

SEMICONDUCTOR®

# HGTG7N60A4D, HGTP7N60A4D, HGT1S7N60A4DS

Data Sheet

January 2005

### 600V, SMPS Series N-Channel IGBT with Anti-Parallel Hyperfast Diode

### The HGTG7N60A4D, HGTP7N60A4D and

HGT1S7N60A4DS are MOS gated high voltage switching devices combining the best features of MOSFETs and bipolar transistors. These devices have the high input impedance of a MOSFET and the low on-state conduction loss of a bipolar transistor. The much lower on-state voltage drop varies only moderately between 25°C and 150°C. The IGBT used is the development type TA49331. The diode used in anti-parallel is the development type TA49370.

This IGBT is ideal for many high voltage switching applications operating at high frequencies where low conduction losses are essential. **This device has been optimized for high frequency switch mode power supplies**.

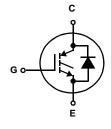
Formerly Developmental Type TA49333.

### **Ordering Information**

PART NUMBER	PACKAGE	BRAND
HGTG7N60A4D	TO-247	G7N60A4D
HGTP7N60A4D	TO-220AB	G7N60A4D
HGT1S7N60A4DS	TO-263AB	G7N60A4D

NOTE: When ordering, use the entire part number. Add the suffix 9A to obtain the TO-263AB variant in tape and reel, e.g., HGT1S7N60A4DS9A.

### Symbol

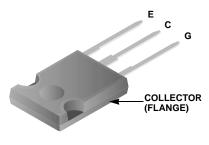


### Features

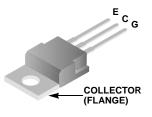
- >100kHz Operation At 390V, 7A
- 200kHz Operation At 390V, 5A
- 600V Switching SOA Capability
- Low Conduction Loss
- *Temperature Compensating* SABER<sup>™</sup> Model www.fairchildsemi.com

### Packaging

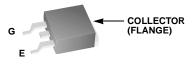




JEDEC TO-220AB



JEDEC TO-263AB



#### Absolute Maximum Ratings $T_C = 25^{\circ}C$ , Unless Otherwise Specified

	ALL TYPES	UNITS
Collector to Emitter Voltage BV <sub>CES</sub>	600	V
Collector Current Continuous		
At $T_{C} = 25^{\circ}C$ $I_{C25}$	34	А
At $T_{C} = 110^{9}C$ $I_{C110}$	14	А
Collector Current Pulsed (Note 1)I <sub>CM</sub>	56	А
Gate to Emitter Voltage Continuous	±20	V
Gate to Emitter Voltage Pulsed	±30	V
Switching Safe Operating Area at T <sub>J</sub> = 150 <sup>0</sup> C, Figure 2	35A at 600V	
Power Dissipation Total at $T_C = 25^{\circ}C$ $P_D$	125	W
Power Dissipation Derating $T_C > 25^{\circ}C$	1.0	W/ <sup>o</sup> C
Operating and Storage Junction Temperature Range	-55 to 150	°C
Maximum Lead Temperature for Soldering		
Leads at 0.063in (1.6mm) from Case for 10s.	300	°C
Package Body for 10s, See Tech Brief 334	260	°C

CAUTION: Stresses above those listed in "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress only rating and operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied.

#### NOTE:

1. Pulse width limited by maximum junction temperature.

PARAMETER	SYMBOL	TEST	CONDITIONS	MIN	TYP	MAX	UNITS
Collector to Emitter Breakdown Voltage	BVCES	$I_{C} = 250 \mu A, V_{GE} = 0 V$		600	-	-	V
Collector to Emitter Leakage Current	ICES	$V_{CE} = 600V$	$T_J = 25^{\circ}C$	-	-	250	μA
			T <sub>J</sub> = 125 <sup>o</sup> C	-	-	2	mA
Collector to Emitter Saturation Voltage	V <sub>CE(SAT)</sub>	I <sub>C</sub> = 7A,	T <sub>J</sub> = 25 <sup>0</sup> C	-	1.9	2.7	V
		V <sub>GE</sub> = 15V	T <sub>J</sub> = 125 <sup>o</sup> C	-	1.6	2.2	V
Gate to Emitter Threshold Voltage	V <sub>GE(TH)</sub>	$I_{C} = 250 \mu A, V_{CE}$	= 600V	4.5	5.9	7	V
Gate to Emitter Leakage Current	IGES	$V_{GE} = \pm 20V$		-	-	±250	nA
Switching SOA	SSOA	$T_J = 150^{\circ}C, R_G$ L = 100µH, V <sub>CE</sub>	= 25Ω, V <sub>GE</sub> = 15V, = 600V	35	-	-	A
Gate to Emitter Plateau Voltage	V <sub>GEP</sub>	I <sub>C</sub> = 7A, V <sub>CE</sub> = 3	00V	-	9	-	V
On-State Gate Charge	Q <sub>g(ON)</sub>	1/ 2001/	V <sub>GE</sub> = 15V	-	37	45	nC
			V <sub>GE</sub> = 20V	-	48	60	nC
Current Turn-On Delay Time	<sup>t</sup> d(ON)I	IGBT and Diode at $T_J = 25^{\circ}C$ , $I_{CE} = 7A$ , $V_{CE} = 390V$ , $V_{GE} = 15V$ , $R_G = 25\Omega$ , L = 1mH, Test Circuit (Figure 24)		-	11	-	ns
Current Rise Time	t <sub>rl</sub>			-	11	-	ns
Current Turn-Off Delay Time	t <sub>d(OFF)</sub> I			-	100	-	ns
Current Fall Time	t <sub>fl</sub>			-	45	-	ns
Turn-On Energy	E <sub>ON1</sub>			-	55	-	μJ
Turn-On Energy	E <sub>ON2</sub>			-	120	150	μJ
Turn-Off Energy (Note 2)	E <sub>OFF</sub>			-	60	75	μJ
Current Turn-On Delay Time	<sup>t</sup> d(ON)I	IGBT and Diode	at T <sub>J</sub> = 125 <sup>o</sup> C,	-	10	-	ns
Current Rise Time	t <sub>rl</sub>	$I_{CE} = 7A,$ Voc = 390V Vo	r = 15V	-	7	-	ns
Current Turn-Off Delay Time	t <sub>d(OFF)</sub> I	$ V_{CE} = 390V, V_{GE} = 15V, \\ R_{G} = 25\Omega, \\ L = 1mH, \\ Test Circuit (Figure 24) \\ - \\ - \\ - \\ - \\ - \\ - \\ - \\ - \\ - \\ $		-	130	150	ns
Current Fall Time	t <sub>fl</sub>			-	75	85	ns
Turn-On Energy (Note 2)	E <sub>ON1</sub>			-	50	-	μJ
Turn-On Energy (Note 2)	E <sub>ON2</sub>			-	200	215	μJ
Turn-Off Energy (Note 3)	EOFF			-	125	170	μJ

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PARAMETER	SYMBOL	TEST CONDITIONS	MIN	ТҮР	MAX	UNITS
Diode Forward Voltage	V <sub>EC</sub>	I <sub>EC</sub> = 7A	-	2.4	-	V
Diode Reverse Recovery Time	t <sub>rr</sub>	$I_{EC} = 7A$ , $dI_{EC}/dt = 200A/\mu s$	-	34	-	ns
		$I_{EC} = 1A$ , $dI_{EC}/dt = 200A/\mu s$	-	22	-	ns
Thermal Resistance Junction To Case	R <sub>θJC</sub>	IGBT	-	-	1.0	°C/W
		Diode	-	-	2.2	°C/W

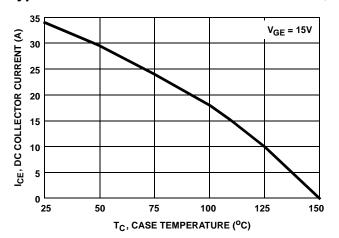
#### Electrical Specifications T<sub>J</sub> = 25<sup>o</sup>C, Unless Otherwise Specified (Continued)

NOTES:

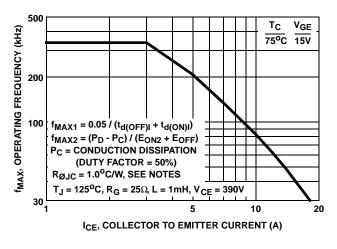
Values for two Turn-On loss conditions are shown for the convenience of the circuit designer. E<sub>ON1</sub> is the turn-on loss of the IGBT only. E<sub>ON2</sub> is the turn-on loss when a typical diode is used in the test circuit and the diode is at the same T<sub>J</sub> as the IGBT. The diode type is specified in Figure 24.

 Turn-Off Energy Loss (E<sub>OFF</sub>) is defined as the integral of the instantaneous power loss starting at the trailing edge of the input pulse and ending at the point where the collector current equals zero (I<sub>CE</sub> = 0A). All devices were tested per JEDEC Standard No. 24-1 Method for Measurement of Power Device Turn-Off Switching Loss. This test method produces the true total Turn-Off Energy Loss.

### Typical Performance Curves Unless Otherwise Specified







#### FIGURE 3. OPERATING FREQUENCY vs COLLECTOR TO EMITTER CURRENT

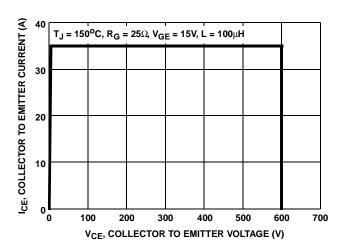
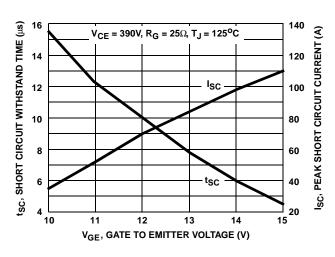


FIGURE 2. MINIMUM SWITCHING SAFE OPERATING AREA





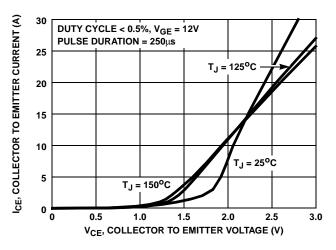


FIGURE 5. COLLECTOR TO EMITTER ON-STATE VOLTAGE

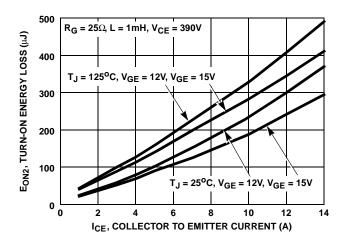
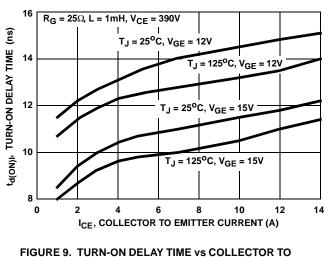


FIGURE 7. TURN-ON ENERGY LOSS vs COLLECTOR TO EMITTER CURRENT



EMITTER CURRENT

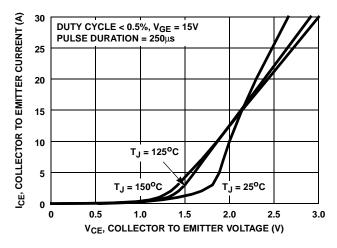


FIGURE 6. COLLECTOR TO EMITTER ON-STATE VOLTAGE

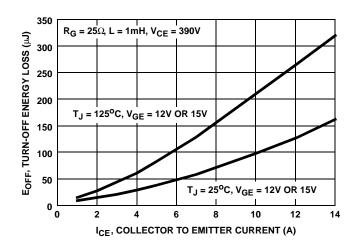
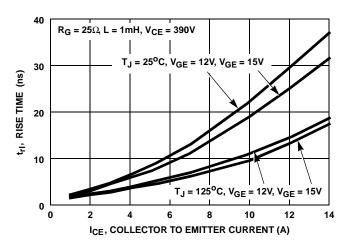
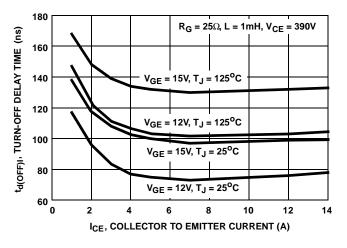


FIGURE 8. TURN-OFF ENERGY LOSS vs COLLECTOR TO EMITTER CURRENT









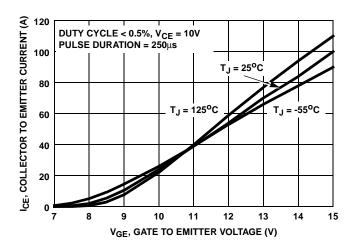
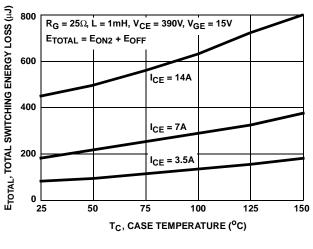


FIGURE 13. TRANSFER CHARACTERISTIC





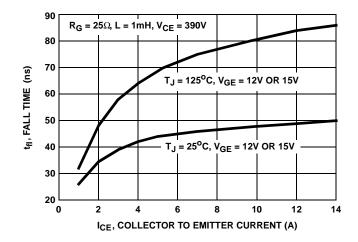


FIGURE 12. FALL TIME vs COLLECTOR TO EMITTER CURRENT

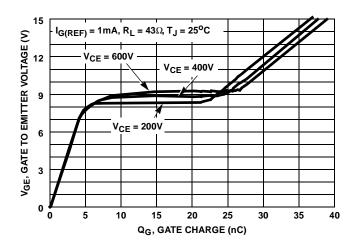
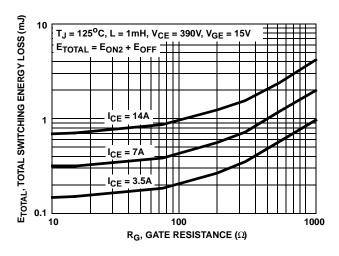
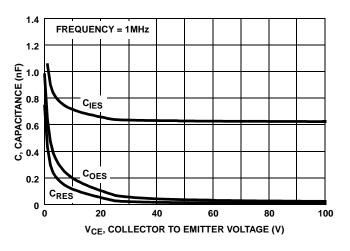


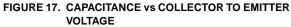
FIGURE 14. GATE CHARGE WAVEFORMS

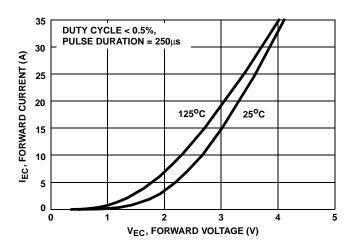




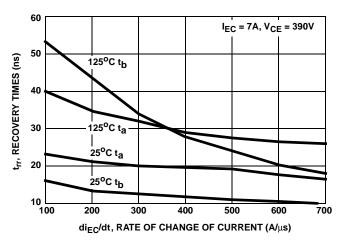
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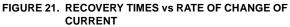












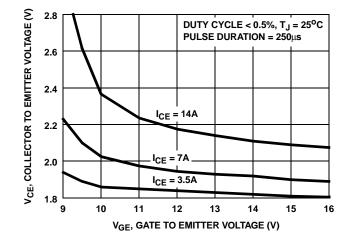
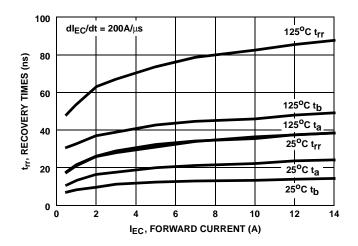
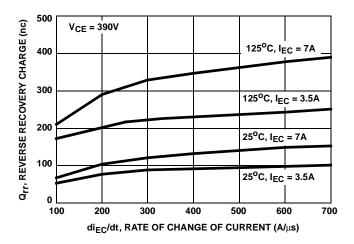


FIGURE 18. COLLECTOR TO EMITTER ON-STATE VOLTAGE vs GATE TO EMITTER VOLTAGE



#### FIGURE 20. RECOVERY TIMES vs FORWARD CURRENT



#### FIGURE 22. STORED CHARGE vs RATE OF CHANGE OF CURRENT

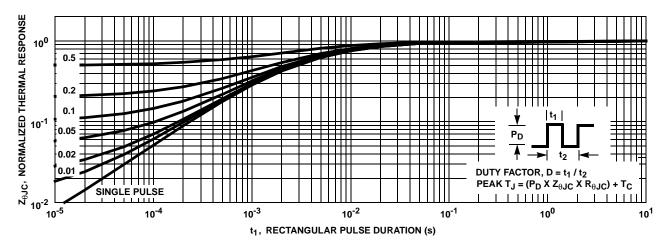


FIGURE 23. IGBT NORMALIZED TRANSIENT THERMAL RESPONSE, JUNCTION TO CASE

### Test Circuit and Waveforms

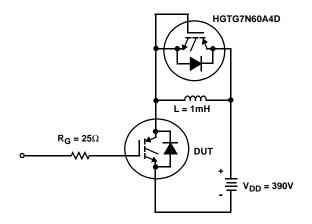


FIGURE 24. INDUCTIVE SWITCHING TEST CIRCUIT

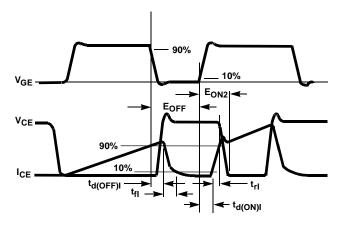


FIGURE 25. SWITCHING TEST WAVEFORMS

# Handling Precautions for IGBTs

Insulated Gate Bipolar Transistors are susceptible to gateinsulation damage by the electrostatic discharge of energy through the devices. When handling these devices, care should be exercised to assure that the static charge built in the handler's body capacitance is not discharged through the device. With proper handling and application procedures, however, IGBTs are currently being extensively used in production by numerous equipment manufacturers in military, industrial and consumer applications, with virtually no damage problems due to electrostatic discharge. IGBTs can be handled safely if the following basic precautions are taken:

- Prior to assembly into a circuit, all leads should be kept shorted together either by the use of metal shorting springs or by the insertion into conductive material such as "ECCOSORBD™ LD26" or equivalent.
- 2. When devices are removed by hand from their carriers, the hand being used should be grounded by any suitable means for example, with a metallic wristband.
- 3. Tips of soldering irons should be grounded.
- 4. Devices should never be inserted into or removed from circuits with power on.
- 5. Gate Voltage Rating Never exceed the gate-voltage rating of  $V_{GEM}$ . Exceeding the rated  $V_{GE}$  can result in permanent damage to the oxide layer in the gate region.
- 6. **Gate Termination** The gates of these devices are essentially capacitors. Circuits that leave the gate opencircuited or floating should be avoided. These conditions can result in turn-on of the device due to voltage buildup on the input capacitor due to leakage currents or pickup.
- 7. **Gate Protection** These devices do not have an internal monolithic Zener diode from gate to emitter. If gate protection is required an external Zener is recommended.

# **Operating Frequency Information**

Operating frequency information for a typical device (Figure 3) is presented as a guide for estimating device performance for a specific application. Other typical frequency vs collector current ( $I_{CE}$ ) plots are possible using the information shown for a typical unit in Figures 6, 7, 8, 9 and 11. The operating frequency plot (Figure 3) of a typical device shows  $f_{MAX1}$  or  $f_{MAX2}$ ; whichever is smaller at each point. The information is based on measurements of a typical device and is bounded by the maximum rated junction temperature.

 $f_{MAX1}$  is defined by  $f_{MAX1} = 0.05/(t_{d(OFF)I} + t_{d(ON)I})$ . Deadtime (the denominator) has been arbitrarily held to 10% of the on-state time for a 50% duty factor. Other definitions are possible.  $t_{d(OFF)I}$  and  $t_{d(ON)I}$  are defined in Figure 25. Device turn-off delay can establish an additional frequency limiting condition for an application other than  $T_{JM}$ .  $t_{d(OFF)I}$  is important when controlling output ripple under a lightly loaded condition.

 $f_{MAX2}$  is defined by  $f_{MAX2}$  = (P\_D - P\_C)/(E\_OFF + E\_ON2). The allowable dissipation (P\_D) is defined by P\_D = (T\_{JM} - T\_C)/R\_{\theta JC}. The sum of device switching and conduction losses must not exceed P\_D. A 50% duty factor was used (Figure 3) and the conduction losses (P\_C) are approximated by P\_C = (V\_{CE} \times I\_{CE})/2.

 $E_{ON2}$  and  $E_{OFF}$  are defined in the switching waveforms shown in Figure 25.  $E_{ON2}$  is the integral of the instantaneous power loss ( $I_{CE} \times V_{CE}$ ) during turn-on and  $E_{OFF}$  is the integral of the instantaneous power loss ( $I_{CE} \times V_{CE}$ ) during turn-off. All tail losses are included in the calculation for  $E_{OFF}$ ; i.e., the collector current equals zero ( $I_{CE} = 0$ ).

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