

# **General Description**

DA9210 is a multi-phase synchronous step-down converter suitable for supplying the CPU power in automotive in-vehicle infotainment systems, smartphones, mobile computing, and other handheld applications, which require high currents to run the processor core.

DA9210 is designed to operate with four phases, each phase using a small external 0.47  $\mu$ H inductor. The buck is capable of delivering up to 12 A continuous output current with an output voltage of 0.3 V to 1.57 V. The input voltage range of 2.8 V to 5.5 V makes it suited for a wide variety of low voltage systems, including all Li-Ion battery supplied applications. Two DA9210s can be used in parallel to deliver an output current of up to 24 A.

The DA9210 point-of-load remote-sensing feature guarantees the highest accuracy while supporting multiple PCB routing scenarios without loss of performance. The highly-integrated design removes the need for external switching FETs or Schottky diodes.

A programmable soft start-up can be enabled, which limits the inrush current from the input node and ensures a slope controlled activation of the rail.

The Dynamic Voltage Control (DVC) supports adaptive adjustment of the supply voltage dependent on the processor load. This is done via direct register write through the communication interface (I<sup>2</sup>C or SPI compatible), via the dedicated DVC control interface, or via a programmable input pin.

DA9210 integrates over-temperature and over-current protection for increased system reliability, without the need for external sensing components. A Power Good and Over-Current Alarm output informs the CPU of an out range voltage output and if the current exceeds a programmable limit, thereby enabling the processor to reduce its consumption before the supply rail collapses.

# **Key Features**

- 2.8 V to 5.5 V input voltage
- 0.3 V to 1.57 V output voltage
- 12 A output current
- 24 A output current in parallel configuration
- 3 MHz nominal switching frequency
  - □ Enables use of low-profile inductors
- Output voltage accuracy ±2.5 %
- Dynamic Voltage Control (DVC)
- Automatic phase shedding
- Integrated power switches
- Remote sensing at point-of-load
- Power Good and Over-Current Alarm signal

- Interfaces:
  - $\Box$  I<sup>2</sup>C and SPI
  - Dedicated DVC
  - 🗆 GPIO
- Adjustable soft-start
- -40 °C to +125 °C junction temperature operation
- Regulator supervision with automatic undervoltage and over-voltage protection
- AEC Q100 grade 3 automotive option
- Package 48 WLCSP (Route Easy<sup>™</sup>, equivalent to 0.8 mm pitch)
- Package 42 VFBGA 0.8 mm pitch (automotive)

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# **Applications**

- High performance multi-core system-on-chip (SoC) applications
- Smartphones
- Mobile phones
- Ultrabooks<sup>™</sup>

- Tablet PCs
- In-car infotainment/dashboard
- Portable navigation devices
- TV and media players
- Embedded industrial systems

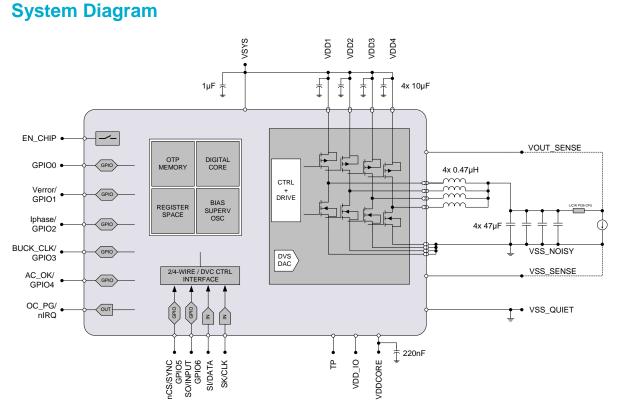


Figure 1: System Diagram

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## DA9210 Multi-Phase 12 A DC-DC Buck Converter

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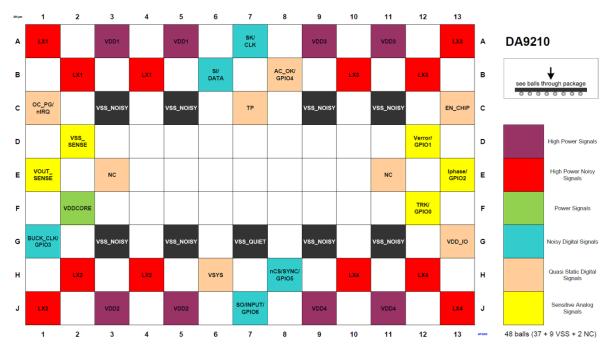
# **1** Terms and Definitions

Central Processing Unit
Dynamic Voltage Control
Field Effect Transistor
General Purpose Input/Output
Interrupt ReQuest
Printed Circuit Board
Pulse Frequency Modulation
Power Management IC
System on (a) Chip



### DA9210 Multi-Phase 12 A DC-DC Buck Converter

## 2 Pinout



## 2.1 Pin Configuration (48 WLCSP)



Pin No.	Pin Name	Alternate Function	Туре	Description
A1, B2, B4	LX1		AO	Switching node for phase 1
J1, H2, H4	LX2		AO	Switching node for phase 2
A13, B10, B12	LX3		AO	Switching node for phase 3
J13, H10, H12	LX4		AO	Switching node for phase 4
A3, A5	VDD1		PS	Supply voltage for phase 1 To be connected to VSYS
J3, J5	VDD2		PS	Supply voltage for phase 2 To be connected to VSYS
A9, A11	VDD3		PS	Supply voltage for phase 3 To be connected to VSYS
J9, J11	VDD4		PS	Supply voltage for phase 4 To be connected to VSYS
C13	EN_CHIP		DI	IC Enable Signal
C1	OC_PG	nIRQ	DO	Output for Over Current Alarm and Power Good signal, IRQ line towards the host

### Table 1: Pin Description (48 WLCSP)

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Pin No.	Pin Name	Alternate Function	Туре	Description
G13	VDD_IO		PS	I/O voltage rail
E1	VOUT_SENSE		AI	Output and sense node for the buck
D2	VSS_SENSE		AI	Ground sense node for the buck
F2	VDDCORE		AO	Regulated supply for internal circuitry 2.5 V (decouple with 220 nF)
F12	GPIO0		AI/DIO	General purpose I/O
D12	VERROR	GPIO1	AIO/DIO	Error amplifier voltage signal for dual parallel mode, general purpose I/O
E13	IPHASE	GPIO2	AIO/DIO	Current distribution signal for dual parallel mode, general purpose I/O
G1	BUCK_CLK	GPIO3	DIO	Buck clock input/output (depending on slave/master function in dual parallel mode), general purpose I/O
B8	AC_OK	GPIO4	DIO	Input from safe charger out to OC_PG signalling, general purpose I/O, input of external 6 MHz clock
H8	nCS/SYNC	GPIO5	DIO	4-WIRE chip select, DVC Interface input clock, general purpose I/O
J7	SO/INPUT	GPIO6	DIO	4-WIRE data output, DVC interface input data, general purpose I/O
B6	SI	DATA	DIO	4-WIRE data input, 2-WIRE data
A7	SK	CLK	DI	4-WIRE/2-WIRE Clock
C7	ТР		DIO	Test pin connect to VSS
E3, E11	NC		VSS	Electrically not connected Connect to VSS
H6	VSYS		PS	Supply for IC and input for voltage supervision
G7	VSS_QUIET		VSS	
C3, C5, G3, G5, C9, C11, G9, G11	VSS_NOISY		VSS	



## DA9210 Multi-Phase 12 A DC-DC Buck Converter

### Table 2: Pin Type Definition

Pin Type	Description	Pin Type	Description
DI	Digital Input	AI	Analog Input
DO	Digital Output	AO	Analog Output
DIO	Digital Input/Output	AIO	Analog Input/Output
DIOD	Digital Input/Output open Drain	BP	Back drive Protection
PU	Fixed Pull-Up resistor	SPU	Switchable Pull-Up resistor
PD	Fixed Pull-Down resistor	SPD	Switchable Pull-Down resistor

## 2.2 Pin Configuration (42 VFBGA)





Pin No.	Pin name	Alternate function	Туре	Description
B1, B2	LX1		AO	Switching node for phase 1
E1, E2	LX2		AO	Switching node for phase 2
B6, B7	LX3		AO	Switching node for phase 3
E6, E7	LX4		AO	Switching node for phase 4

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Pin No.	Pin name	Alternate function	Туре	Description
A1, A2	VDD1		PS	Supply voltage for phase 1
				To be connected to VSYS
F1, F2	VDD2		PS	Supply voltage for phase 2
				To be connected to VSYS
A6, A7	VDD3		PS	Supply voltage for phase 3
				To be connected to VSYS
F6, F7	VDD4		PS	Supply voltage for phase 4
				To be connected to VSYS
A5	EN_CHIP		DI	IC Enable Signal
C5	OC_PG	nIRQ	DO	Output for Over Current Alarm and Power Good signal, IRQ line towards the host
A4	VDD_IO		PS	I/O voltage rail
D4	VOUT_SENSE		AI	Output and Sense node for the buck
C4	VSS_SENSE		AI	Ground Sense node for the buck
F5	VDDCORE		AO	Regulated supply for internal circuitry (decouple with 220 nF)
B5	GPIO0		AI/DIO	General purpose I/O
B4	VERROR	GPIO1	AIO/DIO	Error amplifier voltage signal for dual parallel mode, general purpose I/O
E4	IPHASE	GPIO2	AIO/DIO	Current distribution signal for dual parallel mode, general purpose I/O
E5	BUCK_CLK	GPIO3	DIO	Buck clock input/output (depending on slave/master function in dual parallel mode), general purpose I/O

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Datasheet



### DA9210 Multi-Phase 12 A DC-DC Buck Converter

Pin No.	Pin name	Alternate function	Туре	Description
C3	AC_OK	GPIO4	DIO	Input from safe charger out to OC_PG signalling, general purpose I/O, input of external 6 MHz clock
D3	nCS/SYNC	GPIO5	DIO	4-WIRE chip select, DVC Interface input clock , general purpose I/O
E3	SO/INPUT	GPIO6	DIO	4-WIRE data output, DVC Interface input data, general purpose I/O
B3	SI	DATA	DIO	4-WIRE data input, 2-WIRE data
A3	SK	CLK	DI	4-WIRE/2-WIRE clock
D5	TP		DIO	Test pin, connect to VSS
F3	VSYS		PS	Supply for IC and input for voltage supervision
F4	VSS_QUIET		VSS	
C1, C2, D1, D2, C6, C7, D6, D7	VSS_NOISY		VSS	

### Table 4: Pin Type Definition

Pin type	Description	Pin type	Description
DI	Digital Input	AI	Analog Input
DO	Digital Output	AO	Analog Output
DIO	Digital Input/Output	AIO	Analog Input/Output
DIOD	Digital Input/Output Open Drain	BP	Backdrive Protection
PU	Fixed Pull-Up resistor	SPU	Switchable Pull-Up resistor
PD	Fixed Pull-Down resistor	SPD	Switchable Pull-Down resistor



## 3 Absolute Maximum Ratings

#### **Table 5: Absolute Maximum Ratings**

Parameter	Description	Conditions (Note 1)	Min	Max	Unit
T <sub>STG</sub>	Storage temperature		-65	+150	°C
TJ	Junction temperature		-40	+150 Note 2	°C
V <sub>SYS</sub>	Supply voltage		-0.3	5.5	V
All pins except above			-0.3	V <sub>SYS</sub> + 0.3	V
	ESD protection HBM		2000		V

**Note 1** Stresses beyond those listed under 'Absolute maximum ratings' may cause permanent damage to the device. These are stress ratings only, so functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specification are not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Note 2 See Section 8.6 for more detail.

## 4 **Recommended Operating Conditions**

Parameter	Description	Conditions	Min	Тур	Max	Unit
T <sub>A</sub>	Ambient operating temperature		-40		+85	°C
V <sub>SYS</sub>	Supply voltage		2.8		5.5	V
V <sub>DD_IO</sub>	Supply voltage IO		1.2		3.6 Note 1	V
	Maximum power Dissipation Note 2	Derating factor above T <sub>A</sub> = 70 °C: 33 mW/°C 48 WL-CSP	1815	2310		mW
		Derating factor above T <sub>A</sub> = 70°C: 28 mW/°C 42 VF-BGA	1540	1960		mW

#### **Table 6: Recommended Operating Conditions**

**Note 1** V<sub>DD\_IO</sub> must not exceed V<sub>SYS</sub>

**Note 2** Obtain from simulation on a 2S2P 4L JEDEC Board. Influenced by PCB technology and layout.

All voltages are referenced to  $V_{SS}$  unless otherwise stated. Currents flowing into DA9210 are deemed positive, currents flowing out are deemed negative. All parameters are valid over the recommended temperature range and power supply range unless otherwise noted. Please note that power dissipation must be limited to avoid overheating of DA9210. Maximum power dissipation should not be reached with maximum ambient temperature.

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# **5 Typical Current Consumptions**

### **Table 7: Typical Current Consumption**

Operating Mode	Conditions	Battery (typ)	Unit
OFF mode	EN_CHIP Low	<1	μA
ON mode	EN_CHIP High, BUCK_EN = 0 (excluding the current consumption of the buck)	45	μA

## **6** Electrical Characteristics

#### **Table 8: DC Characteristics**

Parameter	Description	Conditions	Min	Тур	Max	Unit
EN_ON	EN_CHIP Level on		1.1			V
EN_OFF	EN_CHIP Level off				0.35	V
EN_HYST	EN_CHIP Hysteresis			100		mV
I <sub>EN_CHIP</sub>	EN_CHIP Input current	$V_{\text{EN\_CHIP}} \leq 1.1 \text{ V}$		100		nA
t <sub>EN</sub>	IC control start- up time			750		μs
Vddcore	VDDCORE voltage			2.5		V
V <sub>IH</sub>	GPI0-6, SYNC, INPUT, CLK, DATA, (2-WIRE mode) Input High Voltage	VDDCORE mode VDD_IO mode	0.7*V <sub>DDCORE</sub> 0.7*V <sub>DD_IO</sub>			V
VIL	GPI0-6, SYNC, INPUT, CLK, DATA, (2-WIRE mode) Input Low Voltage	VDDCORE mode VDD_IO mode			0.3*Vddcore 0.3*Vdd_io	V
V <sub>IH</sub>	SK, nCS, SI (4- WIRE mode) Input High Voltage		0.7*V <sub>DD_IO</sub>			V



## DA9210 Multi-Phase 12 A DC-DC Buck Converter

Parameter	Description	Conditions	Min	Тур	Max	Unit
VIL	SK, nCS, SI (4- WIRE mode) Input Low Voltage				0.3*V <sub>dd_10</sub>	V
V <sub>OH</sub>	GPO0-6, OC_PG, SO (4- WIRE ode) Output High Voltage	Push-pull mode @1 mA V <sub>DD_IO</sub> ≥1.5 V	0.8*V <sub>DD_IO</sub>			V
V <sub>OL</sub> @1 mA	GP00-6, OC_PG, DATA (2-WIRE mode) SO (4-WIRE mode) Output Low Voltage				0.3	V
V <sub>OL</sub> @3 mA	DATA (2-WIRE mode) Output				0.24	V
V <sub>OL</sub> @20 mA	DATA (2-WIRE mode) Output Low Voltage				0.4	V
C <sub>IN</sub>	CLK, DATA (2- WIRE mode) Input Capacitance			2.5	10	pF
t <sub>SP</sub>	CLK, DATA (2- WIRE mode) Spike Suppression	Fast/Fast+ mode High Speed mode			50 10	ns
t <sub>fDA</sub>	DATA (2-WIRE mode)	Fast @ Cb<550 pF HS @ 10 <cb<100 HS @ Cb&lt;400 pF</cb<100 	20 + 0.1 Cb 10 20		120 40 80	ns
toc_del	OC_PG Delay	From over-current detection to OC_PG port asserted			10	ns
t <sub>OC_ASSERT</sub>	OC_PG Assert	From end of over- current to OC_PG port released	100			ns
t <sub>oc_pd</sub>	OC_PG Power Down	After disabling the buck until OC_PG released			250	μs
f <sub>GPI4</sub>	Digital clock input frequency	At GPIO4	-7 %	6.0	+7 %	MHz
D <sub>GPI4</sub>	Digital clock input duty cycle	At GPIO4		50	10	%

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### 6.1 DC/DC Buck Converter

Unless otherwise noted, the following is valid for VDD = 2.8 V to 5.5 V, COUT = 4x 47  $\mu$ F, local sensing, f = 3 MHz.

#### Table 9: DC/DC Buck Converter Characteristics

Parameter	Description	Conditions	Min	Тур	Мах	Unit
V <sub>DD</sub>	Input voltage	$V_{DD} = V_{SYS}$	2.8		5.5	V
Соит	Output capacitor	Including voltage and temperature coefficient	100	200 (4x47) (8x22)	400	μF
L <sub>BUCK</sub>	Inductor value (per phase)	Including current and temperature dependence	-50 %	0.47	+30 %	μH
V <sub>OUT</sub>	Output voltage Note 1	I <sub>OUT</sub> = 0 mA to IMAX	0.3		1.57	V
		V <sub>DD</sub> = 3.8 V V <sub>OUT</sub> = 0.9 V T <sub>A</sub> = 25 °C no load	-0.5		+0.5	%
	Output voltage accuracy	$V_{DD} = 3.8 V$ $V_{OUT} = 0.9 V$ $T_A = -25 \text{ °C to } +85 \text{ °C}$ no load	-1.0		+1.0	%
		Including static line/load regulation and voltage ripple V <sub>ouT</sub> ≥ 1 V	-2.5		+2.5	%
	_	Including static line/load regulation and voltage ripple V <sub>OUT</sub> < 1 V		25		mV
VTR <sub>LOAD</sub>	Load regulation transient Note 2	I <sub>OUT</sub> = 0 / 5 A dl/dt = 10 A/μs 4-phase operation V <sub>OUT</sub> = 1 V		±2		%
VTR <sub>LINE</sub>	Line regulation transient	$V_{DD}$ = 3.0 to 3.6 V $I_{OUT}$ = $I_{MAX}$ tr = tf = 10 µs		10		mV
	Max resistance PCB for remote sensing Note 3	From output capacitor to sense connection at point of load		10		mΩ
	Max inductance PCB for remote sensing Note 4	From output capacitor to sense connection at point of load		10		nH
R <sub>FB</sub>	Feedback comparator input impedance		500			kΩ
I <sub>MAX</sub>	Output current	4-phase operation	12000			mA

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**Revision 2.1** 

28-Oct-2016



## DA9210 Multi-Phase 12 A DC-DC Buck Converter

Parameter	Description	Conditions	Min	Тур	Max	Unit
ILIM	Current limit per phase	BUCK_ILIM = 0010	-20 %	2000 Note 6	20 %	mA
	(programmable) Note 5	BUCK_ILIM = 1111	-20 %	4600	20 %	mA
Ialarm	Current alarm threshold (programmable)	BUCK_IALARM = 0 and BUCK_ILIM ≥ 1010	ILIM - 600	ILIM – 300	ILIM - 100	mA
		BUCK_IALARM = 1 and BUCK_ILIM ≥ 1010	ILIM - 1000	ILIM – 600	ILIM - 300	mA
IQ <sub>ON</sub>	Quiescent current in synchronous rectification mode	4-phase operation No load V <sub>DD</sub> = 3.7 V		60		mA
	Switching frequency		2.79	3	3.21	MHz
	Minimum on time			20		ns
t <sub>ON</sub>	Turn-on time	STARTUP_CTRL = 011		50		μs
	Output pull-down resistor	Can be switched off via BUCK_PD_DIS		150	200	Ω

**Note 1** Programmable in 10 mV increments

**Note 2** Additional to the DC accuracy. The value is intended measured directly at C<sub>OUT</sub>. In case of remote sensing, parasitics of PCB and external components may affect this value.

- Note 3 10 m $\Omega$  equivalent to a 5 inch (ca 13 cm) copper trace ( $\rho$  = 1.7x10-8  $\Omega/m$ ), width 6 mm, thickness 35  $\mu m$
- Note 4 10 nH equivalent to a 5 inch (ca 13 cm) trace routed over a ground plane (approx. 1.2 nH/cm)
- **Note 5** Peak current on the inductor
- Note 6 Minimum value of the accuracy is ±400 mA under all conditions

### 6.2 2-WIRE Control Bus

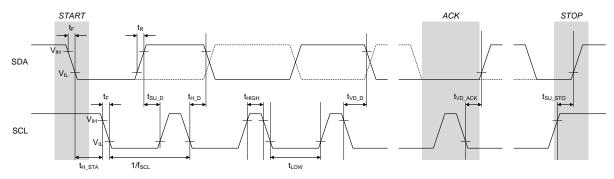


Figure 4: 2-WIRE Interface Timing

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### Table 10: 2-WIRE Interface Electrical Characteristics

Parameter	Description	Test conditions	Min	Тур	Max	Unit
t <sub>BUF</sub>	Bus free time STOP to START		0.5			μs
C <sub>B</sub>	Bus line capacitive load				150	pF
	Standard/Fas	t/Fast mode				
f <sub>SCL</sub>	SCL clock frequency	Note 1	0		1000	kHz
t <sub>SU_STA</sub>	Start condition set-up time		0.26			μs
t <sub>H_STA</sub>	Start condition hold time		0.26			μs
t <sub>W_CL</sub>	SCL low time		0.5			μs
t <sub>w_CH</sub>	SCL high time		0.26			μs
t <sub>R</sub>	2-WIRE SCL and SDA rise time	(input requirement)			1000	ns
t <sub>F</sub>	2-WIRE SCL and SDA fall time	(input requirement)			300	ns
t <sub>SU_D</sub>	Data set-up time		50			ns
t <sub>H_D</sub>	Data hold time		0			ns
t <sub>VD_D</sub>	Data valid time				0.45	μs
t <sub>VD_ACK</sub>	Data valid time acknowledge				0.45	μs
t <sub>SU_STO</sub>	Stop condition set-up time		0.26			μs
	High Spee	ed mode				
f <sub>SCL</sub>	SCL clock frequency	Requires V <sub>DDIO</sub> ≥ 1.8 V Note 1	0		3400	kHz
t <sub>SU_STA</sub>	Start condition set-up time		160			ns
t <sub>H_STA</sub>	Start condition hold time		160			ns
t <sub>W_CL</sub>	SCL low time		160			ns
t <sub>W_CH</sub>	SCL high time		60			ns
t <sub>R</sub>	2-wire SCL and SDA rise time	(input requirement)			160	ns
t <sub>F</sub>	2-wire SCL and SDA fall time	(input requirement)			160	ns
t <sub>SU_D</sub>	Data set-up time		10			ns
t <sub>H_D</sub>	Data hold-time		0			ns
t <sub>SU_STO</sub>	Start condition hold time		160			ns

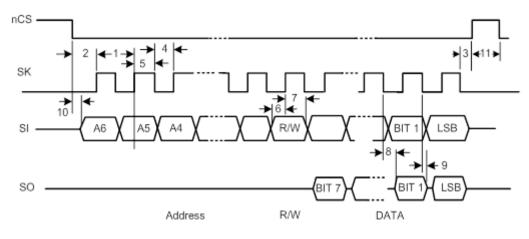
**Note 1** Minimum clock frequency is 10 kHz if 2WIRE\_TO is enabled

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### 6.3 4-WIRE Control Bus



### Figure 5: 4-WIRE Bus Timing

NOTE
The above timing is valid for active low and high CS.

Table 11: 4-WIRE Interface	<b>Electrical Characteristics</b>
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Parameter	Description	Label in Plot	Min	Тур	Max	Unit
C <sub>LD</sub>	Bus line capacitive load				100	pF
t <sub>C</sub>	Cycle time	1	70			ns
t <sub>CSS</sub>	Enable lead time	2, from CS active to first SK edge	20			ns
t <sub>SCS</sub>	Enable lag time	3, from last SK edge to CS idle	20			ns
t <sub>CL</sub>	Clock low time	4	0.4 x tC			ns
t <sub>CH</sub>	Clock high time	5	0.4 x tC			ns
t <sub>SIS</sub>	Data in setup time	6	5			ns
t <sub>SIH</sub>	Data in hold time	7	5			ns
t <sub>SOV</sub>	Data out valid time	8			22	ns
t <sub>soн</sub>	Data out hold time	9	6			ns
t <sub>WCS</sub>	CS inactive time	11	20			ns



### DA9210 Multi-Phase 12 A DC-DC Buck Converter

### 6.4 DVC Interface

#### Table 12: DVC Interface

Parameter	Description	Conditions	Min	Тур	Max	Unit
T <sub>SYNC</sub>	Clock period at SYNC port		40		300	ns
t <sub>PWH</sub>	High pulse width at SYNC port		12		180	ns
t <sub>PWL</sub>	Low pulse width at SYNC port	12			180	ns
D <sub>SYNC</sub>	Duty cycle at SYNC port		40		60	%
t <sub>SETUP</sub>	Set-up time INPUT to rising edge of SYNC		10			ns
thold	Hold time	Hold time INPUT from rising edge 2 of SYNC 2				ns
t <sub>RISE</sub>	Rise time	INPUT and SYNC ports	1			ns
t <sub>FALL</sub>	Fall time	INPUT and SYNC ports	1			ns

### 6.5 **Power Good and Temperature Supervision**

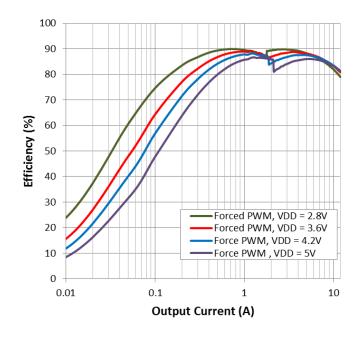
#### Table 13: Power Good and Temperature Supervision

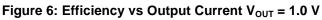
Parameter	Description	Conditions	Min	Тур	Max	Unit
V <sub>GOOD</sub>	Power good low threshold			V <sub>оυт</sub> (Тур) - 0.05		V
V <sub>GOOD</sub> _HYST	Power good low threshold hyst			50		mV
TEMP_WARN Note 1	Thermal warning		110	125	140	°C
TEMP_CRIT	Thermal shutdown threshold		125	140	155	°C
TEMP_POR	Thermal POR threshold		135	150	165	°C

Note 1 Thermal thresholds are non-overlapping



## 7 **Typical Characteristics**





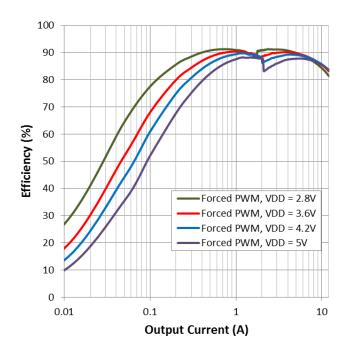


Figure 7: Efficiency vs Output Current  $V_{OUT} = 1.2 V$ 

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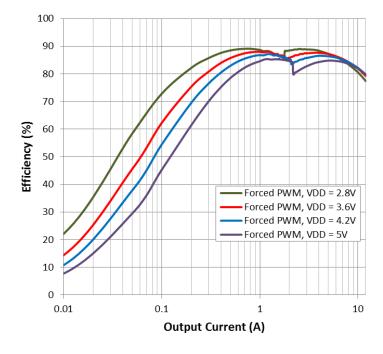


Figure 8: Efficiency vs Output Current V<sub>OUT</sub> = 0.9 V

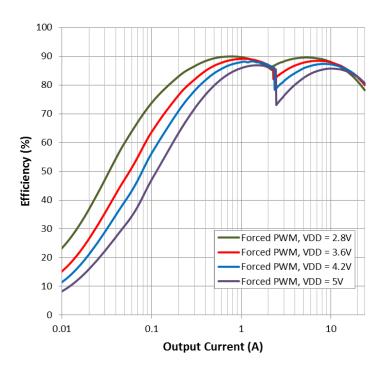


Figure 9: Efficiency vs Output Current  $V_{OUT} = 1.0 V$ , Dual Parallel Mode (24 A max)

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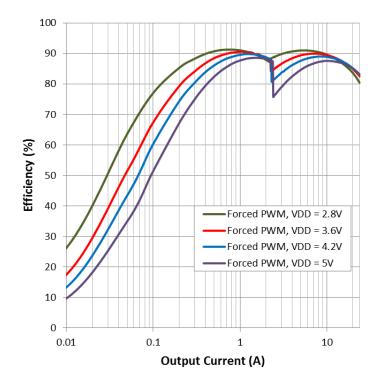


Figure 10: Efficiency vs Output Current V<sub>OUT</sub> = 1.2 V, Dual Parallel Mode (24 A max)

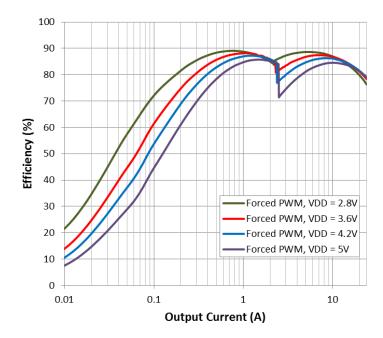


Figure 11: Efficiency vs Output Current  $V_{OUT}$  = 0.9 V, Dual Parallel Mode (24 A max)

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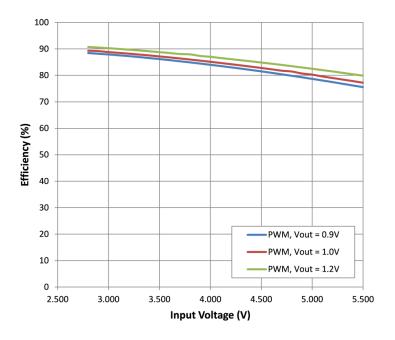
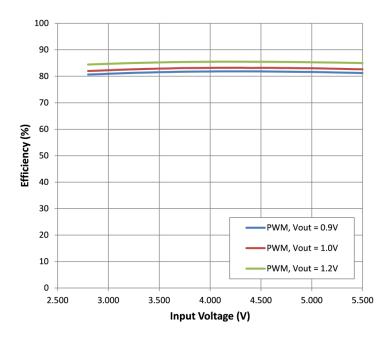


Figure 12: Efficiency vs Input Voltage I<sub>OUT</sub> = 2 A





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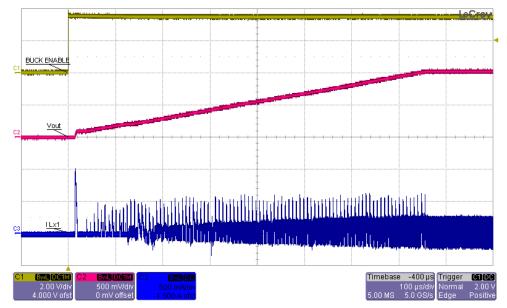


Figure 14: Start-up No Load, STARTUP\_CTRL = 000 (slowest),  $V_{DD}$  = 3.6 V,  $V_{OUT}$  = 1.0 V

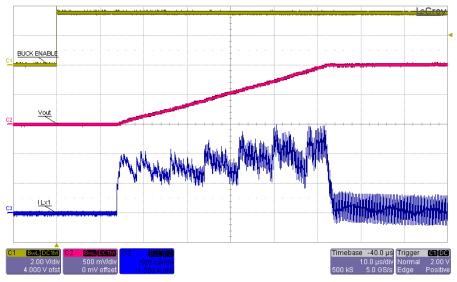


Figure 15: Start-up No Load, STARTUP\_CTRL = 100,  $V_{DD}$  = 3.6 V,  $V_{OUT}$  = 1.0 V

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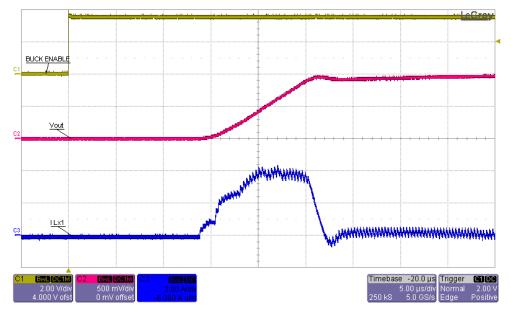


Figure 16: Start-up No Load, STARTUP\_CTRL = 111 (fastest), V<sub>DD</sub> = 3.6 V, V<sub>OUT</sub> = 1.0 V

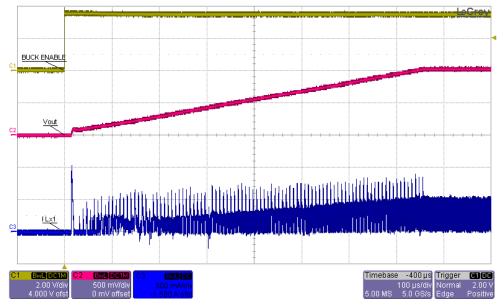


Figure 17: Start-up 1 A Load, STARTUP\_CTRL = 000 (slowest),  $V_{DD}$  = 3.6 V,  $V_{OUT}$  = 1.0 V



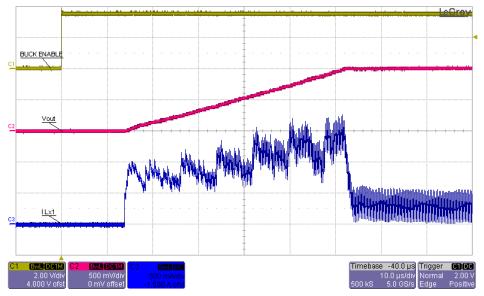


Figure 18: Start-up 1 A Load, STARTUP\_CTRL = 100, V<sub>DD</sub> = 3.6 V, V<sub>OUT</sub> = 1.0 V

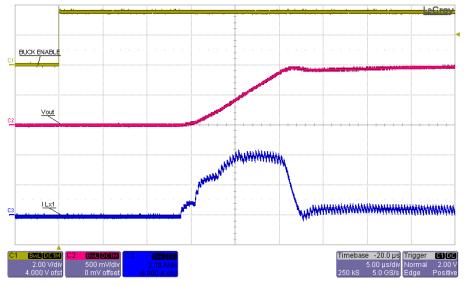


Figure 19: Start-up 1 A Load, STARTUP\_CTRL = 111 (fastest),  $V_{DD}$  = 3.6 V,  $V_{OUT}$  = 1.0 V

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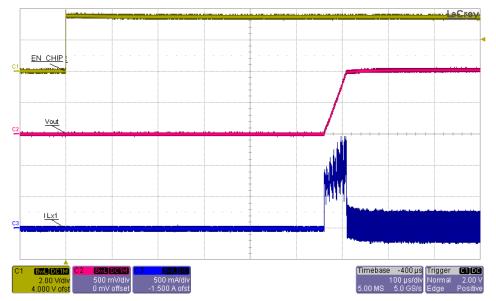


Figure 20: Start-up from EN\_CHIP, no Load, STARTUP\_CTRL = 100, V<sub>DD</sub> = 3.6 V, V<sub>OUT</sub> = 1.0 V

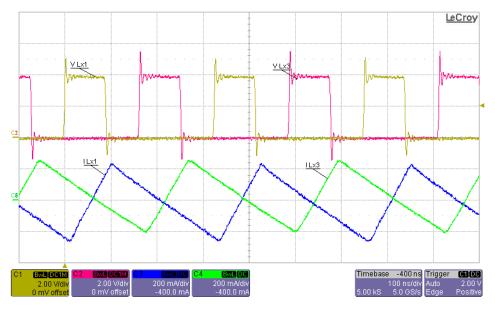


Figure 21: Switching Waveforms, PWM, no Load,  $V_{DD}$  = 3.6 V,  $V_{OUT}$  = 1.0 V



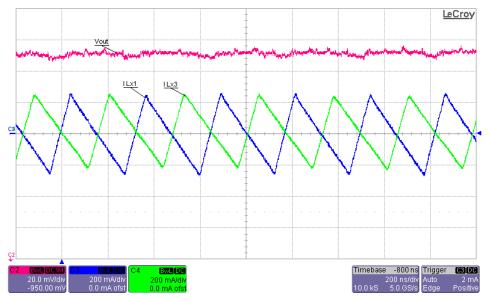


Figure 22: Voltage and Current Ripple, PWM, no Load,  $V_{DD}$  = 3.6 V,  $V_{OUT}$  = 1.0 V

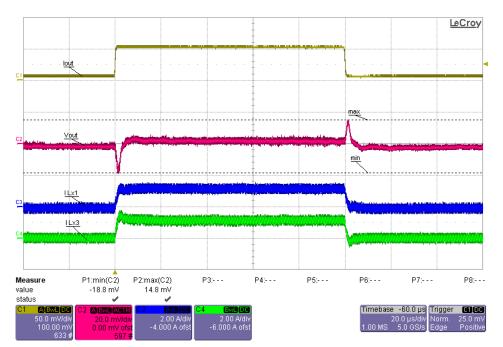


Figure 23: Transient Load, PWM, Dual Mode, 8-phases 5 A to 17 A (12 A/µs), V\_{DD} = 3.7 V, V\_{OUT} = 1.0 V

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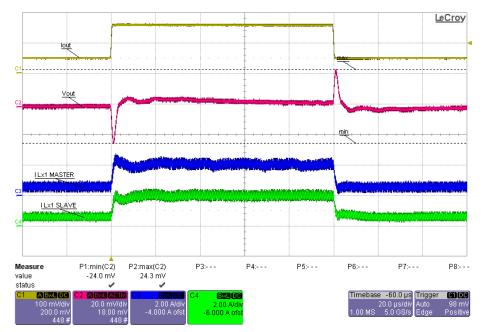
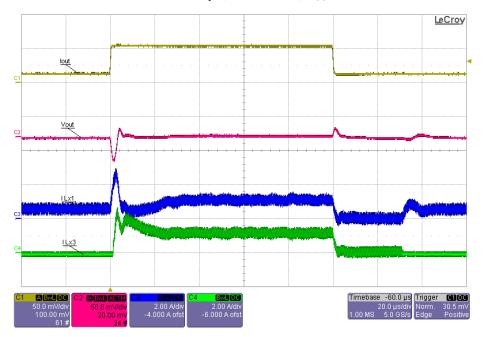
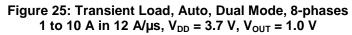


Figure 24: Transient Load, PWM, Dual Mode, 8-phases 5 to 17 A in 12 A/ $\mu$ s, V<sub>DD</sub> = 3.7 V, V<sub>OUT</sub> = 1.0 V





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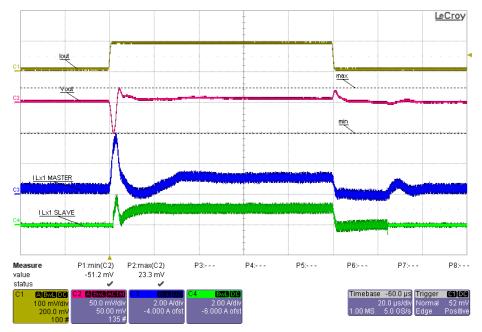
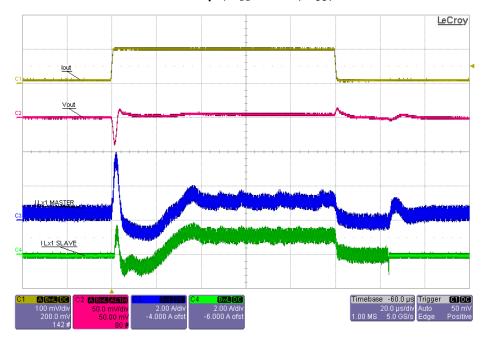
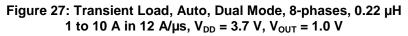


Figure 26: Transient Load, Auto, 4-phases 1 to 5 A in 10 A/ $\mu$ s, V<sub>DD</sub> = 3.7 V, V<sub>OUT</sub> = 1.0 V





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# 8 Functional Description

The DA9210 quad-phase buck converter has been designed to operate either as a high-performance stand-alone regulator or as a sub-PMIC that extends the functionality of an integrated system PMIC such as the Dialog DA9063.

In stand-alone operation, the DA9210 provides:

- high performance, 12 A output current capability,
- high efficiency, quad-phase operation with phase-shedding,
- a small footprint,
- an Over-Current Alarm/ Power Good signal to provide real-time status information to the host processor. In the case of an over-current event or loss of power, the host processor is able to react to maximize system integrity.

When operated with a Dialog system PMIC such as the DA9063, the system also benefits from the following:

- The DA9210 can be enabled as part of the system start-up sequence by utilizing one of the DA9063 sequenced GPIO signals.
- The DA9210 has been designed to operate seamlessly with the DA9063 by sharing the same control interface (same SPI chip select or I<sup>2</sup>C address).
- The DA9210 register map has been designed to interleave with the DA9063 register map. When operated in this way, the two devices appear as a single power management solution to the host processor, thereby simplifying system power control.

By using the general 2-WIRE interface, the DA9210 can easily be integrated into power management systems using system PMICs other than DA9063.

Section 8.1 provides details of the individual blocks of the DA9210 and the configurability available to optimize its performance in any application.

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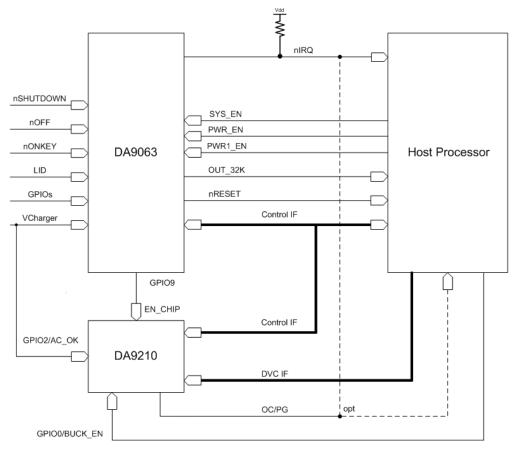


Figure 28: Control Ports and Interface

As shown in Figure 28, a typical application case includes a host processor, a main PMIC (for example DA9063) and the DA9210 used as companion IC for the high power core supply.

The easiest way of controlling the DA9210 is through the control interface. The host processor is the master that initiates communication and reads and writes to and from the main PMIC's and DA9210's registers. To poll the status of DA9210, the host processor must access the dedicated registers area through the control interface.

Additionally, DA9210 can be controlled by means of hardware inputs. A dedicated hardware signal from the DA9210 to the external host processor is implemented through the OC/PG line.

### 8.1 DC/DC Buck Converter

The buck converter is a four-phase, high efficiency, synchronous step down DVC regulator, typically operating at 3 MHz.

The buck converter supports the sensing of the configured voltage directly at the point of load (see VOUT\_SENSE and VSS\_SENSE pins, not supported in tracking mode). The default output voltage is loaded from OTP.

DVC operates in PWM mode (synchronous rectification). The completion of a DVC transition is signalled by GPIO3, assuming the port is configured as GPO and the control bit READY\_EN is asserted.

The buck converter has two voltage registers for output voltage A and B. The appropriate values are stored in the registers VBUCK\_A and VBUCK\_B. The specific output voltage is selected with the bit

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BUCK\_SEL in register BUCK\_CONF. This can be operated either via GPI or via the control interface, according to the configuration of VBUCK\_GPI.

A DVC transition occurs:

- when the selected voltage VBUCK\_A/VBUCK\_B is updated to a new target value
- when the voltage selection is changed via BUCK\_SEL from VBUCK\_A to VBUCK\_B or vice versa

The slew rate of the DVC transition is programmed at 10 mV per (4, 2, 1 or 0.5) µs via SLEW\_RATE control bits.

The typical input current when four phases are enabled is in the order of 60 mA and drops to < 1  $\mu$ A when the buck output is disabled.

### 8.1.1 Switching Frequency

The 3 MHz switching frequency has been chosen to allow the use of a small 0.47 µH inductor (see the complete list of inductors in Section 10). The buck switching frequency can be tuned via register bits OSC\_TUNE. This tunes the internal 6 MHz oscillator frequency in steps of 180 kHz. This impacts the buck converter frequency in steps of 90 kHz. This is used to avoid possible disturbances to other HF systems in the application. If a digital input clock is applied at GPIO4 and the port is configured accordingly on register bit GPIO4\_PIN it is possible to apply an external oscillator delivering 6 MHz to the system, thereby allowing multiple devices to be synchronized to the same clock source.



### 8.1.2 Operation Modes and Phase Selection

The operating mode of the buck converter is selected via bits BUCK\_MODE in register BUCK\_CONF1. The buck converter can be forced to operate in either Normal mode, where all four phases are enabled, or in low-power mode, where the efficiency is optimized for output currents lower than 1 A. In low-power mode, the buck can be forced to operate either with one (default), two, or four phases active.

#### NOTE

The low-power mode configuration is programmed in the device OTP and cannot be changed during normal operation. Please contact your local Dialog Semiconductor support for more information.

The BUCK\_MODE bits also allow an automatic mode based on the output voltage to be selected. With the BUCK\_MODE configured this way the buck will operate in low-power mode as long as the target voltage is lower than the threshold defined in VBUCK\_AUTO, or in Normal mode otherwise.

If BUCK\_MODE bits = 00, the operating mode is selected dependent on the register bit BUCK\_SL\_A/B, so the operating mode is set simultaneously with the output voltage. In low-power mode, the buck operates according to different options as previously described.

The number of active phases can be selected by register bits PHASE\_SEL in register BUCK\_CONF2. This optimizes the efficiency according to the specific output current needed for the application.

If the bit PHASE\_SEL is asserted, automatic phase shedding based on the output current load is enabled. The buck automatically changes between 1-phase, 2-phase, and 4-phase operation, thereby optimizing the efficiency in a wide range of output currents. Automatic phase shedding works only in normal mode and the PHASE\_SEL field configuration is ignored.

When operating in dual parallel mode, the master and the slave device will have up to a total of 8 phases enabled. If a phase is never selected for a certain application, the corresponding LX output pin should be left floating and not connected to any external inductor.

#### NOTE

PFM mode has been withdrawn. Further information on PFM mode limitations in DA9210 can be found in Application Note AN-PM-052.

### 8.1.3 Output Voltage Selection

The switching converter can be configured using the 2-WIRE or 4-WIRE interface, or via the dedicated DVC interface.

The DA9210 provides the capability to set two output voltage levels via registers VBUCK\_A and VBUCK\_B. It is then possible to transition between these two voltages by toggling the register bit BUCK\_SEL, or by using a GPI, selected from GPI0, GPI3, or GPI4 in register bit VBUCK\_GPI. In addition to setting the output voltage, the VBUCK\_A/B registers include the BUCK\_MODE setting which allows the selection between normal and low power modes.

When triggered, the transition will ramp between the set voltages following the slew rate set by the SLEW\_RATE setting in register CONTROL\_A.

The register VBUCK\_MAX will limit the output voltage that can be set for the buck converter.

In addition to triggering an A to B voltage transition, the host is able to modify the output voltage by writing to the currently active VBUCK\_A/B voltage setting. The current SLEW\_RATE setting will be applied to any change triggered by this method.

For security reasons the re-programming of registers that may cause damage if wrongly programmed (for example, voltage settings) can be disabled by asserting the control V\_LOCK in the CONTROL\_A

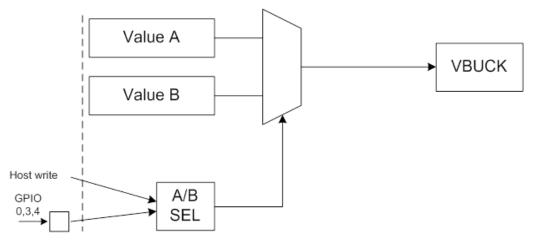
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register. When V\_LOCK is asserted, reprogramming the registers 0xD0 to 0x14F from control interfaces is disabled.

DA9210 implements a dedicated control interface supporting direct DVC requests to the buck converter (see also the detailed description in Section 8.5.3).

When the buck is disabled a 150  $\Omega$  (typ) pull-down resistor is activated for each phase dependent on the value stored in register bit BUCK\_PD\_DIS. Phases disabled via PHASE\_SEL will not have any pull-down. The pull-down resistor is always disabled on all phases when DA9210 is in OFF mode.





The buck current limit should be configured to be at least 40 % higher than the required maximum continuous output current.

Min. ISAT (mA)	Frequency (MHz)	Buck current limit (mA)	Average current (mA)
5060	3	4600	3300
4180	3	3800	2700
3080	3	2800	2000
1760	3	1600	1100

#### Table 14: Selection of Buck Current Limit from Coil Parameters

#### 8.1.4 Soft Start-up

To limit the in-rush current from VSYS input, the buck converter can perform a soft start after being enabled. The start-up behavior is a trade-off between acceptable in-rush current from the battery and turn-on time. In DA9210 different options can be selected using STARTUP bits in BUCK\_CONF1 register. Rates faster than 20 mV/µs may produce overshoot during the start-up phase, so they should be considered carefully.

A ramped power-down can be selected on register bit PWR\_DOWN\_CTRL. If no ramp is selected, the output node will only be discharged by the pull-down resistor, if enabled via BUCK\_PD\_DIS.

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# 8.2 **Ports Description**

#### 8.2.1 VDD\_IO Rail

VDD\_IO is an independent IO supply rail input that can be assigned to the power manager interface and to the GPIOs (see control PM\_IF\_V and GPI\_V). The rail assignment determines the IO voltage levels and logical thresholds (see also the Digital I/O Characteristics).

#### NOTE

The maximum speed of the 4-WIRE interface is only available if the selected supply rail is greater than 1.6 V.

### 8.2.2 EN\_CHIP

EN\_CHIP is a general enable signal for DA9210, turning on and off the internal circuitry (for example the reference, the digital core, and so on). The control of this port has a direct influence on the quiescent current of the whole application and a low level allows the device to reach the minimum quiescent current state. The voltage at this pin is continuously sensed by a dedicated analog circuit.

The EN\_CHIP activation threshold is defined with a built-in hysteresis to avoid erroneous transitions being triggered by unstable rising or falling edges. The EN\_CHIP port has an integrated pull-down current.

#### 8.2.3 GPIO0

The port behaves like a GPIO extender pin (see Section 8.4). In a typical application GPIO0 can be used as GPI port enable for the buck converter, as shown in Figure 28.

#### 8.2.4 VERROR / GPIO1

This port is multi-functional depending on the configuration of GPIO1\_PIN. It can be used as an analog pin to support the dual parallel mode operation of DA9210 (input in case of slave, output in case of master operation, see also Section 8.3.3). Alternatively it behaves like a standard GPIO extender pin, see Section 8.4.

If GPIO1\_PIN = 01 the port will be configured for VERROR operation operating as an input or an output, depending on the voltage level on the VOUT\_SENSE Pin, see Section 8.3.3. In master mode, the port is an output. In slave mode (VOUT\_SENSE must be tied high to VDDCORE), the port is an input.

If GPIO1\_PIN = 01 and master mode is configured, whenever the buck converter is enabled, the signal at the output of the buck error amplifier is internally routed to the VERROR pin and is available externally. This allows the DA9210 to operate as a master together with another slave instance of DA9210 in dual DA9210 operation.

If GPIO1\_PIN = 01 and slave mode is configured, whenever the buck converter is enabled, the signal applied at the VERROR pin will be internally routed as a replacement for the buck error amplifier output, thereby overriding the functionality of the amplifier in the regulation loop. This allows the DA9210 to operate as a slave together with another master instance of DA9210 in dual DA9210 operation.

#### 8.2.5 IPHASE / GPIO2

This port is multi-functional according to the configuration of GPIO2\_PIN. It can be used as an analog pin to support the dual parallel mode operation of DA9210 (input in the case of operation as a slave, output in the case of operation as a master, see Section 8.3.3. Alternatively it behaves like a standard GPIO extender pin, see Section 8.4.

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If GPIO2\_PIN = 01 the port will be configured for IPHASE operation as an input or an output, according to the voltage level on the VOUT\_SENSE Pin, see Section 8.3.3. In master mode the port is an output. In slave mode (VOUT\_SENSE must be tied high to VDDCORE) the port is an input.

If GPIO2\_PIN = 01 and master mode is configured, whenever the buck converter is enabled, the signal from the current sense amplifier of one phase is internally routed to the IPHASE pin and is available externally. This allows the DA9210 to operate as a master together with another slave instance of DA9210 in dual DA9210 operation.

If GPIO2\_PIN = 01 and slave mode is configured, whenever the buck converter is enabled, the signal applied at the IPHASE pin will be internally routed as a replacement for the current sense amplifier of one phase, thereby overriding the function of that current sense amplifier in the current regulation loop. This allows the DA9210 to operate as a slave together with another master instance of DA9210 in dual DA9210 operation.

#### 8.2.6 BUCK\_CLK / GPIO3

This port is multi-functional according to the configuration GPIO3\_PIN. It can be used as a digital pin to support the dual parallel mode operation of DA9210 (input in the case of operation as a slave, output in the case of operation as a master, see Section 8.3.3. Alternatively it behaves like a standard GPIO extender pin, see Section 8.4.

If GPIO3\_PIN = 01 the port will be configured for buck clock operation as an input or an output, according to the voltage level at VOUT\_SENSE pin, see Section 8.3.3. In master mode the port is an output. In slave mode (VOUT\_SENSE must be tied high to VDDCORE) the port is an input.

If GPIO3\_PIN = 01 and master mode is configured, whenever the buck converter is enabled, the clock signal of the buck converter is internally routed to the BUCK\_CLK pin and is available externally. This allows use of the DA9210 in dual operation as a master together with another slave instance of DA9210.

If GPIO3\_PIN = 01 and slave mode is configured, whenever the buck converter is enabled, the signal applied at BUCK\_CLK pin will be internally routed to the clock generation block of the buck converter, thereby overriding the functionality of the internal buck clock. This allows the DA9210 to operate as a slave together with another master instance of DA9210 in dual DA9210 operation.

### 8.2.7 Digital External Clock Input / GPIO4

This port is multi-functional according to the configuration of GPIO4\_PIN. It can be used as digital external clock input for the system, if a signal of typically 6 MHz is applied to the pin.

#### NOTE

The clock must already be running and stable before configuring the port via GPIO4\_PIN and before enabling the buck converter. Missing pulses or signals applied at different frequencies than specified may cause disruption of the IC.

Alternatively it behaves like a standard GPIO extender pin, see Section 8.4.

### 8.2.8 OC\_PG / nIRQ

The OC\_PG (Over Current alarm and Power Good) is an output port shared with nIRQ and can be either push-pull or open-drain (selected via OC\_PG\_IRQ\_TYPE). The port can be configured as active high or active low via control OC\_PG\_IRQ\_LEVEL.

If OC\_PG\_IRQ\_CONF = 0 the port is used as IRQ line. It indicates that an interrupt causing event has occurred and that the event/status information is available in the related registers 0x50 to 0x53. Such information can be a warning of critical temperature, fault conditions, or status changes on GPI ports. The event registers hold information about the events that have occurred. Events are triggered

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by a status change at the monitored signals. When an event bit is set, the nIRQ signal is asserted (unless this interrupt is masked by a corresponding bit in the IRQ mask registers 0x54 and 0x55). The nIRQ is not released until all event registers with asserted bits have been read and cleared. New events that occur during event register reading are held until the event register has been read and cleared, ensuring that the host processor does not miss them.

If OC\_PG\_IRQ\_CONF = 1, the OC\_PG functionality is selected. The OC\_PG function is valid for Lilon battery voltage ranging from 2.8 V to 4.4 V. This can be used for a dedicated communication to the host processor. OC\_PG remains asserted while as at least one of the STATUS indicators is active (see registers 0x50 and 0x51). This allows the host processor to be immediately informed and to promptly react by, for example, reducing its operating frequency. After the fast reaction, the host processor can then check the STATUS registers to see what is causing the OC\_PG port assertion, if the indicator is still active. If the indicator is not active, the host processor is able to track what has caused the OC\_PG port assertion through the EVENT registers.

The STATUS indicators causing the assertion of the OC\_PG port are reported in the following list:

- Not Buck Power Good
- Over Current Alarm
- Warning Temperature
- VBUCK\_MAX Voltage
- GPI Status

When a STATUS bit is set, the OC\_PG signal is asserted, unless the related bit is masked in MASK\_A and MASK\_B registers. An event is produced in parallel when the STATUS bit changes from passive to active state. The OC\_PG port is automatically released when no STATUS indicators are active. The EVENT bit remains asserted and needs to be cleared by the host processor.

When EN\_CHIP is low, the OC\_PG port is configured in high impedance state. When EN\_CHIP is high and the buck converter is disabled, the OC\_PG port is configured to its passive state. Therefore if it is set as open-drain, active low in CONFIG\_A register, a pull-up resistor will be required to achieve a high level on the OC\_PG port.

When disabling the buck converter, if  $OC_PG_REL = 0$ , the  $OC_PG$  port is held asserted during power down for tOC\_PD after the buck has been disabled. When this time expires, the  $OC_PG$  port will be released to its passive state. If  $OC_PG_REL = 1$ , the  $OC_PG$  port is held asserted during power down until the down ramp has completed. After that it will be immediately released to its passive state.

#### Not Buck Power Good

This channel monitors the output voltage of the buck converter and signals an invalid output voltage condition. During power up, the nPWRGOOD indicator is active and is released when the buck output voltage is greater than VGOOD.

During normal operation, the nPWRGOOD indicator is active when the buck output voltage falls below VGOOD – VGOOD\_HYST. The indicator is released when the buck output voltage returns above VGOOD.

#### **Over Current Alarm**

This channel monitors the peak current through the pass devices of the buck converter. The indicator is active as long as the BUCK\_IALARM threshold is hit. This is a pre-warning level information for the host processor.

The control OC\_PG\_MASK can be used to mask the assertion of OC\_PG due to current alarm during the DVC transition of the buck converter until the power up has completed.

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If OC\_PG\_KEEP control bit is asserted, the Over Current Alarm channel is masked for 100 µs after the Power Good condition has been reached, in other words, after the buck rail is valid. This happens when recovering from an out of range condition and during power up of the buck converter.

When triggered due to an over current condition, the OC\_PG port is asserted within tOC\_DEL thereby ensuring fast feedback of information to the host processor.

After being asserted due to Over Current, the OC\_PG port will be released at least tOC\_ASSERT after the current has reduced below the BUCK\_IALARM threshold. This ensures that very short current spikes triggering the OC\_PG will produce pulses with defined minimum width, thereby allowing the host processor to react and take counter-measures.

#### Warning Temperature

This channel monitors the die temperature of DA9210 and the indicator becomes active as soon as the TEMP\_WARN threshold is hit. This is pre-warning level information for the host processor. The indicator returns to inactive only after the die temperature has fallen below TEMP\_WARN – TEMP\_HYST.

#### VBUCK\_MAX Voltage

This channel monitors the output voltage configured for the buck converter either via dedicated DVC interface (see status register VBUCK\_DVC) or via the VBUCK\_A and VBUCK\_B registers. The indicator is active as long as the configured output voltage is greater than or equal to VBUCK\_MAX.

#### **GPI Toggling and External Charger Plugged**

To extend the OC\_PG functionality, if additional external inputs need to be monitored and their activity needs to be combined on the OC\_PG, all DA9210 GPIOs can be configured to trigger the assertion of OC\_PG port if they are configured as GPI and not overridden by any alternative function or masked.

If GPIO1, GPIO2 and GPIO3 are used for dual parallel mode operation (GPIO1\_PIN = 01, GPIO2\_PIN = 01, GPIO3\_PIN = 01), the configuration of these pins is ignored and no signalling on the OC\_PG port is possible.

GPIO4 is assigned for dedicated communication of an External Charger connection signal. If not masked, it asserts when the charger is disconnected from the system. For this purpose, GPIO4 must be configured as GPI, active low.

If GPIO5 and GPIO6 are used as Interface Ports (GPIO5\_PIN = 01, GPIO6\_PIN = 01), the configuration of these pins is ignored and no signalling on the OC\_PG port is possible.

If configured for HW control of the switching regulator, GPIO0, GPIO3 or GPIO4 will not assert OC\_PG when their status is active.

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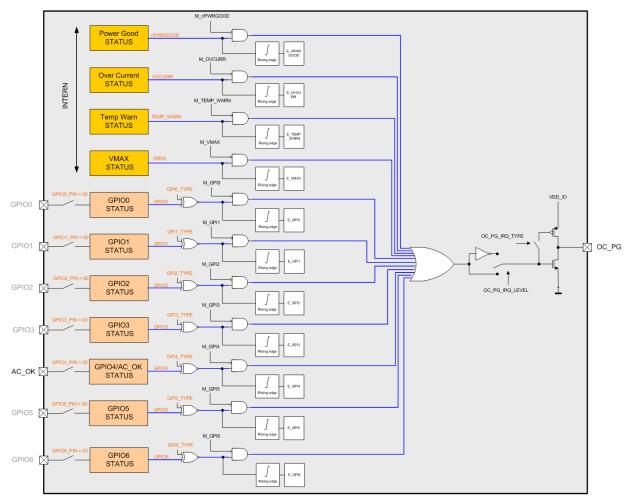


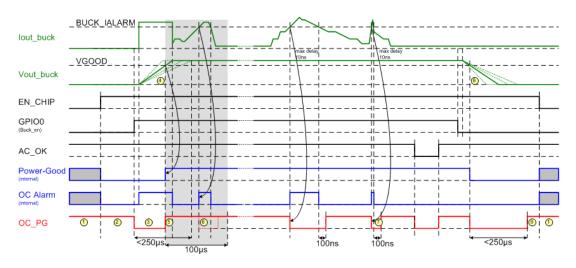
Figure 30: Configuration of OC\_PG Pin Functionality

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# DA9210

### DA9210 Multi-Phase 12 A DC-DC Buck Converter



1 High impedance (ext. pull-up defines the level)

② OC\_PG released to the passive state (ext. pull-up defines the level)

OC\_PG is actively driven low after enabling the buck because Power-Good is low

In STARTUP\_CTRL defines the slope

- OC Alarm ignored during ramping (also successive voltage changes via DVC ramping) if OC\_PG\_MASK=1
- If OC\_PG\_KEEP=1 OC\_PG is not asserted during the first 100µs after Power-Good although an over current occurs.
- The dotted line shows what happens if OC\_PC\_KEEP = 0. In case of short peaks the OC\_PG stays low for at least 100ns

PWR\_DOWN\_CTRL defines the slope

Latest 250µs after the buck is enabled OC\_PG is released to the passive state (ext. pull-up defines the level)

Figure 31: OC\_PG Timing Diagram (CONFIG\_A = 0x16)

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# 8.3 DA9210 Operating Modes

#### 8.3.1 **ON Mode**

DA9210 is in ON mode when the EN\_CHIP port is higher than EN\_ON. Once enabled, the host processor can start communication with DA9210 via control interface after the tEN delay needed for internal circuit start-up.

If BUCK\_EN is asserted when DA9210 is in ON mode, the power up of the buck converter is initiated. If the buck is controlled via GPI (see BUCK\_GPI, VBUCK\_GPI), the level of the controlling ports is checked when entering ON mode, so that an active will immediately affects the buck. If BUCK\_EN is not asserted and all controlling GPI ports are not active, the buck converter remains off with output pull-down resistor enabled or disabled according to BUCK\_PD\_DIS bit.

#### 8.3.2 OFF Mode

DA9210 is in OFF mode when the EN\_CHIP port is lower than EN\_OFF. In OFF mode, the buck is always disabled and the output pull-down resistor is disabled independently of BUCK\_PD\_DIS. All I/O ports of DA9210 are configured to be high impedance.

#### 8.3.3 Dual Parallel Mode

DA9210 is capable of delivering up to 24 A for a high current CPU supply, when operated in dual parallel mode. Two instances of DA9210 are needed with parallel connection of VDD and output node, one acting as master and one as slave (see Figure 32). A DA9210 slave is identified by the connection of VOUT\_SENSE pin to VDDCORE, whilst a master has the VOUT\_SENSE pin in the normal output voltage operational range. A suitable built in filter avoids noise or short spikes on the VOUT\_SENSE line causing the wrong operating mode to be sensed.

To operate in dual parallel mode, both master and slave instances DA9210 must have GPIO1, GPIO2, and GPIO3 configured respectively as VERROR, IPHASE, BUCK\_CLK in the GPIOx\_PIN control field.

The configuration of DA9210 in dual parallel mode is transparent to the host processor and does not require any extra effort in terms of register write and maintenance. Only the master device is visible to the host processor. All commands are sent via the control interface by the host processor into the master. In slave mode, DA9210 will not react to any read command from the control interface. Please contact your local Dialog Semiconductor support for more information and dedicated application note on the Dual Parallel Mode.

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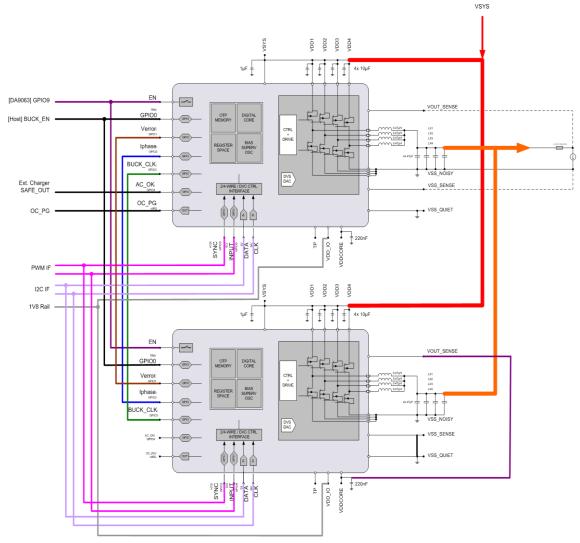


Figure 32: Dual Parallel Mode Configuration



# 8.4 GPIO Extender

DA9210 includes a GPIO extender that offers up to seven 5 V tolerant, GPIO ports; each controlled via registers from the host processor.

The GPIO ports are pin-shared with the control interface, DVC interface, and dual operation. For instance, if GPIO5\_PIN = 01, GPIO6\_PIN = 01 (Interface selected), the GPIO5 and GPIO6 ports will be exclusively dedicated to chip select and output signalling for 4-WIRE purposes or to clock and input for DVC interface, depending on the setting of IF\_TYPE. If the alternate function is selected, all GPIO configurations as per registers 0x58 to 0x5A and 0x145 will be ignored.

GPIs are supplied from the internal rail VDDCORE or VDD\_IO (selected via GPI\_V and can be configured to be active high or active low in GPIOx\_TYPE. The input signals can be debounced (debouncing time configurable via control DEBOUNCING, 10 ms default) can directly change the state of the assigned status register GPIx to high or low, dependent on the setting of GPIOx\_MODE.

If OC\_PG\_IRQ\_CONF = 1, as long as the status is at its configured active state (level sensitive), the OC\_PG port is asserted (unless this is masked, see also Figure 30). Whenever the status changes to the active state (edge sensitive), the assigned event register is set (needs to be cleared by the host processor).

If OC\_PG\_IRQ\_CONF = 0, whenever the status changes to its configured active state (edge sensitive), the assigned event register is set and the nIRQ signal is asserted (unless the nIRQ is masked, see Figure 33).

Whenever DA9210 is enabled and enters ON mode (also when enabled, changing the setting of GPIOx\_PIN), the GPI status bits are initiated towards their configured passive state. This ensures that already active signals are detected and create an event immediately after the GPI comparators are enabled.

If enabled via buck control BUCK\_GPI, port GPI0, GPI3 and GPI4 enable/disable the switching regulator from the rising and falling edges of these signals (changing BUCK\_EN). If GPI ports must be enables for HW control of the switching regulator, do not generate an event and do not assert OC\_PG independently of the OC\_PG\_IRQ\_CONF setting, the relative mask bit should also be set.

GPI0, GPI3 and GPI4 can alternatively be selected to toggle the BUCK\_SEL bit with rising and falling edges at these inputs. In addition to changing the regulator output voltage this also provides a HW control of regulator mode (normal/low power mode) from the settings BUCK\_SL\_A, BUCK\_SL\_B (enabled via BUCK\_MODE = '00').

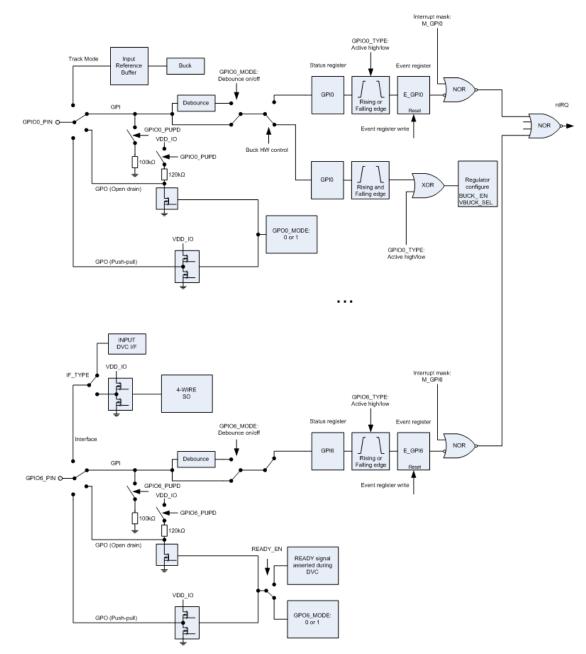
All GPI ports have the additional option of activating a 100 k $\Omega$  pull-down resistor via GPIOx\_PUPD, which ensures a well-defined level in case the input is not actively driven.

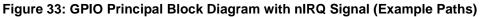
If defined as an output, GPIOs can be configured to be open-drain or push-pull. The supply rail in case of push-pull is VDD\_IO. By disabling the internal 120 k $\Omega$  pull-up resistor in open-drain mode, the GPO can also be supplied from an external rail. The output state will be determined by the GPIO register bit GPIOx\_MODE.

Whenever the GPIO unit is off (POR or OFF mode), all ports are configured as open-drain active high (pass device switched off, high impedance state). When leaving POR, the pull-up or pull-down resistors are configured from register CONFIG\_C

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# 8.5 Control Interfaces

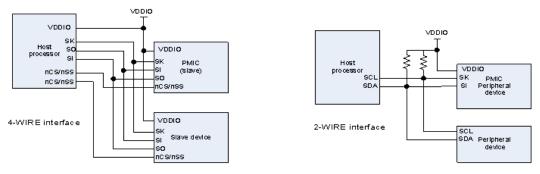
The DA9210 can be SW-controlled by the host. The DA9210 offers access to its registers via a serial control interface. The communication is selectable to be either a 2-WIRE (I<sup>2</sup>C compliant) or a 4-WIRE (SPI compliant) connection via control IF\_TYPE which is selected during the initial OTP read. In both configurations, the DA9210 will act as a slave device, data is shifted into or out from DA9210 under the control of the host processor that also provides the serial clock. The interface is usually only configured once from OTP values, which are loaded during the initial start-up of DA9210.

#### NOTE

DA9210 reacts only on read/write commands where the transmitted register address (using the actual page bits as an MSB address range extension) is within 0x50 to 0x6F, 0xD0 to DF, 0x140 to 0x14F, or (read only) 0x200 to 0x280. Host access to registers outside these ranges will be ignored (no acknowledge after receiving the register address in 2-WIRE mode, SO stays HI-Z in 4-WIRE mode).

DA9210 reacts only to write commands where the transmitted register address is 0x00, 0x80, 0x100, 0x101, 0x105, 0x106, or 0x200. If STAND\_ALONE is asserted (OTP bit), DA9210 also reacts to read commands. If DA9210 is in slave configuration (VOUT\_SENSE tied to VDDCORE) it does not react to any read command, see Section 8.3.3.

DA9210 provides an additional interface supporting direct DVC requests from the host processor to the buck. The DVC interface can be enabled via control DVC\_CTRL\_EN, provided that the GPIO5\_PIN, GPIO6\_PIN, and IF\_TYPE controls are appropriately configured.





### 8.5.1 4-WIRE Communication

In 4-WIRE mode, the interface uses a chip-select line (nCS/nSS), clock line (SK), data input (SI), and data output line (SO).

The DA9210 register map is split into four pages with each page containing up to 128 registers, in order to be transparent to the main PMIC (DA9063) and compliant to its register map. The register at address 0x00 on each page is used as a page control register. The default active page after turn on includes registers 0x50 to 0x6F. Writing to the page control register changes the active page for all subsequent read/write operations unless an automatic return to page 0 was selected by asserting control REVERT. Unless REVERT was asserted, after modifying the active page it is recommended to read back the page control register to ensure that future data exchange is accessing the intended registers.

All registers out of the DA9210 range are write only that is the DA9210 will not answer to a read command and the data bus is tri-state (they are implicitly directed to DA9063). In particular, the information contained in registers 0x105 and 0x106 is used by DA9210 to configure the control interface. They must be the same as the main PMIC (DA9063), so that a write to those registers configures both the main PMIC and DA9210. The default OTP settings need also to be identical for correct operation of the system.

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The 4-WIRE interface features half-duplex operation (data can be transmitted and received within a single 16 bit frame) at an enhanced clock speed (up to 14 MHz). The interface operates at the clock frequency provided by the host.

A transmission begins when initiated by the host. Reading and writing is accomplished using an 8-bit command, sent by the host prior to the exchanged 8-bit data. The byte from the host begins shifting in on the SI pin under the control of the serial clock SK provided from the host. The first 7 bits specify the register address (0x01 to 0x07) which will be written or read by the host. The register address is automatically decoded after receiving the seventh address bit. The command word ends with an R/W bit, this, together with the control bit R/W\_POL, specifies the direction of the following data exchange. During a register write, the host continues sending out data during the following eight SK clocks. For a read, the host stops transmitting and the 8-bit register is clocked out of DA9210 during the following eight SK clocks of the frame. Address and data are transmitted MSB first. The polarity (active state) of nCS is defined by control bit nCS\_POL. nCS resets the interface when inactive and must be released between successive cycles.

The SO output from DA9210 is normally in a high-impedance state and is active only during the second half of a read cycle. A pull-up or pull-down resistor may be needed on the SO line if a floating logic signal could result in unintended current consumption inside other circuits.

The DA9210 4-WIRE interface offers two further configuration bits. Clock polarity (CPOL) and clock phase (CPHA) define when the interface will latch the serial data bits. CPOL determines whether SK idles high (CPOL = 1) or low CPOL= 0). CPHA determines on which SK edge data is shifted in and out. With CPOL = 0 and CPHA = 0, DA9210 latches data on the SK rising edge. If the CPHA is set to 1, the data is latched on the SK falling edge. CPOL and CPHA states allow four different combinations of clock polarity and phase; each setting is incompatible with the other three. The host and DA9210 must be set to the same CPOL and CPHA states to communicate with each other, see Table 15.

CPHA clock polarity	CPOL clock phase	Output data is updated at SK edge	Input data is registered at SK edge	CPHA clock polarity
0 (idle low)	0	falling	rising	0 (idle low)
0 (idle low)	1	rising	falling	0 (idle low)
1 (idle high)	0	rising	falling	1 (idle high)
1 (idle high)	1	falling	rising	1 (idle high)

Table 15: 4-WIRE Clock Configurations





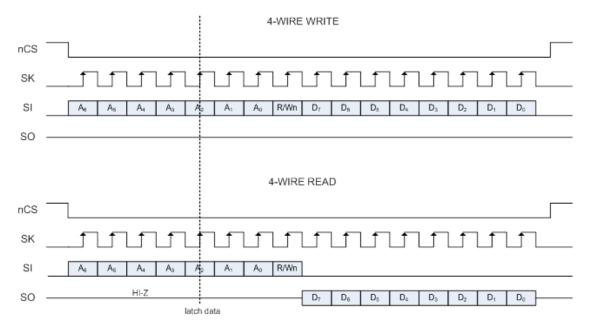


Figure 35: 4-WIRE Host Write and Read Timing (nCS\_POL = '0', CPOL = '0', CPHA = '0')

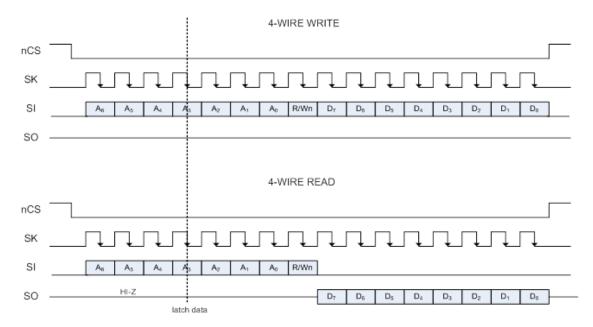


Figure 36: 4-WIRE Host Write and Read Timing (nCS\_POL = '0', CPOL = '0', CPHA = '1')

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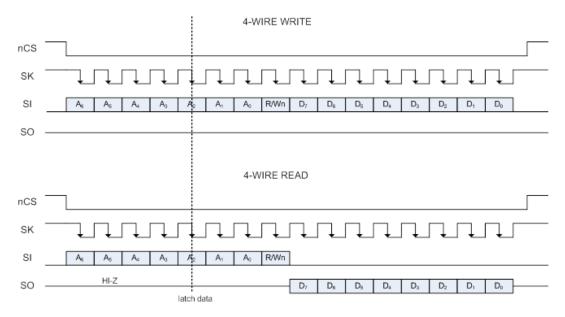


Figure 37: 4-WIRE Host Write and Read Timing (nCS\_POL = '0', CPOL = '1', CPHA = '0')

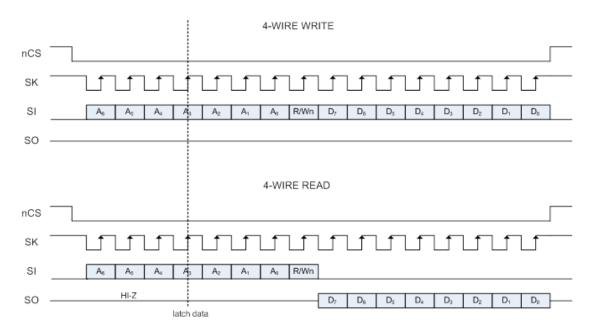


Figure 38: 4-WIRE Host Write and Read Timing (nCS\_POL = '0', CPOL = '1', CPHA = '1')

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Table	16:	4-WIRE	Interface	Summary
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Parameter								
	nCS	Chip select						
Signal Lines	SI Serial input data	Master out Slave in						
Signal Lines	SO Serial output data	Master in Slave out						
	SK	Transmission clock						
Interface	Push-pull with tristate							
Supply voltage	Selected from V <sub>DD_IO</sub>	1.6 V to 3.3 V						
Data rage	Effective read/write data	Up to 7 Mbps						
	Half-duplex	MSB first						
Transmission	16 bit cycles	7bit address, 1 bit read/write, 8 bit data						
	CPOL	clock polarity						
Configuration	СРНА	clock phase						
	nCS_POL	nCS is active low/high						

**Note 1** Reading a register at high clock rates directly after writing to it does not guarantee a correct value. A delay of one frame is recommended before re-accessing a register that has just been written (for example by writing/reading another register address in between)

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#### 8.5.2 2-WIRE Communication

With control IF\_TYPE, the DA9210 power manager interface can be configured towards a 2-WIRE serial data exchange. It has a configurable device write address (default: 0xD0) and a configurable device read address (default: 0xD1). For details of configurable addresses, see control IF\_BASE\_ADDR.

The SK pin provides the 2-WIRE clock (SCL), and SI carries all the power manager bidirectional 2-WIRE data, (SDA). The 2-WIRE interface is open-drain supporting multiple devices on a single line. The bus lines require external pull-up resistors ( $2 k\Omega$  to  $20 k\Omega$  range). The attached devices only drive the bus lines LOW. As a result, two devices cannot conflict if they drive the bus simultaneously. In standard/fast mode, the highest frequency of the bus is 400 kHz. The exact frequency can be determined by the application and does not have any relation to the DA9210 internal clock signals. DA9210 follows the host clock speed within the described limitations and does not initiate any clock arbitration or slow down. An automatic reset of the interface can be triggered via control 2WIRE\_TO if the clock signal stops toggling for more than 35 ms (compatible with SMBus).

The interface supports operation compatible with Standard, Fast, Fast-Plus and High Speed modes of the I<sup>2</sup>C bus specification Rev 4 (UM10204\_4). Operation in high speed mode at 3.4 MHz requires mode changing in order to change spike suppression and slope control characteristics to be compatible with the I<sup>2</sup>C bus specification. The high speed mode can be enabled on a transfer by transfer basis by sending the master code (0000 1XXX) at the beginning of the transfer. DA9210 does not make use of clock stretching and delivers read data up to 3.4 MHz, without additional delay.

Communication on the 2-WIRE bus always takes place between two devices, one acting as the master and the other as the slave. The DA9210 only operates as a slave.

Unlike 4-WIRE mode, the 2-WIRE interface has direct access to two pages of the register map (up to 256 addresses). The register at address 0x00 on each page is used as a page control register (with the 2-WIRE bus ignoring the LSB of control REG\_PAGE). Writing to the page control register changes the active page for all subsequent read/write operations, unless an automatic return to page 0 is selected by asserting control REVERT. After modifying the active page, unless REVERT is asserted, it is recommended to read back the page control register to ensure that future data exchange accesses the intended registers.

In 2-WIRE operation DA9210 offers an alternative way to access register page 2 and 3. It removes the need for the preceding page selection writes by increasing the device write/read address by one (default 0xD2/0xD3) for any direct access of page 2 and 3 (page 0 and 1 access requires the basic write/read device address with the MSB of REG\_PAGE '0').

#### **Details of the 2-WIRE Control Bus Protocol**

The following description uses the standard terms SDA for the serial data, pin SI, and SCL for the serial clock, pin SK.

All data is transmitted across the 2-WIRE bus in 8-bit groups. To send a bit, the SDA line is driven at the intended state while the SCL is low. Once the SDA has settled, the SCL line is brought high and then low. This pulse on SCL stores the SDA bit in the receiver's shift register.

A 2-byte serial protocol is used: one address byte and one data byte. The data and address are transmitted MSB first for both read and write operations. All transmissions begin with the START condition from the master, during which the bus is in IDLE state (the bus is free). It is initiated by a high-to-low transition on the SDA line while the SCL is in high state. A STOP condition is indicated by a low-to-high transition on the SDA line while the SCL is in high state. The START and STOP conditions are illustrated in Figure 39.

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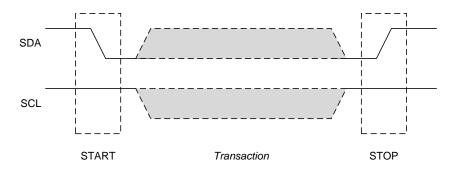


Figure 39: Timing of the START and STOP Conditions

DA9210 monitors the 2-wire bus for a valid slave address whenever the interface is enabled. It responds immediately when it receives its own slave address. This is acknowledged by pulling the SDA line low during the following clock cycle (white blocks marked with 'A' in the following figures).

The protocol for a register write from master to slave consists of a START condition, a slave address, a read/write-bit, 8-bit address, 8-bit data, and a STOP condition. DA9210 responds to all bytes with an ACK. A register write operation is illustrated in Figure 40.

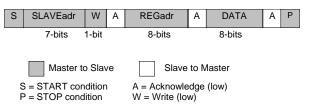


Figure 40: Byte Write Operation

When the host reads register data, the DA9210 first has to access the target register address with write access and then with read access and a repeated START, or alternatively a second START, condition. After receiving the data, the host sends NACK and terminates the transmission with a STOP condition, see Figure 41.

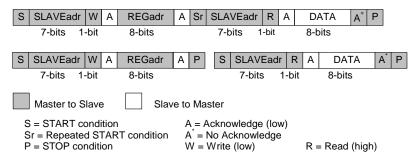


Figure 41: Examples of Byte Read Operations

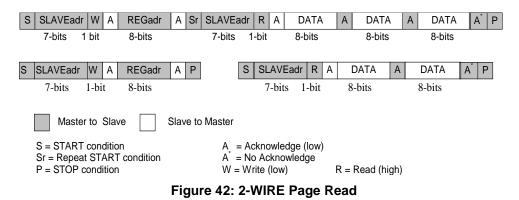
Consecutive (page) read-out mode is initiated from the master by sending an ACK instead of NACK after receiving a byte, see Figure 42. The 2-wire control block then increments the address pointer to the next register address and sends the data to the master. The data bytes are read continuously until the master sends a NACK followed by a subsequent STOP condition directly after receiving the data. If a non-existent 2-wire address is read out, then the DA9210 returns code zero.

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	21		<b>C</b>	n	0		r -
	a	.0	-		-		



# DA9210

# DA9210 Multi-Phase 12 A DC-DC Buck Converter



The slave address after the repeated START condition must be the same as the previous slave address.

Consecutive (page) write mode is supported if the master sends several data bytes after sending the register address. The 2-wire control block then increments the address pointer to the next 2-wire address, stores the received data, and sends an ACK until the master sends a STOP condition. The page write mode is illustrated in Figure 43.

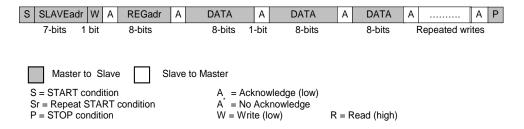


Figure 43: 2-WIRE Page Write

A repeated write mode can be enabled with WRITE\_MODE control. In this mode, the master can execute back-to-back write operations to non-consecutive addresses by transmitting register addresses and data pairs. The data is stored in the address specified by the preceding byte. The repeated write mode is illustrated in Figure 44.

S SLAVEadr W	REGadr	A	DATA	A	REGadr	A	DATA	A		A
7-bits 1 bit	8-bits		8-bits	1-bit	8-bits		8-bits		Repeated wi	rites
Master to Sla	ve 🗌 S	Slave to	Master							
			A _ A	aknow	lades (law)					
S = START condition			A, = P		rledge (low) nowledge					



If a new START or STOP condition occurs within a message, the bus returns to idle mode.

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	a	ια	5		C		



#### 8.5.3 DVC Interface

With control DVC\_CTRL\_EN, GPIO5 and GPIO6 can be dedicated to an alternative mode of defining the buck output voltage (supporting direct DVC) from the host processor to the buck. This is only enabled if GPIO5 and GPIO6 are configured for interface (setting 01) and IF\_TYPE is asserted. Since the logical levels of the I/O are referenced to VDD\_IO, a valid voltage at this port is needed.

After the DVC interface has been activated and synchronized, the buck converter target voltage is derived by adding a delta value to the voltage base register configured in VBUCK\_BASE. The delta value is decoded by DA9210 from the number of clock rising edges at the SYNC pin while INPUT is kept high. During control via the DVC interface, the setting of registers VBUCK\_A, VBUCK\_B are ignored and the target output voltage is calculated as:

 $V_{OUT} = V_{BUCK\_BASE} + (N \times V_{STEPS})$ 

Where  $V_{VSTEPS}$  is 10 mV and N can range from 0 to 32. The principle derivation of the target voltage is shown in Figure 45.

#### Example

The INPUT port is sampled high for 16 of the 32 samples, thus N = 16. If  $V_{BUCK\_BASE} = 0.8$  V, the target voltage is  $V_{OUT} = 0.8$  V + (16 x 10 mV) = 0.96 V. Once the DVC interface is enabled, DA9210 will maintain the voltage set in VBUCK\_A/VBUCK\_B until the valid completion of the first 32 clock sample. The new value of the output voltage is automatically updated after the completion of 32 clock sample. The clocking SYNC signal can be stopped after a multiple of 32 cycles or it can run continuously. The value used for setting the target voltage is always based on the last 32 clock sample frame. There is no explicit frame sync signal. Once the DVC interface is enabled, the frame sync is implicit on the rising edge of the INPUT port, which indicates the start of a 32 clock frame. If a new rising edge occurs before the previous 32 clocks have been sampled, the current accumulated sample is discarded and a new 32 clock sample is started. The previous voltage setting for the buck converter is used as target in the meantime, until a valid sequence is decoded. If a zero value is sent to the INPUT port,  $V_{OUT}$  will be equal to  $V_{BUCK\_BASE}$  so there is no rising edge and thus no frame sync. DA9210 will keep counting the clock edges at the SYNC port. In a similar way, if a high value is continuously sent to the INPUT port,  $V_{OUT}$  will be equal to  $V_{BUCK\_BASE} + (32 \times V_{STEPS})$  although no frame sync takes place.

The host processor can monitor the configured target voltage on the (read only) register VBUCK\_DVC.

The DVC interface is interrupted when DVC\_CTRL\_EN is released or the IF\_TYPE/GPI4/GPI05 configuration is changed. After disabling the DVC interface, the normal control of the output voltage via select register VBUCK\_A, VBUCK\_B will revert automatically. If FORCE\_DVC\_IF is not asserted, the DVC interface is disabled when the buck is disabled (this automatically clears DVC\_CTRL\_EN bit). Otherwise the DVC interface remains armed and, at the next enable of the buck converter, the target output value will be derived from the SYNC and INPUT ports, as soon as a valid completion of the first 32 clock sample takes place. Until a valid 32 clock sample is received, the output voltage is set via select register BUCK\_SEL and VBUCK\_A or VBUCK\_B values.

If DVC\_STEP\_SIZE is asserted the output voltage will increase by double the amount of consecutive asserted bits (V<sub>VSTEPS</sub> will correspond to 20 mV instead of 10 mV).

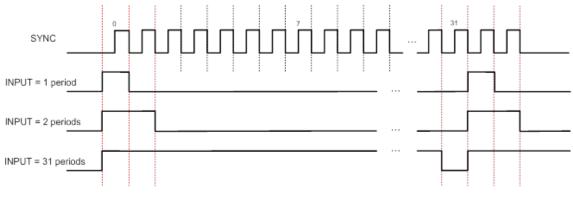
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#### Example

The INPUT port is sampled high for 16 of the 32 samples, thus N = 16. If DVC\_STEP\_SIZE is asserted and  $V_{BUCK_BASE} = 0.8 \text{ V}$ , the target voltage is  $V_{OUT} = 0.8 \text{ V} + (16 \text{ x } 20 \text{ mV}) = 1.12 \text{ V}$ .

To limit the buck output voltage and prevent any failures in case of a communication error, a maximum value for the buck output target voltage can be stored in  $V_{BUCK\_MAX}$ . This value represents the absolute maximum voltage and it has priority over any value resulting from the addition of  $V_{BUCK\_BASE}$  and (N x  $V_{STEPS}$ ) as well as over any value set on  $VBUCK\_A$  and  $VBUCK\_B$ , if the DVC interface is disabled. When the output voltage exceeds  $V_{BUCK\_MAX}$ , an event is generated and the OC\_PG port is asserted, if not masked by M\_VMAX. If the OC\_PG port is configured for IRQ operation, an interrupt is generated instead.







### 8.6 Internal Temperature Supervision

To protect the DA9210 from damage due to excessive power dissipation, the internal temperature is continuously monitored. There are three temperature thresholds, TEMP\_WARN, TEMP\_CRIT, and TEMP\_POR, respectively at typically 125 °C, 140 °C, and 150 °C.

When the junction temperature reaches the TEMP\_WARN threshold, DA9210 asserts the bit TEMP\_WARN and generates the event E\_TEMP\_WARN. If not masked via bit M\_TEMP\_WARN, the output port OC\_PG/nIRQ is asserted. The status bit TEMP\_WARN remains asserted while the junction temperature is higher than TEMP\_WARN.

When the junction temperature increases further over TEMP\_CRIT, the DA9210 immediately disables the buck converter, asserts the bit TEMP\_CRIT and generates the event E\_TEMP\_WARN. If not masked via bit M\_TEMP\_WARN, the output port OC\_PG/nIRQ is asserted. The status bit TEMP\_CRIT remains asserted while the junction temperature is higher than TEMP\_CRIT. The buck converter is disabled as long as the junction temperature is greater than TEMP\_CRIT and is automatically re-enabled after the temperature recovers below the valid threshold (even if the controlling GPI is asserted). A direct write into BUCK\_EN or a toggling of the controlling GPI is needed to enable the buck converter.

Whenever the junction temperature exceeds TEMP\_POR, a power-on reset to the digital core is immediately asserted, which stops all functionality in DA9210. This is needed to prevent possible permanent damage in case of a fast temperature increase.

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# 9 **Register Definitions**

## 9.1 Register Map

All bits loaded from OTP are marked in bold.

Addr	Function	7	6	5	4	3	2	1	0		
	· · · ·				Register Page 0						
0x00	PAGE_CON	REVERT	WRITE_MOD E	Reserved	Reserved	eserved REG_PAGE					
0x50	STATUS_A	Reserved	GPI6	GPI5	GPI4	GPI3	GPI2	GPI1	GPI0		
0x51	STATUS_B	Reserved	Reserved	Reserved	VMAX	TEMP_CRIT	TEMP_WARN	nPWRGOOD	OVCURR		
0x52	EVENT_A	Reserved	E_GPI6	E_GPI5	E_GPI4	E_GPI3	E_GPI2	E_GPI1	E_GPI0		
0x53	EVENT_B	Reserved	Reserved	Reserved	E_VMAX	E_TEMP_CRIT	E_TEMP_WARN	E_nPWRGOOD	E_OVCURR		
0x54	MASK_A	Reserved	M_GPI6	M_GPI5	M_GPI4	M_GPI3	M_GPI2	M_GPI1	M_GPI0		
0x55	MASK_B	Reserved	Reserved	Reserved	M_VMAX	M_TEMP_CRIT	M_TEMP_WARN	M_nPWRGOOD	M_OVCURR		
0x56	CONTROL_A	Reserved	Reserved	V_LOCK	SLEW	/_RATE		DEBOUNCING			
0x57	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved		
0x58	GPIO0-1	GPIO0_MODE	GPIO1_TYPE	GPI	O1_PIN	GPIO0_MODE	GPIO0_TYPE	GPIO	)_PIN		
0x59	GPIO2-3	GPIO3_MODE	GPIO3_TYPE	GPI	O3_PIN	GPIO2_MODE	GPIO2_TYPE	GPIO2	2_PIN		
0x5A	GPIO4-5	GPIO5_MODE	GPIO5_TYPE	GPI	O5_PIN	GPIO4_MODE	GPIO4_TYPE	GPIO4	1_PIN		
0x5B	GPIO6	Reserved	Reserved	Re	served	GPIO6_MODE	GPIO6_TYPE	GPIO6_PIN			
0x5C	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved		
0x5D	BUCK_CONT	DVC_CTRL_EN	VBUCK	_GPI	VBUCK_SEL	BUCK_PD_DIS	BUCK	_GPI	BUCK_EN		

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a	LO.	5		C	C	

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# DA9210 Multi-Phase 12 A DC-DC Buck Converter

Addr	Function	7	6	5	4	3	2	1	0		
				F	Register Page 1						
0x80	PAGE_CON	REVERT	WRITE_MODE	Reserved	Reserved	Reserved		REG_PAGE			
0xD0	BUCK_ILIM	Reserved	Reserved	Reserved	BUCK_IALARM		BUCK_ILIM				
0xD1	BUCK_CONF1	PV	VR_DOWN_CTRL		S	STARTUP_CTRL	BUCK_MODE				
0xD2	BUCK_CONF2	Reserved	Reserved	Reserved	Reserved	PH_SH_EN		PHASE_SEL			
0xD3	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved		
0xD4	VBUCK_AUTO	Reserved				VBUCK_AUTO					
0xD5	VBUCK_BASE	Reserved				VBUCK_BASE					
0xD6	VBUCK_MAX	Reserved				VBUCK_MAX					
0xD7	VBUCK_DVC	Reserved				VBUCK_DVC					
0xD8	VBUCK_A	BUCK_SL_A				VBUCK_A					
0xD9	VBUCK_B	BUCK_SL_B				VBUCK_B					

Addr	Function	7	6	5	4	3	2	1	0		
					Register Page	2					
0x100	PAGE_CON	REVERT	WRITE_MOD E	Reserved	Reserved	Reserved		REG_PAGE			
	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved		
0x105	INTERFACE		IF_BASE_	ADDR		R/W_POL	СРНА	CPOL	nCS_POL		
0x106	INTERFACE 2	IF_TYPE	PM_IF_HSM	PM_IF_FMP	PM_IF_V	Reserved	Reserved	Reserved	Reserved		
0x140	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved		
0x141	Reserved					Reserved					
0x142	Reserved					Reserved					
0x143	CONFIG_A	Reserved	Reserved	Reserved	2WIRE_TO	GPI_V	OC_PG_IRQ_C ONF	OC_PG_IRQ_T YPE	OC_PG_IRQ_LE VEL		
0x144	CONFIG_B	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved		
0x145	CONFIG_C	Reserved	GPIO6_PUPD	GPIO5_PUP D	GPIO4_PUP D	GPIO3_PUPD	GPIO2_PUPD	GPIO1_PUPD	GPIO0_PUPD		
0x146	CONFIG_D	OC_PG_KEEP	OC_PG_REL	OC_PG	6_MASK	READY_EN	FORCE_DVC_IF	DVC_STEP_SIZ E	Reserved		
0x147	CONFIG_E	Reserved	Reserved	OSC_	TUNE	Reserved	Reserved	Reserved	STAND_ALONE		
0x14F	MISC_SUPP	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved	OTP_CLK_ON		

Figure 46: Register Map

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# 9.2 Register Page Control

#### Table 17: PAGE\_CON (0x00)

Bit	Туре	Label	Description
7	R/W	REVERT	Resets REG_PAGE to 000 after read/write access has finished
6	R/W	WRITE_MODE	2-WIRE multiple write mode
		Note 1	0: Page Write mode
			1: Repeated Write mode
5:3	R/W	(Reserved)	
2:0	R/W	REG_PAGE	I <sup>2</sup> C:
			00x: Selects register 0x00 to 0xFF
			01x: Selects register 0x100 to 0x17F
			SPI:
			000: Selects register 0x00 to 0x7F
			001: Selects register 0x80 to 0xFF
			010: Selects register 0x100 to 0x17F
			>010: Reserved for production and test

Note 1 Not used for 4-WIRE-IF

## 9.3 Register Page 0

#### 9.3.1 System Control and Event

The STATUS registers report the current value of the various signals at the time that it is read out.

#### NOTE

All the status bits have the same polarity as their corresponding signals.

#### Table 18: STATUS\_A (0x50)

Bit	Туре	Label	Description
7	R	Reserved	
6	R	GPI6	GPI6 level
5	R	GPI5	GPI5 level
4	R	GPI4	GPI4 level
3	R	GPI3	GPI3 level
2	R	GPI2	GPI2 level
1	R	GPI1	GPI1 level
0	R	GPI0	GPI0 level



#### Table 19: STATUS\_B (0x51)

Bit	Туре	Label	Description
7:5	R	Reserved	
4	R	VMAX	Asserted as long as the voltage configured for the buck converter is equal or greater than VBUCK_MAX
3	R	TEMP_CRIT	Asserted as long as the thermal shutdown threshold is reached
2	R	TEMP_WARN	Asserted as long as the thermal warning threshold is reached
1	R	nPWRGOOD	Asserted as long as the buck output voltage is out of range
0	R	OVCURR	Asserted as long as the buck is in overcurrent

The EVENT registers hold information about events detected by the DA9210.

If the OC\_PG port is configured as IRQ line:

- The events are triggered by a change in the status register which contains the status of monitored signals.
- When an EVENT bit is set in the event register the IRQ signal shall be asserted, unless the event is masked by setting the associated bit in the mask register. The IRQ triggering event register is cleared from the host by writing back its read value. New events occurring during clearing is delayed before they are passed to the event register, ensuring that the host controller does not miss them.

If the OC\_PG port is configured for dedicated communication to the host processor as described in Section 8.2.8, the port will be asserted according to the STATUS bits and the polarity of the STATUS indication (active high/low) can be selected separately (see GPIOxx\_TYPE control). An EVENT is generated only during transition of the STATUS from inactive to active.

Bit	Туре	Label	Description
7	R	Reserved	
6	R	E_GPI6	GPI6 event according to active state setting
5	R	E_GPI5	GPI5 event according to active state setting
4	R	E_GPI4	GPI4 event according to active state setting
3	R	E_GPI3	GPI3 event according to active state setting
2	R	E_GPI2	GPI2 event according to active state setting
1	R	E_GPI1	GPI1 event according to active state setting
0	R	E_GPI0	GPI0 event according to active state setting

#### Table 20: EVENT A (0x52)

#### Table 21: EVENT\_B (0x53)

Bit	Туре	Label	Description
7:5	R	Reserved	
4	R	E_VMAX	VMAX caused event
3	R	E_TEMP_CRIT	TEMP_CRIT caused event
2	R	E_TEMP_WARN	TEMP_WARN caused event

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Bit	Туре	Label	Description
1	R	E_nPWRGOOD	nPWRGOOD caused event
0	R	E_OVCURR	OVCURR caused event

#### Table 22: MASK\_A (0x54)

Bit	Туре	Label	Description
7	R/W	Reserved	
6	R/W	M_GPI6	GPI6 nIRQ Mask
5	R/W	M_GPI5	GPI5 nIRQ Mask
4	R/W	M_GPI4	GPI4 nIRQ Mask
3	R/W	M_GPI3	GPI3 nIRQ Mask
2	R/W	M_GPI2	GPI2 nIRQ Mask
1	R/W	M_GPI1	GPI1 nIRQ Mask
0	R/W	M_GPI0	GPI0 nIRQ Mask

### Table 23: MASK\_B (0x55)

Bit	Туре	Label	Description
7:5	R/W	Reserved	
4	R/W	M_VMAX	VMAX nIRQ / OC_PG Event Mask
3	R/W	M_TEMP_CRIT	Mask TEMP_CRIT nIRQ / OC_PG Event Mask
2	R/W	M_TEMP_WARN	TEMP_WARN nIRQ / OC_PG Event Mask
1	R/W	M_nPWRGOOD	PWRGOOD nIRQ / OC_PG Event Mask
0	R/W	M_OVCURR	OVCURR IRQ / OC_PG Event Mask

## Table 24: CONTROL\_A (0x56)

Bit	Туре	Label	Description
7	R/W	Reserved	
_	5 R/W	V_LOCK	0: Allows host writes into registers 0xD0 to 0x14F
5			1: Disables register 0xD0 to 0x14F re- programming from control interfaces
		SLEW_RATE	DVC slewing is executed at
4:3	4:3 R/W		00: 10 mV every 4.0 μs
			01: 10 mV every 2.0 μs
			10: 10 mV every 1.0 μs
			11: 10 mV every 0.5 μs







Bit	Туре	Label	Description
			GPI debounce time:
2:0 R/W		DEBOUNCING	000: no debounce time
			001: 0.1 ms
			010: 1.0 ms
	R/W		011: 10 ms
			100: 50 ms
			101: 250 ms
			110: 500 ms
			111: 1000 ms

### 9.3.2 GPIO Control

#### Table 25: GPIO0-1 (0x58)

Bit	Туре	Label	Description
7	R/W	GPIO1_MODE	0: GPI: debounce off GPO: Sets output to low level 1: GPI: debounce on GPO: Sets output to high level
6	R/W	GPIO1_TYPE	0: GPI: active low 1: GPI: active high
5:4	R/W	GPIO1_PIN	PIN assigned to: 00: GPI 01: Verror signal 10: GPO (open-drain) 11: GPO (push-pull)
3	R/W	GPIO0_MODE	0: GPI: debounce off GPO: Sets output to low level 1: GPI: debounce on GPO: Sets output to high level
2	R/W	GPIO0_TYPE	0: GPI: active low 1: GPI: active high
1:0	R/W	GPIO0_PIN	PIN assigned to: 00: GPI 01: Reserved 10: GPO (open-drain) 11: GPO (push-pull)





# Table 26: GPIO2-3 (0x59)

Bit	Туре	Label	Description
7	R/W	GPIO3_MODE	0: GPI: debounce off GPO: Sets output to low level 1: GPI: debounce on GPO: Sets output to high level
6	R/W	GPIO3_TYPE	0: GPI: active low 1: GPI: active high
5:4	R/W	GPIO3_PIN	PIN assigned to: 00: GPI 01: BUCK_CLK signal 10: GPO (open-drain) 11: GPO (push-pull)
3	R/W	GPIO2_MODE	00: GPI: debounce off GPO: Sets output to low level 1: GPI: debounce on GPO: Sets output to high level
2	R/W	GPIO2_TYPE	0: GPI: active low 1: GPI: active high
1:0	R/W	GPIO2_PIN	PIN assigned to: 00: GPI 01: Iphase signal 10: GPO (open-drain) 11: GPO (push-pull)

#### Table 27: GPIO4-5 (0x5A)

Bit	Туре	Label	Description
7	R/W	GPIO5_MODE	0: GPI: debounce off GPO: Sets output to low level 1: GPI: debounce on GPO: Sets output to high level
6	R/W	GPIO5_TYPE	0: GPI: active low 1: GPI: active high
5:4	R/W	GPIO5_PIN	PIN assigned to: 00: GPI 01: Interface 10: GPO (open-drain) 11: GPO (push-pull)
3	R/W	GPIO4_MODE	0: GPI: debounce off GPO: Sets output to low level 1: GPI: debounce on GPO: Sets output to high level
2	R/W	GPIO4_TYPE	0: GPI: active low 1: GPI: active high

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Bit	Туре	Label	Description
1:0	R/W	GPIO4_PIN	PIN assigned to: 00: GPI (AC_OK) 01: Digital external clock input 10: GPO (open-drain) 11: GPO (push-pull)

Table 28: GPIO6 (0x5B)

Bit	Туре	Label	Description
7:4	R/W	Reserved	
3	R/W	GPIO6_MODE	0: GPI: debounce off GPO: Sets output to low level 1: GPI: debounce on GPO: Sets output to high level
2	R/W	GPIO6_TYPE	0: GPI: active low 1: GPI: active high
1:0	R/W	GPIO6_PIN	PIN assigned to: 00: GPI 01: Interface 10: GPO (open-drain) 11: GPO (push-pull)

# 9.3.3 Regulator Control

## Table 29: BUCK\_CONT (0x5D)

Bit	Туре	Label	Description
7	R/W	DVC_CTRL_EN	Main control of the dedicated DVC Interface: 0: Disabled 1: Enabled
6:5	R/W	VBUCK_GPI	GPIO select target voltage VBUCK_B on passive to active transition, selects target voltage VBUCK_A on active to passive transition (ramping) 00: Not controlled by GPIO 01: GPIO0 controlled 10: GPIO3 controlled 11: GPIO4 controlled
4	R/W	VBUCK_SEL Note 1	BUCK voltage is selected from (ramping): 0: VBUCK_A 1: VBUCK_B
3	R/W	BUCK_PD_DIS	0: Enable pull-down resistor in disabled mode 1: No pull-down resistor in disabled mode

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Bit	Туре	Label	Description
2:1	R/W	BUCK_GPI	GPIO enables the BUCK on passive to active state transition, disables the BUCK on active to passive state transition 00: Not controlled by GPIO 01: GPIO0 controlled 10: GPIO3 controlled 11: GPIO4 controlled
0	R/W	BUCK_EN	0: BUCK disabled 1: BUCK enabled

**Note 1** Automatically set to 0 by default when the buck converter is enabled, except when the output voltage is controlled via GPI port.

## 9.4 Register Page 1

#### Table 30: PAGE\_CON (0x80)

Bit	Туре	Label	Description
7	R/W	REVERT	Resets REG_PAGE to 000 after read/write access has finished
6	R/W	WRITE_MODE Note 1	<ul><li>2-WIRE multiple write mode</li><li>0: Page write mode</li><li>1: Repeated write mode</li></ul>
5:3	R/W	Reserved	
2:0	R/W	REG_PAGE	I <sup>2</sup> C: 00x: Selects register 0x00 to 0xFF 01x: Selects register 0x100 to 0x17F SPI: 000: Selects register 0x00 to 0x7F 001: Selects register 0x80 to 0xFF 010: Selects register 0x100 to 0x17F >010: Reserved for production and test

**Note 1** Not used for 4-WIRE-IF.





### 9.4.1 Regulators Settings

#### Table 31: BUCK\_ILIM (0xD0)

Bit	Туре	Label	Description
7:5	R/W	Reserved	
4	R/W	BUCK_IALARM	Current Alarm threshold is: 0: BUCKILIM – 300 mA 1: BUCKILIM – 600 mA
3:0	R/W	BUCK_ILIM	Current limit per phase: 0000: 1600 mA 0001: 1800 mA 0010: 2000 mA 0011: 2200 mA 0100: 2400 mA 0100: 2400 mA 0101: 2600 mA 0110: 2800 mA 0111: 3000 mA 1000: 3200 mA 1001: 3400 mA 1011: 3800 mA 1011: 3800 mA 1110: 4000 mA 1110: 4400 mA 1111: 4600 mA

#### Table 32: BUCK\_CONF1 (0xD1)

Bit	Туре	Label	Description
7:5	R/W	PWR_DOWN_CTRL	Voltage ramping during power down 000: 1.25 mV/µs 001: 2.5 mV/µs 010: 5 mV/µs 011: 10 mV/µs 100: 20 mV/µs 101: 30 mV/µs 110: 40 mV/µs 111 : (reserved)





Bit	Туре	Label	Description
4:2	R/W	STARTUP_CTRL	Voltage ramping during start-up 000: 1.25 mV/µs Soft startup with controlled slew rate 001: 2.5 mV/µs 010: 5 mV/µs 011: 10 mV/µs 100: 20 mV/µs Note 1 101: 30 mV/µs 110: 40 mV/µs 111: target voltage applied immediately (no soft start)
1:0	R/W	BUCK_MODE	<ul> <li>00: Low Power/Normal mode controlled via voltage A and B registers</li> <li>01: BUCK always operates in Low power mode</li> <li>10: BUCK always operates in Normal mode</li> <li>11: Automatic mode</li> </ul>

**Note 1** Settings higher than 20 mV/µs may cause significant overshoot

## Table 33: BUCK\_CONF2 (0xD2)

Bit	Туре	Label	Description
7:4	R/W	Reserved	
3	R/W	PH_SH_EN	Enable current dependent phase shedding during Normal mode
2:0	R/W	PHASE_SEL	Phase selection of the multi-phase buck in synchronous mode: 000: 1 phase is selected 001: 2 phases are selected 010: 3 phases are selected (uneven 0/90/180 phase shift) 011: 4 phases are selected 1xx: 8 phases are selected (dual mode only)





## Table 34: VBUCK\_AUTO (0xD4)

Bit	Туре	Label	Description
7	R/W	Reserved	
6:0	R/W	VBUCK_AUTO	Threshold voltage for the Automatic mode: 0000000: 0.30 V 0000001: 0.31 V 0000010: 0.32 V  1000110: 1.0 V  1111101: 1.55 V 1111110: 1.56 V 1111111: 1.57 V

## Table 35: VBUCK\_BASE (0xD5)

Bit	Туре	Label	Description
7	R/W	Reserved	
6:0	R/W	VBUCK_BASE	Sets the base voltage for the buck output when using the DVC interface 0000000: 0.30 V 000001: 0.31 V 0000010: 0.32 V  1000110: 1.0 V  1111101: 1.55 V 1111110: 1.56 V 1111111: 1.57 V





## Table 36: VBUCK\_MAX (0xD6)

Bit	Туре	Label	Description
7	R/W	Reserved	
6:0	R/W	VBUCK_MAX	Sets the maximum voltage allowed for the buck output voltage
			0000000: 0.30 V
			0000001: 0.31 V
			0000010: 0.32 V
			1000110: 1.0 V
			1111101: 1.55 V
			1111110: 1.56 V
			1111111: 1.57 V

## Table 37: VBUCK\_DVC (0xD7)

Туре	Label	Description
R	Reserved	
R	VBUCK_DVC	Internal status of actual target voltage configured via DVC interface 0000000: 0.30 V 0000001: 0.31 V 0000010: 0.32 V  1000110: 1.0 V  1111101: 1.55 V 1111110: 1.56 V 1111111: 1.57 V
	R	R Reserved





## Table 38: VBUCK\_A (0xD8)

Bit	Туре	Label	Description
7	R/W	BUCK_SL_A	<ul><li>0: Configures the BUCK to Normal mode, whenever selecting A voltage settings</li><li>1: Configures the BUCK to Low Power mode, whenever selecting A voltage settings</li></ul>
6:0	R/W	VBUCK_A	0000000: 0.30 V 0000001: 0.31 V 0000010: 0.32 V  1000110: 1.0 V  1111101: 1.55 V 1111110: 1.56 V 1111111: 1.57 V

### Table 39: VBUCK\_B (0xD9)

Bit	Туре	Label	Description
7	R/W	BUCK_SL_B	<ul> <li>0: Configures the BUCK to Normal mode, whenever selecting B voltage settings</li> <li>1: Configures the BUCK to Low Power mode, whenever selecting B voltage settings</li> </ul>
6:0	R/W	VBUCK_B	0000000: 0.30 V 0000001: 0.31 V 0000010: 0.32 V  1000110: 1.0 V  1111101: 1.55 V 1111110: 1.56 V 1111111: 1.57 V





#### **Register Page 2** 9.5

## Table 40: PAGE\_CON (0x100)

Bit	Туре	Label	Description
7	R/W	REVERT	Resets REG_PAGE to 000 after read/write access has finished
6	R/W	WRITE_MODE Note 1	<ul><li>2-WIRE multiple write mode</li><li>0: Page Write mode</li><li>1: Repeated Write mode</li></ul>
5:2	R/W	Reserved	
2:0	R/W	REG_PAGE	I <sup>2</sup> C: 00x: Selects register 0x00 to 0xFF 01x: Selects register 0x100 to 0x17F SPI: 000: Selects register 0x00 to 0x7F 001: Selects register 0x80 to 0xFF 010: Selects register 0x100 to 0x17F >010: Reserved for production and test

Note 1 Not used for 4-WIRE-IF

#### Interface and OTP Settings (shared with DA9063) 9.5.1

	Asserted from Power Commander software after emulated OTP read has finished, automatically cleared when leaving emulated OTP read
	OTP read timing 0: normal read 1: marginal read (for OTP fuse verification)

### Table 42: INTERFACE (0x105)

Bit	Туре	Label	Description
	R		4 MSB of 2-WIRE control interfaces base address XXXX0000
			11010000 = 0xD0 write address of PM 2- WIRE interface (page 0 and 1)
			11010001 = 0xD1 read address of PM 2-WIRE interface (page 0 and 1)
7:4		IF_BASE_ADDR	11010010 = 0xD2 write address of PM-2- WIRE interface (page 2 and 3)
			11010011 = 0xD3 read address of PM-2-WIRE interface (page 2 and 3)
			Code '0000' is reserved for unprogrammed OTP (triggers start-up with hardware default interface address)
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Bit	Туре	Label	Description
3	R	R/W_POL	<ul> <li>4-WIRE: Read/Write bit polarity</li> <li>0: Host indicates reading access via R/W bit = '0'</li> <li>1: Host indicates reading access via R/W bit = '1'</li> </ul>
2	R	СРНА	4-WIRE IF clock phase (see Table 3: 4-WIRE Clock Configurations)
1	R	CPOL	<ul><li>4-WIRE IF clock polarity</li><li>0: SK is low during idle</li><li>1: SK is high during idle</li></ul>
0	R	nCS_POL	<ul><li>4-WIRE chip select polarity</li><li>0: low, (nCS)</li><li>1: high, (CS)</li></ul>

Note 1 Not used for 4-WIRE-IF.

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## Table 43: INTERFACE2 (0x106)

Bit	Туре	Label	Description
7	R22	IF_TYPE	0: Power manager IF is 4-WIRE 1: Power manager IF is 2-WIRE
6	R/W	PM_IF_HSM	Enables continuous high speed mode on 2- WIRE interface if asserted (no master code required)
5	R/W	PM_IF_FMP	Enables 2-WIRE interface operating with fast mode+ timings if asserted
4	R/W	PM_IF_V	Power manager IF in 2-WIRE mode is supplied from: 0: VDDCORE 1: VDD_IO (4-WIRE always from VDD_IO)
3:0	R/W	Reserved	





## **DA9210**

## DA9210 Multi-Phase 12 A DC-DC Buck Converter

## 9.5.2 Application Configuration Settings

### Table 47: CONFIG\_A (0x143)

Bit	Туре	Label	Description
7:5	R	Reserved	
4	R/W	2WIRE_TO	Enables automatic reset of 2-WIRE IF in case of clock stays low for >35 ms 0: Disabled 1: Enabled
3	R/W	GPI_V	GPIs (not configured as Power Manager control inputs) are supplied from: 0: VDDCORE 1: VDD_IO
2	R/W	OC_PG_IRQ_CONF	Configuration for the OC_PG port: 0: Interrupt line 1: Over Current and Power Good
1	R/W	OC_PG_IRQ_TYPE	OC_PG output port is: 0: Push-pull 1: Open-drain (requires external pull-up resistor)
0	R/W	OC_PG_IRQ_LEVEL	OC_PG output port is: 0: Active low 1: Active high





## Table 48: CONFIG\_B (0x144)

Bit	Туре	Label	Description
7:0	R/W	Reserved	

### Table 49: CONFIG\_C (0x145)

Bit	Туре	Label	Description
7	R/W	Reserved	
6	R/W	GPIO6_PUPD	<ul> <li>0: GPI: pull-down resistor disabled</li> <li>GPO (open-drain): pull-up resistor disabled</li> <li>(external pull-up resistor)</li> <li>1: GPI: pull-down resistor enabled</li> <li>GPO (open-drain): pull-up resistor enabled</li> </ul>
5	R/W	GPIO5_PUPD	<ul> <li>0: GPI: pull-down resistor disabled</li> <li>GPO (open drain): pull-up resistor disabled</li> <li>(external pull-up resistor)</li> <li>1: GPI: pull-down resistor enabled</li> <li>GPO (open-drain): pull-up resistor enabled</li> </ul>
4	R/W	GPIO4_PUPD	<ul> <li>0: GPI: pull-down resistor disabled</li> <li>GPO (open drain): pull-up resistor disabled</li> <li>(external pull-up resistor)</li> <li>1: GPI: pull-down resistor enabled</li> <li>GPO (open-drain): pull-up resistor enabled</li> </ul>
3	R/W	GPIO3_PUPD	<ul> <li>0: GPI: pull-down resistor disabled</li> <li>GPO (open drain): pull-up resistor disabled</li> <li>(external pull-up resistor)</li> <li>1: GPI: pull-down resistor enabled</li> <li>GPO (open-drain): pull-up resistor</li> </ul>
2	R/W	GPIO2_PUPD	<ul> <li>0: GPI: pull-down resistor disabled</li> <li>GPO (open drain): pull-up resistor disabled</li> <li>(external pull-up resistor)</li> <li>1: GPI: pull-down resistor enabled</li> <li>GPO (open-drain): pull-up resistor enabled</li> </ul>
1	R/W	GPIO1_PUPD	<ul> <li>0: GPI: pull-down resistor disabled</li> <li>GPO (open drain): pull-up resistor disabled (external pull-up resistor)</li> <li>1: GPI: pull-down resistor enabled</li> <li>GPO (open-drain): pull-up resistor enabled</li> </ul>
0	R/W	GPIO0_PUPD	<ul> <li>0: GPI: pull-down resistor disabled</li> <li>GPO (open drain): pull-up resistor disabled</li> <li>(external pull-up resistor)</li> <li>1: GPI: pull-down resistor enabled</li> <li>GPO (open-drain): pull-up resistor</li> </ul>





### Table 50: CONFIG\_D (0x146)

Bit	Туре	Label	Description
7	R/W	OC_PG_KEEP	0: Normal operation of OC_PG port 1: Over current alarm at OC_PG is masked for 100 µs after the buck rail is valid (when recovering from an out of range condition and during power up)
6	R/W	OC_PG_REL	<ul><li>0: OC_PG port is released 250 µs after BUCK_EN goes low</li><li>1: OC_PG port is released after the ramp down has completed</li></ul>
5:4	R/W	OC_PG_MASK	Over current alarm at OC_PG port is: 00: always active during DVC transitions of the buck converter 01: masked during DVC transitions of the buck converter + 2 µs extra masking at the end 10: masked during DVC transitions of the buck converter + 10 µs extra masking at the end 11: masked during DVC transitions of the buck converter + 50 µs extra masking at the end
3	R/W	READY_EN	GPIO3 is used as READY signal to inform the host processor of DVC ongoing (the GPIO3 needs to be configured as output and any write to GPIO3_MODE will be ignored): 0: Disabled 1: Enabled
2	R/W	FORCE_DVC_IF	0: The DVC_CTRL_EN is automatically reset when the buck converter is disabled 1: The DVC_CTRL_EN is not reset when the buck converter is disabled
1	R/W	DVC_STEP_SIZE	0: $V_{\text{STEPS}}$ is configured to 10 mV when using the DVC interface 1: $V_{\text{STEPS}}$ is configured to 20 mV when using the DVC interface
0	R/W	Reserved	





### Table 51: CONFIG\_E (0x147)

Bit	Туре	Label	Description
7:6	R/W	Reserved	
5:4	R/W	OSC_TUNE	Tune the main 6 MHz oscillator frequency: 00: no tune 01: +180 kHz 10: +360 kHz 11: +540 kHz
3:1	R/W	Reserved	
0	R/W	STAND_ALONE	0: DA9210 is used as companion IC to DA9063 or compliant 1: DA9210 is stand alone or as companion IC with another PMU not DA9063 compliant

### Table 52: MISC\_SUPP (0x14F)

Bit	Туре	Label	Description
7:1	R/W	Reserved	
0	R/W	OTP_CLK_ON	Forces the oscillator and the OTP clock on if asserted

### Table 53: DEVICE\_ID (0x201)

Bit	Туре	Label	Description
7:0	R	DEV_ID	Device ID

### Table 54: DEVICE\_ID (0x202)

Bit	Туре	Label	Description
7:4	R	MRC	Mask Revision Code
3:0	R	VRC	Chip Variant Code

### Table 55: DEVICE-ID (0x203)

Bit	Туре	Label	Description
7:0	R	CUST_ID	Customer ID

### Table 56: CONFIG\_ID (0x204)

Bit	Туре	Label	Description
7:0	R	CONFIG_REV	OTP Settings Revision

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# **10** Application Information

The components recommended in this section are examples selected from requirements of a typical application.

## **10.1 Capacitor Selection**

Ceramic capacitors are used as bypass capacitors on all VDD and output rails. When selecting a capacitor, especially for types with high capacitance and small physical dimension, the DC bias characteristic has to be taken into account.

Application	Value	Size	Temp. Char.	Tol	Rated Voltage	Туре
VDDCORE, output bypass	220 nF	0402	X5R +/- 15%	+/-10%	6.3 V	Murata GRM155R60J224KE01(M E01)
	220 nF	0201	X5R +/- 15%	+/-20%	6.3 V	Murata GRM033R60J224ME90
	200 nF	0201	X5R +/- 15%	+/-10%	16 V	Semco CL03A224KO3NNNC
	220 nF	0402	X7R +/- 15%	+/-10%	16 V	Murata GRM155R71C224KA12D (automotive)
VOUT_SENSE output bypass	47 µF	0805	X5R +/- 15%	+/-20%	4 V	Murata GRM21BR60G476ME15
	47 µF	0603	X5R +/- 15%	+/-20%	4 V	Semco CL10A476MR8NZN
	22 µF	0402	X5R +/- 15%	+/-20%	4 V	Semco CL05A226MR5NZNC
	47 µF	0805	X5R +/- 15%	+/-20%	10 V	Murata GRM21BR61A476ME15L (automotive)
	22 µF	0603	X5R +/- 15%	+/-20%	10 V	Murata GRM188R61A226ME15D (automotive)
	22 µF	0805	X6S +/- 22%	+/-20%	10 V	Murata GRM21BC81A226ME44L (automotive)
VDDx bypass	10 µF	0603	X5R +/- 15%	+/-20%	6.3 V	Murata GRM188R60J106ME84
	10 µF	0805	X7R +/- 15%	+/-10%	6.3 V	Murata GRM21BR70J106KE76L (automotive)
	10 µF	0603	X6S +/- 22%	+/-20%	10 V	Murata GRM188C81A106MA73D (automotive)
VSYS bypass	1 µF	0402	X5R +/- 15%	+/-10%	10 V	Murata GRM155R61A105KE15#

### **Table 57: Recommended Capacitor Types**

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Application	Value	Size	Temp. Char.	Tol	Rated Voltage	Туре
	1 µF	0402	X6S +/- 22%	+/-10%	10 V	Murata GRM155C81A105KA12D (automotive)

## **10.2 Inductor Selection**

Inductors should be selected based upon the following parameters:

- Rated maximum current: Usually a coil provides two current limits: The Isat specifies the maximum current at which the inductance drops by 30 % of the nominal value. The Imax is defined by the maximum power dissipation and is applied to the effective current.
- DC resistance: Critical to converter efficiency and should therefore be minimized.
- Inductance: Given by converter electrical characteristics; 0.47 µH for all DA9210 phases.

### Table 58: Recommended Inductor Types (including only typical values for the parts)

Application	Value	Size	lmax(dc)	lsat	Tolerance	DC Res.	Туре
BUCK	4x 0.47 μΗ	2.0x1.6x1.0 mm	3.3 A	3.5 A	+/-30%	48 mΩ	TOKO DFE201610C 1285AS-H-R47M
	4x 0.47 μΗ	2.0x1.6x1.2 mm	3.8 A	4.2 A	+/-30%	40 mΩ	TOKO DFE201612C 1286AS-H-R47M
	4x 0.47 μΗ	2.5x2.0x1.0 mm	3.6 A	3.9 A	+/-20%	35 mΩ	TOKO DFE252010C 1269AS-H-R47M
	4x 0.47 μΗ	2.5x2.0x1.2 mm	4.4 A	4.7 A	+/-20%	29 mΩ	TOKO DFE252012C 1239AS-H-R47M
	4x0.47 μΗ	2.0x1.6x1.0 mm	2.7 A	3.5 A	+/-20%	38 mΩ	TDK TFM201610A R47M
	4x0.47 μΗ	2.5x2.0x1.0 mm	2.8 A	4.5 A	+/-20%	34 mΩ	TDK TFM252010A R47M
	4x0.47 μΗ	2.0x1.6x1.0 mm	2.7 A	3.56 A	+/-20%	38 mΩ	Cyntec PIFE20161T
	4x0.47 μΗ	2.5x2.0x1.0 mm	3.5 A	4.5 A	+/-20%	34 mΩ	Cyntec PIFE25201T
	4x0.47 μΗ	2.5x2.0x1.2 mm	4.5 A	5.0 A	+/-20%	23 mΩ	Cyntec PIFE25201B
	4x0.47 μΗ	2.5x2.0x1.2 mm	3.7 A	3.9 A	+/-20%	25 mΩ	Cyntec PST25201B
	4x0.47 μΗ	2.0x2.0x1.2 mm	2.8 A	4.2 A	+/-30%	30 mΩ	Taiyo Yuden MDMK2020T R47M
	4x0.47 μΗ	2.5x2.0x1.2 mm	3.9 A	4.8 A	+/-20%	30 mΩ	Taiyo Yuden MAMK2520T R47M
	4x0.47 μΗ	2.0x1.6x1.0 mm	3.2 A	3.6 A	+/-20%	32 mΩ	Murata LQM2MPNR47MGH
	4x 0.47 μΗ	4x4x1.2 mm	8.7 A	6.7 A	+/-20%	14mΩ	Coilcraft XFL4012- 471ME

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## **10.3 Layout Guidelines**

### **10.3.1 General Recommendations**

• Appropriate trace width and amount of vias should be used for all power supply paths. Too high trace resistances can prevent the system operating correctly, for example, efficiency and current ratings of switch mode converters and charger might be degraded. Furthermore the PCB might be exposed to thermal hot spots, which can lead to critical overheating due to the positive temperature coefficient of copper.

Special care must be taken with the DA9210 pad connections. The traces of the outer row should be connected with the same width as the pads and should become wider as soon as possible. For supply pins in the second row, connection to an inner layer is recommended (depending on the maximum current two or more vias might be required).

- A common ground plane should be used, which allows proper electrical and thermal performance. Noise sensitive references like VREF should be referred to a silent ground which is connected at a star point underneath or close to the DA9210 main ground connection.
- Generally all power tracks with discontinuous and / or high currents should be kept as short as possible.
- Noise sensitive analog signals like feedback lines should be kept away from traces carrying pulsed analog or digital signals. This can be achieved by separation (distance) or shielding of quiet signals by ground traces.

### **10.3.2 Switched Mode Supplies**

- The placement of the distributed capacitors at VSYS must ensure that all VDD inputs are connected to a bypass capacitor close to the pads. Using a local power plane underneath the chip for VSYS might be considered.
- The area of switched mode converter transient current loops should be minimized.
- Output capacitors of the LDOCORE should be placed close to DA9210.
- Care must be taken that no current is carried on feedback lines of the buck output voltage VOUT\_SENSE.
- The inductor placement is less critical as parasitic inductances do not matter.

### 10.3.3 DA9210 Thermal Connection, Land Pad and Stencil Design

- The DA9210 provides a center ground area of balls, which are soldered to the PCB's center ground pad. This PCB ground pad must be connected with as many vias and as direct as possible to the PCB's main ground plane in order to achieve good thermal performance.
- Solder mask openings for the ball landing pads must be arranged to prohibit solder balls flowing into vias.

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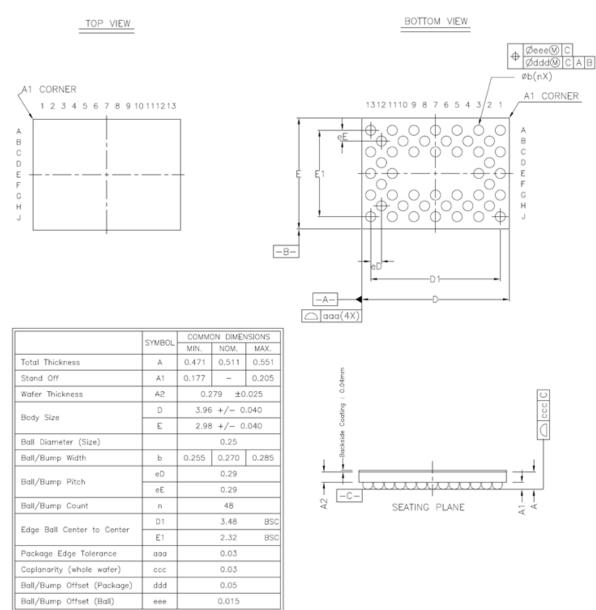
## DA9210



## DA9210 Multi-Phase 12 A DC-DC Buck Converter

## 11 Package Information

## 11.1 Package Outline Drawing (48 WLCSP)



### Figure 47: Package Outline Drawing (48 WLCSP)

The package uses Dialog's innovative RouteEasy<sup>™</sup> technology (Patent pending) implementing a high number of IOs on a small footprint package without the need for cost intensive PCB technology. All signals can be routed within two signal layers of a standard PCB (single trace between IC lands) and the inter layer connection uses drilled VIAs (no need for micro-VIA or VIA-in-land technology). Please contact your local Dialog Semiconductor support for more information about PCB Layout Guidelines.





## 11.1.1 RouteEasy<sup>™</sup> Technology Chart

BGA ball diameter: 0.27 (10 mm) BGA land size: 0.25 (10 mm) Via hole size (FHS):  $\leq$  0.2 - 0.25 (8 - 10 mm) Via pad size:  $\leq$  0.45 (18 mm) Anti-pad size:  $\leq$  0.65 (26 mm) Trace width:  $\geq$  0.1 - 0.125 (4 - 5 mm) Trace/Trace space:  $\geq$  0.1 - 0.125 (4 - 5 mm) Trace/Pad/Land space:  $\geq$  0.1 (4 mm)

### NOTE

Dimensions are in millimeter. The PCB design complexity is compatible with 0.8 mm BGA pin pitch (similar to IPC 6012B Class 2). Better thermal relief and improved high current return paths are achieved by using a reduced copper-to-hole distance for the inner layer reference planes (inner anti-pad size: 0.55 mm or 0.6 mm).





## 11.2 Package Outline Drawing (42 VF-BGA)

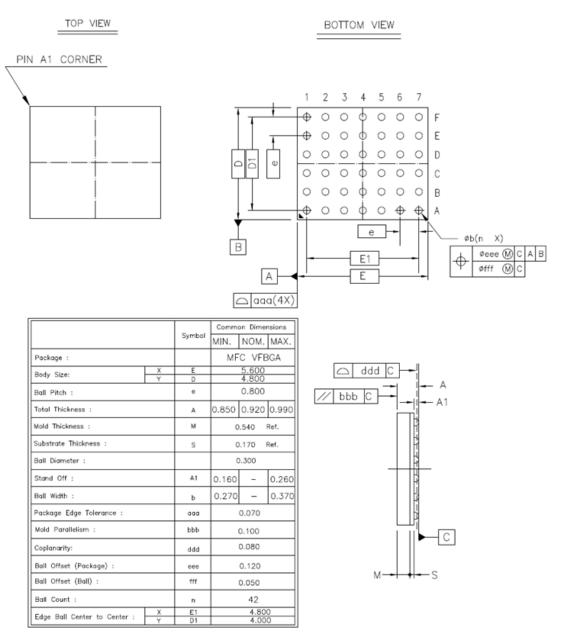


Figure 48: Package Outline Drawing (42 VF-BGA)





## **11.3 Soldering Information**

Refer to the JEDEC standard J-STD-020 for relevant soldering information. This document can be downloaded from http://www.jedec.org.





## **12 Ordering Information**

The ordering number consists of the part number followed by a suffix indicating the packing method. For details and availability, please consult Dialog Semiconductor's customer portal or your local sales representative.

### Table 59: Ordering Information

Part Number	Package	Size (mm)	Shipment Form	Pack Quantity	Comments
DA9210-xxUK2	48 WLCSP		Tape and Reel	5000	
DA9210-xxFN2-A	42 VFBGA		Tape and Real	5000	Automotive AEC-Q100 Grade 3





# **Appendix A Definitions**

## A.1 Power Dissipation and Thermal Design

When designing with the DA9210, consideration must be given to power dissipation as if the device exceeds the package power dissipation, the internal thermal sensor will shut down the device until. until it has cooled sufficiently.

The package includes a thermal management paddle to enable improved heat spreading on the PCB.

Linear regulators operating with a high current and high differential voltage between input and output will dissipate the following power:

$$\mathsf{P}_{\mathsf{diss}} = (\mathsf{V}_{\mathsf{in}} - \mathsf{V}_{\mathsf{out}})^* \mathsf{I}_{\mathsf{out}}$$

Example – a regulator supplying 150 mA @ 2.8 V from a fully charged lithium battery ( $V_{DD} = 4.1 V$ )

P<sub>diss</sub> = (4.1 V - 2.8 V) \* 0.15 A = 195 mW

For switching regulators

 $P_{OUT} = P_{IN} * efficiency$ 

Therefore

 $P_{diss} = P_{IN} - P_{OUT}$ 

$$\mathsf{P}_{\mathsf{diss}} = \frac{\mathsf{P}_{\mathsf{out}}}{\mathsf{Efficiency}} - \mathsf{P}_{\mathsf{out}}$$

$$P_{diss} = P_{out} * \left(\frac{1}{Efficiency} - 1\right)$$

 $P_{diss} = I_{out} * V_{out} * \left(\frac{1}{Efficiency} - 1\right)$ 

Example – an 85 % efficient buck converter supplying 1.2 V@ 400 mA

$$P_{diss} = 1.2V * 0.4A * \left(\frac{1}{0.85} - 1\right) = 85mW$$

As the DA9210 is a multiple regulator configuration each supply must be considered and summed to give the total device dissipation (current drawn from the reference and control circuitry can be considered negligible in these calculations).

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## A.2 Regulator Parameter – Line Regulation

Static line regulation is a measurement that indicates a change in the regulator output voltage  $\Box V_{reg}$  (regulator operating with a constant load current) in response to a change in the input voltage  $\Box V_{in}$ . Transient line regulation is a measurement of the peak change  $\Box V_{reg}$  in regulated voltage seen when the line input voltage changes.

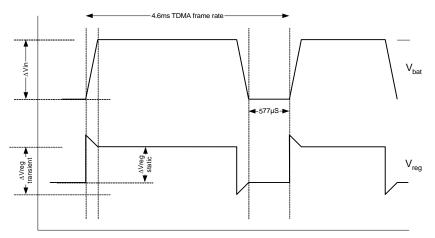


Figure 49: Line Regulation

## A.3 Regulator Parameter – Load Regulation

Static load regulation is a measurement that indicates a change in the regulator output voltage  $\Delta V_{reg}$  in response to a change in the regulator loading  $\Delta$ load whilst the regulator input voltage remains constant. Transient load regulation is a measurement of the peak change in regulated voltage  $\Delta V_{reg}$  seen when the regulator load changes.

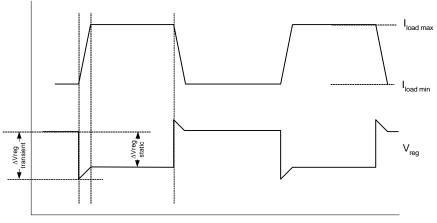


Figure 50: Load Regulation

Please contact Dialog Semiconductor for latest application information on the DA9210 and other power management devices.





# **Revision History**

Revision	Date	Description	
1.0	Oct 2011	Initial release	
1.1 to 1.x	Mar 2015	Regular updates for development and specification alignment	
2.0	Sep 2016	Updated and standardized to latest publishing guidelines	
2.1	28-Oct-2016	PFM mode withdrawn. Further information on PFM mode limitations in DA9210 can be found in Application Note AN-PM-052.	





### **Status Definitions**

Revision	Datasheet Status	Product Status	Definition	
1. <n></n>	Target	Development	This datasheet contains the design specifications for product development. Specifications may be changed in any manner without notice.	
2. <n></n>	Preliminary	Qualification	This datasheet contains the specifications and preliminary characterization data for products in pre-production. Specifications may be changed at any time without notice in order to improve the design.	
3. <n></n>	Final	Production	This datasheet contains the final specifications for products in volume production. The specifications may be changed at any time in order to improve the design, manufacturing and supply. Major specification changes are communicated via Customer Product Notifications. Datasheet changes are communicated via www.dialog-semiconductor.com.	
4. <n></n>	Obsolete	Archived	This datasheet contains the specifications for discontinued products. The information is provided for reference only.	

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Datasheet

Revision 2.1

