



**REVISION HISTORY**

<b>Version</b>	<b>Date</b>	<b>Descriptions</b>
V1.0	July 25, 2018	Initial version
V1.1	August 20, 2018	Add USB PD3.0 certification information. Add 5V PDO voltage sense accuracy data. Remove SSOP-10 package information.
V1.2	August 19, 2019	Update the voltage sense accuracy in Table 1. Update the ordering guide Update package top marking
V1.3	September 14, 2019	Add SSOP-10L package
V1.4	October 31, 2019	Update Fig. 1 and Fig. 5
V1.5	June 20, 2020	Update Fig. 7

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## SPECIFICATIONS

$V_{IN} = 5V$ ,  $T_A = 25^{\circ}C$ , unless otherwise noted.

Table 1.

Parameter	Symbol	Test Conditions	Min	Typ	Max	Unit
<b>GENERAL PARAMETERS</b>						
Supply Voltage	$V_{IN}$		3.3		20	V
Supply Voltage UVLO Threshold	$V_{IN\_UVLO}$	Rising		2.85		V
		Falling		2.75		V
Supply Current at Normal Operating	$I_{CC\_OPR}$			1		mA
Supply Current at Sleep Mode	$I_{CC\_SLEEP}$	GATE pin is not pulled low. CC1 and CC2 floating		250		$\mu A$
Regulator Voltage	$V_{DD}$			1.8		V
Operating Junction Temperature	$T_J$		-40		125	$^{\circ}C$
Operating Ambient Temperature	$T_A$		-40		105	$^{\circ}C$
<b>Type-C</b>						
Default Mode Pullup Current Source	$I_{CC\_DEF}$		64	80	96	$\mu A$
1.5A Mode Pullup Current Source	$I_{CC\_1P5}$		166	180	194	$\mu A$
3.0A Mode Pullup Current Source	$I_{CC\_3P0}$		304	330	356	$\mu A$
UFP Detecting threshold at Default Current	$V_{TH\_DEF}$		1.51	1.6	1.64	V
UFP Detecting threshold at 1.5A current	$V_{TH\_1A5}$		1.51	1.6	1.64	V
UFP Detecting threshold at 3.0A current	$V_{TH\_3A0}$		2.46	2.6	2.74	V
<b>BMC COMMON PARAMETERS</b>						
Bit Rate	$f_{BitRate}$		270	300	330	Kbps
<b>BMC TX PARAMETERS</b>						
Maximum Difference between the Bit-rate during the Part of the Packet Following the Preamble and the Reference Bit-rate.	$P_{BitRate}$				0.25	%
Time to Cease Driving the Line after the End of the Last bit of the Frame.	$t_{EndDriveBMC}$				23	$\mu s$
Fall Time	$t_{Fall}$		300			ns
Time to cease driving the line after the final high-to-low transition.	$t_{HoldLowBMC}$		1			$\mu s$
Time from the End of Last Bit of a Frame until the Start of the First bit of the Next Preamble.	$t_{InterFrameGap}$		25			$\mu s$
Rise Time	$t_{Rise}$		300			ns
Time Before the Start of the First Bit of the Preamble when the Transmitter shall Start Driving the Line.	$t_{StartDrive}$		-1		1	$\mu s$
Voltage Swing	$V_{Swing}$		1.05	1.125	1.2	V
Transmit Low Voltage			-75		75	mV
Transmitter Output Impedance	$Z_{Driver}$		33	54	75	$\Omega$
<b>BMC RX PARAMETERS</b>						
Hysteresis				160		mV
Time Window for Detecting Bus Non-idle	$t_{TransitionWindow}$		12		20	$\mu s$
Number to Count to Detect Bus Non-idle	$n_{Count}$		3			
Time Constant of a Single Pole Filter to Limit Broad-band Noise Ingression <sup>1</sup>	$t_{RxFilter}$		100			ns
Receiver Input Impedance	$Z_{BmcRx}$		1			M $\Omega$
<b>D+ AND D- PINS</b>						
Output Voltage Selection Reference	$V_{SEL\_REF}$		1.8	2.0	2.2	V
Data Detect Voltage Reference	$V_{DAT\_REF}$		0.25	0.325	0.4	V
DatLine Leakage Resistance	$R_{DAT\_LKG}$		300	-	1500	k $\Omega$
D- Pulldown Resistance during HVDCP Mode	$R_{DM\_DWM}$		14.25	19	24.5	k $\Omega$
D+ to D- Resistance During DCP mode	$R_{DCP\_DAT}$			100	200	$\Omega$
D+ High Glitch Filter Time	$T_{GLITCH\_BC\_DONE}$		1000	1250	1500	ms
D- Low Glitch Filter Time	$T_{GLITCH\_DM\_LOW}$		1	2		ms
Output Voltage Glitch Filter Time	$T_{GLITCH\_V\_CHANGE}$		20	40	60	ms
Glitch Filter for D+/- Continuous Change	$T_{GLITCH\_CONT\_CHANGE}$	Continuous Mode	100	150	200	us

VOLTAGE CONTROL(VFB PIN) Voltage Sense Scaling Factor Voltage Sense Accuracy Time from Source issue GoodCRC to Start Voltage Transition	$t_{SrcTransition}$	5V with no load	5.1	10 30	5.25	V ms
CURRENT CONTROL (CS+, CS-, IFB PINS) Current Sense Resistor				5		m $\Omega$
GATE PIN Maximum Sinking Current Pull Low Impedance			2	50	20 150	mA $\Omega$
OPTO PIN Min OPTO Current Max Pull Down Current				30 3		$\mu$ A mA
OV AND OC PROTECTIONS Over-voltage Protection Threshold	$V_{IN\_OV}$	With respect to $V_{IN\_REF}$	115	120	125	%
Under-voltage Protection Threshold	$V_{IN\_UV}$	With respect to $V_{IN\_REF}$	75	80	85	%
Over-current Protection Threshold	$I_{IN\_OC}$	With respect to $I_{IN\_REF}$	110	125	140	%
Thermal Shutdown Rising	$t_{TSD\_RISE}$			130		$^{\circ}$ C
Thermal Shutdown Falling	$t_{TSD\_FALL}$			80		$^{\circ}$ C

## ABSOLUTE MAXIMUM RATING

**Table 2.**

Parameter	Rating
VIN, GATE, VBUS, OPTO	-0.5V to +20V
CC1, CC2	-0.5V to +20V
VDD, D+, D-, CS+, CS-, VFB, IFB	-0.5V to +6V
Storage Temperature Range	-65°C to +150°C
Operating Junction Temperature Range	-40°C to +125°C
ESD HBM (Human Body Model)	±5kV
ESD MM (Machine Model)	500V
Soldering Conditions	JEDEC J-STD-020

## THERMAL RESISTANCE

$\theta_{JA}$  is specified for the worst-case conditions, that is, a device soldered in a circuit board for surface-mount packages.

**Table3. Thermal Resistance**

Package Type	$\theta_{JA}$	$\theta_{JC}$	Unit
SOP-14	83.5	37.7	°C/W
QFN-16	47	4.5	°C/W
SSOP-10L	88	37	°C/W

## ESD CAUTION



**ESD (electrostatic discharge) sensitive device.** Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

PIN CONFIGURATION AND FUNCTION DESCRIPTION

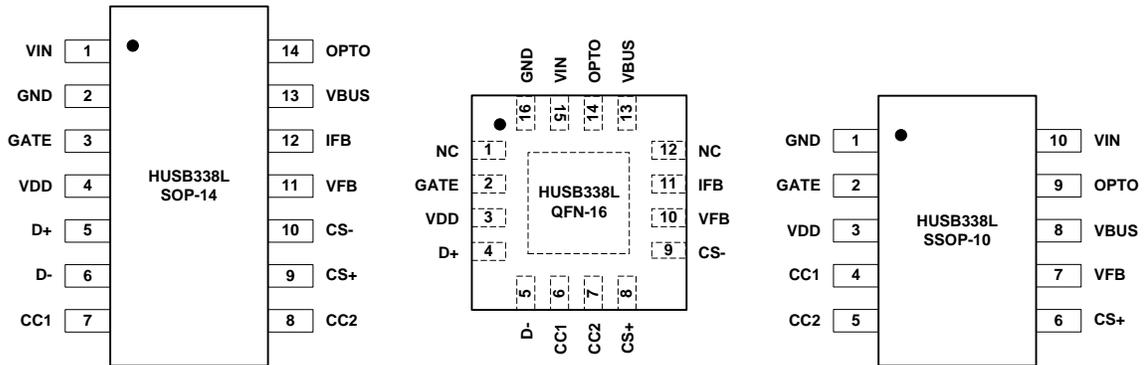


Figure 2. Pin Configuration, View from Top.

Table 4. HUSB338L Pin Function Description

SOP-14 Pin No.	QFN-16 Pin No.	SSOP-10 Pin No.	Pin Name	Pin Type	Voltage Type	Pin Description
1	15	10	VIN	P	HV	Supply input voltage. Connect this pin to GND via the recommended ceramic capacitor.
2	16	1	GND	P	-	Power ground.
-	1	-	NC	-	-	Not connection.
3	2	2	GATE	OD	HV	Open drain gate drive output.
4	3	3	VDD	P	LV	1.8 V regulator output for system power.
5	4	-	D+	DIO	LV	USB D+ line.
6	5	-	D-	DIO	LV	USB D- line.
7	6	4	CC1	AIO	HV	Type-C CC1 line.
8	7	5	CC2	AIO	HV	Type-C CC2 line.
9	8	6	CS+	AI	LV	Positive input of the current sense amplifier.
10	9	-	CS-	AI	LV	Negative input of the current sense amplifier. Provide a low ohmic connection to GND.
11	10	7	VFB	AI	LV	Voltage loop feedback.
12	11	-	IFB	AI	LV	Current loop feedback.
-	12	-	NC	-	-	Not connection.
13	13	8	VBUS	AI	HV	VBUS sense and discharge sink.
14	14	9	OPTO	AI	HV	OPTO driver.

Legend:

HV=High Voltage Pin (Max 20V)

LV=Low Voltage Pin (Max 6V)

OD=Open Drain Pin

A=Analog Pin

P= Power Pin

D=Digital Pin

I=Input Pin

O=Output Pin

## THEORY OF OPERATION

### VIN AND VDD PINS

#### *VIN Power System*

VIN pin is the power supply input, which is derived from the output of the AC-DC or DC-DC converter. Connect a 1 $\mu$ F decoupling MLCC between VIN pin and GND pin, as closer so possible.

#### *VIN Discharge*

The VIN pin is also connected to an internal MOSFET and discharging resistor, which is used as a bleeder to help discharge the output capacitor to vSafe5V upon the detachment of a connected device, or to a lower desired output voltage level upon a UFP request, such as from 15V to 5V.

#### *VIN Voltage Sense*

See the Control Loop Compensation Circuit (VFB, CS+, CS-, IFB, OPTO Pins) section.

#### *VDD System Supply*

An internal liner regulator is used to provide 1.8V system voltage. Connect a 1 $\mu$ F MLCC to VDD pin for decoupling.

### CONTROL LOOP COMPENSATION CIRCUIT (VFB, CS+, CS-, IFB, OPTO PINS)

In the HUSB338L, the constant voltage loop compensation (CV loop) and constant current loop (CC loop) compensation are implemented. The output of the compensator is used to drive the primary side of the opto-coupler and control the AC-DC power loop.

#### *Constant Voltage Loop Compensation Circuit (CV Loop)*

The input of CV loop is connected to VFB pin, which is derived from the internal 90k $\Omega$  and 10k $\Omega$  voltage divider. The voltage divider sense the VIN voltage. Therefore, the VFB voltage is 10% of VIN voltage. The CV loop compensator is implemented by the resistor and capacitor network between OPTO pin and VFB pin.

#### *Constant Current Loop Compensation Circuit (CC Loop)*

In HUSB338L, the CC loop input is derived from a 5m $\Omega$  current sense resistor. The current signal is filtered by an RC network and fed to the CS+ and CS- differential inputs. The CC loop compensator is implemented by the resistor and capacitor network between OPTO pin and IFB pin.

#### *Cable Voltage Drop Compensation (IR Compensation)*

The cable voltage drop compensation is implemented in HUSB338L. Once it is enabled, the output voltage increases with 100mV/A according to the load current. For example, for the 5V 3A condition, the actual output voltage is:

$$5V + 3A * 100mV/A = 5.3V.$$

### CC1 AND CC2 PINS

#### *Type-C CC Function*

CC1 and CC2 are the Configuration Channel pins used for connection and attachment detection, plug orientation determination and system configuration management across USB Type-C cable. CC1 and CC2 only support DFP mode with 500mA, 1.5A and 3A current advertising. The CC1 and CC2 supports 20V high voltage. This is used for protection when the CC1 or CC2 is shorted to the VBUS pin.

#### *BCM Driver*

Through the Type-C detection, one of CC1 or CC2 pins will be connect to the internal BMC block to achieve PD communication.

### VBUS PIN

This pin is used to sense VBUS presence, monitor VBUS voltage and discharge VBUS on USB Type-C receptacle side. Connect directly to VBUS at the Type-C Receptacle.

#### *vSafe0V Detection*

When the HUSB338 enter AttachWait.SRC for t<sub>CCDebounce</sub>, it detects whether the VBUS voltage is within vSafe0V. If yes, HUSB338 pulls low GATE pin and enters Attached.SRC state. If no, it will stay at AttachWait.SRC state.

**VBUS Discharge**

The VBUS pin is also connected to an internal MOSFET and discharging resistor, which is used as a bleeder to help discharge the output capacitor to vSafe0V upon the detachment of a connected device, or to a lower desired output voltage level upon a UFP request, such as from 15V to 5V.

**GATE PIN**

GATE pin is open-drain output allows to drive directly a PMOS load switch. When HUSB338L enters Type-C attached state (Attached.SRC), GATE pin is pulled low. When HUSB338L exits from Type-C attached state, the GATE pin is removed from being pulled low.

**OVER-VOLTAGE PROTECTION (VIN\_OV)**

The HUSB338L detects the VIN pin voltage to achieve over-voltage protection function. The threshold to trigger over-voltage protection is 120% of the  $V_{IN\_REF}$ . When the over-voltage condition occurs, the HUSB338L stops to pull the GATE pin low. When the over-voltage condition is removed, HUSB338L is reset to standby mode and will automatic recover again.

**UNDER-VOLTAGE PROTECTION (VIN\_UV)**

The HUSB338L detects the VIN pin voltage to achieve under-voltage protection function. The threshold to trigger under-voltage protection is 80% of the  $V_{IN\_REF}$ . When the under-voltage condition occurs, the HUSB338L stops to pull the GATE pin low. When the over-voltage condition is removed, HUSB338L is reset to standby mode and will automatic recover again.

**OVER-TEMPERATURE PROTECTION (OT)**

When the junction temperature rises across 130°C, over-temperature protection takes action and the load switch is turned off. When the junction temperature falls across 80°C, HUSB338L is reset to standby mode and will automatic recover again.

**OVER-CURRENT PROTECTION (IIN\_OC)**

When the current sensed by the sense resistor exceeds the 125% of  $I_{IN\_REF}$ , the over-current protection takes action and the load switch is turned off. When the over-current condition is removed, the HUSB338L is reset to standby mode and will automatic recover again.

**D+ AND D- PINS**

The HUSB338L with SOP-14L package and QFN-16L package supports BC1.2 DCP and HVDCP fast charger protocols. The HVDCP hardware diagram is shown in Figure 3. The voltage scaling truth cable is shown is Table 5.

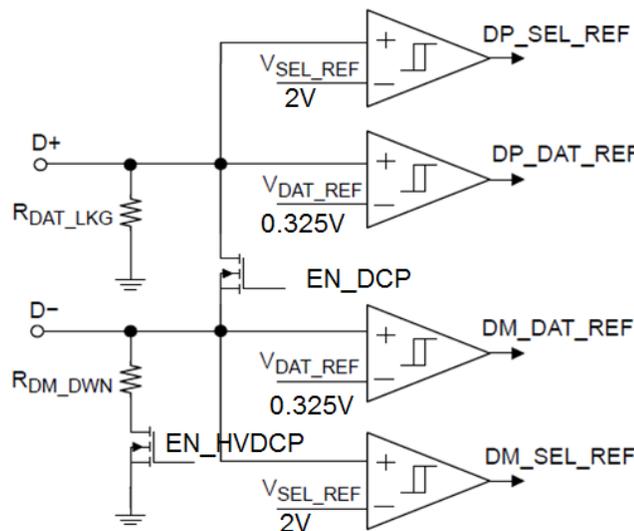


Figure 3. HVDCP Hardware Implementation

Table 5. HVDCP Voltage Scaling Truth Table

Portable Device		Class A HVDCP
D+	D-	Adapter Output Voltage

---

0.6V	0.6V	12V
3.3 V	0.6 V	9V
0.6 V	3.3 V	Continuous Mode
0.6 V	GND	5V

TYPICAL APPLICATIONS CIRCUITS

Figure 4 to Figure 5 provide some application circuits where the HUSB338L drives an optocoupler.

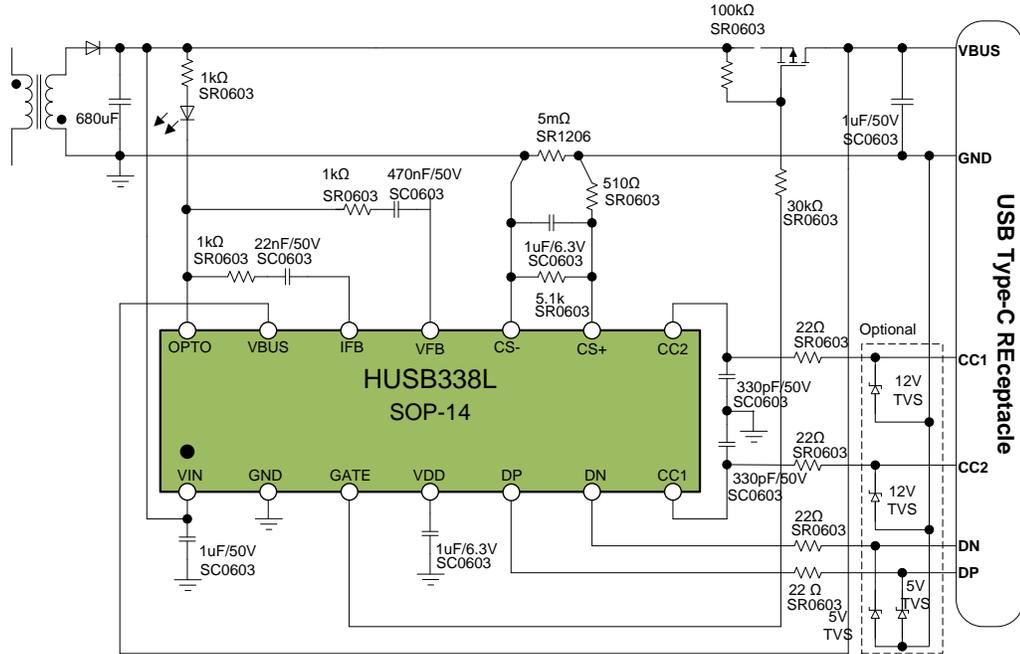


Figure 4. HUSB338L SOP-14L Package Application Circuit

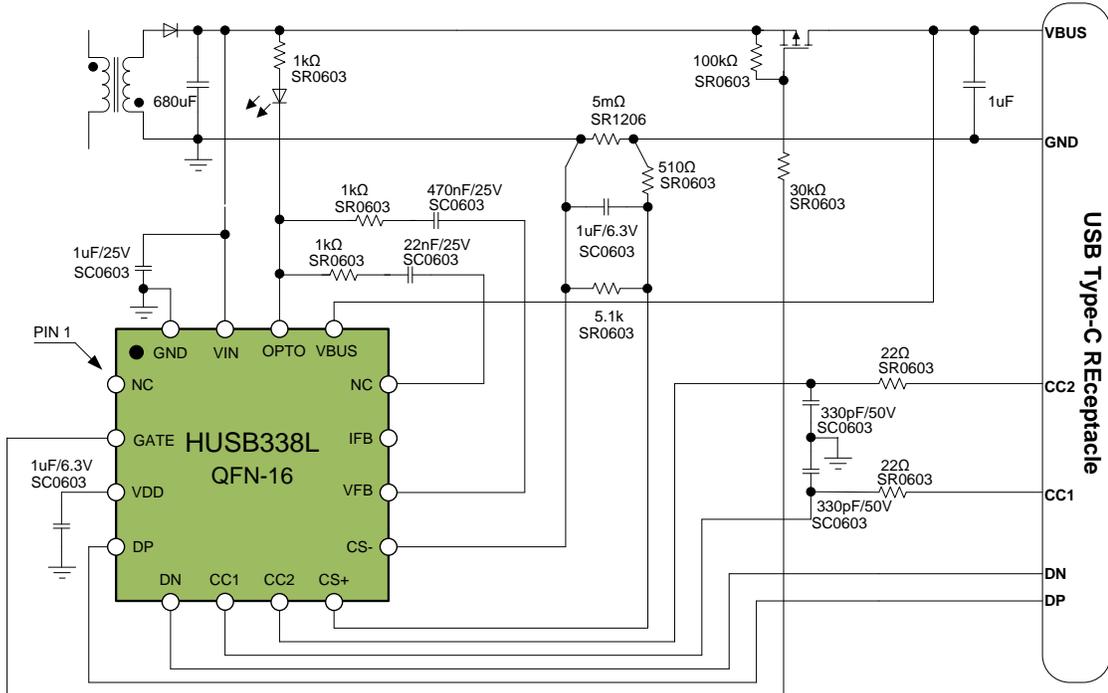
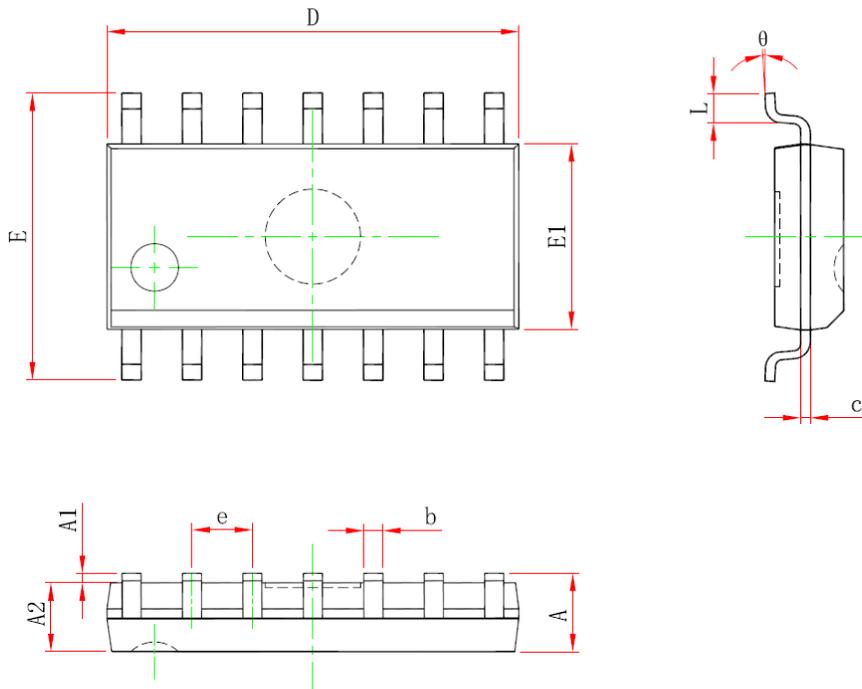


Figure 5. HUSB338L QFN-16L Package Application Circuit

PACKAGE OUTLINE DEMENSIONS

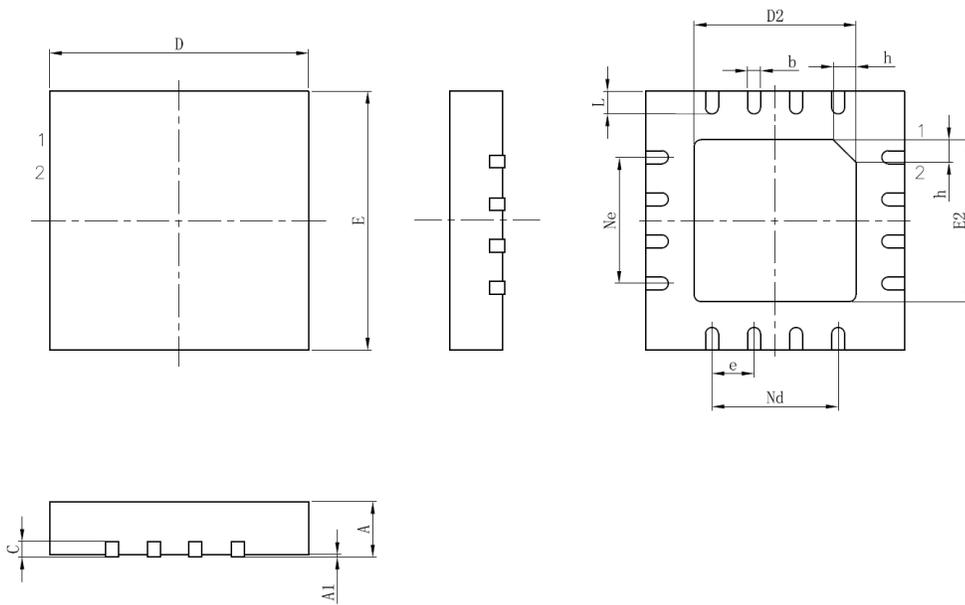
SOP-14L PACKAGE



Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min	Max	Min	Max
A	--	1.750	--	0.069
A1	0.100	0.250	0.004	0.010
A2	1.250	--	0.049	--
b	0.310	0.510	0.012	0.020
c	0.100	0.250	0.004	0.010
D	8.450	8.850	0.333	0.348
E	5.800	6.200	0.228	0.244
E1	3.800	4.000	0.150	0.157
e	1.270(BSC)		0.050(BSC)	
L	0.400	1.270	0.016	0.050
theta	0°	8°	0°	8°

Figure 6. SOP-14L Package, 8.65 mm x 6 mm

QFN-16L PACKAGE

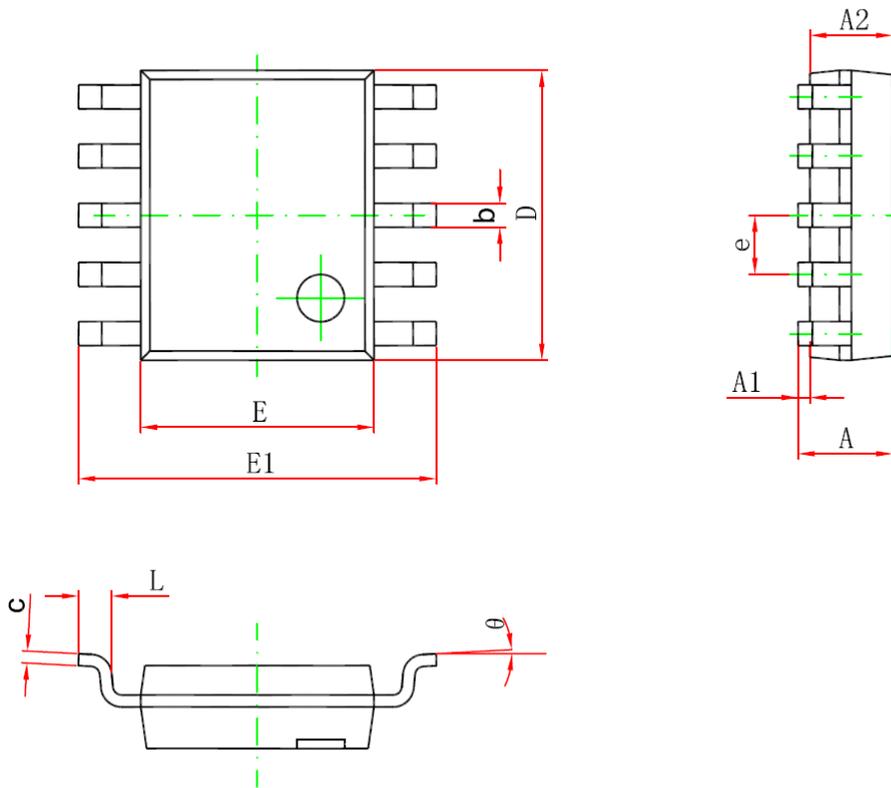


SYMBOL	MILLIMETER		
	MIN	NOM	MAX
A	0.70	0.75	0.80
A1	—	0.02	0.05
b	0.25	0.30	0.35
c	0.18	0.20	0.25
D	3.90	4.00	4.10
D2	2.00	2.15	2.30
e	0.650BSC		
Ne	1.95BSC		
Nd	1.95BSC		
E	3.90	4.00	4.10
E2	2.00	2.15	2.30
L	0.45	0.55	0.65
h	0.30	0.35	0.40

Figure 7. QFN-16L Package, 4 mm x 4 mm

SSOP-10L PACKAGE

SSOP-10L PACKAGE OUTLINE DIMENSIONS



Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min	Max	Min	Max
A	1.350	1.750	0.053	0.069
A1	0.100	0.250	0.004	0.010
A2	1.350	1.550	0.053	0.061
b	0.300	0.450	0.012	0.018
c	0.170	0.250	0.007	0.010
D	4.700	5.100	0.185	0.201
E	3.800	4.000	0.150	0.157
E1	5.800	6.200	0.228	0.244
e	1.000 (BSC)		0.039 (BSC)	
L	0.400	1.270	0.016	0.050
θ	0°	8°	1°	8°

PACKAGE TOP MARKING

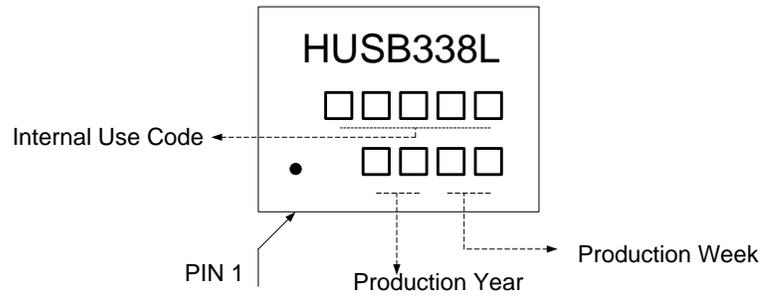


Figure 8. Package Top Marking

ORDERING GUIDE

Package	Model	Power	FPDO	APDO	QC2.0/3.0	Package Option
SOP-14L	See HUSB338L model list	Configurable	Configurable	Configurable	Configurable	Tape & Reel, 4k
QFN-16L	See HUSB338L model list	Configurable	Configurable	Configurable	Configurable	Tape & Reel, 5k
SSOP-10L	See HUSB338L model list	Configurable	Configurable	Configurable	Configurable	Tape & Reel, 4k

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