DP3842A(AM)/43A(AM)/44A(AM)/45A(AM)

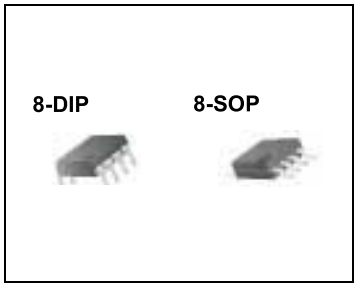
Current Mode PWM Controller

Features

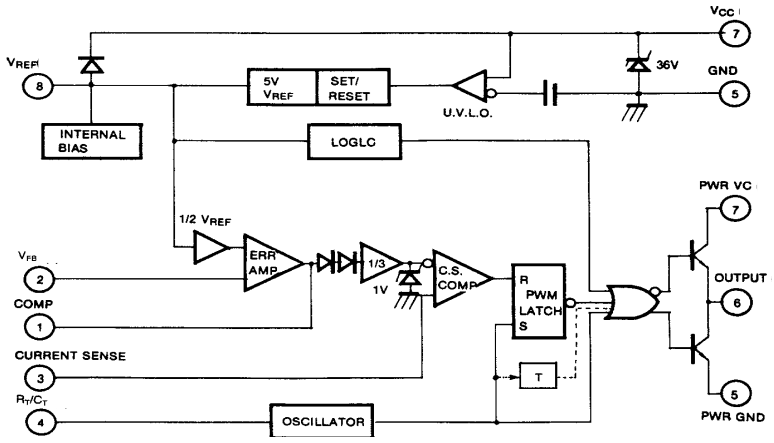
- Low Start up Current
- Maximum Duty Clamp
- UVLO With Hysteresis
- 384xA Operating Frequency up to 300KHz
- 384xAM Operating Frequency up to 500KHz

Description

The DP3842/DP3843/DP3844/DP3845 are fixed frequencycurrent-mode PWM controller. They are specially designed for Off-Line and DC to DC converter applications with minimum external components. These integrated circuits feature a trimmed oscillator for precise duty cycle control, a temperature compensated reference, high gain error amplifier, current sensing comparator and a high current totempole output for driving a Power MOSFET. The DP3842 and DP3844 have UVLO thresholds of 16V (on) and 10V (off). The DP3843 and DP3845 are 8.5V(on) and 7.9V (off). The DP3842 and DP3843 can operate within 100% duty cycle. The DP3844 and DP3845 can operate with 50% duty cycle.



Internal Block Diagram



* NORMALLY 8DIP/8SOP PIN NO.

* TOGGLE FLIP FLOP USED ONLY IN DP3844, DP3845

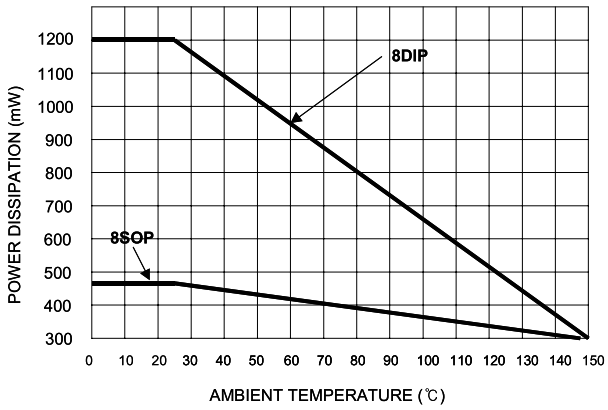
Absolute Maximum Ratings

Parameter	Symbol	Value	Unit
Supply Voltage	V_{CC}	30	V
Output Current	I_O	±1	A
Analog Inputs (Pin 2.3)	$V_{(ANA)}$	-0.3 to 5.5	V
Error Amp Output Sink Current	$I_{SINK\ (E.A)}$	10	mA
Power Dissipation at T_A≤25°C (8DIP)	P_D(Note1,2)	--	mW
Power Dissipation at T_A≤25°C (8SOP)	P_D(Note1,2)	--	mW
Power Dissipation at T_A≤25°C (8SOP)	P_D(Note1,2)	--	mW
Storage Temperature Range	T_{STG}	-65 ~ +150	°C
Lead Temperature (Soldering, 10sec)	T_{LEAD}	+260	°C

Note:

1. Board Thickness 1.6mm, Board Dimension 76.2mm ×114.3mm, (Reference EIA / JSED51-3, 51-7)
2. Do not exceeed P_D and SOA (Safe Operation Area)

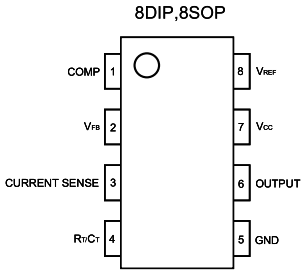
Power Dissipation Curve



Thermal Data

Characteristic	Symbol	8-DIP	8-SOP	14-SOP	Unit
Thermal Resistance Junction-ambient	$R_{thj\text{-}amb(MAX)}$	100	265	180	°C/W

Pin Array



Electrical Characteristics

(V_{CC}=15V, R_T=10kΩ, C_T=3.3nF, T_A= 0°C to +70°C, unless otherwise specified)

Parameter	Symbol	Conditions	Min.	Typ.	Max.	Unit
REFERENCE SECTION						
Reference Output Voltage	V_{REF}	T_J = 25°C, I_{REF} = 1mA	4.90	5.00	5.10	V
Line Regulation	ΔV_{REF}	12V ≤ V_{CC} ≤ 25V	-	6	20	mV
Load Regulation	ΔV_{REF}	1mA ≤ I_{REF} ≤ 20mA	-	6	25	mV
Short Circuit Output Current	I_{SC}	T_A = 25°C	-	-100	-180	mA
OSCILLATOR SECTION						
Oscillation Frequency	f	T_J = 25°C	47	52	57	kHz
Frequency Change with Voltage	$\Delta f/\Delta V_{CC}$	12V ≤ V_{CC} ≤ 25V	-	0.05	1	%
Oscillator Amplitude	V_{OSC}	-	-	1.6	-	V_{P-P}
ERROR AMPLIFIER SECTION						
Input Bias Current	I_{BIAS}	-	-	-0.1	-2	μA
Input Voltage	$V_{I(E>A)}$	V_{pin1} = 2.5V	2.42	2.50	2.58	V
Open Loop Voltage Gain	G_{VO}	2V ≤ V_O ≤ 4V (Note3)	65	90	-	dB
Power Supply Rejection Ratio	PSRR	12V ≤ V_{CC} ≤ 25V (Note3)	60	70	-	dB
Output Sink Current	I_{SINK}	V_{pin2} = 2.7V, V_{pin1} = 1.1V	2	7	-	mA
Output Source Current	I_{SOURCE}	V_{pin2} = 2.3V, V_{pin1} = 5V	-0.5	-1.0	-	mA
High Output Voltage	V_{OH}	V_{pin2} = 2.3V, R_L = 15kΩ to GND	5	6	-	V
Low Output Voltage	V_{OL}	V_{pin2} = 2.7V, R_L = 15kΩ to Pin 8	-	0.8	1.1	V
CURRENT SENSE SECTION						
Gain	G_V	(Note 1 & 2)	2.85	3	3.15	V/V
Maximum Input Signal	$V_{I(MAX)}$	V_{pin1} = 5V(Note 1)	0.9	1	1.1	V
Power Supply Rejection Ratio	PSRR	12V ≤ V_{CC} ≤ 25V (Note 1,3)	-	70	-	dB
Input Bias Current	I_{BIAS}	-	-	-3	-10	μA
OUTPUT SECTION						
Low Output Voltage	V_{OL}	I_{SINK} = 20mA	-	0.08	0.4	V
		I_{SINK} = 200mA	-	1.4	2.2	V
High Output Voltage	V_{OH}	I_{SOURCE} = 20mA	13	13.5	-	V
		I_{SOURCE} = 200mA	12	13.0	-	V
Rise Time	t_R	T_J = 25°C, C_L= 1nF (Note 3)	-	45	150	ns
Fall Time	t_F	T_J = 25°C, C_L= 1nF (Note 3)	-	35	150	ns
UNDER-VOLTAGE LOCKOUT SECTION						
	$V_{TH(ST)}$	DP3842/DP3844	14.5	16.0	17.5	V
		DP3843/DP3845	7.8	8.4	9.0	V
Min. Operating Voltage (After Turn On)		DP3842/DP3844	8.5	10.0	11.5	V
		DP3843/DP3845	7.0	7.6	8.2	V

Electrical Characteristics (Continued)

(V_{CC}=15V, R_T=10kΩ, C_T=3.3nF, T_A= 0°C to +70°C, unless otherwise specified)

Parameter	Symbol	Conditions	Min.	Typ.	Max.	Unit
PWM SECTION						
Max. Duty Cycle	$D_{(Max)}$	DP3842/DP3843	95	97	100	%
	$D_{(Max)}$	DP3844/DP3845	47	48	50	%
Min. Duty Cycle	$D_{(MIN)}$	-	-	-	0	%
TOTAL STANDBY CURRENT						
Start-Up Current	I_{ST}	-	-	0.17	0.3	mA
Operating Supply Current	$I_{CC(OPR)}$	V_{pin3}=V_{pin2}=ON	-	14	17	mA
Zener Voltage	V_Z	I_{CC} = 25mA	30	38	-	V

Adjust V_{CC} above the start threshould before setting at 15V

Note:

1. Parameter measured at trip point of latch
2. Gain defined as:

$$A = \frac{\Delta V_{pin1}}{\Delta V_{pin3}} \quad ,0 \le V_{pin3} \le 0.8V$$

3. These parameters, although guaranteed, are not 100 tested in production.

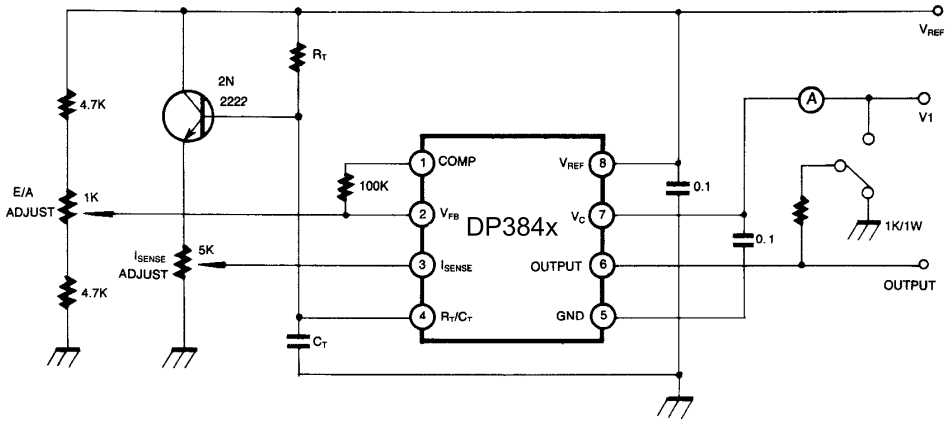


Figure 1. Open Loop Test Circuit

High peak currents associated with capacitive loads necessitate careful grounding techniques Timing and bypass capacitors should be connected close to pin 5 in a single point ground. The transistor and 5kΩ potentiometer are used to sample the oscillator waveform and apply an adjustable ramp to pin 3.

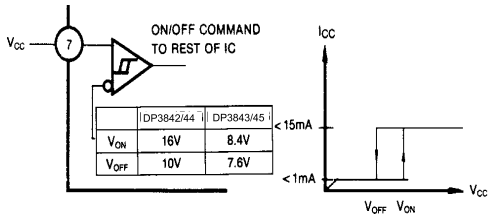


Figure 2. Under Voltage Lockout

During Under-Voltage Lock-Out, the output driver is biased to a high impedance state. Pin 6 should be shunted to ground with a bleeder resistor to prevent activating the power switch with output leakage current.

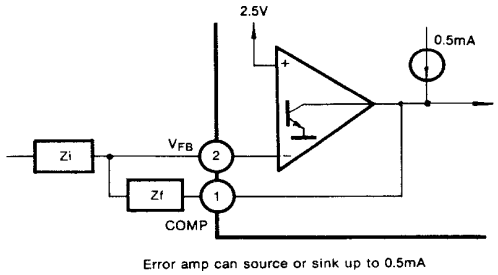


Figure 3. Error Amp Configuration

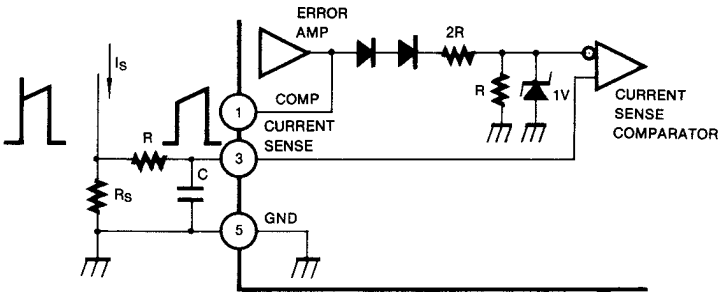


Figure 4. Current Sense Circuit

Peak current (I_S) is determined by the formula:

$$I_{S(MAX)} = \frac{1.0V}{R_S}$$

A small RC filter may be required to suppress switch transients.

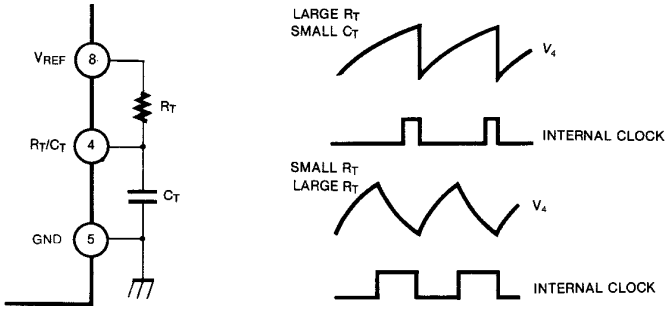


Figure 5. Oscillator Waveforms and Maximum Duty Cycle

Oscillator timing capacitor, C_T, is charged by V_{REF} through R_T and discharged by an internal current source. During the discharge time, the internal clock signal blanks the output to the low state. Selection of R_T and C_T therefore determines both oscillator frequency and maximum duty cycle. Charge and discharge times are determined by the formulas:

$t_c = 0.55\ R_T\ C_T$

$$t_D = R_T C_T \ln\left(\frac{0.0063R_T - 2.7}{0.0063R_T - 4}\right)$$

Frequency, then, is: $f=(t_c + t_d)^{-1}$

$$\text{For } RT > 5K\Omega, f = \frac{1.8}{R_T C_T}$$

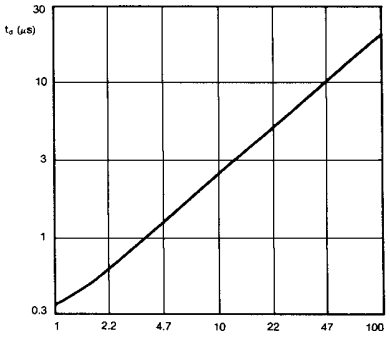


Figure 6. Oscillator Dead Time & Frequency

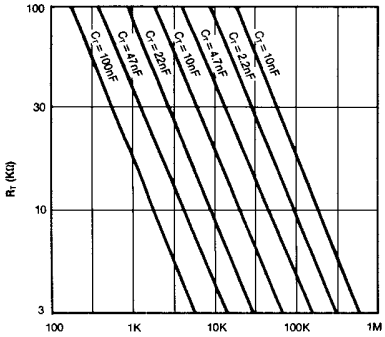


Figure 7. Timing Resistance vs Frequency

(Deadtime vs C_T RT > 5kΩ)

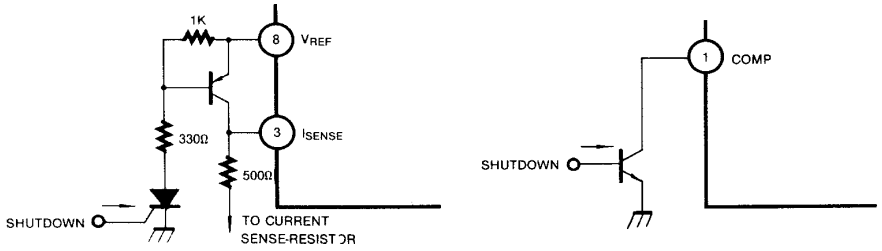


Figure 8. Shutdown Techniques

Shutdown of the DP3842 can be accomplished by two methods; either raise pin 3 above 1V or pull pin 1 below a voltage two diode drops above ground. Either method causes the output of the PWM comparator to be high (refer to block diagram). The PWM latch is reset dominant so that the output will remain low until the next clock cycle after the shutdown condition at pins 1 and/or 3 is removed. In one example, an externally latched shutdown may be accomplished by adding an SCR which will be reset by cycling V_{CC} below the lower UVLO threshold. At this point the reference turns off, allowing the SCR to reset.

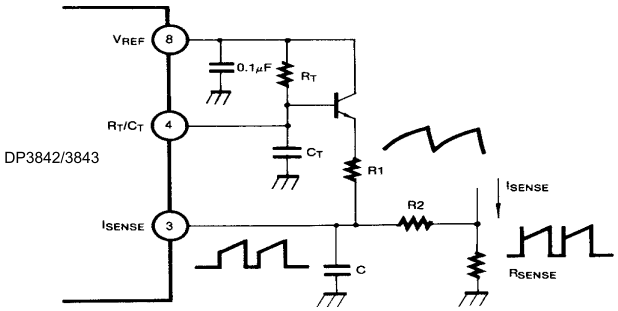


Figure 9. Slope Compensation

A fraction of the oscillator ramp can be resistively summed with the current sense signal to provide slope compensation for converters requiring duty cycles over 50%. Note that capacitor, C_T, forms a filter with R2 to suppress the leading edge switch spikes.

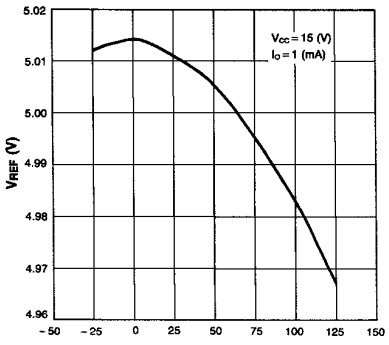


Figure 10. Temperature Drift (Vref)

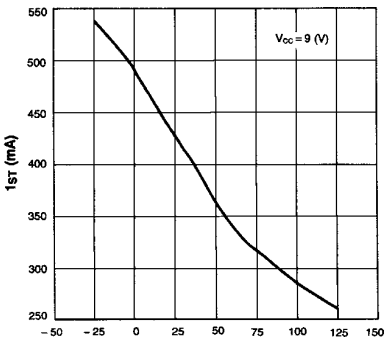


Figure 11. Temperature Drift (Ist)

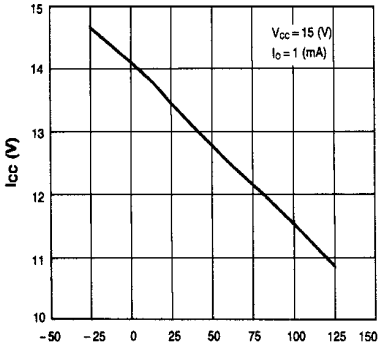
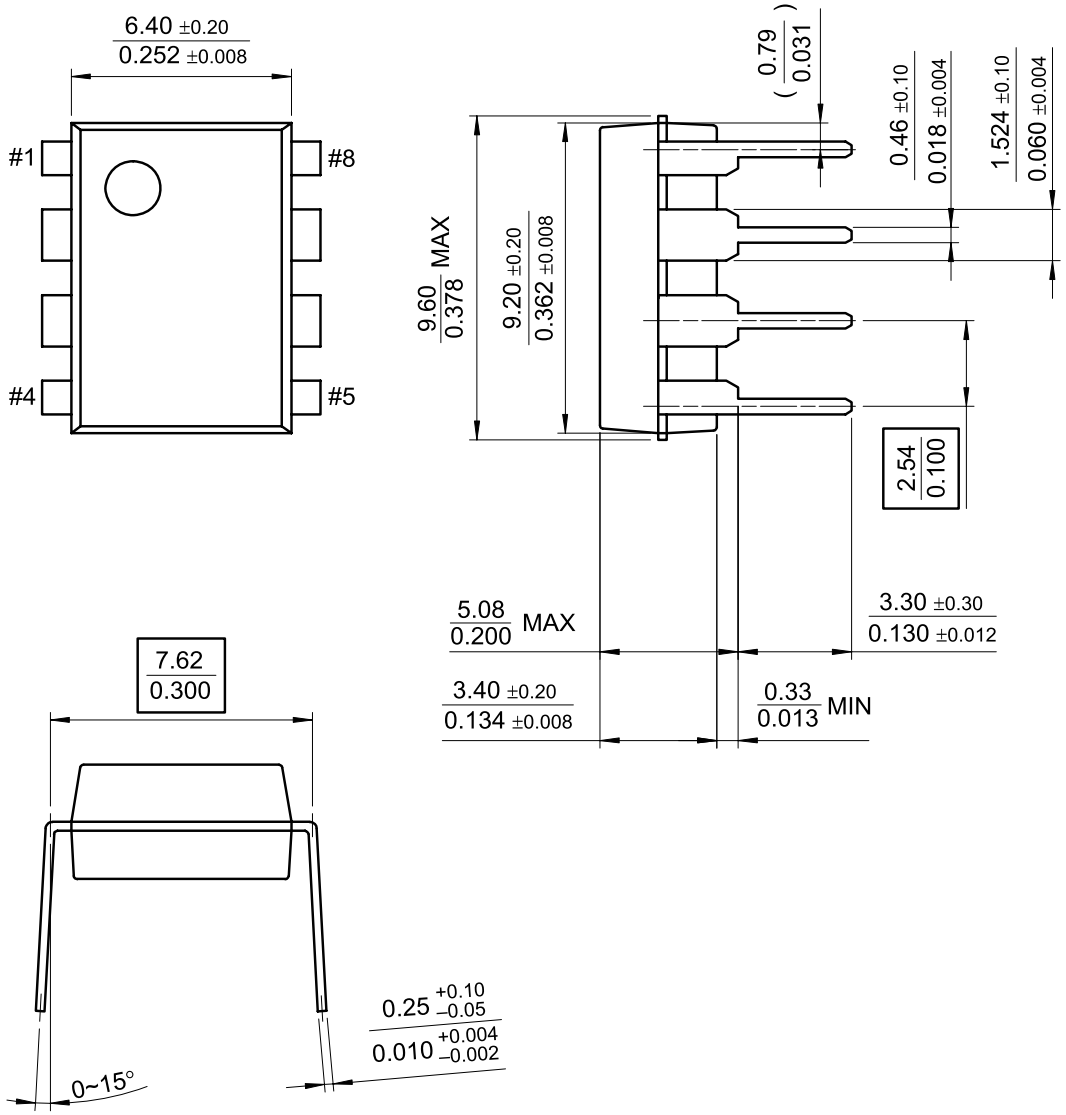


Figure 12. Temperature Drift (Icc)

Mechanical Dimensions

Package

8-DIP



Mechanical Dimensions (Continued)

Package

8-SOP