voltage can be one of the following:

- External GPIO
- Internal OP1_OUT connected to CMP1_INM7 (See [Figure 1](#))
- Internal reference voltage (V_{INTRV}) or submultiple (1/4, 1/2, 3/4)

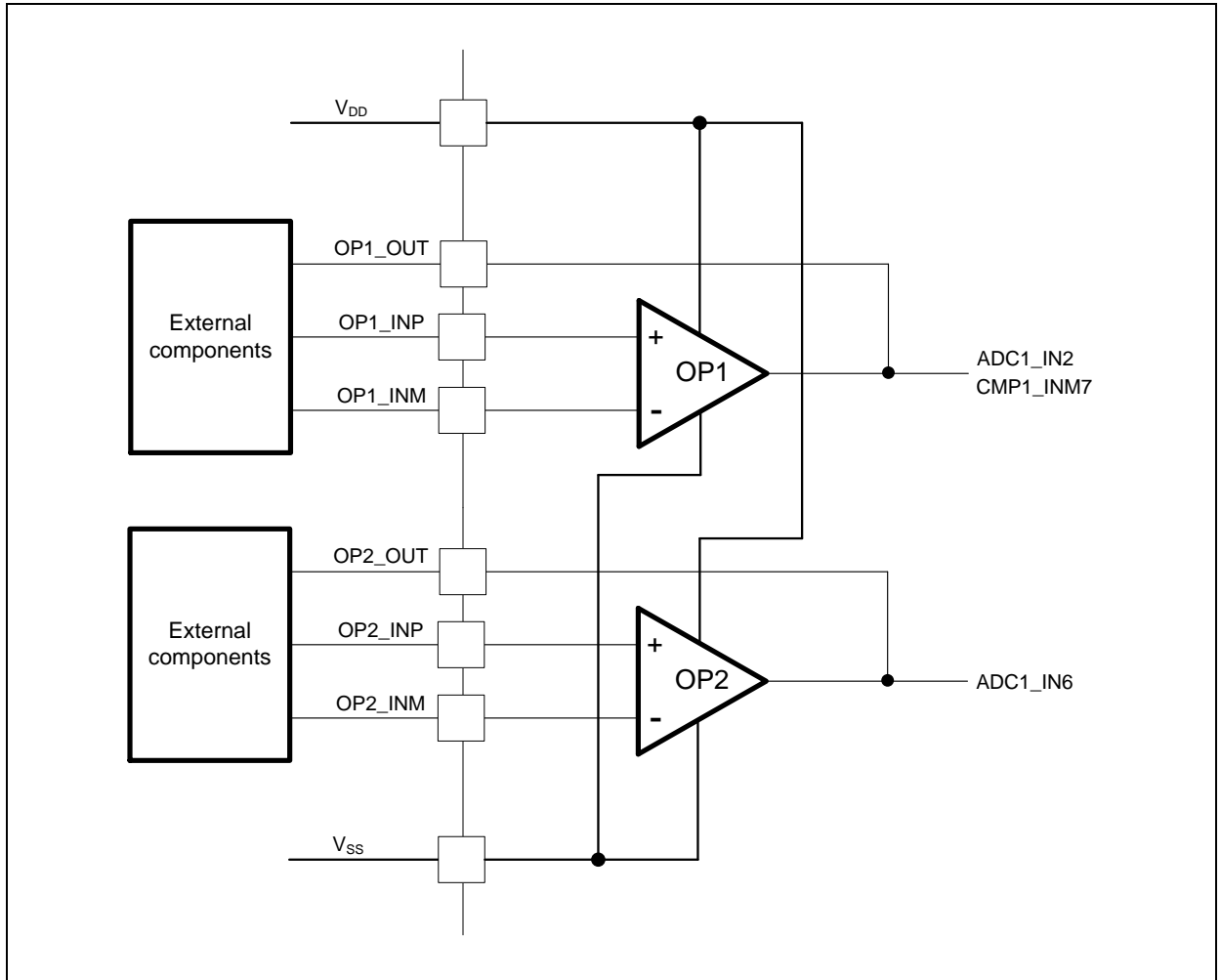
The comparator can wake up from DeepSleep mode. It can also remap output to timers.

2.17 Operational amplifier (OP)

The AT32F4212 series embeds two rail-to-rail operational amplifiers (OP). Both OPs provide up to 6 MHz bandwidth and offset voltage drift of less than 4.5 mV.

As the OP is always switched on as MCU is powered on, it is not controlled by software. And OP_OUT is internally connected to ADC and CMP. Thus this sets a limit on the use of GPIOs and other functions that are multiplexed with OP pins. Refer to the pin descriptions and notes in [Table 5](#) for details.

Figure 1. OP input/output and internal connection diagram

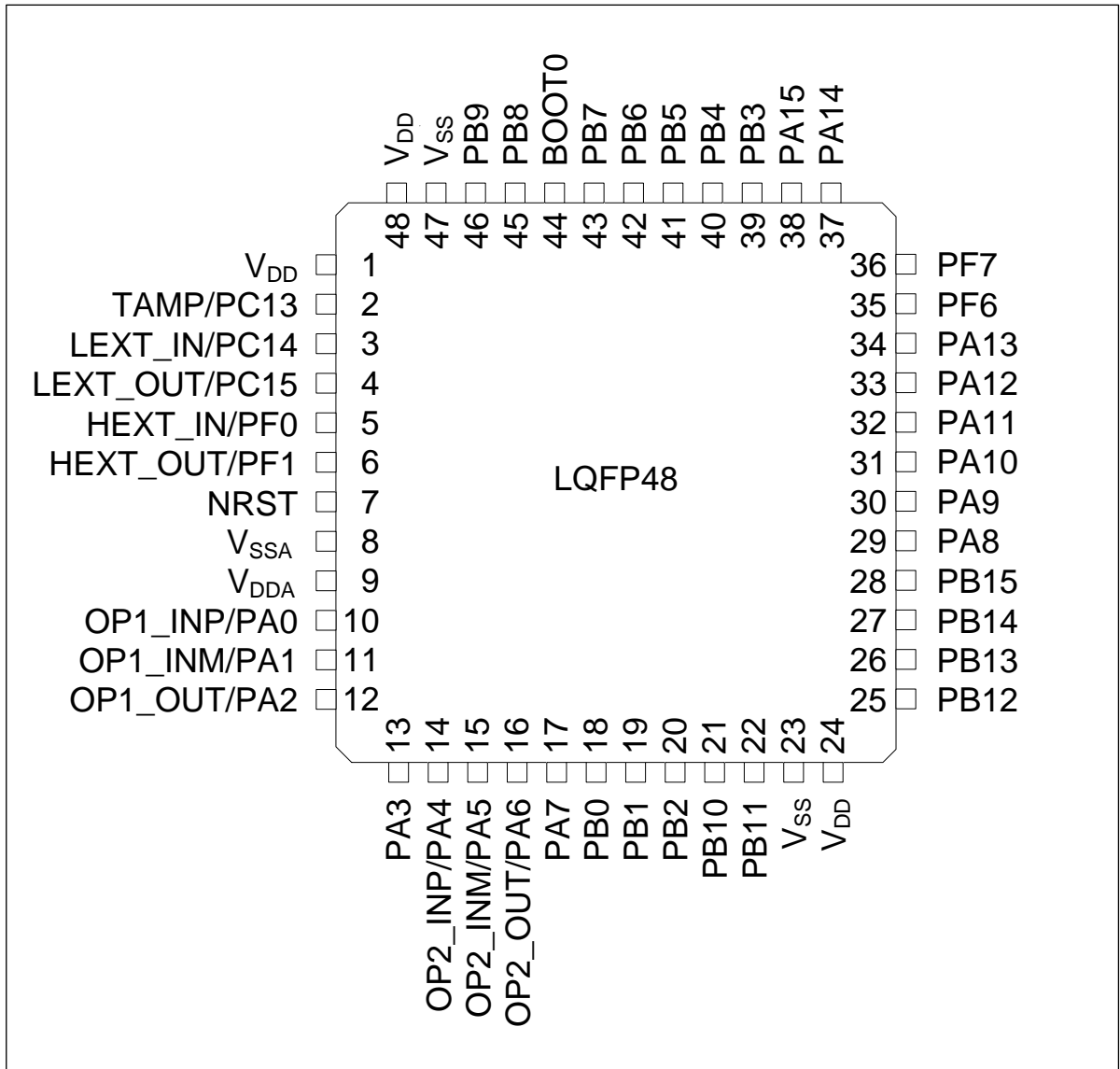


2.18 Serial wire debug (SWD)

The ARM® SWD interface is embedded, and is a serial wire debug port that enables a serial wire debug to be connected to the target for programming and debugging purposes.

3 Pin functional definitions

Figure 2. AT32F4212 LQFP48 pinout



The table below is the pin definitions of the AT32F4212. “-” presents there is no such pinout on the related package. Unless otherwise specified, the pins during and after reset have the same functions as those of actual ones, and all GPIOs are configured as floating input during and after reset. Alternate functions of pins are enabled through the GPIOx_MUXx register, and additional functions are selected through peripheral registers.

Table 5. AT32F4212 series pin definitions

Pin number	Pin name (after reset)	Pin type ⁽¹⁾	GPIO structure ⁽²⁾	Alternate function	Additional function
1	V _{DD}	S	-	Digital power supply	
2	PC13	I/O	FT	-	TAMP1 / WKUP2
3	PC14 / LEXT_IN (PC14)	I/O	TC	-	LEXT_IN
4	PC15 / LEXT_OUT (PC15)	I/O	TC	-	LEXT_OUT
5	PF0 / HEXT_IN (PF0)	I/O	TC	I2C1_SDA	HEXT_IN
6	PF1 / HEXT_OUT (PF1)	I/O	TC	I2C1_SCL	HEXT_OUT
7	NRST	I/O	R	Device reset input / internal reset input (active low)	
8	V _{SSA}	S	-	Analog ground	
9	V _{DDA}	S	-	Analog power supply	
10	PA0 / OP1_INP ⁽³⁾ (OP1_INP)	I/O	TC	TMR1_EXT / USART2_CTS / I2C2_SCL / CMP1_OUT	ADC1_IN0 / CMP1_INP2 / CMP1_INM6 / WKUP1 / OP1_INP
11	PA1 / OP1_INM ⁽³⁾ (OP1_INM)	I/O	TC	TMR15_CH1C / USART2_RTS / I2C2_SDA / EVENTOUT	ADC1_IN1 / CMP1_INP1 / OP1_INM
12	PA2 / OP1_OUT ⁽⁴⁾ (OP1_OUT)	I/O	TC	-	OP1_OUT
13	PA3	I/O	FTa	TMR15_CH2 / USART2_RX / I2S2_MCK	ADC1_IN3
14	PA4 / OP2_INP ⁽³⁾ (OP2_INP)	I/O	TC	TMR14_CH1 / USART2_CK / SPI1_CS / I2S1_WS	ADC1_IN4 / CMP1_INM4 / OP2_INP
15	PA5 / OP2_INM ⁽³⁾ (OP2_INM)	I/O	TC	SPI1_SCK / I2S1_CK	ADC1_IN5 / CMP1_INP0 / CMP1_INM5 / OP2_INM
16	PA6 / OP2_OUT ⁽⁴⁾ (OP2_OUT)	I/O	TC	-	OP2_OUT
17	PA7	I/O	FTa	TMR1_CH1C / TMR3_CH2 / TMR14_CH1 / TMR17_CH1 / SPI1_MOSI / I2S1_SD / EVENTOUT	ADC1_IN7
18	PB0	I/O	FTa	TMR1_CH2C / TMR3_CH3 / USART2_RX / I2S1_MCK / EVENTOUT	ADC1_IN8
19	PB1	I/O	FTa	TMR1_CH3C / TMR3_CH4 / TMR14_CH1 / SPI2_SCK / I2S2_CK	ADC1_IN9
20	PB2	I/O	FTa	TMR3_EXT	ADC1_IN10

Pin number	Pin name (after reset)	Pin type ⁽¹⁾	GPIO structure ⁽²⁾	Alternate function	Additional function
21	PB10	I/O	FT	SPI2_SCK / I2S2_CK / I2C2_SCL	-
22	PB11	I/O	FT	I2C2_SDA / EVENTOUT	-
23	V _{SS}	S	-	Digital ground	
24	V _{DD}	S	-	Digital power supply	
25	PB12	I/O	FTa	TMR1_BRK / TMR15_BRK / SPI2_CS / I2S2_WS / I2C2_SMBA / EVENTOUT	ADC1_IN11
26	PB13	I/O	FTa	TMR1_CH1C / SPI2_SCK / I2S2_CK / I2C2_SCL	ADC1_IN12
27	PB14	I/O	FTa	TMR1_CH2C / TMR15_CH1 / SPI2_MISO / I2S2_MCK / I2C2_SDA	ADC1_IN13
28	PB15	I/O	FTa	TMR1_CH3C / TMR15_CH2 / TMR15_CH1C / SPI2_MOSI / I2S2_SD	ADC1_IN14 / ERTC_REFIN / WKUP7
29	PA8	I/O	FT	TMR1_CH1 / USART1_CK / USART2_TX / I2C2_SCL / CLKOUT / EVENTOUT	-
30	PA9	I/O	FT	TMR1_CH2 / TMR15_BRK / USART1_TX / I2C1_SCL / I2C2_SMBA / CLKOUT	-
31	PA10	I/O	FT	TMR1_CH3 / TMR17_BRK / USART1_RX / I2C1_SDA	-
32	PA11	I/O	FT	TMR1_CH4 / USART1_CTS / I2C1_SMBA / I2C2_SCL / CMP1_OUT / EVENTOUT	-
33	PA12	I/O	FT	TMR1_EXT / USART1_RTS / I2C2_SDA / EVENTOUT	-
34	PA13 (SWDIO ⁽⁵⁾)	I/O	FT	PA13 / IR_OUT / SPI2_MISO / I2S2_MCK	-
35	PF6	I/O	FT	I2C2_SCL	-
36	PF7	I/O	FT	I2C2_SDA	-
37	PA14 (SWCLK ⁽⁵⁾)	I/O	FT	PA14 / USART2_TX / SPI2_MOSI / I2S2_SD	-
38	PA15	I/O	FT	USART2_RX / SPI1_CS / I2S1_WS / SPI2_CS / I2S2_WS / EVENTOUT	-
39	PB3	I/O	FT	SPI1_SCK / I2S1_CK / SPI2_SCK / I2S2_CK / EVENTOUT	-
40	PB4	I/O	FT	TMR3_CH1 / TMR17_BRK / SPI1_MISO / I2S1_MCK / SPI2_MISO / I2S2_MCK / I2C2_SDA / EVENTOUT	-

Pin number	Pin name (after reset)	Pin type ⁽¹⁾	GPIO structure ⁽²⁾	Alternate function	Additional function
41	PB5	I/O	FT	TMR3_CH2 / TMR16_BRK / SPI1_MOSI / I2S1_SD / SPI2_MOSI / I2S2_SD / I2C1_SMBA	WKUP6
42	PB6	I/O	FT	TMR16_CH1C / USART1_TX / I2S1_MCK / I2C1_SCL	-
43	PB7	I/O	FT	TMR17_CH1C / USART1_RX / I2C1_SDA	-
44	BOOT0	I	B	Boot mode selection 0	
45	PB8	I/O	FT	TMR16_CH1 / I2C1_SCL	-
46	PB9	I/O	FT	TMR17_CH1 / IR_OUT / SPI2_CS / I2S2_WS / I2S1_MCK / I2C1_SDA / EVENTOUT	-
47	V _{SS}	S	-	Digital ground	
48	V _{DD}	S	-	Digital power supply	
-	EPAD (V _{SS} /V _{SSA})	S	-	Digital ground / Analog ground	

- (1) I = input, O = output, S = supply.
- (2) TC = standard level, FT = general 5 V-tolerant, FTa = 5 V-tolerant with analog functions, R = bidirectional reset pin with embedded weak pull-up resistor, B = dedicated BOOT0 pin with embedded weak pull-down resistor. FTa is a 5 V-tolerant pin when used as input floating, input pull-up, or input pull-down; when used as analog mode, it loses 5 V tolerant characteristic, and in this case, the input level must be less than VDD + 0.3 V.
- (3) OPx_INy pin can still be used as GPIO or multiplexed function. As OP input, users are recommended to set this pin as analog mode. Note that the input level of these pins must be lower than VDD + 0.3V. If OPx_INy pin is unused, it is recommended for users to set it as GPIO output low.
- (4) It is recommended for users to set OPx_OUT as analog mode. Such pins are likely to output voltage because of external circuit and OP_IN pin level, and thus they can no longer be used as external input pins. However, OPx_OUT pin are internally connected to ADC and CMP.
- (5) After reset, PA13/PA14 pin are configured as alternate function SWDIO/SWCLK. In this case, the internal pull-up on the SWDIO pin and the internal pull-down on the SWCLK pin are activated.

4 Electrical characteristics

4.1 Parameter conditions

4.1.1 Minimum and maximum values

The minimum and maximum values are obtained in the worst conditions. Data based on characterization results, design simulation and/or technology characteristics are indicated in the table footnotes and are not tested in production. The minimum and maximum values represent the mean value plus or minus three times the standard deviation ($\text{mean} \pm 3\sigma$).

4.1.2 Typical values

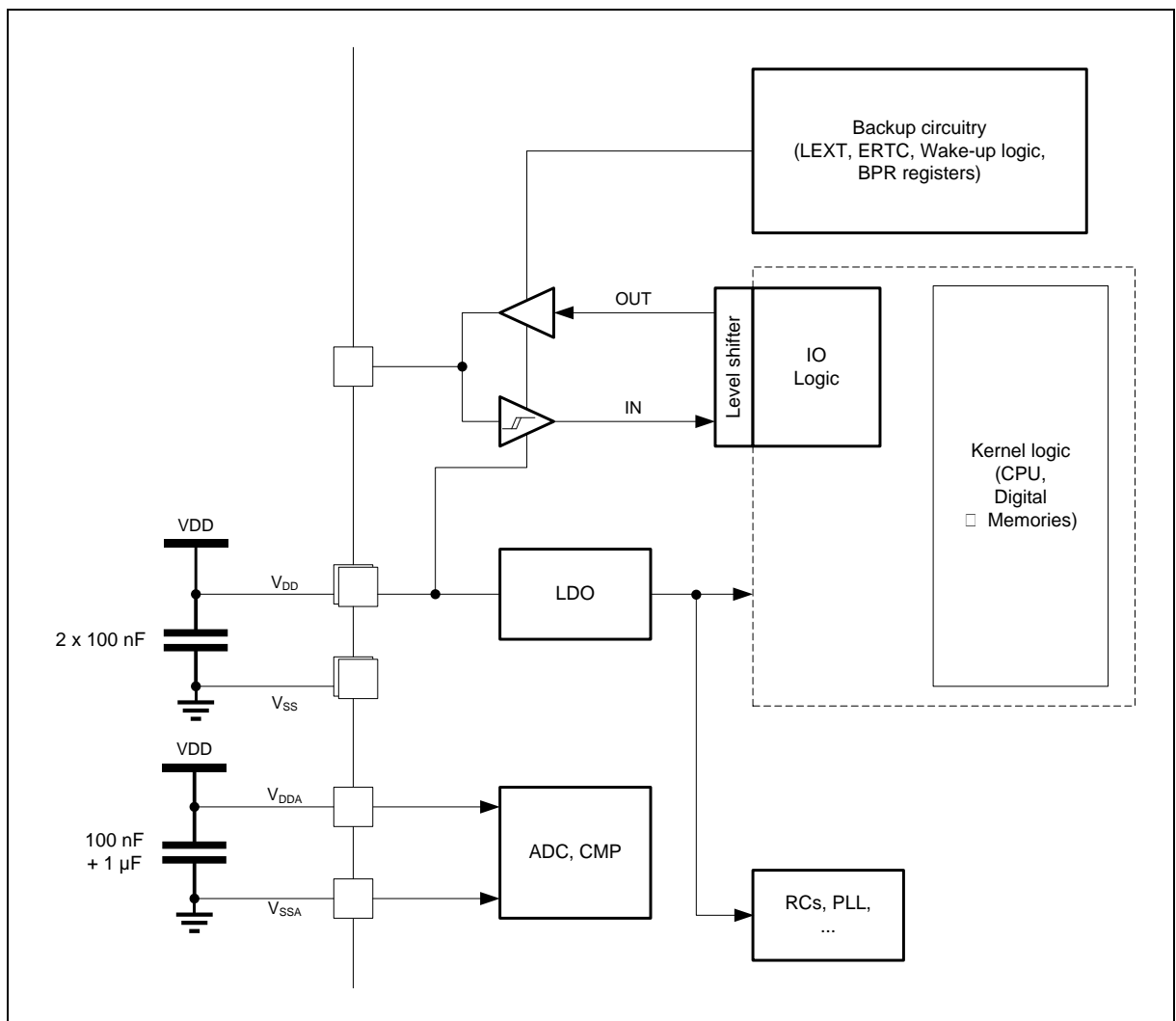
Typical data are based on $T_A = 25\text{ }^\circ\text{C}$, $V_{DD} = 3.3\text{ V}$.

4.1.3 Typical curves

All typical curves are given only as design guidelines and are not tested.

4.1.4 Power supply scheme

Figure 3. Power supply scheme



4.2 Absolute maximum values

4.2.1 Ratings

If stresses were out of the absolute maximum ratings listed [Table 6](#), [Table 7](#) and [Table 8](#), it may cause permanent damage to the device. These are maximum stress ratings only that the device could bear, but the functional operation of the device under these conditions is not implied. Exposure to maximum rating conditions for an extended period of times may affect device reliability.

Table 6. Voltage characteristics

Symbol	Ratings	Min	Max	Unit
$V_{DD}-V_{SS}$	External main supply voltage (including V_{DDA} and V_{DD})	-0.3	4.0	V
V_{IN}	Input voltage on FT GPIO	$V_{SS}-0.3$	6.0	
	Input voltage on FTa GPIO (set as input floating, input pull-up or input pull-down mode)			
	Input voltage on TC GPIO	$V_{SS}-0.3$	4.0	
Input voltage on FTa GPIO (set as analog mode)				
$ \Delta V_{DDx} $	Variations between different V_{DD} power pins	-	50	mV
$ V_{SSx}-V_{SS} $	Variations between all the different ground pins	-	50	

Table 7. Current characteristics

Symbol	Ratings	Max	Unit
I_{VDD}	Total current into V_{DD} power lines (source)	150	mA
I_{VSS}	Total current out of V_{SS} ground lines (sink)	150	
I_{IO}	Output current sunk by any GPIO and control pin	25	
	Output current source by any GPIO and control pin	-25	

Table 8. Temperature characteristics

Symbol	Ratings	Value	Unit
T_{STG}	Storage temperature range	-60 ~ +150	°C
T_J	Maximum junction temperature	125	

4.2.2 Electrical sensitivity

Based on three different tests (HBM, CDM, and LU) using specific measurement methods, the device is stressed in order to determine its performance in terms of electrical sensitivity.

Electrostatic discharge (ESD)

Electrostatic discharges are applied to the pins of each sample according to each pin combination. This test conforms to the JS-001-2017/JS-002-2018 standard.

Table 9. ESD values

Symbol	Parameter	Conditions	Class	Max	Unit
$V_{ESD(HBM)}$	Electrostatic discharge voltage (human body model)	$T_A = +25\text{ }^\circ\text{C}$, conforms to JS-001-2017	3A	± 6000	V
$V_{ESD(CDM)}$	Electrostatic discharge voltage (charge device model)	$T_A = +25\text{ }^\circ\text{C}$, conforms to JS-002-2018	III	± 1000	

Static latch-up

Tests compliant with EIA/JESD78E IC latch-up standard are required to assess the latch-up performance:

- A supply overvoltage is applied to each power supply pin;
- A current injection is applied to each input, output and configurable GPIO pin.

Table 10. Latch-up values

Symbol	Parameter	Conditions	Level/Class
LU	Static latch-up	$T_A = +105\text{ }^\circ\text{C}$, conforming to EIA/JESD78E	II level A ($\pm 200\text{ mA}$)

4.3 Specifications

4.3.1 General operating conditions

Table 11. General operating conditions

Symbol	Parameter	Conditions	Min	Max	Unit
f _{HCLK}	Internal AHB clock frequency	-	0	120	MHz
f _{PCLK1/2}	Internal APB1/2 clock frequency	-	0	f _{HCLK}	MHz
V _{DD}	Digital operating voltage	-	2.4	3.6	V
V _{DDA}	Analog operating voltage	Must be the same potential as V _{DD}	V _{DD}		V
P _D	Power dissipation: T _A = 105 °C	-	-	230	mW
T _A	Ambient temperature	-	-40	105	°C

4.3.2 Operating conditions at power-up/power-down

Table 12. Operating conditions at power-up/power-down

Symbol	Parameter	Conditions	Min	Max	Unit
t _{VDD}	V _{DD} rise time rate	-	0	∞	ms/V
	V _{DD} fall time rate		20	∞	µs/V

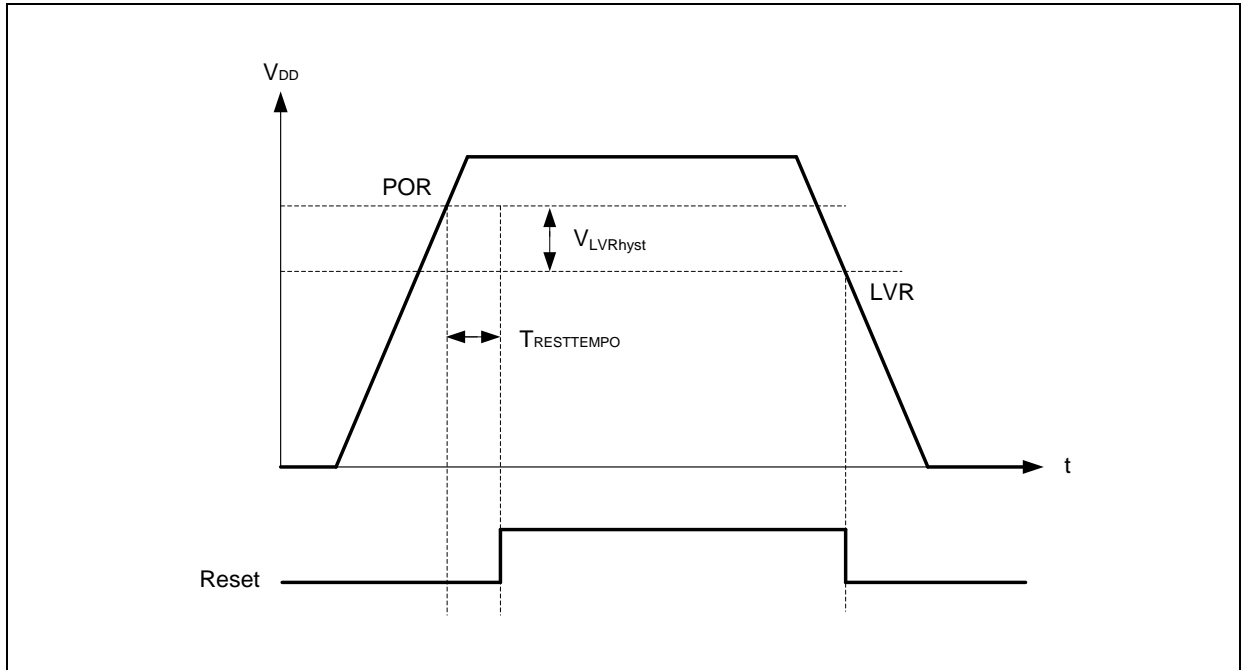
4.3.3 Embedded reset and power control block characteristics

Table 13. Embedded reset and power control block characteristics⁽¹⁾

Symbol	Parameter	Min	Typ	Max	Unit
V _{POR}	Power-on reset threshold	1.73	2.06	2.4	V
V _{LVR}	Low voltage reset threshold	1.62 ⁽²⁾	1.88	2.16	V
V _{LVRhyst}	LVR hysteresis	-	180	-	mV
T _{RESTEMPO}	Reset temporization: CPU starts execution after V _{DD} keeps higher than V _{POR} for T _{RSTEMPO}	-	4.5	-	ms

(1) Guaranteed by design, not tested in production.

(2) The product behavior is guaranteed by design down to the minimum V_{LVR} value.

Figure 4. Power-on reset and low voltage reset waveform

Table 14. Programmable voltage regulator characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V_{PVM1}	PVM threshold 1 (PVMSEL[2:0] = 001)	Rising edge ⁽¹⁾	2.19	2.28	2.37	V
		Falling edge ⁽¹⁾	2.09	2.18	2.27	V
V_{PVM2}	PVM threshold 2 (PVMSEL[2:0] = 010)	Rising edge ⁽²⁾	2.28	2.38	2.48	V
		Falling edge ⁽²⁾	2.18	2.28	2.38	V
V_{PVM3}	PVM threshold 3 (PVMSEL[2:0] = 011)	Rising edge ⁽²⁾	2.38	2.48	2.58	V
		Falling edge ⁽²⁾	2.28	2.38	2.48	V
V_{PVM4}	PVM threshold 4 (PVMSEL[2:0] = 100)	Rising edge ⁽²⁾	2.47	2.58	2.69	V
		Falling edge ⁽²⁾	2.37	2.48	2.59	V
V_{PVM5}	PVM threshold 5 (PVMSEL[2:0] = 101)	Rising edge ⁽²⁾	2.57	2.68	2.79	V
		Falling edge ⁽²⁾	2.47	2.58	2.69	V
V_{PVM6}	PVM threshold 6 (PVMSEL[2:0] = 110)	Rising edge ⁽²⁾	2.66	2.78	2.9	V
		Falling edge ⁽²⁾	2.56	2.68	2.8	V
V_{PVM7}	PVM threshold 7 (PVMSEL[2:0] = 111)	Rising edge	2.76	2.88	3	V
		Falling edge	2.66	2.78	2.9	V
$V_{HYS_P}^{(2)}$	PVM hysteresis	-	-	100	-	mV
$I_{DD(PVM)}^{(2)}$	PVM current consumption	-	-	20	30 ⁽²⁾	μ A

(1) PVMSEL [2:0] = 001 may not be used because it is lower than V_{POR} .

(2) Guaranteed by characterization results, not tested in production.

4.3.4 Memory characteristics

Table 15. Internal Flash memory characteristics⁽¹⁾

Symbol	Parameter	Typ	Max	Unit
T _{PROG}	Programming time	60	65	μs
t _{ERASE}	Sector erase time	6.6	8	ms
t _{ME}	Mass erase time	8.2	10	ms

(1) Guaranteed by design, not tested in production.

Table 16. Internal Flash memory endurance and data retention⁽¹⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
N _{END}	Endurance	T _A = -40 ~ 105 °C	100	-	-	kcycles
t _{RET}	Data retention	T _A = 105 °C	10	-	-	years

(1) Guaranteed by design, not tested in production.

4.3.5 Supply current characteristics

The current consumption is subject to several parameters and factors such as the operating voltage, ambient temperature, GPIO pin loading, device software configuration, operating frequencies, GPIO pin switching rate, and executed binary code. The current consumption is obtained by characterization results, not tested in production.

Typical and maximum current consumption

The MCU is placed under the following conditions:

- All GPIO pins are in analog mode.
- Flash memory access time depends on f_{HCLK} (0 ~ 32 MHz: zero wait state; 33 ~ 64 MHz: one wait state; 65 ~ 96 MHz: two wait states; over 96 MHz: three wait states).
- Prefetch is ON.
- f_{PCLK1} = f_{HCLK}, f_{PCLK2} = f_{HCLK}, f_{ADCCLK} = f_{PCLK2}/8.
- Unless otherwise specified, the typical values are measured with V_{DD} = 3.3 V and T_A = 25 °C condition and the maximum values are measured with V_{DD} = 3.6 V.

Table 17. Typical current consumption in Run mode

Symbol	Parameter	Conditions	f _{HCLK}	Typ		Unit
				All peripherals enabled	All peripherals disabled	
I _{DD}	Supply current in Run mode	High-speed external crystal (HEXT) ⁽¹⁾⁽²⁾	120 MHz	17.7	12.3	mA
			108 MHz	16.2	11.3	
			72 MHz	11.5	8.19	
			48 MHz	8.62	6.44	
			36 MHz	6.98	5.34	
			24 MHz	5.65	4.54	
			16 MHz	4.45	3.71	
			8 MHz	2.96	2.57	
			4 MHz	2.50	2.30	
			2 MHz	2.27	2.16	
			1 MHz	2.16	2.10	
			500 kHz	2.10	2.07	
		125 kHz	2.06	2.04		
		High-speed internal clock (HICK)	120 MHz	17.7	12.3	mA
			108 MHz	16.1	11.3	
			72 MHz	11.4	8.14	
			48 MHz	8.52	6.38	
			36 MHz	6.88	5.27	
			24 MHz	5.53	4.47	
			16 MHz	4.34	3.63	
			8 MHz	2.83	2.48	
			4 MHz	2.37	2.20	
			2 MHz	2.15	2.06	
			1 MHz	2.03	1.99	
500 kHz	1.97		1.95			
125 kHz	1.93	1.93				

(1) External clock is 8 MHz.

(2) PLL is ON when f_{HCLK} > 8 MHz.

Table 18. Typical current consumption in Sleep mode

Symbol	Parameter	Conditions	f _{HCLK}	Typ		Unit
				All peripherals enabled	All peripherals disabled	
I _{DD}	Supply current in Sleep mode	High-speed external crystal (HEXT) ⁽¹⁾⁽²⁾	120 MHz	13.2	5.99	mA
			108 MHz	12.1	5.59	
			72 MHz	8.76	4.38	
			48 MHz	6.81	3.89	
			36 MHz	5.60	3.42	
			24 MHz	4.70	3.25	
			16 MHz	3.80	2.83	
			8 MHz	2.60	2.11	
			4 MHz	2.30	2.05	
			2 MHz	2.15	2.02	
			1 MHz	2.07	2.00	
			500 kHz	2.03	1.99	
		125 kHz	2.00	1.98		
		High-speed internal clock (HICK)	120 MHz	13.1	5.88	mA
			108 MHz	12.0	5.47	
			72 MHz	8.65	4.27	
			48 MHz	6.70	3.78	
			36 MHz	5.49	3.30	
			24 MHz	4.59	3.13	
			16 MHz	3.68	2.71	
			8 MHz	2.47	1.98	
			4 MHz	2.17	1.93	
			2 MHz	2.02	1.89	
			1 MHz	1.94	1.88	
500 kHz	1.90		1.87			
125 kHz	1.87	1.86				

(1) External clock is 8 MHz.

(2) PLL is ON when f_{HCLK} > 8 MHz.

Table 19. Maximum current consumption in Run mode

Symbol	Parameter	Conditions	f _{HCLK}	Max		Unit
				T _A = 85 °C	T _A = 105 °C	
I _{DD}	Supply current in Run mode	High-speed external crystal (HEXT) ⁽¹⁾ All peripherals enabled	120 MHz	20.4	22.3	mA
			108 MHz	18.8	20.7	
			72 MHz	14.1	16.0	
			48 MHz	11.2	13.1	
			36 MHz	9.54	11.4	
			24 MHz	8.19	10.1	
			16 MHz	6.99	8.84	
			8 MHz	5.49	7.33	
		High-speed external crystal (HEXT) ⁽¹⁾ All peripherals disabled	120 MHz	15.0	16.8	mA
			108 MHz	13.9	15.8	
			72 MHz	10.8	12.6	
			48 MHz	9.02	10.9	
			36 MHz	7.91	9.74	
			24 MHz	7.10	8.93	
16 MHz	6.26	8.09				
8 MHz	5.11	6.95				

(1) External clock is 8 MHz, and PLL is ON when f_{HCLK} > 8 MHz.

Table 20. Maximum current consumption in Sleep mode

Symbol	Parameter	Conditions	f _{HCLK}	Max		Unit
				T _A = 85 °C	T _A = 105 °C	
I _{DD}	Supply current in Sleep mode	High-speed external crystal (HEXT) ⁽¹⁾ All peripherals enabled	120 MHz	15.9	17.7	mA
			108 MHz	14.8	16.6	
			72 MHz	11.4	13.2	
			48 MHz	9.39	11.2	
			36 MHz	8.19	9.96	
			24 MHz	7.29	9.05	
			16 MHz	6.38	8.13	
			8 MHz	5.18	6.91	
		High-speed external crystal (HEXT)(1) All peripherals disabled	120 MHz	8.56	10.3	mA
			108 MHz	8.16	9.90	
			72 MHz	6.95	8.69	
			48 MHz	6.46	8.17	
			36 MHz	5.98	7.72	
			24 MHz	5.81	7.53	
			16 MHz	5.39	7.13	
			8 MHz	4.68	6.41	

(1) External clock is 8 MHz, and PLL is ON when f_{HCLK} > 8 MHz.

Table 21. Typical and maximum current consumption in Deepsleep and Standby modes

Symbol	Parameter	Conditions	Typ ⁽¹⁾		Max ⁽²⁾			Unit
			V _{DD} = 2.4 V	V _{DD} = 3.3 V	T _A = 25 °C	T _A = 85 °C	T _A = 105 °C	
I _{DD}	Supply current in Deepsleep mode	LDO in normal mode, HICK and HEXT OFF (no WDT)	1245	1290	See ⁽⁴⁾	5600	8350	μA
		LDO in low-power mode, VREXPEN=1, HICK and HEXT OFF (no WDT)	1005	1050		3500	4915	
	Supply current in Standby mode	LEXT and ERTC OFF	802.4	843.6	944.6	1505.7	1607.6	μA
		LEXT and ERTC ON	803.2	845.1	945.9	1507.2	1609.2	

(1) OP is always ON, which products extra static power consumption that compromises power-saving effect in Deepsleep and Standby modes (that is, power-saving effect in both modes is not so obvious).

(2) Typical values are measured at T_A = 25 °C.

(3) Guaranteed by characterization results, not tested in production.

(4) This value may be several times the typical value due to process variations.

On-chip peripheral current consumption

The MCU is placed under the following conditions:

- All GPIO pins are in analog mode.
- The given values are calculated by measuring the current consumption difference between “all peripherals clocked OFF” and “only one peripheral clocked ON”.

Table 22. Peripheral current consumption

Peripheral		Typ	Unit
AHB	DMA1	2.15	μA/MHz
	SRAM	1.06	
	Flash	12.08	
	GPIOA	0.50	
	GPIOB	0.50	
	GPIOC	0.50	
	GPIOF	0.50	
	CRC	0.70	
APB1	TMR3	6.29	
	TMR6	0.49	
	TMR14	2.28	
	SPI2/I ² S2	2.26	
	USART2	2.11	
	I ² C1	1.71	
	I ² C2	1.68	
	WWDT	0.20	
	PWC	0.39	
APB2	SCFG/CMP1	0.29	
	SPI1/I ² S1	2.03	
	USART1	2.12	
	TMR1	7.68	
	TMR15	4.65	
	TMR16	3.19	
	TMR17	3.41	
	ADC1	5.17	

4.3.6 External clock source characteristics

High-speed external clock generated from a crystal / ceramic resonator

The high-speed external (HEXT) clock can be supplied with a 4 to 25 MHz crystal/ceramic resonator oscillator. All the information given in this paragraph are based on characterization results obtained with typical external components specified in the table below. In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy).

Table 23. HEXT 4 ~ 25 MHz crystal characteristics ⁽¹⁾⁽²⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$f_{\text{HEXT_IN}}$	Oscillator frequency	-	4	8	25	MHz
$t_{\text{SU(HEXT)}}^{(3)}$	Startup time	V_{DD} is stabilized	-	2	-	ms

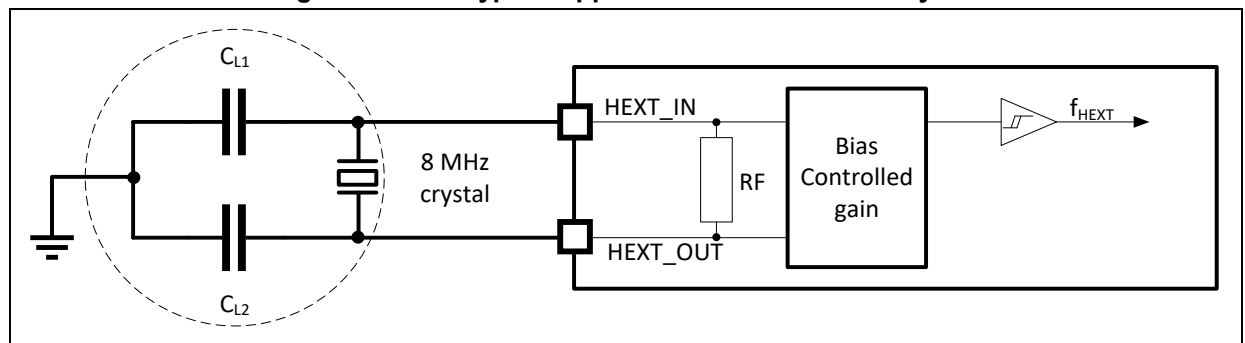
(1) Oscillator characteristics given by the crystal/ceramic resonator manufacturer.

(2) Guaranteed by characterization results, not tested in production.

(3) $t_{\text{SU(HEXT)}}$ is the startup time measured from the moment HEXT is enabled (by software) to a stabilized 8 MHz oscillation is reached. This value is measured for a standard crystal resonator and it can vary significantly from one crystal manufacturer to another manufacturer.

For C_{L1} and C_{L2} , it is recommended to use high-quality external ceramic capacitors in the 5 pF to 25 pF range (typ.), designed for high-frequency applications, and select a crystal or resonator meeting the requirements. C_{L1} and C_{L2} are usually the same size. The crystal manufacturer typically specifies a load capacitance which is the series combination of C_{L1} and C_{L2} . PCB and MCU pin capacitance must be included (10 pF can be used as a rough estimate of the combined pin and board capacitance) when sizing C_{L1} and C_{L2} .

Figure 5. HEXT typical application with an 8 MHz crystal



High-speed external clock generated from an external source

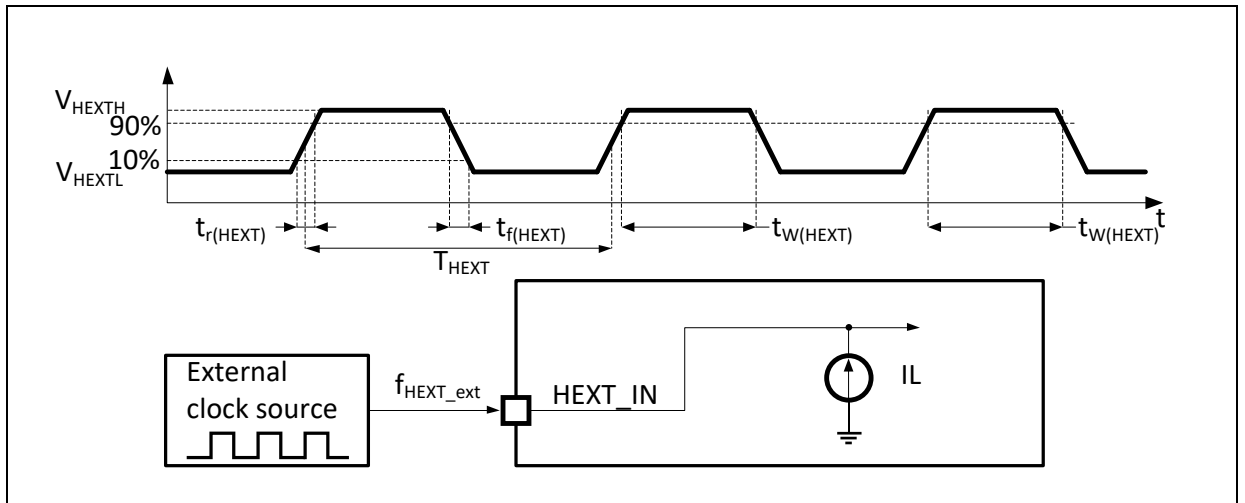
The characteristics given in the table below result from tests performed using a high-speed external clock source.

Table 24. HEXT external source characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$f_{\text{HEXT_ext}}$	User external clock source frequency ⁽¹⁾		1	8	25	MHz
V_{HEXTH}	HEXT_IN input pin high level voltage		$0.7V_{\text{DD}}$	-	V_{DD}	V
V_{HEXTL}	HEXT_IN input pin low level voltage		V_{SS}	-	$0.3V_{\text{DD}}$	
$t_{\text{w(HEXT)}}$ $t_{\text{w(HEXT)}}$	HEXT_IN high or low time ⁽¹⁾	-	5	-	-	ns
$t_{\text{r(HEXT)}}$ $t_{\text{r(HEXT)}}$	HEXT_IN rise or fall time ⁽¹⁾	-	-	-	20	
$C_{\text{in(HEXT)}}$	HEXT_IN input capacitance ⁽¹⁾	-	-	5	-	pF
Duty(HEXT)	Duty cycle	-	45	-	55	%
I_{L}	HEXT_IN input leakage current	$V_{\text{SS}} \leq V_{\text{IN}} \leq V_{\text{DD}}$	-	-	± 1	μA

(1) Guaranteed by design, not tested in production.

Figure 6. HEXT external source AC timing diagram



Low-speed external clock generated from a crystal / ceramic resonator

The low-speed external (LEXT) clock can be supplied with a 32.768 kHz crystal/ceramic resonator oscillator. All the information given in this paragraph are based on characterization results obtained with typical external components specified in the table below. In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy).

Table 25. LEXT 32.768 kHz crystal characteristics⁽¹⁾⁽²⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$t_{SU(LEXT)}$	Startup time	V_{DD} is stabilized	-	180	-	ms

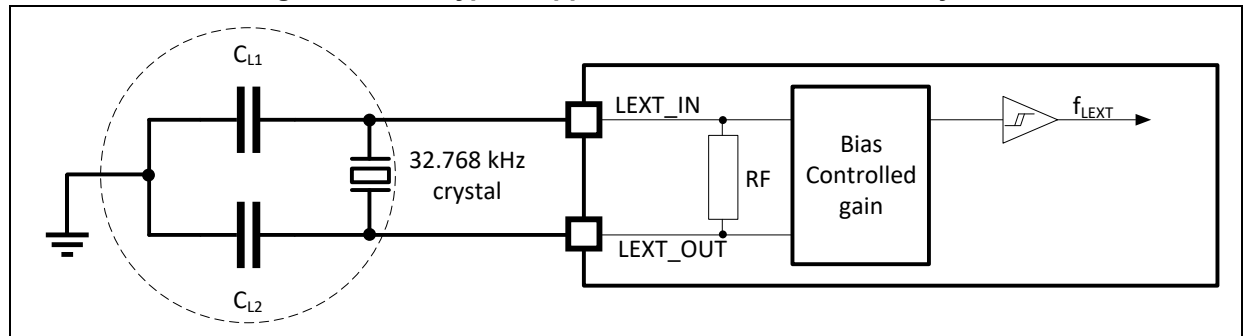
(1) Oscillator characteristics are given by the crystal/ceramic resonator manufacturer.

(2) Guaranteed by characterization results, not tested in production.

For C_{L1} and C_{L2} , it is recommended to use high-quality ceramic capacitors in the 5 pF to 20 pF range selected to match the requirements of the crystal or resonator. C_{L1} and C_{L2} , are usually the same size. The crystal manufacturer typically specifies a load capacitance which is the series combination of C_{L1} and C_{L2} .

Load capacitance C_L has the following formula: $C_L = C_{L1} \times C_{L2} / (C_{L1} + C_{L2}) + C_{stray}$, where C_{stray} is the pin capacitance and board or trace PCB-related capacitance. Typically, it is between 2 pF and 7 pF.

Figure 7. LEXT typical application with a 32.768 kHz crystal



Note: No external resistor is required between LEXT_IN and LEXT_OUT, and it is also prohibited to add it.

Low-speed external clock generated from an external source

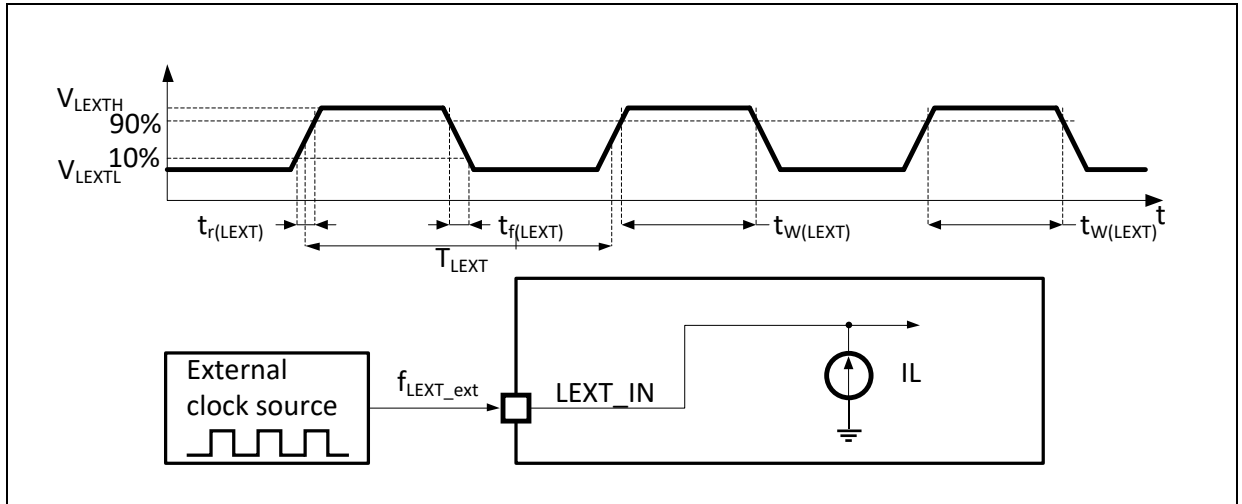
The characteristics given in the table below result from tests performed using a low-speed external clock source.

Table 26. LEXT external source characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit	
f_{LEXT_ext}	User external clock source frequency ⁽¹⁾		-	32.768	1000	kHz	
V_{LEXTH}	LEXT_IN input pin high level voltage		0.7V _{DD}	-	V _{DD}		V
V_{LEXTL}	LEXT_IN input pin low level voltage		V _{SS}	-	0.3V _{DD}		
$t_{w(LEXT)}$ $t_{w(LEXT)}$	LEXT_IN high or low time ⁽¹⁾	-	450	-	-	ns	
$t_{r(LEXT)}$ $t_{f(LEXT)}$	LEXT_IN rise or fall time ⁽¹⁾	-	-	-	50		
$C_{in(LEXT)}$	LEXT_IN input capacitance ⁽¹⁾	-	-	5	-	pF	
Duty(LEXT)	Duty cycle	-	30	-	70	%	
I_L	LEXT_IN input leakage current	$V_{SS} \leq V_{IN} \leq V_{DD}$	-	-	±1	µA	

(1) Guaranteed by design, not tested in production.

Figure 8. LEXT external source AC timing diagram



4.3.7 Internal clock source characteristics

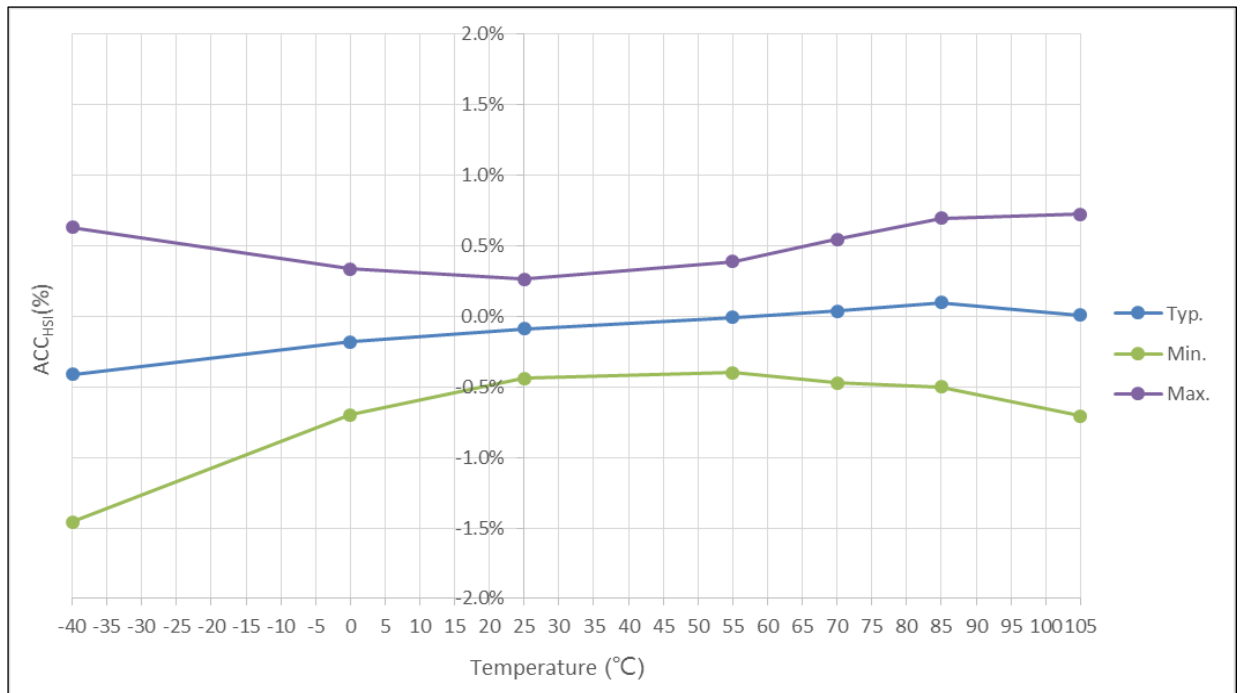
High-speed internal clock (HICK)

Table 27. HICK clock characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit	
f_{HICK}	Frequency	-	-	48	-	MHz	
$DuCy_{(HICK)}$	Duty cycle	-	45	-	55	%	
ACC_{HICK}	Accuracy of the HICK oscillator	User-trimmed with the CRM_CTRL register ⁽¹⁾	-1	-	1	%	
		Factory-calibrated ⁽²⁾	$T_A = -40 \sim 105 \text{ }^\circ\text{C}$	-2	-	1.5	%
			$T_A = -40 \sim 85 \text{ }^\circ\text{C}$	-2	-	1.2	%
			$T_A = 0 \sim 70 \text{ }^\circ\text{C}$	-1.5	-	1.2	%
		$T_A = 25 \text{ }^\circ\text{C}$	-1	-	1	%	
$t_{SU(HICK)}^{(2)}$	HICK oscillator startup time	-	-	10	12	μs	
$I_{DD(HICK)}^{(2)}$	HICK oscillator power consumption	-	-	220	290	μA	

(1) Guaranteed by design, not tested in production.

(2) Guaranteed by characterization results, not tested in production.

Figure 9. HICK clock accuracy vs. temperature


Low-speed internal clock (LICK)

Table 28. LICK clock characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$f_{LICK}^{(1)}$	Frequency	-	25	35	45	kHz

(1) Guaranteed by characterization results, not tested in production.

4.3.8 PLL characteristics

Table 29. PLL characteristics

Symbol	Parameter	Min ⁽¹⁾	Typ	Max ⁽¹⁾	Unit
f _{PLL_IN}	PLL input clock ⁽²⁾	2	8	16	MHz
	PLL input clock duty cycle	40	-	60	%
f _{PLL_OUT}	PLL multiplier output clock	16	-	120	MHz
t _{LOCK}	PLL lock time	-	-	200	μs
Jitter	Cycle-to-cycle jitter	-	-	300	ps

(1) Guaranteed by design, not tested in production.

(2) Take care of using the appropriate multiplier factors so as to have PLL input clock values compatible with the range defined by f_{PLL_OUT}.

4.3.9 Wakeup time from low-power mode

The wakeup times given in the table below is measured on a wakeup phase with the HICK. The clock source used to wake up the device depends on the current operating mode:

- Sleep mode: the clock source is the clock that is set before entering Sleep mode.
- Deepsleep or Standby mode: the clock source is HICK.

Table 30. Low-power mode wakeup time

Symbol	Parameter	Typ	Unit
t _{WUSLEEP}	Wakeup from Sleep mode	3.3	μs
t _{WUDEEPSLEEP}	Wakeup from Deepsleep mode (LDO in normal mode)	380	μs
	Wakeup from Deepsleep mode (LDO in low-power mode)	450	
t _{WUSTDBY}	Wakeup from Standby mode	1250	μs

4.3.10 EMC characteristics

Susceptibility tests are performed on a sample basis during device characterization.

Functional EMS (electromagnetic susceptibility)

- **EFT:** A Burst of Fast Transient voltage (positive and negative) is applied to V_{DD} and V_{SS} through a coupling/decoupling network, until a functional disturbance occurs. This test is compliant with the IEC 61000-4-4 standard.

Table 31. EMS characteristics

Symbol	Parameter	Conditions	Level/Class
V _{EFT}	Fast transient voltage burst limit is applied to V _{DD} and V _{SS} via coupling/decoupling network conforming to IEC 61000-4-4 to induce functional disturbance, V _{DD} and V _{SS} input has one 47 μF capacitor and each V _{DD} and V _{SS} pin pair has 0.1 μF bypass resistor.	V _{DD} = 3.3 V, LQFP48, T _A = +25 °C, f _{HCLK} = 120 MHz, conforms to IEC 61000-4-4 V _{DD} = 3.3 V, LQFP48, T _A = +25 °C, f _{HCLK} = 72 MHz, conforms to IEC 61000-4-4	3/A (3.5 kV)

EMC characterization and optimization are performed at component level with a typical application environment. It should be noted that good EMC performance is highly dependent on the user application and the software in particular. Therefore it is recommended that the user applies EMC software optimization and prequalification tests in relation with the EMC level requested for his application.

4.3.11 GPIO port characteristics

General input / output characteristics

All GPIOs are CMOS and TTL compliant.

Table 32. GPIO static characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V _{IL}	GPIO input low level voltage	-	-0.3	-	0.28 × V _{DD} + 0.1	V
V _{IH}	TC GPIO input high level voltage	-	0.31 × V _{DD} + 0.8	-	V _{DD} + 0.3	V
	FTa GPIO input high level voltage	Analog mode		-	5.5	
	FT GPIO input high level voltage	-		-		
	FTa GPIO input high level voltage	Input floating, input pull-up, or input pull-down mode		-		
V _{hys}	Schmitt trigger voltage hysteresis ⁽¹⁾	-	200	-	-	mV
		-	5% V _{DD}	-	-	-
I _{lkg}	Input leakage current ⁽²⁾	V _{SS} ≤ V _{IN} ≤ V _{DD} TC GPIO	-	-	±1	μA
		V _{SS} ≤ V _{IN} ≤ 5.5 V FT and FTa GPIO	-	-	±1	
R _{PU}	Weak pull-up equivalent resistor ⁽³⁾	V _{IN} = V _{SS}	65	80	130	kΩ
R _{PD}	Weak pull-down equivalent resistor ⁽³⁾⁽⁴⁾	V _{IN} = V _{DD}	65	70	130	kΩ
C _{IO}	GPIO pin capacitance	-	-	9	-	pF

(1) Hysteresis voltage between Schmitt trigger switching levels. Guaranteed by characterization results, not tested in production.

(2) Leakage could be higher than max if negative current is injected on adjacent pins.

(3) When the input is higher than V_{DD} + 0.3 V, the internal pull-up and pull-down resistors must be disabled for FT, and FTa pins.

(4) The pull-down resistor of BOOT0 exists permanently.

All GPIOs are CMOS and TTL compliant (no software configuration required). Their characteristics cover more than the strict CMOS-technology or TTL parameters

Output driving current

In the user application, the number of GPIO pins which can drive current must be controlled to respect the absolute maximum rating defined in [4.2.1](#):

- The sum of the currents sourced by all GPIOs on V_{DD} plus the maximum Run consumption of the MCU sourced on V_{DD}, cannot exceed the absolute maximum rating I_{VDD} (see [Table 7](#)).
- The sum of the currents sunk by all GPIOs on V_{SS} plus the maximum Run consumption of the MCU sunk on V_{SS}, cannot not exceed the absolute maximum rating I_{VSS} (see [Table 7](#)).

Output voltage levels

All GPIOs are CMOS and TTL compliant.

Table 33. Output voltage characteristics

Symbol	Parameter	Conditions	Min	Max	Unit
Normal sourcing/sinking strength					
$V_{OL}^{(1)}$	Output low level voltage	CMOS standard, $I_{IO} = 4\text{ mA}$ $2.7\text{ V} \leq V_{DD} \leq 3.6\text{ V}$	-	0.4	V
$V_{OH}^{(1)}$	Output high level voltage		$V_{DD}-0.4$	-	
$V_{OL}^{(1)}$	Output low level voltage	TTL standard, $I_{IO} = 2\text{ mA}$ $2.7\text{ V} \leq V_{DD} \leq 3.6\text{ V}$	-	0.4	V
$V_{OH}^{(1)}$	Output high level voltage		2.4	-	
$V_{OL}^{(1)}$	Output low level voltage	$I_{IO} = 9\text{ mA}$ $2.7\text{ V} \leq V_{DD} \leq 3.6\text{ V}$	-	1.3	V
$V_{OH}^{(1)}$	Output high level voltage		$V_{DD}-1.3$	-	
$V_{OL}^{(1)}$	Output low level voltage	$I_{IO} = 2\text{ mA}$ $2.4\text{ V} \leq V_{DD} < 2.7\text{ V}$	-	0.4	V
$V_{OH}^{(1)}$	Output high level voltage		$V_{DD}-0.4$	-	
Large sourcing/sinking strength					
V_{OL}	Output low level voltage	CMOS standard, $I_{IO} = 6\text{ mA}$ $2.7\text{ V} \leq V_{DD} \leq 3.6\text{ V}$	-	0.4	V
V_{OH}	Output high level voltage		$V_{DD}-0.4$	-	
$V_{OL}^{(1)}$	Output low level voltage	TTL standard, $I_{IO} = 5\text{ mA}$ $2.7\text{ V} \leq V_{DD} \leq 3.6\text{ V}$	-	0.4	V
$V_{OH}^{(1)}$	Output high level voltage		2.4	-	
$V_{OL}^{(1)}$	Output low level voltage	$I_{IO} = 18\text{ mA}$ $2.7\text{ V} \leq V_{DD} \leq 3.6\text{ V}$	-	1.3	V
$V_{OH}^{(1)}$	Output high level voltage		$V_{DD}-1.3$	-	
$V_{OL}^{(1)}$	Output low level voltage	$I_{IO} = 4\text{ mA}$ $2.4\text{ V} \leq V_{DD} < 2.7\text{ V}$	-	0.4	V
$V_{OH}^{(1)}$	Output high level voltage		$V_{DD}-0.4$	-	
Maximum sourcing/sinking strength					
$V_{OL}^{(1)}$	Output low level voltage	CMOS standard, $I_{IO} = 15\text{ mA}$ $2.7\text{ V} \leq V_{DD} \leq 3.6\text{ V}$	-	0.4	V
$V_{OH}^{(1)}$	Output high level voltage		$V_{DD}-0.4$	-	
$V_{OL}^{(1)}$	Output low level voltage	TTL standard, $I_{IO} = 12\text{ mA}$ $2.7\text{ V} \leq V_{DD} \leq 3.6\text{ V}$	-	0.4	V
$V_{OH}^{(1)}$	Output high level voltage		2.4	-	
$V_{OL}^{(1)}$	Output low level voltage	$I_{IO} = 12\text{ mA}$ $2.4\text{ V} \leq V_{DD} < 2.7\text{ V}$	-	0.4	V
$V_{OH}^{(1)}$	Output high level voltage		$V_{DD}-0.4$	-	

(1) Guaranteed by characterization results, not tested in production.

Input AC characteristics

The definition and values of input AC characteristics are given as follows.

Table 34. Input AC characteristics

Symbol	Parameter	Min	Max	Unit
$t_{EXINTPW}$	Pulse width of external signals detected by EXINT controller	10	-	ns

4.3.12 NRST pin characteristics

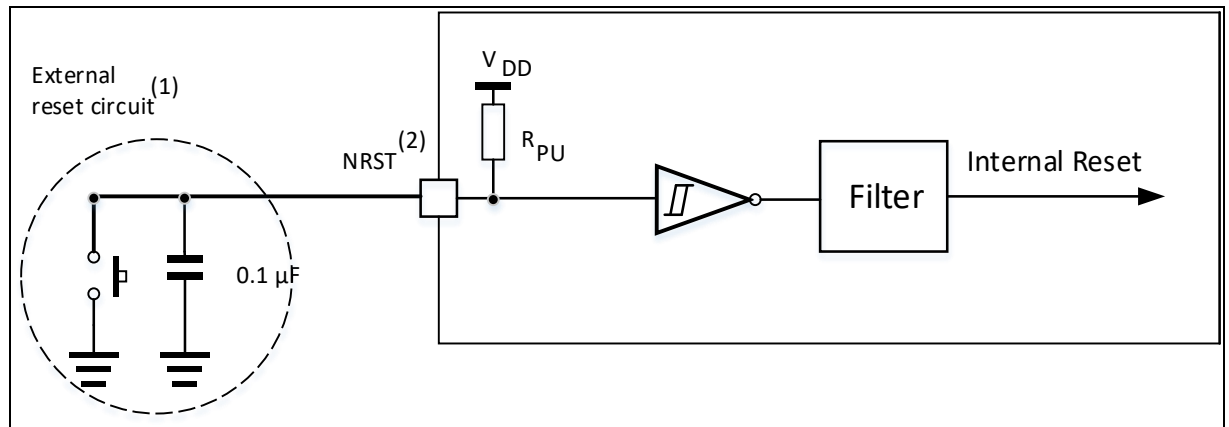
The NRST pin input driver uses CMOS technology. It is connected to a permanent pull-up resistor, R_{PU} (see the table below).

Table 35. NRST pin characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{IL(NRST)}^{(1)}$	NRST input low level voltage	-	-0.3	-	0.72	V
$V_{IH(NRST)}^{(1)}$	NRST input high level voltage	-	2	-	$V_{DD} + 0.3$	
$V_{hys(NRST)}$	NRST Schmitt trigger voltage hysteresis	-	-	400	-	mV
R_{PU}	Weak pull-up equivalent resistor	$V_{IN} = V_{SS}$	30	40	50	k Ω
$t_{iLV(NRST)}^{(1)}$	NRST input low level inactive	-	-	-	40	μ s
$t_{iLV(NRST)}^{(1)}$	NRST input low level active	-	80	-	-	μ s

(1) Guaranteed by design, not tested in production.

Figure 10. Recommended NRST pin protection



(1) The reset network protects the device against parasitic resets.

(2) The user must ensure that the level on the NRST pin can go below the $V_{IL(NRST)max}$ level specified in [Table 35](#). Otherwise the reset will not be taken into account by the device.

4.3.13 TMR timer characteristics

The parameters given in the table below are guaranteed by design.

Table 36. TMR characteristics

Symbol	Parameter	Conditions	Min	Max	Unit
$t_{res(TMR)}$	Timer resolution time	-	1	-	$t_{TMRxCLK}$
		$f_{TMRxCLK} = 120$ MHz	8.3	-	ns
f_{EXT}	Timer external clock frequency on CH1 to CH4	-	0	$f_{TMRxCLK}/2$	MHz

4.3.14 SPI characteristics

Table 37. SPI characteristics

Symbol	Parameter	Conditions	Min	Max	Unit
f_{SCK} $(1/t_{c(SCK)})^{(1)}$	SPI clock frequency ⁽²⁾⁽³⁾	Master mode	-	36	MHz
		Slave receive mode	-	36	
		Slave transmit mode	-	32	
$t_{su(CS)}^{(1)}$	CS setup time	Slave mode	$4t_{PCLK}$	-	ns
$t_{h(CS)}^{(1)}$	CS hold time	Slave mode	$2t_{PCLK}$	-	ns
$t_{w(SCKH)}^{(1)}$ $t_{w(SCKL)}^{(1)}$	SCK high and low time	Master mode, prescaler factor = 4	$2t_{PCLK} - 3$	$2t_{PCLK} + 3$	ns
$t_{su(MI)}^{(1)}$	Data input setup time	Master mode	6	-	ns
$t_{su(SI)}^{(1)}$		Slave mode	5	-	
$t_{h(MI)}^{(1)}$	Data input hold time	Master mode	4	-	ns
$t_{h(SI)}^{(1)}$		Slave mode	5	-	
$t_{a(SO)}^{(1)(4)}$	Data output access time	Slave mode	$t_{PCLK} - 2$	$2t_{PCLK} + 2$	ns
$t_{dis(SO)}^{(1)(5)}$	Data output disable time	Slave mode	$t_{PCLK} - 2$	$2t_{PCLK} + 2$	ns
$t_{v(SO)}^{(1)}$	Data output valid time	Slave mode (after enable edge)	-	25	ns
$t_{v(MO)}^{(1)}$	Data output valid time	Master mode (after enable edge)	-	10	ns
$t_{h(SO)}^{(1)}$	Data output hold time	Slave mode (after enable edge)	9	-	ns
$t_{h(MO)}^{(1)}$		Master mode (after enable edge)	2	-	

(1) Guaranteed by design, not tested in production.

(2) The maximum SPI clock frequency in slave mode should not exceed $f_{PCLK}/2$.

(3) The maximum SPI clock frequency is highly related with devices and the PCB layout. For more details about the complete solution, please contact your local Artery sales representative.

(4) Min time is for the minimum time to drive the output and the max time is for the maximum time to validate the data.

(5) Min time is for the minimum time to invalidate the output and the max time is for the maximum time to put the data in Hi-Z.

Figure 11. SPI timing diagram – slave mode and CPHA = 0

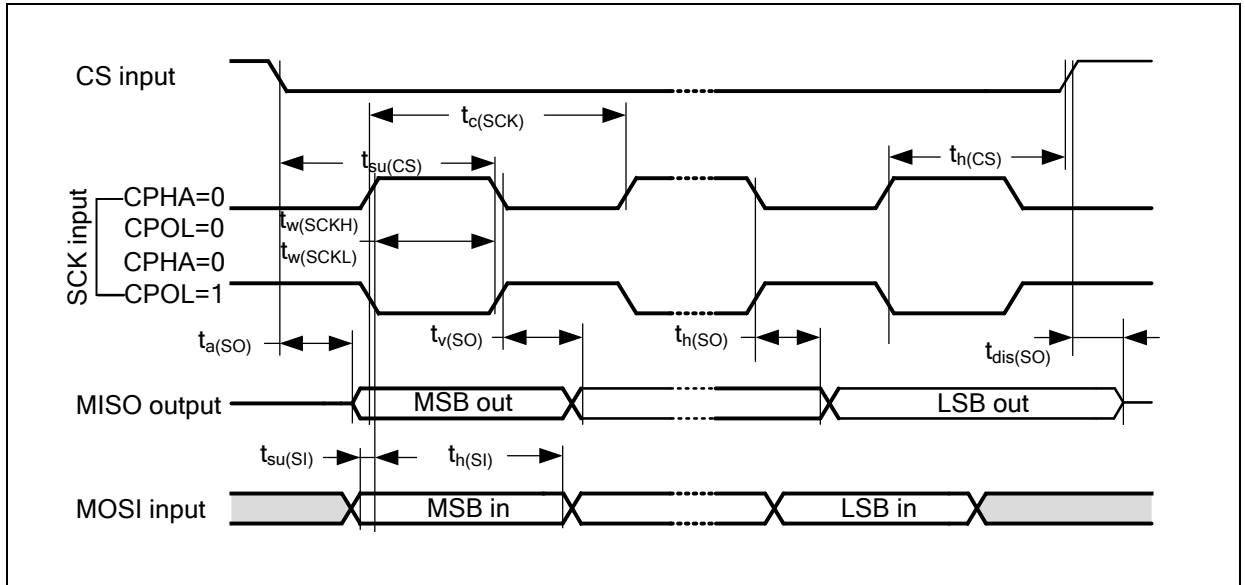


Figure 12. SPI timing diagram – slave mode and CPHA = 1

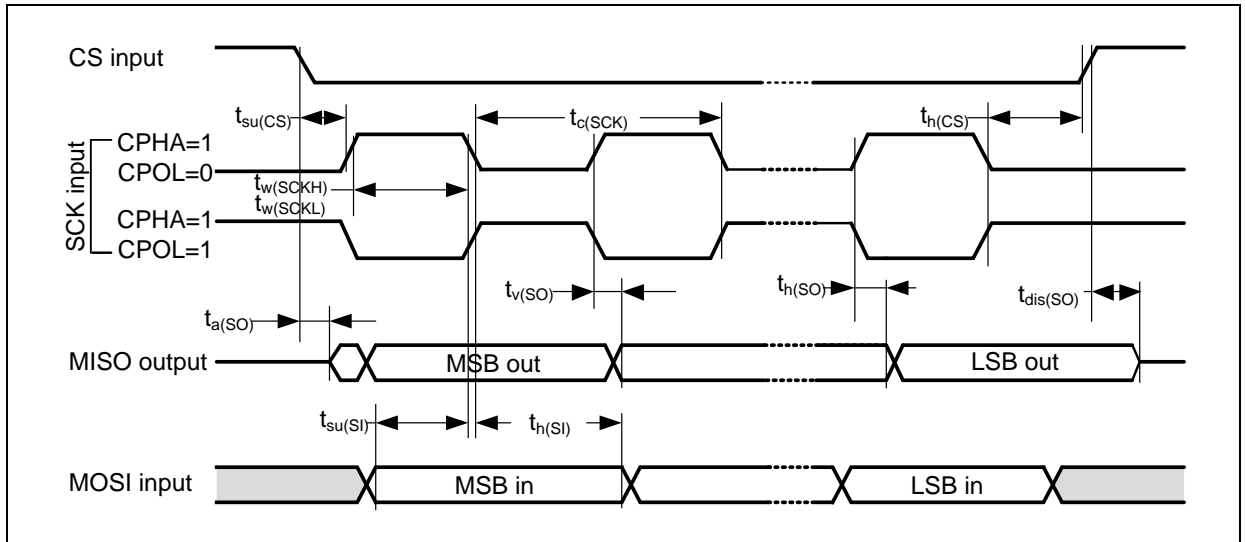
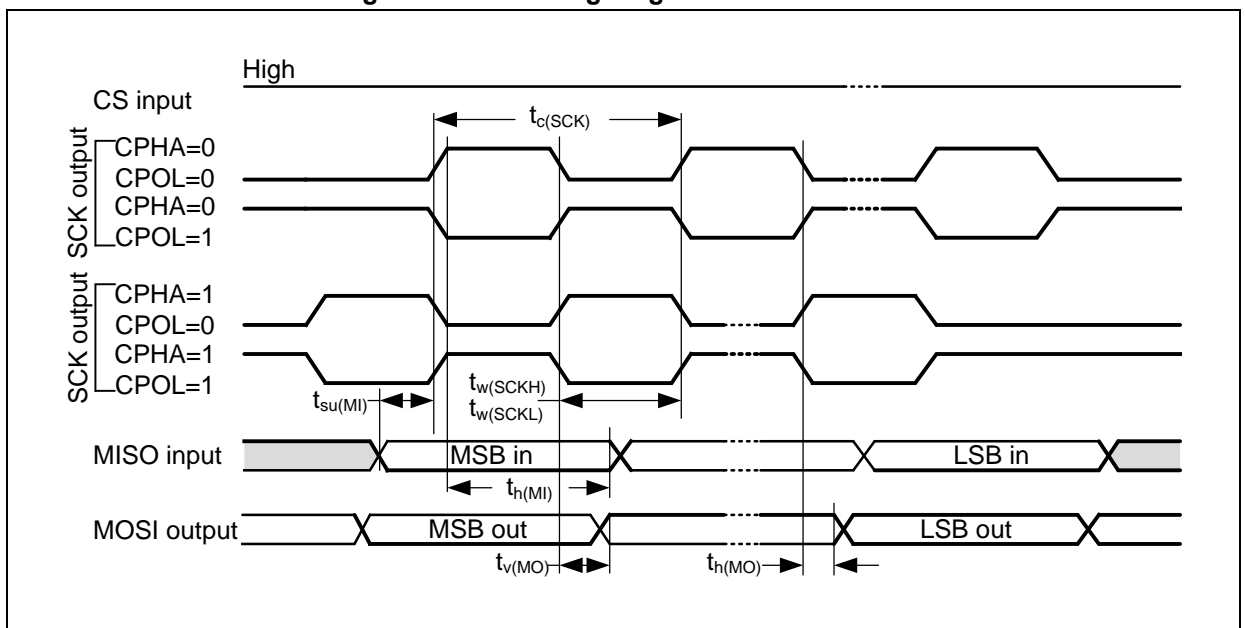


Figure 13. SPI timing diagram – master mode



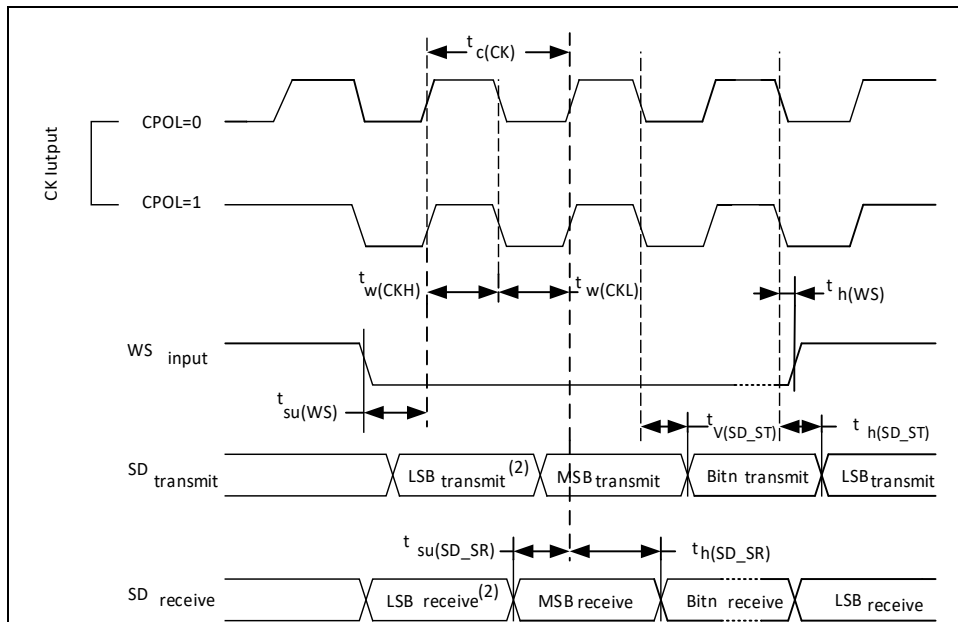
4.3.15 I²S characteristics

Table 38. I²S characteristics

Symbol	Parameter	Conditions	Min	Max	Unit
$t_r(\text{CK})$ $t_f(\text{CK})$	I ² S clock rise and fall time	Capacitive load: C = 15 pF	-	12	ns
$t_{v(\text{WS})}^{(1)}$	WS valid time	Master mode	0	4	
$t_{h(\text{WS})}^{(1)}$	WS hold time	Master mode	0	4	
$t_{su(\text{WS})}^{(1)}$	WS setup time	Slave mode	9	-	
$t_{h(\text{WS})}^{(1)}$	WS hold time	Slave mode	0	-	
$t_{su(\text{SD_MR})}^{(1)}$	Data input setup time	Master receiver	6	-	
$t_{su(\text{SD_SR})}^{(1)}$		Slave receiver	2	-	
$t_{h(\text{SD_MR})}^{(1)(2)}$	Data input hold time	Master receiver	0.5	-	
$t_{h(\text{SD_SR})}^{(1)(2)}$		Slave receiver	0.5	-	
$t_{v(\text{SD_ST})}^{(1)(2)}$	Data output valid time	Slave transmitter (after edge enable)	-	20	
$t_{h(\text{SD_ST})}^{(1)}$	Data output hold time	Slave transmitter (after edge enable)	9	-	
$t_{v(\text{SD_MT})}^{(1)(2)}$	Data output valid time	Master transmitter (after edge enable)	-	15	
$t_{h(\text{SD_MT})}^{(1)}$	Data output hold time	Master transmitter (after edge enable)	0	-	

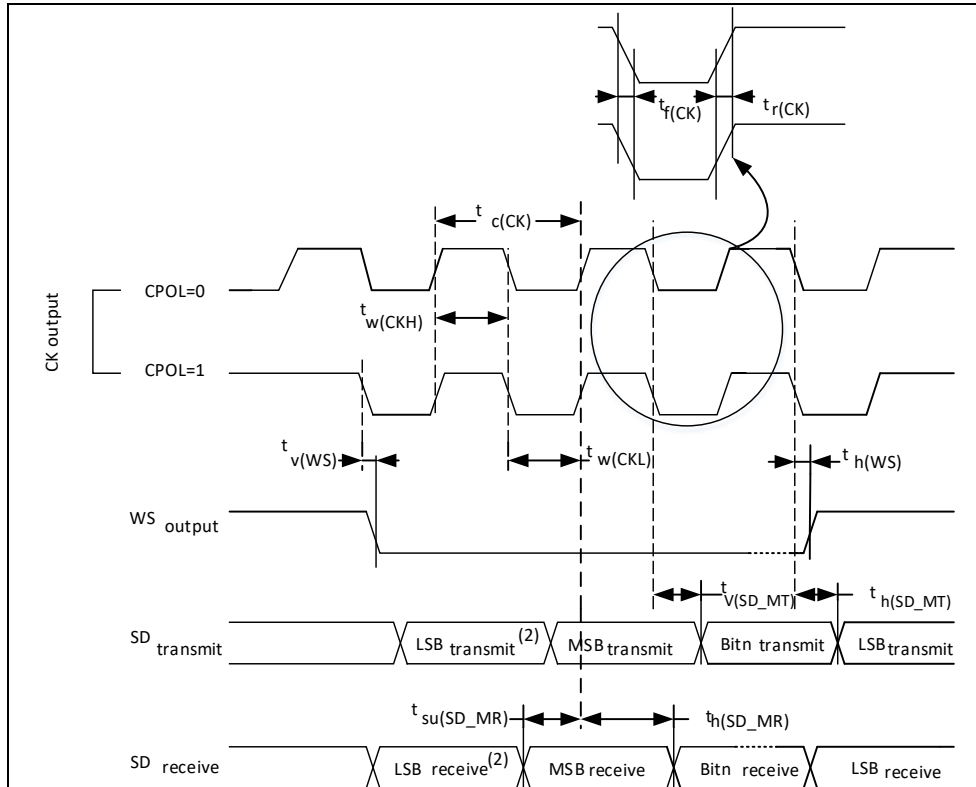
(1) Guaranteed by design, not tested in production.

(2) Depends on f_{PCLK} . For example, if $f_{\text{PCLK}} = 8 \text{ MHz}$, then $t_{\text{PCLK}} = 1/f_{\text{PCLK}} = 125 \text{ ns}$.

Figure 14. I²S slave timing diagram (Philips protocol)


(1) LSB transmit/receive of the previously transmitted byte. No LSB transmit/receive is sent before the first byte.

Figure 15. I²S master timing diagram (Philips protocol)



(1) LSB transmit/receive of the previously transmitted/received byte. No LSB transmit/receive is sent before the first byte.

4.3.16 I²C characteristics

GPIO pins SDA and SCL have limitation as follows: they are not “true” open-drain. When configured as open-drain, the PMOS connected between the GPIO pin and V_{DD} is disabled, but is still present.

I²C bus interface can support standard mode (max. 100 kHz) and fast mode (max. 400 kHz). The I²C bus frequency can be increased up to 1 MHz. For more complete information, please contact your local Artery sales office for technical support.

4.3.17 12-bit ADC characteristics

Unless otherwise specified, the parameters given in the table below are preliminary values derived from tests performed under ambient temperature, f_{PCLK2} frequency and V_{DDA} supply voltage conditions summarized in [Table 11](#).

Note: It is recommended to perform a calibration after each power-up.

Table 39. ADC characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V_{DDA}	Power supply	-	2.4	-	3.6	V
I_{DDA}	Current on the V_{DDA} input pin	-	-	480 ⁽¹⁾	560	μ A
f_{ADC}	ADC clock frequency	-	0.6	-	28	MHz
$f_s^{(2)}$	Sampling rate	-	0.05	-	2	MHz
$f_{TRIG}^{(2)}$	External trigger frequency	$f_{ADC} = 28$ MHz	-	-	1.65	MHz
		-	-	-	17	$1/f_{ADC}$
V_{AIN}	Conversion voltage range ⁽³⁾	-	0 (V_{REF-} tied to ground)	-	V_{REF+}	V
$R_{AIN}^{(2)}$	External input impedance	-	See Table 40 and Table 41			Ω
$C_{ADC}^{(2)}$	Internal sample and hold capacitor	-	-	8.5	13	pF
$t_{CAL}^{(2)}$	Calibration time	$f_{ADC} = 28$ MHz	6.61			μ s
		-	185			$1/f_{ADC}$
$t_{lat}^{(2)}$	Trigger conversion latency	$f_{ADC} = 28$ MHz	-	-	71.4	μ s
		-	-	-	2 ⁽⁴⁾	$1/f_{ADC}$
$t_s^{(2)}$	Sampling time	$f_{ADC} = 28$ MHz	0.053	-	8.55	μ s
		-	1.5	-	239.5	$1/f_{ADC}$
$t_{STAB}^{(2)}$	Power-on time	-	42			$1/f_{ADC}$
$t_{CONV}^{(2)}$	Total conversion time(including sampling time)	$f_{ADC} = 28$ MHz	0.5	-	9	μ s
		-	14~252 (t_s for sampling + 12.5 for successive approximation)			$1/f_{ADC}$

- (1) Obtained by characterization results, not tested in production.
- (2) Guaranteed by design, not tested in production.
- (3) V_{REF+} is internally connected to V_{DDA} , and V_{REF-} is internally connected to V_{SSA} .
- (4) For external triggers, a delay of $1/f_{PCLK2}$ must be added to the latency specified in [Table 39](#).

Table 40 and Table 41 are used to determine the maximum external impedance allowed for an error below 1 LSB.

Table 40. R_{AIN} max for $f_{ADC} = 14$ MHz ⁽¹⁾

T_s (Cycle)	t_s (μs)	R_{AIN} max (k Ω)
1.5	0.11	0.35
7.5	0.54	3.9
13.5	0.96	7.4
28.5	2.04	16.3
41.5	2.96	24.0
55.5	3.96	32.3
71.5	5.11	41.8
239.5	17.11	50.0

(1) Guaranteed by design.

Table 41. R_{AIN} max for $f_{ADC} = 28$ MHz ⁽¹⁾

T_s (Cycle)	t_s (μs)	R_{AIN} max (k Ω)
1.5	0.05	0.1
7.5	0.27	1.6
13.5	0.48	3.4
28.5	1.02	7.9
41.5	1.48	11.7
55.5	1.98	15.9
71.5	2.55	20.6
239.5	8.55	50.0

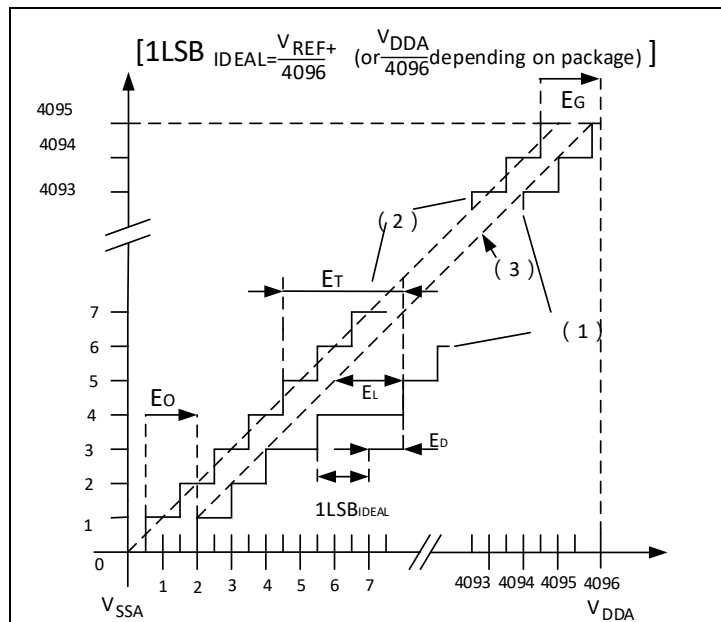
(1) Guaranteed by design.

Table 42. ADC accuracy ⁽¹⁾⁽²⁾

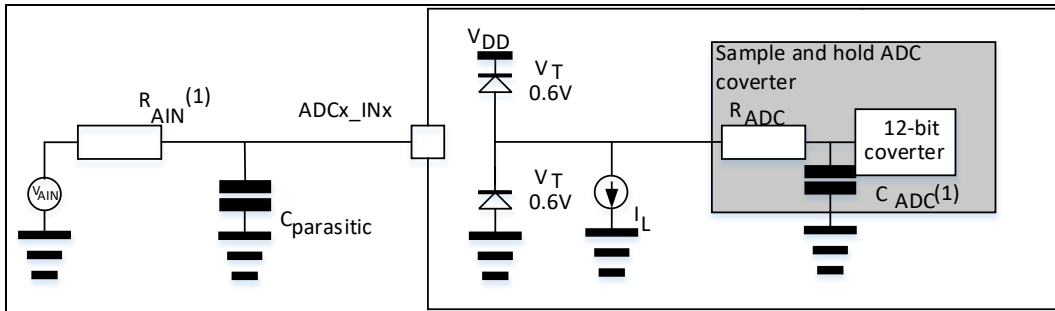
Symbol	Parameter	Test Conditions	Typ ⁽²⁾	Max ⁽²⁾	Unit
ET	Total unadjusted error	f _{PCLK2} = 56 MHz, f _{ADC} = 28 MHz, R _{AIN} < 10 kΩ, V _{DDA} = 3.0~3.6 V, T _A = 25 °C	+2	+3.5	LSB
EO	Offset error		+1	+2.5	
EG	Gain error		+1.5	+3	
ED	Differential linearity error		±0.7	±1	
EL	Integral linearity error		±0.8	±1.5	
ET	Total unadjusted error	f _{PCLK2} = 56 MHz, f _{ADC} = 28 MHz, R _{AIN} < 10 kΩ, V _{DDA} = 2.4~3.6 V, T _A = -40 ~ 105 °C	±2	+4	LSB
EO	Offset error		+1	+3	
EG	Gain error		+1.5	+3.5	
ED	Differential linearity error		±0.6	+1.5/-1	
EL	Integral linearity error		±1	±2.5	

- (1) ADC DC accuracy values are measured after internal calibration.
 (2) Obtained by characterization results, not tested in production.

Figure 16. ADC accuracy characteristics



- (1) Example of an actual transfer curve.
 (2) Ideal transfer curve.
 (3) End point correlation line.
 (4) ET = Maximum deviation between the actual and the ideal transfer curves.
 EO = Deviation between the first actual transition and the first ideal one.
 EG = Deviation between the last ideal transition and the last actual one.
 ED = Maximum deviation between actual steps and the ideal one.
 EL = Maximum deviation between any actual transition and the end point correlation line.

Figure 17. Typical connection diagram using the ADC


(1) Refer to [Table 39](#) for the values of R_{AIN} and C_{ADC} .

(2) $C_{parasitic}$ represents the capacitance of the PCB (dependent on soldering and PCB layout quality) plus the pad capacitance (roughly 7 pF). A high $C_{parasitic}$ value will downgrade conversion accuracy. To remedy this, f_{ADC} should be reduced.

General PCB design guidelines

Power supply decoupling should be performed as shown in [Figure 3](#). The 100 nF capacitors should be ceramic (good quality). They should be placed as close as possible to the chip.

4.3.18 Internal reference voltage (V_{INTRV}) characteristics

Table 43. Internal reference voltage characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{INTRV}^{(1)}$	Internal reference voltage	-	1.16	1.20	1.24	V
$T_{Coeff}^{(1)}$	Temperature coefficient	-	-	50	100	ppm/°C
$T_{S_VINTRV}^{(2)}$	ADC sampling time when reading the internal reference voltage	-	5.1	-	-	μs

(1) Obtained by characterization results, not tested in production.

(2) Guaranteed by design, not tested in production.

4.3.19 Temperature sensor (V_{TS}) characteristics

Table 44. Temperature sensor characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$T_L^{(1)}$	V_{TS} linearity with temperature	$T_A = -20 \sim 85 \text{ }^\circ\text{C}$	-	± 1	± 1.5	°C
		$T_A = -40 \sim 105 \text{ }^\circ\text{C}$	-	-	± 2	
$Avg_Slope^{(1)(2)}$	Average slope		-4.17	-4.30	-4.44	mV/°C
$V_{25}^{(1)(2)}$	Voltage at $T=25 \text{ }^\circ\text{C}$		1.22	1.28	1.34	V
$t_{START}^{(3)}$	Setup time		-	-	100	μs
$T_{S_temp}^{(3)}$	ADC sampling time when reading the temperature		5.1	-	-	μs

(1) Obtained by characterization results, not tested in production.

(2) The temperature sensor output voltage changes linearly with temperature. The offset of this line varies from chip to chip due to process variation (up to 50 °C from one chip to another). The internal temperature sensor is more suited to applications that detect temperature variations instead of absolute temperatures. If accurate temperature readings are needed, an external temperature sensor part should be used.

(3) Guaranteed by characterization design, not tested in production.

Obtain the temperature using the following formula:

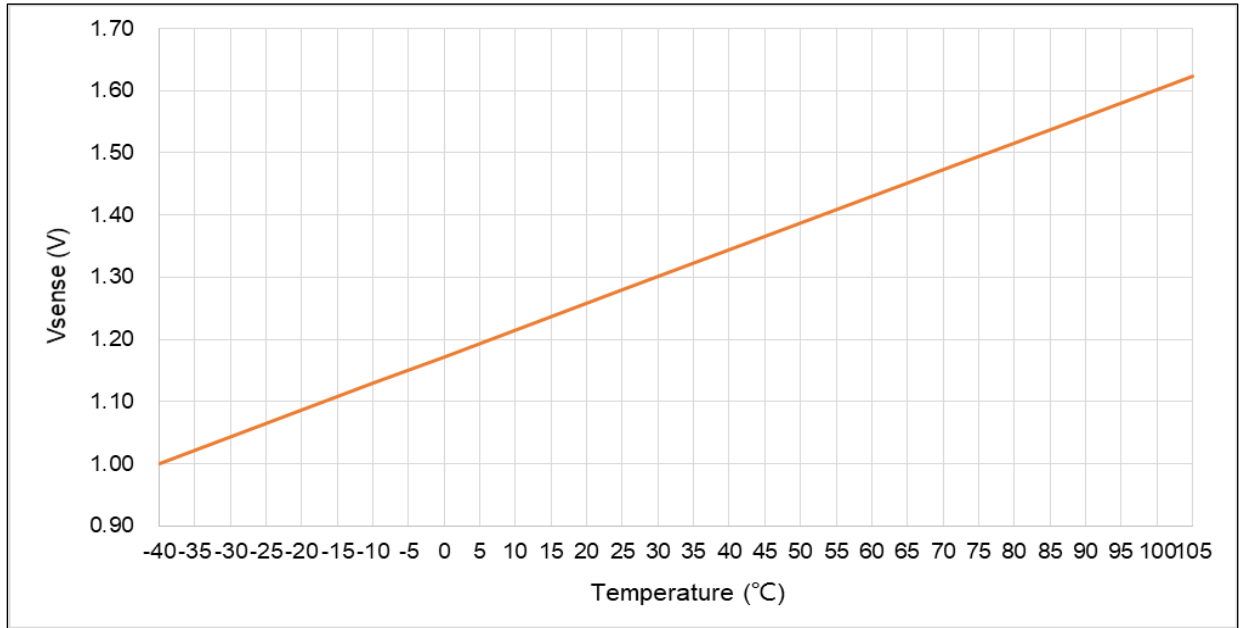
$$\text{Temperature (}^\circ\text{C)} = \{(V_{25} - V_{TS}) / Avg_Slope\} + 25$$

Where,

$$V_{25} = V_{TS} \text{ value for } 25 \text{ }^\circ\text{C}$$

Avg_Slope = average slope for curve between temperature vs. V_{SENSE} (given in mV/°C)

Figure 18. V_{TS} vs. temperature (ideal curve)

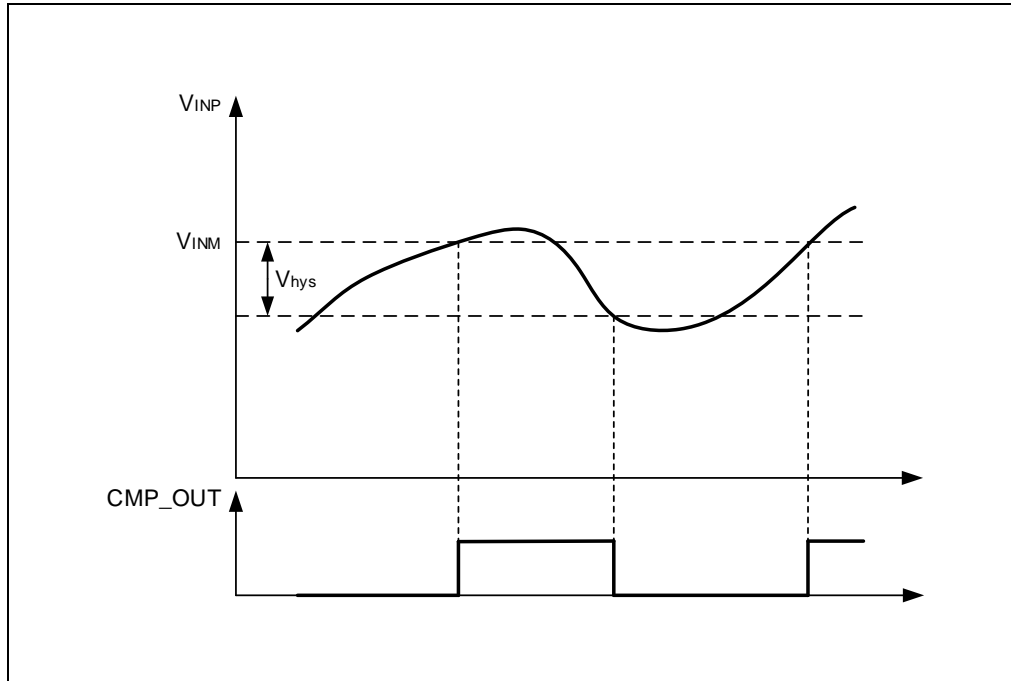


4.3.20 Comparator (CMP) characteristics

Table 45. Comparator characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V_{DDA}	Power supply voltage	-	2.4	-	3.6	V
V_{IN}	Input voltage range	-	0	-	V_{DDA}	V
t_{START}	Startup time	High speed mode	-	1.0	3.5	μs
		Medium speed mode	-	2.8	5	
		Low power mode	-	8	13	
		Ultra low power mode	-	12	18	
t_D	Propagation delay for 200 mV step with 100 mV overdrive	High speed mode	-	40	100	ns
		Medium speed mode	-	240	320	
		Low power mode	-	500	820	
		Ultra low power mode	-	800	1800	
V_{offset}	Offset voltage	-	-	± 4	± 15	mV
V_{hys}	Hysteresis	No hysteresis	-	0	1	mV
		Low hysteresis	5	8	17	
		Medium hysteresis	10	18	37	
		High hysteresis	18	38	70	
I_{DDA}	Current on V_{DDA} input pin	High speed mode	-	40	61	μA
		Medium speed mode	-	9.7	13.9	
		Low power mode	-	3.2	4.7	
		Ultra low power mode	-	1.9	2.8	

(1) Guaranteed by characterization results, not tested in production.

Figure 19. Comparator hysteresis


4.3.21 Operational amplifier (OP) characteristics

Table 46. OP characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min ⁽¹⁾	Typ	Max ⁽¹⁾	Unit
V _{DD}	Power supply voltage	-	2.4	-	3.6	V
I _{DD}	Current on V _{DD} input pin	Non-loaded, quiescent mode	-	940	1600	μA
V _{offset}	Offset voltage	T _A = 25 °C	-	0.8	3.5	mV
		T _A = -40 ~ 105 °C	-	-	4.5	
V _{IN}	Input voltage range	-	0	-	V _{DD}	V
V _{COMM}	Common mode input range	-	0.1	-	V _{DD} - 0.1	mV
V _{OUT}	Output voltage range	-	0.1	-	V _{DD} - 0.1	V
I _{OUT}	Output current	-	38	53	-	mA
A _o	Open loop gain	-	82	97	-	dB
GBW	Gain bandwidth	R _L = 10 kΩ, C _L = 100 pF	-	6	-	MHz
SR	Slew rate	Gain = +1, 2 V successive, R _L = 10 kΩ	-	4.2	-	V/μs
CMRR	Common mode rejection ratio	-	67	90	-	dB
PSRR	Power supply rejection ratio	-	70	91	-	dB

(1) Guaranteed by design, not tested in production.

5 Package information

5.1 LQFP48 – 7 x 7 mm

Figure 20. LQFP48 – 7 x 7 mm 48 pin low-profile quad flat package outline

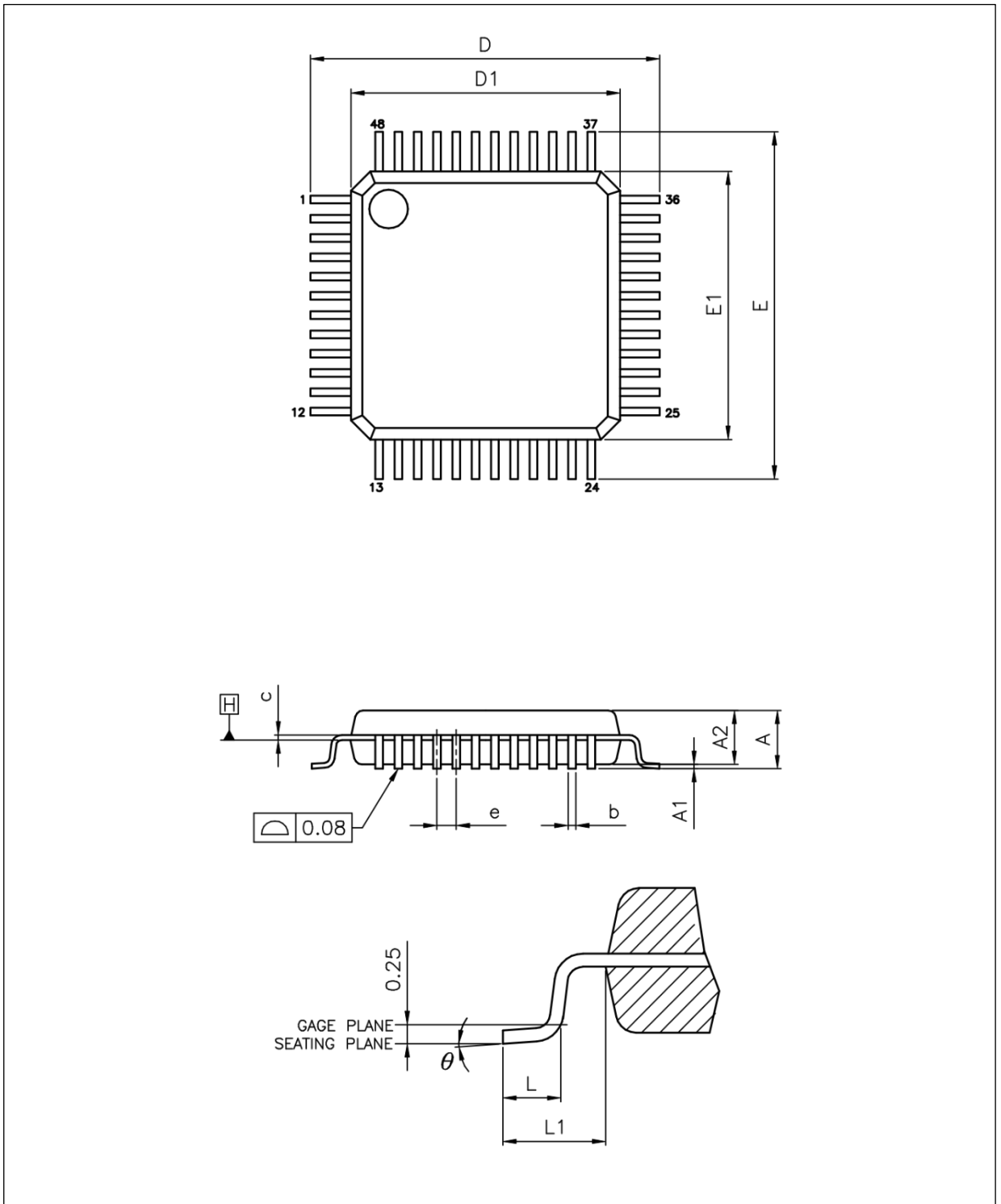
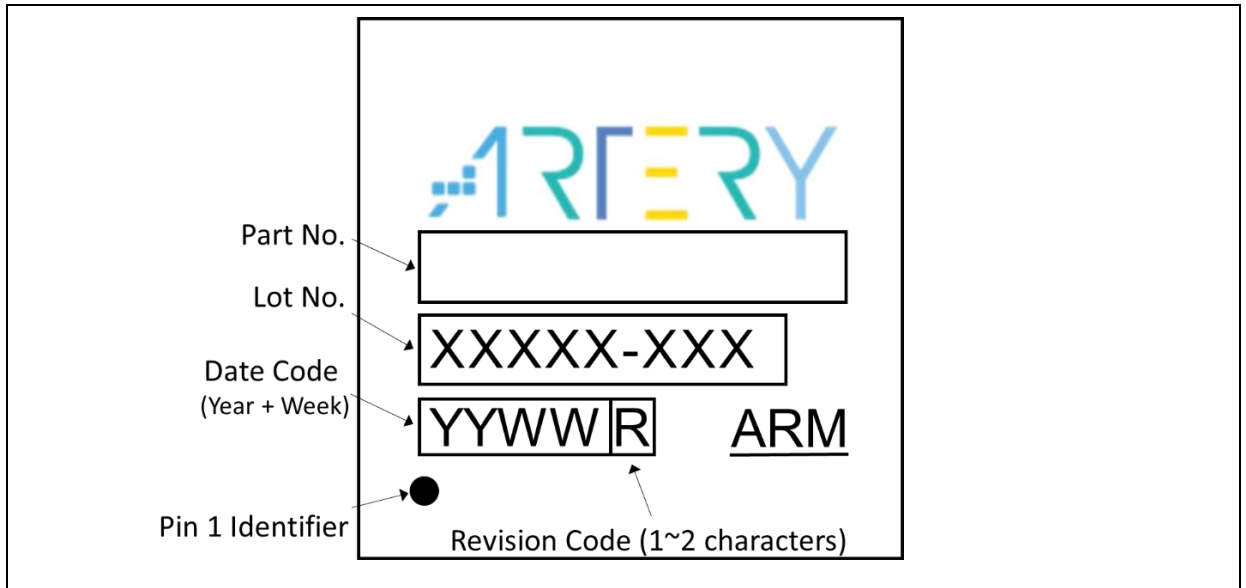


Table 47. LQFP48 – 7 x 7 mm 48 pin low-profile quad flat package mechanical data

Symbol	Millimeters		
	Min	Typ	Max
A	-	-	1.60
A1	0.05	-	0.15
A2	1.35	1.40	1.45
b	0.17	0.22	0.27
c	0.09	-	0.20
D	8.80	9.00	9.20
D1	6.90	7.00	7.10
E	8.80	9.00	9.20
E1	6.90	7.00	7.10
e	0.50 BSC.		
Θ	0°	3.5°	7°
L	0.45	0.60	0.75
L1	1.00 REF.		

5.2 Device marking

Figure 21. Marking example


(1) Not to scale.

5.3 Thermal characteristics

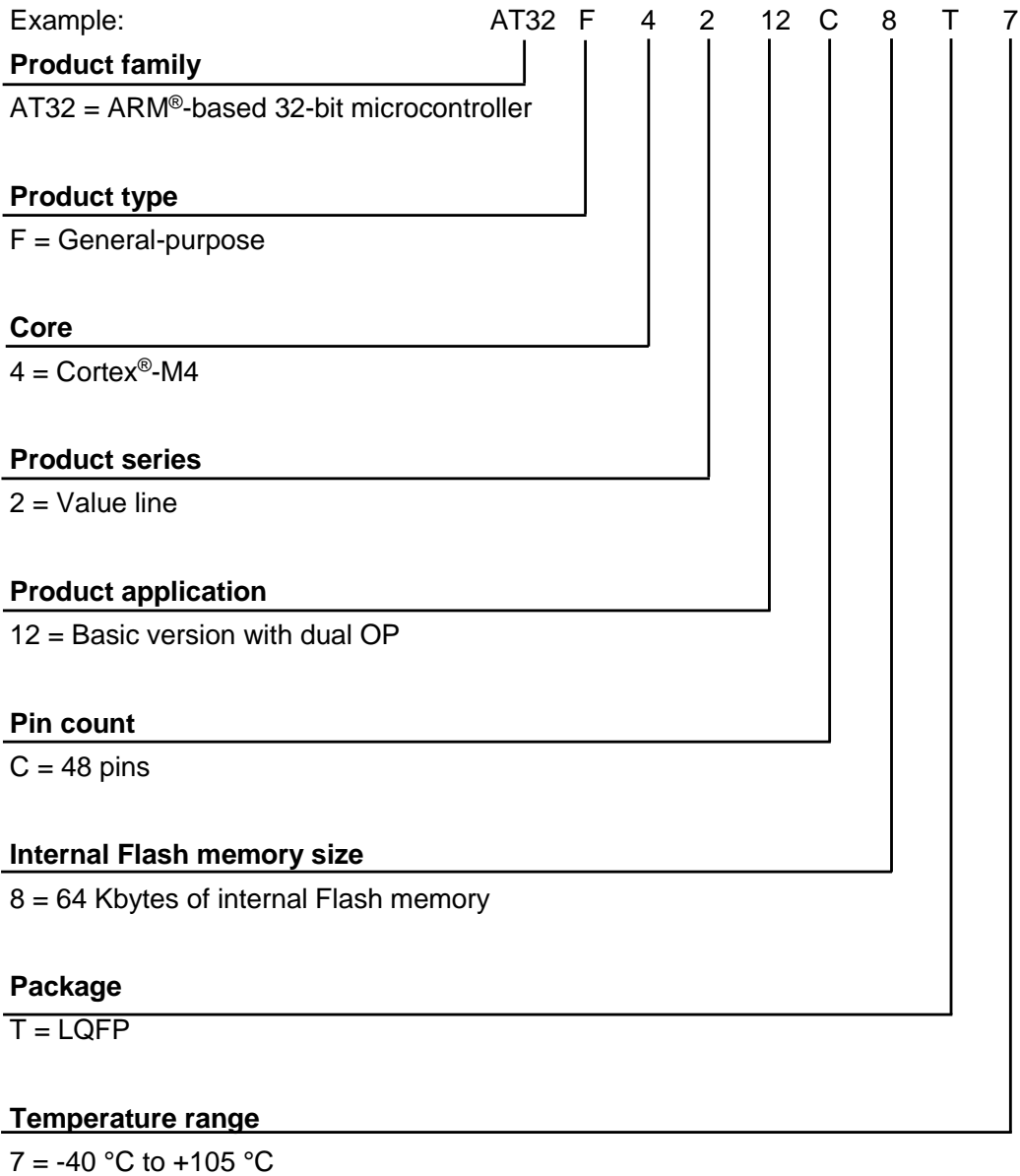
Thermal characteristics are calculated based on a two-layer board that uses FR-4 material of 1.6mm thickness. They are guaranteed by design, not tested in production.

Table 48. Package thermal characteristics

Symbol	Parameter	Value	Unit
Θ_{JA}	Thermal resistance junction-ambient – LQFP48 – 7 x 7 mm	87.0	°C/W

6 Part numbering

Table 49. AT32F4212 series part numbering



For a list of available options (speed, package, etc.) or for more information on this device, please contact your local Artery sales office.

7 Document revision history

Table 50. Document revision history

Date	Version	Change
2022.10.12	2.00	Initial release
2023.10.17	2.01	1. Added note [3] to Table 32 2. Modified Table 37 and Table 38 3. Updated "IMPORTANT NOTICE" at the end of this file

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