

DS90C241/DS90C124

5-35MHz DC-Balanced 24-Bit LVDS Serializer and Deserializer

General Description

The DS90C241/124 Chipset translates a 24-bit parallel bus into a fully transparent data/control LVDS serial stream with embedded clock information. This single serial stream simplifies transferring a 24-bit bus over PCB traces and cable by eliminating the skew problems between parallel data and clock paths. It saves system cost by narrowing data paths that in turn reduce PCB layers, cable width, and connector size and pins.

The DS90C241/124 incorporates LVDS signaling on the high-speed I/O. LVDS provides a low power and low noise environment for reliably transferring data over a serial transmission path. By optimizing the serializer output edge rate for the operating frequency range EMI is further reduced.

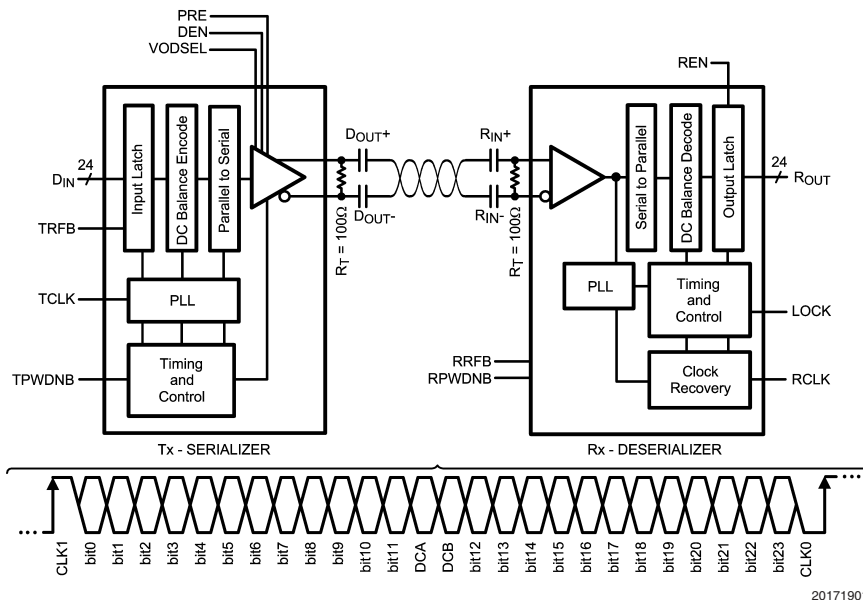
In addition the device features pre-emphasis to boost signals over longer distances using lossy cables. Internal DC balanced encoding/decoding is used to support AC-Coupled interconnects.

Features

- 5 MHz–35 MHz clock embedded and DC-Balancing 24:1 and 1:24 data transmissions
- User defined Pre-Emphasis driving ability through external resistor on LVDS outputs and capable to drive up to 10 meters shielded twisted-pair cable

- User selectable clock edge for parallel data on both Transmitter and Receiver
- Internal DC Balancing encode/decode – Supports AC-coupling interface with no external coding required
- Individual power-down controls for both Transmitter and Receiver
- Embedded clock CDR (clock and data recovery) on Receiver and no external source of reference clock needed
- All codes RDL (random data lock) to support hot-pluggable applications
- LOCK output flag to ensure data integrity at Receiver side
- Balanced T_{SETUP}/T_{HOLD} between RCLK and RDATA on Receiver side
- PTO (progressive turn-on) LVCMOS outputs to reduce EMI and minimize SSO effects
- All LVCMOS inputs and control pins have internal pulldown
- On-chip filters for PLLs on Transmitter and Receiver
- 48-pin TQFP package
- Pure CMOS .35 μ m process
- Power supply range $3.3V \pm 10\%$
- Temperature range $-40^{\circ}C$ to $+105^{\circ}C$
- 8 kV HBM ESD structure

Block Diagram



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Absolute Maximum Ratings (Note 1)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/Distributors for availability and specifications.

Supply Voltage (V_{CC})	-0.3V to +4V
LVC MOS/LVTTL Input Voltage	-0.3V to ($V_{CC} + 0.3V$)
LVC MOS/LVTTL Output Voltage	-0.3V to ($V_{CC} + 0.3V$)
LVDS Receiver Input Voltage	-0.3V to 3.9V
LVDS Driver Output Voltage	-0.3V to 3.9V
LVDS Output Short Circuit Duration	10 ms
Junction Temperature	+150°C
Storage Temperature	-65°C to +150°C
Lead Temperature (Soldering, 4 seconds)	+260°C
Maximum Package Power Dissipation Capacity Package De-rating:	
48L TQFP	$1/\theta_{JA}$ °C/W above +25°C
DS90C241	
θ_{JA}	45.8 (4L*); 75.4 (2L*) °C/W
θ_{JC}	21.0°C/W

DS90C124

θ_{JA}	45.4 (4L*); 75.0 (2L*) °C/W
θ_{JC}	21.1°C/W
	*JEDEC

ESD Rating (HBM)	>8 kV
ESD Rating (ISO10605)	DS90C241 meets ISO 10605
$R_D = 2$ k Ω , $C_S = 330$ pF	
Contact Discharge (D_{OUT+} , D_{OUT-}) to GND	±8 kV
Air Discharge (D_{OUT+} , D_{OUT-}) to GND	±25 kV

Recommended Operating Conditions

	Min	Nom	Max	Units
Supply Voltage (V_{CC})	3.0	3.3	3.6	V
Operating Free Air Temperature (T_A)	-40	+25	+105	°C
Clock Rate	5		35	MHz
Supply Noise			±100	mV _{P-P}

Electrical Characteristics

Over recommended operating supply and temperature ranges unless otherwise specified.

Symbol	Parameter	Conditions	Pin/Freq.	Min	Typ	Max	Units
LVC MOS/LVTTL DC SPECIFICATIONS							
V_{IH}	High Level Voltage		Tx: DIN[23:0], TCLK, TPWDNB, DEN, TRFB, DCAOFF, DCBOFF, VODSEL	2.0		V_{CC}	V
V_{IL}	Low Level Input Voltage		Rx: RPWDNB, RRFB, REN	GND		0.8	V
V_{CL}	Input Clamp Voltage	$I_{CL} = -18$ mA (Note 8)			-0.8	-1.5	V
I_{IN}	Input Current	$V_{IN} = 0V$ or 3.6V	Tx: DIN[23:0], TCLK, TPWDNB, DEN, TRFB, DCAOFF, DCBOFF, VODSEL	-10	±5	+10	μA
			Rx: RPWDNB, RRFB, REN	-20	±5	+20	μA
V_{OH}	High Level Output Voltage	$I_{OH} = -4$ mA	Rx: ROUT[23:0], RCLK, LOCK	2.3	3.0	V_{CC}	V
V_{OL}	Low Level Output Voltage	$I_{OL} = +4$ mA		GND	0.33	0.5	V
I_{OS}	Output Short Circuit Current	$V_{OUT} = 0V$ (Note 8)		-40	-70	-110	mA
I_{OZ}	TRI-STATE® Output Current	RPWDNB, REN = 0V $V_{OUT} = 0V$ or 2.4V	Rx: ROUT[23:0], RCLK, LOCK	-30	±0.4	+30	μA
LVDS DC SPECIFICATIONS							
V_{TH}	Differential Threshold High Voltage	$V_{CM} = +1.2V$	Rx: R_{IN+} , R_{IN-}			+100	mV
V_{TL}	Differential Threshold Low Voltage			-100		mV	
I_{IN}	Input Current	$V_{IN} = +2.4V$, $V_{CC} = 3.6V$ or 0V				±200	μA
		$V_{IN} = 0V$, $V_{CC} = 3.6V$ or 0V				±200	μA

Electrical Characteristics (Continued)

Over recommended operating supply and temperature ranges unless otherwise specified.

Symbol	Parameter	Conditions	Pin/Freq.	Min	Typ	Max	Units
LVDS DC SPECIFICATIONS							
V_{OD}	Output Differential Voltage (D_{OUT+}) - (D_{OUT-})	$R_L = 100\Omega$, w/o Pre-emphasis VODSEL = L (VODSEL = H) (Figure 16)	Tx: D_{OUT+} , D_{OUT-}	250 (450)	400 (750)	600 (1200)	mV
ΔV_{OD}	Output Differential Voltage Unbalance	$R_L = 100\Omega$, w/o Pre-emphasis			10	50	mV
V_{OS}	Offset Voltage	$R_L = 100\Omega$, w/o Pre-emphasis		1.00	1.25	1.50	V
ΔV_{OS}	Offset Voltage Unbalance	$R_L = 100\Omega$, w/o Pre-emphasis			1	50	mV
I_{OS}	Output Short Circuit Current	DOUT = 0V, DIN = H, TPWDNB, DEN = 2.4V, VODSEL = L		-2		-8	mA
		DOUT = 0V, DIN = H, TPWDNB, DEN = 2.4V, VODSEL = H	-7		-13	mA	
I_{OZ}	TRI-STATE Output Current	TPWDNB, DEN = 0V, DOUT = 0V or 2.4V	-15	± 1	+15	μA	

SER/DES SUPPLY CURRENT (DVDD*, PVDD* and AVDD* pins) *Digital, PLL, and Analog VDDs

I_{CCT}	Serializer (Tx) Total Supply Current (includes load current)	$R_L = 100\Omega$ $R_{PRE} = OFF$ VODSEL = H/L Checker-board pattern (Figure 1)	f = 35 MHz		40	65	mA
		$R_L = 100\Omega$ $R_{PRE} = 6 k\Omega$ VODSEL = H/L Checker-board pattern (Figure 1)	f = 35 MHz		45	70	mA
	Serializer (Tx) Total Supply Current (includes load current)	$R_L = 100\Omega$ $R_{PRE} = OFF$ VODSEL = H/L	f = 35 MHz		40	65	mA
		$R_L = 100\Omega$ $R_{PRE} = 6 k\Omega$ VODSEL = H/L Random pattern	f = 35 MHz		45	70	mA
I_{CCTZ}	Serializer (Tx) Supply Current Power-down	TPWDNB = 0V (All other LVCMOS Inputs = 0V)				800	μA
I_{CCR}	Deserializer (Rx) Total Supply Current (includes load current)	$C_L = 8 pF$ LVCMOS Output Checker-board pattern (Figure 2)	f = 35 MHz			75	mA
	Deserializer (Rx) Total Supply Current (includes load current)	$C_L = 8 pF$ LVCMOS Output Random pattern	f = 35 MHz			50	mA
I_{CCRZ}	Deserializer (Rx) Supply Current Power-down	RPWDNB = 0V (All other LVCMOS Inputs = 0V, $R_{IN+}/R_{IN-} = 0V$)				50	μA

Serializer Timing Requirements for TCLK

Over recommended operating supply and temperature ranges unless otherwise specified.

Symbol	Parameter	Conditions	Min	Typ	Max	Units
t_{TCP}	Transmit Clock Period	(Figure 6)	28.6	T	200	ns
t_{TCIH}	Transmit Clock High Time		0.4T	0.5T	0.6T	ns
t_{TCIL}	Transmit Clock Low Time		0.4T	0.5T	0.6T	ns
t_{CLKT}	TCLK Input Transition Time	(Figure 5)		3	6	ns

Serializer Timing Requirements for TCLK (Continued)

Over recommended operating supply and temperature ranges unless otherwise specified.

Symbol	Parameter	Conditions	Min	Typ	Max	Units
t_{JIT}	TCLK Input Jitter	(Note 9)			33	ps (RMS)

Serializer Switching Characteristics

Over recommended operating supply and temperature ranges unless otherwise specified.

Symbol	Parameter	Conditions	Min	Typ	Max	Units
t_{LLHT}	LVDS Low-to-High Transition Time	$R_L = 100\Omega$, (Figure 3) $C_L = 10$ pF to GND			0.6	ns
t_{LHLT}	LVDS High-to-Low Transition Time	VODSEL = L			0.6	ns
t_{DIS}	DIN (23:0) Setup to TCLK	$R_L = 100\Omega$, $C_L = 10$ pF to GND	5			ns
t_{DIH}	DIN (23:0) Hold from TCLK	(Note 8)	5			ns
t_{HZD}	DOUT \pm HIGH to TRI-STATE Delay	$R_L = 100\Omega$,			15	ns
t_{LZD}	DOUT \pm LOW to TRI-STATE Delay	$C_L = 10$ pF to GND			15	ns
t_{ZHD}	DOUT \pm TRI-STATE to HIGH Delay	(Figure 7) (Note 4)			200	ns
t_{ZLD}	DOUT \pm TRI-STATE to LOW Delay				200	ns
t_{PLD}	Serializer PLL Lock Time	$R_L = 100\Omega$, (Figure 8)			10	ms
t_{SD}	Serializer Delay	$R_L = 100\Omega$, (Figure 9) VODSEL = L, TRFB = H		3.5T + 2.85	3.5T + 10	ns
		$R_L = 100\Omega$, (Figure 9) VODSEL = L, TRFB = L		3.5T + 2.85	3.5T + 10	ns
TxOUT_E_O	TxOUT_Eye_Opening (respect to ideal)	5–35 MHz (Figure 14) (Notes 8, 9, 13)	0.75			UI (Note 10)

Deserializer Switching Characteristics

Over recommended operating supply and temperature ranges unless otherwise specified.

Symbol	Parameter	Conditions	Pin/Freq.	Min	Typ	Max	Units
t_{RCP}	Receiver out Clock Period	$t_{RCP} = t_{TCP}$ (Note 8)	RCLK	28.6		200	ns
t_{RDC}	RCLK Duty Cycle		RCLK	45	50	55	%
t_{CLH}	LVC MOS Low-to-High Transition Time	$C_L = 8$ pF (lumped load) (Figure 4)	ROUT [23:0], LOCK, RCLK		2.5	3.5	ns
t_{CHL}	LVC MOS High-to-Low Transition Time				2.5	3.5	ns
t_{ROS}	ROUT (7:0) Setup Data to RCLK (Group 1)	(Figure 11)	ROUT [7:0]	(0.40)* t_{RCP}	(29/56)* t_{RCP}		ns
t_{ROH}	ROUT (7:0) Hold Data to RCLK (Group 1)			(0.40)* t_{RCP}	(27/56)* t_{RCP}		ns
t_{ROS}	ROUT (15:8) Setup Data to RCLK (Group 2)	(Figure 11)	ROUT [15:8], LOCK	(0.40)* t_{RCP}	0.5* t_{RCP}		ns
t_{ROH}	ROUT (15:8) Hold Data to RCLK (Group 2)			(0.40)* t_{RCP}	0.5* t_{RCP}		ns
t_{ROS}	ROUT (23:16) Setup Data to RCLK (Group 3)	(Figure 11)	ROUT [23:16]	(0.40)* t_{RCP}	(27/56)* t_{RCP}		ns
t_{ROH}	ROUT (23:16) Hold Data to RCLK (Group 3)			(0.40)* t_{RCP}	(29/56)* t_{RCP}		ns

Deserializer Switching Characteristics (Continued)

Over recommended operating supply and temperature ranges unless otherwise specified.

Symbol	Parameter	Conditions	Pin/Freq.	Min	Typ	Max	Units
t_{HZR}	HIGH to TRI-STATE Delay	(Figure 12)	ROUT [23:0], RCLK, LOCK		3	10	ns
t_{LZR}	LOW to TRI-STATE Delay				3	10	ns
t_{ZHR}	TRI-STATE to HIGH Delay				3	10	ns
t_{ZLR}	TRI-STATE to LOW Delay				3	10	ns
t_{DD}	Deserializer Delay	(Figure 10)	RCLK		$[4+(3/56)]T$ +5.9	$[4+(3/56)]T$ +14	ns
t_{DRDL}	Deserializer PLL Lock Time from Powerdown	(Figure 13) (Notes 7, 8)	5 MHz		5	50	ms
			35 MHz		5	50	ms
RxIN_TOL_L	Receiver INput TOLerance Left,	(Figure 15) (Notes 6, 8, 10)	5 MHz–35 MHz			0.25	UI
RxIN_TOL_R	Receiver INput TOLerance Right,	(Figure 15) (Notes 6, 8, 10)	5 MHz–35 MHz			0.25	UI

Note 1: “Absolute Maximum Ratings” are those values beyond which the safety of the device cannot be guaranteed. They are not meant to imply that the devices should be operated at these limits. The table of “Electrical Characteristics” specifies conditions of device operation.

Note 2: Typical values are given for $V_{CC} = 3.3V$ and $T_A = +25^{\circ}C$.

Note 3: Current into device pins is defined as positive. Current out of a device pin is defined as negative. Voltages are referenced to ground except VOD, ΔVOD , VTH and VTL which are differential voltages.

Note 4: When the Serializer output is tri-stated, the Deserializer will lose PLL lock. Resynchronization MUST occur before data transfer.

Note 5: t_{DRDL} is the time required by the deserializer to obtain lock when exiting powerdown mode. t_{DRDL} is specified with an external synchronization pattern.

Note 6: RxIN_TOL is a measure of how much phase noise (jitter) the deserializer can tolerate in the incoming data stream before bit errors occur. It is a measurement in reference with the ideal bit position, please see National’s AN-1217 for detail.

Note 7: The Deserializer PLL lock time (t_{DRDL}) may vary depending on input data patterns and the number of transitions within the pattern.

Note 8: Parameter is guaranteed by design and characterization using statistical analysis.

Note 9: t_{JIT} (@BER of $10e-9$) specifies the allowable jitter on TCLK. t_{JIT} not included in TxOUT_E_O parameter.

Note 10: UI – Unit Interval, equivalent to one ideal serialized data bit width. The UI scales with frequency.

Note 11: Figures 1, 2, 9, 10, 13 show a falling edge data strobe (TCLK IN/RCLK OUT).

Note 12: Figures 6, 11 show a rising edge data strobe (TCLK IN/RCLK OUT).

Note 13: TxOUT_E_O is affected by pre-emphasis value.

AC Timing Diagrams and Test Circuits

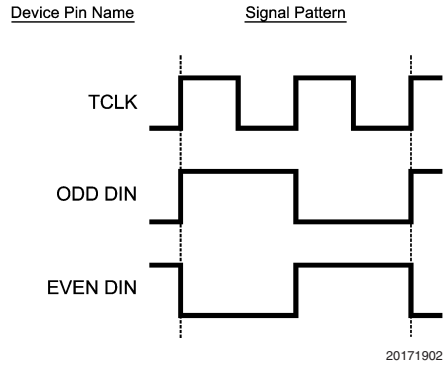


FIGURE 1. Serializer Input Checker-board Pattern

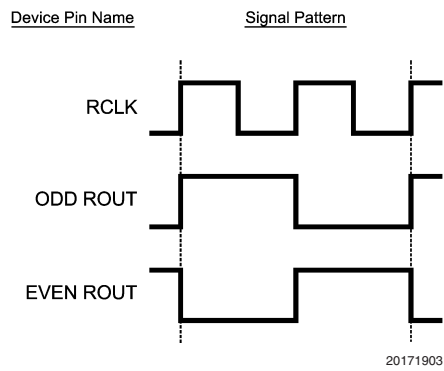


FIGURE 2. Deserializer Output Checker-board Pattern

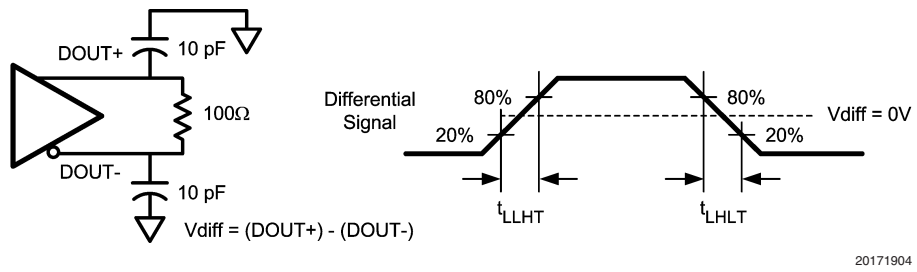


FIGURE 3. Serializer LVDS Output Load and Transition Times

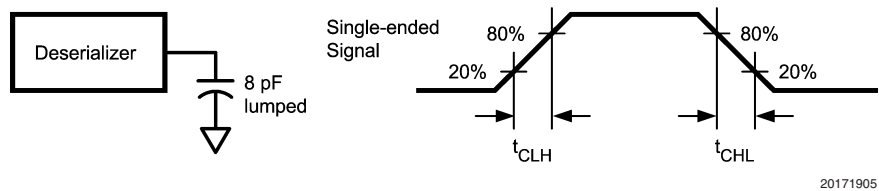


FIGURE 4. Deserializer LVCMOS/LVTTL Output Load and Transition Times

AC Timing Diagrams and Test Circuits (Continued)

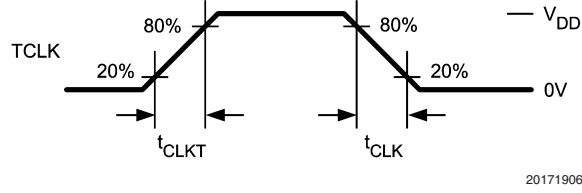


FIGURE 5. Serializer Input Clock Transition Times

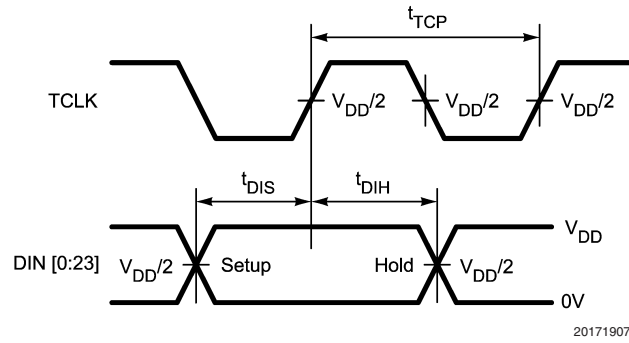
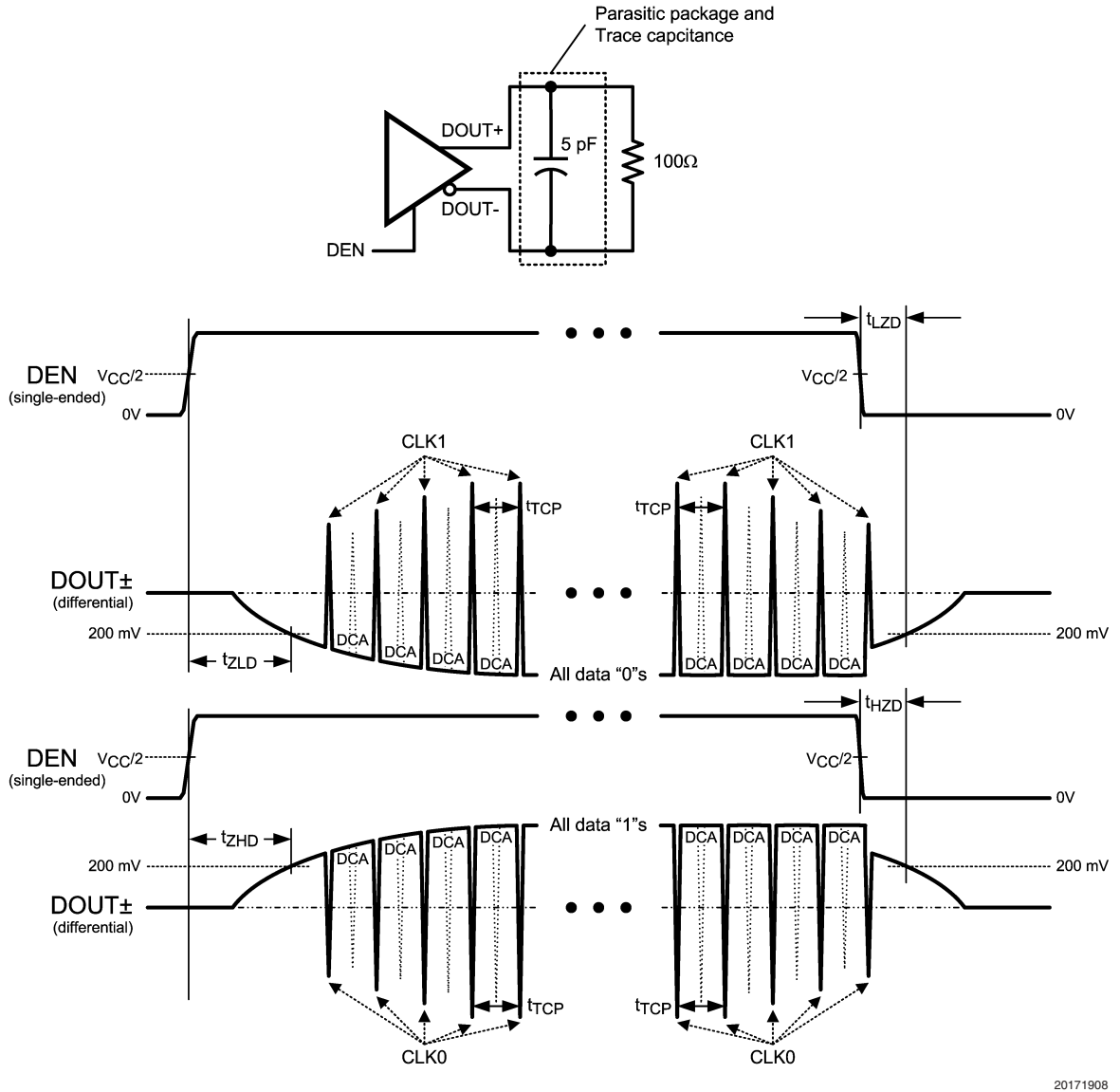


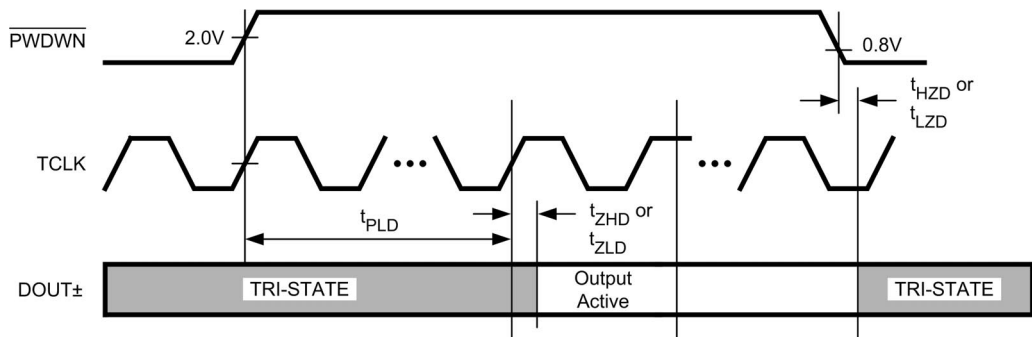
FIGURE 6. Serializer Setup/Hold Times

AC Timing Diagrams and Test Circuits (Continued)



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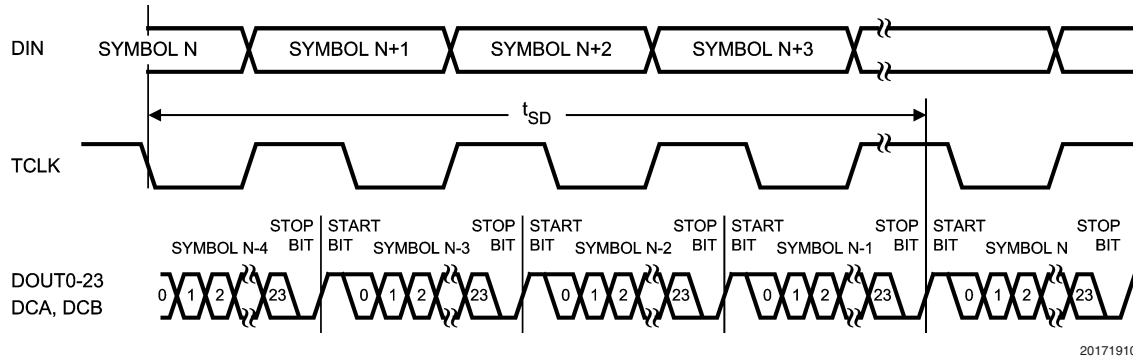
FIGURE 7. Serializer TRI-STATE Test Circuit and Delay



20171909

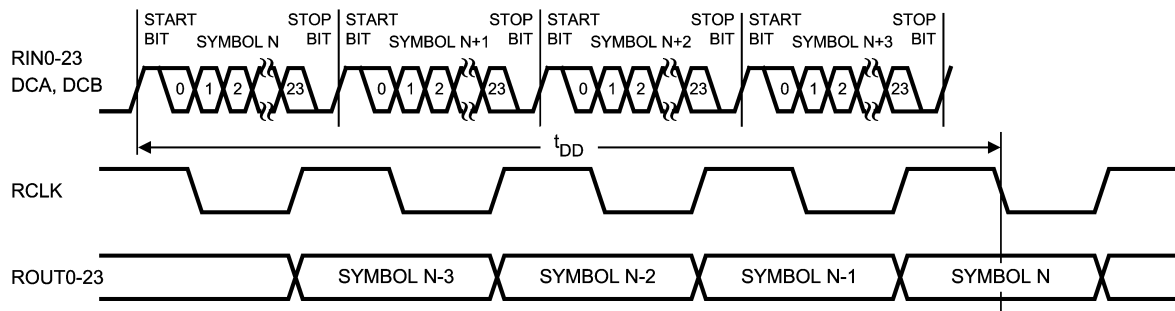
FIGURE 8. Serializer PLL Lock Time, and TPWDNB TRI-STATE Delays

AC Timing Diagrams and Test Circuits (Continued)



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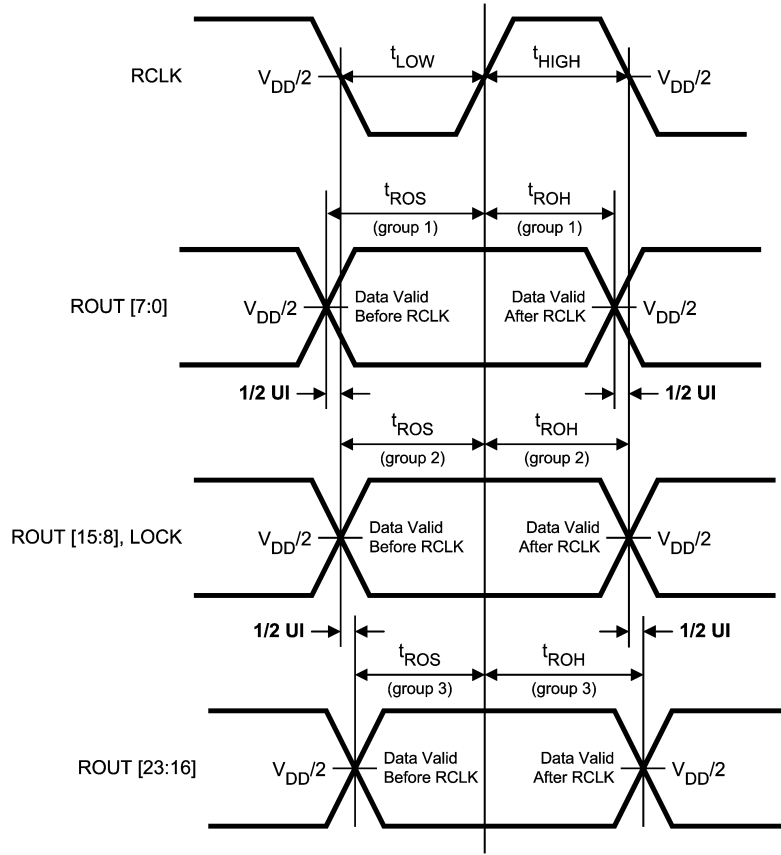
FIGURE 9. Serializer Delay



20171911

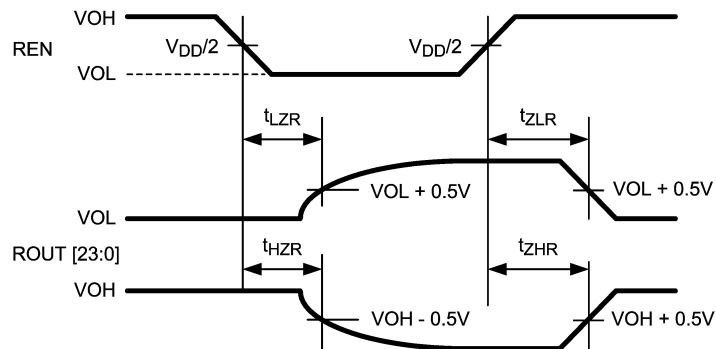
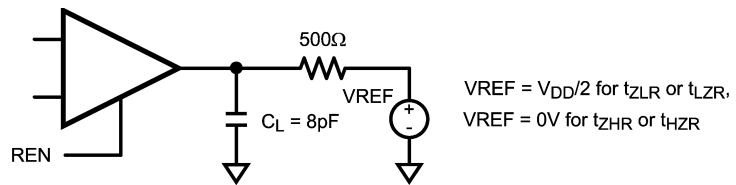
FIGURE 10. Deserializer Delay

AC Timing Diagrams and Test Circuits (Continued)



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FIGURE 11. Deserializer Setup and Hold Times



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Note: C_L includes instrumentation and fixture capacitance within 6 cm of ROUT[23:0]

FIGURE 12. Deserializer TRI-STATE Test Circuit and Timing

AC Timing Diagrams and Test Circuits (Continued)

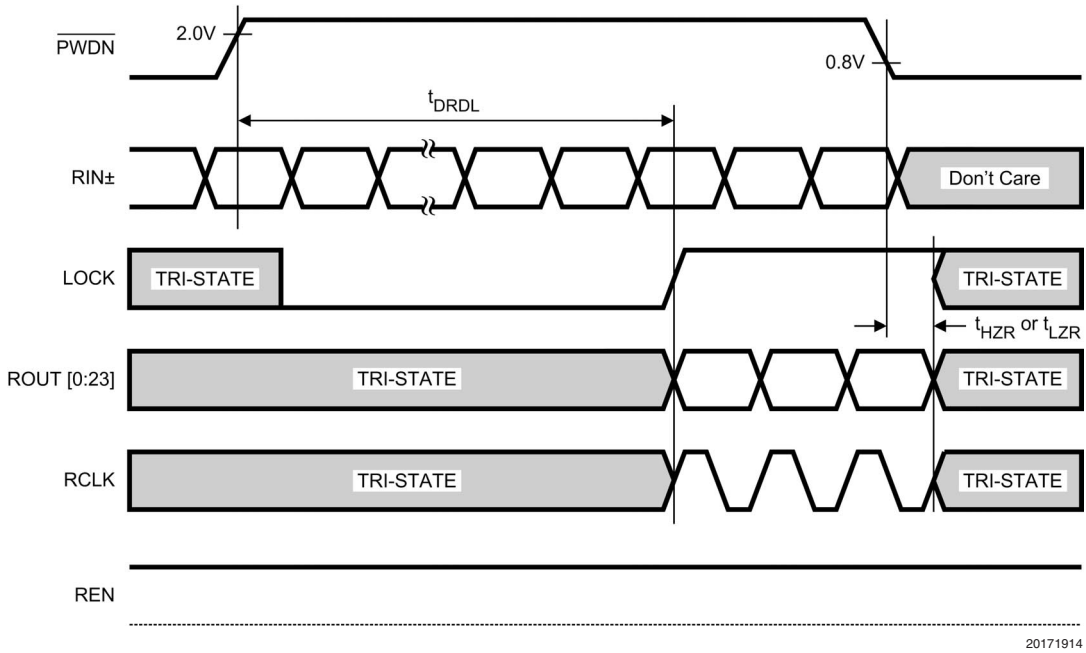


FIGURE 13. Deserializer PLL Lock Times and RPWDNB TRI-STATE Delay

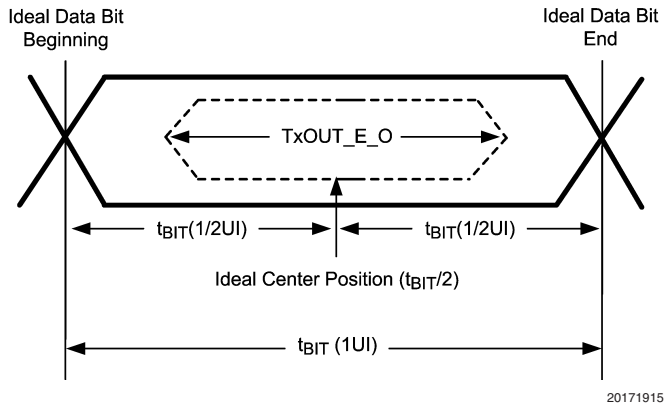
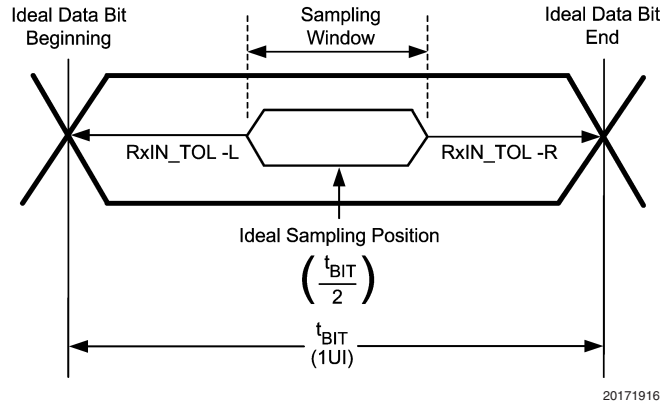


FIGURE 14. Transmitter Output Eye Opening (TxOUT_E_O)

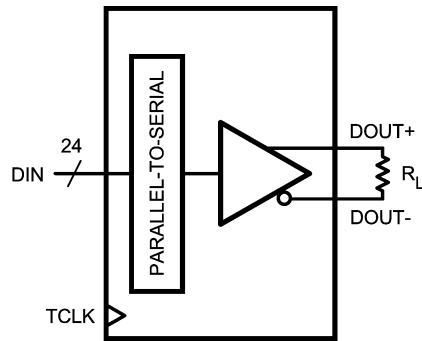
AC Timing Diagrams and Test Circuits (Continued)



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RxIN_TOL_L is the ideal noise margin on the left of the figure, with respect to ideal.
 RxIN_TOL_R is the ideal noise margin on the right of the above figure, with respect to ideal.

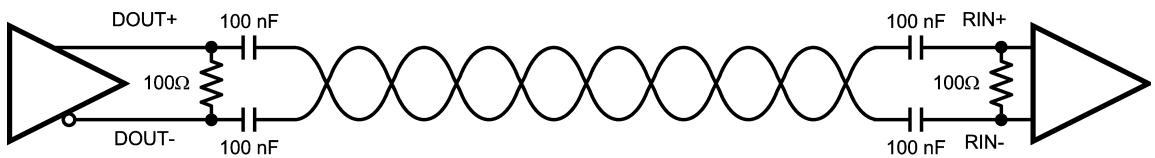
FIGURE 15. Receiver Input Tolerance (RxIN_TOL) and Sampling Window



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$VOD = (D_{OUT+}) - (D_{OUT-})$
 Differential output signal is shown as $(D_{OUT+}) - (D_{OUT-})$, device in Data Transfer mode.

FIGURE 16. Serializer VOD Diagram



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FIGURE 17. AC Coupled Application

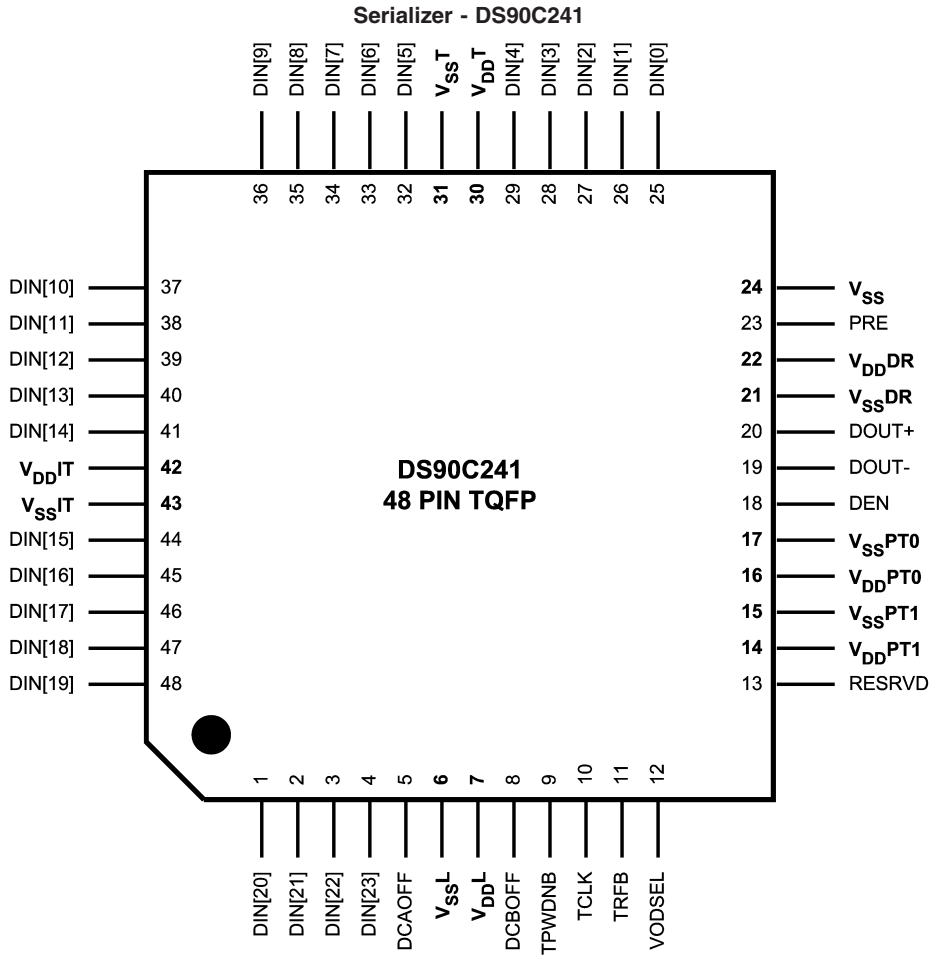
Pin Descriptions

Pin #	Pin Name	I/O	Description
DS90C241 SERIALIZER PIN DESCRIPTIONS			
22	VDDDR	VDD	Analog Voltage Supply, LVDS Output Power
21	VSSDR	GND	Analog Ground, LVDS Output Ground
16	VDDPT0	VDD	Analog Voltage supply, VCO Power
17	VSSPT0	GND	Analog Ground, VCO Ground
14	VDDPT1	VDD	Analog Voltage supply, PLL Power
15	VSSPT1	GND	Analog Ground, PLL Ground
30	VDDT	VDD	Digital Voltage supply, Tx Serializer Power
31	VSST	GND	Digital Ground, Tx Serializer Ground
7	VDDL	VDD	Digital Voltage supply, Tx Logic Power
6	VSSL	GND	Digital Ground, Tx Logic Ground
42	VDDIT	VDD	Digital Voltage supply, Tx Input Power
43	VSSIT	GND	Digital Ground, Tx Input Ground
24	VSS	GND	ESD Ground
4-1, 48-44, 41-32, 29-25	DIN[23:0]	CMOS_I	Transmitter D ata I nputs
10	TCLK	CMOS_I	T ransmitter reference C lock. Used to strobe data at the DIN inputs and to drive the transmitter PLL
9	TPWDNB	CMOS_I	T ransmitter P o W er D o W n B ar (ACTIVE L). TPWDNB = L; Disabled, DOUT (+/-) are TRI-STATED stand-by mode, PLL is shutdown TPWDNB = H; Enabled
18	DEN	CMOS_I	D ata E nable (ACTIVE H) DEN = L; Disabled, DOUT (+/-) are TRI-STATED, PLL still operational DEN = H; Enabled
13	RESRVD	CMOS_I	RESERVED - tie Low
23	PRE	CMOS_I	PRE -emphasis select pin. $PRE = (R_{PRE} \geq 3 \text{ k}\Omega); I_{max} = [(1.2/R)*20], R_{min} = 3 \text{ k}\Omega$ PRE = H or floating; pre-emphasis is disabled
11	TRFB	CMOS_I	T ransmitter R ising/ F alling B ar TRFB = H; DIN LVCMOS Input clocked on Rising TCLK TRFB = L; DIN LVCMOS Input clocked on Falling TCLK
12	VODSEL	CMOS_I	VOD level S ELect VODSEL = L; IOD $\approx 3.5 \text{ mA}$, (default). e.g. $3.5 \text{ mA} * 100\Omega = 350 \text{ mV}$ VODSEL = H; IOD $\approx 7.0 \text{ mA}$, VOD doubles approximately. e.g. $7 \text{ mA} * 100\Omega \approx 700 \text{ mV}$
5	DCAOFF	CMOS_I	RESERVED — tie Low
8	DCBOFF	CMOS_I	RESERVED — tie Low
20	DOUT+	LVDS_O	Transmitter LVDS true (+) O U T put
19	DOUT-	LVDS_O	Transmitter LVDS inverted (-) O U T put
DS90C124 DESERIALIZER PIN DESCRIPTIONS			
39	VDDIR	VDD	Analog LVDS Voltage supply, Power
40	VSSIR	GND	Analog LVDS Ground
47	VDDPR0	VDD	Analog Voltage supply, PLL Power
46	VSSPR0	GND	Analog Ground, PLL Ground
45	VDDPR1	VDD	Analog Voltage supply, PLL VCO Power
44	VSSPR1	GND	Analog Ground, PLL VCO Ground
37	VDDR1	VDD	Digital Voltage supply, Logic Power
38	VSSR1	GND	Digital Ground, Logic Ground
36	VDDR0	VDD	Digital Voltage supply, Logic Power

Pin Descriptions (Continued)

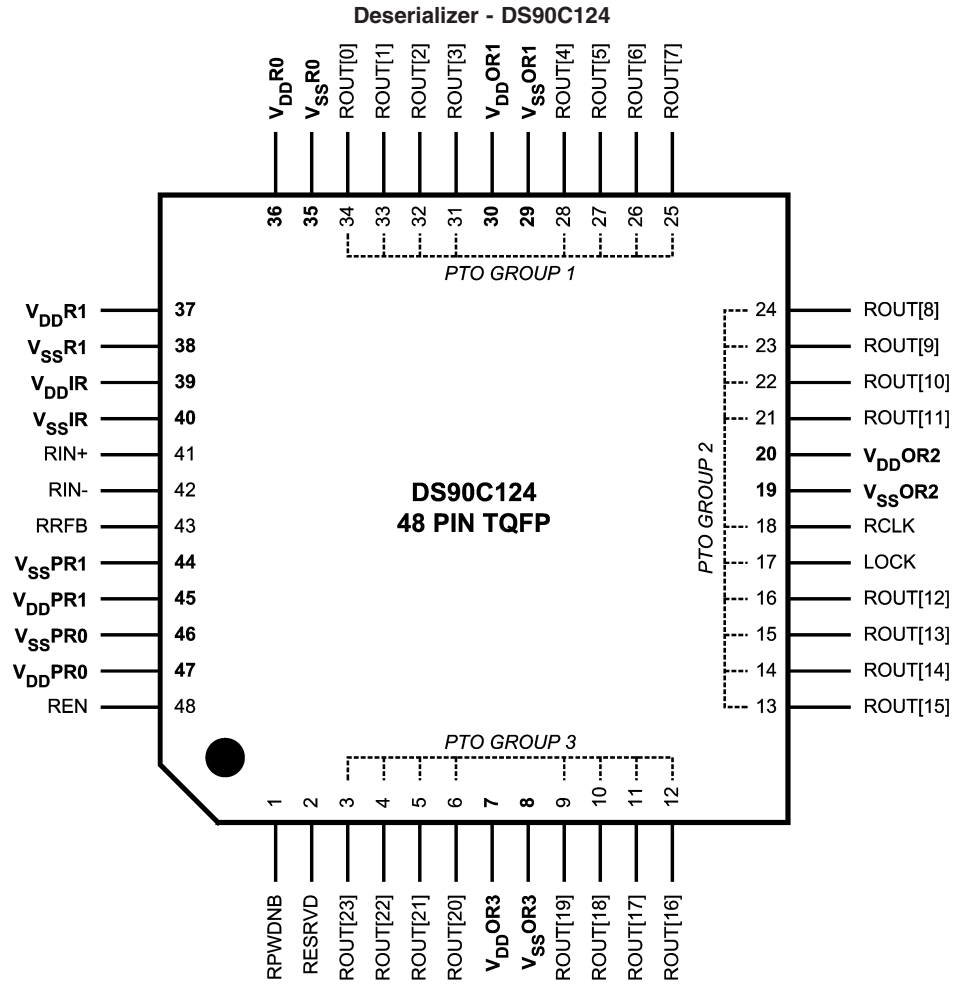
Pin #	Pin Name	I/O	Description
DS90C124 DESERIALIZER PIN DESCRIPTIONS			
35	VSSR0	GND	Digital Ground, Logic Ground
30	VDDOR1	VDD	Digital Voltage supply, LVCMOS Output Power
29	VSSOR1	GND	Digital Ground, LVCMOS Output Ground
20	VDDOR2	VDD	Digital Voltage supply, LVCMOS Output Power
19	VSSOR2	GND	Digital Ground, LVCMOS Output Ground
7	VDDOR3	VDD	Digital Voltage supply, LVCMOS Output Power
8	VSSOR3	GND	Digital Ground, LVCMOS Output Ground
41	RIN+	LVDS_I	Receiver LVDS true (+) Input
42	RIN-	LVDS_I	Receiver LVDS inverted (-) Input
2	RESRVD	CMOS_I	RESERVED - tie Low
43	RRFB	CMOS_I	Receiver Rising Falling Bar clock Edge Select RRFB = H; ROUT LVCMOS Output clocked on Rising RCLK RRFB = L; ROUT LVCMOS Output clocked on Falling RCLK
48	REN	CMOS_I	Receiver EN able, (ACTIVE H) REN = L; Disabled, ROUT[23-0] and RCLK TRI-STATED, PLL still operational REN = H; Enabled
1	RPWDB	CMOS_I	Receiver Po Wer Dow N B ar (ACTIVE L) RPWDB = L; Disabled, ROUT[23-0], RCLK, and LOCK are TRI-STATED in stand-by mode, PLL is shutdown RPWDB = H; Enabled
17	LOCK	CMOS_O	LOCK indicates the status of the receiver PLL LOCK = L; receiver PLL is unlocked, ROUT[23-0] and RCLK are TRI-STATED LOCK = H; receiver PLL is locked
25-28, 31-34	ROUT[7:0]	CMOS_O	Receiver LVCMOS level O utputs – Group 1
13-16, 21-24	ROUT[15:8]	CMOS_O	Receiver LVCMOS level O utputs – Group 2
3-6, 9-12	ROUT[23:16]	CMOS_O	Receiver LVCMOS level O utputs – Group 3
18	RCLK	CMOS_O	Recovered C lock. Parallel data rate clock recovered from the embedded clock. Used to strobe ROUT, LVCMOS level output.

Pin Diagrams



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Pin Diagrams (Continued)



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Functional Description

The DS90C241 Serializer and DS90C124 Deserializer chipset is an easy-to-use transmitter and receiver pair that sends 24-bits of parallel LVCMOS data over a single serial LVDS link from 120 Mbps to 840 Mbps throughput. The DS90C241 transforms a 24-bit wide parallel LVCMOS data into a single high speed LVDS serial data stream with embedded clock. The DS90C124 receives the LVDS serial data stream and converts it back into a 24-bit wide parallel data and recovered clock. The 24-bit Serializer/Deserializer chipset is designed to transmit data up to 10 meters over shielded twisted pair (STP) at clock speeds from 5 MHz to 35 MHz.

The Deserializer can attain lock to a data stream without the use of a separate reference clock source. The Deserializer synchronizes to the Serializer regardless of data pattern, delivering true automatic “plug and lock” performance. The Deserializer recovers the clock and data by extracting the embedded clock information and validating data integrity from the incoming data stream and then deserializes the data. The Deserializer monitors the incoming clock information, determines lock status, and asserts the LOCK output high when lock occurs. Each has a power down control to enable efficient operation in various applications.

INITIALIZATION AND LOCKING MECHANISM

Initialization of the DS90C241 and DS90C124 must be established before each device sends or receives data. Initialization refers to synchronizing the Serializer's and Deserializer's PLL's together. After the Serializers locks to the input clock source, the Deserializer synchronizes to the Serializers as the second and final initialization step.

Step 1: When V_{CC} is applied to both Serializer and/or Deserializer, the respective outputs are held in TRI-STATE and internal circuitry is disabled by on-chip power-on circuitry. When V_{CC} reaches $V_{CC\ OK}$ (2.2V) the PLL in Serializer begins locking to a clock input. For the Serializer, the local clock is the transmit clock, TCLK. The Serializer outputs are held in TRI-STATE while the PLL locks to the TCLK. After locking to TCLK, the Serializer block is now ready to send data patterns. The Deserializer output will remain in TRI-STATE while its PLL locks to the embedded clock information in serial data stream. Also, the Deserializer LOCK output will remain low until its PLL locks to incoming data and sync-pattern on the RIN_{\pm} pins.

Step 2: The Deserializer PLL acquires lock to a data stream without requiring the Serializer to send special patterns. The Serializer that is generating the stream to the Deserializer will automatically send random (non-repetitive) data patterns during this step of the Initialization State. The Deserializer will lock onto embedded clock within the specified amount of time. An embedded clock and data recovery (CDR) circuit locks to the incoming bit stream to recover the high-speed receive bit clock and re-time incoming data. The CDR circuit expects a coded input bit stream. In order for the Deserializer to lock to a random data stream from the Serializer, it performs a series of operations to identify the rising clock edge and validates data integrity, then locks to it. Because this locking procedure is independent on the data pattern, total random locking duration may vary. At the point when the Deserializer's CDR locks to the embedded clock, the LOCK pin goes high and valid RCLK/data appears on the outputs. Note that the LOCK signal is synchronous to valid data appearing on the outputs. The Deserializer's LOCK pin is a convenient way to ensure data integrity is achieved on receiver side.

DATA TRANSFER

After lock is established, the Serializer inputs DIN_0 – DIN_{23} may be used to input data to the Serializer. Data is clocked into the Serializer by the TCLK input. The edge of TCLK used to strobe the data is selectable via the TRFB pin. TRFB high selects the rising edge for clocking data and low selects the falling edge. The Serializer outputs ($DOUT_{\pm}$) are intended to drive point-to-point connections or limited multi-point applications.

CLK1, CLK0, DCA, DCB are four overhead bits transmitted along the single LVDS serial data stream. The CLK1 bit is always high and the CLK0 bit is always low. The CLK1 and CLK0 bits function as the embedded clock bits in the serial stream. DCB functions as the DC Balance control bit. It does not require any pre-coding of data on transmit side. The DC Balance bit is used to minimize the short and long-term DC bias on the signal lines. This bit operates by selectively sending the data either unmodified or inverted. The DCA bit is used to validate data integrity in the embedded data stream. Both DCA and DCB coding schemes are integrated and automatically performed within Serializer and Deserializer.

Serialized data and clock/control bits (24+4 bits) are transmitted from the serial data output ($DOUT_{\pm}$) at 28 times the TCLK frequency. For example, if TCLK is 35 MHz, the serial rate is $35 \times 28 = 980$ Mega bits per second. Since only 24 bits are from input data, the serial “payload” rate is 24 times the TCLK frequency. For instance, if $TCLK = 35$ MHz, the payload data rate is $35 \times 24 = 840$ Mbps. TCLK is provided by the data source and must be in the range of 5 MHz to 35 MHz nominal. The Serializer outputs ($DOUT_{\pm}$) can drive a point-to-point connection. The outputs transmit data when the enable pin (DEN) is high, TPWDBN is high. The DEN pin may be used to TRI-STATE the outputs when driven low.

When the Deserializer channel attains lock to the input from a Serializer, it drives its LOCK pin high and synchronously delivers valid data and recovered clock on the output. The Deserializer locks onto the embedded clock, uses it to generate multiple internal data strobes, and then drives the recovered clock to the RCLK pin. The recovered clock (RCLK output pin) is synchronous to the data on the $ROUT[23:0]$ pins. While LOCK is high, data on $ROUT[23:0]$ is valid. Otherwise, $ROUT[23:0]$ is invalid. The polarity of the RCLK edge is controlled by the RRFB input. $ROUT(0-23)$, LOCK and RCLK outputs will each drive a maximum of 8 pF load with a 35 MHz clock. REN controls TRI-STATE for $ROUT_n$ and the RCLK pin on the Deserializer.

RESYNCHRONIZATION

If the Deserializer loses lock, it will automatically try to re-establish lock. For example, if the embedded clock edge is not detected one time in succession, the PLL loses lock and the LOCK pin is driven low. The Deserializer then enters the operating mode where it tries to lock to a random data stream. It looks for the embedded clock edge, identifies it and then proceeds through the locking process. The logic state of the LOCK signal indicates whether the data on $ROUT$ is valid; when it is high, the data is valid. The system must monitor the LOCK pin to determine whether data on the $ROUT$ is valid.

POWERDOWN

The Powerdown state is a low power sleep mode that the Serializer and Deserializer may use to reduce power when no data is being transferred. The TPWDBN and RPWDBN are used to set each device into power down mode, which

Functional Description (Continued)

reduces supply current to the μA range. The Serializer enters powerdown when the TPWDNB pin is driven low. In powerdown, the PLL stops and the outputs go into TRI-STATE, disabling load current and reducing supply. To exit Powerdown, TPWDNB must be driven high. When the Serializer exits Powerdown, its PLL must lock to TCLK before it is ready for the Initialization state. The system must then allow time for Initialization before data transfer can begin. The Deserializer enters powerdown mode when RPWDNB is driven low. In powerdown mode, the PLL stops and the outputs enter TRI-STATE. To bring the Deserializer block out of the powerdown state, the system drives RPWDNB high.

Both the Serializer and Deserializer must reinitialize and relock before data can be transferred. The Deserializer will initialize and assert LOCK high until it is locked to the input clock.

TRI-STATE

For the Serializer, TRI-STATE is entered when the DEN or TPWDNB pin is driven low. This will TRI-STATE both driver output pins (DOUT+ and DOUT-). When DEN is driven high, the serializer will return to the previous state as long as all other control pins remain static (TPWDNB, TRFB).

When you drive the REN or RPWDNB pin low, the Deserializer enters TRI-STATE. Consequently, the receiver output pins (ROUT0–ROUT23) and RCLK will enter TRI-STATE. The LOCK output remains active, reflecting the state of the PLL. The Deserializer input pins are high impedance during receiver powerdown (RPWDNB low) and power-off ($V_{CC} = 0\text{V}$).

PRE-EMPHASIS

The DS90C241 features a Pre-Emphasis mode used to compensate for long or lossy transmission media. Cable drive is enhanced with a user selectable Pre-Emphasis feature that provides additional output current during transitions to counteract cable loading effects. The transmission distance will be limited by the loss characteristics and quality of the media. Pre-Emphasis adds extra current during LVDS logic transition to reduce the cable loading effects and increase driving distance. In addition, Pre-Emphasis helps provide faster transitions, increased eye openings, and improved signal integrity. To enable the Pre-Emphasis function, the "PRE" pin requires one external resistor (R_{pre}) to V_{SS} in order to set the additional current level. Pre-Emphasis strength is set via an external resistor (R_{pre}) applied from min to max (floating to $3\text{k}\Omega$) at the "PRE" pin. A lower input resistor value on the "PRE" pin increases the magnitude of dynamic current during data transition. There is an internal current source based on the following formula: $PRE = (R_{pre} \geq 3\text{k}\Omega); I_{MAX} = [(1.2/R_{pre}) \times 20]$. The ability of the DS90C241 to use the Pre-Emphasis feature will extend the transmission distance up to 10 meters in most cases.

AC-COUPLING AND TERMINATION

The DS90C241 and DS90C124 supports AC-coupled interconnects through integrated DC balanced encoding/decoding scheme. To use AC coupled connection between the Serializer and Deserializer, insert external AC coupling capacitors in series in the LVDS signal path as illustrated in *Figure 17*. The Deserializer input stage is designed for AC-coupling by providing a built-in AC bias network which sets the internal V_{CM} to $+1.2\text{V}$. With AC signal coupling, capacitors provide the ac-coupling path to the signal input.

For the high-speed LVDS transmissions, the smallest available package should be used for the AC coupling capacitor. This will help minimize degradation of signal quality due to package parasitics. The most common used capacitor value for the interface is 100 nF ($0.1\ \mu\text{F}$) capacitor.

A termination resistor across $\text{DOUT}\pm$ and $\text{RIN}\pm$ is also required for proper operation to be obtained. The termination resistor should be equal to the differential impedance of the media being driven. This should be in the range of 90 to 132 Ohms. 100 Ohms is a typical value common used with standard 100 Ohm transmission media. This resistor is required for control of reflections and also to complete the current loop. It should be placed as close to the Serializer $\text{DOUT}\pm$ outputs and Deserializer $\text{RIN}\pm$ inputs to minimize the stub length from the pins. To match with the differential impedance on the transmission line, the LVDS I/O are terminated with 100 ohm resistors on Serializer $\text{DOUT}\pm$ outputs pins and Deserializer $\text{RIN}\pm$ input pins.

PROGRESSIVE TURN-ON (PTO)

Deserializer $\text{ROUT}[23:0]$ outputs are grouped into three groups of eight, with each group switching about 0.5UI apart in phase to reduce EMI, simultaneous switching noise, and system ground bounce.

Applications Information

USING THE DS90C241 AND DS90C124

The DS90C241/DS90C124 Serializer/Deserializer (SER-DES) pair sends 24 bits of parallel LVCMOS data over a serial LVDS link up to 840 Mbps. Serialization of the input data is accomplished using an on-board PLL at the Serializer which embeds clock with the data. The Deserializer extracts the clock/control information from the incoming data stream and deserializes the data. The Deserializer monitors the incoming clock information to determine lock status and will indicate lock by asserting the LOCK output high.

POWER CONSIDERATIONS

An all CMOS design of the Serializer and Deserializer makes them inherently low power devices. Additionally, the constant current source nature of the LVDS outputs minimize the slope of the speed vs. I_{CC} curve of CMOS designs.

NOISE MARGIN

The Deserializer noise margin is the amount of input jitter (phase noise) that the Deserializer can tolerate and still reliably recover data. Various environmental and systematic factors include:

Serializer: TCLK jitter, V_{CC} noise (noise bandwidth and out-of-band noise)

Media: ISI, V_{CM} noise

Deserializer: V_{CC} noise

For a graphical representation of noise margin, please see *Figure 15*.

TRANSMISSION MEDIA

The Serializer and Deserializer can be used in point-to-point configuration, through a PCB trace, or through twisted pair cable. In a point-to-point configuration, the transmission media needs be terminated at both ends of the transmitter and receiver pair. Interconnect for LVDS typically has a differential impedance of 100 Ohms. Use cables and connectors that have matched differential impedance to minimize impedance discontinuities. In most applications that involve

Applications Information (Continued)

cables, the transmission distance will be determined on data rates involved, acceptable bit error rate and transmission medium.

HOT PLUG INSERTION

The Serializer and Deserializer devices support hot pluggable applications. The "Hot Inserted" operation on the serial interface does not disrupt communication data on the active data lines. The automatic receiver lock to random data "plug & go" hot insertion capability allows the DS90C124 to attain lock to the active data stream during a live insertion event.

PCB LAYOUT AND POWER SYSTEM CONSIDERATIONS

Circuit board layout and stack-up for the LVDS SERDES devices should be designed to provide low-noise power feed to the device. Good layout practice will also separate high frequency or high-level inputs and outputs to minimize unwanted stray noise pickup, feedback and interference. Power system performance may be greatly improved by using thin dielectrics (2 to 4 mils) for power / ground sandwiches. This arrangement provides plane capacitance for the PCB power system with low-inductance parasitics, which has proven especially effective at high frequencies, and makes the value and placement of external bypass capacitors less critical. External bypass capacitors should include both RF ceramic and tantalum electrolytic types. RF capacitors may use values in the range of 0.01 uF to 0.1 uF. Tantalum capacitors may be in the 2.2 uF to 10 uF range. Voltage rating of the tantalum capacitors should be at least 5X the power supply voltage being used.

Surface mount capacitors are recommended due to their smaller parasitics. When using multiple capacitors per supply pin, locate the smaller value closer to the pin. A large bulk capacitor is recommended at the point of power entry. This is typically in the 50uF to 100uF range and will smooth low frequency switching noise. It is recommended to connect power and ground pins directly to the power and ground planes with bypass capacitors connected to the plane with via on both ends of the capacitor. Connecting power or ground pins to an external bypass capacitor will increase the inductance of the path.

A small body size X7R chip capacitor, such as 0603, is recommended for external bypass. Its small body size reduces the parasitic inductance of the capacitor. The user must pay attention to the resonance frequency of these external bypass capacitors, usually in the range of 20-30 MHz range. To provide effective bypassing, multiple capaci-

tors are often used to achieve low impedance between the supply rails over the frequency of interest. At high frequency, it is also a common practice to use two vias from power and ground pins to the planes, reducing the impedance at high frequency.

Some devices provide separate power and ground pins for different portions of the circuit. This is done to isolate switching noise effects between different sections of the circuit. Separate planes on the PCB are typically not required. Pin Description tables typically provide guidance on which circuit blocks are connected to which power pin pairs. In some cases, an external filter may be used to provide clean power to sensitive circuits such as PLLs.

Use at least a four layer board with a power and ground plane. Locate LVCMOS (LVTTTL) signals away from the LVDS lines to prevent coupling from the CMOS lines to the LVDS lines. Closely-coupled differential lines of 100 Ohms are typically recommended for LVDS interconnect. The closely coupled lines help to ensure that coupled noise will appear as common-mode and thus is rejected by the receivers. The tightly coupled lines will also radiate less.

Termination of the LVDS interconnect is required. For point-to-point applications, termination should be located at both ends of the devices. Nominal value is 100 Ohms to match the line's differential impedance. Place the resistor as close to the transmitter DOUT± outputs and receiver RIN± inputs as possible to minimize the resulting stub between the termination resistor and device.

LVDS INTERCONNECT GUIDELINES

See AN-1108 and AN-905 for full details.

- Use 100Ω coupled differential pairs
- Use the S/2S/3S rule in spacings
 - S = space between the pair
 - 2S = space between pairs
 - 3S = space to LVCMOS/LVTTTL signal
- Minimize the number of VIA
- Use differential connectors when operating above 500Mbps line speed
- Maintain balance of the traces
- Minimize skew within the pair
- Terminate as close to the TX outputs and RX inputs as possible

Additional general guidance can be found in the LVDS Owner's Manual - available in PDF format from the National web site at: www.national.com/lvds

Truth Tables

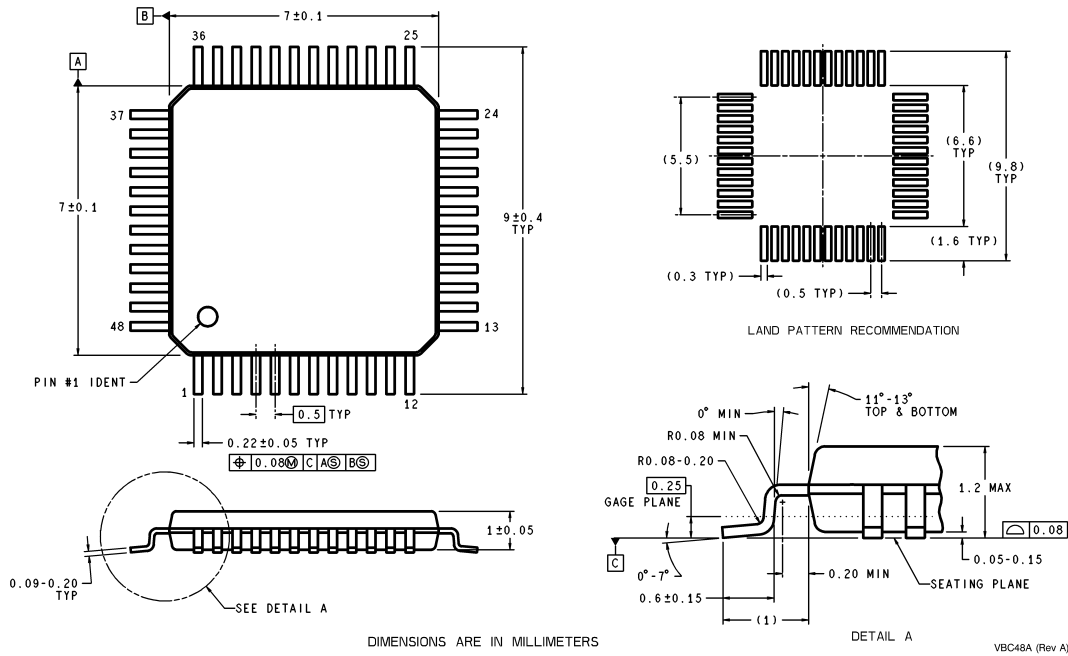
TABLE 1. DS90C241 Serializer Truth Table

Pin 9	Pin 18	(Internal)	Pins 19 and 20
TPWDNB	DEN	Tx PLL Status	LVDS Outputs
L	X	X	Hi Z
H	L	X	Hi Z
H	H	Not Locked	Hi Z
H	H	Locked	Serialized Data with Embedded Clock

TABLE 2. DS90C124 Deserializer Truth Table

Pin 1	Pin 48	(Internal)	(See Pin Diagram)	Pin 17
RPWDNB	REN	Rx PLL Status	ROUTn and RCLK	LOCK
L	X	X	Hi Z	Hi Z
H	L	X	Hi Z	L = PLL Unocked; H = PLL Locked
H	H	Not Locked	Hi Z	L
H	H	Locked	Data and RCLK Active	H

Physical Dimensions inches (millimeters) unless otherwise noted



Dimensions show in millimeters only
NS Package Number IVS48

Ordering Information

NSID	Package Type	Package ID
DS90C241	48 Lead TQFP style, 7.0 X 7.0 X 1.0 mm, 0.5 mm pitch, DS90C241IVS	IVS48
DS90C241	48 Lead TQFP style, 7.0 X 7.0 X 1.0 mm, 0.5 mm pitch, 1000 std reel, DS90C241IVSX	IVS48
DS90C124	48 Lead TQFP style, 7.0 X 7.0 X 1.0 mm, 0.5 mm pitch, DS90C124IVS	IVS48
DS90C124	48 Lead TQFP style, 7.0 X 7.0 X 1.0 mm, 0.5 mm pitch, 1000 std reel, DS90C124IVSX	IVS48

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