

LDO Regulators with Watch-Dog Timer

500mA Output LDO Regulators with Voltage Detector and Watchdog Timer

BD3020HFP-M BD3021HFP-M

General Description

BD3020HFP-M BD3021HFP-M is a regulator IC with integrated WDT (Watch Dog Timer), high output voltage accuracy ± 2.0 % and 80 μA (Typ) low circuit current consumption. These are supports usage of low ESR ceramic capacitor for output stability. The reset detection voltage can be adjusted by connecting resistors on the Vs terminal (BD3020HFP-M). They can be a stable power supply for any applications while detecting malfunction of microcontrollers.

Features

- Integrated WDT Reset Circuit
 [BD3020HFP-M]: Adjustable Detection Voltage through Vs pin
 [BD3021HFP-M]: WDT Can be Switched ON / OFF by Using INH Pin
- Low saturation voltage by using PMOS output transistor
- VCC Max Voltage: 50 V
- Integrated Over Current Protection and Thermal Shut Down
- HRP7 package

Key specification

- Low Circuit Current: 80 µA (Typ)
 Output Voltage: 5.0 V (Typ)
 Output Current: 500 mA
 High Output Voltage Accuracy: ±2 %
- Low ESR ceramic capacitor can be used as output capacitor

Package W (Typ) × D (Typ) × H (Max) ■ HRP7 9.395 mm × 10.540 mm × 2.005 mm

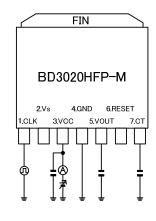


Figure 1. Package Outlook

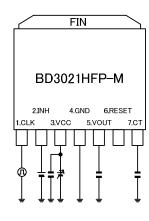
Applications

■ Automotive (body, audio system, navigation system, etc.)

Typical Application Circuits



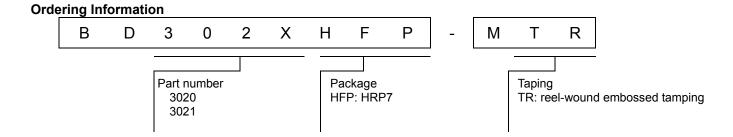
BD3020HFP-M



BD3021HFP-M

Figure 2. Typical Application Circuits

OProduct structure: Silicon monolithic integrated circuit OThis product is not designed protection against radioactive rays



Pin Configuration

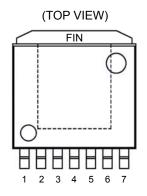


Figure 3. Pin Configuration

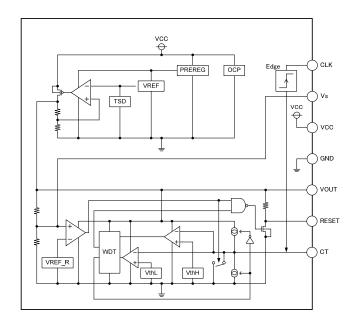
Pin Description

Pin No.	Pin Name	Function	
1	CLK	Clock Input from Microcontroller	
2	Vs (BD3020HFP-M)	Reset Detection Voltage Set Pin	
2	INH (BD3021HFP-M)	WDT ON/OFF Function Pin	
3	VCC	Power Supply Pin	
4	GND	GND	
5	VOUT	Voltage Output Pin	
6	RESET	Reset Output Pin	
7	СТ	External Capacitance for Reset Output Delay Time, WDT Monitor Time Setting Connection Pin	
Fin	GND	GND	

Block Diagram

<BD3020HFP-M>

<BD3021HFP-M>



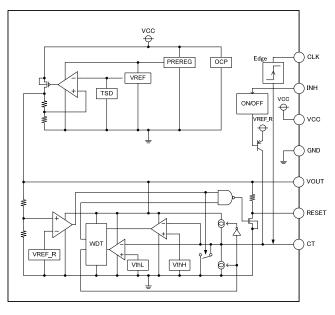


Figure 4. Block Diagrams

Absolute Maximum Ratings (Ta = 25 °C)

Parameter	Symbol	Ratings	Unit
Supply Voltage (1)	Vcc	-0.3 to +50.0	V
Vs pin Voltage (BD3020HFP-M)	Vs	-0.3 to +15.0	V
INH pin Voltage (BD3021HFP-M)	Vinh	-0.3 to +15.0	V
Regulator Output pin Voltage	V _{OUT}	-0.3 to +15.0	V
Reset Output pin Voltage	V _{RESET}	-0.3 to +15.0	V
Watchdog Input pin Voltage	Vclk	-0.3 to +15.0	V
Reset Delay Setting pin Voltage	Vст	-0.3 to +15.0	V
Power Dissipation (2)	Pd	1.6	W
Operating Temperature Range	Topr	-40 to +125	°C
Storage Temperature Range	Tstg	-55 to +150	°C
Maximum Junction Temperature	Tjmax	150	°C

⁽¹⁾ Pd should not be exceeded.

(1-layer PCB: Copper foil area on the reverse side of PCB: 0 mm × 0 mm)

Caution: Operating the IC over the absolute maximum ratings may damage the IC. The damage can either be a short circuit between pins or an open circuit between pins and the internal circuitry. Therefore, it is important to consider circuit protection measures, such as adding a fuse, in case the IC is operated over the absolute maximum ratings.

Operating Conditions (-40°C ≤ Ta ≤ +125 °C)

Parameter	Symbol	Min	Max	Unit
Supply Voltage (3)	Vcc	5.6	36.0	V
Output Current	lo	0	500	mA

⁽³⁾ For the output voltage, consider the voltage drop (dropout voltage) due to the output current.

⁽²⁾ HRP7 mounted on 70.0 mm × 70.0 mm × 1.6 mmt Glass-Epoxy PCB. If Ta ≥ 25 °C, reduce by 12.8 mW / °C.

Electrical Characteristics (Unless otherwise specified, -40°C ≤ Ta ≤ +125 °C, V_{CC} = 13.5 V, V_{CLK} = GND)

Devenuetor	Company of	Limit		l lmi4	O and this are		
Parameter	Symbol	Min	Тур	Max	Unit	Conditions	
Overall Device							
Bias Current 1	Icc1	_	80	130	μA	Io = 0 mA	
Bias Current 2	lcc2	_	150	300	μA	lo = 50 mA (Ta = 25 °C)	
Regulator							
Output Voltage	V _{OUT}	4.90	5.00	5.10	V	Io = 200 mA	
Line Regulation	Line.Reg	_	5	35	mV	5.6 V ≤ V _{CC} ≤ 36 V	
Load Regulation	Load.Reg	_	30	70	mV	5 mA ≤ lo ≤ 200 mA	
Dropout Voltage	ΔVd	-	0.3	0.6	V	V _{CC} = 4.75 V, Io = 200 mA	
Ripple Rejection	R.R.	45	55	_	dB	f = 120Hz, ein = 1 Vrms, lo = 100 mA	
Reset							
Detection Voltage (BD3020HFP-M)	Vdet	4.02	4.10	4.18	V		
Detection Voltage (BD3021HFP-M)	Vdet	4.40	4.50	4.60	V		
Hysteresis Width	V _{HS}	50	100	150	mV		
Output Delay Time Low→High (Power On Reset Time)	t _{аLН}	1.1	1.9	2.7	ms	V_{CC} = Vdet ±0.5 V (V_{CC} = V_{OUT}) INH = open ⁽¹⁾ , C_{CT} = 0.01 μ F	
Output Delay Time High→Low	taHL	_	100	300	μs	V_{CC} = Vdet ±0.5 V (V_{CC} = V_{OUT}) INH = open ⁽¹⁾ , C_{CT} = 0.01 μ F	
RESET Discharge Current	IRESET	0.2	_	_	mA	V _{CC} = 1.5 V, V _{RESET} = 0.5 V (V _{CC} = V _{OUT})	
CT Discharge Current	Ict	0.1	_	_	mA	V _{CC} = 1.5 V, V _{CT} = 0.5 V (V _{CC} = V _{OUT})	
Low Output Voltage	V _{RST}	_	0.1	0.2	٧	V _{OUT} = 4.0 V	
Min Operating Voltage	Vopl	1.5	_	_	V		

⁽¹⁾ BD3021HFP-M only

Electrical Characteristics (Unless otherwise specified, -40°C, ≤ Ta ≤ +125 °C, V_{CC} = 13.5 V, V_{CLK} = GND)

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Parameter	Symbol	Limit		Unit	Conditions		
Parameter	Symbol	Min	Тур	Max	Unit	Conditions	
Watchdog timer							
CT Switching Threshold Voltage High	VthH	1.08	1.15	1.25	V	WDT ON ⁽¹⁾ , INH = open ⁽¹⁾	
CT Switching Threshold Voltage Low	VthL	0.13	0.15	0.17	V	WDT ON ⁽¹⁾ , INH = open ⁽¹⁾	
WDT Charge Current	Ictc	3.5	6.0	8.5	μA	WDT ON ⁽¹⁾ , INH = open ⁽¹⁾ V _{CT} = 0 V	
WDT Discharge Current	lctd	1.2	2.0	2.8	μA	WDT ON ⁽¹⁾ , INH = open ⁽¹⁾ V _{CT} = 1.3 V	
Watchdog Monitor Time Low	twн	3.0	5.0	7.0	ms	WDT ON ⁽¹⁾ , INH = open ⁽¹⁾ , C _{CT} = 0.01 µF (Ceramic Capacitor) ⁽²⁾	
Watchdog Reset Time	t _{WL}	1.0	1.7	2.4	ms		
CLK Input Pulse Width	twclk	500	-	-	ns		
INH*							
WDT OFF Threshold Voltage	V_{HINH}	V _{OUT} × 0.8	-	V _{OUT}	V		
WDT ON Threshold Voltage	VLINH	0	-	V _{ОUТ} × 0.3	V	INH is pulled down inside the IC when INH open.	
INH Input current	linh	-	10	20	μA	V _{INH} = 5 V	

⁽¹⁾ BD3021HFP-M only(2) Characteristics of ceramic capacitor not considered.

Reference data

Unless otherwise specified, Ta = 25 °C, V_{CC} = 13.5 V, V_{CLK} = GND

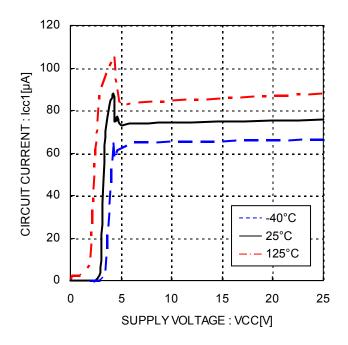


Figure 5. Circuit Current 1

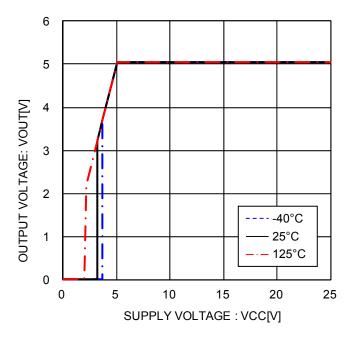


Figure 7. Input Stability

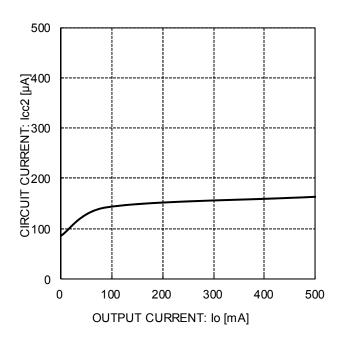


Figure 6. Circuit Current 2

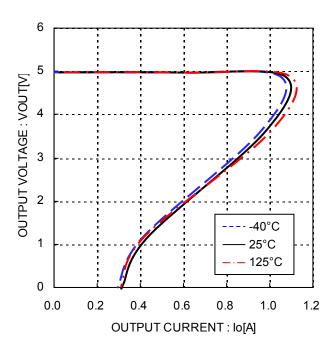
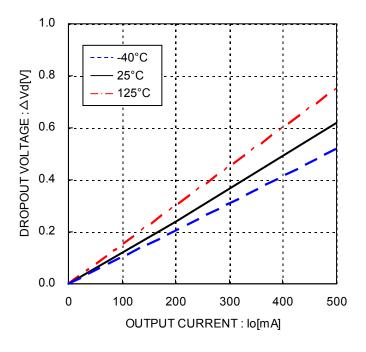


Figure 8. Lode Stability

Reference data

Unless otherwise specified, Ta = 25 °C, V_{CC} = 13.5 V, V_{CLK} = GND



80 70 -40°C 25°C RIPPLE REJECTION : R.R.[dB] 60 125°C 50 40 30 20 10 10 100 1000 10000 100000 1000000 FREQUENCY: f[Hz]

Figure 9. Dropout Voltage

Figure 10. Ripple Rejection

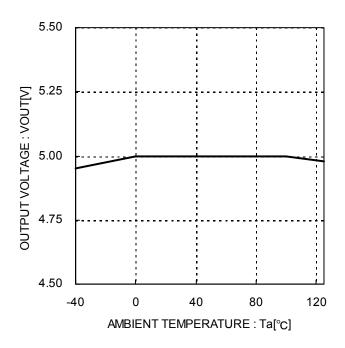


Figure 11. Output Voltage Temperature Characteristics

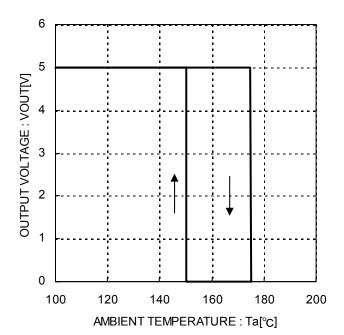
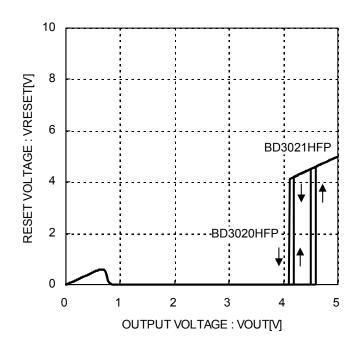


Figure 12. Thermal Shutdown Circuit Characteristics

Reference data

Unless otherwise specified, Ta = 25 °C, Vcc = 13.5 V, Vclk = GND



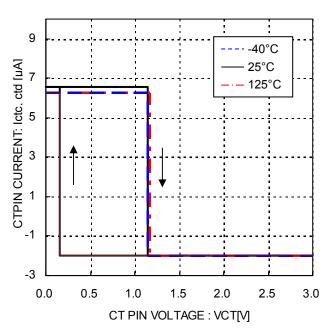


Figure 13. Detection Voltage (V_{CC}=V_{OUT})

Figure 14. CT Pin Charge / Discharge Current (V_{CC}=5V)

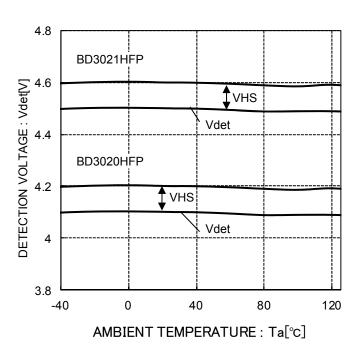


Figure 15. Reset Detection Voltage vs. Temperature

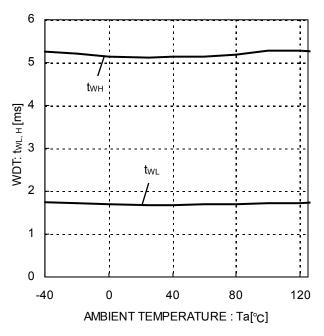
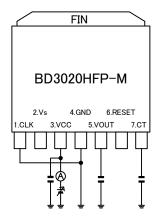
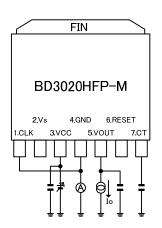


Figure 16. WDT Time vs. Temperature $(C_{CT}=0.01\mu F)$

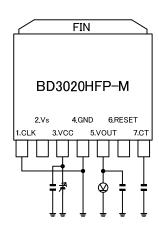
Measurement Circuits (BD3020HFP-M)



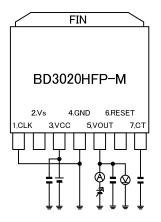
Measurement setup for Figure 5.



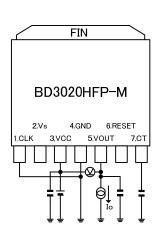
Measurement setup for Figure 6.



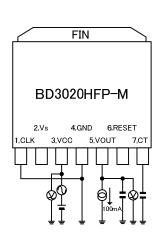
Measurement setup for Figure 7, 11, 12.



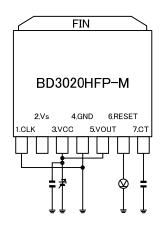
Measurement setup for Figure 8.



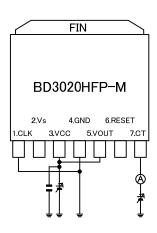
Measurement setup for Figure 9.



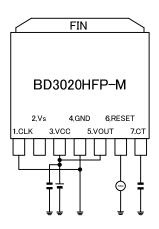
Measurement setup for Figure 10.



Measurement setup for Figure 13, 15.

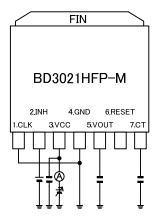


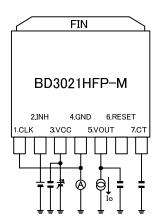
Measurement setup for Figure 14.

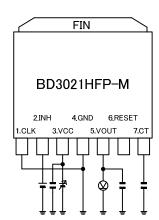


Measurement setup for Figure 16.

Measurement Circuits (BD3021HFP-M)



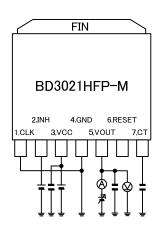


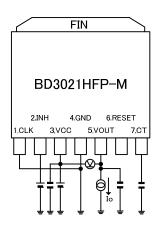


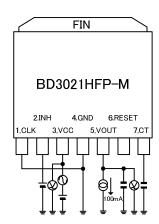
Measurement setup for Figure 5.

Measurement setup for Figure 6.

Measurement setup for Figure 7, 11, 12.



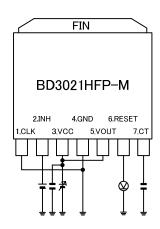


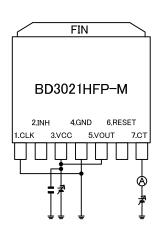


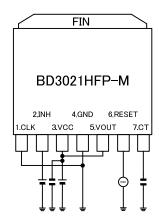
Measurement setup for Figure 8.

Measurement setup for Figure 9.

Measurement setup for Figure 10.





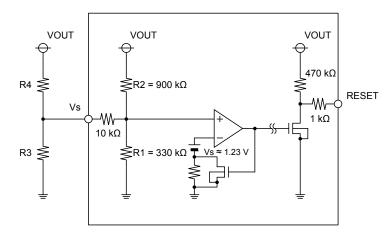


Measurement setup for Figure 13, 15.

Measurement setup for Figure 14.

Measurement setup for Figure 16.

BD3020HFP-M Detection Voltage Adjustment (Resistance value is typical value)



IC Internal Block Diagram

When typical detection voltage is 4.1 V

Vdet ≈ Vs × (R1 + R2) / R1

Vdet : Reset detection voltage

Vs : Internal reference voltage (MOS input)

• R1, R2 : IC internal resistor

(Voltage detection precision is tightened up to ±2 % by laser-trimming the R1 and R2)

Vs will fluctuate 1.23 V ±6.0 %.

The reset detection voltage can be adjusted by connecting resistors on the Vs terminal.

Insert pull down resistor R3 (lower resistance than R1) in between Vs-GND, and pull down resistor R4 (lower resistance than R2) in between Vs-VOUT to adjust the detection voltage.

By doing so, the detection voltage can be adjusted by the calculation below.

$$Vdet = Vs \times [{R2 \times R4 / (R2 + R4)} + {R1 \times R3 / (R1 + R3)}] / {R1 \times R3 / (R1+R3)}$$

When the output resistance value is as small enough to ignore the IC internal resistance, you can find the detection voltage by the calculation below.

$$Vdet \approx Vs \times (R3 + R4) / R3$$

Adjust the resistance value by application as the circuit current will increase due to the added resistor.

BD3020/21HFP-M Power on Reset / Watchdog Timer

Power ON reset (output delay time) is adjustable by CT pin capacitor.

 t_{dLH} (S) \approx (1.15 V × CT capacitance (μ F)) / Ictc (μ A) (Typ)

t_{dLH} : Output delay time (power ON reset)
 1.15 V : Upper switching threshold voltage (Typ)

• CT capacitance : Capacitor connected to CT pin

Ictc : WDT charge current

Calculation example) with 0.01 µF CT pin capacitor

$$t_{dLH}$$
 (S) = 1.15 V × 0.01 μF / 6 μA

≈ 1.9 ms

*If the CT capacitance is not the same as the condition on the electrical characteristics table, i.e., 0.01 µF, choose the capacitance value in ratio referring to the above equation.

Watch Dog Timer (WDT twh, twl) is adjustable by the CT pin capacitor

 $t_{W \ H}$ (S) $\approx 1.00 \ V \times CT$ capacitance (μF)) / Ictd(μA) (Typ) $t_{W \ L}$ (S) $\approx 1.00 \ V \times CT$ capacitance (μF)) / Ictc(μA) (Typ)

• t_{WH} : Watchdog monitor time Low (delay time to turn the reset ON)

• t_{WL} : Watchdog reset time (time the reset is ON)

1.00 V : Upper switching threshold voltage - lower switching threshold voltage

• CT capacitance : CT pin capacitor *Shared with power ON reset

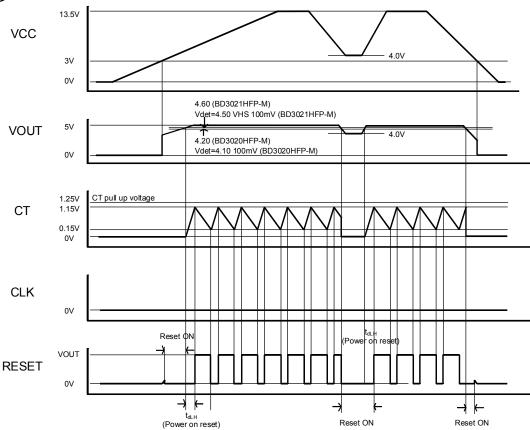
Ictc : WDT charge currentIctd : WDT discharge current

Calculation example) with 0.01 µF CT pin capacitor

$$t_{W H}$$
 (S) $\approx 1.00 \text{ V} \times 0.01 \,\mu\text{F} / 2 \,\mu\text{A} \approx 5.0 \,\text{ms}$ (Typ)

$$t_{W L}$$
 (S) $\approx 1.00 \text{ V} \times 0.01 \,\mu\text{F} / 6 \,\mu\text{A} \approx 1.7 \,\text{ms}$ (Typ)

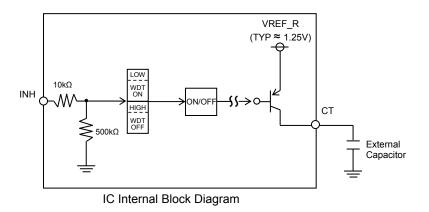
<Timing Chart>



^{*}If the CT capacitance is not the same as the condition on the electrical characteristics table, i.e., 0.01 µF, choose the capacitance value in ratio referring to the above equation.

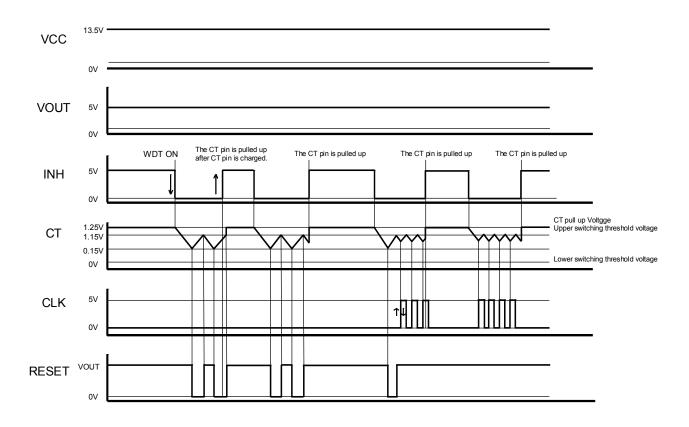
WDT timer ON / OFF switch INH (Resistance value is typical value)

BD3021HFP-M has a switch INH to turn the WDT ON / OFF.

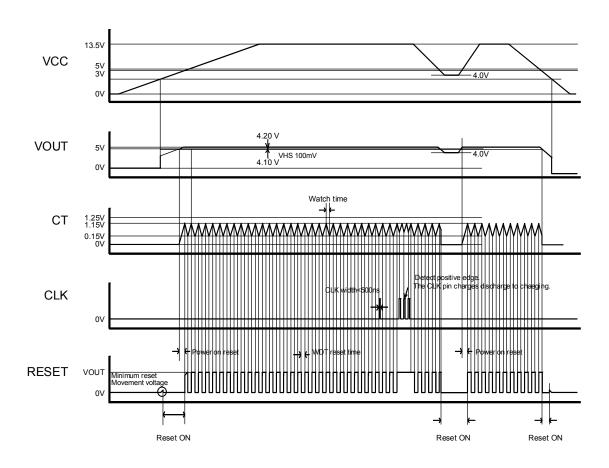


By using INH ON, CT potential can be pulled up to internal voltage VREF_R (invalid with power ON reset).

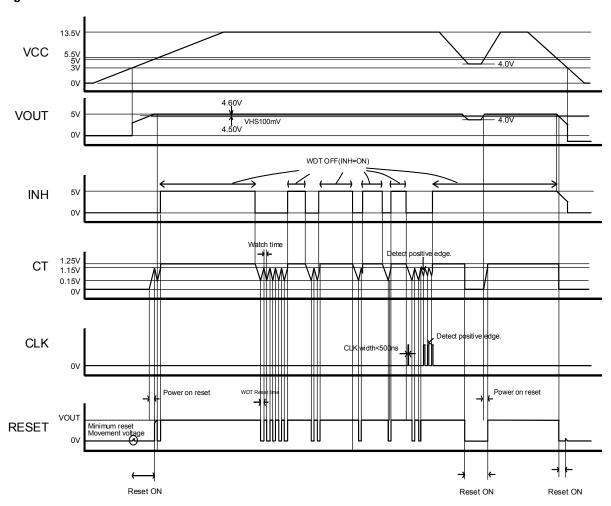
<Timing Chart> BD3021HFP-M



<Timing Chart> BD3020HFP-M



<Timing Chart> BD3021HFP-M



Pin Settings / Precautions

1. VCC Pin

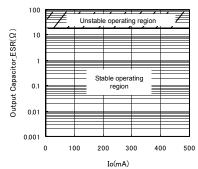
Insert a 0.33 µF to 1000 µF capacitor between the VCC and GND. The appropriate capacitance value varies by application. Be sure to allow a sufficient margin for input voltage levels.

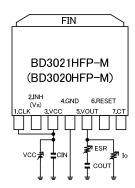
2. Output pins

In order to prevent oscillation, a capacitor needs to be placed between the output pin and GND. We recommend using a capacitor with a capacitance of $0.1~\mu\text{F}$ to $1000~\mu\text{F}$. Electrolytic, tantalum and ceramic capacitors can be used. When selecting the capacitor ensure that the capacitance of $0.1\mu\text{F}$ to $1000\mu\text{F}$ is maintained at the intended applied voltage and temperature range. Due to changes in temperature the capacitor's capacitance can fluctuate possibly resulting in oscillation. For selection of the capacitor refer to the Cout_ESR vs. lo data. The stable operation range given in the reference data is based on the standalone IC and resistive load. For actual applications the stable operating range is influenced by the PCB impedance, input supply impedance and load impedance. Therefore verification of the final operating environment is needed.

Also, in case of rapidly changing input voltage and load current, select the capacitance in accordance with verifying that the actual application meets with the required specification.







Output Capacitor ESR vs lo (reference data)

*Pin Settings / Precautions2 Measurement circuit

3. CT pin

Connecting a capacitance of 0.01 μF to 1 μF on the CT pin is recommended.

Power Dissipation

■HRP7

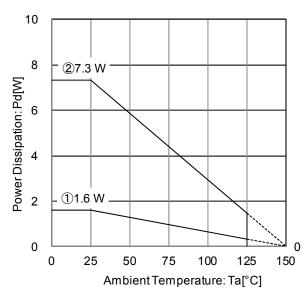


Figure 17. Package Data (HRP7)

IC mounted on ROHM standard board.

Board material: FR4

Board size: 70.0 mm × 70.0 mm × 1.6 mmt

(with thermal via on the board)

Mount condition: PCB and exposed pad are soldered. Top copper foil: The footprint ROHM recommend.

+ wiring to measure.

①: 1-layer PCB

(Back surface copper foil area: 0mm × 0 mm)

2: 4-layer PCB

(Back surface copper foil area: 70.0mm × 70.0 mm)

Condition ①: θ ja = 78.1 °C / W Condition ②: θ ja = 17.1 °C / W

Refer to Figure 17 thermal dissipation characteristics for usage above Ta = 25 °C. The IC's characteristics are affected heavily by the temperature, and if is exceeds its max junction temperature (Tjmax), the chip may degrade or destruct. Thermal design is critical in terms of avoiding Instantaneous destruction and reliability in long term usage.

The IC needs to be operated below its max junction temperature (Tjmax) to avoid thermal destruction. Refer to Figure 17 for HRP7 package thermal dissipation characteristics. Operate the IC within power dissipation (Pd) when using this IC.

Power consumption Pc (W) calculation will be as below

Vcc : Input Voltage
VouT : Output Voltage
Io : Load Current
Icc : Circuit Current

Pc = $(V_{CC} - V_{OUT}) \times Io + V_{CC} \times Icc$ Power dissipation Pd \geq Pc

If load current lo is calculated to operate within power dissipation, it will be as below, where you can find the max load current lomax for the applied voltage V_{CC} of the thermal design.

$$lo \le \frac{Pd - V_{CC} \times Icc}{V_{CC} - V_{OUT}}$$
 (Refer to Figure 6 for the Icc)

■Example) at Ta = 125 °C, V_{CC} = 12 V, V_{OUT} = 5 V

lo ≤
$$\frac{1.46 - 12 \times lcc}{12 - 5}$$

$$\begin{cases} \theta ja = 17.1 \text{ °C / W} \rightarrow -58.4 \text{ mV / °C} \\ 25 \text{ °C} = 7.30 \text{ W} \rightarrow 125 \text{ °C} = 1.452 \text{ W} \end{cases}$$
lo ≤ 208 mA (lcc: 150 µA)

At Ta = 125 °C with Figure 17 ② condition, the calculation shows that ca 207 mA of output current is possible at 7 V potential difference across input and output.

I/O Equivalence Circuit (Resistance value is typical value) CLK (1pin) Vs (BD3020HFP-M 2pin) VOUT VOUT **≧** 900kΩ CLK 10kΩ 10kΩ **≷** 330kΩ INH (BD3021HFP-M 2pin) VCC (3pin) VOUT VCC INH 10kΩ $500k\Omega$ VOUT (5pin) RESET (6pin) VOUT 470kΩ RESET 1kΩ VOUT $3.7M\Omega$ 1.25ΜΩ CT (7pin)

Figure 18. I / O equivalence circuit

Operational Notes

1. Electrical characteristics

Electrical characteristics described in these specifications may vary, depending on temperature, supply voltage, external circuits and other conditions. Therefore, be sure to check all relevant factors, including transient characteristics.

2. Reverse Connection of Power Supply

Connecting the power supply in reverse polarity can damage the IC. Take precautions against reverse polarity when connecting the power supply, such as mounting an external diode between the power supply and the IC's power supply pins.

3. Power Supply Lines

Design the PCB layout pattern to provide low impedance supply lines. Separate the ground and supply lines of the digital and analog blocks to prevent noise in the ground and supply lines of the digital block from affecting the analog block. Furthermore, connect a capacitor to ground at all power supply pins. Consider the effect of temperature and aging on the capacitance value when using electrolytic capacitors.

4. Ground Voltage

Ensure that no pins are at a voltage below that of the ground pin at any time, even during transient condition.

5. Ground Wiring Pattern

When using both small-signal and large-current ground traces, the two ground traces should be routed separately but connected to a single ground at the reference point of the application board to avoid fluctuations in the small-signal ground caused by large currents. Also ensure that the ground traces of external components do not cause variations on the ground voltage. The ground lines must be as short and thick as possible to reduce line impedance.

6. Thermal Consideration

Should by any chance the power dissipation rating be exceeded the rise in temperature of the chip may result in deterioration of the properties of the chip. In case of exceeding this absolute maximum rating, increase the board size and copper area to prevent exceeding the Pd rating.

Use a thermal design that allows for a sufficient margin in light of the Pd in actual operating conditions. Consider Pc that does not exceed Pd in actual operating conditions. (Pd ≥ Pc)

Tjmax: Maximum junction temperature = 150 °C, Ta: Peripheral temperature [°C], θja: Thermal resistance of package-ambience [°C / W], Pd: Package Power dissipation [W], Pc: Power dissipation [W], V_{CC}: Input Voltage, V_{OUT}: Output Voltage, Io: Load, I_{CC2}: Bias Current2

Package Power dissipation : Pd (W) = $(Tjmax - Ta) / \theta ja$

Power dissipation : $Pc(W) = (V_{CC} - V_{OUT}) \times I_0 + V_{CC} \times I_{CC2}$

7. Inrush Current

When power is first supplied to the IC, it is possible that the internal logic may be unstable and inrush current may flow instantaneously due to the internal powering sequence and delays, especially if the IC has more than one power supply. Therefore, give special consideration to power coupling capacitance, power wiring, width of ground wiring, and routing of connections.

8. Testing on Application Boards

When testing the IC on an application board, connecting a capacitor directly to a low-impedance output pin may subject the IC to stress. Always discharge capacitors completely after each process or step. The IC's power supply should always be turned off completely before connecting or removing it from the test setup during the inspection process. To prevent damage from static discharge, ground the IC during assembly and use similar precautions during transport and storage.

9. Inter-pin Short and Mounting Errors

Ensure that the direction and position are correct when mounting the IC on the PCB. Incorrect mounting may result in damaging the IC. Avoid nearby pins being shorted to each other especially to ground, power supply and output pin. Inter-pin shorts could be due to many reasons such as metal particles, water droplets (in very humid environment) and unintentional solder bridge deposited in between pins during assembly to name a few.

10. Unused Input Pins

Input pins of an IC are often connected to the gate of a MOS transistor. The gate has extremely high impedance and extremely low capacitance. If left unconnected, the electric field from the outside can easily charge it. The small charge acquired in this way is enough to produce a significant effect on the conduction through the transistor and cause unexpected operation of the IC. So unless otherwise specified, unused input pins should be connected to the power supply or ground line.

Operational Notes - continued

11. Regarding the Input Pin of the IC

This monolithic IC contains P+ isolation and P substrate layers between adjacent elements in order to keep them isolated. P-N junctions are formed at the intersection of the P layers with the N layers of other elements, creating a parasitic diode or transistor. For example (refer to figure below):

When GND > Pin A and GND > Pin B, the P-N junction operates as a parasitic diode.

When GND > Pin B, the P-N junction operates as a parasitic transistor.

Parasitic diodes inevitably occur in the structure of the IC. The operation of parasitic diodes can result in mutual interference among circuits, operational faults, or physical damage. Therefore, conditions that cause these diodes to operate, such as applying a voltage lower than the GND voltage to an input pin (and thus to the P substrate) should be

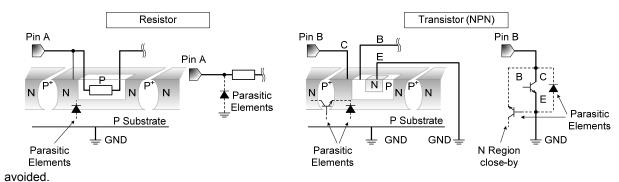


Figure 19. Example of monolithic IC structure

12. Thermal Shutdown Circuit(TSD)

This IC has a built-in thermal shutdown circuit that prevents heat damage to the IC. Normal operation should always be within the IC's power dissipation rating. If however the rating is exceeded for a continued period, the junction temperature (Tj) will rise which will activate the TSD circuit that will turn OFF all output pins. When the Tj falls below the TSD threshold, the circuits are automatically restored to normal operation.

Note that the TSD circuit operates in a situation that exceeds the absolute maximum ratings and therefore, under no circumstances, should the TSD circuit be used in a set design or for any purpose other than protecting the IC from heat damage.

13. Over Current Protection Circuit (OCP)

This IC incorporates an integrated overcurrent protection circuit that is activated when the load is shorted. This protection circuit is effective in preventing damage due to sudden and unexpected incidents. However, the IC should not be used in applications characterized by continuous operation or transitioning of the protection circuit.

14. Applications or inspection processes where the potential of the VCC pin or other pins may be reversed from their normal state may cause damage to the IC's internal circuitry or elements. Use an output pin capacitance of 1000µF or lower in case VCC is shorted with the GND pin while the external capacitor is charged. Insert a diode in series with VCC to prevent reverse current flow, or insert bypass diodes between VCC and each pin.

15. Positive voltage surges on VCC pin

A power zener diode should be inserted between VCC and GND for protection against voltage surges of more than 50 V on the VCC pin.

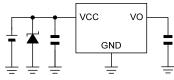


Figure 20. Application Examples 1

16. Negative voltage surges on VCC pin

A schottky barrier diode should be inserted between VCC and GND for protection against voltages lower than GND on the VCC pin.

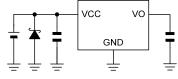


Figure 21. Application Examples 2

Operational Notes - continued

17. Output protection diode

Loads with large inductance components may cause reverse current flow during startup or shutdown. In such cases, a protection diode should be inserted on the output to protect the IC.

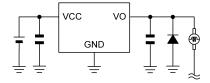
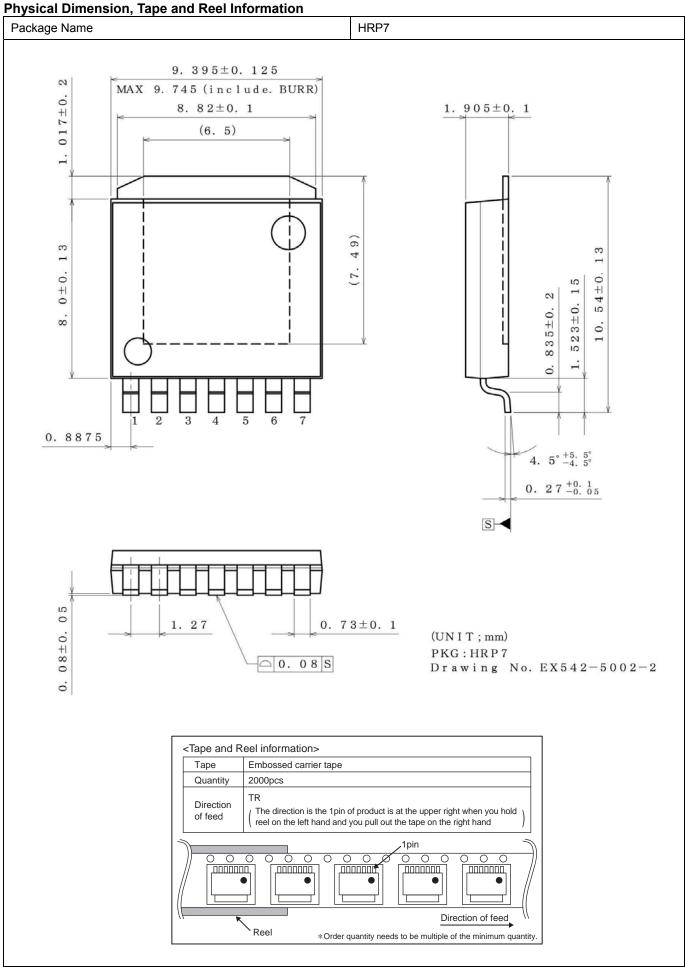
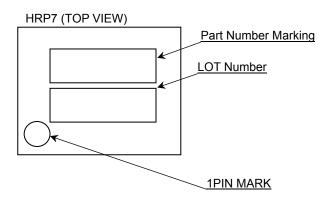


Figure 22. Application Examples 3



Marking Diagram



Product Name	Part Number Marking		
BD3020HFP-M	BD3020		
BD3021HFP-M	BD3021		

Revision History

Date	Revision	Changes
24.Jun.2013	001	New Release
20.Apr.2015	002	Applied the ROHM standard Style

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Ì	JAPAN USA		EU	CHINA
Γ	CLASSⅢ	CL ACCTI	CLASS II b	СГУССШ
Γ	CLASSIV	CLASSⅢ	CLASSⅢ	CLASSⅢ

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 - [d] Use of our Products in places where the Products are exposed to static electricity or electromagnetic waves
 - [e] Use of our Products in proximity to heat-producing components, plastic cords, or other flammable items
 - [f] Sealing or coating our Products with resin or other coating materials
 - [g] Use of our Products without cleaning residue of flux (even if you use no-clean type fluxes, cleaning residue of flux is recommended); or Washing our Products by using water or water-soluble cleaning agents for cleaning residue after soldering
 - [h] Use of the Products in places subject to dew condensation
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 - [b] the temperature or humidity exceeds those recommended by ROHM
 - [c] the Products are exposed to direct sunshine or condensation
 - [d] the Products are exposed to high Electrostatic
- 2. Even under ROHM recommended storage condition, solderability of products out of recommended storage time period may be degraded. It is strongly recommended to confirm solderability before using Products of which storage time is exceeding the recommended storage time period.
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