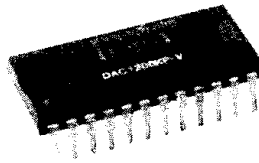




# DAC1200KP-V



## Integrated Circuit 12-Bit Resolution DIGITAL-TO-ANALOG CONVERTER

### FEATURES

- COMPLETE D/A CONVERTER:  
INTERNAL REFERENCE  
 $\pm 10V$  OUTPUT OPERATIONAL AMPLIFIER
- $\pm 1/2$ LSB LINEARITY ERROR
- MONOTONICITY GUARANTEED  $0^{\circ}C$  TO  $+70^{\circ}C$
- SETTLING TIME  $7\mu s$ , MAX
- $\pm 12V$  TO  $\pm 15V$  POWER SUPPLY OPERATION
- 24-PIN MOLDED PLASTIC DIP
- LOWEST COST 12-BIT DAC

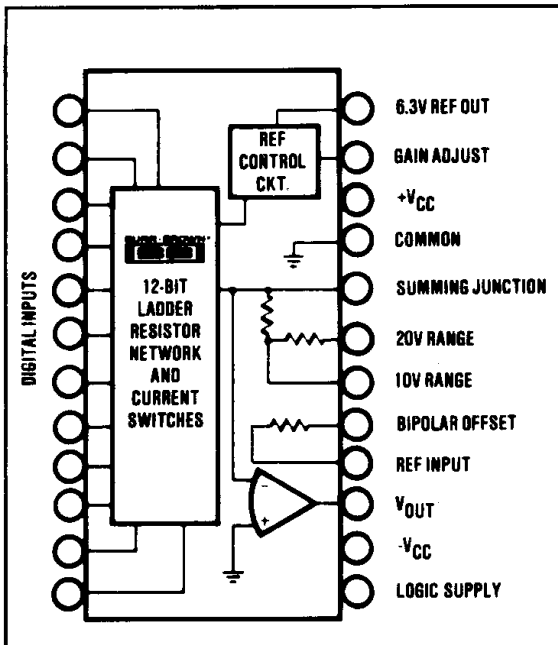
### DESCRIPTION

The low price of DAC1200KP-V makes this 12-bit resolution D/A converter the best value available for commercial applications.

The DAC1200 offers TTL input compatibility, guaranteed monotonicity over  $0^{\circ}C$  to  $+70^{\circ}C$  and settling time of  $7\mu sec$  maximum. It comes complete with internal reference and output operational amplifier.

This precision component is made possible using Burr-Brown's proprietary monolithic integrated circuit process which has been optimized for converter circuits. A stable subsurface reference zener, laser-trimmed thin-film ladder resistors, and high speed current switches combine to give superior performance over the rated temperature range.

DAC1200 is priced and specified for applications where high resolution and monotonicity are the key application parameters and where tightly specified performance over temperature is not required. Because of the low price, it is feasible to use a 12-bit D/A converter for new applications in communications systems, control systems, medical systems, electronic games and personal computer peripherals.



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DAC1200KP-V

6.1

INSTRUMENTATION D/A CONVERTERS

# SPECIFICATIONS

## ELECTRICAL

Typical at +25°C and  $\pm V_{CC} = 12V$  or  $15V$ ,  $V_{DD} = +5V$  unless otherwise noted.

MODEL	DAC1200KP-V	UNITS
<b>INPUTS</b>		
<b>DIGITAL INPUTS</b>		
Input Code <sup>(1)</sup>	CSB, COB	Bits
Resolution	12	
Digital Logic Inputs <sup>(2)</sup> :		
$V_{IH}$ , min to max	+2.4 to $+V_{DD}$	V
$V_{IL}$ , min to max	0 to +0.8	V
$I_{IH}$ , $V_I = +2.7V$ , max	+20	$\mu A$
$I_{IL}$ , $V_I = +0.4V$ , max	-400	$\mu A$
<b>TRANSFER CHARACTERISTICS</b>		
<b>ACCURACY</b>		
Linearity Error, max <sup>(3)</sup>	$\pm 0.018$	% of FSR <sup>(4)</sup>
Differential Linearity Error, max	$\pm 0.024$	% of FSR
Gain Error, max <sup>(5)(6)</sup>	$\pm 0.3$	%
Unipolar Offset Error <sup>(5)(7)</sup>	$\pm 20$	mV
Bipolar Offset Error, max <sup>(5)(8)</sup>	$\pm 40$	mV
Monotonicity Over 0°C to +70°C <sup>(9)</sup>	12	Bits
Sensitivity of Gain to Power Supply Variations:		
+ $V_{CC}$ and - $V_{CC}$	$\pm 0.003$	% of FSR/ $\%V_{CC}$
$V_{DD}$	$\pm 0.0002$	% of FSR/ $\%V_{DD}$
<b>TEMPERATURE COEFFICIENTS</b>		
Gain	$\pm 10$	ppm/°C
Bipolar Offset	$\pm 8$	ppm of FSR/°C
<b>SETTLING TIME to <math>\pm 0.012\%</math> of FSR<sup>(10)</sup></b>		
20V Step and 2k $\Omega$ Load, max	7	$\mu sec$
<b>OUTPUT</b>		
<b>ANALOG OUTPUT</b>		
Voltage Range, min	$\pm 2.5, \pm 5, \pm 10, +5, +10$	V
Current, min <sup>(11)(12)</sup>	$\pm 5$	mA
Impedance	0.05	$\Omega$
<b>REFERENCE OUTPUT</b>		
Voltage <sup>(13)</sup>	+6.3	$V_{DD}$
Source Current Available for External Loads, max	+1.5	mA
Temperature Coefficient	$\pm 10$	ppm/°C
<b>POWER SUPPLY REQUIREMENTS</b>		
<b>RATED VOLTAGE</b>		
+ $V_{CC}$ /- $V_{CC}$ <sup>(14)</sup>	+15/-15	V
$V_{DD}$ <sup>(15)</sup>	+5	V
<b>CURRENT (no load), max<sup>(16)</sup></b>		
+ $V_{CC}$ /- $V_{CC}$	+12/-25	mA
$V_{DD}$	+10	mA
<b>TEMPERATURE RANGE</b>		
For parameters specified over temp, min to max	0 to +70	°C
Storage, min to max	-60 to +100	°C

NOTES: (1) CSB = Complementary Straight Binary (unipolar), COB = Complementary Offset Binary (bipolar). (2) Digital inputs are TTL-compatible for  $V_{DD}$  over the range of +4.5V to  $+V_{CC}$ . Digital input specs are guaranteed over 0°C to +70°C. These specs are tested at 25°C only. (3)  $\pm 0.018\%$  of FSR is 3/4LSB at 12 bits. (4) FSR means Full Scale Range and is 20V for a  $\pm 10V$  range. (5) Adjustable to zero with external potentiometer. (6) Adjusting the Gain Adjust potentiometer rotates the transfer function about 0V for unipolar operation and about minus full scale (-FS) for bipolar operation. (7) Error at input code FFF<sub>H</sub> for unipolar operation (output at 0V). (8) Error at input code FFF<sub>H</sub> for bipolar operation (output at minus full scale, -FS). (9) Guaranteed. Tested at 25°C only. (10) Guaranteed. Not tested. (11) For operation with supply voltages of less than  $\pm 13V$ , load current must be limited to 1mA. (12) Output may be indefinitely shorted to Common without damage. (13) Tolerance is  $\pm 5\%$ . (14) Range of operation is  $\pm 11.4V$  to  $\pm 16.5V$ . (15)  $V_{DD}$  may be operated up to  $+V_{CC}$ . Digital input logic threshold remains at +1.4V over the  $V_{DD}$  range. (16) Typical power supply currents are about 70% of the maximum.

## ABSOLUTE MAXIMUM RATINGS

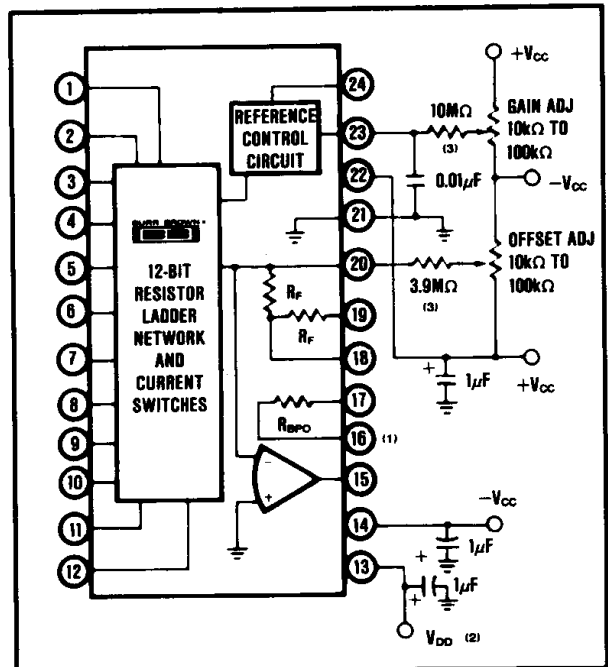
+ $V_{CC}$ to Common	0 to +18V
- $V_{CC}$ to Common	0 to -18V
$V_{DD}$ to Common	0 to +7V
Digital Inputs (pins 1-12) to Common	-0.4V to +18V
External Voltage Applied to Range Resistors	$\pm 12V$
REF OUT	Indefinite short to Common
External Voltage Applied to Analog Output	-5V to +5V
Power Dissipation	1000mW
Operating Temperature	0 to +70°C
Storage Temperature	-60°C to +100°C

NOTE: Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. Exposure to absolute maximum conditions for extended periods may affect device reliability.

## PIN ASSIGNMENTS

Pin	Description	Pin	Description
1	Bit 1 (MSB)	13	Logic Supply, $V_{DD}$
2	Bit 2	14	- $V_{CC}$
3	Bit 3	15	$V_{OUT}$
4	Bit 4	16	Reference Input
5	Bit 5	17	Bipolar Offset
6	Bit 6	18	10V Range
7	Bit 7	19	20V Range
8	Bit 8	20	Summing Junction
9	Bit 9	21	Common
10	Bit 10	22	+ $V_{CC}$
11	Bit 11	23	Gain Adjust
12	Bit 12 (LSB)	24	6.3V Reference Out

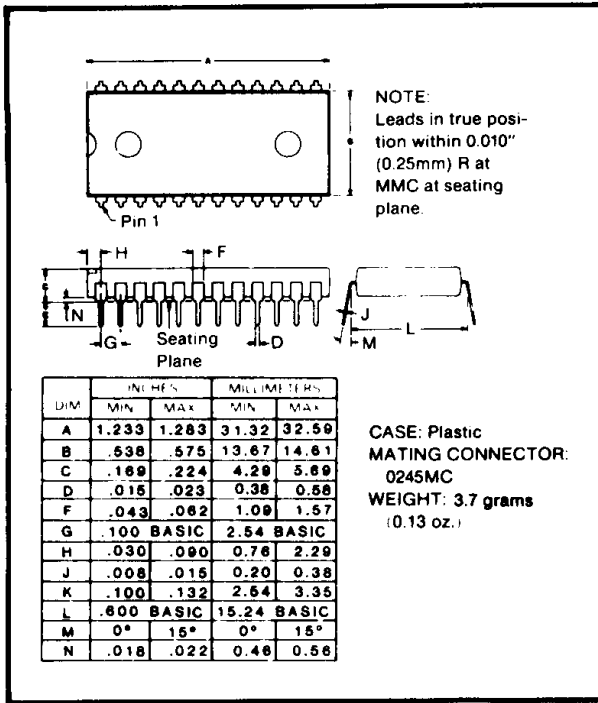
## CONNECTION DIAGRAM



Output Range	Output Voltage Range Connections				
	Digital Input Codes	Connect Pin 15 to	Connect Pin 17 to	Connect Pin 19 to	Connect Pin 16 to
$\pm 10$	COB	19	20	15	24
$\pm 5$	COB	18	20	NC	24
$\pm 2.5V$	COB	18	20	20	24
0 to +10V	CSB	18	21	NC	24
0 to +5V	CSB	18	21	20	24

NOTES: (1) Pin 16 is used only to connect the bipolar offset resistor. An external reference voltage may not be used. (2) If connected to  $+V_{CC}$ , which is permissible, power dissipation increases 75mW typ, 100mW max. (3) Values shown are for  $\pm 15V$  supplies. For supplies below  $\pm 13.5V$  use 2.7M $\Omega$  in place of 3.9M $\Omega$  and 7.5M $\Omega$  in place of 10M $\Omega$ .

## MECHANICAL



## INSTALLATION AND OPERATING INSTRUCTIONS

### POWER SUPPLY CONNECTIONS

**Decoupling:** For optimum performance and noise rejection, power supply decoupling capacitors should be added as shown in the Connection Diagrams. These capacitors (1 $\mu$ F to 10 $\mu$ F tantalum) should be located close to the DAC1200.

### $\pm 12$ V OPERATION

The DAC1200 is fully specified for operation on  $\pm 12$ V power supplies. However, to use the  $\pm 10$ V and 0 to +10V ranges of the voltage output models, the power supplies must be  $\pm 13$ V or greater. All other voltage output ranges and all current output ranges provide satisfactory operation with  $\pm 11.4$ V supplies. The supplies should be balanced to obtain optimum performance.

### EXTERNAL OFFSET AND GAIN ADJUSTMENT

Offset and Gain may be trimmed by installing external Offset and Gain potentiometers. Connect these potentiometers as shown in the connection diagrams and adjust as described below. TCR of the potentiometers should be 100ppm/ $^{\circ}$ C or less. The 3.9M $\Omega$  and 10M $\Omega$  resistors (20% carbon or better) should be located close to the DAC1200 to prevent noise pick-up. For operation with

supplies of less than  $\pm 13.5$ V, use 2.7M $\Omega$  and 7.5M $\Omega$  resistors in place of the 3.9M $\Omega$  and 10M $\Omega$  resistors, respectively. If it is not convenient to use these high value resistors, an equivalent "T" network, as shown in Figure 1, may be substituted in each case. The Gain Adjust (pin 23) is a high impedance point and a 0.001 $\mu$ F to 0.01 $\mu$ F ceramic capacitor should be connected from this pin to Common (pin 21) to reduce noise pick-up. Figures 2 and 3 illustrate the relationship of Offset and Gain adjustments to unipolar and bipolar D/A converter output.

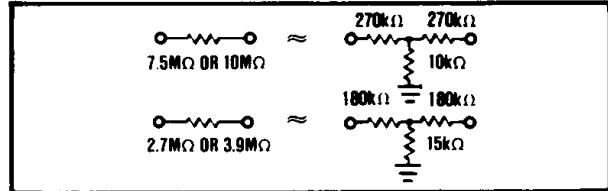


FIGURE 1. Equivalent Resistances.

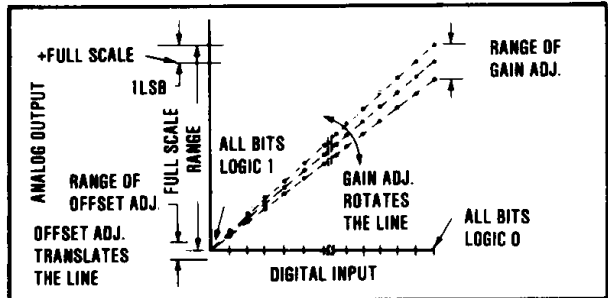


FIGURE 2. Relationship of Offset and Gain Adjustments for a Unipolar D/A Converter.

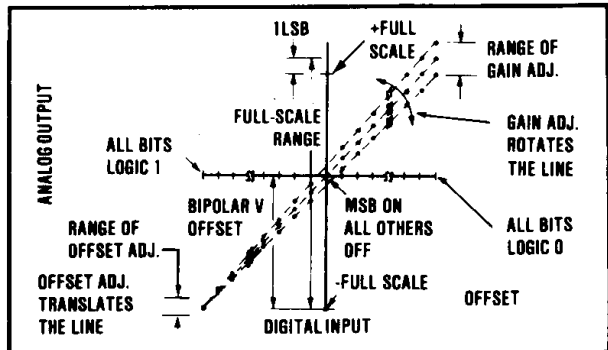


FIGURE 3. Relationship of Offset and Gain Adjustments for a Bipolar D/A Converter.

**Offset Adjustment:** For unipolar (CSB) configurations, apply the digital input code that should produce zero potential output and adjust the Offset potentiometer for zero output.

For bipolar (COB, CTC) configurations, apply the digital input code that should produce the maximum negative output voltage and adjust the Offset potentiometer for minus full-scale voltage. Example: If the Full Scale Range is connected for 20V, the maximum negative output voltage is -10V. See Table I for corresponding codes. Offset should be adjusted before gain.

**Gain Adjustment:** For either unipolar or bipolar configurations, apply the digital input that should give the

maximum positive voltage output. Adjust the Gain potentiometer for this positive full-scale voltage. See Table I for positive full-scale voltages.

TABLE I. Digital Input/Analog Output.

Digital Input		Analog Output	
		0 to +10V	±10V
MSB	LSB		
0000000000		+9.9976V	+9.9951V
0111111111		+5.0000V	0.0000V
1000000000		+4.9976V	-0.0049V
1111111111		0.0000V	-10.0000V
One LSB		2.44mV	4.88mV

To obtain values for other ranges:

0 to +5V range: divide 0 to +10V range values by 2.

±5V range: divide ±10V range values by 2.

±2.5V range: divide ±10V range values by 4.