

## 2 Channel Synchronous Buck PWM Controller



### General Description

The FP5148 is a dual channel synchronous buck switching controller with adjustable over-current protection for DC-DC applications. The FP5148 includes two stage 180° out of phase for driving NMOS, power down under-voltage lockout circuit (UVLO) , the inverting input of amplifier connects to a 0.8V precision reference regulator , short circuit and over current shutdown protection circuit (SCP / OCP)and programmable soft start function.

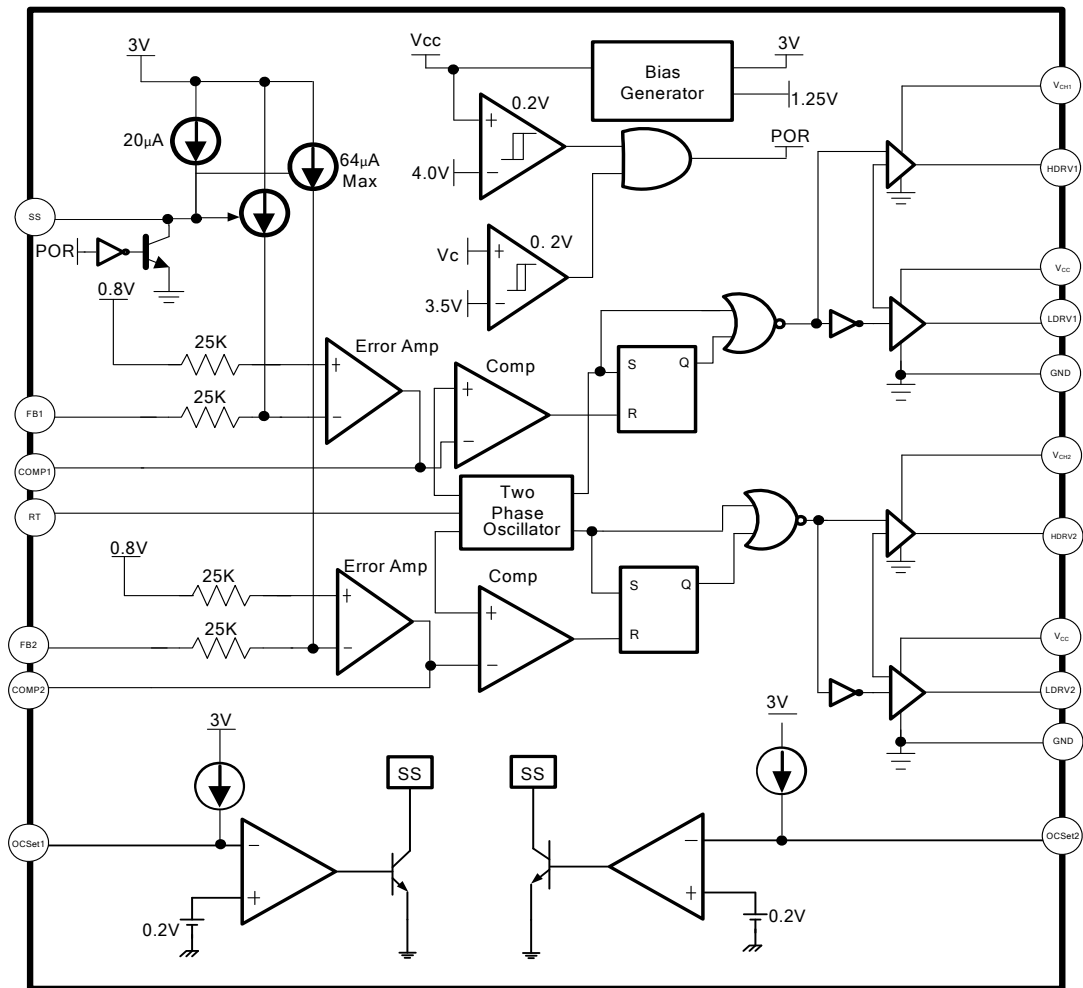
### Features

- Wide Supply Voltage Operating Range: 4.2~15V
- Precision Reference Voltage: 0.8V  $\pm$ 2%
- Peak Output Source / Drive Capability: 500mA
- Totem-pole Output for MOS Driving
- Programming Oscillation Frequency
- Programmable Soft Start Function (SS)
- OC/SC Protection Function (OCP / SCP)
- UVLO Protection Function
- Package: SOP-16L, SOP-16L (EP)

### Applications

- VGA Card
- Battery Charger
- Hard Disk Drive
- Multi-Output Application

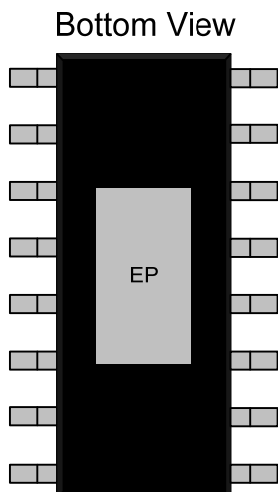
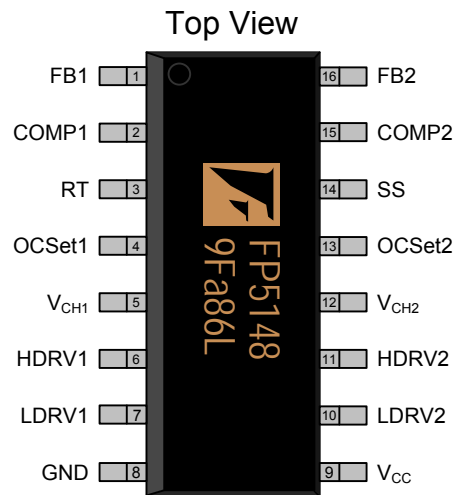
### Function Block Diagram



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## Pin Descriptions

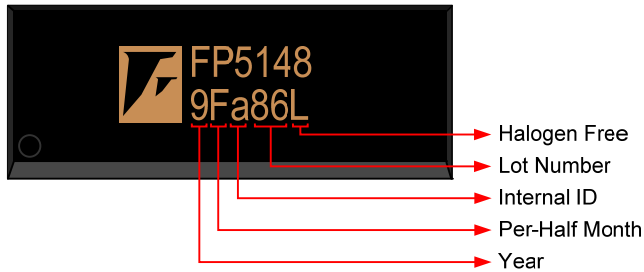
### SOP-16L & SOP-16L (EP)



Name	No.	I / O	Description
FB1	1	I	Error Amplifier 1 Inverting Input
COMP1	2	O	Error Amplifier 1 Compensation Output
RT	3	O	Oscillator Resistor
OCSet1	4	I	Over-current Protection 1
V <sub>CH1</sub>	5	P	Supply Voltage for Output Driver 1
HDRV1	6	O	Output Driver 1 for High Side NMOS
LDRV1	7	O	Output Driver 1 for Low Side NMOS
GND	8	P	IC Ground
V <sub>CC</sub>	9	P	IC Power Supply
LDRV2	10	O	Output Driver 2 for Low Side NMOS
HDRV2	11	O	Output driver 2 for High Side NMOS
V <sub>CH2</sub>	12	P	Supply Voltage for Output Driver 2
OCSet2	13	I	Over-current Protection 2
SS	14	I	To Connect a Capacitor for Soft-start
COMP2	15	O	Error Amplifier 2 Compensation Output
FB2	16	I	Error Amplifier 2 Inverting Input
EP	17	P	Exposed PAD - must connect to Ground

## Marking Information

### SOP-16L & SOP-16L (EP)



**Halogen Free:** Halogen free product indicator

**Lot Number:** Wafer lot number's last two digits

For Example: 132386TB → 86

**Internal ID:** Internal Identification Code

**Per-Half Month:** Production period indicated in half month time unit

For Example: January → A (Front Half Month), B (Last Half Month)

February → C (Front Half Month), D (Last Half Month)

**Year:** Production year's last digit

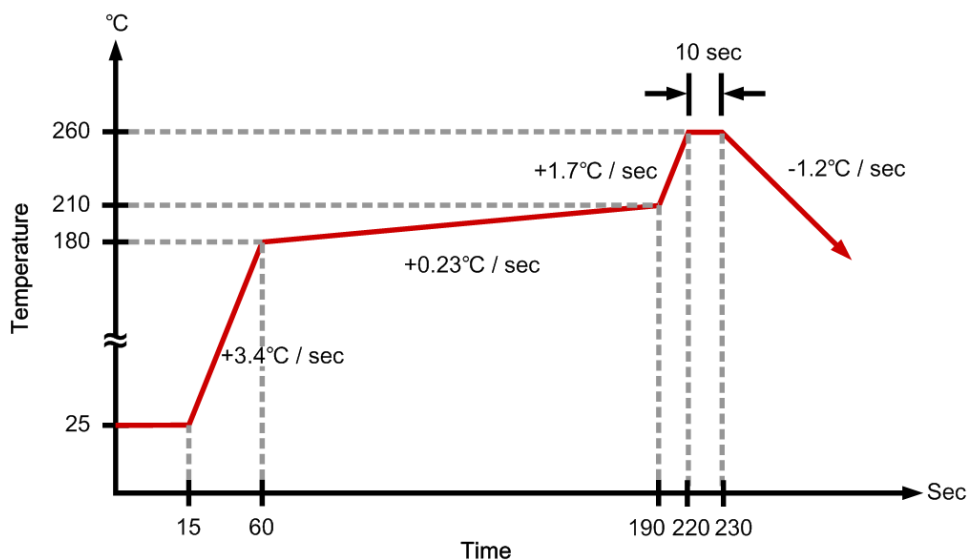
## Ordering Information

Part Number	Operating Temperature	Package	MOQ	Description
FP5148DR-LF	-10°C ~ +85°C	SOP-16L	2500EA	Tape & Reel
FP5148XR-LF	-10°C ~ +85°C	SOP-16L (EP)	2500EA	Tape & Reel

## Absolute Maximum Ratings

Parameter	Symbol	Conditions	Min.	Typ.	Max.	Unit
Power Supply Voltage	$V_{IN}$				15	V
Power Supply Voltage for High Side Driver		$V_{CH1}, V_{CH2}$			25	V
Output Source Current					-500	mA
Output Sink Current					500	mA
Allowable Power Dissipation		SOP-16L, $T_A \leq +25^\circ\text{C}$			830	mW
		SOP-16L (EP), $T_A \leq +25^\circ\text{C}$			650	mW
Operating Temperature			-10		+85	°C
Storage Temperature			-55		+125	°C
Operating Junction Temperature Range			+0		+125	°C
Lead Temperature SOP-16L		SOP-16L, soldering, 10 sec			+260	°C
Lead Temperature SOP-16L (EP)		SOP-16L (EP), soldering, 10 sec			+260	°C

## Suggested IR Re-flow Soldering Curve



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## Recommended Operating Conditions

Parameter	Symbol	Conditions	Min.	Typ.	Max.	Unit
Supply Voltage			4.2		15	V
Operating Temperature			-10		+85	°C

## DC Electrical Characteristics ( $V_{CC}=5V$ , $V_{CH1}=V_{CH2}=12V$ , unless otherwise noted)

Parameter	Symbol	Test Conditions	Min.	Typ.	Max.	Unit
<b>Under Voltage Lock-Out Section (UVLO)</b>						
UVLO Threshold Voltage	$V_{CC(UVLO)}$	Sweep up $V_{CC}$	4.0	4.2	4.4	V
Hysteresis Voltage	$V_{CC(HYS)}$			0.25		V
UVLO Threshold Voltage	$V_{CH1}, V_{CH2(UVLO)}$	Sweep up $V_{CH1}, V_{CH2}$	3.1	3.3	3.5	V
Hysteresis Voltage	$V_{CH1}, V_{CH2(HYS)}$			0.25		V
UVLO Threshold Voltage	$V_{FB1}, V_{FB2(UVLO)}$	Sweep down $V_{FB1}, V_{FB2}$	0.3	0.4	0.5	V
Hysteresis Voltage	$V_{FB1}, V_{FB2(HYS)}$			0.1		V
<b>Soft Start Section (SS)</b>						
Input Source Current	$I_{SS}$	$V_{SS}=0V$	-10	-20	-30	$\mu A$
<b>Oscillator Section</b>						
Oscillation Frequency	f	$R_T=30K\Omega$	180	200	220	KHz
Frequency Change with Voltage	$\Delta f / \Delta V$	$V_{CC}=5V$ to 14V		2	10	%
		$V_{CH1}=V_{CH2}=5V$ to 25V				
Frequency change with temperature	$\Delta f / \Delta V$	$T_A = -10^\circ C$ to $85^\circ C$		5		%
<b>Period Adjustment Section</b>						
Maximum Duty Cycle	$T_{DUTY(ON)}$	$V_{FB}=0.7V$	85	90		%
Minimum Duty Cycle	$T_{DUTY(OFF)}$	$V_{FB}=0.9V$	0			%
<b>Total Device Section</b>						
Dynamic $V_{CC}$ Supply Current	$I_{CC(DYN)}$	$C_L=1500pF$	2	5	8	mA
Static $V_{CC}$ Supply Current	$I_{CCQ}$	$V_{SS}=0V$	1	3.3	6	mA
Dynamic $V_{CH1}$ Supply Current	$I_{CH1(DYN)}$	$C_L=1500pF$	2	7	10	mA
Static $V_{CH1}$ Supply Current	$I_{CH1Q}$	$V_{SS}=0V$	0.5	1	4.5	mA
Dynamic $V_{CH2}$ Supply Current	$I_{CH2(DYN)}$	$C_L=1500pF$	2	7	10	mA
Static $V_{CH2}$ Supply Current	$I_{CH2Q}$	$V_{SS}=0V$	0.5	1	4.5	mA

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Parameter	Symbol	Test Conditions	Min.	Typ.	Max.	Unit
<b>Error Amplifier Section</b>						
Input Threshold Voltage	$V_{FB1}, V_{FB2}$		0.784	0.8	0.816	V
$V_{FB}$ change with voltage	$\Delta V_{FB} / \Delta V$	$V_{CC}=5V$ to 14V		5	20	mV
$V_T$ change with temperature	$\Delta V_{FB} / \Delta T$	$T_A = -10^\circ C$ to $85^\circ C$		1		%
Input bias current	$I_{FB1}, I_{FB2}$	$V_{SS}=3V, V_{FB}=1V$		-0.1		$\mu A$
		$V_{SS}=0V, V_{FB}=1V$		-64		$\mu A$
CH1 Trans-conductance	$g_{m1}$		450	600	750	$\mu mho$
CH2 Trans-conductance	$g_{m2}$		450	600	750	$\mu mho$
<b>Output Section</b>						
Rise Time	$T_r$	$C_L=1500pF$		50	100	ns
Fall Time	$T_f$	$C_L=1500pF$		50	100	ns
Dead Band Time	$T_{db}$		50	100	150	ns
<b>Current Limit Section</b>						
OC Threshold Set Current	$I_{OCSET}$		20	30	40	$\mu A$
OC Threshold Voltage	$V_{OC}$	Sweep down		0.2		V
OC Comp Off-Set Voltage	$V_{OC(OFFSET)}$		-5	0	+5	mV

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## Function Description

### Voltage Reference

A 3.0V regulator operating from  $V_{CC}$  is used to power the internal circuitry of the FP5148. An internal resistive divider provides 0.8V reference for the error amplifier.

### Error Amplifier

The error amplifier compares a sample of the dc-dc converter output voltage with the 0.8V reference and generates an error signal for the error comparator. Output voltage of dc-dc converter is setting with the resistor divider using the following equation (see figure 1):

$$V_{OUT} = \left(1 + \frac{R1}{R2}\right) \times 0.8$$

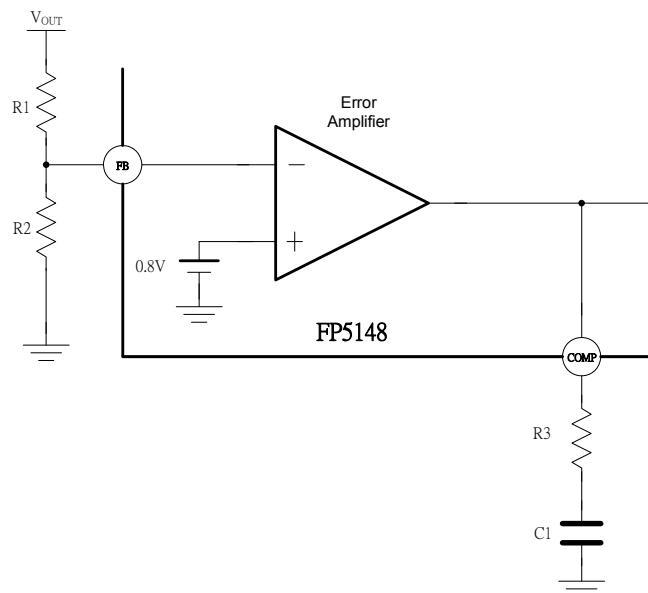


Figure 1 Error Amplifier with Feedback Resistance Divider

### Oscillator

The oscillator frequency ( $f_{osc}$ ) can be set from 20KHz to 500KHz by connecting a resistor  $R_T$  from  $R_T$  pin to GND.

$$f = \frac{6000}{R_T(K\Omega)} \text{ (KHz)}$$



## Under Voltage Lockout (UVLO)

FP5148 has a different under voltage lockout voltage for  $V_{CC}$  (typ. 4.2V) and  $V_{CH1} / V_{CH2}$  (typ. 3.3V). The controller starts to work when supply voltage is higher than the lockout point; otherwise the output drivers of FP5148 are turned off.

## Soft Start

When the  $V_{CC}$  and  $V_{CH1} / V_{CH2}$  of FP5148 are ready, the power on reset (POR) signal is turned off. The internal current source (20 $\mu$ A typ.) will charge the external capacitor, which is connected to the soft-start(SS) pin, to about 3V. Also, another internal current source (64 $\mu$ A max.) will control feedback voltage input of error amplifier and enable output driver for soft-start function. The soft-start timing can be decided by following equation:

$$T_{ss}(ms) = 75 \times C_{ss}(\mu F)$$

FP5148 will be shutdown whenever a voltage under 0.5V is forced on soft-start pin.

## Short-Circuit Protection

FP5148 will be shutdown immediately whenever COMP voltage is lower than 0.4V. The mean is hard to maintain the output voltage during maximum duty cycle under short-circuit, and the PWM output are off until power restart.

## Output Transistor and Boost Voltage $V_c$

FP5148 uses four NMOS and their turn-on VGS voltage near 4.5V, the source voltage of high side NMOS is near  $V_{IN}$ , and it is necessary to supply a boost voltage higher than  $V_{IN}$  for high side NMOS gate drive, the following figure explains the relation of MOS gate to source voltage. Application circuits can use different method to solve the boost power supply  $V_{CH1} / V_{CH2}$  for high side driver.

**(Recommend boost voltage  $V_{CH1} / V_{CH2}$  is the sum of  $V_{IN}$  and MOS  $V_{GS}$ )**

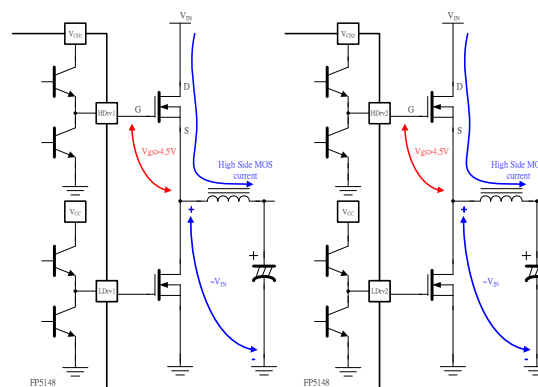
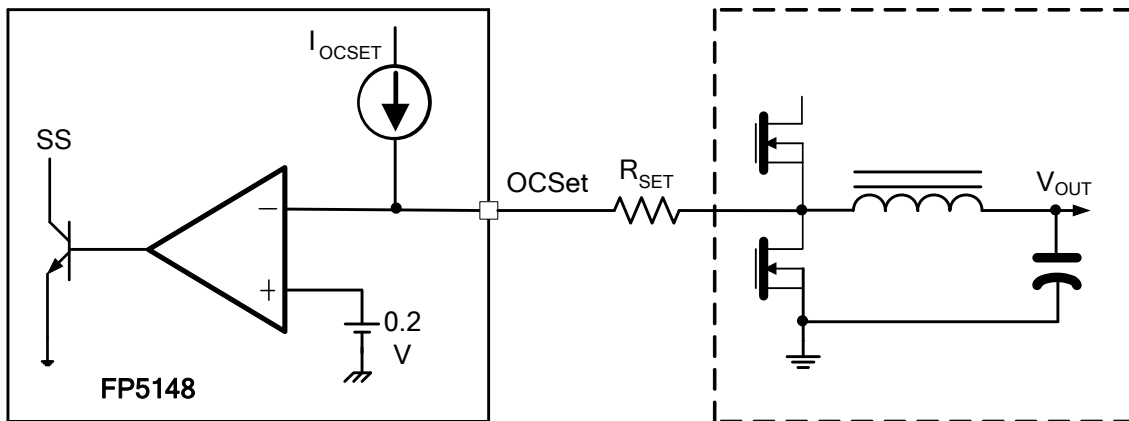


Figure 2 IC Output Stage with NMOS

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### Over-Current Protection

Over-current protection is achieved with a cycle by cycle scheme. It is performed by sensing current through the  $R_{DS(ON)}$  of low side MOSFET. As shown in the figure below, an external resistor  $R_{SET}$  is connected between OCSet pin and the drain of low side MOSFET (Q2) to set the current limit point. The internal current source develops a voltage across  $R_{SET}$ . When the low side switch is turned on, the inductor current flows through the Q2 and results in a voltage which is given by:

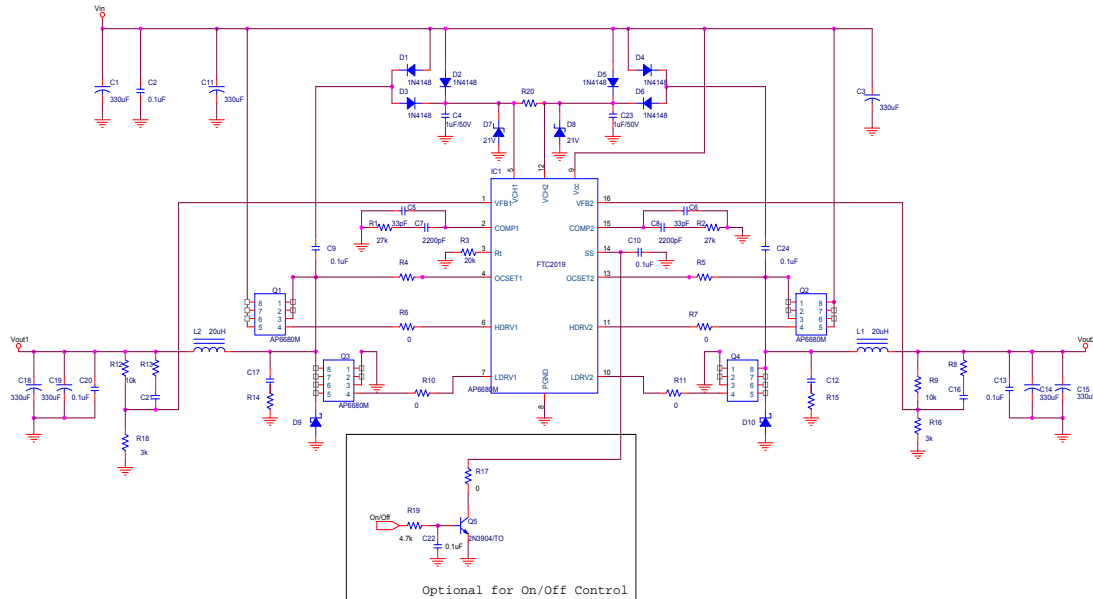


When Voltage  $V_{OCSET}$  is below 0.2V, the current sensing comparator flips and pull SS pin low. The high side MOSFET is turned off and the low side MOSFET is turned on until the inductor current reduces to below the over current setting value. The critical inductor current can be calculated by

$$V_{OCSET} = I_{OCSET} \times R_{SET} - R_{DS(ON)} \times I_L = 0.2$$

$$I_{SET} = I_{L(CRITICAL)} = \frac{R_{SET} \times I_{OCSET} - 0.2}{R_{DS(ON)}}$$

## Application Information



Above is a simple application circuit using power supply +5V. The converting voltage is from +5V to +3.3V (dual output) , and  $V_{CH1} / V_{CH2}$  pin voltage are high enough to make high side NMOS turn-on.

The output1 voltage is set by following equation:

$$V_{OUT1} = \left(1 + \frac{R12}{R18}\right) \times V_{REF} = \left(1 + \frac{10K}{3K}\right) \times 0.8V = 3.47V$$

The output2 voltage is set by following equation:

$$V_{OUT2} = \left(1 + \frac{R9}{R16}\right) \times V_{REF} = \left(1 + \frac{10K}{3K}\right) \times 0.8V = 3.47V$$

The soft-start time can be calculated by following equation:

$$T_{SS} = 75 \times C10 = 75 \times 0.1(\mu F) = 7.5ms$$

The over current protection that set by  $R_{SET}$  to limit the max inductance current is:

$$V_{OCSET} = I_{OCSET} \times R_{SET} - R_{DS(ON)} \times I_L = 0.2$$

$$I_{SET} = I_{L(CRITICAL)} = \frac{R_{SET} \times I_{OCSET} - 0.2}{R_{DS(ON)}}$$

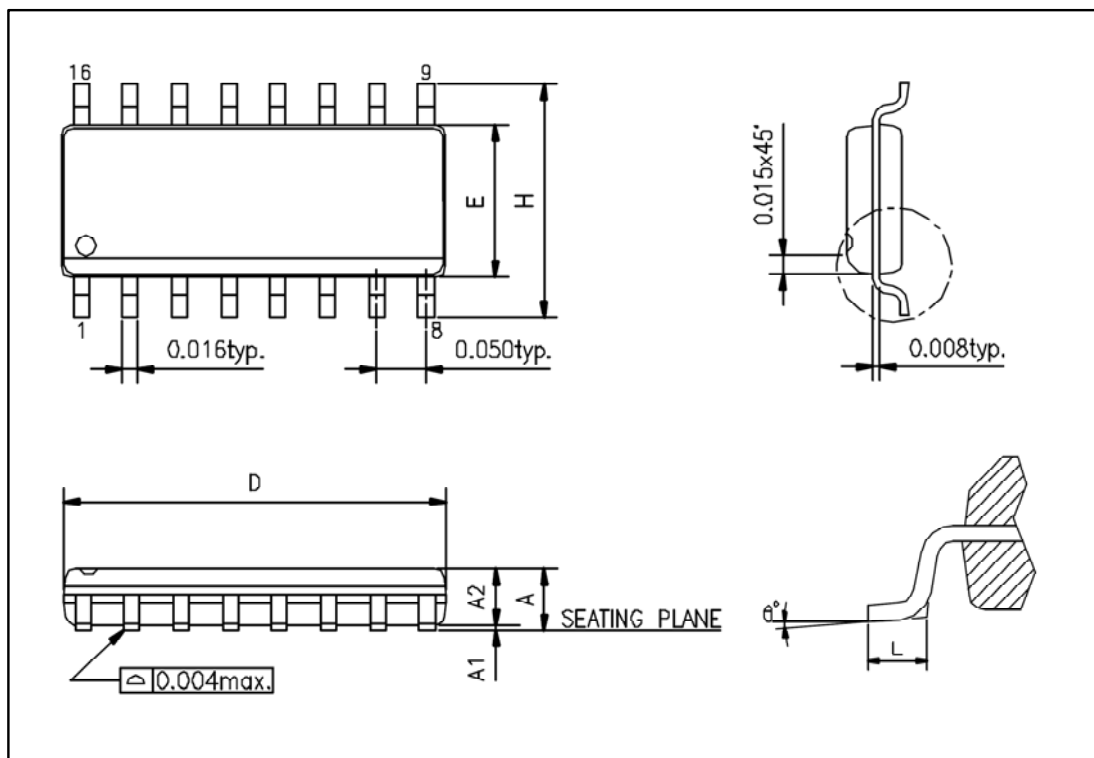
$$I_{SET1} = I_{L(CRITICAL)} = \frac{R4 \times I_{OCSET} - 0.2}{R_{DS(ON)}}$$

$$I_{SET2} = I_{L(CRITICAL)} = \frac{R5 \times I_{OCSET} - 0.2}{R_{DS(ON)}}$$

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## Package Outline

SOP-16L



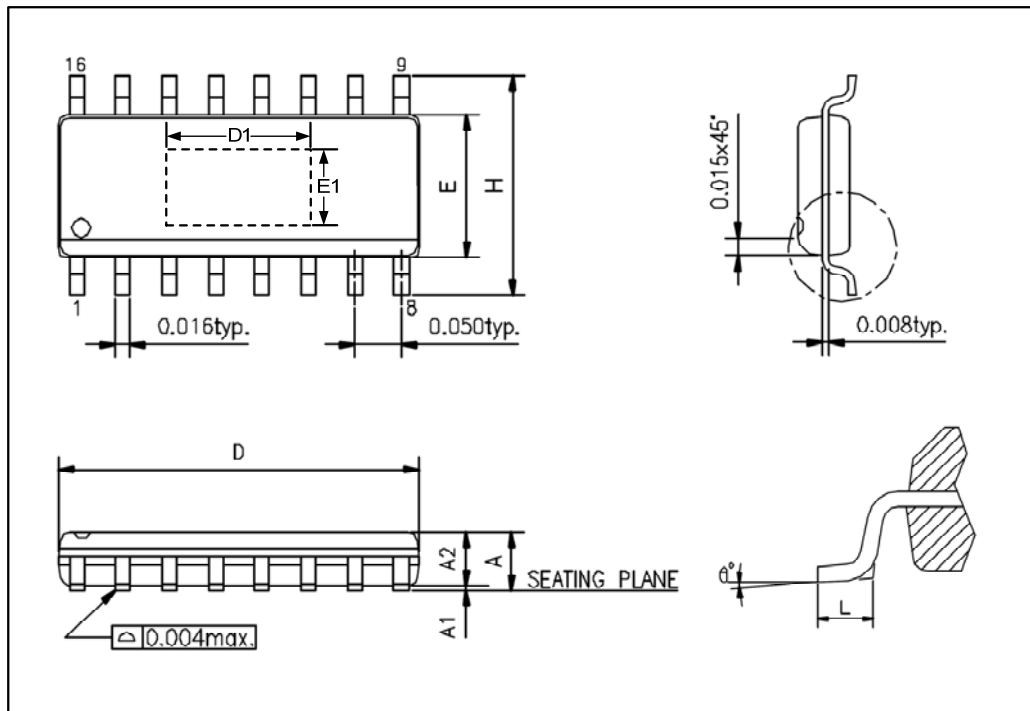
UNIT: mm

Symbols	Min. (mm)	Max. (mm)
A	1.346	1.752
A1	0.101	0.254
A2	1.244	1.651
D	9.804	10.007
E	3.810	3.987
H	5.791	6.197
L	0.406	1.270
$\theta^\circ$	0°	8°

**Note:**

1. Package dimensions are in compliance with JEDEC outline: MS-012 AC.
2. Dimension "D" does not include molding flash, protrusions or gate burrs.
3. Dimension "E" does not include inter-lead flash or protrusions.

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**SOP-16L (EP)**

**UNIT: mm**

Symbols	Min. (mm)	Max. (mm)
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A1	0.101	0.254
A2	1.244	1.651
D	9.804	10.007
E	3.810	3.987
H	5.791	6.197
L	0.406	1.270
$\theta^\circ$	0°	8°

**Exposed PAD Dimensions:**

Symbols	Min. (mm)	Max. (mm)
E1		2.184 REF
D1		4.114 REF

**Note:**

1. Dimension "D" does not include molding flash, protrusions or gate burrs.
2. Dimension "E" does not include inter-lead flash or protrusions.