

# LM2672EP Enhanced Plastic SIMPLE SWITCHER<sup>®</sup> Power Converter High Efficiency 1A Step-Down Voltage Regulator with Features

### **General Description**

The LM2672EP series of regulators are monolithic integrated circuits built with a LMDMOS process. These regulators provide all the active functions for a step-down (buck) switching regulator, capable of driving a 1A load current with excellent line and load regulation. These devices are available in fixed output voltages of 3.3V, 5.0V, 12V, and an adjustable output version.

Requiring a minimum number of external components, these regulators are simple to use and include patented internal frequency compensation (Patent Nos. 5,382,918 and 5,514,947), fixed frequency oscillator, external shutdown, soft-start, and frequency synchronization.

The LM2672EP series operates at a switching frequency of 260 kHz, thus allowing smaller sized filter components than what would be needed with lower frequency switching regulators. Because of its very high efficiency (>90%), the copper traces on the printed circuit board are the only heat sinking needed.

A family of standard inductors for use with the LM2672EP are available from several different manufacturers. This feature greatly simplifies the design of switch-mode power supplies using these advanced ICs. Also included in the datasheet are selector guides for diodes and capacitors designed to work in switch-mode power supplies.

Other features include a guaranteed  $\pm 1.5\%$  tolerance on output voltage within specified input voltages and output load conditions, and  $\pm 10\%$  on the oscillator frequency. External shutdown is included, featuring typically 50 µA stand-by current. The output switch includes current limiting, as well as thermal shutdown for full protection under fault conditions.

To simplify the LM2672EP buck regulator design procedure, there exists computer design software, *LM267X Made Simple* version 6.0.

#### ENHANCED PLASTIC

- Extended Temperature Performance of -40°C to +125°C
- Baseline Control Single Fab & Assembly Site
- Process Change Notification (PCN)
- Qualification & Reliability Data
- Solder (PbSn) Lead Finish is standard
- Enhanced Diminishing Manufacturing Sources (DMS)
   Support

#### **Features**

- Efficiency up to 96%
- Available in SO-8, 8-pin DIP and LLP packages
- Computer Design Software LM267X Made Simple version 6.0
- Simple and easy to design with
- Requires only 5 external components
- Uses readily available standard inductors
- 3.3V, 5.0V, 12V, and adjustable output versions
- Adjustable version output voltage range: 1.21V to 37V
- ±1.5% max output voltage tolerance over line and load conditions
- Guaranteed 1A output load current
- 0.25Ω DMOS Output Switch
- Wide input voltage range: 8V to 40V
- 260 kHz fixed frequency internal oscillator
- TTL shutdown capability, low power standby mode
- Soft-start and frequency synchronization
- Thermal shutdown and current limit protection

# **Typical Applications**

- Simple High Efficiency (>90%) Step-Down (Buck) Regulator
- Efficient Pre-Regulator for Linear Regulators
- Selected Military Applications
- Selected Avionics Applications

# **Ordering Information**

PART NUMBER	VID PART NUMBER	NS PACKAGE NUMBER (Note 3)
LM2672MX-ADJEP	V62/04629-01	M08A
(Notes 1, 2)	TBD	TBD

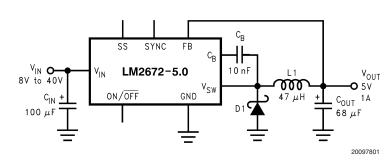
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#### Ordering Information (Continued)

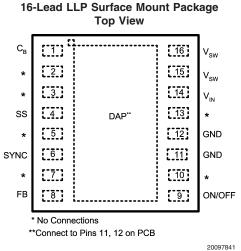
Note 1: For the following (Enhanced Plastic) versions, check for availability: LM2672LD-12EP, LM2672LD-3.3EP, LM2672LD-5.0EP, LM2672LD-ADJEP, LM2672LDX-12EP, LM2672LDX-3.3EP, LM2672LDX-5.0EP, LM2672LDX-ADJEP, LM2672M-12EP, LM2672M-3.3EP, LM2672M-5.0EP, LM2672M-ADJEP, LM2672MX-12EP, LM2672MX-3.3EP, LM2672MX-5.0EP, LM2672N-12EP, LM2672N-3.3EP, LM2672N-5.0EP, LM2672N-ADJEP. Parts listed with an "X" are provided in Tape & Reel and parts without an "X" are in Rails.

Note 2: FOR ADDITIONAL ORDERING AND PRODUCT INFORMATION, PLEASE VISIT THE ENHANCED PLASTIC WEB SITE AT: www.national.com/mil Note 3: Refer to package details under Physical Dimensions

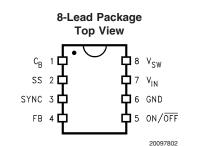
Typical Application (Fixed Output Voltage Versions)



#### **Connection Diagrams**



LLP Package See NSC Package Drawing Number LDA16A



SO-8/DIP Package See NSC Package Drawing Number MO8A/N08E

# Absolute Maximum Ratings (Note 4)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/ Distributors for availability and specifications.

Supply Voltage	45V
ON/OFF Pin Voltage	$-0.1V \leq V_{SH} \leq 6V$
Switch Voltage to Ground	-1V
Boost Pin Voltage	$V_{SW} + 8V$
Feedback Pin Voltage	$-0.3V \le V_{FB} \le 14V$
ESD Susceptibility	
Human Body Model (Note 5)	2 kV
Power Dissipation	Internally Limited
Storage Temperature Range	-65°C to +150°C

Lead Temperature	
M Package	
Vapor Phase (60s)	+215°C
Infrared (15s)	+220°C
N Package (Soldering, 10s)	+260°C
LLP Package (see AN-1187)	
Maximum Junction Temperature	+150°C

# **Operating Ratings**

Supply Voltage	6.5V to 40V
Temperature Range	$-40^{\circ}C \le T_{J} \le +125^{\circ}C$

# **Electrical Characteristics**

**LM2672-3.3EP** Specifications with standard type face are for  $T_J = 25^{\circ}C$ , and those in **bold type face** apply over full Operating Temperature Range.

Symbol	Parameter	Conditions	Typical	Min	Max	Units		
			(Note 7)	(Note 8)	(Note 8)			
SYSTEM	SYSTEM PARAMETERS Test Circuit Figure 2 (Note 6)							
V <sub>OUT</sub>	Output Voltage	$V_{IN} = 8V$ to 40V, $I_{LOAD} = 20$ mA to 1A	3.3	3.251/ <b>3.201</b>	3.350/ <b>3.399</b>	V		
V <sub>OUT</sub>	Output Voltage	$V_{IN} = 6.5V$ to 40V, $I_{LOAD} = 20$ mA to 500 mA	3.3	3.251/ <b>3.201</b>	3.350/ <b>3.399</b>	V		
η	Efficiency	$V_{IN} = 12V, I_{LOAD} = 1A$	86			%		
LM2	672-5.0EP							
LM2	672-5.0EP							
LM2	672-5.0EP Parameter	Conditions	Typical	Min	Мах	Units		
		Conditions	Typical (Note 7)	Min (Note 8)	Max (Note 8)	Units		
Symbol	Parameter	Conditions est Circuit <i>Figure 2</i> (Note 6)				Units		
Symbol	Parameter					Units		
Symbol SYSTEM	Parameter PARAMETERS Te	est Circuit <i>Figure 2</i> (Note 6)	(Note 7)	(Note 8)	(Note 8)			

# LM2672-12EP

Symbol	Parameter	Conditions	Typical	Min	Мах	Units	
			(Note 7)	(Note 8)	(Note 8)		
SYSTEM	SYSTEM PARAMETERS Test Circuit Figure 2 (Note 6)						
V <sub>OUT</sub>	Output Voltage	$V_{IN} = 15V$ to 40V, $I_{LOAD} = 20$ mA to 1A	12	11.82/ <b>11.64</b>	12.18/ <b>12.36</b>	V	
η	Efficiency	$V_{IN} = 24V, I_{LOAD} = 1A$	94			%	

# LM2672-ADJEP

Symbol	Parameter	Conditions	Тур	Min	Max	Units		
			(Note 7)	(Note 8)	(Note 8)			
SYSTEM	SYSTEM PARAMETERS Test Circuit Figure 3 (Note 6)							
$V_{FB}$	Feedback Voltage	$V_{IN} = 8V$ to 40V, $I_{LOAD} = 20$ mA to 1A	1.210	1.192/ <b>1.174</b>	1.228/ <b>1.246</b>	V		
		V <sub>OUT</sub> Programmed for 5V (see Circuit of <i>Figure 3</i> )						
$V_{FB}$	Feedback Voltage	$V_{IN}$ = 6.5V to 40V, $I_{LOAD}$ = 20 mA to 500 mA	1.210	1.192/ <b>1.174</b>	1.228/ <b>1.246</b>	V		
		V <sub>OUT</sub> Programmed for 5V						
		(see Circuit of <i>Figure 3</i> )						

#### LM2672-ADJEP (Continued)

Symbol	Parameter	Conditions	Typ (Note 7)	Min (Note 8)	Max (Note 8)	Units
η	Efficiency	$V_{IN} = 12V, I_{LOAD} = 1A$	90			%

### All Output Voltage Versions

Specifications with standard type face are for  $T_J = 25$ °C, and those in **bold type face** apply over **full Operating Temperature Range**. Unless otherwise specified,  $V_{IN} = 12V$  for the 3.3V, 5V, and Adjustable versions and  $V_{IN} = 24V$  for the 12V version, and  $I_{LOAD} = 100$  mA.

Symbol	Parameters	Conditions	Тур	Min	Max	Units
DEVICE	PARAMETERS					
Ι <sub>Q</sub>	Quiescent Current	$V_{\text{FEEDBACK}} = 8V$	2.5		3.6	mA
		For 3.3V, 5.0V, and ADJ Versions				
		V <sub>FEEDBACK</sub> = 15V	2.5			mA
		For 12V Versions				
I <sub>STBY</sub>	Standby Quiescent Current	ON/OFF Pin = 0V	50		100/ <b>150</b>	μA
I <sub>CL</sub>	Current Limit		1.55	1.25/ <b>1.2</b>	2.1/ <b>2.2</b>	A
IL	Output Leakage Current	$V_{IN} = 40V, ON/\overline{OFF}$ Pin = 0V	1		25	μA
		$V_{SWITCH} = 0V$				
		$V_{SWITCH} = -1V, ON/\overline{OFF}$ Pin = 0V	6		15	mA
R <sub>DS(ON)</sub>	Switch On-Resistance	I <sub>SWITCH</sub> = 1A	0.25		0.30/ <b>0.50</b>	Ω
f <sub>o</sub>	Oscillator Frequency	Measured at Switch Pin	260	225	275	kHz
D	Maximum Duty Cycle		95			%
	Minimum Duty Cycle		0			%
I <sub>BIAS</sub>	Feedback Bias	V <sub>FEEDBACK</sub> = 1.3V	85			nA
	Current	ADJ Version Only				
V <sub>S/D</sub>	ON/OFF Pin		1.4	0.8	2.0	V
	Voltage Thresholds					
I <sub>S/D</sub>	ON/OFF Pin Current	ON/OFF Pin = 0V	20	7	37	μA
F <sub>SYNC</sub>	Synchronization Frequency	V <sub>SYNC</sub> = 3.5V, 50% duty cycle	400			kHz
V <sub>SYNC</sub>	Synchronization Threshold Voltage		1.4			V
V <sub>SS</sub>	Soft-Start Voltage		0.63	0.53	0.73	V
I <sub>ss</sub>	Soft-Start Current		4.5	1.5	6.9	μA
θ <sub>JA</sub>	Thermal Resistance	N Package, Junction to Ambient (Note 9)	95			°C/W
		M Package, Junction to Ambient (Note 9)	105			

Note 4: Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but device parameter specifications may not be guaranteed under these conditions. For guaranteed specifications and test conditions, see the Electrical Characteristics.

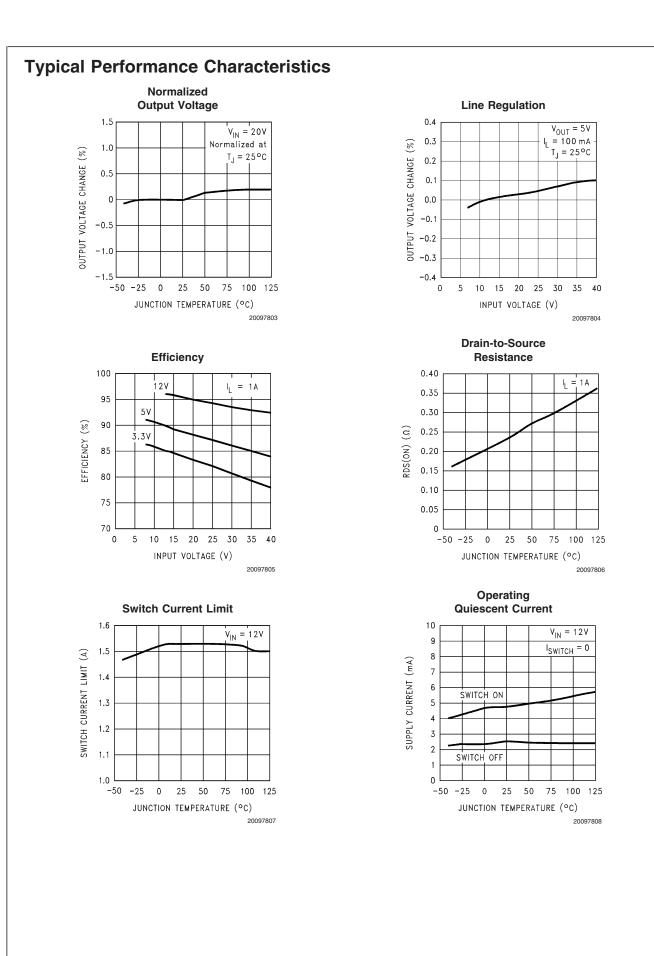
Note 5: The human body model is a 100 pF capacitor discharged through a 1.5 k $\Omega$  resistor into each pin.

Note 6: External components such as the catch diode, inductor, input and output capacitors, and voltage programming resistors can affect switching regulator performance. When the LM2672EP is used as shown in *Figure 2* and *Figure 3* test circuits, system performance will be as specified by the system parameters section of the Electrical Characteristics.

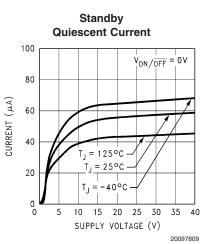
Note 7: Typical numbers are at 25°C and represent the most likely norm.

Note 8: All limits guaranteed at room temperature (standard type face) and at temperature extremes (bold type face). All room temperature limits are 100% production tested. All limits at temperature extremes are guaranteed via correlation using standard Statistical Quality Control (SQC) methods. All limits are used to calculate Average Outgoing Quality Level (AOQL).

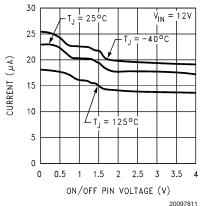
**Note 9:** Junction to ambient thermal resistance with approximately 1 square inch of printed circuit board copper surrounding the leads. Additional copper area will lower thermal resistance further. See Application Information section in the application note accompanying this datasheet and the thermal model in *LM267X Made Simple* version 6.0 software. The value  $\theta_{J-A}$  for the LLP (LD) package is specifically dependent on PCB trace area, trace material, and the number of layers and thermal vias. For improved thermal resistance and power dissipation for the LLP package, refer to Application Note AN-1187.



# Typical Performance Characteristics (Continued)









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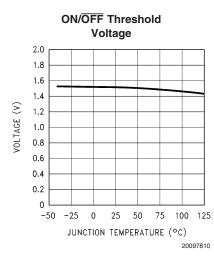
-50 -25

0 25

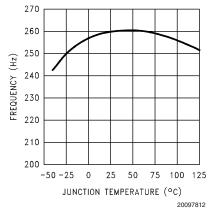
50 75 100 125

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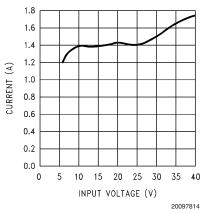
JUNCTION TEMPERATURE (°C)

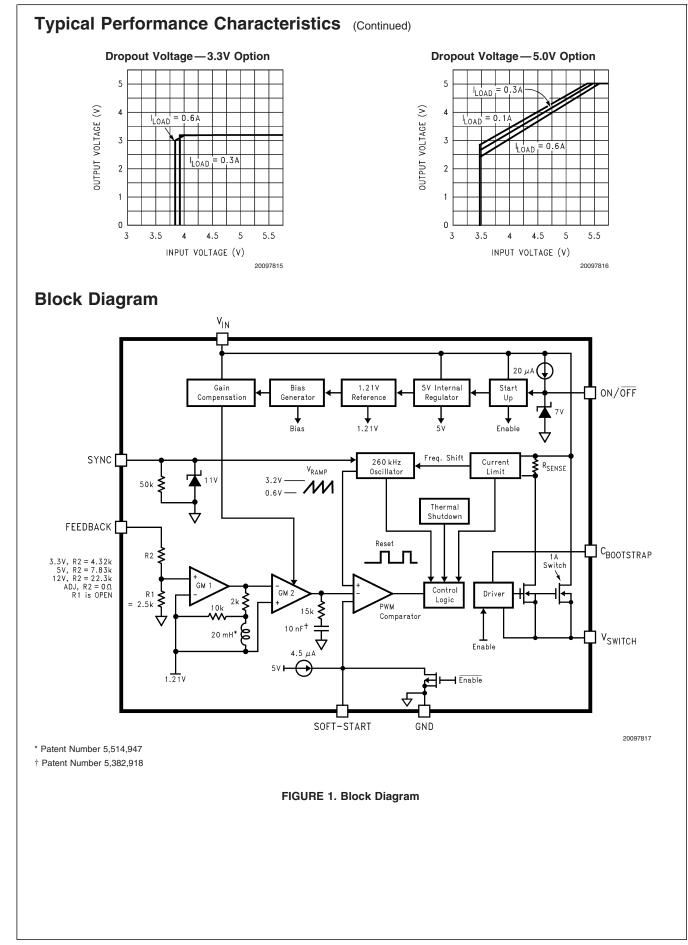


**Switching Frequency** 



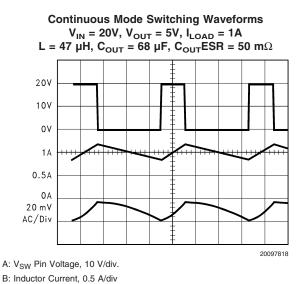
**Peak Switch Current** 

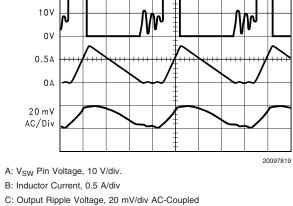






#### **Test Circuit and Layout Guidelines** (Circuit of Figure 2)



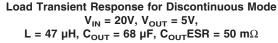


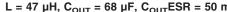
**Discontinuous Mode Switching Waveforms** 

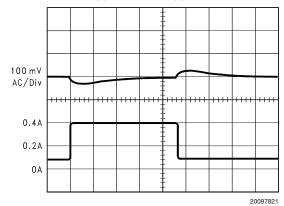
 $V_{\text{IN}} = 20V, V_{\text{OUT}} = 5V, I_{\text{LOAD}} = 300 \text{ mA}$ L = 15  $\mu$ H, C<sub>OUT</sub> = 68  $\mu$ F (2x), C<sub>OUT</sub>ESR = 25 m $\Omega$ 

20V

Horizontal Time Base: 1 µs/div







A: Output Voltage, 100 mV/div, AC-Coupled

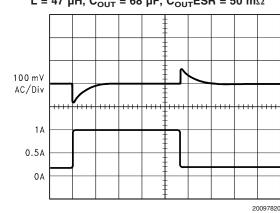
B: Load Current: 100 mA to 300 mA Load Pulse

Horizontal Time Base: 200 µs/div

Load Transient Response for Continuous Mode 
$$\label{eq:VIN} \begin{split} V_{\text{IN}} &= 20V, \, V_{\text{OUT}} = 5V, \, I_{\text{LOAD}} = 1A\\ L &= 47 \; \mu\text{H}, \, C_{\text{OUT}} = 68 \; \mu\text{F}, \, C_{\text{OUT}} \text{ESR} = 50 \; \text{m}\Omega \end{split}$$

Horizontal Time Base: 1 µs/div

C: Output Ripple Voltage, 20 mV/div AC-Coupled

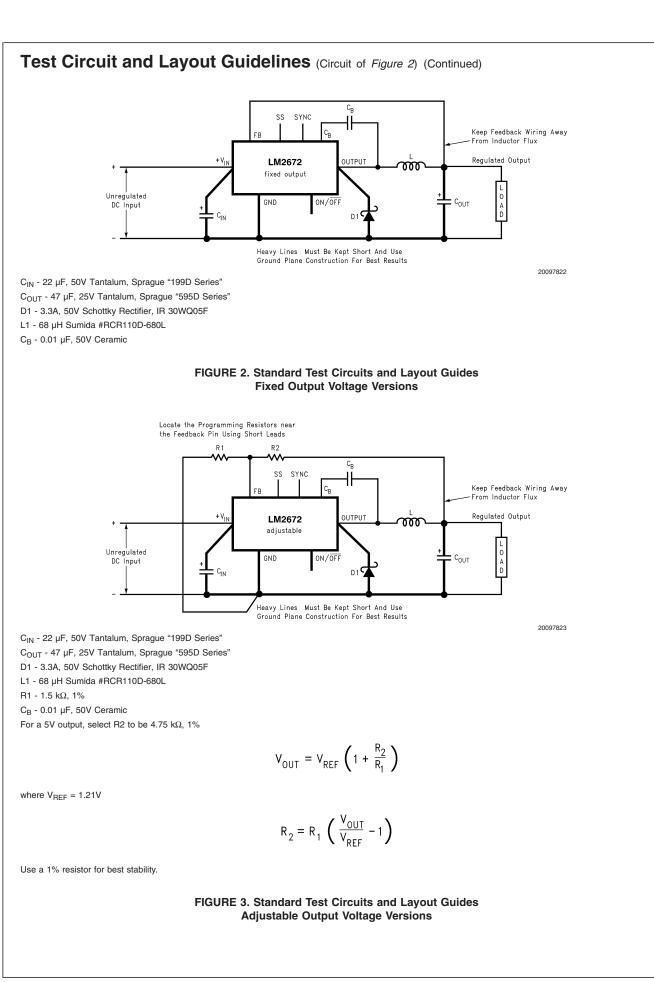


A: Output Voltage, 100 mV/div, AC-Coupled

B: Load Current: 200 mA to 1A Load Pulse

Horizontal Time Base: 50 µs/div

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# LM2672EP Series Buck Regulator Design Procedure (Fixed Output)

PROCEDURE (Fixed Output Voltage Version)	EXAMPLE (Fixed Output Voltage Version)
To simplify the buck regulator design procedure, National	
Semiconductor is making available computer design software to	
be used with the SIMPLE SWITCHER® line of switching	
regulators. LM267X Made Simple version 6.0 is available on	
Windows <sup>®</sup> 3.1, NT, or 95 operating systems.	
Given:	Given:
V <sub>OUT</sub> = Regulated Output Voltage (3.3V, 5V, or 12V)	$V_{OUT} = 5V$
V <sub>IN</sub> (max) = Maximum DC Input Voltage	$V_{IN}(max) = 12V$
I <sub>LOAD</sub> (max) = Maximum Load Current	$I_{LOAD}(max) = 1A$
1. Inductor Selection (L1)	1. Inductor Selection (L1)
A. Select the correct inductor value selection guide from <i>Figure</i>	<b>A.</b> Use the inductor selection guide for the 5V version shown in
4 and Figure 5 or Figure 6 (output voltages of 3.3V, 5V, or 12V	Figure 5.
respectively). For all other voltages, see the design procedure	
for the adjustable version.	
B. From the inductor value selection guide, identify the	<b>B.</b> From the inductor value selection guide shown in <i>Figure 5</i> ,
inductance region intersected by the Maximum Input Voltage	the inductance region intersected by the 12V horizontal line and
line and the Maximum Load Current line. Each region is	the 1A vertical line is 33 $\mu$ H, and the inductor code is L23.
identified by an inductance value and an inductor code (LXX).	
<b>C.</b> Select an appropriate inductor from the four manufacturer's	<b>C.</b> The inductance value required is 33 $\mu$ H. From the table in
part numbers listed in <i>Figure 8</i> . Each manufacturer makes a	Figure 8, go to the L23 line and choose an inductor part number
different style of inductor to allow flexibility in meeting various	from any of the four manufacturers shown. (In most instances,
design requirements. Listed below are some of the differentiating characteristics of each manufacturer's inductors:	both through hole and surface mount inductors are available.)
<i>Schott:</i> ferrite EP core inductors; these have very low leakage	
magnetic fields to reduce electro-magnetic interference (EMI)	
and are the lowest power loss inductors	
<i>Renco:</i> ferrite stick core inductors; benefits are typically lowest	
cost inductors and can withstand E•T and transient peak	
currents above rated value. Be aware that these inductors have	
an external magnetic field which may generate more EMI than	
other types of inductors.	
Pulse: powered iron toroid core inductors; these can also be low	
cost and can withstand larger than normal E•T and transient	
peak currents. Toroid inductors have low EMI.	
Coilcraft: ferrite drum core inductors; these are the smallest	
physical size inductors, available only as SMT components. Be	
aware that these inductors also generate EMI-but less than	
stick inductors.	
Complete specifications for these inductors are available from	
the respective manufacturers. A table listing the manufacturers'	
phone numbers is located in Figure 9.	
2. Output Capacitor Selection (C <sub>OUT</sub> )	2. Output Capacitor Selection (C <sub>OUT</sub> )
A. Select an output capacitor from the output capacitor table in	<b>A.</b> Use the 5.0V section in the output capacitor table in <i>Figure</i>
Figure 10. Using the output voltage and the inductance value	10. Choose a capacitor value and voltage rating from the line
found in the inductor selection guide, step 1, locate the	that contains the inductance value of 33 $\mu$ H. The capacitance
appropriate capacitor value and voltage rating.	and voltage rating values corresponding to the 33 $\mu H$

# LM2672EP Series Buck Regulator Design Procedure (Fixed Output) (Continued)

PROCEDURE (Fixed Output Voltage Version)	EXAMPLE (Fixed Output Voltage Version)
The capacitor list contains through-hole electrolytic capacitors	Surface Mount:
rom four different capacitor manufacturers and surface mount	68 μF/10V Sprague 594D Series.
tantalum capacitors from two different capacitor manufacturers.	100 µF/10V AVX TPS Series.
It is recommended that both the manufacturers and the	Through Hole:
manufacturer's series that are listed in the table be used. A	68 μF/10V Sanyo OS-CON SA Series.
table listing the manufacturers' phone numbers is located in	220 µF/35V Sanyo MV-GX Series.
Figure 11.	220 µF/35V Nichicon PL Series.
	220 µF/35V Panasonic HFQ Series.
3. Catch Diode Selection (D1)	3. Catch Diode Selection (D1)
A. In normal operation, the average current of the catch diode is	A. Refer to the table shown in Figure 12. In this example, a 1A
the load current times the catch diode duty cycle, 1-D (D is the	20V Schottky diode will provide the best performance. If the
switch duty cycle, which is approximately the output voltage	circuit must withstand a continuous shorted output, a higher
divided by the input voltage). The largest value of the catch	current Schottky diode is recommended.
diode average current occurs at the maximum load current and	
maximum input voltage (minimum D). For normal operation, the	
catch diode current rating must be at least 1.3 times greater	
than its maximum average current. However, if the power supply	
design must withstand a continuous output short, the diode	
should have a current rating equal to the maximum current limit	
of the LM2672EP. The most stressful condition for this diode is	
a shorted output condition.	
B. The reverse voltage rating of the diode should be at least	
1.25 times the maximum input voltage.	
C. Because of their fast switching speed and low forward	
voltage drop, Schottky diodes provide the best performance and	
efficiency. This Schottky diode must be located close to the	
LM2672EP using short leads and short printed circuit traces.	

# LM2672EP Series Buck Regulator Design Procedure (Fixed Output) (Continued)

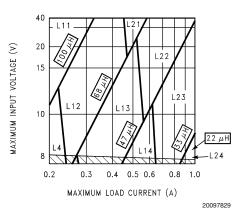
PROCEDURE (Fixed Output Voltage Version)	EXAMPLE (Fixed Output Voltage Version)
4. Input Capacitor (C <sub>IN</sub> )	4. Input Capacitor (C <sub>IN</sub> )
<b>4. Input Capacitor (C</b> <sub>IN</sub> ) A low ESR aluminum or tantalum bypass capacitor is needed between the input pin and ground to prevent large voltage transients from appearing at the input. This capacitor should be located close to the IC using short leads. In addition, the RMS current rating of the input capacitor should be selected to be at least $\frac{1}{2}$ the DC load current. The capacitor manufacturer data sheet must be checked to assure that this current rating is not exceeded. The curves shown in <i>Figure 14</i> show typical RMS current ratings for several different aluminum electrolytic capacitor values. A parallel connection of two or more capacitors may be required to increase the total minimum RMS current rating to suit the application requirements. For an aluminum electrolytic capacitor, the voltage rating should be at least 1.25 times the maximum input voltage. Caution must be exercised if solid tantalum capacitors are used. The tantalum capacitor voltage rating should be twice the maximum input voltage. The tables in <i>Figure 15</i> show the recommended application voltage for AVX TPS and Sprague 594D tantalum capacitors. It is also recommended that they be surge current tested by the manufacturer. The TPS series available from AVX, and the 593D and 594D series from Sprague are all surge current tested. Another approach to minimize the surge current stresses on the input capacitor is to add a small inductor in series with the input supply line. Use caution when using ceramic capacitors for input bypassing, because it may cause severe ringing at the V <sub>IN</sub> pin.	<b>4. Input Capacitor (C</b> <sub>IN</sub> ) The important parameters for the input capacitor are the input voltage rating and the RMS current rating. With a maximum input voltage of 12V, an aluminum electrolytic capacitor with a voltage rating greater than 15V (1.25 x V <sub>IN</sub> ) would be needed. The next higher capacitor voltage rating is 16V. The RMS current rating requirement for the input capacitor in a buck regulator is approximately ½ the DC load current. In this example, with a 1A load, a capacitor with a RMS current rating of at least 500 mA is needed. The curves shown in <i>Figure 14</i> can be used to select an appropriate input capacitor. From the curves, locate the 16V line and note which capacitor values have RMS current ratings greater than 500 mA. For a through hole design, a 330 µF/16V electrolytic capacitor (Panasonic HFQ series, Nichicon PL, Sanyo MV-GX series or equivalent) would be adequate. Other types or other manufacturers' capacitors can be used provided the RMS ripple current ratings are adequate. Additionally, for a complete surface mount design, electrolytic capacitors such as the Sanyo CV-C or CV-BS and the Nichicon WF or UR and the NIC Components NACZ series could be considered. For surface mount designs, solid tantalum capacitors can be used, but caution must be exercised with regard to the capacitor surge current rating and voltage rating. In this example, checking <i>Figure 15</i> , and the Sprague 594D series datasheet, a Sprague 594D 15 µF, 25V capacitor is adequate.
5. Boost Capacitor ( $C_B$ ) This capacitor develops the necessary voltage to turn the switch gate on fully. All applications should use a 0.01 $\mu$ F, 50V ceramic capacitor.	5. Boost Capacitor ( $C_B$ ) For this application, and all applications, use a 0.01 $\mu$ F, 50V ceramic capacitor.
6. Soft-Start Capacitor (C <sub>SS</sub> - optional)	6. Soft-Start Capacitor (C <sub>SS</sub> - optional)
This capacitor controls the rate at which the device starts up. The formula for the soft-start capacitor $C_{SS}$ is:	For this application, selecting a start-up time of 10 ms and using the formula for $C_{SS}$ results in a value of:
$C_{SS} \approx (I_{SS} \cdot t_{SS}) / [V_{SSTH} + 2.6V \cdot (\frac{V_{OUT} + V_{SCHOTTKY}}{V_{IN}})]$	$C_{SS} ≈ (4.5 μA \cdot 10 ms) / [0.63V + 2.6V \cdot (\frac{5V + 0.4V}{12V})]$ = 25 nF ≈ 0.022 μF.
where: $I_{SS} = Soft-Start Current :4.5 \ \mu A \ typical.$ $t_{SS} = Soft-Start Time :Selected.$ $V_{SSTH} = Soft-Start Threshold Voltage :0.63V \ typical.$ $V_{OUT} = Output \ Voltage :Selected.$ $V_{SCHOTTKY} = Schottky \ Diode \ Voltage \ Drop :0.4V \ typical.$ $V_{IN} = Input \ Voltage \ :Selected.$ If this feature is not desired, leave this pin open.	

# LM2672EP Series Buck Regulator Design Procedure (Fixed Output)

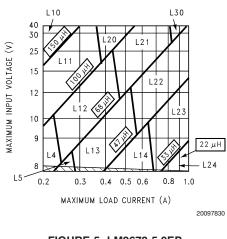
(Continued)

PROCEDURE (Fixed Output Voltage Version)	EXAMPLE (Fixed Output Voltage Version)
7. Frequency Synchronization (optional)	7. Frequency Synchronization (optional)
The LM2672EP (oscillator) can be synchronized to run with an	For all applications, use a 1 k $\Omega$ resistor and a 100 pF capacito
external oscillator, using the sync pin (pin 3). By doing so, the	for the RC filter.
LM2672EP can be operated at higher frequencies than the	
standard frequency of 260 kHz. This allows for a reduction in	
the size of the inductor and output capacitor.	
As shown in the drawing below, a signal applied to a RC filter at	
the sync pin causes the device to synchronize to the frequency	
of that signal. For a signal with a peak-to-peak amplitude of 3V	
or greater, a 1 k $\!\Omega$ resistor and a 100 pF capacitor are suitable	
values.	
$V_{S} \ge 3.0V$ LM2672	

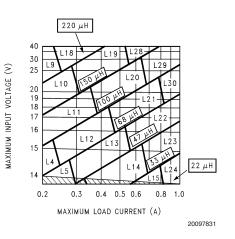
#### Inductor Value Selection Guides (For Continuous Mode Operation)



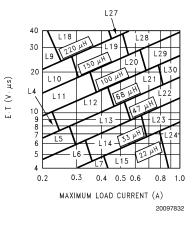








#### FIGURE 6. LM2672-12EP



#### FIGURE 7. LM2672-ADJEP

Ind.	Inductors	Current	Scl	hott	Rend	0	Pulse E	ngineering	Coilcraft
Ref. Desg.	Inductance (µH)	Current (A)	Through Hole	Surface Mount	Through Hole	Surface Mount	Through Hole	Surface Mount	Surface Mount
L4	68	0.32	67143940	67144310	RL-1284-68-43	RL1500-68	PE-53804	PE-53804-S	DO1608-68
			67148310	67144310	RL-1204-00-43 RL-1284-47-43	RL1500-66 RL1500-47	PE-53804 PE-53805		DO1608-66
L5 L6	47	0.37	67148310	67148420	RL-1284-47-43 RL-1284-33-43	RL1500-47 RL1500-33		PE-53805-S PE-53806-S	DO1608-47
-	33	0.44					PE-53806		
L7	22	0.52	67148330	67148440	RL-1284-22-43	RL1500-22	PE-53807	PE-53807-S	DO1608-22
L9	220	0.32	67143960	67144330	RL-5470-3	RL1500-220		PE-53809-S	DO3308-22
L10	150	0.39	67143970	67144340	RL-5470-4	RL1500-150		PE-53810-S	DO3308-15
L11	100	0.48	67143980	67144350	RL-5470-5	RL1500-100		PE-53811-S	DO3308-10
L12	68	0.58	67143990	67144360	RL-5470-6	RL1500-68	PE-53812	PE-53812-S	DO3308-68
L13	47	0.70	67144000	67144380	RL-5470-7	RL1500-47	PE-53813	PE-53813-S	DO3308-47
L14	33	0.83	67148340	67148450	RL-1284-33-43	RL1500-33	PE-53814	PE-53814-S	DO3308-33
L15	22	0.99	67148350	67148460	RL-1284-22-43	RL1500-22	PE-53815	PE-53815-S	DO3308-22
L18	220	0.55	67144040	67144420	RL-5471-2	RL1500-220	PE-53818	PE-53818-S	DO3316-22
L19	150	0.66	67144050	67144430	RL-5471-3	RL1500-150	PE-53819	PE-53819-S	DO3316-15
L20	100	0.82	67144060	67144440	RL-5471-4	RL1500-100	PE-53820	PE-53820-S	DO3316-10
L21	68	0.99	67144070	67144450	RL-5471-5	RL1500-68	PE-53821	PE-53821-S	DO3316-68
L22	47	1.17	67144080	67144460	RL-5471-6		PE-53822	PE-53822-S	DO3316-47
L23	33	1.40	67144090	67144470	RL-5471-7		PE-53823	PE-53823-S	DO3316-33
L24	22	1.70	67148370	67148480	RL-1283-22-43	_	PE-53824	PE-53824-S	DO3316-22
L27	220	1.00	67144110	67144490	RL-5471-2		PE-53827	PE-53827-S	DO5022P-2
L28	150	1.20	67144120	67144500	RL-5471-3		PE-53828	PE-53828-S	DO5022P-1
L29	100	1.47	67144130	67144510	RL-5471-4		PE-53829	PE-53829-S	DO5022P-1
L30	68	1.78	67144140	67144520	RL-5471-5		PE-53830	PE-53830-S	DO5022P-6

FIGURE 8. Inductor Manufacturers' Part Numbers

Coilcraft Inc.	Phone	(800) 322-2645
	FAX	(708) 639-1469
Coilcraft Inc., Europe	Phone	+44 1236 730 595
	FAX	+44 1236 730 627
Pulse Engineering Inc.	Phone	(619) 674-8100
	FAX	(619) 674-8262
Pulse Engineering Inc.,	Phone	+353 93 24 107
Europe	FAX	+353 93 24 459
Renco Electronics Inc.	Phone	(800) 645-5828
	FAX	(516) 586-5562
Schott Corp.	Phone	(612) 475-1173
	FAX	(612) 475-1786

FIGURE 9. Inductor Manufacturers' Phone Numbers

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#### Inductor Manufacturers' Part Numbers (Continued)

		Output Capacitor					
Output		Surface Mount		Through Hole			
Voltage	Inductance (µH)	Sprague	AVX TPS	Sanyo OS-CON	Sanyo MV-GX	Nichicon	Panasonic
(V)	(μπ)	594D Series	Series	SA Series	Series	PL Series	HFQ Series
		(µF/V)	(µF/V)	(µF/V)	(µF/V)	(µF/V)	(µF/V)
	22	120/6.3	100/10	100/10	330/35	330/35	330/35
	33	120/6.3	100/10	68/10	220/35	220/35	220/35
3.3	47	68/10	100/10	68/10	150/35	150/35	150/35
3.3	68	120/6.3	100/10	100/10	120/35	120/35	120/35
	100	120/6.3	100/10	100/10	120/35	120/35	120/35
	150	120/6.3	100/10	100/10	120/35	120/35	120/35
5.0	22	100/16	100/10	100/10	330/35	330/35	330/35
	33	68/10	10010	68/10	220/35	220/35	220/35
	47	68/10	100/10	68/10	150/35	150/35	150/35
5.0	68	100/16	100/10	100/10	120/35	120/35	120/35
	100	100/16	100/10	100/10	120/35	120/35	120/35
	150	100/16	100/10	100/10	120/35	120/35	120/35
	22	120/20	(2x) 68/20	68/20	330/35	330/35	330/35
	33	68/25	68/20	68/20	220/35	220/35	220/35
	47	47/20	68/20	47/20	150/35	150/35	150/35
12	68	47/20	68/20	47/20	120/35	120/35	120/35
	100	47/20	68/20	47/20	120/35	120/35	120/35
	150	47/20	68/20	47/20	120/35	120/35	120/35
	220	47/20	68/20	47/20	120/35	120/35	120/35

FIGURE 10. Output Capacitor Table

Nichicon Corp.	Phone	(847) 843-7500
	FAX	(847) 843-2798
Panasonic	Phone	(714) 373-7857
	FAX	(714) 373-7102
AVX Corp.	Phone	(803) 448-9411
	FAX	(803) 448-1943
Sprague/Vishay	Phone	(207) 324-4140
	FAX	(207) 324-7223
Sanyo Corp.	Phone	(619) 661-6322
	FAX	(619) 661-1055

FIGURE 11. Capacitor Manufacturers' Phone Numbers

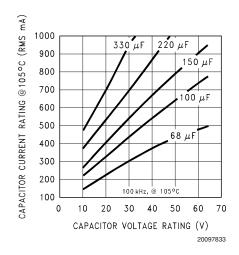
#### Inductor Manufacturers' Part Numbers (Continued)

	1A Diodes		3A Di	iodes
VR	Surface	Through	Surface	Through
	Mount	Hole	Mount	Hole
20V	SK12	1N5817	SK32	1N5820
	B120	SR102		SR302
30V	SK13	1N5818	SK33	1N5821
	B130	11DQ03	30WQ03F	31DQ03
	MBRS130	SR103		
40V	SK14	1N5819	SK34	1N5822
	B140	11DQ04	30BQ040	MBR340
	MBRS140	SR104	30WQ04F	31DQ04
	10BQ040		MBRS340	SR304
	10MQ040		MBRD340	
	15MQ040			
50V	SK15	MBR150	SK35	MBR350
	B150	11DQ05	30WQ05F	31DQ05
	10BQ050	SR105		SR305

FIGURE 12. Schottky Diode Selection Table

International Rectifier Corp.	Phone	(310) 322-3331
	FAX	(310) 322-3332
Motorola, Inc.	Phone	(800) 521-6274
	FAX	(602) 244-6609
General Instruments Corp.	Phone	(516) 847-3000
	FAX	(516) 847-3236
Diodes, Inc.	Phone	(805) 446-4800
	FAX	(805) 446-4850

FIGURE 13. D	Diode Manufacturers'	<b>Phone Numbers</b>
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### Inductor Manufacturers' Part Numbers (Continued)

AVX	TPS
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Recommended Application Voltage	Voltage Rating
+85°C Rati	ng
3.3	6.3
5	10
10	20
12	25
15	35

#### Sprague 594D

Recommended Application Voltage	Voltage Rating	
+85°C Rating		
2.5	4	
3.3	6.3	
5	10	
8	16	
12	20	
18	25	
24	35	
29	50	

FIGURE 15. Recommended Application Voltage for AVX TPS and Sprague 594D Tantalum Chip Capacitors Derated for 85°C.

# LM2672EP Series Buck Regulator Design Procedure (Adjustable Output)

PROCEDURE (Adjustable Output Voltage Version)	EXAMPLE (Adjustable Output Voltage Version)
To simplify the buck regulator design procedure, National	
Semiconductor is making available computer design software to	
be used with the SIMPLE SWITCHER line of switching	
regulators. LM267X Made Simple version 6.0 is available on	
Windows 3.1, NT, or 95 operating systems.	
Given:	Given:
V <sub>OUT</sub> = Regulated Output Voltage	$V_{OUT} = 20V$
V <sub>IN</sub> (max) = Maximum Input Voltage	$V_{IN}(max) = 28V$
I <sub>LOAD</sub> (max) = Maximum Load Current	$I_{LOAD}(max) = 1A$
F = Switching Frequency (Fixed at a nominal 260 kHz).	F = Switching Frequency (Fixed at a nominal 260 kHz).
<b>1. Programming Output Voltage</b> (Selecting $R_1$ and $R_2$ , as shown in <i>Figure 3</i> )	<b>1. Programming Output Voltage</b> (Selecting $R_1$ and $R_2$ , as shown in <i>Figure 3</i> )
Use the following formula to select the appropriate resistor	Select $R_1$ to be 1 k $\Omega$ , 1%. Solve for $R_2$ .
values.	
$V_{OUT} = V_{REF} \left(1 + \frac{R_2}{R_1}\right)$ where $V_{REF} = 1.21V$	$R_{2} = R_{1} \left( \frac{V_{OUT}}{V_{REF}} - 1 \right) = 1 k\Omega \left( \frac{20V}{1.23V} - 1 \right)$
Select a value for $R_1$ between 240 $\Omega$ and 1.5 k $\Omega.$ The lower	$R_2 = 1 k\Omega (16.53 - 1) = 15.53 kΩ$ , closest 1% value is 15.4 kΩ.
resistor values minimize noise pickup in the sensitive feedback	$R_2 = 15.4 \text{ k}\Omega.$
pin. (For the lowest temperature coefficient and the best stability	
with time, use 1% metal film resistors.)	

# LM2672EP Series Buck Regulator Design Procedure (Adjustable Output) (Continued)

PROCEDURE (Adjustable Output Voltage Version)	EXAMPLE (Adjustable Output Voltage Version)
$R_{2} = R_{1} \left( \frac{V_{OUT}}{V_{REF}} - 1 \right)$	
2. Inductor Selection (L1)	2. Inductor Selection (L1)
A. Calculate the inductor Volt • microsecond constant E • T	A. Calculate the inductor Volt • microsecond constant (E • T),
(V • μs), from the following formula:	
$E \cdot T = (V_{\text{IN}(\text{MAX})} - V_{\text{OUT}} - V_{\text{SAT}}) \cdot \frac{V_{\text{OUT}} + V_{\text{D}}}{V_{\text{IN}(\text{MAX})} - V_{\text{SAT}} + V_{\text{D}}} \cdot \frac{1000}{260} (V \cdot \mu s)$	$E \cdot T = (28 - 20 - 0.25) \cdot \frac{20 + 0.5}{28 - 0.25 + 0.5} \cdot \frac{1000}{260} (V \cdot \mu s)$
	$E \cdot T = (7.75) \cdot \frac{20.5}{28.25} \cdot 3.85 (V \cdot \mu s) = 21.6 (V \cdot \mu s)$
where V <sub>SAT</sub> =internal switch saturation voltage=0.25V and	
$V_{\rm D}$ = diode forward voltage drop = 0.5V	
<b>B.</b> Use the E • T value from the previous formula and match it	<b>B.</b> E • T = 21.6 (V • μs)
with the E • T number on the vertical axis of the Inductor Value	- (
Selection Guide shown in Figure 7.	
<b>C.</b> On the horizontal axis, select the maximum load current.	<b>C.</b> I <sub>LOAD</sub> (max) = 1A
<b>D.</b> Identify the inductance region intersected by the $E \bullet T$ value	<b>D.</b> From the inductor value selection guide shown in <i>Figure 7</i> ,
and the Maximum Load Current value. Each region is identified	the inductance region intersected by the 21.6 (V $\bullet$ µs) horizontal
by an inductance value and an inductor code (LXX).	line and the 1A vertical line is 68 $\mu$ H, and the inductor code is
	L30.
E. Select an appropriate inductor from the four manufacturer's	E. From the table in <i>Figure 8</i> , locate line L30, and select an
part numbers listed in Figure 8. For information on the different	inductor part number from the list of manufacturers' part
types of inductors, see the inductor selection in the fixed output	numbers.
voltage design procedure.	
3. Output Capacitor Selection (C <sub>OUT</sub> )	3. Output Capacitor Selection (C <sub>OUT</sub> )
A. Select an output capacitor from the capacitor code selection	A. Use the appropriate row of the capacitor code selection
guide in Figure 16. Using the inductance value found in the	guide, in Figure 16. For this example, use the 15-20V row. The
inductor selection guide, step 1, locate the appropriate capacitor	capacitor code corresponding to an inductance of 68 µH is C20.
code corresponding to the desired output voltage.	
<b>B.</b> Select an appropriate capacitor value and voltage rating,	<b>B.</b> From the output capacitor selection table in <i>Figure 17</i> ,
using the capacitor code, from the output capacitor selection	choose a capacitor value (and voltage rating) that intersects the
table in <i>Figure 17</i> . There are two solid tantalum (surface mount)	capacitor code(s) selected in section A, C20.
capacitor manufacturers and four electrolytic (through hole)	The capacitance and voltage rating values corresponding to the
capacitor manufacturers to choose from. It is recommended that	capacitor code C20 are the:
both the manufacturers and the manufacturer's series that are	Surface Mount:
listed in the table be used. A table listing the manufacturers'	33 μF/25V Sprague 594D Series.
phone numbers is located in <i>Figure 11</i> .	33 μF/25V AVX TPS Series.
	Through Hole:
	33 μF/25V Sanyo OS-CON SC Series. 120 μF/35V Sanyo MV-GX Series.
	$120 \ \mu\text{F}/35V$ Saliyo WV-GX Series.
	$120 \mu\text{F}/35V$ Panasonic HFQ Series.
	Other manufacturers or other types of capacitors may also be
	used, provided the capacitor specifications (especially the 100
	kHz ESR) closely match the characteristics of the capacitors
	listed in the output capacitor table. Refer to the capacitor
	manufacturers' data sheet for this information.

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#### LM2672EP Series Buck Regulator Design Procedure (Adjustable Output) (Continued)

PROCEDURE (Adjustable Output Voltage Version)	EXAMPLE (Adjustable Output Voltage Version)
4. Catch Diode Selection (D1)	4. Catch Diode Selection (D1)
<b>A.</b> In normal operation, the average current of the catch diode is the load current times the catch diode duty cycle, 1-D (D is the switch duty cycle, which is approximately $V_{OUT}/V_{IN}$ ). The largest value of the catch diode average current occurs at the maximum input voltage (minimum D). For normal operation, the catch diode current rating must be at least 1.3 times greater than its maximum average current. However, if the power supply design must withstand a continuous output short, the diode should have a current rating greater than the maximum current limit of the LM2672EP. The most stressful condition for this diode is a shorted output condition. <b>B.</b> The reverse voltage rating of the diode should be at least 1.25 times the maximum input voltage.	<b>A.</b> Refer to the table shown in <i>Figure 12</i> . Schottky diodes provide the best performance, and in this example a 1A, 40V Schottky diode would be a good choice. If the circuit must withstand a continuous shorted output, a higher current (at leas 2.2A) Schottky diode is recommended.
<b>C.</b> Because of their fast switching speed and low forward voltage drop, Schottky diodes provide the best performance and efficiency. The Schottky diode must be located close to the LM2672EP using short leads and short printed circuit traces.	
5. Input Capacitor (C <sub>IN</sub> )	5. Input Capacitor (C <sub>IN</sub> )
A low ESR aluminum or tantalum bypass capacitor is needed between the input pin and ground to prevent large voltage transients from appearing at the input. This capacitor should be located close to the IC using short leads. In addition, the RMS current rating of the input capacitor should be selected to be at least $\frac{1}{2}$ the DC load current. The capacitor manufacturer data sheet must be checked to assure that this current rating is not exceeded. The curves shown in <i>Figure 14</i> show typical RMS current ratings for several different aluminum electrolytic capacitor values. A parallel connection of two or more capacitors may be required to increase the total minimum RMS current rating to suit the application requirements. For an aluminum electrolytic capacitor, the voltage rating should be at least 1.25 times the maximum input voltage. Caution must be exercised if solid tantalum capacitors are used. The tantalum capacitor voltage rating should be twice the maximum input voltage. The tables in <i>Figure 15</i> show the recommended application voltage for AVX TPS and Sprague 594D tantalum capacitors. It is also recommended that they be surge current tested by the manufacturer. The TPS series available from AVX, and the 593D and 594D series from Sprague are all surge current tested. Another approach to minimize the surge current stresses on the input capacitor is to add a small inductor in series with the input supply line. Use caution when using ceramic capacitors for input bypassing, because it may cause severe ringing at the V <sub>IN</sub> pin. <b>6. Boost Capacitor (C<sub>B</sub>)</b>	The important parameters for the input capacitor are the input voltage rating and the RMS current rating. With a maximum input voltage of 28V, an aluminum electrolytic capacitor with a voltage rating of at least 35V ( $1.25 \times V_{IN}$ ) would be needed. The RMS current rating requirement for the input capacitor in a buck regulator is approximately ½ the DC load current. In this example, with a 1A load, a capacitor with a RMS current rating of at least 500 mA is needed. The curves shown in <i>Figure 14</i> can be used to select an appropriate input capacitor. From the curves, locate the 35V line and note which capacitor values have RMS current ratings greater than 500 mA. For a through hole design, a 330 µF/35V electrolytic capacitor (Panasonic HFQ series, Nichicon PL, Sanyo MV-GX series or equivalent) would be adequate. Other types or other manufacturers' capacitors can be used provided the RMS ripple current ratings are adequate. Additionally, for a complete surface mount design, electrolytic capacitors such as the Sanyo CV-C or CV-BS and the Nichicon WF or UR and the NIC Components NACZ series could be considered. For surface mount designs, solid tantalum capacitors can be used, but caution must be exercised with regard to the capacito surge current rating and voltage rating. In this example, checking <i>Figure 15</i> , and the Sprague 594D series datasheet, a Sprague 594D 15 µF, 50V capacitor is adequate.
This capacitor develops the necessary voltage to turn the switch gate on fully. All applications should use a 0.01 $\mu$ F, 50V ceramic capacitor. If the soft-start and frequency synchronization features are desired, look at steps 6 and 7 in the fixed output design procedure.	For this application, and all applications, use a 0.01 $\mu$ F, 50V ceramic capacitor.

# LM2672EP Series Buck Regulator Design Procedure (Adjustable Output) (Continued)

Case	Output	Inductance (µH)							
Style (Note 10)	Voltage (V)	22	33	47	68	100	150	220	
SM and TH	1.21-2.50		_	_	_	C1	C2	C3	
SM and TH	2.50-3.75	_	_	_	C1	C2	C3	C3	
SM and TH	3.75-5.0	_	_	C4	C5	C6	C6	C6	
SM and TH	5.0-6.25	_	C4	C7	C6	C6	C6	C6	
SM and TH	6.25-7.5	C8	C4	C7	C6	C6	C6	C6	
SM and TH	7.5–10.0	C9	C10	C11	C12	C13	C13	C13	
SM and TH	10.0-12.5	C14	C11	C12	C12	C13	C13	C13	
SM and TH	12.5-15.0	C15	C16	C17	C17	C17	C17	C17	
SM and TH	15.0–20.0	C18	C19	C20	C20	C20	C20	C20	
SM and TH	20.0-30.0	C21	C22	C22	C22	C22	C22	C22	
TH	30.0-37.0	C23	C24	C24	C25	C25	C25	C25	

Note 10: SM - Surface Mount, TH - Through Hole

FIGURE 16. Capacitor Code Selection Guide

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#### LM2672EP Series Buck Regulator Design Procedure (Adjustable Output) (Continued)

Output Capacitor								
Cap.	Surface Mount			Through Hole				
Ref.	Sprague	AVX TPS	Sanyo OS-CON	Sanyo MV-GX	Nichicon	Panasonic		
Desg.	594D Series	Series	SA Series	Series	PL Series	HFQ Series		
#	(µF/V)	(µF/V)	(µF/V)	(µF/V)	(µF/V)	(µF/V)		
C1	120/6.3	100/10	100/10	220/35	220/35	220/35		
C2	120/6.3	100/10	100/10	150/35	150/35	150/35		
C3	120/6.3	100/10	100/35	120/35	120/35	120/35		
C4	68/10	100/10	68/10	220/35	220/35	220/35		
C5	100/16	100/10	100/10	150/35	150/35	150/35		
C6	100/16	100/10	100/10	120/35	120/35	120/35		
C7	68/10	100/10	68/10	150/35	150/35	150/35		
C8	100/16	100/10	100/10	330/35	330/35	330/35		
C9	100/16	100/16	100/16	330/35	330/35	330/35		
C10	100/16	100/16	68/16	220/35	220/35	220/35		
C11	100/16	100/16	68/16	150/35	150/35	150/35		
C12	100/16	100/16	68/16	120/35	120/35	120/35		
C13	100/16	100/16	100/16	120/35	120/35	120/35		
C14	100/16	100/16	100/16	220/35	220/35	220/35		
C15	47/20	68/20	47/20	220/35	220/35	220/35		
C16	47/20	68/20	47/20	150/35	150/35	150/35		
C17	47/20	68/20	47/20	120/35	120/35	120/35		
C18	68/25	(2x) 33/25	47/25 (Note 11)	220/35	220/35	220/35		
C19	33/25	33/25	33/25 (Note 11)	150/35	150/35	150/35		
C20	33/25	33/25	33/25 (Note 11)	120/35	120/35	120/35		
C21	33/35	(2x) 22/25	(Note 12)	150/35	150/35	150/35		
C22	33/35	22/35	(Note 12)	120/35	120/35	120/35		
C23	(Note 12)	(Note 12)	(Note 12)	220/50	100/50	120/50		
C24	(Note 12)	(Note 12)	(Note 12)	150/50	100/50	120/50		
C25	(Note 12)	(Note 12)	(Note 12)	150/50	82/50	82/50		

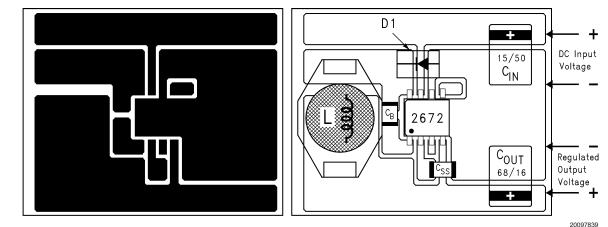
Note 11: The SC series of Os-Con capacitors (others are SA series)

Note 12: The voltage ratings of the surface mount tantalum chip and Os-Con capacitors are too low to work at these voltages.

FIGURE 17. Output Capacitor Selection Table

# **Application Information**

TYPICAL SURFACE MOUNT PC BOARD LAYOUT, FIXD OUTPUT (4X SIZE)



 $C_{IN}$  - 15  $\mu F,$  50V, Solid Tantalum Sprague, "594D series"  $C_{OUT}$  - 68  $\mu F,$  16V, Solid Tantalum Sprague, "594D series"

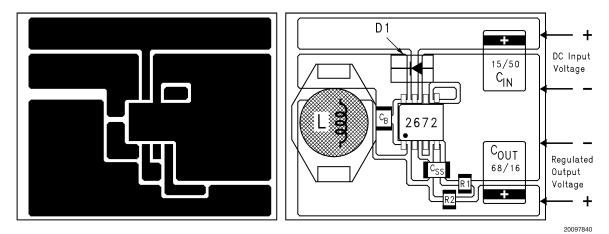
COUT - 00 µF, 10V, Solid Tantaidin Sprague, 594D St

D1 - 1A, 40V Schottky Rectifier, Surface Mount

L1 - 33 µH, L23, Coilcraft DO3316

 $C_B$  - 0.01  $\mu F,\,50V,\,Ceramic$ 

#### TYPICAL SURFACE MOUNT PC BOARD LAYOUT, ADJUSTABLE OUTPUT (4X SIZE)



C<sub>IN</sub> - 15 µF, 50V, Solid Tantalum Sprague, "594D series"

- $C_{OUT}$  33  $\mu\text{F},$  25V, Solid Tantalum Sprague, "594D series"
- D1 1A, 40V Schottky Rectifier, Surface Mount
- L1 68 µH, L30, Coilcraft DO3316
- $C_B$  0.01  $\mu F,\,50V,\,Ceramic$
- R1 1k, 1%

R2 - Use formula in Design Procedure

#### FIGURE 18. PC Board Layout

Layout is very important in switching regulator designs. Rapidly switching currents associated with wiring inductance can generate voltage transients which can cause problems. For minimal inductance and ground loops, the wires indicated by heavy lines (in *Figure 2* and *Figure 3*) should be wide printed circuit traces and should be kept as short as possible. For best results, external components should be located as close to the switcher IC as possible using ground plane construction or single point grounding. If **open core inductors are used**, special care must be taken as to the location and positioning of this type of inductor. Allowing the inductor flux to intersect sensitive feedback, IC ground path, and  $C_{OUT}$  wiring can cause problems.

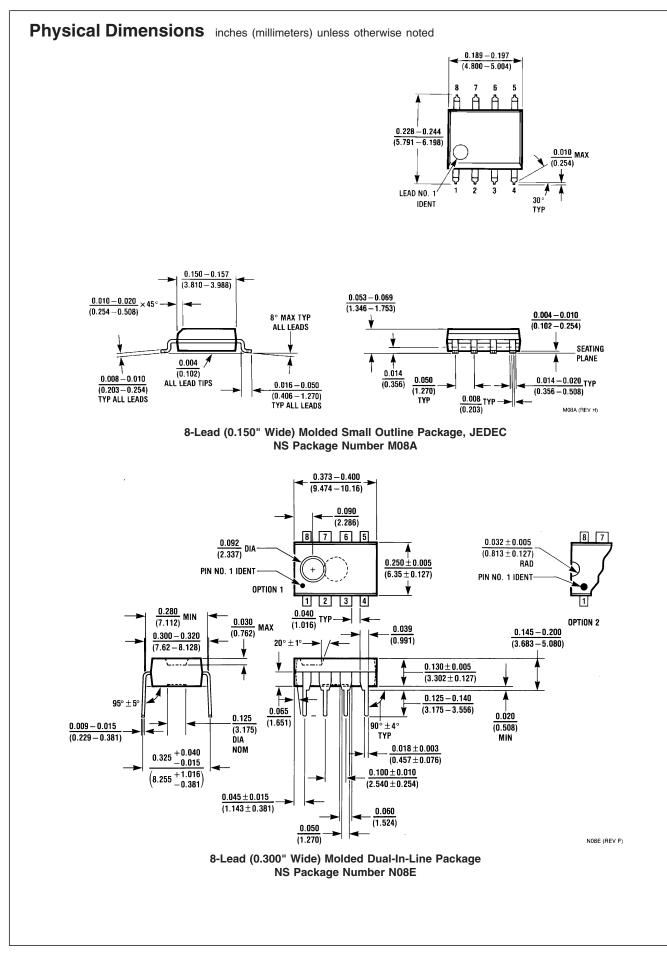
When using the adjustable version, special care must be taken as to the location of the feedback resistors and the associated wiring. Physically locate both resistors near the IC, and route the wiring away from the inductor, especially an open core type of inductor.

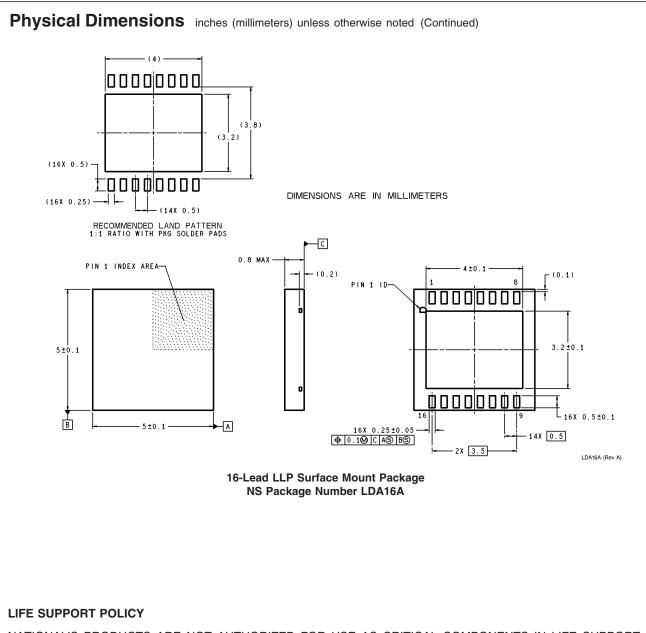
# Application Information (Continued)

#### LLP PACKAGE DEVICES

The LM2672EP may be offered in the 16 lead LLP surface mount package to allow for increased power dissipation compared to the SO-8 and DIP.

The Die Attach Pad (DAP) can and should be connected to PCB Ground plane/island. For CAD and assembly guidelines refer to Application Note AN-1187 at http://power.national.com.





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