Normally – OFF Silicon Carbide Junction Transistor

V_{DS}	=	600 V
$R_{DS(ON)}$	=	120 mΩ
I_D @ 25 °C	=	25 A
h_{FE}	=	120

Features

- 210°C maximum operating temperature
- Gate Oxide Free SiC switch
- Exceptional Safe Operating Area
- Excellent Gain Linearity
- Temperature Independent Switching Performance
- Low Output Capacitance
- Positive Temperature Co-efficient of $R_{DS,ON}$
- Suitable for connecting an anti-parallel diode
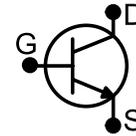
Die Size = 2.10 mm x 2.10 mm

Advantages

- Compatible with Si MOSFET/IGBT gate-drivers
- > 20 µs Short-Withstand Capability
- Lowest-in-class Conduction Losses
- High Circuit Efficiency
- Minimal Input Signal Distortion
- High Amplifier Bandwidth

Applications

- Down Hole Oil Drilling, Geothermal Instrumentation
- Hybrid Electric Vehicles (HEV)
- Solar Inverters
- Switched-Mode Power Supply (SMPS)
- Power Factor Correction (PFC)
- Induction Heating
- Uninterruptible Power Supply (UPS)
- Motor Drives

Absolute Maximum Ratings (T_C = 25 °C unless otherwise specified)

Parameter	Symbol	Conditions	Value	Unit	Notes
Drain – Source Voltage	V_{DS}	V_{GS} = 0 V	600	V	
Continuous Drain Current	I_D	T_C = 25°C	25	A	
Continuous Drain Current	I_D	T_C > 125°C, assumes R_{thJC} < 0.88 °C/W	10	A	
Continuous Gate Current	I_{GM}		1.3	A	
Turn-Off Safe Operating Area	RBSOA	T_{VJ} = 210 °C, Clamped Inductive Load	$I_{D,max}$ = 10 @ $V_{DS} \leq V_{DSmax}$	A	
Short Circuit Safe Operating Area	SCSOA	T_{VJ} = 210 °C, I_G = 1 A, V_{DS} = 400 V, Non Repetitive	> 20	µs	
Reverse Gate – Source Voltage	V_{SG}		30	V	
Reverse Drain – Source Voltage	V_{SD}		25	V	
Operating Junction and Storage Temperature	T_j, T_{stg}		-55 to 210	°C	
Maximum Processing Temperature	T_{Proc}	10 min. maximum	325	°C	

Electrical Characteristics

| Parameter	Symbol	Conditions	Value Min.	Value Typical	Value Max.	Unit	Notes
On State Characteristics							
Drain – Source On Resistance	$R_{DS(ON)}$	I_D = 10 A, T_j = 25 °C I_D = 10 A, T_j = 125 °C I_D = 10 A, T_j = 175 °C I_D = 10 A, T_j = 210 °C		120 180 240 320		mΩ	Fig. 5
Gate – Source Saturation Voltage	$V_{GS,SAT}$	I_D = 10 A, I_D/I_G = 40, T_j = 25 °C I_D = 10 A, I_D/I_G = 30, T_j = 175 °C		3.50 3.27		V	Fig. 4
DC Current Gain	h_{FE}	V_{DS} = 5 V, I_D = 10 A, T_j = 25 °C V_{DS} = 5 V, I_D = 10 A, T_j = 125 °C V_{DS} = 5 V, I_D = 10 A, T_j = 175 °C V_{DS} = 10 V, I_D = 10 A, T_j = 210 °C		120 87 80 76		–	Fig. 5
Off State Characteristics							
Drain Leakage Current	I_{DSS}	V_R = 600 V, V_{GS} = 0 V, T_j = 25 °C V_R = 600 V, V_{GS} = 0 V, T_j = 125 °C V_R = 600 V, V_{GS} = 0 V, T_j = 210 °C		10 50 100		µA	Fig. 6
Gate Leakage Current	I_{SG}	V_{SG} = 20 V, T_j = 25 °C		20		nA	

Figures
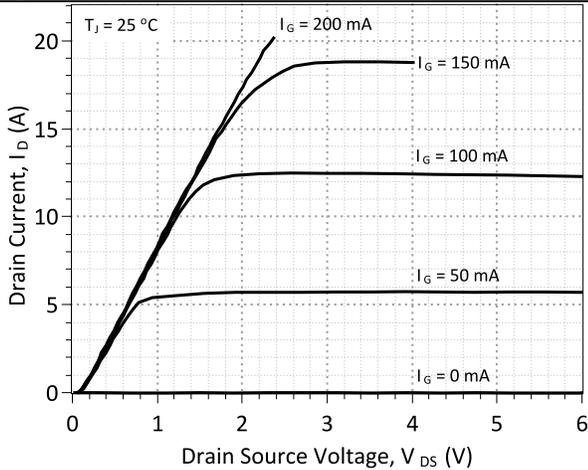
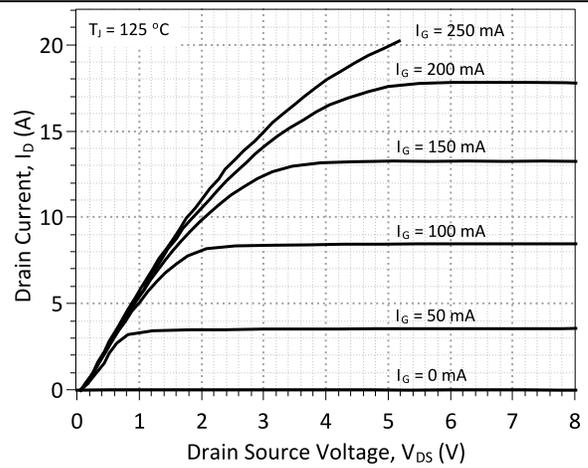Figure 1: Typical Output Characteristics at 25 °C

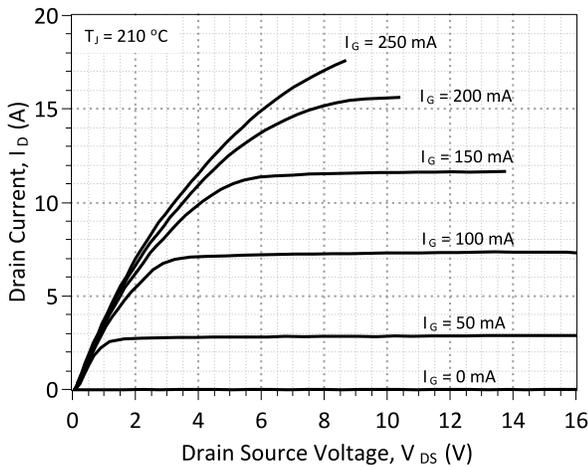Figure 2: Typical Output Characteristics at 125 °C

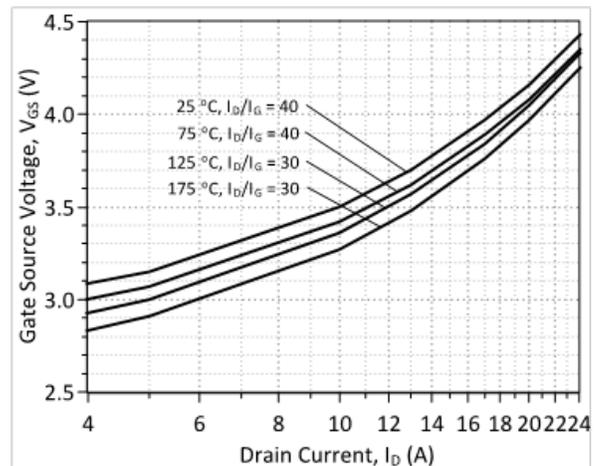Figure 3: Typical Output Characteristics at 210 °C
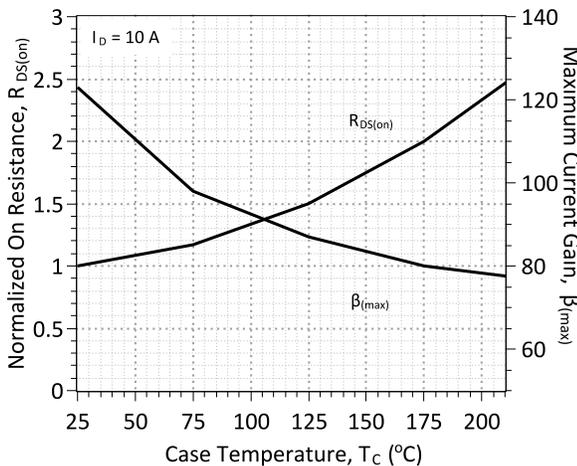

Figure 4: Typical Gate – Source Saturation Voltage

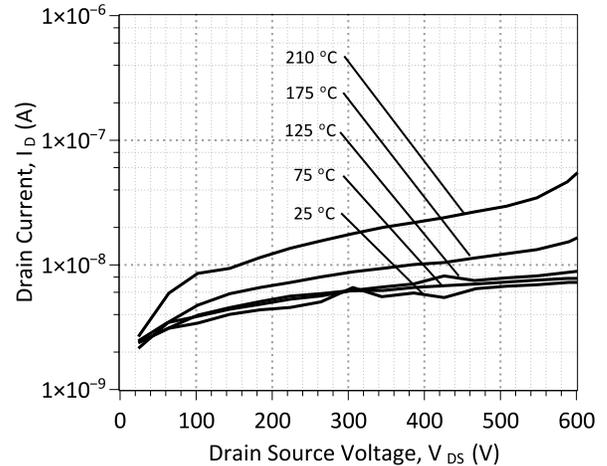Figure 5: Normalized On-Resistance and Current Gain vs. Temperature

Figure 6: Typical Blocking Characteristics

Driving the GA10JT06-CAL

Drive Topology	Gate Drive Power Consumption	Switching Frequency	Application Emphasis	Availability
TTL Logic	High	Low	Wide Temperature Range	Coming Soon
Constant Current	Medium	Medium	Wide Temperature Range	Coming Soon
High Speed – Boost Capacitor	Medium	High	Fast Switching	Production
High Speed – Boost Inductor	Low	High	Ultra Fast Switching	Coming Soon
Proportional	Lowest	High	Wide Drain Current Range	Coming Soon
Pulsed Power	Medium	N/A	Pulse Power	Coming Soon

A: Static TTL Logic Driving

The GA10JT06-CAL may be driven using direct (5 V) TTL logic after current amplification. The (amplified) current level of the supply must meet or exceed the steady state gate current ($I_{G,steady}$) required to operate the GA10JT06- CAL. The power level of the supply can be estimated from the target duty cycle of the particular application. $I_{G,steady}$ is dependent on the anticipated drain current ID through the SJT and the DC current gain h_{FE}, it may be calculated from the following equation. An accurate value of the h_{FE} may be read from Figure 5.

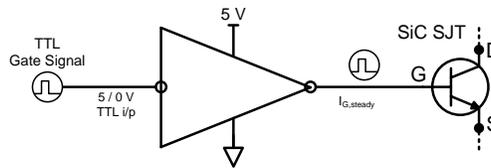$$I_{G,steady} \approx \frac{I_D}{h_{FE}(T, I_D)} * 1.5$$

Figure 7: TTL Gate Drive Schematic

B: High Speed Driving

The SJT is a current controlled transistor which requires a positive gate current for turn-on as well as to remain in on-state. An ideal gate current waveform for ultra-fast switching of the SJT, while maintaining low gate drive losses, is shown in Figure 8 which features a positive current peak during turn-on, a negative current peak during turn-off, and continuous gate current to remain on.
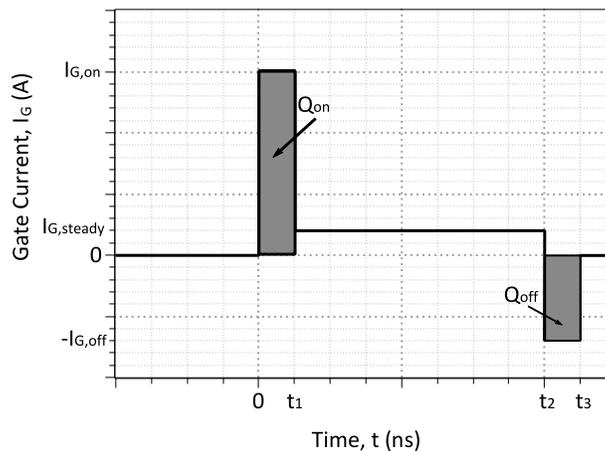

Figure 8: An idealized gate current waveform for fast switching of an SJT.

An SJT is rapidly switched from its blocking state to on-state, when the necessary gate charge, Q_G, for turn-on is supplied by a burst of high gate current, $I_{G,on}$, until the gate-source capacitance, C_{GS}, and gate-drain capacitance, C_{GD}, are fully charged.

$$Q_{on} = I_{G,on} * t_1$$

$$Q_{on} \geq Q_{gs} + Q_{gd}$$

Ideally, $I_{G,pon}$ should terminate when the drain voltage falls to its on-state value in order to avoid unnecessary drive losses during the steady on-state. In practice, the rise time of the $I_{G,on}$ pulse is affected by the parasitic inductances, L_{par} in the device package and drive circuit. A voltage developed across the parasitic inductance in the source path, L_s, can de-bias the gate-source junction, when high drain currents begin to flow through the device. The voltage applied to the gate pin should be maintained high enough, above the $V_{GS,sat}$ (see Figure 4) level to counter these effects.

A high negative peak current, $-I_{G,off}$ is recommended at the start of the turn-off transition, in order to rapidly sweep out the injected carriers from the gate, and achieve rapid turn-off. While satisfactory turn off can be achieved with $V_{GS} = 0$ V, a negative gate voltage V_{GS} may be used in order to speed up the turn-off transition.

Two high-speed drive topologies for the SiC SJTs are presented below.

B:1: High Speed, Low Loss Drive with Boost Capacitor, GA03IDDJT30-FR4

The GA10JT06- CAL may be driven using a High Speed, Low Loss Drive with Boost Capacitor topology in which multiple voltage levels, a gate resistor, and a gate capacitor are used to provide fast switching current peaks at turn-on and turn-off and a continuous gate current while in on-state. A 3 kV isolated evaluation gate drive board (GA03IDDJT30-FR4) utilizing this topology is commercially available for high and low-side driving, its datasheet provides additional details about this drive topology.
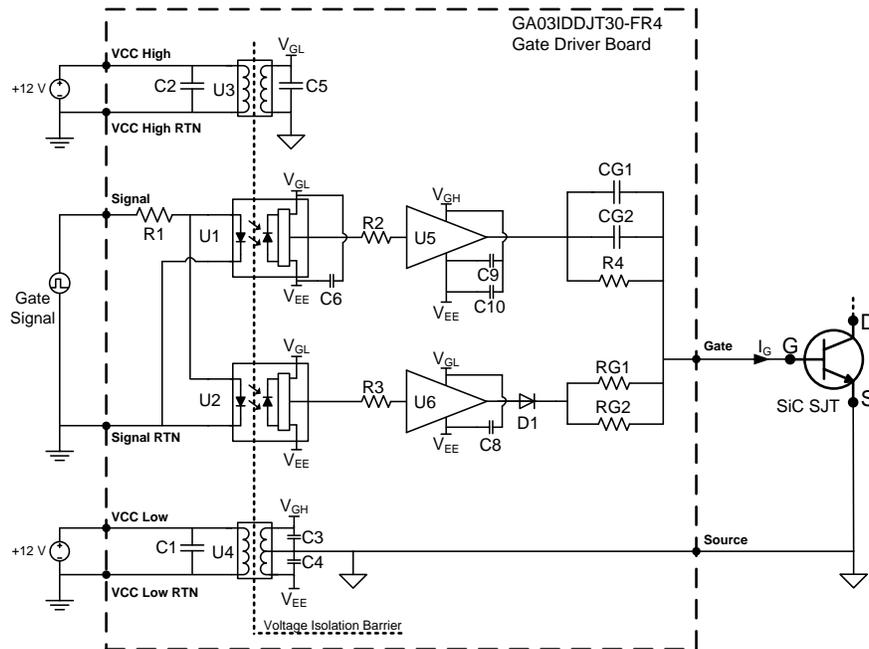

Figure 9: Topology of the GA03IDDJT30-FR4 Two Voltage Source gate driver.

The GA03IDDJT30-FR4 evaluation board comes equipped with two on board gate drive resistors (RG1, RG2) pre-installed for an effective gate resistance[3] of $R_G = 3.75\,\Omega$. It may be necessary for the user to reduce RG1 and RG2 under high drain current conditions f or safe operation of the GA10JT06- CAL. The steady state current supplied to the gate pin of the GA10JT06- CAL with on-board $R_G = 3.75\,\Omega$, is shown in Figure 10. The maximum allowable safe value of R_G for the user's required drain current can be read from Figure 11.

For the GA10JT06-CAL, R_G must be reduced for $I_D \geq$ ~10 A for safe operation with the GA03IDDJT30-FR4.

For operation at $I_D \approx$ ~10 A, R_G may be calculated from the following equation, which contains the DC current gain h_{FE} (Figure 5) and the gate-source saturation voltage $V_{GS,sat}$ (Figure 4).

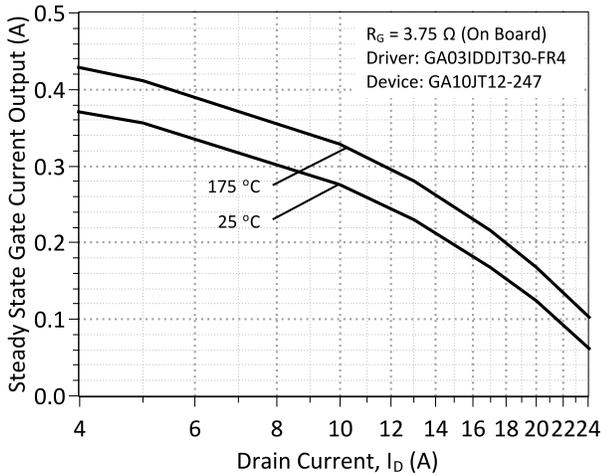$$R_{G,max} = \frac{\left(4.7V - V_{GS,sat}\right) * h_{FE}(T, I_D)}{I_D * 1.5} - 0.6\Omega$$

Figure 10: Typical steady state gate current supplied by the GA03IDDJT30-FR4 board for the GA10JT06-CAL with the on board resistance of 3.75 Ω
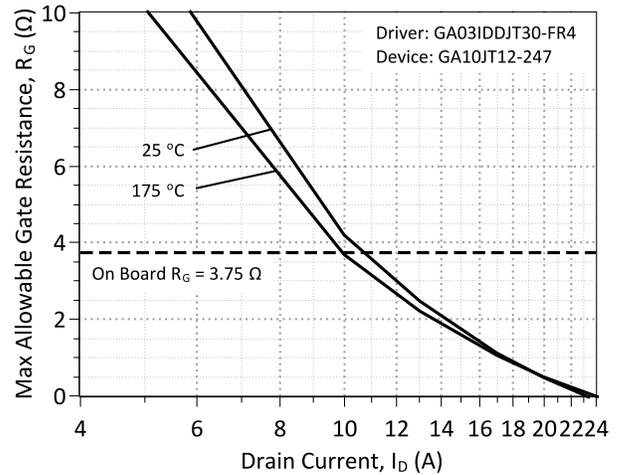

Figure 11: Maximum gate resistance for safe operation of the GA10JT06-CAL at different drain currents using the GA03IDDJT30-FR4 board.

B:2: High Speed, Low Loss Drive with Boost Inductor

A High Speed, Low-Loss Driver with Boost Inductor is also capable of driving the GA10JT06- CAL at high-speed. It utilizes a gate drive inductor instead of a capacitor to provide the high-current gate current pulses $I_{G,on}$ and $I_{G,off}$. During operation, inductor L is charged to a specified $I_{G,on}$ current value then made to discharge I_L into the SJT gate pin using logic control of S_1, S_2, S_3, and S_4, as shown in Figure 12. After turn on, while the device remains on the necessary steady state gate current $I_{G,steady}$ is supplied from source VCC through RG. Please refer to the article "A current-source concept for fast and efficient driving of silicon carbide transistors" by Dr. Jacek Rąbkowski for additional information on this driving topology.[4]
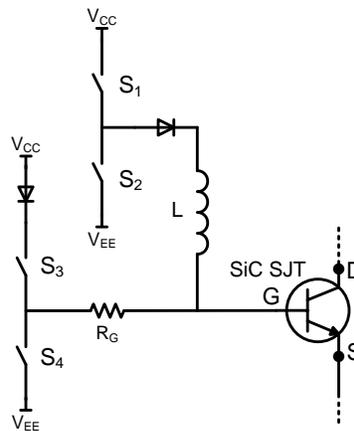

Figure 12: Simplified Inductive Pulsed Drive Topology

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[3] – $R_G = (1/RG1 +1/RG2)^{-1}$. Driver is pre-installed with RG1 = RG2 = 7.5 Ω

[4] – Archives of Electrical Engineering. Volume 62, Issue 2, Pages 333–343, ISSN (Print) 0004-0746, DOI: 10.2478/aee-2013-0026, June 2013

C: Proportional Gate Current Driving

For applications in which the GA10JT06- CAL will operate over a wide range of drain current conditions, it may be beneficial to drive the device using a proportional gate drive topology to optimize gate drive power consumption. A proportional gate driver relies on instantaneous drain current I_D feedback to vary the steady state gate current $I_{G,steady}$ supplied to the GA10JT06- CAL

C:1: Voltage Controlled Proportional Driver

The voltage controlled proportional driver relies on a gate drive IC to detect the GA10JT06- CAL drain-source voltage V_{DS} during on-state to sense I_D. The gate drive IC will then increase or decrease $I_{G,steady}$ in response to I_D. This allows $I_{G,steady}$, and thus the gate drive power consumption, to be reduced while I_D is relatively low or for $I_{G,steady}$ to increase when is I_D higher. A high voltage diode connected between the drain and sense protects the IC from high-voltage when the driver and GA10JT06- CAL are in off-state. A simplified version of this topology is shown in Figure1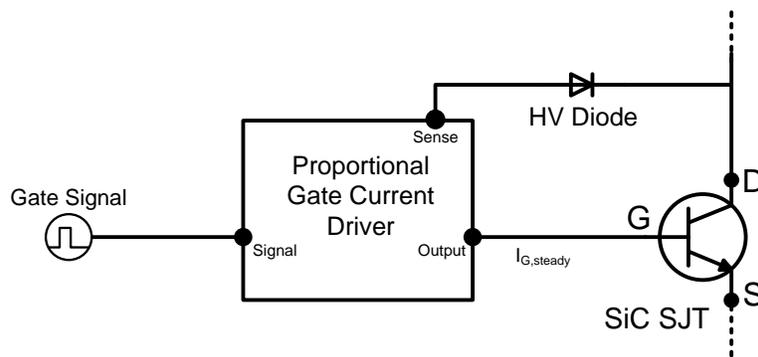3, additional information will be available in the future at http://www.genesicsemi.com/commercial-sic/sic-junction-transistors/

Figure 13: Simplified Voltage Controlled Proportional Driver

C:2: Current Controlled Proportional Driver

The current controlled proportional driver relies on a low-loss transformer in the drain or source path to provide feedback I_D of the GA10JT06-CAL during on-state to supply $I_{G,steady}$ into the device gate. $I_{G,steady}$ will then increase or decrease in response to I_D at a fixed forced current gain which is set be the turns ratio of the transformer, $h_{force} = I_D / I_G = N_2 / N_1$. GA10JT06- CAL is initially tuned-on using a gate current pulse supplied into an RC drive circuit to allow I_D current to begin flowing. This topology allows $I_{G,steady}$, and thus the gate drive power consumption, to be reduced while I_D is relatively low or for $I_{G,steady}$ to increase when is I_D higher. A simplified version of this topology is shown in Figure14, additional information will be available in the future at http://www.genesicsemi.com/commercial-sic/sic-junction-transistors/.
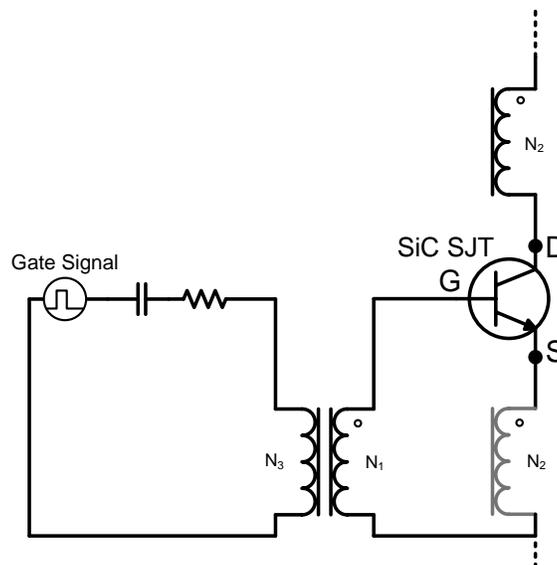

Figure 14: Simplified Current Controlled Proportional Driver

Mechanical Parameters

Die Dimensions	2.10 x 2.10	mm^2	83 x 83	mil^2
Die Area total / active	4.41/3.31	mm^2	6836/5134	mil^2
Die Thickness	360	µm	14	mil
Wafer Size	100	mm	3937	mil
Flat Position	0	deg	0	deg
Die Frontside Passivation	Polyimide			
Gate/Source Pad Metallization	4000 nm Al			
Bottom Drain Pad Metallization	400 nm Ni + 200 nm Au			
Die Attach	Electrically conductive glue or solder			
Wire Bond	Al ≤ 10 mil (Source) Al ≤ 3 mil (Gate)			
Reject ink dot size	Φ ≥ 0.3 mm			
Recommended storage environment	Store in original container, in dry nitrogen, < 6 months at an ambient temperature of 23 °C			

Chip Dimensions
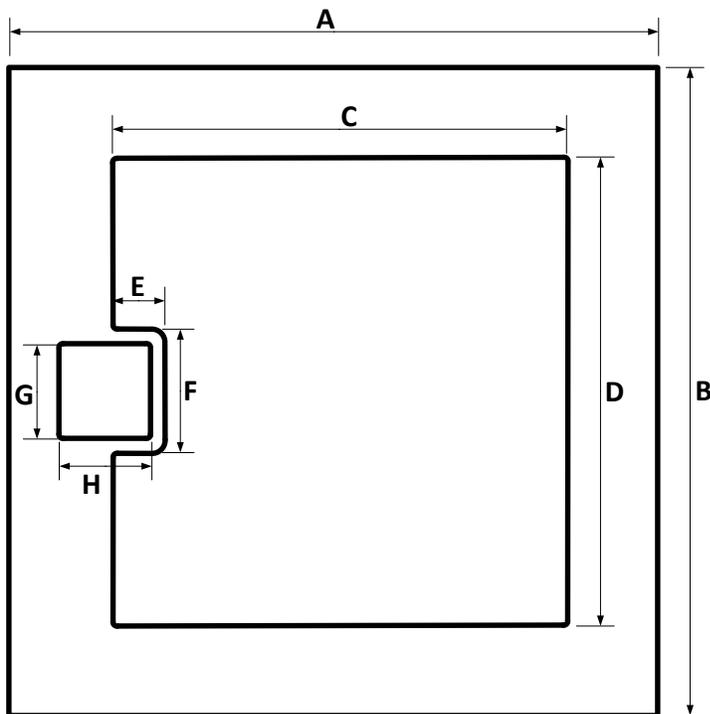
		mm	mil
DIE	A	2.10	83
	B	2.10	83
SOURCE WIREBONDABLE	C	1.47	58
	D	1.52	60
	E	0.17	7
	F	0.40	16
GATE WIREBONDABLE	G	0.30	12
	H	0.30	12

Revision History

Date	Revision	Comments	Supersedes
2015/02/06	1	Updated Electrical Characteristics	
2014/08/26	0	Initial Release	

GeneSiC
S E M I C O N D U C T O R

SPICE Model Parameters

This is a secure document. Please copy this code from the SPICE model PDF file on our website (http://www.genesicsemi.com/images/hit_sic/baredie/sjt/GA10JT06-CAL_SPICE.pdf) into LTSPICE (version 4) software for simulation of the GA10JT06-CAL.

```
*      MODEL OF GeneSiC Semiconductor Inc.
*
*      $Revision:   2.0            $
*      $Date:   26-AUG-2014        $
*
*      GeneSiC Semiconductor Inc.
*      43670 Trade Center Place Ste. 155
*      Dulles, VA 20166
*
*      COPYRIGHT (C) 2014 GeneSiC Semiconductor Inc.
*      ALL RIGHTS RESERVED
*
* These models are provided "AS IS, WHERE IS, AND WITH NO WARRANTY
* OF ANY KIND EITHER EXPRESSED OR IMPLIED, INCLUDING BUT NOT LIMITED
* TO ANY IMPLIED WARRANTIES OF MERCHANTABILITY AND FITNESS FOR A
* PARTICULAR PURPOSE."
* Models accurate up to 2 times rated drain current.
*
.model GA10JT06 NPN
+ IS       4.4E-48
+ ISE      1.858E-28
+ EG       3.23
+ BF       125
+ BR       0.55
+ IKF      400
+ NF       1
+ NE       2
+ RB       7.0
+ RE       0.005
+ RC       0.105
+ CJC      3.96E-10
+ VJC      3.189
+ MJC      0.469
+ CJE      8.083E-10
+ VJE      3.1441
+ MJE      0.4308
+ XTI      3
+ XTB      -0.9
+ TRC1     1.0E-2
+ VCEO     600
+ ICRATING 10
+ MFG      GeneSiC_Semiconductor
*
* End of GA10JT06 SPICE Model
```