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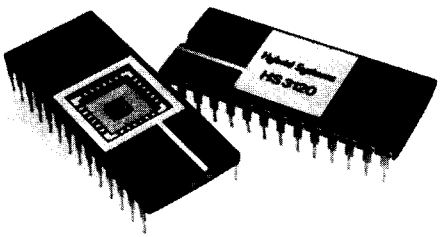
Hybrid Systems
Sipex Corporation

HS 3120
Double Buffered 12-Bit MDAC

FEATURES

- Monolithic Construction
- 12 Bit Resolution
- 0.01% Non-Linearity
- μP Compatible
- 4-Quadrant Multiplication
- Latch-up Protected



DESCRIPTION

The HS 3120 is a precision monolithic 12-bit multiplying DAC with internal two-stage input storage registers for easy interfacing with microprocessor busses. It is packaged in a 28-pin DIP to give high I/O design flexibility.

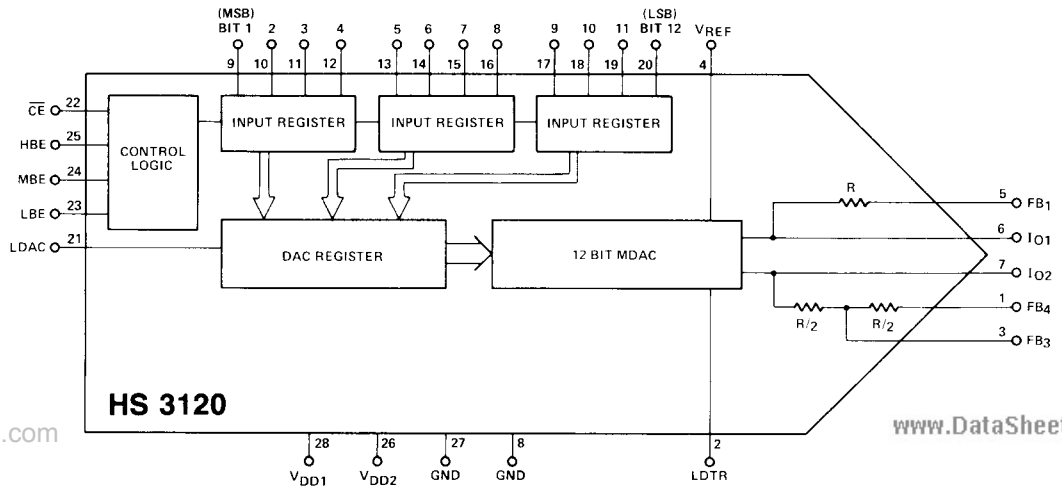
DOUBLE BUFFERED – The input registers are sectioned into 3 segments of 4 bits each, all individually addressable. The DAC-register, following the input registers, is a parallel 12-bit register for holding the DAC data while the input registers are updated. Only the data held in the DAC register determines the analog output value of the converter.

MICRO PROCESSOR COMPATIBLE – The HS 3120 has been designed for great flexibility in connecting to bus-oriented systems. The 12 data inputs are organized into 3 independent addressable 4-bit input registers such that the HS 3120 can be connected to either a 4, 8 or 16-bit data bus. The control logic of the HS 3120 includes chip enable and latch enable inputs for flexible memory mapping. All controls are level-triggered to allow static or dynamic operation.

VERSATILE OUTPUTS – A total of 5 output lines are provided by the HS 3120 to allow unipolar and bipolar output connection with a minimum of external components. The feedback resistor is internal. The resistor ladder network termination is externally available, thus eliminating an external resistor for the 1 LSB offset in bipolar mode.

MONOLITHIC CMOS CONSTRUCTION – The HS 3120 is a one-chip CMOS circuit with a resistor ladder network designed for 0.01% linearity without laser trimming. Small chip size and high manufacturing yields result in greatly reduced cost.

FUNCTIONAL DIAGRAM



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SPECIFICATIONS

(Typical @ 25°C, nominal power supply, V_{REF} = +10V, unipolar unless otherwise noted).

MODEL	HS 3120-2	HS 3120-0
TYPE	**MULTIPLYING, DOUBLE BUFFERED INPUTS**	*
DIGITAL INPUT		
Resolution	12-Bits	*
2-Quad. Unipolar Coding	Binary[1], Comp. Binary[1]	*
4-Quad. Bipolar Coding	Offset Binary	*
Logic Compatibility[2]	CMOS, TTL	*
Input Current	±1 μA (max)	*
Data Set-up Time[3]	250nS (min)	*
Strobe Width[3]	250nS (min)	*
Data Hold Time[3]	0nS (min)	*
REFERENCE INPUT		
Voltage Range	±25V (max)	*
Input Impedance	8kΩ ±50%	*
ANALOG OUTPUT		
Scale Factor	$125\mu A/V_{Ref}$ ±50%	*
Scale Factor Accuracy[4]	±0.4%	*
Output Leakage[5]		
@ 25°C	<10nA (max)	*
@ 125°C	<200nA (max)	*
Output Capacitance		
C_{OUT} 1, all inputs high	80pF	*
C_{OUT} 1, all inputs low	40pF	*
C_{OUT} 2, all inputs high	40pF	*
C_{OUT} 2, all inputs low	80pF	*
STATIC PERFORMANCE		
Integral Linearity	±0.015% F.S.R. (max)	±0.05 % F.S.R. (max)
Differential Linearity	±0.024% F.S.R. (max)	±0.097% F.S.R. (max)
Monotonicity	Guaranteed to 12 bits	Guaranteed to 10 bits
Monotonicity Temp. Range		
C-Models	0°C to +70°C	*
B-Models	-55°C to +125°C	*
DYNAMIC PERFORMANCE		
Digital Small Signal Settling	1.0μsec	*
Full Scale Transition Settling to 0.01% (strobed)	2.0μsec	*
Reference Feedthrough Error (V_{Ref} = $20V_{pp}$)		
@ 1kHz	<1mV	*
@ 10kHz	2mV	*
Delay to output		
from Bits input	100nS[6]	*
from $\overline{LDAC}$	200nS[6]	*
from $\overline{CE}$	120nS[6]	*
STABILITY (Over Specified Temp. Range)		
Scale Factor[4]	2 ppm F.S.R./°C (max)	*
Integral Linearity	0.2 ppm F.S.R./°C (max)	*
Differential Linearity	0.2 ppm F.S.R./°C (max)	*
Monotonicity Temp. Range		
C-Option	0°C to +70°C	*
B-Option	-55°C to +125°C	*
POWER SUPPLY (V_{DD})		
Operating Voltage (specifications guaranteed)	+15V ±5%	*
Maximum Voltage Range	+5V to 16V	*
Current	2.5mA (max)	*
Rejection Ratio	0.002%/% (max)	*
TEMPERATURE RANGE		
Operating C-Option	0°C to +70°C	*
Operating B-Option	-55°C to +125°C	*
Storage	-65°C to +150°C	*
MECHANICAL		
Case Style	28-pin double DIP	*
C-Option	plastic or ceramic	*
B-Option	ceramic	*

NOTES:

* Same as HS 3120-2

1. The input coding is complementary binary if I_{O2} is used.
2. Digital input voltage must not exceed supply voltage or go below -0.5V. "0" <0.8V, 2.4V < "1" ≤V_{DD}.
3. All strobes are level triggered. See TIMING DIAGRAM.
4. Using the internal feedback resistor and an external opamp.
5. The output leakage current will create an offset voltage at the external opamps output. It doubles every 10°C temperature increase.
6. Delay times are twice the amount shown at T_A = +125°C

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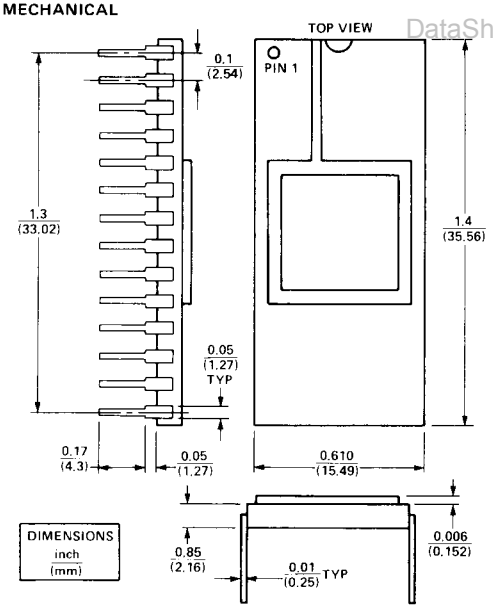
PIN ASSIGNMENTS

PIN	FUNCTION
1	FB_4, Feedback Bipolar Operation
2	LDTR, Ladder Termination
3	FB_3, Feedback Bipolar Operation
4	V_{REF}, Reference Voltage Input
5	FB_1, Feedback, Unipolar/Bipolar
6	I_{O1}, Current out into virtual ground
7	I_{O2}, Current out-complement of I_{O1}
8	V_{SS}, Ground, Analog and DAC Register
9	Bit 1, MSB
10	Bit 2
11	Bit 3
12	Bit 4
13	Bit 5
14	Bit 6
15	Bit 7
16	Bit 8
17	Bit 9
18	Bit 10
19	Bit 11
20	Bit 12
21	LDAC, Transfers data from input to DAC register
22	$\overline{CE}$, Chip Enable, active low
23	LBE, Bit 12 to Bit 9 Enable
24	MBE, Bit 8 to Bit 5 Enable
25	HBE, Bit 4 to Bit 1 Enable
26	V_{DD2}, Supply Analog and DAC Register
27	V_{SS1}, Ground input latches
28	V_{DD1}, Supply input latches

NOTE: Pins 8 and 27 and pins 26 and 28 must be connected externally.

MECHANICAL



CONNECTIONS

Unipolar Operation: Connect I_{O1} and FB_1 as shown in diagram. Tie I_{O2} (Pin 7), FB_3 (Pin 3), FB_4, (Pin 1) all to Ground (Pin 8)

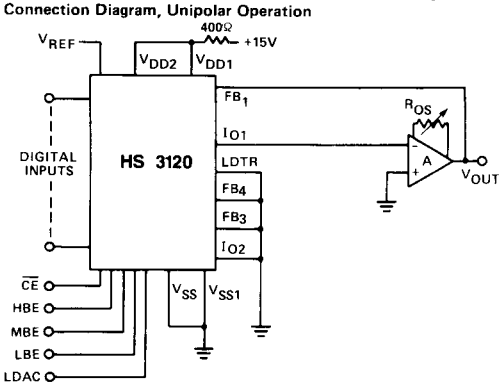
Bipolar Operation: Connect I_{O1}, I_{O2}, FB_1, FB_3, FB_4 as shown in diagram. Tie LDTR to I_{O2}

Grounding: Connect all GRD to system analog ground and tie this to digital ground.

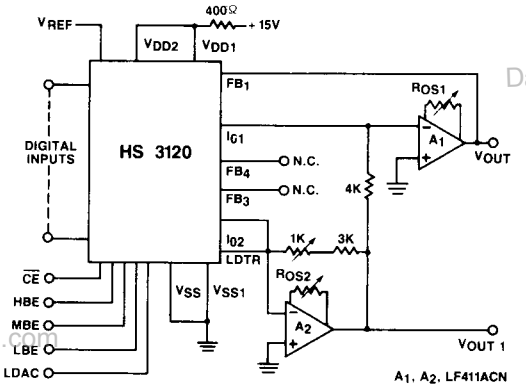
NOTE: All unused input pins must be grounded.

APPLICATIONS INFORMATION

Connection Diagram, Unipolar Operation



Connection Diagram, Bipolar Operation



Connection Diagram, Bipolar Operation (for applications where bipolar offset temperature drift (≈ 10 ppm/°C) is not critical)

NOTE: To maintain specified linearity, external amplifiers must be zeroed. This is best done with V_{REF} set to zero and,

Unipolar: load the DAC register with all bits at zero and adjust R_{OS} for V_{OUT} = OV

Bipolar: load the DAC register with 10 . . .0 (MSB =1) and set R_{OS2} for $V_{OUT\,1}$ = OV. Then set R_{OS1} for

TRANSFER FUNCTION (N=12)

BINARY INPUT	UNIPOLAR OUTPUT	BIPOLAR OUTPUT
111 . . . 111	$-V_{REF}\,(1 - 2^{-N})$	$-V_{REF}\,(1 - 2^{-(N-1)})$
100 . . . 001	$-V_{REF}\,(1/2 + 2^{-N})$	$-V_{REF}\,(2^{-(N-1)})$
100 . . . 000	$\frac{-V_{REF}}{2}$	0
011 . . . 111	$-V_{REF}\,(1/2 - 2^{-N})$	$V_{REF}\,(2^{-(N-1)})$
000 . . . 000	0	V_{REF}

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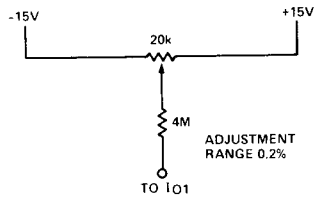
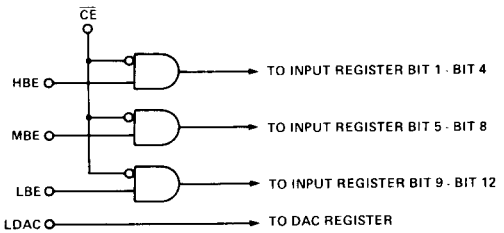
BIPOLAR OFFSET ADJUST (external)



NOTE: External opamps have to be zeroed before the bipolar offset adjust circuit is connected.

CONTROL LOGIC

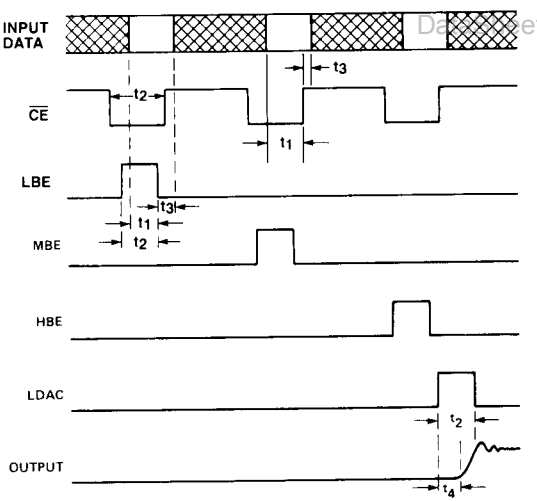


NOTE: The transfer from input register to DAC register can be performed without Enabling Chip.

STROBE LOGIC

Strobe	Function
0	data latched (held)
1	data changing (transfer)

TIMING DIAGRAM



TIME AXIS NOT TO SCALE. ALL STROBES ARE LEVEL TRIGGERED.

t_1: Data Setup Time. Time data must be stable before strobe (byte enable/LDAC) goes to "0", t_1 (min) = 250 nsec.

t_2: Strobe Width. t_2 (min) = 250 nsec. (CE, LBE, MBE, HBE, LDAC).

t_3: Hold Time. Time data must be stable after strobe goes to "0", t_3 = 0 nsec.

t_4: Delay from LDAC to Output. t_4 = 200 nsec.

NOTE: Minimum common active time for $\overline{CE}$ and any byte enable is 250 nsec.

ORDERING INFORMATION

MODEL	DESCRIPTION
HS 3120C-0	Double Buffered 12-Bit MDAC, Commercial
HS 3120C-2	Double Buffered 12-Bit MDAC, Commercial
HS 3120B-0	Double Buffered 12-Bit MDAC, MIL-STD-883C
HS 3120B-2	Double Buffered 12-Bit MDAC, MIL-STD-883C

CAUTION: ESD (Electro-Static Discharge) sensitive device. Permanent damage may occur when unconnected devices are subjected to high energy electrostatic fields. Unused devices must be stored in conductive foam or shunts. Protective foam should be discharged to the destination socket before devices are removed. Devices should be handled at static safe workstations only. Unused digital inputs must be grounded or tied to the logic supply voltage. Unless otherwise noted, the supply voltage at any digital input should never exceed the supply voltage by more than 0.5 volts or go below -0.5 volts. If this condition cannot be maintained, limit input current on digital inputs by using series resistors or contact Hybrid Systems for technical assistance.

Specifications subject to change without notice.

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