SPECIFICATION

L2273

Current Mode PWM Controller VERSION 1.0



Current Mode PWM Controller

Description

L2273 is highly integrated current mode PWM control IC optimized for performance, low standby power and cost effective offline flyblack converter applications in sub 60W range.

Fixed 65kHz PWM switching frequency at normal operation. At no load or light load condition, the IC operates in extended 'burst mode' to minimize switching loss.Lower higher standby power and conversion efficiency in thus achieved.

VDD low startup current and low operating current contribute to a reliable power on startup design with L2273. A large value resistor could thus be used in the startup circuit to minimize the standby power.

L2273 offers complete protection cov-erage with automatic self- recovery feature including Cycle-by-Cycle current limiting (OCP), over temperature protection(OTP), VDD over voltage clamp and under voltage lockout (UVLO). The Gate output is clamped to maximum 18V to protect the power MOSFET.

Features

- Proprietary frequency shuffling technology for improved EMI performance.
- Fixed 65kHz PWM switching frequency.

- **X** Leading edge Blanking on current sense input.
- \mathbb{H} Internal synchronized slope compensation.
- ₩ Extended burst mode control for improved efficiency and minimum standby power design(75mW)
- ₩ Low VDD startup current and low operating current.
- ₩ Gate output maximum voltage clamp
- ₩ Cycle-by-Cycle Current Limiting, Built-in Adaptive Current Peak Regulat-ion
- ₩ Power on Soft-start, Programmable CV and CC Regulation
- ₩ VDD Under Voltage Lockout with Hysteresis (UVLO), VDD OVP, OLP, OTP, OCP, VDD Clamp
- Internal over temperature protection (OTP)

 \mathfrak{R}

Applications

- ₩ Cell Phone Charger
- ${\mathbb H}$ Digital Cameras Charger
- Power adptor
- **¥** Set_top box power supplies
- Open frame SMPS
- \mathbb{H} Battery charger

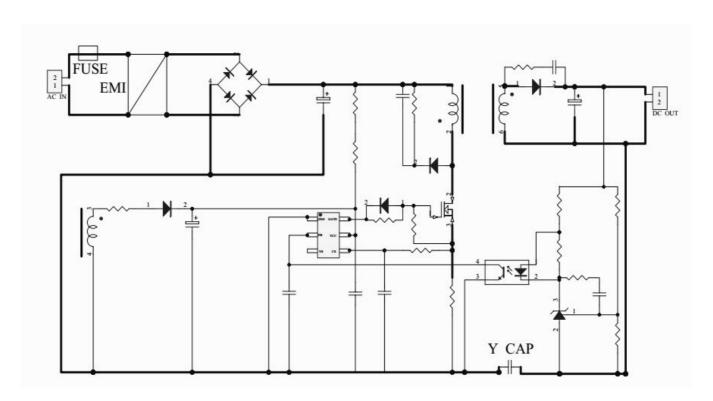


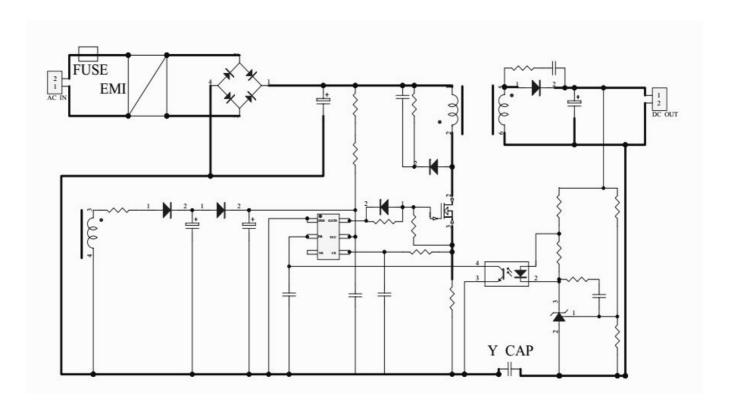
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Application Ciruit

1: Two large value resistors are connected with VCC as startup circuit and Pin3 is inpending .Standby-power (<100 mW)

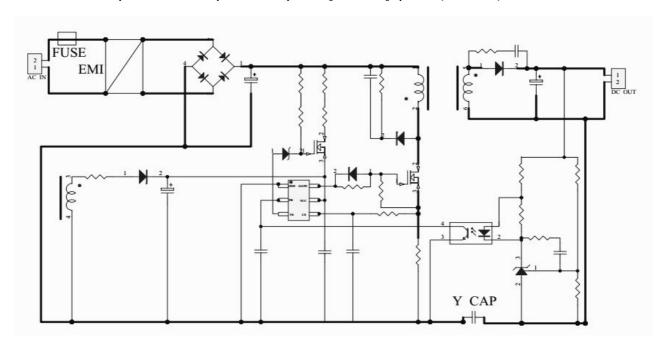


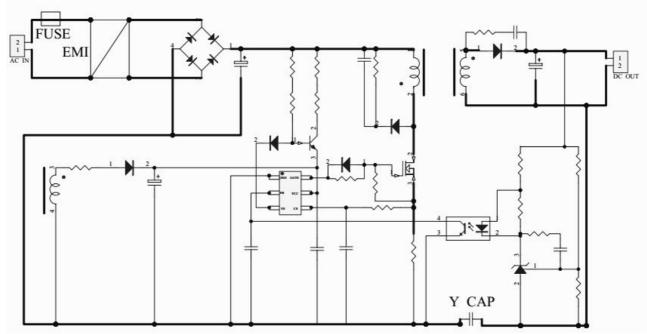




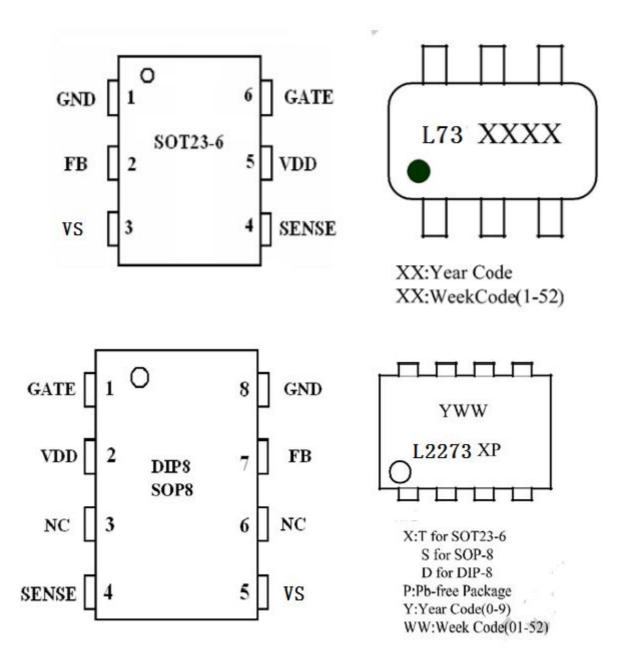
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2: L2273 Pin3 provide start-up circuit especially.Standby-power(<75mW)





Pin Assignment & Marking Information



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Current Mode PWM Controller

Ordering Information

Part number	Package	version number
L2273TP	SOT23-6	
L2273SP	SOP-8	
L2273DP	DIP-8	

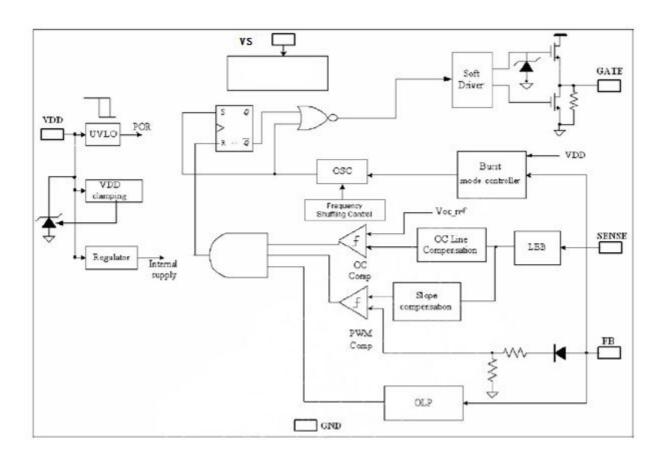
Pin Description

Symbol	Type	Description
GATE	О	Totem-pole gate diver output for the power MOSFET
VDD	P	Chip DC power supply pin
SENSE	I	Current sense intput pin. Connected to MOSFET current resistor node.
VS	Z	Active Start-Up Pin
FB	Ι	Feedback input pin. The PWM duty cycle is determined by voltage level into this pin and SENSE pin input.
Gnd	P	Ground.

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Block Diagram



Absolute Maximum Rating

Parameter	Value	Unit
VDD supply voltage	30	V
VDD clamp voltage	29	V
VDD clamp current	10	mA
VFB input voltage	-0.3 to7	V
VSENSE input voltage to SENSE pin	-0.3 to7	V
Vvs input voltage to VS Pin	-0.3 to7	V
Min/Max operating junction temperature	-55 to 150	$^{\circ}$
Operating ambient temperature	-20 to 85	$^{\circ}$



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Recommended Operating Conditions

Symbol	Parameter	Min. Max.	Unit
VDD	Supply Voltage Vcc	8 to 25	V
Тоа	Operating Ambient Tmperature	-20 to 85	C

Electrical Characteristics($T_A = 25$ °C, if not otherwise noted)

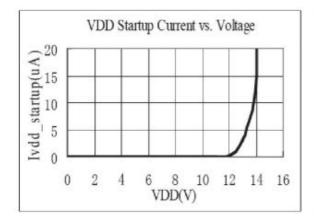
Symbol	Parameter	Conditions	Value			Unit
			Min	Тур	Max	
	Su	pply Voltage(V _{dd} Pin)	_			
Idd_startup	VDD start up current	VDD=12.5V,RI=100K		3	10	uA
Idd	VDD Operatio current	VDD=16V RT=100K Ω , FB=3V		1.5		mA
UVLO(ON)	VDD under voltage lockout enter	7.0 8.0		8.0	10.2	V
UVLO(OFF)	VDD under voltage lockout exit		13	14	16.5	V
VDD_clamp	VDD zener clamp voltage	Idd=10mA	27	29	31	V
	Vol	tage Feedback (FB Pin)	1	,		
AVCS	PWM input gain	VFB/ VSENSE		2		V/V
VFB_open	VFB open loop voltage			5.7		V
IFB_short	FB pin short current	Short FB pin to GND and measure current	0.05	0.1	0.15	mA
VFB_burst	Burst mode voltage			1.1		V
VTH_PL	Power limiting FB threshold voltage	Iout=-10mA		3.7		V
TD_PL	Power limiting debounce time			65		mS
DC_MAX	Maximum duty cycl	VDD=18V,SENSE=0V RI=100KΩ,FB=3V		75		%

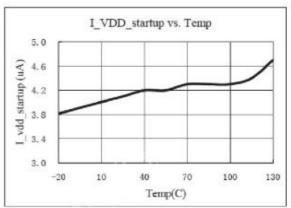
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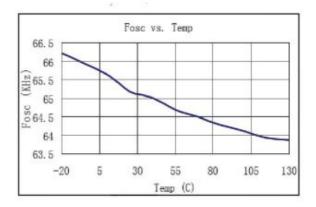
	(Current Sensing (SEN Pin)				
T_blanking	Leading edge	our tent sensing (SEI (Tim)		400		nS
ZSENSE_I N	Input impedance			40		ΚΩ
VTH_sense	Over current thresholdvoltage			0.67		V
		Oscillator				
Fosc	Normal oscillation frequency		50	64	80	Khz
Δf_temp	Frequency temperature stability	TA= -20℃ to 100℃		5		%
f_VDD	Frequency voltage stability	VDD=12V to 25V		1		%
Fosc_BM	Burst mode base frequency			25		Khz
Δf_OSC	Frequency modulation range Base frequency		-7		+7	%
		Gate Drive Output				
VOL	Output low level	VDD=16V,IO=-20mA			0.8	V
VOH	Output high level	VDD=16V,IO=20mA	10			V
V_Clamp	output clamp voltage level			18		V
T_r	Output rising time	VDD=16V,CL=1nF		220		nS
T_f	Output falling time	VDD=16V,CL=1nF		70		nS
		Active Start-Up (VS Pin)				
VS-High	Voltage of VS		VDD+ 0.7			V
VS-Low	Voltage of VS				1	V

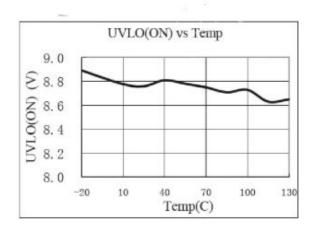
Characterization Plots

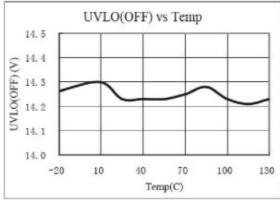
VDD = 16V, TA = 25oC condition applies if not otherwise noted.











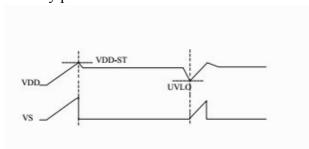


Application Information

The L2273 is a highly integrated PWM controller IC optimized for offline flyback converter applications in sub 60W power range. The extended burst mode control greatly reduces the standby power consumption and helps the design easily meet the international power conservation requirements.

Start up Control

L2273 has especially Start-up circuit .As application circuit, Vdd active Start-up circuits using external N-mos and BJT respectively Prior to start-up, the N-mos or the BJT is turned on, allowing the start-up current to charge the VDD bypass capacitor.r. When the VDD bypass capacitor is charged to a voltage higher than the start-up threshold VDD(ST), the N-mos and BJT will be turned off .The N-mos and BJT will be turned off all the time until the voltage of VDD bypass capacitor lower than UVLO. The voltage of VS will control the N-mos and BJT can low Standby power.



Operating Current

The Operating current of L2273 is low at 1.5mA.Good efficiency is achieved with L2273 low operating current together with extended burst mode control features.

Frequency shuffling for EMI **improvement**

The frequency Shuffling (switching frequency modulation) is implemented in L2273. The oscillation frequency is modulated with a random source so that the tone energy is spread out. The spread spectrum minimizes the conduction band EMI and therefore reduces system design challenge.

ended Burst Mode Operation

At zero load or light load condition, majority of the power dissipation in a switching mode power supply is from switching loss on the MOSFET transistor. the core loss transformer and the loss on the snubber circuit. The magnitude of power loss is in proportion to the number of switching events within a fixed period of time. Reducing switching events leads to the reduction on the power loss and thus conserves the energy.L2273 self adjusts the switching mode according to the loading condition. At from no load to light/medium load condition, the FB input drops below burst mode threshold level. Device enters Burst Mode control. The Gate drive output switches only when VDD voltage drops below a preset level and FB input is active to output an on state. Otherwise the gate drive remains at off state to minimize the switching loss and reduces the standby power consumption to the greatest extend. The frequency control also eliminates the audio noise at any loading conditions.

Oscillator Operation

The switching frequency of briard is internally fixed at 65kHz.No external frequence setting components are required for PCB desigh simplifecation.

Current Sensing and Leading Edge Blanking

Cycle-by-Cycle current limiting is offered in L2273 current mode PWM control. The switch current is detected by a sense resistor into the sense pin. An internal leading edge blanking circuit chops off the sense voltage spike at initial MOSFET on state due to Snubber diode reverse recovery so that the external RC filtering on sense input is no longer required. The current limit comparator is disabled and thus cannot turn off the external MOSFET during the blanking period. PWM duty cycle is determined by the current sense input voltage and the FB input voltage.



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Internal Synchronized Slope Compensation

Built-in slope compensation circuit adds voltage ramp onto the current sense input voltage for PWM generation. This greatly improves the close loop stability at CCM and prevents the sub-harmonic oscillation and thus reduces the output ripple voltage.

Gate Drive

L2273 Gate is connected to an external MOSFET gate for power switch control. Too weak the gate drive strength results in higher conduction and switch loss of MOSFET while too strong gate drive output compromises the EMI.A good tradeoff is achieved through the built-in totem pole gate design with right output strength and dead time control. The low idle loss and good EMI system design is easier to achieve with this dedicated control scheme. An internal 18V clamp is added for MOSFET gate protection at higher than expected VDD input.

Protection Controls

Good power supply system reliability is achieved with its rich protection features including Cycle-by- Cycle current limiting (OCP), Over Load Protection (OLP) and over voltage clamp, Under Voltage Lockout on VDD (UVLO). With On-Bright Proprietary technology, the OCP threshold tracks PWM Duty cycles and is line voltage compensated to achieve constant output power limit over the universal input voltage range with recommended reference design. At overload condition when FB input voltage exceeds power limit threshold value for more than TD PL, control circuit reacts to shut down the output power MOSFET. Device restarts when VDD voltage drops below UVLO limit. VDD is supplied by transformer auxiliary winding output. It is clamped when VDD is higher than threshold value. The power MOSFET is shut down when VDD drops below UVLO limit and deviceenters power on start-up sequence thereafter

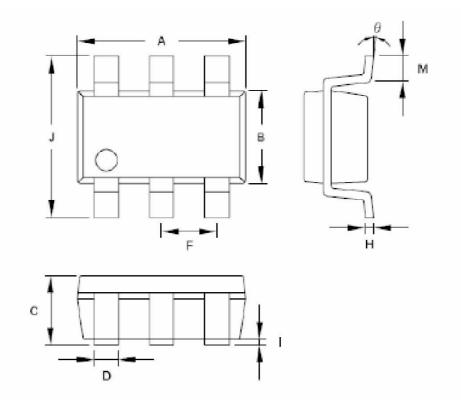




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Package Information

SOT-23-6

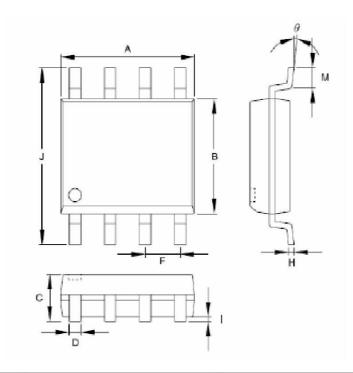


Symbol	Dimension	in Millimeters	Dimensio	ons in Inches
Symbol	Min	Max	Min	Max
А	2.692	3.099	0.106	0.122
В	1.397	1.803	0.055	0.071
С		1.450		0.058
D	0.300	0.550	0.012	0.022
F	0.838	1.041	0.033	0.041
Н	0.080	0.254	0.003	0.010
1	0.050	0.150	0.002	0.006
J	2.600	3.000	0.102	0.118
М	0.300	0.600	0.012	0.024
е	0°	10°	0°	10°





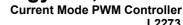
SOP-8



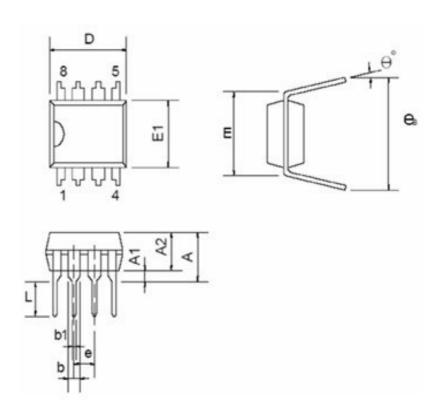
Chl-	Dimensions i	n Millimeters	Dimensio	ns in Inch
Symbols	MIN	MAX	MIN	MAX
А	4.801	5.004	0.189	0.197
В	3.810	3.988	0.150	0.157
С	1.346	1.753	0.053	0.069
D	0.330	0.508	0.013	0.020
F	1.194	1.346	0.047	0.053
Н	0.178	0.229	0.007 0.009	
Ţ	0.102	0.254	0.004	0.010
J	5.791	6.198	0.228	0.244
М	0.406	1.270	0.016	0.050
θ	0°	8°	0°	8°



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Cumbal		Millimeters			Inches	
Symbol	Min.	Тур.	Max.	Min.	Typ.	Max.
A			5.334			0.210
A1	0.381			0.015		
A2	3.175	3.302	3.429	0.125	0.130	0.135
b		1.524			0.060	
b1		0.457			0.018	
D	9.017	9.271	10.160	0.355	0.365	0.400
E		7.620			0.300	
E1	6.223	6.350	6.477	0.245	0.250	0.255
e		2.540			0.100	
L.	2.921	3.302	3.810	0.115	0.130	0.150
eB	8.509	9.017	9.525	0.335	0.355	0.375
θ'	0"	7'	15*	0,	7*	15'

