

1.5MHz 1A, Synchronous Step-Down Regulator

General Description

EML3493 is a high efficiency step down DC/DC converter. It features an extremely low quiescent current, which is suitable for reducing standby power consumption, especially for portable applications.

The device can accept input voltage from 2.5V to 5.5V and deliver up to 1A output current. High 1.5MHz switching frequency allows the use of small surface mount inductors and capacitors to reduce overall PCB board space. Furthermore, the built-in synchronous switch improves efficiency and eliminates external Schottky diode. EML3493 uses different modulation algorithms for various loading conditions: (1) Pulse Width Modulation (PWM) for low output voltage ripple and fixed frequency noise, (2) Pulse Frequency Modulation (PFM) for improving light load efficiency, and (3) Low Dropout (LDO) Mode for providing 100% duty cycle operation during heavy loading. Adopting low reference voltage design reduces regulated output to 0.6V. The adjustable version of this device is available in SOT-23-5 package.

Features

- Achieve 97% efficiency
- Input voltage : 2.5V to 5.5V
- Output current up to 1A
- Reference voltage: 0.6V
- Quiescent current 15 μ A with no load
- Internal switching frequency: 1.5MHz
- No Schottky diode needed
- Low dropout operation: 100% duty cycle
- Shutdown current < 1 μ A
- Excellent line and load transient response
- Over-temperature protection

Applications

- Blue-Tooth devices
- Cellular and Smart Phones
- Personal Multi-media Player (PMP)
- Wireless networking
- Digital Still Cameras
- Portable applications

Typical Application

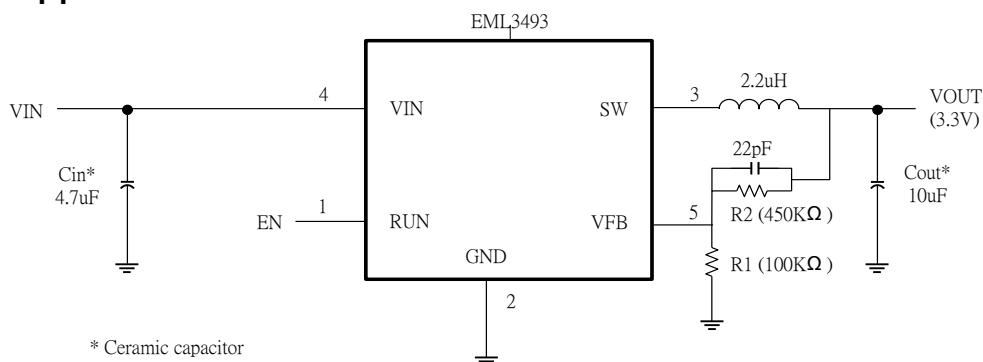
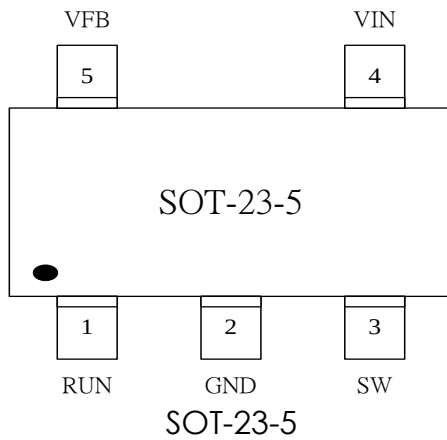


Fig. 1

Package Configuration



EML3493-00VN05NRR

00 Adjustable
VN05 SOT-23-5 Package
NRR RoHS & Halogen free package
Commercial Grade Temperature
Rating: -40 to 85°C
Package in Tape & Reel

Order, Mark & Packing information

Package	Vout(V)	Product ID	Marking	Packing
SOT-23-5	adjustable	EML3493-00VN05NRR		Tape & Reel 3K units

Pin Functions

Pin Name	SOT-23-5	Function
RUN	1	Enable Pin. Minimum 1.2V to enable the device. Maximum 0.4V to shut down the device.
VIN	4	Power Input Pin. Must be closely decoupled to GND pin with a 4.7μF or greater ceramic capacitor.
SW	3	Switch Pin. Must be connected to Inductor. This pin connects to the drains of the internal main and synchronous power MOSFET switches.
GND	2	Ground Pin.
VFB (Adjustable)	5	Feedback Pin. Receives the feedback voltage from an external resistive divider across the output.

Absolute Maximum Ratings

Devices are subjected to fail if they stay above absolute maximum ratings.

Input Voltage ----- -0.3V to 6V
 RUN, VFB Voltages ----- -0.3V to V_{IN}
 SW(DC) Voltage ----- -0.3V to ($V_{IN} + 0.3V$)
 SW(AC, less than 20ns) Voltage ----- -3V to 7.5V
 Lead Temperature (Soldering, 10 sec)----- 260°C

Operating Temperature Range ----- -40°C to 85°C
 Junction Temperature (Notes 1, 2) ----- 150°C
 Storage Temperature Range ----- -65°C to 150°C
 ESD Susceptibility HBM ----- 2KV
 CDM ----- 750V

Recommended Operating Conditions

Input Voltage (V_{IN}) ----- +2.5V to +5.5V

Junction Operating Temperature ----- -40°C to 125°C

Thermal data

Package	Thermal resistance	Parameter	Value
SOT-23-5	θ_{JA} (Note 3)	Junction-ambient	134.5°C/W
	θ_{JC} (Note 4)	Junction-case	81°C/W

Electrical Characteristics

$V_{IN} = 3.6V$, $T_A = 25^\circ C$, unless otherwise specified.

Symbol	Parameter	Conditions	Min	Typ	Max	Units
V_{IN}	Input Voltage Range		2.5		5.5	V
I_{VFB}	Feedback Current				±100	nA
V_{FB}	Regulated Feedback Voltage		0.588	0.600	0.612	V
$V_{OUT} \%$	Output Voltage Accuracy	$I_{OUT} = 100mA$	-3		+3	%
ΔV_{FB}	Reference Voltage Line Regulation	$V_{IN} = 2.5V$ to $5.5V$			0.4	%/V
I_{PK}	Peak Inductor Current (Note 5)	$V_{FB} = 0.5V$ or $V_{OUT} = 90\%$	1.5	2.3		A
I_Q	PWM Quiescent Current	$V_{FB} = 0.5V$ or $V_{OUT} = 90\%$		130		μA
	PFM Quiescent Current	$V_{FB} = 0.65V$ or $V_{OUT} = 108\%$		15		μA
	Shutdown	$V_{RUN} = 0V$, $V_{IN} = 4.2V$		0.1	1	μA
f_{OSC}	Oscillator Frequency	$V_{FB} = 0.6V$ or $V_{OUT} = 100\%$	1.2	1.5	1.8	MHz
	Short-Circuit Oscillator Frequency	$V_{FB} = 0V$ or $V_{OUT} = 0V$		750		kHz
R_{PFET}	$R_{DS(ON)}$ of PMOS	$I_{SW} = 100mA$		0.25		Ω
R_{NFET}	$R_{DS(ON)}$ of NMOS	$I_{SW} = -100mA$		0.18		Ω
V_{UVLO}	VIN UVLO Threshold			1.8		V
	VIN UVLO Hysteresis			250		mV
I_{LSW}	SW Leakage	$V_{RUN} = 0V$, $V_{SW} = 0V$ or $5V$, $V_{IN} = 5V$			±1	μA
V_{RUN}	Enable Threshold		1.2			V
	Shutdown Threshold				0.4	V
I_{RUN}	RUN Leakage Current				±1	μA
T_{SD}	Thermal Shutdown			170		°C
	Thermal Shutdown Hysteresis			30		°C

Note 1: T_J is a function of the ambient temperature T_A and power dissipation P_D ($T_J = T_A + (P_D) * (134.5^\circ C/W)$).

Note 2: This IC has a built-in over-temperature protection to avoid damage from overloaded conditions.

Note 3: θ_{JA} is measured in the natural convection at $T_A = 25^\circ C$ on a highly effective thermal conductivity test board (2 layers, 250P) according to the JEDEC 51-7 thermal measurement standard.

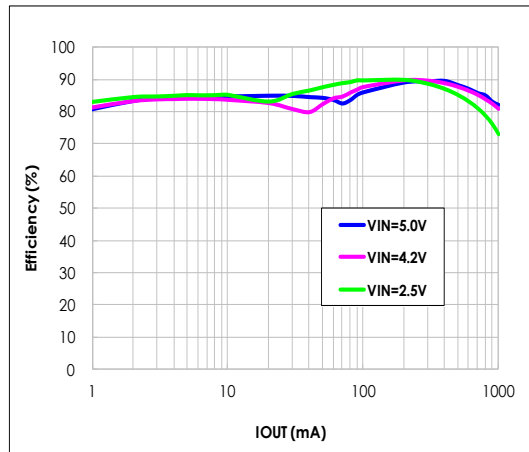
Note 4: θ_{JC} represents the heat resistance between the chip and the package top case.

Note 5: Design guaranteed.

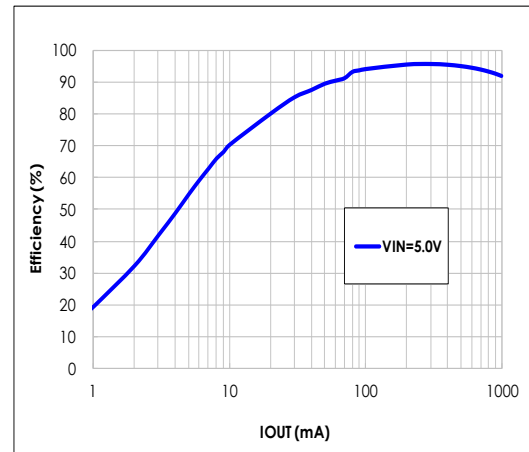
Typical Performance Characteristics

$V_{IN}=3.6V$, $V_{OUT}=1.2V$, $C_{IN}=4.7\mu F$, $C_{OUT}=10\mu F$, $L=2.2\mu H$, $T_A=25^\circ C$, unless otherwise specified

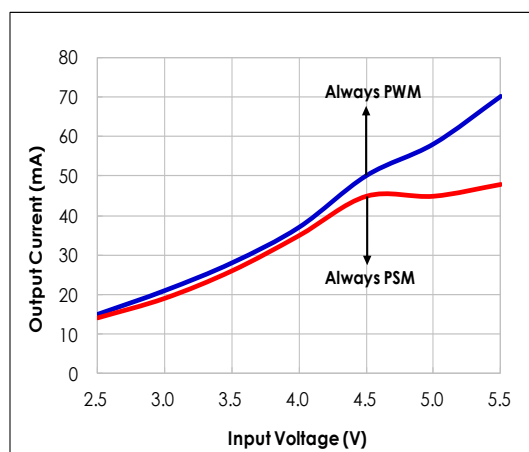
Efficiency vs. Output Current ($V_{OUT}=1.2V$)



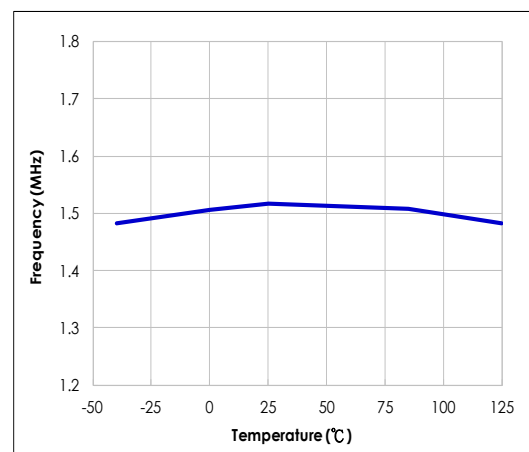
Efficiency vs. Output Current ($V_{OUT}=3.3V$)



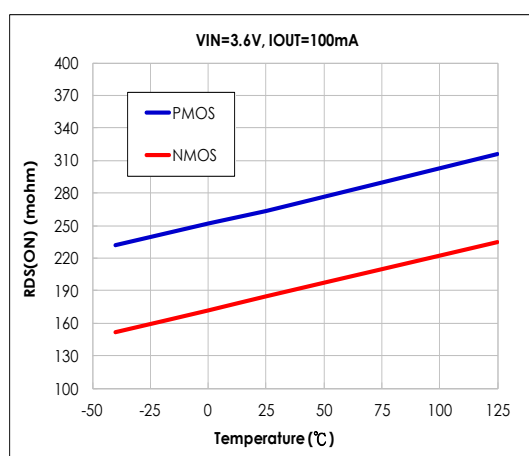
PSM/PWM Boundaries ($V_{OUT}=1.2V$)



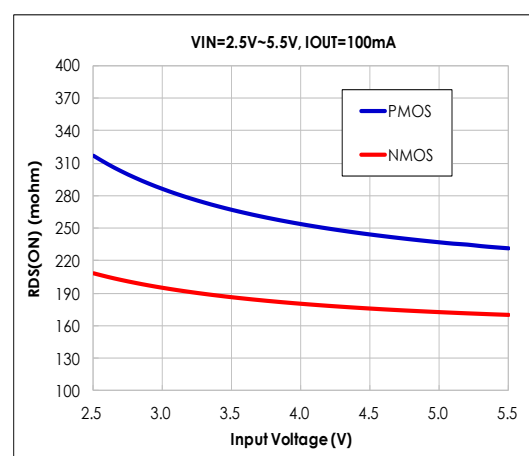
Oscillator Frequency vs. Temperature



$R_{DS(ON)}$ vs. Temperature

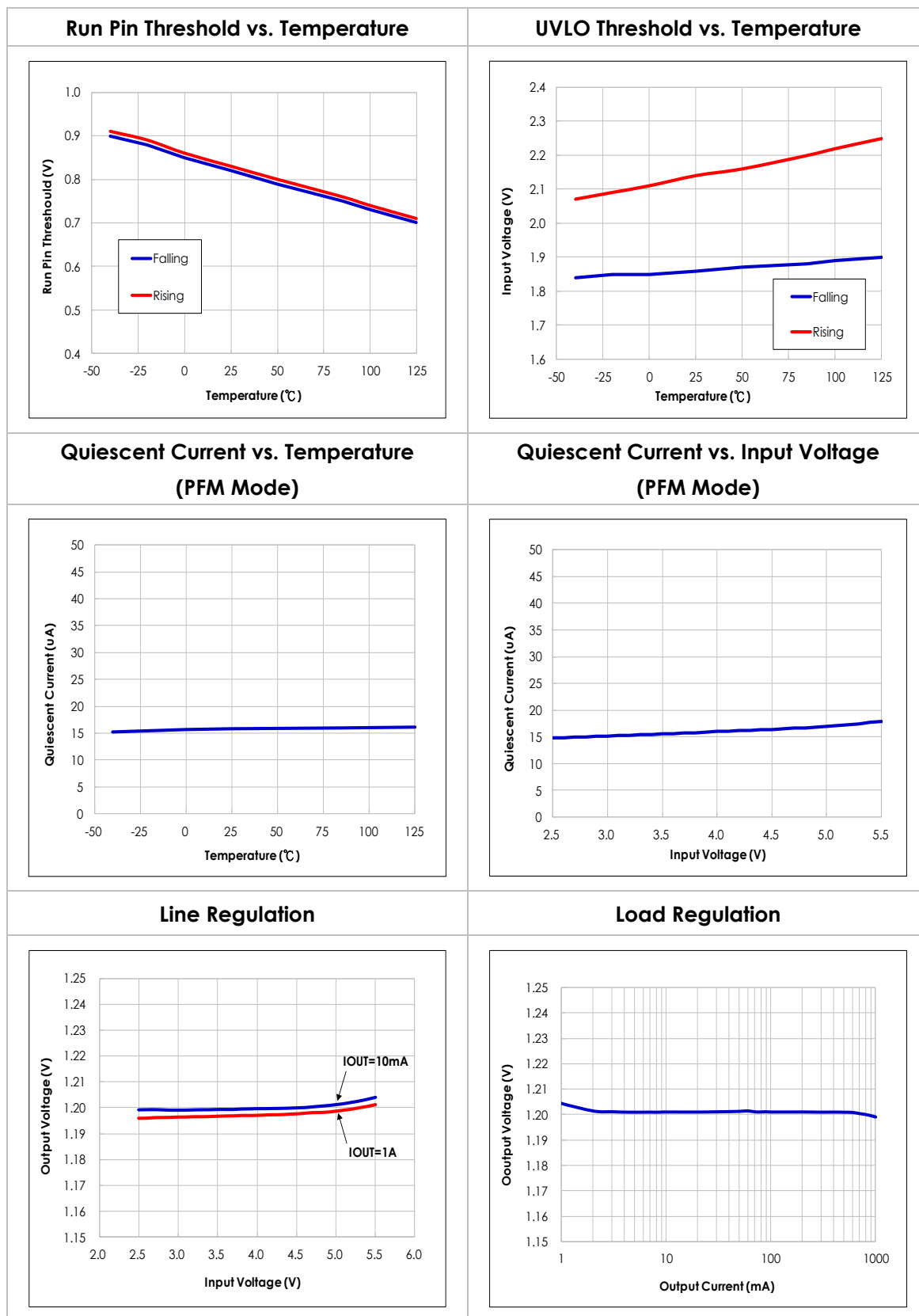


$R_{DS(ON)}$ vs. Input Voltage



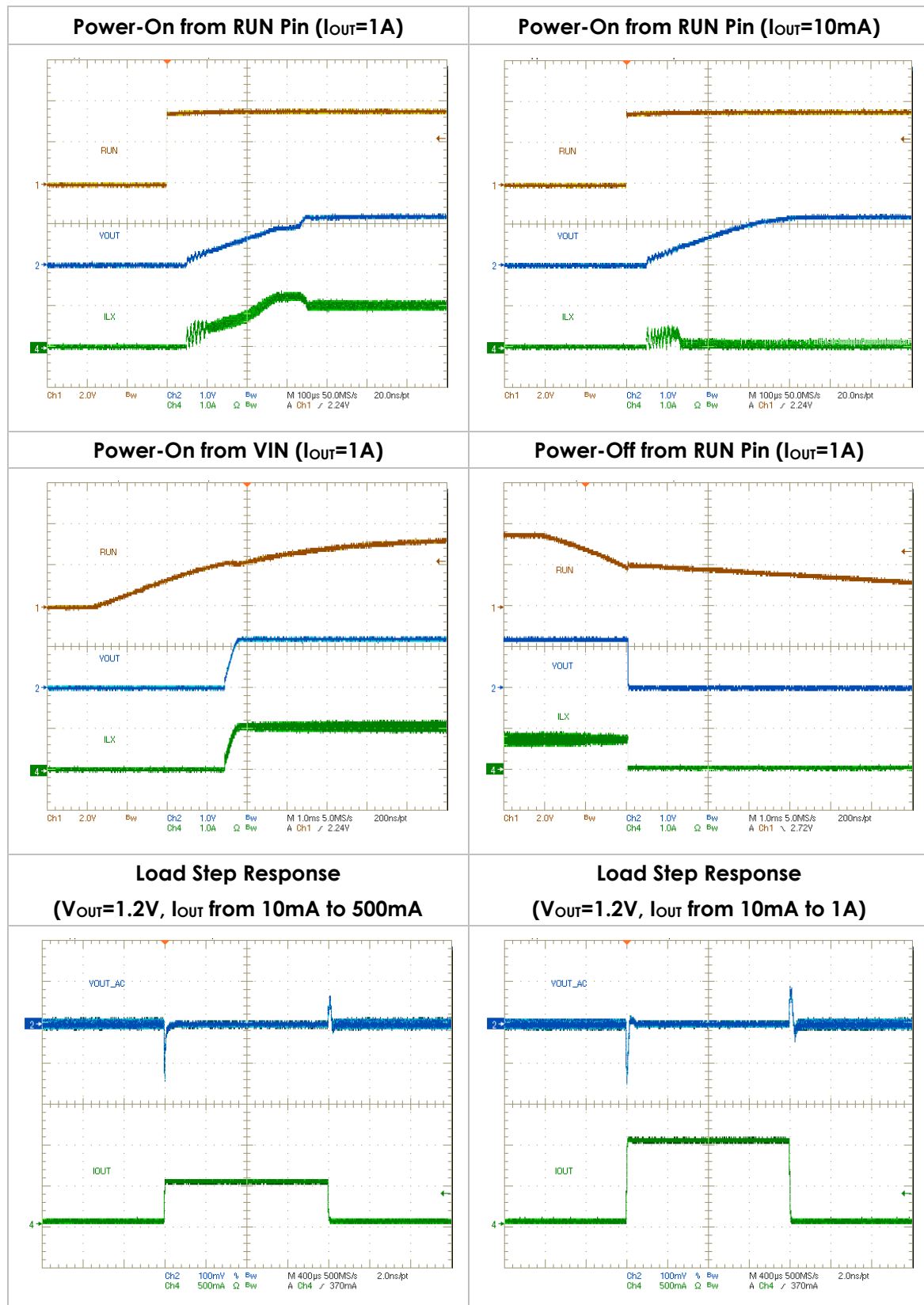
Typical Performance Characteristics (cont.)

$V_{IN}=3.6V$, $V_{OUT}=1.2V$, $C_{IN}=4.7\mu F$, $C_{OUT}=10\mu F$, $L=2.2\mu H$, $T_A=25^\circ C$, unless otherwise specified



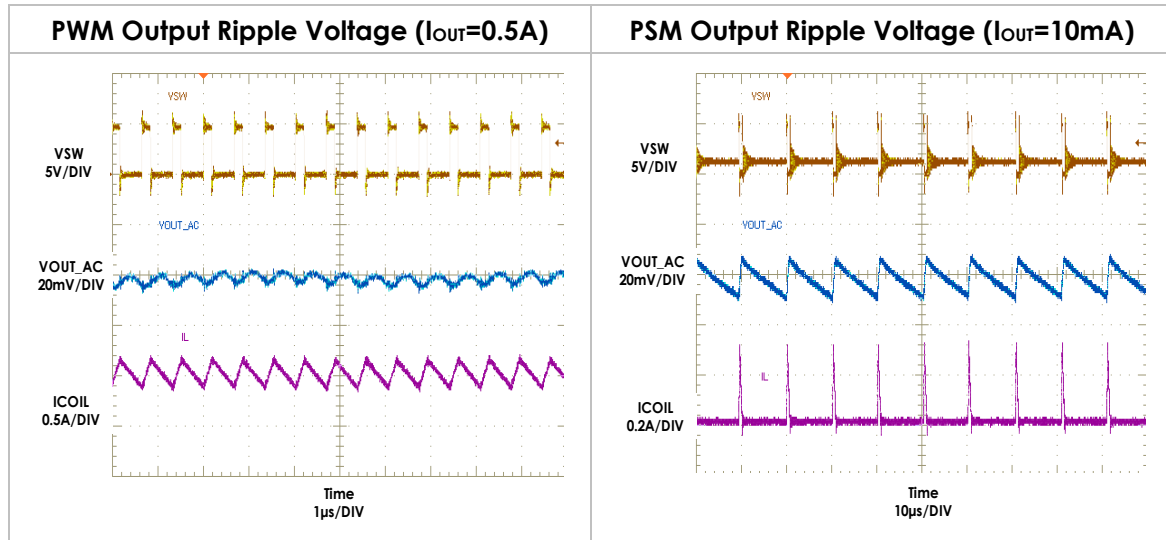
Typical Performance Characteristics (cont.)

$V_{IN}=3.6V$, $V_{OUT}=1.2V$, $C_{IN}=4.7\mu F$, $C_{OUT}=10\mu F$, $L=2.2\mu H$, $T_A=25^\circ C$, unless otherwise specified

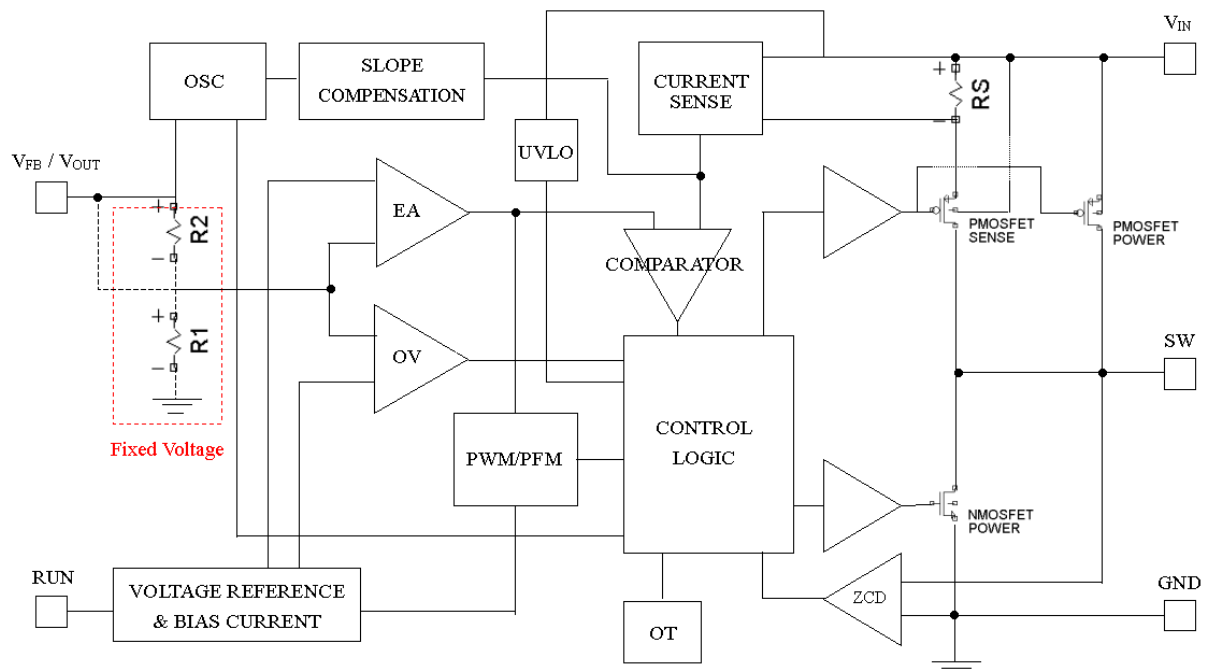


Typical Performance Characteristics (cont.)

$V_{IN}=3.6V$, $V_{OUT}=1.2V$, $C_{IN}=4.7\mu F$, $C_{OUT}=10\mu F$, $L=2.2\mu H$, $T_A=25^\circ C$, unless otherwise specified



Functional Block Diagram



Applications

The typical application circuit of adjustable version is shown in Fig.1.

Inductor Selection

Inductor ripple current and core saturation current are the two main factors that decide the Inductor value. A low DCR inductor is preferred.

C_{IN} and C_{OUT} Selection

A low ESR input capacitor can prevent large voltage transients at V_{IN}. The RMS current of input capacitor is required larger than I_{RMS} calculated by:

$$I_{RMS} \cong I_{OMAX} \frac{\sqrt{V_{OUT}(V_{IN} - V_{OUT})}}{V_{IN}} \quad \text{Eq. 1}$$

ESR is an important parameter to select C_{OUT}, which can be seen in the following output ripple V_{OUT} equation:

$$\Delta V_{OUT} \cong \Delta I_L \left(ESR + \frac{1}{8 \cdot f \cdot C_{OUT}} \right) \quad \text{Eq. 2}$$

Cheaper and smaller ceramic capacitors with higher capacitance values are now commercially available. These ceramic capacitors have low ripple currents, high voltage ratings and low ESR which make them suitable for switching regulator applications. It is feasible to optimize very low output ripples by Cout since Cout does not affect the internal control loop stability. X5R or X7R types are recommended since they have the best temperature and voltage characteristics of all ceramics capacitors.

Output Voltage (EML3493 adjustable)

In the adjustable version, the output voltage can be determined by:

$$V_{OUT} = 0.6V \left(1 + \frac{R_2}{R_1} \right) \quad \text{Eq. 3}$$

Thermal Considerations

Although the thermal shutdown circuit is designed in EML3493 to protect the device from thermal damage, the total power dissipation that EML3493 can sustain depends on the thermal capability of the package. The formula to ensure the safe operation is shown in note 1 on page 5.

To avoid the EML3493 from exceeding the maximum junction temperature, the user should perform some thermal analysis during PCB design.

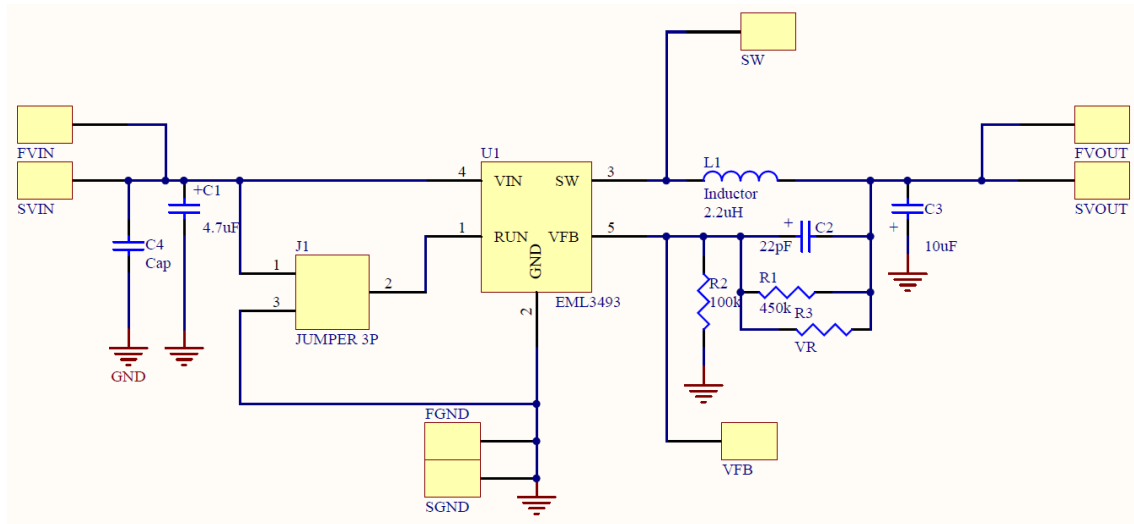
Guidelines for PCB Layout

To ensure proper operation of the EML3493, please note the following PCB layout guidelines:

1. The GND, SW and the VIN trace should be kept short, direct and wide.
2. VFB pin must be connected directly to the feedback resistors. Resistive divider R1/R1 must be connected parallel to the output capacitor C_{OUT}.
3. The Input capacitor C_{IN} must be connected to the pin VIN as close as possible.
4. Keep SW node away from the sensitive VFB node since this node has high frequency and voltage swing.
5. Keep the (–) plates of C_{IN} and C_{OUT} as close as possible.

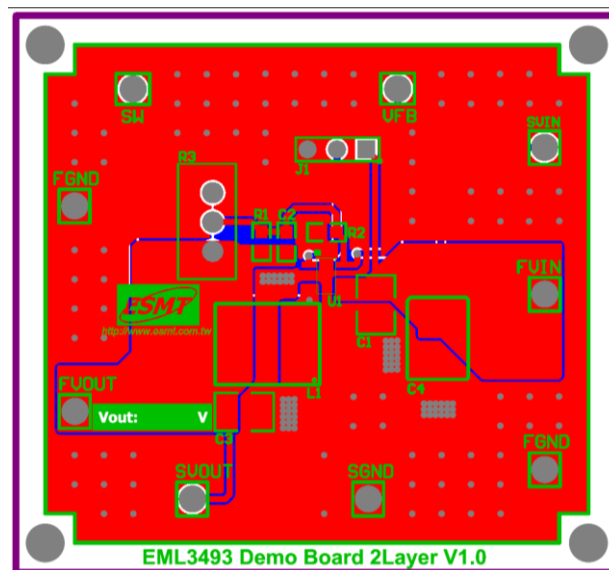
Applications

Typical schematic for PCB layout

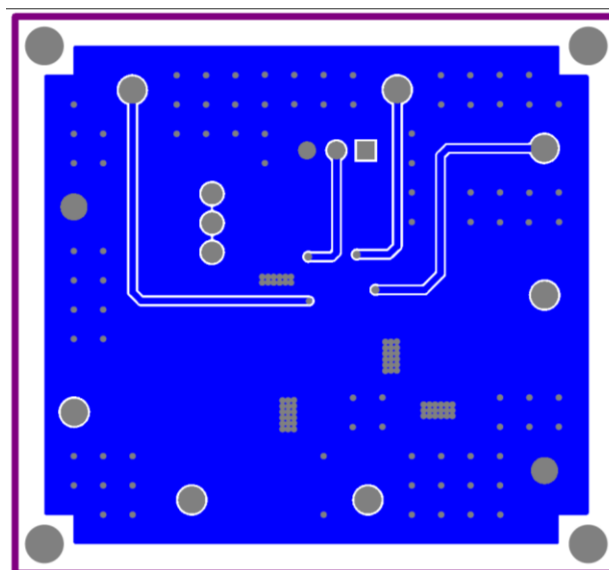


Note.

R3 and C4 are reserved locations for testing purposes. They are removed during normal applications.

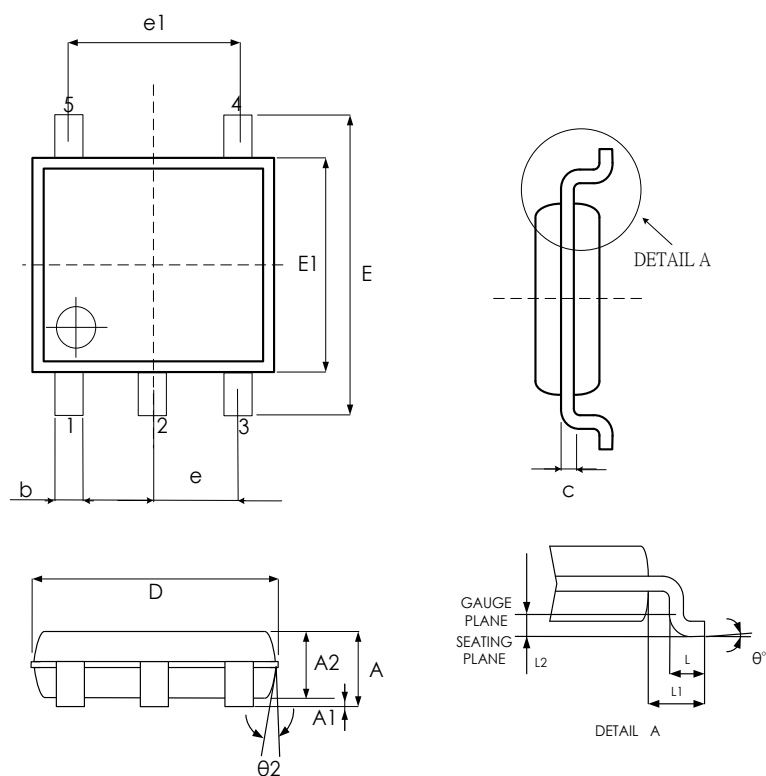
Typical schematic for PCB layout (cont.)

Top Layer



Bottom Layer

Package Outline Drawing SOT-23-5



SYMBPLS	MIN.	NOM.	MAX.
A	1.05	1.20	1.35
A1	0.05	0.10	0.15
A2	1.00	1.10	1.20
B	0.30	—	0.50
C	0.08	—	0.20
D	2.80	2.90	3.00
E	2.60	2.80	3.00
E1	1.50	1.60	1.70
E	0.95 BSC		
e1	1.90 BSC		
L	0.30	0.45	0.55
L1	0.60 REF		
θ°	0	5	10
$\theta2^\circ$	6	8	10

UNIT: mm

Revision History

Revision	Date	Description
0.1	2023.03.21	Initial version.
1.0	2023.06.19	Remove preliminary

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