Features

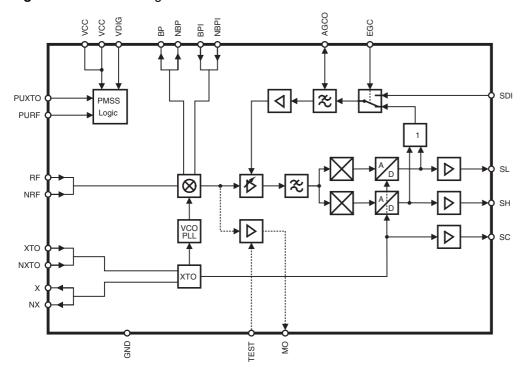
- Very Low Power Design (40 mW)
- Single IF Architecture
- Excellent Noise Performance
- 1.5-bit ADC On Chip
- Small QFN Package (4 mm × 4 mm, 24 pins)
- . Highly integrated, Few External Components
- Advanced BiCMOS Technology (UHF6s)
- Non ESD Sensitive Device



1. Description

The ATR0601 is a single IF GPS front-end IC, designed to meet the requirements of mobile and automotive applications. Excellent RF performance combined with low noise figure enables high quality GPS solutions and it's very low power consumption fits perfectly to portable devices. Featuring a balanced XTO and a fully integrated balanced frequency synthesizer, only few external components are required. The ATR0601 offers a complete autonomous mode, utilizing the on chip AGC in closed loop operation, to set the gain of the IF VGA. Alternatively, in combination with the baseband processor ATR0621 the optimum gain of the IF VGA can be computed and set by software, using the digital SDI interface.

Figure 1-1. Block Diagram





GPS Front-end IC

ATR0601

Preliminary

Rev. 4866A-GPS-08/05





2. Pin Configuration

Figure 2-1. Pinning QFN24

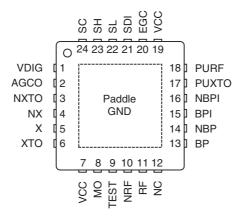


Table 2-1. Pin Description

Pin	Symbol	Type ⁽¹⁾	Function	
Paddle	GND	S	Common ground	
1	VDIG	S	Digital supply	
2	AGCO	A_I/O	AGC: gain control voltage output/corner frequency determination	
3	NXTO	A_I	XTO interface (optional: TCXO input)	
4	NX	A_O	XTO interface	
5	Х	A_O	XTO interface	
6	XTO	A_I	XTO interface (optional: TCXO input)	
7	VCC	S	Analog supply	
8	MO	A_O	Testbuffer output (f _{IF})	
9	TEST	A_I	Enable testbuffer	
10	NRF	A_I	RF input complementary	
11	RF	A_I	RF input	
12	NC	_	Not connected	
13	BP	A_O	IF-Filter interface (mixer output, open collector)	
14	NBP	A_O	IF-Filter interface (mixer output complementary, open collector)	
15	BPI	A_I	IF-Filter interface (IF-input)	
16	NBPI	A_I	IF-Filter interface (IF-input complementary)	
17	PUXTO	D_I	Power-up XTO	
18	PURF	D_I	Power-up RF	
19	VCC	S	Analog supply	
20	EGC	D_I	Enable external gain control (high = external; low = internal)	
21	SDI	D_I	Input for external gain control signal ($\Sigma\Delta$ modulation)	
22	SL	D_O	Data output: "low"	
23	SH	D_O	Data output: "high"	
24	SC	D_O	Sample clock	

Note: 1. Type: A_I Analog input, A_O Analog output, D_I Digital input, D_O Digital output, S Supply

2 ATR0601 [Preliminary]

3. Functional Description

3.1 General Description

The ATR0601 GPS receiver IC has been especially designed for GPS applications in both mobile phone and automotive applications. From this system point of view, it incorporates highest isolation between GPS and cellular bands, as well as very low power consumption.

The L1 input signal (f_{RF}) is a Direct Sequence Spread Spectrum (DSSS) signal with a centre frequency of: $f_{RF} = 1575.42$ MHz. The digital modulation scheme is Bi-Phase-Shift-Keying (BPSK) with a chip rate of 1.023 Mbps. As the input signal power at the antenna is approximately -140 dBm, the desired signal is below the thermal noise floor.

3.2 PMSS Logic

The Power Management, Startup and Shutdown Logic ensures reliable operation within the recommended operating conditions. The external power control signals PUrf and PUxto are passed thru Schmitt-trigger inputs to eliminate voltage ripple and prevent undesired behaviour during startup and shutdown. Digital and analog supply voltages are analyzed by a monitoring circuit, enabling the startup of the IC only within a secure operating area.

3.3 XTO

The XTO is designed for minimum phase noise and frequency perturbations. The balanced topology gives maximum isolation from external and ground coupled noise. The built-in jump start circuitry ensures reliable start-up behaviour of any specified crystal. For use with an external TCXO, the XTO circuitry can be used as a single-ended or balanced input buffer.

The recommended reference frequency is: $f_{XTO} = 23.104$ MHz.

3.4 VCO/PLL

The frequency synthesizer features a balanced VCO and a fully integrated loop filter, thus no external components are required. The VCO combines very good phase noise behaviour and excellent spurious suppression. The relation between the reference frequency (f_{XTO}) and the VCO centre frequency (f_{VCO}) is given by: $f_{VCO} = f_{XTO} \times 64 = 23.104 \text{ MHz} \times 64 = 1478.656 \text{ MHz}$.

3.5 RF-Mixer/Image-filter

Combined with the antenna an external LNA provides a first band path filtering of the signal. For the LNA, Atmel's ATR0610 is recommended, due to it's low Noise Figure, high linearity an low power consumption. The output of the LNA drives an SAW filter, which provides image rejection for the mixer and the required isolation of all GSM bands. The output of the SAW filter is fed into a highly linear mixer with high conversion gain and excellent noise performance.

The IF frequency (f_{IF}) is given by: $f_{IF} = f_{RF} - f_{VCO} = 1575.42 \text{ MHz} - 1478.656 \text{ MHz} = 96.764 \text{ MHz}$.

3.6 IF-filter

The mixer directly drives an external LC-Bandpath filter via open collector outputs. In order to provide highest selectivity and conversion gain, it is recommended to design the external filter, according to the application proposal, as a 2-pole filter with a quality factor Q > 25.





3.7 VGA/AGC

The output of the IF-Filter drives an on-chip Variable Gain Amplifier (VGA) which is combined with additional low-pass filtering. The on-chip Automatic Gain Control (AGC) stage sets the gain of the VGA in order to optimally charge the input of the following analog-to-digital converter. The AGC control loop can be selected for on-chip closed loop operation or for external gain control mode. For external gain control mode, the loop needs to be closed by the baseband IC ATR0621.

3.8 A/D Converter

The analog-to-digital converter stage has a total resolution of 1.5 bit. It comprises balanced comparators and a sub sampling unit, clocked by the reference frequency (f_{XTO}). The frequency spectrum of the digital output signal (f_{OUT}), present at the data outputs SL and SH, is then given by: $f_{OUT} = \left| f_{IF} - f_{XTO} \times n \right|$. The selected sub sampling factor (n = 4) leads to the designated digital output signal, with a centre frequency given by:

 $f_{OUT} = f_{IF} - f_{XTO} \times \, 4 = 96.764 \; MHz - 23.104 \; MHz - 4 = 4.348 \; MHz. \label{eq:four_four_four_four}$

3.9 Clock and Data Driver

CMOS output drivers are providing sign and magnitude bits as well as the system clock to the baseband IC ATR0621. The rail-to-rail output signal level is determined by the digital supply voltage (VDIG).

4. Absolute Maximum Ratings

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Parameters	Symbol	Value	Unit
Analog supply voltage	V _{CC}	−0.3 to +3.7	V
Digital supply voltage	V _{DIG}	-0.3 to +3.7	V
Input voltage	V _{in}	−0.3 to +3.7	V
Operating temperature	T _{op}	-40 to +85	°C
Storage temperature	T _{stg}	-55 to +125	°C

5. Thermal Resistance

Parameters	Symbol	Value	Unit
Junction ambient	R _{th}	125	K/W

6. Operating Range

Parameters	Symbol	Value	Unit
Analog supply voltage	V _{CC}	2.70 to 3.30	V
Digital supply voltage	V _{DIG}	1.65 to 2.00	V
Supply voltage difference $(V_{\Delta} = V_{CC} - V_{DIG})$	$V_{\scriptscriptstyle{\Delta}}$	≥ 0.80	V
Temperature range	Temp	-40 to +85	°C
Input frequency	f _{RF}	1575.42	MHz
Reference frequency	f _{XTO}	23.104	MHz

7. Electrical Characteristics

No.	Parameters	Test Conditions	Pin	Symbol	Min.	Тур.	Max.	Unit	Type*
1	Common								
1.1	Analog supply current ⁽¹⁾	$V_{PUxto} = V_{PUrf} = V_{PU,on}$	7, 19	I _S		14.2		mA	Α
1.2	Digital supply current (1),(2)	$V_{PUxto} = V_{PUrf} = V_{PU,on}$	1	I _{DIG}		700		μΑ	Α
1.3	Analog supply current in XTO mode ⁽¹⁾	$V_{PUxto} = V_{PU,on},$ $V_{PUrf} = V_{PU,off}$	7, 19	I _{S_XTO}		2.9		mA	Α
1.4	Digital supply current in XTO mode ^{(1),(3)}	$V_{PUxto} = V_{PU,on},$ $V_{PUrf} = V_{PU,off}$	1	I _{DIG_XTO}		500		μΑ	Α
1.5	Supply current in power down mode ⁽¹⁾	$V_{PUxto} = V_{PUrf} = V_{PU,off}$	1, 7, 19	I _{PD}			2	μΑ	Α
1.6	Maximum total gain	$V_{AGCO} = 2.2V$		G _{max_tot}		90		dB	В
1.7	Noise figure (SSB)			NF _{tot}			6.8	dB	С

^{*)} Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter

Notes: 1. Conditions: $V_{CC} = 2.7V$; $V_{DIG} = 1.65V$; Temperature = $27^{\circ}C$

- 2. Capacitive load ($C_L = 3.3 \, pF$) at pins 22, 23, 24
- 3. Capacitive load ($C_L = 3.3 \text{ pF}$) at pin 24





7. Electrical Characteristics (Continued)

No.	Parameters	Test Conditions	Pin	Symbol	Min.	Тур.	Max.	Unit	Type*
2	Mixer								
2.1	Output frequency	f _{XTO} = 23.104 MHz	13, 14	f _{IF}		96.764		MHz	Α
2.2	Input impedance (balanced)	f _{RF} = 1575.42 MHz	10, 11	Z ₁₁		10-j80		Ω	С
2.3	Conversion Gain	Recommended IF-filter	8	G _{MIX}		20		dB	В
2.4	Noise figure (SSB)		8	NF _{MIX}		5.8		dB	С
3	VGA/AGC								
3.1	Minimum gain	$V_{AGCO} = 1.0V$		$G_{VGA,min}$		0		dB	В
3.2	Maximum gain	$V_{AGCO} = 2.2V$		$G_{VGA,max}$		70		dB	В
3.3	Control-voltage sensitivity	$V_{AGCO} = 2.2V$		$N_{\text{VGA},\text{min}}$		6.6		dB/V	D
0.0	Control-voltage sensitivity	$V_{AGCO} = 1.0V$		$N_{\text{VGA},\text{max}}$		150		dB/V	D
3.4	AGC cut-off frequency	C _{ext} = open	2	f _{3dB_AGC}		250		kHz	D
3.5	AGC cut-off frequency	C _{ext} = 100 pF	2	f _{3dB_AGC}		33		kHz	D
3.6	Gain-control output voltage		2	V_{AGCO}	0.9		2.3	V	В
4	Reference Oscillator								
4.1	XTO phase noise at 100 Hz	With specified crystal	24	Pn ₁₀₀		-80		dBc/Hz	С
4.2	XTO phase noise at 1 kHz	With specified crystal	24	Pn _{1k}		-100		dBc/Hz	С
5	Clock and Data Driver								
5.1	Clock driver frequency	f _{XTO} = 23.104 MHz	24	f _{CLK}		23.104		MHz	Α
5.2	Clock output level	C _{load,max} = 10 pF	24	$V_{CLK,high}$		$0.9 \times V_{DIG}$		V	В
5.3	Clock output level	C _{load.max} = 10 pF	24	$V_{\rm CLK,low}$		$0.1 \times V_{DIG}$		V	В
5.4	Data output level	C _{load,max} = 10 pF	22, 23	$V_{\text{Data,high}}$		$0.9 \times V_{DIG}$		V	В
5.5	Data output level	C _{load,max} = 10 pF	22, 23	$V_{\text{Data,low}}$		$0.1 \times V_{DIG}$		V	В
6	PMSS								
6.1	Voltage level power-on		17, 18	$V_{PU,on}$	1.3			V	Α
6.2	Voltage level power-off		17, 18	$V_{PU,off}$			0.5	V	Α

^{*)} Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter

Notes: 1. Conditions: $V_{CC} = 2.7V$; $V_{DIG} = 1.65V$; Temperature = $27^{\circ}C$

- 2. Capacitive load ($C_L = 3.3 \text{ pF}$) at pins 22, 23, 24
- 3. Capacitive load ($C_L = 3.3 \text{ pF}$) at pin 24

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8. Timing

Figure 8-1. Recommended Power-up/down Sequence

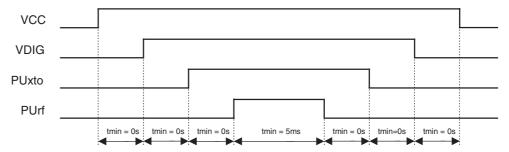


Figure 8-2. Recommended Sleep-mode Sequence

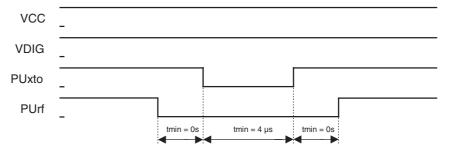


Figure 8-3. Recommended XTO Start-up/Shut-down Sequence

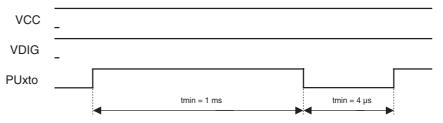


Figure 8-4. Sample Clock Start-up Delay

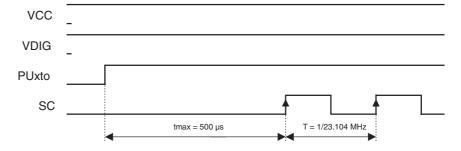






Figure 8-5. Synchronous Shut-down Behaviour of SC with Respect to PUxto

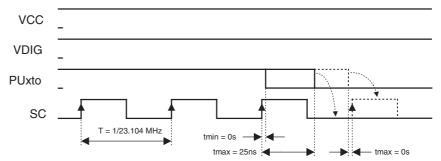
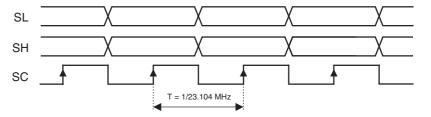
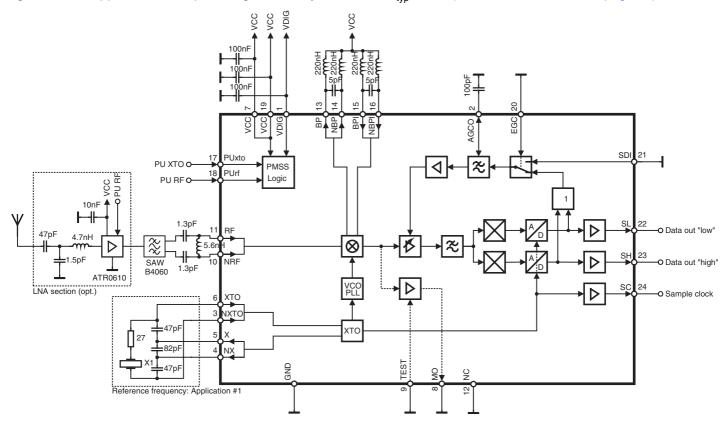


Figure 8-6. Data Outputs SL and SH are Valid with Rising Edge of Sample Clock SC



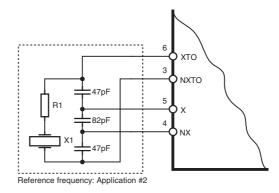
9. Application Circuit

Figure 9-1. Application Example Using a GPS Crystal with $ESR_{typ} = 12\Omega$ (Please see Table 9-1 on page 11)



Note: Please consider the recommended IF-filter layout, shown in Figure 9-5 on page 11.

Figure 9-2. Application Example Using a GPS Crystal with $ESR_{typ} \neq 12Ω$ (Please see Table 9-2 on page 11)

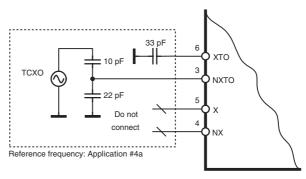


Note: The external series resistor R1 has to be selected depending on the typical value of the crystal ESR. Please refer to Application Note "ATR0601: Crystal and TXCO selection".





Figure 9-3. Equivalent Application Examples Using a GPS TCXO (Please see Table 9-3 on page 12)



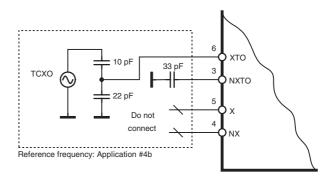


Figure 9-4. Application Example Using an External Reference and Balanced Inputs (Please see Table 9-4 on page 12)

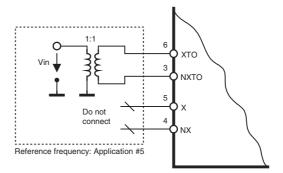
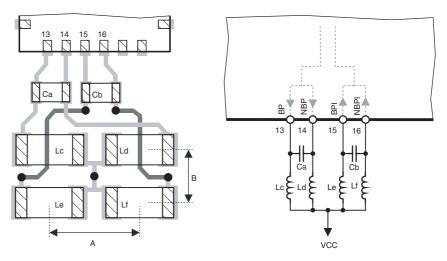


Figure 9-5. Recommended IF-filter: Layout versus Schematic



Note: Mutual inductance between the four inductors Lc - Lf plays an important role in the IF-filter characteristics. In any design, the layout arrangement shown in Figure 9-5 on page 11 should be resembled as close as possible. Measures: A = 2.8 mm; B = 1.4 mm; Lc - Lf: Wirewound SMD inductors, 0603 size. (Please see Table 10-1).

Table 9-1. Specification of GPS Crystals Appropriate for the Application Example Shown in Figure 9-1 on page 9

Parameter	Comment	Min.	Тур.	Max.	Units
Frequency Characteristics	•				
Fundamental Frequency	Nominal frequency referenced to 25°C		23.104		MHz
Calibration tolerance	Frequency at 23°C ±2°C			7.0	±ppm
Frequency deviation	Over operating temperature range			15.0	±ppm
Temperature range	Operating temperature range	-40.0		+85.0	°C
Electrical					
Load capacitance (CL)		18.5		19.5	pF
Equivalent Series Resistance (ESR)					
Fundamental	Specification	7	12	23	Ω

Table 9-2. Specification of GPS Crystals Appropriate for the Application Example Shown in Figure 9-2 on page 9

Parameter Comment		Min.	Тур.	Max.	Units
Equivalent Series Resistance (ESR)					
Fundamental	Specification	7		40	Ω

Note: All other parameters as specified in Table 9-1.





Table 9-3. Specification of GPS TCXOs Appropriate for the Application Example Shown in Figure 9-3 on page 10

Parameter	Comment	Min.	Тур.	Max.	Units
Frequency Characteristics					
Nominal Frequency	Nominal frequency referenced to 25°C		23.104		MHz
Frequency deviation	Over operating temperature range			2.0	±ppm
Temperature range	Operating temperature range	-40.0		+85.0	°C
Electrical					
Output waveform	DC coupled clipped sinewave				
Output voltage (peak-to-peak)	At minimum supply voltage		0.8		V
Output load capacitance	Tolerable load capacitance	10			pF

Table 9-4. Specification of an external reference signal for the application example shown in Figure 9-4 on page 10

Parameter	Comment	Min.	Тур.	Max.	Units	
Signal Characteristics						
Nominal Frequency			23.104		MHz	
Waveform	eform Sinewave or clipped sinewave					
Amplitude	Voltage peak-to-peak	0.6	0.8	1.0	V	

10. Demonstration Board

Figure 10-1. Schematic of Demonstration Board

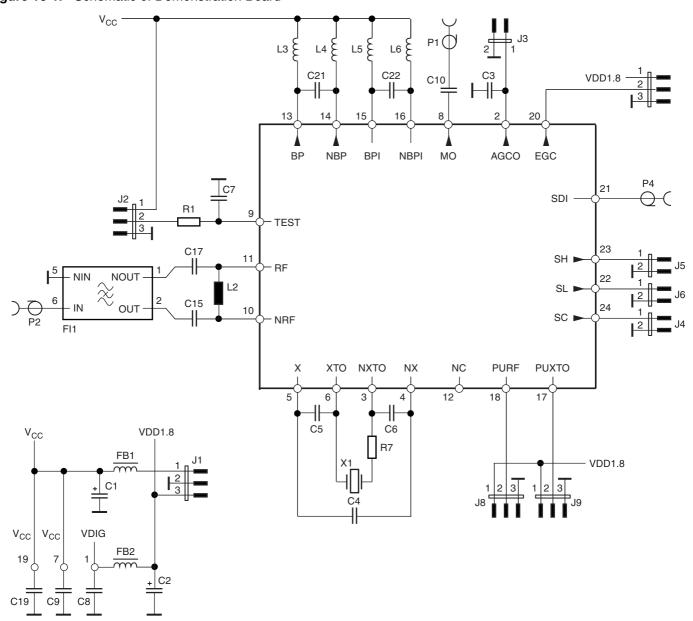






Figure 10-2. Illustration of Demonstration Board

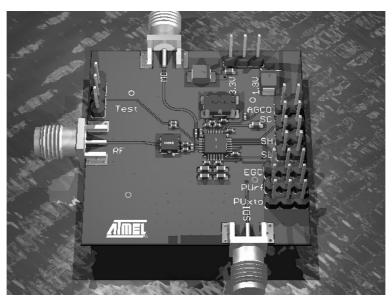


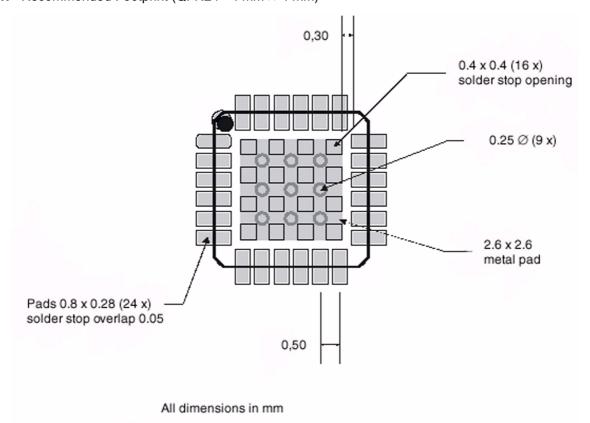
Table 10-1. BOM of Demonstration Board

Qty	Value	Device	Parts	Tolerance	Manufacturer	Mfr. Order Code
4		JP2E	J3, J4, J5, J6		Molex	90120-0762
5		JP3E	J1, J2, J7, J8, J9		Molex	90120-0763
1	0	RESISTOR-0402	R1		Vishay	CRCW0402000Z
1	2n2	CAPACITOR-0402	C3	5%	Vishay	VJ0402Y222JXJA
2	1p3	CAPACITOR-0402	C15, C17	0,1 pF	Tayo Yuden	EVK105CH1R3BW
1	5n6 2% Multilayer	INDUCTOR-0402	L2	2%	Würth Elektronik	744784056G
2	5p0 ±0p1	CAPACITOR-0402	C21, C22	±0p1	Yageo America	0402CG509C9B200
2	10μ	ELKO-B	C1, C2	20%	Vishay	293D106X0016B2
1	27	RESISTOR-0402	R7	5%	Vishay	CRCW040227RJ
2	47p	CAPACITOR-0402	C5, C6	5%	Vishay	VJ0402A470JXXA.
1	82p	CAPACITOR-0402	C4		Vishay	VJ0402A820JXXA
4	100n	CAPACITOR-0402	C7, C8, C9, C19	5%	Vishay	VJ0402V104JXJ
1	100p	CAPACITOR-0402	C10	5%	Vishay	VJ0402A101JXXA.
3	142-0711-841	COAX-SMA	P1, P2, P4		Johnson Comp.	142-0711-841
4	220n 2%	INDUCTOR_WIRE- WOUND-0603	L3, L4, L5, L6	2%	Würth Elektronik	744761222G
2	74279266	FERRITE_BEAD-0603	FB1, FB2		Würth Elektronik	74279266
1	ATR0601-1	ATR0601-1	IC1		Atmel	ATR0601
1	B4060	FILTER-BALANCED	FI1		Epcos	B4060
1	RSX-5 23.104 MHz	XTAL-4PIN-6035	X1		Rakon	XZC736 IEC19RSX-5 23.104 MHz

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11. Recommended Footprint

Figure 11-1. Recommended Footprint (QFN24 - 4 mm × 4 mm)





12. Ordering Information

Extended Type Number	Package	Remarks
ATR0601-PFQW	QFN24, 4 × 4	Taped and reeled

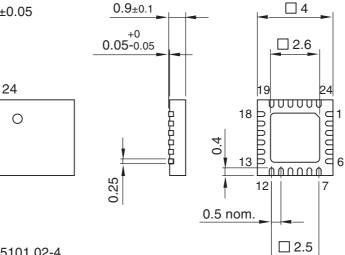
13. Package Information

Package: QFN 24 - 4 x 4 Exposed pad 2.6 x 2.6

(acc. JEDEC OUTLINE No. MO - 220)

Dimensions in mm

Not indicated tolerances ±0.05



technical drawings according to DIN specifications

Drawing-No.: 6.543-5101.02-4

1

6

Issue: 1; 03.06.05

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