

HDMI Receiver Port Protection and Interface Device

CM2030

Features

- HDMI 1.3 compliant
- Supports thin dielectric and 2-layer boards
- Minimizes TMDS skew with 0.05pF matching
- Long HDMI cable support with integrated I²C accelerator
- Active termination and slew rate limiting for CEC
- Supports direct connection to CEC microcontroller
- Integrated I²C level shifting to CMOS level including low logic level voltages
- Integrated 8kV ESD protection and backdrive protection on all external I/O lines
- Integrated overcurrent output protection per HDMI 1.3
- Multiport I²C support eliminates need for analog mux on DDC lines
- Simplified layout with matched 0.5mm trace spacing
- RoHS-compliant, lead-free packaging

Applications

- PC and consumer electronics
- · Set top box, DVD RW, PC, graphics cards

Product Description

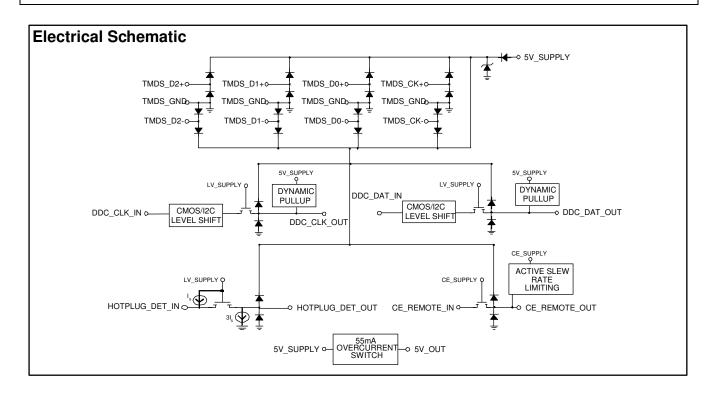
The CM2030 HDMI Transmitter Port Protection and Interface Device is specifically designed for next generation HDMI Host interface protection.

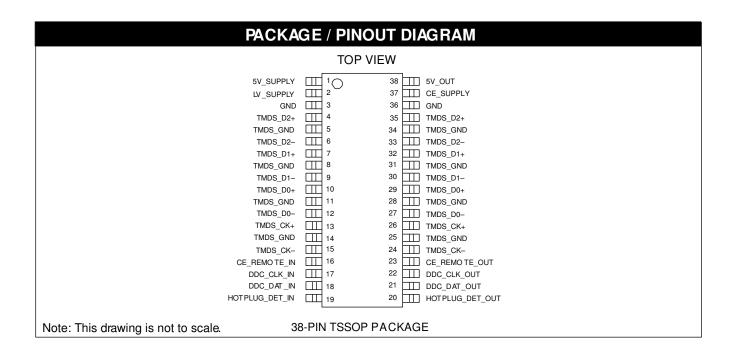
An integrated package provides all ESD, slew rate limiting on CEC line, level shifting/isolation, overcurrent output protection and backdrive protection for an HDMI port in a single 38-pin TSSOP package.

The CM2030 part is specifically designed to provide the designer with the most reliable path to HDMI 1.3 CTS compliance.

The CM2030 also incorporates a silicon overcurrent protection device for +5V supply voltage output to the connector.







	PIN DESCRIPTIONS							
		ESD Level	DESCRIPTION					
4, 35	TMDS_D2+	8kV³	TMDS 0.9pF ESD protection. ¹					
6, 33	TMDS_D2-	8kV ³	TMDS 0.9pF ESD protection.1					
7, 32	TMDS_D1+	8kV³	TMDS 0.9pF ESD protection.1					
9, 30	TMDS_D1-	8kV ³	TMDS 0.9pF ESD protection. ¹					
10, 29	TMDS_D0+	8kV³	TMDS 0.9pF ESD protection.1					
12, 27	TMDS_D0-	8kV ³	TMDS 0.9pF ESD protection.1					
13, 26	TMDS_CK+	8kV ³	TMDS 0.9pF ESD protection. ¹					
15, 24	TMDS_CK-	8kV ³	TMDS 0.9pF ESD protection.1					
16	CE_REMOTE_IN	2kV⁴	CE_SUPPLY referenced logic level in.					
23	CE_REMOTE_OUT	8kV ³	5V_SUPPLY referenced logic level out plus 10pF ESD.6					
17	DDC_CLK_IN	2kV⁴	LV_SUPPLY referenced logic level in.					
22	DDC_CLK_OUT	8kV ³	5V_SUPPLY referenced logic level out plus 10pF ESD.6					
18	DDC_DAT_IN	2kV⁴	LV_SUPPLY referenced logic level in.					
21	DDC_DAT_OUT	8kV ³	5V_SUPPLY referenced logic level out plus 10pF ESD.6					
19	HOTPLUG_DET_IN	2kV⁴	LV_SUPPLY referenced logic level in.					
20	HOTPLUG_DET_OUT	8kV ³	5V_SUPPLY referenced logic level out plus 10pF ESD. A 0.1μF					
			bypass ceramic capacitor is recommended on this pin. ²					
2	LV_SUPPLY	2kV⁴	Bias for CE / DDC / HOTPLUG level shifters.					
37	CE_SUPPLY	2kV ^{4,2}	CEC bias voltage. Previously CM2020 ESD_BYP pin.					
1	5V_SUPPLY	2kV⁴	Current source for 5V_OUT, VREF for DDC I ² C voltage references,					
			and bias for 8kV ESD pins.					
38	5V_OUT	8kV ³	55mA minimum overcurrent protected 5V output. This output must be					
			bypassed with a 0.1μF ceramic capacitor.					
3, 5, 8, 11,	GND / TMDS_GND	N/A	GND reference.					
14, 25,								
28, 31, 34, 36								

- Note 1: These 2 pins need to be connected together in-line on the PCB. See recommended layout diagram.
- Note 2: This output can be connected to an external $0.1\mu F$ ceramic capacitor/pads to maintain backward compatibility with the CM2020.
- Note 3: Standard IEC 61000-4-2, $C_{\text{DISCHARGE}}$ =150pF, $R_{\text{DISCHARGE}}$ =330 Ω , 5V_SUPPLY and LV_SUPPLY within recommended operating conditions, GND=0V, 5V_OUT (pin 38), and HOTPLUG_DET_OUT (pin 20) each bypassed with a 0.1 μ F ceramic capacitor connected to GND.
- Note 4: Human Body Model per MIL-STD-883, Method 3015, $C_{\text{DISCHARGE}} = 100 \text{pF}$, $R_{\text{DISCHARGE}} = 1.5 \text{k}\Omega$, 5V_SUPPLY and LV_SUPPLY within recommended operating conditions, GND=0V, 5V_OUT (pin 38), and HOTPLUG_DET_OUT (pin 20) each bypassed with a $0.1 \mu\text{F}$ ceramic capacitor connected to GND.
- Note 5: These pins should be routed directly to the associated GND pins on the HDMI connector with single point ground vias at the connector.
- Note 6: The slew-rate control and active acceleration circuitry dynamically offsets the system capacitive load on these pins.

Backdrive Protection and Isolation

Backdrive current is defined as the undesirable current flow through an I/O pin when that I/O pin's voltage exceeds the related local supply voltage for that circuitry. This is a potentially common occurrence in multimedia entertainment systems with multiple components and several power plane domains in each system.

For example, if a DVD player is switched off and an HDMI connected TV is powered on, there is a possibility of reverse current flow back into the main power supply rail of the DVD player from pull-ups in the TV. As little as a few milliamps of backdrive current flowing back into the power rail can charge the DVD player's bulk bypass capacitance on the power rail to some intermediate level. If this level rises above the power-on-reset (POR) voltage level of some of the integrated circuits in the DVD player, then these devices may not reset properly when the DVD player is turned back on.

If any SOC devices are incorporated in the design which have built-in level shifter and/or ESD protection structures, there can be a risk of permanent damage due to backdrive. In this case, backdrive current can forward bias the on-chip ESD protection structure. If the current flow is high enough, even as little as a few milliamps, it could destroy one of the SOC chip's internal DRC diodes, as they are not designed for passing DC.

To avoid either of these situations, the CM2030 was designed to block backdrive current, guaranteeing less than 5μA into any I/O pin when the I/O pin voltage exceeds its related operating CM2030 supply voltage.

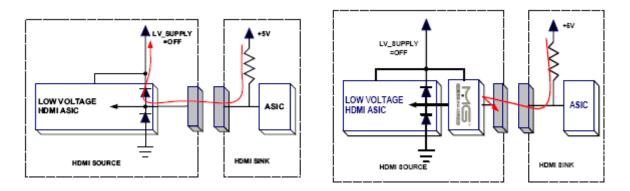


Figure 1. Backdrive Protection Diagram.

Display Data Channel (DDC) lines

The DDC interface is based on the I²C serial bus protocol for EDID configuration.

DYNAMIC PULLUPS

Based on the HDMI specification, the maximum capacitance of the DDC line can approach 800pF (50pF from source, 50pF from sink, and 700pF from cable). At the upper range of capacitance values (i.e. long cables), it becomes impossible for the DDC lines to meet the I^2 C timing specifications with the minimum pull-up resistor of 1.5k Ω .

For this reason, the CM2030 was designed with an internal I²C accelerator to meet the AC timing specification even with very long and non-compliant cables.

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The internal accelerator increases the positive slew rate of the DDC_CLK_OUT and DDC_DAT_OUT lines whenever the sensed voltage level exceeds 0.3*5V_SUPPLY (approximately 1.5V). This provides faster overall risetime in heavily loaded situations without overloading the multi-drop open drain I²C outputs elsewhere.

DYNAMIC PULLUPS (CONT'D)

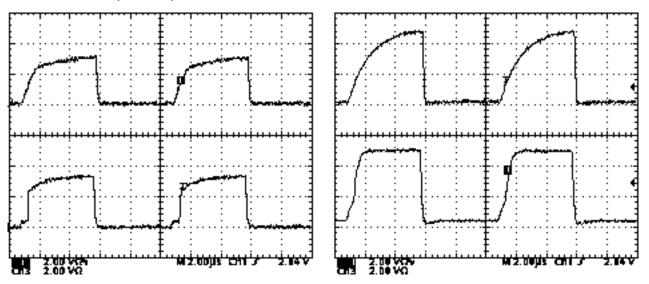


Figure 2. Dynamic DDC Pullups (Discrete - Top, CM2030 - Bottom; 3.3V ASIC - Left, 5V Cable - Right.)

Figure 2 demonstrates the "worst case" operation of the dynamic CM2030 DDC level shifting circuitry (bottom) against a discrete NFET common-gate level shifter circuit with a typical 1.5kW pullup at the source (top.) Both are shown driving an off-spec, but unfortunately readily available 31m HDMI cable which exceeds the 700pF HDMI specification. Some widely available HDMI cables have been measured at *over 4nF*.

When the standard I/OD cell releases the NFET discrete shifter, the risetime is limited by the pullup and the parasitics of the cable, source and sink. For long cables, this can extend the risetime and reduce the margin for reading a valid "high" level on the data line. In this case, an HDMI source may not be able to read uncorrupted data and will not be able to initiate a link.

With the CM2030's dynamic pullups, when the ASIC driver releases its DDC line and the "OUT" line reaches at least 0.3*VDD (of 5V_SUPPLY), then the "OUT" active pullups are enabled and the CM2030 takes over driving the cable until the "OUT" voltage approaches the 5V_SUPPLY rail.

The internal pass element and the dynamic pullups also work together to damp reflections on the longer cables and keep them from glitching the local ASIC.

I²C LOW LEVEL SHIFTING

In addition to the Dynamic Pullups described in the previous section, the CM2030 also incorporates improved I²C low-level shifting on the DDC_CLK_IN and DDC_DAT_IN lines for enhanced compatibility.

Typical discrete NFET level shifters can advertise specifications for low $R_{\scriptscriptstyle DS}$ [on], but usually state relatively high $V_{\scriptscriptstyle [GS]}$ test parameters, requiring a 'switch' signal (gate voltage) as high as 10V or more. At a sink current of 4mA for the ASIC on DDC_XX_IN, the CM2030 guarantees no more than 140mV increase to DDC_XX_OUT, even with a switching control of 2.5V on LV_SUPPLY.

When I²C devices are driving the external cable, an internal pulldown on DDC_XX_IN guarantees that the VOL seen by the ASIC on DDC_XX_IN is equal to or lower than DDC_XX_OUT.

Multiport DDC Multiplexing

By switching LV_SUPPLY, the DDC/HPD blocks can be independently disabled by engaging their inherent "backdrive" protection. This allows N:1 multiplexing of the low-speed HDMI signals without any additional FET switches.

Consumer Electronics Control (CEC)

The Consumer Electronics Control (CEC) line is a high level command and control protocol, based on a single wire multidrop open drain communication bus running at approximately 1kHz (See Figure 3). While the HDMI link provides only a single point-to-point connection, up to ten (10) CEC devices may reside on the bus, and they may be daisy chained out through other physical connectors including other HDMI ports or other dedicated CEC links. The high level protocol of CEC can be implemented in a simple microcontroller or other interface with any I/OD (input/open-drain) GPIO.

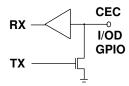


Figure 3. Typical μC I/OD Driver

To limit possible EMI and ringing in this potentially complex connection topology, the rise- and fall-time of this line are limited by the specification. However, meeting the slew-rate limiting requirements with additional discrete circuitry in this bi-directional block is not trivial without an additional RX/TX control line to limit the output slew-rate without affecting the input sensing (See Figure 4).

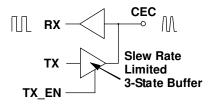


Figure 4. Three-Pin External Buffer Control

Simple CMOS buffers cannot be used in this application since the load can vary so much (total pullup of $27k\Omega$ to less than $2k\Omega$, and up to 7.3nF total capacitance.) The CM2030 targets an output drive slew-rate of less than 100mV/ms regardless of static load for the CEC line. Additionally, the same internal circuitry will perform active termination, thus reducing ringing and overshoot in entertainment systems connected to legacy or poorly designed CEC nodes.

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The CM2030's bi-directional slew rate limiting is integrated into the CEC level-shifter functionality thus allowing the designer to directly interface a simple low voltage CMOS GPIO directly to the CEC bus and simultaneously guarantee meeting all CEC output logic levels and HDMI slew-rate and isolation specifications (See Figure 5).

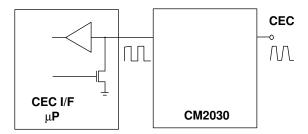


Figure 5. Integrated CM2030 Solution

The CM2030 also includes an internal backdrive protected static pullup 120μA current source from the CE_SUPPLY rail in addition to the dynamic slew rate control circuitry.

Figure 6 shows a typical shaped CM2030 CEC output (bottom) against a ringing uncontrolled discrete solution (top).

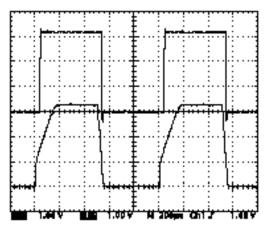


Figure 6. CM2030 CEC Output

Hotplug Detect Logic

The CM2030 ensures that the local ASIC will properly detect an HDMI compliant Sink. The current sink maintains a local logic "low" when no system is connected.

A valid pullup on the HDMI connector pin will overdrive the internal pulldown and deliver a logic "high" to the local ASIC.

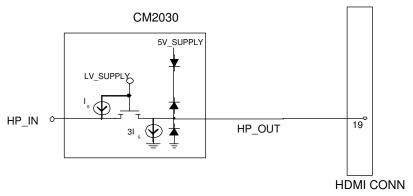


Figure 7. Hotplug Detect Circuit

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Ordering Information

PART NUMBERING INFORMATION						
38	TSSOP-38	CM2030-A0TR	CM2030-A0TR			

Note 1: Parts are shipped in Tape & Reel form unless otherwise specified.

Specifications

ABSOLUTE MAXIMUM RATINGS						
PARAMETER		_				
VCC5, VCCLV	6.0	٧				
DC Voltage at any Channel Input	[GND - 0.5] to [VCC + 0.5]	٧				
Storage Temperature Range	65 to +150	S				

STANDARD (RECOMMENDED) OPERATING CONDITIONS								
SYMBOL	PARAMETER				1			
5V_SUPPLY	Operating Supply Voltage		5	5.5	V			
LV_SUPPLY	Bias Supply Voltage	1	3.3	5.5	V			
CE_SUPPLY	Bias Supply Voltage	3	3.3	3.6	V			
	Operating Temperature Range	40		85	$^{\circ}$			

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
ICC5	Operating Supply Current	5V_SUPPLY = 5.0V,		300	350	μΑ
		CEC_OUT = 3.3V, LV_SUPPLY=				
		3.3V,CE_SUPPLY= 3.3V, DDC=5V;				
		Note6				
ICCLV	Bias Supply Current	LV_SUPPLY=3.3V; Note 6		60	150	μΑ
ICCCE	Bias Supply Current	CE_SUPPLY=3.3V, CEC_OUT=0V;		60	150	μΑ
		Notes 6 and 7				
ICEC	Current source on CEC pin	CE_SUPPLY=3.3V,	111	120	128	μ A
VDROP	5V_OUT Overcurrent Out put	5V_SUPPLY=5.0V, IOUT=55mA		65	100	mV
	Drop					
ISC	5V_OUT Short Circuit Cur	5V_SUPPLY=5.0V, 5V_OUT=GND	90	135	175	mA
	rent Limit					
IOFF	OFF state leakage current,	LV_SUPPLY=0V		0.1	5	μΑ
	level shifting NFET					
IBACKDRIVECEC	Current through CE-	CE-REMOTE_IN = CE_SUPPLY <		0.1	1.8	μΑ
	REMOTE_OUT when	CE_REMOTE_OUT				
	powered down					
IBACKDRIVETMDS	Current through TMDS pins	All Supplies = 0V; TMDS_[2:0]+/,		0.1	5	μΑ
	when powered down	TMDS_CK+/ = 4V				·
IBACKDRIVE5V OUT	Current through 5V OUT	All Supplies = 0V; 5V_OUT_PIN =		0.1	5	μΑ
_	when powered down	5V				·
IBACKDRIVEDDC	Current through	All Supplies = 0V;		0.1	5	μΑ
	DDC DAT/CLK OUT when	DDC_DAT/CLK_OUT = 5V;				•
	powered down	DDC_DAT/CLK_IN = 0V				
IBACKDRIVEHOTPLUG	Current through	All Supplies = 0V;		0.1	5	μΑ
	HOTPLUG_DET_OUT when	HOTPLUG_DET_OUT = 5V;				·
	powered down	HOTPLUG IN = 0V				
CECSL	CEC Slew Limit	Measured from 10-90% or 90-10%		0.26	0.65	V/µs
CECRT	CEC Rise Time	Measured from 10-90%	26.4		250	μs
		Assumes a signal swing from 0-				•
		3.3V				
CECFT	CEC Fall Time	Measured from 90-10%	4		50	μs
		Assumes a signal swing from 0-	-			μ
		3.3V				
VACC	Turn On Threshold of I2C/	Voltage is 0.3 ±10% X 5V_Supply;	1.35	1.5	1.65	V
77.00	DDC Accelerator	Note 2	1.00	1.0	1.00	
VON(DDC OUT)	Voltage drop across DDC	LV SUPPLY=3.3V, 3mA Sink at		150	225	mV
7011(220_001)	level shifter	DDCIN, DDCOUT < VACC				
VOL(DDC IN)	Logic Level (ASIC side) when			0.3	0.4	V
VOL(DDO_IIV)	I2C/DDC Logic Low Applied;	LV SUPPLY=3.3V, 1.5k Ω pullup on		0.3	0.4	v
	(I2C pass-through compatibility)	- ' '				
		DDC_OUT to 5.0V; Note 2				
tr(DDC)	DDC_OUT Line Risetime,	DDC_IN floating,			1	μs
	VACC < VDDC_OUT <	LV_SUPPLY=3.3V, 1.5k Ω pullup on			1	

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	(5V_Supply-0.5V)	DDC_OUT to 5.0V, Bus				
		Capacitance = 1500pF				
VF	Diode Forward Voltage	IF = 8mA, TA = 25 ℃				
	Top Diode		0.6	0.85	0.95	V
	Bottom Diode		0.6	0.85	0.95	V
VESD	ESD Withstand Voltage (IEC)	Pins 4, 7, 10, 13, 20, 21, 22, 23, 24,	±8			kV
		27, 30, 33, TA = 25°C; Note 2				
VESD	ESD Withstand Voltage	Pins 1, 2, 16, 17, 18, 19, 37, 38, TA	±2			kV
	(HBM)	= 25℃				
VCL	Channel Clamp Voltage	TA=25 °C, IPP = 1A, $tP = 8/20\mu S$;				
	Positive Transients	Note 5		11.0		V
	Negative Transients			2.0		V
RDYN	Dynamic Resistance	TA=25°C, IPP = 1A, tP = 8/20μS				
	Positive Transients	Any I/O pin to Ground; Note5		1.4		Ω
	Negative Transients			0.9		Ω
ILEAK	TMDS Channel Leakage	TA = 25℃		0.01	1	μΑ
	Current					
CIN, TMDS	TMDS Channel Input	5V_SUPPLY=5.0V, Measured at		0.9	1.2	рF
	Capacitance	1MHz, VBIAS=2.5V				
Δ CIN, TMDS	TMDS Channel Input	5V_SUPPLY=5.0V, Measured at		0.05		рF
	Capacitance Matching	1MHz, VBIAS=2.5V; Note 4				
CMUTUAL	Mutual Capacitance between	5V SUPPLY=0V, Measured at		0.07		pF
	signal pin and adja cent signal	1MHz, VBIAS=2.5V				·
	pin					
CIN, DDCOUT	Level Shifting Input Capaci	5V SUPPLY=0V,		10		рF
,	tance, Capacitance to GND	Measured at 100KHz, VBIAS=2.5V				•
CIN, CECOUT	Level Shifting Input Capaci	5V SUPPLY=0V,		10		pF
, 0_00	tance, Capacitance to GND	Measured at 100KHz,				۳.
	,	VBIAS=1.65V				
CIN, HPOUT	Level Shifting Input Capaci	5V SUPPLY=0V,		10		pF
3,	tance, Capacitance to GND	Measured at 100KHz,				ρ,
	and, Sapasianos is GIVD	VBIAS=2.5V				

- Note 1: Operating Characteristics are over Standard Operating Conditions unless otherwise specified.
- Note 2: Standard IEC61000-4-2, $C_{\mbox{\scriptsize DISCHARGE}}$ =150pF, $R_{\mbox{\tiny DISCHARGE}}$ =330 Ω , 5V_SUPPLY=5V, 3.3V_SUPPLY=3.3V, LV_SUPPLY=3.3V, GND=0V.
- Note 3: Human Body Model per MIL-STD-883, Method 3015, $C_{\text{DISCHARGE}} = 100 \text{pF}$, $R_{\text{DISCHARGE}} = 1.5 \text{k}\Omega$, 5V_SUPPLY=5V, 3.3V_SUPPLY=3.3V, LV_SUPPLY=3.3V, GND=0V.
- Note 4: Intra-pair matching, each TMDS pair (i.e. D+, D-)
- Note 5 These measurements performed with no external capacitor on $V_p(V_p \text{floating})$
- Note 6: These static measurements do not include AC activity on controlled I/O lines.
- Note 7: This measurement does not inclue supply current for the $120\mu A$ current source on the CEC pin.

Performance Information

Typical Filter Performance (T_A=25 °C, DC Bias=0V, 50 Ohm Environment)

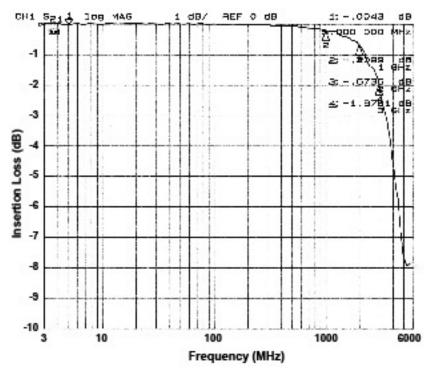


Figure 8. Insertion Loss vs. Frequency (TMDS_D1- to GND)

Application Information

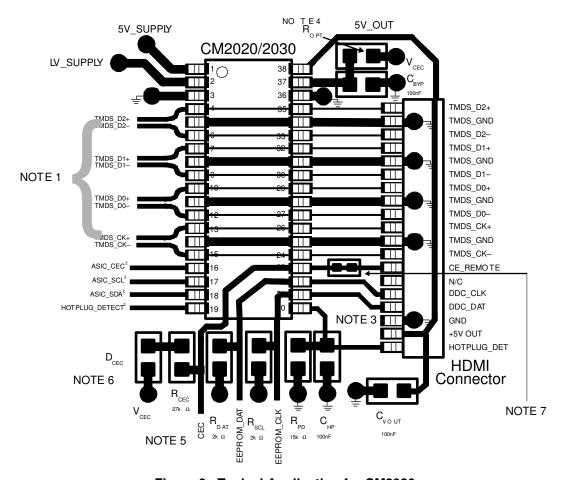


Figure 9. Typical Application for CM2030

LAYOUT NOTES

1. Differential TMDS Pairs should be designed as normal 100Ω HDMI Microstrip. Single Ended (decoupled) TMDS traces underneath MediaGuard[™], and traces between MediaGuard[™] and Connector should be tuned to match chip/connector IBIS parasitics.

² Level Shifter signals should be biased with a weak pullup to the desired local LV_SUPPLY. If the local ASIC includes sufficient pullups to register a logic high, then external pullups may not be needed.

³ Place MediaGuard[™] as close to the connector as possible, and as with any controlled impedance line always avoid placing any silk-screen printing over TMDS traces.

 $^{^4}$ CM2020/CM2030 footprint compatibility - For the CM2030, Pin 37 becomes the V_{CEC} power supply pin for the slew-rate limiting circuitry. This can be supplied by a 0W jumper to V_{CEC} which should be depopulated to utilize the CM2020. The 100nF C_{BYP} is recommended for all applications.

⁵ CEC pullup isolation. The 27k R_{CEC} and a Schottky D_{CEC} provide the necessary isolation for the CEC pullup.



Note: *This circuitry is used only in the CM2020.* Depopulate the components for CM2030 applications in a CM2020/ CM2030 dual footprint layout.

The CM2020 does not not have internal backdrive protection and requires the external R_{cec} and D_{cec} components.

 7 (For CM2030) If CEC firmware *is not* implemented, *do not* populate with $0~\Omega$ resistor. If CEC firmware is implemented, then populate with $0~\Omega$ resistor.

(For CM2020) Populate with 0 Ω resistor in either case.

Application Information (cont'd)

Design Considerations

1. 5V out (pin 38)

Maximum overcurrent protection output drop at 55mA on 5V_OUT is 100mV. To meet HDMI output requirements of 4.8-5.3V, an input of greater than 4.9V should be used (i.e. 5.1V ±4%)

2. DUT On vs. DUT Off

Many HDMI CTS tests require a power off condition on the System Under Test. Many discrete ESD diode configurations can be forward biased when their VDD rail is lower than the I/O pin bias, thereby exhibiting extremely high apparent capacitance measurements, for example. The *MediaGuard*[™] backdrive isolation circuitry limits this current to less than 5mA, and will help ensure HDMI compliance.

Please review all of the current HDMI design guidelines available at:

http://www.calmicro.com/applications/customer/downloads/current-cmd-mediaguard-design-guidelines.zip

⁶ Footprint compatibility - The CM2030 has (built-in) internal backdrive protection.

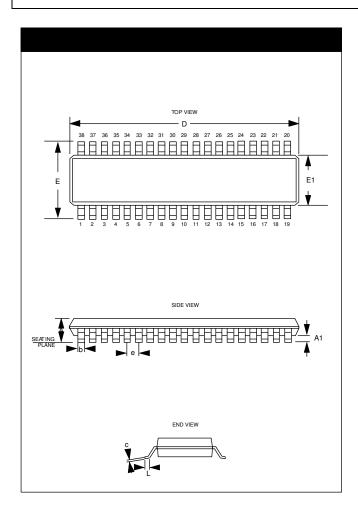
CM2030

Mechanical Details

TSSOP-38 Mechanical Specifications

CM2030 devices are supplied in 38-pin TSSOP packages. Dimensions are presented below.

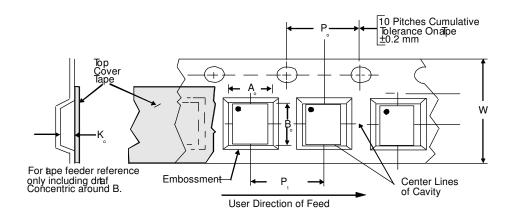
PACKAGE DIMENSIONS								
		TSSOP						
1	N	MO-153 (Variation BD-1)						
		3	8					
İ		I		İ				
	_							
	1	1.20	1	0.047				
	0.05	0.15	0.002	0.006				
	0.17	0.27	0.007	0.011				
	0.09	0.20	0.004	0.008				
1	9.60	9.80	0.378	0.386				
1	6.40	BSC	0.252	2 BSC				
	4.30	4.50	0.169	0.177				
	0.50 BSC 0.020 BSC							
	0.45	0.75	0.018	0.030				
	2500 pieces							
		ì						



Package Dimensions for TSSOP-38

Tape and Reel Specifications

PART NUMBER	PACKAGE SIZE	POCKET SIZE (mm) TAPE WIDTH		REEL	QTY PER	P _o	P ₁
	(mm)	B ₀ X A ₀ X K ₀	W	DIAMETER	REEL		
CM2030	9.70 X 6.40 X 1.20	10.20 X 6.90 X 1.80	16mm	330mm (13")	2500	4mm	12mm



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