



Smoke and CO/GAS Detector Flash MCU with LCD/LED Driver

BA45F6956

Revision: V1.12 Date: March 04, 2025

www.holtek.com

Features

CPU Features

- Operating voltage
 - ♦ $f_{SYS}=2\text{MHz}$: 2.2V~5.5V
 - ♦ $f_{SYS}=4\text{MHz}$: 2.2V~5.5V
 - ♦ $f_{SYS}=8\text{MHz}$: 2.2V~5.5V
- Up to 0.5 μs instruction cycle with 8MHz system clock at $V_{DD}=5\text{V}$
- Power down and wake-up functions to reduce power consumption
- Oscillator types
 - ♦ Internal High Speed 2/4/8MHz RC – HIRC
 - ♦ Internal Low Speed 32kHz RC – LIRC
- Fully integrated internal oscillators require no external components
- Multi-mode operation: FAST, SLOW, IDLE and SLEEP
- All instructions executed in 1~3 instruction cycles
- Table read instructions
- 115 powerful instructions
- 8-level subroutine nesting
- Bit manipulation instruction

Peripheral Features

- Flash Program Memory: 8K $\times$ 16
- RAM Data Memory: 1024 $\times$ 8
- True EEPROM Memory: 256 $\times$ 8
- In Application Programming – IAP
- Watchdog Timer function
- Up to 29 bidirectional I/O lines
- 4-level programmable I/O port source current for LED applications
- Two external interrupt lines shared with I/O pins
- Internal LDO provides fixed 2.2V/2.5V/3.0V output
- 7 external channel 12-bit resolution A/D converter with internal reference voltage V_{BREF}
- Temperature Sensor with internal reference voltage
- Multiple Timer Modules for time measurement, compare match output, PWM output or single pulse output function
- Dual Time Base functions for generation of fixed time interrupt signals
- Universal Serial Interface Module – USIM for SPI, I²C or UART communication
- Fully-duplex or Half-duplex Universal Asynchronous Receiver and Transmitter Interface – UART
- CO/Gas detector AFE including an operational amplifier
- Smoke detector
 - ♦ Operating voltage: 2.2V~5.5V
 - ♦ Standby current: 1.5 μA @ 3V
 - ♦ Dual-channel sink current generator for constant current output
 - ♦ Smoke detector AFE including two operational amplifiers
 - ♦ One-wire communication interface

- 16-bit Voice D/A converter
- LCD Driver function
 - ♦ SEGs × COMs: 13×4
 - ♦ Duty type: 1/4 duty
 - ♦ Bias level: 1/3 bias
 - ♦ Bias type: R type or C type
 - ♦ Waveform type: type A or type B
- Low voltage reset function
- Low voltage detect function
- Package type: 48-pin LQFP

Development Tools

For rapid product development and to simplify device parameter setting, Holtek has provided relevant development tools which users can download from the following link:

https://www.holtek.com/page/tool-detail/dev_plat/voice/voice_workshop

General Description

The BA45F6956 is a Flash Memory A/D type 8-bit high performance RISC architecture microcontroller with a fully integrated LCD driver, especially designed for smoke and CO/Gas detector applications which require an LCD/LED display function.

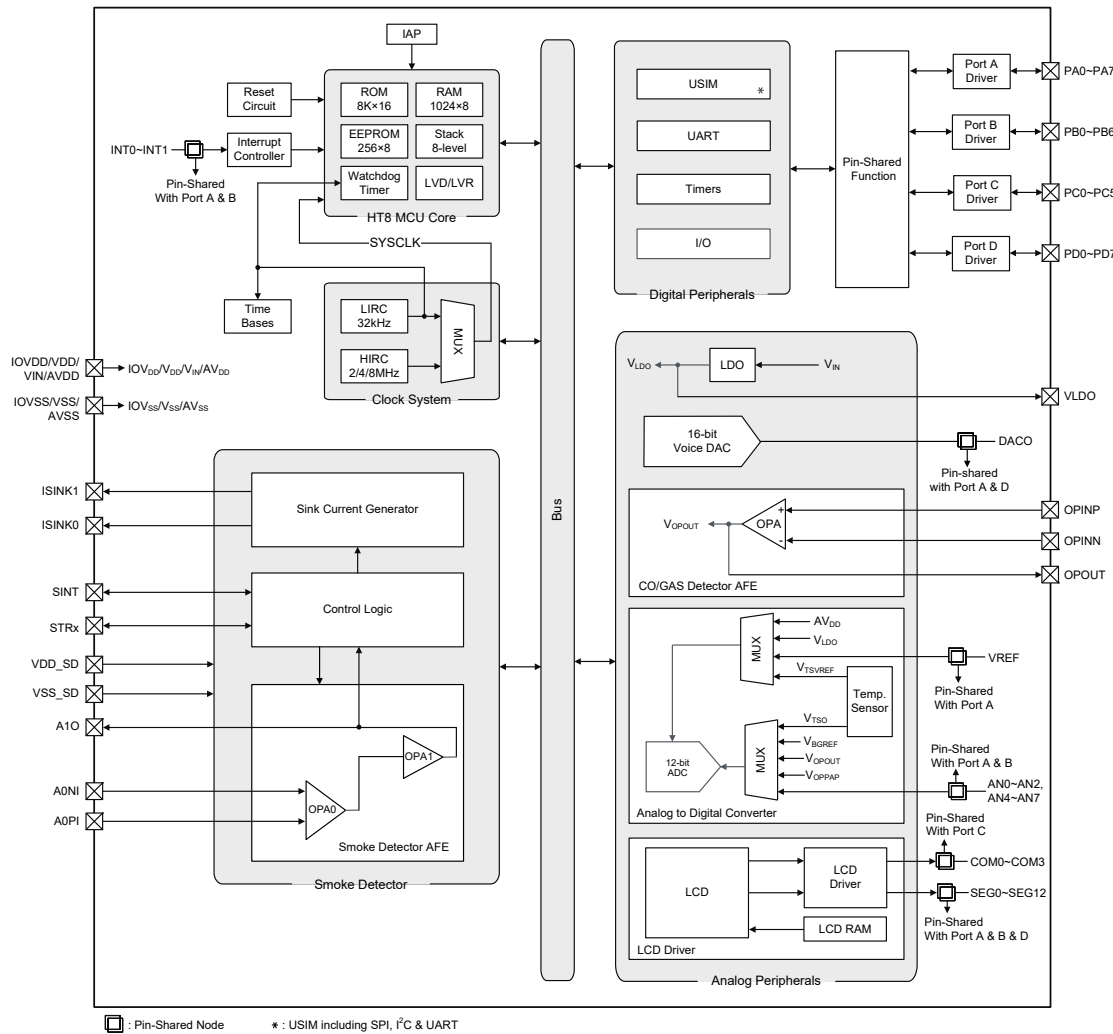
For memory features, the Flash Memory offers users the convenience of multi-programming features. Other memory includes an area of RAM Data Memory as well as an area of True EEPROM memory for storage of non-volatile data such as serial numbers, calibration data etc. By using the In Application Programming technology, user have a convenient means to directly store their measured data in the Flash Program Memory as well as having the ability to easily update their application programs.

Analog features include a multi-channel Analog to Digital converter, an integrated temperature sensor circuitry, a smoke detector AFE, a sink current generator, a CO/GAS detector AFE and a D/A converter. Multiple and extremely flexible Timer Modules provide timing, pulse generation and PWM generation functions. Communication with the outside world is catered for by including fully integrated SPI, I²C and UART interface functions, three popular interfaces which provide designers with a means of easy communication with external peripheral hardware. In addition, an internal LDO function provides various fixed voltage options to the internal modules and external devices. Protective features such as an internal Watchdog Timer, Low Voltage Reset and Low Voltage Detector coupled with excellent noise immunity and ESD protection ensure that reliable operation is maintained in hostile electrical environments.

The device also includes fully integrated high and low speed oscillators which require no external components for their implementation. The ability to operate and switch dynamically between a range of operating modes using different clock sources gives users the ability to optimize microcontroller operation and minimize power consumption.

The inclusion of flexible I/O programming features, Time Base functions and 16-bit Voice D/A converter along with many other features ensure that the device will find excellent use in combined smoke detection products, such as LCD smoke and CO/Gas detector alarms and voice LCD smoke and CO/Gas detector alarms, etc.

Block Diagram



Pin configuration diagram for the BA45F6956/BA45V6956 48 LQFP-A package. The package is shown with pins numbered 1 to 48. The pin functions are listed as follows:

- Pin 1:** STRx
- Pin 2:** A1O
- Pin 3:** SINT
- Pin 4:** VSS_SD
- Pin 5:** PA0/UTX/SDO/ICPDA/OCDSDA
- Pin 6:** VLDO
- Pin 7:** IOVDD/VDD/VIN/AVDD
- Pin 8:** PA1/INT0/SCK/SCL/PTPB/AN0
- Pin 9:** PA3/STCK/SCS/PTP/AN1
- Pin 10:** PA4/INT1/UTX/SDO/VREF/DACO
- Pin 11:** IOVSS/VSS/AVSS
- Pin 12:** PB6/PTP/AN2/VDDIO
- Pin 13:** OPINP
- Pin 14:** OPINN
- Pin 15:** OPOUT
- Pin 16:** PD0/PTP/DACO
- Pin 17:** PD1/STP/SCS
- Pin 18:** PD2/STPB/SCK/SCL
- Pin 19:** PD3/PTP/PTPB/SDI/SDA/URX/UTX
- Pin 20:** PC4/UTX/SDO
- Pin 21:** PA5/SEG0/RX/TX/AN4
- Pin 22:** PA6/PTCK/SEG1/TX/AN5
- Pin 23:** PA7/INT1/SEG3/RX/TX/AN7
- Pin 24:** PB0/INT0/SEG3/RX/TX/AN7
- Pin 25:** PC3/COM3
- Pin 26:** PC2/COM2
- Pin 27:** PC1/COM1
- Pin 28:** PC0/COM0
- Pin 29:** PD5/SEG10/C2
- Pin 30:** PD4/SEG9/C1
- Pin 31:** PD6/SEG11/V2
- Pin 32:** V1
- Pin 33:** VMAX
- Pin 34:** PLCD
- Pin 35:** PD7/SEG12
- Pin 36:** PB1/SEG4
- Pin 37:** PB2/SEG5
- Pin 38:** PB3/SEG6
- Pin 39:** PB4/SEG7
- Pin 40:** PB5/SEG8
- Pin 41:** PC5/SDI/SDA/URX/UTX/PTP
- Pin 42:** PA2/SDI/SDA/URX/UTX/ICPCK/OCDSCK
- Pin 43:** ISINK1
- Pin 44:** ISINK0
- Pin 45:** VDD_SD
- Pin 46:** VSS_SD
- Pin 47:** A0NI
- Pin 48:** A0PI

2. The OCSDSA and OCDSCK pins are supplied for the OCDS dedicated pins and as such only available for the BA45V6956 device which is the OCDS EV chip for the BA45F6956 device.

Pin Description

The function of each pin is listed in the following table, however the details behind how each pin is configured is contained in other sections of the datasheet.

Pin Name	Function	OPT	I/T	O/T	Description
PA0/UTX/SDO/ICPDA/ OCDSDA	PA0	PAPU PAWU PAS0	ST	CMOS	General purpose I/O. Register enabled pull-high and wake-up
	UTX	PAS0	—	CMOS	USIM UART serial data output
	SDO	PAS0	—	CMOS	SPI serial data output
	ICPDA	—	ST	CMOS	ICP Data/Address pin
	OCDSDA	—	ST	CMOS	OCDS Data/Address pin, for EV chip only
PA1/INT0/SCK/SCL/ PTPB/AN0	PA1	PAPU PAWU PAS0	ST	CMOS	General purpose I/O. Register enabled pull-high and wake-up
	INT0	PAS0 INTC0 INTEG IFS1	ST	—	External interrupt input
	SCK	PAS0 IFS0	ST	CMOS	SPI serial clock
	SCL	PAS0 IFS0	ST	NMOS	I ² C clock line
	PTPB	PAS0	—	CMOS	PTM inverted output
	AN0	PAS0	AN	—	A/D Converter external input channel
PA2/SDI/SDA/URX/ UTX/ICPCK/OCDSCK	PA2	PAPU PAWU PAS0	ST	CMOS	General purpose I/O. Register enabled pull-high and wake-up
	SDI	PAS0 IFS0	ST	—	SPI serial data input
	SDA	PAS0 IFS0	ST	NMOS	I ² C data line
	URX/UTX	PAS0 IFS0	ST	CMOS	USIM UART serial data input in full-duplex communication or USIM UART serial data input/output in Single Wire Mode communication
	ICPCK	—	ST	—	ICP clock pin
	OCDSCK	—	ST	—	OCDS clock pin, for EV chip only
PA3/STCK/ $\overline{\text{SCS}}$ /PTP/ AN1	PA3	PAPU PAWU PAS0	ST	CMOS	General purpose I/O. Register enabled pull-high and wake-up
	STCK	PAS0 IFS1	ST	—	STM clock input
	$\overline{\text{SCS}}$	PAS0 IFS0	ST	CMOS	SPI slave select
	PTP	PAS0	—	CMOS	PTM output
	AN1	PAS0	AN	—	A/D Converter external input channel

Pin Name	Function	OPT	I/T	O/T	Description
PA4/INT1/UTX/SDO/ VREF/DACO	PA4	PAPU PAWU PAS1	ST	CMOS	General purpose I/O. Register enabled pull-high and wake-up
	INT1	PAS1 INTC0 INTEG IFS1	ST	—	External interrupt input
	UTX	PAS1	—	CMOS	USIM UART serial data output
	SDO	PAS1	—	CMOS	SPI serial data output
	VREF	PAS1	AN	—	A/D Converter external reference voltage input
	DACO	PAS1	—	AN	16-bit D/A Converter output
PA5/SEG0/RX/TX/AN4	PA5	PAPU PAWU PAS1	ST	CMOS	General purpose I/O. Register enabled pull-high and wake-up
	SEG0	PAS1	—	AN	LCD segment output
	RX/TX	PAS1 IFS1	ST	CMOS	UART serial data input in full-duplex communication or UART serial data input/output in Single Wire Mode communication
	AN4	PAS1	AN	—	A/D Converter external input channel
PA6/PTCK/SEG1/TX/ AN5	PA6	PAPU PAWU PAS1	ST	CMOS	General purpose I/O. Register enabled pull-high and wake-up
	PTCK	PAS1	ST	—	PTM clock input
	SEG1	PAS1	—	AN	LCD segment output
	TX	PAS1	—	CMOS	UART serial data output
	AN5	PAS1	AN	—	A/D Converter external input channel
PA7/INT1/SEG2/TX/ AN6	PA7	PAPU PAWU PAS1	ST	CMOS	General purpose I/O. Register enabled pull-high and wake-up
	INT1	PAS1 INTC0 INTEG IFS1	ST	—	External interrupt input
	SEG2	PAS1	—	AN	LCD segment output
	TX	PAS1	—	CMOS	UART serial data output
	AN6	PAS1	AN	—	A/D Converter external input channel
PB0/INT0/SEG3/RX/TX/ AN7	PB0	PBPU PBS0	ST	CMOS	General purpose I/O. Register enabled pull-high
	INT0	PBS0 INTC0 INTEG IFS1	ST	—	External interrupt input
	SEG3	PBS0	—	AN	LCD segment output
	RX/TX	PBS0 IFS1	ST	CMOS	UART serial data input in full-duplex communication or UART serial data input/output in Single Wire Mode communication
	AN7	PBS0	AN	—	A/D Converter external input channel
PB1/SEG4	PB1	PBPU PBS0	ST	CMOS	General purpose I/O. Register enabled pull-high
	SEG4	PBS0	—	AN	LCD segment output
PB2/SEG5	PB2	PBPU PBS0	ST	CMOS	General purpose I/O. Register enabled pull-high
	SEG5	PBS0	—	AN	LCD segment output

Pin Name	Function	OPT	I/T	O/T	Description
PB3/SEG6	PB3	PBPU PBS0	ST	CMOS	General purpose I/O. Register enabled pull-high
	SEG6	PBS0	—	AN	LCD segment output
PB4/SEG7	PB4	PBPU PBS1	ST	CMOS	General purpose I/O. Register enabled pull-high
	SEG7	PBS1	—	AN	LCD segment output
PB5/SEG8	PB5	PBPU PBS1	ST	CMOS	General purpose I/O. Register enabled pull-high
	SEG8	PBS1	—	AN	LCD segment output
PB6/PTP/AN2/VDDIO	PB6	PBPU PBS1	ST	CMOS	General purpose I/O. Register enabled pull-high
	PTP	PBS1	—	CMOS	PTM output
	AN2	PBS1	AN	—	A/D Converter external input channel
	VDDIO	PBS1 PMPS	PWR	—	Power for PA1, PA3, PA4
PC0/COM0	PC0	PCPU PCS0	ST	CMOS	General purpose I/O. Register enabled pull-high
	COM0	PCS0	—	AN	LCD common output
PC1/COM1	PC1	PCPU PCS0	ST	CMOS	General purpose I/O. Register enabled pull-high
	COM1	PCS0	—	AN	LCD common output
PC2/COM2	PC2	PCPU PCS0	ST	CMOS	General purpose I/O. Register enabled pull-high
	COM2	PCS0	—	AN	LCD common output
PC3/COM3	PC3	PCPU PCS0	ST	CMOS	General purpose I/O. Register enabled pull-high
	COM3	PCS0	—	AN	LCD common output
PC4/UTX/SDO	PC4	PCPU PCS1	ST	CMOS	General purpose I/O. Register enabled pull-high
	UTX	PCS1	—	CMOS	USIM UART serial data output
	SDO	PCS1	—	CMOS	SPI serial data output
PC5/SDI/SDA/URX/UTX/ PTP	PC5	PCPU PCS1	ST	CMOS	General purpose I/O. Register enabled pull-high
	SDI	PCS1 IFS0	ST	—	SPI serial data input
	SDA	PCS1 IFS0	ST	NMOS	I ² C data line
	URX/UTX	PCS1 IFS0	ST	CMOS	USIM UART serial data input in full-duplex communication or USIM UART serial data input/output in Single Wire Mode communication
	PTP	PCS1	—	CMOS	PTM output
PD0/PTP/DACO	PD0	PDPUPDS0	ST	CMOS	General purpose I/O. Register enabled pull-high
	PTP	PDS0	—	CMOS	PTM output
	DACO	PDS0	—	AN	16-bit D/A Converter output
PD1/STP/SCS	PD1	PDPUPDS0	ST	CMOS	General purpose I/O. Register enabled pull-high
	STP	PDS0	—	CMOS	STM output
	SCS	PDS0 IFS0	ST	CMOS	SPI slave select

Pin Name	Function	OPT	I/T	O/T	Description
PD2/STPB/SCK/SCL	PD2	PDP PDS0	ST	CMOS	General purpose I/O. Register enabled pull-high
	STPB	PDS0	—	CMOS	STM inverted output
	SCK	PDS0 IFS0	ST	CMOS	SPI serial clock
	SCL	PDS0 IFS0	ST	NMOS	I ² C clock line
PD3/PTP/PTPB/SDI/ SDA/URX/UTX	PD3	PDP PDS0	ST	CMOS	General purpose I/O. Register enabled pull-high
	PTP	PDS0	—	CMOS	PTM output
	PTPB	PDS0	—	CMOS	PTM inverted output
	SDI	PDS0 IFS0	ST	—	SPI serial data input
	SDA	PDS0 IFS0	ST	NMOS	I ² C data line
	URX/UTX	PDS0 IFS0	ST	CMOS	USIM UART serial data input in full-duplex communication or USIM UART serial data input/output in Single Wire Mode communication
PD4/SEG9/C1	PD4	PDP PDS1	ST	CMOS	General purpose I/O. Register enabled pull-high
	SEG9	PDS1	—	AN	LCD segment output
	C1	PDS1	AN	AN	LCD voltage pump
PD5/SEG10/C2	PD5	PDP PDS1	ST	CMOS	General purpose I/O. Register enabled pull-high
	SEG10	PDS1	—	AN	LCD segment output
	C2	PDS1	AN	AN	LCD voltage pump
PD6/SEG11/V2	PD6	PDP PDS1	ST	CMOS	General purpose I/O. Register enabled pull-high
	SEG11	PDS1	—	AN	LCD segment output
	V2	PDS1	PWR	AN	LCD voltage pump
PD7/SEG12	PD7	PDP PDS1	ST	CMOS	General purpose I/O. Register enabled pull-high
	SEG12	PDS1	—	AN	LCD segment output
VLDO	VLDO	—	—	PWR	LDO output
V1	V1	—	PWR	AN	LCD voltage pump
VMAX	VMAX	—	PWR	—	LCD maximum voltage, connected to VDD or V1
PLCD	PLCD	—	PWR	AN	LCD power supply
OPINP	OPINP	—	AN	—	OPAMP external positive input pin
OPINN	OPINN	—	AN	—	OPAMP external negative input pin
OPOUT	OPOUT	—	—	AN	OPAMP output
ISINK0	ISINK0	—	—	AN	Sink current 0 source
ISINK1	ISINK1	—	—	AN	Sink current 1 source
A0NI	A0NI	—	AN	—	OPA0 inverting input
A0PI	A0PI	—	AN	—	OPA0 non-inverting input
A1O	A1O	—	—	AN	OPA1 output
SINT	SINT	—	ST	CMOS	Smoke detector wake-up pin
STRx	STRx	—	ST	CMOS	Smoke detector one-wire communication pin
VDD_SD	VDD_SD	—	PWR	—	Smoke detector positive power supply
VSS_SD	VSS_SD	—	PWR	—	Smoke detector negative power supply

Pin Name	Function	OPT	I/T	O/T	Description
IOVDD/VDD/VIN/AVDD	IOVDD	—	PWR	—	I/O pad positive power supply
	VDD	—	PWR	—	Digital positive power supply
	VIN	—	PWR	—	LDO positive power supply
	AVDD	—	PWR	—	Analog positive power supply
IOVSS/VSS/AVSS	IOVSS	—	PWR	—	I/O pad ground
	VSS	—	PWR	—	Digital negative power supply, ground
	AVSS	—	PWR	—	Analog negative power supply

Legend: I/T: Input type;

OPT: Optional by register option;

ST: Schmitt Trigger input;

NMOS: NMOS output;

O/T: Output type;

PWR: Power;

AN: Analog signal;

CMOS: CMOS output.

Absolute Maximum Ratings

Supply Voltage	V _{SS} - 0.3V to 6.0V
Input Voltage	V _{SS} - 0.3V to V _{DD} +0.3V
Storage Temperature.....	-60°C to 150°C
Operating Temperature.....	-40°C to 85°C
I _{OH} Total	-80mA
I _{OL} Total	80mA
Total Power Dissipation	500mW

Note: These are stress ratings only. Stresses exceeding the range specified under “Absolute Maximum Ratings” may cause substantial damage to the device. Functional operation of the device at other conditions beyond those listed in the specification is not implied and prolonged exposure to extreme conditions may affect device reliability.

D.C. Characteristics

For data in the following tables, note that factors such as oscillator type, operating voltage, operating frequency, pin load conditions, temperature and program instruction type, etc., can all exert an influence on the measured values.

Operating Voltage Characteristics

T_a=-40°C~85°C

Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
V _{DD}	Operating Voltage – HIRC	f _{SYS} =f _{HIRC} =2MHz	2.2	—	5.5	V
		f _{SYS} =f _{HIRC} =4MHz	2.2	—	5.5	
		f _{SYS} =f _{HIRC} =8MHz	2.2	—	5.5	
	Operating Voltage – LIRC	f _{SYS} =f _{LIRC} =32kHz	2.2	—	5.5	V

Operating Current Characteristics

Ta=-40°C~85°C

Symbol	Operation Mode	Test Conditions		Min.	Typ.	Max.	Unit
		V _{DD}	Conditions				
I _{DD}	SLOW Mode – LIRC	2.2V	f _{sys} =32kHz	—	8	16	μA
		3V		—	10	20	
		5V		—	30	50	
	FAST Mode – HIRC	2.2V	f _{sys} =2MHz	—	0.15	0.20	mA
		3V		—	0.2	0.3	
		5V		—	0.4	0.6	
		2.2V	f _{sys} =4MHz	—	0.3	0.5	mA
		3V		—	0.4	0.6	
		5V		—	0.8	1.2	
		2.2V	f _{sys} =8MHz	—	0.6	1.0	mA
		3V		—	0.8	1.2	
		5V		—	1.6	2.4	

Note: When using the characteristic table data, the following notes should be taken into consideration:

1. Any digital input is setup in a non-floating condition.
2. All measurements are taken under conditions of no load and with all peripherals in an off state.
3. There are no DC current paths.
4. All Operating Current values are measured using a continuous NOP instruction program loop.

Standby Current Characteristics

Ta=-40°C~85°C, unless otherwise specified

Symbol	Operation Mode	Test Conditions		Typ.	Max.	Max. @85°C	Unit
		V _{DD}	Conditions				
I _{STB}	SLEEP Mode	2.2V	WDT on	1.2	2.4	2.9	μA
		3V		1.5	3.0	3.6	
		5V		3	5	6	
	IDLE0 Mode – LIRC	2.2V	f _{SUB} on	2.4	4.0	4.8	μA
		3V		3	5	6	
		5V		5	10	12	
	IDLE1 Mode – HIRC	2.2V	f _{SUB} on, f _{sys} =2MHz	60	120	140	μA
		3V		70	140	160	
		5V		130	260	280	
		2.2V	f _{SUB} on, f _{sys} =4MHz	144	200	240	μA
		3V		180	250	300	
		5V		400	600	720	
		2.2V	f _{SUB} on, f _{sys} =8MHz	288	400	480	μA
		3V		360	500	600	
		5V		600	800	960	

Note: When using the characteristic table data, the following notes should be taken into consideration:

1. Any digital input is setup in a non-floating condition.
2. All measurements are taken under conditions of no load and with all peripherals in an off state.
3. There are no DC current paths.
4. All Standby Current values are taken after a HALT instruction executed thus stopping all instruction execution.

A.C. Characteristics

For data in the following tables, note that factors such as oscillator type, operating voltage, operating frequency and temperature etc., can all exert an influence on the measured values.

High Speed Internal Oscillator – HIRC – Frequency Accuracy

During the program writing operation the writer will trim the HIRC oscillator at a user selected HIRC frequency and user selected voltage of either 3V or 5V.

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V _{DD}	Temp.				
f _{HIRC}	2MHz Writer Trimmed HIRC Frequency	3V/5V	25°C	-1%	2	+1%	MHz
			-20°C~60°C	-2%	2	+2%	
			-40°C~85°C	-3%	2	+3%	
		2.2V~5.5V	25°C	-6%	2	+9%	
			-40°C~85°C	-6%	2	+10%	
	4MHz Writer Trimmed HIRC Frequency	3V/5V	25°C	-1%	4	+1%	MHz
			-40°C~85°C	-2.5%	4	+2.5%	
		2.2V~5.5V	25°C	-2.5%	4	+2.5%	
			-40°C~85°C	-3%	4	+3%	
	8MHz Writer Trimmed HIRC Frequency	3V/5V	25°C	-1%	8	+1%	MHz
			-40°C~85°C	-10%	8	+2%	
		2.2V~5.5V	25°C	-10%	8	+3%	
			-40°C~85°C	-15%	8	+5%	

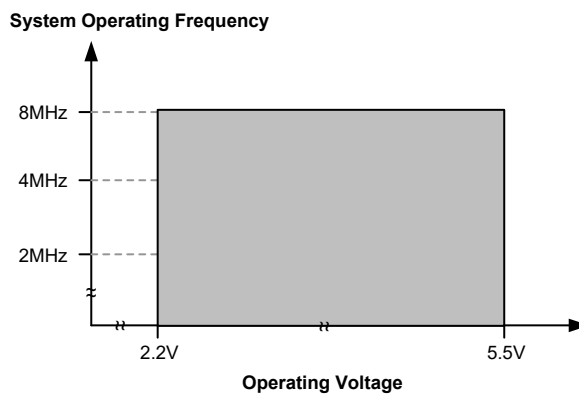
Note: 1. The 3V/5V values for V_{DD} are provided as these are the two selectable fixed voltages at which the HIRC frequency is trimmed by the writer.

2. The row below the 3V/5V trim voltage row is provided to show the values for the full V_{DD} range operating voltage. It is recommended that the trim voltage is fixed at 3V for application voltage ranges from 2.2V to 3.6V and fixed at 5V for application voltage ranges from 3.3V to 5.5V.
3. The minimum and maximum tolerance values provided in the table are only for the frequency at which the writer trims the HIRC oscillator. After trimming at this chosen specific frequency any change in HIRC oscillator frequency using the oscillator register control bits by the application program will give a frequency tolerance to within ±20%.

Low Speed Internal Oscillator Characteristics – LIRC

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V _{DD}	Temp.				
f _{LIRC}	LIRC Frequency	2.2V~5.5V	-40°C~85°C	-7%	32	+7%	kHz
t _{START}	LIRC Start-up Time	—	-40°C~85°C	—	—	100	μs

Operating Frequency Characteristic Curves



System Start Up Time Characteristics

Ta=-40°C~85°C

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V _{DD}	Conditions				
t _{SST}	System Start-up Time	—	f _{SYS} =f _H ~f _H /64, f _H =f _{HIRC}	—	16	—	t _{HIRC}
	Wake-up from conditions where f _{SYS} is off	—	f _{SYS} =f _{SUB} =f _{LIRC}	—	2	—	t _{LIRC}
	System Start-up Time	—	f _{SYS} =f _H ~f _H /64, f _H =f _{HIRC}	—	2	—	t _H
	Wake-up from conditions where f _{SYS} is on	—	f _{SYS} =f _{SUB} =f _{LIRC}	—	2	—	t _{SUB}
	System Speed Switch Time FAST to SLOW Mode or SLOW to FAST Mode	—	f _{HIRC} switches from off → on	—	16	—	t _{HIRC}
t _{RSTD}	System Reset Delay Time Reset source from Power-on Reset or LVR Hardware Reset	—	RR _{POR} =5V/ms	14	16	18	ms
	System Reset Delay Time WDTC Software Reset	—	—				
	System Reset Delay Time Reset source from WDT Overflow	—	—	14	16	18	ms
t _{SRESET}	Minimum Software Reset Width to Reset	—	—	45	90	120	μs

- Note: 1. For the System Start-up time values, whether f_{SYS} is on or off depends upon the mode type and the chosen f_{SYS} system oscillator. Details are provided in the System Operating Modes section.
2. The time units, shown by the symbol t_{HIRC} etc. are the inverse of the corresponding frequency values as provided in the frequency tables. For example, t_{HIRC}=1/f_{HIRC}, t_{SYS}=1/f_{SYS} etc.
3. If the LIRC is used as the system clock and if it is off when in the SLEEP Mode, then an additional LIRC start up time, t_{START}, as provided in the LIRC frequency table, must be added to the t_{SST} time in the table above.
4. The System Speed Switch Time is effectively the time taken for the newly activated oscillator to start up.

Input/Output Characteristics

Ta=-40°C~85°C

Input/Output (without Multi-power) D.C. Characteristics

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V _{DD}	Conditions				
V _{IL}	Input Low Voltage for I/O Ports (Except PA1, PA3 and PA4 Pins)	5V	—	0	—	1.5	V
		—		0	—	0.2V _{DD}	
V _{IH}	Input High Voltage for I/O Ports (Except PA1, PA3 and PA4 Pins)	5V	—	3.5	—	5.0	V
		—		0.8V _{DD}	—	V _{DD}	
I _{OL}	Sink Current for I/O Ports (Except PA1, PA3 and PA4 Pins)	3V	V _{OL} =0.1V _{DD}	16	32	—	mA
		5V		32	65	—	
I _{OH}	Source Current for I/O Ports (Except PA1, PA3 and PA4 Pins)	3V	V _{OH} =0.9V _{DD} , SLEDCn[m+1:m]=00 (n=0, 1, m=0, 2, 4, 6)	-0.7	-1.5	—	mA
		5V		-1.5	-2.9	—	
		3V	V _{OH} =0.9V _{DD} , SLEDCn[m+1:m]=01 (n=0, 1, m=0, 2, 4, 6)	-1.3	-2.5	—	
		5V		-2.5	-5.1	—	
		3V	V _{OH} =0.9V _{DD} , SLEDCn[m+1:m]=10 (n=0, 1, m=0, 2, 4, 6)	-1.8	-3.6	—	
		5V		-3.6	-7.3	—	
		3V	V _{OH} =0.9V _{DD} , SLEDCn[m+1:m]=11 (n=0, 1, m=0, 2, 4, 6)	-4	-8	—	
		5V		-8	-16	—	
R _{PH}	Pull-high Resistance for I/O Ports ^(Note) (Except PA1, PA3 and PA4 Pins)	3V	—	20	60	100	kΩ
		5V		10	30	50	
I _{LEAK}	Input Leakage Current (Except PA1, PA3 and PA4 Pins)	5V	V _{IN} =V _{DD} or V _{IN} =V _{SS}	—	—	±1	μA
t _{TCK}	TM Clock Input Pin Minimum Pulse Width	—	—	0.3	—	—	μs
t _{INT}	Interrupt Pin Minimum Pulse Width	—	—	10	—	—	μs
f _{TMCLK}	PTM Maximum Timer Clock Source Frequency	5V	—	—	—	1	f _{sys}

Note: The R_{PH} internal pull-high resistance value is calculated by connecting to ground and enabling the input pin with a pull-high resistor and then measuring the pin current at the specified supply voltage level. Dividing the voltage by this measured current provides the R_{PH} value.

Input/Output (with Multi-power) D.C. Characteristics

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V _{DD}	Conditions				
V _{DD}	V _{DD} Power Supply for PA1, PA3 and PA4 Pins	—	—	2.2	5.0	5.5	V
V _{DDIO}	V _{DDIO} Power Supply for PA1, PA3 and PA4 Pins	—	—	1.8	—	V _{DD}	V
V _{IL}	Input Low Voltage for PA1, PA3 and PA4 Pins	5V	Pin power=V _{DD} or V _{DDIO} V _{DDIO} =V _{DD}	0	—	1.5	V
		—	Pin power=V _{DD} or V _{DDIO}	0	—	0.2 (V _{DD} / V _{DDIO})	
V _{IH}	Input High Voltage for PA1, PA3 and PA4 Pins	5V	Pin power=V _{DD} or V _{DDIO} V _{DDIO} =V _{DD}	3.5	—	5.0	V
		—	Pin power=V _{DD} or V _{DDIO}	0.8 (V _{DD} / V _{DDIO})	—	V _{DD} / V _{DDIO}	
I _{OL}	Sink Current for PA1, PA3 and PA4 Pins	3V	V _{OL} =0.1(V _{DD} /V _{DDIO}), V _{DDIO} =V _{DD}	16	32	—	mA
		5V		32	65	—	mA
		5V	V _{OL} =0.1V _{DDIO} , V _{DDIO} =3V	20	40	—	mA
I _{OH}	Source Current for PA1, PA3 and PA4 Pins	3V	V _{OH} =0.9 (V _{DD} /V _{DDIO}), V _{DDIO} =V _{DD} SLEDCn[m+1:m]=00 (n=0, m=0 or 2)	-0.7	-1.5	—	mA
		5V		-1.5	-2.9	—	mA
		5V	V _{OH} =0.9 (V _{DD} /V _{DDIO}), V _{DDIO} =3V SLEDCn[m+1:m]=00 (n=0, m=0 or 2)	-0.40	-0.85	—	mA
		3V	V _{OH} =0.9 (V _{DD} /V _{DDIO}), V _{DDIO} =V _{DD} SLEDCn[m+1:m]=01 (n=0, m=0 or 2)	-1.3	-2.5	—	mA
		5V		-2.5	-5.1	—	mA
		5V	V _{OH} =0.9 (V _{DD} /V _{DDIO}), V _{DDIO} =3V SLEDCn[m+1:m]=01 (n=0, m=0 or 2)	-0.70	-1.35	—	mA
		3V	V _{OH} =0.9 (V _{DD} /V _{DDIO}), V _{DDIO} =V _{DD} SLEDCn[m+1:m]=10 (n=0, m=0 or 2)	-1.8	-3.6	—	mA
		5V		-3.6	-7.3	—	mA
		5V	V _{OH} =0.9 (V _{DD} /V _{DDIO}), V _{DDIO} =3V SLEDCn[m+1:m]=10 (n=0, m=0 or 2)	-0.95	-1.90	—	mA
		3V	V _{OH} =0.9 (V _{DD} /V _{DDIO}), V _{DDIO} =V _{DD} SLEDCn[m+1:m]=11 (n=0, m=0 or 2)	-4	-8	—	mA
		5V		-8	-16	—	mA
		5V	V _{OH} =0.9 (V _{DD} /V _{DDIO}), V _{DDIO} =3V SLEDCn[m+1:m]=11 (n=0, m=0 or 2)	-2.5	-5.0	—	mA
R _{PH}	Pull-high Resistance for PA1, PA3 and PA4 Pins ^(Note)	3V	Pin power=V _{DD} or V _{DDIO}	20	60	100	kΩ
		5V	V _{DDIO} =V _{DD}	10	30	50	kΩ
		5V	Pin power=V _{DD} or V _{DDIO} V _{DDIO} =3V	36	110	180	kΩ
I _{LEAK}	Input Leakage Current for PA1, PA3 and PA4 Pins	5V	V _{IN} =V _{SS} or V _{IN} =V _{DD} or V _{DDIO}	—	—	±1	μA

Note: The R_{PH} internal pull-high resistance value is calculated by connecting to ground and enabling the input pin with a pull-high resistor and then measuring the pin current at the specified supply voltage level. Dividing the voltage by this measured current provides the R_{PH} value.

Memory Characteristics

Ta=-40°C~85°C, unless otherwise specified

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V _{DD}	Conditions				
V _{DD}	Read/Write Operating Voltage	—	—	V _{DDmin}	—	V _{DDmax}	V
Flash Program/Data EEPROM Memory							
t _{DEW}	Erase/Write Cycle Time – Flash Program Memory	—	—	—	2	3	ms
	Write Cycle Time – Data EEPROM Memory	—	—	—	4	6	
I _{DDPGM}	Programming/Erase Current on V _{DD}	—	—	—	—	5.0	mA
E _P	Cell Endurance – Flash Program Memory	—	—	10K	—	—	E/W
	Cell Endurance – Data EEPROM Memory	—	—	100K	—	—	
t _{RETD}	ROM Data Retention Time	—	Ta=25°C	—	40	—	Year
RAM Data Memory							
V _{DR}	RAM Data Retention Voltage	—	—	1.0	—	—	V

Note: “E/W” means Erase/Write times.

LVD & LVR Electrical Characteristics

Ta=-40°C~85°C

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V _{DD}	Conditions				
V _{LVR}	Low Voltage Reset Voltage	—	LVR enable	-5%	2.1	+5%	V
V _{LVD}	Low Voltage Detection Voltage	—	LVD enable, voltage select 2.0V	-5%	2.0	+5%	V
			LVD enable, voltage select 2.2V		2.2		
			LVD enable, voltage select 2.4V		2.4		
			LVD enable, voltage select 2.7V		2.7		
			LVD enable, voltage select 3.0V		3.0		
			LVD enable, voltage select 3.3V		3.3		
			LVD enable, voltage select 3.6V		3.6		
			LVD enable, voltage select 4.0V		4.0		
I _{LVR/LVDBG}	Operating Current	3V	LVD enable, LVR enable, VBGEN=0	—	—	20	μA
		5V	LVD enable, LVR enable, VBGEN=0	—	20	25	
		3V	LVD enable, LVR enable, VBGEN=1	—	—	25	
		5V	LVD enable, LVR enable, VBGEN=1	—	25	30	
t _{LVDS}	LVDO Stable Time	—	For LVR enable, VBGEN=0, LVD off → on	—	—	18	μs
			For LVR disable, VBGEN=0, LVD off → on	—	—	150	
t _{LVR}	Minimum Low Voltage Width to Reset	—	—	120	240	480	μs
t _{LVD}	Minimum Low Voltage Width to Interrupt	—	—	60	120	240	μs
I _{LVR}	Additional Current for LVR Enable	—	LVD disable, VBGEN=0	—	—	24	μA
I _{LVD}	Additional Current for LVD Enable	—	LVR disable, VBGEN=0	—	—	24	μA

A/D Converter Electrical Characteristics

Ta=-40°C~85°C

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V _{DD}	Conditions				
V _{ADI}	A/D Converter Input Voltage	—	—	0	—	V _{REF}	V
V _{REF}	A/D Converter Reference Voltage	—	—	2.01	—	V _{DD}	V
N _R	Resolution	—	—	—	—	12	Bit
DNL	A/D Converter Differential Non-linearity	—	V _{REF} =AV _{DD} , t _{ADCK} =0.5μs	-3	—	+3	LSB
INL	A/D Converter Integral Non-linearity	—	V _{REF} =AV _{DD} , t _{ADCK} =0.5μs	-4	—	+4	LSB
I _{ADC}	Additional Current for A/D Converter Enable	2.2V	No load, t _{ADCK} =0.5μs	—	300	420	μA
		3V		—	340	500	
		5V		—	500	700	
t _{ADCK}	A/D Converter Clock Period	—	AN≠temperature sensor output	0.5	—	10.0	μs
		2.2V~5.5V	AN=temperature sensor output	1	—	2	
t _{ON2ST}	A/D Converter On-to-Start Time	—	—	4	—	—	μs
t _{ADS}	A/D Converter Sampling Time	—	AN≠temperature sensor output	—	4	—	t _{ADCK}
		2.2V~5.5V	AN=temperature sensor output	—	46	—	
t _{ADC}	A/D Conversion Time (Including A/D Sample and Hold Time)	—	AN≠temperature sensor output	—	16	—	t _{ADCK}
		2.2V~5.5V	AN=temperature sensor output	—	58	—	
GERR	A/D Conversion Gain Error	—	V _{REF} =AV _{DD}	-4	—	4	LSB
OSRR	A/D Conversion Offset Error	—	V _{REF} =AV _{DD}	-4	—	4	LSB
t _{START}	A/D Conversion START=1 Width	—	AN=temperature sensor only	2	—	—	μs

Reference Voltage Characteristics

Ta=-40°C~85°C, unless otherwise specified

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V _{DD}	Conditions				
V _{BGREF}	Bandgap Reference Voltage	—	Ta=25°C	-1%	1.2	+1%	V
			Ta=-40°C~85°C	-2%	1.2	+2%	
I _{BGREF}	Additional Current for Bandgap Enable	5.5V	Ta=-40°C~85°C	—	25	40	μA
PSRR	Power Supply Rejection Ratio	—	Ta=25°C, V _{RIIPPLE} =1V _{P-P} , f _{RIIPPLE} =100Hz	75	—	—	dB
En	Output Noise	—	Ta=25°C, no load current, f=0.1Hz~10Hz	—	300	—	μV _{RMS}
I _{DRV}	Buffer Driving Capability	—	ΔV _{BGREF} =-1%	1	—	—	mA
I _{SD}	Shutdown Current	—	VBGREN=0	—	—	0.1	μA
t _{START}	Startup Time	2.2V~5.5V	Ta=25°C	—	—	400	μs

Note: The V_{BGREF} voltage is used as the A/D converter internal signal input.

Temperature Sensor Characteristics

Ta=-40°C~85°C, unless otherwise specified

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V _{DD}	Conditions				
V _{DD}	Temperature Sensor Operating Voltage	—	—	2.2	—	5.5	V
I _{TS}	Temperature Sensor Operating Current	3V	TSEN=1, t _{ADCK} =1μs, A/D Converter Included	—	1260	1950	μA
		5V		—	1490	2250	
t _{TSS}	Temperature Sensor Turn On Stable Time	3V	—	—	—	100	μs
		5V	—	—	—	100	
V _{TSVREF}	Temperature Sensor Reference Voltage	3V	—	-5%	2.01	+5%	V
		5V	—	-5%	2.01	+5%	
T _{LE}	Temperature Linearity Error	—	—	—	±1	±2	°C
T _{ACC}	Temperature Accuracy (Error)	2.7V~4.5V	V _{REF} =V _{TSVREF} , Ta=0°C~70°C, with linear calibration (Note)	-2	—	+2	°C
		2.7V~5.5V		-2.5	—	+2.5	
		—	V _{REF} =V _{TSVREF} , Ta=-40°C~85°C, with linear calibration (Note)	—	±4	—	
		2.7V~4.5V		-4	—	+4	
TS _{Noise}	Temperature Noise	3V	No average	—	0.4	—	°C(p-p)
		5V		—	0.6	—	

Note: Linear calibration is implemented using the linear formula which is established on the relation between the two calibrated temperatures and their corresponding ADC code. The temperature accuracy T_{ACC} is defined as the error between the actual temperature and the temperature obtained by the conversion of the ADC code through the formula.

LDO Electrical Characteristics

V_{IN}=V_{OUT}+0.3V, C_{LOAD}=4.7μF, Ta=-40°C~85°C, unless otherwise specified

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V _{DD}	Conditions				
V _{IN}	LDO Input Voltage	—	—	2.5	—	5.5	V
V _{OUT}	LDO Output Voltage	—	Ta=25°C, I _{LOAD} =1mA, V _{OUT} =2.2V	-3%	2.2	+3%	V
			Ta=-40°C~85°C, I _{LOAD} =1mA, V _{OUT} =2.2V	-5%	2.2	+5%	
		—	Ta=25°C, I _{LOAD} =1mA, V _{OUT} =2.5V	-2%	2.5	+2%	V
			Ta=-40°C~85°C, I _{LOAD} =1mA, V _{OUT} =2.5V	-5%	2.5	+5%	
		—	Ta=25°C, I _{LOAD} =1mA, V _{OUT} =3.0V	-2%	3.0	+2%	V
			Ta=-40°C~85°C, I _{LOAD} =1mA, V _{OUT} =3.0V	-5%	3.0	+5%	
I _Q	LDO Quiescent Current	5V	No load	—	2.3	5.0	μA
I _{OUT}	LDO Output Current	—	V _{IN} =2.5V, ΔV _{OUT} =0.1V, V _{OUT} =2.2V	10	—	—	mA
		—	V _{IN} =2.8V, ΔV _{OUT} =0.1V, V _{OUT} =2.5V	15	—	—	
		—	V _{IN} =3.4V, ΔV _{OUT} =0.1V, V _{OUT} =3.0V	30	—	—	
TC	Temperature Coefficient	—	Ta=-40°C~85°C, I _{LOAD} =10mA	—	±1.5	±2.0	mV/°C

Note: Load regulation is measured at a constant junction temperature, using pulse testing with a low ON time and is guaranteed up to the maximum power dissipation. Power dissipation is determined by the input/output differential voltage and the output current. Guaranteed maximum power dissipation will not be available over the full input/output range. The maximum allowable power dissipation at any ambient temperature is $P_D = (T_{J(MAX)} - T_a) / \theta_{JA}$

CO/Gas Detector AFE Electrical Characteristics

Ta=-40°C~85°C

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V _{DD}	Conditions				
I _{OPA}	Additional Current for OPAMP Enable	—	OPBW[1:0]=00B, no load	—	3	5	μA
		—	OPBW[1:0]=01B, no load	—	10	16	
		—	OPBW[1:0]=10B, no load	—	80	128	
		—	OPBW[1:0]=11B, no load	—	200	320	
V _{OS}	Input Offset Voltage	5V	Without calibration (OPOF[5:0]=100000B)	-15	—	+15	mV
		5V	With calibration	-4	—	+4	
I _{OS}	Input Offset Current	5V	V _{IN} =(1/2)V _{CM}	—	1	10	nA
V _{CM}	Common Mode Voltage Range	—	OPBW[1:0]=00B/01B/10B/11B	V _{SS}	—	V _{DD} -1.4	V
PSRR	Power Supply Rejection Ratio	5V	OPBW[1:0]=00B/01B/10B/11B	50	70	—	dB
CMRR	Common Mode Rejection Ratio	5V	OPBW[1:0]=00B/01B/10B/11B	50	80	—	dB
A _{OL}	Open Loop Gain	—	OPBW[1:0]=00B/01B/10B/11B	60	80	—	dB
SR	Slew Rate	5V	R _{LOAD} =1MΩ, C _{LOAD} =60pF, OPBW[1:0]=00B	0.5	1.5	—	V/ms
		5V	R _{LOAD} =1MΩ, C _{LOAD} =60pF, OPBW[1:0]=01B	5	15	—	
		5V	R _{LOAD} =1MΩ, C _{LOAD} =60pF, OPBW[1:0]=10B	180	500	—	
		5V	R _{LOAD} =1MΩ, C _{LOAD} =60pF, OPBW[1:0]=11B	600	1800	—	
GBW	Gain Bandwidth	5V	R _{LOAD} =1MΩ, C _{LOAD} =60pF, OPBW[1:0]=00B	1.5	5.0	—	kHz
		5V	R _{LOAD} =1MΩ, C _{LOAD} =60pF, OPBW[1:0]=01B	15	40	—	
		5V	R _{LOAD} =1MΩ, C _{LOAD} =60pF, OPBW[1:0]=10B	400	600	—	
		5V	R _{LOAD} =1MΩ, C _{LOAD} =60pF, OPBW[1:0]=11B	1000	2000	—	
V _{OR}	Maximum Output Voltage Range	5V	OPBW[1:0]=00B/01B, R _L =10kΩ connected to V _{DD} /2	V _{SS} +140	—	V _{DD} -160	mV
		5V	OPBW[1:0]=10B/11B, R _L =10kΩ connected to V _{DD} /2	V _{SS} +120	—	V _{DD} -140	
I _{SC}	Output Short Circuit Current	5V	R _{LOAD} =5.1Ω, OPBW[1:0]=00B/01B	±6	±12	—	mA
		5V	R _{LOAD} =5.1Ω, OPBW[1:0]=10B/11B	±10	±20	—	

Note: These parameters are characterized but not tested.

$V_{DD}=2.2V\sim 5.5V$, $T_a=-40^{\circ}C\sim 85^{\circ}C$

Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
I_{OPA}	Additional Current for OPAMP Enable	OPBW[1:0]=00B, no load	—	2.5	4.0	μA
		OPBW[1:0]=01B, no load	—	10	16	
		OPBW[1:0]=10B, no load	—	80	128	
		OPBW[1:0]=11B, no load	—	200	320	
V_{OS}	Input Offset Voltage	Without calibration (OPOF[5:0]=100000B)	-15	—	+15	mV
		With calibration	-6	—	+6	
I_{OS}	Input Offset Current	$V_{IN}=(1/2)V_{CM}$	—	1	10	nA
V_{CM}	Common Mode Voltage Range	OPBW[1:0]=00B/01B/10B/11B	V_{SS}	—	V_{DD} -1.4	V
PSRR	Power Supply Rejection Ratio	OPBW[1:0]=00B/01B/10B/11B	50	70	—	dB
CMRR	Common Mode Rejection Ratio	OPBW[1:0]=00B/01B/10B/11B	50	80	—	dB
A_{OL}	Open Loop Gain	OPBW[1:0]=00B/01B/10B/11B	60	80	—	dB
SR	Slew Rate	$R_{LOAD}=1M\Omega$, $C_{LOAD}=60pF$, OPBW[1:0]=00B	0.5	1.5	—	V/ms
		$R_{LOAD}=1M\Omega$, $C_{LOAD}=60pF$, OPBW[1:0]=01B	5	15	—	
		$R_{LOAD}=1M\Omega$, $C_{LOAD}=60pF$, OPBW[1:0]=10B	180	500	—	
		$R_{LOAD}=1M\Omega$, $C_{LOAD}=60pF$, OPBW[1:0]=11B	600	1800	—	
GBW	Gain Bandwidth	$R_{LOAD}=1M\Omega$, $C_{LOAD}=60pF$, OPBW[1:0]=00B	1	5	—	kHz
		$R_{LOAD}=1M\Omega$, $C_{LOAD}=60pF$, OPBW[1:0]=01B	10	40	—	
		$R_{LOAD}=1M\Omega$, $C_{LOAD}=60pF$, OPBW[1:0]=10B	250	600	—	
		$R_{LOAD}=1M\Omega$, $C_{LOAD}=60pF$, OPBW[1:0]=11B	800	2000	—	
V_{OR}	Maximum Output Voltage Range	OPBW[1:0]=00B/01B, $R_L=10k\Omega$ connected to $V_{DD}/2$	V_{SS} +140	—	V_{DD} -160	mV
		OPBW[1:0]=10B/11B, $R_L=10k\Omega$ connected to $V_{DD}/2$	V_{SS} +120	—	V_{DD} -140	
I_{SC}	Output Short Circuit Current	$R_{LOAD}=5.1\Omega$, OPBW[1:0]=00B/01B	± 1.2	± 12	—	mA
		$R_{LOAD}=5.1\Omega$, OPBW[1:0]=10B/11B	± 2	± 20	—	

Note: These parameters are characterized but not tested.

Smoke Detector Electrical Characteristics

Ta=25°C, unless otherwise specified

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V _{DD}	Conditions				
V _{DD_SD}	Operating Voltage	—	—	2.2	—	5.5	V
I _{STB_SD}	Standby Current	3V	—	—	1.5	—	μA
V _{IL}	Input Low Voltage for STRx and SINT Pins	5V	—	0	—	1.5	V
		—		0	—	0.2V _{DD_SD}	
V _{IH}	Input High Voltage for STRx and SINT Pins	5V	—	3.5	—	5.0	V
		—		0.8V _{DD_SD}	—	V _{DD_SD}	
I _{OL}	Sink Current for STRx and SINT Pins	3V	V _{OL} =0.1V _{DD_SD}	16	32	—	mA
		5V		32	65	—	
I _{OH}	Source Current for STRx and SINT Pins	3V	V _{OH} =0.9V _{DD_SD}	-4	-8	—	mA
		5V		-8	-16	—	
I _{SINK0}	Sink Current for ISINK0 Pin	5V	V _{ISINK0} =3.0V, ISGDATA0[4:0]=00000B	47.5	50.0	52.5	mA
		—	Ta=-40°C~85°C, V _{ISINK0} =1.0V~4.5V, ISGDATA0[4:0]=00000B	41	50	59	
		—	Ta=-40°C~85°C, V _{ISINK0} =0.7V~1.0V, ISGDATA0[4:0]=00000B	37.5	50.0	50.0	
		5V	V _{ISINK0} =3.0V, ISGDATA0[4:0]=11111B	330	360	390	
		—	Ta=-40°C~85°C, V _{ISINK0} =1.0V~4.5V, ISGDATA0[4:0]=11111B	295	360	425	
		—	Ta=-40°C~85°C, V _{ISINK0} =0.7V~1.0V, ISGDATA0[4:0]=11111B	270	360	360	
I _{SINK1}	Sink Current for ISINK1 Pin	5V	V _{ISINK1} =3.0V, ISGDATA1[4:0]=00000B	43.5	50.0	56.5	mA
		—	Ta=-40°C~85°C, V _{ISINK1} =1.0V~4.5V, ISGDATA1[4:0]=00000B	41	50	59	
		—	Ta=-40°C~85°C, V _{ISINK1} =0.7V~1.0V, ISGDATA1[4:0]=00000B	37.5	50.0	50.0	
		5V	V _{ISINK1} =3.0V, ISGDATA1[4:0]=11111B	178	205	231	
		—	Ta=-40°C~85°C, V _{ISINK1} =1.0V~4.5V, ISGDATA1[4:0]=11111B	168	205	242	
		—	Ta=-40°C~85°C, V _{ISINK1} =0.7V~1.0V, ISGDATA1[4:0]=11111B	154	205	205	
T _{SST}	Power-on Start-up Time	—	—	47	50	60	ms
T _{SINT-WU}	SINT Wake-up Pulse Time	—	—	12	15	18	μs
T _{SINT-WU(D)}	SINT Wake-up Signal Duration	—	Running variable[0x00]=0x00	1382.4	1536.0	1689.6	μs
T _{SINT-S}	SINT Sampling Pulse Time	—	—	12	15	18	μs

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V _{DD}	Conditions				
T _{CHn-OPAW}	Channel n Waiting Time after OPA is on	—	n=A, running variable[0x05]=0x00 n=B, running variable[0x0C]=0x00	1382.4	1536.0	1689.6	μs
T _{CHn-ES}	Channel n Emission Stable Time	—	n=A, running variable[0x06]=0x00 n=B, running variable[0x0D]=0x00	691.2	768.0	844.8	μs
T _{CHn-ADCST00}	Channel n A/D Sample Timeout Time 0	—	n=A, running variable[0x07]=0x00 n=B, running variable[0x0E]=0x00	345.6	384.0	442.4	μs
T _{CHn-ADCST01}	Channel n A/D Sample Timeout Time 1	—	n=A, running variable[0x07]=0x00 n=B, running variable[0x0E]=0x00	576	640	704	μs
T _{SINT-ISINKOFF}	Trigger Time for the SINT Pin to Turn off the ISINK in Advance	—	—	5	—	—	μs
T _{SD}	Smoke Detection Period	—	Running variable[0x11]=0x01	0.9	1.0	1.1	s
T _{TB}	Time Base Wake-up Period	—	Running variable[0x12]=0x01	0.9	1.0	1.1	s
T _{STRx-CR}	STRx Communication Request Time	—	—	9	10	20	μs
T _{STRx-S}	STRx Send Start Signal Time	—	—	10	—	T _{TIMEOUT}	μs
T _{STRx-TS}	STRx Trigger Sampling Time	—	—	38	40	50	μs
T _{STRx-TSP}	STRx Trigger Respond Time	—	—	—	60	80	μ
T _{STRx-R}	STRx Slave Response Time	—	—	12	15	18	μs
T _{DATA1(H)}	Data 1 High-level Time	—	—	26	30	36	μs
T _{DATA1(L)}	Data 1 Low-level Time	—	—	8	10	12	μs
T _{DATA0(H)}	Data 0 High-level Time	—	—	8	10	12	μs
T _{DATA0(L)}	Data 0 Low-level Time	—	—	26	30	36	μs
T _{TIMEOUT}	Communication Timeout Time	—	—	—	454	520	μs
T _{OPACALI}	OPA Calibration Time	—	—	—	300	600	ms

Operational Amplifier Electrical Characteristics

Ta=25°C

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V _{DD}	Conditions				
I _{OPA}	Operating Current	5V	SDAnBW[1:0]=00B, no load	—	3	5	μA
			SDAnBW[1:0]=01B, no load	—	10	16	
			SDAnBW[1:0]=10B, no load	—	80	128	
			SDAnBW[1:0]=11B, no load	—	200	320	
V _{OS}	Input Offset Voltage	5V	With calibration	-2	—	2	mV
I _{OS}	Input Offset Current	5V	V _{IN} =1/2V _{CM}	—	1	10	nA
V _{CM}	Common Mode Voltage Range	5V	SDAnBW[1:0]=00, 01, 10, 11	V _{SS_SD}	—	V _{DD_SD} -1.4	V
PSRR	Power Supply Rejection Ratio	5V	SDAnBW[1:0]=00, 01, 10, 11	50	70	—	dB
CMRR	Common Mode Rejection Ratio	5V	SDAnBW[1:0]=00, 01, 10, 11	50	80	—	dB
A _{OL}	Open Loop Gain	5V	SDAnBW[1:0]=00, 01, 10, 11	60	80	—	dB
SR	Slew Rate	5V	R _{LOAD} =1MΩ, C _{LOAD} =60pF, SDAnBW[1:0]=00	0.5	1.5	—	V/ms
			R _{LOAD} =1MΩ, C _{LOAD} =60pF, SDAnBW[1:0]=01	5	15	—	
			R _{LOAD} =1MΩ, C _{LOAD} =60pF, SDAnBW[1:0]=10	180	500	—	
			R _{LOAD} =1MΩ, C _{LOAD} =60pF, SDAnBW[1:0]=11	600	1800	—	
GBW	Gain Bandwidth	5V	R _{LOAD} =1MΩ, C _{LOAD} =60pF, SDAnBW[1:0]=00	2.5	5.0	—	kHz
			R _{LOAD} =1MΩ, C _{LOAD} =60pF, SDAnBW[1:0]=01	20	40	—	
			R _{LOAD} =1MΩ, C _{LOAD} =60pF, SDAnBW[1:0]=10	400	600	—	
			R _{LOAD} =1MΩ, C _{LOAD} =60pF, SDAnBW [1:0]=11	1300	2000	—	
V _{OR}	Maximum Output Voltage Range	5V	SDAnBW[1:0]=00, 01, R _{LOAD} =5kΩ to V _{DD_SD} /2	V _{SS_SD} +140	—	V _{DD_SD} -160	mV
			SDAnBW[1:0]=10, 11, R _{LOAD} =5kΩ to V _{DD_SD} /2	V _{SS_SD} +120	—	V _{DD_SD} -140	
I _{SC}	Output Short Circuit Current	5V	R _{LOAD} =5.1Ω, SDAnBW[1:0]=00,01	±6	±12	—	mA
			R _{LOAD} =5.1Ω, SDAnBW[1:0]=10,11	±10	±20	—	

Ta=25°C

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V _{DD}	Conditions				
I _{OPA}	Operating Current	—	SDAnBW[1:0]=00B, no load	—	2.5	4.0	μA
			SDAnBW[1:0]=01B, no load	—	10	16	
			SDAnBW[1:0]=10B, no load	—	80	128	
			SDAnBW[1:0]=11B, no load	—	200	320	
V _{OS}	Input Offset Voltage	—	With calibration	-2	—	2	mV
I _{OS}	Input Offset Current	—	V _{IN} =1/2V _{CM}	—	1	10	nA
V _{CM}	Common Mode Voltage Range	—	SDAnBW[1:0]=00, 01, 10, 11	V _{SS_SD}	—	V _{DD_SD} -1.4	V
PSRR	Power Supply Rejection Ratio	—	SDAnBW[1:0]=00, 01, 10, 11	50	70	—	dB
CMRR	Common Mode Rejection Ratio	—	SDAnBW[1:0]=00, 01, 10, 11	50	80	—	dB
A _{OL}	Open Loop Gain	—	SDAnBW[1:0]=00, 01, 10, 11	60	80	—	dB
SR	Slew Rate	—	R _{LOAD} =1MΩ, C _{LOAD} =60pF, SDAnBW[1:0]=00	0.5	1.5	—	V/ms
			R _{LOAD} =1MΩ, C _{LOAD} =60pF, SDAnBW[1:0]=01	5	15	—	
			R _{LOAD} =1MΩ, C _{LOAD} =60pF, SDAnBW[1:0]=10	180	500	—	
			R _{LOAD} =1MΩ, C _{LOAD} =60pF, SDAnBW[1:0]=11	600	1800	—	
GBW	Gain Bandwidth	—	R _{LOAD} =1MΩ, C _{LOAD} =60pF, SDAnBW[1:0]=00	2	5	—	kHz
			R _{LOAD} =1MΩ, C _{LOAD} =60pF, SDAnBW[1:0]=01	15	40	—	
			R _{LOAD} =1MΩ, C _{LOAD} =60pF, SDAnBW[1:0]=10	250	600	—	
			R _{LOAD} =1MΩ, C _{LOAD} =60pF, SDAnBW[1:0]=11	800	2000	—	
V _{OR}	Maximum Output Voltage Range	—	SDAnBW[1:0]=00, 01 R _{LOAD} =5kΩ to V _{DD_SD} /2	V _{SS_SD} +140	—	V _{DD_SD} -160	mV
			SDAnBW[1:0]=10, 11 R _{LOAD} =5kΩ to V _{DD_SD} /2	V _{SS_SD} +120	—	V _{DD_SD} -140	
I _{SC}	Output Short Circuit Current	—	R _{LOAD} =5.1Ω, SDAnBW[1:0]=00, 01	±1.2	±12	—	mA
			R _{LOAD} =5.1Ω, SDAnBW[1:0]=10, 11	±2	±20	—	

Note: These parameters are characterized but not tested.

LCD Electrical Characteristics

Ta=-40°C~85°C

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V _{DD}	Conditions				
V _{IN}	LCD Operating Voltage	—	R type, power supply from PLCD pin, PLCD[3:0]=1xxxB	3.0	—	5.5	V
		—	C type, power supply from PLCD pin	2.0	—	3.7	V
		—	C type, power supply from V1 pin	3.0	—	5.5	V
		—	C type, power supply from V2 pin	1.0	—	1.8	V
		—	C type, power supply from V _A	3.0	—	5.5	V
		—	C type, power supply from V _B	2.0	—	3.7	V
		2.2V~5.5V	C type, power supply from V _C	-10%	1.04	+10%	V
I _{LCD}	Additional Current for LCD Enable (R type)	3V	R _T =1170kΩ, no load, V _A =V _{PLCD} =V _{DD} , 1/3 bias	—	3	6	μA
		5V		—	5	10	
		3V	R _T =225kΩ, no load, V _A =V _{PLCD} =V _{DD} , 1/3 bias	—	16	28	μA
		5V		—	21	40	
	Additional Current for LCD Enable (C type)	3V	R _T =60kΩ, no load, V _A =V _{PLCD} =V _{DD} , 1/3 bias	—	50	75	μA
		5V		—	80	120	
		3V	No load, V _A =V ₁ =V _{DD} , 1/3 bias, LCDP[1:0]=11B	—	0.6	1.2	μA
		5V	No load, V _A =V ₁ =V _{DD} , 1/3 bias, LCDP[1:0]=01B, V _C =V _{REFIN}	—	1	2	
I _{LCDOL}	LCD Common and Segment Sink Current	3V	V _{OL} =0.1V _{DD}	210	420	—	μA
		5V		350	700	—	
I _{LCDOH}	LCD Common and Segment Source Current	3V	V _{OH} =0.9V _{DD}	-80	-160	—	μA
		5V		-180	-360	—	

16-bit Vioces D/A Converter Electrical Characteristics

Ta=-40°C~85°C

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V _{DD}	Conditions				
I _{DAC}	Additional Current for D/A Converter Enable with Buffer	3V	—	—	—	3	mA
		5V	—	—	—	3	
I _{STB(DAC)}	Standby Current	5V	DACEN=0	—	—	1	μA
THD+N	Total Harmonic Distortion + Noise ^(Note)	3V	10kΩ load	—	-55	—	dB
V _{OUT}	Output Voltage Range	5V	No load	0.01	—	0.99	V _{DD}
t _{DACS}	D/A Converter Circuit Turn on Stable Time	5V	—	—	—	1	ms

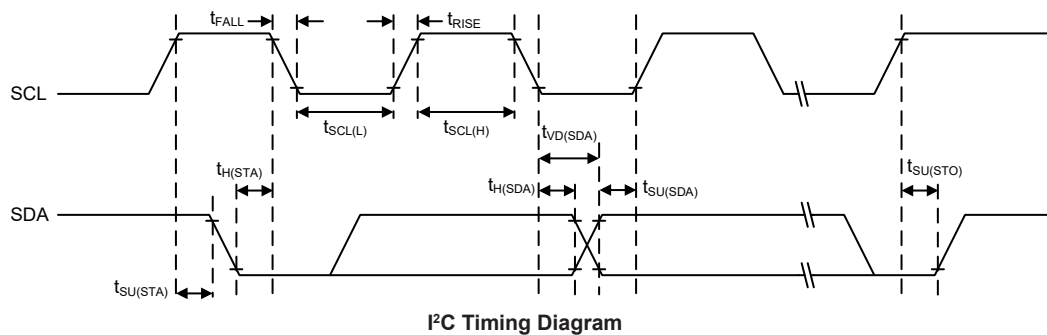
Note: Sin wave input @ 1kHz, -6dBFS.

I²C Electrical Characteristics

Ta=25°C

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V _{DD}	Conditions				
f _{I2C}	I ² C Standard Mode (100kHz) f _{sys} Frequency ^(Note)	—	No debounce	2	—	—	MHz
			2 system clock debounce	4	—	—	
			4 system clock debounce	4	—	—	
	I ² C Fast Mode (400kHz) f _{sys} Frequency ^(Note)	—	No debounce	4	—	—	MHz
			2 system clock debounce	8	—	—	
			4 system clock debounce	8	—	—	
f _{SCL}	SCL Clock Frequency	3V/5V	Standard mode	—	—	100	kHz
			Fast mode	—	—	400	
t _{SCL(H)}	SCL Clock High Time	3V/5V	Standard mode	3.5	—	—	μs
			Fast mode	0.9	—	—	
t _{SCL(L)}	SCL Clock Low Time	3V/5V	Standard mode	3.5	—	—	μs
			Fast mode	0.9	—	—	
t _{FALL}	SCL and SDA Fall Time	3V/5V	Standard mode	—	—	1.3	μs
			Fast mode	—	—	0.34	
t _{RISE}	SCL and SDA Rise Time	3V/5V	Standard mode	—	—	1.3	μs
			Fast mode	—	—	0.34	
t _{SU(SDA)}	SDA Data Setup Time	3V/5V	Standard mode	0.25	—	—	μs
			Fast mode	0.1	—	—	
t _{H(SDA)}	SDA Data Hold Time	3V/5V	—	0.1	—	—	μs
t _{VD(SDA)}	SDA Data Valid Time	3V/5V	—	—	—	0.6	μs
t _{SU(STA)}	Start Condition Setup Time	3V/5V	Standard mode	3.5	—	—	μs
			Fast mode	0.6	—	—	
t _{H(STA)}	Start Condition Hold Time	3V/5V	Standard mode	4.0	—	—	μs
			Fast mode	0.6	—	—	
t _{SU(STO)}	Stop Condition Setup Time	3V/5V	Standard mode	3.5	—	—	μs
			Fast mode	0.6	—	—	

Note: Using the debounce function can make the transmission more stable and reduce the probability of communication failure due to interference.

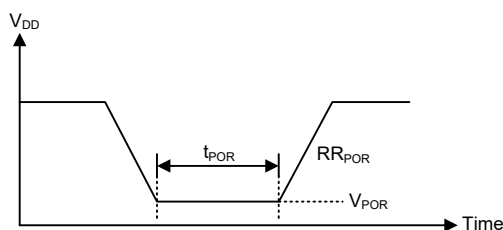


I²C Timing Diagram

Power-on Reset Characteristics

$T_a = -40^{\circ}\text{C} \sim 85^{\circ}\text{C}$

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V_{DD}	Conditions				
V_{POR}	V_{DD} Start Voltage to Ensure Power-on Reset	—	—	—	—	100	mV
RR_{POR}	V_{DD} Rising Rate to Ensure Power-on Reset	—	—	0.035	—	—	V/ms
t_{POR}	Minimum Time for V_{DD} Stays at V_{POR} to Ensure Power-on Reset	—	—	1	—	—	ms



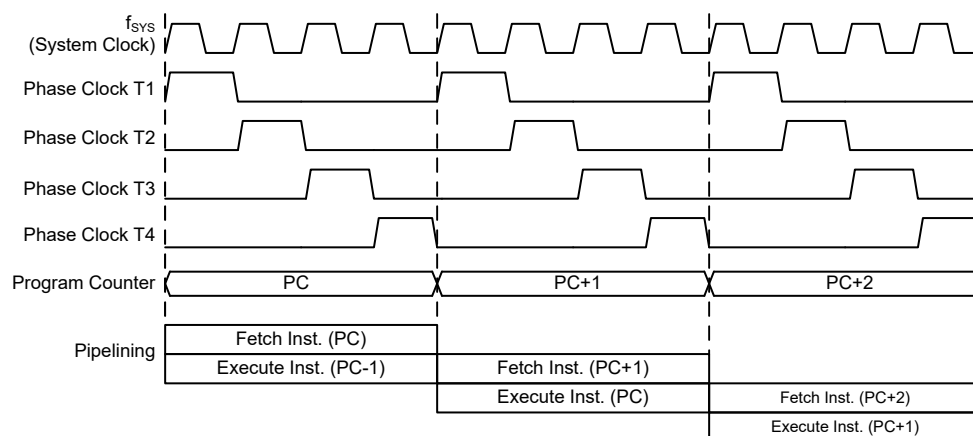
System Architecture

A key factor in the high-performance features of the Holtek range of microcontrollers is attributed to their internal system architecture. The range of the device take advantage of the usual features found within RISC microcontrollers providing increased speed of operation and enhanced performance. The pipelining scheme is implemented in such a way that instruction fetching and instruction execution are overlapped, hence instructions are effectively executed in one or two cycles for most of the standard or extended instructions respectively, with the exception of branch or call instructions which need one more cycle. An 8-bit wide ALU is used in practically all instruction set operations, which carries out arithmetic operations, logic operations, rotation, increment, decrement, branch decisions, etc. The internal data path is simplified by moving data through the Accumulator and the ALU. Certain internal registers are implemented in the Data Memory and can be directly or indirectly addressed. The simple addressing methods of these registers along with additional architectural features ensure that a minimum of external components is required to provide a functional I/O and A/D control system with maximum reliability and flexibility. This makes the device suitable for low-cost, high-volume production for controller applications.

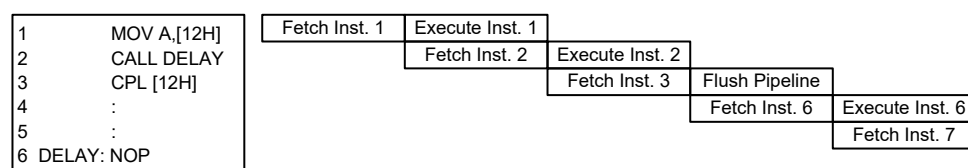
Clocking and Pipelining

The main system clock, derived from either a HIRC or LIRC oscillator is subdivided into four internally generated non-overlapping clocks, T1~T4. The Program Counter is incremented at the beginning of the T1 clock during which time a new instruction is fetched. The remaining T2~T4 clocks carry out the decoding and execution functions. In this way, one T1~T4 clock cycle forms one instruction cycle. Although the fetching and execution of instructions takes place in consecutive instruction cycles, the pipelining structure of the microcontroller ensures that instructions are effectively executed in one instruction cycle. The exception to this are instructions where the contents of the Program Counter are changed, such as subroutine calls or jumps, in which case the instruction will take one more instruction cycle to execute.

For instructions involving branches, such as jump or call instructions, two machine cycles are required to complete instruction execution. An extra cycle is required as the program takes one cycle to first obtain the actual jump or call address and then another cycle to actually execute the branch. The requirement for this extra cycle should be taken into account by programmers in timing sensitive applications.



System Clocking and Pipelining



Instruction Fetching

Program Counter

During program execution, the Program Counter is used to keep track of the address of the next instruction to be executed. It is automatically incremented by one each time an instruction is executed except for instructions, such as “JMP” or “CALL” that demand a jump to a non-consecutive Program Memory address. Only the lower 8 bits, known as the Program Counter Low Register, are directly addressable by the application program.

When executing instructions requiring jumps to non-consecutive addresses such as a jump instruction, a subroutine call, interrupt or reset, etc., the microcontroller manages program control by loading the required address into the Program Counter. For conditional skip instructions, once the condition has been met, the next instruction, which has already been fetched during the present instruction execution, is discarded and a dummy cycle takes its place while the correct instruction is obtained.

Program Counter	
Program Counter High Byte	PCL Register
PC12~PC8	PCL7~PCL0

Program Counter

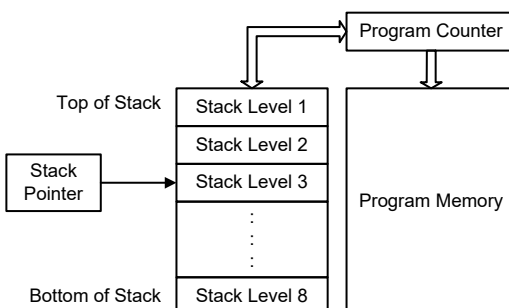
The lower byte of the Program Counter, known as the Program Counter Low register or PCL, is available for program control and is a readable and writeable register. By transferring data directly into this register, a short program jump can be executed directly; however, as only this low byte is available for manipulation, the jumps are limited to the present page of memory that is 256 locations. When such program jumps are executed it should also be noted that a dummy cycle will be inserted. Manipulating the PCL register may cause program branching, so an extra cycle is needed to pre-fetch.

Stack

This is a special part of the memory which is used to save the contents of the Program Counter only. The stack is organized into 8 levels and neither part of the data nor part of the program space, and is neither readable nor writeable. The activated level is indexed by the Stack Pointer, and is neither readable nor writeable. At a subroutine call or interrupt acknowledge signal, the contents of the Program Counter are pushed onto the stack. At the end of a subroutine or an interrupt routine, signaled by a return instruction, RET or RETI, the Program Counter is restored to its previous value from the stack. After a device reset, the Stack Pointer will point to the top of the stack.

If the stack is full and an enabled interrupt takes place, the interrupt request flag will be recorded but the acknowledge signal will be inhibited. When the Stack Pointer is decremented, by RET or RETI, the interrupt will be serviced. This feature prevents stack overflow allowing the programmer to use the structure more easily. However, when the stack is full, a CALL subroutine instruction can still be executed which will result in a stack overflow. Precautions should be taken to avoid such cases which might cause unpredictable program branching.

If the stack is overflow, the first Program Counter save in the stack will be lost.



Arithmetic and Logic Unit – ALU

The arithmetic-logic unit or ALU is a critical area of the microcontroller that carries out arithmetic and logic operations of the instruction set. Connected to the main microcontroller data bus, the ALU receives related instruction codes and performs the required arithmetic or logical operations after which the result will be placed in the specified register. As these ALU calculation or operations may result in carry, borrow or other status changes, the status register will be correspondingly updated to reflect these changes. The ALU supports the following functions:

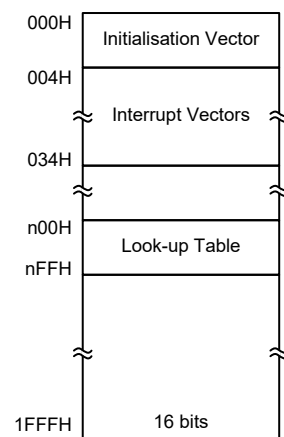
- Arithmetic operations:
 ADD, ADDM, ADC, ADCM, SUB, SUBM, SBC, SBCM, DAA,
 LADD, LADDM, LADC, LADCM, LSUB, LSUBM, LSBC, LSBCM, LDAA
- Logic operations:
 AND, OR, XOR, ANDM, ORM, XORM, CPL, CPLA,
 LAND, LANDM, LOR, LORM, LXOR, LXORM, LCPL, LCPLA
- Rotation:
 RRA, RR, RRCA, RRC, RLA, RL, RLCA, RLC,
 LRR, LRRCA, LRRCA, LRRCA, LRLA, LRL, LRLCA, LRLC
- Increment and Decrement:
 INCA, INC, DECA, DEC,
 LINCA, LINC, LDECA, LDEC
- Branch decision:
 JMP, SZ, SZA, SNZ, SIZ, SDZ, SIZA, SDZA, CALL, RET, RETI,
 LSNZ, LSZ, LSZA, LSIZ, LSIZA, LSDZ, LSDZA

Flash Program Memory

The Program Memory is the location where the user code or program is stored. For the device the Program Memory is Flash type, which means it can be programmed and re-programmed a large number of times, allowing the user the convenience of code modification on the same device. By using the appropriate programming tools, the Flash device offers users the flexibility to conveniently debug and develop their applications while also offering a means of field programming and updating.

Structure

The Program Memory has a capacity of 8K×16 bits. The Program Memory is addressed by the Program Counter and also contains data, table information and interrupt entries. Table data, which can be setup in any location within the Program Memory, is addressed by a separate table pointer register.



Program Memory Structure

Special Vectors

Within the Program Memory, certain locations are reserved for the reset and interrupts. The location 000H is reserved for use by the device reset for program initialisation. After a device reset is initiated, the program will jump to this location and begin execution.

Look-up Table

Any location within the Program Memory can be defined as a look-up table where programmers can store fixed data. To use the look-up table, the table pointer must first be setup by placing the address of the look up data to be retrieved in the table pointer registers, TBLP and TBHP. These registers define the total address of the look-up table.

After setting up the table pointer, the table data can be retrieved from the Program Memory using the “ITABRD [m]”, “ITABRDL [m]”, “TABRD [m]” or “ITABRDL [m]” instructions respectively when the memory [m] is located in Sector 0. If the memory [m] is located in other sectors except sector 0, the data can be retrieved from the program memory using the corresponding extended table read instruction such as “LITABRD [m]”, “LITABRDL [m]”, “LTABRD [m]” or “LITABRDL [m]” respectively. When the instruction is executed, the lower order table byte from the Program Memory will be transferred to the user defined Data Memory register [m] as specified in the instruction. The higher order table data byte from the Program Memory will be transferred to the TBLH special register. Any unused bits in this transferred higher order byte will be read as “0”.

The accompanying diagram illustrates the addressing data flow of the look-up table.

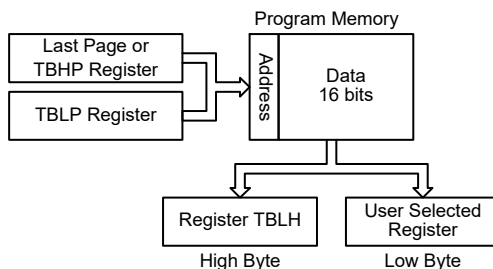


Table Program Example

The following example shows how the table pointer and table data is defined and retrieved from the microcontroller. This example uses raw table data located in the Program Memory which is stored there using the ORG statement. The value at this ORG statement is “1F00H” which refers to the start address of the last page within the 8K words Program Memory of the device. The table pointer low byte register is setup here to have an initial value of “06H”. This will ensure that the first data read from the data table will be at the Program Memory address “1F06H” or 6 locations after the start of the last page. Note that the value for the table pointer is referenced to the specific address pointed by the TBLP and TBHP registers if the “TABRD [m]” or “LTABRD [m]” instruction is being used. The high byte of the table data which in this case is equal to zero will be transferred to the TBLH register automatically when the “TABRD [m]” or “LTABRD [m]” instruction is executed.

Because the TBLH register is a read/write register and can be restored, care should be taken to ensure its protection if both the main routine and Interrupt Service Routine use table read instructions. If using the table read instructions, the Interrupt Service Routines may change the value of the TBLH and subsequently cause errors if used again by the main routine. As a rule it is recommended that simultaneous use of the table read instructions should be avoided. However, in situations where simultaneous use cannot be avoided, the interrupts should be disabled prior to the execution of any main routine table-read instructions. Note that all table related instructions require two instruction cycles to complete their operation.

Table Read Program Example

```

tempreg1 db?      ; temporary register #1
tempreg2 db?      ; temporary register #2
:
mov a,06h         ; initialise low table pointer - note that this address is referenced
mov tblp,a        ; to the last page or the page that tbhp pointed
mov a,1Fh         ; initialise high table pointer
mov tbhp,a        ; it is not necessary to set tbhp if executing tabrdl or ltabrdl
:
:
tabrd tempreg1     ; transfers value in table referenced by table pointer
                  ; data at program memory address "1F06H" transferred to tempreg1 and
                  ; TBLH
dec tblp           ; reduce value of table pointer by one
tabrd tempreg2     ; transfers value in table referenced by table pointer
                  ; data at program memory address "1F05H" transferred to tempreg2 and
                  ; TBLH
                  ; in this example the data "1AH" is transferred to tempreg1 and data "0FH"
                  ; to tempreg2
                  ; the value "00H" will be transferred to the high byte register TBLH
:
org 1F00h          ; sets initial address of program memory
dc 00Ah,00Bh,00Ch,00Dh,00Eh,00Fh,01Ah,01Bh
  
```

In Circuit Programming – ICP

The provision of Flash type Program Memory provides the user with a means of convenient and easy upgrades and modifications to their programs on the same device.

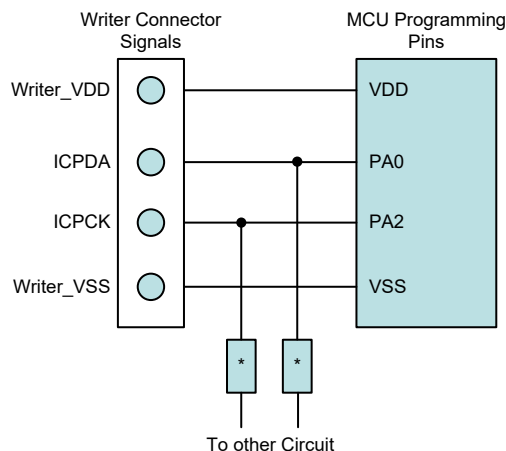
As an additional convenience, Holtek has provided a means of programming the microcontroller in-circuit using a 4-pin interface. This provides manufacturers with the possibility of manufacturing their circuit boards complete with a programmed or un-programmed microcontroller, and then programming or upgrading the program at a later stage. This enables product manufacturers to easily keep their manufactured products supplied with the latest program releases without removal and re-insertion of the device.

The Holtek Flash MCU to Writer Programming Pin correspondence table is as follows:

Holtek Writer Pins	MCU Programming Pins	Pin Description
ICPDA	PA0	Programming Serial Data/Address
ICPCK	PA2	Programming Clock
VDD	VDD	Power Supply
VSS	VSS	Ground

The Program Memory can be programmed serially in-circuit using this 4-wire interface. Data is downloaded and uploaded serially on a single pin with an additional line for the clock. Two additional lines are required for the power supply. The technical details regarding the in-circuit programming of the device is beyond the scope of this document and will be supplied in supplementary literature.

During the programming process, the user must take care of the ICPDA and ICPCK pins for data and clock programming purposes to ensure that no other outputs are connected to these two pins.



Note: * may be resistor or capacitor. The resistance of * must be greater than 1kΩ or the capacitance of * must be less than 1nF.

On-Chip Debug Support – OCDS

There is an EV chip named BA45V6956 which is used to emulate the real MCU device named BA45F6956. The EV chip device also provides the “On-Chip Debug” function to debug the real MCU device during the development process. The EV chip and the real MCU device are almost functionally compatible except for “On-Chip Debug” function. Users can use the EV chip device to emulate the real chip device behavior by connecting the OCSDA and OCDSCK pins to the Holtek HT-IDE development tools. The OCSDA pin is the OCDS Data/Address input/output pin while the OCDSCK pin is the OCDS clock input pin. When users use the EV chip device for debugging,

the corresponding pin functions shared with the OCSDSA and OCDSCCK pins will have no effect in the EV chip. However, the two OCDS pins which are pin-shared with the ICP programming pins are still used as the Flash Memory programming pins for ICP. For more detailed OCDS information, refer to the corresponding document named “Holtek e-Link for 8-bit MCU OCDS User’s Guide”.

Holtek e-Link Pins	EV Chip OCDS Pins	Pin Description
OCSDSA	OCSDSA	On-Chip Debug Support Data/Address input/output
OCDSCCK	OCDSCCK	On-Chip Debug Support Clock input
VDD	VDD	Power Supply
GND	VSS	Ground

In Application Programming – IAP

Flash type Program Memory provides the user with a means of convenient and easy upgrades and modifications to their programs on the same device. The provision of the IAP function offers users the convenience of Flash Memory multi-programming features. The convenience of the IAP function is that it can execute the updated program procedure using its internal firmware, without requiring an external Program Writer or PC. In addition, the IAP interface can also be any type of communication protocol, such as UART, using I/O pins. Regarding the internal firmware, the user can select versions provided by Holtek or create their own. The following section illustrates the procedures regarding how to implement the IAP firmware.

Flash Memory Read/Write Size

The Flash memory Erase and Write operations are carried out in a page format while the Read operation is carried out in a word format. The page size and write buffer size are both assigned with a capacity of 32 words. Note that the Erase operation should be executed before the Write operation is executed.

When the Flash Memory Erase/Write Function is successfully enabled, the CFWEN bit will be set high. When the CFWEN bit is set high, the data can be written into the write buffer. The FWT bit is used to initiate the write process and then indicate the write operation status. This bit is set high by application programs to initiate a write process and will be cleared by hardware if the write process is finished.

The Read operation can be carried out by executing a specific read procedure. The FRDEN bit is used to enable the read function and the FRD bit is used to initiate the read process by application programs and then indicate the read operation status. When the read process is finished, this bit will be cleared by hardware.

Operations	Format
Erase	32 words/page/time
Write	32 words/time
Read	1 word/time
Note: Page size=Write buffer size=32 words.	

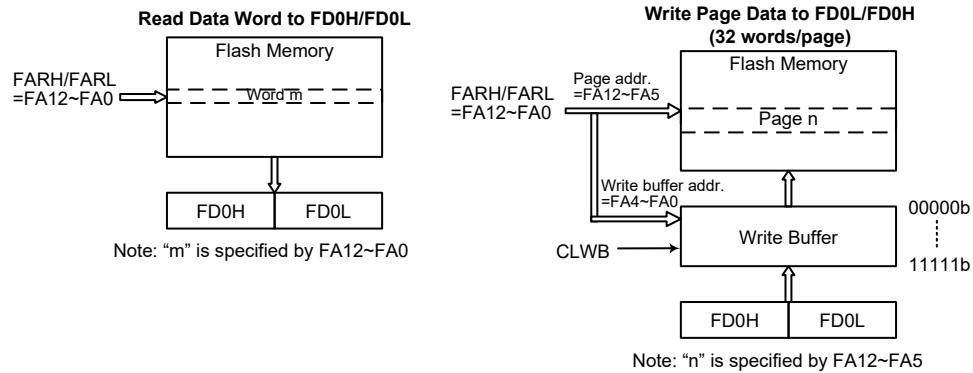
IAP Operation Format

Erase Page	FARH	FARL[7:5]	FARL[4:0]
0	0000 0000	000	x xxxx
1	0000 0000	001	x xxxx
2	0000 0000	010	x xxxx
3	0000 0000	011	x xxxx
4	0000 0000	100	x xxxx

Erase Page	FARH	FARL[7:5]	FARL[4:0]
⋮	⋮	⋮	⋮
254	0001 1111	110	x xxxx
255	0001 1111	111	x xxxx

"x": Don't care

Erase Page Number and Selection



Flash Memory IAP Read/Write Structure

Write Buffer

The write buffer is used to store the written data temporarily when executing the write operation. The Write Buffer can be filled with written data after the Flash Memory Erase/Write Function has been successfully enabled by executing the Flash Memory Erase/Write Function Enable procedure. The write buffer can be cleared by configuring the CLWB bit in the FC2 register. The CLWB bit can be set high to enable the Clear Write Buffer procedure. When the procedure is finished this bit will be cleared to low by the hardware. It is recommended that the write buffer should be cleared by setting the CLWB bit high before the write buffer is used for the first time or when the data in the write buffer is updated.

The write buffer size is 32 words corresponding to a page. The write buffer address is mapped to a specific flash memory page specified by the memory address bits, FA12~FA5. The data written into the FD0L and FD0H registers will be loaded into the write buffer. When data is written into the high byte data register, FD0H, it will result in the data stored in the high and low byte data registers both being written into the write buffer. It will also cause the flash memory address to be incremented by one, after which the new address will be loaded into the FARH and FARL address registers. When the flash memory address reaches the page boundary, 11111b of a page with 32 words, the address will now not be incremented but will stop at the last address of the page. At this point a new page address should be specified for any other erase/write operations.

After a write process is finished, the write buffer will automatically be cleared by the hardware. Note that the write buffer should be cleared manually by the application program when the data written into the flash memory is incorrect in the data verification step. The data should again be written into the write buffer after the write buffer has been cleared when the data is found to be incorrect during the data verification step.

IAP Flash Program Memory Registers

There are two address registers, four 16-bit data registers and three control registers. All the registers are located in Sector 0. Read and Write operations to the Flash memory are carried out by 16-bit data operations using the address and data registers and the control register. Several registers control the overall operation of the internal Flash Program Memory. The address registers are named FARL and FARH, the data registers are named FDnL and FDnH and the control registers are named FC0, FC1 and FC2.

Register Name	Bit							
	7	6	5	4	3	2	1	0
FC0	CFWEN	FMOD2	FMOD1	FMOD0	FWPEN	FWT	FRDEN	FRD
FC1	D7	D6	D5	D4	D3	D2	D1	D0
FC2	—	—	—	—	—	—	—	CLWB
FARL	FA7	FA6	FA5	FA4	FA3	FA2	FA1	FA0
FARH	—	—	—	FA12	FA11	FA10	FA9	FA8
FD0L	D7	D6	D5	D4	D3	D2	D1	D0
FD0H	D15	D14	D13	D12	D11	D10	D9	D8
FD1L	D7	D6	D5	D4	D3	D2	D1	D0
FD1H	D15	D14	D13	D12	D11	D10	D9	D8
FD2L	D7	D6	D5	D4	D3	D2	D1	D0
FD2H	D15	D14	D13	D12	D11	D10	D9	D8
FD3L	D7	D6	D5	D4	D3	D2	D1	D0
FD3H	D15	D14	D13	D12	D11	D10	D9	D8

IAP Register List

• FARL Register

Bit	7	6	5	4	3	2	1	0
Name	FA7	FA6	FA5	FA4	FA3	FA2	FA1	FA0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **FA7~FA0**: Flash Memory Address bit 7 ~ bit 0

• FARH Register

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	FA12	FA11	FA10	FA9	FA8
R/W	—	—	—	R/W	R/W	R/W	R/W	R/W
POR	—	—	—	0	0	0	0	0

Bit 7~5 Unimplemented, read as “0”

Bit 4~0 **FA12~FA8**: Flash Memory Address bit 12 ~ bit 8

• FD0L Register

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D7~D0**: The first Flash Memory data bit 7 ~ bit 0

Note that data written into the low byte data register FD0L will only be stored in the FD0L register and not loaded into the lower 8-bit write buffer.

• **FD0H Register**

Bit	7	6	5	4	3	2	1	0
Name	D15	D14	D13	D12	D11	D10	D9	D8
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D15~D8:** The first Flash Memory data bit 15 ~ bit 8

Note that when 8-bit data is written into the high byte data register FD0H, the whole 16 bits of data stored in the FD0H and FD0L registers will simultaneously be loaded into the 16-bit write buffer after which the contents of the Flash memory address register pair, FARH and FARL, will be incremented by one.

• **FD1L Register**

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D7~D0:** The second Flash Memory data bit 7 ~ bit 0

• **FD1H Register**

Bit	7	6	5	4	3	2	1	0
Name	D15	D14	D13	D12	D11	D10	D9	D8
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D15~D8:** The second Flash Memory data bit 15 ~ bit 8

• **FD2L Register**

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D7~D0:** The third Flash Memory data bit 7 ~ bit 0

• **FD2H Register**

Bit	7	6	5	4	3	2	1	0
Name	D15	D14	D13	D12	D11	D10	D9	D8
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D15~D8:** The second Flash Memory data bit 15 ~ bit 8

• **FD3L Register**

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D7~D0:** The fourth Flash Memory data bit 7 ~ bit 0

• **FD3H Register**

Bit	7	6	5	4	3	2	1	0
Name	D15	D14	D13	D12	D11	D10	D9	D8
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D15~D8**: The fourth Flash Memory data bit 15 ~ bit 8

• **FC0 Register**

Bit	7	6	5	4	3	2	1	0
Name	CFWEN	FMOD2	FMOD1	FMOD0	FWPEN	FWT	FRDEN	FRD
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7 **CFWEN**: Flash Memory Erase/Write function enable control

0: Flash Memory erase/write function is disabled

1: Flash Memory erase/write function has been successfully enabled

When this bit is cleared to zero by application program, the Flash Memory erase/write function is disabled. Note that this bit cannot be set high by application programs. Writing “1” into this bit results in no action. This bit is used to indicate that the Flash Memory erase/write function status. When this bit is set high by hardware, it means that the Flash Memory erase/write function is enabled successfully. Otherwise, the Flash Memory erase/write function is disabled as the bit content is zero.

Bit 6~4 **FMOD2~FMOD0**: Flash Memory Mode selection

000: Write Mode

001: Page Erase Mode

010: Reserved

011: Read Mode

100: Reserved

101: Reserved

110: Flash Memory Erase/Write function Enable Mode

111: Reserved

These bits are used to select the Flash Memory operation modes. Note that the “Flash memory Erase/Write function Enable Mode” should first be successfully enabled before the Erase or Write Flash memory operation is executed.

Bit 3 **FWPEN**: Flash Memory Erase/Write function enable procedure trigger

0: Erase/Write function enable procedure is not triggered or procedure timer times out

1: Erase/Write function enable procedure is triggered and procedure timer starts to count

This bit is used to activate the flash memory Erase/Write function enable procedure and an internal timer. It is set by the application programs and then cleared to zero by the hardware when the internal timer times out. The correct patterns must be written into the FD1L/FD1H, FD2L/FD2H and FD3L/FD3H register pairs respectively as soon as possible after the FWPEN bit is set high.

Bit 2 **FWT**: Flash Memory write initiate control

0: Do not initiate Flash Memory write or indicating that a Flash Memory write process has completed

1: Initiate a Flash Memory write process

This bit is set by software and cleared to zero by the hardware when the Flash memory write process has completed. Note that all CPU operations will be stopped cease when this bit is set to 1.

- Bit 1 **FRDEN**: Flash Memory read enabled bit
 0: Flash Memory read disable
 1: Flash Memory read enable
 This is the Flash memory Read Enable bit which must be set high before any Flash memory read operations are carried out. Clearing this bit to zero will inhibit Flash memory read operations.
- Bit 0 **FRD**: Flash Memory read control bit
 0: Do not initiate Flash Memory read or indicating that a Flash Memory read process has completed
 1: Initiate a Flash Memory read process
 This bit is set by software and cleared to zero by the hardware when the Flash memory read process has completed. Note that all CPU operations will be stopped cease when this bit is set to 1.

- Note: 1. The FWT, FRDEN and FRD bits cannot be set to “1” at the same time with a single instruction.
 2. Ensure that the f_{SUB} clock is stable before executing the erase or write operation.
 3. Note that the CPU will be stopped when a read, erase or write operation is successfully activated.
 4. Ensure that the read, erase or write operation is totally complete before executing other operations.

• **FC1 Register**

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

- Bit 7~0 **D7~D0**: Chip Reset Pattern
 When a specific value of “55H” is written into this register, a reset signal will be generated to reset the whole chip.

• **FC2 Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	—	CLWB
R/W	—	—	—	—	—	—	—	R/W
POR	—	—	—	—	—	—	—	0

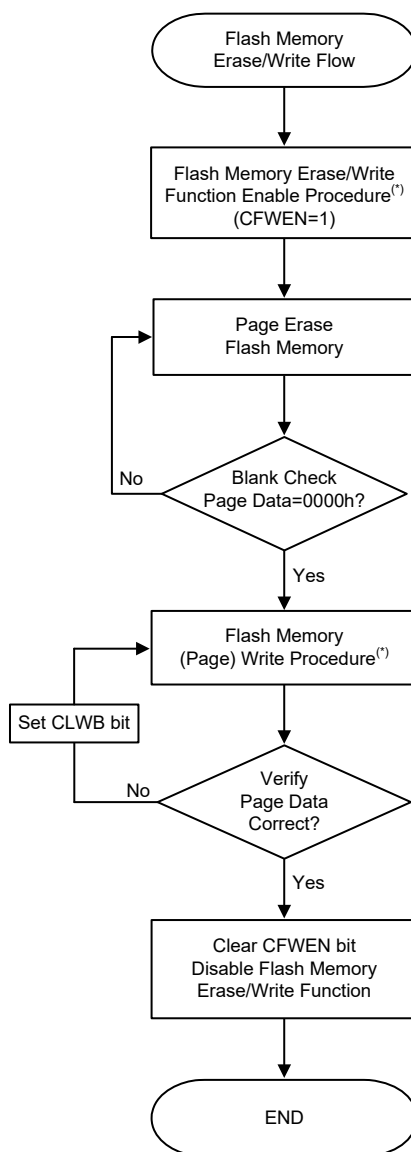
- Bit 7~1 Unimplemented, read as “0”
- Bit 0 **CLWB**: Flash Memory Write buffer clear control
 0: Do not initiate a Write Buffer Clear process or indicating that a Write Buffer Clear process has completed
 1: Initiate a Write Buffer Clear process
 This bit is set by software and cleared to zero by hardware when the Write Buffer Clear process has completed.

Flash Memory Erase/Write Flow

It is important to understand the Flash memory Erase/Write flow before the Flash memory contents are updated. Users can refer to the corresponding operation procedures when developing their IAP program to ensure that the flash memory contents are correctly updated.

Flash Memory Erase/Write Flow Descriptions

1. Activate the “Flash Memory Erase/Write function enable procedure” first. When the Flash Memory Erase/Write function is successfully enabled, the CFWEN bit in the FC0 register will automatically be set high by hardware. After this, Erase or Write operations can be executed on the Flash memory. Refer to the “Flash Memory Erase/Write Function Enable Procedure” for details.
2. Configure the flash memory address to select the desired erase page and then erase this page.
3. Execute a Blank Check operation to ensure whether the page erase operation is successful or not. The “TABRD” instruction should be executed to read the flash memory contents and to check if the contents is 0000h or not. If the flash memory page erase operation fails, users should go back to Step 2 and execute the page erase operation again.
4. Write data into the specific page. Refer to the “Flash Memory Write Procedure” for details.
5. Execute the “TABRD” instruction to read the flash memory contents and check if the written data is correct or not. If the data read from the flash memory is different from the written data, it means that the page write operation has failed. The CLWB bit should be set high to clear the write buffer and then write the data into the specific page again if the write operation has failed.
6. Clear the CFWEN bit to disable the Flash Memory Erase/Write function enable mode if the current page Erase and Write operations are completed and no more pages need to be erased or written.



Flash Memory Erase/Write Flow

Note: The Flash Memory Erase/Write Function Enable procedure and Flash Memory Write procedure will be described in the following sections.

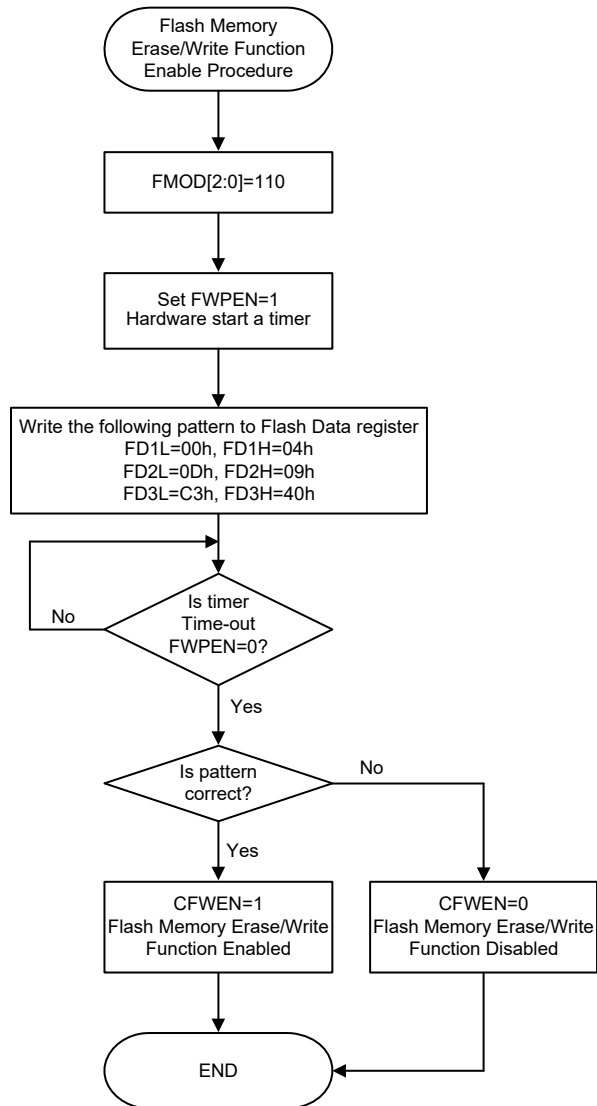
Flash Erase/Write Function Enable Procedure

The Flash Memory Erase/Write Function Enable Mode is specially designed to prevent the flash memory contents from being wrongly modified. In order to allow users to change the Flash memory data using the IAP control registers, users must first enable the Flash memory Erase/Write function.

Flash Memory Erase/Write Function Enable Procedure Description

1. Write data “110” to the FMODE [2:0] bits in the FC0 register to select the Flash Memory Erase/Write Function Enable Mode.
2. Set the FWPEN bit in the FC0 register to “1” to activate the Flash Memory Erase/Write Enable Function. This will also activate an internal timer.
3. Write the correct data pattern into the Flash data registers, FD1L~FD3L and FD1H~FD3H, as soon as possible after the FWPEN bit is set high. The enable Flash memory erase/write function data pattern is 00H, 0DH, C3H, 04H, 09H and 40H corresponding to the FD1L~FD3L and FD1H~FD3H registers respectively.
4. Once the timer has timed out, the FWPEN bit will automatically be cleared to zero by hardware regardless of the input data pattern.
5. If the written data pattern is incorrect, the Flash memory erase/write function will not be enabled successfully and the above steps should be repeated. If the written data pattern is correct, the Flash memory erase/write function will be enabled successfully.
6. Once the Flash memory erase/write function is enabled, the Flash memory contents can be updated by executing the page erase and write operations using the IAP control registers.

To disable the Flash memory erase/write function, the CFWEN bit in the FC0 register can be cleared. There is no need to execute the above procedure.



Flash Memory Erase/Write Function Enable Procedure

Flash Memory Write Procedure

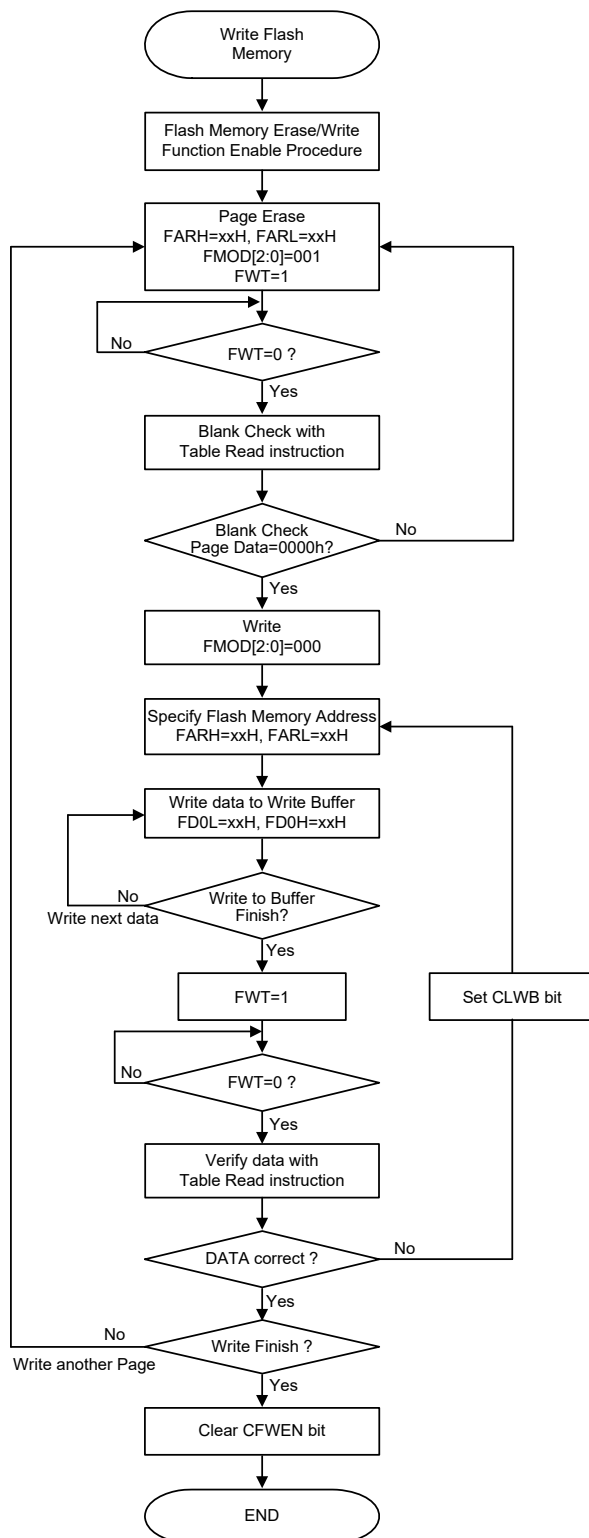
After the Flash memory erase/write function has been successfully enabled as the CFWEN bit is set high, the data to be written into the flash memory can be loaded into the write buffer. The selected flash memory page data should be erased by properly configuring the IAP control registers before the data write procedure is executed.

The write buffer size is 32 words, known as a page, whose address is mapped to a specific flash memory page specified by the memory address bits, FA12~FA5. It is important to ensure that the page where the write buffer data is located is the same one which the memory address bits, FA12~FA5, specify.

Flash Memory Consecutive Write Description

The maximum amount of write data is 32 words for each write operation. The write buffer address will be automatically incremented by one when consecutive write operations are executed. The start address of a specific page should first be written into the FARL and FARH registers. Then the data word should first be written into the FD0L register and then the FD0H register. At the same time the write buffer address will be incremented by one and then the next data word can be written into the FD0L and FD0H registers for the next address without modifying the address register pair, FARH and FARL. When the write buffer address reaches the page boundary the address will not be further incremented but will stop at the last address of the page.

1. Activate the “Flash Memory Erase/Write function enable procedure”. Check the CFWEN bit value and then execute the erase/write operations if the CFWEN bit is set high. Refer to the “Flash Memory Erase/Write function enable procedure” for more details.
2. Set the FMOD field to “001” to select the erase operation. Set the FWT bit high to erase the desired page which is specified by the FARH and FARL registers. Wait until the FWT bit goes low.
3. Execute a Blank Check operation using the table read instruction to ensure that the erase operation has successfully completed.
Go to step 2 if the erase operation is not successful.
Go to step 4 if the erase operation is successful.
4. Set the FMOD field to “000” to select the write operation.
5. Setup the desired start address in the FARH and FARL registers. Write the desired data words consecutively into the FD0L and FD0H registers within a page as specified by their consecutive addresses. The maximum written data number is 32 words.
6. Set the FWT bit high to write the data words from the write buffer to the flash memory. Wait until the FWT bit goes low.
7. Verify the data using the table read instruction to ensure that the write operation has successfully completed.
If the write operation has not successfully completed, set the CLWB bit high to clear the write buffer and then go to step 5.
Go to step 8 if the write operation is successful.
8. Clear the CFWEN bit low to disable the Flash memory erase/write function.



Flash Memory Consecutive Write Procedure

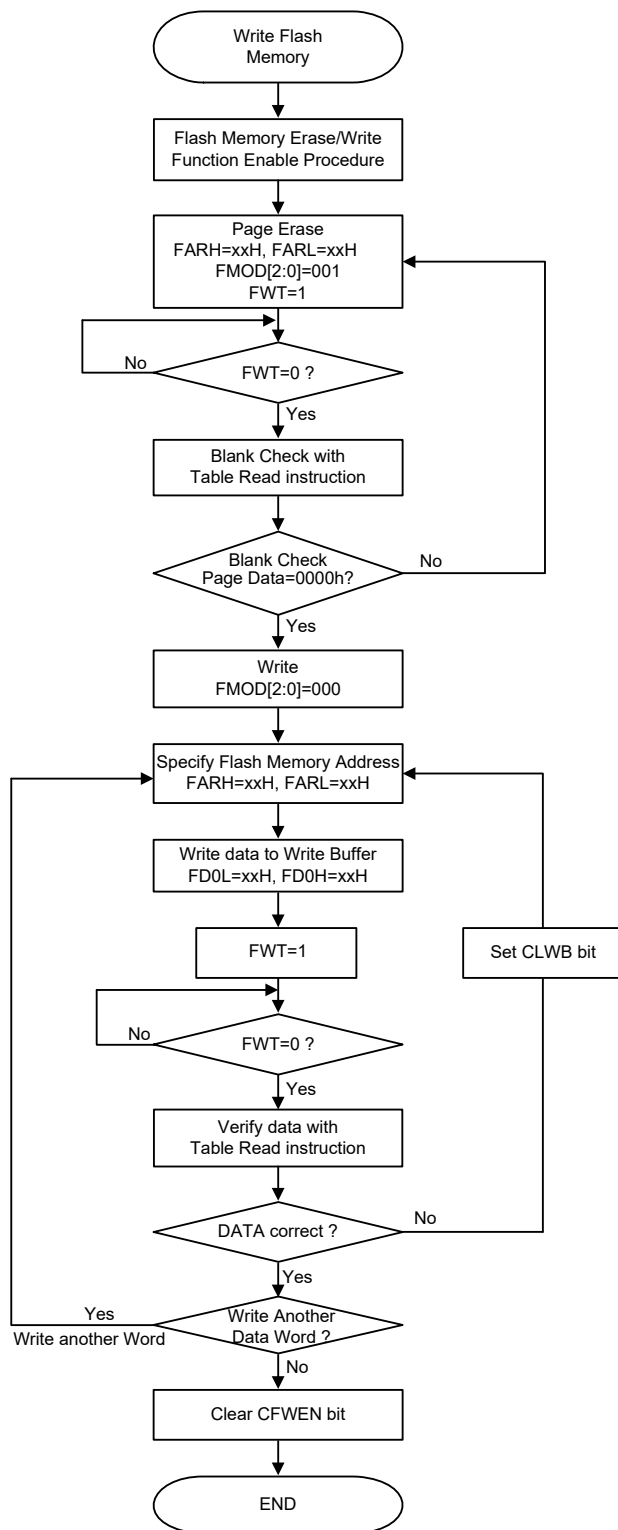
Note: 1. When the erase or write operation is successfully activated, all CPU operations will temporarily cease.
 2. It will take a typical time of 2.2ms for the FWT bit state changing from high to low.

Flash Memory Non-Consecutive Write Description

The main difference between Flash Memory Consecutive and Non-Consecutive Write operations is whether the data words to be written are located in consecutive addresses or not. If the data to be written is not located in consecutive addresses the desired address should be re-assigned after a data word is successfully written into the Flash Memory.

A two data word non-consecutive write operation is taken as an example here and described as follows:

1. Activate the “Flash Memory Erase/Write function enable procedure”. Check the CFWEN bit value and then execute the erase/write operation if the CFWEN bit is set high. Refer to the “Flash Memory Erase/Write function enable procedure” for more details.
2. Set the FMOD field to “001” to select the erase operation. Set the FWT bit high to erase the desired page which is specified by the FARH and FARL registers. Wait until the FWT bit goes low.
3. Execute a Blank Check operation using the table read instruction to ensure that the erase operation has successfully completed.
Go to step 2 if the erase operation is not successful.
Go to step 4 if the erase operation is successful.
Set the FMOD field to “000” to select the write operation.
5. Setup the desired address ADDR1 in the FARH and FARL registers. Write the desired data word DATA1 first into the FD0L register and then into the FD0H register.
6. Set the FWT bit high to transfer the data word from the write buffer to the flash memory. Wait until the FWT bit goes low.
7. Verify the data using the table read instruction to ensure that the write operation has successfully completed.
If the write operation has not successfully completed, set the CLWB bit high to clear the write buffer and then go to step 5.
Go to step 8 if the write operation is successful.
8. Setup the desired address ADDR2 in the FARH and FARL registers. Write the desired data word DATA2 first into the FD0L register and then into the FD0H register.
9. Set the FWT bit high to transfer the data word from the write buffer to the flash memory. Wait until the FWT bit goes low.
10. Verify the data using the table read instruction to ensure that the write operation has successfully completed.
If the write operation has not successfully completed, set the CLWB bit high to clear the write buffer and then go to step 8.
Go to step 11 if the write operation is successful.
11. Clear the CFWEN bit low to disable the Flash memory erase/write function.



Flash Memory Non-Consecutive Write Procedure

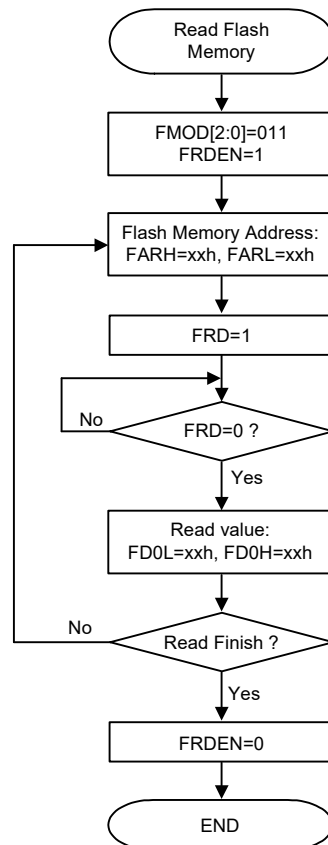
Note: 1. When the erase or write operation is successfully activated, all CPU operations will temporarily cease.
 2. It will take a typical time of 2.2ms for the FWT bit state changing from high to low.

Important Points to Note for Flash Memory Write Operations

1. The “Flash Memory Erase/Write Function Enable Procedure” must be successfully activated before the Flash Memory erase/write operation is executed.
2. The Flash Memory erase operation is executed to erase a whole page.
3. The whole write buffer data will be written into the flash memory in a page format. The corresponding address cannot exceed the page boundary.
4. After the data is written into the flash memory the flash memory contents must be read out using the table read instruction, TABRD, and checked if it is correct or not. If the data written into the flash memory is incorrect, the write buffer should be cleared by setting the CLWB bit high and then write the data again into the write buffer. Then activate a write operation on the same flash memory page without erasing it. The data check, buffer clear and data re-write steps should be repeatedly executed until the data written into the flash memory is correct.
5. The system frequency should be setup to the maximum application frequency when data write and data check operations are executed using the IAP function.

Flash Memory Read Procedure

To activate the Flash Memory Read procedure, the FMOD field should be set to “011” to select the flash memory read mode and the FRDEN bit should be set high to enable the read function. The desired flash memory address should be written into the FARH and FARL registers and then the FRD bit should be set high. After this the flash memory read operation will be activated. The data stored in the specified address can be read from the data registers, FD0H and FD0L, when the FRD bit goes low. There is no need to first activate the Flash Memory Erase/Write Function Enable Procedure before the flash memory read operation is executed.



Flash Memory Read Procedure

- Note: 1. When the read operation is successfully activated, all CPU operations will temporarily cease.
2. It will take a typical time of three instruction cycles for the FRD bit state changing from high to low.

Data Memory

The Data Memory is a volatile area of 8-bit wide RAM internal memory and is the location where temporary information is stored.

Categorized into two types, the first of these is an area of RAM where special function register are located. These registers have fixed locations and are necessary for correct operation of the device. Many of these registers can be read from and written to directly under program control, however, some remain protected from user manipulation. The second area of Data Memory is reserved for general purpose use. All locations within this area are read and write accessible under program control. All location within this area are read and write accessible under program control.

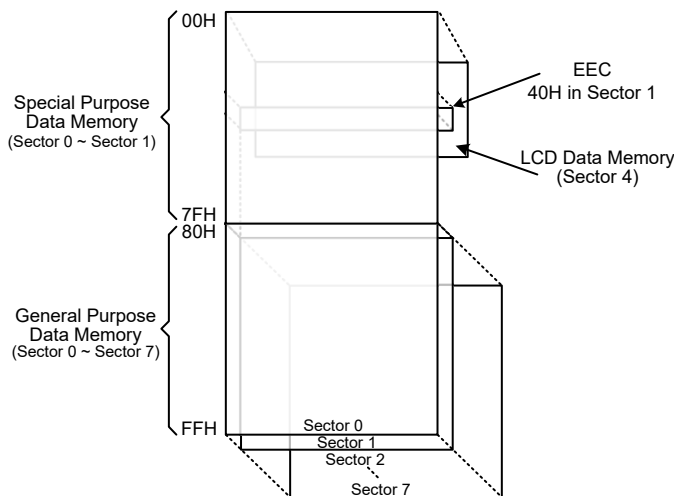
The device also provides a dedicated memory area for the LCD display data storage. In this chapter, only the General Purpose Data Memory and the Special Function Register Data Memory are introduced. Details about the LCD can be obtained in the LCD Driver section.

Structure

The Data Memory is subdivided into several sectors, all of which are implemented in 8-bit wide RAM. Switching between the different Data Memory sectors can be achieved by properly setting the Memory Pointers to correct value. The special purpose registers are accessible in Sector 0, with the exception of the EEC register at address 40H, which is only accessible in Sector 1. The address range of the Special Purpose Data Memory for the device is from 00H to 7FH while the General Purpose Data Memory address range is from 80H to FFH.

Special Purpose Data Memory	General Purpose Data Memory		LCD Data Memory
Located Sectors	Capacity	Sector: Address	Sector: Address
Sector 0: 00H~7FH Sector 1: 40H (EEC Only)	1024×8	Sector 0: 80H~FFH Sector 1: 80H~FFH : Sector 7: 80H~FFH	Sector 4: 00H~0FH

Data Memory Summary



Data Memory Structure

Data Memory Addressing

For this device that supports the extended instructions, there is no Bank Pointer for Data Memory addressing. For Data Memory the desired Sector is pointed by the MP1H or MP2H register and the certain Data Memory address in the selected sector is specified by the MP1L or MP2L register when using indirect addressing access.

Direct Addressing can be used in all sectors using the extended instructions which can address all available data memory space. For the accessed data memory which is located in any data memory sectors except sector 0, the extended instructions can be used to access the data memory instead of using the indirect addressing access. The main difference between standard instructions and extended instructions is that the data memory address “m” in the extended instructions has 11 valid bits for this device, the high byte indicates a sector and the low byte indicates a specific address.

General Purpose Data Memory

All microcontroller programs require an area of read/write memory where temporary data can be stored and retrieved for use later. It is this area of RAM memory that is known as General Purpose Data Memory. This area of Data Memory is fully accessible by the user programming for both reading and writing operations. By using the bit operation instructions individual bits can be set or reset under program control giving the user a large range of flexibility for bit manipulation in the Data Memory.

Special Purpose Data Memory

This area of Data Memory is where registers, necessary for the correct operation of the microcontroller, are stored. Most of the registers are both readable and writeable but some are protected and are readable only, the details of which are located under the relevant Special Function Register section. Note that for locations that are unused, any read instruction to these addresses will return the value “00H”.

Sector 0		Sector 1	Sector 0		Sector 1
00H	IAR0		40H	SIMC0	EEC
01H	MP0		41H	SIMC1/UUCR1	
02H	IAR1		42H	SIMD/UTXR_RXR	
03H	MP1L		43H	SIMC2/SIMA/UUCR2	
04H	MP1H		44H	UUCR3	
05H	ACC		45H	SIMTOC/UBRG	
06H	PCL		46H	UUSR	
07H	TBLP		47H	INTEG	
08H	TBLH		48H	INTC0	
09H	TBHP		49H	INTC1	
0AH	STATUS		4BH	INTC2	
0BH	VBGRC		4CH	INTC3	
0CH	IAR2		4DH	PTMC0	
0DH	MP2L		4EH	PTMC1	
0EH	MP2H		4FH		
0FH	RSTFC		50H	PTMDL	
10H	TB0C		51H	PTMDH	
11H	TB1C		52H	PTMAL	
12H	SCC		53H	PTMAH	
13H	HIRCC		54H		
14H	PA		55H		
15H	PAC		56H	PTMRPL	
16H	PAPU		57H	PTMRPH	
17H	PAWU		58H	FC0	
18H	PB		59H	FC1	
19H	PBC		5AH	FC2	
1AH	PBPU		5BH	FARL	
1BH	IFS0		5CH	FARH	
1CH	IFS1		5DH	FD0L	
1DH	PSCR		5EH	FD0H	
1EH	LVDC		5FH	FD1L	
1FH	REGC		60H	FD1H	
20H	PC		61H	FD2L	
21H	PCC		62H	FD2H	
22H	PCPU		63H	FD3L	
23H	PD		64H	FD3H	
24H	PDC		65H	OPSW0	
25H	PDPU		66H	OPSW1	
26H	STMC0		67H	OPPW	
27H	STMC1		68H	OPC	
28H	STMDL		69H	OPVOS	
29H	STMDH		6AH	OPPGAC0	
2AH	STMAL		6BH	OPPGAC1	
2BH	STMAH		6CH	LMSADOH	
2CH	SADOL		6DH	LMSADOL	
2DH	SADOH		6EH	ORMC	
2EH	SADC0		6FH	PE	
2FH	SADC1		70H	PEC	
30H	SADC2		71H	PEPU	
31H	LCDC0		72H	DAH	
32H	LCDC1		73H	DAL	
33H	PAS0		74H	DACC	
34H	PAS1		75H	SLEDC0	
35H	PBS0		76H	SLEDC1	
36H	PBS1		77H		
37H	PCS0		78H	USR	
38H	PCS1		79H	UCR1	
39H	PDS0		7AH	UCR2	
3AH	PDS1		7BH	UCR3	
3BH			7CH	TXR_RXR	
3CH			7DH	BRG	
3DH	WDTC		7EH	PMPs	
3EH	EEA		7FH		
3FH	EED				

□ : Unused, read as 00H

▤ : Reserved, cannot be changed, unless otherwise specified

Special Purpose Data Memory

Special Function Register Description

Most of the Special Function Register details will be described in the relevant functional section. However, several registers require a separate description in this section.

Indirect Addressing Registers – IAR0, IAR1, IAR2

The Indirect Addressing Registers, IAR0, IAR1 and IAR2, although having their locations in normal RAM register space, do not actually physically exist as normal registers. The method of indirect addressing for RAM data manipulation uses these Indirect Addressing Registers and Memory Pointers, in contrast to direct memory addressing, where the actual memory address is specified. Actions on the IAR0, IAR1 and IAR2 registers will result in no actual read or write operation to these registers but rather to the memory location specified by their corresponding Memory Pointers, MP0, MP1L/MP1H or MP2L/MP2H. Acting as a pair, IAR0 and MP0 can together access data only from Sector 0 while the IAR1 register together with the MP1L/MP1H register pair and IAR2 register together with the MP2L/MP2H register pair can access data from any Data Memory Sector. As the Indirect Addressing Registers are not physically implemented, reading the Indirect Addressing Registers will return a result of “00H” and writing to the registers will result in no operation.

Memory Pointers – MP0, MP1L/MP1H, MP2L/ MP2H

Five Memory Pointers, known as MP0, MP1L, MP1H, MP2L, MP2H, are provided. These Memory Pointers are physically implemented in the Data Memory and can be manipulated in the same way as normal registers providing a convenient way with which to address and track data. When any operation to the relevant Indirect Addressing Registers is carried out, the actual address that the microcontroller is directed to is the address specified by the related Memory Pointer. MP0, together with Indirect Addressing Register, IAR0, are used to access data from Sector 0, while MP1L/MP1H together with IAR1 and MP2L/MP2H together with IAR2 are used to access data from all sectors according to the corresponding MP1H or MP2H register. Direct Addressing can be used in all sectors using the extended instructions which can address all available data memory space.

The following example shows how to clear a section of four Data Memory locations already defined as locations adres1 to adres4.

Indirect Addressing Program Example 1

```
data .section 'data'
adres1 db ?
adres2 db ?
adres3 db ?
adres4 db ?
block db?
code .section at 0 'code'
org 00h
start:
    mov a, 04h          ; setup size of block
    mov block, a
    mov a, offset adres1 ; Accumulator loaded with first RAM address
    mov mp0, a          ; setup memory pointer with first RAM address
loop:
    clr IAR0            ; clear the data at address defined by MP0
    inc mp0             ; increase memory pointer
    sdz block           ; check if last memory location has been cleared
    jmp loop
continue:
```

Indirect Addressing Program Example 2

```
data .section 'data'
adres1 db ?
adres2 db ?
adres3 db ?
adres4 db ?
block db ?
code .section at 0 'code'
org 00h
start:
    mov a, 04h          ; setup size of block
    mov block, a
    mov a, 01h          ; setup the memory sector
    mov mplh, a
    mov a, offset adres1 ; Accumulator loaded with first RAM address
    mov mp1l, a          ; setup memory pointer with first RAM address
loop:
    clr IAR1            ; clear the data at address defined by MP1L
    inc mp1l             ; increase memory pointer MP1L
    sdz block            ; check if last memory location has been cleared
    jmp loop
continue:
```

The important point to note here is that in the examples shown above, no reference is made to specific Data Memory addresses.

Direct Addressing Program Example using extended instructions

```
data .section 'data'
temp db ?
code .section at 0 'code'
org 00h
start:
    lmov a, [m]          ; move [m] data to acc
    lsub a, [m+1]         ; compare [m] and [m+1] data
    snz c                ; [m]>[m+1]?
    jmp continue         ; no
    lmov a, [m]           ; yes, exchange [m] and [m+1] data
    mov temp, a
    lmov a, [m+1]
    lmov [m], a
    mov a, temp
    lmov [m+1], a
continue:
```

Note: Here “m” is a data memory address located in any data memory sectors. For example, m=1F0H, it indicates address F0H in Sector 1.

Accumulator – ACC

The Accumulator is central to the operation of any microcontroller and is closely related with operations carried out by the ALU. The Accumulator is the place where all intermediate results from the ALU are stored. Without the Accumulator it would be necessary to write the result of each calculation or logical operation such as addition, subtraction, shift, etc., to the Data Memory resulting in higher programming and timing overheads. Data transfer operations usually involve the temporary storage function of the Accumulator; for example, when transferring data between one user-defined register and another, it is necessary to do this by passing the data through the Accumulator as no direct transfer between two registers is permitted.

Program Counter Low Byte Register – PCL

To provide additional program control functions, the low byte of the Program Counter is made accessible to programmers by locating it within the Special Purpose area of the Data Memory. By manipulating this register, direct jumps to other program locations are easily implemented. Loading a value directly into this PCL register will cause a jump to the specified Program Memory location, however, as the register is only 8-bit wide, only jumps within the current Program Memory page are permitted. When such operations are used, note that a dummy cycle will be inserted.

Look-up Table Registers – TBLP, TBHP, TBLH

These three special function registers are used to control operation of the look-up table which is stored in the Program Memory. TBLP and TBHP are the table pointers and indicate the location where the table data is located. Their value must be setup before any table read commands are executed. Their value can be changed, for example using the “INC” or “DEC” instructions, allowing for easy table data pointing and reading. TBLH is the location where the high order byte of the table data is stored after a table read data instruction has been executed. Note that the lower order table data byte is transferred to a user defined location.

Option Memory Mapping Register – ORMC

The ORMC register is used to enable Option Memory Mapping function. The Option Memory capacity is 32 words. When a specific pattern of 55H and AAH is consecutively written into this register, the Option Memory Mapping function will be enabled and then the Option Memory code can be read by using the table read instruction. The Option Memory addresses 00H~1FH will be mapped to Program Memory last page addresses E0H~FFH.

To successfully enable the Option Memory Mapping function, the specific pattern of 55H and AAH must be written into the ORMC register in two consecutive instruction cycles. It is therefore recommended that the global interrupt bit EMI should first be cleared before writing the specific pattern, and then set high again at a proper time according to users’ requirements after the pattern is successfully written. An internal timer will be activated when the pattern is successfully written. The mapping operation will be automatically finished after a period of $4 \times t_{LIRC}$. Therefore, users should read the data in time, otherwise the Option Memory Mapping function needs to be restarted. After the completion of each consecutive write operation to the ORMC register, the timer will recount.

When the table read instructions are used to read the Option Memory code, both “TABRD [m]” and “TABRDL [m]” instructions can be used. However, care must be taken if the “TABRD [m]” instruction is used, the table pointer defined by the TBHP register must be referenced to the last page. Refer to corresponding sections about the table read instruction for more details.

• ORMC Register

Bit	7	6	5	4	3	2	1	0
Name	ORMC7	ORMC6	ORMC5	ORMC4	ORMC3	ORMC2	ORMC1	ORMC0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0

ORMC7~ORMC0: Option Memory Mapping specific pattern

When a specific pattern of 55H and AAH is written into this register, the Option Memory Mapping function will be enabled.

Status Register – STATUS

This 8-bit register contains the SC flag, CZ flag, zero flag (Z), carry flag (C), auxiliary carry flag (AC), overflow flag (OV), power down flag (PDF), and watchdog time-out flag (TO). These arithmetic/logical operation and system management flags are used to record the status and operation of the microcontroller.

With the exception of the TO and PDF flags, bits in the status register can be altered by instructions like most other registers. Any data written into the status register will not change the TO or PDF flag. In addition, operations related to the status register may give different results due to the different instruction operations. The TO flag can be affected only by a system power-up, a WDT time-out or by executing the “CLR WDT” or “HALT” instruction. The PDF flag is affected only by executing the “HALT” or “CLR WDT” instruction or during a system power-up.

The Z, OV, AC, C, SC and CZ flags generally reflect the status of the latest operations.

- C is set if an operation results in a carry during an addition operation or if a borrow does not take place during a subtraction operation; otherwise C is cleared. C is also affected by a rotate through carry instruction.
- AC is set if an operation results in a carry out of the low nibbles in addition, or no borrow from the high nibble into the low nibble in subtraction; otherwise AC is cleared.
- Z is set if the result of an arithmetic or logical operation is zero; otherwise Z is cleared.
- OV is set if an operation results in a carry into the highest-order bit but not a carry out of the highest-order bit, or vice versa; otherwise OV is cleared.
- PDF is cleared by a system power-up or executing the “CLR WDT” instruction. PDF is set by executing the “HALT” instruction.
- TO is cleared by a system power-up or executing the “CLR WDT” or “HALT” instruction. TO is set by a WDT time-out.
- CZ is the operational result of different flags for different instructions. Refer to register definitions for more details.
- SC is the result of the “XOR” operation which is performed by the OV flag and the MSB of the current instruction operation result.

In addition, on entering an interrupt sequence or executing a subroutine call, the status register will not be pushed onto the stack automatically. If the contents of the status registers are important and if the subroutine can corrupt the status register, precautions must be taken to correctly save it.

• STATUS Register

Bit	7	6	5	4	3	2	1	0
Name	SC	CZ	TO	PDF	OV	Z	AC	C
R/W	R/W	R/W	R	R	R/W	R/W	R/W	R/W
POR	x	x	0	0	x	x	x	x

“x”: unknown

Bit 7 **SC:** The result of the “XOR” operation which is performed by the OV flag and the MSB of the instruction operation result

Bit 6 **CZ:** The operational result of different flags for different instructions
For SUB/SUBM/LSUB/LSUBM instructions, the CZ flag is equal to the Z flag.
For SBC/SBCM/LSBC/LSBCM instructions, the CZ flag is the “AND” operation result which is performed by the previous operation CZ flag and current operation zero flag.
For other instructions, the CZ flag will not be affected.

Bit 5	TO: Watchdog Time-out flag 0: After power up or executing the “CLR WDT” or “HALT” instruction 1: A watchdog time-out occurred
Bit 4	PDF: Power down flag 0: After power up or executing the “CLR WDT” instruction 1: By executing the “HALT” instruction
Bit 3	OV: Overflow flag 0: No overflow 1: An operation results in a carry into the highest-order bit but not a carry out of the highest-order bit or vice versa
Bit 2	Z: Zero flag 0: The result of an arithmetic or logical operation is not zero 1: The result of an arithmetic or logical operation is zero
Bit 1	AC: Auxiliary flag 0: No auxiliary carry 1: An operation results in a carry out of the low nibbles in addition, or no borrow from the high nibble into the low nibble in subtraction
Bit 0	C: Carry flag 0: No carry-out 1: An operation results in a carry during an addition operation or if a borrow does not take place during a subtraction operation The “C” flag is also affected by a rotate through carry instruction.

EEPROM Data Memory

This device contains an area of internal EEPROM Data Memory. EEPROM is by its nature a non-volatile form of re-programmable memory, with data retention even when its power supply is removed. By incorporating this kind of memory, a whole new host of application possibilities are made available to the designer. The availability of EEPROM storage allows information such as product identification numbers, calibration values, specific user data, system setup data or other product information to be stored directly within the product microcontroller. The process of reading and writing data to the EEPROM memory has been reduced to a very trivial affair.

EEPROM Data Memory Structure

The EEPROM Data Memory capacity is 256×8 bits for the device. Unlike the Program Memory and RAM Data Memory, the EEPROM Data Memory is not directly mapped into memory space and is therefore not directly addressable in the same way as the other types of memory. Read and Write operations to the EEPROM are carried out in single byte operations using an address and a data register in Sector 0 and a single control register in Sector 1.

EEPROM Registers

Three registers control the overall operation of the internal EEPROM Data Memory. These are the address register, EEA, the data register, EED and a single control register, EEC. As both the EEA and EED registers are located in Sector 0, they can be directly accessed in the same way as any other Special Function Register. The EEC register however, being located in only Sector 1, can only be read from or written to indirectly using the MP1L/MP1H or MP2L/MP2H Memory Pointer and Indirect Addressing Register, IAR1/IAR2. Because the EEC control register is located at address 40H in Sector 1, the MP1L or MP2L Memory Pointer must first be set to the value 40H and the MP1H or MP2H Memory Pointer high byte set to the value, 01H, before any operations on the EEC register is executed.

Register Name	Bit							
	7	6	5	4	3	2	1	0
EEA	EEA7	EEA6	EEA5	EEA4	EEA3	EEA2	EEA1	EEA0
EED	D7	D6	D5	D4	D3	D2	D1	D0
EEC	—	—	—	—	WREN	WR	RDEN	RD

EEPROM Register List

• **EEA Register**

Bit	7	6	5	4	3	2	1	0
Name	EEA7	EEA6	EEA5	EEA4	EEA3	EEA2	EEA1	EEA0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **EEA7~EEA0**: Data EEPROM address bit 7 ~ bit 0

• **EED Register**

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D7~D0**: Data EEPROM data bit 7 ~ bit 0

• **EEC Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	WREN	WR	RDEN	RD
R/W	—	—	—	—	R/W	R/W	R/W	R/W
POR	—	—	—	—	0	0	0	0

Bit 7~4 Unimplemented, read as “0”

Bit 3 **WREN**: Data EEPROM Write Enable
0: Disable
1: Enable

This is the Data EEPROM Write Enable Bit which must be set high before Data EEPROM write operations are carried out. Clearing this bit to zero will inhibit Data EEPROM write operations.

Bit 2 **WR**: EEPROM Write Control
0: Write cycle has finished
1: Activate a write cycle

This is the Data EEPROM Write Control Bit and when set high by the application program will activate a write cycle. This bit will be automatically reset to zero by the hardware after the write cycle has finished. Setting this bit high will have no effect if the WREN has not first been set high.

Bit 1 **RDEN**: Data EEPROM Read Enable
0: Disable
1: Enable

This is the Data EEPROM Read Enable Bit which must be set high before Data EEPROM read operations are carried out. Clearing this bit to zero will inhibit Data EEPROM read operations.

Bit 0 **RD:** EEPROM Read Control
 0: Read cycle has finished
 1: Activate a read cycle
This is the Data EEPROM Read Control Bit and when set high by the application program will activate a read cycle. This bit will be automatically reset to zero by the hardware after the read cycle has finished. Setting this bit high will have no effect if the RDEN has not first been set high.

- Note: 1. The WREN, WR, RDEN and RD bits cannot be set high at the same time in one instruction.
2. Ensure that the f_{SUB} clock is stable before executing the write operation.
3. Ensure that the write operation is totally complete before changing the contents of the EEPROM related registers.

Reading Data from the EEPROM

To read data from the EEPROM, the EEPROM address of the data to be read must first be placed in the EEA register. The read enable bit, RDEN, in the EEC register must then be set high to enable the read function. If the RD bit in the EEC register is now set high, a read cycle will be initiated. Setting the RD bit high will not initiate a read operation if the RDEN bit has not been set. When the read cycle terminates, the RD bit will be automatically cleared to zero, after which the data can be read from the EED register. The data will remain in the EED register until another read or write operation is executed. The application program can poll the RD bit to determine when the data is valid for reading.

Writing Data to the EEPROM

To write data to the EEPROM, the EEPROM address of the data to be written must first be placed in the EEA register and the data placed in the EED register. To initiate a write cycle the write enable bit, WREN, in the EEC register must first be set high to enable the write function. After this, the WR bit in the EEC register must be immediately set high to initiate a write cycle successfully. These two instructions must be executed in two consecutive instruction cycles. The global interrupt bit EMI should also first be cleared before implementing any write operations, and then set high again after the write cycle has started. Note that setting the WR bit high will not initiate a write cycle if the WREN bit has not been set. As the EEPROM write cycle is controlled using an internal timer whose operation is asynchronous to microcontroller system clock, a certain time will elapse before the data will have been written into the EEPROM. Detecting when the write cycle has finished can be implemented either by polling the WR bit in the EEC register or by using the EEPROM interrupt. When the write cycle terminates, the WR bit will be automatically cleared to zero by the microcontroller, informing the user that the data has been written to the EEPROM. The application program can therefore poll the WR bit to determine when the write cycle has ended.

Write Protection

Protection against inadvertent write operation is provided in several ways. After the device is powered-on the Write Enable bit in the control register will be cleared preventing any write operations. Also at power-on the Memory Pointer high byte register, MP1H or MP2H, will be reset to zero, which means that Data Memory Sector 0 will be selected. As the EEPROM control register EEC is located in Sector 1, this adds a further measure of protection against spurious write operations. During normal program operation, ensuring that the Write Enable bit in the control register is cleared will safeguard against incorrect write operations.

EEPROM Interrupt

The EEPROM write interrupt is generated when an EEPROM write cycle has ended. The EEPROM interrupt must first be enabled by setting the DEE bit in the relevant interrupt register. When an EEPROM write cycle ends, the DEF request flag will be set. If the global and EEPROM interrupts are enabled and the stack is not full, a jump to the Data EEPROM Interrupt vector will take place. When the interrupt is serviced, the EEPROM Interrupt flag will be automatically reset. More details can be obtained in the Interrupts section.

Programming Considerations

Care must be taken that data is not inadvertently written to the EEPROM. Protection can be enhanced by ensuring that the Write Enable bit is normally cleared to zero when not writing. Also the Memory Pointer high byte register, MP1H or MP2H, could be normally cleared to zero as this would inhibit access to Sector 1 where the EEPROM control register exists. Although certainly not necessary, consideration might be given in the application program to the checking of the validity of new write data by a simple read back process.

When writing data the WR bit must be set high immediately after the WREN bit has been set high, to ensure the write cycle executes correctly. The global interrupt bit EMI should also be cleared before a write cycle is executed and then set high again after a valid write activation procedure has completed. Note that the device should not enter the IDLE or SLEEP mode until the EEPROM read or write operation is totally complete. Otherwise, the EEPROM read or write operation will fail.

Programming Examples

Reading data from the EEPROM – polling method

```
MOV A, EEPROM_ADRES ; user defined address
MOV EEA, A
MOV A, 040H          ; set memory pointer MP1L
MOV MP1L, A          ; MP1L points to EEC register
MOV A, 01H           ; set memory pointer MP1H
MOV MP1H, A
SET IAR1.1           ; set RDEN bit, enable read operations
SET IAR1.0           ; start Read Cycle - set RD bit
BACK:
SZ IAR1.0            ; check for read cycle end
JMP BACK
CLR IAR1             ; disable EEPROM read if no more read operations are required
CLR MP1H
MOV A, EED           ; move read data to register
MOV READ_DATA, A
```

Note: For each read operation, the address register should be re-specified followed by setting the RD bit high to activate a read cycle even if the target address is consecutive.

Writing Data to the EEPROM – polling method

```
MOV A, EEPROM_ADRES ; user defined address
MOV EEA, A
MOV A, EEPROM_DATA   ; user defined data
MOV EED, A
MOV A, 040H          ; set memory pointer MP1L
MOV MP1L, A          ; MP1L points to EEC register
MOV A, 01H           ; set memory pointer MP1H
MOV MP1H, A
CLR EMI
SET IAR1.3           ; set WREN bit, enable write operations
```

```
SET  IAR1.2          ; start Write Cycle - set WR bit - executed immediately
                                ; after setting WREN bit

SET  EMI
BACK:
SZ   IAR1.2          ; check for write cycle end
JMP  BACK
CLR  MP1H
```

Oscillators

Various oscillator types offer the user a wide range of functions according to their various application requirements. The flexible features of the oscillator functions ensure that the best optimisation can be achieved in terms of speed and power saving. Oscillator selections and operation are selected through the combination of configuration option and relevant control registers.

Oscillator Overview

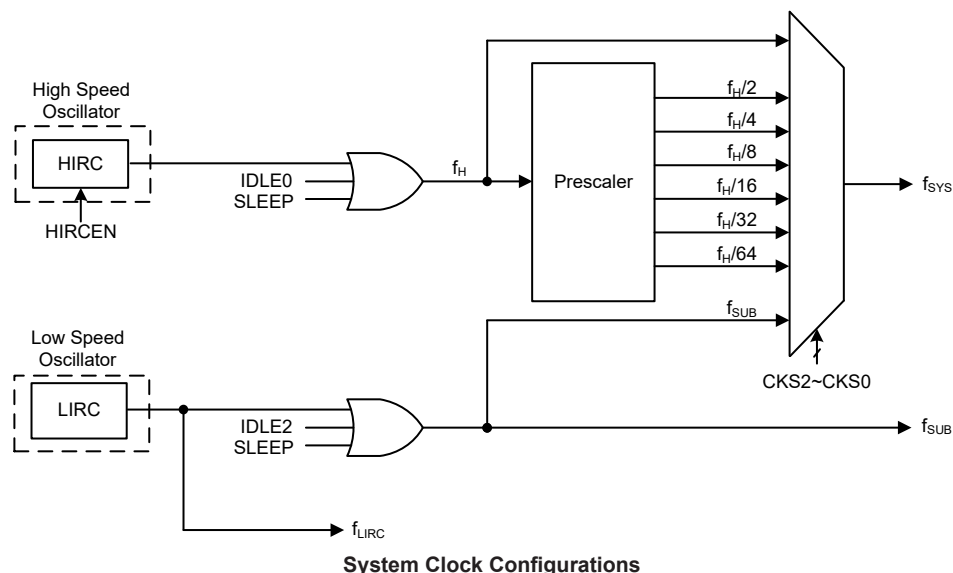
In addition to being the source of the main system clock, the oscillators also provide clock sources for the Watchdog Timer and Time Base Interrupts. Two fully integrated internal oscillators, requiring no external components, are provided to form a wide range of both fast and slow system oscillators. The higher frequency oscillator provides higher performance but carry with it the disadvantage of higher power requirements, while the opposite is of course true for the lower frequency oscillators. With the capability of dynamically switching between fast and slow system clock, the device has the flexibility to optimize the performance/power ratio, a feature especially important in power sensitive portable applications.

Type	Name	Frequency
Internal High Speed RC	HIRC	2/4/8MHz
Internal Low Speed RC	LIRC	32kHz

Oscillator Types

System Clock Configurations

There are two oscillator sources, a high speed oscillator and a low speed oscillator. The high speed system clock is sourced from the internal 2/4/8MHz RC oscillator, HIRC. The low speed oscillator is the internal 32kHz RC oscillator, LIRC. Selecting whether the low or high speed oscillator is used as the system oscillator is implemented using the CKS2~CKS0 bits in the SCC register and as the system clock can be dynamically selected.



Internal High Speed RC Oscillator – HIRC

The internal RC oscillator is a fully integrated system oscillator requiring no external components. The internal RC oscillator has three fixed frequencies of 2MHz, 4MHz and 8MHz, which are selected by the HIRC1~HIRC0 bits in the HIRCC register. These bits must also be setup to match the selected configuration option frequency to ensure that the HIRC frequency accuracy specified in the A.C. Characteristics is achieved. Device trimming during the manufacturing process and the inclusion of internal frequency compensation circuits are used to ensure that the influence of the power supply voltage, temperature and process variations on the oscillation frequency are minimised.

Internal 32kHz Oscillator – LIRC

The Internal 32kHz oscillator is a fully integrated RC oscillator with a typical frequency of 32kHz at full voltage range, requiring no external components for its implementation. Device trimming during the manufacturing process and the inclusion of internal frequency compensation circuits are used to ensure that the influence of the power supply voltage, temperature and process variations on the oscillation frequency are minimised.

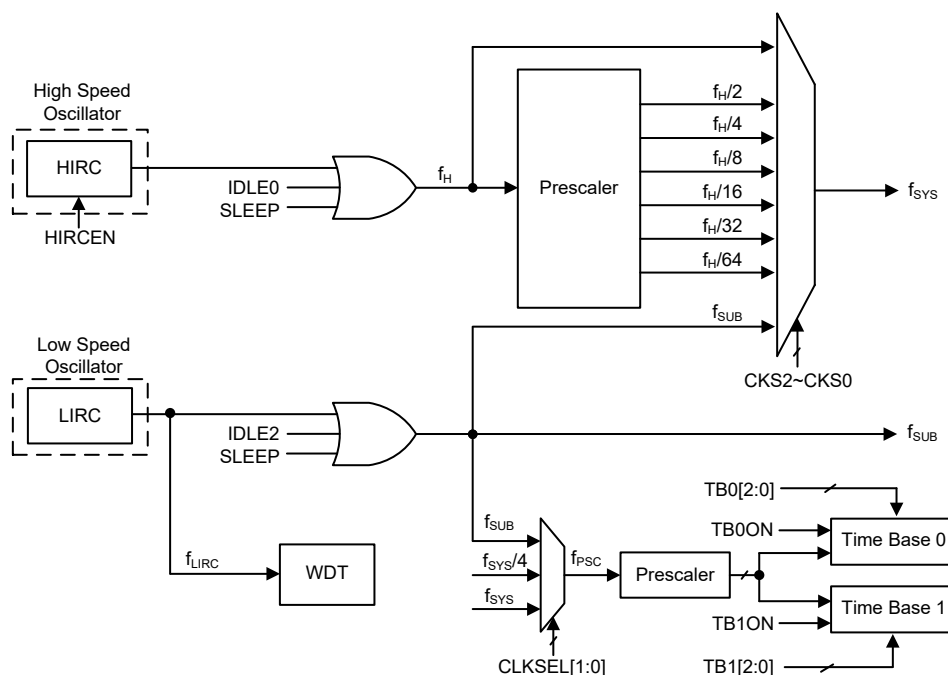
Operating Modes and System Clocks

Present day applications require that their microcontrollers have high performance but often still demand that they consume as little power as possible, conflicting requirements that are especially true in battery powered portable applications. The fast clocks required for high performance will by their nature increase current consumption and of course vice versa, lower speed clocks reduce current consumption. As Holtek has provided the device with both high and low speed clock sources and the means to switch between them dynamically, the user can optimise the operation of their microcontroller to achieve the best performance/power ratio.

System Clocks

The device has many different clock sources for both the CPU and peripheral function operation. By providing the user with a wide range of clock options using register programming, a clock system can be configured to obtain maximum application performance.

The main system clock, can come from a high frequency, f_H , or low frequency, f_{SUB} , source, and is selected using the CKS2~CKS0 bits in the SCC register. The high speed system clock is sourced from the HIRC oscillator. The low speed system clock source is sourced from the LIRC oscillator. The other choice, which is a divided version of the high speed system oscillator has a range of $f_H/2 \sim f_H/64$.



Device Clock Configurations

Note: When the system clock source f_{SYS} is switched to f_{SUB} from f_H , the high speed oscillator can be stopped to conserve the power or continue to oscillate to provide the clock source, $f_H \sim f_H/64$, for peripheral circuits to use, which is determined by configuring the corresponding high speed oscillator enable control bit.

System Operation Modes

There are six different modes of operation for the microcontroller, each one with its own special characteristics and which can be chosen according to the specific performance and power requirements of the application. There are two modes allowing normal operation of the microcontroller, the FAST Mode and SLOW Mode. The remaining four modes, the SLEEP, IDLE0, IDLE1 and IDLE2 Modes are used when the microcontroller CPU is switched off to conserve power.

Operation Mode	CPU	Register Setting			f_{SYS}	f_H	f_{SUB}	f_{LIRC}
		FHIDEN	FSIDEN	CKS2~CKS0				
FAST	On	x	x	000~110	$f_H \sim f_H/64$	On	On	On
SLOW	On	x	x	111	f_{SUB}	On/Off ⁽¹⁾	On	On
IDLE0	Off	0	1	000~110	Off	Off	On	On
				111	On			
IDLE1	Off	1	1	xxx	On	On	On	On
IDLE2	Off	1	0	000~110	On	On	Off	On
				111	Off			
SLEEP	Off	0	0	xxx	Off	Off	Off	On ⁽²⁾

"x": Don't care

Note: 1. The f_H clock will be switched on or off by configuring the corresponding oscillator enable bit in the SLOW mode.

2. In the SLEEP mode, the f_{LIRC} clock is still on as the WDT function is always enabled.

FAST Mode

This is one of the main operating modes where the microcontroller has all of its functions operational and where the system clock is provided by the internal high speed oscillator. This mode operates allowing the microcontroller to operate normally with a clock source coming from the HIRC oscillator. The high speed oscillator will however first be divided by a ratio ranging from 1 to 64, the actual ratio being selected by the CKS2~CKS0 bits in the SCC register. Although a high speed oscillator is used, running the microcontroller at a divided clock ratio reduces the operating current.

SLOW Mode

This is also a mode where the microcontroller operates normally although now with a slower speed clock source. The clock source used will be from f_{SUB} . The f_{SUB} clock is derived from the LIRC oscillator.

SLEEP Mode

The SLEEP Mode is entered when a HALT instruction is executed and when the FHIDEN and FSIDEN bits in the SCC register are both low. In the SLEEP mode the CPU will be stopped. The f_{SUB} clock provided to the peripheral function will also be stopped, too. However the f_{LIRC} clock continues to operate since the WDT function is always enabled.

IDLE0 Mode

The IDLE0 Mode is entered when a HALT instruction is executed and when the FHIDEN bit in the SCC register is low and the FSIDEN bit in the SCC register is high. In the IDLE0 Mode the CPU will be switched off but the low speed oscillator will be on to drive some peripheral functions.

IDLE1 Mode

The IDLE1 Mode is entered when a HALT instruction is executed and when the FHIDEN and FSIDEN bits in the SCC register are both high. In the IDLE1 Mode the CPU will be switched off but both the high and low speed oscillators will be on to provide a clock source to keep some peripheral functions operational.

IDLE2 Mode

The IDLE2 Mode is entered when a HALT instruction is executed and when the FHIDEN bit in the SCC register is high and the FSIDEN bit in the SCC register is low. In the IDLE2 Mode the CPU will be switched off but the high speed oscillator will be on to provide a clock source to keep some peripheral functions operational.

Control Registers

The registers, SCC and HIRCC, are used to control the system clock and the corresponding oscillator configurations.

Register Name	Bit							
	7	6	5	4	3	2	1	0
SCC	CKS2	CKS1	CKS0	—	—	—	FHIDEN	FSIDEN
HIRCC	—	—	—	—	HIRC1	HIRC0	HIRCF	HIRCEN

System Operating Mode Control Register List

• SCC Register

Bit	7	6	5	4	3	2	1	0
Name	CKS2	CKS1	CKS0	—	—	—	FHIDEN	FSIDEN
R/W	R/W	R/W	R/W	—	—	—	R/W	R/W
POR	0	0	0	—	—	—	0	0

Bit 7~5 **CKS2~CKS0**: System clock selection

000: f_H
 001: $f_H/2$
 010: $f_H/4$
 011: $f_H/8$
 100: $f_H/16$
 101: $f_H/32$
 110: $f_H/64$
 111: f_{SUB}

These three bits are used to select which clock is used as the system clock source. In addition to the system clock source directly derived from f_H or f_{SUB} , a divided version of the high speed system oscillator can also be chosen as the system clock source.

Bit 4~2 Unimplemented, read as 0.

Bit 1 **FHIDEN**: High Frequency oscillator control when CPU is switched off

0: Disable
 1: Enable

This bit is used to control whether the high speed oscillator is activated or stopped when the CPU is switched off by executing a “HALT” instruction.

Bit 0 **FSIDEN**: Low Frequency oscillator control when CPU is switched off

0: Disable
 1: Enable

This bit is used to control whether the low speed oscillator is activated or stopped when the CPU is switched off by executing a “HALT” instruction.

Note: A certain delay is required before the relevant clock is successfully switched to the target clock source after any clock switching setup using the CKS2~CKS0 bits. A proper delay time must be arranged before executing the following operations which require immediate reaction with the target clock source.

Clock switching delay time = $4 \times t_{SYS} + [0 \sim (1.5 \times t_{curr.} + 0.5 \times t_{tar.})]$, where $t_{curr.}$ indicates the current clock period, $t_{tar.}$ indicates the target clock period and t_{SYS} indicates the current system clock period.

• **HIRCC Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	HIRC1	HIRC0	HIRCF	HIRCEN
R/W	—	—	—	—	R/W	R/W	R	R/W
POR	—	—	—	—	0	0	0	1

Bit 7~4 Unimplemented, read as “0”

Bit 3~2 **HIRC1~HIRC0**: HIRC frequency selection

00: 2MHz

01: 4MHz

10: 8MHz

11: 2MHz

When the HIRC oscillator is enabled or the HIRC frequency selection is changed by application program, the clock frequency will automatically be changed after the HIRCF flag is set to 1. It is recommended that the HIRC frequency selected by these bits is the same as the frequency determined by the configuration option to keep the HIRC frequency accuracy specified in the A.C. characteristics.

Bit 1 **HIRCF**: HIRC oscillator stable flag

0: HIRC unstable

1: HIRC stable

This bit is used to indicate whether the HIRC oscillator is stable or not. When the HIRCEN bit is set to 1 to enable the HIRC oscillator, the HIRCF bit will first be cleared to 0 and then set to 1 after the HIRC oscillator is stable.

Bit 0 **HIRCEN**: HIRC oscillator enable control

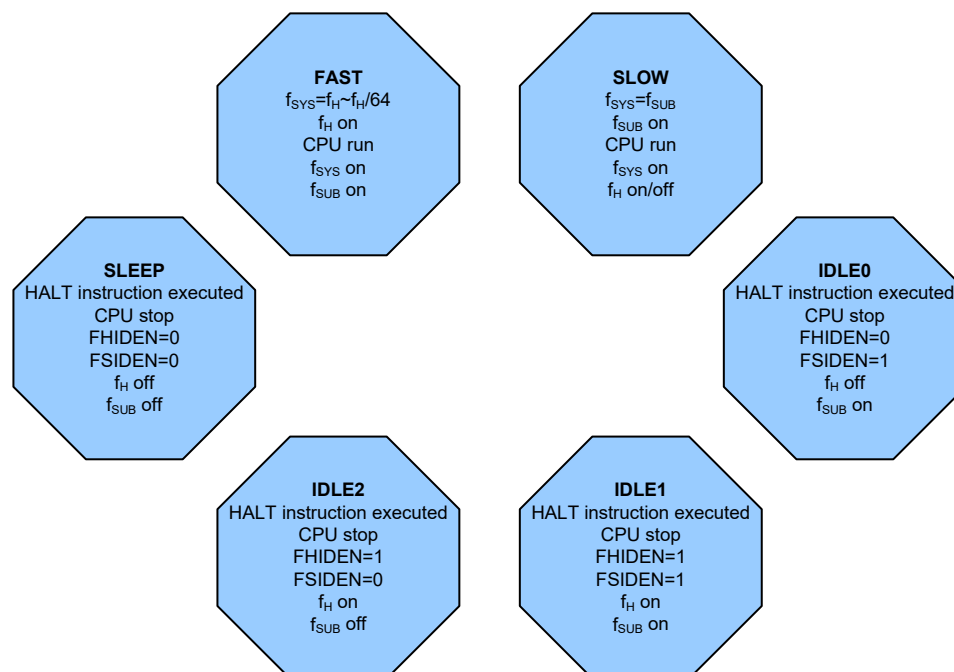
0: Disable

1: Enable

Operating Mode Switching

The device can switch between operating modes dynamically allowing the user to select the best performance/power ratio for the present task in hand. In this way microcontroller operations that do not require high performance can be executed using slower clocks thus requiring less operating current and prolonging battery life in portable applications.

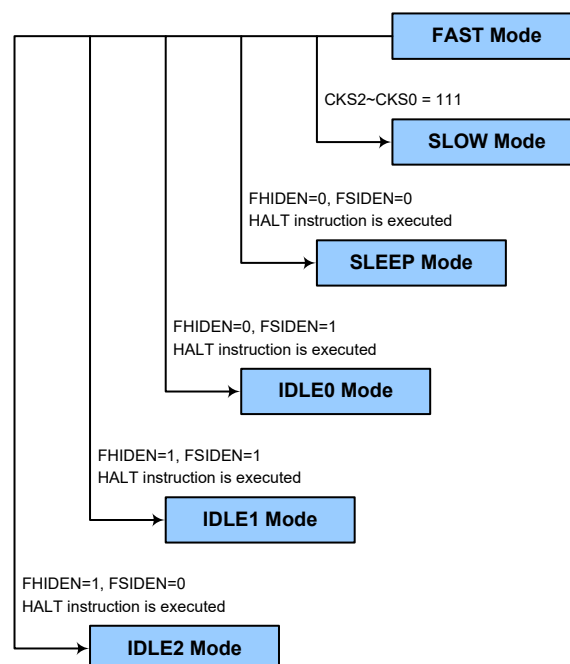
In simple terms, mode switching between the FAST mode and SLOW mode is executed using the CKS2~CKS0 bits in the SCC register while Mode Switching from the FAST/SLOW modes to the SLEEP/IDLE modes is executed via the HALT instruction. When a HALT instruction is executed, whether the device enters the IDLE mode or the SLEEP mode is determined by the condition of the FHIDEN and FSIDEN bits in the SCC register.



FAST Mode to SLOW Mode Switching

When running in the FAST Mode, which uses the high speed system oscillator, and therefore consumes more power, the system clock can switch to run in the SLOW Mode by setting the CKS2~CKS0 bits to “111” in the SCC register. This will then use the low speed system oscillator and consume less power. Users may decide to do this for certain operations which do not require high performance and can subsequently reduce power consumption.

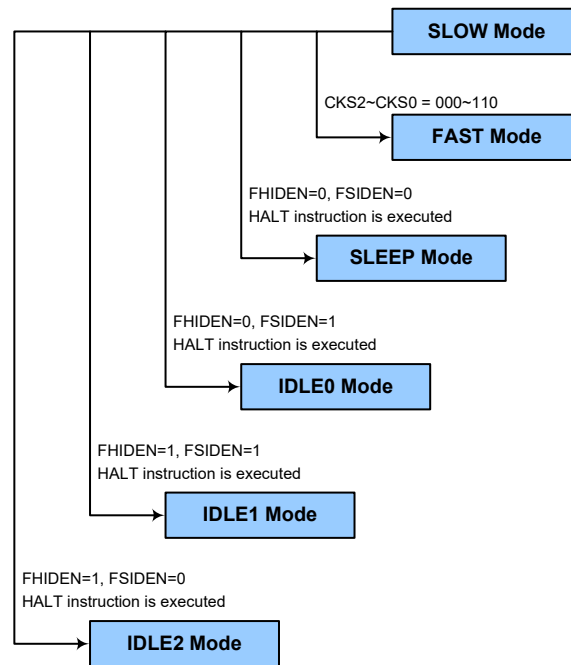
The SLOW Mode is sourced from the LIRC oscillator and therefore requires this oscillator to be stable before full mode switching occurs.



SLOW Mode to FAST Mode Switching

In SLOW Mode the system clock is derived from f_{SUB} . When system clock is switched back to the FAST Mode from f_{SUB} , the CKS2~CKS0 bits should be set to “000”~“110” and then the system clock will respectively be switched to $f_H \sim f_H/64$.

However, if f_H is not used in SLOW Mode and thus switched off, it will take some time to re-oscillate and stabilise when switching to the FAST Mode from the SLOW Mode. This is monitored using the HIRCF bit in the HIRCC register. The time duration required for the high speed system oscillator stabilization is specified in the System Start Up Time Characteristics.



Entering the SLEEP Mode

There is only one way for the device to enter the SLEEP Mode and that is to execute the “HALT” instruction in the application program with both the FHIDEN and FSIDEN bits in the SCC register equal to “0”. In this mode all the clocks and functions will be switched off except the WDT function. When this instruction is executed under the conditions described above, the following will occur:

- The system clock will be stopped and the application program will stop at the “HALT” instruction.
- The Data Memory contents and registers will maintain their present condition.
- The I/O ports will maintain their present conditions.
- In the status register, the Power Down flag, PDF, will be set and the Watchdog time-out flag, TO, will be cleared.
- The WDT will be cleared and resume counting as the WDT function is always enabled.
-

Entering the IDLE0 Mode

There is only one way for the device to enter the IDLE0 mode and that is to execute the “HALT” instruction in the application program with the FHIDEN bit in the SCC register equal to “0” and the FSIDEN bit in the SCC register equal to “1”. When this instruction is executed under the conditions described above, the following will occur:

- The f_H clock will be stopped and the application program will stop at the “HALT” instruction, but the f_{SUB} clock will be on.
- The Data Memory contents and registers will maintain their present condition.
- The I/O ports will maintain their present conditions.
- In the status register, the Power Down flag, PDF, will be set and the Watchdog time-out flag, TO, will be cleared.
- The WDT will be cleared and resume counting as the WDT function is always enabled.

Entering the IDLE1 Mode

There is only one way for the device to enter the IDLE1 mode and that is to execute the “HALT” instruction in the application program with both the FHIDEN and FSIDEN bits in the SCC register equal to “1”. When this instruction is executed under the conditions described above, the following will occur:

- The f_H and f_{SUB} clocks will be on but the application program will stop at the “HALT” instruction.
- The Data Memory contents and registers will maintain their present condition.
- The I/O ports will maintain their present conditions.
- In the status register, the Power Down flag, PDF, will be set and the Watchdog time-out flag, TO, will be cleared.
- The WDT will be cleared and resume counting as the WDT function is always enabled.

Entering the IDLE2 Mode

There is only one way for the device to enter the IDLE2 mode and that is to execute the “HALT” instruction in the application program with the FHIDEN bit in the SCC register equal to “1” and the FSIDEN bit in the SCC register equal to “0”. When this instruction is executed under the conditions described above, the following will occur:

- The f_H clock will be on but the f_{SUB} clock will be off and the application program will stop at the “HALT” instruction.
- The Data Memory contents and registers will maintain their present condition.
- The I/O ports will maintain their present conditions.
- In the status register, the Power Down flag, PDF, will be set and the Watchdog time-out flag, TO, will be cleared.
- The WDT will be cleared and resume counting as the WDT function is always enabled.

Standby Current Considerations

As the main reason for entering the SLEEP or IDLE mode is to keep the current consumption of the device to as low a value as possible, perhaps only in the order of several micro-amps except in the IDLE1 and IDLE2 mode, there are other considerations which must also be taken into account by the circuit designer if the power consumption is to be minimised. Special attention must be made to the I/O pins on the device. All high-impedance input pins must be connected to either a fixed high or low level as any floating input pins could create internal oscillations and result in increased current consumption. This also applies to the device which has different package types, as there may

be unbonded pins. These must either be setup as outputs or if setup as inputs must have pull-high resistors connected.

Care must also be taken with the loads, which are connected to I/O pins, which are setup as outputs. These should be placed in a condition in which minimum current is drawn or connected only to external circuits that do not draw current, such as other CMOS inputs. Also note that additional standby current will also be required if the LIRC oscillator has enabled.

In the IDLE1 and IDLE2 mode the high speed oscillator is on, if the peripheral function clock source is derived from the high speed oscillator, the additional standby current will also be perhaps in the order of several hundred micro-amps.

Wake-up

To minimise power consumption the device can enter the SLEEP or any IDLE mode, where the CPU will be switched off. However, when the device is woken up again, it will take a considerable time for the original system oscillator to restart, stabilise and allow normal operation to resume.

After the system enters the SLEEP or IDLE mode, it can be woken up from one of various sources listed as follows:

- An external falling edge on Port A
- A system interrupt
- A WDT overflow

When the device executes the “HALT” instruction, it will enter the SLEEP or IDLE mode and the PDF flag will be set high. The PDF flag will be cleared to 0 if the device experiences a system power-up or executes the clear Watchdog Timer instruction. If the system is woken up by a WDT overflow, a Watchdog Timer hardware reset will be initiated and the TO flag will be set to 1. The TO flag is set if a WDT time-out occurs and causes a wake-up that only resets the Program Counter and Stack Pointer, other flags remain in their original status.

Each pin on Port A can be setup using the PAWU register to permit a negative transition on the pin to wake-up the system. When a pin wake-up occurs, the program will resume execution at the instruction following the “HALT” instruction. If the system is woken up by an interrupt, then two possible situations may occur. The first is where the related interrupt is disabled or the interrupt is enabled but the stack is full, in which case the program will resume execution at the instruction following the “HALT” instruction. In this situation, the interrupt which woke up the device will not be immediately serviced, but will rather be serviced later when the related interrupt is finally enabled or when a stack level becomes free. The other situation is where the related interrupt is enabled and the stack is not full, in which case the regular interrupt response takes place. If an interrupt request flag is set high before entering the SLEEP or IDLE mode, the wake-up function of the related interrupt will be disabled.

Watchdog Timer

The Watchdog Timer is provided to prevent program malfunctions or sequences from jumping to unknown locations, due to certain uncontrollable external events such as electrical noise.

Watchdog Timer Clock Source

The Watchdog Timer clock source is provided by the internal clock, f_{LIRC} , which is sourced from the LIRC oscillator. The LIRC internal oscillator has an approximate frequency of 32kHz and this specified internal clock period can vary with V_{DD} , temperature and process variations. The Watchdog Timer source clock is then subdivided by a ratio of 2^8 to 2^{18} to give longer timeouts, the actual value being chosen using the WS2~WS0 bits in the WDTC register.

Watchdog Timer Control Register

A single register, WDTC, controls the required time-out period as well as the Watchdog Timer the enable and the MCU reset operation.

• WDTC Register

Bit	7	6	5	4	3	2	1	0
Name	WE4	WE3	WE2	WE1	WE0	WS2	WS1	WS0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	1	0	1	0	0	1	1

Bit 7~3 **WE4~WE0**: WDT function software control

10101 or 01010: Enable

Other values: Reset MCU

When these bits are changed to any other value due to environmental noise the microcontroller will be reset; this reset operation will be activated after a delay time, t_{SRESET} , and the WRF bit in the RSTFC register will be set high.

Bit 2~0 **WS2~WS0**: WDT time-out period selection

000: $2^8/f_{LIRC}$

001: $2^{10}/f_{LIRC}$

010: $2^{12}/f_{LIRC}$

011: $2^{14}/f_{LIRC}$

100: $2^{15}/f_{LIRC}$

101: $2^{16}/f_{LIRC}$

110: $2^{17}/f_{LIRC}$

111: $2^{18}/f_{LIRC}$

These three bits determine the division ratio of the Watchdog Timer source clock, which in turn determines the timeout period.

• RSTFC Register

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	LVRF	—	WRF
R/W	—	—	—	—	—	R/W	—	R/W
POR	—	—	—	—	—	x	—	0

"x": unknown

Bit 7~3 Unimplemented, read as "0"

Bit 2 **LVRF**: LVR function reset flag

Refer to the Low Voltage Reset section.

Bit 1 Unimplemented, read as "0"

Bit 0 **WRF**: WDTC register software reset flag
 0: Not occurred
 1: Occurred
 This bit is set to 1 by the WDTC register software reset and cleared to zero by the application program. Note that this bit can only be cleared to zero by the application program.

Watchdog Timer Operation

The Watchdog Timer operates by providing a device reset when its timer overflows. This means that in the application program and during normal operation the user has to strategically clear the Watchdog Timer before it overflows to prevent the Watchdog Timer from executing a reset. This is done using the clear watchdog instruction. If the program malfunctions for whatever reason, jumps to an unknown location, or enters an endless loop, the clear instruction will not be executed in the correct manner, in which case the Watchdog Timer will overflow and reset the device. There are five bits, WE4~WE0, in the WDTC register to offer the enable control of the Watchdog Timer and the MCU reset. The WDT function will be enabled if the WE4~WE0 bits are equal to 01010B or 10101B. If the WE4~WE0 bits are set to any other values, other than 01010B and 10101B, it will reset the device after a delay time, t_{SRESET} . After power-on these bits will have a value of 01010B.

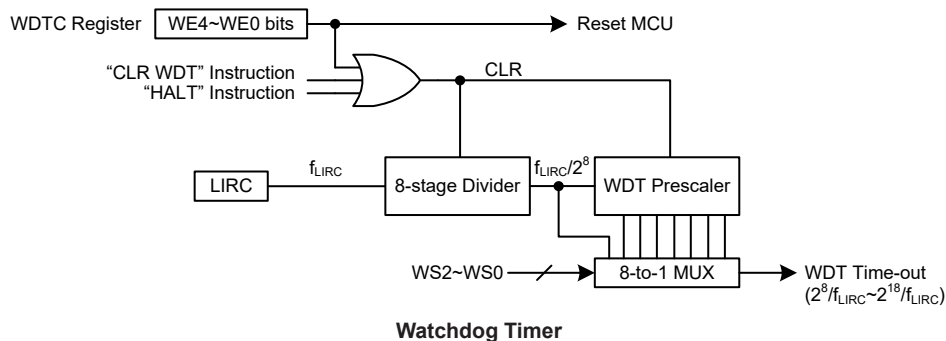
WE4~WE0 Bits	WDT Function
01010B or 10101B	Enable
Any other value	Reset MCU

Watchdog Timer Function Control

Under normal program operation, a Watchdog Timer time-out will initialise a device reset and set the status bit TO. However, if the system is in the SLEEP or IDLE Mode, when a Watchdog Timer time-out occurs, the TO and PDF bits in the STATUS register will be set and only the Program Counter and Stack Pointer will be reset. Three methods can be adopted to clear the contents of the Watchdog Timer. The first is a WDTC register software reset, which means a certain value except 01010B and 10101B written into the WE4~WE0 bits, the second is using the Watchdog Timer software clear instruction, the third is via a HALT instruction.

There is only one method of using software instruction to clear the Watchdog Timer. That is to use the single “CLR WDT” instruction to clear the WDT.

The maximum time-out period is when the 2^{18} division ratio is selected. As an example, with a 32kHz LIRC oscillator as its source clock, this will give a maximum watchdog period of around 8s for the 2^{18} division ratio, and a minimum timeout of 8ms for the 2^8 division ration.



Reset and Initialisation

A reset function is a fundamental part of any microcontroller ensuring that the device can be set to some predetermined condition irrespective of outside parameters. The most important reset condition is after power is first applied to the microcontroller. In this case, internal circuitry will ensure that the microcontroller, after a short delay, will be in a well-defined state and ready to execute the first program instruction. After this power-on reset, certain important internal registers will be set to defined states before the program commences. One of these registers is the Program Counter, which will be reset to zero forcing the microcontroller to begin program execution from the lowest Program Memory address.

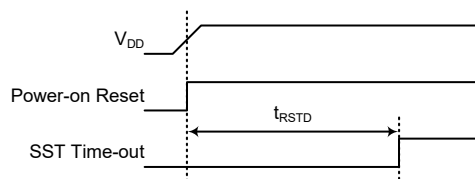
In addition to the power-on reset, another reset exists in the form of a Low Voltage Reset, LVR, where a full reset is implemented in situations where the power supply voltage falls below a certain threshold. Another type of reset is when the Watchdog Timer overflows and resets the microcontroller. All types of reset operations result in different register conditions being setup.

Reset Functions

There are several ways in which a microcontroller reset can occur, through events occurring internally.

Power-on Reset

The most fundamental and unavoidable reset is the one that occurs after power is first applied to the microcontroller. As well as ensuring that the Program Memory begins execution from the first memory address, a power-on reset also ensures that certain other registers are preset to known conditions. All the I/O port and port control registers will power up in a high condition ensuring that all pins will be first set to inputs.

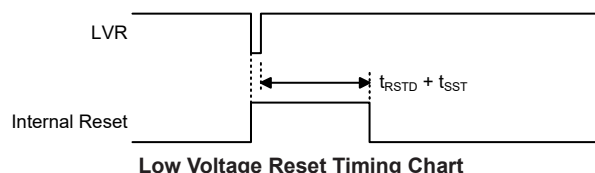


Power-on Reset Timing Chart

Low Voltage Reset – LVR

The microcontroller contains a low voltage reset circuit in order to monitor the supply voltage of the device and provide an MCU reset when the value falls below a certain predefined level.

The LVR function is always enabled in FAST and SLOW mode with a specific LVR voltage V_{LVR} . For the device the V_{LVR} value is fixed at 2.1V. If the supply voltage of the device drops to within a range of $0.9V \sim V_{LVR}$ such as might occur when changing the battery in battery powered applications, the LVR will automatically reset the device internally and the LVRF bit in the RSTFC register will also be set to 1. For a valid LVR signal, a low supply voltage, i.e., a voltage in the range between $0.9V \sim V_{LVR}$ must exist for a time greater than that specified by t_{LVR} in the LVD & LVR Electrical Characteristics. If the low supply voltage state does not exceed this value, the LVR will ignore the low supply voltage and will not perform a reset function. Note that the LVR function will be automatically disabled when the device enters the SLEEP or IDLE mode.



Low Voltage Reset Timing Chart

• **RSTFC Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	LVRF	—	WRF
R/W	—	—	—	—	—	R/W	—	R/W
POR	—	—	—	—	—	x	—	0

“x”: unknown

Bit 7~3 Unimplemented, read as “0”

Bit 2 **LVRF**: LVR function reset flag
 0: Not occurred
 1: Occurred

This bit is set high when an actual Low Voltage Reset condition occurs. This bit can be cleared to zero only by the application program.

Bit 1 Unimplemented, read as “0”

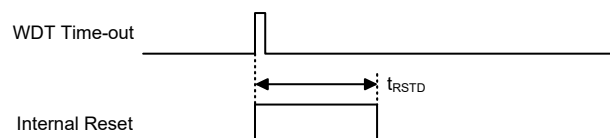
Bit 0 **WRF**: WDTC register software reset flag
 Refer to the Watchdog Timer Control Register section.

IAP Reset

When a specific value of “55H” is written into the FC1 register, a reset signal will be generated to reset the whole device. Refer to the In Application Programming section for more associated details.

Watchdog Time-out Reset during Normal Operation

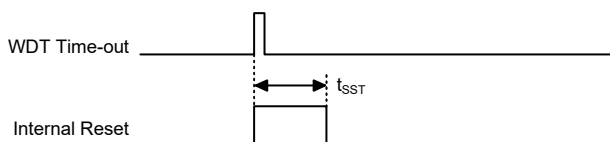
After a Watchdog time-out reset during normal operation, the Watchdog time-out flag TO will be set to “1”.



WDT Time-out Reset during Normal Operation Timing Chart

Watchdog Time-out Reset during SLEEP or IDLE Mode

The Watchdog time-out Reset during SLEEP or IDLE Mode is a little different from other kinds of reset. Most of the conditions remain unchanged except that the Program Counter and the Stack Pointer will be cleared to “0” and the TO and PDF flags will be set to “1”. Refer to the System Start Up Time Characteristics for t_{SST} details.



WDT Time-out Reset during SLEEP or IDLE Timing Chart

Reset Initial Conditions

The different types of reset described affect the reset flags in different ways. These flags, known as PDF and TO are located in the status register and are controlled by various microcontroller operations, such as the SLEEP or IDLE Mode function or Watchdog Timer. The reset flags are shown in the table:

TO	PDF	Reset Conditions
0	0	Power-on reset
u	u	LVR reset during FAST or SLOW Mode operation
1	u	WDT time-out reset during FAST or SLOW Mode operation
1	1	WDT time-out reset during IDLE or SLEEP Mode operation

“u” stands for unchanged

The following table indicates the way in which the various components of the microcontroller are affected after a power-on reset occurs.

Item	Condition after Reset
Program Counter	Reset to zero
Interrupts	All interrupts will be disabled
WDT, Time Bases	Cleared after reset, WDT begins counting
Timer Modules	Timer Modules will be turned off
Input/Output Ports	I/O ports will be setup as inputs
Stack Pointer	Stack Pointer will point to the top of the stack

The different kinds of resets all affect the internal registers of the microcontroller in different ways. To ensure reliable continuation of normal program execution after a reset occurs, it is important to know what condition the microcontroller is in after a particular reset occurs. The following table describes how each type of reset affects each of the microcontroller internal registers.

Register	Power-On Reset	WDT Time-out (Normal Operation)	WDT Time-out (IDLE/SLEEP)
IAR0	0000 0000	0000 0000	uuuu uuuu
MP0	0000 0000	0000 0000	uuuu uuuu
IAR1	0000 0000	0000 0000	uuuu uuuu
MP1L	0000 0000	0000 0000	uuuu uuuu
MP1H	0000 0000	0000 0000	uuuu uuuu
ACC	xxxx xxxx	uuuu uuuu	uuuu uuuu
PCL	0000 0000	0000 0000	0000 0000
TBLP	xxxx xxxx	uuuu uuuu	uuuu uuuu
TBLH	xxxx xxxx	uuuu uuuu	uuuu uuuu
TBHP	---x xxxx	---u uuuu	---u uuuu
STATUS	xx00 xxxx	uu1u uuuu	uu11 uuuu
VBGRC	---- ---0	---- ---0	---- ---u
IAR2	0000 0000	0000 0000	uuuu uuuu
MP2L	0000 0000	0000 0000	uuuu uuuu
MP2H	0000 0000	0000 0000	uuuu uuuu
RSTFC	---- -x -0	---- -u -u	---- -u -u
TB0C	0--- -000	0--- -000	u--- -uuu
TB1C	0--- -000	0--- -000	u--- -uuu
SCC	000- --00	000- --00	uuu- --uu
HIRCC	---- 0001	---- 0001	---- uuuu
PA	1111 1111	1111 1111	uuuu uuuu
PAC	1111 1111	1111 1111	uuuu uuuu

Register	Power-On Reset	WDT Time-out (Normal Operation)	WDT Time-out (IDLE/SLEEP)
PAPU	0000 0000	0000 0000	uuuu uuuu
PAWU	0000 0000	0000 0000	uuuu uuuu
PB	1111 1111	1111 1111	uuuu uuuu
PBC	1111 1111	1111 1111	uuuu uuuu
PBPU	0000 0000	0000 0000	uuuu uuuu
IFS0	0000 0000	0000 0000	uuuu uuuu
IFS1	0000 0000	0000 0000	uuuu uuuu
PSCR	---- --00	---- --00	---- --uu
LVDC	--00 0000	--00 0000	--uu uuuu
REGC	0--- -000	0--- -000	u--- -uuu
PC	1111 1111	1111 1111	uuuu uuuu
PCC	1111 1111	1111 1111	uuuu uuuu
PCPU	0000 0000	0000 0000	uuuu uuuu
PD	1111 1111	1111 1111	uuuu uuuu
PDC	1111 1111	1111 1111	uuuu uuuu
PDPU	0000 0000	0000 0000	uuuu uuuu
STMC0	0000 0000	0000 0000	uuuu uuuu
STMC1	0000 0000	0000 0000	uuuu uuuu
STMDL	0000 0000	0000 0000	uuuu uuuu
STMDH	---- --00	---- --00	---- --uu
STMAL	0000 0000	0000 0000	uuuu uuuu
STMAH	---- --00	---- --00	---- --uu
SADOL	xxxx ----	xxxx ----	uuuu ---- (ADRFs=0)
			uuuu uuuu (ADRFs=1)
SADOH	xxxx xxxx	xxxx xxxx	uuuu uuuu (ADRFs=0)
			---- uuuu (ADRFs=1)
SADC0	0000 0000	0000 0000	uuuu uuuu
SADC1	0000 0000	0000 0000	uuuu uuuu
SADC2	---- --10	---- --10	---- --uu
LCDC0	0000 0000	0000 0000	uuuu uuuu
LCDC1	000- 0000	000- 0000	uuu- uuuu
PAS0	0000 0000	0000 0000	uuuu uuuu
PAS1	0000 0000	0000 0000	uuuu uuuu
PBS0	0000 0000	0000 0000	uuuu uuuu
PBS1	0000 0000	0000 0000	uuuu uuuu
PCS0	0000 0000	0000 0000	uuuu uuuu
PCS1	0000 0000	0000 0000	uuuu uuuu
PDS0	0000 0000	0000 0000	uuuu uuuu
PDS1	0000 0000	0000 0000	uuuu uuuu
WDTC	0101 0011	0101 0011	uuuu uuuu
EEA	0000 0000	0000 0000	uuuu uuuu
EED	0000 0000	0000 0000	uuuu uuuu
EEC	---- 0000	---- 0000	---- uuuu
SIMC0	1110 0000	1110 0000	uuuu uuuu

Register	Power-On Reset	WDT Time-out (Normal Operation)	WDT Time-out (IDLE/SLEEP)
SIMC1 (UMD=0)	1000 0001	1000 0001	uuuu uuuu
UUCR1* (UMD=1)	0000 00x0	0000 00x0	uuuu uuuu
SIMD/UTXR_RXR	xxxx xxxx	xxxx xxxx	uuuu uuuu
SIMA/SIMC2/UUCR2	0000 0000	0000 0000	uuuu uuuu
UUCR3	---- --0	---- --0	---- --u
SIMTOC (UMD=0)	0000 0000	0000 0000	uuuu uuuu
UBRG* (UMD=1)	xxxx xxxx	xxxx xxxx	uuuu uuuu
UUSR	0000 1011	0000 1011	uuuu uuuu
INTEG	---- 0000	---- 0000	---- uuuu
INTC0	-000 0000	-000 0000	-uuu uuuu
INTC1	0000 0000	0000 0000	uuuu uuuu
INTC2	0000 0000	0000 0000	uuuu uuuu
INTC3	---0 ---0	---0 ---0	---u ---u
PTMC0	0000 0---	0000 0---	uuuu u---
PTMC1	0000 0000	0000 0000	uuuu uuuu
PTMDL	0000 0000	0000 0000	uuuu uuuu
PTMDH	---- --00	---- --00	---- --uu
PTMAL	0000 0000	0000 0000	uuuu uuuu
PTMAH	---- --00	---- --00	---- --uu
PTMRPL	0000 0000	0000 0000	uuuu uuuu
PTMRPH	---- --00	---- --00	---- --uu
FC0	0000 0000	0000 0000	uuuu uuuu
FC1	0000 0000	0000 0000	uuuu uuuu
FC2	---- --0	---- --0	---- --u
FARL	0000 0000	0000 0000	uuuu uuuu
FARH	---0 0000	---0 0000	---u uuuu
FD0L	0000 0000	0000 0000	uuuu uuuu
FD0H	0000 0000	0000 0000	uuuu uuuu
FD1L	0000 0000	0000 0000	uuuu uuuu
FD1H	0000 0000	0000 0000	uuuu uuuu
FD2L	0000 0000	0000 0000	uuuu uuuu
FD2H	0000 0000	0000 0000	uuuu uuuu
FD3L	0000 0000	0000 0000	uuuu uuuu
FD3H	0000 0000	0000 0000	uuuu uuuu
OPSW0	0000 0000	0000 0000	uuuu uuuu
OPSW1	---- --0	---- --0	---- --u
OPPW	---- --00	---- --00	---- --uu
OPC	-00- --00	-00- --00	-uu- --uu
OPVOS	0010 0000	0010 0000	uuuu uuuu
OPPGAC0	0000 0000	0000 0000	uuuu uuuu
OPPGAC1	---- --0	---- --0	---- --u
LMSADOH	xxxx xxxx	uuuu uuuu	uuuu uuuu
LMSADOL	xxxx ----	uuuu ----	uuuu ----
ORMC	0000 0000	0000 0000	0000 0000
PE	---- 1111	---- 1111	---- uuuu
PEC	---- 1111	---- 1111	---- uuuu
PEPU	---- 0000	---- 0000	---- uuuu

Register	Power-On Reset	WDT Time-out (Normal Operation)	WDT Time-out (IDLE/SLEEP)
DAH	0000 0000	0000 0000	uuuu uuuu
DAL	0000 0000	0000 0000	uuuu uuuu
DACC	---- --0	---- --0	---- --u
SLEDC0	0000 0000	0000 0000	uuuu uuuu
SLEDC1	0000 0000	0000 0000	uuuu uuuu
USR	0000 1011	0000 1011	uuuu uuuu
UCR1	0000 00x0	0000 00x0	uuuu uuuu
UCR2	0000 0000	0000 0000	uuuu uuuu
UCR3	---- --0	---- --0	---- --u
TXR_RXR	xxxx xxxx	xxxx xxxx	uuuu uuuu
BRG	xxxx xxxx	xxxx xxxx	uuuu uuuu
PMPS	0000 0000	0000 0000	uuuu uuuu

Note: “u” stands for unchanged

“x” stands for unknown

“-” stands for unimplemented

“*”: The UUCR1 and SIMC1 registers share the same memory address while the UBRG and SIMTOC registers share the same memory address. The default value of the UUCR1 or UBRG register can be obtained when the UMD bit is set high by application program after a reset.

Input/Output Ports

Holtek microcontrollers offer considerable flexibility on their I/O ports. With the input or output designation of every pin fully under user program control, pull-high selections for all ports and wake-up selections on certain pins, the user is provided with an I/O structure to meet the needs of a wide range of application possibilities.

The device provides bidirectional input/output lines. These I/O ports are mapped to the Data Memory with specific addresses as shown in the Special Purpose Data Memory table. All of these I/O ports can be used for input and output operations. For input operation, these ports are non-latching, which means the inputs must be ready at the T2 rising edge of instruction “MOV A, [m]”, where m denotes the port address. For output operation, all the data is latched and remains unchanged until the output latch is rewritten.

Register Name	Bit							
	7	6	5	4	3	2	1	0
PA	PA7	PA6	PA5	PA4	PA3	PA2	PA1	PA0
PAC	PAC7	PAC6	PAC5	PAC4	PAC3	PAC2	PAC1	PAC0
PAPU	PAPU7	PAPU6	PAPU5	PAPU4	PAPU3	PAPU2	PAPU1	PAPU0
PAWU	PAWU7	PAWU6	PAWU5	PAWU4	PAWU3	PAWU2	PAWU1	PAWU0
PB	PB7	PB6	PB5	PB4	PB3	PB2	PB1	PB0
PBC	PBC7*	PBC6	PBC5	PBC4	PBC3	PBC2	PBC1	PBC0
PBPU	PBPU7**	PBPU6	PBPU5	PBPU4	PBPU3	PBPU2	PBPU1	PBPU0
PC	PC7	PC6	PC5	PC4	PC3	PC2	PC1	PC0
PCC	PCC7*	PCC6*	PCC5	PCC4	PCC3	PCC2	PCC1	PCC0
PCPU	PCPU7**	PCPU6**	PCPU5	PCPU4	PCPU3	PCPU2	PCPU1	PCPU0
PD	PD7	PD6	PD5	PD4	PD3	PD2	PD1	PD0
PDC	PDC7	PDC6	PDC5	PDC4	PDC3	PDC2	PDC1	PDC0
PDPU	PDPU7	PDPU6	PDPU5	PDPU4	PDPU3	PDPU2	PDPU1	PDPU0

Register Name	Bit							
	7	6	5	4	3	2	1	0
PE	—	—	—	—	PE3	PE2	PE1	PE0
PEC	—	—	—	—	PEC3*	PEC2*	PEC1*	PEC0*
PEPU	—	—	—	—	PEPU3**	PEPU2**	PEPU1**	PEPU0**

“—”: Unimplemented, read as “0”

I/O Logic Function Register List

Note: The control bit with an asterisk * should be cleared to 0 and the control bit with an asterisk ** should remain the POR value after power-on reset. This can set the unbounded and not internally used pin as output to prevent additional power consumption.

Pull-high Resistors

Many product applications require pull-high resistors for their switch inputs usually requiring the use of an external resistor. To eliminate the need for these external resistors, all I/O pins, when configured as a digital input have the capability of being connected to an internal pull-high resistor. These pull-high resistors are selected using registers, namely PAPU~PEPU, and are implemented using weak PMOS transistors.

Note that the pull-high resistor can be controlled by the relevant pull-high control register only when the pin-shared functional pin is selected as a digital input or NMOS output. Otherwise, the pull-high resistors cannot be enabled

• PxPU Register

Bit	7	6	5	4	3	2	1	0
Name	PxPU7	PxPU6	PxPU5	PxPU4	PxPU3	PxPU2	PxPU1	PxPU0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

PxPUn: I/O Port x Pin pull-high function control

0: Disable

1: Enable

The PxPUn bit is used to control the pin pull-high function. Here the “x” can be A, B, C, D or E. However, the actual available bits for each I/O Port may be different.

Port A Wake-up

The HALT instruction forces the microcontroller into the SLEEP or IDLE Mode which preserves power, a feature that is important for battery and other low-power applications. Various methods exist to wake up the microcontroller, one of which is to change the logic condition on one of the Port A pins from high to low. This function is especially suitable for applications that can be woken up via external switches. Each pin on Port A can be selected individually to have this wake-up feature using the PAWU register.

Note that the wake-up function can be controlled by the wake-up control register only when the pin is selected as a general purpose input and the MCU enters the IDLE or SLEEP mode.

• **PAWU Register**

Bit	7	6	5	4	3	2	1	0
Name	PAWU7	PAWU6	PAWU5	PAWU4	PAWU3	PAWU2	PAWU1	PAWU0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **PAWU7~PAWU0**: PA7~PA0 wake-up function control

0: Disable

1: Enable

I/O Port Control Registers

Each I/O port has its own control register known as PAC~PEC, to control the input/output configuration. With this control register, each CMOS output or input can be reconfigured dynamically under software control. Each pin of the I/O ports is directly mapped to a bit in its associated port control register. For the I/O pin to function as an input, the corresponding bit of the control register must be written as a “1”. This will then allow the logic state of the input pin to be directly read by instructions. When the corresponding bit of the control register is written as a “0”, the I/O pin will be setup as a CMOS output. If the pin is currently setup as an output, instructions can still be used to read the output register. However, it should be noted that the program will in fact only read the status of the output data latch and not the actual logic status of the output pin.

• **PxC Register**

Bit	7	6	5	4	3	2	1	0
Name	PxC7	PxC6	PxC5	PxC4	PxC3	PxC2	PxC1	PxC0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	1	1	1	1	1	1	1	1

PxCn: I/O Port x Pin type selection

0: Output

1: Input

The PxCn bit is used to control the pin type selection. Here the “x” can be A, B, C, D or E. However, the actual available bits for each I/O Port may be different.

I/O Port Source Current Selection

As for LED driving applications, the source current of each I/O pin in this device can be configured using its corresponding source current selection bits. These source current selection bits are available only when the corresponding pin is configured as a CMOS output. Otherwise, these select bits have no effect. Users should refer to the Input/Output Characteristics section to obtain the exact value for different applications.

Register Name	Bit							
	7	6	5	4	3	2	1	0
SLEDC0	SLEDC07	SLEDC06	SLEDC05	SLEDC04	SLEDC03	SLEDC02	SLEDC01	SLEDC00
SLEDC1	SLEDC17	SLEDC16	SLEDC15	SLEDC14	SLEDC13	SLEDC12	SLEDC11	SLEDC10

Source Current Selection Register List

• **SLEDC0 Register**

Bit	7	6	5	4	3	2	1	0
Name	SLEDC07	SLEDC06	SLEDC05	SLEDC04	SLEDC03	SLEDC02	SLEDC01	SLEDC00
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

- Bit 7~6 **SLEDC07~SLEDC06:** PB6~PB4 source current selection
00: Source current=Level 0 (Min.)
01: Source current=Level 1
10: Source current=Level 2
11: Source current=Level 3 (Max.)
- Bit 5~4 **SLEDC05~SLEDC04:** PB3~PB0 source current selection
00: Source current=Level 0 (Min.)
01: Source current=Level 1
10: Source current=Level 2
11: Source current=Level 3 (Max.)
- Bit 3~2 **SLEDC03~SLEDC02:** PA7~PA4 source current selection
00: Source current=Level 0 (Min.)
01: Source current=Level 1
10: Source current=Level 2
11: Source current=Level 3 (Max.)
- Bit 1~0 **SLEDC01~SLEDC00:** PA3~PA0 source current selection
00: Source current=Level 0 (Min.)
01: Source current=Level 1
10: Source current=Level 2
11: Source current=Level 3 (Max.)

• **SLEDC1 Register**

Bit	7	6	5	4	3	2	1	0
Name	SLEDC17	SLEDC16	SLEDC15	SLEDC14	SLEDC13	SLEDC12	SLEDC11	SLEDC10
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

- Bit 7~6 **SLEDC17~SLEDC16:** PD7~PD4 source current selection
00: Source current=Level 0 (Min.)
01: Source current=Level 1
10: Source current=Level 2
11: Source current=Level 3 (Max.)
- Bit 5~4 **SLEDC15~SLEDC14:** PD3~PD0 source current selection
00: Source current=Level 0 (Min.)
01: Source current=Level 1
10: Source current=Level 2
11: Source current=Level 3 (Max.)
- Bit 3~2 **SLEDC13~SLEDC12:** PC5~PC4 source current selection
00: Source current=Level 0 (Min.)
01: Source current=Level 1
10: Source current=Level 2
11: Source current=Level 3 (Max.)
- Bit 1~0 **SLEDC11~SLEDC10:** PC3~PC0 source current selection
00: Source current=Level 0 (Min.)
01: Source current=Level 1
10: Source current=Level 2
11: Source current=Level 3 (Max.)

I/O Port Power Source Control

This device supports different I/O port power source selections for PA1, PA3 and PA4. With the exception of $\overline{\text{RES}}/\text{OCDS}$, the multi-power function is only effective when the pin is set to have a digital input or output function.

The port power can come from either the power pin VDD or VDDIO, which is determined using the PMPS7~PMPS0 bits in the PMPS register. The VDDIO power pin function should first be selected using the corresponding pin-shared function selection bits if the port power is supposed to come from the VDDIO pin.

An important point to know is that the input power voltage on the VDDIO pin should be equal to or less than the device supply power voltage V_{DD} when the VDDIO pin is selected as the port power supply pin. However, when either V_{DD} or V_{DDIO} is less than 2.2V, it is recommended that the V_{DDIO} power should be equal to V_{DD} .

• PMPS Register

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	PMPS5	PMPS4	PMPS3	PMPS2	PMPS1	PMPS0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~6 **D7~D6:** Reserved bits, cannot be changed

Bit 5~4 **PMPS5~PMPS4:** PA4 pin power supply selection

0x: V_{DD}

1x: V_{DDIO}

Bit 3~2 **PMPS3~PMPS2:** PA3 pin power supply selection

0x: V_{DD}

1x: V_{DDIO}

Bit 1~0 **PMPS1~PMPS0:** PA1 pin power supply selection

0x: V_{DD}

1x: V_{DDIO}

If the PB6 pin is switched to the VDDIO function and the PMPS[5:0] bits are set to “1x”, the VDDIO pin input voltage can be used for PA1, PA3 and PA4 pin power.

Pin-shared Functions

The flexibility of the microcontroller range is greatly enhanced by the use of pins that have more than one function. Limited numbers of pins can force serious design constraints on designers but by supplying pins with multi-functions, many of these difficulties can be overcome. For these pins, the desired function of the multi-function I/O pins is selected by a series of registers via the application program control.

Pin-shared Function Selection Registers

The limited number of supplied pins in a package can impose restrictions on the amount of functions a certain device can contain. However by allowing the same pins to share several different functions and providing a means of function selection, a wide range of different functions can be incorporated into even relatively small package sizes. The device includes Port “x” Output Function Selection register “n”, labeled as P_xS_n, and Input Function Selection register, labeled as IFS_i, which can select the desired functions of the multi-function pin-shared pins.

The most important point to note is to make sure that the desired pin-shared function is properly selected and also deselected. For most pin-shared functions, to select the desired pin-shared function, the pin-shared function should first be correctly selected using the corresponding pin-shared control

register. After that the corresponding peripheral functional setting should be configured and then the peripheral function can be enabled. However, a special point must be noted for digital input pins, such as INTn, xTCK, etc, which share the same pin-shared control configuration with their corresponding general purpose I/O functions when setting the relevant pin-shared control bits. To select these pin functions, in addition to the necessary pin-shared control and peripheral functional setup aforementioned, they must also be set as an input by setting the corresponding bits in the I/O port control register. To correctly deselect the pin-shared function, the peripheral function should first be disabled and then the corresponding pin-shared function control register can be modified to select other pin-shared functions.

Register Name	Bit							
	7	6	5	4	3	2	1	0
IFS0	D7	D6	IFS05	IFS04	IFS03	IFS02	IFS01	IFS00
IFS1	IFS17	IFS16	IFS15	IFS14	IFS13	IFS12	IFS11	IFS10
PAS0	PAS07	PAS06	PAS05	PAS04	PAS03	PAS02	PAS01	PAS00
PAS1	PAS17	PAS16	PAS15	PAS14	PAS13	PAS12	PAS11	PAS10
PBS0	PBS07	PBS06	PBS05	PBS04	PBS03	PBS02	PBS01	PBS00
PBS1	D7	D6	PBS15	PBS14	PBS13	PBS12	PBS11	PBS10
PCS0	PCS07	PCS06	PCS05	PCS04	PCS03	PCS02	PCS01	PCS00
PCS1	D7	D6	D5	D4	PCS13	PCS12	PCS11	PCS10
PDS0	PDS07	PDS06	PDS05	PDS04	PDS03	PDS02	PDS01	PDS00
PDS1	PDS17	PDS16	PDS15	PDS14	PDS13	PDS12	PDS11	PDS10

Pin-shared Function Selection Register List

• **IFS0 Register**

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	IFS05	IFS04	IFS03	IFS02	IFS01	IFS00
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~6 **D7~D6:** Reserved, must be fixed at “00”

Bit 5~4 **IFS05~IFS04:** $\overline{\text{SCS}}$ input source pin selection

00: PA3
01: Reserved
10: PD1
11: PA3

Bit 3~2 **IFS03~IFS02:** SCK/SCL input source pin selection

00: PA1
01: Reserved
10: PD2
11: PA1

Note: If the SPI Master Mode is selected, when the SIMEN bit is set high, the PA1 and PD2 pins all can be used as the SCK pin function ignoring the IFS0[3:2] bit settings.

Bit 1~0 **IFS01~IFS00:** SDI/SDA/URX/UTX input source pin selection

00: PA2
01: PC5
10: Reserved
11: PD3

• **IFS1 Register**

Bit	7	6	5	4	3	2	1	0
Name	IFS17	IFS16	IFS15	IFS14	IFS13	IFS12	IFS11	IFS10
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

- Bit 7~6 **IFS17~IFS16:** RX/TX input source pin selection
00: PA5
01: PB0
10: PA5
11: PA5
- Bit 5~4 **IFS15~IFS14:** STCK input source pin selection
00: PA3
01: Reserved
10: PA3
11: PA3
- Bit 3~2 **IFS13~IFS12:** INT1 input source pin selection
00: PA4
01: PA7
10: PA4
11: PA4
- Bit 1~0 **IFS11~IFS10:** INT0 input source pin selection
00: PA1
01: PB0
10: PA1
11: PA1

• **PAS0 Register**

Bit	7	6	5	4	3	2	1	0
Name	PAS07	PAS06	PAS05	PAS04	PAS03	PAS02	PAS01	PAS00
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

- Bit 7~6 **PAS07~PAS06:** PA3 Pin-Shared function selection
00: PA3/STCK
01: SCS
10: PTP
11: AN1
- Bit 5~4 **PAS05~PAS04:** PA2 Pin-Shared function selection
00: PA2
01: SDI/SDA/URX/UTX
10: PA2
11: PA2
- Bit 3~2 **PAS03~PAS02:** PA1 Pin-Shared function selection
00: PA1/INT0
01: SCK/SCL
10: PTPB
11: AN0
- Bit 1~0 **PAS01~PAS00:** PA0 Pin-Shared function selection
00: PA0
01: UTX/SDO
10: PA0
11: PA0

• **PAS1 Register**

Bit	7	6	5	4	3	2	1	0
Name	PAS17	PAS16	PAS15	PAS14	PAS13	PAS12	PAS11	PAS10
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~6 **PAS17~PAS16:** PA7 Pin-Shared function selection

00: PA7/INT1
01: SEG2
10: TX
11: AN6

Bit 5~4 **PAS15~PAS14:** PA6 Pin-Shared function selection

00: PA6/PTCK
01: SEG1
10: TX
11: AN5

Bit 3~2 **PAS13~PAS12:** PA5 Pin-Shared function selection

00: PA5
01: SEG0
10: RX/TX
11: AN4

Bit 1~0 **PAS11~PAS10:** PA4 Pin-Shared function selection

00: PA4/INT1
01: UTX/SDO
10: VREF
11: DACO

• **PBS0 Register**

Bit	7	6	5	4	3	2	1	0
Name	PBS07	PBS06	PBS05	PBS04	PBS03	PBS02	PBS01	PBS00
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~6 **PBS07~PBS06:** PB3 Pin-Shared function selection

00: PB3
01: SEG6
10: PB3
11: PB3

Bit 5~4 **PBS05~PBS04:** PB2 Pin-Shared function selection

00: PB2
01: SEG5
10: PB2
11: PB2

Bit 3~2 **PBS03~PBS02:** PB1 Pin-Shared function selection

00: PB1
01: SEG4
10: PB1
11: PB1

Bit 1~0 **PBS01~PBS00:** PB0 Pin-Shared function selection

00: PB0/INT0
01: SEG3
10: RX/TX
11: AN7

• **PBS1 Register**

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	PBS15	PBS14	PBS13	PBS12	PBS11	PBS10
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

- Bit 7~6 **D7~D6:** Reserved bits, cannot be changed
- Bit 5~4 **PBS15~PBS14:** PB6 Pin-Shared function selection
 00: PB6
 01: PTP
 10: AN2
 11: VDDIO
- Bit 3~2 **PBS13~PBS12:** PB5 Pin-Shared function selection
 00: PB5
 01: SEG8
 10: PB5
 11: PB5
- Bit 1~0 **PBS11~PBS10:** PB4 Pin-Shared function selection
 00: PB4
 01: SEG7
 10: PB4
 11: PB4

• **PCS0 Register**

Bit	7	6	5	4	3	2	1	0
Name	PCS07	PCS06	PCS05	PCS04	PCS03	PCS02	PCS01	PCS00
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

- Bit 7~6 **PCS07~PCS06:** PC3 Pin-Shared function selection
 00: PC3
 01: COM3
 10: PC3
 11: PC3
- Bit 5~4 **PCS05~PCS04:** PC2 Pin-Shared function selection
 00: PC2
 01: COM2
 10: PC2
 11: PC2
- Bit 3~2 **PCS03~PCS02:** PC1 Pin-Shared function selection
 00: PC1
 01: COM1
 10: PC1
 11: PC1
- Bit 1~0 **PCS01~PCS00:** PC0 Pin-Shared function selection
 00: PC0
 01: COM0
 10: PC0
 11: PC0

• **PCS1 Register**

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	PCS13	PCS12	PCS11	PCS10
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~4 **D7~D4:** Reserved bits, cannot be changed

Bit 3~2 **PCS13~PCS12:** PC5 Pin-Shared function selection
 00: PC5
 01: SDI/SDA/URX/UTX
 10: PTP
 11: PC5

Bit 1~0 **PCS11~PCS10:** PC4 Pin-Shared function selection
 00: PC4
 01: UTX/SDO
 10: PC4
 11: PC4

• **PDS0 Register**

Bit	7	6	5	4	3	2	1	0
Name	PDS07	PDS06	PDS05	PDS04	PDS03	PDS02	PDS01	PDS00
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~6 **PDS07~PDS06:** PD3 Pin-Shared function selection
 00: PD3
 01: PTP
 10: PTPB
 11: SDI/SDA/URX/UTX

Bit 5~4 **PDS05~PDS04:** PD2 Pin-Shared function selection
 00: PD2
 01: STPB
 10: SCK/SCL
 11: PD2

Bit 3~2 **PDS03~PDS02:** PD1 Pin-Shared function selection
 00: PD1
 01: STP
 10: SCS
 11: PD1

Bit 1~0 **PDS01~PDS00:** PD0 Pin-Shared function selection
 00: PD0
 01: PTP
 10: DACO
 11: PD0

• **PDS1 Register**

Bit	7	6	5	4	3	2	1	0
Name	PDS17	PDS16	PDS15	PDS14	PDS13	PDS12	PDS11	PDS10
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~6 **PDS17~PDS16:** PD7 Pin-Shared function selection

00: PD7
01: SEG12
10: PD7
11: PD7

Bit 5~4 **PDS15~PDS14:** PD6 Pin-Shared function selection

00: PD6
01: SEG11
10: V2
11: PD6

Bit 3~2 **PDS13~PDS12:** PD5 Pin-Shared function selection

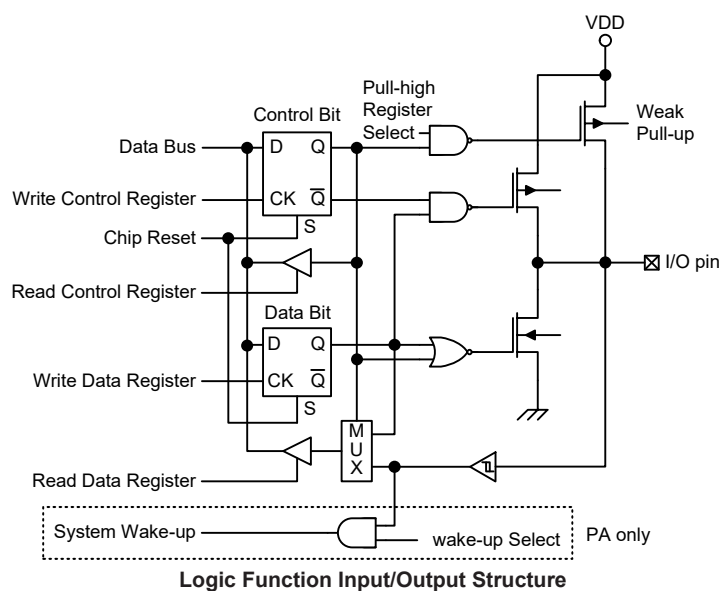
00: PD5
01: SEG10
10: C2
11: PD5

Bit 1~0 **PDS11~PDS10:** PD4 Pin-Shared function selection

00: PD4
01: SEG9
10: C1
11: PD4

I/O Pin Structures

The accompanying diagram illustrates the internal structure of the I/O logic function. As the exact logical construction of the I/O pin will differ from this drawing, it is supplied as a guide only to assist with the functional understanding of the I/O logic function. The wide range of pin-shared structures does not permit all types to be shown.



Programming Considerations

Within the user program, one of the first things to consider is port initialisation. After a reset, all of the I/O data and port control registers will be set high. This means that all I/O pins will default to an input state, the level of which depends on the other connected circuitry and whether pull-high selections have been chosen. If the port control registers are then programmed to setup some pins as outputs, these output pins will have an initial high output value unless the associated port data registers are first programmed. Selecting which pins are inputs and which are outputs can be achieved byte-wide by loading the correct values into the appropriate port control register or by programming individual bits in the port control register using the “SET [m].i” and “CLR [m].i” instructions. Note that when using these bit control instructions, a read-modify-write operation takes place. The microcontroller must first read in the data on the entire port, modify it to the required new bit values and then rewrite this data back to the output ports.

Port A has the additional capability of providing wake-up function. When the device is in the SLEEP or IDLE mode, various methods are available to wake the device up. One of these is a high to low transition of any of the Port A pins. Single or multiple pins on Port A can be setup to have this function.

Timer Modules – TM

One of the most fundamental functions in any microcontroller device is the ability to control and measure time. To implement time related functions the device includes two Timer Modules, abbreviated to the name TM. The TMs are multi-purpose timing units and serve to provide operations such as Timer/Counter, Compare Match Output and Single Pulse Output as well as being the functional unit for the generation of PWM signals. Each of the TMs has two individual interrupts. The addition of input and output pins for each TM ensures that users are provided with timing units with a wide and flexible range of features.

The common features of the different TM types are described here with more detailed information provided in the individual Standard and Periodic Type TM sections.

Introduction

The device contains two TMs and each individual TM can be categorised as a certain type, namely Standard Type TM and Periodic Type TM. Although similar in nature, the different TM types vary in their feature complexity. The common features to both of the Standard and Periodic TMs will be described in this section and the detailed operation regarding each of the TM types will be described in separate sections. The main features and differences between the two types of TMs are summarised in the accompanying table.

TM Function	STM	PTM
Timer/Counter	√	√
Compare Match Output	√	√
PWM Output	√	√
Single Pulse Output	√	√
PWM Alignment	Edge	Edge
PWM Adjustment Period & Duty	Duty or Period	Duty or Period

TM Function Summary

TM Operation

The different types of TM offer a diverse range of functions, from simple timing operations to PWM signal generation. The key to understanding how the TM operates is to see it in terms of a free running counter whose value is then compared with the value of pre-programmed internal comparators. When the free running count-up counter has the same value as the pre-programmed comparator, known as a compare match situation, a TM interrupt signal will be generated which can clear the counter and perhaps also change the condition of the TM output pin. The internal TM counter is driven by a user selectable clock source, which can be an internal clock or an external pin.

TM Clock Source

The clock source which drives the main counter in each TM can originate from various sources. The selection of the required clock source is implemented using the xTCK2~xTCK0 bits in the xTM control registers, where “x” stands for S or P type TM. The clock source can be a ratio of the system clock f_{SYS} or the internal high frequency clock f_{IH} , the f_{SUB} clock source or the external xTCK pin. The xTCK pin clock source is used to allow an external signal to drive the TM as an external clock source or for event counting.

TM Interrupts

The Standard and Periodic type TMs each have two internal interrupts, one for each of the internal comparator A or comparator P, which generate a TM interrupt when a compare match condition occurs. When a TM interrupt is generated it can be used to clear the counter and also to change the state of the TM output pin.

TM External Pins

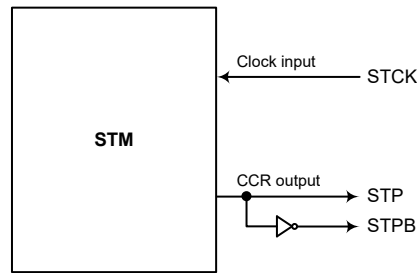
Each of the TMs, irrespective of what type, has one TM input pin, with the label xTCK. The xTM input pin, xTCK, is essentially a clock source for the xTM and is selected using the xTCK2~xTCK0 bits in the xTMC0 register. This external TM input pin allows an external clock source to drive the internal TM. The xTCK input pin can be chosen to have either a rising or falling active edge. The xTCK pin is also used as the external trigger input pin in single pulse output mode.

The TMs each have two output pins, xTP and xTPB. The xTPB is the inverted signal of the xTP output. When the TM is in the Compare Match Output Mode, these pins can be controlled by the TM to switch to a high or low level or to toggle when a compare match situation occurs. The external xTP and xTPB output pin are also the pins where the TM generates the PWM output waveform.

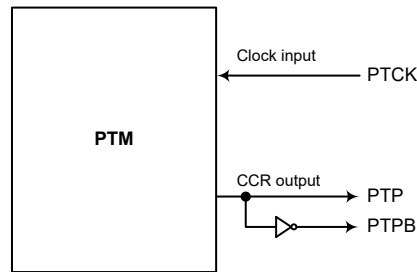
As the TM input and output pins are pin-shared with other functions, the TM input and output functions must first be setup using the relevant pin-shared function selection bits described in the Pin-shared Function section. The details of the pin-shared function selection are described in the pin-shared function section.

STM		PTM	
Input	Output	Input	Output
STCK	STP, STPB	PTCK	PTP, PTPB

TM External Pins



STM Function Pin Block Diagram

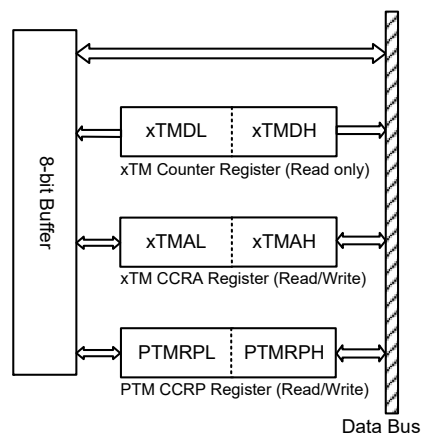


PTM Function Pin Block Diagram

Programming Considerations

The TM Counter Registers and the Compare CCRA and CCRP registers, all have a low and high byte structure. The high bytes can be directly accessed, but as the low bytes can only be accessed via an internal 8-bit buffer, reading or writing to these register pairs must be carried out in a specific way. The important point to note is that data transfer to and from the 8-bit buffer and its related low byte only takes place when a write or read operation to its corresponding high byte is executed.

As the CCRA and CCRP registers are implemented in the way shown in the following diagram and accessing these register pairs is carried out in a specific way as described above, it is recommended to use the “MOV” instruction to access the CCRA and CCRP low byte registers, named xTMAL, PTMRPL, using the following access procedures. Accessing the CCRA or CCRP low byte register without following these access procedures will result in unpredictable values.

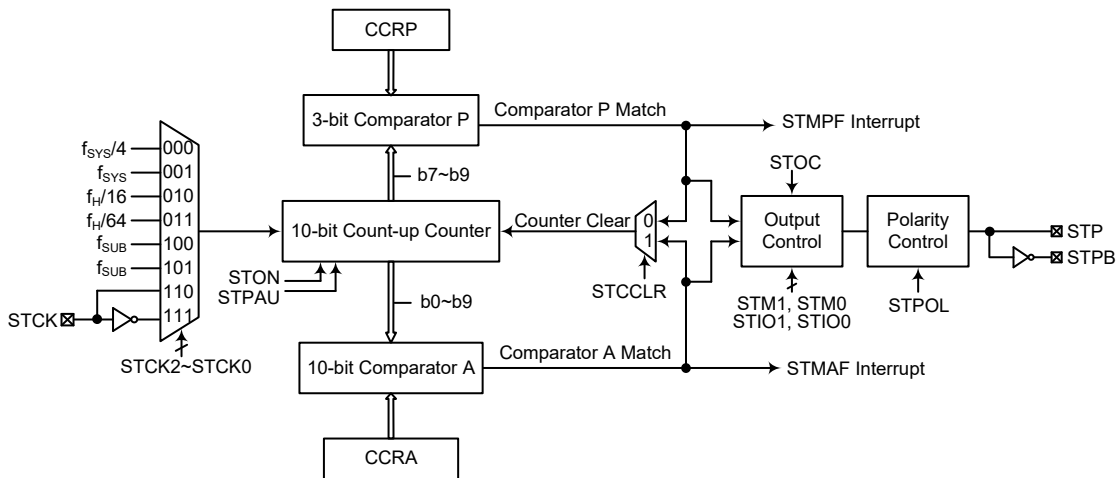


The following steps show the read and write procedures:

- Writing Data to CCRA or CCRP
 - ♦ Step 1. Write data to Low Byte xTMAL or PTMRPL
 - Note that here data is only written to the 8-bit buffer.
 - ♦ Step 2. Write data to High Byte xTMAH or PTMRPH
 - Here data is written directly to the high byte registers and simultaneously data is latched from the 8-bit buffer to the Low Byte registers.
- Reading Data from the Counter Registers, CCRA or CCRP
 - ♦ Step 1. Read data from the High Byte xTMDH, xTMAH or PTMRPH
 - Here data is read directly from the High Byte registers and simultaneously data is latched from the Low Byte register into the 8-bit buffer.
 - ♦ Step 2. Read data from the Low Byte xTMDL, xTMAL or PTMRPL
 - This step reads data from the 8-bit buffer.

Standard Type TM – STM

The Standard Type TM contains four operating modes, which are Compare Match Output, Timer/Event Counter, Single Pulse Output and PWM Output modes. The Standard TM can also be controlled with one external input pin and can drive two external output pins.



Note: The STM external pins are pin-shared with other functions, so before using the STM function, ensure that the pin-shared function registers have been set properly to enable the STM pin function. The STCK pin, if used, must also be set as an input by setting the corresponding bits in the port control register.

10-bit Standard Type TM Block Diagram

Standard TM Operation

The size of Standard type TM is 10-bit wide and its core is a 10-bit count-up counter which is driven by a user selectable internal or external clock source. There are also two internal comparators with the names, Comparator A and Comparator P. These comparators will compare the value in the counter with CCRP and CCRA registers. The CCRP comparator is 3-bit wide whose value is compared with the highest 3 bits in the counter while the CCRA is 10 bits and therefore compares all counter bits.

The only way of changing the value of the 10-bit counter using the application program, is to clear the counter by changing the STON bit from low to high. The counter will also be cleared

automatically by a counter overflow or a compare match with one of its associated comparators. When these conditions occur, an STM interrupt signal will also usually be generated. The Standard Type TM can operate in a number of different operational modes, can be driven by different clock sources including an input pin and can also control more than one output pins. All operating setup conditions are selected using relevant internal registers.

Standard Type TM Register Description

Overall operation of the Standard TM is controlled using a series of registers. A read only register pair exists to store the internal counter 10-bit value, while a read/write register pair exists to store the internal 10-bit CCRA value. The remaining two registers are control registers which setup the different operating and control modes as well as the 3-bit CCRP value.

Register Name	Bit							
	7	6	5	4	3	2	1	0
STMC0	STPAU	STCK2	STCK1	STCK0	STON	STRP2	STRP1	STRP0
STMC1	STM1	STM0	STIO1	STIO0	STOC	STPOL	STDPX	STCCLR
STMDL	D7	D6	D5	D4	D3	D2	D1	D0
STMDH	—	—	—	—	—	—	D9	D8
STMAL	D7	D6	D5	D4	D3	D2	D1	D0
STMAH	—	—	—	—	—	—	D9	D8

10-bit Standard Type TM Register List

• STMC0 Register

Bit	7	6	5	4	3	2	1	0
Name	STPAU	STCK2	STCK1	STCK0	STON	STRP2	STRP1	STRP0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7 **STPAU**: STM Counter Pause Control
0: Run
1: Pause

The counter can be paused by setting this bit high. Clearing the bit to zero restores normal counter operation. When in a Pause condition the STM will remain powered up and continue to consume power. The counter will retain its residual value when this bit changes from low to high and resume counting from this value when the bit changes to a low value again.

Bit 6~4 **STCK2~STCK0**: Select STM Counter Clock
000: $f_{SYS}/4$
001: f_{SYS}
010: $f_H/16$
011: $f_H/64$
100: f_{SUB}
101: f_{SUB}
110: STCK rising edge clock
111: STCK falling edge clock

These three bits are used to select the clock source for the STM. The external pin clock source can be chosen to be active on the rising or falling edge. The clock source f_{SYS} is the system clock, while f_H and f_{SUB} are other internal clocks, the details of which can be found in the “Operating Modes and System Clocks” section.

Bit 3 **STON**: STM Counter On/Off Control
0: Off
1: On

This bit controls the overall on/off function of the STM. Setting the bit high enables the counter to run while clearing the bit disables the STM. Clearing this bit to zero will stop the counter from counting and turn off the STM which will reduce its power consumption. When the bit changes state from low to high the internal counter value will be reset to zero, however when the bit changes from high to low, the internal counter will retain its residual value until the bit returns high again.

If the STM is in the Compare Match Output Mode then the STM output pin will be reset to its initial condition, as specified by the STOC bit, when the STON bit changes from low to high.

Bit 2~0 **STRP2~STRP0**: STM CCRP 3-bit Register, compared with the STM counter bit 9 ~ bit 7

Comparator P Match Period=

- 000: 1024 STM clocks
- 001: 128 STM clocks
- 010: 256 STM clocks
- 011: 384 STM clocks
- 100: 512 STM clocks
- 101: 640 STM clocks
- 110: 768 STM clocks
- 111: 896 STM clocks

These three bits are used to setup the value on the internal CCRP 3-bit register, which are then compared with the internal counter's highest three bits. The result of this comparison can be selected to clear the internal counter if the STCCLR bit is set to zero. Clearing the STCCLR bit to zero ensures that a compare match with the CCRP values will reset the internal counter. As the CCRP bits are only compared with the highest three counter bits, the compare values exist in 128 clock cycle multiples. Clearing all three bits to zero is in effect allowing the counter to overflow at its maximum value.

• STMC1 Register

Bit	7	6	5	4	3	2	1	0
Name	STM1	STM0	STIO1	STIO0	STOC	STPOL	STDPX	STCCLR
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~6 **STM1~STM0**: Select STM Operating Mode

- 00: Compare Match Output Mode
- 01: Undefined
- 10: PWM Output Mode or Single Pulse Output Mode
- 11: Timer/Counter Mode

These bits setup the required operating mode for the STM. To ensure reliable operation the STM should be switched off before any changes are made to the STM1 and STM0 bits. In the Timer/Counter Mode, the STM output pin state is undefined.

Bit 5~4 **STIO1~STIO0**: Select STM External Pins Function

Compare Match Output Mode

- 00: No change
- 01: Output low
- 10: Output high
- 11: Toggle output

PWM Output Mode/Single Pulse Output Mode

- 00: PWM output inactive state
- 01: PWM output active state
- 10: PWM output
- 11: Single Pulse Output

Timer/Counter Mode

Unused

These two bits are used to determine how the STM external pin changes state when a certain condition is reached. The function that these bits select depends upon in which mode the STM is running.

In the Compare Match Output Mode, the STIO1 and STIO0 bits determine how the STM output pin changes state when a compare match occurs from the Comparator A. The TM output pin can be setup to switch high, switch low or to toggle its present state when a compare match occurs from the Comparator A. When the bits are both zero, then no change will take place on the output. The initial value of the STM output pin should be setup using the STOC bit in the STMC1 register. Note that the output level requested by the STIO1 and STIO0 bits must be different from the initial value setup using the STOC bit otherwise no change will occur on the STM output pin when a compare match occurs. After the STM output pin changes state, it can be reset to its initial level by changing the level of the STON bit from low to high.

In the PWM Output Mode, the STIO1 and STIO0 bits determine how the STM output pin changes state when a certain compare match condition occurs. The PWM output function is modified by changing these two bits. It is necessary to change the values of the STIO1 and STIO0 bits only after the STM has been switched off. Unpredictable PWM outputs will occur if the STIO1 and STIO0 bits are changed when the STM is running.

Bit 3 STOC: STM STP Output Control

Compare Match Output Mode

0: Initial low

1: Initial high

PWM Output Mode/Single Pulse Output Mode

0: Active low

1: Active high

This is the output control bit for the STM output pin. Its operation depends upon whether STM is being used in the Compare Match Output Mode or in the PWM Output Mode/Single Pulse Output Mode. It has no effect if the STM is in the Timer/Counter Mode. In the Compare Match Output Mode it determines the logic level of the STM output pin before a compare match occurs. In the PWM Output Mode it determines if the PWM signal is active high or active low. In the Single Pulse Output Mode it determines the logic level of the STP output pin when the STON bit changes from low to high.

Bit 2 STPOL: STP Output Polarity Control

0: Non-invert

1: Invert

This bit controls the polarity of the STP output pin. When the bit is set high the STM output pin will be inverted and not inverted when the bit is zero. It has no effect if the STM is in the Timer/Counter Mode.

Bit 1 STDPX: STM PWM Duty/Period Control

0: CCRP – period; CCRA – duty

1: CCRP – duty; CCRA – period

This bit determines which of the CCRA and CCRP registers are used for period and duty control of the PWM waveform.

Bit 0 STCCLR: STM Counter Clear Condition Selection

0: Comparator P match

1: Comparator A match

This bit is used to select the method which clears the counter. Remember that the Standard TM contains two comparators, Comparator A and Comparator P, either of which can be selected to clear the internal counter. With the STCCLR bit set high, the counter will be cleared when a compare match occurs from the Comparator A. When the bit is low, the counter will be cleared when a compare match occurs from the Comparator P or with a counter overflow. A counter overflow clearing method can only be implemented if the CCRP bits are all cleared to zero. The STCCLR bit is not used in the PWM Output or Single Pulse Output Mode.

• **STMDL Register**

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R	R	R	R	R	R	R	R
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D7~D0:** STM Counter Low Byte Register bit 7 ~ bit 0
STM 10-bit Counter bit 7 ~ bit 0

• **STMDH Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	D9	D8
R/W	—	—	—	—	—	—	R	R
POR	—	—	—	—	—	—	0	0

Bit 7~2 Unimplemented, read as “0”
Bit 1~0 **D9~D8:** STM Counter High Byte Register bit 1 ~ bit 0
STM 10-bit Counter bit 9 ~ bit 8

• **STMAL Register**

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D7~D0:** STM CCRA Low Byte Register bit 7 ~ bit 0
STM 10-bit CCRA bit 7 ~ bit 0

• **STMAH Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	D9	D8
R/W	—	—	—	—	—	—	R/W	R/W
POR	—	—	—	—	—	—	0	0

Bit 7~2 Unimplemented, read as “0”
Bit 1~0 **D9~D8:** STM CCRA High Byte Register bit 1 ~ bit 0
STM 10-bit CCRA bit 9 ~ bit 8

Standard Type TM Operation Modes

The Standard Type TM can operate in one of four operating modes, Compare Match Output Mode, PWM Output Mode, Single Pulse Output Mode or Timer/Counter Mode. The operating mode is selected using the STM1 and STM0 bits in the STMC1 register.

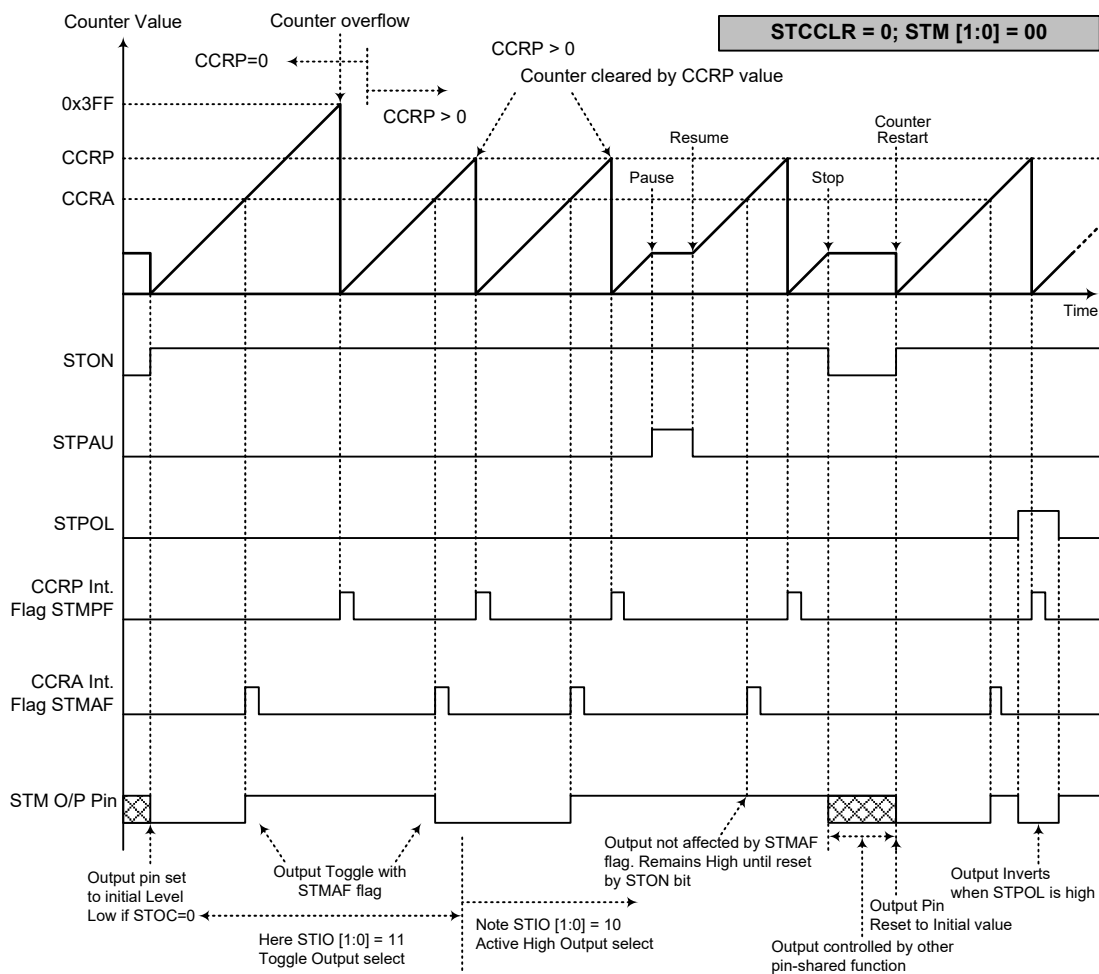
Compare Match Output Mode

To select this mode, bits STM1 and STM0 in the STMC1 register, should be set to 00 respectively. In this mode once the counter is enabled and running it can be cleared by three methods. These are a counter overflow, a compare match from Comparator A and a compare match from Comparator P. When the STCCLR bit is low, there are two ways in which the counter can be cleared. One is when a compare match from Comparator P, the other is when the CCRP bits are all zero which allows the counter to overflow. Here both STMAF and STMPF interrupt request flags for Comparator A and Comparator P respectively, will both be generated.

If the STCCLR bit in the STMC1 register is high then the counter will be cleared when a compare match occurs from Comparator A. However, here only the STMAF interrupt request flag will be generated even if the value of the CCRP bits is less than that of the CCRA registers. Therefore when STCCLR is high no STMPF interrupt request flag will be generated. In the Compare Match Output Mode, the CCRA can not be set to “0”.

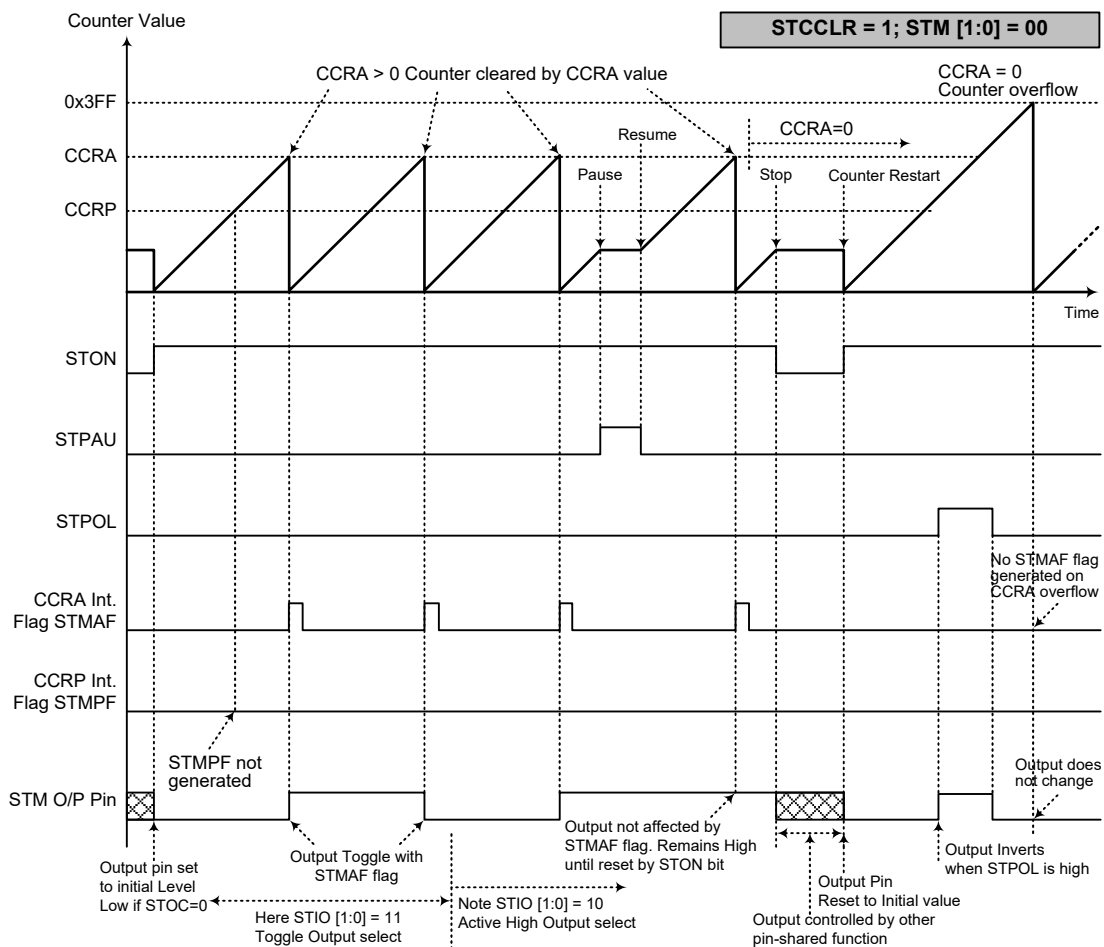
If the CCRA bits are all zero, the counter will overflow when it reaches its maximum 10-bit, 3FF Hex, value, however here the STMAF interrupt request flag will not be generated.

As the name of the mode suggests, after a comparison is made, the STM output pin, will change state. The STM output pin condition however only changes state when a STMAF interrupt request flag is generated after a compare match occurs from Comparator A. The STMPF interrupt request flag, generated from a compare match occurs from Comparator P, will have no effect on the STM output pin. The way in which the STM output pin changes state are determined by the condition of the STIO1 and STIO0 bits in the STMC1 register. The STM output pin can be selected using the STIO1 and STIO0 bits to go high, to go low or to toggle from its present condition when a compare match occurs from Comparator A. The initial condition of the STM output pin, which is setup after the STON bit changes from low to high, is setup using the STOC bit. Note that if the STIO1 and STIO0 bits are zero then no pin change will take place.



Compare Match Output Mode – STCCLR=0

- Note: 1. With STCCLR=0 a Comparator P match will clear the counter
2. The STM output pin is controlled only by the STMAF flag
3. The output pin is reset to its initial state by an STON bit rising edge



Compare Match Output Mode – STCCLR=1

- Note: 1. With STCCLR=1 a Comparator A match will clear the counter
2. The STM output pin is controlled only by the STMAF flag
3. The output pin is reset to its initial state by an STON bit rising edge
4. An STMPF flag is not generated when STCCLR=1

Timer/Counter Mode

To select this mode, bits STM1 and STM0 in the STMC1 register should be set to 11 respectively. The Timer/Counter Mode operates in an identical way to the Compare Match Output Mode generating the same interrupt flags. The exception is that in the Timer/Counter Mode the STM output pin is not used. Therefore the above description and Timing Diagrams for the Compare Match Output Mode can be used to understand its function. As the STM output pin is not used in this mode, the pin can be used as a normal I/O pin or other pin-shared function.

PWM Output Mode

To select this mode, bits STM1 and STM0 in the STMC1 register should be set to 10 respectively. The PWM function within the STM is useful for applications which require functions such as motor control, heating control, illumination control etc. By providing a signal of fixed frequency but of varying duty cycle on the STM output pin, a square wave AC waveform can be generated with varying equivalent DC RMS values.

As both the period and duty cycle of the PWM waveform can be controlled, the choice of generated waveform is extremely flexible. In the PWM Output mode, the STCCLR bit has no effect as the PWM period. Both of the CCRA and CCRP registers are used to generate the PWM waveform, one register is used to clear the internal counter and thus control the PWM waveform frequency, while the other one is used to control the duty cycle. Which register is used to control either frequency or duty cycle is determined using the STDPX bit in the STMC1 register. The PWM waveform frequency and duty cycle can therefore be controlled by the values in the CCRA and CCRP registers.

An interrupt flag, one for each of the CCRA and CCRP, will be generated when a compare match occurs from either Comparator A or Comparator P. The STOC bit in the STMC1 register is used to select the required polarity of the PWM waveform while the two STIO1 and STIO0 bits are used to enable the PWM output or to force the STM output pin to a fixed high or low level. The STPOL bit is used to reverse the polarity of the PWM output waveform.

• 10-bit STM, PWM Output Mode, Edge-aligned Mode, STDPX=0

CCRP	1~7	0
Period	CCRP×128	1024
Duty	CCRA	

If $f_{SYS}=4\text{MHz}$, STM clock source is $f_{SYS}/4$, CCRP=4 and CCRA=128,

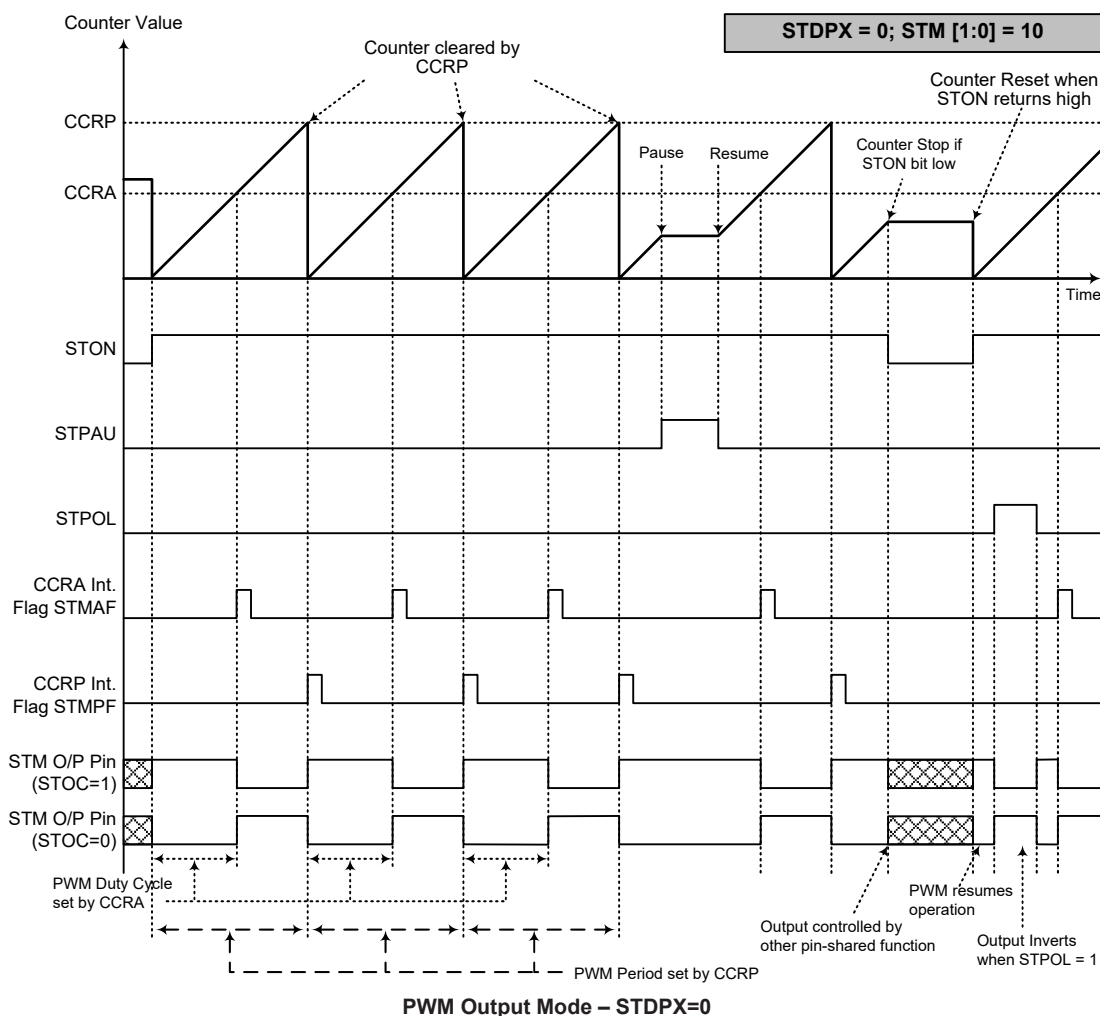
The STM PWM output frequency= $(f_{SYS}/4)/(4 \times 128)=f_{SYS}/2048=2\text{kHz}$, duty= $128/(4 \times 128)=25\%$.

If the Duty value defined by the CCRA register is equal to or greater than the Period value, then the PWM output duty is 100%.

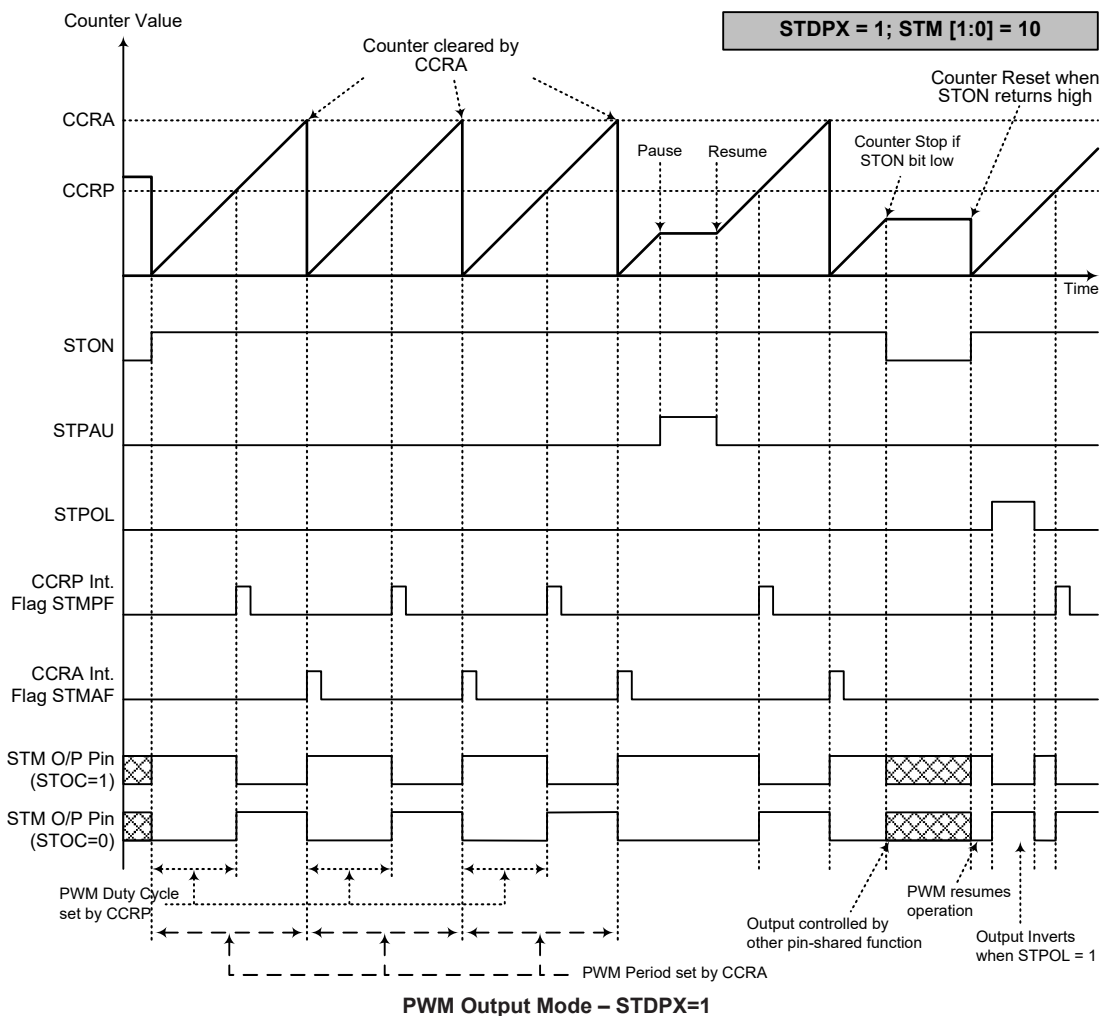
• 10-bit STM, PWM Output Mode, Edge-aligned Mode, STDPX=1

CCRP	1~7	0
Period	CCRA	
Duty	CCRP×128	1024

The PWM output period is determined by the CCRA register value together with the STM clock while the PWM duty cycle is defined by the CCRP register value.



- Note: 1. Here STDPX=0 – Counter cleared by CCRP
2. A counter clear sets the PWM Period
3. The internal PWM function continues running even when STIO[1:0]=00 or 01
4. The STCCLR bit has no influence on PWM operation



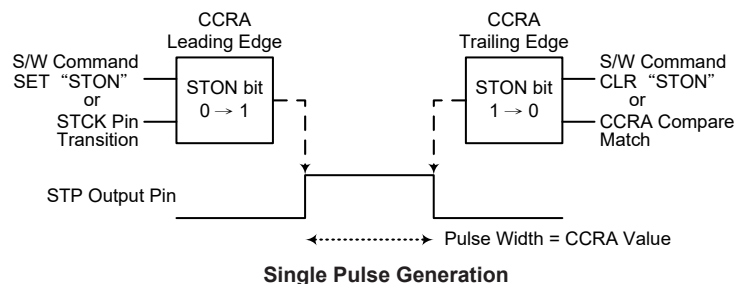
- Note: 1. Here STDPX=1 – Counter cleared by CCRA
 2. A counter clear sets the PWM Period
 3. The internal PWM function continues even when STIO[1:0]=00 or 01
 4. The STCCLR bit has no influence on PWM operation

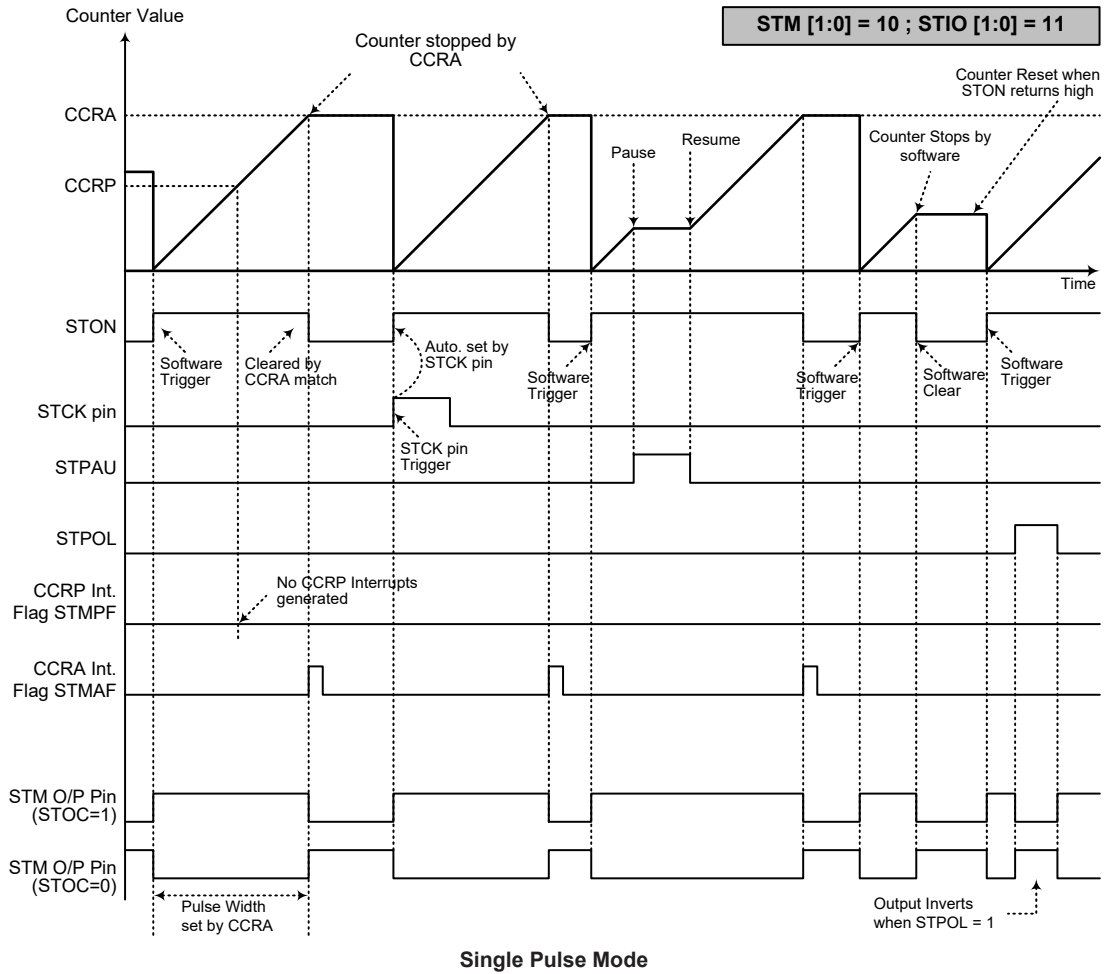
Single Pulse Output Mode

To select this mode, bits STM1 and STM0 in the STMC1 register should be set to 10 respectively and also the STIO1 and STIO0 bits should be set to 11 respectively. The Single Pulse Output Mode, as the name suggests, will generate a single shot pulse on the STM output pin.

The trigger for the pulse output leading edge is a low to high transition of the STON bit, which can be implemented using the application program. However in the Single Pulse Mode, the STON bit can also be made to automatically change from low to high using the external STCK pin, which will in turn initiate the Single Pulse output. When the STON bit transitions to a high level, the counter will start running and the pulse leading edge will be generated. The STON bit should remain high when the pulse is in its active state. The generated pulse trailing edge will be generated when the STON bit is cleared to zero, which can be implemented using the application program or when a compare match occurs from Comparator A.

However a compare match from Comparator A will also automatically clear the STON bit and thus generate the Single Pulse output trailing edge. In this way the CCRA value can be used to control the pulse width. A compare match from Comparator A will also generate a STM interrupt. The counter can only be reset back to zero when the STON bit changes from low to high when the counter restarts. In the Single Pulse Mode CCRP is not used. The STCCLR and STDPX bits are not used in this mode.

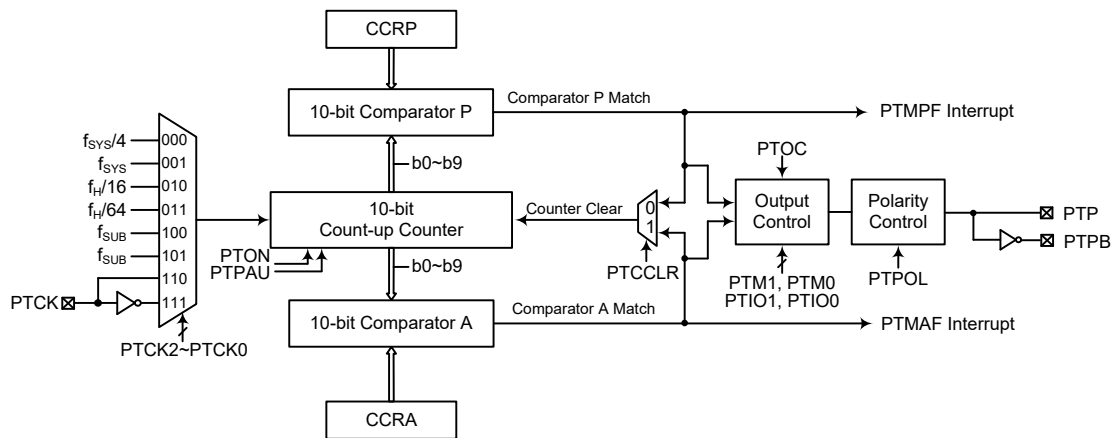




- Note:
1. Counter stopped by CCRA
 2. CCRP is not used
 3. The pulse triggered by the STCK pin or by setting the STON bit high
 4. An STCK pin active edge will automatically set the STON bit high
 5. In the Single Pulse Mode, STIO[1:0] must be set to "11" and can not be changed

Periodic Type TM – PTM

The Periodic Type TM contains four operating modes, which are Compare Match Output, Timer/Event Counter, Single Pulse Output and PWM Output modes. The Periodic TM can be controlled with one external input pin and can drive two external output pins.



Note: The PTM external pins are pin-shared with other functions, so before using the PTM function, ensure that the pin-shared function registers have been set properly to ensure the PTM pin function. The PTCK pin, if used, must also be set as an input by setting the corresponding bits in the port control register.

10-bit Periodic Type TM Block Diagram

Periodic TM Operation

The Periodic Type TM core is a 10-bit count-up counter which is driven by a user selectable internal or external clock source. There are also two internal comparators with the names, Comparator A and Comparator P. These comparators will compare the value in the counter with CCRP and CCRA registers. The CCRA and CCRP comparators are 10-bit wide whose value is respectively compared with all counter bits.

The only way of changing the value of the 10-bit counter using the application program, is to clear the counter by changing the PTON bit from low to high. The counter will also be cleared automatically by a counter overflow or a compare match with one of its associated comparators. When these conditions occur, a PTM interrupt signal will also usually be generated. The Periodic Type TM can operate in a number of different operational modes and can be driven by different clock sources and also control more than one output pins. All operating setup conditions are selected using relevant internal registers.

Periodic Type TM Register Description

Overall operation of the Periodic Type TM is controlled using a series of registers. A read only register pair exists to store the internal 10-bit counter value, while two read/write register pairs exist to store the internal 10-bit CCRA value and CCRP value. The remaining two registers are control registers which setup the different operating and control modes.

Register Name	Bit							
	7	6	5	4	3	2	1	0
PTMC0	PTPAU	PTCK2	PTCK1	PTCK0	PTON	—	—	—
PTMC1	PTM1	PTM0	PTIO1	PTIO0	PTOC	PTPOL	D1	PTCCLR
PTMDL	D7	D6	D5	D4	D3	D2	D1	D0
PTMDH	—	—	—	—	—	—	D9	D8
PTMAL	D7	D6	D5	D4	D3	D2	D1	D0
PTMAH	—	—	—	—	—	—	D9	D8
PTMRPL	D7	D6	D5	D4	D3	D2	D1	D0
PTMRPH	—	—	—	—	—	—	D9	D8

10-bit Periodic TM Register List

• **PTMC0 Register**

Bit	7	6	5	4	3	2	1	0
Name	PTPAU	PTCK2	PTCK1	PTCK0	PTON	—	—	—
R/W	R/W	R/W	R/W	R/W	R/W	—	—	—
POR	0	0	0	0	0	—	—	—

Bit 7 **PTPAU**: PTM Counter Pause Control

0: Run
1: Pause

The counter can be paused by setting this bit high. Clearing the bit to zero restores normal counter operation. When in a Pause condition the PTM will remain powered up and continue to consume power. The counter will retain its residual value when this bit changes from low to high and resume counting from this value when the bit changes to a low value again.

Bit 6~4 **PTCK2~PTCK0**: Select PTM Counter Clock

000: $f_{SYS}/4$
001: f_{SYS}
010: $f_H/16$
011: $f_H/64$
100: f_{SUB}
101: f_{SUB}
110: PTCK rising edge clock
111: PTCK falling edge clock

These three bits are used to select the clock source for the PTM. The external pin clock source can be chosen to be active on the rising or falling edge. The clock source f_{SYS} is the system clock, while f_H and f_{SUB} are other internal clocks, the details of which can be found in the “Operating Modes and System Clocks” section.

Bit 3 **PTON**: PTM Counter On/Off Control

0: Off
1: On

This bit controls the overall on/off function of the PTM. Setting the bit high enables the counter to run, clearing the bit disables the PTM. Clearing this bit to zero will stop the counter from counting and turn off the PTM which will reduce its power consumption. When the bit changes state from low to high the internal counter value will be reset to zero, however when the bit changes from high to low, the internal counter will retain its residual value until the bit returns high again.

If the PTM is in the Compare Match Output Mode, PWM output Mode or Single Pulse Output Mode then the PTM output pin will be reset to its initial condition, as specified by the PTOC bit, when the PTON bit changes from low to high.

Bit 2~0 Unimplemented, read as “0”

• **PTMC1 Register**

Bit	7	6	5	4	3	2	1	0
Name	PTM1	PTM0	PTIO1	PTIO0	PTOC	PTPOL	D1	PTCCLR
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~6 **PTM1~PTM0**: Select PTM Operating Mode

- 00: Compare Match Output Mode
- 01: Undefined
- 10: PWM Output Mode or Single Pulse Output Mode
- 11: Timer/Counter Mode

These bits setup the required operating mode for the PTM. To ensure reliable operation the PTM should be switched off before any changes are made to the PTM1 and PTM0 bits. In the Timer/Counter Mode, the PTM output pin state is undefined.

Bit 5~4 **PTIO1~PTIO0**: Select PTM External Pins Function

Compare Match Output Mode

- 00: No change
- 01: Output low
- 10: Output high
- 11: Toggle output

PWM Output Mode/Single Pulse Output Mode

- 00: PWM output inactive state
- 01: PWM output active state
- 10: PWM output
- 11: Single pulse output

Timer/Counter Mode

Unused

These two bits are used to determine how the PTM external pin functions when a certain condition is reached. The function that these bits select depends upon in which mode the PTM is running.

In the Compare Match Output Mode, the PTIO1 and PTIO0 bits determine how the PTM output changes state when a compare match occurs from the Comparator A. The PTM output can be setup to switch high, switch low or to toggle its present state when a compare match occurs from the Comparator A. When the bits are both zero, then no change will take place on the output. The initial value of the PTM output should be setup using the PTOC bit in the PTMC1 register. Note that the output level requested by the PTIO1 and PTIO0 bits must be different from the initial value setup using the PTOC bit otherwise no change will occur on the PTM output when a compare match occurs. After the PTM output changes state, it can be reset to its initial level by changing the level of the PTON bit from low to high.

In the PWM Output Mode, the PTIO1 and PTIO0 bits determine how the PTM output changes state when a certain compare match condition occurs. The PWM output function is modified by changing these two bits. It is necessary to only change the values of the PTIO1 and PTIO0 bits only after the PTM has been switched off. Unpredictable PWM outputs will occur if the PTIO1 and PTIO0 bits are changed when the PTM is running.

Bit 3 **PTOC**: PTM PTP Output Control Bit

Compare Match Output Mode

- 0: Initial low
- 1: Initial high

PWM Output Mode/Single Pulse Output Mode

- 0: Active low
- 1: Active high

This is the output control bit for the PTP output. Its operation depends upon whether PTM is being used in the Compare Match Output Mode or in the PWM Output Mode/Single Pulse Output Mode. It has no effect if the PTM is in the Timer/Counter Mode. In the Compare Match Output Mode it determines the logic level of the PTM output before a compare match occurs. In the PWM Output Mode it determines if the PWM signal is active high or active low. In the Single Pulse Output Mode it determines the logic level of the PTM output when the PTON bit changes from low to high.

Bit 2 **PTPOL**: PTM PTP Output Polarity Control
 0: Non-invert
 1: Invert

This bit controls the polarity of the PTP output pin. When the bit is set high the PTM output pin will be inverted and not inverted when the bit is zero. It has no effect if the PTM is in the Timer/Counter Mode.

Bit 1 **D1**: Reserved, must be fixed at “0”

Bit 0 **PTCCLR**: Select PTM Counter Clear Condition
 0: PTM Comparator P match
 1: PTM Comparator A match

This bit is used to select the method which clears the counter. Remember that the Periodic TM contains two comparators, Comparator A and Comparator P, either of which can be selected to clear the internal counter. With the PTCCLR bit set high, the counter will be cleared when a compare match occurs from the Comparator A. When the bit is low, the counter will be cleared when a compare match occurs from the Comparator P or with a counter overflow. A counter overflow clearing method can only be implemented if the CCRP bits are all cleared to zero. The PTCCLR bit is not used in the PWM Output Mode or Single Pulse Output Mode.

• **PTMDL Register**

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R	R	R	R	R	R	R	R
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D7~D0**: PTM Counter Low Byte Register bit 7 ~ bit 0
 PTM 10-bit Counter bit 7 ~ bit 0

• **PTMDH Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	D9	D8
R/W	—	—	—	—	—	—	R	R
POR	—	—	—	—	—	—	0	0

Bit 7~2 Unimplemented, read as “0”

Bit 1~0 **D9~D8**: PTM Counter High Byte Register bit 1 ~ bit 0
 PTM 10-bit Counter bit 9 ~ bit 8

• **PTMAL Register**

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D7~D0**: PTM CCRA Low Byte Register bit 7 ~ bit 0
PTM 10-bit CCRA bit 7 ~ bit 0

• **PTMAH Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	D9	D8
R/W	—	—	—	—	—	—	R/W	R/W
POR	—	—	—	—	—	—	0	0

Bit 7~2 Unimplemented, read as “0”
Bit 1~0 **D9~D8**: PTM CCRA High Byte Register bit 1 ~ bit 0
PTM 10-bit CCRA bit 9 ~ bit 8

• **PTMRPL Register**

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D7~D0**: PTM CCRP Low Byte Register bit 7 ~ bit 0
PTM 10-bit CCRP bit 7 ~ bit 0

• **PTMRPH Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	D9	D8
R/W	—	—	—	—	—	—	R/W	R/W
POR	—	—	—	—	—	—	0	0

Bit 7~2 Unimplemented, read as “0”
Bit 1~0 **D9~D8**: PTM CCRP High Byte Register bit 1 ~ bit 0
PTM 10-bit CCRP bit 9 ~ bit 8

Periodic Type TM Operating Modes

The Periodic Type TM can operate in one of four operating modes, Compare Match Output Mode, PWM Output Mode, Single Pulse Output Mode or Timer/Counter Mode. The operating mode is selected using the PTM1 and PTM0 bits in the PTMC1 register.

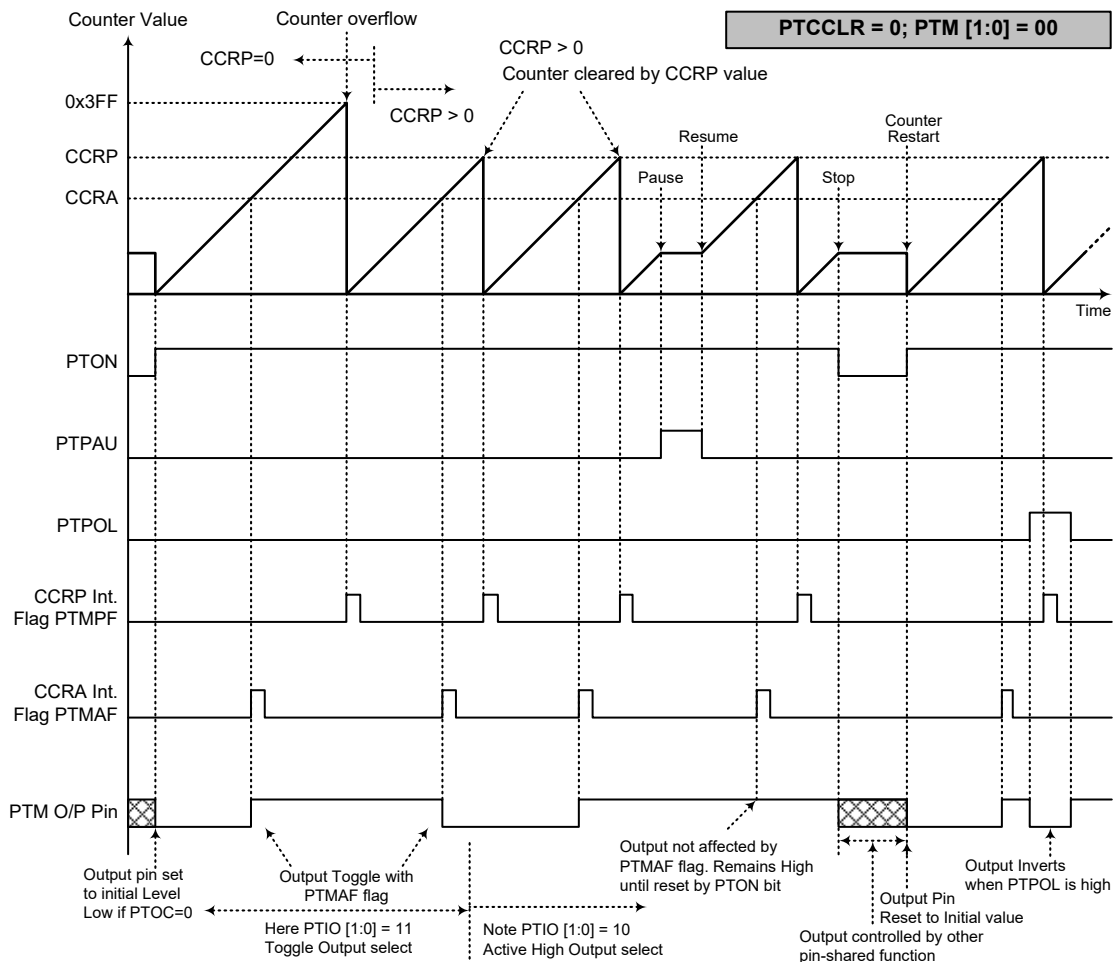
Compare Match Output Mode

To select this mode, bits PTM1 and PTM0 in the PTMC1 register should be set to 00 respectively. In this mode once the counter is enabled and running it can be cleared by three methods. These are a counter overflow, a compare match from Comparator A and a compare match from Comparator P. When the PTCCLR bit is low, there are two ways in which the counter can be cleared. One is when a compare match from Comparator P, the other is when the CCRP bits are all zero which allows the counter to overflow. Here both PTMAF and PTMPF interrupt request flags for Comparator A and Comparator P respectively, will both be generated.

If the PTCCLR bit in the PTMC1 register is high then the counter will be cleared when a compare match occurs from Comparator A. However, here only the PTMAF interrupt request flag will be generated even if the value of the CCRP bits is less than that of the CCRA registers. Therefore when PTCCLR is high no PTMPF interrupt request flag will be generated. In the Compare Match Output Mode, the CCRA can not be cleared to zero.

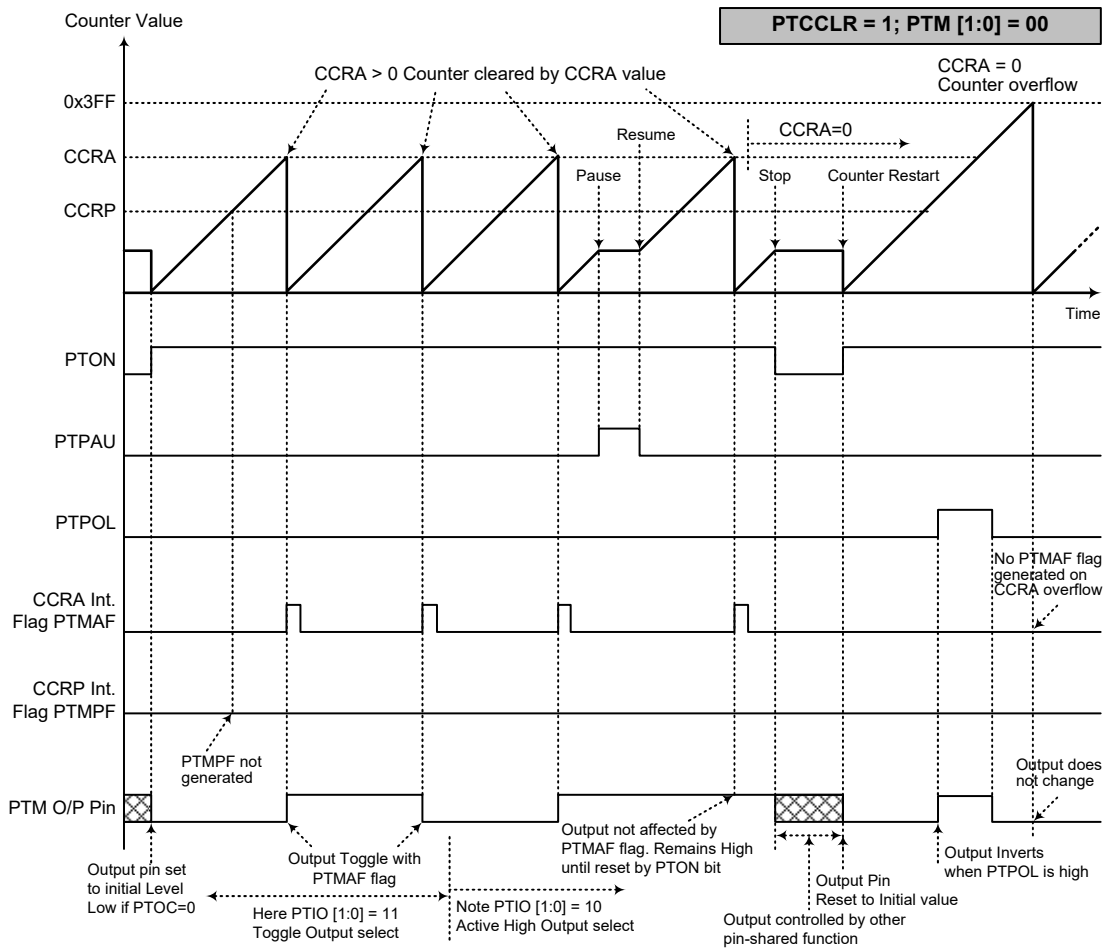
If the CCRA bits are all zero, the counter will overflow when it reaches its maximum 10-bit, 3FF Hex value, however here the PTMAF interrupt request flag will not be generated.

As the name of the mode suggests, after a comparison is made, the PTM output will change state. The PTM output condition however only changes state when a PTMAF interrupt request flag is generated after a compare match occurs from Comparator A. The PTMPF interrupt request flag, generated from a compare match occurs from Comparator P, will have no effect on the PTM output. The way in which the PTM output changes state are determined by the condition of the PTIO1 and PTIO0 bits in the PTMC1 register. The PTM output can be selected using the PTIO1 and PTIO0 bits to go high, to go low or to toggle from its present condition when a compare match occurs from Comparator A. The initial condition of the PTM output pin, which is setup after the PTON bit changes from low to high, is setup using the PTOC bit. Note that if the PTIO1 and PTIO0 bits are zero then no output change will take place.



Compare Match Output Mode – PTCCLR=0

- Note: 1. With PTCCLR=0 a Comparator P match will clear the counter
2. The PTM output pin is controlled only by the PTMAF flag
3. The output is reset to its initial state by a PTON bit rising edge



Compare Match Output Mode – PTCCLR=1

- Note: 1. With PTCCLR=1 a Comparator A match will clear the counter
2. The PTM output pin is controlled only by the PTMAF flag
3. The output pin is reset to its initial state by a PTON bit rising edge
4. A PTMPF flag is not generated when PTCCLR=1

Timer/Counter Mode

To select this mode, bits PTM1 and PTM0 in the PTMC1 register should be set to 11 respectively. The Timer/Counter Mode operates in an identical way to the Compare Match Output Mode generating the same interrupt flags. The exception is that in the Timer/Counter Mode the PTM output pins are not used. Therefore the above description and Timing Diagrams for the Compare Match Output Mode can be used to understand its function. As the PTM output pins are not used in this mode, the pins can be used as normal I/O pins or other pin-shared function.

PWM Output Mode

To select this mode, bits PTM1 and PTM0 in the PTMC1 register should be set to 10 respectively. The PWM function within the PTM is useful for applications which require functions such as motor control, heating control, illumination control etc. By providing a signal of fixed frequency but of varying duty cycle on the PTM output pin, a square wave AC waveform can be generated with varying equivalent DC RMS values.

As both the period and duty cycle of the PWM waveform can be controlled, the choice of generated waveform is extremely flexible. In the PWM Output Mode, the PTCCLR bit has no effect on the PWM operation. Both of the CCRA and CCRP registers are used to generate the PWM waveform, the CCRP is used to clear the internal counter and thus control the PWM waveform frequency, while the CCRA is used to control the duty cycle. The PWM waveform frequency and duty cycle can therefore be controlled by the values in the CCRP and CCRA registers.

An interrupt flag, one for each of the CCRA and CCRP, will be generated when a compare match occurs from either Comparator A or Comparator P. The PTOC bit in the PTMC1 register is used to select the required polarity of the PWM waveform while the two PTIO1 and PTIO0 bits are used to enable the PWM output or to force the PTM output pin to a fixed high or low level. The PTPOL bit is used to reverse the polarity of the PWM output waveform.

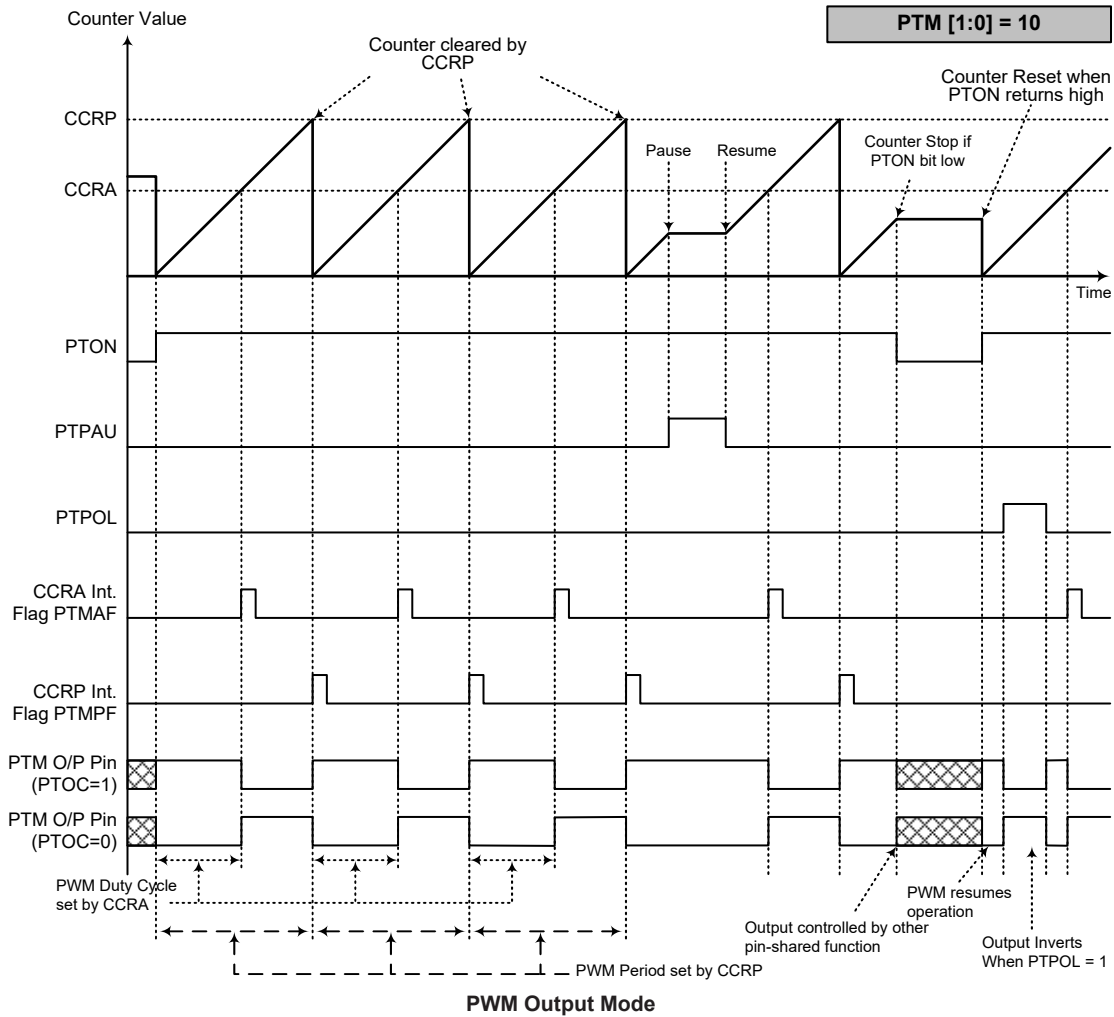
• 10-bit PTM, PWM Output Mode, Edge-aligned Mode

CCRP	1~1023	0
Period	1~1023	1024
Duty	CCRA	

If $f_{SYS}=8\text{MHz}$, PTM clock source select $f_{SYS}/4$, CCRP=512 and CCRA=128,

The PTM PWM output frequency= $(f_{SYS}/4)/512=f_{SYS}/2048=4\text{kHz}$, duty=128/512=25%.

If the Duty value defined by the CCRA register is equal to or greater than the Period value, then the PWM output duty is 100%.



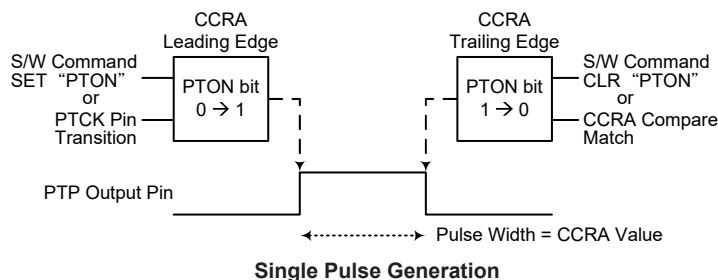
- Note:
1. Counter cleared by CCRP
 2. A counter clear sets PWM Period
 3. The internal PWM function continues running even when PTIO[1:0]=00 or 01
 4. The PTCCLR bit has no influence on PWM operation

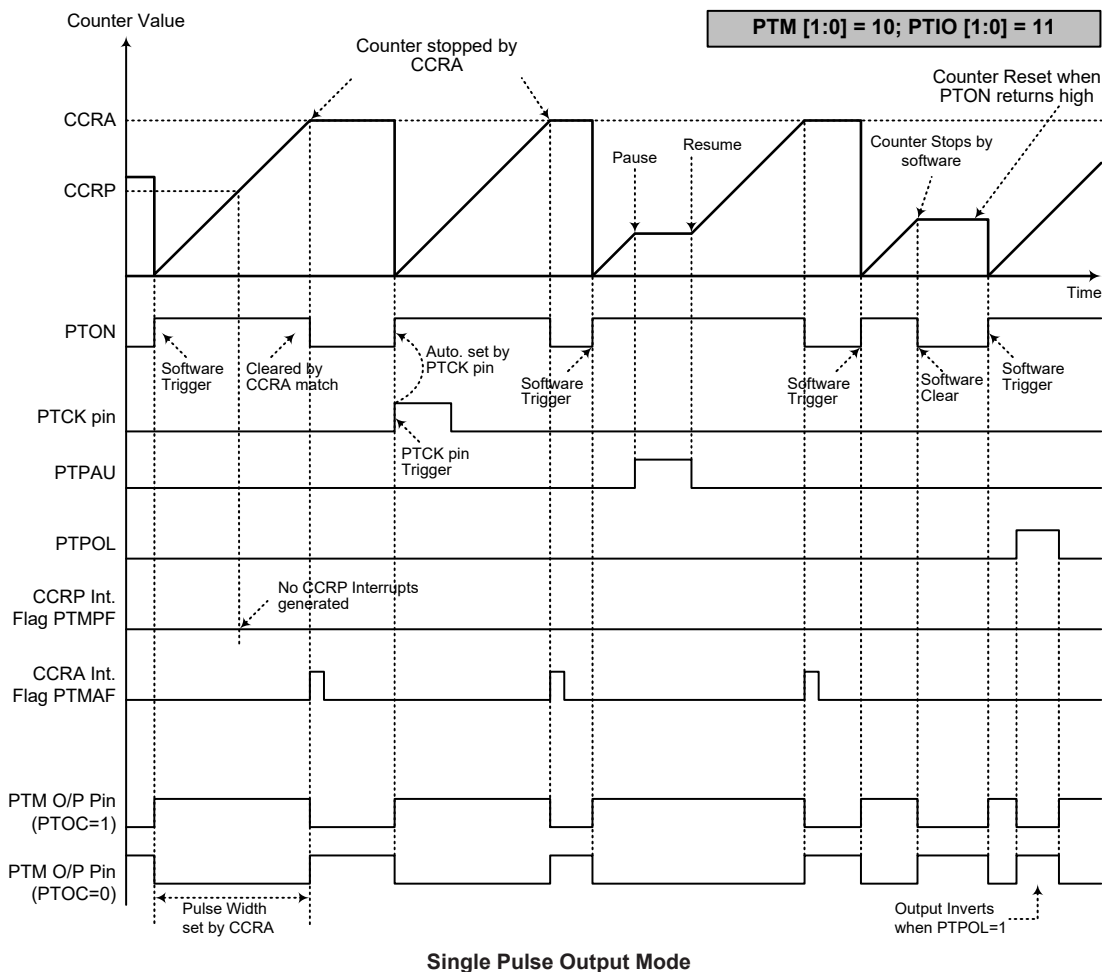
Single Pulse Output Mode

To select this mode, bits PTM1 and PTM0 in the PTMC1 register should be set to 10 respectively and also the PTIO1 and PTIO0 bits should be set to 11 respectively. The Single Pulse Output Mode, as the name suggests, will generate a single shot pulse on the PTM output pin.

The trigger for the pulse output leading edge is a low to high transition of the PTON bit, which can be implemented using the application program. However in the Single Pulse Output Mode, the PTON bit can also be made to automatically change from low to high using the external PTCK pin, which will in turn initiate the Single Pulse output. When the PTON bit transitions to a high level, the counter will start running and the pulse leading edge will be generated. The PTON bit should remain high when the pulse is in its active state. The generated pulse trailing edge will be generated when the PTON bit is cleared to zero, which can be implemented using the application program or when a compare match occurs from Comparator A.

However a compare match from Comparator A will also automatically clear the PTON bit and thus generate the Single Pulse output trailing edge. In this way the CCRA value can be used to control the pulse width. A compare match from Comparator A will also generate a PTM interrupt. The counter can only be reset back to zero when the PTON bit changes from low to high when the counter restarts. In the Single Pulse Output Mode CCRP is not used. The PTCCLR bit is not used in this mode.

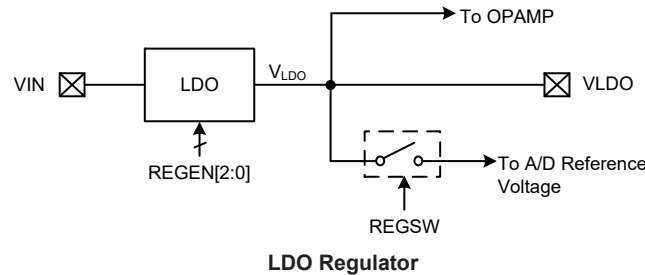




- Note:
1. Counter stopped by CCRA
 2. CCRP is not used
 3. The pulse is triggered by the PTCK pin or by setting the PTON bit high
 4. A PTCK pin active edge will automatically set the PTON bit high
 5. In the Single Pulse Mode, PTIO[1:0] must be set to "11" and cannot be changed

Voltage Regulator – LDO

The device includes a voltage regulator, LDO. The REGC register controls the regulator module to work in five modes. In the Hi-impedance mode, the LDO will be turned off and the VLDO pin will be floating. In the bypass mode, the LDO is turned off and the V_{IN} will bypass the LDO circuit and be connected to the VLDO pin directly. In the third mode the regulator is turned on, when the input voltage is larger than 2.5V, the LDO will output a fixed voltage of 2.2V on the VLDO pin. In the fourth mode the regulator is turned on, when the input voltage is larger than 2.8V, the LDO will output a fixed voltage of 2.5V. In the fifth mode the regulator is turned on, when the input voltage is larger than 3.3V, the LDO will output a fixed voltage of 3.0V. The LDO output can be used as the OPAMP power supply and A/D converter reference input.



• REGC Register

Bit	7	6	5	4	3	2	1	0
Name	REGSW	—	—	—	—	REGEN2	REGEN1	REGEN0
R/W	R/W	—	—	—	—	R/W	R/W	R/W
POR	0	—	—	—	—	0	0	0

Bit 7 **REGSW**: Switch on/off control
0: Off
1: On

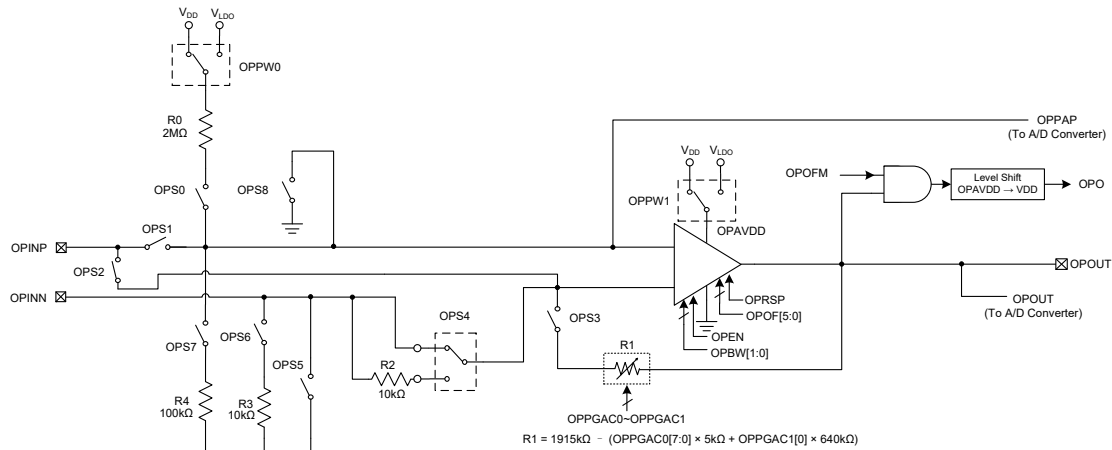
When the A/D converter reference voltage is selected to come from V_{LDO} , this bit should be set high. However when the A/D converter reference voltage is selected to come from any other voltage, other than V_{LDO} , this bit should be cleared to zero to turn off the switch, otherwise V_{LDO} will be connected together with other selected A/D reference voltage to the A/D converter simultaneously, which will result in unpredictable situations such as an irreversible damage.

Bit 6~3 Unimplemented, read as “0”

Bit 2~0 **REGEN2~REGEN0**: Regulator mode selection
x00: Regulator off in Hi-impedance mode, LDO output is floating
x01: Regulator off in Bypass mode, LDO output= V_{IN} input voltage
010: Regulator on, V_{LDO} =2.2V
011: Regulator on, V_{LDO} =2.5V
11x: Regulator on, V_{LDO} =3.0V

CO/Gas Detector AFE

The device includes a CO/Gas Detector AFE module which is mainly composed of a software configurable resistor and an Operational Amplifier, OPAMP. The operational amplifier can be used for signal amplification according to specific user requirements. This OPAMP features include enable/disable control, multiple switch and input path selections, input offset voltage calibration and four bandwidth options. In addition, the positive input and the output of the OPAMP can be converted using the internal A/D converter.



CO/Gas Detector AFE Block Diagram

CO/Gas Detector AFE Registers

The overall CO/Gas Detector AFE circuits are controlled by a series of registers. The OPSW0~OPSW1 registers are used to configure the paths by controlling a series of switches. The OPPW register is used to select the OPAMP power supply source or the input end voltage. The OPC register is used for the OPAMP enable/disable control, output status indication and bandwidth selection. The OPVOS register is used for OPAMP input offset voltage calibration control. The OPPGAC0~OPPGAC1 registers are used to setup the R1 resistance.

[illegible]

CO/Gas Detector AFE Register List

• **OPSW0 Register**

Bit	7	6	5	4	3	2	1	0
Name	OPS7	OPS6	OPS5	OPS4	OPS3	OPS2	OPS1	OPS0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

- Bit 7 **OPS7:** OPS7 switch On/Off control
0: Off
1: On
- Bit 6 **OPS6:** OPS6 switch On/Off control
0: Off
1: On
- Bit 5 **OPS5:** OPS5 switch On/Off control
0: Off
1: On
- Bit 4 **OPS4:** OPS4 switch connection selection
0: OPAMP negative input is connected to OPINN
1: OPAMP negative input is connected to R2
- Bit 3 **OPS3:** OPS3 switch On/Off control.
0: Off
1: On
- Bit 2 **OPS2:** OPS2 switch On/Off control
0: Off
1: On
- Bit 1 **OPS1:** OPS1 switch On/Off control
0: Off
1: On
- Bit 0 **OPS0:** OPS0 switch On/Off control
0: Off
1: On

• **OPSW1 Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	—	OPS8
R/W	—	—	—	—	—	—	—	R/W
POR	—	—	—	—	—	—	—	0

- Bit 7~1 Unimplemented, read as “0”
- Bit 0 **OPS8:** OPS8 switch On/Off control
0: Off
1: On

• **OPPW Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	OPPW1	OPPW0
R/W	—	—	—	—	—	—	R/W	R/W
POR	—	—	—	—	—	—	0	0

- Bit 7~2 Unimplemented, read as “0”
- Bit 1 **OPPW1:** OPAMP power selection switch 1
0: V_{DD}
1: V_{LDO}

Bit 0 **OPPW0**: OPAMP power selection switch 0
0: V_{DD}
1: V_{LDO}

• **OPC Register**

Bit	7	6	5	4	3	2	1	0
Name	—	OPEN	OPO	—	—	—	OPBW1	OPBW0
R/W	—	R/W	R	—	—	—	R/W	R/W
POR	—	0	0	—	—	—	0	0

Bit 7 Unimplemented, read as “0”

Bit 6 **OPEN**: OPAMP enable/disable control
0: Disable
1: Enable

Bit 5 **OPO**: OPAMP output status (positive logic)
This bit is read only.
When OPOFM=1, the OPO bit value indicates the OPA output status, refer to the Operational Amplifier Input Calibration section. When OPOFM=0, this bit is fixed at 0.

Bit 4~2 Unimplemented, read as “0”

Bit 1~0 **OPBW1~OPBW0**: OPAMP bandwidth selection
00: 5kHz
01: 40kHz
10: 600kHz
11: 2MHz

When the operational amplifier is used together with the 12-bit A/D converter, its bandwidth selection and A/D converter clock frequency configuration should follow this table to ensure a correct measurement. Refer to the CO/Gas Detector AFE Electrical Characteristics for more details. In this table, “√” represents that can be used.

OPBW[1:0]	A/D Converter Clock Frequency (kHz)							
	15.625	31.25	62.5	125	250	500	1000	2000
00	√	×	×	×	×	×	×	×
01	√	√	√	√	×	×	×	×
10	√	√	√	√	√	√	√	√
11	√	√	√	√	√	√	√	√

• **OPVOS Register**

Bit	7	6	5	4	3	2	1	0
Name	OPOFM	OPRSP	OPOF5	OPOF4	OPOF3	OPOF2	OPOF1	OPOF0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	1	0	0	0	0	0

Bit 7 **OPOFM**: OPAMP normal operation or input offset calibration mode selection
0: Normal operation mode
1: Input offset calibration mode

Bit 6 **OPRSP**: OPAMP input offset voltage calibration reference selection
0: OPINN is selected as reference input
1: OPINP is selected as reference input

Bit 5~0 **OPOF5~OPOF0**: OPAMP input offset voltage calibration value
This bit field is used to perform the OPAMP input offset calibration operation and the value after the input offset calibration can be restored into this bit field. Refer to the “Input Offset Calibration” section for more detailed information.

• **OPPGAC0 Register**

Bit	7	6	5	4	3	2	1	0
Name	OPPGA7	OPPGA6	OPPGA5	OPPGA4	OPPGA3	OPPGA2	OPPGA1	OPPGA0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **OPPGA7~OPPGA0**: R1 resistance control code
 $R1 = 1915 \text{ k}\Omega - (\text{OPPGAC0}[7:0] \times 5 \text{ k}\Omega + \text{OPPGAC1}[0] \times 640 \text{ k}\Omega)$

• **OPPGAC1 Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	—	OPPGA8
R/W	—	—	—	—	—	—	—	R/W
POR	—	—	—	—	—	—	—	0

Bit 7~1 Unimplemented, read as “0”

Bit 0 **OPPGA8**: R1 resistance control code
 $R1 = 1915 \text{ k}\Omega - (\text{OPPGAC0}[7:0] \times 5 \text{ k}\Omega + \text{OPPGAC1}[0] \times 640 \text{ k}\Omega)$

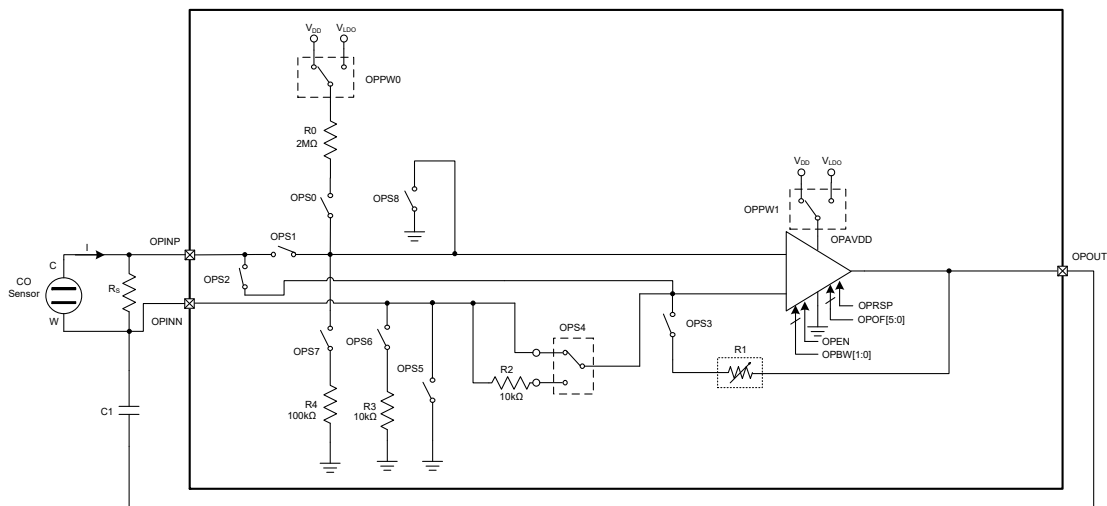
Input Offset Calibration

To operate in the input offset calibration mode, the OPAMP input pins to be used should first be enabled by setting the OPEN bit high.

- Step1: Set OPOFM=1, the OPAMP is now under offset calibration mode. To make sure V_{OS} as minimise as possible after calibration, the input reference voltage in calibration mode should be the same as input DC operating voltage in normal mode operation.
- Step2: Set OPOF[5:0]=000000, then read the OPO bit.
- Step3: Let OPOF[5:0]=OPOF[5:0]+1 then read the OPO bit.
 If the OPO bit state is not changed, repeat Step3 until the OPO bit state is changed.
 If the OPO bit state is changed, record the current OPOF[5:0] data as V_{OS1} , then go to Step4.
- Step4: Set OPOF[5:0]=111111, then read OPO bit.
- Step5: Let OPOF[5:0]=OPOF[5:0] - 1 then read the OPO bit.
 If the OPO bit state is not changed, repeat Step5 until the OPO bit state is changed.
 If the OPO bit state is changed, record the current OPOF[5:0] data as V_{OS2} , then go to Step6.
- Step6: Restore $V_{OS} = (V_{OS1} + V_{OS2}) / 2$ to OPOF[5:0] bits, the calibration is finished.
 If $(V_{OS1} + V_{OS2}) / 2$ is not integral, discard the decimal.

CO/Gas Detector AFE Application Description

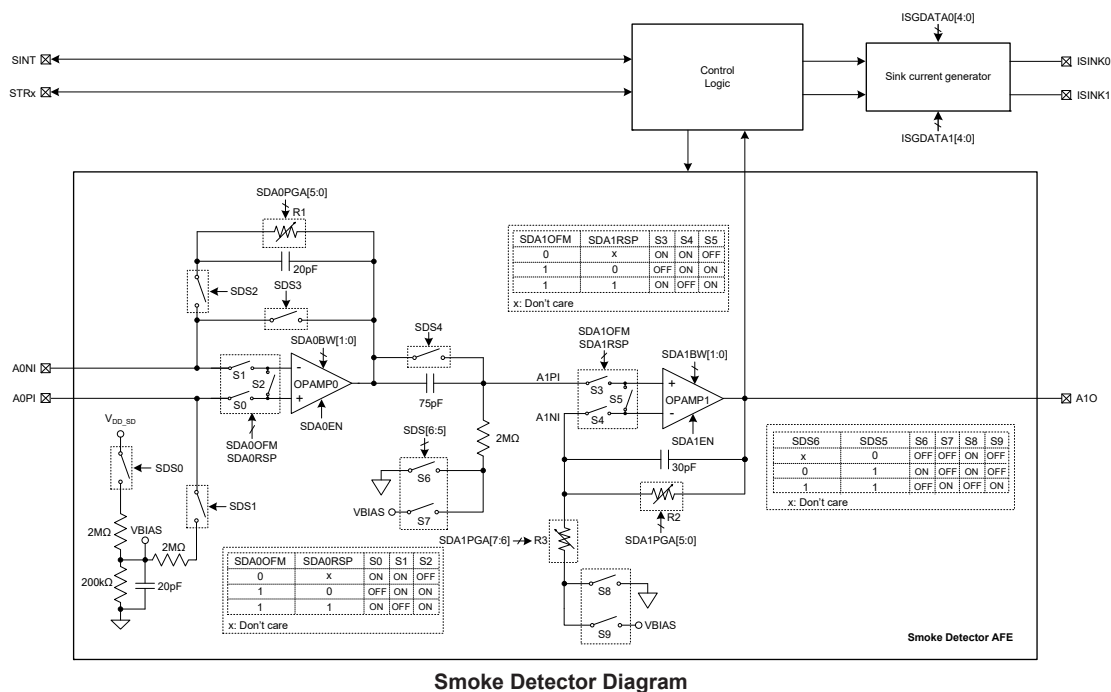
Together with proper peripheral circuits, this CO/Gas detector AFE module can be used for CO or other gas detector applications. In the application circuit below, a 100kΩ resistor (R_S) is connected to the sensor in parallel by connecting its two ends to the OPINN and OPINP pins respectively. When the sensor has detected the CO or other gases, the current will flow out of the C end. At the same time a same magnitude of current flows from the OPOUT end through the R1 resistor and then to the W end of this CO sensor. The voltage on the OPOUT pin should be equal to $(V_{OPINP} + I \times R1)$.



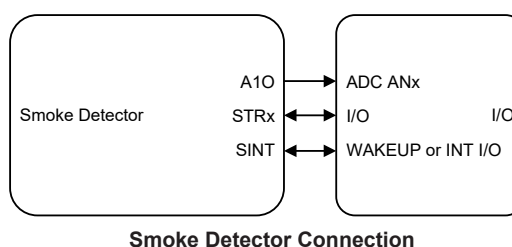
CO/Gas Detector Application Schematic

Smoke Detector – SD

The device contains a smoke detector, the block diagram is shown below. There are two Operational Amplifiers which are used for smoke detection, a programmable constant current generator providing two outputs. Users can implement the smoke detection function after using the one-wire communication interface to complete simple settings, so that customers can quickly complete smoke detector product development.



Smoke Detector Interface



The smoke detector connection is shown in the figure above. The A1O pin is the smoke detector AFE output, which can be connected to any A/D analog input pin. The STRx pin is the one-wire communication pin, which can be connected to any general purpose I/O port without interrupt requirement. The SINT pin is used to generate a wake-up signal, which should be connected to a wake-up or interrupt I/O port.

The SINT wake-up pin is valid only when the smoke detection period or time base wake-up period is larger than 0. If the period is set to 0, it is disabled, refer to the "Parameter Configuration" for details. The smoke detection period and the time base wake-up period which allow to be set with different values, are distinguished by the STRx pulse, SINT pulse and their respective low level duration. The wake-up pulse width is fixed at 15μs and the wake-up duration can be modified by relevant running variables. The wake-up signal for each case are as follows:



STRx

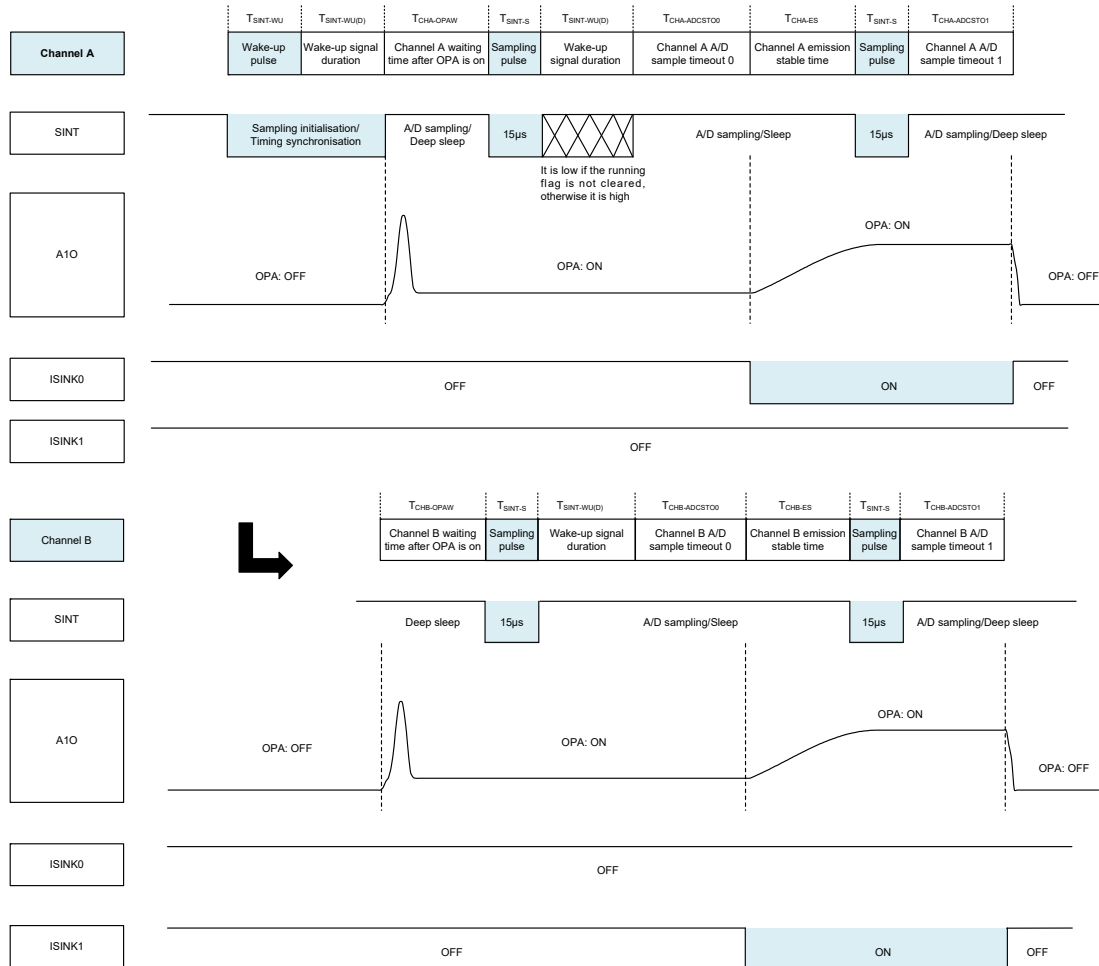
$T_{STRx-TS}$

$>40\mu s$

Trigger Sampling

Wake-up/Synchronous Sampling Timing – Method A

The smoke detection generally works in pulse mode with low power consumption. There are strict timing requirements for the AFE operation and ISINK emission, otherwise it will cause additional power consumption. In addition to the wake-up function, the SINT pin also has a synchronisation function. If the channels A and B are enabled using relevant running variables and the detection period setting is larger than 0, the SINT pin will generate sampling pulses with a pulse width of 15 μ s after executing the wake-up function to notify the A/D converter to perform sampling. Each channel has two sampling pulses, corresponding to the A1O output in the ISINK off state and the A1O output in the ISINK on state. The specific time can be modified by the relevant running variables. The following is the timing when both channels A and B are enabled.

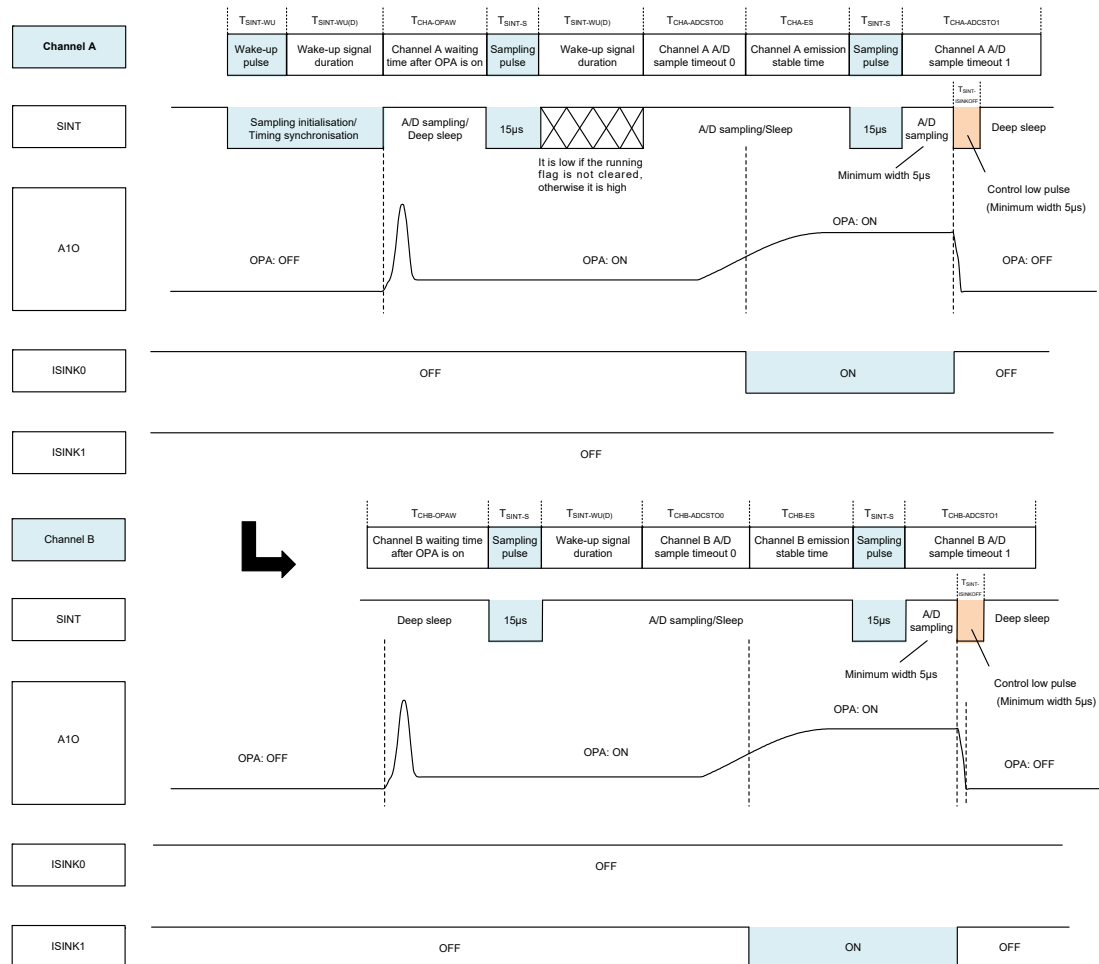


Note: 1. It is necessary to enable the channel A and channel B using the running variables and set the smoke detection period to be larger than 0 to generate a synchronous pulse. Refer to the parameter configuration section for details.

2. If only one of channel A and B is enabled, there is only one channel timing.

Wake-up/Synchronous Sampling Timing – Method B

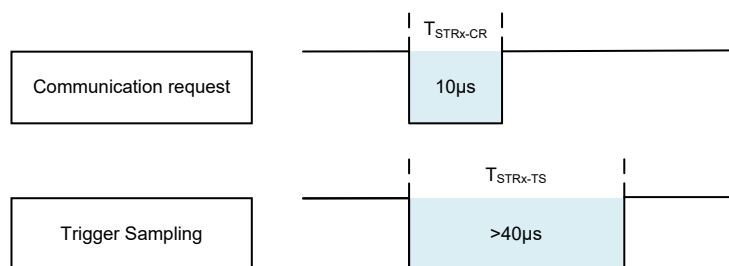
There is an additional ISINK early shutdown function for the method B to avoid unnecessary power consumption, such as delay waiting. During the ISINK emission, after the smoke detector has sent the sampling pulse, the SINT pin will remain pulled high. At this time, if the A/D sampling has been completed, send a control low pulse via the SINT pin, the ISINK can be turned off in advance without waiting until the A/D sampling timeout ends. If no such low pulse is sent, it is method A. Note that the A/D sampling time (SINT high level width) should not be less than 5 μ s and the control low pulse width should not be less than 5 μ s, otherwise they will not be recognized. The following is the timing when both channels A and B are enabled.



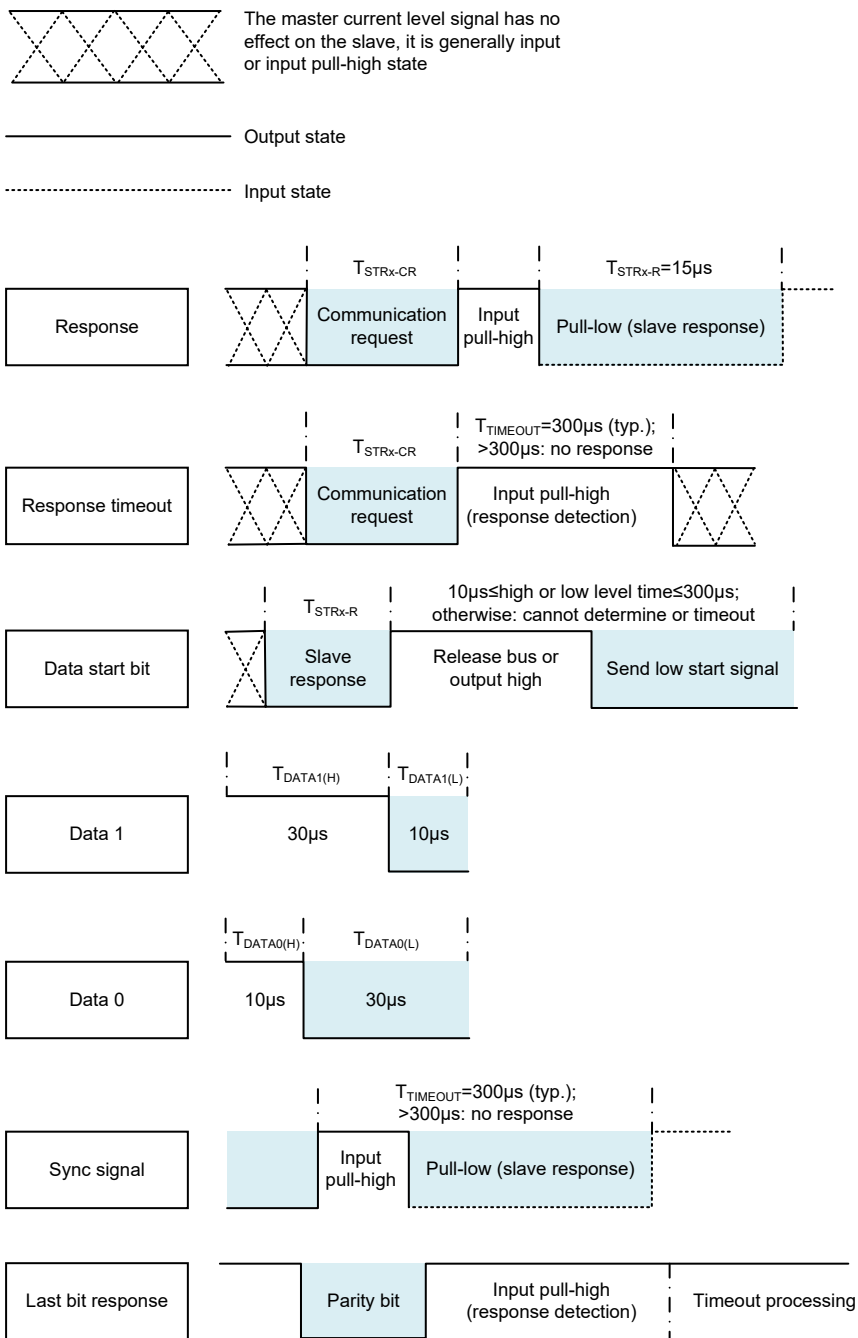
- Note: 1. It is necessary to enable the channel A and channel B using the running variables and set the smoke detection period to be larger than 0 to generate a synchronous pulse. Refer to the parameter configuration section for details.
2. If only one of channel A and B is enabled, there is only one channel timing.

One-wire Communication Protocol

The one-wire communication can only be initiated by the master. A complete communication process includes a request signal, a request response signal, a data start signal, data 0, data 1, a parity bit (even parity) and a last bit response. The parity bit is data 0 or 1. Note that if the initial low level time is larger than $40\mu\text{s}$, it is a trigger sampling rather than a communication request. The request response signal is used to determine whether the slave is online and the last bit response is determined by the slave after verifying the previous data, by which the master can know whether the slave has received the data correctly, the master can consider retransmission if there is an abnormality.



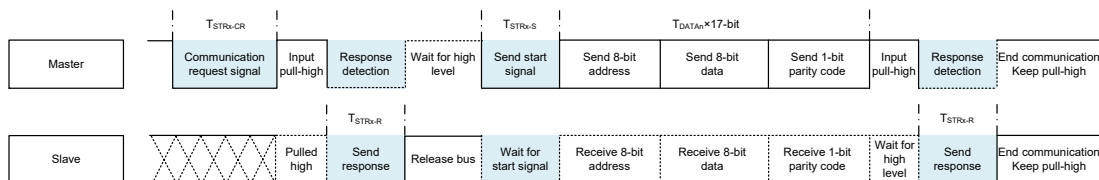
Note: If the initial low level time is larger than $40\mu\text{s}$, it is a trigger sampling timing rather than an one-wire communication request.



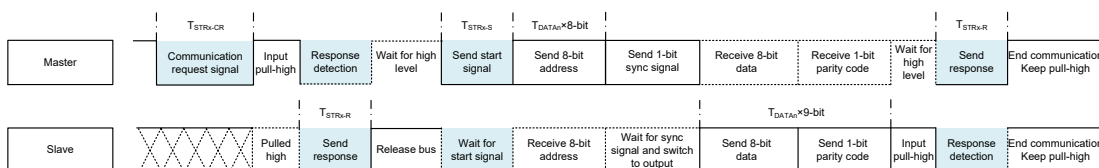
Data Frame

The order of data transmission is the LSB first. The bit 7, which is the MSB of the address byte, is used to distinguish the read or write instruction. When this bit is 0, it indicates a send instruction; when this bit is 1, it indicates a receive instruction. The parity bit is the even parity of the address byte and data byte. If the address byte is 0x01 and the data byte is 0x77, the parity bit is 1.

Send one frame of data



Receive one frame of data



Parameter Configuration

Running Variable

The running variable has two channels, A and B. The AFE parameters and emission current of each channel are independent. After the master has configured the corresponding parameters and enabled the channel, it does not need to execute communicate processing again. The smoke detector will automatically send out a wake-up signal, configure AFE, turn on ISINK with the set parameters and then enter low-power standby mode. This mode can effectively avoid frequent communication from the master and therefore reduce standby power consumption. The specific operation timing are shown in the above section. The following data are all 8-bit wide.

Function	Address	Type	Initial Value	Description
Write protection	0x1A	R/W	0x00	It needs to set this function to 0xA5 to execute write operation of the running variables, otherwise, disable the write operation.
Wake-up signal duration (low level)	0x00	R/W	0x00	$T_{SINT-WU(D)} = (1.5 + 1.5N) \times 4\mu s$. When it is 0x00, $N=255$.
Channel A SDSW	0x01	R/W	0x00	Refer to the Smoke Detector SDSW register.
Channel A SDPGAC0	0x02	R/W	0x00	Refer to the Smoke Detector SDPGAC0 register.
Channel A SDPGAC1	0x03	R/W	0x00	Refer to the Smoke Detector SDPGAC1 register.
Channel A ISGDATA0 (emission current)	0x04	R/W	0x00	Refer to the Smoke Detector ISGDATA0 register.
Channel A waiting time after OPA is on	0x05	R/W	0x00	$T_{CHA-OPAW} = (1.5 + 1.5N) \times 4\mu s$. When it is 0x00, $N=255$.
Channel A emission stable time	0x06	R/W	0x00	$T_{CHA-ES} = (1.5 + 1.5N) \times 2\mu s$. When it is 0x00, $N=255$.
Channel A A/D sample timeout	0x07	R/W	0x00	$T_{CHA-ADCST00} = (1.5 + 1.5N) \mu s$. When it is 0x00, $N=255$. $T_{CHA-ADCST01} = (2.5 + 2.5N) \mu s$. When it is 0x00, $N=255$.
Channel B SDSW	0x08	R/W	0x00	Refer to the Smoke Detector SDSW register.
Channel B SDPGAC0	0x09	R/W	0x00	Refer to the Smoke Detector SDPGAC0 register.
Channel B SDPGAC1	0x0A	R/W	0x00	Refer to the Smoke Detector SDPGAC1 register.
Channel B ISGDATA1 (emission current)	0x0B	R/W	0x00	Refer to the Smoke Detector ISGDATA1 register.
Channel B waiting time after OPA is on	0x0C	R/W	0x00	$T_{CHB-OPAW} = (1.5 + 1.5N) \times 4\mu s$. When it is 0x00, $N=255$.
Channel B emission stable time	0x0D	R/W	0x00	$T_{CHB-ES} = (1.5 + 1.5N) \times 2\mu s$. When it is 0x00, $N=255$.
Channel B A/D sample timeout	0x0E	R/W	0x00	$T_{CHB-ADCST00} = (1.5 + 1.5N) \mu s$. When it is 0x00, $N=255$. $T_{CHB-ADCST01} = (2.5 + 2.5N) \mu s$. When it is 0x00, $N=255$.

Function	Address	Type	Initial Value	Description
Channel A enable	0x0F	R/W	0x00	When it is 0x00, channel A is disabled. When it is larger than 0, channel A is enabled.
Channel B enable	0x10	R/W	0x00	When it is 0x00, channel B is disabled. When it is larger than 0, channel B is enabled.
Smoke detection period	0x11	R/W	0x00	$T_{SD}=N$. Unit: s. When it is 0x00, it is disabled.
Time base wake-up period	0x12	R/W	0x00	$T_{TB}=N$. Unit: s. When it is 0x00, it is disabled.
Running flag	0x1D	R/W	0xBA	After power on, the default value is 0xBA. It needs to write a 0x00 to clear. If the flag is not the 0x00, the wake-up pin, SINT, will output a low level during the wake-up signal duration of the sampling timing, otherwise it will be high, to indicate whether the operating state is normal as shown in the sampling timing diagrams. During normal operation, the flag remains at the last value written.
OPA calibration	0x1E	R/W	0x00	Writing a 0xA0 to trigger the OPA calibration. After calibration, the value is 0x55. During the OPA calibration, do not response the communication. There requires about 300ms to calibrate.
SDA0VOS	0x5D	R/W	—	Refer to the Smoke Detector SDA0VOS register.
SDA1VOS	0x5F	R/W	—	Refer to the Smoke Detector SDA1VOS register.

Smoke Detector Registers

The overall operation of the smoke detector AFE circuit is controlled using a series of registers. The SDSW register is used to control the switches on or off thus controlling the Operational Amplifier operating mode. The SDPGAC0 and SDPGAC1 registers are used to select the R1, R2 and R3 resistance. The sink current generator could provide constant current no matter what V_{ISINK} voltage is ranging from 0.7V to 4.5V. The constant current value is controlled by the ISGDATA0/ISGDATA1 register and the sink current range is 50mA~360mA.

Register Name	Bit							
	7	6	5	4	3	2	1	0
SDSW	—	SDS6	SDS5	SDS4	SDS3	SDS2	SDS1	SDS0
SDPGAC0	—	—	SDA0PGA5	SDA0PGA4	SDA0PGA3	SDA0PGA2	SDA0PGA1	SDA0PGA0
SDPGAC1	SDA1PGA7	SDA1PGA6	SDA1PGA5	SDA1PGA4	SDA1PGA3	SDA1PGA2	SDA1PGA1	SDA1PGA0
SDA0VOS	SDA0OFM	SDA0RSP	SDA0OF5	SDA0OF4	SDA0OF3	SDA0OF2	SDA0OF1	SDA0OF0
SDA1VOS	SDA1OFM	SDA1RSP	SDA1OF5	SDA1OF4	SDA1OF3	SDA1OF2	SDA1OF1	SDA1OF0
ISGDATA0	—	—	—	D4	D3	D2	D1	D0
ISGDATA1	—	—	—	D4	D3	D2	D1	D0

Smoke Detector Register List

• SDSW Register

Bit	7	6	5	4	3	2	1	0
Name	—	SDS6	SDS5	SDS4	SDS3	SDS2	SDS1	SDS0
R/W	—	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	—	0	0	0	0	0	0	0

Bit 7 Unimplemented, read as “0”

Bit 6~5 **SDS6~SDS5**: Mode Control

00: External mode
 01: AC coupling mode
 10: External mode
 11: DC coupling mode

Note: 1. When the Switch S6 and S8 are both on to select the AC couple mode, the OPAMP1 amplification factor will be decreased as the Switch S8 has an on-resistance R_{ON_S8} . The OPAMP1 Gain = $1 + R2 / (R3 + R_{ON_S8})$, where $R_{ON_S8} \approx 150\Omega @ V_{DD_SD} = 5V$, $R_{ON_S8} \approx 170\Omega @ V_{DD_SD} = 3V$.

2. When the Switch S7 and S9 are both on, connecting to V_{BIAS} , to select the DC couple mode, the original settings of the resistance amplification factor will not be available for reference, which is an especially important point users should pay attention to.

Bit 4 **SDS4**: Switch on/off control

0: Off
 1: On

Bit 3 **SDS3**: Switch on/off control

0: Off
 1: On

Bit 2 **SDS2**: Switch on/off control

0: Off
 1: On

Bit 1 **SDS1**: Switch on/off control

0: Off
 1: On

Bit 0 **SDS0**: Switch on/off control

0: Off
 1: On

• SDPGAC0 Register

Bit	7	6	5	4	3	2	1	0
Name	—	—	SDA0PGA5	SDA0PGA4	SDA0PGA3	SDA0PGA2	SDA0PGA1	SDA0PGA0
R/W	—	—	R/W	R/W	R/W	R/W	R/W	R/W
POR	—	—	0	0	0	0	0	0

Bit 7~6 Unimplemented, read as “0”

Bit 5~0 **SDA0PGA5~SDA0PGA0**: R1 resistance control

$R1 = SDA0PGA[5:0] \times 100k\Omega$

• SDPGAC1 Register

Bit	7	6	5	4	3	2	1	0
Name	SDA1PGA7	SDA1PGA6	SDA1PGA5	SDA1PGA4	SDA1PGA3	SDA1PGA2	SDA1PGA1	SDA1PGA0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~6 **SDA1PGA7~SDA1PGA6**: R3 resistance control

00: 10k Ω
 01: 20k Ω
 10: 30k Ω
 11: 40k Ω

Bit 5~0 **SDA1PGA5~SDA1PGA0**: R2 resistance control

$R2 = SDA1PGA[5:0] \times 100k\Omega$

• **SDA0VOS Register**

Bit	7	6	5	4	3	2	1	0
Name	SDA0OFM	SDA0RSP	SDA0OF5	SDA0OF4	SDA0OF3	SDA0OF2	SDA0OF1	SDA0OF0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	1	0	0	0	0	0

- Bit 7 **SDA0OFM**: SD OPA0 normal operation or input offset voltage calibration mode selection
0: Normal operation
1: Offset calibration mode
- Bit 6 **SDA0RSP**: SD OPA0 input offset voltage calibration reference selection
0: Input reference voltage comes from A0NI
1 : Input reference voltage comes from A0PI
- Bit 5~0 **SDA0OF5~SDA0OF0**: SD OPA0 input offset voltage calibration control

• **SDA1VOS Register**

Bit	7	6	5	4	3	2	1	0
Name	SDA1OFM	SDA1RSP	SDA1OF5	SDA1OF4	SDA1OF3	SDA1OF2	SDA1OF1	SDA1OF0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	1	0	0	0	0	0

- Bit 7 **SDA1OFM**: SD OPA1 normal operation or input offset voltage calibration mode selection
0: Normal operation
1: Offset calibration mode
- Bit 6 **SDA1RSP**: SD OPA1 input offset voltage calibration reference selection
0: Input reference voltage comes from A1NI
1 : Input reference voltage comes from A1PI
- Bit 5~0 **SDA1OF5~SDA1OF0**: SD OPA1 input offset voltage calibration control

• **ISGDATA0 Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	D4	D3	D2	D1	D0
R/W	—	—	—	R/W	R/W	R/W	R/W	R/W
POR	—	—	—	0	0	0	0	0

- Bit 7~5 Unimplemented, read as “0”
- Bit 4~0 **D4~D0**: Sink current generator control for ISINK0 pin
Current value (mA)=50+10×D[4:0]
Refer to “Smoke Detector Electrical Characteristics” table for more details.

• **ISGDATA1 Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	D4	D3	D2	D1	D0
R/W	—	—	—	R/W	R/W	R/W	R/W	R/W
POR	—	—	—	0	0	0	0	0

- Bit 7~5 Unimplemented, read as “0”
- Bit 4~0 **D4~D0**: Sink current generator control for ISINK1 pin
Current value (mA)=50+5×D[4:0]
Refer to “Smoke Detector Electrical Characteristics” table for more details.

Analog to Digital Converter

The need to interface to real world analog signals is a common requirement for many electronic systems. However, to properly process these signals by a microcontroller, they must first be converted into digital signals by A/D converters. By integrating the A/D conversion electronic circuitry into the microcontroller, the need for external components is reduced significantly with the corresponding follow-on benefits of lower costs and reduced component space requirements.

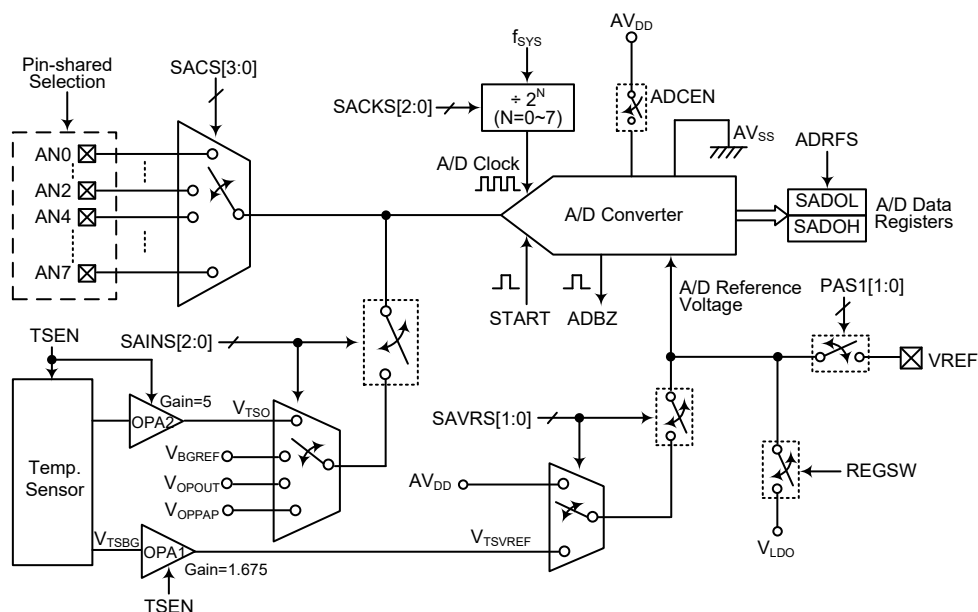
A/D Converter Overview

This device contains a multi-channel analog to digital converter which can directly interface to external analog signals, such as that from external sensors or other control signals and convert these signals directly into a 12-bit digital value. It also can convert the internal signals, such as the bandgap reference voltage or the temperature sensor output, into a 12-bit digital value. The external or internal analog signal to be converted is determined by the SAINS2~SAINS0 bits together with the SACS3~SACS0 bits. When the external analog signal is to be converted, the corresponding pin-shared control bits should first be properly configured and then desired external channel input should be selected using SAINS2~SAINS0 and SACS3~SACS0 bits. Note that when the internal analog signal is to be converted, the SAINS and SACS fields should be properly configured to avoid external channel input. More detailed information about the A/D input signal is described in the “A/D Converter Control Registers” and “A/D Converter Input Signals” sections respectively.

This A/D converter also includes a temperature sensor circuitry which contains a temperature sensor, two operational amplifiers and an internal reference voltage. The temperature sensor can detect the temperature and then output a voltage proportional to the temperature. The output voltage can be amplified by the OPA and then converted to a 12-bit digital data using the A/D converter.

External Input Channels	Internal Signals	A/D Input Select Bits
AN0~AN2, AN4~AN7	V _{BGREF} , V _{OPOUT} , V _{OPPAP} , V _{TSO}	SAINS2~SAINS0, SACS3~SACS0

The accompanying block diagram shows the overall internal structure of the A/D converter with temperature sensor together with their associated registers and control bits.



A/D Converter with Temperature Sensor Diagram

Register Descriptions

Overall operation of the A/D converter with Temperature sensor is controlled using a series of registers. A read only register pair exists to store the A/D converter data 12-bit value. Two registers, SADC0 and SADC1, are control registers which setup the operating and control function of the A/D converter. The VBGRC register is used to enable/disable the A/D converter internal bandgap reference voltage output. The SADC2 is the temperature sensor circuitry control register. The remaining two registers, LMSADOH and LMSADOL, are read only registers and store a 12-bit A/D conversion result of certain temperature.

Register Name	Bit							
	7	6	5	4	3	2	1	0
SADOL (ADRF=0)	D3	D2	D1	D0	—	—	—	—
SADOL (ADRF=1)	D7	D6	D5	D4	D3	D2	D1	D0
SADOH (ADRF=0)	D11	D10	D9	D8	D7	D6	D5	D4
SADOH (ADRF=1)	—	—	—	—	D11	D10	D9	D8
SADC0	START	ADBZ	ADCEN	ADRF	SACS3	SACS2	SACS1	SACS0
SADC1	SAINS2	SAINS1	SAINS0	SAVRS1	SAVRS0	SACKS2	SACKS1	SACKS0
SADC2	—	—	—	—	—	—	D1	TSEN
VBGRC	—	—	—	—	—	—	—	VBGREN
LMSADOH	D11	D10	D9	D8	D7	D6	D5	D4
LMSADOL	D3	D2	D1	D0	—	—	—	—

A/D Converter with Temperature Sensor Register List

A/D Converter Data Registers – SADOL, SADOH

As the internal A/D converter provides a 12-bit digital conversion value, it requires two data registers to store the converted value. These are a high byte register, known as SADOH, and a low byte register, known as SADOL. After the conversion process takes place, these registers can be directly read by the microcontroller to obtain the digitised conversion value. As only 12 bits of the 16-bit register space is utilised, the format in which the data is stored is controlled by the ADRF bit in the SADC0 register, as shown in the accompanying table. Any unused bits will be read as zero. Note that A/D data registers contents will be unchanged if the A/D converter is disabled.

ADRF	SADOH								SADOL							
	7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0
0	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0	0	0	0	0
1	0	0	0	0	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0

A/D Converter Data Registers

A/D Converter Control Registers – SADC0, SADC1

To control the function and operation of the A/D converter, two control registers known as SADC0 and SADC1 are provided. These 8-bit registers define functions such as the selection of which analog channel is connected to the internal A/D converter, the digitised data format, the A/D clock source as well as controlling the start function and monitoring the A/D converter busy status. As the device contains only one actual analog to digital converter hardware circuit, each of the external or internal analog signals must be routed to the converter. The SACS3~SACS0 bits in the SADC0 register are used to determine which external channel input is selected to be converted.

The SAINS2~SAINS0 bits in the SADC1 register are used to determine that the analog signal to be converted comes from the internal analog signal or external analog channel input.

• **SADC0 Register**

Bit	7	6	5	4	3	2	1	0
Name	START	ADBZ	ADCEN	ADRF5	SACS3	SACS2	SACS1	SACS0
R/W	R/W	R	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

- Bit 7** **START:** Start the A/D conversion
0→1→0: Start A/D conversion
This bit is used to initiate an A/D conversion process. The bit is normally low but if set high and then cleared low again, the A/D converter will initiate a conversion process.
- Bit 6** **ADBZ:** A/D converter busy flag
0: No A/D conversion is in progress
1: A/D conversion is in progress
This read only flag is used to indicate whether the A/D conversion is in progress or not. When the START bit is set from low to high and then to low again, the ADBZ flag will be set to 1 to indicate that the A/D conversion is initiated. The ADBZ flag will be cleared to 0 after the A/D conversion is complete.
- Bit 5** **ADCEN:** A/D converter function enable control
0: Disable
1: Enable
This bit controls the A/D internal function. This bit should be set to one to enable the A/D converter. If the bit is cleared to zero, then the A/D converter will be switched off reducing the device power consumption. When the A/D converter function is disabled, the contents of the A/D data register pair known as SADOH and SADOL will remain unchanged.
- Bit 4** **ADRF5:** A/D converter data format select
0: A/D converter data format → SADOH=D[11:4]; SADOL=D[3:0]
1: A/D converter data format → SADOH=D[11:8]; SADOL=D[7:0]
This bit controls the format of the 12-bit converted A/D value in the two A/D data registers. Details are provided in the A/D converter data register section.
- Bit 3~0** **SACS3~SACS0:** A/D converter external analog channel input select
0000: AN0
0001: AN1
0010: AN2
0011: Reserved
0100: AN4
0101: AN5
0110: AN6
0111: AN7
1000~1111: Undefined, input floating if selected
These bits are used to select which external analog input channel is to be converted.

• **SADC1 Register**

Bit	7	6	5	4	3	2	1	0
Name	SAINS2	SAINS1	SAINS0	SAVRS1	SAVRS0	SACKS2	SACKS1	SACKS0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

- Bit 7~5** **SAINS2~SAINS0:** A/D converter input signal select
000: External source – External analog channel input, ANn
001: Internal source – Bandgap reference voltage, V_{BGREF}

010: Internal source – Internal temperature Sensor output, V_{TSO}
 011: Internal source – OPAMP output, V_{OPOUT}
 100: Internal source – OPAMP positive input, V_{OPPAP}
 101~110: External source – External analog channel input, ANn
 111: Forbidden data, SAINS2~SAINS0 bits can not be written with “111”

Care must be taken if the SAINS2~SAINS0 bits are set to “001~100” to select the internal analog signal to be converted. When the internal analog signal is selected to be converted, the external channel input pin must never be selected as the A/D input signal by properly setting the SACS3~SACS0 bits with a value from 1000 to 1111. Otherwise, the external channel input will be connected together with the internal analog signal. This will result in unpredictable situations such as an irreversible damage.

Bit 4~3 **SAVRS1~SAVRS0**: A/D converter reference voltage select

00: VREF pin input or LDO output
 01: Internal A/D converter power supply, AV_{DD}
 10: Internal Temperature Sensor reference voltage, V_{TSVREF}
 11: Internal A/D converter power supply, AV_{DD}

These bits are used to select the A/D converter reference voltage source. Care must be taken if the SAVRS1~SAVRS0 bits are set to “01~11” to select the internal A/D converter power or Temperature Sensor reference voltage as the reference voltage source. When the internal A/D converter power or Temperature Sensor reference voltage is selected as the reference voltage, the VREF pin cannot be configured as the reference voltage input by properly configuring the corresponding pin-shared function control bits. Additionally, the LDO output path to the A/D converter reference voltage must also be switched off by clearing the REGSW bit in the REGC register. Otherwise, the external input voltage on VREF pin or the LDO output voltage together with the selected internal reference voltage will be simultaneously connected to the A/D converter reference voltage input. This will result in unpredictable situations. Refer to the “A/D Converter Reference Voltage” section for more details.

Bit 2~0 **SACKS2~SACKS0**: A/D conversion clock rate select

000: f_{SYS}
 001: $f_{SYS}/2$
 010: $f_{SYS}/4$
 011: $f_{SYS}/8$
 100: $f_{SYS}/16$
 101: $f_{SYS}/32$
 110: $f_{SYS}/64$
 111: $f_{SYS}/128$

These bits are used to select the clock source for the A/D converter. If the internal OPAMP signal is selected to be converted, care must be taken for the A/D clock rate selection limitation. Refer to the OPC Register description in the CO/Gas Detector AFE section for details.

Bandgap Reference Voltage Control Register – VBGRC

A high performance bandgap voltage reference is included in the device. It has an accurate voltage reference output, V_{BGREF} , when input supply voltage change or temperature variation. The VBGRC register is used to control the bandgap reference voltage circuit enable or disable.

• VBGRC Register

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	—	VBGREN
R/W	—	—	—	—	—	—	—	R/W
POR	—	—	—	—	—	—	—	0

Bit 7~1 Unimplemented, read as “0”

Bit 0 **VBGREN**: Bandgap enable/disable control
 0: Disable
 1: Enable
 This bit is used to enable/disable the internal Bandgap reference circuit. The internal Bandgap reference circuit should first be enabled before the V_{BGREF} voltage is selected to be used.

Temperature Sensor Control Register – SADC2

To control the enable and disable of the integrated temperature sensor circuitry, a control register known as SADC2 is provided.

• SADC2 Register

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	D1	TSEN
R/W	—	—	—	—	—	—	R/W	R/W
POR	—	—	—	—	—	—	1	0

Bit 7~2 Unimplemented, read as “0”

Bit 1 **D1**: Reserved bit, should be fixed to 1

Bit 0 **TSEN**: Temperature sensor circuitry enable control
 0: Disable
 1: Enable

This bit controls the internal temperature sensor circuitry. If the temperature sensor output will be converted or the temperature sensor reference voltage will be selected as the A/D conversion reference voltage, the temperature sensor circuitry should be turned on by setting the TSEN bit high first. When the temperature sensor is enabled by setting the TSEN bit to 1, a time named as t_{TSS} should be allowed for the temperature sensor circuit to stabilise before implementing relevant temperature sensor operation.

85°C A/D Conversion Value Registers – LMSADOH, LMSADOL

A pair of read-only registers, LMSADOH and LMSADOL, are provided to store the 12-bit A/D converted value of 85°C temperature.

Register	LMSADOH								LMSADOL							
	7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0
Name	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0	—	—	—	—
R/W	R	R	R	R	R	R	R	R	R	R	R	R	—	—	—	—
POR	85°C A/D Converted Value												—	—	—	—

“—”: Unimplemented, read as “0”

85°C A/D Conversion Value Registers

A/D Converter Operation

The START bit in the SADC0 register is used to start the A/D conversion. When the microcontroller sets this bit from low to high and then low again, an analog to digital conversion cycle will be initiated.

The ADBZ bit in the SADC0 register is used to indicate whether the analog to digital conversion process is in progress or not. This bit will be automatically set to 1 by the microcontroller after an A/D conversion is successfully initiated. When the A/D conversion is complete, the ADBZ will be cleared to 0. In addition, the corresponding A/D interrupt request flag will be set in the interrupt control register, and if the interrupts are enabled, an appropriate internal interrupt signal will be generated. This A/D internal interrupt signal will direct the program flow to the associated A/D internal interrupt address for processing. If the A/D internal interrupt is disabled, the microcontroller

can poll the ADBZ bit in the SADC0 register to check whether it has been cleared as an alternative method of detecting the end of an A/D conversion cycle.

The clock source for the A/D converter, which originates from the system clock f_{SYS} , can be chosen to be either f_{SYS} or a subdivided version of f_{SYS} . The division ratio value is determined by the SACKS2~SACKS0 bits in the SADC1 register. Although the A/D clock source is determined by the system clock f_{SYS} and SACKS2~SACKS0 bits, there are some limitations on the maximum A/D clock source speed that can be selected. As the recommended range of permissible A/D clock period, t_{ADCK} , is from 0.5 μ s to 10 μ s for all input analog signals except the temperature sensor output, care must be taken for system clock frequencies. For example, as the system clock operates at a frequency of 8MHz and the analog input is not selected from the temperature sensor output, the SACKS2~SACKS0 bits should not be set to 000, 001 or 111. Doing so will give A/D clock periods that are less than the minimum A/D clock period or larger than the maximum A/D clock period, which may result in inaccurate A/D conversion values. Refer to the following table for examples, where values marked with an asterisk * show where special care must be taken.

If the input signal to be converted is the temperature sensor output voltage, the permissible A/D clock period is from 1 μ s to 2 μ s.

f_{SYS}	A/D Clock Period (t_{ADCK})							
	SACKS[2:0] =000 (f_{SYS})	SACKS[2:0] =001 ($f_{SYS}/2$)	SACKS[2:0] =010 ($f_{SYS}/4$)	SACKS[2:0] =011 ($f_{SYS}/8$)	SACKS[2:0] =100 ($f_{SYS}/16$)	SACKS[2:0] =101 ($f_{SYS}/32$)	SACKS[2:0] =110 ($f_{SYS}/64$)	SACKS[2:0] =111 ($f_{SYS}/128$)
1MHz	1 μ s	2 μ s	4 μ s	8 μ s	16 μ s *	32 μ s *	64 μ s *	128 μ s *
2MHz	500ns	1 μ s	2 μ s	4 μ s	8 μ s	16 μ s *	32 μ s *	64 μ s *
4MHz	250ns *	500ns	1 μ s	2 μ s	4 μ s	8 μ s	16 μ s *	32 μ s *
8MHz	125ns *	250ns *	500ns	1 μ s	2 μ s	4 μ s	8 μ s	16 μ s *

A/D Clock Period Examples for External Analog Inputs

Controlling the power on/off function of the A/D converter circuitry is implemented using the ADCEN bit in the SADC0 register. This bit must be set high to power on the A/D converter. When the ADCEN bit is set high to power on the A/D converter internal circuitry a certain delay, as indicated in the timing diagram, must be allowed before an A/D conversion is initiated. Even if no pins are selected for use as A/D inputs, if the ADCEN bit is high, then some power will still be consumed. In power conscious applications it is therefore recommended that the ADCEN is set low to reduce power consumption when the A/D converter function is not being used.

A/D Converter Reference Voltage

The A/D converter has its own external reference voltage input pin, VREF. However, the reference voltage can also be supplied from the LDO output, or from the A/D converter power supply, AV_{DD} , or from the temperature sensor reference voltage, V_{TSVREF} . The choice is made using the SAVRS1 and SAVRS0 bits in the SADC1 register together with other control bits.

If the external VREF pin input voltage is required to use, firstly the SAVRS bit field should be set to “00” to select “VREF pin input or LDO output” option. As the VREF pin is pin-shared with other functions, then the relevant pin-shared control bits PAS1[1:0] should be set as “10” to enable the VREF pin function. Additionally, the LDO output path to the A/D converter reference voltage must be switched off by clearing the REGSW bit in the REGC register.

If the LDO output is required to use, firstly the SAVRS bit field should be set to “00” to select “VREF pin input or LDO output” option. Then set the REGEN[2:0] bits in the REGC register to select the required LDO voltage as A/D reference voltage and set the REGSW bit high to switch on the LDO output path to the A/D converter reference voltage. Additionally, ensure that the VREF pin is

not configured as the reference voltage input function by setting the VREF pin-shared control bits PAS1[1:0] to any other value except “10”.

If the A/D converter power supply, AV_{DD} is required to use, the SAVRS bit field should be set to “01” or “11”. If the temperature sensor reference voltage is required to use, the SAVRS bit field should be set to “10”. As the temperature sensor circuitry is controlled by the TSEN bit, the TSEN bit should be set high to enable the temperature sensor. However, it is important to note that even when the AV_{DD} or V_{TSVREF} has been selected as the A/D converter reference voltage by correctly setting the SAVRS bit field, the LDO output path to the A/D converter reference voltage must be switched off and the VREF pin should not be configured as the reference voltage input function by properly configuring the REGSW bit and the PAS1[1:0] bits. Otherwise these signals may be input to the A/D converter reference simultaneously, which will result in unpredictable situations such as an irreversible damage.

Note that the analog input values must not be allowed to exceed the value of the selected A/D conversion reference voltage.

SAVRS[1:0]	Other Relevant Bits	Reference	Description
00	PAS1[1:0]=10 & REGSW=0	VREF pin	External VREF pin input voltage
	PAS1[1:0]≠10 & REGSW=1 & REGEN[2:0]=010/011/11x	V_{LDO}	LDO output voltage
01/11	PAS1[1:0]≠10 & REGSW=0	AV_{DD}	Internal A/D converter power supply
10	TSEN=1 & PAS1[1:0]≠10 & REGSW=0	V_{TSVREF}	Temperature sensor reference voltage

A/D Converter Reference Voltage Selection

A/D Converter Input Signals

All of the external A/D analog input pins are pin-shared with the I/O pins as well as other functions. The corresponding pin-shared function selection bits in the PAS0/1 and PBS0/1 registers, determine whether the external pins are setup as A/D converter analog inputs or whether they have other functions. If the pin is setup to be an A/D analog channel input, the other pin functions will be disabled. In this way, pins can be changed under program control to change their function between A/D inputs and other functions. All pull-high resistors, which are setup through register programming, will be automatically disconnected if the pins are setup as A/D inputs. Note that it is not necessary to first setup the A/D pin as an input in the port control register to enable the A/D input as when the ANn pin function is enabled by the relevant pin-shared function selection bits, the status of the port control register will be overridden.

If the SAINS2~SAINS0 bits are set to one value of “000” and “101~110”, the external analog channel input is selected to be converted and the SACS3~SACS0 bits can determine which actual external channel is selected to be converted. If the SAINS2~SAINS0 bits are set to “001~100”, one of the internal signals, including the internal Bandgap reference voltage, temperature sensor output voltage, OPAMP output voltage and the OPAMP positive input voltage is selected to be converted. Note that if the internal analog signal is selected to be converted, the external input channel determined by the SACS3~SACS0 bits must be switched to a non-existed A/D input channel by properly setting the SACS bit field with a value from 1000 to 1111. Otherwise, the external channel input will be connected together with the internal analog signal, which will result in unpredictable errors.

SAINS[2:0]	SACS[3:0]	Input Signals	Description
000, 101~110	0000~0010, 0100~0111	AN0~AN2, AN4~AN7	External channel input
	1000~1111	—	Non-existed channel, input is floating
001	1000~1111	V _{BGREF}	Internal Bandgap reference voltage
010	1000~1111	V _{T_{SO}}	Internal Temperature Sensor output voltage
011	1000~1111	V _{O_{POUT}}	Internal OPAMP output voltage
100	1000~1111	V _{O_{PPAP}}	Internal OPAMP positive input voltage
111	Forbidden data, SAINS2~SAINS0 bits can not be written with “111”		

A/D Converter Input Signal Selection

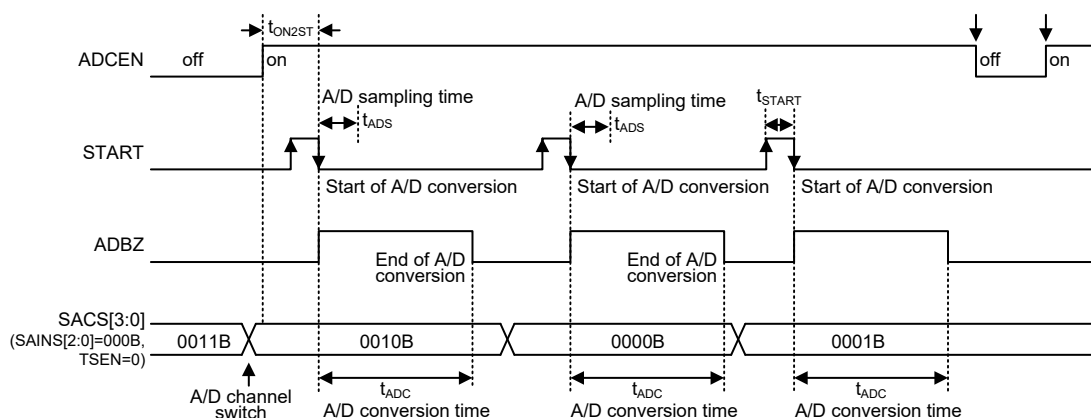
Conversion Rate and Timing Diagram

A complete A/D conversion contains two phases, data sampling and data conversion. If the conversion input signal is not the temperature sensor output, the data sampling which is defined as t_{ADS} takes 4 A/D clock periods and the data conversion takes 12 A/D clock periods. Therefore a total of 16 A/D clock periods for an A/D conversion which is defined as t_{ADC} are necessary. However, an A/D conversion for an internal temperature sensor signal will take a total of 58 A/D clock periods, which includes 46 A/D clock periods for data sampling and 12 A/D clock periods for data conversion.

Maximum single A/D conversion rate = $1/(A/D \text{ clock period} \times 16)$ (Temperature sensor output signal is not used)

Maximum single A/D conversion rate = $1/(A/D \text{ clock period} \times 58)$ (Internal Temperature sensor output signal is used)

The accompanying diagram shows graphically the various stages involved in an external channel input signal analog to digital conversion process and its associated timing. After an A/D conversion process has been initiated by the application program, the microcontroller internal hardware will begin to carry out the conversion, during which time the program can continue with other functions. The time taken for the A/D conversion is $16 t_{ADCK}$ where t_{ADCK} is equal to the A/D clock period.



A/D Conversion Timing – External Channel Inputs

Summary of A/D Conversion Steps

The following summarises the individual steps that should be executed in order to implement an A/D conversion process.

- Step 1
Select the required A/D conversion clock by correctly programming bits SACKS2~SACKS0 in the SADC1 register.
- Step 2
Enable the A/D by setting the ADCEN bit in the SADC0 register to one.
- Step 3
Select which signal is to be connected to the internal A/D converter by correctly configuring the SAINS2~SAINS0 bits.
Select the external channel input to be converted, go to Step 4.
Select the internal analog signal to be converted, go to Step 5.
- Step 4
If the A/D input signal comes from the external channel input selected by configuring the SAINS bit field, the corresponding pins should be configured as A/D input function by configuring the relevant pin-shared function control bits. The desired analog channel then should be selected by configuring the SACS bit field. After this step, go to Step 6.
- Step 5
Before the A/D input signal is selected to come from the internal analog signal by configuring the SAINS bit field, the corresponding external input pin must be switched to a non-existed channel input by setting the SACS3~SACS0 bits with a value from 1000 to 1111. The desired internal analog signal then can be selected by configuring the SAINS bit field. After this step, go to Step 6.
- Step 6
Select the reference voltage source by configuring the SAVRS1~SAVRS0 bits in the SADC1 register. Care should be taken in this step which can refer to the A/D Converter Reference Voltage section for details.
- Step 7
Select A/D converter output data format by setting the ADRFS bit in the SADC0 register.
- Step 8
If the A/D conversion interrupt is used, the interrupt control registers must be correctly configured to ensure the A/D interrupt function is active. The master interrupt control bit, EMI, and the A/D conversion interrupt control bit, ADE, must both be set high in advance.
- Step 9
The A/D conversion procedure can now be initialized by setting the START bit from low to high and then low again.
- Step 10
If A/D conversion is in progress, the ADBZ flag will be set high. After the A/D conversion process is complete, the ADBZ flag will go low and then the output data can be read from SADOH and SADOL registers.

Note: When checking for the end of the conversion process, if the method of polling the ADBZ bit in the SADC0 register is used, the interrupt enable step above can be omitted.

Programming Considerations

During microcontroller operations where the A/D converter is not being used, the A/D internal circuitry can be switched off to reduce power consumption, by clearing ADCEN bit in the SADC0 register. When this happens, the internal A/D converter circuits will not consume power irrespective of what analog voltage is applied to their input lines. If the A/D converter input lines are used as normal I/Os, then care must be taken as if the input voltage is not at a valid logic level, then this may lead to some increase in power consumption.

A/D Conversion Function

As the device contains a 12-bit A/D converter, its full-scale converted digitised value is equal to FFFH. Since the full-scale analog input value is equal to the actual A/D converter reference voltage, V_{REF} , this gives a single bit analog input value of V_{REF} divided by 4096.

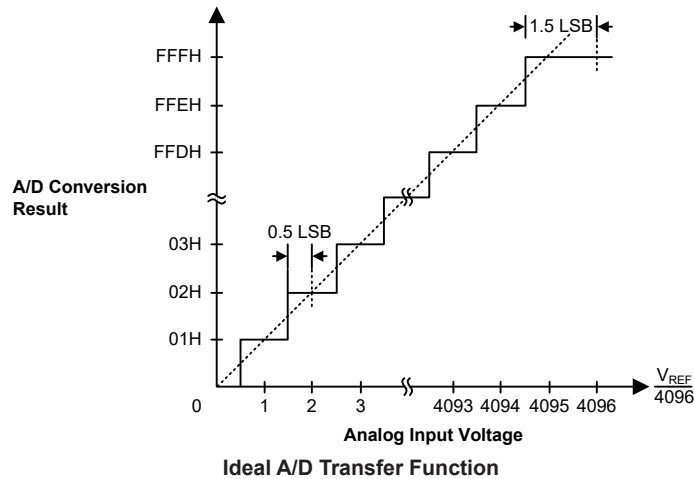
$$1 \text{ LSB} = V_{REF} \div 4096$$

The A/D Converter input voltage value can be calculated using the following equation:

$$\text{A/D input voltage} = \text{A/D output digital value} \times V_{REF} \div 4096$$

The diagram shows the ideal transfer function between the analog input value and the digitised output value for the A/D converter. Except for the digitised zero value, the subsequent digitised values will change at a point 0.5 LSB below where they would change without the offset, and the last full scale digitised value will change at a point 1.5 LSB below the V_{REF} level.

Note that here the V_{REF} voltage is the actual A/D converter reference voltage determined by the SAVRS field.



Temperature Measurement Function

As the temperature sensor output voltage, V_{TSO} , has a linear relationship with temperature, the V_{TSO} A/D converted data value will also have a linear relationship with temperature. The current temperature T_x can be proportionally calculated from its A/D converted value ADC_x using the following formula.

$$T_x (^\circ\text{C}) = \text{slope} \times (ADC_x - ADC_2) + T_2$$

As the device has provided two sets of values which are (ADC_1, T_1) and (ADC_2, T_2) . The T_1 and T_2 are two values of temperature and the ADC_1 and ADC_2 are their A/D converted values respectively. The slope can be calculated using the following formula.

$$\text{slope} = (T_1 - T_2) / (ADC_1 + 4096 - ADC_2)$$

For the device, the T1 has a fixed value of 175°C. The ADC1 can be read from the LMSAD0H and LMSAD0L registers. The ADC2 and T2 code are stored in the Option Memory and can be read from the Program Memory last page using the table read instruction when the Option Memory mapping function is enabled.

Name	Mapped Address in Program Memory	Description
T1	—	A fixed value of 175°C
ADC1	—	12-bit T1 A/D converted value in LMSAD0H & LMSAD0L registers
T2	1FF5H	T2 code (00H (0°C)~FFH (51°C)) Temperature value can be converted from the code with 0.2°C/step
ADC2	1FF6H	12-bit T2 A/D converted value bit 11 ~ bit4
	1FF7H	12-bit T2 A/D converted value bit 3 ~ bit 0

Temperature Measurement Reference Items

The Option Memory mapping function is enabled by using the ORMC register. For more details, refer to the “Option Memory Mapping Register – ORMC” in the Special Function Register Description section.

A/D Conversion Programming Examples

The following two programming examples illustrate how to setup and implement an A/D conversion. In the first example, the method of polling the ADBZ bit in the SADC0 register is used to detect when the conversion cycle is complete, whereas in the second example, the A/D interrupt is used to determine when the conversion is complete.

Example: using an ADBZ polling method to detect the end of conversion

```

clr ADE                ; disable ADC interrupt
clr TSEN               ; disable temperature sensor circuitry
mov a,0Bh              ; select fsys/8 as A/D clock, external channel as A/D input signal
mov SADC1,a            ; and A/D internal power as reference voltage
mov a,0Ch              ; setup PAS0 register to configure pin AN0
mov PAS0,a
mov a,20h
mov SADC0,a            ; enable and connect AN0 channel to A/D converter
:
start_conversion:
clr START              ; high pulse on start bit to initiate conversion
set START              ; reset A/D
clr START              ; start A/D
polling_EOC:
sz ADBZ                ; poll the SADC0 register ADBZ bit to detect end of A/D conversion
jmp polling_EOC        ; continue polling
mov a,SAD0L            ; read low byte conversion result value
mov SAD0L_buffer,a     ; save result to user defined register
mov a,SAD0H            ; read high byte conversion result value
mov SAD0H_buffer,a     ; save result to user defined register
:
:
jmp start_conversion   ; start next A/D conversion

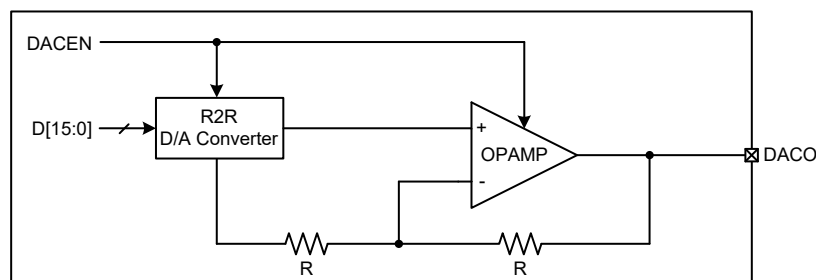
```

Example: using the interrupt method to detect the end of conversion

```
clr ADE                ; disable ADC interrupt
clr TSEN               ; disable temperature sensor circuitry
mov a,0Bh              ; select fsys/8 as A/D clock, external channel as A/D input signal
mov SADC1,a            ; and A/D internal power as reference voltage
mov a,0Ch              ; setup PAS0 register to configure pin AN0
mov PAS0,a
mov a,20h
mov SADC0,a            ; enable and connect AN0 channel to A/D converter
:
start_conversion:
clr START              ; high pulse on START bit to initiate conversion
set START              ; reset A/D
clr START              ; start A/D
clr ADF                ; clear ADC interrupt request flag
set ADE                ; enable ADC interrupt
set EMI                ; enable global interrupt
:
:
; ADC interrupt service routine
ADC_ISR:
mov acc_stack,a        ; save ACC to user defined memory
mov a,STATUS
mov status_stack,a     ; save STATUS to user defined memory
:
:
mov a,SADOL             ; read low byte conversion result value
mov SADOL_buffer,a     ; save result to user defined register
mov a,SADOH             ; read high byte conversion result value
mov SADOH_buffer,a     ; save result to user defined register
:
:
EXIT_INT_ISR:
mov a,status_stack
mov STATUS,a           ; restore STATUS from user defined memory
mov a,acc_stack        ; restore ACC from user defined memory
reti
```

16-bit Voice D/A Converter

The device has a 16-bit D/A Converter. The circuit is a 16-bit R2R D/A Converter for audio application. Its reference voltage comes from analog supply voltage only, and can be power down to save power. The 16-bit D/A Converter is good for voice or audio application. Although this D/A Converter is not general one-to-one digital to analog conversion, it provides not bad and same audio quality no matter what small or big voice. Note that the D/A Converter voltage is amplified and buffer output by OPAMP.



16-bit D/A Converter Block Diagram

D/A Converter Registers

Overall operation of the D/A Converter is controlled by using three registers. There are a 16-bit D/A Converter data high byte register, DAH, a 16-bit D/A Converter data low byte register, DAL, and a control register named as DACC is used to control the function and operation of the D/A converter.

Register Name	Bit							
	7	6	5	4	3	2	1	0
DAH	D15	D14	D13	D12	D11	D10	D9	D8
DAL	D7	D6	D5	D4	D3	D2	D1	D0
DACC	—	—	—	—	—	—	—	DACEN

16-bit D/A Converter Register List

• DAH register

Bit	7	6	5	4	3	2	1	0
Name	D15	D14	D13	D12	D11	D10	D9	D8
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0

D15~D8: 16-bit D/A converter data high byte

The 16-bit D/A converter Data low byte register, known as DAL, should first be modified and then followed by the DAH register modification. Each time when the DAH register is written, the whole 16-bit data will be loaded into the D/A converter and a conversion cycle will be initiated. Note that the D/A converter should first be enabled before the D/A converter data is updated.

• **DAL register**

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D7~D0**: 16-bit D/A converter data low byte
 Writing this register will only write the data to the shadow buffer and writing the DAH register will simultaneously copy the shadow buffer data to the DAL register. Note that the D/A converter should first be enabled before the D/A converter is updated.

• **DACC register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	—	DACEN
R/W	—	—	—	—	—	—	—	R/W
POR	—	—	—	—	—	—	—	0

Bit 7~1 Unimplemented, read as “0”
 Bit 0 **DACEN**: D/A converter enable or disable control
 0: Disable
 1: Enable
 If the D/A converter is enable, users must wait t_{DACS} time to ensure the D/A converter circuit is stable. A time t_{DACS} should be allowed for the D/A converter circuit to stabilize. And the 16-bit D/A converter data register should be updated after D/A converter circuit is stable.

Universal Serial Interface Module – USIM

The device contains a Universal Serial Interface Module, which includes the four-line SPI interface, the two-line I²C interface and the two-line/single-wire UART interface types, to allow an easy method of communication with external peripheral hardware. Having relatively simple communication protocols, these serial interface types allow the microcontroller to interface to external SPI, I²C or UART based hardware such as sensors, Flash or EEPROM memory, etc. The USIM interface pins are pin-shared with other I/O pins therefore the USIM interface functional pins must first be selected using the corresponding pin-shared function selection bits. As all the interface types share the same pins and registers, the choice of whether the UART, SPI or I²C type is used is made using the UART mode selection bit, named UMD, and the SPI/I²C operating mode control bits, named SIM2~SIM0, in the SIMC0 register. These pull-high resistors of the USIM pin-shared I/O are selected using pull-high control registers when the USIM function is enabled and the corresponding pins are used as USIM input pins.

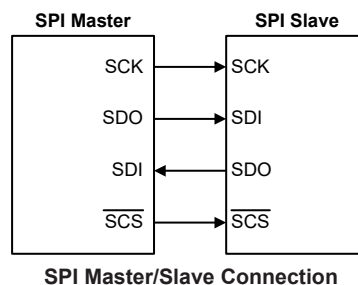
SPI Interface

The SPI interface is often used to communicate with external peripheral devices such as sensors, Flash or EEPROM memory devices etc. Originally developed by Motorola, the four-line SPI interface is a synchronous serial data interface that has a relatively simple communication protocol simplifying the programming requirements when communicating with external hardware devices.

The communication is full duplex and operates as a slave/master type, where the device can be either master or slave. Although the SPI interface specification can control multiple slave devices from a single master, but the device provides only one $\overline{SCS}$ pin. If the master needs to control multiple slave devices from a single master, the master can use I/O pin to select the slave devices.

SPI Interface Operation

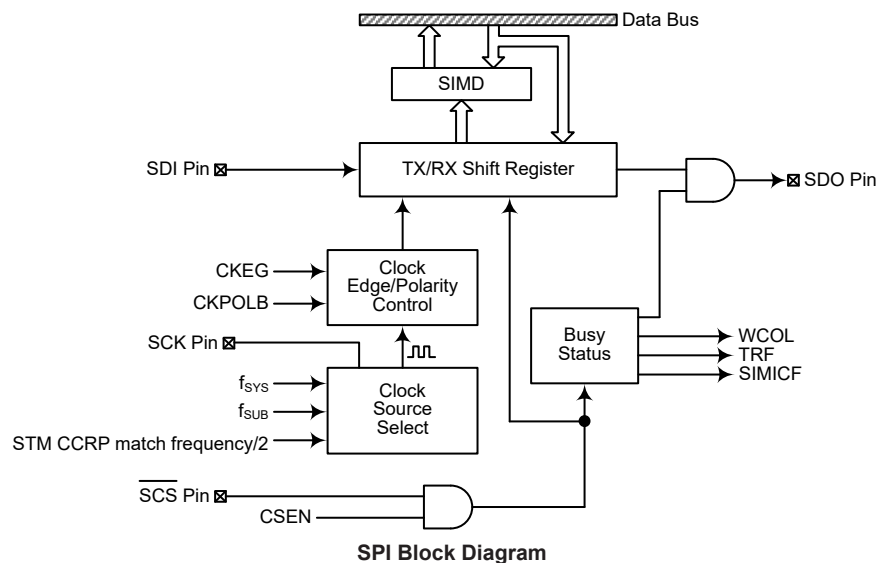
The SPI interface is a full duplex synchronous serial data link. It is a four-line interface with pin names SDI, SDO, SCK and $\overline{\text{SCS}}$. Pins SDI and SDO are the Serial Data Input and Serial Data Output lines, the SCK pin is the Serial Clock line and $\overline{\text{SCS}}$ is the Slave Select line. As the SPI interface pins are pin-shared with normal I/O pins and with the I²C/UART function pins, the SPI interface pins must first be selected by configuring the pin-shared function selection bits and setting the correct bits in the SIMC0 and SIMC2 registers. Communication between devices connected to the SPI interface is carried out in a slave/master mode with all data transfer initiations being implemented by the master. The Master also controls the clock signal. As the device only contains a single $\overline{\text{SCS}}$ pin only one slave device can be utilized. The $\overline{\text{SCS}}$ pin is controlled by software, set CSEN bit to 1 to enable $\overline{\text{SCS}}$ pin function, set CSEN bit to 0 the $\overline{\text{SCS}}$ pin will be floating state.



The SPI function in the device offers the following features:

- Full duplex synchronous data transfer
- Both Master and Slave modes
- LSB first or MSB first data transmission modes
- Transmission complete flag
- Rising or falling active clock edge

The status of the SPI interface pins is determined by a number of factors such as whether the device is in the master or slave mode and upon the condition of certain control bits such as CSEN and SIMEN.



SPI Registers

There are three internal registers which control the overall operation of the SPI interface. These are the SIMD data register and two control registers, SIMC0 and SIMC2. Note that the SIMC2 and SIMD registers and their POR values are only available when the SPI mode is selected by properly configuring the UMD and SIM2~SIM0 bits in the SIMC0 register.

Register Name	Bit							
	7	6	5	4	3	2	1	0
SIMC0	SIM2	SIM1	SIM0	UMD	SIMDEB1	SIMDEB0	SIMEN	SIMICF
SIMC2	D7	D6	CKPOLB	CKEG	MLS	CSEN	WCOL	TRF
SIMD	D7	D6	D5	D4	D3	D2	D1	D0

SPI Register List

SPI Data Register

The SIMD register is used to store the data being transmitted and received. The same register is used by both the SPI and I²C functions. Before the device writes data to the SPI bus, the actual data to be transmitted must be placed in the SIMD register. After the data is received from the SPI bus, the device can read it from the SIMD register. Any transmission or reception of data from the SPI bus must be made via the SIMD register.

• SIMD Register

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	x	x	x	x	x	x	x	x

“x”: unknown

Bit 7~0 **D7~D0**: USIM SPI/I²C data register bit 7 ~ bit 0

SPI Control Registers

There are also two control registers for the SPI interface, SIMC0 and SIMC2. The SIMC0 register is used to control the enable/disable function and to set the data transmission clock frequency. The SIMC2 register is used for other control functions such as LSB/MSB selection, write collision flag etc.

• SIMC0 Register

Bit	7	6	5	4	3	2	1	0
Name	SIM2	SIM1	SIM0	UMD	SIMDEB1	SIMDEB0	SIMEN	SIMICF
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	1	1	1	0	0	0	0	0

Bit 7~5 **SIM2~SIM0**: USIM SPI/I²C Operating Mode Control

- 000: SPI master mode; SPI clock is $f_{SYS}/4$
- 001: SPI master mode; SPI clock is $f_{SYS}/16$
- 010: SPI master mode; SPI clock is $f_{SYS}/64$
- 011: SPI master mode; SPI clock is f_{SUB}
- 100: SPI master mode; SPI clock is STM CCRP match frequency/2
- 101: SPI slave mode
- 110: I²C slave mode
- 111: Unused mode

When the UMD bit is cleared to zero, these bits setup the SPI or I²C operating mode of the USIM function. As well as selecting if the I²C or SPI function, they are used

to control the SPI Master/Slave selection and the SPI Master clock frequency. The SPI clock is a function of the system clock but can also be chosen to be sourced from STM and f_{SUB} . If the SPI Slave Mode is selected then the clock will be supplied by an external Master device.

- Bit 4 **UMD**: UART mode selection bit
 0: SPI or I²C mode
 1: UART mode

This bit is used to select the UART mode. When this bit is cleared to zero, the actual SPI or I²C mode can be selected using the SIM2~SIM0 bits. Note that the UMD bit must be cleared to zero for SPI or I²C mode.

- Bit 3~2 **SIMDEB1~SIMDEB0**: I²C Debounce Time Selection

These bits are only available when the USIM is configured to operate in the I²C mode. Refer to the I²C register section.

- Bit 1 **SIMEN**: USIM SPI/I²C Enable Control
 0: Disable
 1: Enable

The bit is the overall on/off control for the USIM SPI/I²C interface. When the SIMEN bit is cleared to zero to disable the USIM SPI/I²C interface, the SDI, SDO, SCK and $\overline{SCS}$, or SDA and SCL lines will lose their SPI or I²C function and the USIM operating current will be reduced to a minimum value. When the bit is high the USIM SPI/I²C interface is enabled. If the USIM is configured to operate as an SPI interface via the UMD and SIM2~SIM0 bits, the contents of the SPI control registers will remain at the previous settings when the SIMEN bit changes from low to high and should therefore be first initialised by the application program. If the USIM is configured to operate as an I²C interface via the UMD and SIM2~SIM0 bits and the SIMEN bit changes from low to high, the contents of the I²C control bits such as HTX and TXAK will remain at the previous settings and should therefore be first initialised by the application program while the relevant I²C flags such as HCF, HAAS, HBB, SRW and RXAK will be set to their default states.

- Bit 0 **SIMICF**: USIM SPI Incomplete Flag
 0: USIM SPI incomplete condition is not occurred
 1: USIM SPI incomplete condition is occurred

This bit is only available when the USIM is configured to operate in an SPI slave mode. If the SPI operates in the slave mode with the SIMEN and CSEN bits both being set high but the $\overline{SCS}$ line is pulled high by the external master device before the SPI data transfer is completely finished, the SIMICF bit will be set high together with the TRF bit. When this condition occurs, the corresponding interrupt will occur if the interrupt function is enabled. However, the TRF bit will not be set high if the SIMICF bit is set high by software application program.

• SIMC2 Register

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	CKPOLB	CKEG	MLS	CSEN	WCOL	TRF
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

- Bit 7~6 **D7~D6**: Undefined bits

These bits can be read or written by application program.

- Bit 5 **CKPOLB**: SPI clock line base condition selection
 0: The SCK line will be high when the clock is inactive
 1: The SCK line will be low when the clock is inactive

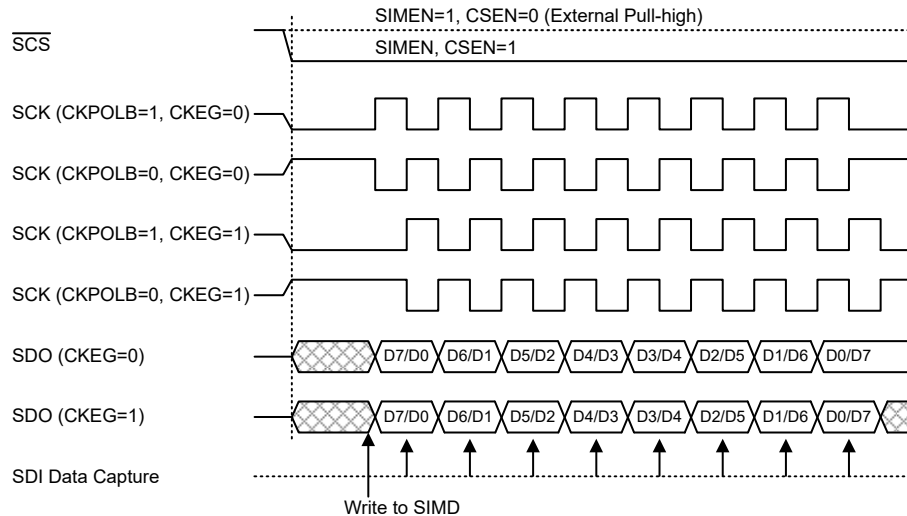
The CKPOLB bit determines the base condition of the clock line, if the bit is high, then the SCK line will be low when the clock is inactive. When the CKPOLB bit is low, then the SCK line will be high when the clock is inactive.

- Bit 4 **CKEG**: SPI SCK active clock edge type selection
CKPOLB=0
0: SCK is high base level and data capture at SCK rising edge
1: SCK is high base level and data capture at SCK falling edge
CKPOLB=1
0: SCK is low base level and data capture at SCK falling edge
1: SCK is low base level and data capture at SCK rising edge
The CKEG and CKPOLB bits are used to setup the way that the clock signal outputs and inputs data on the SPI bus. These two bits must be configured before data transfer is executed otherwise an erroneous clock edge may be generated. The CKPOLB bit determines the base condition of the clock line, if the bit is high, then the SCK line will be low when the clock is inactive. When the CKPOLB bit is low, then the SCK line will be high when the clock is inactive. The CKEG bit determines active clock edge type which depends upon the condition of CKPOLB bit.
- Bit 3 **MLS**: SPI data shift order
0: LSB first
1: MSB first
This is the data shift select bit and is used to select how the data is transferred, either MSB or LSB first. Setting the bit high will select MSB first and low for LSB first.
- Bit 2 **CSEN**: SPI $\overline{SCS}$ pin control
0: Disable
1: Enable
The CSEN bit is used as an enable/disable for the $\overline{SCS}$ pin. If this bit is low, then the $\overline{SCS}$ pin will be disabled and placed into a floating condition. If the bit is high the $\overline{SCS}$ pin will be enabled and used as a select pin.
- Bit 1 **WCOL**: SPI write collision flag
0: No collision
1: Collision
The WCOL flag is used to detect if a data collision has occurred. If this bit is high it means that data has been attempted to be written to the SIMD register during a data transfer operation. This writing operation will be ignored if data is being transferred. The bit can be cleared to zero by the application program.
- Bit 0 **TRF**: SPI Transmit/Receive complete flag
0: SPI data is being transferred
1: SPI data transmission is completed
The TRF bit is the Transmit/Receive Complete flag and is set “1” automatically when an SPI data transmission is completed, but must set to “0” by the application program. It can be used to generate an interrupt.

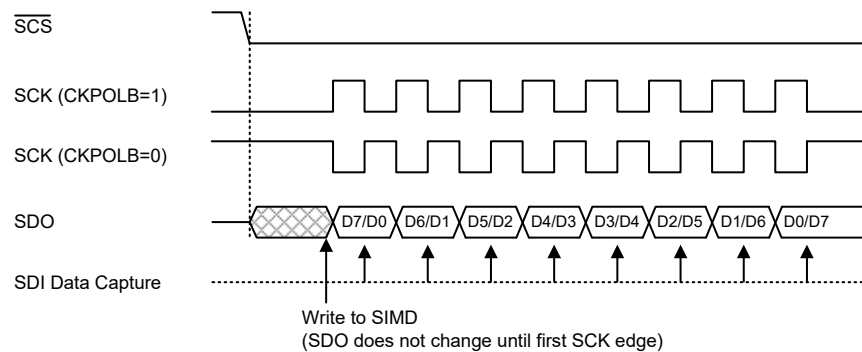
SPI Communication

After the SPI interface is enabled by setting the SIMEN bit high, then in the Master Mode, when data is written to the SIMD register, transmission/reception will begin simultaneously. When the data transfer is completed, the TRF flag will be set high automatically, but must be cleared using the application program. In the Slave Mode, when the clock signal from the master has been received, any data in the SIMD register will be transmitted and any data on the SDI pin will be shifted into the SIMD register. The master should output an $\overline{SCS}$ signal to enable the slave devices before a clock signal is provided. The slave data to be transferred should be well prepared at the appropriate moment relative to the SCK signal depending upon the configurations of the CKPOLB bit and CKEG bit. The accompanying timing diagram shows the relationship between the slave data and SCK signal for various configurations of the CKPOLB and CKEG bits.

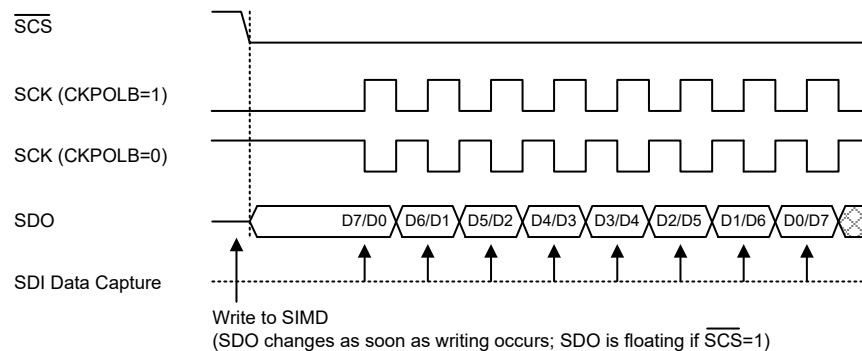
The SPI Master mode will continue to function if the SPI clock is running.



SPI Master Mode Timing

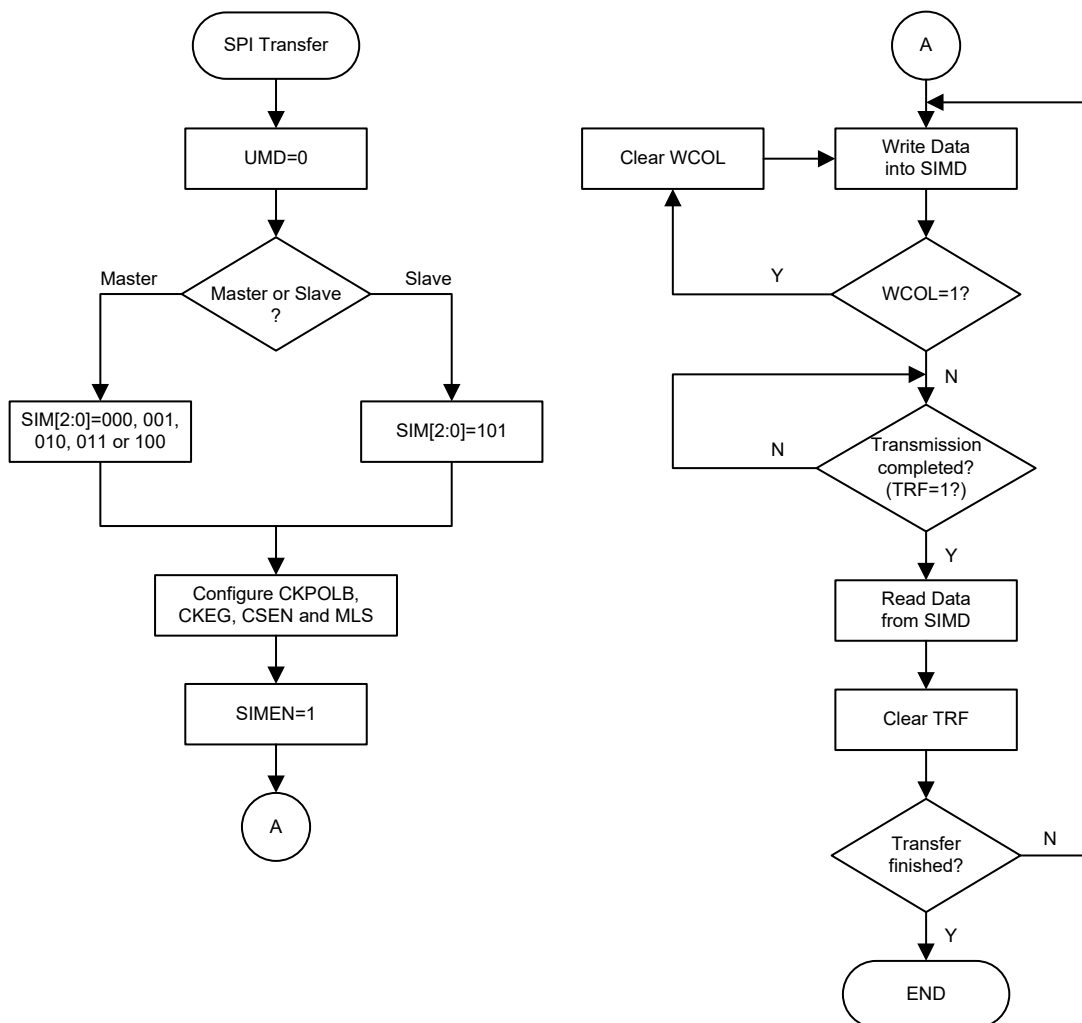


SPI Slave Mode Timing – CKEG=0



Note: For SPI slave mode, if SIMEN=1 and CSEN=0, SPI is always enabled and ignores the SCS level.

SPI Slave Mode Timing – CKEG=1



SPI Transfer Control Flowchart

SPI Bus Enable/Disable

To enable the SPI bus, set CSEN=1 and $\overline{SCS}$ =0, then wait for data to be written into the SIMD (TXRX buffer) register. For the Master Mode, after data has been written to the SIMD (TXRX buffer) register, then transmission or reception will start automatically. When all the data has been transferred, the TRF bit should be set. For the Slave Mode, when clock pulses are received on SCK, data in the TXRX buffer will be shifted out or data on SDI will be shifted in.

When the SPI bus is disabled, SCK, SDI, SDO and $\overline{SCS}$ can become I/O pins or other pin-shared functions using the corresponding pin-shared control bits.

SPI Operation Steps

All communication is carried out using the 4-line interface for either Master or Slave Mode.

The CSEN bit in the SIMC2 register controls the $\overline{SCS}$ pin function of the SPI interface. Setting this bit high will enable the SPI interface by allowing the $\overline{SCS}$ line to be active, which can then be used to control the SPI interface. If the CSEN bit is low, the SPI interface will be disabled and the $\overline{SCS}$ line will be in a floating condition and can therefore not be used for control of the SPI interface. If the CSEN bit and the SIMEN bit in the SIMC0 are set high, this will place the SDI

line in a floating condition and the SDO line high. If in Master Mode the SCK line will be either high or low depending upon the clock polarity selection bit CKPOLB in the SIMC2 register. If in Slave Mode the SCK line will be in a floating condition. If the SIMEN bit is low, then the bus will be disabled and $\overline{SCS}$, SDI, SDO and SCK will all become I/O pins or the other functions using the corresponding pin-shared control bits. In the Master Mode the Master will always generate the clock signal. The clock and data transmission will be initiated after data has been written into the SIMD register. In the Slave Mode, the clock signal will be received from an external master device for both data transmission and reception. The following sequences show the order to be followed for data transfer in both Master and Slave Mode.

Master Mode:

- Step 1
Select the SPI Master mode and clock source using the UMD and SIM2~SIM0 bits in the SIMC0 control register.
- Step 2
Setup the CSEN bit and setup the MLS bit to choose if the data is MSB or LSB first, this setting must be the same with the Slave devices.
- Step 3
Setup the SIMEN bit in the SIMC0 control register to enable the SPI interface.
- Step 4
For write operations: write the data to the SIMD register, which will actually place the data into the TXRX buffer. Then use the SCK and SDO lines to output the data. After this, go to step 5.
For read operations: the data transferred in on the SDI line will be stored in the TXRX buffer until all the data has been received at which point it will be latched into the SIMD register.
- Step 5
Check the WCOL bit if set high then a collision error has occurred so return to step 4. If equal to zero then go to the following step.
- Step 6
Check the TRF bit or wait for an USIM SPI serial bus interrupt.
- Step 7
Read data from the SIMD register.
- Step 8
Clear TRF.
- Step 9
Go to step 4.

Slave Mode:

- Step 1
Select the SPI Slave mode using the UMD and SIM2~SIM0 bits in the SIMC0 control register.
- Step 2
Setup the CSEN bit and setup the MLS bit to choose if the data is MSB or LSB first, this setting must be the same with the Master devices.
- Step 3
Setup the SIMEN bit in the SIMC0 control register to enable the SPI interface.
- Step 4
For write operations: write the data to the SIMD register, which will actually place the data into the TXRX buffer. Then wait for the master clock SCK and SDO signal. After this, go to step 5.

For read operations: the data transferred in on the SDI line will be stored in the TXRX buffer until all the data has been received at which point it will be latched into the SIMD register.

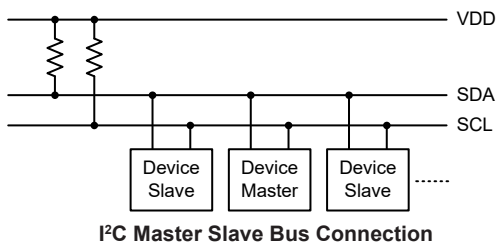
- Step 5
Check the WCOL bit if set high then a collision error has occurred so return to step 4. If equal to zero then go to the following step.
- Step 6
Check the TRF bit or wait for an USIM SPI serial bus interrupt.
- Step 7
Read data from the SIMD register.
- Step 8
Clear TRF.
- Step 9
Go to step 4.

Error Detection

The WCOL bit in the SIMC2 register is provided to indicate errors during data transfer. The bit is set by the SPI serial Interface but must be cleared by the application program. This bit indicates that a data collision has occurred which happens if a write to the SIMD register takes place during a data transfer operation and will prevent the write operation from continuing.

I²C Interface

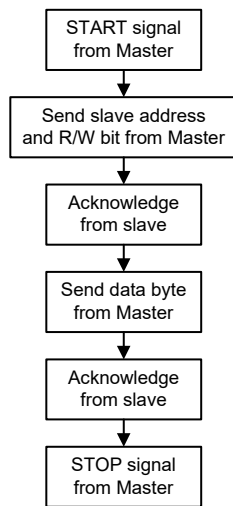
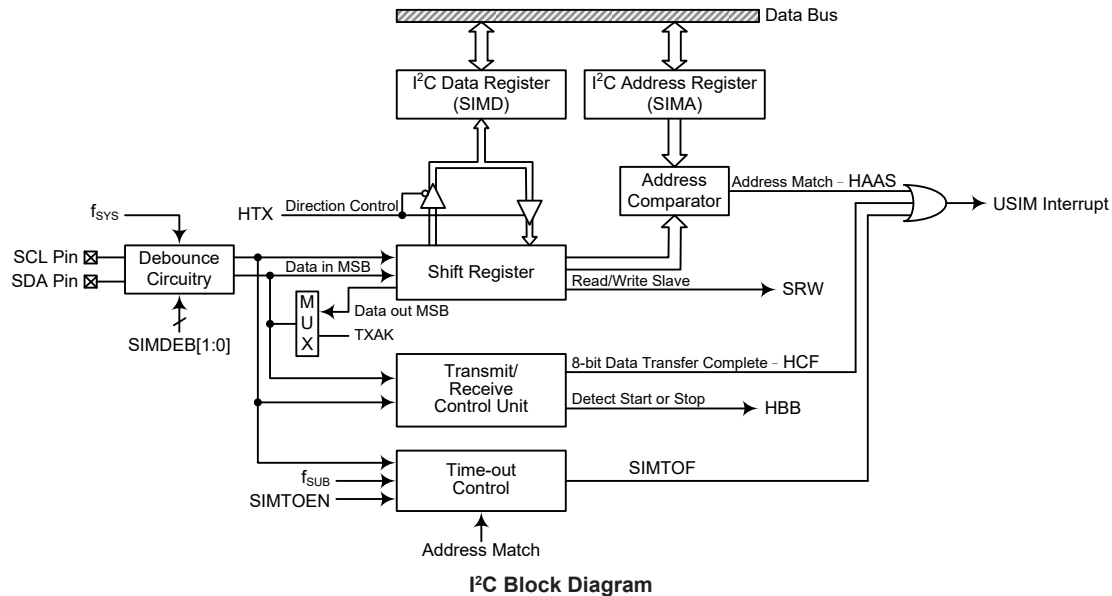
The I²C interface is used to communicate with external peripheral devices such as sensors, EEPROM memory etc. Originally developed by Philips, it is a two-line low speed serial interface for synchronous serial data transfer. The advantage of only two-lines for communication, relatively simple communication protocol and the ability to accommodate multiple devices on the same bus has made it an extremely popular interface type for many applications.



I²C Interface Operation

The I²C serial interface is a two-line interface, a serial data line, SDA, and serial clock line, SCL. As many devices may be connected together on the same bus, their outputs are both open drain types. For this reason it is necessary that external pull-high resistors are connected to these outputs. Note that no chip select line exists, as each device on the I²C bus is identified by a unique address which will be transmitted and received on the I²C bus.

When two devices communicate with each other on the bidirectional I²C bus, one is known as the master device and one as the slave device. Both master and slave can transmit and receive data, however, it is the master device that has overall control of the bus. For the device, which only operates in slave mode, there are two methods of transferring data on the I²C bus, the slave transmit mode and the slave receive mode. The pull-high control function pin-shared with SCL/SDA pin is still applicable even if I²C device is activated and the related internal pull-high register could be controlled by its corresponding pull-high control register.



I²C Interface Operation

The SIMDEB1 and SIMDEB0 bits determine the debounce time of the I²C interface. This uses the internal clock to in effect add a debounce time to the external clock to reduce the possibility of glitches on the clock line causing erroneous operation. The debounce time, if selected, can be chosen to be either 2 or 4 system clocks. To achieve the required I²C data transfer speed, there exists a relationship between the system clock, f_{SYS} , and the I²C debounce time. For either the I²C Standard or Fast mode operation, users must take care of the selected system clock frequency and the configured debounce time to match the criterion shown in the following table.

I²C Debounce Time Selection	I²C Standard Mode (100kHz)	I²C Fast Mode (400kHz)
No Debounce	$f_{SYS} > 2\text{MHz}$	$f_{SYS} > 4\text{MHz}$
2 system clock debounce	$f_{SYS} > 4\text{MHz}$	$f_{SYS} > 8\text{MHz}$
4 system clock debounce	$f_{SYS} > 4\text{MHz}$	$f_{SYS} > 8\text{MHz}$

I²C Minimum f_{SYS} Frequency Requirements

I²C Registers

There are three control registers associated with the I²C bus, SIMC0, SIMC1 and SIMTOC, one address register SIMA and one data register, SIMD. Note that the SIMC1, SIMD, SIMA and SIMTOC registers and their POR values are only available when the I²C mode is selected by properly configuring the UMD and SIM2~SIM0 bits in the SIMC0 register.

Register Name	Bit							
	7	6	5	4	3	2	1	0
SIMC0	SIM2	SIM1	SIM0	UMD	SIMDEB1	SIMDEB0	SIMEN	SIMICF
SIMC1	HCF	HAAS	HBB	HTX	TXAK	SRW	IAMWU	RXAK
SIMD	D7	D6	D5	D4	D3	D2	D1	D0
SIMA	SIMA6	SIMA5	SIMA4	SIMA3	SIMA2	SIMA1	SIMA0	D0
SIMTOC	SIMTOEN	SIMTOF	SIMTOS5	SIMTOS4	SIMTOS3	SIMTOS2	SIMTOS1	SIMTOS0

I²C Register List

I²C Data Register

The SIMD register is used to store the data being transmitted and received. The same register is used by both the SPI and I²C functions. Before the device writes data to the I²C bus, the actual data to be transmitted must be placed in the SIMD register. After the data is received from the I²C bus, the device can read it from the SIMD register. Any transmission or reception of data from the I²C bus must be made via the SIMD register.

• SIMD Register

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	x	x	x	x	x	x	x	x

"x": unknown

Bit 7~0 **D7~D0:** USIM SPI/I²C data register bit 7 ~ bit 0

I²C Address Register

The SIMA register is also used by the SPI interface but has the name, SIMC2. The SIMA register is the location where the 7-bit slave address of the slave device is stored. Bit 7~1 of the SIMA register define the device slave address. Bit 0 is not defined. When a master device, which is connected to the I²C bus, sends out an address, which matches the slave address in the SIMA register, the slave device will be selected.

• SIMA Register

Bit	7	6	5	4	3	2	1	0
Name	SIMA6	SIMA5	SIMA4	SIMA3	SIMA2	SIMA1	SIMA0	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~1 **SIMA6~SIMA0:** I²C slave address

SIMA6~SIMA0 is the 7-bit I²C slave address.

Bit 0 **D0:** Reserved bit, can be read or written by application program

I²C Control Registers

There are three control registers for the I²C interface, SIMC0, SIMC1 and SIMTOC. The SIMC0 register is used to control the enable/disable function and to select the I²C slave mode and debounce time. The SIMC1 register contains the relevant flags which are used to indicate the I²C communication status. Another register, SIMTOC, is used to control the I²C time-out function and is described in the corresponding section.

• SIMC0 Register

Bit	7	6	5	4	3	2	1	0
Name	SIM2	SIM1	SIM0	UMD	SIMDEB1	SIMDEB0	SIMEN	SIMICF
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	1	1	1	0	0	0	0	0

Bit 7~5 **SIM2~SIM0**: USIM SPI/I²C Operating Mode Control
 000: SPI master mode; SPI clock is $f_{SYS}/4$
 001: SPI master mode; SPI clock is $f_{SYS}/16$
 010: SPI master mode; SPI clock is $f_{SYS}/64$
 011: SPI master mode; SPI clock is f_{SUB}
 100: SPI master mode; SPI clock is STM CCRP match frequency/2
 101: SPI slave mode
 110: I²C slave mode
 111: Unused mode

When the UMD bit is cleared to zero, these bits setup the SPI or I²C operating mode of the USIM function. As well as selecting if the I²C or SPI function, they are used to control the SPI Master/Slave selection and the SPI Master clock frequency. The SPI clock is a function of the system clock but can also be chosen to be sourced from STM and f_{SUB} . If the SPI Slave Mode is selected then the clock will be supplied by an external Master device.

Bit 4 **UMD**: UART mode selection bit
 0: SPI or I²C mode
 1: UART mode

This bit is used to select the UART mode. When this bit is cleared to zero, the actual SPI or I²C mode can be selected using the SIM2~SIM0 bits. Note that the UMD bit must be set low for SPI or I²C mode.

Bit 3~2 **SIMDEB1~SIMDEB0**: I²C Debounce Time Selection
 00: No debounce
 01: 2 system clock debounce
 10: 4 system clock debounce
 11: 4 system clock debounce

These bits are used to select the I²C debounce time when the USIM is configured as the I²C interface function by setting the UMD bit to “0” and SIM2~SIM0 bits to “110”.

Bit 1 **SIMEN**: USIM SPI/I²C Enable Control
 0: Disable
 1: Enable

The bit is the overall on/off control for the USIM SPI/I²C interface. When the SIMEN bit is cleared to zero to disable the USIM SPI/I²C interface, the SDI, SDO, SCK and SCS, or SDA and SCL lines will lose their SPI or I²C function and the USIM operating current will be reduced to a minimum value. When the bit is high the USIM SPI/I²C interface is enabled. If the USIM is configured to operate as an SPI interface via the UMD and SIM2~SIM0 bits, the contents of the SPI control registers will remain at the previous settings when the SIMEN bit changes from low to high and should therefore be first initialised by the application program. If the USIM is configured to operate as an I²C interface via the UMD and SIM2~SIM0 bits and the SIMEN bit changes from low to high, the contents of the I²C control bits such as HTX and TXAK will remain

at the previous settings and should therefore be first initialised by the application program while the relevant I²C flags such as HCF, HAAS, HBB, SRW and RXAK will be set to their default states.

Bit 0 **SIMICF:** USIM SPI Incomplete Flag
This bit is only available when the USIM is configured to operate in an SPI slave mode. Refer to the SPI register section.

• **SIMC1 Register**

Bit	7	6	5	4	3	2	1	0
Name	HCF	HAAS	HBB	HTX	TXAK	SRW	IAMWU	RXAK
R/W	R	R	R	R/W	R/W	R	R/W	R
POR	1	0	0	0	0	0	0	1

Bit 7 **HCF:** I²C Bus data transfer completion flag
0: Data is being transferred
1: Completion of an 8-bit data transfer
The HCF flag is the data transfer flag. This flag will be zero when data is being transferred. Upon completion of an 8-bit data transfer the flag will go high and an interrupt will be generated.

Bit 6 **HAAS:** I²C Bus address match flag
0: Not address match
1: Address match
The HAAS flag is the address match flag. This flag is used to determine if the slave device address is the same as the master transmit address. If the addresses match then this bit will be high, if there is no match then the flag will be low.

Bit 5 **HBB:** I²C Bus busy flag
0: I²C Bus is not busy
1: I²C Bus is busy
The HBB flag is the I²C busy flag. This flag will be “1” when the I²C bus is busy which will occur when a START signal is detected. The flag will be set to “0” when the bus is free which will occur when a STOP signal is detected.

Bit 4 **HTX:** I²C slave device is transmitter or receiver selection
0: Slave device is the receiver
1: Slave device is the transmitter

Bit 3 **TXAK:** I²C Bus transmit acknowledge flag
0: Slave send acknowledge flag
1: Slave do not send acknowledge flag
The TXAK bit is the transmit acknowledge flag. After the slave device receipt of 8 bits of data, this bit will be transmitted to the bus on the 9th clock from the slave device. The slave device must always set TXAK bit to “0” before further data is received.

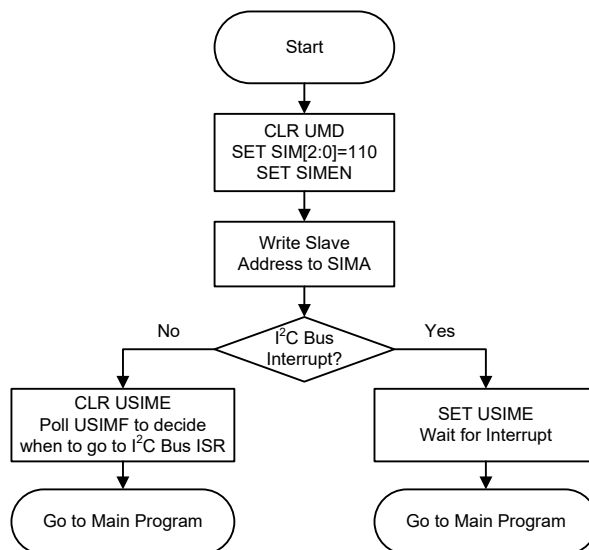
Bit 2 **SRW:** I²C Slave Read/Write flag
0: Slave device should be in receive mode
1: Slave device should be in transmit mode
The SRW flag is the I²C Slave Read/Write flag. This flag determines whether the master device wishes to transmit or receive data from the I²C bus. When the transmitted address and slave address is match, that is when the HAAS flag is set high, the slave device will check the SRW flag to determine whether it should be in transmit mode or receive mode. If the SRW flag is high, the master is requesting to read data from the bus, so the slave device should be in transmit mode. When the SRW flag is zero, the master will write data to the bus, therefore the slave device should be in receive mode to read this data.

Bit 1	IAMWU: I²C Address Match Wake-up control 0: Disable 1: Enable This bit should be set high to enable the I²C address match wake-up from the SLEEP or IDLE Mode. If the IAMWU bit has been set before entering either the SLEEP or IDLE mode to enable the I²C address match wake-up, then this bit must be cleared to zero by the application program after wake-up to ensure correction device operation.
Bit 0	RXAK: I²C Bus Receive acknowledge flag 0: Slave receive acknowledge flag 1: Slave does not receive acknowledge flag The RXAK flag is the receiver acknowledge flag. When the RXAK flag is “0”, it means that a acknowledge signal has been received at the 9th clock, after 8 bits of data have been transmitted. When the slave device in the transmit mode, the slave device checks the RXAK flag to determine if the master receiver wishes to receive the next byte. The slave transmitter will therefore continue sending out data until the RXAK flag is “1”. When this occurs, the slave transmitter will release the SDA line to allow the master to send a STOP signal to release the I²C Bus.

I²C Bus Communication

Communication on the I²C bus requires four separate steps, a START signal, a slave device address transmission, a data transmission and finally a STOP signal. When a START signal is placed on the I²C bus, all devices on the bus will receive this signal and be notified of the imminent arrival of data on the bus. The first seven bits of the data will be the slave address with the first bit being the MSB. If the address of the slave device matches that of the transmitted address, the HAAS bit in the SIMC1 register will be set and an USIM interrupt will be generated. After entering the interrupt service routine, the slave device must first check the condition of the HAAS and SIMTOF bits to determine whether the interrupt source originates from an address match or from the completion of an 8-bit data transfer completion or from the I²C bus time-out occurrence. During a data transfer, note that after the 7-bit slave address has been transmitted, the following bit, which is the 8th bit, is the read/write bit whose value will be placed in the SRW bit. This bit will be checked by the slave device to determine whether to go into transmit or receive mode. Before any transfer of data to or from the I²C bus, the microcontroller must initialise the bus, the following are steps to achieve this:

- Step 1
Set the UMD, SIM2~SIM0 and SIMEN bits in the SIMC0 register to “0”, “110” and “1” respectively to enable the I²C bus.
- Step 2
Write the slave address of the device to the I²C bus address register SIMA.
- Step 3
Set the USIME interrupt enable bit of the interrupt control register to enable the USIM interrupt.



I²C Bus Initialisation Flow Chart

I²C Bus Start Signal

The START signal can only be generated by the master device connected to the I²C bus and not by the slave device. This START signal will be detected by all devices connected to the I²C bus. When detected, this indicates that the I²C bus is busy and therefore the HBB bit will be set. A START condition occurs when a high to low transition on the SDA line takes place when the SCL line remains high.

I²C Slave Address

The transmission of a START signal by the master will be detected by all devices on the I²C bus. To determine which slave device the master wishes to communicate with, the address of the slave device will be sent out immediately following the START signal. All slave devices, after receiving this 7-bit address data, will compare it with their own 7-bit slave address. If the address sent out by the master matches the internal address of the microcontroller slave device, then an internal USIM I²C bus interrupt signal will be generated. The next bit following the address, which is the 8th bit, defines the read/write status and will be saved to the SRW bit of the SIMC1 register. The slave device will then transmit an acknowledge bit, which is a low level, as the 9th bit. The slave device will also set the status flag HAAS when the addresses match.

As an USIM I²C bus interrupt signal can come from three sources, when the program enters the interrupt subroutine, the HAAS and SIMTOF bits should be examined to see whether the interrupt source has come from a matching slave address or from the completion of a data byte transfer or from the I²C bus time-out occurrence. When a slave address is matched, the device must be placed in either the transmit mode and then write data to the SIMD register, or in the receive mode where it must implement a dummy read from the SIMD register to release the SCL line.

I²C Bus Read/Write Signal

The SRW bit in the SIMC1 register defines whether the master device wishes to read data from the I²C bus or write data to the I²C bus. The slave device should examine this bit to determine if it is to be a transmitter or a receiver. If the SRW flag is "1" then this indicates that the master device wishes to read data from the I²C bus, therefore the slave device must be setup to send data to the I²C bus as a transmitter. If the SRW flag is "0" then this indicates that the master wishes to send data to the I²C bus, therefore the slave device must be setup to read data from the I²C bus as a receiver.

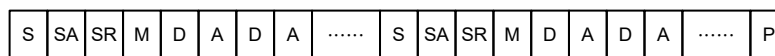
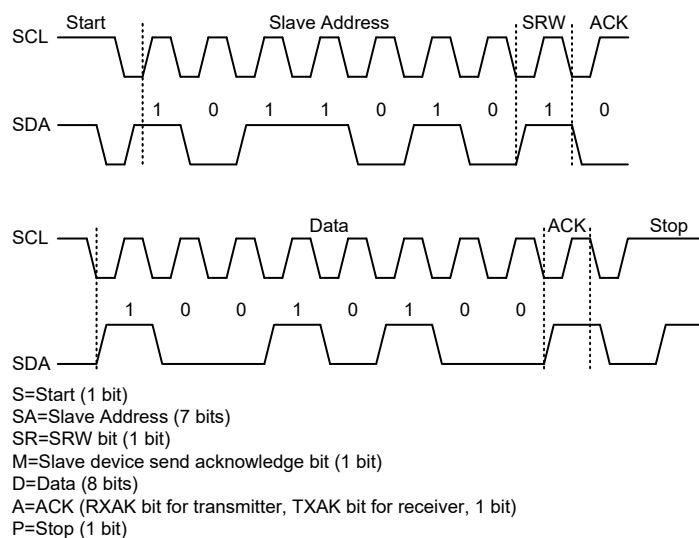
I²C Bus Slave Address Acknowledge Signal

After the master has transmitted a calling address, any slave device on the I²C bus, whose own internal address matches the calling address, must generate an acknowledge signal. The acknowledge signal will inform the master that a slave device has accepted its calling address. If no acknowledge signal is received by the master then a STOP signal must be transmitted by the master to end the communication. When the HAAS flag is high, the addresses have matched and the slave device must check the SRW flag to determine if it is to be a transmitter or a receiver. If the SRW flag is high, the slave device should be setup to be a transmitter so the HTX bit in the SIMC1 register should be set to “1”. If the SRW flag is low, then the microcontroller slave device should be setup as a receiver and the HTX bit in the SIMC1 register should be set to “0”.

I²C Bus Data and Acknowledge Signal

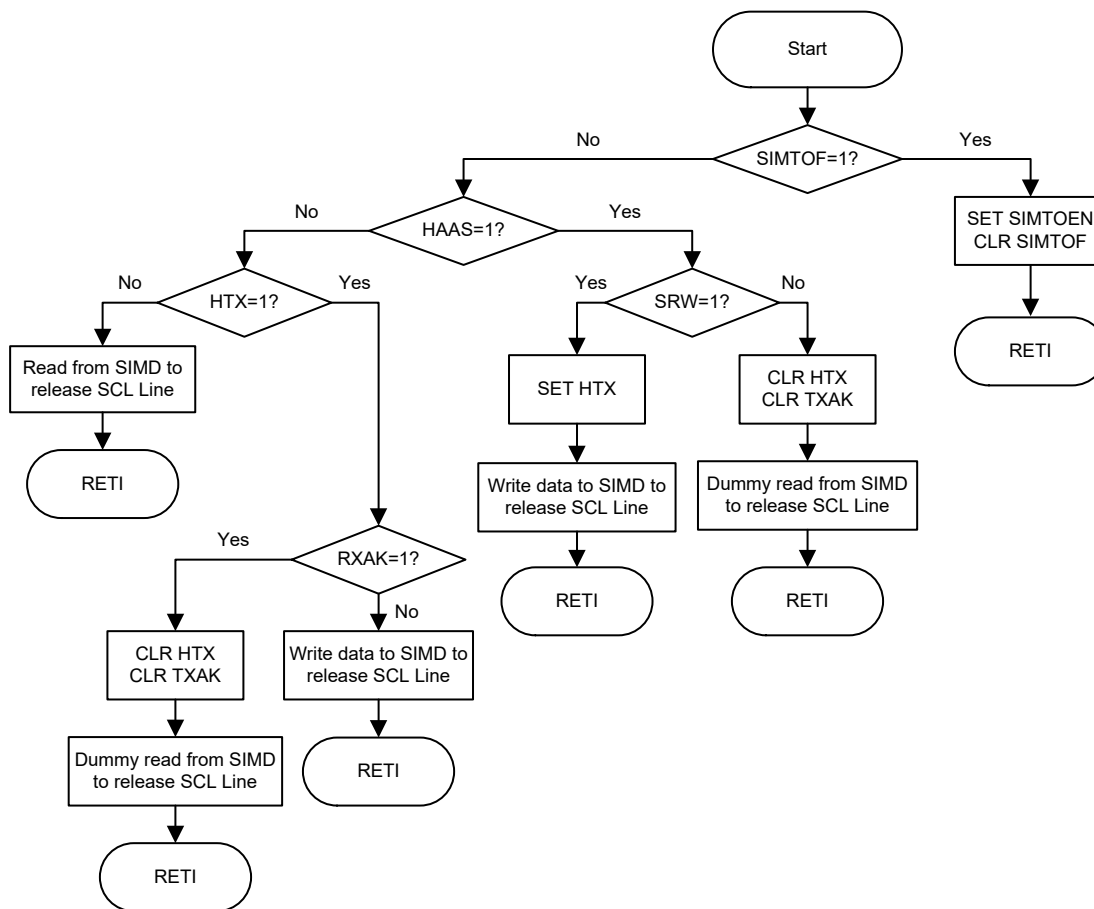
The transmitted data is 8-bit wide and is transmitted after the slave device has acknowledged receipt of its slave address. The order of serial bit transmission is the MSB first and the LSB last. After receipt of 8 bits of data, the receiver must transmit an acknowledge signal, level “0”, before it can receive the next data byte. If the slave transmitter does not receive an acknowledge bit signal from the master receiver, then the slave transmitter will release the SDA line to allow the master to send a STOP signal to release the I²C Bus. The corresponding data will be stored in the SIMD register. If setup as a transmitter, the slave device must first write the data to be transmitted into the SIMD register. If setup as a receiver, the slave device must read the transmitted data from the SIMD register.

When the slave receiver receives the data byte, it must generate an acknowledge bit, known as TXAK, on the 9th clock. The slave device, which is setup as a transmitter will check the RXAK bit in the SIMC1 register to determine if it is to send another data byte, if not then it will release the SDA line and await the receipt of a STOP signal from the master.



I²C Communication Timing Diagram

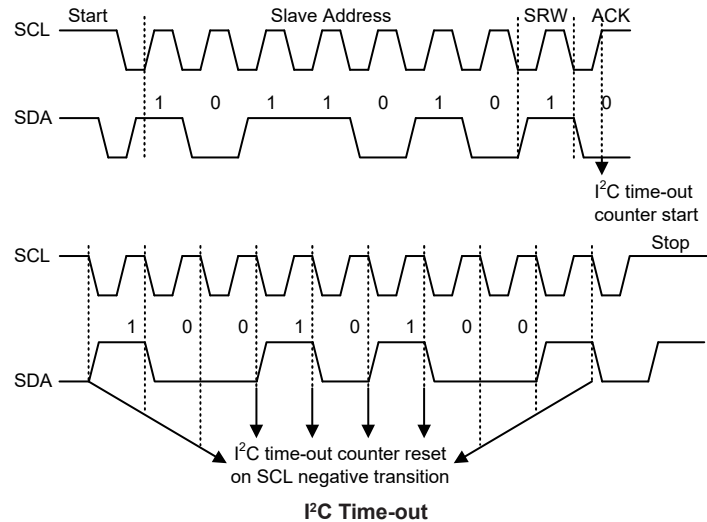
Note: When a slave address is matched, the device must be placed in either the transmit mode and then write data to the SIMD register, or in the receive mode where it must implement a dummy read from the SIMD register to release the SCL line.



I²C Bus ISR Flow Chart

I²C Time-out Control

In order to reduce the problem of I²C lockup due to reception of erroneous clock sources, a time-out function is provided. If the clock source to the I²C is not received for a while, then the I²C circuitry and registers will be reset after a certain time-out period. The time-out counter starts counting on an I²C bus “START” & “address match” condition, and is cleared by an SCL falling edge. Before the next SCL falling edge arrives, if the time elapsed is greater than the time-out setup by the SIMTOC register, then a time-out condition will occur. The time-out function will stop when an I²C “STOP” condition occurs.



When an I²C time-out counter overflow occurs, the counter will stop and the SIMTOEN bit will be cleared to zero and the SIMTOF bit will be set high to indicate that a time-out condition has occurred. The time-out condition will also generate an interrupt which uses the USIM interrupt vector. When an I²C time-out occurs, the I²C internal circuitry will be reset and the registers will be reset into the following condition:

Registers	After I ² C Time-out
SIMD, SIMA, SIMC0	No change
SIMC1	Reset to POR condition

I²C Registers after Time-out

The SIMTOF flag can be cleared by the application program. There are 64 time-out periods which can be selected using SIMTOS bit field in the SIMTOC register. The time-out time is given by the formula: $((1 \sim 64) \times 32) / f_{SUB}$. This gives a time-out period which ranges from about 1ms to 64ms.

• SIMTOC Register

Bit	7	6	5	4	3	2	1	0
Name	SIMTOEN	SIMTOF	SIMTOS5	SIMTOS4	SIMTOS3	SIMTOS2	SIMTOS1	SIMTOS0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7 **SIMTOEN**: USIM I²C Time-out control
0: Disable
1: Enable

Bit 6 **SIMTOF**: USIM I²C Time-out flag
0: No time-out occurred
1: Time-out occurred

This bit is set high when time-out occurs and can only be cleared to zero by application program.

Bit 5~0 **SIMTOS5~SIMTOS0**: USIM I²C Time-out period selection
I²C time-out clock source is $f_{SUB}/32$.
I²C time-out time is equal to $(SIMTOS[5:0]+1) \times (32/f_{SUB})$.

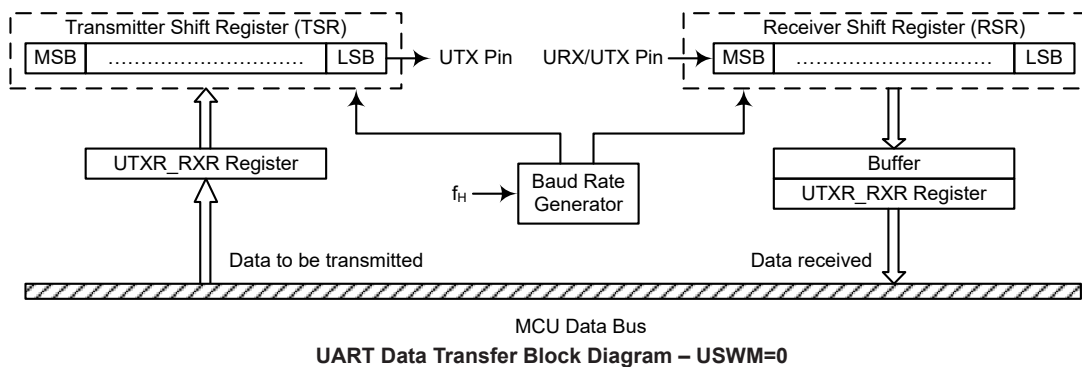
UART Interface

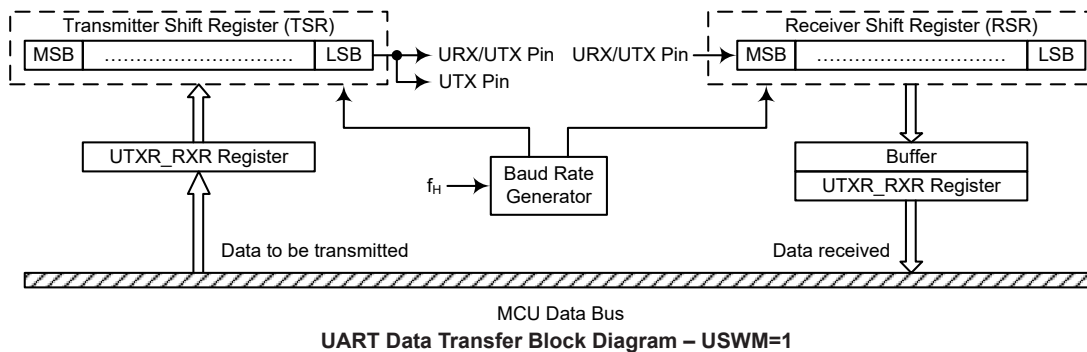
This UART interface function, which is part of the Serial Interface Module, should not be confused with the other independent UART function, which is described in another section of this datasheet.

The device contains an integrated full-duplex or half-duplex asynchronous serial communication UART interface that enables communication with external devices that contain a serial interface. The UART function has many features and can transmit and receive data serially by transferring a frame of data with eight or nine data bits per transmission as well as being able to detect errors when the data is overwritten or incorrectly framed. The UART function shares the same internal interrupt vector with the SPI and I²C interfaces which can be used to indicate when a reception occurs or when a transmission terminates.

The integrated UART function contains the following features:

- Full-duplex or half-duplex (single wire mode) asynchronous communication
- 8 or 9 bits character length
- Even, odd or no parity options
- One or two stop bits
- Baud rate generator with 8-bit prescaler
- Parity, framing, noise and overrun error detection
- Support for interrupt on address detect (last character bit=1)
- Separately enabled transmitter and receiver
- 2-byte Deep FIFO Receive Data Buffer
- URX/UTX pin wake-up function
- Transmit and receive interrupts
- Interrupts can be triggered by the following conditions:
 - ♦ Transmitter Empty
 - ♦ Transmitter Idle
 - ♦ Receiver Full
 - ♦ Receiver Overrun
 - ♦ Address Mode Detect





UART External Pins

To communicate with an external serial interface, the internal UART has two external pins known as UTX pin and URX/UTX pin, which are pin-shared with I/O or other pin functions. The UTX and URX/UTX pin function should first be selected by the corresponding pin-shared function selection register before the UART function is used. Along with the UMD bit, the UREN bit, the UTXEN or URXEN bits, if set, will setup these pins to transmitter output and receiver input conditions. At this time the internal pull-high resistor related to the transmitter output pin will be disabled, while the internal pull-high resistor related to the receiver input pin is controlled by the corresponding I/O pull-high function control bit. When the UTX or URX/UTX pin function is disabled by clearing the UMD, UREN, UTXEN or URXEN bit, the UTX or URX/UTX pin will be set to a floating state. At this time whether the internal pull-high resistor is connected to the UTX or URX/UTX pin or not is determined by the corresponding I/O pull-high function control bit.

UART Single Wire Mode

The UART function also supports a Single Wire Mode communication which is selected using the USWM bit in the UUCR3 register. When the USWM bit is set high, the UART function will be in the single wire mode. In the single wire mode, a single URX/UTX pin can be used to transmit and receive data depending upon the corresponding control bits. When the URXEN bit is set high, the URX/UTX pin is used as a receiver pin. When the URXEN bit is cleared to zero and the UTXEN bit is set high, the URX/UTX pin will act as a transmitter pin.

It is recommended not to set both the URXEN and UTXEN bits high in the single wire mode. If both the URXEN and UTXEN bits are set high, the URXEN bit will have the priority and the UART will act as a receiver.

It is important to note that the functional description in this UART chapter, which is described from the full-duplex communication standpoint, also applies to the half-duplex (single wire mode) communication except the pin usage. In the single wire mode, the UTX pin mentioned in this chapter should be replaced by the URX/UTX pin to understand the whole UART single wire mode function.

In the single wire mode, the data can also be transmitted on the UTX pin in a transmission operation with proper software configurations. Therefore, the data will be output on the URX/UTX and UTX pins.

UART Data Transfer Scheme

The UART Data Transfer Block Diagram shows the overall data transfer structure arrangement for the UART. The actual data to be transmitted from the MCU is first transferred to the UTXR_RXR register by the application program. The data will then be transferred to the Transmit Shift Register from where it will be shifted out, LSB first, onto the UTX pin at a rate controlled by the Baud Rate Generator. Only the UTXR_RXR register is mapped onto the MCU Data Memory, the Transmit Shift Register is not mapped and is therefore inaccessible to the application program.

Data to be received by the UART is accepted on the external URX/UTX pin, from where it is shifted in, LSB first, to the Receiver Shift Register at a rate controlled by the Baud Rate Generator. When the shift register is full, the data will then be transferred from the shift register to the internal UTXR_RXR register, where it is buffered and can be manipulated by the application program. Only the UTXR_RXR register is mapped onto the MCU Data Memory, the Receiver Shift Register is not mapped and is therefore inaccessible to the application program.

It should be noted that the actual register for data transmission and reception only exists as a single shared register in the Data Memory. This shared register known as the UTXR_RXR register is used for both data transmission and data reception.

UART Status and Control Registers

There are seven control registers associated with the UART function. The UMD bit in the SIMC0 register can be used to select the UART interface. The USWM bit in the UUCR3 register is used to enable/disable the UART Single Wire Mode. The UUSR, UUCR1 and UUCR2 registers control the overall function of the UART, while the UBRG register controls the Baud rate. The actual data to be transmitted and received on the serial interface is managed through the UTXR_RXR data register. Note that UART related registers and their POR values are only available when the UART mode is selected by setting the UMD bit in the SIMC0 register to "1".

Register Name	Bit							
	7	6	5	4	3	2	1	0
SIMC0	SIM2	SIM1	SIM0	UMD	SIMDEB1	SIMDEB0	SIMEN	SIMICF
UUSR	UPERR	UNF	UFERR	UOERR	URIDLE	URXIF	UTIDLE	UTXIF
UUCR1	UREN	UBNO	UPREN	UPRT	USTOPS	UTXBRK	URX8	UTX8
UUCR2	UTXEN	URXEN	UBRGH	UADDEN	UWAKE	URIE	UTIIE	UTEIE
UUCR3	—	—	—	—	—	—	—	USWM
UTXR_RXR	UTXRX7	UTXRX6	UTXRX5	UTXRX4	UTXRX3	UTXRX2	UTXRX1	UTXRX0
UBRG	UBRG7	UBRG6	UBRG5	UBRG4	UBRG3	UBRG2	UBRG1	UBRG0

UART Register List

• SIMC0 Register

Bit	7	6	5	4	3	2	1	0
Name	SIM2	SIM1	SIM0	UMD	SIMDEB1	SIMDEB0	SIMEN	SIMICF
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	1	1	1	0	0	0	0	0

- Bit 7~5 **SIM2~SIM0**: USIM SPI/I²C Operating Mode Control
When the UMD bit is cleared to zero, these bits setup the SPI or I²C operating mode of the USIM function. Refer to the SPI or I²C register section for more details.
- Bit 4 **UMD**: UART mode selection bit
0: SPI or I²C mode
1: UART mode
This bit is used to select the UART mode. When this bit is cleared to zero, the actual SPI or I²C mode can be selected using the SIM2~SIM0 bits. Note that the UMD bit must be set low for SPI or I²C mode.
- Bit 3~2 **SIMDEB1~SIMDEB0**: I²C Debounce Time Selection
Refer to the I²C register section.
- Bit 1 **SIMEN**: USIM SPI/I²C Enable Control
This bit is only available when the USIM is configured to operate in an SPI or I²C mode with the UMD bit set low. Refer to the SPI or I²C register section for more details.

Bit 0 **SIMICF**: USIM SPI Incomplete Flag
Refer to the SPI register section.

• **UUSR Register**

The UUSR register is the status register for the UART, which can be read by the program to determine the present status of the UART. All flags within the UUSR register are read only. Further explanation on each of the flags is given below:

Bit	7	6	5	4	3	2	1	0
Name	UPERR	UNF	UFERR	UOERR	URIDLE	URXIF	UTIDLE	UTXIF
R/W	R	R	R	R	R	R	R	R
POR	0	0	0	0	1	0	1	1

Bit 7 **UPERR**: Parity error flag
0: No parity error is detected
1: Parity error is detected
The UPERR flag is the parity error flag. When this read only flag is “0”, it indicates a parity error has not been detected. When the flag is “1”, it indicates that the parity of the received word is incorrect. This error flag is applicable only if Parity mode (odd or even) is selected. The flag can also be cleared to zero by a software sequence which involves a read to the status register UUSR followed by an access to the UTXR_RXR data register.

Bit 6 **UNF**: Noise flag
0: No noise is detected
1: Noise is detected
The UNF flag is the noise flag. When this read only flag is “0”, it indicates no noise condition. When the flag is “1”, it indicates that the UART has detected noise on the receiver input. The UNF flag is set during the same cycle as the URXIF flag but will not be set in the case of an overrun. The UNF flag can be cleared to zero by a software sequence which will involve a read to the status register UUSR followed by an access to the UTXR_RXR data register.

Bit 5 **UFERR**: Framing error flag
0: No framing error is detected
1: Framing error is detected
The UFERR flag is the framing error flag. When this read only flag is “0”, it indicates that there is no framing error. When the flag is “1”, it indicates that a framing error has been detected for the current character. The flag can also be cleared to zero by a software sequence which will involve a read to the status register UUSR followed by an access to the UTXR_RXR data register.

Bit 4 **UOERR**: Overrun error flag
0: No overrun error is detected
1: Overrun error is detected
The UOERR flag is the overrun error flag which indicates when the receiver buffer has overflowed. When this read only flag is “0”, it indicates that there is no overrun error. When the flag is “1”, it indicates that an overrun error occurs which will inhibit further transfers to the UTXR_RXR receive data register. The flag is cleared to zero by a software sequence, which is a read to the status register UUSR followed by an access to the UTXR_RXR data register.

Bit 3 **URIDLE**: Receiver status
0: Data reception is in progress (Data being received)
1: No data reception is in progress (Receiver is idle)
The URIDLE flag is the receiver status flag. When this read only flag is “0”, it indicates that the receiver is between the initial detection of the start bit and the completion of the stop bit. When the flag is “1”, it indicates that the receiver is idle. Between the completion of the stop bit and the detection of the next start bit, the

URIDLE bit is “1” indicating that the UART receiver is idle and the URX/UTX pin stays in logic high condition.

Bit 2 **URXIF**: Receive UTXR_RXR data register status

0: UTXR_RXR data register is empty

1: UTXR_RXR data register has available data

The URXIF flag is the receive data register status flag. When this read only flag is “0”, it indicates that the UTXR_RXR read data register is empty. When the flag is “1”, it indicates that the UTXR_RXR read data register contains new data. When the contents of the shift register are transferred to the UTXR_RXR register, an interrupt is generated if URIE=1 in the UUCR2 register. If one or more errors are detected in the received word, the appropriate receive-related flags UNF, UFERR, and/or UPERR are set within the same clock cycle. The URXIF flag will eventually be cleared to zero when the UUSR register is read with URXIF set, followed by a read from the UTXR_RXR register, and if the UTXR_RXR register has no more new data available.

Bit 1 **UTIDLE**: Transmission idle

0: Data transmission is in progress (Data being transmitted)

1: No data transmission is in progress (Transmitter is idle)

The UTIDLE flag is known as the transmission complete flag. When this read only flag is “0”, it indicates that a transmission is in progress. This flag will be set high when the UTXIF flag is “1” and when there is no transmit data or break character being transmitted. When UTIDLE is equal to “1”, the UTX pin becomes idle with the pin state in logic high condition. The UTIDLE flag is cleared to zero by reading the UUSR register with UTIDLE set and then writing to the UTXR_RXR register. The flag is not generated when a data character or a break is queued and ready to be sent.

Bit 0 **UTXIF**: Transmit UTXR_RXR data register status

0: Character is not transferred to the transmit shift register

1: Character has transferred to the transmit shift register (UTXR_RXR data register is empty)

The UTXIF flag is the transmit data register empty flag. When this read only flag is “0”, it indicates that the character is not transferred to the transmitter shift register. When the flag is “1”, it indicates that the transmitter shift register has received a character from the UTXR_RXR data register. The UTXIF flag is cleared to zero by reading the UART status register (UUSR) with UTXIF set and then writing to the UTXR_RXR data register. Note that when the UTXEN bit is set, the UTXIF flag bit will also be set since the transmit data register is not yet full.

• UUCR1 Register

The UUCR1 register together with the UUCR2 and UUCR3 registers are the three UART control registers that are used to set the various options for the UART function, such as overall on/off control, parity control, data transfer bit length, single wire mode communication etc. Further explanation on each of the bits is given below:

Bit	7	6	5	4	3	2	1	0
Name	UREN	UBNO	UPREN	UPRT	USTOPS	UTXBRK	URX8	UTX8
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R	W
POR	0	0	0	0	0	0	x	0

“x”: unknown

Bit 7 **UREN**: UART function enable control

0: Disable UART. UTX and URX/UTX pins are in a floating state

1: Enable UART. UTX and URX/UTX pins function as UART pins

The UREN bit is the UART enable bit. When this bit is equal to “0”, the UART will be disabled and the URX/UTX pin as well as the UTX pin will be set in a floating state. When the bit is equal to “1”, the UART will be enabled if the UMD bit is set and the UTX and URX/UTX pins will function as defined by the USWM mode selection bit together with the UTXEN and URXEN enable control bits.

When the UART is disabled, it will empty the buffer so any character remaining in the buffer will be discarded. In addition, the value of the baud rate counter will be reset. If the UART is disabled, all error and status flags will be reset. Also the UTXEN, URXEN, UTXBRK, URXIF, UOERR, UFERR, UPERR and UNF bits will be cleared to zero, while the UTIDLE, UTXIF and URIDLE bits will be set high. Other control bits in UUCR1, UUCR2, UUCR3 and UBRG registers will remain unaffected. If the UART is active and the UREN bit is cleared to zero, all pending transmissions and receptions will be terminated and the module will be reset as defined above. When the UART is re-enabled, it will restart in the same configuration.

- Bit 6 UBNO:** Number of data transfer bits selection
 0: 8-bit data transfer
 1: 9-bit data transfer
- This bit is used to select the data length format, which can have a choice of either 8-bit or 9-bit format. When this bit is equal to “1”, a 9-bit data length format will be selected. If the bit is equal to “0”, then an 8-bit data length format will be selected. If 9-bit data length format is selected, then bits URX8 and UTX8 will be used to store the 9th bit of the received and transmitted data respectively.
- Bit 5 UPREN:** Parity function enable control
 0: Parity function is disabled
 1: Parity function is enabled
- This is the parity enable bit. When this bit is equal to “1”, the parity function will be enabled. If the bit is equal to “0”, then the parity function will be disabled.
- Bit 4 UPRT:** Parity type selection bit
 0: Even parity for parity generator
 1: Odd parity for parity generator
- This bit is the parity type selection bit. When this bit is equal to “1”, odd parity type will be selected. If the bit is equal to “0”, then even parity type will be selected.
- Bit 3 USTOPS:** Number of Stop bits selection for transmitter
 0: One stop bit format is used
 1: Two stop bits format is used
- This bit determines if one or two stop bits are to be used for transmitter. When this bit is equal to “1”, two stop bits are used. If this bit is equal to “0”, then only one stop bit is used.
- Bit 2 UTXBRK:** Transmit break character
 0: No break character is transmitted
 1: Break characters transmit
- The UTXBRK bit is the Transmit Break Character bit. When this bit is “0”, there are no break characters and the UTX pin operates normally. When the bit is “1”, there are transmit break characters and the transmitter will send logic zeros. When this bit is equal to “1”, after the buffered data has been transmitted, the transmitter output is held low for a minimum of a 13-bit length and until the UTXBRK bit is reset.
- Bit 1 URX8:** Receive data bit 8 for 9-bit data transfer format (read only)
- This bit is only used if 9-bit data transfers are used, in which case this bit location will store the 9th bit of the received data known as URX8. The UBNO bit is used to determine whether data transfers are in 8-bit or 9-bit format.
- Bit 0 UTX8:** Transmit data bit 8 for 9-bit data transfer format (write only)
- This bit is only used if 9-bit data transfers are used, in which case this bit location will store the 9th bit of the transmitted data known as UTX8. The UBNO bit is used to determine whether data transfers are in 8-bit or 9-bit format.

• UUCR2 Register

The UUCR2 register is the second of the UART control registers and serves several purposes. One of its main functions is to control the basic enable/disable operation of the UART Transmitter and Receiver as well as enabling the various USIM UART mode interrupt sources. The register also serves to control the baud rate speed, receiver wake-up enable and the address detect enable. Further explanation on each of the bits is given below:

Bit	7	6	5	4	3	2	1	0
Name	UTXEN	URXEN	UBRGH	UADDEN	UWAKE	URIE	UTIIE	UTEIE
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7 **UTXEN**: UART Transmitter enabled control

0: UART transmitter is disabled

1: UART transmitter is enabled

The bit named UTXEN is the Transmitter Enable Bit. When this bit is equal to “0”, the transmitter will be disabled with any pending data transmissions being aborted. In addition the buffers will be reset. In this situation the UTX pin will be set in a floating state.

If the UTXEN bit is equal to “1” and the UMD and UREN bit are also equal to “1”, the transmitter will be enabled and the UTX pin will be controlled by the UART. Clearing the UTXEN bit during a transmission will cause the data transmission to be aborted and will reset the transmitter. If this situation occurs, the UTX pin will be set in a floating state.

Bit 6 **URXEN**: UART Receiver enabled control

0: UART receiver is disabled

1: UART receiver is enabled

The bit named URXEN is the Receiver Enable Bit. When this bit is equal to “0”, the receiver will be disabled with any pending data receptions being aborted. In addition the receive buffers will be reset. In this situation the URX/UTX pin will be set in a floating state. If the URXEN bit is equal to “1” and the UMD and UREN bit are also equal to “1”, the receiver will be enabled and the URX/UTX pin will be controlled by the UART. Clearing the URXEN bit during a reception will cause the data reception to be aborted and will reset the receiver. If this situation occurs, the URX/UTX pin will be set in a floating state.

Bit 5 **UBRGH**: Baud Rate speed selection

0: Low speed baud rate

1: High speed baud rate

The bit named UBRGH selects the high or low speed mode of the Baud Rate Generator. This bit, together with the value placed in the baud rate register UBRG, controls the Baud Rate of the UART. If this bit is equal to “1”, the high speed mode is selected. If the bit is equal to “0”, the low speed mode is selected.

Bit 4 **UADDEN**: Address detect function enable control

0: Address detect function is disabled

1: Address detect function is enabled

The bit named UADDEN is the address detect function enable control bit. When this bit is equal to “1”, the address detect function is enabled. When it occurs, if the 8th bit, which corresponds to UTXRX7 if UBNO=0 or the 9th bit, which corresponds to URX8 if UBNO=1, has a value of “1”, then the received word will be identified as an address, rather than data. If the corresponding interrupt is enabled, an interrupt request will be generated each time the received word has the address bit set, which is the 8th or 9th bit depending on the value of UBNO. If the address bit known as the 8th or 9th bit of the received word is “0” with the address detect function being enabled, an interrupt will not be generated and the received data will be discarded.

- Bit 3 **UWAKE**: URX/UTX pin wake-up UART function enable control
 0: URX/UTX pin wake-up UART function is disabled
 1: URX/UTX pin wake-up UART function is enabled
 This bit is used to control the wake-up UART function when a falling edge on the URX/UTX pin occurs. Note that this bit is only available when the UART clock (f_{H1}) is switched off. There will be no URX/UTX pin wake-up UART function if the UART clock (f_{H1}) exists. If the UWAKE bit is set high as the UART clock (f_{H1}) is switched off, a UART wake-up request will be initiated when a falling edge on the URX/UTX pin occurs. When this request happens and the corresponding interrupt is enabled, an URX/UTX pin wake-up UART interrupt will be generated to inform the MCU to wake up the UART function by switching on the UART clock (f_{H1}) via the application program. Otherwise, the UART function cannot resume even if there is a falling edge on the URX/UTX pin when the UWAKE bit is cleared to zero.
- Bit 2 **URIE**: Receiver interrupt enable control
 0: Receiver related interrupt is disabled
 1: Receiver related interrupt is enabled
 This bit enables or disables the receiver interrupt. If this bit is equal to “1” and when the receiver overrun flag UOERR or receive data available flag URXIF is set, the USIM interrupt request flag USIMF will be set. If this bit is equal to “0”, the USIM interrupt request flag USIMF will not be influenced by the condition of the UOERR or URXIF flags.
- Bit 1 **UTIE**: Transmitter Idle interrupt enable control
 0: Transmitter idle interrupt is disabled
 1: Transmitter idle interrupt is enabled
 This bit enables or disables the transmitter idle interrupt. If this bit is equal to “1” and when the transmitter idle flag UTIDLE is set, due to a transmitter idle condition, the USIM interrupt request flag USIMF will be set. If this bit is equal to “0”, the USIM interrupt request flag USIMF will not be influenced by the condition of the UTIDLE flag.
- Bit 0 **UTEIE**: Transmitter Empty interrupt enable control
 0: Transmitter empty interrupt is disabled
 1: Transmitter empty interrupt is enabled
 This bit enables or disables the transmitter empty interrupt. If this bit is equal to “1” and when the transmitter empty flag UTXIF is set, due to a transmitter empty condition, the USIM interrupt request flag USIMF will be set. If this bit is equal to “0”, the USIM interrupt request flag USIMF will not be influenced by the condition of the UTXIF flag.

• UUCR3 Register

The UUCR3 register is used to enable the UART Single Wire Mode communication. As the name suggests in the single wire mode the UART communication can be implemented in one single line, URX/UTX, together with the control of the URXEN and UTXEN bits in the UUCR2 register.

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	—	USWM
R/W	—	—	—	—	—	—	—	R/W
POR	—	—	—	—	—	—	—	0

Bit 7~1 Unimplemented, read as “0”

- Bit 0 **USWM**: Single Wire Mode enable control
 0: Disable, the URX/UTX pin is used as UART receiver function only
 1: Enable, the URX/UTX pin can be used as UART receiver or transmitter function controlled by the URXEN and UTXEN bits
 Note that when the Single Wire Mode is enabled, if both the URXEN and UTXEN bits are high, the URX/UTX pin will just be used as UART receiver input.

• UTXR_RXR Register

The UTXR_RXR register is the data register which is used to store the data to be transmitted on the UTX pin or being received from the URX/UTX pin.

Bit	7	6	5	4	3	2	1	0
Name	UTXRX7	UTXRX6	UTXRX5	UTXRX4	UTXRX3	UTXRX2	UTXRX1	UTXRX0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	x	x	x	x	x	x	x	x

"x": unknown

Bit 7~0 **UTXRX7~UTXRX0**: UART Transmit/Receive Data bit 7 ~ bit 0

• UBRG Register

Bit	7	6	5	4	3	2	1	0
Name	UBRG7	UBRG6	UBRG5	UBRG4	UBRG3	UBRG2	UBRG1	UBRG0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	x	x	x	x	x	x	x	x

"x": unknown

Bit 7~0 **UBRG7~UBRG0**: Baud Rate values

By programming the UBRGH bit in UUCR2 register which allows selection of the related formula described above and programming the required value in the UBRG register, the required baud rate can be setup.

Note: Baud rate= $f_H/[64 \times (N+1)]$ if UBRGH=0.

Baud rate= $f_H/[16 \times (N+1)]$ if UBRGH=1.

Baud Rate Generator

To setup the speed of the serial data communication, the UART function contains its own dedicated baud rate generator. The baud rate is controlled by its own internal free running 8-bit timer, the period of which is determined by two factors. The first of these is the value placed in the baud rate register UBRG and the second is the value of the UBRGH bit in the control register UUCR2. The UBRGH bit decides if the baud rate generator is to be used in a high speed mode or low speed mode, which in turn determines the formula that is used to calculate the baud rate. The value N in the UBRG register which is used in the following baud rate calculation formula determines the division factor. Note that N is the decimal value placed in the UBRG register and has a range of between 0 and 255.

UUCR2 UBRGH Bit	0	1
Baud Rate (BR)	$f_H/[64 (N+1)]$	$f_H/[16 (N+1)]$

By programming the UBRGH bit which allows selection of the related formula and programming the required value in the UBRG register, the required baud rate can be setup. Note that because the actual baud rate is determined using a discrete value, N, placed in the UBRG register, there will be an error associated between the actual and requested value. The following example shows how the UBRG register value N and the error value can be calculated.

Calculating the Baud Rate and Error Values

For a clock frequency of 4MHz, and with UBRGH cleared to zero determine the UBRG register value N, the actual baud rate and the error value for a desired baud rate of 4800.

From the above table the desired baud rate $BR = f_H/[64 (N+1)]$

Re-arranging this equation gives $N = [f_H/(BR \times 64)] - 1$

Giving a value for $N = [4000000/(4800 \times 64)] - 1 = 12.0208$

To obtain the closest value, a decimal value of 12 should be placed into the UBRG register. This gives an actual or calculated baud rate value of $BR = 4000000 / [64 \times (12 + 1)] = 4808$

Therefore the error is equal to $(4808 - 4800) / 4800 = 0.16\%$

UART Setup and Control

For data transfer, the UART function utilizes a non-return-to-zero, more commonly known as NRZ, format. This is composed of one start bit, eight or nine data bits, and one or two stop bits. Parity is supported by the UART hardware, and can be setup to be even, odd or no parity. For the most common data format, 8 data bits along with no parity and one stop bit, denoted as 8, N, 1, is used as the default setting, which is the setting at power-on. The number of data bits and stop bits, along with the parity, are setup by programming the corresponding UBNO, UPRT, UPREN, and USTOPS bits in the UUCR1 register. The baud rate used to transmit and receive data is setup using the internal 8-bit baud rate generator, while the data is transmitted and received LSB first. Although the UART transmitter and receiver are functionally independent, they both use the same data format and baud rate. In all cases stop bits will be used for data transmission.

Enabling/Disabling the UART Interface

The basic on/off function of the internal UART function is controlled using the UREN bit in the UUCR1 register. When the UART mode is selected by setting the UMD bit in the SIMC0 register to "1", if the UREN, UTXEN and URXEN bits are set, then these two UART pins will act as normal UTX output pin and URX/UTX input pin respectively. If no data is being transmitted on the UTX pin, then it will default to a logic high value.

Clearing the UREN bit will disable the UTX and URX/UTX pin and allow these pins to be used as normal I/O or other pin-shared functional pins by configuring the corresponding pin-shared control bits. When the UART function is disabled the buffer will be reset to an empty condition, at the same time discarding any remaining residual data. Disabling the UART will also reset the error and status flags with bits UTXEN, URXEN, UTXBRK, URXIF, UOERR, UFERR, UPERR and UNF being cleared while bits UTIDLE, UTXIF and URIDLE will be set. The remaining control bits in the UUCR1, UUCR2, UUCR3 and UBRG registers will remain unaffected. If the UREN bit in the UUCR1 register is cleared while the UART is active, then all pending transmissions and receptions will be immediately suspended and the UART will be reset to a condition as defined above. If the UART is then subsequently re-enabled, it will restart again in the same configuration.

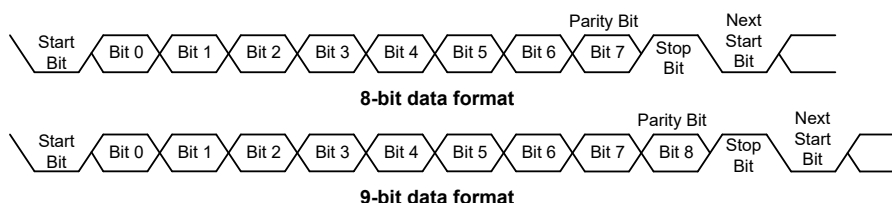
Data, Parity and Stop Bit Selection

The format of the data to be transferred is composed of various factors such as data bit length, parity on/off, parity type, address bits and the number of stop bits. These factors are determined by the setup of various bits within the UUCR1 register. The UBNO bit controls the number of data bits which can be set to either 8 or 9, the UPRT bit controls the choice of odd or even parity, the UPREN bit controls the parity on/off function and the USTOPS bit decides whether one or two stop bits are to be used. The following table shows various formats for data transmission. The address bit, which is the MSB of the data byte, identifies the frame as an address character or data if the address detect function is enabled. The number of stop bits, which can be either one or two, is independent of the data length and is only used for the transmitter. There is only one stop bit for the receiver.

Start Bit	Data Bits	Address Bit	Parity Bit	Stop Bit
Example of 8-bit Data Formats				
1	8	0	0	1
1	7	0	1	1
1	7	1	0	1
Example of 9-bit Data Formats				
1	9	0	0	1
1	8	0	1	1
1	8	1	0	1

Transmitter Receiver Data Format

The following diagram shows the transmit and receive waveforms for both 8-bit and 9-bit data formats.



UART Transmitter

Data word lengths of either 8 or 9 bits can be selected by programming the UBNO bit in the UUCR1 register. When UBNO bit is set, the word length will be set to 9 bits. In this case the 9th bit, which is the MSB, needs to be stored in the UTX8 bit in the UUCR1 register. At the transmitter core lies the Transmitter Shift Register, more commonly known as the TSR, whose data is obtained from the transmit data register, which is known as the UTXR_RXR register. The data to be transmitted is loaded into this UTXR_RXR register by the application program. The TSR register is not written to with new data until the stop bit from the previous transmission has been sent out. As soon as this stop bit has been transmitted, the TSR can then be loaded with new data from the UTXR_RXR register, if it is available. It should be noted that the TSR register, unlike many other registers, is not directly mapped into the Data Memory area and as such is not available to the application program for direct read/write operations. An actual transmission of data will normally be enabled when the UTXEN bit is set, but the data will not be transmitted until the UTXR_RXR register has been loaded with data and the baud rate generator has defined a shift clock source. However, the transmission can also be initiated by first loading data into the UTXR_RXR register, after which the UTXEN bit can be set. When a transmission of data begins, the TSR is normally empty, in which case a transfer to the UTXR_RXR register will result in an immediate transfer to the TSR. If during a transmission the UTXEN bit is cleared, the transmission will immediately cease and the transmitter will be reset. The UTX output pin can then be configured as the I/O or other pin-shared function by configuring the corresponding pin-shared control bits.

Transmitting Data

When the UART is transmitting data, the data is shifted on the UTX pin from the shift register, with the least significant bit first. In the transmit mode, the UTXR_RXR register forms a buffer between the internal bus and the transmitter shift register. It should be noted that if 9-bit data format has been selected, then the MSB will be taken from the UTX8 bit in the UUCR1 register. The steps to initiate a data transfer can be summarized as follows:

- Make the correct selection of the UBNO, UPRT, UPREN and USTOPS bits to define the required word length, parity type and number of stop bits.

- Setup the UBRG register to select the desired baud rate.
- Set the UTXEN bit ensure that the UTX pin is used as a UART transmitter pin.
- Access the UUSR register and write the data that is to be transmitted into the UTXR_RXR register. Note that this step will clear the UTXIF bit.

This sequence of events can now be repeated to send additional data.

It should be noted that when UTXIF=0, data will be inhibited from being written to the UTXR_RXR register. Clearing the UTXIF flag is always achieved using the following software sequence:

1. A UUSR register access
2. A UTXR_RXR register write execution

The read-only UTXIF flag is set by the UART hardware and if set indicates that the UTXR_RXR register is empty and that other data can now be written into the UTXR_RXR register without overwriting the previous data. If the UTEIE bit is set then the UTXIF flag will generate an interrupt.

During a data transmission, a write instruction to the UTXR_RXR register will place the data into the UTXR_RXR register, which will be copied to the shift register at the end of the present transmission. When there is no data transmission in progress, a write instruction to the UTXR_RXR register will place the data directly into the shift register, resulting in the commencement of data transmission, and the UTXIF bit being immediately set. When a frame transmission is complete, which happens after stop bits are sent or after the break frame, the UTIDLE bit will be set. To clear the UTIDLE bit the following software sequence is used:

1. A UUSR register access
2. A UTXR_RXR register write execution

Note that both the UTXIF and UTIDLE bits are cleared by the same software sequence.

Transmitting Break

If the UTXBRK bit is set high and the state keeps for a time greater than $[(BRG+1) \times t_H]$ while UTIDLE=1, then the break characters will be sent on the next transmission. Break character transmission consists of a start bit, followed by $13 \times N$ '0' bits and stop bits, where $N=1, 2$, etc. If a break character is to be transmitted then the UTXBRK bit must be first set by the application program, and then cleared to generate the stop bits. Transmitting a break character will not generate a transmit interrupt. Note that a break condition length is at least 13 bits long. If the UTXBRK bit is continually kept at a logic high level then the transmitter circuitry will transmit continuous break characters. After the application program has cleared the UTXBRK bit, the transmitter will finish transmitting the last break character and subsequently send out one or two stop bits. The automatic logic highs at the end of the last break character will ensure that the start bit of the next frame is recognized.

UART Receiver

The UART is capable of receiving word lengths of either 8 or 9 bits. If the UBNO bit is set, the word length will be set to 9 bits with the MSB being stored in the URX8 bit of the UUCR1 register. At the receiver core lies the Receive Serial Shift Register, commonly known as the RSR. The data which is received on the URX/UTX pin input is sent to the data recovery block. The data recovery block operating speed is 16 times that of the baud rate, while the main receive serial shifter operates at the baud rate. After the URX/UTX pin is sampled for the stop bit, the received data in RSR is transferred to the receive data register, if the register is empty. The data which is received on the external URX/UTX pin input is sampled three times by a majority detect circuit to determine the logic level that has been placed onto the URX/UTX pin. It should be noted that the RSR register,

unlike many other registers, is not directly mapped into the Data Memory area and as such is not available to the application program for direct read/write operations.

Receiving Data

When the UART receiver is receiving data, the data is serially shifted in on the external URX/UTX pin input, LSB first. In the read mode, the UTXR_RXR register forms a buffer between the internal bus and the receiver shift register. The UTXR_RXR register is a two-byte deep FIFO data buffer, where two bytes can be held in the FIFO while a third byte can continue to be received. Note that the application program must ensure that the data is read from UTXR_RXR before the third byte has been completely shifted in, otherwise this third byte will be discarded and an overrun error UOERR will be subsequently indicated. The steps to initiate a data transfer can be summarized as follows:

- Make the correct selection of UBNO, UPRT and UPREN bits to define the word length, parity type.
- Setup the UBRG register to select the desired baud rate.
- Set the URXEN bit to ensure that the URX/UTX pin is used as a UART receiver pin.

At this point the receiver will be enabled which will begin to look for a start bit.

When a character is received the following sequence of events will occur:

- The URXIF bit in the UUSR register will be set when the UTXR_RXR register has data available. There will be at most one more character available before an overrun error occurs.
- When the contents of the shift register have been transferred to the UTXR_RXR register, then if the URIE bit is set, an interrupt will be generated.
- If during reception, a frame error, noise error, parity error, or an overrun error has been detected, then the error flags can be set.

The URXIF bit can be cleared using the following software sequence:

1. A UUSR register access
2. A UTXR_RXR register read execution

Receiving Break

Any break character received by the UART will be managed as a framing error. The receiver will count and expect a certain number of bit times as specified by the values programmed into the UBNO bit plus one stop bit. If the break is much longer than 13 bit times, the reception will be considered as complete after the number of bit times specified by UBNO plus one stop bit. The URXIF bit is set, UFERR is set, zeros are loaded into the receive data register, interrupts are generated if appropriate and the URIDLE bit is set. A break is regarded as a character that contains only zeros with the UFERR flag set. If a long break signal has been detected, the receiver will regard it as a data frame including a start bit, data bits and the invalid stop bit and the UFERR flag will be set. The receiver must wait for a valid stop bit before looking for the next start bit. The receiver will not make the assumption that the break condition on the line is the next start bit. The break character will be loaded into the buffer and no further data will be received until stop bits are received. It should be noted that the URIDLE read only flag will go high when the stop bits have not yet been received. The reception of a break character on the UART registers will result in the following:

- The framing error flag, UFERR, will be set.
- The receive data register, UTXR_RXR, will be cleared.
- The UOERR, UNF, UPERR, URIDLE or URXIF flags will possibly be set.

Idle Status

When the receiver is reading data, which means it will be in between the detection of a start bit and the reading of a stop bit, the receiver status flag in the UUSR register, otherwise known as the URIDLE flag, will have a zero value. In between the reception of a stop bit and the detection of the next start bit, the URIDLE flag will have a high value, which indicates the receiver is in an idle condition.

Receiver Interrupt

The read only receive interrupt flag URXIF in the UUSR register is set by an edge generated by the receiver. An interrupt is generated if URIE=1, when a word is transferred from the Receive Shift Register, RSR, to the Receive Data Register, UTXR_RXR. An overrun error can also generate an interrupt if URIE=1.

Managing Receiver Errors

Several types of reception errors can occur within the UART module, the following section describes the various types and how they are managed by the UART.

Overrun Error – UOERR

The UTXR_RXR register is composed of a two-byte deep FIFO data buffer, where two bytes can be held in the FIFO register, while a third byte can continue to be received. Before this third byte has been entirely shifted in, the data should be read from the UTXR_RXR register. If this is not done, the overrun error flag UOERR will be consequently indicated.

In the event of an overrun error occurring, the following will happen:

- The UOERR flag in the UUSR register will be set.
- The UTXR_RXR contents will not be lost.
- The shift register will be overwritten.
- An interrupt will be generated if the URIE bit is set.

The UOERR flag can be cleared by an access to the UUSR register followed by a read to the UTXR_RXR register.

Noise Error – UNF

Over-sampling is used for data recovery to identify valid incoming data and noise. If noise is detected within a frame the following will occur:

- The read only noise flag, UNF, in the UUSR register will be set on the rising edge of the URXIF bit.
- Data will be transferred from the Shift register to the UTXR_RXR register.
- No interrupt will be generated. However this bit rises at the same time as the URXIF bit which itself generates an interrupt.

Note that the UNF flag is reset by a UUSR register read operation followed by a UTXR_RXR register read operation.

Framing Error – UFERR

The read only framing error flag, UFERR, in the UUSR register, is set if a zero is detected instead of stop bits. If two stop bits are selected, both stop bits must be high; otherwise the UFERR flag will be set. The UFERR flag and the received data will be recorded in the UUSR and UTXR_RXR registers respectively, and the flag is cleared in any reset.

Parity Error – UPERR

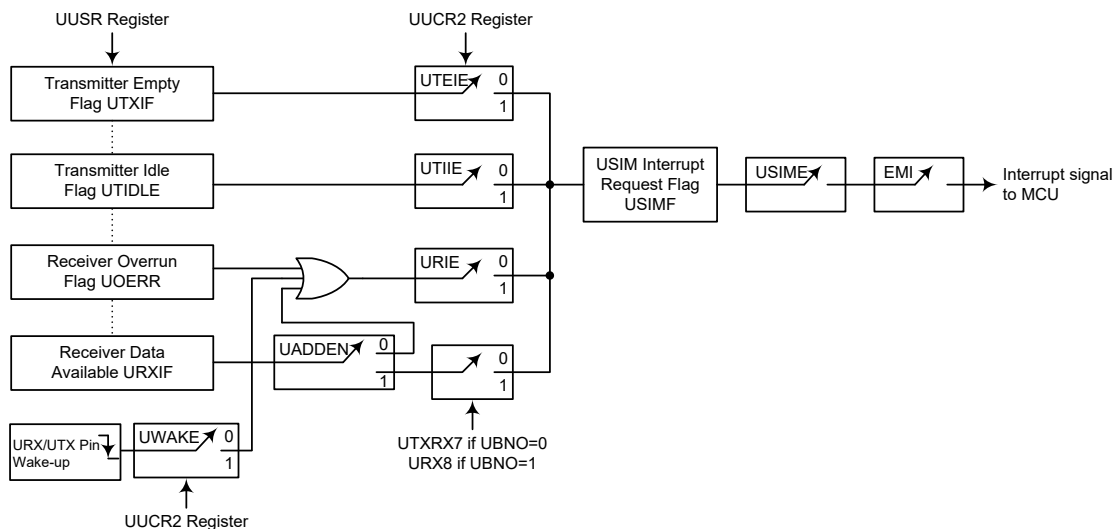
The read only parity error flag, UPERR, in the UUSR register, is set if the parity of the received word is incorrect. This error flag is only applicable if the parity is enabled, UPREN=1, and if the parity type, odd or even is selected. The read only UPERR flag and the received data will be recorded in the UUSR and UTXR_RXR registers respectively. It is cleared on any reset, it should be noted that the flags, UFERR and UPERR, in the UUSR register should first be read by the application program before reading the data word.

UART Interrupt Structure

Several individual UART conditions can trigger an USIM interrupt. When these conditions exist, a low pulse will be generated to get the attention of the microcontroller. These conditions are a transmitter data register empty, transmitter idle, receiver data available, receiver overrun, address detect and an URX/UTX pin wake-up. When any of these conditions are created, if the global interrupt enable bit and the USIM interrupt control bit are enabled and the stack is not full, the program will jump to its corresponding interrupt vector where it can be serviced before returning to the main program. Four of these conditions have the corresponding UUSR register flags which will generate an USIM interrupt if its associated interrupt enable control bit in the UUCR2 register is set. The two transmitter interrupt conditions have their own corresponding enable control bits, while the two receiver interrupt conditions have a shared enable control bit. These enable bits can be used to mask out individual USIM UART mode interrupt sources.

The address detect condition, which is also an USIM UART mode interrupt source, does not have an associated flag, but will generate an USIM interrupt when an address detect condition occurs if its function is enabled by setting the UADDEN bit in the UUCR2 register. An URX/UTX pin wake-up, which is also an USIM UART mode interrupt source, does not have an associated flag, but will generate an USIM interrupt if the UART clock (f_H) source is switched off and the UWAKE and URUE bits in the UUCR2 register are set when a falling edge on the URX/UTX pin occurs. Note that in the event of an URX/UTX wake-up interrupt occurring, there will be a certain period of delay, commonly known as the System Start-up Time, for the oscillator to restart and stabilize before the system resumes normal operation.

Note that the UUSR register flags are read only and cannot be cleared or set by the application program, neither will they be cleared when the program jumps to the corresponding interrupt servicing routine, as is the case for some of the other interrupts. The flags will be cleared automatically when certain actions are taken by the UART, the details of which are given in the UART register section. The overall UART interrupt can be disabled or enabled by the USIM interrupt enable control bit in the interrupt control register of the microcontroller to decide whether the interrupt requested by the UART module is masked out or allowed.



UART Interrupt Structure

Address Detect Mode

Setting the Address Detect Mode bit, UADDEN, in the UUCR2 register, enables this special mode. If this bit is enabled then an additional qualifier will be placed on the generation of a Receiver Data Available interrupt, which is requested by the URXIF flag. If the UADDEN bit is enabled, then when data is available, an interrupt will only be generated, if the highest received bit has a high value. Note that the USIME and EMI interrupt enable bits must also be enabled for correct interrupt generation. This highest address bit is the 9th bit if UBNO=1 or the 8th bit if UBNO=0. If this bit is high, then the received word will be defined as an address rather than data. A Data Available interrupt will be generated every time the last bit of the received word is set. If the UADDEN bit is not enabled, then a Receiver Data Available interrupt will be generated each time the URXIF flag is set, irrespective of the data last bit status. The address detect mode and parity enable are mutually exclusive functions. Therefore if the address detect mode is enabled, then to ensure correct operation, the parity function should be disabled by resetting the parity enable bit UPREN to zero.

UADDEN	9th bit if UBNO=1, 8th bit if UBNO=0	USIM Interrupt Generated
0	0	√
	1	√
1	0	×
	1	√

UADDEN Bit Function

UART Power Down and Wake-up

When the UART clock (f_{H1}) is off, the UART will cease to function, all clock sources to the module are shutdown. If the UART clock (f_{H1}) is off while a transmission is still in progress, then the transmission will be paused until the UART clock source derived from the microcontroller is activated. In a similar way, if the MCU enters the IDLE or SLEEP mode while receiving data, then the reception of data will likewise be paused. When the MCU enters the IDLE or SLEEP mode, note that the UUSR, UUCR1, UUCR2, UUCR3, UTXR_RXR, as well as the UBRG register will not be affected. It is recommended to make sure first that the UART data transmission or reception has been finished before the microcontroller enters the IDLE or SLEEP mode.

The UART function contains a receiver URX/UTX pin wake-up function, which is enabled or disabled by the UWAKE bit in the UUCR2 register. If this bit, along with the UART mode selection bit, UMD, the UART enable bit, UREN, the receiver enable bit, URXEN and the receiver interrupt bit, URIE, are all set when the UART clock (f_{H1}) is off, then a falling edge on the URX/UTX pin will trigger an URX/UTX pin wake-up UART interrupt. Note that as it takes certain system clock cycles after a wake-up, before normal microcontroller operation resumes, any data received during this time on the URX/UTX pin will be ignored.

For a UART wake-up interrupt to occur, in addition to the bits for the wake-up being set, the global interrupt enable bit, EMI, and the USIM interrupt enable bit, USIME, must be set. If the EMI and USIME bits are not set then only a wake-up event will occur and no interrupt will be generated. Note also that as it takes certain system clock cycles after a wake-up before normal microcontroller resumes, the USIM interrupt will not be generated until after this time has elapsed.

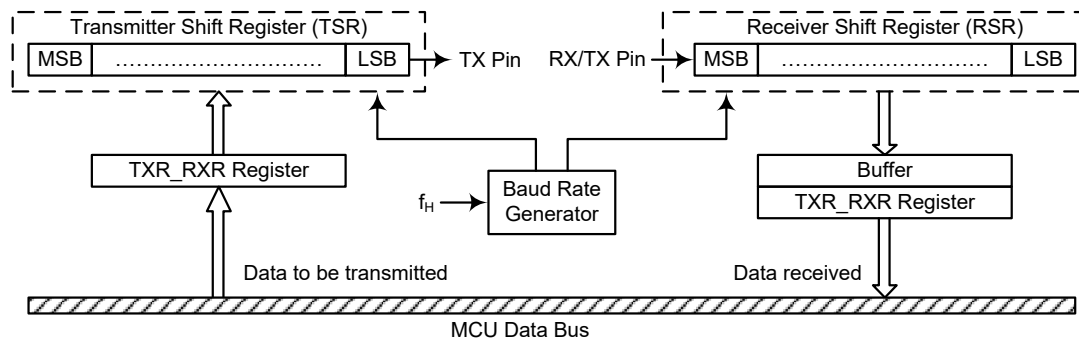
UART Interface

The device contains an independent UART function. It is important not to confuse this independent UART function with the additional one contained within the combined SIM function, which is described in another section of this datasheet.

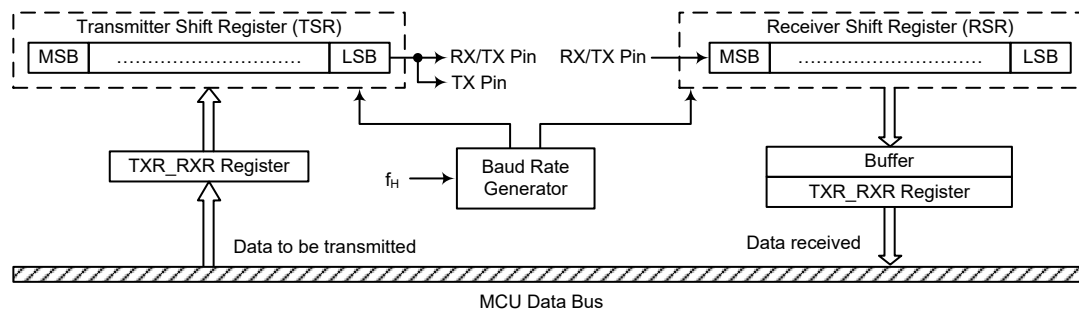
The device contains one integrated full-duplex or half-duplex asynchronous serial communications UART interface that enables communication with external devices that contain a serial interface. The UART function has many features and can transmit and receive data serially by transferring a frame of data with eight or nine data bits per transmission as well as being able to detect errors when the data is overwritten or incorrectly framed. The UART function possesses its own internal interrupt which can be used to indicate when a reception occurs or when a transmission terminates.

The integrated UART function contains the following features:

- Full-duplex or half-duplex (single wire mode) asynchronous communication
- 8 or 9 bits character length
- Even, odd or no parity options
- One or two stop bits
- Baud rate generator with 8-bit prescaler
- Parity, framing, noise and overrun error detection
- Support for interrupt on address detect (last character bit=1)
- Separately enabled transmitter and receiver
- 2-byte Deep FIFO Receive Data Buffer
- RX/TX pin wake-up function
- Transmit and receive interrupts
- Interrupts can be triggered by the following conditions:
 - ♦ Transmitter Empty
 - ♦ Transmitter Idle
 - ♦ Receiver Full
 - ♦ Receiver Overrun
 - ♦ Address Mode Detect



UART Data Transfer Block Diagram – SWM=0



UART Data Transfer Block Diagram – SWM=1

UART External Pins

To communicate with an external serial interface, the internal UART has two external pins known as TX and RX/TX, which are pin-shared with I/O or other pin functions. The TX and RX/TX pin function should first be selected by the pin-shared function selection register before the UART function is used. Along with the UARTEN bit, the TXEN and RXEN bits, if set, will setup these pins to transmitter output and receiver input conditions. At this time the internal pull-high resistor related to the transmitter output pin will be disabled, while the internal pull-high resistor related to the receiver input pin is controlled by the corresponding I/O pull-high function control bit. When the TX or RX/TX pin function is disabled by clearing the UARTEN, TXEN or RXEN bit, the TX or RX/TX pin will be set to a floating state. At this time whether the internal pull-high resistor is connected to the TX or RX/TX pin or not is determined by the corresponding I/O pull-high function control bit.

UART Single Wire Mode

The UART function also supports a Single Wire Mode communication which is selected using the SWM bit in the UCR3 register. When the SWM bit is set high, the UART function will be in the single wire mode. In the single wire mode, a single RX/TX pin can be used to transmit and receive data depending upon the corresponding control bits. When the RXEN bit is set high, the RX/TX pin is used as a receiver pin. When the RXEN bit is cleared to zero and the TXEN bit is set high, the RX/TX pin will act as a transmitter pin.

It is recommended not to set both the RXEN and TXEN bits high in the single wire mode. If both the RXEN and TXEN bits are set high, the RXEN bit will have the priority and the UART will act as a receiver.

It is important to note that the functional description in this UART chapter, which is described from the full-duplex communication standpoint, also applies to the half-duplex (single wire mode) communication except the pin usage. In the single wire mode, the TX pin mentioned in this chapter should be replaced by the RX/TX pin to understand the whole UART single wire mode function.

In the single wire mode, the data can also be transmitted on the TX pin in a transmission operation with proper software configurations. Therefore, the data will be output on the RX/TX and TX pins.

UART Data Transfer Scheme

The following block diagram shows the overall data transfer structure arrangement for the UART. The actual data to be transmitted from the MCU is first transferred to the TXR_RXR register by the application program. The data will then be transferred to the Transmit Shift Register from where it will be shifted out, LSB first, onto the TX pin at a rate controlled by the Baud Rate Generator. Only the TXR_RXR register is mapped onto the MCU Data Memory, the Transmit Shift Register is not mapped and is therefore inaccessible to the application program.

Data to be received by the UART is accepted on the external RX/TX pin, from where it is shifted in, LSB first, to the Receiver Shift Register at a rate controlled by the Baud Rate Generator. When the shift register is full, the data will then be transferred from the shift register to the internal TXR_RXR register, where it is buffered and can be manipulated by the application program. Only the TXR_RXR register is mapped onto the MCU Data Memory, the Receiver Shift Register is not mapped and is therefore inaccessible to the application program.

It should be noted that the actual register for data transmission and reception only exists as a single shared register in the Data Memory. This shared register known as the TXR_RXR register is used for both data transmission and data reception.

UART Status and Control Registers

There are six control registers associated with the UART function. The SWM bit in the UCR3 register is used to enable/disable the UART Single Wire Mode. The USR, UCR1 and UCR2 registers control the overall function of the UART, while the BRG register controls the Baud rate. The actual data to be transmitted and received on the serial interface is managed through the TXR_RXR data register.

Register Name	Bit							
	7	6	5	4	3	2	1	0
USR	PERR	NF	FERR	OERR	RIDLE	RXIF	TIDLE	TXIF
UCR1	UARTEN	BNO	PREN	PRT	STOPS	TXBRK	RX8	TX8
UCR2	TXEN	RXEN	BRGH	ADDEN	WAKE	RIE	TIIE	TEIE
UCR3	—	—	—	—	—	—	—	SWM
TXR_RXR	TXRX7	TXRX6	TXRX5	TXRX4	TXRX3	TXRX2	TXRX1	TXRX0
BRG	D7	D6	D5	D4	D3	D2	D1	D0

UART Register List

• **USR Register**

The USR register is the status register for the UART, which can be read by the program to determine the present status of the UART. All flags within the USR register are read only. Further explanation on each of the flags is given below:

Bit	7	6	5	4	3	2	1	0
Name	PERR	NF	FERR	OERR	RIDLE	RXIF	TIDLE	TXIF
R/W	R	R	R	R	R	R	R	R
POR	0	0	0	0	1	0	1	1

- Bit 7 PERR:** Parity error flag
0: No parity error is detected
1: Parity error is detected
The PERR flag is the parity error flag. When this read only flag is “0”, it indicates a parity error has not been detected. When the flag is “1”, it indicates that the parity of the received word is incorrect. This error flag is applicable only if Parity mode (odd or even) is selected. The flag can also be cleared to zero by a software sequence which involves a read to the status register USR followed by an access to the TXR_RXR data register.
- Bit 6 NF:** Noise flag
0: No noise is detected
1: Noise is detected
The NF flag is the noise flag. When this read only flag is “0”, it indicates no noise condition. When the flag is “1”, it indicates that the UART has detected noise on the receiver input. The NF flag is set during the same cycle as the RXIF flag but will not be set in the case of an overrun. The NF flag can be cleared to zero by a software sequence which will involve a read to the status register USR followed by an access to the TXR_RXR data register.
- Bit 5 FERR:** Framing error flag
0: No framing error is detected
1: Framing error is detected
The FERR flag is the framing error flag. When this read only flag is “0”, it indicates that there is no framing error. When the flag is “1”, it indicates that a framing error has been detected for the current character. The flag can also be cleared to zero by a software sequence which will involve a read to the status register USR followed by an access to the TXR_RXR data register.
- Bit 4 OERR:** Overrun error flag
0: No overrun error is detected
1: Overrun error is detected
The OERR flag is the overrun error flag which indicates when the receiver buffer has overflowed. When this read only flag is “0”, it indicates that there is no overrun error. When the flag is “1”, it indicates that an overrun error occurs which will inhibit further transfers to the TXR_RXR receive data register. The flag is cleared to zero by a software sequence, which is a read to the status register USR followed by an access to the TXR_RXR data register.
- Bit 3 RIDLE:** Receiver status
0: Data reception is in progress (Data being received)
1: No data reception is in progress (Receiver is idle)
The RIDLE flag is the receiver status flag. When this read only flag is “0”, it indicates that the receiver is between the initial detection of the start bit and the completion of the stop bit. When the flag is “1”, it indicates that the receiver is idle. Between the completion of the stop bit and the detection of the next start bit, the RIDLE bit is “1” indicating that the UART receiver is idle and the RX/TX pin stays in logic high condition.

- Bit 2 **RXIF:** Receive TXR_RXR data register status
 0: TXR_RXR data register is empty
 1: TXR_RXR data register has available data
 The RXIF flag is the receive data register status flag. When this read only flag is “0”, it indicates that the TXR_RXR read data register is empty. When the flag is “1”, it indicates that the TXR_RXR read data register contains new data. When the contents of the shift register are transferred to the TXR_RXR register, an interrupt is generated if RIE=1 in the UCR2 register. If one or more errors are detected in the received word, the appropriate receive-related flags NF, FERR, and/or PERR are set within the same clock cycle. The RXIF flag will eventually be cleared to zero when the USR register is read with RXIF set, followed by a read from the TXR_RXR register, and if the TXR_RXR register has no more new data available.
- Bit 1 **TIDLE:** Transmission idle
 0: Data transmission is in progress (Data being transmitted)
 1: No data transmission is in progress (Transmitter is idle)
 The TIDLE flag is known as the transmission complete flag. When this read only flag is “0”, it indicates that a transmission is in progress. This flag will be set high when the TXIF flag is “1” and when there is no transmit data or break character being transmitted. When TIDLE is equal to “1”, the TX pin becomes idle with the pin state in logic high condition. The TIDLE flag is cleared to zero by reading the USR register with TIDLE set and then writing to the TXR_RXR register. The flag is not generated when a data character or a break is queued and ready to be sent.
- Bit 0 **TXIF:** Transmit TXR_RXR data register status
 0: Character is not transferred to the transmit shift register
 1: Character has transferred to the transmit shift register (TXR_RXR data register is empty)
 The TXIF flag is the transmit data register empty flag. When this read only flag is “0”, it indicates that the character is not transferred to the transmitter shift register. When the flag is “1”, it indicates that the transmitter shift register has received a character from the TXR_RXR data register. The TXIF flag is cleared to zero by reading the UART status register (USR) with TXIF set and then writing to the TXR_RXR data register. Note that when the TXEN bit is set, the TXIF flag bit will also be set since the transmit data register is not yet full.

• UCR1 Register

The UCR1 register together with the UCR2 and UCR3 registers are the three UART control registers that are used to set the various options for the UART function, such as overall on/off control, parity control, data transfer bit length, single wire mode communication etc. Further explanation on each of the bits is given below:

Bit	7	6	5	4	3	2	1	0
Name	UARTEN	BNO	PREN	PRT	STOPS	TXBRK	RX8	TX8
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R	W
POR	0	0	0	0	0	0	x	0

“x”: unknown

- Bit 7 **UARTEN:** UART function enable control
 0: Disable UART. TX and RX/TX pins are in a floating state
 1: Enable UART. TX and RX/TX pins can function as UART pins
 The UARTEN bit is the UART enable bit. When this bit is equal to “0”, the UART will be disabled and the RX/TX pin as well as the TX pin will be in a floating state. When the bit is equal to “1”, the UART will be enabled and the TX and RX/TX pins will function as defined by the SWM mode selection bit together with the TXEN and RXEN enable control bits.
 When the UART is disabled, it will empty the buffer so any character remaining in the buffer will be discarded. In addition, the value of the baud rate counter will be

reset. If the UART is disabled, all error and status flags will be reset. Also the TXEN, RXEN, TXBRK, RXIF, OERR, FERR, PERR and NF bits will be cleared to zero, while the TIDLE, TXIF and RIDLE bits will be set high. Other control bits in UCR1, UCR2, UCR3 and BRG registers will remain unaffected. If the UART is active and the UARTEN bit is cleared to zero, all pending transmissions and receptions will be terminated and the module will be reset as defined above. When the UART is re-enabled, it will restart in the same configuration.

- Bit 6** **BNO:** Number of data transfer bits selection
 0: 8-bit data transfer
 1: 9-bit data transfer
- This bit is used to select the data length format, which can have a choice of either 8-bit or 9-bit format. When this bit is equal to “1”, a 9-bit data length format will be selected. If the bit is equal to “0”, then an 8-bit data length format will be selected. If 9-bit data length format is selected, then bits RX8 and TX8 will be used to store the 9th bit of the received and transmitted data respectively.
- Note that the 9th bit of data if BNO=1, or the 8th bit of data if BNO=0, which is used as the parity bit, does not transfer to RX8 or TXRX7 respectively when the parity function is enabled.
- Bit 5** **PREN:** Parity function enable control
 0: Parity function is disabled
 1: Parity function is enabled
- This is the parity enable bit. When this bit is equal to “1”, the parity function will be enabled. If the bit is equal to “0”, then the parity function will be disabled.
- Bit 4** **PRT:** Parity type selection bit
 0: Even parity for parity generator
 1: Odd parity for parity generator
- This bit is the parity type selection bit. When this bit is equal to “1”, odd parity type will be selected. If the bit is equal to “0”, then even parity type will be selected.
- Bit 3** **STOPS:** Number of Stop bits selection
 0: One stop bit format is used
 1: Two stop bits format is used
- This bit determines if one or two stop bits are to be used. When this bit is equal to “1”, two stop bits are used. If this bit is equal to “0”, then only one stop bit is used.
- Bit 2** **TXBRK:** Transmit break character
 0: No break character is transmitted
 1: Break characters transmit
- The TXBRK bit is the Transmit Break Character bit. When this bit is “0”, there are no break characters and the TX pin operates normally. When the bit is “1”, there are transmit break characters and the transmitter will send logic zeros. When this bit is equal to “1”, after the buffered data has been transmitted, the transmitter output is held low for a minimum of a 13-bit length and until the TXBRK bit is reset.
- Bit 1** **RX8:** Receive data bit 8 for 9-bit data transfer format (read only)
- This bit is only used if 9-bit data transfers are used, in which case this bit location will store the 9th bit of the received data known as RX8. The BNO bit is used to determine whether data transfers are in 8-bit or 9-bit format.
- Bit 0** **TX8:** Transmit data bit 8 for 9-bit data transfer format (write only)
- This bit is only used if 9-bit data transfers are used, in which case this bit location will store the 9th bit of the transmitted data known as TX8. The BNO bit is used to determine whether data transfers are in 8-bit or 9-bit format.

• UCR2 Register

The UCR2 register is the second of the two UART control registers and serves several purposes. One of its main functions is to control the basic enable/disable operation of the UART Transmitter and Receiver as well as enabling the various UART interrupt sources. The register also serves to control the baud rate speed, receiver wake-up enable and the address detect enable. Further explanation on each of the bits is given below:

Bit	7	6	5	4	3	2	1	0
Name	TXEN	RXEN	BRGH	ADDEN	WAKE	RIE	TIIE	TEIE
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7 **TXEN**: UART Transmitter enabled control

- 0: UART transmitter is disabled
- 1: UART transmitter is enabled

The bit named TXEN is the Transmitter Enable Bit. When this bit is equal to “0”, the transmitter will be disabled with any pending data transmissions being aborted. In addition the buffers will be reset. In this situation the TX pin will be in a floating state. If the TXEN bit is equal to “1” and the UARTEN bit is also equal to “1”, the transmitter will be enabled and the TX pin will be controlled by the UART. Clearing the TXEN bit during a transmission will cause the data transmission to be aborted and will reset the transmitter. If this situation occurs, the TX pin will be in a floating state.

Bit 6 **RXEN**: UART Receiver enabled control

- 0: UART receiver is disabled
- 1: UART receiver is enabled

The bit named RXEN is the Receiver Enable Bit. When this bit is equal to “0”, the receiver will be disabled with any pending data receptions being aborted. In addition the receive buffers will be reset. In this situation the RX/TX pin will be in a floating state. If the RXEN bit is equal to “1” and the UARTEN bit is also equal to “1”, the receiver will be enabled and the RX/TX pin will be controlled by the UART. Clearing the RXEN bit during a reception will cause the data reception to be aborted and will reset the receiver. If this situation occurs, the RX/TX pin will be in a floating state.

Bit 5 **BRGH**: Baud Rate speed selection

- 0: Low speed baud rate
- 1: High speed baud rate

The bit named BRGH selects the high or low speed mode of the Baud Rate Generator. This bit, together with the value placed in the baud rate register BRG, controls the Baud Rate of the UART. If this bit is equal to “1”, the high speed mode is selected. If the bit is equal to “0”, the low speed mode is selected.

Bit 4 **ADDEN**: Address detect function enable control

- 0: Address detect function is disabled
- 1: Address detect function is enabled

The bit named ADDEN is the address detect function enable control bit. When this bit is equal to “1”, the address detect function is enabled. When it occurs, if the 8th bit, which corresponds to TXR_RXR.7 if BNO=0 or the 9th bit, which corresponds to RX8 if BNO=1, has a value of “1”, then the received word will be identified as an address, rather than data. If the corresponding interrupt is enabled, an interrupt request will be generated each time the received word has the address bit set, which is the 8th or 9th bit depending on the value of BNO. If the address bit known as the 8th or 9th bit of the received word is “0” with the address detect function being enabled, an interrupt will not be generated and the received data will be discarded.

Bit 3 **WAKE**: RX/TX pin wake-up UART function enable control

- 0: RX/TX pin wake-up UART function is disabled
- 1: RX/TX pin wake-up UART function is enabled

This bit is used to control the wake-up UART function when a falling edge on the RX/TX pin occurs. Note that this bit is only available when the UART clock (f_H) is switched off. There will be no RX/TX pin wake-up UART function if the UART clock (f_H) exists. If the WAKE bit is set to 1 as the UART clock (f_H) is switched off, a UART wake-up request will be initiated when a falling edge on the RX/TX pin occurs. When this request happens and the corresponding interrupt is enabled, an RX/TX pin wake-up UART interrupt will be generated to inform the MCU to wake up the UART function by switching on the UART clock (f_H) via the application program. Otherwise, the UART function can not resume even if there is a falling edge on the RX/TX pin when the WAKE bit is cleared to 0.

Bit 2 **RIE**: Receiver interrupt enable control
 0: Receiver related interrupt is disabled
 1: Receiver related interrupt is enabled

This bit enables or disables the receiver interrupt. If this bit is equal to “1” and when the receiver overrun flag OERR or receive data available flag RXIF is set, the UART interrupt request flag will be set. If this bit is equal to “0”, the UART interrupt request flag will not be influenced by the condition of the OERR or RXIF flags.

Bit 1 **TIIE**: Transmitter Idle interrupt enable control
 0: Transmitter idle interrupt is disabled
 1: Transmitter idle interrupt is enabled

This bit enables or disables the transmitter idle interrupt. If this bit is equal to “1” and when the transmitter idle flag TIDLE is set, due to a transmitter idle condition, the UART interrupt request flag will be set. If this bit is equal to “0”, the UART interrupt request flag will not be influenced by the condition of the TIDLE flag.

Bit 0 **TEIE**: Transmitter Empty interrupt enable control
 0: Transmitter empty interrupt is disabled
 1: Transmitter empty interrupt is enabled

This bit enables or disables the transmitter empty interrupt. If this bit is equal to “1” and when the transmitter empty flag TXIF is set, due to a transmitter empty condition, the UART interrupt request flag will be set. If this bit is equal to “0”, the UART interrupt request flag will not be influenced by the condition of the TXIF flag.

• UCR3 Register

The UCR3 register is used to enable the UART Single Wire Mode communication. As the name suggests in the single wire mode the UART communication can be implemented in one single line, RX/TX, together with the control of the RXEN and TXEN bits in the UCR2 register.

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	—	SWM
R/W	—	—	—	—	—	—	—	R/W
POR	—	—	—	—	—	—	—	0

Bit 7~1 Unimplemented, read as “0”

Bit 0 **SWM**: Single Wire Mode enable control
 0: Disable, the RX/TX pin is used as UART receiver function only
 1: Enable, the RX/TX pin can be used as UART receiver or transmitter function controlled by the RXEN and TXEN bits

Note that when the Single Wire Mode is enabled, if both the RXEN and TXEN bits are high, the RX/TX pin will just be used as UART receiver input.

• TXR_RXR Register

The TXR_RXR register is the data register which is used to store the data to be transmitted on the TX pin or being received from the RX/TX pin.

Bit	7	6	5	4	3	2	1	0
Name	TXRX7	TXRX6	TXRX5	TXRX4	TXRX3	TXRX2	TXRX1	TXRX0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	x	x	x	x	x	x	x	x

"x": unknown

Bit 7~0 **TXRX7~TXRX0**: UART Transmit/Receive Data bit 7 ~ bit 0

• BRG Register

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	x	x	x	x	x	x	x	x

"x": unknown

Bit 7~0 **D7~D0**: Baud Rate values

By programming the BRGH bit in UCR2 Register which allows selection of the related formula described above and programming the required value in the BRG register, the required baud rate can be setup.

Note: Baud rate= $f_H/[64 \times (N+1)]$ if BRGH=0;

Baud rate= $f_H/[16 \times (N+1)]$ if BRGH=1.

Baud Rate Generator

To setup the speed of the serial data communication, the UART function contains its own dedicated baud rate generator. The baud rate is controlled by its own internal free running 8-bit timer, the period of which is determined by two factors. The first of these is the value placed in the baud rate register BRG and the second is the value of the BRGH bit with the control register UCR2. The BRGH bit decides if the baud rate generator is to be used in a high speed mode or low speed mode, which in turn determines the formula that is used to calculate the baud rate. The value N in the BRG register which is used in the following baud rate calculation formula determines the division factor. Note that N is the decimal value placed in the BRG register and has a range of between 0 and 255.

UCR2 BRGH Bit	0	1
Baud Rate (BR)	$f_H/[64 \times (N+1)]$	$f_H/[16 \times (N+1)]$

By programming the BRGH bit which allows selection of the related formula and programming the required value in the BRG register, the required baud rate can be setup. Note that because the actual baud rate is determined using a discrete value, N, placed in the BRG register, there will be an error associated between the actual and requested value. The following example shows how the BRG register value N and the error value can be calculated.

Calculating the Baud Rate and Error Values

For a clock frequency of 4MHz, and with BRGH cleared to zero determine the BRG register value N, the actual baud rate and the error value for a desired baud rate of 4800.

From the above table the desired baud rate $BR = f_H/[64 \times (N+1)]$

Re-arranging this equation gives $N = [f_H/(BR \times 64)] - 1$

Giving a value for $N = [4000000/(4800 \times 64)] - 1 = 12.0208$

To obtain the closest value, a decimal value of 12 should be placed into the BRG register. This gives an actual or calculated baud rate value of $BR = 4000000 / [64 \times (12 + 1)] = 4808$

Therefore the error is equal to $(4808 - 4800) / 4800 = 0.16\%$.

UART Setup and Control

For data transfer, the UART function utilizes a non-return-to-zero, more commonly known as NRZ, format. This is composed of one start bit, eight or nine data bits, and one or two stop bits. Parity is supported by the UART hardware, and can be setup to be even, odd or no parity. For the most common data format, 8 data bits along with no parity and one stop bit, denoted as 8, N, 1, is used as the default setting, which is the setting at power-on. The number of data bits and stop bits, along with the parity, are setup by programming the corresponding BNO, PRT, PREN, and STOPS bits in the UCR1 register. The baud rate used to transmit and receive data is setup using the internal 8-bit baud rate generator, while the data is transmitted and received LSB first. Although the UART transmitter and receiver are functionally independent, they both use the same data format and baud rate. In all cases stop bits will be used for data transmission.

Enabling/Disabling the UART Interface

The basic on/off function of the internal UART function is controlled using the UARTEN bit in the UCR1 register. If the UARTEN, TXEN and RXEN bits are set, then these two UART pins will act as normal TX output pin and RX/TX input pin respectively. If no data is being transmitted on the TX pin, then it will default to a logic high value.

Clearing the UARTEN bit will disable the TX and RX/TX pins and allow these two pins to be used as normal I/O or other pin-shared functional pins. When the UART function is disabled the buffer will be reset to an empty condition, at the same time discarding any remaining residual data. Disabling the UART will also reset the error and status flags with bits TXEN, RXEN, TXBRK, RXIF, OERR, FERR, PERR and NF being cleared while bits TIDLE, TXIF and RIDLE will be set. The remaining control bits in the UCR1, UCR2, UCR3 and BRG registers will remain unaffected. If the UARTEN bit in the UCR1 register is cleared while the UART is active, then all pending transmissions and receptions will be immediately suspended and the UART will be reset to a condition as defined above. If the UART is then subsequently re-enabled, it will restart again in the same configuration.

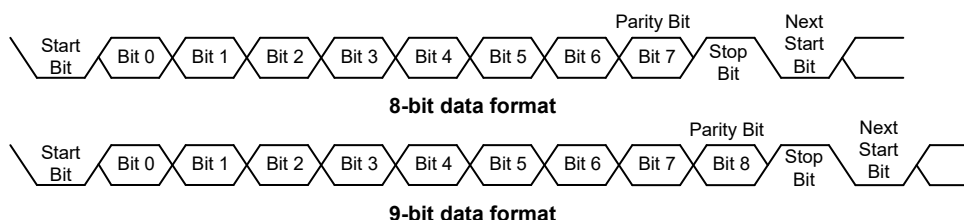
Data, Parity and Stop Bit Selection

The format of the data to be transferred is composed of various factors such as data bit length, parity on/off, parity type, address bits and the number of stop bits. These factors are determined by the setup of various bits within the UCR1 register. The BNO bit controls the number of data bits which can be set to either 8 or 9, the PRT bit controls the choice of odd or even parity, the PREN bit controls the parity on/off function and the STOPS bit decides whether one or two stop bits are to be used. The following table shows various formats for data transmission. The address bit, which is the MSB of the data byte, identifies the frame as an address character or data if the address detect function is enabled. The number of stop bits, which can be either one or two, is independent of the data length and is only used for the transmitter. There is only one stop bit for the receiver.

Start Bit	Data Bits	Address Bit	Parity Bit	Stop Bit
Example of 8-bit Data Formats				
1	8	0	0	1
1	7	0	1	1
1	7	1	0	1
Example of 9-bit Data Formats				
1	9	0	0	1
1	8	0	1	1
1	8	1	0	1

Transmitter Receiver Data Format

The following diagram shows the transmit and receive waveforms for both 8-bit and 9-bit data formats.



UART Transmitter

Data word lengths of either 8 or 9 bits can be selected by programming the BNO bit in the UCR1 register. When BNO bit is set, the word length will be set to 9 bits. In this case the 9th bit, which is the MSB, needs to be stored in the TX8 bit in the UCR1 register. At the transmitter core lies the Transmitter Shift Register, more commonly known as the TSR, whose data is obtained from the transmit data register, which is known as the TXR_RXR register. The data to be transmitted is loaded into this TXR_RXR register by the application program. The TSR register is not written to with new data until the stop bit from the previous transmission has been sent out. As soon as this stop bit has been transmitted, the TSR can then be loaded with new data from the TXR_RXR register, if it is available. It should be noted that the TSR register, unlike many other registers, is not directly mapped into the Data Memory area and as such is not available to the application program for direct read/write operations. An actual transmission of data will normally be enabled when the TXEN bit is set, but the data will not be transmitted until the TXR_RXR register has been loaded with data and the baud rate generator has defined a shift clock source. However, the transmission can also be initiated by first loading data into the TXR_RXR register, after which the TXEN bit can be set. When a transmission of data begins, the TSR is normally empty, in which case a transfer to the TXR_RXR register will result in an immediate transfer to the TSR. If during a transmission the TXEN bit is cleared, the transmission will immediately cease and the transmitter will be reset. The TX output pin can then be configured as the I/O or other pin-shared function by configuring the corresponding pin-shared control bits.

Transmitting Data

When the UART is transmitting data, the data is shifted on the TX pin from the shift register, with the least significant bit first. In the transmit mode, the TXR_RXR register forms a buffer between the internal bus and the transmitter shift register. It should be noted that if 9-bit data format has been selected, then the MSB will be taken from the TX8 bit in the UCR1 register. The steps to initiate a data transfer can be summarized as follows:

- Make the correct selection of the BNO, PRT, PREN and STOPS bits to define the required word length, parity type and number of stop bits.
- Setup the BRG register to select the desired baud rate.
- Set the TXEN bit to ensure that the TX pin is used as a UART transmitter pin.
- Access the USR register and write the data that is to be transmitted into the TXR_RXR register. Note that this step will clear the TXIF bit.

This sequence of events can now be repeated to send additional data.

It should be noted that when TXIF=0, data will be inhibited from being written to the TXR_RXR register. Clearing the TXIF flag is always achieved using the following software sequence:

1. A USR register access
2. A TXR_RXR register write execution

The read-only TXIF flag is set by the UART hardware and if set indicates that the TXR_RXR register is empty and that other data can now be written into the TXR_RXR register without overwriting the previous data. If the TEIE bit is set then the TXIF flag will generate an interrupt.

During a data transmission, a write instruction to the TXR_RXR register will place the data into the TXR_RXR register, which will be copied to the shift register at the end of the present transmission. When there is no data transmission in progress, a write instruction to the TXR_RXR register will place the data directly into the shift register, resulting in the commencement of data transmission, and the TXIF bit being immediately set. When a frame transmission is complete, which happens after stop bits are sent or after the break frame, the TIDLE bit will be set. To clear the TIDLE bit the following software sequence is used:

1. A USR register access
2. A TXR_RXR register write execution

Note that both the TXIF and TIDLE bits are cleared by the same software sequence.

Transmit Break

If the TXBRK bit is set high and the state keeps for a time greater than $[(BRG+1) \times t_H]$ while TIDLE=1, then the break characters will be sent on the next transmission. Break character transmission consists of a start bit, followed by $13 \times N$ '0' bits and stop bits, where $N=1, 2$, etc. If a break character is to be transmitted then the TXBRK bit must be first set by the application program, and then cleared to generate the stop bits. Transmitting a break character will not generate a transmit interrupt. Note that a break condition length is at least 13 bits long. If the TXBRK bit is continually kept at a logic high level then the transmitter circuitry will transmit continuous break characters. After the application program has cleared the TXBRK bit, the transmitter will finish transmitting the last break character and subsequently send out one or two stop bits. The automatic logic highs at the end of the last break character will ensure that the start bit of the next frame is recognized.

UART Receiver

The UART is capable of receiving word lengths of either 8 or 9 bits. If the BNO bit is set, the word length will be set to 9 bits with the MSB being stored in the RX8 bit of the UCR1 register. At the receiver core lies the Receive Serial Shift Register, commonly known as the RSR. The data which is received on the RX/TX external input pin is sent to the data recovery block. The data recovery block operating speed is 16 times that of the baud rate, while the main receive serial shifter operates at the baud rate. After the RX/TX pin is sampled for the stop bit, the received data in RSR is transferred to the receive data register, if the register is empty. The data which is received on the external RX/TX input pin is sampled three times by a majority detect circuit to determine the logic level that has been

placed onto the RX/TX pin. It should be noted that the RSR register, unlike many other registers, is not directly mapped into the Data Memory area and as such is not available to the application program for direct read/write operations.

Receiving Data

When the UART receiver is receiving data, the data is serially shifted in on the external RX/TX pin input, LSB first. In the read mode, the TXR_RXR register forms a buffer between the internal bus and the receiver shift register. The TXR_RXR register is a two-byte deep FIFO data buffer, where two bytes can be held in the FIFO while a third byte can continue to be received. Note that the application program must ensure that the data is read from TXR_RXR before the third byte has been completely shifted in, otherwise this third byte will be discarded and an overrun error OERR will be subsequently indicated. The steps to initiate a data transfer can be summarized as follows:

- Make the correct selection of BNO, PRT and PREN bits to define the word length, parity type.
- Setup the BRG register to select the desired baud rate.
- Set the RXEN bit to ensure that the RX/TX pin is used as a UART receiver pin.

At this point the receiver will be enabled which will begin to look for a start bit.

When a character is received the following sequence of events will occur:

- The RXIF bit in the USR register will be set when the TXR_RXR register has data available. There will be at most one more character available before an overrun error occurs.
- When the contents of the shift register have been transferred to the TXR_RXR register, then if the RIE bit is set, an interrupt will be generated.
- If during reception, a frame error, noise error, parity error, or an overrun error has been detected, then the error flags can be set.

The RXIF bit can be cleared using the following software sequence:

1. A USR register access
2. A TXR_RXR register read execution

Receiving Break

Any break character received by the UART will be managed as a framing error. The receiver will count and expect a certain number of bit times as specified by the values programmed into the BNO bit plus one stop bit. If the break is much longer than 13 bit times, the reception will be considered as complete after the number of bit times specified by BNO plus one stop bit. The RXIF bit is set, FERR is set, zeros are loaded into the receive data register, interrupts are generated if appropriate and the RIDLE bit is set. A break is regarded as a character that contains only zeros with the FERR flag set. If a long break signal has been detected, the receiver will regard it as a data frame including a start bit, data bits and the invalid stop bit and the FERR flag will be set. The receiver must wait for a valid stop bit before looking for the next start bit. The receiver will not make the assumption that the break condition on the line is the next start bit. The break character will be loaded into the buffer and no further data will be received until stop bits are received. It should be noted that the RIDLE read only flag will go high when the stop bits have not yet been received. The reception of a break character on the UART registers will result in the following:

- The framing error flag, FERR, will be set.
- The receive data register, TXR_RXR, will be cleared.
- The OERR, NF, PERR, RIDLE or RXIF flags will possibly be set.

Idle Status

When the receiver is reading data, which means it will be in between the detection of a start bit and the reading of a stop bit, the receiver status flag in the USR register, otherwise known as the RIDLE flag, will have a zero value. In between the reception of a stop bit and the detection of the next start bit, the RIDLE flag will have a high value, which indicates the receiver is in an idle condition.

Receiver Interrupt

The read only receive interrupt flag RXIF in the USR register is set by an edge generated by the receiver. An interrupt is generated if RIE=1, when a word is transferred from the Receive Shift Register, RSR, to the Receive Data Register, TXR_RXR. An overrun error can also generate an interrupt if RIE=1.

Managing Receiver Errors

Several types of reception errors can occur within the UART module, the following section describes the various types and how they are managed by the UART.

Overrun Error – OERR

The TXR_RXR register is composed of a two-byte deep FIFO data buffer, where two bytes can be held in the FIFO register, while a third byte can continue to be received. Before this third byte has been entirely shifted in, the data should be read from the TXR_RXR register. If this is not done, the overrun error flag OERR will be consequently indicated.

In the event of an overrun error occurring, the following will happen:

- The OERR flag in the USR register will be set.
- The TXR_RXR contents will not be lost.
- The shift register will be overwritten.
- An interrupt will be generated if the RIE bit is set.

The OERR flag can be cleared by an access to the USR register followed by a read to the TXR_RXR register.

Noise Error – NF

Over-sampling is used for data recovery to identify valid incoming data and noise. If noise is detected within a frame the following will occur:

- The read only noise flag, NF, in the USR register will be set on the rising edge of the RXIF bit.
- Data will be transferred from the Shift register to the TXR_RXR register.
- No interrupt will be generated. However this bit rises at the same time as the RXIF bit which itself generates an interrupt.

Note that the NF flag is reset by a USR register read operation followed by a TXR_RXR register read operation.

Framing Error – FERR

The read only framing error flag, FERR, in the USR register, is set if a zero is detected instead of stop bits. If two stop bits are selected, both stop bits must be high; otherwise the FERR flag will be set. The FERR flag and the received data will be recorded in the USR and TXR_RXR registers respectively, and the flag is cleared in any reset.

Parity Error – PERR

The read only parity error flag, PERR, in the USR register, is set if the parity of the received word is incorrect. This error flag is only applicable if the parity is enabled, PREN=1, and if the parity type,

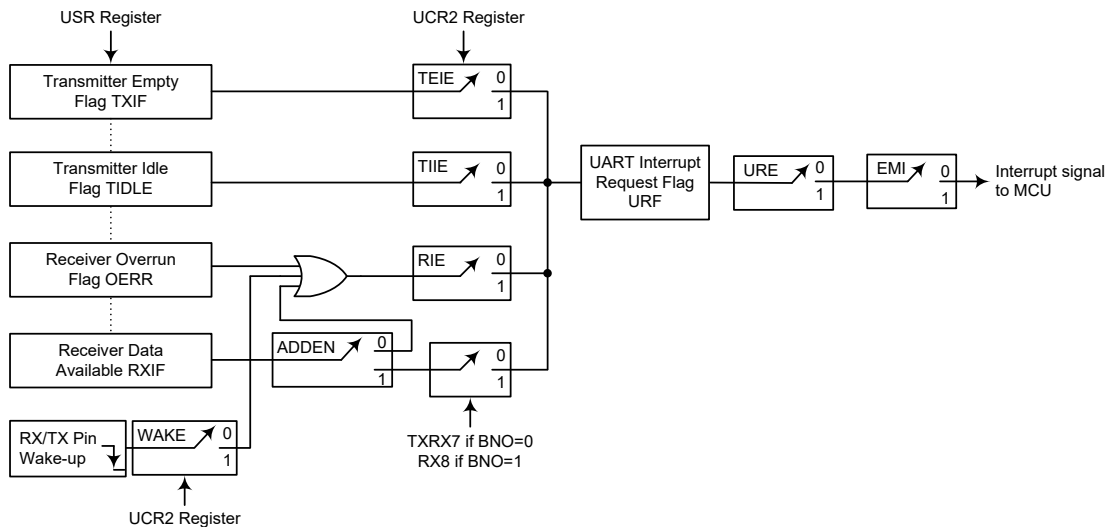
odd or even is selected. The read only PERR flag and the received data will be recorded in the USR and TXR_RXR registers respectively. It is cleared on any reset, it should be noted that the flags, FERR and PERR, in the USR register should first be read by the application program before reading the data word.

UART Interrupt Structure

Several individual UART conditions can generate a UART interrupt. When these conditions exist, a low pulse will be generated to get the attention of the microcontroller. These conditions are a transmitter data register empty, transmitter idle, receiver data available, receiver overrun, address detect and an RX/TX pin wake-up. When any of these conditions are created, if the global interrupt enable bit and its corresponding interrupt control bit are enabled and the stack is not full, the program will jump to its corresponding interrupt vector where it can be serviced before returning to the main program. Four of these conditions have the corresponding USR register flags which will generate a UART interrupt if its associated interrupt enable control bit in the UCR2 register is set. The two transmitter interrupt conditions have their own corresponding enable control bits, while the two receiver interrupt conditions have a shared enable control bit. These enable bits can be used to mask out individual UART interrupt sources.

The address detect condition, which is also a UART interrupt source, does not have an associated flag, but will generate a UART interrupt when an address detect condition occurs if its function is enabled by setting the ADDEN bit in the UCR2 register. An RX/TX pin wake-up, which is also a UART interrupt source, does not have an associated flag, but will generate a UART interrupt if the UART clock (f_{H1}) source is switched off and the WAKE and RIE bits in the UCR2 register are set when a falling edge on the RX/TX pin occurs.

Note that the USR register flags are read only and cannot be cleared or set by the application program, neither will they be cleared when the program jumps to the corresponding interrupt servicing routine, as is the case for some of the other interrupts. The flags will be cleared automatically when certain actions are taken by the UART, the details of which are given in the UART register section. The overall UART interrupt can be disabled or enabled by the related interrupt enable control bits in the interrupt control registers of the microcontroller to decide whether the interrupt requested by the UART module is masked out or allowed.



UART Interrupt Structure

Address Detect Mode

Setting the Address Detect Mode bit, ADDEN, in the UCR2 register, enables this special mode. If this bit is enabled then an additional qualifier will be placed on the generation of a Receiver Data Available interrupt, which is requested by the RXIF flag. If the ADDEN bit is enabled, then when data is available, an interrupt will only be generated, if the highest received bit has a high value. Note that the URE and EMI interrupt enable bits must also be enabled for correct interrupt generation. This highest address bit is the 9th bit if BNO=1 or the 8th bit if BNO=0. If this bit is high, then the received word will be defined as an address rather than data. A Data Available interrupt will be generated every time the last bit of the received word is set. If the ADDEN bit is not enabled, then a Receiver Data Available interrupt will be generated each time the RXIF flag is set, irrespective of the data last bit status. The address detect mode and parity enable are mutually exclusive functions. Therefore if the address detect mode is enabled, then to ensure correct operation, the parity function should be disabled by resetting the parity enable bit PREN to zero.

ADDEN	9th Bit if BNO=1, 8th Bit if BNO=0	UART Interrupt Generated
0	0	√
	1	√
1	0	×
	1	√

ADDEN Bit Function

UART Power Down and Wake-up

When the UART clock, f_{H} , is switched off, the UART will cease to function. If the MCU switches off the UART clock, f_{H} , and enters the power down mode while a transmission is still in progress, then the transmission will be paused until the UART clock source derived from the microcontroller is activated. In a similar way, if the MCU switches off the UART clock f_{H} and enters the IDLE or SLEEP mode by executing the “HALT” instruction while receiving data, then the reception of data will likewise be paused. When the MCU enters the IDLE or SLEEP mode, note that the USR, UCR1, UCR2, UCR3, TXR_RXR, as well as the BRG register will not be affected. It is recommended to make sure first that the UART data transmission or reception has been finished before the microcontroller enters the IDLE or SLEEP mode.

The UART function contains a receiver RX/TX pin wake-up function, which is enabled or disabled by the WAKE bit in the UCR2 register. If this bit, along with the UART enable bit, UARTEN, the receiver enable bit, RXEN and the receiver interrupt bit, RIE, are all set when the MCU enters the power down mode with the UART clock f_{H} being switched off, then a falling edge on the RX/TX pin will initiate an RX/TX pin wake-up UART interrupt. Note that as it takes certain system clock cycles after a wake-up, before normal microcontroller operation resumes, any data received during this time on the RX/TX pin will be ignored.

For a UART wake-up interrupt to occur, in addition to the bits for the wake-up being set, the global interrupt enable bit, EMI, and the UART interrupt enable bit, URE, must be set. If the EMI, the multi-function interrupt enable bit, MFnE, and URE bits are not set then only a wake up event will occur and no interrupt will be generated. Note also that as it takes certain system clock cycles after a wake-up before normal microcontroller resumes, the UART interrupt will not be generated until after this time has elapsed.

LCD Driver

For large volume applications, which incorporate an LCD in their design, the use of a custom display rather than a more expensive character based display reduces costs significantly. However, the corresponding COM and SEG signals required, which vary in both amplitude and time, to drive such a custom display require many special considerations for proper LCD operation to occur.

This device contains an LCD Driver function, which with its internal LCD signal generating circuitry and various options will automatically generate these time and amplitude varying signals to provide a means of direct driving and easy interfacing to a range of custom LCDs.

Driver No.	Duty	Bias	Bias Type	Wave Type
13x4	1/4	1/3	R or C type	A or B

LCD Driver Output Selection

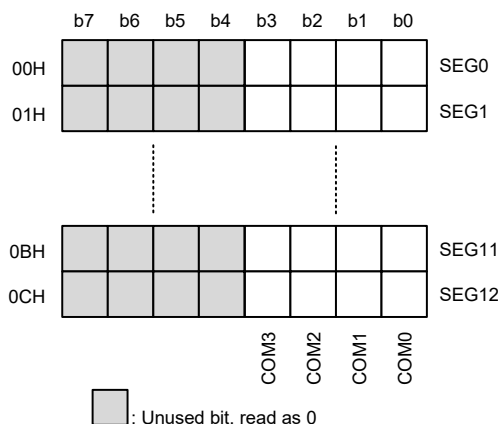
LCD Display Data Memory

An area of Data Memory is especially reserved for use for the LCD display data. This data area is known as the LCD Display Data Memory. Any data written here will be automatically read by the internal display driver circuits, which will in turn automatically generate the necessary LCD driving signals. Therefore any data written into this Memory will be immediately reflected into the actual display connected to the microcontroller.

This device provides an area of embedded data memory for the LCD display. This area is located at 00H to 0FH in Sector 4 of the Data Memory. The LCD display memory can be read and written by indirect addressing mode using MP1L/MP1H and MP2L/MP2H, or by direct addressing mode using the corresponding extended instructions. If using the indirect addressing to access the LCD Display Data Memory therefore requires first that Sector 4 is selected by writing a value of 04H to MP1H or MP2H. After this, the memory can then be accessed by using indirect addressing through the use of MP1L or MP2L. With Sector 4 selected, then using MP1L/MP2L to read or write to the memory area, from 00H, will result in operations to the LCD memory.

When data is written into the display data area, it is automatically read by the LCD driver which then generates the corresponding LCD driving signals. To turn the display on or off, a “1” or a “0” is written to the corresponding bit of the display memory, respectively.

The accompanying LCD Memory Map diagrams shows how the internal LCD Memory is mapped to the Segments and Commons of the display for the device. The unimplemented LCD RAM bits cannot be used as general purpose RAM for application. For example, if the LCD duty is selected as 1/4 duty (4COM), the COM b4~b7 will be read as 0 only.



LCD Memory Map

LCD Clock Source

The LCD clock source is the internal clock signal, f_{SUB} , divided by 8 using an internal divider circuit. The f_{SUB} internal clock is supplied from the LIRC oscillator. For proper LCD operation, this arrangement is provided to generate an ideal LCD clock source frequency of 4 kHz.

LCD Registers

There are two control registers, named as LCDC0 and LCDC1, in the Data Memory which are used to control the various setup features of the LCD Driver. Various bits in these registers control functions such as LCD wave type, bias type, supply power selection, total bias resistor selection together with the overall LCD enable and disable control.

The LCDEN bit in the LCDC0 register, which provides the overall LCD enable/disable function, will only be effective when the device is in the FAST, SLOW or IDLE Mode. If the device is in the SLEEP Mode, then the display will always be disabled. Bits, RSEL2~RSEL0, in the LCDC0 register select the internal total bias resistors to supply the LCD panel with the proper bias current. A choice to best match the LCD panel used in the application can be selected also to minimise bias current. The TYPE bit in the LCDC0 register is used to select whether Type A or Type B LCD waveform signals are used. The RCT bit in the same register is used to select whether R Type or C Type LCD bias is used. The LCDP1 and LCDP0 bits are used to select that the LCD supply power comes from either the external pin or internal power supply for C type bias application.

The PLCD3~PLCD0 bits in the LCDC1 register are used to select the V_A voltage for R type bias circuitry. The QCT2~QCT0 bits in the same register are used to determine the quick charging time period.

Register Name	Bit							
	7	6	5	4	3	2	1	0
LCDC0	TYPE	RCT	LCDP1	LCDP0	RSEL2	RSEL1	RSEL0	LCDEN
LCDC1	QCT2	QCT1	QCT0	—	PLCD3	PLCD2	PLCD1	PLCD0

LCD Control Register List

• LCDC0 Register

Bit	7	6	5	4	3	2	1	0
Name	TYPE	RCT	LCDP1	LCDP0	RSEL2	RSEL1	RSEL0	LCDEN
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7 **TYPE**: LCD waveform type selection

0: Type A

1: Type B

Bit 6 **RCT**: LCD bias type selection

0: R type

1: C type

When the RCT bit is cleared to 0, the LCDP[1:0] bits should be fixed at “00” and the power source is from the PLCD pin.

If the C1, C2 and V2 pin has pin-shared I/O or other pin-shared functions, when the RCT=1, the selected I/O or other pin-shared functions will interference with the C1, C2 and V2 functions.

Bit 5~4 **LCDP1~LCDP0**: C type bias LCD power source selection

00: From external pin PLCD, V1 or V2

01: From internal reference voltage V_{REFIN} supplied to VC

10: From internal voltage V_{DD} supplied to VB

11: From internal voltage V_{DD} supplied to V_A

The V_{REFIN} is an internal reference voltage with an approximate level of 1.04V.

Bit 3~1 **RSEL2~RSEL0**: R type bias LCD total bias resistors (R_T) selection

000: 1170k Ω

001: 225k Ω

010: 60k Ω

011: Quick charging mode – switching between 60k Ω and 1170k Ω

1xx: Quick charging mode – switching between 60k Ω and 225k Ω

The device provides the low power quick charging mode for R type bias LCD display.

In quick charging mode the LCD will provide more bias current with 60k Ω R_T used at the beginning of each COMn phase as LCD display refreshes and then provide less bias current with 1170k Ω or 225k Ω R_T used to reduce the bias current consumption in the remaining time duration in the same COMn phase.

Bit 0 **LCDEN**: LCD Enable control

0: Disable

1: Enable

In the FAST, SLOW or IDLE mode, the LCD on/off function can be controlled by this bit. However, in the SLEEP mode, the LCD function is always off.

• LCDC1 Register

Bit	7	6	5	4	3	2	1	0
Name	QCT2	QCT1	QCT0	—	PLCD3	PLCD2	PLCD1	PLCD0
R/W	R/W	R/W	R/W	—	R/W	R/W	R/W	R/W
POR	0	0	0	—	0	0	0	0

Bit 7~5 **QCT2~QCT0**: R type quick charging time selection

000: 1 t_{SUB}

001: 2 t_{SUB}

010: 3 t_{SUB}

011: 4 t_{SUB}

100: 5 t_{SUB}

101: 6 t_{SUB}

110: 7 t_{SUB}

111: 8 t_{SUB}

The t_{SUB} is the period of the LCD clock source f_{SUB} , i.e., $1/f_{SUB}$.

Bit 4 Unimplemented, read as “0”

Bit 3~0 **PLCD3~PLCD0**: R type bias supply voltage selection for V_A

0000: $8/16 \times V_{PLCD}$

0001: $9/16 \times V_{PLCD}$

0010: $10/16 \times V_{PLCD}$

0011: $11/16 \times V_{PLCD}$

0100: $12/16 \times V_{PLCD}$

0101: $13/16 \times V_{PLCD}$

0110: $14/16 \times V_{PLCD}$

0111: $15/16 \times V_{PLCD}$

1XXX: V_{PLCD}

Note that the V_A voltage level must be equal to or greater than 2.1V.

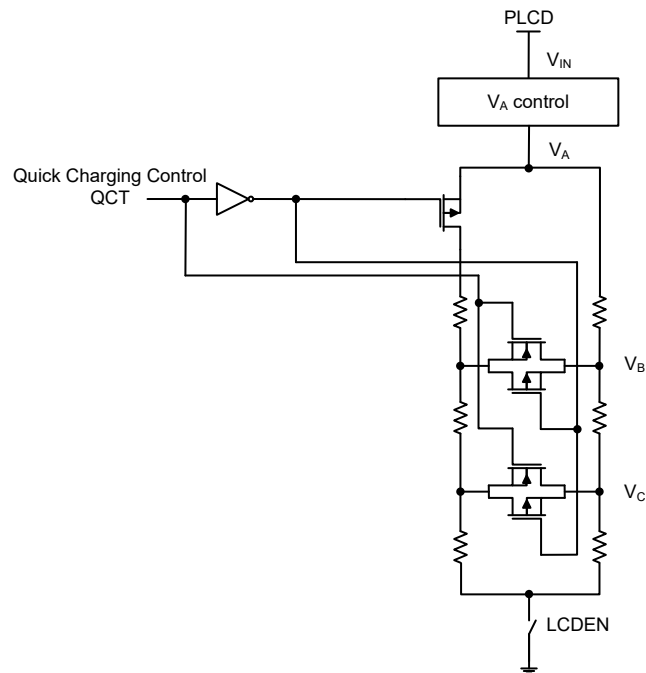
LCD Voltage Source and Biasing

The time and amplitude varying signals generated by the LCD Driver function require the generation of several voltage levels for their operation. The device can have either R type or C type biasing selected via a software control bit RCT. Selecting the C type biasing will enable C type internal charge pump circuitry.

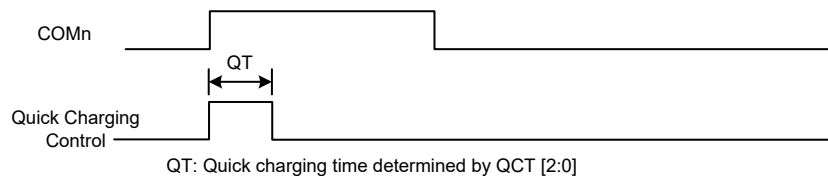
R Type Biasing

For R type biasing an external LCD voltage source must be supplied on pin PLCD to generate the internal biasing voltages. This could be the microcontroller power supply V_{DD} or some other voltage source equal to or less than V_{DD} . For the R type 1/3 bias scheme, four voltage levels V_{SS} , V_A , V_B and V_C are utilised. The voltage V_A which is selected by the PLCD3~PLCD0 bits in the LCDC1 register can be equal to a specific ratio of PLCD voltage, varying from $8/16 V_{PLCD}$ to V_{PLCD} . The voltage V_B is equal to $V_A \times 2/3$ while the voltage V_C is equal to $V_A \times 1/3$.

Different values of internal bias resistors can be selected using the RSEL2~RSEL0 bits in the LCDC0 register. This along with the voltage on pin PLCD will determine the bias current. The VMAX pin should be connected to the VDD pin since the available maximum voltage applied to the PLCD pin is equal to V_{DD} .



Note: When the R type LCD is disabled, the DC path will be switched off.



R Type Bias Configuration – 1/3 Bias

C Type Biasing

For C type biasing the LCD voltage source can be supplied on the external pin PLCD, V1 or V2 or can be derived from the internal power source to generate the required biasing voltages. The C type bias voltage source is selected using the LCDP1~LCDP0 bits in the LCDC0 register.

When the LCD voltage source is from the PLCD or V2 pin, the C type biasing scheme uses an internal charge pump circuit, which can generate voltages higher than what is supplied on PLCD or V2. This feature is useful in applications where the microcontroller supply voltage is less than the

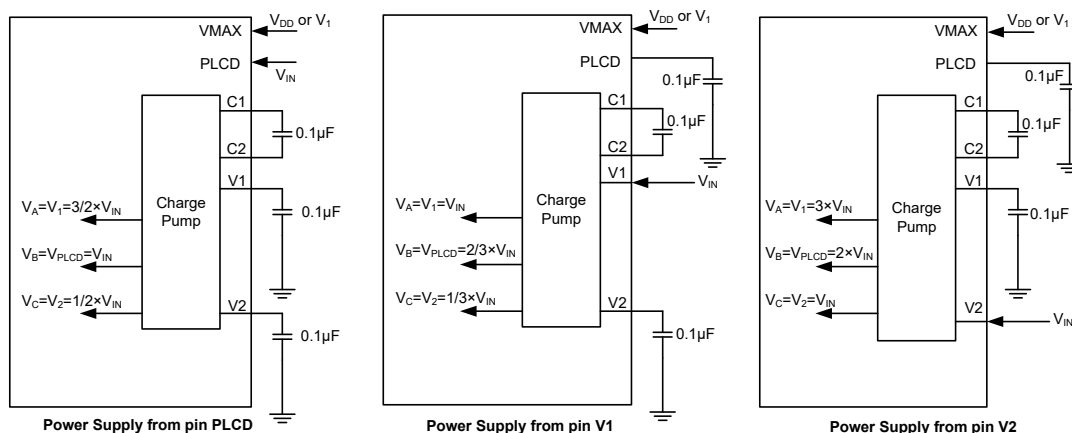
supply voltage required by the LCD. An additional charge pump capacitor must also be connected between pins C1 and C2 to generate the necessary voltage levels.

For C type 1/3 bias scheme and whether the LCD power is selected from external pin or internal voltage, four voltage levels V_{SS} , V_A , V_B and V_C are utilised. These bias voltages have different levels depending upon different LCD power supply schemes.

LCD Power Supply		V_A Voltage	V_B Voltage	V_C Voltage
External Power Supply	V_{IN} From V1 pin	V_{IN}	$2/3 \times V_{IN}$	$1/3 \times V_{IN}$
	V_{IN} From PLCD pin	$3/2 \times V_{IN}$	V_{IN}	$1/2 \times V_{IN}$
	V_{IN} From V2 pin	$3 \times V_{IN}$	$2 \times V_{IN}$	V_{IN}
Internal Power Supply	V_A ($V_A = V_{DD}$)	V_{DD}	$2/3 \times V_{DD}$	$1/3 \times V_{DD}$
	V_B ($V_B = V_{DD}$)	$3/2 \times V_{DD}$	V_{DD}	$1/2 \times V_{DD}$
	V_C ($V_C = V_{REFIN}$) ^{Note}	$V_{REFIN} \times 3$	$V_{REFIN} \times 2$	V_{REFIN}

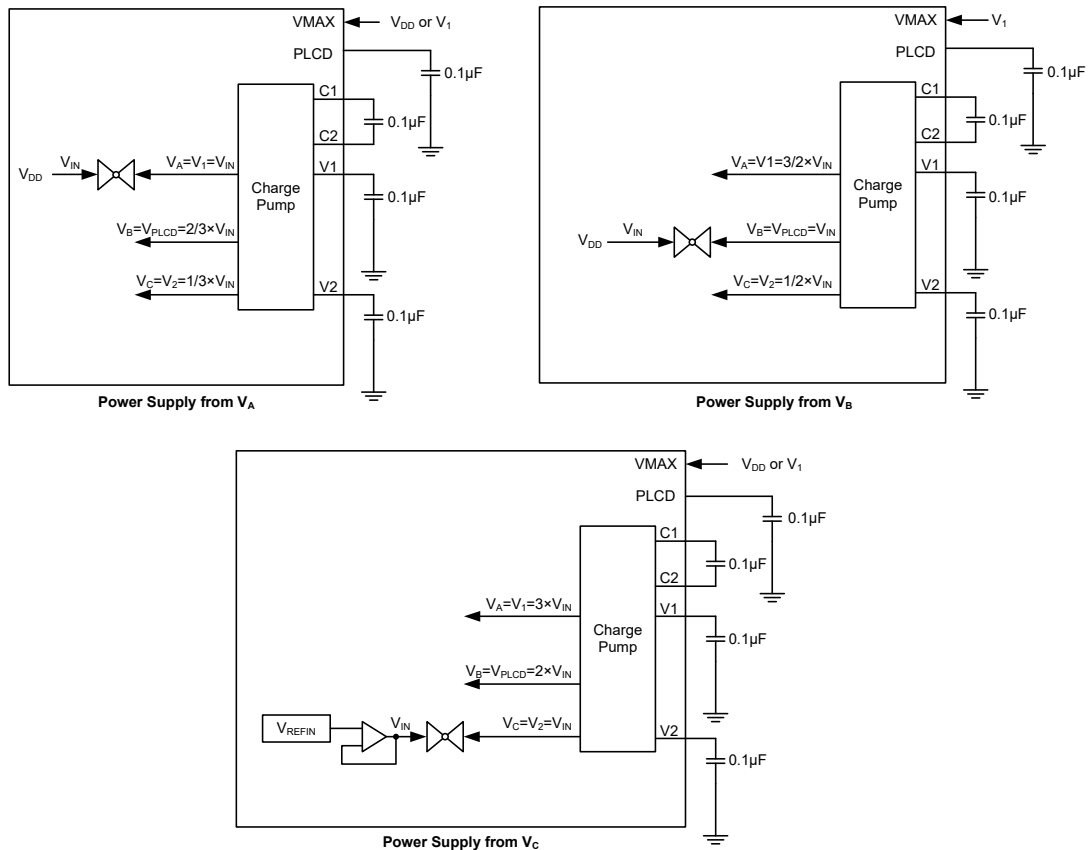
Note: The V_{REFIN} voltage is from the device ingenerated depletion circuit and has an approximate level of 1.04V.

C Type Bias Power Supply Scheme



Note: The pin VMAX must be connected to the maximum voltage to prevent from the pad leakage.

C Type Bias External Power Supply Configuration – 1/3 Bias



Note: The pin VMAX must be connected to the maximum voltage to prevent from the pad leakage.

C Type Bias Internal Power Supply Configuration – 1/3 Bias

The connection to the VMAX pin depends upon the LCD power supply scheme. The details are shown in the table. It is extremely important to ensure that these charge pump generated internal voltages do not exceed the maximum V_{DD} voltage of 5.5V.

Condition	VMAX Connection
$V_{DD} > V_{PLCD} \times 1.5$	Connect VMAX to VDD
Otherwise	Connect VMAX to V1

C Type Bias VMAX Pin Connection

LCD Reset Status

The LCD has an internal reset function that is an OR function of the inverted LCDEN bit in the LCDEN0 register and the SLEEP function. Clearing the LCDEN bit to zero will reset the LCD function. The LCD function will also be reset after the device enters the SLEEP mode even if the LCDEN bit is set to “1” to enable the LCD driver function.

When the LCDEN bit is set to “1” to enable the LCD driver and then an MCU reset occurs, the LCD driver will be reset and the COM and SEG output will be in a floating state during the MCU reset duration. The reset operation will take a time of $t_{RSTD} + t_{SST}$. Refer to the System Start Up Time Characteristics for t_{RSTD} and t_{SST} details.

MCU Reset	SLEEP Mode	LCDEN	LCD Reset	COM & SEG Voltage Level
No	Off	1	No	Normal Operation
No	Off	0	Yes	Low
No	On	x	Yes	Low
Yes	x	x	Yes	Floating

- Note: 1. The watchdog time-out reset in the IDLE or SLEEP Mode is excluded from the MCU Reset conditions.
2. “x”: Don’t care.

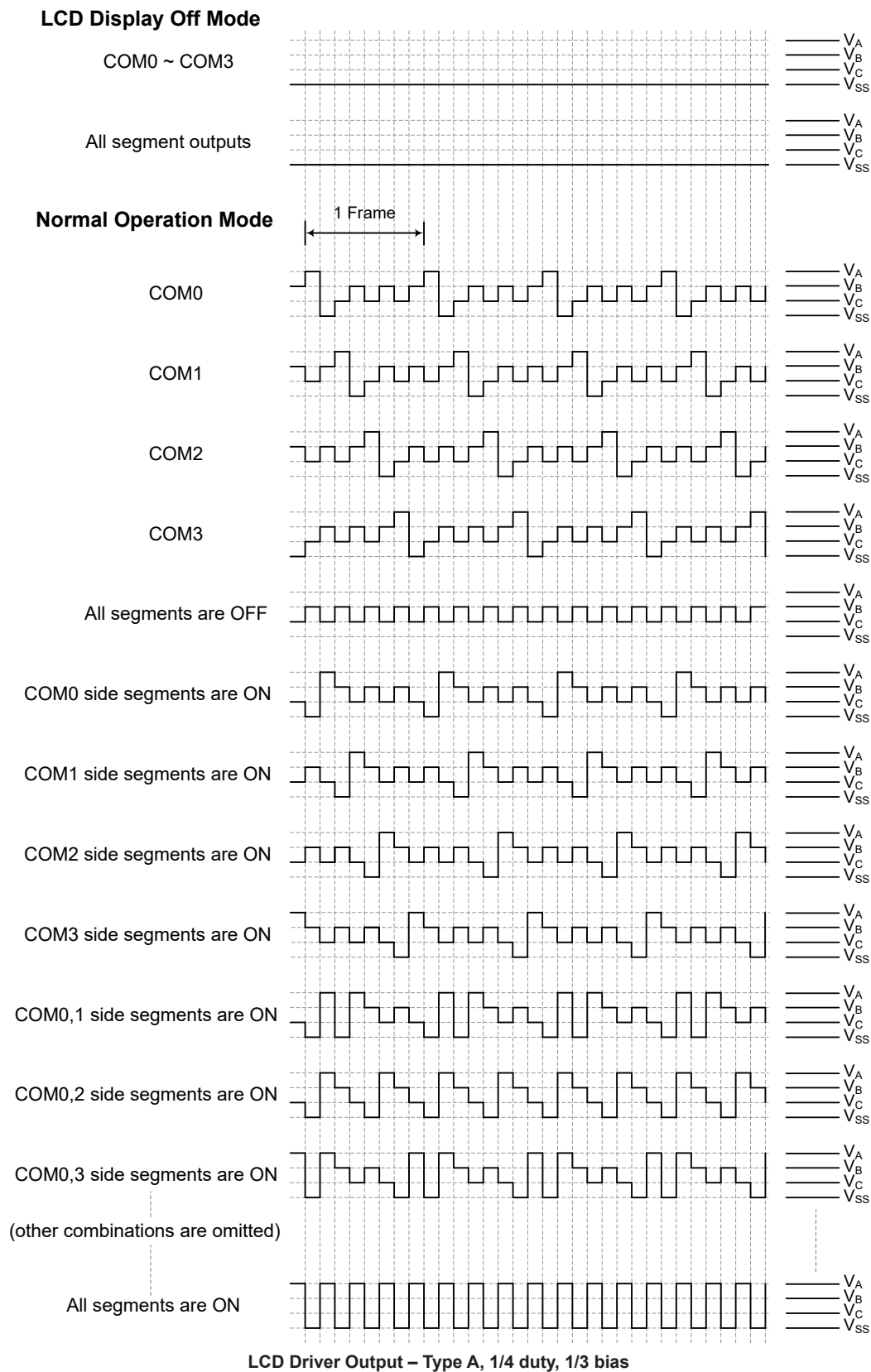
LCD Reset Status

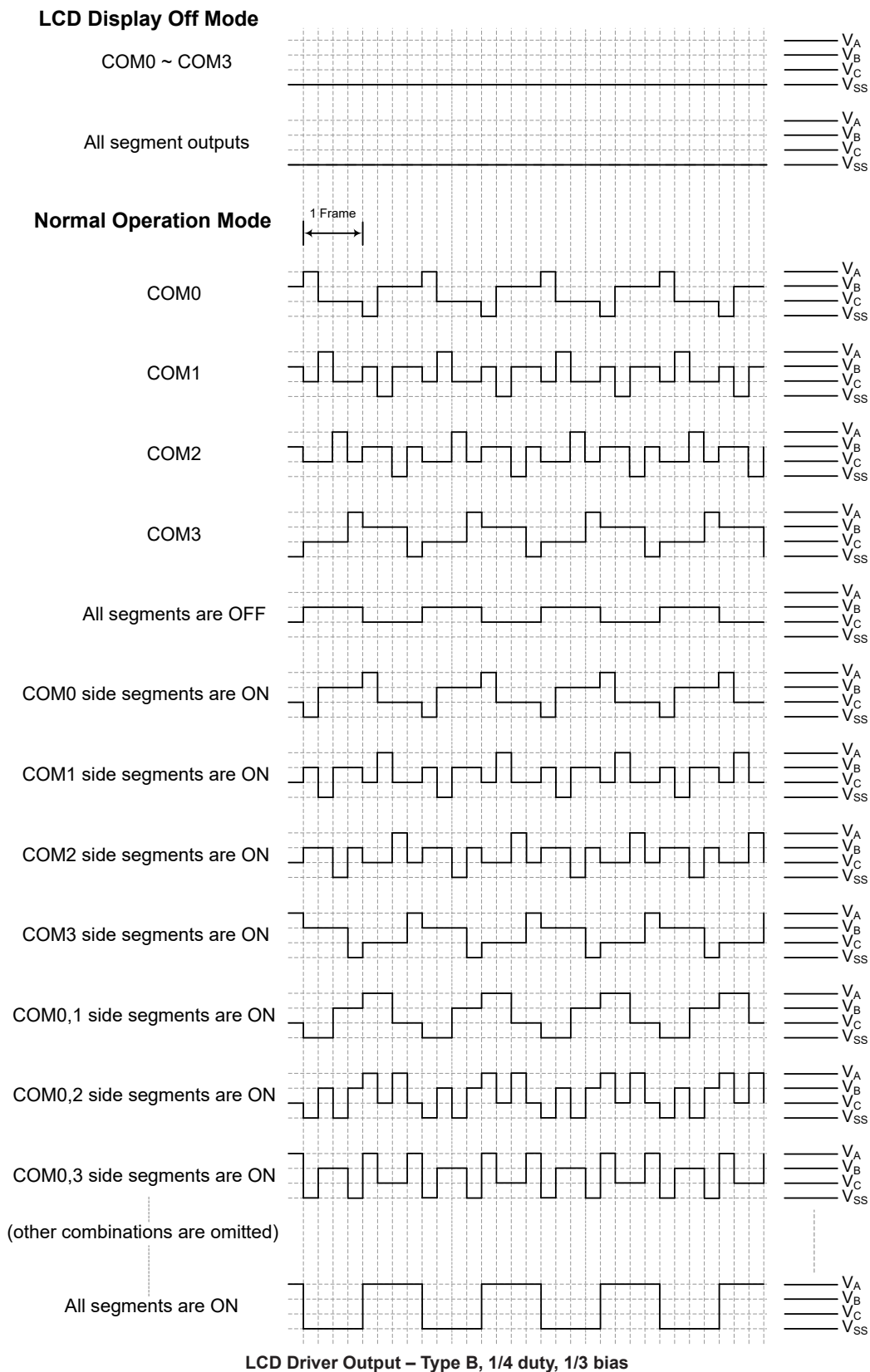
LCD Driver Output

The number of COM and SEG outputs supplied by the LCD driver, as well as its biasing and wave type selections, are dependent upon how the LCD control bits are programmed. The Bias Type, whether C or R type is also selected by a software control bit.

The nature of Liquid Crystal Displays requires that only AC voltages can be applied to their pixels as the application of DC voltages to LCD pixels may cause permanent damage. For this reason the relative contrast of an LCD display is controlled by the actual RMS voltage applied to each pixel, which is equal to the RMS value of the voltage on the COM pin minus the voltage applied to the SEG pin. This differential RMS voltage must be greater than the LCD saturation voltage for the pixel to be on and less than the threshold voltage for the pixel to be off.

The requirement to limit the DC voltage to zero and to control as many pixels as possible with a minimum number of connections requires that both a time and amplitude signal is generated and applied to the application LCD. These time and amplitude varying signals are automatically generated by the LCD driver circuits in the microcontroller. What is known as the duty determines the number of common lines used, which are also known as backplanes or COMs. The duty, which is to have a value of 1/4 and which equates to a COM number of 4, therefore defines the number of time divisions within each LCD signal frame. Two types of signal generation are also provided, known as Type A and Type B, the required type is selected via the TYPE bit in the LCDC0 register. Type B offers lower frequency signals, however, lower frequencies may introduce flickering and influence display clarity.





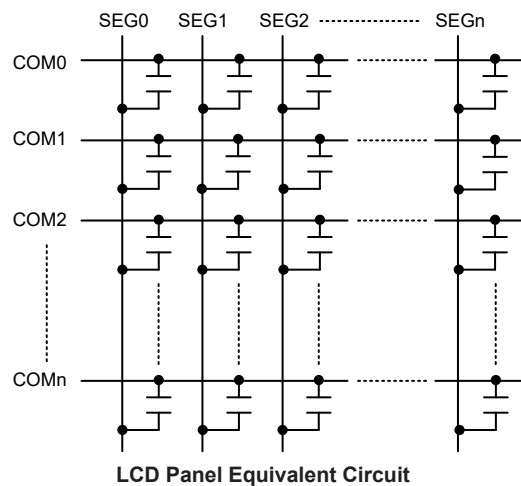
Programming Considerations

Certain precautions must be taken when programming the LCD. One of these is to ensure that the LCD Memory is properly initialised after the microcontroller is powered on. Like the General Purpose Data Memory, the contents of the LCD Memory are in an unknown condition after power-on. As the contents of the LCD Memory will be mapped into the actual display, it is important to initialise this memory area into a known condition soon after applying power to obtain a proper display pattern.

Consideration must also be given to the capacitive load of the actual LCD used in the application. As the load presented to the microcontroller by LCD pixels can be generally modeled as mainly capacitive in nature, it is important that this is not excessive, a point that is particularly true in the case of the COM lines which may be connected to many LCD pixels. The accompanying diagram depicts the equivalent circuit of the LCD.

One additional consideration that must be taken into account is what happens when the LCD driver clock source is turned off. The LCDEN control bit in the LCDC0 register permits the display to be powered off to reduce power consumption. If this bit is zero, the driving signals to the display will cease, producing a blank display pattern but reducing any power consumption associated with the LCD.

After Power-on, note that as the LCDEN bit will be cleared to zero, the display function will be disabled.



Low Voltage Detector – LVD

The device has a Low Voltage Detector function, also known as LVD. This enabled the device to monitor the power supply voltage, V_{DD} , and provide a warning signal should it fall below a certain level. This function may be especially useful in battery applications where the supply voltage will gradually reduce as the battery ages, as it allows an early warning battery low signal to be generated. The Low Voltage Detector also has the capability of generating an interrupt signal.

LVD Register

The Low Voltage Detector function is controlled using a single register with the name LVDC. Three bits in this register, VLVD2~VLVD0, are used to select one of eight fixed voltages below which a low voltage condition will be determined. A low voltage condition is indicated when the LVDO bit is set. If the LVDO bit is low, this indicates that the V_{DD} voltage is above the preset low voltage value. The LVDEN bit is used to control the overall on/off function of the low voltage detector. Setting the bit high will enable the low voltage detector. Clearing the bit to zero will switch off the internal low voltage detector circuits. As the low voltage detector will consume a certain amount of power, it may be desirable to switch off the circuit when not in use, an important consideration in power sensitive battery powered applications.

• LVDC Register

Bit	7	6	5	4	3	2	1	0
Name	—	—	LVDO	LVDEN	VBGEN	VLVD2	VLVD1	VLVD0
R/W	—	—	R	R/W	R/W	R/W	R/W	R/W
POR	—	—	0	0	0	0	0	0

Bit 7~6 Unimplemented, read as “0”

Bit 5 **LVDO**: LVD Detection Output flag
 0: No Low Voltage Detected
 1: Low Voltage Detected

Bit 4 **LVDEN**: Low Voltage Detector control
 0: Disable
 1: Enable

In the FAST, SLOW or IDLE mode, the LVD function can be controlled by this bit. However, in the SLEEP mode, the LVD function is always off.

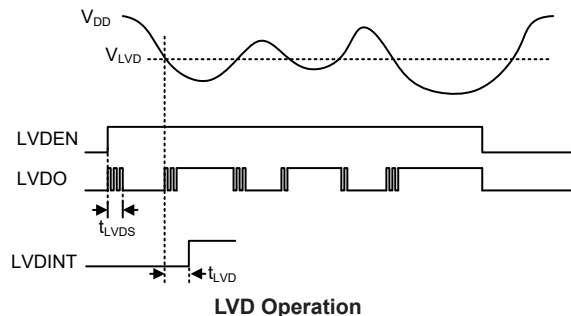
Bit 3 **VBGEN**: Bandgap Buffer control
 0: Disable
 1: Enable

Note that the Bandgap circuit is enabled when the LVD or LVR function is enabled or when the VBGEN bit is set to 1.

Bit 2~0 **VLVD2~VLVD0**: Select LVD Reference Voltage
 000: 2.0V
 001: 2.2V
 010: 2.4V
 011: 2.7V
 100: 3.0V
 101: 3.3V
 110: 3.6V
 111: 4.0V

LVD Operation

The Low Voltage Detector function operates by comparing the power supply voltage, V_{DD} , with a pre-specified voltage level stored in the LVDC register. This has a range of between 2.0V and 4.0V. When the power supply voltage, V_{DD} , falls below this pre-determined value, the LVDO bit will be set high indicating a low power supply voltage condition. When the device is in the SLEEP mode, the low voltage detector will be disabled even if the LVDEN bit is high. After enabling the Low Voltage Detector, a time delay t_{LVDS} should be allowed for the circuitry to stabilise before reading the LVDO bit. Note also that as the V_{DD} voltage may rise and fall rather slowly, at the voltage nears that of V_{LVD} , there may be multiple bit LVDO transitions.



The Low Voltage Detector also has its own interrupt, providing an alternative means of low voltage detection, in addition to polling the LVDO bit. The interrupt will only be generated after a delay of t_{LVD} after the LVDO bit has been set high by a low voltage condition. In this case, the LVF interrupt request flag will be set, causing an interrupt to be generated if V_{DD} falls below the preset LVD voltage. This will cause the device to wake up from the IDLE Mode, however if the Low Voltage Detector wake-up function is not required then the LVF flag should be first set high before the device enters the IDLE Mode.

Interrupts

Interrupts are an important part of any microcontroller system. When an external event or an internal function such as a Timer Module or an A/D converter requires microcontroller attention, their corresponding interrupt will enforce a temporary suspension of the main program allowing the microcontroller to direct attention to their respective needs. The device contains several external interrupt and internal interrupt functions. The external interrupts are generated by the action of the external INT0 and INT1 pins, while the internal interrupts are generated by various internal functions such as the TMs, Time Base, USIM, UART, LVD, EEPROM and the A/D converter, etc.

Interrupt Registers

Overall interrupt control, which basically means the setting of request flags when certain microcontroller conditions occur and the setting of interrupt enable bits by the application program, is controlled by a series of registers, located in the Special Purpose Data Memory, as shown in the accompanying table. The number of registers falls into two categories. The first is the INTC0~INTC3 registers which setup the primary interrupts, the second is the INTEG register to setup the external interrupt trigger edge type.

Each register contains a number of enable bits to enable or disable individual registers as well as interrupt flags to indicate the presence of an interrupt request. The naming convention of these follows a specific pattern. First is listed an abbreviated interrupt type, then the (optional) number of that interrupt followed by either an "E" for enable/disable bit or "F" for request flag.

Function	Enable Bit	Request Flag	Notes
Global	EMI	—	—
External Interrupt	INTnE	INTnF	n=0~1
USIM	USIME	USIMF	—
UART	UARTE	UARTF	—
LVD	LVE	LVF	—
A/D Converter	ADE	ADF	—
EEPROM	DEE	DEF	—
Time Base	TBnE	TBnF	n=0~1
PTM	PTMPE	PTMPF	—
	PTMAE	PTMAF	
STM	STMPE	STMPF	—
	STMAE	STMAF	

Interrupt Register Bit Naming Conventions

Register Name	Bit							
	7	6	5	4	3	2	1	0
INTEG	—	—	—	—	INT1S1	INT1S0	INT0S1	INT0S0
INTC0	—	USIMF	INT1F	INT0F	USIME	INT1E	INT0E	EMI
INTC1	TB0F	DEF	ADF	LVF	TB0E	DEE	ADE	LVE
INTC2	STMAF	STMPF	PTMAF	PTMPF	STMAE	STMPE	PTMAE	PTMPE
INTC3	—	—	UARTF	TB1F	—	—	UARTE	TB1E

Interrupt Register List

• **INTEG Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	INT1S1	INT1S0	INT0S1	INT0S0
R/W	—	—	—	—	R/W	R/W	R/W	R/W
POR	—	—	—	—	0	0	0	0

- Bit 7~4 Unimplemented, read as “0”
- Bit 3~2 **INT1S1~INT1S0**: Interrupt edge control for INT1 pin
 00: Disable
 01: Rising edge
 10: Falling edge
 11: Rising and falling edges
- Bit 1~0 **INT0S1~INT0S0**: Interrupt edge control for INT0 pin
 00: Disable
 01: Rising edge
 10: Falling edge
 11: Rising and falling edges

• **INTC0 Register**

Bit	7	6	5	4	3	2	1	0
Name	—	USIMF	INT1F	INT0F	USIME	INT1E	INT0E	EMI
R/W	—	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	—	0	0	0	0	0	0	0

- Bit 7 Unimplemented, read as “0”
- Bit 6 **USIMF**: USIM interrupt request flag
0: No request
1: Interrupt request
- Bit 5 **INT1F**: INT1 interrupt request flag
0: No request
1: Interrupt request
- Bit 4 **INT0F**: INT0 interrupt request flag
0: No request
1: Interrupt request
- Bit 3 **USIME**: USIM interrupt control
0: Disable
1: Enable
- Bit 2 **INT1E**: INT1 interrupt control
0: Disable
1: Enable
- Bit 1 **INT0E**: INT0 interrupt control
0: Disable
1: Enable
- Bit 0 **EMI**: Global interrupt control
0: Disable
1: Enable

• **INTC1 Register**

Bit	7	6	5	4	3	2	1	0
Name	TB0F	DEF	ADF	LVF	TB0E	DEE	ADE	LVE
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

- Bit 7 **TB0F**: Time Base 0 interrupt request flag
0: No request
1: Interrupt request
- Bit 6 **DEF**: Data EEPROM interrupt request flag
0: No request
1: Interrupt request
- Bit 5 **ADF**: A/D Converter interrupt request flag
0: No request
1: Interrupt request
- Bit 4 **LVF**: LVD interrupt request flag
0: No request
1: Interrupt request
- Bit 3 **TB0E**: Time Base 0 interrupt control
0: Disable
1: Enable
- Bit 2 **DEE**: Data EEPROM interrupt control
0: Disable
1: Enable

- Bit 1 **ADE**: A/D Converter interrupt control
0: Disable
1: Enable
- Bit 0 **LVE**: LVD interrupt control
0: Disable
1: Enable

• **INTC2 Register**

Bit	7	6	5	4	3	2	1	0
Name	STMAF	STMPF	PTMAF	PTMPF	STMAE	STMPE	PTMAE	PTMPE
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

- Bit 7 **STMAF**: STM Comparator A match interrupt request flag
0: No request
1: Interrupt request
- Bit 6 **STMPF**: STM Comparator P match interrupt request flag
0: No request
1: Interrupt request
- Bit 5 **PTMAF**: PTM Comparator A match interrupt request flag
0: No request
1: Interrupt request
- Bit 4 **PTMPF**: PTM Comparator P match interrupt request flag
0: No request
1: Interrupt request
- Bit 3 **STMAE**: STM Comparator A match interrupt control
0: Disable
1: Enable
- Bit 2 **STMPE**: STM Comparator P match interrupt control
0: Disable
1: Enable
- Bit 1 **PTMAE**: PTM Comparator A match interrupt control
0: Disable
1: Enable
- Bit 0 **PTMPE**: PTM Comparator P match interrupt control
0: Disable
1: Enable

• **INTC3 Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	UARTF	TB1F	—	—	UARTE	TB1E
R/W	—	—	R/W	R/W	—	—	R/W	R/W
POR	—	—	0	0	—	—	0	0

- Bit 7~6 Unimplemented, read as “0”
- Bit 5 **UARTF**: UART transfer interrupt request flag
0: No request
1: Interrupt request
- Bit 4 **TB1F**: Time Base 1 interrupt request flag
0: No request
1: Interrupt request
- Bit 3~2 Unimplemented, read as “0”

Bit 1	UARTE : UART transfer interrupt control 0: Disable 1: Enable
Bit 0	TBIE : Time Base 1 interrupt control 0: Disable 1: Enable

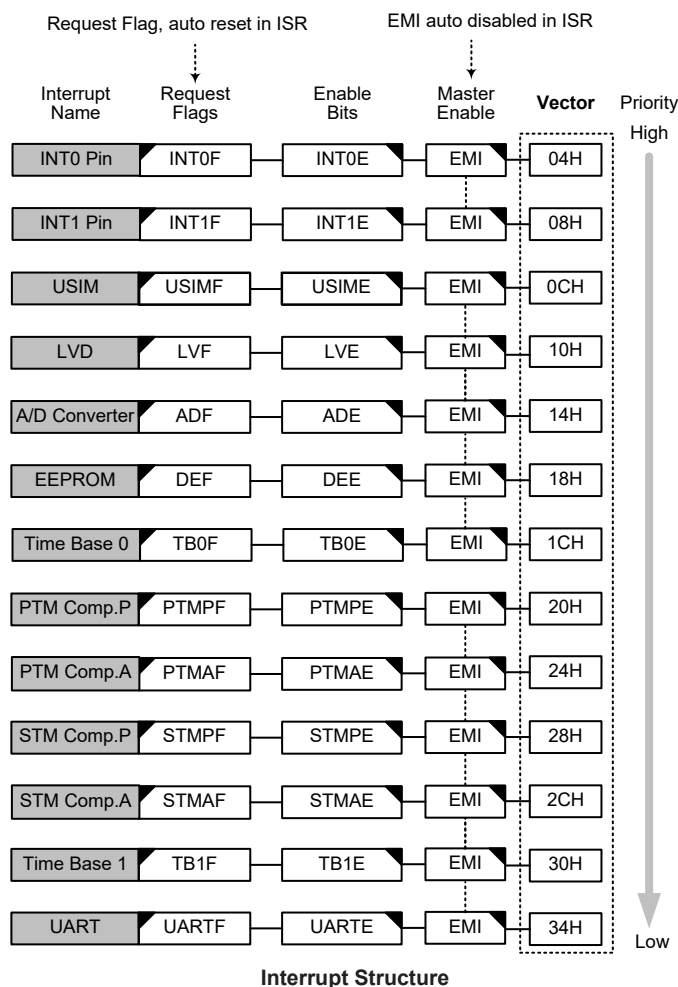
Interrupt Operation

When the conditions for an interrupt event occur, such as a TM Comparator P or Comparator A match or A/D conversion completion etc., the relevant interrupt request flag will be set. Whether the request flag actually generates a program jump to the relevant interrupt vector is determined by the condition of the interrupt enable bit. If the enable bit is set high then the program will jump to its relevant vector; if the enable bit is zero then although the interrupt request flag is set an actual interrupt will not be generated and the program will not jump to the relevant interrupt vector. The global interrupt enable bit, if cleared to zero, will disable all interrupts.

When an interrupt is generated, the Program Counter, which stores the address of the next instruction to be executed, will be transferred onto the stack. The Program Counter will then be loaded with a new address which will be the value of the corresponding interrupt vector. The microcontroller will then fetch its next instruction from this interrupt vector. The instruction at this vector will usually be a “JMP” which will jump to another section of program which is known as the interrupt service routine. Here is located the code to control the appropriate interrupt. The interrupt service routine must be terminated with a “RETI”, which retrieves the original Program Counter address from the stack and allows the microcontroller to continue with normal execution at the point where the interrupt occurred.

The various interrupt enable bits, together with their associated request flags, are shown in the accompanying diagrams with their order of priority. All interrupt sources have their own individual vector. Once an interrupt subroutine is serviced, all the other interrupts will be blocked, as the global interrupt enable bit, EMI bit will be cleared automatically. This will prevent any further interrupt nesting from occurring. However, if other interrupt requests occur during this interval, although the interrupt will not be immediately serviced, the request flag will still be recorded.

If an interrupt requires immediate servicing while the program is already in another interrupt service routine, the EMI bit should be set after entering the routine, to allow interrupt nesting. If the stack is full, the interrupt request will not be acknowledged, even if the related interrupt is enabled, until the Stack Pointer is decremented. If immediate service is desired, the stack must be prevented from becoming full. In case of simultaneous requests, the accompanying diagram shows the priority that is applied. All of the interrupt request flags when set will wake up the device if it is in SLEEP or IDLE Mode, however to prevent a wake-up from occurring the corresponding flag should be set before the device is in SLEEP or IDLE Mode.



External Interrupts

The external interrupts are controlled by signal transitions on the pins INT0 and INT1. An external interrupt request will take place when the external interrupt request flag, INTnF, is set, which will occur when a transition, whose type is chosen by the edge select bits, appears on the external interrupt pin. To allow the program to branch to their respective interrupt vector addresses, the global interrupt enable bit, EMI, and respective external interrupt enable bit, INTnE, must first be set. Additionally the correct interrupt edge type must be selected using the INTEG register to enable the external interrupt function and to choose the trigger edge type. As the external interrupt pins are pin-shared with I/O pins, they can only be configured as external interrupt pins if their external interrupt enable bit in the corresponding interrupt register has been set and the external interrupt pin is selected by the corresponding pin-shared function selection bits. The pin must also be set as an input by setting the corresponding bit in the port control register. When the interrupt is enabled, the stack is not full and the correct transition type appears on the external interrupt pin, a subroutine call to the external interrupt vector, will take place. When the interrupt is serviced, the external interrupt request flag, INTnF, will be automatically reset and the EMI bit will be automatically cleared to disable other interrupts. Note that any pull-high resistor selections on the external interrupt pins will remain valid even if the pin is used as an external interrupt input.

The INTEG register is used to select the type of active edge that will trigger the external interrupt. A choice of either rising or falling or both edge types can be chosen to trigger an external interrupt. Note that the INTEG register can also be used to disable the external interrupt function.

Universal Serial Interface Module Interrupt

The Universal Serial Interface Module Interrupt, also known as the USIM interrupt, will take place when the USIM Interrupt request flag, USIMF, is set. As the USIM interface can operate in three modes which are SPI mode, I²C mode and UART mode, the USIMF flag can be set by different conditions depending on the selected interface mode.

If the SPI or I²C mode is selected, the USIM interrupt can be triggered when a byte of data has been received or transmitted by the SPI/I²C interface, or an I²C slave address match occurs, or an I²C bus time-out occurs. If the UART mode is selected, several individual UART conditions including a transmitter data register empty, transmitter idle, receiver data available, receiver overrun, address detect and an URX/UTX pin wake-up, can generate an USIM interrupt with the USIMF flag bit set high.

To allow the program to branch to its interrupt vector address, the global interrupt enable bit, EMI, and the Serial Interface Interrupt enable bit, USIME, must first be set. When the interrupt is enabled, the stack is not full and any of the above described situations occurs, a subroutine call to the USIM Interrupt vector, will take place. When the interrupt is serviced, the Universal Serial Interface Interrupt flag, USIMF, will be automatically cleared. The EMI bit will also be automatically cleared to disable other interrupts.

Note that if the USIM interrupt is triggered by the UART interface, after the interrupt has been serviced, the UUSR register flags will only be cleared when certain actions are taken by the UART, the details of which are given in the UART section.

LVD Interrupt

An LVD Interrupt request will take place when the LVD Interrupt request flag, LVF, is set, which occurs when the Low Voltage Detector function detects a low power supply voltage. To allow the program to branch to its interrupt vector address, the global interrupt enable bit, EMI, and the Low Voltage Interrupt enable bit, LVE, must first be set. When the interrupt is enabled, the stack is not full and a low voltage condition occurs, a subroutine call to the LVD Interrupt vector, will take place. When the Low Voltage Interrupt is serviced, the LVF flag will be automatically cleared. The EMI bit will also be automatically cleared to disable other interrupts.

A/D Converter Interrupt

An A/D Converter Interrupt request will take place when the A/D Converter Interrupt request flag, ADF, is set, which occurs when the A/D conversion process finishes. To allow the program to branch to its interrupt vector address, the global interrupt enable bit, EMI, and A/D Interrupt enable bit, ADE, must first be set. When the interrupt is enabled, the stack is not full and the A/D conversion process has ended, a subroutine call to the A/D Converter Interrupt vector, will take place. When the A/D Converter Interrupt is serviced, the A/D Interrupt flag, ADF, will be automatically cleared. The EMI bit will also be automatically cleared to disable other interrupts.

EEPROM Interrupt

An EEPROM Interrupt request will take place when the EEPROM Interrupt request flag, DEF, is set, which occurs when an EEPROM Write cycle ends. To allow the program to branch to its interrupt vector address, the global interrupt enable bit, EMI, and EEPROM Interrupt enable bit, DEE, must first be set. When the interrupt is enabled, the stack is not full and an EEPROM Write cycle ends,

a subroutine call to the EEPROM Interrupt vector will take place. When the EEPROM Interrupt is serviced, the DEF flag will be automatically cleared. The EMI bit will also be automatically cleared to disable other interrupts.

TM Interrupts

The Standard and Periodic Type TMs each have two interrupts, one comes from the comparator A match situation and the other comes from the comparator P match situation. For all of the TM types there are two interrupt request flags and two enable control bits. A TM interrupt request will take place when any of the TM request flags is set, a situation which occurs when a TM comparator P or A match situation happens.

To allow the program to branch to their respective interrupt vector addresses, the global interrupt enable bit, EMI and the respective TM Interrupt enable bit must first be set. When the interrupt is enabled, the stack is not full and a TM comparator match situation occurs, a subroutine call to the relevant TM Interrupt vector location, will take place. When the TM interrupt is serviced, the TM interrupt flag will be automatically cleared. The EMI bit will also be automatically cleared to disable other interrupts.

UART Transfer Interrupt

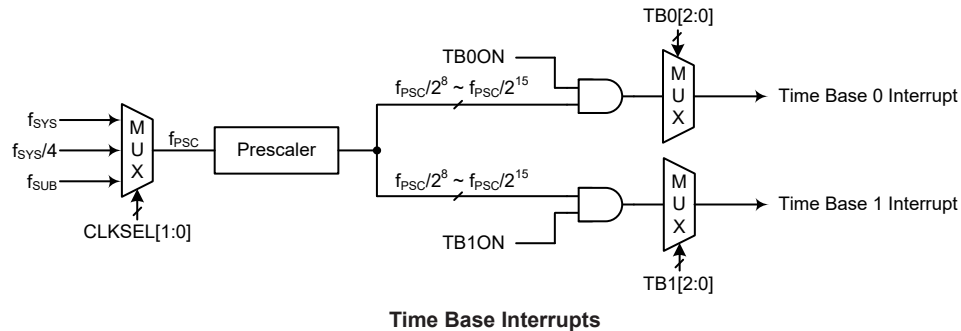
The UART Transfer Interrupt is controlled by several UART transfer conditions. When one of these conditions occurs, an interrupt pulse will be generated to get the attention of the microcontroller. These conditions are a transmitter data register empty, transmitter idle, receiver data available, receiver overrun, address detect and an RX/TX pin wake-up. To allow the program to branch to its interrupt vector address, the global interrupt enable bit, EMI, and UART Interrupt enable bit, UARTE, must first be set. When the interrupt is enabled, the stack is not full and any of the conditions described above occurs, a subroutine call to the UART Interrupt vector, will take place. When the interrupt is serviced, the UART Interrupt flag, UARTEF, will be automatically cleared. The EMI bit will also be automatically cleared to disable other interrupts.

However, the USR register flags will be cleared automatically when certain actions are taken by the UART, the details of which are given in the UART section.

Time Base Interrupts

The function of the Time Base Interrupts is to provide regular time signals in the form of an internal interrupt. They are controlled by the overflow signals from their respective timer functions. When these happens their respective interrupt request flags, TB0F or TB1F will be set. To allow the program to branch to their respective interrupt vector addresses, the global interrupt enable bit, EMI and Time Base enable bits, TB0E or TB1E, must first be set. When the interrupt is enabled, the stack is not full and the Time Base overflows, a subroutine call to their respective vector locations will take place. When the interrupt is serviced, the respective interrupt request flag, TB0F or TB1F, will be automatically reset and the EMI bit will be cleared to disable other interrupts.

The purpose of the Time Base Interrupt is to provide an interrupt signal at fixed time periods. The clock source, f_{PSC} , originates from the internal clock source f_{SYS} , $f_{SYS}/4$ or f_{SUB} and then passes through a divider, the division ratio of which is selected by programming the appropriate bits in the TB0C and TB1C registers to obtain longer interrupt periods whose value ranges. The clock source that generates f_{PSC} , which in turn controls the Time Base interrupt period, is selected using the CLKSEL1~CLKSEL0 bits in the PSCR register.



• **PSCR Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	CLKSEL1	CLKSEL0
R/W	—	—	—	—	—	—	R/W	R/W
POR	—	—	—	—	—	—	0	0

Bit 7~2 Unimplemented, read as “0”

Bit 1~0 **CLKSEL1~CLKSEL0**: Prescaler clock source selection

00: f_{SYS}

01: $f_{SYS}/4$

1x: f_{SUB}

• **TB0C Register**

Bit	7	6	5	4	3	2	1	0
Name	TB0ON	—	—	—	—	TB02	TB01	TB00
R/W	R/W	—	—	—	—	R/W	R/W	R/W
POR	0	—	—	—	—	0	0	0

Bit 7 **TB0ON**: Time Base 0 Control

0: Disable

1: Enable

Bit 6~3 Unimplemented, read as “0”

Bit 2~0 **TB02~TB00**: Select Time Base 0 Time-out Period

000: $2^8/f_{PSC}$

001: $2^9/f_{PSC}$

010: $2^{10}/f_{PSC}$

011: $2^{11}/f_{PSC}$

100: $2^{12}/f_{PSC}$

101: $2^{13}/f_{PSC}$

110: $2^{14}/f_{PSC}$

111: $2^{15}/f_{PSC}$

• **TB1C Register**

Bit	7	6	5	4	3	2	1	0
Name	TB1ON	—	—	—	—	TB12	TB11	TB10
R/W	R/W	—	—	—	—	R/W	R/W	R/W
POR	0	—	—	—	—	0	0	0

Bit 7 **TB1ON**: Time Base 1 Control

0: Disable

1: Enable

Bit 6~3 Unimplemented, read as “0”

Bit 2~0	TB12~TB10: Select Time Base 1 Time-out Period
	000: $2^8/f_{PSC}$
	001: $2^9/f_{PSC}$
	010: $2^{10}/f_{PSC}$
	011: $2^{11}/f_{PSC}$
	100: $2^{12}/f_{PSC}$
	101: $2^{13}/f_{PSC}$
	110: $2^{14}/f_{PSC}$
	111: $2^{15}/f_{PSC}$

Interrupt Wake-up Function

Each of the interrupt functions has the capability of waking up the microcontroller when in the SLEEP or IDLE mode. A wake-up is generated when an interrupt request flag changes from low to high and is independent of whether the interrupt is enabled or not. Therefore, even though the device is in the SLEEP or IDLE mode and its system oscillator stopped, situations such as external edge transitions on the external interrupt pins or a low power supply voltage may cause their respective interrupt flag to be set high and consequently generate an interrupt. Care must therefore be taken if spurious wake-up situations are to be avoided. If an interrupt wake-up function is to be disabled then the corresponding interrupt request flag should be set high before the device enters the SLEEP or IDLE mode. The interrupt enable bits have no effect on the interrupt wake-up function.

Programming Considerations

By disabling the relevant interrupt enable bits, a requested interrupt can be prevented from being serviced, however, once an interrupt request flag is set, it will remain in this condition in the interrupt register until the corresponding interrupt is serviced or until the request flag is cleared by the application program.

It is recommended that programs do not use the “CALL” instruction within the interrupt service subroutine. Interrupts often occur in an unpredictable manner or need to be serviced immediately. If only one stack is left and the interrupt is not well controlled, the original control sequence will be damaged once a CALL subroutine is executed in the interrupt subroutine.

Every interrupt has the capability of waking up the microcontroller when it is in the SLEEP or IDLE mode, the wake-up being generated when the interrupt request flag changes from low to high. If it is required to prevent a certain interrupt from waking up the microcontroller then its respective request flag should be first set high before enter SLEEP or IDLE mode.

As only the Program Counter is pushed onto the stack, then when the interrupt is serviced, if the contents of the accumulator, status register or other registers are altered by the interrupt service program, their contents should be saved to the memory at the beginning of the interrupt service routine.

To return from an interrupt subroutine, either a RET or RETI instruction may be executed. The RETI instruction in addition to executing a return to the main program also automatically sets the EMI bit high to allow further interrupts. The RET instruction however only executes a return to the main program leaving the EMI bit in its present zero state and therefore disabling the execution of further interrupts.

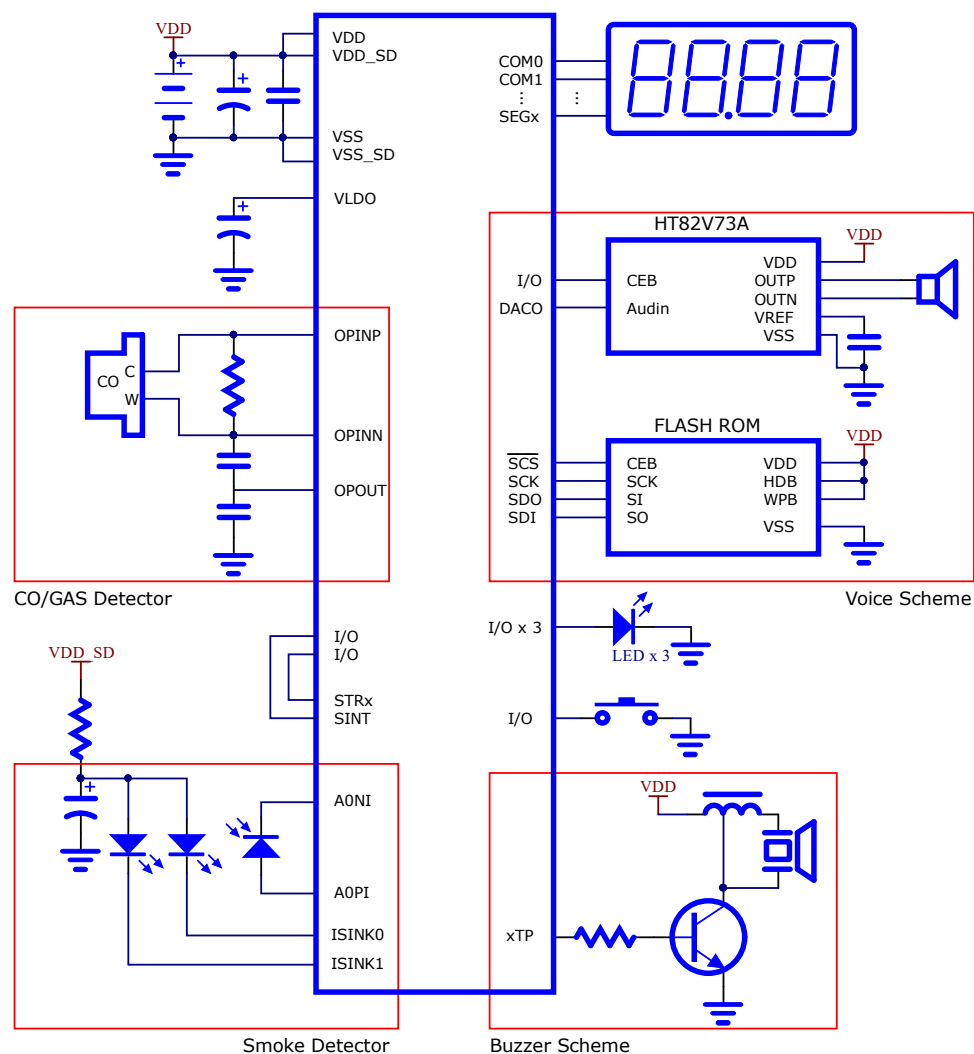
Configuration Options

Configuration options refer to certain options within the MCU that are programmed into the device during the programming process. During the development process, these options are selected using the HT-IDE software development tools. All options must be defined for proper system function, the details of which are shown in the table.

No.	Option
Oscillator Option	
1	HIRC frequency selection – f_{HIRC} : 2MHz, 4MHz or 8MHz

Note: When the HIRC has been configured at a frequency shown in this table, the HIRC1 and HIRC0 bits should also be setup to select the same frequency to achieve the HIRC frequency accuracy specified in the A.C. Characteristics.

Application Circuits



Instruction Set

Introduction

Central to the successful operation of any microcontroller is its instruction set, which is a set of program instruction codes that directs the microcontroller to perform certain operations. In the case of Holtek microcontroller, a comprehensive and flexible set of over 60 instructions is provided to enable programmers to implement their application with the minimum of programming overheads.

For easier understanding of the various instruction codes, they have been subdivided into several functional groupings.

Instruction Timing

Most instructions are implemented within one instruction cycle. The exceptions to this are branch, call, or table read instructions where two instruction cycles are required. One instruction cycle is equal to 4 system clock cycles, therefore in the case of an 8MHz system oscillator, most instructions would be implemented within 0.5 μ s and branch or call instructions would be implemented within 1 μ s. Although instructions which require one more cycle to implement are generally limited to the JMP, CALL, RET, RETI and table read instructions, it is important to realize that any other instructions which involve manipulation of the Program Counter Low register or PCL will also take one more cycle to implement. As instructions which change the contents of the PCL will imply a direct jump to that new address, one more cycle will be required. Examples of such instructions would be "CLR PCL" or "MOV PCL, A". For the case of skip instructions, it must be noted that if the result of the comparison involves a skip operation then this will also take one more cycle, if no skip is involved then only one cycle is required.

Moving and Transferring Data

The transfer of data within the microcontroller program is one of the most frequently used operations. Making use of several kinds of MOV instructions, data can be transferred from registers to the Accumulator and vice-versa as well as being able to move specific immediate data directly into the Accumulator. One of the most important data transfer applications is to receive data from the input ports and transfer data to the output ports.

Arithmetic Operations

The ability to perform certain arithmetic operations and data manipulation is a necessary feature of most microcontroller applications. Within the Holtek microcontroller instruction set are a range of add and subtract instruction mnemonics to enable the necessary arithmetic to be carried out. Care must be taken to ensure correct handling of carry and borrow data when results exceed 255 for addition and less than 0 for subtraction. The increment and decrement instructions such as INC, INCA, DEC and DECA provide a simple means of increasing or decreasing by a value of one of the values in the destination specified.

Logical and Rotate Operation

The standard logical operations such as AND, OR, XOR and CPL all have their own instruction within the Holtek microcontroller instruction set. As with the case of most instructions involving data manipulation, data must pass through the Accumulator which may involve additional programming steps. In all logical data operations, the zero flag may be set if the result of the operation is zero. Another form of logical data manipulation comes from the rotate instructions such as RR, RL, RRC and RLC which provide a simple means of rotating one bit right or left. Different rotate instructions exist depending on program requirements. Rotate instructions are useful for serial port programming applications where data can be rotated from an internal register into the Carry bit from where it can be examined and the necessary serial bit set high or low. Another application which rotate data operations are used is to implement multiplication and division calculations.

Branches and Control Transfer

Program branching takes the form of either jumps to specified locations using the JMP instruction or to a subroutine using the CALL instruction. They differ in the sense that in the case of a subroutine call, the program must return to the instruction immediately when the subroutine has been carried out. This is done by placing a return instruction “RET” in the subroutine which will cause the program to jump back to the address right after the CALL instruction. In the case of a JMP instruction, the program simply jumps to the desired location. There is no requirement to jump back to the original jumping off point as in the case of the CALL instruction. One special and extremely useful set of branch instructions are the conditional branches. Here a decision is first made regarding the condition of a certain data memory or individual bits. Depending upon the conditions, the program will continue with the next instruction or skip over it and jump to the following instruction. These instructions are the key to decision making and branching within the program perhaps determined by the condition of certain input switches or by the condition of internal data bits.

Bit Operations

The ability to provide single bit operations on Data Memory is an extremely flexible feature of all Holtek microcontrollers. This feature is especially useful for output port bit programming where individual bits or port pins can be directly set high or low using either the “SET [m].i” or “CLR [m].i” instructions respectively. The feature removes the need for programmers to first read the 8-bit output port, manipulate the input data to ensure that other bits are not changed and then output the port with the correct new data. This read-modify-write process is taken care of automatically when these bit operation instructions are used.

Table Read Operations

Data storage is normally implemented by using registers. However, when working with large amounts of fixed data, the volume involved often makes it inconvenient to store the fixed data in the Data Memory. To overcome this problem, Holtek microcontrollers allow an area of Program Memory to be setup as a table where data can be directly stored. A set of easy to use instructions provides the means by which this fixed data can be referenced and retrieved from the Program Memory.

Other Operations

In addition to the above functional instructions, a range of other instructions also exist such as the “HALT” instruction for Power-down operations and instructions to control the operation of the Watchdog Timer for reliable program operations under extreme electric or electromagnetic environments. For their relevant operations, refer to the functional related sections.

Instruction Set Summary

The instructions related to the data memory access in the following table can be used when the desired data memory is located in Data Memory sector 0.

Table Conventions

x: Bits immediate data
m: Data Memory address
A: Accumulator
i: 0~7 number of bits
addr: Program memory address

Mnemonic	Description	Cycles	Flag Affected
Arithmetic			
ADD A,[m]	Add Data Memory to ACC	1	Z, C, AC, OV, SC
ADDM A,[m]	Add ACC to Data Memory	↑Note	Z, C, AC, OV, SC
ADD A,x	Add immediate data to ACC	1	Z, C, AC, OV, SC
ADC A,[m]	Add Data Memory to ACC with Carry	1	Z, C, AC, OV, SC
ADCM A,[m]	Add ACC to Data memory with Carry	↑Note	Z, C, AC, OV, SC
SUB A,x	Subtract immediate data from the ACC	1	Z, C, AC, OV, SC, CZ
SUB A,[m]	Subtract Data Memory from ACC	1	Z, C, AC, OV, SC, CZ
SUBM A,[m]	Subtract Data Memory from ACC with result in Data Memory	↑Note	Z, C, AC, OV, SC, CZ
SBC A,x	Subtract immediate data from ACC with Carry	1	Z, C, AC, OV, SC, CZ
SBC A,[m]	Subtract Data Memory from ACC with Carry	1	Z, C, AC, OV, SC, CZ
SBCM A,[m]	Subtract Data Memory from ACC with Carry, result in Data Memory	↑Note	Z, C, AC, OV, SC, CZ
DAA [m]	Decimal adjust ACC for Addition with result in Data Memory	↑Note	C
Logic Operation			
AND A,[m]	Logical AND Data Memory to ACC	1	Z
OR A,[m]	Logical OR Data Memory to ACC	1	Z
XOR A,[m]	Logical XOR Data Memory to ACC	1	Z
ANDM A,[m]	Logical AND ACC to Data Memory	↑Note	Z
ORM A,[m]	Logical OR ACC to Data Memory	↑Note	Z
XORM A,[m]	Logical XOR ACC to Data Memory	↑Note	Z
AND A,x	Logical AND immediate Data to ACC	1	Z
OR A,x	Logical OR immediate Data to ACC	1	Z
XOR A,x	Logical XOR immediate Data to ACC	1	Z
CPL [m]	Complement Data Memory	↑Note	Z
CPLA [m]	Complement Data Memory with result in ACC	1	Z
Increment & Decrement			
INCA [m]	Increment Data Memory with result in ACC	1	Z
INC [m]	Increment Data Memory	↑Note	Z
DECA [m]	Decrement Data Memory with result in ACC	1	Z
DEC [m]	Decrement Data Memory	↑Note	Z
Rotate			
RRA [m]	Rotate Data Memory right with result in ACC	1	None
RR [m]	Rotate Data Memory right	↑Note	None
RRCA [m]	Rotate Data Memory right through Carry with result in ACC	1	C
RRC [m]	Rotate Data Memory right through Carry	↑Note	C
RLA [m]	Rotate Data Memory left with result in ACC	1	None
RL [m]	Rotate Data Memory left	↑Note	None
RLCA [m]	Rotate Data Memory left through Carry with result in ACC	1	C
RLC [m]	Rotate Data Memory left through Carry	↑Note	C

Mnemonic	Description	Cycles	Flag Affected
Data Move			
MOV A,[m]	Move Data Memory to ACC	1	None
MOV [m],A	Move ACC to Data Memory	1 ^{Note}	None
MOV A,x	Move immediate data to ACC	1	None
Bit Operation			
CLR [m].i	Clear bit of Data Memory	1 ^{Note}	None
SET [m].i	Set bit of Data Memory	1 ^{Note}	None
Branch Operation			
JMP addr	Jump unconditionally	2	None
SZ [m]	Skip if Data Memory is zero	1 ^{Note}	None
SZA [m]	Skip if Data Memory is zero with data movement to ACC	1 ^{Note}	None
SZ [m].i	Skip if bit i of Data Memory is zero	1 ^{Note}	None
SNZ [m]	Skip if Data Memory is not zero	1 ^{Note}	None
SNZ [m].i	Skip if bit i of Data Memory is not zero	1 ^{Note}	None
SIZ [m]	Skip if increment Data Memory is zero	1 ^{Note}	None
SDZ [m]	Skip if decrement Data Memory is zero	1 ^{Note}	None
SIZA [m]	Skip if increment Data Memory is zero with result in ACC	1 ^{Note}	None
SDZA [m]	Skip if decrement Data Memory is zero with result in ACC	1 ^{Note}	None
CALL addr	Subroutine call	2	None
RET	Return from subroutine	2	None
RET A,x	Return from subroutine and load immediate data to ACC	2	None
RETI	Return from interrupt	2	None
Table Read Operation			
TABRD [m]	Read table (specific page) to TBLH and Data Memory	2 ^{Note}	None
TABRDL [m]	Read table (last page) to TBLH and Data Memory	2 ^{Note}	None
ITABRD [m]	Increment table pointer TBLP first and Read table (specific page) to TBLH and Data Memory	2 ^{Note}	None
ITABRDL [m]	Increment table pointer TBLP first and Read table (last page) to TBLH and Data Memory	2 ^{Note}	None
Miscellaneous			
NOP	No operation	1	None
CLR [m]	Clear Data Memory	1 ^{Note}	None
SET [m]	Set Data Memory	1 ^{Note}	None
CLR WDT	Clear Watchdog Timer	1	TO, PDF
SWAP [m]	Swap nibbles of Data Memory	1 ^{Note}	None
SWAPA [m]	Swap nibbles of Data Memory with result in ACC	1	None
HALT	Enter power down mode	1	TO, PDF

Note: 1. For skip instructions, if the result of the comparison involves a skip then two cycles are required, if no skip takes place only one cycle is required.

2. Any instruction which changes the contents of the PCL will also require 2 cycles for execution.

Extended Instruction Set

The extended instructions are used to support the full range address access for the data memory. When the accessed data memory is located in any data memory sector except sector 0, the extended instruction can be used to directly access the data memory instead of using the indirect addressing access. This can not only reduce the use of Flash memory space but also improve the CPU execution efficiency.

Mnemonic	Description	Cycles	Flag Affected
Arithmetic			
LADD A,[m]	Add Data Memory to ACC	2	Z, C, AC, OV, SC
LADDM A,[m]	Add ACC to Data Memory	2 ^{Note}	Z, C, AC, OV, SC
LADC A,[m]	Add Data Memory to ACC with Carry	2	Z, C, AC, OV, SC
LADCM A,[m]	Add ACC to Data memory with Carry	2 ^{Note}	Z, C, AC, OV, SC
LSUB A,[m]	Subtract Data Memory from ACC	2	Z, C, AC, OV, SC, CZ
LSUBM A,[m]	Subtract Data Memory from ACC with result in Data Memory	2 ^{Note}	Z, C, AC, OV, SC, CZ
LSBC A,[m]	Subtract Data Memory from ACC with Carry	2	Z, C, AC, OV, SC, CZ
LSBCM A,[m]	Subtract Data Memory from ACC with Carry, result in Data Memory	2 ^{Note}	Z, C, AC, OV, SC, CZ
LDAA [m]	Decimal adjust ACC for Addition with result in Data Memory	2 ^{Note}	C
Logic Operation			
LAND A,[m]	Logical AND Data Memory to ACC	2	Z
LOR A,[m]	Logical OR Data Memory to ACC	2	Z
LXOR A,[m]	Logical XOR Data Memory to ACC	2	Z
LANDM A,[m]	Logical AND ACC to Data Memory	2 ^{Note}	Z
LORM A,[m]	Logical OR ACC to Data Memory	2 ^{Note}	Z
LXORM A,[m]	Logical XOR ACC to Data Memory	2 ^{Note}	Z
LCPL [m]	Complement Data Memory	2 ^{Note}	Z
LCPLA [m]	Complement Data Memory with result in ACC	2	Z
Increment & Decrement			
LINCA [m]	Increment Data Memory with result in ACC	2	Z
LINC [m]	Increment Data Memory	2 ^{Note}	Z
LDECA [m]	Decrement Data Memory with result in ACC	2	Z
LDEC [m]	Decrement Data Memory	2 ^{Note}	Z
Rotate			
LRRA [m]	Rotate Data Memory right with result in ACC	2	None
LRR [m]	Rotate Data Memory right	2 ^{Note}	None
LRRCA [m]	Rotate Data Memory right through Carry with result in ACC	2	C
LRRC [m]	Rotate Data Memory right through Carry	2 ^{Note}	C
LRLA [m]	Rotate Data Memory left with result in ACC	2	None
LRL [m]	Rotate Data Memory left	2 ^{Note}	None
LRLCA [m]	Rotate Data Memory left through Carry with result in ACC	2	C
LRLC [m]	Rotate Data Memory left through Carry	2 ^{Note}	C
Data Move			
LMOV A,[m]	Move Data Memory to ACC	2	None
LMOV [m],A	Move ACC to Data Memory	2 ^{Note}	None
Bit Operation			
LCLR [m].i	Clear bit of Data Memory	2 ^{Note}	None
LSET [m].i	Set bit of Data Memory	2 ^{Note}	None

Mnemonic	Description	Cycles	Flag Affected
Branch			
LSZ [m]	Skip if Data Memory is zero	2 ^{Note}	None
LSZA [m]	Skip if Data Memory is zero with data movement to ACC	2 ^{Note}	None
LSNZ [m]	Skip if Data Memory is not zero	2 ^{Note}	None
LSZ [m].i	Skip if bit i of Data Memory is zero	2 ^{Note}	None
LSNZ [m].i	Skip if bit i of Data Memory is not zero	2 ^{Note}	None
LSIZ [m]	Skip if increment Data Memory is zero	2 ^{Note}	None
LSIZ [m]	Skip if decrement Data Memory is zero	2 ^{Note}	None
LSIZA [m]	Skip if increment Data Memory is zero with result in ACC	2 ^{Note}	None
LSIZA [m]	Skip if decrement Data Memory is zero with result in ACC	2 ^{Note}	None
Table Read			
LTABRD [m]	Read table (specific page) to TBLH and Data Memory	3 ^{Note}	None
LTABRDL [m]	Read table (last page) to TBLH and Data Memory	3 ^{Note}	None
LITABRD [m]	Increment table pointer TBLP first and Read table (specific page) to TBLH and Data Memory	3 ^{Note}	None
LITABRDL [m]	Increment table pointer TBLP first and Read table (last page) to TBLH and Data Memory	3 ^{Note}	None
Miscellaneous			
LCLR [m]	Clear Data Memory	2 ^{Note}	None
LSET [m]	Set Data Memory	2 ^{Note}	None
LSWAP [m]	Swap nibbles of Data Memory	2 ^{Note}	None
LSWAPA [m]	Swap nibbles of Data Memory with result in ACC	2	None

Note: 1. For these extended skip instructions, if the result of the comparison involves a skip then three cycles are required, if no skip takes place two cycles is required.

2. Any extended instruction which changes the contents of the PCL register will also require three cycles for execution.

Instruction Definition

ADC A,[m]	Add Data Memory to ACC with Carry
Description	The contents of the specified Data Memory, Accumulator and the carry flag are added. The result is stored in the Accumulator.
Operation	$ACC \leftarrow ACC + [m] + C$
Affected flag(s)	OV, Z, AC, C, SC
ADCM A,[m]	Add ACC to Data Memory with Carry
Description	The contents of the specified Data Memory, Accumulator and the carry flag are added. The result is stored in the specified Data Memory.
Operation	$[m] \leftarrow ACC + [m] + C$
Affected flag(s)	OV, Z, AC, C, SC
ADD A,[m]	Add Data Memory to ACC
Description	The contents of the specified Data Memory and the Accumulator are added. The result is stored in the Accumulator.
Operation	$ACC \leftarrow ACC + [m]$
Affected flag(s)	OV, Z, AC, C, SC
ADD A,x	Add immediate data to ACC
Description	The contents of the Accumulator and the specified immediate data are added. The result is stored in the Accumulator.
Operation	$ACC \leftarrow ACC + x$
Affected flag(s)	OV, Z, AC, C, SC
ADDM A,[m]	Add ACC to Data Memory
Description	The contents of the specified Data Memory and the Accumulator are added. The result is stored in the specified Data Memory.
Operation	$[m] \leftarrow ACC + [m]$
Affected flag(s)	OV, Z, AC, C, SC
AND A,[m]	Logical AND Data Memory to ACC
Description	Data in the Accumulator and the specified Data Memory perform a bitwise logical AND operation. The result is stored in the Accumulator.
Operation	$ACC \leftarrow ACC \text{ "AND" } [m]$
Affected flag(s)	Z
AND A,x	Logical AND immediate data to ACC
Description	Data in the Accumulator and the specified immediate data perform a bit wise logical AND operation. The result is stored in the Accumulator.
Operation	$ACC \leftarrow ACC \text{ "AND" } x$
Affected flag(s)	Z
ANDM A,[m]	Logical AND ACC to Data Memory
Description	Data in the specified Data Memory and the Accumulator perform a bitwise logical AND operation. The result is stored in the Data Memory.
Operation	$[m] \leftarrow ACC \text{ "AND" } [m]$
Affected flag(s)	Z

CALL addr	Subroutine call
Description	Unconditionally calls a subroutine at the specified address. The Program Counter then increments by 1 to obtain the address of the next instruction which is then pushed onto the stack. The specified address is then loaded and the program continues execution from this new address. As this instruction requires an additional operation, it is a two cycle instruction.
Operation	Stack $\leftarrow$ Program Counter + 1 Program Counter $\leftarrow$ addr
Affected flag(s)	None
CLR [m]	Clear Data Memory
Description	Each bit of the specified Data Memory is cleared to 0.
Operation	[m] $\leftarrow$ 00H
Affected flag(s)	None
CLR [m].i	Clear bit of Data Memory
Description	Bit i of the specified Data Memory is cleared to 0.
Operation	[m].i $\leftarrow$ 0
Affected flag(s)	None
CLR WDT	Clear Watchdog Timer
Description	The TO, PDF flags and the WDT are all cleared.
Operation	WDT cleared TO $\leftarrow$ 0 PDF $\leftarrow$ 0
Affected flag(s)	TO, PDF
CPL [m]	Complement Data Memory
Description	Each bit of the specified Data Memory is logically complemented (1's complement). Bits which previously contained a 1 are changed to 0 and vice versa.
Operation	[m] $\leftarrow$ $\overline{[m]}$
Affected flag(s)	Z
CPLA [m]	Complement Data Memory with result in ACC
Description	Each bit of the specified Data Memory is logically complemented (1's complement). Bits which previously contained a 1 are changed to 0 and vice versa. The complemented result is stored in the Accumulator and the contents of the Data Memory remain unchanged.
Operation	ACC $\leftarrow$ $\overline{[m]}$
Affected flag(s)	Z
DAA [m]	Decimal-Adjust ACC for addition with result in Data Memory
Description	Convert the contents of the Accumulator value to a BCD (Binary Coded Decimal) value resulting from the previous addition of two BCD variables. If the low nibble is greater than 9 or if AC flag is set, then a value of 6 will be added to the low nibble. Otherwise the low nibble remains unchanged. If the high nibble is greater than 9 or if the C flag is set, then a value of 6 will be added to the high nibble. Essentially, the decimal conversion is performed by adding 00H, 06H, 60H or 66H depending on the Accumulator and flag conditions. Only the C flag may be affected by this instruction which indicates that if the original BCD sum is greater than 100, it allows multiple precision decimal addition.
Operation	[m] $\leftarrow$ ACC + 00H or [m] $\leftarrow$ ACC + 06H or [m] $\leftarrow$ ACC + 60H or [m] $\leftarrow$ ACC + 66H
Affected flag(s)	C

DEC [m]	Decrement Data Memory
Description	Data in the specified Data Memory is decremented by 1.
Operation	$[m] \leftarrow [m] - 1$
Affected flag(s)	Z
DECA [m]	Decrement Data Memory with result in ACC
Description	Data in the specified Data Memory is decremented by 1. The result is stored in the Accumulator. The contents of the Data Memory remain unchanged.
Operation	$ACC \leftarrow [m] - 1$
Affected flag(s)	Z
HALT	Enter power down mode
Description	This instruction stops the program execution and turns off the system clock. The contents of the Data Memory and registers are retained. The WDT and prescaler are cleared. The power down flag PDF is set and the WDT time-out flag TO is cleared.
Operation	$TO \leftarrow 0$ $PDF \leftarrow 1$
Affected flag(s)	TO, PDF
INC [m]	Increment Data Memory
Description	Data in the specified Data Memory is incremented by 1.
Operation	$[m] \leftarrow [m] + 1$
Affected flag(s)	Z
INCA [m]	Increment Data Memory with result in ACC
Description	Data in the specified Data Memory is incremented by 1. The result is stored in the Accumulator. The contents of the Data Memory remain unchanged.
Operation	$ACC \leftarrow [m] + 1$
Affected flag(s)	Z
JMP addr	Jump unconditionally
Description	The contents of the Program Counter are replaced with the specified address. Program execution then continues from this new address. As this requires the insertion of a dummy instruction while the new address is loaded, it is a two cycle instruction.
Operation	Program Counter $\leftarrow$ addr
Affected flag(s)	None
MOV A,[m]	Move Data Memory to ACC
Description	The contents of the specified Data Memory are copied to the Accumulator.
Operation	$ACC \leftarrow [m]$
Affected flag(s)	None
MOV A,x	Move immediate data to ACC
Description	The immediate data specified is loaded into the Accumulator.
Operation	$ACC \leftarrow x$
Affected flag(s)	None
MOV [m],A	Move ACC to Data Memory
Description	The contents of the Accumulator are copied to the specified Data Memory.
Operation	$[m] \leftarrow ACC$
Affected flag(s)	None

NOP	No operation
Description	No operation is performed. Execution continues with the next instruction.
Operation	No operation
Affected flag(s)	None
OR A,[m]	Logical OR Data Memory to ACC
Description	Data in the Accumulator and the specified Data Memory perform a bitwise logical OR operation. The result is stored in the Accumulator.
Operation	$ACC \leftarrow ACC \text{ "OR" } [m]$
Affected flag(s)	Z
OR A,x	Logical OR immediate data to ACC
Description	Data in the Accumulator and the specified immediate data perform a bitwise logical OR operation. The result is stored in the Accumulator.
Operation	$ACC \leftarrow ACC \text{ "OR" } x$
Affected flag(s)	Z
ORM A,[m]	Logical OR ACC to Data Memory
Description	Data in the specified Data Memory and the Accumulator perform a bitwise logical OR operation. The result is stored in the Data Memory.
Operation	$[m] \leftarrow ACC \text{ "OR" } [m]$
Affected flag(s)	Z
RET	Return from subroutine
Description	The Program Counter is restored from the stack. Program execution continues at the restored address.
Operation	$\text{Program Counter} \leftarrow \text{Stack}$
Affected flag(s)	None
RET A,x	Return from subroutine and load immediate data to ACC
Description	The Program Counter is restored from the stack and the Accumulator loaded with the specified immediate data. Program execution continues at the restored address.
Operation	$\text{Program Counter} \leftarrow \text{Stack}$ $ACC \leftarrow x$
Affected flag(s)	None
RETI	Return from interrupt
Description	The Program Counter is restored from the stack and the interrupts are re-enabled by setting the EMI bit. EMI is the master interrupt global enable bit. If an interrupt was pending when the RETI instruction is executed, the pending Interrupt routine will be processed before returning to the main program.
Operation	$\text{Program Counter} \leftarrow \text{Stack}$ $EMI \leftarrow 1$
Affected flag(s)	None
RL [m]	Rotate Data Memory left
Description	The contents of the specified Data Memory are rotated left by 1 bit with bit 7 rotated into bit 0.
Operation	$[m].(i+1) \leftarrow [m].i; (i=0\sim 6)$ $[m].0 \leftarrow [m].7$
Affected flag(s)	None

RLA [m]	Rotate Data Memory left with result in ACC
Description	The contents of the specified Data Memory are rotated left by 1 bit with bit 7 rotated into bit 0. The rotated result is stored in the Accumulator and the contents of the Data Memory remain unchanged.
Operation	$ACC.(i+1) \leftarrow [m].i; (i=0\sim6)$ $ACC.0 \leftarrow [m].7$
Affected flag(s)	None
RLC [m]	Rotate Data Memory left through Carry
Description	The contents of the specified Data Memory and the carry flag are rotated left by 1 bit. Bit 7 replaces the Carry bit and the original carry flag is rotated into bit 0.
Operation	$[m].(i+1) \leftarrow [m].i; (i=0\sim6)$ $[m].0 \leftarrow C$ $C \leftarrow [m].7$
Affected flag(s)	C
RLCA [m]	Rotate Data Memory left through Carry with result in ACC
Description	Data in the specified Data Memory and the carry flag are rotated left by 1 bit. Bit 7 replaces the Carry bit and the original carry flag is rotated into the bit 0. The rotated result is stored in the Accumulator and the contents of the Data Memory remain unchanged.
Operation	$ACC.(i+1) \leftarrow [m].i; (i=0\sim6)$ $ACC.0 \leftarrow C$ $C \leftarrow [m].7$
Affected flag(s)	C
RR [m]	Rotate Data Memory right
Description	The contents of the specified Data Memory are rotated right by 1 bit with bit 0 rotated into bit 7.
Operation	$[m].i \leftarrow [m].(i+1); (i=0\sim6)$ $[m].7 \leftarrow [m].0$
Affected flag(s)	None
RRA [m]	Rotate Data Memory right with result in ACC
Description	Data in the specified Data Memory is rotated right by 1 bit with bit 0 rotated into bit 7. The rotated result is stored in the Accumulator and the contents of the Data Memory remain unchanged.
Operation	$ACC.i \leftarrow [m].(i+1); (i=0\sim6)$ $ACC.7 \leftarrow [m].0$
Affected flag(s)	None
RRC [m]	Rotate Data Memory right through Carry
Description	The contents of the specified Data Memory and the carry flag are rotated right by 1 bit. Bit 0 replaces the Carry bit and the original carry flag is rotated into bit 7.
Operation	$[m].i \leftarrow [m].(i+1); (i=0\sim6)$ $[m].7 \leftarrow C$ $C \leftarrow [m].0$
Affected flag(s)	C

RRCA [m]	Rotate Data Memory right through Carry with result in ACC
Description	Data in the specified Data Memory and the carry flag are rotated right by 1 bit. Bit 0 replaces the Carry bit and the original carry flag is rotated into bit 7. The rotated result is stored in the Accumulator and the contents of the Data Memory remain unchanged.
Operation	$ACC.i \leftarrow [m].(i+1); (i=0\sim6)$ $ACC.7 \leftarrow C$ $C \leftarrow [m].0$
Affected flag(s)	C
SBC A,[m]	Subtract Data Memory from ACC with Carry
Description	The contents of the specified Data Memory and the complement of the carry flag are subtracted from the Accumulator. The result is stored in the Accumulator. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.
Operation	$ACC \leftarrow ACC - [m] - \overline{C}$
Affected flag(s)	OV, Z, AC, C, SC, CZ
SBC A, x	Subtract immediate data from ACC with Carry
Description	The immediate data and the complement of the carry flag are subtracted from the Accumulator. The result is stored in the Accumulator. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.
Operation	$ACC \leftarrow ACC - [m] - \overline{C}$
Affected flag(s)	OV, Z, AC, C, SC, CZ
SBCM A,[m]	Subtract Data Memory from ACC with Carry and result in Data Memory
Description	The contents of the specified Data Memory and the complement of the carry flag are subtracted from the Accumulator. The result is stored in the Data Memory. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.
Operation	$[m] \leftarrow ACC - [m] - \overline{C}$
Affected flag(s)	OV, Z, AC, C, SC, CZ
SDZ [m]	Skip if decrement Data Memory is 0
Description	The contents of the specified Data Memory are first decremented by 1. If the result is 0 the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction.
Operation	$[m] \leftarrow [m] - 1$ Skip if $[m]=0$
Affected flag(s)	None
SDZA [m]	Skip if decrement Data Memory is zero with result in ACC
Description	The contents of the specified Data Memory are first decremented by 1. If the result is 0, the following instruction is skipped. The result is stored in the Accumulator but the specified Data Memory contents remain unchanged. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0, the program proceeds with the following instruction.
Operation	$ACC \leftarrow [m] - 1$ Skip if $ACC=0$
Affected flag(s)	None

SET [m]	Set Data Memory
Description	Each bit of the specified Data Memory is set to 1.
Operation	$[m] \leftarrow FFH$
Affected flag(s)	None
SET [m].i	Set bit of Data Memory
Description	Bit i of the specified Data Memory is set to 1.
Operation	$[m].i \leftarrow 1$
Affected flag(s)	None
SIZ [m]	Skip if increment Data Memory is 0
Description	The contents of the specified Data Memory are first incremented by 1. If the result is 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction.
Operation	$[m] \leftarrow [m] + 1$ Skip if $[m]=0$
Affected flag(s)	None
SIZA [m]	Skip if increment Data Memory is zero with result in ACC
Description	The contents of the specified Data Memory are first incremented by 1. If the result is 0, the following instruction is skipped. The result is stored in the Accumulator but the specified Data Memory contents remain unchanged. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction.
Operation	$ACC \leftarrow [m] + 1$ Skip if $ACC=0$
Affected flag(s)	None
SNZ [m].i	Skip if bit i of Data Memory is not 0
Description	If bit i of the specified Data Memory is not 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is 0 the program proceeds with the following instruction.
Operation	Skip if $[m].i \neq 0$
Affected flag(s)	None
SNZ [m]	Skip if Data Memory is not 0
Description	The contents of the specified Data Memory are read out and then written back to the specified Data Memory again. If the specified Data Memory is not 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is 0 the program proceeds with the following instruction.
Operation	Skip if $[m] \neq 0$
Affected flag(s)	None
SUB A,[m]	Subtract Data Memory from ACC
Description	The specified Data Memory is subtracted from the contents of the Accumulator. The result is stored in the Accumulator. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.
Operation	$ACC \leftarrow ACC - [m]$
Affected flag(s)	OV, Z, AC, C, SC, CZ

SUBM A,[m]	Subtract Data Memory from ACC with result in Data Memory
Description	The specified Data Memory is subtracted from the contents of the Accumulator. The result is stored in the Data Memory. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.
Operation	$[m] \leftarrow ACC - [m]$
Affected flag(s)	OV, Z, AC, C, SC, CZ
SUB A,x	Subtract immediate data from ACC
Description	The immediate data specified by the code is subtracted from the contents of the Accumulator. The result is stored in the Accumulator. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.
Operation	$ACC \leftarrow ACC - x$
Affected flag(s)	OV, Z, AC, C, SC, CZ
SWAP [m]	Swap nibbles of Data Memory
Description	The low-order and high-order nibbles of the specified Data Memory are interchanged.
Operation	$[m].3 \sim [m].0 \leftrightarrow [m].7 \sim [m].4$
Affected flag(s)	None
SWAPA [m]	Swap nibbles of Data Memory with result in ACC
Description	The low-order and high-order nibbles of the specified Data Memory are interchanged. The result is stored in the Accumulator. The contents of the Data Memory remain unchanged.
Operation	$ACC.3 \sim ACC.0 \leftarrow [m].7 \sim [m].4$ $ACC.7 \sim ACC.4 \leftarrow [m].3 \sim [m].0$
Affected flag(s)	None
SZ [m]	Skip if Data Memory is 0
Description	The contents of the specified Data Memory are read out and then written back to the specified Data Memory again. If the contents of the specified Data Memory is 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction.
Operation	Skip if $[m]=0$
Affected flag(s)	None
SZA [m]	Skip if Data Memory is 0 with data movement to ACC
Description	The contents of the specified Data Memory are copied to the Accumulator. If the value is zero, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction.
Operation	$ACC \leftarrow [m]$ Skip if $[m]=0$
Affected flag(s)	None
SZ [m].i	Skip if bit i of Data Memory is 0
Description	If bit i of the specified Data Memory is 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0, the program proceeds with the following instruction.
Operation	Skip if $[m].i=0$
Affected flag(s)	None

TABRD [m]	Read table (specific page) to TBLH and Data Memory
Description	The low byte of the program code (specific page) addressed by the table pointer (TBLP and TBHP) is moved to the specified Data Memory and the high byte moved to TBLH.
Operation	[m] ← program code (low byte) TBLH ← program code (high byte)
Affected flag(s)	None
TABRDL [m]	Read table (last page) to TBLH and Data Memory
Description	The low byte of the program code (last page) addressed by the table pointer (TBLP) is moved to the specified Data Memory and the high byte moved to TBLH.
Operation	[m] ← program code (low byte) TBLH ← program code (high byte)
Affected flag(s)	None
ITABRD [m]	Increment table pointer low byte first and read table (specific page) to TBLH and Data Memory
Description	Increment table pointer low byte, TBLP, first and then the program code (specific page) addressed by the table pointer (TBHP and TBLP) is moved to the specified Data Memory and the high byte moved to TBLH.
Operation	[m] ← program code (low byte) TBLH ← program code (high byte)
Affected flag(s)	None
ITABRDL [m]	Increment table pointer low byte first and read table (last page) to TBLH and Data Memory
Description	Increment table pointer low byte, TBLP, first and then the low byte of the program code (last page) addressed by the table pointer (TBLP) is moved to the specified Data Memory and the high byte moved to TBLH.
Operation	[m] ← program code (low byte) TBLH ← program code (high byte)
Affected flag(s)	None
XOR A,[m]	Logical XOR Data Memory to ACC
Description	Data in the Accumulator and the specified Data Memory perform a bitwise logical XOR operation. The result is stored in the Accumulator.
Operation	ACC ← ACC "XOR" [m]
Affected flag(s)	Z
XORM A,[m]	Logical XOR ACC to Data Memory
Description	Data in the specified Data Memory and the Accumulator perform a bitwise logical XOR operation. The result is stored in the Data Memory.
Operation	[m] ← ACC "XOR" [m]
Affected flag(s)	Z
XOR A,x	Logical XOR immediate data to ACC
Description	Data in the Accumulator and the specified immediate data perform a bitwise logical XOR operation. The result is stored in the Accumulator.
Operation	ACC ← ACC "XOR" x
Affected flag(s)	Z

Extended Instruction Definition

The extended instructions are used to directly access the data stored in any data memory sections.

LADC A,[m]	Add Data Memory to ACC with Carry
Description	The contents of the specified Data Memory, Accumulator and the carry flag are added. The result is stored in the Accumulator.
Operation	$ACC \leftarrow ACC + [m] + C$
Affected flag(s)	OV, Z, AC, C, SC
LADCM A,[m]	Add ACC to Data Memory with Carry
Description	The contents of the specified Data Memory, Accumulator and the carry flag are added. The result is stored in the specified Data Memory.
Operation	$[m] \leftarrow ACC + [m] + C$
Affected flag(s)	OV, Z, AC, C, SC
LADD A,[m]	Add Data Memory to ACC
Description	The contents of the specified Data Memory and the Accumulator are added. The result is stored in the Accumulator.
Operation	$ACC \leftarrow ACC + [m]$
Affected flag(s)	OV, Z, AC, C, SC
LADDM A,[m]	Add ACC to Data Memory
Description	The contents of the specified Data Memory and the Accumulator are added. The result is stored in the specified Data Memory.
Operation	$[m] \leftarrow ACC + [m]$
Affected flag(s)	OV, Z, AC, C, SC
LAND A,[m]	Logical AND Data Memory to ACC
Description	Data in the Accumulator and the specified Data Memory perform a bitwise logical AND operation. The result is stored in the Accumulator.
Operation	$ACC \leftarrow ACC \text{ "AND" } [m]$
Affected flag(s)	Z
LANDM A,[m]	Logical AND ACC to Data Memory
Description	Data in the specified Data Memory and the Accumulator perform a bitwise logical AND operation. The result is stored in the Data Memory.
Operation	$[m] \leftarrow ACC \text{ "AND" } [m]$
Affected flag(s)	Z
LCLR [m]	Clear Data Memory
Description	Each bit of the specified Data Memory is cleared to 0.
Operation	$[m] \leftarrow 00H$
Affected flag(s)	None
LCLR [m].i	Clear bit of Data Memory
Description	Bit i of the specified Data Memory is cleared to 0.
Operation	$[m].i \leftarrow 0$
Affected flag(s)	None

LCPL [m]	Complement Data Memory
Description	Each bit of the specified Data Memory is logically complemented (1's complement). Bits which previously contained a 1 are changed to 0 and vice versa.
Operation	$[m] \leftarrow \overline{[m]}$
Affected flag(s)	Z
LCPLA [m]	Complement Data Memory with result in ACC
Description	Each bit of the specified Data Memory is logically complemented (1's complement). Bits which previously contained a 1 are changed to 0 and vice versa. The complemented result is stored in the Accumulator and the contents of the Data Memory remain unchanged.
Operation	$ACC \leftarrow \overline{[m]}$
Affected flag(s)	Z
LDAA [m]	Decimal-Adjust ACC for addition with result in Data Memory
Description	Convert the contents of the Accumulator value to a BCD (Binary Coded Decimal) value resulting from the previous addition of two BCD variables. If the low nibble is greater than 9 or if AC flag is set, then a value of 6 will be added to the low nibble. Otherwise the low nibble remains unchanged. If the high nibble is greater than 9 or if the C flag is set, then a value of 6 will be added to the high nibble. Essentially, the decimal conversion is performed by adding 00H, 06H, 60H or 66H depending on the Accumulator and flag conditions. Only the C flag may be affected by this instruction which indicates that if the original BCD sum is greater than 100, it allows multiple precision decimal addition.
Operation	$[m] \leftarrow ACC + 00H$ or $[m] \leftarrow ACC + 06H$ or $[m] \leftarrow ACC + 60H$ or $[m] \leftarrow ACC + 66H$
Affected flag(s)	C
LDEC [m]	Decrement Data Memory
Description	Data in the specified Data Memory is decremented by 1.
Operation	$[m] \leftarrow [m] - 1$
Affected flag(s)	Z
LDECA [m]	Decrement Data Memory with result in ACC
Description	Data in the specified Data Memory is decremented by 1. The result is stored in the Accumulator. The contents of the Data Memory remain unchanged.
Operation	$ACC \leftarrow [m] - 1$
Affected flag(s)	Z
LINC [m]	Increment Data Memory
Description	Data in the specified Data Memory is incremented by 1.
Operation	$[m] \leftarrow [m] + 1$
Affected flag(s)	Z
LINCA [m]	Increment Data Memory with result in ACC
Description	Data in the specified Data Memory is incremented by 1. The result is stored in the Accumulator. The contents of the Data Memory remain unchanged.
Operation	$ACC \leftarrow [m] + 1$
Affected flag(s)	Z

LMOV A,[m]	Move Data Memory to ACC
Description	The contents of the specified Data Memory are copied to the Accumulator.
Operation	$ACC \leftarrow [m]$
Affected flag(s)	None
LMOV [m],A	Move ACC to Data Memory
Description	The contents of the Accumulator are copied to the specified Data Memory.
Operation	$[m] \leftarrow ACC$
Affected flag(s)	None
LOR A,[m]	Logical OR Data Memory to ACC
Description	Data in the Accumulator and the specified Data Memory perform a bitwise logical OR operation. The result is stored in the Accumulator.
Operation	$ACC \leftarrow ACC \text{ "OR" } [m]$
Affected flag(s)	Z
LORM A,[m]	Logical OR ACC to Data Memory
Description	Data in the specified Data Memory and the Accumulator perform a bitwise logical OR operation. The result is stored in the Data Memory.
Operation	$[m] \leftarrow ACC \text{ "OR" } [m]$
Affected flag(s)	Z
LRL [m]	Rotate Data Memory left
Description	The contents of the specified Data Memory are rotated left by 1 bit with bit 7 rotated into bit 0.
Operation	$[m].(i+1) \leftarrow [m].i; (i=0\sim6)$ $[m].0 \leftarrow [m].7$
Affected flag(s)	None
LRLA [m]	Rotate Data Memory left with result in ACC
Description	The contents of the specified Data Memory are rotated left by 1 bit with bit 7 rotated into bit 0. The rotated result is stored in the Accumulator and the contents of the Data Memory remain unchanged.
Operation	$ACC.(i+1) \leftarrow [m].i; (i=0\sim6)$ $ACC.0 \leftarrow [m].7$
Affected flag(s)	None
LRLC [m]	Rotate Data Memory left through Carry
Description	The contents of the specified Data Memory and the carry flag are rotated left by 1 bit. Bit 7 replaces the Carry bit and the original carry flag is rotated into bit 0.
Operation	$[m].(i+1) \leftarrow [m].i; (i=0\sim6)$ $[m].0 \leftarrow C$ $C \leftarrow [m].7$
Affected flag(s)	C
LRLCA [m]	Rotate Data Memory left through Carry with result in ACC
Description	Data in the specified Data Memory and the carry flag are rotated left by 1 bit. Bit 7 replaces the Carry bit and the original carry flag is rotated into the bit 0. The rotated result is stored in the Accumulator and the contents of the Data Memory remain unchanged.
Operation	$ACC.(i+1) \leftarrow [m].i; (i=0\sim6)$ $ACC.0 \leftarrow C$ $C \leftarrow [m].7$
Affected flag(s)	C

LRR [m]	Rotate Data Memory right
Description	The contents of the specified Data Memory are rotated right by 1 bit with bit 0 rotated into bit 7.
Operation	$[m].i \leftarrow [m].(i+1); (i=0\sim6)$ $[m].7 \leftarrow [m].0$
Affected flag(s)	None
LRRA [m]	Rotate Data Memory right with result in ACC
Description	Data in the specified Data Memory is rotated right by 1 bit with bit 0 rotated into bit 7. The rotated result is stored in the Accumulator and the contents of the Data Memory remain unchanged.
Operation	$ACC.i \leftarrow [m].(i+1); (i=0\sim6)$ $ACC.7 \leftarrow [m].0$
Affected flag(s)	None
LRRC [m]	Rotate Data Memory right through Carry
Description	The contents of the specified Data Memory and the carry flag are rotated right by 1 bit. Bit 0 replaces the Carry bit and the original carry flag is rotated into bit 7.
Operation	$[m].i \leftarrow [m].(i+1); (i=0\sim6)$ $[m].7 \leftarrow C$ $C \leftarrow [m].0$
Affected flag(s)	C
LRRCA [m]	Rotate Data Memory right through Carry with result in ACC
Description	Data in the specified Data Memory and the carry flag are rotated right by 1 bit. Bit 0 replaces the Carry bit and the original carry flag is rotated into bit 7. The rotated result is stored in the Accumulator and the contents of the Data Memory remain unchanged.
Operation	$ACC.i \leftarrow [m].(i+1); (i=0\sim6)$ $ACC.7 \leftarrow C$ $C \leftarrow [m].0$
Affected flag(s)	C
LSBC A,[m]	Subtract Data Memory from ACC with Carry
Description	The contents of the specified Data Memory and the complement of the carry flag are subtracted from the Accumulator. The result is stored in the Accumulator. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.
Operation	$ACC \leftarrow ACC - [m] - \overline{C}$
Affected flag(s)	OV, Z, AC, C, SC, CZ
LSBCM A,[m]	Subtract Data Memory from ACC with Carry and result in Data Memory
Description	The contents of the specified Data Memory and the complement of the carry flag are subtracted from the Accumulator. The result is stored in the Data Memory. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.
Operation	$[m] \leftarrow ACC - [m] - \overline{C}$
Affected flag(s)	OV, Z, AC, C, SC, CZ

LSDZ [m]	Skip if decrement Data Memory is 0
Description	The contents of the specified Data Memory are first decremented by 1. If the result is 0 the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a three cycle instruction. If the result is not 0 the program proceeds with the following instruction.
Operation	$[m] \leftarrow [m] - 1$ Skip if $[m]=0$
Affected flag(s)	None
LSDZA [m]	Skip if decrement Data Memory is zero with result in ACC
Description	The contents of the specified Data Memory are first decremented by 1. If the result is 0, the following instruction is skipped. The result is stored in the Accumulator but the specified Data Memory contents remain unchanged. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a three cycle instruction. If the result is not 0, the program proceeds with the following instruction.
Operation	$ACC \leftarrow [m] - 1$ Skip if $ACC=0$
Affected flag(s)	None
LSET [m]	Set Data Memory
Description	Each bit of the specified Data Memory is set to 1.
Operation	$[m] \leftarrow FFH$
Affected flag(s)	None
LSET [m].i	Set bit of Data Memory
Description	Bit i of the specified Data Memory is set to 1.
Operation	$[m].i \leftarrow 1$
Affected flag(s)	None
LSIZ [m]	Skip if increment Data Memory is 0
Description	The contents of the specified Data Memory are first incremented by 1. If the result is 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a three cycle instruction. If the result is not 0 the program proceeds with the following instruction.
Operation	$[m] \leftarrow [m] + 1$ Skip if $[m]=0$
Affected flag(s)	None
LSIZA [m]	Skip if increment Data Memory is zero with result in ACC
Description	The contents of the specified Data Memory are first incremented by 1. If the result is 0, the following instruction is skipped. The result is stored in the Accumulator but the specified Data Memory contents remain unchanged. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a three cycle instruction. If the result is not 0 the program proceeds with the following instruction.
Operation	$ACC \leftarrow [m] + 1$ Skip if $ACC=0$
Affected flag(s)	None
LSNZ [m].i	Skip if bit i of Data Memory is not 0
Description	If bit i of the specified Data Memory is not 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a three cycle instruction. If the result is 0 the program proceeds with the following instruction.
Operation	Skip if $[m].i \neq 0$
Affected flag(s)	None

LSNZ [m]	Skip if Data Memory is not 0
Description	The contents of the specified Data Memory are read out and then written to the specified Data Memory again. If the content of the specified Data Memory is not 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a three cycle instruction. If the result is 0 the program proceeds with the following instruction.
Operation	Skip if [m] $\neq$ 0
Affected flag(s)	None
LSUB A,[m]	Subtract Data Memory from ACC
Description	The specified Data Memory is subtracted from the contents of the Accumulator. The result is stored in the Accumulator. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.
Operation	$ACC \leftarrow ACC - [m]$
Affected flag(s)	OV, Z, AC, C, SC, CZ
LSUBM A,[m]	Subtract Data Memory from ACC with result in Data Memory
Description	The specified Data Memory is subtracted from the contents of the Accumulator. The result is stored in the Data Memory. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.
Operation	$[m] \leftarrow ACC - [m]$
Affected flag(s)	OV, Z, AC, C, SC, CZ
LSWAP [m]	Swap nibbles of Data Memory
Description	The low-order and high-order nibbles of the specified Data Memory are interchanged.
Operation	$[m].3 \sim [m].0 \leftrightarrow [m].7 \sim [m].4$
Affected flag(s)	None
LSWAPA [m]	Swap nibbles of Data Memory with result in ACC
Description	The low-order and high-order nibbles of the specified Data Memory are interchanged. The result is stored in the Accumulator. The contents of the Data Memory remain unchanged.
Operation	$ACC.3 \sim ACC.0 \leftarrow [m].7 \sim [m].4$ $ACC.7 \sim ACC.4 \leftarrow [m].3 \sim [m].0$
Affected flag(s)	None
LSZ [m]	Skip if Data Memory is 0
Description	The contents of the specified Data Memory are read out and then written to the specified Data Memory again. If the contents of the specified Data Memory is 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a three cycle instruction. If the result is not 0 the program proceeds with the following instruction.
Operation	Skip if [m]=0
Affected flag(s)	None
LSZA [m]	Skip if Data Memory is 0 with data movement to ACC
Description	The contents of the specified Data Memory are copied to the Accumulator. If the value is zero, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a three cycle instruction. If the result is not 0 the program proceeds with the following instruction.
Operation	$ACC \leftarrow [m]$ Skip if [m]=0
Affected flag(s)	None

LSZ [m].i	Skip if bit i of Data Memory is 0
Description	If bit i of the specified Data Memory is 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a three cycle instruction. If the result is not 0, the program proceeds with the following instruction.
Operation	Skip if [m].i=0
Affected flag(s)	None
LTABRD [m]	Read table (specific page) to TBLH and Data Memory
Description	The low byte of the program code (specific page) addressed by the table pointer (TBHP and TBLP) is moved to the specified Data Memory and the high byte moved to TBLH.
Operation	[m] ← program code (low byte) TBLH ← program code (high byte)
Affected flag(s)	None
LTABRDL [m]	Read table (last page) to TBLH and Data Memory
Description	The low byte of the program code (last page) addressed by the table pointer (TBLP) is moved to the specified Data Memory and the high byte moved to TBLH.
Operation	[m] ← program code (low byte) TBLH ← program code (high byte)
Affected flag(s)	None
LITABRD [m]	Increment table pointer low byte first and read table (specific page) to TBLH and Data Memory
Description	Increment table pointer low byte, TBLP, first and then the program code (specific page) addressed by the table pointer (TBHP and TBLP) is moved to the specified Data Memory and the high byte moved to TBLH.
Operation	[m] ← program code (low byte) TBLH ← program code (high byte)
Affected flag(s)	None
LITABRDL [m]	Increment table pointer low byte first and read table (last page) to TBLH and Data Memory
Description	Increment table pointer low byte, TBLP, first and then the low byte of the program code (last page) addressed by the table pointer (TBLP) is moved to the specified Data Memory and the high byte moved to TBLH.
Operation	[m] ← program code (low byte) TBLH ← program code (high byte)
Affected flag(s)	None
LXOR A,[m]	Logical XOR Data Memory to ACC
Description	Data in the Accumulator and the specified Data Memory perform a bitwise logical XOR operation. The result is stored in the Accumulator.
Operation	ACC ← ACC "XOR" [m]
Affected flag(s)	Z
LXORM A,[m]	Logical XOR ACC to Data Memory
Description	Data in the specified Data Memory and the Accumulator perform a bitwise logical XOR operation. The result is stored in the Data Memory.
Operation	[m] ← ACC "XOR" [m]
Affected flag(s)	Z

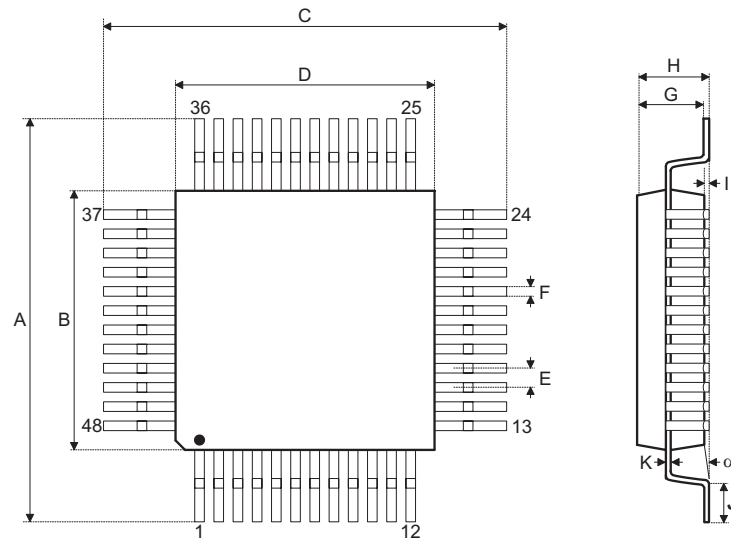
Package Information

Note that the package information provided here is for consultation purposes only. As this information may be updated at regular intervals users are reminded to consult the [Holtek website](#) for the latest version of the [Package/Carton Information](#).

Additional supplementary information with regard to packaging is listed below. Click on the relevant section to be transferred to the relevant website page.

- [Package Information \(include Outline Dimensions, Product Tape and Reel Specifications\)](#)
- [The Operation Instruction of Packing Materials](#)
- [Carton information](#)

48-pin LQFP (7mm×7mm) Outline Dimensions



Symbol	Dimensions in inch		
	Min.	Nom.	Max.
A	0.354 BSC		
B	0.276 BSC		
C	0.354 BSC		
D	0.276 BSC		
E	0.020 BSC		
F	0.007	0.009	0.011
G	0.053	0.055	0.057
H	—	—	0.063
I	0.002	—	0.006
J	0.018	0.024	0.030
K	0.004	—	0.008
α	0°	—	7°

Symbol	Dimensions in mm		
	Min.	Nom.	Max.
A	9.00 BSC		
B	7.00 BSC		
C	9.00 BSC		
D	7.00 BSC		
E	0.50 BSC		
F	0.17	0.22	0.27
G	1.35	1.40	1.45
H	—	—	1.60
I	0.05	—	0.15
J	0.45	0.60	0.75
K	0.09	—	0.20
α	0°	—	7°

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