

Inverter for motor control

600V IGBT Intelligent Power Module (IPM) for high speed switching drive

BM63764S-VA BM63764S-VC

General Description

BM63764S-VA /-VC is an Intelligent Power Module composed of gate drivers, bootstrap diodes, IGBTs, fly wheel diodes. Small switching loss IGBTs optimized for high speed switching drive such as a washing machine or a fan motor is adopted.

Features

- 3phase DC/AC Inverter
- 600V/15A
- Low Side IGBT Open Emitter
- Built -in Bootstrap Diode
- High Side IGBT Gate Driver(HVIC):
 SOI (Silicon On Insulator) Process,
 Drive Circuit, High Voltage Level Shifting,
 Current Limit for Bootstrap Diode,
 Control Supply Under-Voltage Locked Out (UVLO)
- Low Side IGBT Gate Driver(LVIC):
 Drive Circuit, Short Circuit Current Protection (SCP),
 Control Supply Under Voltage Locked Out (UVLO),
 Thermal Shutdown (TSD)
- Fault Signal(LVIC)
 Corresponding to SCP (Low Side IGBT), TSD, UVLO
 Fault
- Input Interface 3.3V, 5V Line

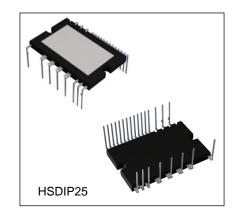
Application

- High Speed Switching Drive of AC100 to 240Vrms(DC Voltage: Less Than 400V) Class Motor
- High Speed Switching Drive of Motor for Washing Machine, Fan

Key Specifications

■ IGBT Collector-Emitter Voltage V_{CESAT}: 1.7V(Typ)
 ■ FWD Forward Voltage V_F: 1.5V(Typ)
 ■ FWD Reverse Recovery Time t_{rr}: 100ns(Typ)
 ■ Module Case Temperature T_C: -25 to +100°C
 ■ Junction Temperature T_{imax}: 150°C

Package HSDIP25 HSDIP25VC W(Typ) x D(Typ) x H(Typ) 38.0mm x 24.0mm x 3.5mm 38.0mm x 24.0mm x 3.5mm



Typical Application Circuit

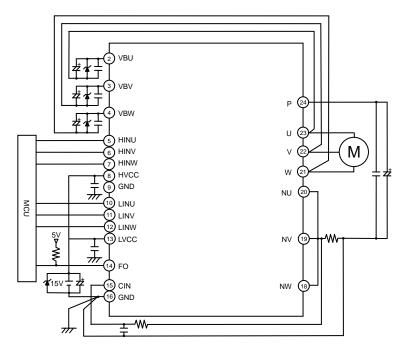


Figure 1. Example of Application Circuit

Pin Configuration

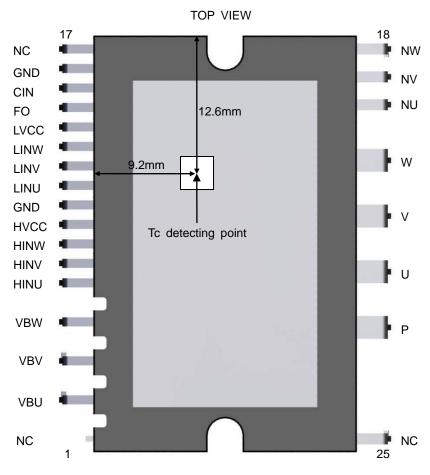


Figure 2. Pin Configuration and Tc Detecting Point

Pin Description

Description					
Pin No.	Pin Name	Function	Pin No.	Pin Name	Function
1	NC	No connection(GND potential)	14	FO	Alarm output
2	VBU	U phase floating control supply	15	CIN	Detecting of short circuit current trip voltage
3	VBV	V phase floating control supply	16	GND	Ground (Note 1)
4	VBW	W phase floating control supply	17	NC	No connection (Note 2)
5	HINU	U phase high side IGBT control	18	NW	W phase low side IGBT emitter
6	HINV	V phase high side IGBT control	19	NV	V phase low side IGBT emitter
7	HINW	W phase high side IGBT control	20	NU	U phase low side IGBT emitter
8	HVCC	Control supply for HVIC	21	W	W phase output
9	GND	Ground (Note 1)	22	V	V phase output
10	LINU	U phase low side IGBT control	23	U	U phase output
11	LINV	V phase low side IGBT control	24	Р	Inverter supply
12	LINW	W phase low side IGBT control	25	NC	No connection (Note 2)
13	LVCC	Control supply for LVIC		•	

⁽Note 1) Two GND pins (9 & 16pin) are connected inside IPM, please connect one pin (16pin is recommended) to the 15V power supply GND outside and leave the other open.

(Note 2) NC pins (17 $\&\,25\text{pin})$ are not electrically connected to any other potential inside.

Block Diagram

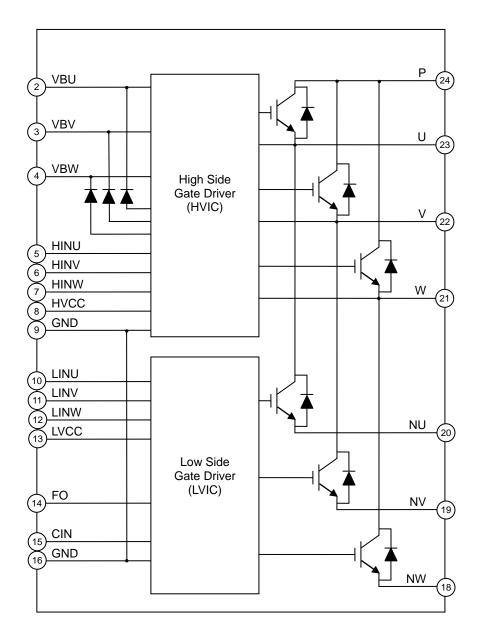


Figure 3. Block Diagram

Description of Block

1) High Side IGBT Drive (HVIC, Bootstrap Diode)

High voltage level shifting circuit drives high side IGBT.

Built-in bootstrap diode and current limit function for bootstrap diode enable HVIC to drive high side IGBT without external component (bootstrap diode, resistor). There is under-voltage-locked-out (UVLO) function for floating control power supply.

2) Low Side IGBT Drive (LVIC)

There is short circuit current protection (SCP), under-voltage locked out (UVLO) for control power supply LVCC, thermal shutdown (TSD) function. Alarm signal (FO) will output when these protection circuits work.

Absolute Maximum Ratings (Unless otherwise specified, Tj=25°C)

Inverter Part

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Item		Symbol	Ratings	Unit	Conditions			
Supply Voltage	Supply Voltage		450	V	Applied between P-NU,NV,NW			
Supply Voltage(Surge)		V _{P(surge)}	500	V	Applied between P-NU,NV,NW			
Collector-Emitter Voltage		V _{CES}	600	V				
Collector Current	DC	Ic	±15 (Note 1)	Α	T _C =25°C			
Collector Current	PEAK	I _{CP}	±30 ^(Note 1)	Α	T _C =25°C, less than 1ms			
Collector Power Dissipation		Pc	41	W	T _C =25°C, per 1 chip			
Junction Temperature		T _{jmax}	150	°C				

(Note 1) Do not, however exceed P_{C} , ASO.

Control part

Item	Symbol	Ratings	Unit	Conditions
Control Power Supply	V _{CC}	20	V	Applied between HVCC-GND, LVCC-GND
Floating Control Power Supply	V _{BS}	20	V	Applied between VBU-U, VBV-V, VBW-W
Control Input Voltage	V _{IN}	-0.5 to V _{CC} +0.5	V	Applied between HINX, LINX-GND (X=U,V,W)
Fault Output Supply Voltage	V_{FO}	-0.5 to V _{CC} +0.5	V	Applied between FO-GND
Fault Output Current	I _{FO}	1	mA	Sink current at FO pin
Current Sensing Input Voltage	V _{CIN}	-0.5 to +7.0	V	Applied between CIN-GND

Bootstrap diode part

Item	Symbol	Ratings	Unit	Conditions
Reverse Voltage	V_{RB}	600	V	
Junction Temperature	T _{jmaxD}	150	°C	

Total system

Item	Symbol	Ratings	Unit	Conditions
Self Protection Supply Voltage (SCP Capability)	V _{P(PROT)}	400	V	V _{CC} =13.5 to 16.5V, Inverter part T _j =125°C, non-repetitive, less than 2μs
Module Case Temperature	T _C	-25 to +100	°C	Measurement point of T _C is provided in Figure 2
Storage Temperature	T _{stg}	-40 to +125	°C	
Isolation Voltage	V _{iso}	1500	V _{rms}	Sinusoidal, 60Hz, AC 1minute, between connected all pins and heat sink plate

Thermal resistance

Itom	Symbol		Limit		Unit	Conditions
Item	Symbol	Min	Тур	Max	Offic	Conditions
Junction to Case Thermal	R _{th(j-c)_IGBT}	-	-	3.0	°C /W	Inverter IGBT(1/6 module)
Resistance (Note 2)	R _{th(j-c)_FWD}	-	-	3.9	°C /W	Inverter FWD(1/6 module)

(Note 2) Grease with good conductivity and high reliability should be applied evenly with +100 to +200µm on the contacting surface of IPM and heat sink. Use a torque wrench to fasten up to the specified torque rating. The contacting thermal resistance between IPM case and heat sink is determined by the thickness and the thermal conductivity of the applied grease.

Caution: Operating the IPM over the absolute maximum ratings may damage the IPM. The damage can either be a short circuit between pins or an open circuit between pins and the internal circuitry. Therefore, it is important to consider circuit protection measures, such as adding a fuse, in case the IPM is operated over the absolute maximum ratings.

Recommended Operating Conditions

Itam	Cumbal		Limit		Lloit	Conditions	
Item	Symbol	Min	Тур	Max	Unit	Conditions	
Supply Voltage	V_P	0	300	400	V	Applied between P-NU,NV,NW	
Control Power Supply	V _{CC}	13.5	15.0	16.5	V	Applied between HVCC-GND, LVCC-GND	
Floating Control Power Supply	V_{BS}	13.0	15.0	18.5	V	Applied between VBU-U, VBV-V, VBW-W	
Control Power Supply Variation	⊿V _{CC} ⊿V _{BS}	-1	-	+1	V/µs		
Control Input Voltage	V _{IN}	0	-	5.5	V		
Current Sensing Input Voltage	V _{CIN}	0	-	5.5	V		
Blanking Time for Preventing Arm-short	t _{dead}	1.5	-	-	μs	For each input signal	
PWM Input Frequency	f _{PWM}	-	-	20	kHz	T _C ≤ 100°C, T _j ≤ 125°C	
High Side IGBT	PW _{ONH}	0.8	-	-	μs		
Minimum Input Pulse Width ^(Note1)	PW _{OFFH}	0.8	-	-	μs		
Low Side IGBT	PWonL	2.5	-	-	μs		
Minimum Input Pulse Width ^(Note1)	PW _{OFFL}	0.8	-	-	μs		
Voltage Variation Between GND- NU, NV, NW	V _N	-5	-	+5	V	Between GND-NU, NV, NW (Including surge voltage)	
Junction Temperature	Tj	-25		+125	°C		

(Note 1) IPM might not respond if the input signal pulse width is less than PW_{ON}, PW_{OFF}.

Electrical Characteristics (Unless otherwise specified, Tj=25°C, V_{CC} = V_{BS} =15V, V_P =300V) Inverter Part

Item	Cumbal		Limit		Unit	Conditions	
item	Symbol	Min	Тур	Max	Offic	Conditions	
Collector Emitter Seturation Voltage	\/	-	1.70	2.10	V	I _C =15A	
Collector-Emitter Saturation Voltage	V _{CESAT}	-	0.90	1.20	V	I _C =1.5A	
Collector-Emitter Cut-off Current	I _{CES}	-	-	100	μA	V _{CE} =V _{CES}	
FWD Forward Voltage	V _F	-	1.50	2.00	V	I _F =15A	
FWD Reverse Recovery Time	t _{rr}	-	100	-	ns	Inductive Load, I _C =15A	
High Side IGBT Turn on Delay Time	t _{onH}	0.50	0.90	1.40	μs	Inductive Load, I _C =15A	
High Side IGBT Turn on Switching Time	t _{c(on)H}	-	0.20	-	μs	Inductive Load, I _C =15A	
High Side IGBT Turn off Delay Time	t _{offH}	-	1.00	1.70	μs	Inductive Load, I _C =15A	
High Side IGBT Turn off Switching Time	t _{c(off)H}	-	0.20	-	μs	Inductive Load, I _C =15A	
Low Side IGBT Turn on Delay Time	t _{onL}	0.80	1.70	2.50	μs	Inductive Load, I _C =15A	
Low Side IGBT Turn on Switching Time	t _{c(on)L}	-	0.25	-	μs	Inductive Load, I _C =15A	
Low Side IGBT Turn off Delay Time	t _{offL}	-	0.65	1.15	μs	Inductive Load, I _C =15A	
Low Side IGBT Turn off Switching Time	t _{c(off)L}	-	0.20	-	μs	Inductive Load, I _C =15A	

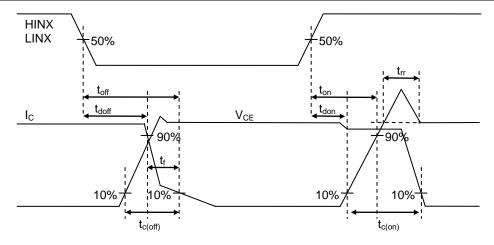


Figure 4. Switching Time Definition

Control part

Item	Cumbal		Limit		Unit	Conditions
nem	Symbol	Min	Тур	Max	Unit	Conditions
Whole						
VCC Circuit Current 1	I _{CC1}	-	0.70	1.60	mA	VIN=0V
VCC Circuit Current 2	I _{CC2}	-	0.90	2.00	mA	VIN=5V
VBS Circuit Current 1	I _{BS1}	-	0.15	0.35	mA	VIN=0V
VBS Circuit Current 2	I _{BS2}	-	0.15	0.35	mA	VIN=5V
Control Input(HINU,HINV,HINW	LINU,LINV	,LINW)				
H Level Input Current	I _{INH}	0.7	1.0	1.5	mA	V _{IN} =5V
L Level Input Current	I _{INL}	-10	-	-	μΑ	V _{IN} =0V
H Level Input Threshold Voltage	V _{INH}	-	-	2.6	V	
L Level Input Threshold Voltage	V_{INL}	0.8	-	-	V	
Input Hysteresis Voltage	V _{HYS}	-	0.25	-	V	
Short Circuit Current Protectio	n		l .	ll .		·
CIN Input Bias Current	I _{CIN}	-2	-	-	μΑ	CIN=0V
Trip Voltage	V _{SC}	0.43	0.48	0.53	V	
Under Voltage Locked Out	1		l .	ll .		·
VCC Trip Voltage	Vccuvt	10.5	11.5	12.5	V	
VCC Release Voltage	V_{CCUVR}	11	12	13	V	
VBS Trip Voltage	V_{BSUVT}	10	11	12	V	
VBS Release Voltage	V_{BSUVR}	10.5	11.5	12.5	V	
Thermal Shutdown	1		II.	II.	1	ı
Trip Temperature	T _{SDT}	100	130	-	°C	Monitor LVIC temperature
Hysteresis Temperature	T _{SDHYS}	-	20	-	°C	Monitor LVIC temperature
Fault Output(FO)	- "		•	•	•	
Output low Voltage	V_{FO}	-	-	0.95	V	I _{FO} =1mA
Leak Current	I _{FOLEAK}	-	-	10	μΑ	V _{FO} =5V
Output Pulse Width	t _{FO}	20	-	-	μs	

Bootstrap Diode Part

otstrap Diode Part								
Item	Cumbal		Limit		Unit	Conditions		
item	Symbol	Min	Тур	Max	Unit	Conditions		
Forward Voltage	V_{FB1}	0.3	0.6	0.9	V	I _{FB} =1mA Voltage drop between HVCC-VBX (X=U,V,W)		
	V _{FB2}	1.1	2.0	2.9	V	I _{FB} =100mA Voltage drop between HVCC-VBX (X=U,V,W)		
Reverse Current	I _{RB}	-	-	10	μA	V _{RB} =600V		
Reverse Recovery Time	t _{rrB}	-	80	-	ns	I _{FB} =0.1A		

Mechanical Characteristics And Ratings

Item	Limit			Unit	Following Standard	Conditions
item	Min	Тур	Max	Unit	Following Standard	Conditions
Mounting Torque	0.59	0.69	0.78	N∙m	-	Mounting Screw M3 (Note 1) Recommended 0.69N·m (Note 2)
Pin Pulling Strength	10	-	-	s	EIAJ-ED-4701/400	Load Control Pin:4.9N Power Pin:9.8N
Pin Bending Strength	2	-	-	times	EIAJ-ED-4701/400	Load Control Pin: 2.45N Power Pin:4.9N 90deg. Bend
Weight	-	10	-	g	-	
Heat Sink Flatness	0	-	+200	μm	-	Measurement point is provided in Figure 6.

⁽Note 1) Plain washers (ISO 7089 to 7094) are recommended.

⁽Note 2) When installing a module to a heat sink, excessive uneven fastening force might apply stress to inside chips or ceramic of heat sink plate, which will lead to a broken or cracked or degraded device. An example of recommended fastening sequence is shown in Figure 5. The temporary fastening torque is set to 20 to 30% of the maximum torque rating. Evenly apply thermally-conductive grease with 100µm to 200µm thickness over the contact surface between the module and the heat sink. Also, pay attention not to have any dirt left on the contact surface between the module and the heat sink.

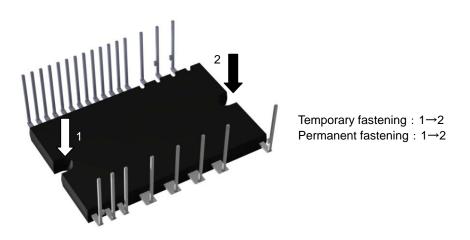


Figure 5. Example of Recommended Fastening Sequence

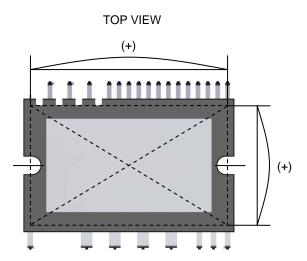


Figure 6. Measurement Point of Heat Sink Flatness

Typical Performance Curve

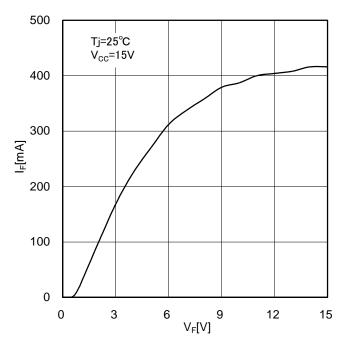


Figure 7. I_F vs V_F
Characteristic of Bootstrap Diode I_F-V_F Curve
Between HVCC-VBX pin (X=U,V,W)

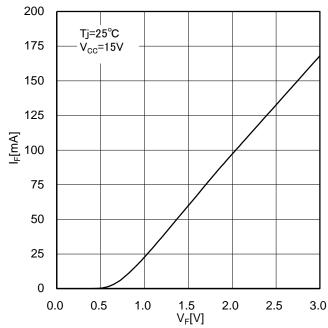
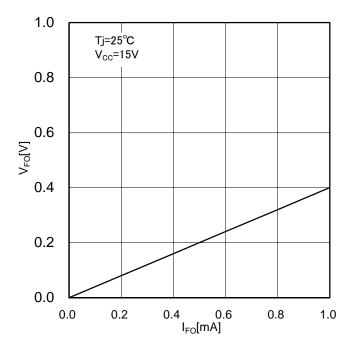


Figure 8. Magnification of Figure 7



 $\label{eq:Figure 9.VFO} Figure 9. \ V_{FO} \ vs \ I_{FO} \\ (Characteristic of FO pin \ V_{FO}\text{-}I_{FO} \ Curve)$

Timing Chart

- 1) Short Circuit Current Protection (protection with the external shunt resistor and RC filter)
 - a1. Normal operation: IGBT ON and outputs current Ic.
 - a2. Short circuit current detection (SCP trigger)
 - It is recommended to set RC time constant of 1.0 to 2.0µs so that IGBT shuts down within 2.0µs when SCP is triggered.
 - a3. All low side IGBT's gates are shut down (soft turn off).
 - a4. All low side IGBTs turn off.
 - a5. FO outputs for t_{FO} =20 μ s (Min).
 - a6. LIN=L
 - a7. LIN=H, but all IGBTs keep OFF during SCP=H.
 - a8. FO finishes output , but IGBTs don't turn on until inputting the next ON signal(LIN=L→H) IGBT of each phase can return to normal state by inputting ON signal to each phase.
 - a9. Normal operation: IGBT ON and outputs current Ic.

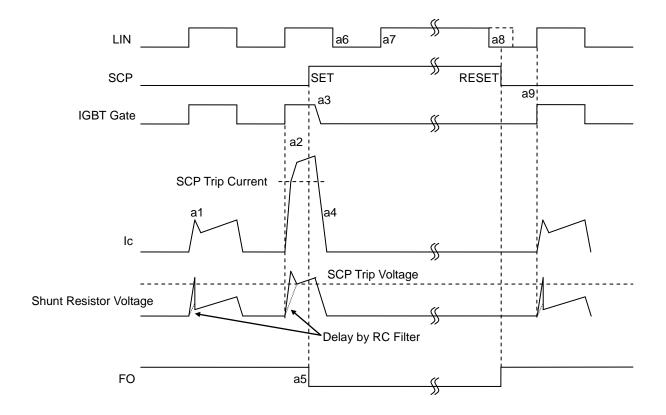


Figure 10. SCP Timing Chart

Notice

SCP works only for low side IGBT only.

In case of SCP trip and FO output, please stop controlling IPM quickly to avoid the abnormal state.

- 2) Control Supply (LVCC) Under Voltage Locked Out (UVLO)
 - b1 Control supply(LVCC) voltage exceeds UVLO release level (V_{CCUVR}), but IGBT turns on by the next ON signal (LIN=L→ H).IGBT of each phase can return to normal state by inputting ON signal to each phase.
 - b2 Normal operation: IGBT ON and outputs current Ic.
 - b3 LVCC drops to UVLO trip level (V_{CCUVT}).
 - b4 All low side IGBTs turn off in spite of control input condition.
 - b5 FO outputs for t_{FO} =minimum 20 μ s, but output is extended while LVCC is below V_{CCUVR} .
 - b6 LVCC reaches V_{CCUVR}.
 - b7 Even if LVCC reaches V_{CCUVR} during LIN=H, IGBTs don't turn on until inputting the next ON signal (LIN=L→H).
 - b8 Normal operation: IGBT ON and outputs current Ic.

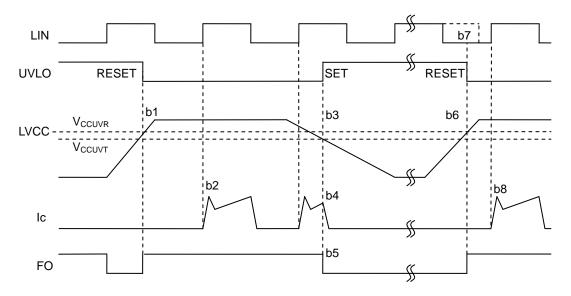


Figure 11. LVCC UVLO Timing Chart

- 3) Control supply (VBS) Under Voltage Locked Out (UVLO)
 - c1 Control supply(VBS) voltage exceeds UVLO release level (V_{BSUVR}), but IGBT turns on by the next ON signal (HIN=L→ H).
 - c2 Normal operation: IGBT ON and outputs current Ic.
 - c3 VBS drops to UVLO trip level (V_{BSUVT}).
 - c4 Only IGBT of the corresponding phase turns off in spite of control input signal, there is no FO signal output.
 - c5 VBS reaches V_{BSUVR}.
 - c6 Even if VBS reaches V_{BSUVR} during HIN=H, IGBTs don't turn on until inputting the next ON signal (HIN=L \rightarrow H).
 - c7 Normal operation: IGBT ON and outputs current Ic.

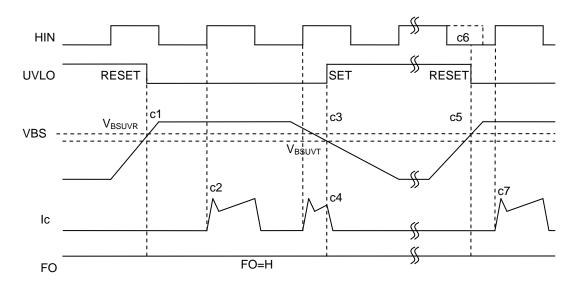


Figure 12. VBS UVLO Timing Chart

- 4) Thermal Shutdown (TSD), monitoring LVIC temperature
 - d1 Normal operation: IGBT ON and outputs current Ic.
 - d2 LVIC temperature (T_i) exceeds thermal shutdown trip level (T_{SDT}).
 - d3 All low side IGBTs turn off in spite of control input condition.
 - d4 FO outputs for t_{FO}=20µs (Min),

but output is extended while T_i is above thermal shutdown release level (T_{SDT}-T_{SDHYS}).

- d5 T_i drops to T_{SDT} - T_{SDHYS}
- d6 Even if T_j reaches T_{SDT}-T_{SDHYS} during LIN=H, IGBTs don't turn on until inputting the next ON signal (LIN=L→H). IGBT of each phase can return to normal state by inputting ON signal to each phase.
- d7 Normal operation: IGBT ON and outputs current Ic.

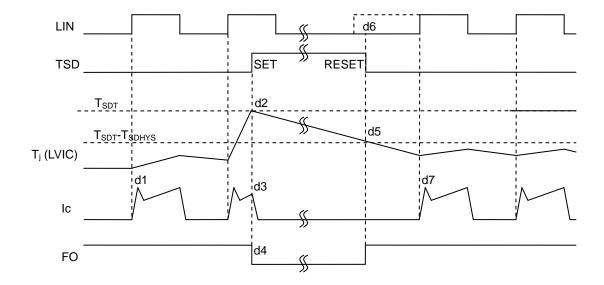


Figure 13. TSD Timing Chart

Notice

- 1) In case of TSD trip and FO output, please stop controlling IPM quickly to avoid the abnormal state.
- 2) If the cooling system is in abnormal state (e.g. heat sink comes off, fixed loosely, or cooling fan stops) when TSD trips, don't reuse IPM. This may cause the junction temperature of power chips to exceed its maximum rating of T_{jmax} (150°C).
- 3) TSD function detects LVIC temperature, so it cannot respond to rapid temperature rise of power chip. Therefore, TSD will not work properly in the case of rapid temperature rise like motor lock or over current.

Application Example(one shunt resistor drive)

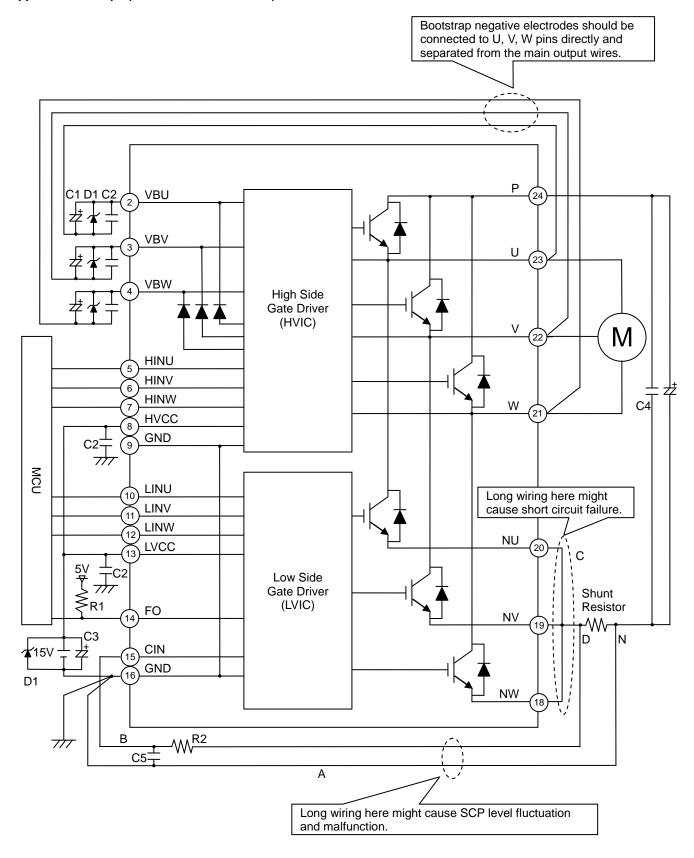


Figure 14. Example of Application Circuit

Selection of Components Externally Connected (Refer to Figure 14)

1) VBU, VBV, VBW pin

- The bypass capacitor(good temperature, frequency characteristic electrolytic type C1: 22µF to 100µF) should be mounted as close as possible to the pin in order to prevent malfunction or destruction due to switching noise and power supply ripple. In addition, for the purpose of reducing of the power supply's impedance in wide frequency bandwidth, ceramic capacitor (good temperature, frequency and DC bias characteristic ceramic type C2: 0.1µF to 0.22µF) should also be mounted.
- · Zenner diode D1(1W) should be mounted between each pair of control supply pins to prevent surge destruction.
- Line ripple voltage should meet dV/dt ≤1V/µs, V_{ripple} ≤ 2V_{p-p.}
- The wiring from U, V, W pin should be as thick and as short as possible. They should be connected directly and separated from the main output wires.

2) HVCC, LVCC pin

- The bypass capacitor(good temperature, frequency characteristic electrolytic type C3) should be mounted as close as possible to the pin in order to prevent malfunction or destruction due to switching noise and power supply ripple. In addition, for the purpose of reducing of the power supply's impedance in wide frequency bandwidth, ceramic capacitor (good temperature, frequency and DC bias characteristic ceramic type C2: 0.1µF to 0.22µF) should also be mounted.
- Zenner diode D1(1W) should be mounted between each pair of control supply pins to prevent surge destruction.
- Line ripple voltage should meet dV/dt ≤ 1V/µs, V_{ripple} ≤ 2V_{p-p.}

3) P pin

- To prevent surge destruction, the wiring between the smoothing capacitor and P, N pins should be as short as possible.
- Snubber capacitor(C4: 0.1µF to 0.22µF) should be mounted between the P-N pin.

4) Control Input pin (HINU, HINV, HINW, LINU, LINV, LINW)

- The wiring should be as short as possible to prevent malfunction.
- Input drive is active-high type. There is a 3.3kΩ(Min) pull-down resistor in the input circuit of IPM. When using RC coupling circuit, make sure the input signal level meet the input threshold voltage.
- Dead time of input signal should be more than specified value.

5) FO pin

• FO output is open drain type. It should be pulled up to control power supply(e.g. 5V, 15V) by a resistor that makes I_{FO} up to 1mA. I_{FO} is estimated roughly by the formula of control power supply voltage divided by pull-up resistance(R1). In the case of pulled up to 5V, R1=10k Ω is recommended.

6) CIN pin

- RC filter(R2, C5) should be mounted as close as possible to the pin in order to prevent malfunction by recovery current or switching noise. It is recommended to select tight tolerance, temp-compensated type for RC filter (R2, C5). The time constant R2C5 (1.0µs to 2.0µs is recommended) should be set so that SCP current is shut down within 2µs. Please confirm operation on the actual application since SCP shutdown time changes depending on the PCB wiring pattern.
- The point D at which the wiring to CIN filter is divided should be near the pin of shunt resistor. NU, NV, NW pin should be connected at near NU, NV, NW pin.
- To prevent malfunction, the wiring of B should be as short as possible.

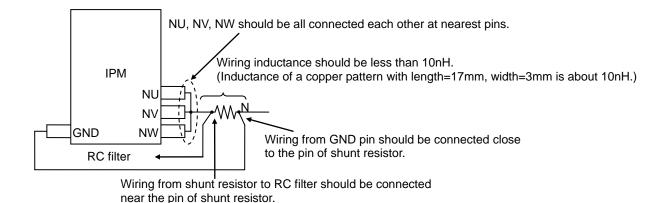
7) GND pin

- Two GND pins (9 & 16 pin) are connected inside IPM. Please connect one pin (16 pin is recommended.) to the 15V power supply GND outside and leave the other open.
- If control GND is connected with power GND by common broad pattern, it may cause malfunction by power GND fluctuation. It is recommended to connect control GND and power GND at only a point N (near the pin of shunt resistor).
- To prevent malfunction, the wiring of A should be as short as possible.

8) NU, NV, NW pin

• When operating with one-shunt resistor, please short the three pins(NU, NV, NW). In addition, to prevent malfunction, the wiring of C should be as short as possible.

9) One-shunt Resistor Drive



Low inductance shunt resistor like surface mounted (SMD) type is recommended.

Figure 15. Wiring Pattern around the Shunt Resistor when Operating with One-shunt Resistor

10) Three-shunt Resistors Drive

- It is not recommended to input the voltage of each shunt resistor directly to the CIN pin when IPM is operated with three shunt resistor. In that case, it is necessary to use the external protection circuit as below.
- It is necessary to set the time constant R_uf_f(1.0µs to 2.0µs is recommended) of external comparator input so that IGBT stops within 2µs when short circuit occurs. Please confirm operation on the actual application since SCP shutdown time changes depending on the PCB wiring pattern.
- It is recommended for the threshold voltage VREF to be set to the same rating of short circuit trip level(V_{SC}=0.48V(Typ))
- To prevent malfunction, the wiring of A, B, C should be as short as possible.
- OR output high level when protection works should be 0.53V(maximum V_{SC} rating) to 7V(CIN absolute maximum rating).

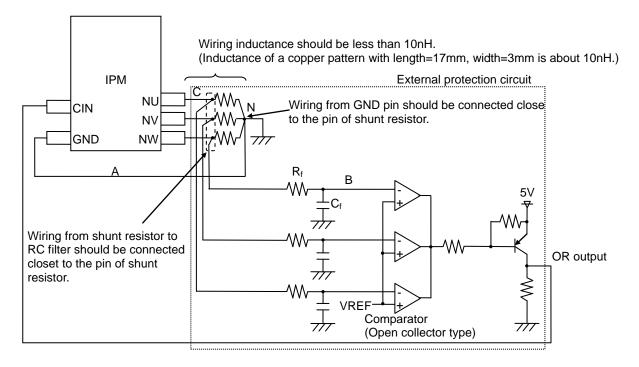


Figure 16. Wiring Pattern around the Shunt Resistor when operating with Three-shunt Resistors

I/O Equivalence Circuit

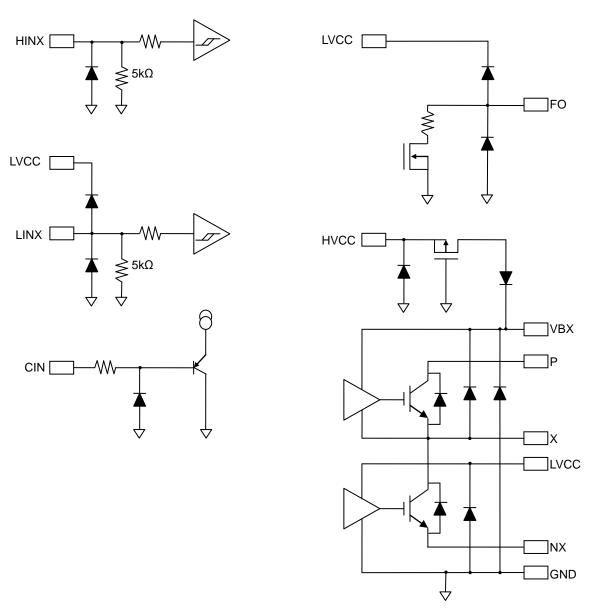


Figure 17. Input / Output Equivalent Circuit (X=U, V, W)

Operational Notes

1. Reverse Connection of Power Supply

Connecting the power supply in reverse polarity can damage the IPM. Take precautions against reverse polarity when connecting the power supply, such as mounting an external diode between the power supply and the IPM's power supply pins.

2. Power Supply Lines

Design the PCB layout pattern to provide low impedance supply lines. Separate the ground and supply lines of the digital and analog blocks to prevent noise in the ground and supply lines of the digital block from affecting the analog block. Furthermore, connect a capacitor to ground at all power supply pins. Consider the effect of temperature and aging on the capacitance value when using electrolytic capacitors.

3. Ground Voltage

Ensure that no pins are at a voltage below that of the ground pin at any time, even during transient condition.

4. Ground Wiring Pattern

When using both small-signal and large-current ground traces, the two ground traces should be routed separately but connected to a single ground at the reference point of the application board to avoid fluctuations in the small-signal ground caused by large currents. Also ensure that the ground traces of external components do not cause variations on the ground voltage. The ground lines must be as short and thick as possible to reduce line impedance.

5. Thermal Consideration

Should by any chance the power dissipation rating be exceeded the rise in temperature of the chip may result in deterioration of the properties of the chip.

6. Recommended Operating Conditions

These conditions represent a range within which the expected characteristics of the IPM can be approximately obtained. The electrical characteristics are guaranteed under the conditions of each parameter.

7. Inrush Current

When power is first supplied to the IPM, it is possible that the internal logic may be unstable and inrush current may flow instantaneously due to the internal powering sequence and delays, especially if the IPM has more than one power supply. Therefore, give special consideration to power coupling capacitance, power wiring, width of ground wiring, and routing of connections.

8. Operation Under Strong Electromagnetic Field

Operating the IPM in the presence of a strong electromagnetic field may cause the IPM to malfunction.

9. Testing on Application Boards

When testing the IPM on an application board, connecting a capacitor directly to a low-impedance output pin may subject the IPM to stress. Always discharge capacitors completely after each process or step. The IPM's power supply should always be turned off completely before connecting or removing it from the test setup during the inspection process. To prevent damage from static discharge, ground the IPM during assembly and use similar precautions during transport and storage.

10. Inter-pin Short and Mounting Errors

Ensure that the direction and position are correct when mounting the IPM on the PCB. Incorrect mounting may result in damaging the IPM. Avoid nearby pins being shorted to each other especially to ground, power supply and output pin. Inter-pin shorts could be due to many reasons such as metal particles, water droplets (in very humid environment) and unintentional solder bridge deposited in between pins during assembly to name a few.

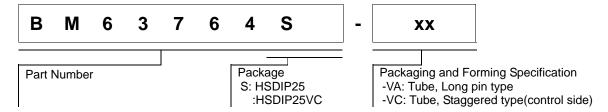
11. Unused Input Pins

Input pins of an IPM are often connected to the gate of a MOS transistor. The gate has extremely high impedance and extremely low capacitance. If left unconnected, the electric field from the outside can easily charge it. The small charge acquired in this way is enough to produce a significant effect on the conduction through the transistor and cause unexpected operation of the IPM. So unless otherwise specified, unused input pins should be connected to the power supply or ground line.

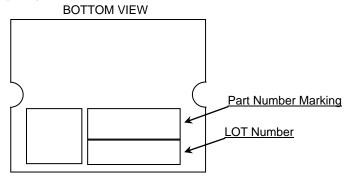
12. Area of Safe Operation (ASO)

Operate the IPM such that the output voltage, output current, and power dissipation are all within the Area of Safe Operation (ASO).

Ordering Information



Marking Diagram



Physical Dimension, Tape and Reel Information Package Name HSDIP25 38 ± 0.5 3. 5±0. 25 20X1.778 (=35.56) 1. 5±0. 05 0.28 16-0.5 0 1. 778±0. 2 3 Type name Lot No HEAT SINK SIDE Specular surface finish 1. 5MIN 8-0.6 2. 54±0. 2 14X2.54 (= 35.56) 0.5 0.5 0.5 HEAT SINK SIDE (2. 756) DETAIL A DETAIL B (UNIT:mm)PKG: HSDIP25 Drawing No. EX641-5002 <Tape and Reel information> Container Tube Quantity 60pcs Direction of feed | Direction of products is fixed in a container tube

*When you order, please order in times the amount of package quantity.

Physical Dimension, Tape and Reel Information - continued **Package Name HSDIP25VC** 38.0±0.5 20×1.778 (=35.56) 3. 5±0. 25 0.28 35.0±0.3 1. 5±0. 05 1. 778±0. 2 16-0.5 2 ± 0 . 0±0. 4±0.5 29. Type name Lot No 0.8 HEAT SINK SIDE Specular surface finish 4-C1. 2 1. 5MIN 8-0.6 2. 54±0. 2 14×2.54 (=35.56) 5. 5±0. 5 9. 5±0. 5 HEAT SINK SIDE ij DETAIL A DETAIL B (UNIT:mm) PKG: HSDIP25VC Drawing No. 641-5003 <Tape and Reel information> Container Tube Quantity 60pcs *When you order, please order in times the amount of package quantity.

Revision History

Date	Revision	Changes
05.Aug.2014	001	New Release
11.Nov.2014	002	BM63764S-VC added

Notice

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(Note1) Medical Equipment Classification of the Specific Applications

JAPAN	USA	EU	CHINA
CLASSⅢ	CLASSII	CLASS II b	CLASSIII
CLASSIV		CLASSⅢ	

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 - [d] Use of our Products in places where the Products are exposed to static electricity or electromagnetic waves
 - [e] Use of our Products in proximity to heat-producing components, plastic cords, or other flammable items
 - [f] Sealing or coating our Products with resin or other coating materials
 - [g] Use of our Products without cleaning residue of flux (even if you use no-clean type fluxes, cleaning residue of flux is recommended); or Washing our Products by using water or water-soluble cleaning agents for cleaning residue after soldering
 - [h] Use of the Products in places subject to dew condensation
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- 7. De-rate Power Dissipation (Pd) depending on Ambient temperature (Ta). When used in sealed area, confirm the actual ambient temperature.
- 8. Confirm that operation temperature is within the specified range described in the product specification.
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- 2. In principle, the reflow soldering method must be used on a surface-mount products, the flow soldering method must be used on a through hole mount products. If the flow soldering method is preferred on a surface-mount products, please consult with the ROHM representative in advance.

For details, please refer to ROHM Mounting specification

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- 1. If change is made to the constant of an external circuit, please allow a sufficient margin considering variations of the characteristics of the Products and external components, including transient characteristics, as well as static characteristics.
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This Product is electrostatic sensitive product, which may be damaged due to electrostatic discharge. Please take proper caution in your manufacturing process and storage so that voltage exceeding the Products maximum rating will not be applied to Products. Please take special care under dry condition (e.g. Grounding of human body / equipment / solder iron, isolation from charged objects, setting of lonizer, friction prevention and temperature / humidity control).

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 - [b] the temperature or humidity exceeds those recommended by ROHM
 - [c] the Products are exposed to direct sunshine or condensation
 - [d] the Products are exposed to high Electrostatic
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- 4. Use Products within the specified time after opening a humidity barrier bag. Baking is required before using Products of which storage time is exceeding the recommended storage time period.

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