KMA220

Dual channel programmable angle sensor

Rev. 1 — 24 May 2012

Product data sheet

Product profile 1.

1.1 General description

The KMA220 is a dual channel magnetic angle sensor module. The MagnetoResistive (MR) sensor bridges, the mixed signal Integrated Circuits (ICs) and the required capacitors are integrated into a single package. Both integrated channels are working full independently. The supply voltage has common terminals. Further explanations are referring to a single channel of the KMA220.

This angular measurement module KMA220 is pre-programmed, pre-calibrated and therefore, ready to use.

The KMA220 allows user-specific adjustments of angular range, zero angle and clamping voltages for each channel separately. The settings are stored permanently in a non-volatile memory.

1.2 Features and benefits

- High precision sensor for magnetic angular measurement
- Single package dual channel sensor module with integrated filters for improved ElectroMagnetic Compatibility (EMC)
- Automotive qualified in accordance with
 Overvoltage protection up to 16 V AEC-Q100 Rev-G
- Programmable user adjustments, including zero angle and angular range
- Fail-safe non-volatile memory with write Each channel includes protection using lock bit
- Independent from magnetic field strength above 35 kA/m
- Ready to use without external components
- One common supply line

- High temperature range up to 160 °C
- Dual electric independent sensor channels with analog ratiometric output voltages
- Independent programming via separated One-Wire Interfaces (OWI)
- user-programmable 32-bit identifier
- Magnet-loss and power-loss detection
- Factory calibrated

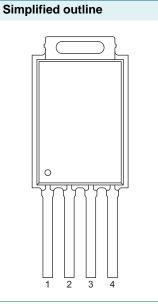


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2. Pinning information

Table 1. Pinning

Pin	Symbol	Description	5
1	OUT1/DATA1	analog output 1 or data interface 1	
2	GND	ground	
3	V_{DD}	supply voltage	
4	OUT2/DATA2	analog output 2 or data interface 2	

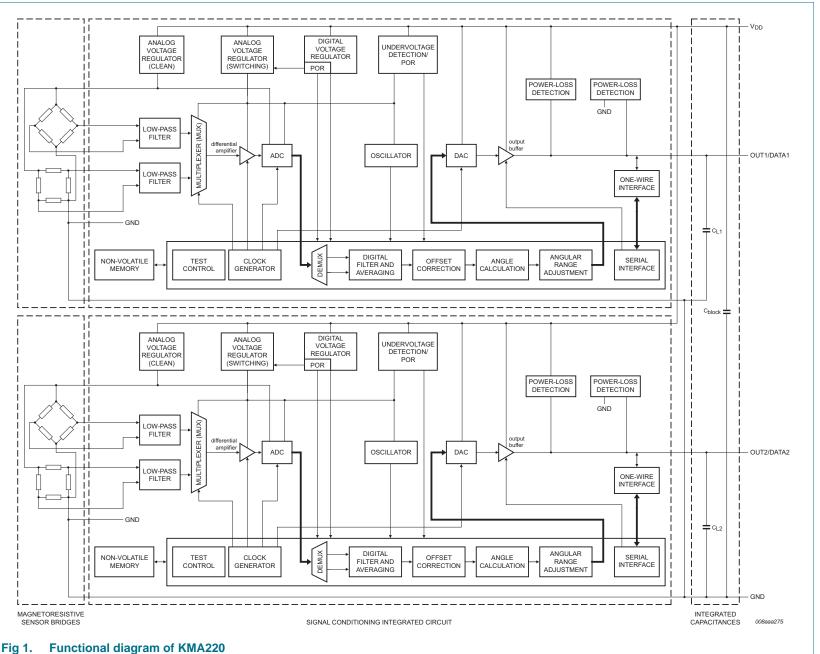


3. Ordering information

Table 2. Ordering information

Type number	Package	Package					
	Name	Description	Version				
KMA220	SIL4	plastic, single in-line package	SOT1188-1				

Functional diagram



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5. Functional description

Each channel of the KMA220 amplifies two orthogonal differential signals from MR sensor bridges and converts them into the digital domain. The angle is calculated using the COordinate Rotation DIgital Computer (CORDIC) algorithm. After a digital-to-analog conversion, the analog signal is provided to the output as a linear representation of the angular value. Zero angle, clamping voltages and angular range are programmable. In addition, two 16-bit registers are available for customer purposes, such as sample identification.

Each channel of the KMA220 comprises a Cyclic Redundancy Check (CRC) and an Error Detection and Correction (EDC), as well as magnet-loss to ensure a fail-safe operation. If either the supply voltage or the ground line of the mixed signal IC is interrupted, a power-loss detection circuit pulls the analog output to the remaining connection.

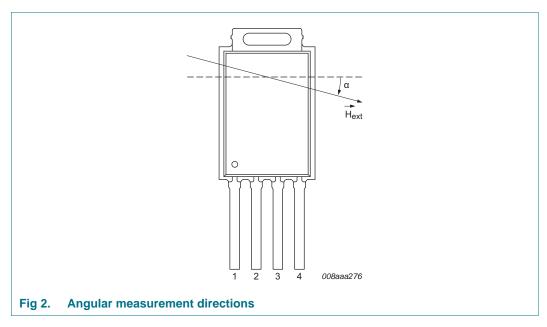
After multiplexing the two MR Wheatstone bridge signals and their successive amplification, the signal is converted into the digital domain by an Analog-to-Digital Converter (ADC). Further processing is done within an on-chip state machine. This state machine controls offset cancelation, calculation of the mechanical angle using the CORDIC algorithm, as well as zero angle and angular range adjustment. The internal Digital-to-Analog Converter (DAC) and the analog output stage are used for conversion of the angle information into an analog output voltage, which is ratiometric to the supply voltage.

The configuration parameters of each channel are stored independently in a user-programmable non-volatile memory. The OWI (accessible using pin OUTn/DATAn) is used for accessing the memory. In order to protect the memory content a lock bit can be set. After locking the non-volatile memory, its content cannot be changed anymore.

5.1 Angular measurement directions

The differential signals of the MR sensor bridges depend only on the direction of the external magnetic field strength H_{ext} , which is applied parallel to the plane of the sensor. In order to obtain a correct output signal, exceed the minimum saturation field strength.

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Since the Anisotropic MR (AMR) effect is periodic over 180°, the sensor output is also 180°-periodic. The angle is calculated relative to a freely programmable zero angle. The dashed line indicates the mechanical zero degree position.

6. Analog output

The KMA220 provides two analog output signals on pin OUT1/DATA1 and on pin OUT2/DATA2. The measured angle α is converted linearly into a value, which is ratiometric to the supply voltage $V_{DD}.$ Either a positive or a negative slope can be programmed independently for each pin by the customer for this purpose.

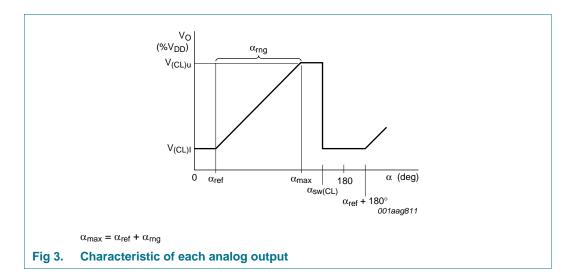
<u>Table 3</u> describes the analog output behavior for a positive slope. If for example, a magnetic field angle, above the programmed maximum angle α_{max} , but below the clamp switch angle $\alpha_{\text{sw}(CL)}$ is applied to the sensor, then analog output is set to the upper clamping voltage. If the magnetic field angle is larger than the clamp switch angle, the analog output switches from upper to lower clamping voltage. If there is a negative slope, the clamping voltages are changed.

Table 3. Analog output behavior for a positive slope

Magnetic field angle	Analog output
$\alpha_{\text{max}} < \alpha < \alpha_{\text{sw(CL)}}$	$V_{(CL)u}$
$\alpha_{\rm sw(CL)} < \alpha < \alpha_{\rm ref}$ + 180°	V _{(CL)I}

The analog output voltage range encodes both angular and diagnostic information. A valid angle value is between the upper and lower clamping voltage. If the analog output is in the diagnostic range, that is below 4 $\rm \%V_{DD}$ or above 96 $\rm \%V_{DD}$, an error condition has been detected. The analog output repeats every 180°.

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7. Diagnostic features

Each channel provides several diagnostic features:

7.1 CRC and EDC supervision

Each channel of the KMA220 includes a supervision of the programmed data. At power-on, a CRC of the non-volatile memory is performed. Furthermore the memory is protected against bit errors. Every 16-bit data word is saved internally as a 22-bit word for this purpose. The protection logic corrects any single-bit error in a data word, while the sensor continues in normal operation mode. Furthermore the logic detects double-bit error per word and switches the output into diagnostic mode.

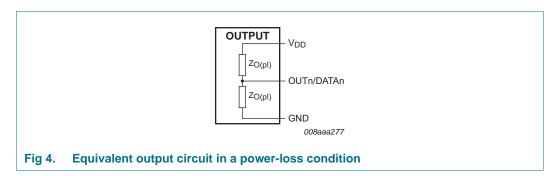
7.2 Magnet-loss detection

If the applied magnetic field strength is not sufficient, each channel of the KMA220 can raise a diagnostic condition. In order to enter the diagnostic mode, due to magnet-loss, enable the detection first. Both channels can be programmed independently into active diagnostic mode, where the output is driven below 4 %V_{DD} or above 96 %V_{DD}.

7.3 Power-loss detection

The power-loss detection circuit enables the detection of an interrupted supply or ground line of the mixed signal IC. If there is a power-loss condition, two internal switches in the sensor are closed, connecting the pin of the analog output to the supply voltage and the ground pins.

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<u>Table 4</u> describes the power-loss behavior and gives the resulting output voltage depending on the interrupted supply or ground line and the load resistance.

Table 4. Power-loss behavior

Load resistance	Interrupted supply line	Interrupted ground line
$R_{L(ext)} > 5 k\Omega$	$V_O \leq 4~\%V_{DD}$	$V_O \ge 96~\%V_{DD}$

7.4 Low supply voltage detection and overvoltage protection

If the supply voltage is below the switch-off threshold voltage, a status bit is set in each signal conditioning integrated circuit and both channels go into diagnostic mode. If the supply voltage is above the overvoltage switch-on threshold voltage, both channel enter diagnostic mode. Table 5 describes the system behavior depending on the voltage range of the supply voltage.

Table 5. System behavior for each output

Supply voltage	State	Description
0 V to \cong 1.8 V	start-up power	The output buffer drives an active LOW or is powered down, but the switches of the power-loss detection circuit are not fully opened and set the output to a level between ground and half the supply voltage.
\cong 1.8 V to V_{POR}	power-on reset	The power-loss charge pump is fully operational and turns the switches of the detection circuit off. The output buffer drives an active LOW and sets the output to the lower diagnostic level. During the reset phase, all circuits are in reset and/or Power-down mode.
V_{POR} to $V_{th(on)}$ or $V_{th(off)}$	initialization	The digital core and the oscillator are active. After reset, the content of the non-volatile memory is copied into the shadow registers. The output buffer drives an active LOW and sets the output to the lower diagnostic level.
$V_{th(on)}$ or $V_{th(off)}$ to minimum V_{DD}	functional operation	All analog circuits are active and the measured angle is available at the analog output. Not all parameters are within the specified limits.
Minimum V_{DD} to maximum V_{DD}	normal operation	All analog circuits are active and the measured angle is available at the analog output. All parameters are within the specified limits.
Maximum V_{DD} to $V_{th(ov)}$	functional operation	All analog circuits are active and the measured angle is available at the analog output. Not all parameters are within the specified limits.
V _{th(ov)} to 16 V	overvoltage	Both digital cores and oscillators are active but all other circuits are in Power-down mode. The outputs are set to the lower diagnostic level.

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<u>Table 6</u> describes the diagnostic behavior and the resulting output voltage depending on the error case. Furthermore the duration and termination condition to enter and leave the diagnostic mode are given, respectively.

Table 6. Diagnostic behavior

Diagnostic condition	Duration	Analog output	Termination condition
Low voltage	1 μs < t < 10 μs	\leq 4 % V_{DD}	functional or normal operation
Overvoltage	1 μs < t < 10 μs	\leq 4 % V_{DD}	functional or normal operation
Checksum error	n/a	$\leq 4~\text{\%V}_{DD}~\text{or} \geq 96~\text{\%V}_{DD} {\color{red} \underline{^{[2]}}}$	power-on reset[1]
Double-bit error	n/a	$\leq 4~\text{\%V}_{DD}~\text{or} \geq 96~\text{\%V}_{DD} {\color{red} \underline{^{[2]}}}$	power-on reset[1]
Magnet-loss	0.5 ms < t < 6 ms	$\leq 4 \text{ %V}_{DD} \text{ or } \geq 96 \text{ %V}_{DD}^{[2]}$	magnet present[1]
Power-loss	≤ 2 ms	$\leq 4~\%V_{DD}~or \geq 96~\%V_{DD}^{\begin{subarray}{c} \begin{subarray}{c} subarray$	power-on reset

^[1] Status bit stays set in command register until power-on reset.

8. Limiting values

Table 7. Limiting values
In accordance with the Absolute Maximum Rating System (IEC 60134).

Symbol	Parameter	Conditions	Min	Max	Unit
V_{DD}	supply voltage		-0.3	+16	V
V _O	output voltage		-0.3	+16	V
V _{O(ov)}	overvoltage output voltage	T_{amb} < 140 °C at t < 1 h	$[1]$ $V_{th(ov)}$	16	V
I _r	reverse current	T _{amb} < 70 °C	-	150	mA
T _{amb}	ambient temperature		-40	+160	°C
$T_{amb(pr)}$	programming ambient temperature		10	70	°C
T_{stg}	storage temperature		-40	+125	°C
Non-volatile	e memory				
t _{ret(D)}	data retention time	T _{amb} = 50 °C	17	-	year
$N_{\text{endu}(W_ER)}$	write or erase endurance	$T_{amb(pr)} = 70 ^{\circ}C$	100	-	cycle

^[1] Overvoltage on analog output and supply within the specified operating voltage range.

^[2] Depending on the diagnostic level setting.

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9. Recommended operating conditions

Table 8. Operating conditions

In a homogenous magnetic field.

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
V_{DD}	supply voltage		<u>[1]</u>	4.5	5.0	5.5	V
T _{amb}	ambient temperature			-40	-	+160	°C
T _{amb(pr)}	programming ambient temperature			10	-	70	°C
C _{L(ext)}	external load capacitance		[1][2]	0	-	22	nF
			[2][3]	0	-	6.8	nF
R _{L(ext)}	external load resistance		[4]	5	-	∞	kΩ
H _{ext}	external magnetic field strength			35	-	-	kA/m

^[1] Normal operation mode.

10. Thermal characteristics

Table 9. Thermal characteristics

Symbol	Parameter	Conditions	Тур	Unit
$R_{th(j-a)}$	thermal resistance from junction to ambient		100	K/W

11. Characteristics

Table 10. Mechanical characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
F _{lead}	mechanical force to the leads	T_{amb} = 25 °C	-	-	10	N
F _{fin}	mechanical force to the fin holder	T _{amb} = 25 °C	-	-	15	N

Table 11. Supply current

Characteristics are valid for the operating conditions, as specified in Section 9.

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
I_{DD}	supply current	[1]	10	-	21	mΑ
		[3][4] _	-	26	mA
$I_{\text{off(ov)}}$	overvoltage switch-off current		[5] _	-	12	mA

^[1] Normal operation and diagnostic mode excluding overvoltage and undervoltage within the specified operating supply voltage range.

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^[2] Between ground and analog outputs.

^[3] Command mode.

^[4] Power-loss detection is only possible with a load resistance within the specified range connected to the supply or ground line.

^[2] Without load current at the analog output.

^[3] Normal operation and diagnostic mode over full voltage range up to limiting supply voltage at steady state.

^[4] With minimum load resistance at the analog outputs.

^[5] Diagnostic mode for a supply voltage above the overvoltage threshold voltage up to the limiting supply voltage.

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Table 12. Power-on resetCharacteristics are valid for the operating conditions, as specified in <u>Section 9</u>.

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
$V_{th(on)} \\$	switch-on threshold voltage	analog output switches on, if $V_{DD} > V_{th(on)}$	-	4.30	4.45	V
$V_{\text{th(off)}}$	switch-off threshold voltage	analog output switches off, if $V_{DD} < V_{th(off)}$	3.90	4.10	-	V
V_{hys}	hysteresis voltage	$V_{hys} = V_{th(on)} - V_{th(off)}$	0.1	0.2	-	V
V_{POR}	power-on reset voltage	IC is initialized	-	3.3	3.6	V
$V_{\text{th(ov)}}$	overvoltage threshold voltage	analog output switches off, if $V_{DD} > V_{th(ov)}$	6.5	7.5	8.0	V
V _{hys(ov)}	overvoltage hysteresis voltage		0.1	0.3	-	V

Table 13. Module performance

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
α_{res}	angle resolution		<u>[1]</u>	-	-	0.04	deg
α_{max}	maximum angle	programmable angular range for $V_{(CL)u} - V_{(CL)I} \ge 80~\%V_{DD}$	[2]	5	-	180	deg
α_{ref}	reference angle	programmable zero angle	[2]	0	-	180	deg
$V_{O(nom)}$	nominal output voltage	at full supply operating range		5	-	95	$%V_{DD}$
$V_{O(udr)}$	upper diagnostic range output voltage		[3][4][5]	96	-	100	%V _{DD}
$V_{O(Idr)}$	lower diagnostic range output voltage		[3][4][5]	0	-	4	$%V_{DD}$
$V_{(CL)u}$	upper clamping voltage		[4][5][6]	40	-	95	$%V_{DD}$
$V_{(CL)I}$	lower clamping voltage		[4][5][6]	5	-	30.5	$%V_{DD}$
$\Delta V_{(CL)}$	clamping voltage variation	deviation from programmed value	[4][5]	-0.3	-	+0.3	$%V_{DD}$
$V_{n(o)(RMS)}$	RMS output noise voltage	equivalent power noise	[1][4]	-	0.4	2.5	mV
$\Delta \phi_{lin}$	linearity error	temperature range –40 °C to +160 °C	[4][7][8]	-1.2	-	+1.2	deg
		temperature range -40 °C to +140 °C	[4][7][8]	-1	-	+1	deg
$\Delta \phi_{temp}$	temperature drift error	temperature range -40 °C to +160 °C	[1][4][7] [9]	-	-	0.8	deg
		temperature range -40 °C to +140 °C	[1][4][7] [9]	-	-	0.65	deg
$\Delta \phi_{temp} _{RT}$	temperature drift error at room temperature	temperature range -40 °C to +160 °C	[7][9] [10]	-	-	0.65	deg
		temperature range -40 °C to +140 °C	[7][9] [10]	-	-	0.55	deg
$\Delta \phi_{hys}$	hysteresis error	referred to input	[4][7]	-	-	0.09	deg
$\Delta \phi_{\mu lin}$	microlinearity error	referred to input	[4][7]	-0.1	-	+0.1	deg

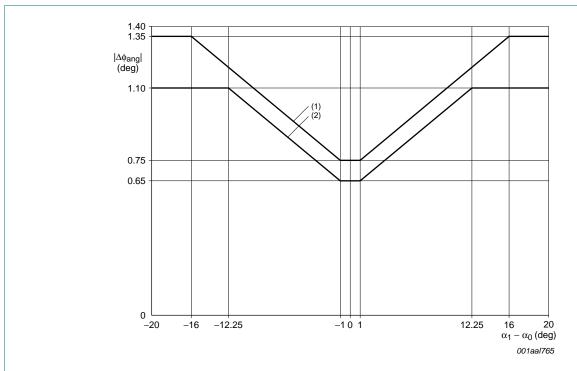
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Table 13. Module performance ...continued

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
$\Delta \phi_{ ext{ang}}$	angular error	temperature range -40 °C to +160 °C	[4][7][8] -1.35 [11]	-	+1.35	deg
		temperature range -40 °C to +140 °C	[4][7][8] -1.1 [11]	-	+1.1	deg
m _{ang}	slope of angular error		[4][7] - [11]	-	0.04	deg/deg
$Z_{O(pl)}$	power-loss output impedance	impedance to remaining supply line in case of lost supply voltage or lost ground	-	-	210	Ω

- [1] At a nominal output voltage between 5 %V_{DD} and 95 %V_{DD} and a maximum angle of α_{max} = 180°.
- [2] In steps of resolution < 0.022°.
- [3] Activation is dependent on the programmed diagnostic mode.
- [4] At a low-pass filtered analog output with a cut-off frequency of 0.7 kHz.
- [5] Settling to these values is limited by 0.7 kHz low-pass filtering of analog output.
- [6] In steps of 0.02 %V_{DD}.
- [7] Definition of errors is given in Section 12.
- [8] Inhomogeneity of a 15 mm diameter disc magnet can increase the linearity error by < 0.1°.
- [9] Based on a 3σ standard deviation.
- [10] Room temperature is given for an ambient temperature of 25 $^{\circ}$ C.
- [11] Graph of angular error is shown in Figure 5.

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- (1) -40 °C to +160 °C
- (2) -40 °C to +140 °C

Fig 5. Envelope curve for the magnitude of angular error

Table 14. Dynamics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
ton	turn-on time	until first valid result	-	-	5	ms
f _{upd}	update frequency		2.4	3.125	-	kHz
ts	settling time	after an ideal mechanical angle step of 45°, until 90 % of the final value is reached	-	-	1.8	ms
t _{cmd(ent)}	enter command mode time	after power-on	20	-	30	ms
t _{rec(ov)}	overvoltage recovery time	after overvoltage	-	-	4	ms

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Table 15. Digital interface

Characteristics are valid for the operating conditions, as specified in Section 9.

				_		
Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V_{IH}	HIGH-level input voltage		80	-	-	$%V_{DD}$
V_{IL}	LOW-level input voltage		-	-	20	$%V_{DD}$
V_{OH}	HIGH-level output voltage	$I_O = 2 \text{ mA}$	80	-	-	$%V_{DD}$
V_{OL}	LOW-level output voltage	$I_O = 2 \text{ mA}$	-	-	20	$%V_{DD}$
l _{od}	overdrive current	absolute value for overdriving the output buffer	-	-	20	mA
t _{start}	start time	LOW level before rising edge	5	-	-	μS
t _{stop}	stop time	HIGH level before falling edge	5	-	-	μS
T _{bit}	bit period	the load capacitance limits the minimum period	10	-	100	μS
ΔT_{bit}	bit period deviation	deviation between received clock and sent clock	0.8T _{bit}	1T _{bit}	1.2T _{bit}	μS
t _{w0}	pulse width 0		0.175T _{bit}	0.25T _{bit}	0.375T _{bit}	μS
t _{w1}	pulse width 1		0.625T _{bit}	0.75T _{bit}	0.825T _{bit}	μS
t _{to}	time-out time	communication reset guaranteed after maximum \mathbf{t}_{to}	-	-	220	μS
t _{tko(slv)}	slave takeover time	duration of LOW level for slave takeover	1	-	5	μS
t _{tko(mas)}	master takeover time	duration of LOW level for master takeover	0T _{bit}	-	0.5T _{bit}	μS
t _{prog}	programming time	for a single memory address	20	-	-	ms
t _{cp}	charge pump time	waiting time after enabling the non-volatile memory charge pump clock	1	-	-	ms
		· · ·				

Table 16. Internal capacitances

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
C _{block}	blocking capacitance		<u>[1]</u> 50	100	150	nF
C _L	load capacitance		<u>11</u> 1.1	2.2	3.3	nF

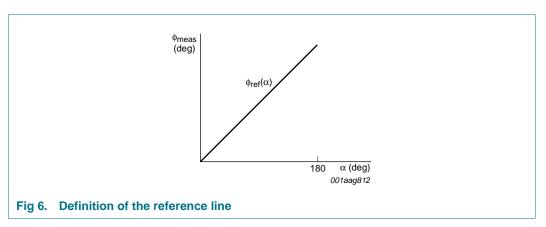
^[1] Measured at 1 MHz.

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12. Definition of errors

12.1 General

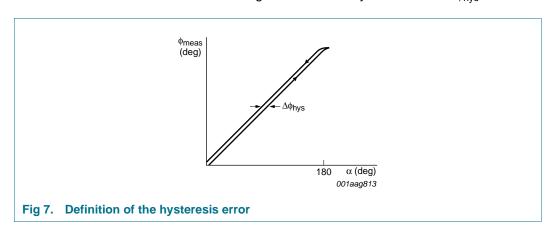
Angular measurement errors by the KMA220 result from linearity errors, temperature drift errors and hysteresis errors and die displacement error. Figure 6 shows the output signal of an ideal sensor, where the measured angle ϕ_{meas} corresponds ideally to the magnetic field angle α . This curve represents the angle reference line $\phi_{ref}(\alpha)$ with a slope of 0.5 %V_{DD}/degree.



The angular range is set to α_{max} = 180° and the clamping voltages are programmed to $V_{(CL)I}$ = 5 % V_{DD} and $V_{(CL)I}$ = 95 % V_{DD} for a valid definition of errors.

12.2 Hysteresis error

The device output performs a positive (clockwise) rotation and negative (counter clockwise) rotation over an angular range of 180° at a constant temperature. The maximum difference between the angles defines the hysteresis error $\Delta\phi_{hvs}$.



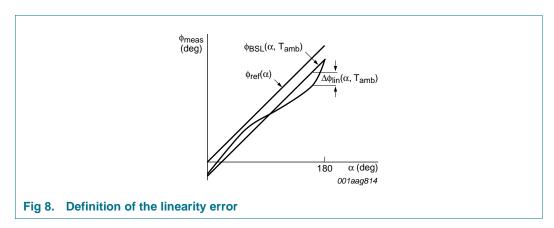
Equation 1 gives the mathematical description for the hysteresis value $\Delta \phi_{hys}$:

$$\Delta \phi_{hys}(\alpha) = \left| \phi_{meas}(\alpha \to 180^{\circ}) - \phi_{meas}(\alpha \to 0^{\circ}) \right| \tag{1}$$

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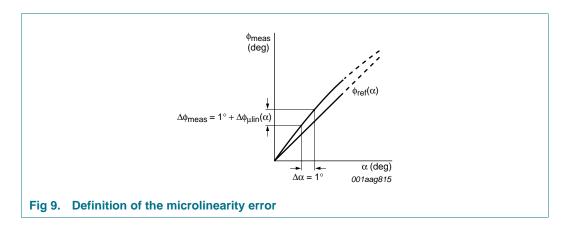
12.3 Linearity error

The KMA220 output signal deviation from a best straight line ϕ_{BSL} , with the same slope as the reference line, is defined as linearity error. The magnetic field angle is varied at fixed temperatures for measurement of this linearity error. The output signals deviation from the best straight line at the given temperature is the linearity error $\Delta \phi_{lin}$. It is a function of the magnetic field angle α and the temperature of the device T_{amb} .



12.4 Microlinearity error

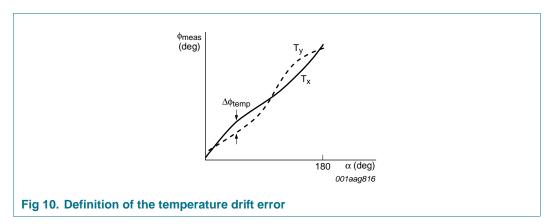
 α is the magnetic field angle. If $\Delta\alpha$ = 1°, the microlinearity error $\Delta\phi_{\mu lin}$ is the device output deviation from 1°.



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12.5 Temperature drift error

The temperature drift $\Delta\phi_{temp}$ is defined as the envelope over the deviation of the angle versus the temperature range. It is considered as the pure thermal effect.



Equation 2 gives the mathematical description for temperature drift value $\Delta \phi_{\text{temp}}$:

$$\Delta \phi_{temp}(\alpha) = \left| \phi_{meas}(\alpha, T_x) - \phi_{meas}(\alpha, T_y) \right| \tag{2}$$

with:

 T_x : temperature for maximum ϕ_{meas} at angle α

 T_y : temperature for minimum ϕ_{meas} at angle α

The deviation from the value at room temperature $\Delta\phi_{temp|RT}$ describes the temperature drift of the angle, compared to the value, which the sensor provides at room temperature:

$$\Delta \phi_{temp|RT}(\alpha, T_{amb}) = |\phi_{meas}(\alpha, T_{amb}) - \phi_{meas}(\alpha, T_{RT})| \tag{3}$$

with:

T_{RT}: room temperature (25 °C)

12.6 Angular error

The angular error $\Delta\phi_{ang}$ is the difference between mechanical angle and sensor output during a movement from α_0 to α_1 . Here α_0 and α_1 are arbitrary angles within the angular range. The customer initially programs the angle measurement at α_0 at room temperature and zero hour upon production. The angle measurement at α_1 is made at any temperature within the ambient temperature range:

$$\Delta \phi_{ang} = (\phi_{meas}(\alpha_1, T_{amb}) - \phi_{meas}(\alpha_0, T_{RT})) - (\alpha_1 - \alpha_0) \tag{4}$$

with:

 α_0 , α_1 : arbitrary mechanical angles within the angular range

 $\phi_{meas}(\alpha_0, T_{RT})$: programmed angle at α_0, T_{RT} = 25 °C and zero hour upon production

 $\phi_{meas}(\alpha_1, T_{amb})$: the sensor measures angle at α_1 and any temperature within T_{amb}

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This error comprises non-linearity and temperature drift related to the room temperature.

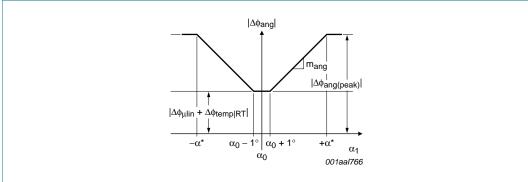


Fig 11. Envelope curve for the magnitude of angular error

Figure 11 shows the envelope curve for the magnitude of angular error $|\Delta\phi_{ang}|$ versus α_1 for all angles α_0 and all temperatures T_{amb} within the ambient temperature range. If α_1 is in the range of $\pm 1^\circ$ around α_0 , $|\Delta\phi_{ang}|$ has its minimum. Here only the microlinearity error $\Delta\phi_{\mu lin}$ and the temperature drift related to the room temperature $|\Delta\phi_{temp}|_{RT}|$ occurs. If α_1 deviates from α_0 by more than 1° in either direction, $|\Delta\phi_{ang}|$ can increase. Slope m_{ang} defines the gradient.

Equation 5 to Equation 8 express the angular error:

for $|\alpha_1 - \alpha_0| \le 1^{\circ}$

$$\left|\Delta\phi_{ang}\right| = \left|\Delta\phi_{\mu lin} + \Delta\phi_{temp|RT}\right| \tag{5}$$

for $1^{\circ} < |\alpha_1 - \alpha_0| < \alpha^*$

$$\left|\Delta\phi_{ang}\right| = \left|\Delta\phi_{\mu lin} + \Delta\phi_{temp|RT}\right| + m_{ang} \times \left(\left|\alpha_{I} - \alpha_{0}\right| - I^{\circ}\right) \tag{6}$$

for $|\alpha_1 - \alpha_0| \ge \alpha^*$

$$|\Delta\phi_{ang}| = \sqrt{(\Delta\phi_{lin})^2 + (\Delta\phi_{temp|RT})^2}$$
 (7)

with:

$$\alpha^* = \frac{\left|\Delta\phi_{ang(peak)}\right| - \left|\Delta\phi_{\mu lin} + \Delta\phi_{temp|RT}\right|}{m_{ang}} + \alpha_0 + 1^{\circ}$$
(8)

Dual channel programmable angle sensor

13. Programming

13.1 General description

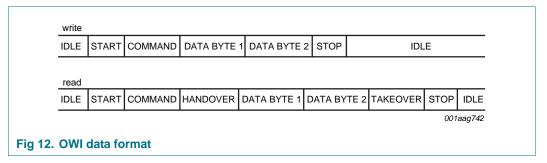
Each channel of the KMA220 provides an OWI to enable programming of the device which uses pin OUT1/DATA1 and pin OUT2/DATA2 bidirectionally.

In general the device runs in analog output mode, the normal operation mode. The embedded programming data configures this mode. After a power-on reset once time ton has elapsed, it starts. In this mode, the magnetic field angle is converted into the corresponding output voltage per channel.

A second mode, the command mode enables programming. In this mode, the customer can adjust all required parameters (for example zero angle and angular range) to meet the application requirements. After enabling the internal charge pump and waiting for t_{cp} , the data is stored in the non-volatile memory. After changing the contents of the memory, recalculate and write the checksum (see Section 13.4).

In order to enter the command mode, send a specific command sequence after a power-on reset and during the time slot $t_{cmd(ent)}$. The external source used to send the command sequence must overdrive the output buffers of the KMA220. In doing so, it provides current I_{od} . This signature can be sent to each channel separately or in parallel.

During communication, the channels of the KMA220 are always the slaves and the external programming hardware is always the master. <u>Figure 12</u> illustrates the structure of the OWI data format.



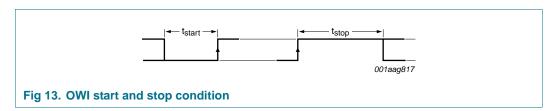
The master provides the start condition, which is a rising edge after a LOW level. Then a command byte which can be either a read or a write command is sent. Depending on the command, the master or the slave has to send the data immediately after the command sequence. If there is a read command, an additional handover or takeover bit is inserted before and after the data bytes. The master must close each communication with a stop condition. If the slave does not receive a rising edge for a time longer than t_{to} , a time-out condition occurs. The bus is reset to the idle state and waits for a start condition and a new command. This behavior can be used to synchronize the device regardless of the previous state.

All communication is based on this structure (see <u>Figure 12</u>), even for entering the command mode. The customer can access the non-volatile memory, CTRL1, TESTCTRL0 and SIGNATURE registers (described in <u>Section 13.5</u>). Only a power-on reset leaves the command mode. A more detailed description of the programming is given in the next sections.

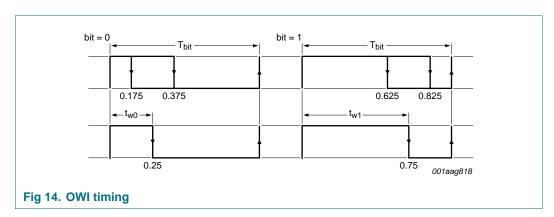
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13.2 Timing characteristics

As described in the previous section, a start and stop condition is necessary for communication. The LOW-level duration before the rising edge of the start condition is defined as t_{start} . The HIGH-level duration after the rising edge of the stop condition is defined as t_{stop} . These parameters, together with all other timing characteristics are shown in Table 15.



<u>Figure 14</u> shows the coding of a single bit with a HIGH level of V_{IH} and a LOW level of V_{IL} . Here the pulse width t_{w1} or t_{w0} represents a logic 1 or a logic 0 of a full bit period T_{bit} , respectively.



13.3 Sending and receiving data

The master has to control the communication during sending or receiving data. The command byte defines the region, address and type of command the master requests. Read commands need an additional handover or takeover bit. Insert this bit before and after the two data bytes (see Figure 12). However the OWI is a serial data transmission, whereas the Most Significant Byte (MSB) send at first.

Table 17. Format of a command byte

7	6	5	4	3	2	1	0
CMD7	CMD6	CMD5	CMD4	CMD3	CMD2	CMD1	CMD0

Table 18. Command byte bit description

Bit	Symbol	Description				
7 to 5	CMD[7:5]	region bits				
	000 = 16-bit non-volatile memory					
		001 to 011 = reserved				
		100 = 16-bit register				
		101 to 111 = reserved				

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Table 18. Command byte bit description ... continued

Bit	Symbol	Description
4 to 1	CMD[4:1]	address bits
0	CMD0	read/write
		0 = write
		1 = read

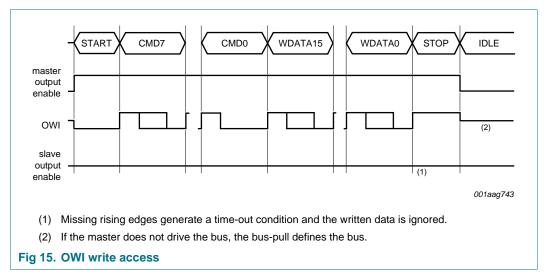
A more detailed description of all customer accessible registers is given in <u>Section 13.5</u>. Both default value and the complete command including the address and write or read request are also listed.

13.3.1 Write access

To write data to the non-volatile memory, enable the internal charge pump. Set bits CP_CLOCK_EN and $WRITE_EN$ and wait for t_{cp} enables the internal charge pump. Perform the following procedure for write access:

- 1. Start condition: The master drives a rising edge after a LOW level
- 2. Command: The master sends a write command (CMD0 = 0)
- 3. Data: The master sends two data bytes
- 4. Stop condition: The master drives a rising edge after a LOW level

<u>Figure 15</u> shows the write access of the digital interface. The signal OWI represents the data on the bus from the master or slave. The signals: master output enable and slave output enable indicate when the master or the slave output is enabled or disabled, respectively.



Note: As already mentioned in Section 13.1, use the write procedure to enter the command mode. If command mode is not entered, communication is not possible and the sensor operates in normal operation mode. After changing an address, the time t_{prog} must elapse before changing another address. After changing the contents of the non-volatile memory, recalculate and write the checksum (see Section 13.4).

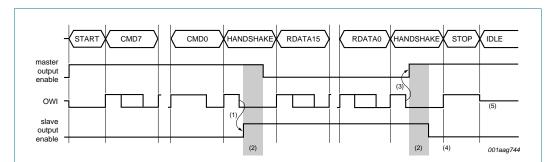
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13.3.2 Read access

Perform the following procedure, to read data from the sensor:

- 1. Start condition: The master drives a rising edge after a LOW level
- 2. Command: The master sends a read command (CMD0 = 1)
- 3. Handover: The master sends a handover bit, that is a logic 0 and disables the output after a three-quarter bit period
- 4. Takeover: The slave drives a LOW level after the falling edge for ttko(slv)
- 5. Data: The slave sends two data bytes
- 6. Handover: The slave sends a handover bit, that is a logic 0 and disables the output after a three-quarter bit period
- 7. Takeover: The master drives a LOW level after the falling edge for t_{tko(mas)}
- 8. Stop condition: The master drives a rising edge after a LOW level

<u>Figure 16</u> shows the read access of the digital interface. The signal OWI represents the data on the bus from the master or slave. The signals: master output enable and slave output enable indicate when the master or the slave output is enabled or disabled, respectively.



- (1) Duration of LOW level for slave takeover $t_{tko(slv)}$.
- (2) The master output enable and the slave output enable overlap, because both drive a LOW level. However this behavior ensures the independency from having a pull-up or pull-down on the bus. In addition, it improves the EMC robustness, because all levels are actively driven.
- (3) Duration of LOW level for master takeover t_{tko(mas)}.
- (4) If the master does not take over, the pull-up generates the stop condition. Otherwise a time-out is generated if there is a pull-down and the slave waits for a rising edge as start condition.
- (5) If the master does not drive the bus, the bus-pull defines the bus.

Fig 16. OWI read access

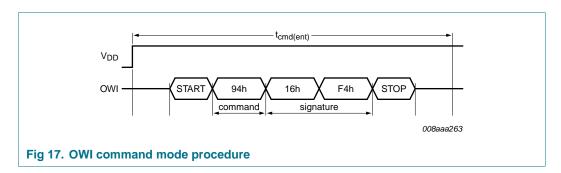
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13.3.3 Entering the command mode

After a power-on reset, the sensor provides a time slot $t_{cmd(ent)}$ for entering the command mode. Send a specific command sequence (see Figure 17). If command mode is not entered, the sensor starts in the normal operation mode. If the sensor stays in the diagnostic mode, the master can write the signature without a power-on reset.

During the command mode sequence, the analog output is enabled. The external programming hardware has to overdrive the output with current I_{od} . If command mode is activated, the analog output is disabled and pin OUT1/DATA1 and pin OUT2/DATA2 operates as a digital interface.



13.4 Cyclic redundancy check

As already mentioned in <u>Section 7</u>, there is an 8-bit checksum for the non-volatile memory data. To calculate this value, the MSB of the memory data word generates the CRC at first over all corresponding addresses in increasing order.

Read out all addresses from 8h to Fh for calculating the checksum. The Least Significant Byte (LSB) of address Fh which contains the previous checksum must be overwritten with 0h before the calculation can be started.

Setting bits CP_CLOCK_EN and WRITE_EN (see Section 13.5.1) and waiting for t_{cp} enables the internal charge pump for programming.

The generator polynomial for the calculation of the checksum is:

$$G(x) = x^8 + x^2 + x + 1 (9)$$

With a start value of FFh and the data bits are XOR at the x⁸ point.

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13.4.1 Software example in C

```
#include <stdio.h.>
1
     // calc_crc accepts unsigned 16-bit data in data
3
     int calc crc(int crc, unsigned int data)
5
6
           const int gpoly = 0x107; // generator polynomial
                                       //index variable
7
           for (i = 15; i >= 0; i--)
9
                                      //shift left
10
                crc <<= 1;
11
                crc |= (int) ((data & (1u<<i))>>i);
                // XOR of with generator polynomial when MSB(9) = HIGH
12
13
                if (crc & 0x100) crc ^= gpoly;
14
15
           return crc;
16
17
     int main(void)
18
19
           int crc, crc_res, i;
20
           // 8 LSB are CRC field filled with 0
           unsigned int data_seq[] = {0x0000, 0xFFC1, 0x0400, 0x0100,
2.1
                                       0x1300, 0x0000, 0x0000, 0x0000);
22
23
           // calculate checksum over all data
24
           crc = 0xFF;
                                       // start value of crc register
25
           printf("Address\tValue\n");
26
           for (i = 0; i <= 7; i++)
2.7
                printf("0x%1X\t0x%04X\n", i, data_seq[i]);
28
29
                crc = calc_crc(crc, data_seq[i]);
30
31
           crc_res = crc;
                                       // crc_res = 0xA9
           printf("\nChecksum\n0x%02X\n", crc_res);
32
33
           // check procedure for preceding data sequence
34
           crc = 0xFF;
35
           for (i = 0; i <= 6; i++)
36
                crc = calc_crc(crc, data_seq[i]);
37
           // last word gets crc inserted
38
           crc = calc_crc(crc, data_seq[i] | crc_res);
39
           printf("\nCheck procedure for data sequence: must be 0x00 is 0x%02X.\n", crc);
4 0
           return 1;
41
```

The checksum of this data sequence is A9h.

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13.5 Registers

13.5.1 Command registers

To enter the command mode, write the signature given in <u>Table 19</u> into the specific register using the OWI. Do this procedure as described in <u>Section 13.3.3</u>, with a write command, the signature follows it, but after a power-on reset and not later than $t_{cmd(ent)}$.

Table 19. Command registers

Command write/read	Register	Bit	Access	Field	Description
82h/83h	CTRL1	15	R	IN_DIAG_MODE	shows if there is a diagnostic condition present; the setting of register field FORCE_DIAG_OFF does not affect this bit
		14	W	FORCE_DIAG_OFF	force diagnostic mode off; default: 0b
		13	-	-	reserved
		12	R	LOW_VOLTAGE_DET	low voltage condition detected
		11	R/W	CP_CLOCK_EN	charge pump clock enabled (must be set after setting write enable signal for writing to non-volatile memory); default: 0b
		10 and 9	-	-	reserved
		8	R	ERR_CORRECT	single-bit error of non-volatile memory has been detected and corrected; updated every memory readout; remains set until the diagnostic condition disappears and a power-on reset is done
		7	R	UNCORR_ERR	double-bit error of non-volatile memory has been detected; updated every memory readout; remains set until the diagnostic condition disappears and a power-on reset is done
		6	R	MAGNET_LOSS_DET	magnet-loss detected; bit remains set until the diagnostic condition disappears and a power-on reset is done; enable magnet-loss detection for entering diagnostic mode
		5	-	-	reserved
		4	R	CRC_BAD	checksum error detected; updated every start-up
		3 to 0	-	-	reserved
94h/-	SIGNATURE	15 to 0	W	SIGNATURE	write signature 16F4h within t _{cmd(ent)} to enter command mode; see Section 13.3.3 for more details
96h/97h	TESTCTRL0	15 to 12	-	-	reserved
		11	W	WRITE_EN	write enable signal; set before writing to non-volatile memory; default: 0b
		10 to 0	-	-	reserved

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13.5.2 Non-volatile memory registers

The device includes several internal registers which are used for customization and identification.

The initial signature allows read access to all areas but only write access to customer registers. Write accesses to reserved areas are ignored. Since these registers are implemented as non-volatile memory cells, writing to the registers needs a specific time t_{prog} after each write access to complete.

As there is no check for the programming time, make sure that no other accesses to the non-volatile memory are made during the programming cycle. Do not address the non-volatile memory during the time t_{proq} .

Note: Before data can be stored in the non-volatile memory, switch on the internal charge pump for the programming duration by setting register CTRL1, bit 11 CP_CLOCK_EN and register TESTCTRL0, bit 11 WRITE_EN. Read out and consult register addresses 8h to Fh to calculate the checksum.

Table 20. Non-volatile memory registers

Address	Command write/read	Register	Bit	Description	Default MSB/LSB				
0h	-/01h	reserved	-	addresses are reserved for calibration purposes	<u>[1]</u>				
1h	-/03h								
2h	-/05h								
3h	-/07h								
4h	-/09h								
5h	-/0Bh								
6h	-/0Dh								
7h	-/0Fh								
8h	10h/11h	ZERO_ANGLE	15 to 0	mechanical zero degree position; see Table 21	00h/00h				
9h	12h/13h	ANG_RNG_MULT_MSB	15 to 6	CLAMP_SW_ANGLE; when the measured angle is bigger than CLAMP_SW_ANGLE the output switches to CLAMP_LO for a positive slope; see Table 26	FFh/C1h				
			5 to 0	ANG_RNG_MULT_MSB; most significant bits of the angular range multiplicator; see Table 24					
Ah	14h/15h	ANG_RNG_MULT_LSB	DIAGNOSTIC_LEVEL; diagnostic level behavior of the analog output; see Table 25	04h/00h					
			00b — active LOW (in lower diagnostic range) with driver strength of the analog output						
			01b — active HIGH (in upper diagnostic range) with driver strength of the analog output						
				10b — reserved					
				11b — reserved					
			13	SLOPE_DIR; slope of analog output					
			<pre>0b — rising (not inverted)</pre>						
				1b — falling (inverted)					
			12 to 0	ANG_RNG_MULT_LSB; least significant bits of the angular range multiplicator					

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Table 20. Non-volatile memory registers ...continued

Address	Command write/read	Register	Bit	Description	Default MSB/LSB	
Bh	16h/17h	CLAMP_LO	15	0b — reserved	01h/00h	
			14 and 13	undefined[2]		
			12 to 0	lower clamping level; see <u>Table 22</u>		
Ch	18h/19h	CLAMP_HI	15 to 13	undefined[2]	13h/00h	
			12 to 0	upper clamping level; see <u>Table 23</u>		
Dh	1Ah/1Bh	ID_LO	15 to 0	lower 16 bits of identification code	00h/00h	
Eh	1Ch/1Dh	ID_HI	15 to 0	upper 16 bits of identification code	00h/00h	
Fh	1Eh/1Fh	CTRL_CUST	15	LOCK; irreversible write protection of non-volatile memory 1b — enabled	00h/[1]	
			14 to 8	MAGNET_LOSS; magnet-loss detection 00h — disabled 49h — enabled		
			7 to 0	CRC; checksum (see Section 13.4)	_	

^[1] Variable and individual for each device.

Table 21. ZERO_ANGLE - mechanical zero degree position (address 8h) bit allocation

Data format: unsigned fixed point; resolution: 2⁻¹⁶.

Bit																
Value	2-1	2-2	2-3	2-4	2-5	2-6	2-7	2-8	2 ⁻⁹	2-10	2-11	2-12	2-13	2-14	2-15	2-16

Mechanical angular range $0000h = 0^{\circ}$ to FFFFh = $180^{\circ} - 1$ LSB.

Examples:

- Mechanical zero angle 0° = 0000h
- Mechanical zero angle 10° = 0E38h
- Mechanical zero angle 45° = 4000h

Table 22. CLAMP_LO - lower clamping level (address Bh) bit allocation

Data format: unsigned integer (DAC values 256 to 4864); resolution: 20.

Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Value	0	U <u>[1]</u>	U <u>[1]</u>	2 ¹²	2 ¹¹	2 ¹⁰	2 ⁹	2 ⁸	2 ⁷	2 ⁶	2 ⁵	2 ⁴	2 ³	2 ²	2 ¹	2 ⁰

[1] Undefined; write as zero for default; returns any value when read.

Values 0 to 255 are reserved. It is not permitted to use such values.

Examples:

- 100 %V_{DD} = 5120 (reserved)
- 10 %V_{DD} = 512
- 5 %V_{DD} = 256

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^[2] Undefined; write as zero for default.

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Table 23. CLAMP_HI - upper clamping level (address Ch) bit allocation

Data format: unsigned integer (DAC values 256 to 4864); resolution: 20.

Bit																
Value	U <u>[1]</u>	U <u>[1]</u>	<u>U[1]</u>	2 ¹²	2 ¹¹	2 ¹⁰	2 ⁹	2 ⁸	27	2 ⁶	2 ⁵	2 ⁴	2 ³	2 ²	21	20

[1] Undefined; write as zero for default; returns any value when read.

Values 4865 to 5120 are reserved. It is not permitted to use such values.

Examples:

- 100 %V_{DD} = 5120 (reserved)
- 95 %V_{DD} = 4864
- 90 %V_{DD} = 4608

Table 24. ANG_RNG_MULT_MSB - most significant bits of angular range multiplicator (address 9h) bit allocation Data format: unsigned fixed point; resolution: 2⁻¹.

Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Value	CLAMP_SW_ANGLE							2 ⁴	2 ³	2 ²	2 ¹	20	2^{-1}			

$$ANG_RNG_MULT = \frac{CLAMP_HI - CLAMP_LO}{8192} \times \frac{180^{\circ}}{ANGULAR_RANGE}$$
 (10)

Examples:

•
$$ANG_RNG_MULT = \frac{4864 - 256}{8192} \times \frac{180^{\circ}}{180^{\circ}} = 0.5625$$

•
$$ANG_RNG_MULT = \frac{4864 - 256}{8192} \times \frac{180^{\circ}}{90^{\circ}} = 1.125$$

Table 25. ANG_RNG_MULT_LSB - least significant bits of angular range multiplicator (address Ah) bit allocation Data format: unsigned fixed point; resolution: 2⁻¹⁴.

																0
Value	V	<u>1]</u>	V <u>[1]</u>	2-2	2-3	2-4	2-5	2-6	2-7	2-8	2-9	2-10	2-11	2-12	2-13	2-14

[1] Variable; depending on the setting of diagnostic level and slope of analog output.

$$ANG_RNG_MULT = \frac{CLAMP_HI - CLAMP_LO}{8192} \times \frac{180^{\circ}}{ANGULAR\ RANGE}$$
(11)

Table 26. CLAMP_SW_ANGLE - clamp switch angle (address 9h) bit allocation

Data format: unsigned fixed point; resolution: 2^{-10} .

Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Value	2-1	2-2	2-3	2-4	2-5	2-6	2-7	2-8	2-9	2-10		ANG	_RNG_	MULT_	MSB	

Mechanical angular range $0000h = 0^{\circ}$ to 3FFh = $180^{\circ} - 1$ LSB.

$$CLAMP_SW_ANGLE = \frac{1}{2} \times \left(1 + \frac{CLAMP_HI - CLAMP_LO}{8192} \times \frac{1}{ANG_RNG_MULT} \right)$$
 (12)

If the magnetic field angle is larger than the CLAMP_SW_ANGLE, the output switches to CLAMP_LO for a positive slope. Program the value of CLAMP_SW_ANGLE, which can be calculated from other non-volatile memory constants.

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14. Electromagnetic compatibility

EMC is verified in an independent and certified test laboratory.

14.1 Emission (CISPR 25)

Tests according to CISPR 25 were fulfilled.

14.1.1 Conducted radio disturbance

Test of the device according to CISPR 25, third edition (2008-03), Chapter 6.2.

Classification level: 5.

14.1.2 Radiated radio disturbance

Test of the device according to CISPR 25, third edition (2008-03), Chapter 6.4.

Classification level: 5 (without addition of 6 dB in FM band).

14.2 Radiated disturbances (ISO 11452-1 third edition (2005-02), ISO 11452-2, ISO 11452-4 and ISO 11452-5)

The common understanding of the requested function is that an effect is tolerated as described in <u>Table 27</u> during the disturbance. The reachable values are setup-dependent and differ from the final application.

Table 27. Failure condition for radiated disturbances

Parameter	Comment	Min	Max	Unit
Variation of output signal in analog output mode	value measured relative to the output at test start	-	±0.9	%V _{DD}

14.2.1 Absorber lined shielded enclosure

Tests according to ISO 11452-2, second edition (2004-11), were fulfilled.

Test level: 200 V/m; extended up to 4 GHz.

State: A.

14.2.2 Bulk-current injection

Tests according to ISO 11452-4, third edition (2005-04), were fulfilled.

Test level: 200 mA.

State: A.

14.2.3 Strip line

Tests according to ISO 11452-5, second edition (2002-04), were fulfilled.

Test level: 200 V/m; extended up to 1 GHz.

State: A.

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14.2.4 Immunity against mobile phones

Tests according to ISO 11452-2, second edition (2004-11), were fulfilled.

State: A.

Definition of Global System for Mobile Communications (GSM) signal:

- Pulse modulation: per GSM specification (217 Hz; 12.5 % duty cycle)
- Modulation grade: ≥ 60 dB
- Sweep: linear 800 MHz to 3 GHz (duration 10 s at 890 MHz, 940 MHz and 1.8 GHz band)
- Antenna polarization: vertical, horizontal
- Field strength: 200 V/m during on-time [calibration in Continuous Wave (CW)]

In deviation of ISO 11452-2, a GSM signal instead of an AM signal was used.

14.3 Electrical transient transmission by capacitive coupling [ISO 7637-3, second edition (2007-07)]

The common understanding of the requested function is that an effect is tolerated as described in $\underline{\text{Table 28}}$ during the disturbance.

Table 28. Failure condition for electrical transient transmission

Parameter	Comment	Min	Max	Unit
Variation of output signal in analog output mode	value measured relative to the output at test start	-	±0.9	$%V_{DD}$

Tests according to ISO 7637-3 were fulfilled.

Test level: IV (for 12 V electrical system).

Classification level: B for pulse Fast a, B for pulse Fast b.

15. ElectroStatic Discharge (ESD)

15.1 Human body model (AEC-Q100-002)

The KMA220 is protected up to 8 kV, according to the human body model at 100 pF and 1.5 k Ω . This protection is ensured at all pins.

Classification level: H3B.

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15.2 Human metal model (ANSI/ESD SP5.6-2009)

The KMA220 is protected up to 8 kV, according to the human metal model at 150 pF and 330 Ω inside the ESD gun. This test utilizes waveforms of the IEC 61000-4-2 standard on component level. Apply the contact discharge in an unsupplied state at pins OUT1/DATA1, OUT2/DATA2 and V_{DD} referred to GND which is connected directly to the ground plane.

Test setup: A.

Test level: 5.

15.3 Machine model (AEC-Q100-003)

The KMA220 is protected up to 400 V, according to the machine model. This protection is ensured at all pins.

Classification level: M4.

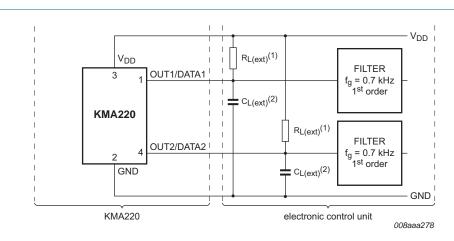
All pins have latch-up protection.

15.4 Charged-device model (AEC-Q100-011)

The KMA220 is protected up to 750 V, according to the charged-device model. This protection is ensured at all pins.

Classification level: C4.

16. Application information



- (1) Power-loss detection is only possible with a load resistance within the specified range connected to the supply or ground line.
- (2) The load capacitance between ground and analog output can be used to improve the electromagnetic immunity of the device. A blocking capacitance to suppress noise on the supply line of the device is integrated into the package and thus not required externally.

Fig 18. Application diagram of KMA220

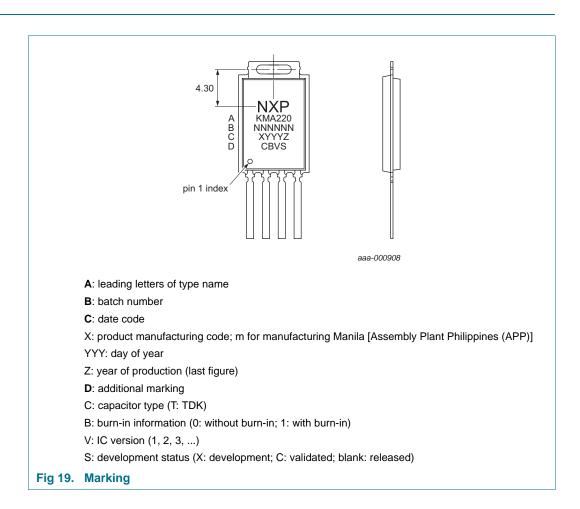
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17. Test information

17.1 Quality information

This product has been qualified in accordance with the Automotive Electronics Council (AEC) standard *Q100 Rev-G - Failure mechanism based stress test qualification for integrated circuits*, and is suitable for use in automotive applications.

18. Marking



19. Terminals

Lead frame material: CuZr with 99.9 % Cu and 0.1 % Zr.

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20. Package outline

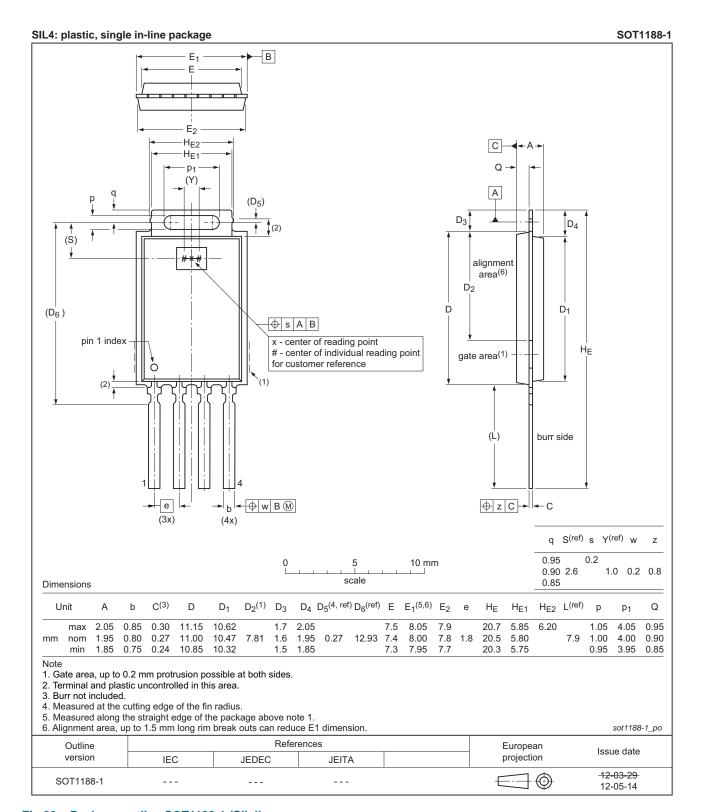


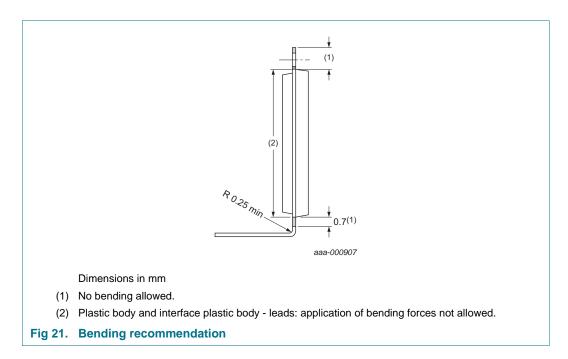
Fig 20. Package outline SOT1188-1 (SIL4)

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Dual channel programmable angle sensor

21. Handling information



22. Solderability information

The solderability qualification is according to AEC-Q100 Rev-G. Recommended soldering process for leaded devices is wave soldering. The maximum soldering temperature is 260 °C for maximum 5 s. Device terminals are compatible with laser and electrical welding.

23. Revision history

Table 29. Revision history

Document ID	Release date	Data sheet status	Change notice	Supersedes
KMA220 v.1	20120524	Product data sheet	-	-

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24. Legal information

24.1 Data sheet status

Document status[1][2]	Product status[3]	Definition
Objective [short] data sheet	Development	This document contains data from the objective specification for product development.
Preliminary [short] data sheet	Qualification	This document contains data from the preliminary specification.
Product [short] data sheet	Production	This document contains the product specification.

- [1] Please consult the most recently issued document before initiating or completing a design.
- [2] The term 'short data sheet' is explained in section "Definitions"
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