



AX1017

1.8A high Efficiency Step-down Converter

Features

- Up to 90% Conversion Efficiency
- Typical Quiescent Current: 40 μ A
- Load Current: up to 1.8A
- Operating Input Voltage Range: 2.5 V to 5.5 V
- Switching Frequency up to 1MHz
- Adjustable and Fixed Output Voltage
- 100% Duty Cycle for Lowest Dropout
- Internal Soft Start
- Thermal Shutdown
- Short-Circuit Protection
- SOT-23-6L Package

Application

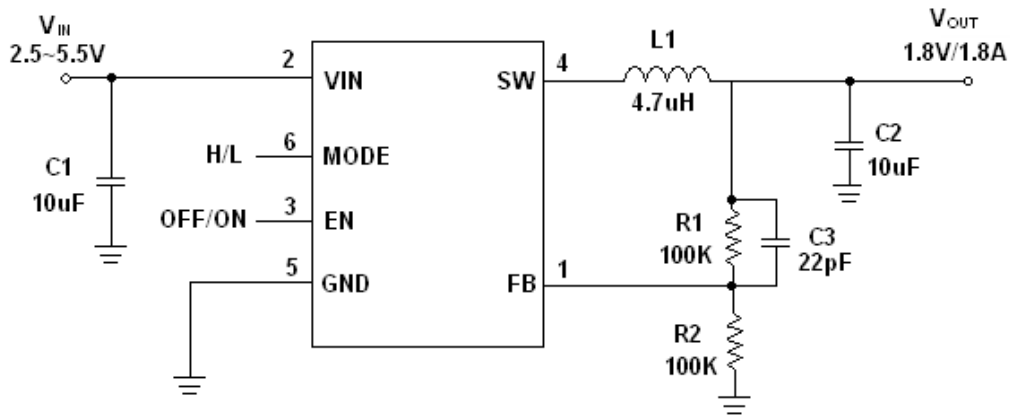
- CPUs and DSPS
- PC Cards and Notebooks
- Standard 5V to 3.3V Conversion
- PDA, Pocket PC and Smart Phones
- USB Powered Modems

Description

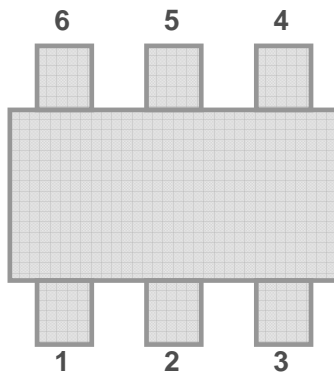
The AX1017 family of devices is high efficiency synchronous step-down dc-dc converters optimized for battery powered portable applications. The devices are ideal for portable applications powered by a single Li-Ion battery cell or by 3-cell NIMH/NICD batteries. With an output voltage range from 5.0 V down to 0.7 V, the devices support low voltage DSPS and processors in PDAS, pocket PCs, as well as notebooks computers. The AX1017 operates at a fixed switching frequency of 1MAz. The AX1017 supports up to 1.8A load current.

Typical Applications

ADJUSTABLE OUTPUT VOLTAGE



Pin Assignment



SOT-23-6L

PIN NUMBER SOT-23-6	PIN NAME	FUNCTION
1	FB	Output Feedback
2	VIN	Input
3	EN	ON/OFF Control (High Enable)
4	SW	Switch Output
5	GND	Ground
6	MODE	H/L MODE

Absolute Maximum Ratings

- Power Dissipation.....Internally limited
- V_{IN} - 0.3 V ~ + 6 V
- $V_{ON/OFF}$ - 0.3 V ~ ($V_{IN} + 0.3$) V
- V_{SW} - 0.3 V ~ ($V_{IN} + 0.3$) V
- V_{FB}- 0.3 V ~ + 6 V
- I_{SW} 2A
- Operating Temperature Range- 40°C ~ + 85°C
- Lead Temperature (Soldering 10 sec.)+ 300°C
- Storage Temperature Range- 65°C ~ + 150°C
- Junction Temperature+ 125°C

Electrical Characteristics

Operating Conditions: $T_A=25^\circ\text{C}$, $V_{IN}=3.6\text{V}$ unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
V_{IN}	Operating Voltage Range		2.5		5.5	V
V_{FB}	Regulated Voltage	$T_A = 25^\circ\text{C}$	0.565	0.58	0.598	V
I_{FB}	Feedback Current				± 30	nA
ΔV_{FB}	V_{REF}	$V_{IN}=2.5\text{V}\sim 5.5\text{V}$		0.03	0.4	%/V
F_{OSC}	Oscillator Frequency	$V_{FB} = 0.58\text{V}$ or $V_{OUT} = 100\%$	0.8	1	1.25	MHz
I_Q	Quiescent Current	$V_{FB} = 0.5\text{V}$ or $V_{OUT} = 90\%$, $I_{LOAD} = 0\text{A}$,Mode=0V		40		μA
I_S	Shutdown Current	$V_{EN} = 0\text{V}$, $V_{IN} = 4.2\text{V}$		0.1	1	μA
I_{PK}	Peak Inductor Current	$V_{IN} = 3.6\text{V}$, $V_{FB} = 0.5\text{V}$ or $V_{OUT} = 90\%$,		1.85		A
R_{PFET}	$R_{DS(ON)}$ of P-Channel FET	$I_{SW} = 500\text{mA}$		0.1		Ω
R_{NFET}	$R_{DS(ON)}$ of N-Channel FET	$I_{SW} = -500\text{mA}$		0.3		Ω
EFFI*	Efficiency	$V_{IN}=EN=3.6\text{V}$, $I_{OUT}=100\text{mA}$		90		%
ΔV_{OUT}	V_{OUT} Line Regulation	$V_{IN}=2.5\text{V}\sim 5.5\text{V}$		0.04	0.3	%/V
$V_{LOADREG}$	V_{OUT} Load Regulation			0.33		%

*EFFI = [(Output Voltage x Output Current) / (Input Voltage x Input Current)] x 100%

Typical Performance Characteristics

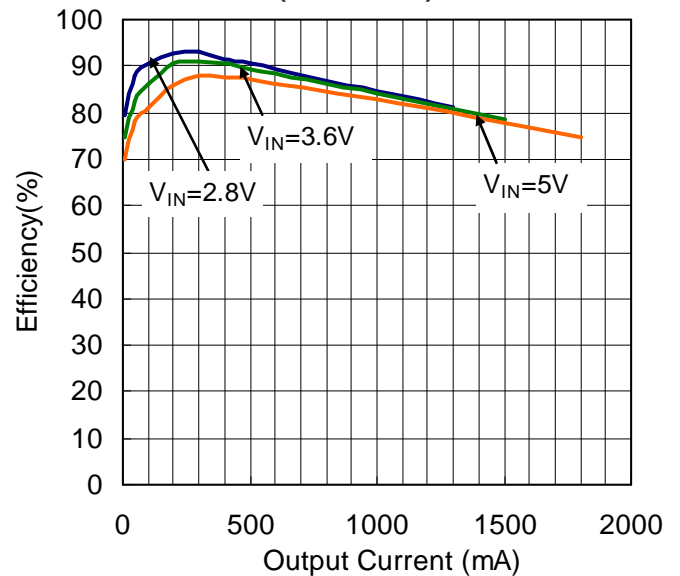
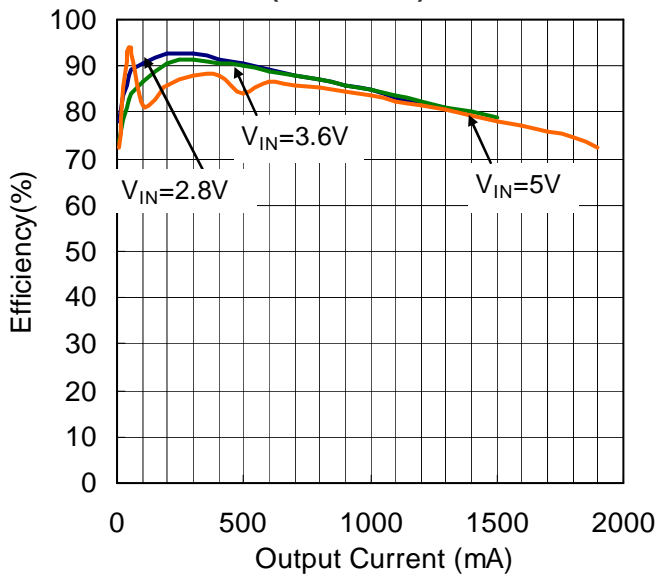
TA=25°C, C_{IN} =10 μF, C_{OUT} =10 μF, L=4.7 μH, R1=200KΩ, R2=100KΩ, unless otherwise noted.

MODE Connect GND

MODE Connect VIN

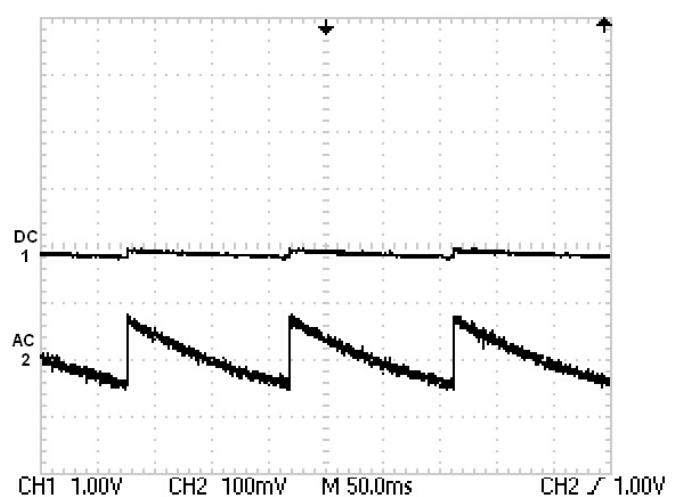
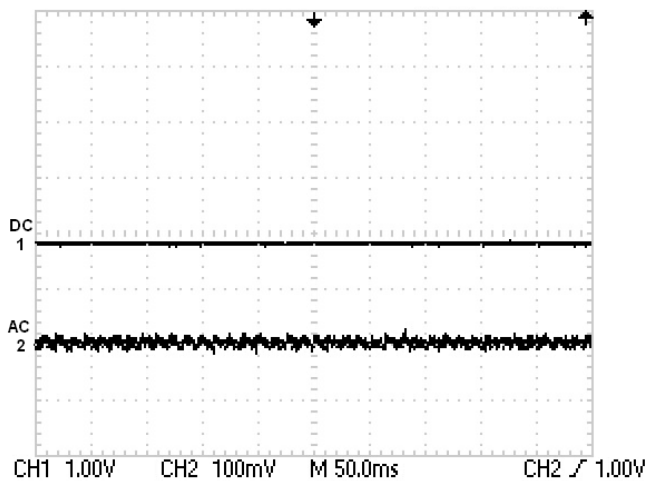
**Efficiency vs. Output Current
(V_{out}=1.8V)**

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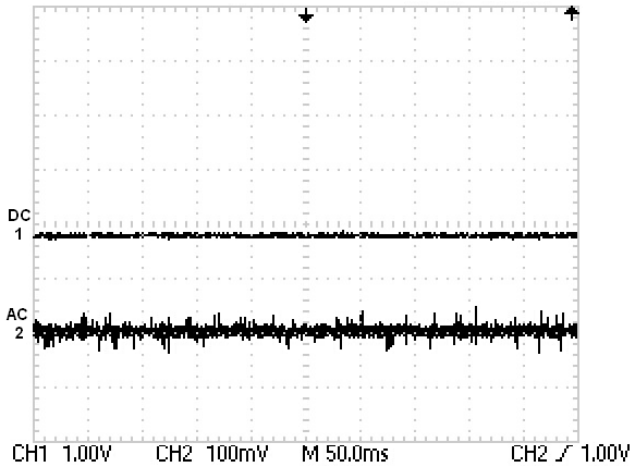
**Output Noise
(V_{IN}=5V, I_{LOAD}=0, MODE Connect VIN)**

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(V_{IN}=5V, I_{LOAD}=0, MODE Connect GND)**



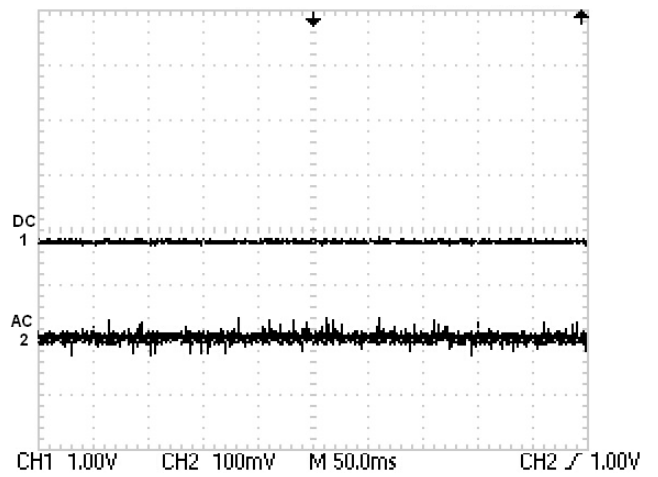
Output Noise

($V_{IN}=5V$, $I_{LOAD}=1A$, MODE Connect VIN)



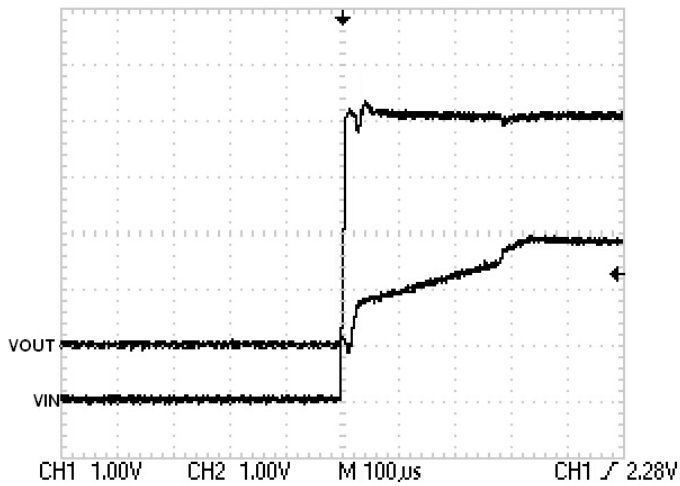
Output Noise

($V_{IN}=5V$, $I_{LOAD}=1A$, MODE Connect GND)



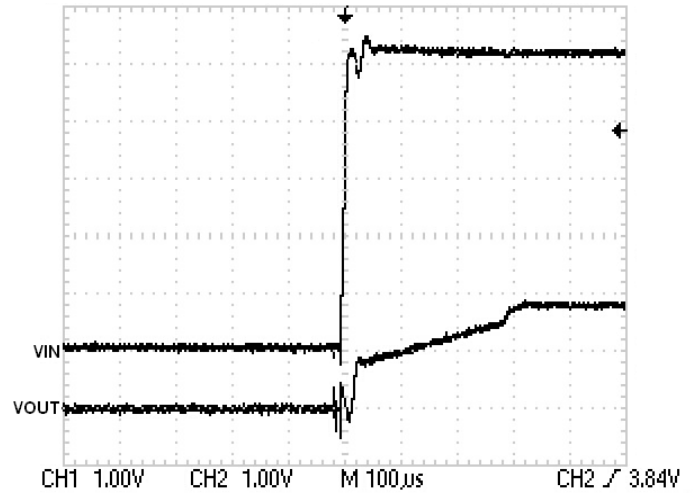
Start-up from power on

($V_{IN}=5V$, $I_{LOAD}=0A$, MODE Connect GND)



Start-up from power on

($V_{IN}=5V$, $I_{LOAD}=0A$, MODE Connect VIN)



Application Information

PIN ASSIGNMENT

FB (Pin 1): Feedback Pin. Receives the feedback voltage from an external resistive divider across the output. In the adjustable version, the output voltage is set by a resistive divider according to the following formula.

$$V_{out} = 0.58 \text{ V} \times \left(1 + \frac{R1}{R2} \right)$$

VIN (Pin 2): Supply voltage input.

EN (Pin 3): En Control Input. Forcing this pin above 1.5V enables the part. Forcing this pin below 0.4V shuts down the device. In shutdown, all functions are disabled drawing <1uA supply current. Do not leave EN floating.

MODE(PIN 4): When MODE connects VIN, high power consumption; When MODE connects GND, low power consumption.

GND (Pin 5): Ground Pin.

SW (Pin 6): Switch Node Connection to inductor. This pin connects to the drains of the internal main and synchronous power MOSFET switches.

PCB LAYOUT GUIDELINES

When laying out the printed circuit board, the following checklist should be used to ensure proper operation of the HX1017. Check the following in your layout:

1. The power traces, consisting of the GND trace, the SW trace and the V_{IN} trace should be kept short, direct and wide. closely as possible? This capacitor provides the AC current to the internal power MOSFETs.
2. Does the V_{FB} pin connect directly to the feedback resistors? The resistive divider R1/R2 must be connected between the (+) plate of C_{OUT} and ground.
3. Does the (+) plate of C_{IN} connects to V_{IN} as
4. Keep the switching node, SW, away from the sensitive V_{FB} node.
5. Keep the (-) plates of C_{IN} and C_{OUT} as close as possible.

INDUCTOR SELECTION

For most applications, the value of the inductor will fall in the range of 1μH to 4.7μH. Its value is chosen based on the desired ripple current. Large value inductors lower ripple current and small value inductors result in higher ripple currents. Higher V_{IN} or V_{OUT} also increases the ripple current as shown in equation 1. A reasonable starting point for setting ripple current is $\Delta I_L = 240\text{mA}$ (40% of 600mA).

$$\Delta I_L = \frac{1}{(f)(L)} V_{OUT} \left(1 - \frac{V_{OUT}}{V_{IN}} \right)$$

The DC current rating of the inductor should be at least equal to the maximum load current plus half the ripple current to prevent core saturation. Thus, a 1.8A rated inductor should be enough for most applications (1.5A + 300mA). For better efficiency, choose a low DC-resistance inductor.

Different core materials and shapes will change the size/current and price/current relationship of an inductor. Toroid or shielded pot cores in ferrite or perm alloy materials are small and don't radiate much energy, but generally cost more than powdered iron core inductors with similar electrical characteristics. The choice of which style inductor to use often depends more on the price vs. size requirements and any radiated field/EMI requirements than on what the AX1017 requires to operate. Table 1 shows some typical surface mount inductors that work well in AX1017 applications.

Table1. Recommended Inductors

Part	L (μH)	Max DCR (mΩ)	Max DC Current (A)	Size W x L x H (mm ³)	Vendor
CDRH3D16	2.2	75	1.20	3.8 x 3.8 x 1.8	Sumida
CDH3B16	2.2	70	1.20	4.0 x 4.0 x 1.8	Ceaiya

OUTPUT AND INPUT CAPACITOR SELECTION

In continuous mode, the source current of the top MOSFET is a square wave of duty cycle V_{OUT}/V_{IN} . To prevent large voltage transients, a low ESR input capacitor sized for the maximum RMS current must be used. The maximum RMS capacitor current is given by:

$$C_{IN} \text{ required } I_{RMS} \cong I_{OMAX} \frac{[V_{OUT}(V_{IN} - V_{OUT})]^{1/2}}{V_{IN}}$$

This formula has a maximum at $V_{IN} = 2V_{OUT}$, where $I_{RMS} = I_{OUT}/2$. This simple worst-case condition is commonly used for design because even significant deviations do not offer much relief. Note that the capacitor manufacturer's ripple current ratings are often based on 2000 hours of life. This makes it advisable to further derate the capacitor, or choose a capacitor rated at a higher temperature than required. Always consult the manufacturer if there is any question.

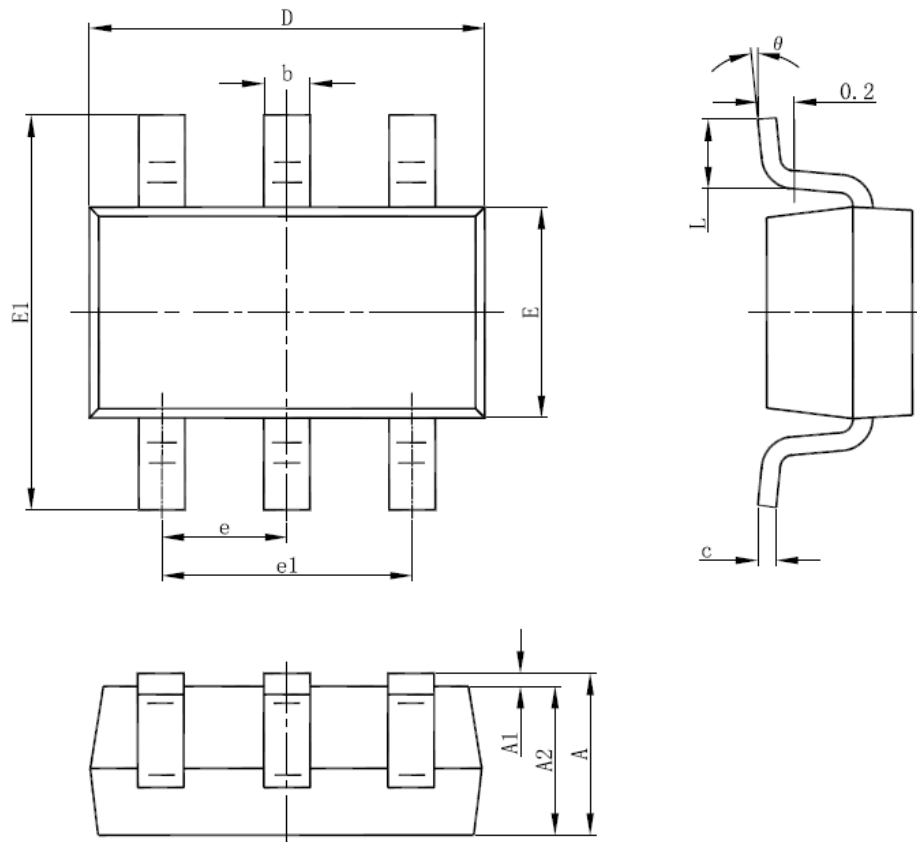
The selection of C_{OUT} is driven by the required effective series resistance (ESR).

Typically, once the ESR requirement for C_{OUT} has been met, the RMS current rating generally far exceeds the $I_{RIPPLE(P-P)}$ requirement. The output ripple ΔV_{OUT} is determined by:

$$\Delta V_{OUT} \cong \Delta I_L \left(ESR + \frac{1}{8fC_{OUT}} \right)$$

Where f = operating frequency, C_{OUT} = output capacitance and ΔI_L = ripple current in the inductor. For a fixed output voltage, the output ripple is highest at maximum input voltage since ΔI_L increases with input voltage.

Aluminum electrolytic and dry tantalum capacitors are both available in surface mount configurations. In the case of tantalum, it is critical that the capacitors are surge tested for use in switching power supplies. An excellent choice is the AVX TPS series of surface mount tantalum. These are specially constructed and tested for low ESR so they give the lowest ESR for a given volume. Other capacitor types include Sanyo POSCAP, Kemet T510 and T495 series, and Sprague 593D and 595D series. Consult the manufacturer for other specific recommendations.

Packaging Information
SOT-23-6L Package Outline Dimension


Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min	Max	Min	Max
A	1.050	1.250	0.041	0.049
A1	0.000	0.100	0.000	0.004
A2	1.050	1.150	0.041	0.045
b	0.300	0.500	0.012	0.020
c	0.100	0.200	0.004	0.008
D	2.820	3.020	0.111	0.119
E	1.500	1.700	0.059	0.067
E1	2.650	2.950	0.104	0.116
e	0.950(BSC)		0.037(BSC)	
e1	1.800	2.000	0.071	0.079
L	0.300	0.600	0.012	0.024
θ	0°	8°	0°	8°