

H.F. POWER TRANSISTOR

N-P-N silicon planar epitaxial transistor intended for use in class-AB operated high power industrial and military transmitting equipment in the h.f. band. The transistor presents excellent performance as a linear amplifier in s.s.b. applications. It is resistance stabilized and is guaranteed to withstand severe load mismatch conditions. Matched h_{FE} groups are available on request.

The transistor has a $\frac{1}{2}$ " flange envelope with a ceramic cap. All leads are isolated from the flange.

QUICK REFERENCE DATA

R.F. performance up to $T_h = 25\text{ }^\circ\text{C}$

mode of operation	V_{CE} V	$I_{C(ZS)}$ A	f MHz	P_L W	G_p dB	η_{dt} %	d_3 dB
s.s.b. (class-AB)	50	0,1	1,6 – 28	20 – 160 (P.E.P.)	> 14	> 40*	< -30

* At 160 W P.E.P.

MECHANICAL DATA

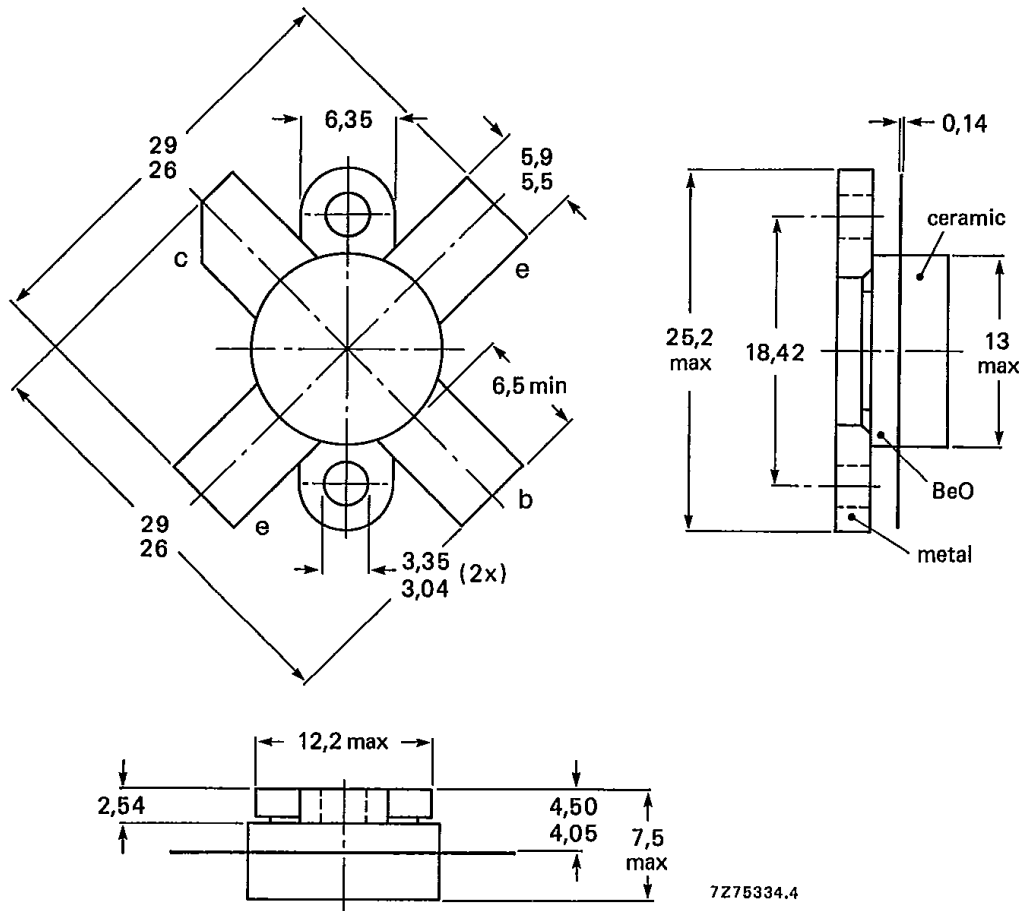
SOT-121A (see Fig. 1).

PRODUCT SAFETY This device incorporates beryllium oxide, the dust of which is toxic. The device is entirely safe provided that the BeO disc is not damaged.

MECHANICAL DATA

Fig. 1 SOT-121.

Dimensions in mm



Torque on screw: min. 0,6 Nm (6 kg cm)
 max. 0,75 Nm (7,5 kg cm)

Recommended screw: cheese-head 4-40 UNC/2A

Heatsink compound must be applied sparingly and evenly distributed.

RATINGS

Limiting values in accordance with the Absolute Maximum System (IEC 134)

Collector-emitter voltage ($V_{BE} = 0$)
peak value V_{CESM} max. 110 V

Collector-emitter voltage (open base)

 V_{CEO} max. 53 V

Emitter-base voltage (open collector)

 V_{EBO} max. 4 V

Collector current (average)

 $I_{C(AV)}$ max. 8 ACollector current (peak value); $f > 1$ MHz I_{CM} max. 20 AR.F. power dissipation ($f > 1$ MHz); $T_{mb} = 25$ °C P_{rf} max. 245 W

Storage temperature

 T_{stg} -65 to +150 °C

Operating junction temperature

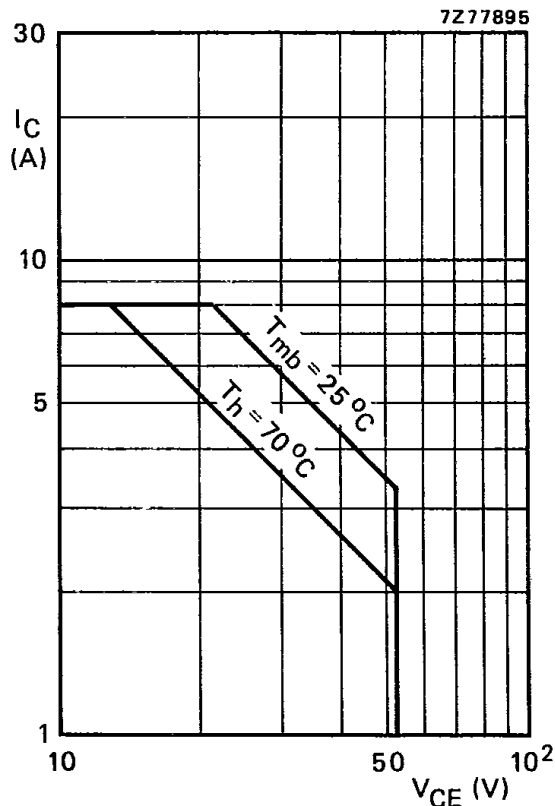
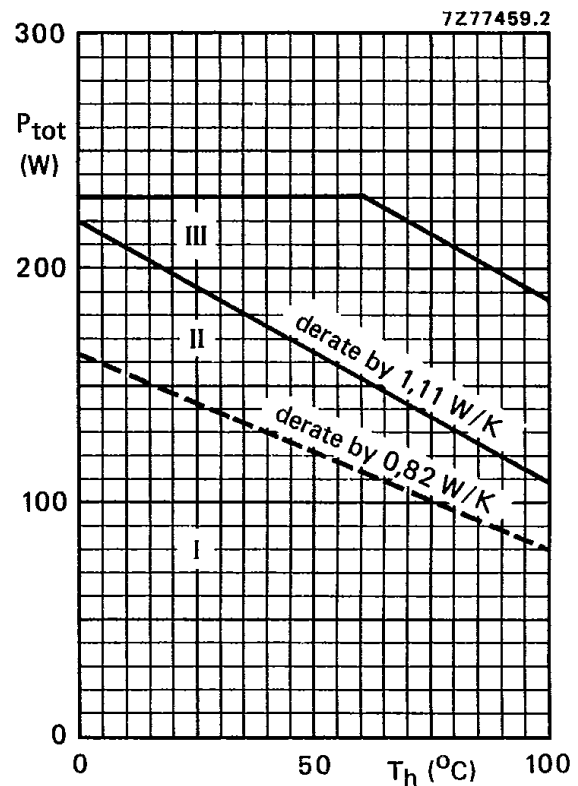
 T_j max. 200 °C

Fig. 2 D.C. SOAR.

Fig. 3 R.F. power dissipation; $V_{CE} \leq 50$ V;
 $f \geq 1$ MHz.

- I Continuous d.c. operation
- II Continuous r.f. operation
- III Short-time operation during mismatch

THERMAL RESISTANCE (dissipation = 100 W; $T_{mb} = 90$ °C, i.e. $T_h = 70$ °C)

From junction to mounting base (d.c. dissipation)

 $R_{th\ j-mb(dc)}$ = 1,0 K/W

From junction to mounting base (r.f. dissipation)

 $R_{th\ j-mb(rf)}$ = 0,7 K/W

From mounting base to heatsink

 $R_{th\ mb-h}$ = 0,2 K/W

CHARACTERISTICS

 $T_j = 25\text{ }^\circ\text{C}$

Collector-emitter breakdown voltage

 $V_{BE} = 0; I_C = 25\text{ mA}$ $V_{(BR)CES} > 110\text{ V}$

Collector-emitter breakdown voltage

open base; $I_C = 100\text{ mA}$ $V_{(BR)CEO} > 53\text{ V}$

Emitter-base breakdown voltage

open collector; $I_E = 20\text{ mA}$ $V_{(BR)EBO} > 4\text{ V}$

Collector cut-off current

 $V_{BE} = 0; V_{CE} = 53\text{ V}$ $I_{CES} < 10\text{ mA}$ Second breakdown energy; $L = 25\text{ mH}; f = 50\text{ Hz}$

open base

 $R_{BE} = 10\ \Omega$ $E_{SBO} > 12,5\text{ mJ}$ $E_{SBR} > 12,5\text{ mJ}$

D.C. current gain *

 $I_C = 4\text{ A}; V_{CE} = 5\text{ V}$ h_{FE} typ. 30
15 to 50

D.C. current gain ratio of matched devices *

 $I_C = 4\text{ A}; V_{CE} = 5\text{ V}$ $h_{FE1}/h_{FE2} \leq 1,2$

Collector-emitter saturation voltage *

 $I_C = 12,5\text{ A}; I_B = 2,5\text{ A}$ V_{CEsat} typ. 2,2 VTransition frequency at $f = 100\text{ MHz}$ * $-I_E = 4\text{ A}; V_{CB} = 40\text{ V}$ $-I_E = 12,5\text{ A}; V_{CB} = 40\text{ V}$ f_T typ. 270 MHz f_T typ. 285 MHzCollector capacitance at $f = 1\text{ MHz}$ $I_E = I_e = 0; V_{CB} = 50\text{ V}$ C_c typ. 185 pFFeedback capacitance at $f = 1\text{ MHz}$ $I_C = 150\text{ mA}; V_{CE} = 50\text{ V}$ C_{re} typ. 115 pF

Collector-flange capacitance

 C_{cf} typ. 3 pFMeasured under pulse conditions: $t_p \leq 200\ \mu\text{s}; \delta \leq 0,02$.

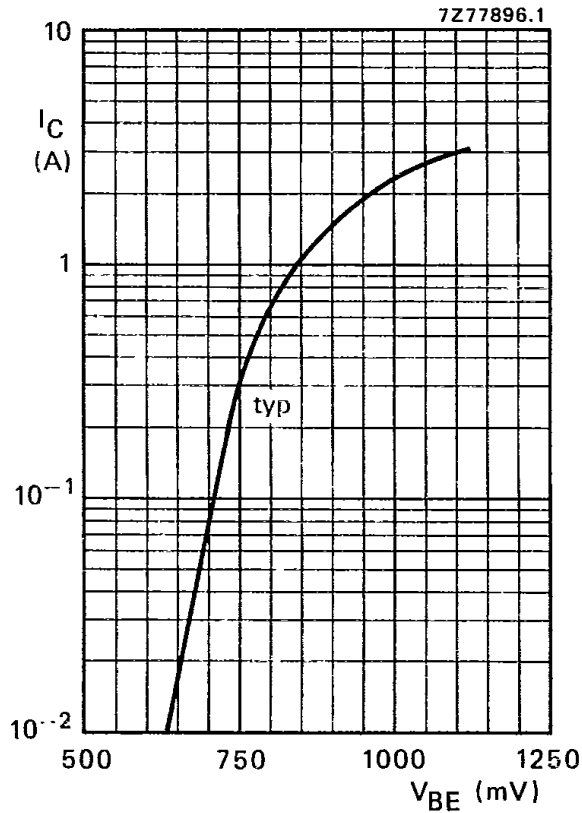


Fig. 4 $V_{CE} = 40 \text{ V}; T_h = 25 \text{ }^\circ\text{C}.$

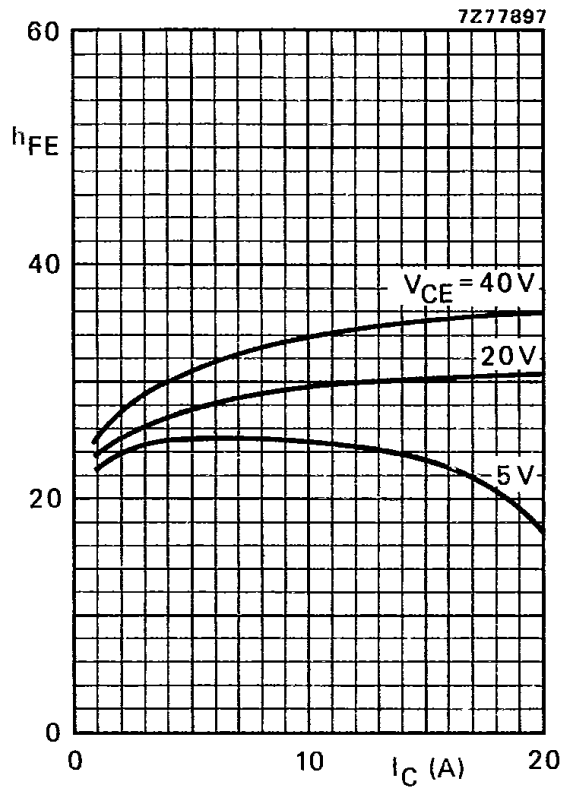


Fig. 5 Typical values; $T_j = 25 \text{ }^\circ\text{C}.$

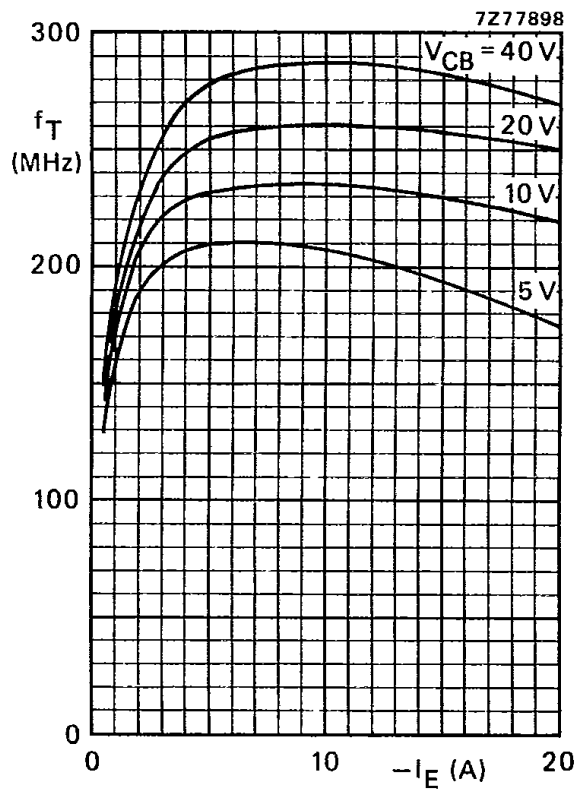


Fig. 6 Typical values; $f = 100 \text{ MHz}; T_j = 25 \text{ }^\circ\text{C}.$

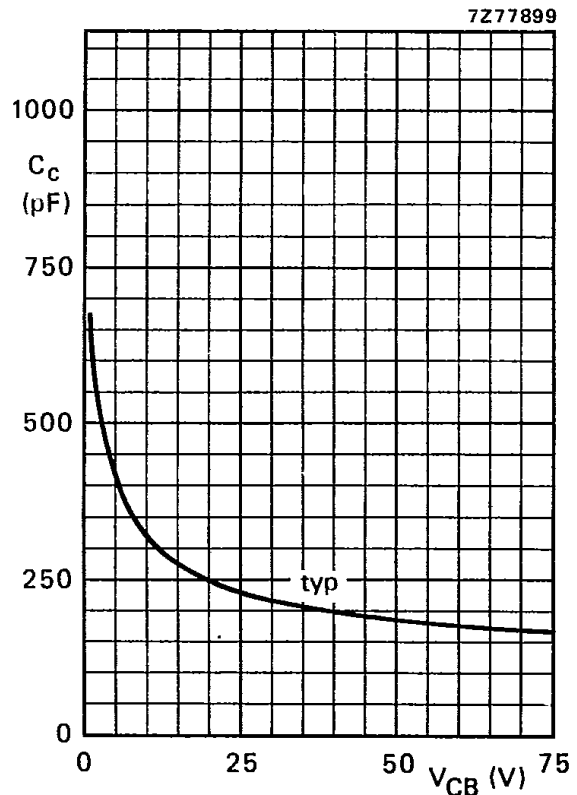


Fig. 7 $I_E = I_e = 0; f = 1 \text{ MHz}; T_j = 25 \text{ }^\circ\text{C}.$

APPLICATION INFORMATION

R.F. performance in s.s.b. class-AB operation (linear power amplifier)

 $V_{CE} = 50 \text{ V}$; $T_h = 25 \text{ }^\circ\text{C}$; $f_1 = 28,000 \text{ MHz}$; $f_2 = 28,001 \text{ MHz}$

output power W	G_p dB	$\eta_{dt}(\%)$ at 160 W (P.E.P.)	I_C (A)	d_3 dB *	d_5 dB *	$I_C(Z_S)$ A
20 to 160 (P.E.P.)	> 14	> 40	< 4,0	< -30	< -30	0,1

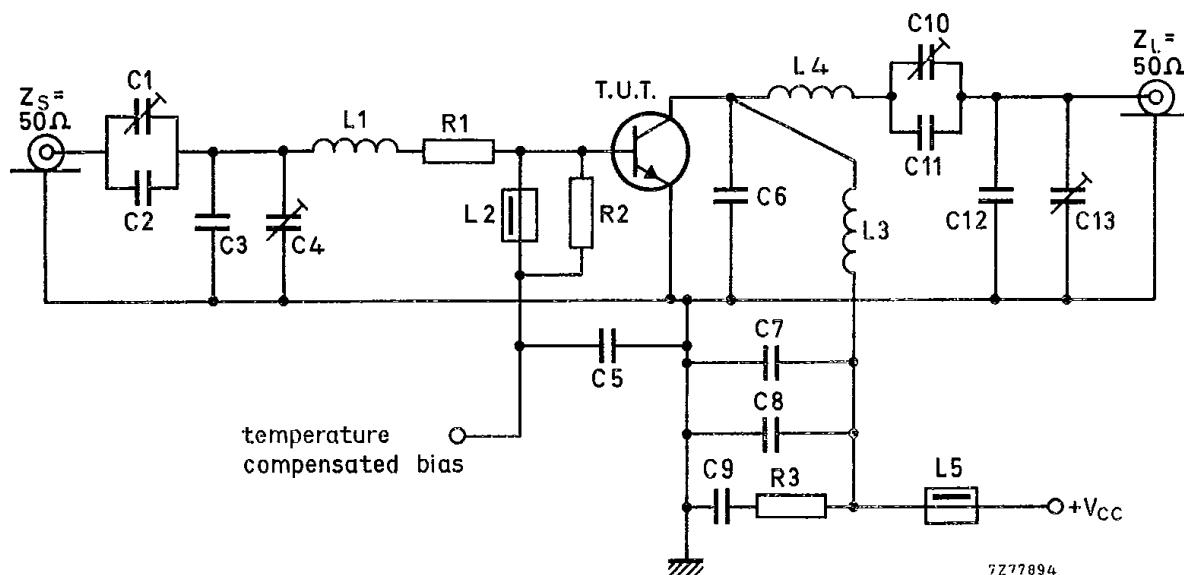


Fig. 8 Test circuit; s.s.b. class-AB.

List of components:

C1 = C10 = 100 pF film dielectric trimmer

C2 = C6 = 27 pF ceramic capacitor (500 V)

C3 = 220 pF polystyrene capacitor

C4 = C13 = 100 pF film dielectric trimmer

C5 = C7 = 3,9 nF ceramic capacitor

C8 = 100 nF polyester capacitor

C9 = 2,2 μ F moulded metallized polyester capacitor

C11 = 68 pF ceramic capacitor (500 V)

C12 = 220 pF polystyrene capacitor

L1 = 88 nH; 3 turns Cu wire (1,0 mm); int. dia. 9,0 mm; length 6,1 mm; leads 2 x 5 mm

L2 = L5 = Ferroxcube wide-band h.f. choke, grade 3B (cat. no. 4312 020 36640)

L3 = 180 nH; 4 turns enamelled Cu wire (1,6 mm); int. dia. 12,0 mm; length 9,9 mm; leads 2 x 10 mm

L4 = 350 nH; 7 turns enamelled Cu wire (1,6 mm); int. dia. 12,0 mm; length 19,1 mm; leads 2 x 10 mm

R1 = 0,66 Ω ; parallel connection of 5 x 3,3 Ω carbon resistors ($\pm 5\%$; 0,5 W each)R2 = 27 Ω carbon resistor ($\pm 5\%$; 0,5 W)R3 = 4,7 Ω carbon resistor ($\pm 5\%$; 0,5 W)

* Stated intermodulation distortion figures are referred to the according level of either of the equal amplified tones. Relative to the according peak envelope powers these figures should be increased by 6 dB.

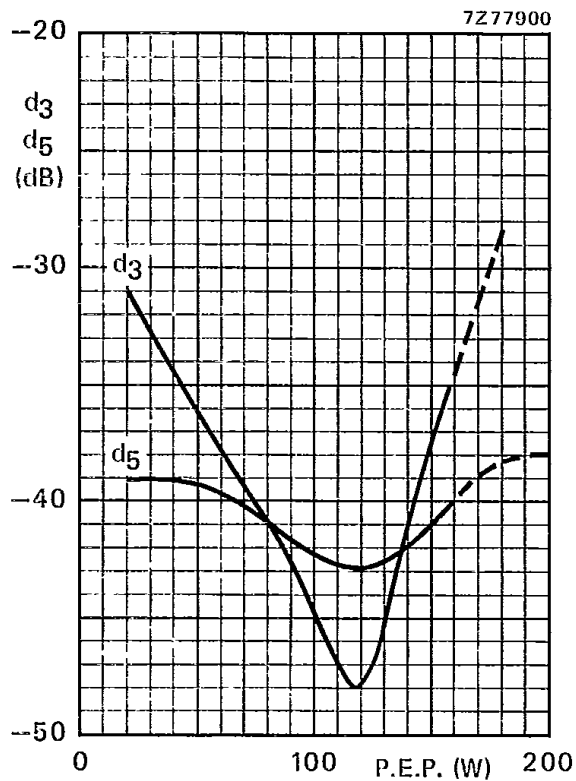


Fig. 9 Intermodulation distortion as a function of output power.*

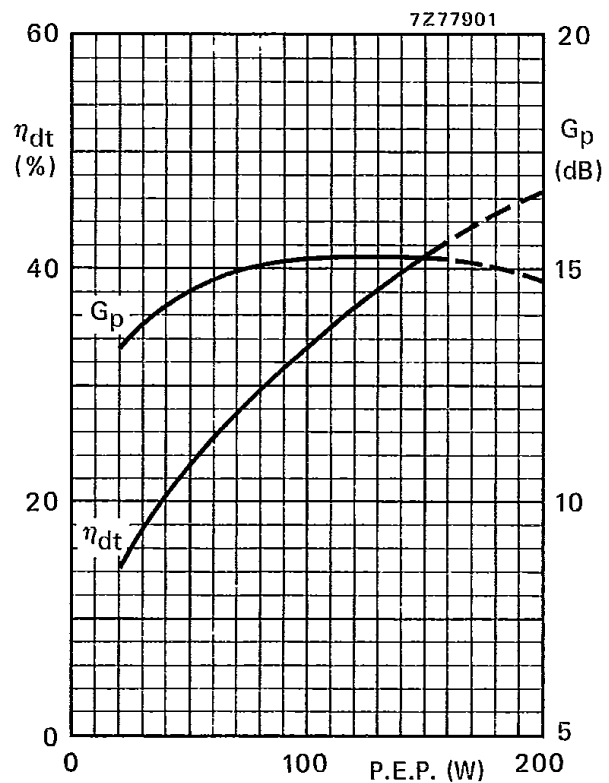


Fig. 10 Double-tone efficiency and power gain as a function of output power.

Conditions for Figs 9 and 10:

$V_{CE} = 50 \text{ V}$; $I_{C(ZS)} = 0,1 \text{ A}$; $f_1 = 28,000 \text{ MHz}$; $f_2 = 28,001 \text{ MHz}$; $T_h = 25 \text{ }^\circ\text{C}$; typical values.

Ruggedness

The BLW95 is capable of withstanding full load mismatch ($V_{SWR} = 50$) up to 150 W (P.E.P.) under the following conditions:

$V_{CE} = 45\text{V}$; $f = 28 \text{ MHz}$; $T_h = 70 \text{ }^\circ\text{C}$; $R_{th \text{ mb-h}} = 0,2 \text{ K/W}$.

* See note on previous page.

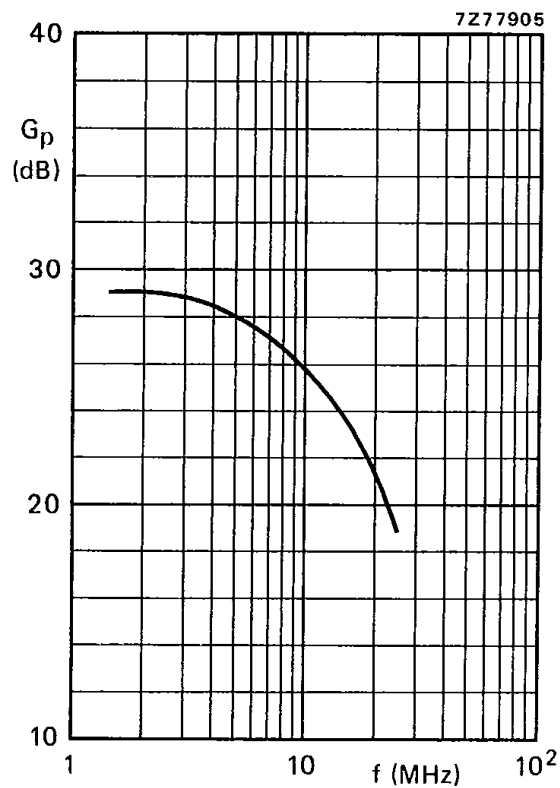


Fig. 11 Power gain as a function of frequency.

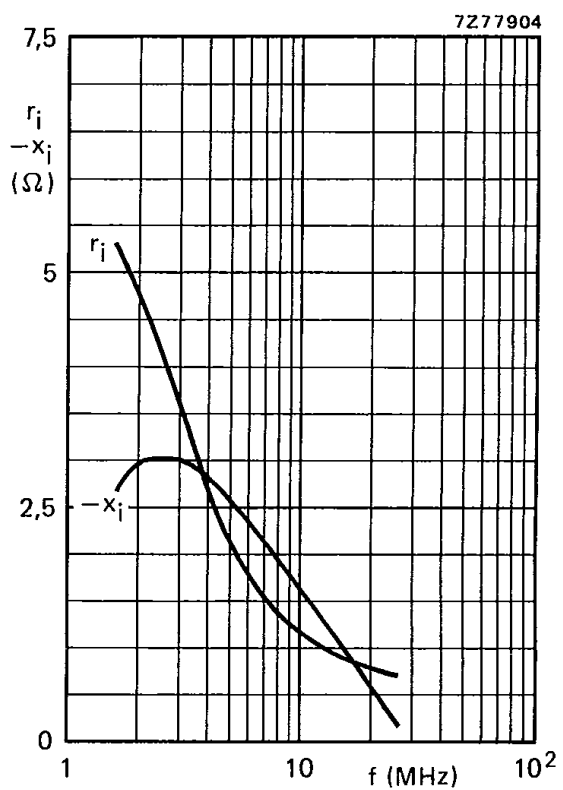


Fig. 12 Input impedance (series components) as a function of frequency.

Figs 11 and 12 are typical curves and hold for an unneutralized amplifier in s.s.b. class-AB operation.

Conditions:

$V_{CE} = 50 \text{ V}$; $I_C(ZS) = 0,1 \text{ A}$; $P_L = 160 \text{ W (P.E.P.)}$; $T_h = 25 \text{ }^\circ\text{C}$; $Z_L = 6,25 \text{ } \Omega$ in series with $7,3 \text{ nH}$ (in parallel with -188 pF).

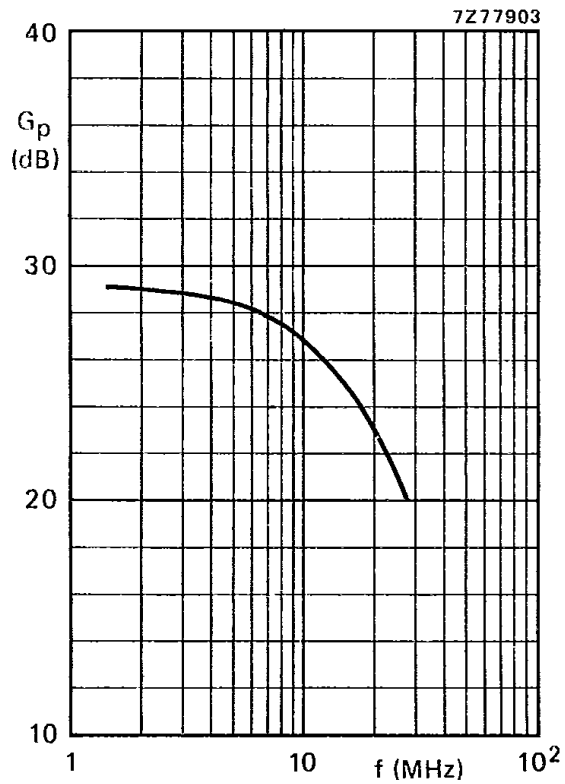


Fig. 13 Power gain as a function of frequency.

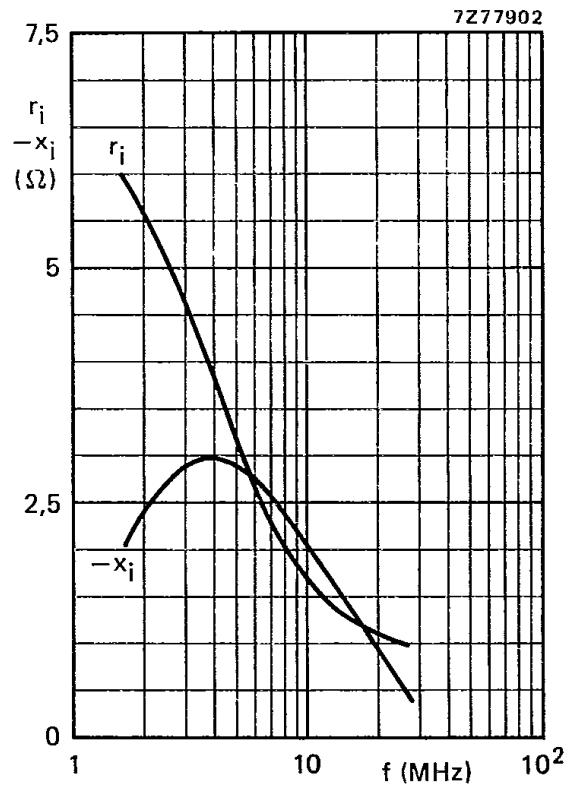


Fig. 14 Input impedance (series components) as a function of frequency.

Figs 13 and 14 are typical curves and hold for one transistor of a push-pull amplifier with cross-neutralization in s.s.b. class-AB operation.

Conditions:

$V_{CE} = 50 \text{ V}$; $I_{C(ZS)} = 0,1 \text{ A}$; $P_L = 160 \text{ W (P.E.P.)}$; $T_h = 25 \text{ }^\circ\text{C}$; $Z_L = 6,25 \text{ } \Omega$ in series with $10,4 \text{ nH}$ (in parallel with -267 pF); neutralizing capacitor: 82 pF .