

LM144/LM344 High Voltage, High Slew Rate Operational Amplifier

General Description

The LM144 is a general purpose high voltage, uncompensated operational amplifier featuring operation to $\pm 36\text{V}$, complete input overvoltage protection up to the supply voltages and input currents comparable to those of other super- β op amps. Increased slew rate, together with high common-mode and supply rejection, insure excellent performance at high supply voltages. Operating characteristics, in particular supply current, slew rate and gain, are virtually independent of supply voltage and temperature. Furthermore, due to thermal symmetry on the die, gain is unaffected by output loading at high supply voltages.

With the unique advantages of low input current, high gain, and high slew rate, the LM144 can increase accuracy and useful frequency range in many existing applications. For example, the LM144 is a plug-in replacement for the LM101A, as well as other general purpose op amps.

The LM144 can be compensated with a single capacitor, thus giving the user the ability to optimize ac parameters to suit the application. For example, in applications such as audio power amplifiers, the LM144 with a gain of 10 can provide a $\pm 30 \text{V}$ output swing, a slew rate of approximately $30 \text{V}/\mu s$, and a 120 kHz full power bandwidth.

In applications where capacitive loads or cables must be driven, the LM144 can be overcompensated for increased stability.

The LM344 is similar to the LM144 for applications in less severe supply voltage and temperature environments.

Features

- External compensation provides large power bandwidth (A_V ≥ 10)
- 120 kHz
- Wide operating voltage range
- $\pm 4.0 V$ to $\pm 36 V$
- Large output voltage swing

±30V

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- Wide input common-mode range
- Input overvoltage protection
- Electrical characteristics independent of supply voltage and temperature

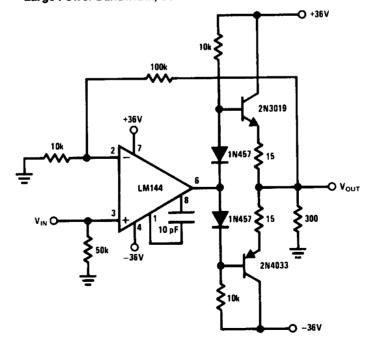
Unique Characteristics

- Low input bias current 8.0 nA Low input offset current 1.0 nA High slew rate (A $_{V} \ge 10$) 30V/ μ s
- High slew rate (A_V ≥ 10)
 High voltage gain
 100k min
- Offset voltage null capability

Typical Application

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Large Power Bandwidth, Current Boosted Audio Line Driver



TL/H/7784-1

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Absolute Maximum Ratings (These ratings are not concurrent)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/Distributors for availability and specifications. (Note 4)

LM144 LM344 Supply Voltage ±40V ±34V Power Dissipation (Note 1) 680 mW 680 mW Differential Input Voltage (Note 2) 80V 68V Input Voltage (Note 2) ±40V ±34V Operating Temperature Range -55°C to +125°C 0°C to +70°C Storage Temperature Range -65° C to $+150^{\circ}$ C -65°C to +150°C Output Short Circuit Duration 5 seconds 5 seconds Lead Temperature (Soldering, 10 sec) 300°C 300°C ESD rating to be determined.

Electrical Characteristics (Note 3)

Parameter	Conditions	LM144			LM344			Units
		Min	Тур	Max	Min	Тур	Max	Onito
Input Offset Voltage	T _A = 25°C		2.0	5.0		2.0	8.0	m∨
Input Offset Current	$T_A = 25^{\circ}C$		1.0	3.0		1.0	10	nA
Input Bias Current	T _A = 25°C		8.0	20		8.0	40	nA
Supply Voltage Rejection Ratio	T _A = 25°C		10	100		10	200	μV/V
Output Voltage Swing	$T_A = 25^{\circ}C, R_L \ge 5 k\Omega$	22	25		20	25		V
Large Signal Voltage Gain	$T_A = 25^{\circ}C$, $V_{OUT} = \pm 10V$, $R_L \ge 100 \text{ k}\Omega$	100k	180k		70k	180k		V/V
Common-Mode Rejection Ratio	T _A = 25°C	80	90		70	90		dB
Cinput Voltage Range	T _A = 25°C Data	Sh 24 t4l	J.C 26 1		22	26		V
Supply Current	$T_A = 25^{\circ}C$		2.0	4.0		2.0	5.0	mA
Short Circuit Current	T _A = 25°C		20			20		mA
Slew Rate	$T_A = 25^{\circ}C, A_V = 1$ $T_A = 25^{\circ}C, A_V = 10, C1 = 3 pF$		2.5 30			2.5 30		V/μs V/μs
Power Bandwidth	$T_A = 25^{\circ}\text{C}, V_{OUT} = 40 V_{p-p},$ $R_L = 5 \text{ k}\Omega, \text{THD} \le 1\%, A_V = 1$		20			20		kHz
Unity Gain Frequency	T _A = 25°C		1.0			1.0		MHz
Input Offset Voltage	T _A = Max T _A = Min			6.0 6.0			10 10	mV mV
Input Offset Current	T _A = Max T _A = Min		0.8 1.8	4.5 7.0		0.8 1.8	14 14	nA nA
Input Bias Current	T _A = Max T _A = Min		5.0 16	35 35		5.0 16	55 55	nA nA
Large Signal Voltage Gain	$R_L \ge 100 \text{ k}\Omega$, $T_A = \text{Max}$ $R_L \ge 100 \text{ k}\Omega$, $T_A = \text{Min}$	50k 50k	150k 220k		50k 50k	150k 220k		V/V V/V
Output Voltage Swing	$\begin{aligned} & \textbf{R}_{L} \geq 5.0 \text{ k}\Omega, \textbf{T}_{\textbf{A}} = \textbf{Max} \\ & \textbf{R}_{L} \geq 5.0 \text{ k}\Omega, \textbf{T}_{\textbf{A}} = \textbf{Min} \end{aligned}$	22 22	26 25		20 20	26 25		V V

Note 1: The maximum junction temperature of the LM144 is 150°C, while that of the LM344 is 100°C. For operating at elevated temperatures, devices in the H08 package must be derated based on a thermal resistance of 150°C/W, junction to ambient, or 20°C/W, junction to case.

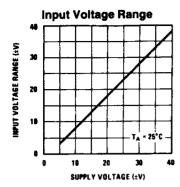
Note 2: For supply voltage less than \pm 40V for the LM144 and less than \pm 34V for the LM344, the absolute maximum input voltage is equal to the supply voltage. Note 3: These specifications apply for $V_S = \pm 28V$. For the LM144, $T_A = max = 125^{\circ}C$ and $T_A = min = -55^{\circ}C$. For the LM344, $T_A = max = 70^{\circ}C$ and $T_A = min = 0^{\circ}C$.

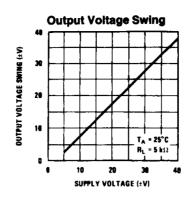
Note 4: Refer to RETS144X for LM144H specifications.

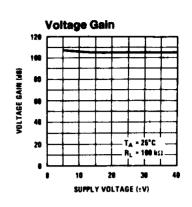
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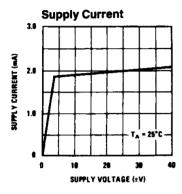


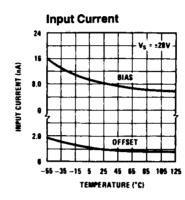
Typical Performance Characteristics

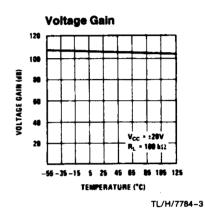








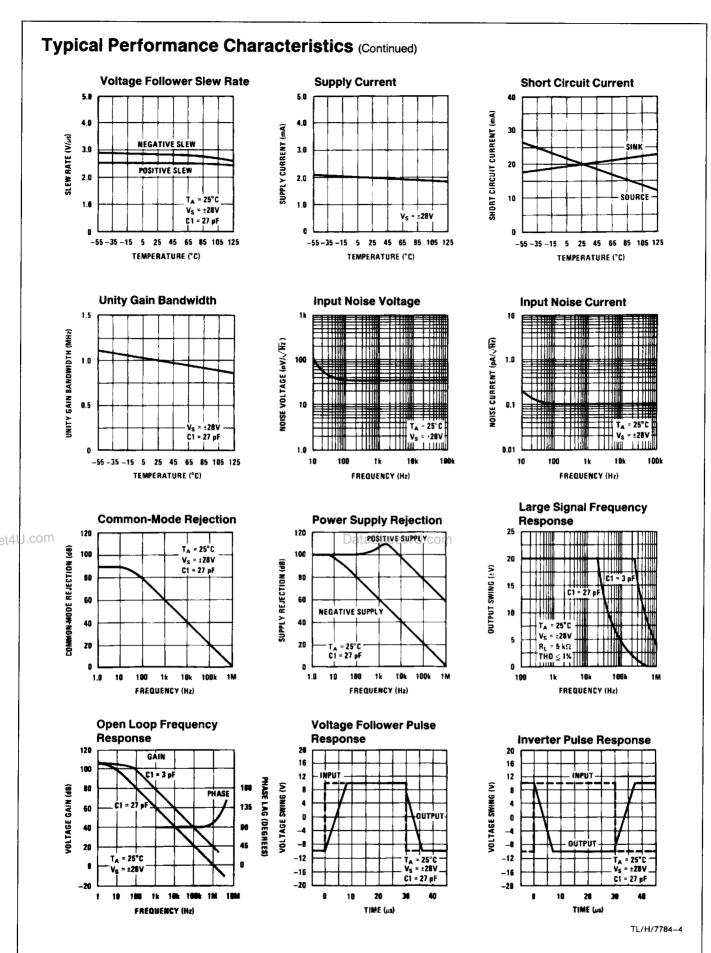




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Application Hints (See Also AN-127)

The LM144 is designed for trouble-free operation at any supply voltage up to a maximum of ± 40 V. Input overvoltage protection, both common-mode and differential, is 100% tested and guaranteed at the maximum supply voltage. Furthermore, all possible high voltage destructive modes during supply voltage turn-on have been eliminated by design. As with most IC op amps, however, certain precautions should be observed to insure that the LM144 remains virtually blow-out proof.

Although output short circuits to ground or either supply can be sustained indefinitely for supply voltages, below $\pm\,18V$, these short circuits should be of limited duration when operating at higher supply voltages. Units can be destroyed by any combination of high ambient temperature, high supply voltages, and high power dissipation which results in excessive die temperature. This is also true when driving low impedance or reactive loads or loads that can revert to low impedance; for example, the LM144 can drive most general purpose op amps outside of their maximum input voltage range, causing heavy current to flow and possibly destroying both devices.

Precautions should be taken to insure that the power supplies never become reversed in polarity—even under transient conditions. With reverse voltage, the IC will conduct excessive current, fusing the internal aluminum interconnects. Voltage reversal between the power supplies will almost always result in a destroyed unit.

In high voltage applications which are sensitive to very low input currents, special precautions should be exercised.

For example, with high source resistances, care should be taken to prevent the magnitude of the PC board leakage currents, although quite small, from approaching those of the op amp input currents. These leakage currents become larger at 125°C and are made worse by high supply voltages. To prevent this, PC boards should be properly cleaned and coated to prevent contamination and to provide protection from condensed water vapor when operating below 0°C. A guard ring is also recommended to significantly reduce leakage currents from the op amp input pins to the adjacent high voltge pins in the standard op amp pin connection as shown in *Figure 1. Figures 2, 3 and 4* show how the guard ring is connected for the three most common op amp configurations.

The minimum values given for the frequency compensation capacitor are stable only for source resistances less than 10 $k\Omega$, stray capacitances on the summing junction less than 5 pF and capacitive loads smaller than 100 pF. If any of these conditions are not met, it becomes necessary to overcompensate the amplifier with a larger compensation capacitor. Alternately, lead capacitors can be used in the feedback network to negate the effect of stray capacitance and large feedback resistors or an RC network can be added to isolate capacitive loads. See *Figures 5, 6 and 7*.

Finally, caution should be exercised in high voltage applications as electrical shock hazards are present. Since the negative supply is connected to the case, users may inadvertantly contact voltages equal to those across the power supplies.

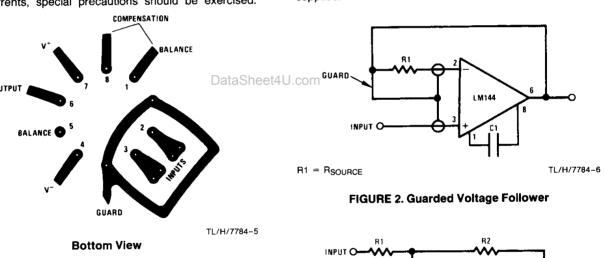


FIGURE 1. Printed Circuit Layout for Input Guarding with TO-5 Package

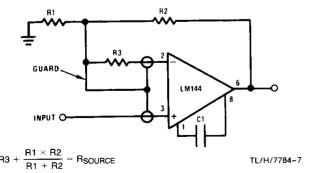


FIGURE 3. Guarded Non-Inverting Amplifier

PUT O R1

GUARD

R3

R3

R1 \times R2

FIGURE 4. Guarded Inverting Amplifier

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Application Hints (Continued)

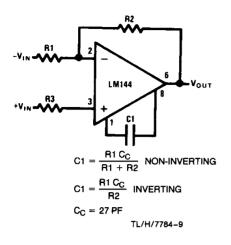
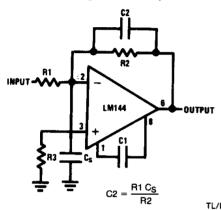


FIGURE 5. Single Pole Compensation



or Large Feedback Resistor

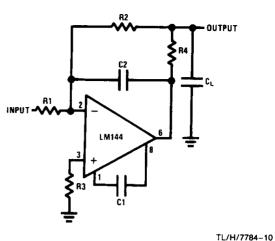


FIGURE 6. Isolating Large Capacitive Loads

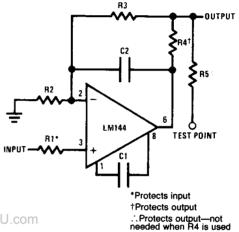


FIGURE 8. Protecting Against Gross Fault Conditions

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27 pF R2 10M 27 pF R2 5.1M TL/H/7784-13

FIGURE 9. Balancing Circuit

Connection Diagram

Metal Can Package

COMPENSATION

BALANCE 1 7 V+

INVERTING INPUT 2 6 OUTPUT

NON-INVERTING INPUT 3 4 5 BALANCE

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Top View

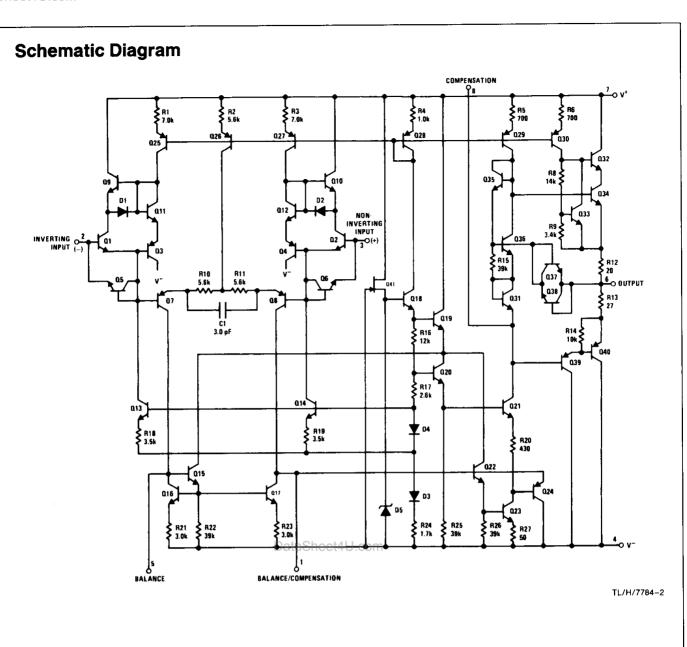
Pin 4 is connected to case

Order Number LM144H, LM144H/883* or LM344H See NS Package Number H08C

*Available per SMD# 7800301

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