

## 150-mW STEREO AUDIO POWER AMPLIFIER

### General Description

LN4812 is a differential input stereo audio power amplifier circuit for mobile phones and other portable audio devices built-in speaker. It provides a stable output power of 150mW 16Ω loads.

The shutdown current of the LN4812 is less than 100nA, that can save energy for the system. At the same time the amplifier gain can be set by external resistors, that it is easy to use.

The LN4812 using small MSOP-10 package, to facilitate high-density mounting .

### Applications

- Mobile Phones
- PDA
- Bluetooth headset

### Key Specifications

- 150mW Stereo Output
- Differential Inputs

### Ordering Information

Ordering Number	Package
LN4812SR	MSOP-10

### Pin Function Description

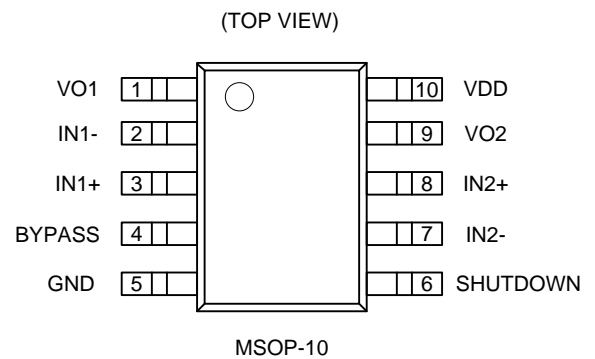
Pin Number	Pin Name	Function Description
1	VO1	VO1 is the audio output for channel 1.
2	IN1-	IN1- is the negative input for channel 1.
3	IN1+	IN1+is the positive input for channel 1.
4	BYPASS	Bypass Capacitance Input Pin
5	GND	GND is the ground connection.
6	SHUTDOWN	Puts the device in a low quiescent current mode when held high.
7	IN2+	IN2+ is the positive input for channel 2.
8	IN2-	IN2- is the negative input for channel 2.
9	VO2	VO2 is the audio output for channel 2.
10	VDD	VDD is the supply voltage terminal.

- shutdown current is less than 0.1uA
- Built-in " wave " sound canceling circuit
- Wide input voltage range : 2.2V-5.0V

### Package

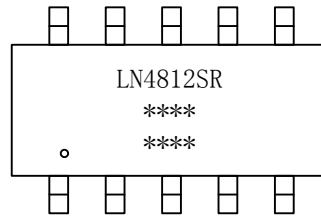
- MSOP-10

### Pin Configuration



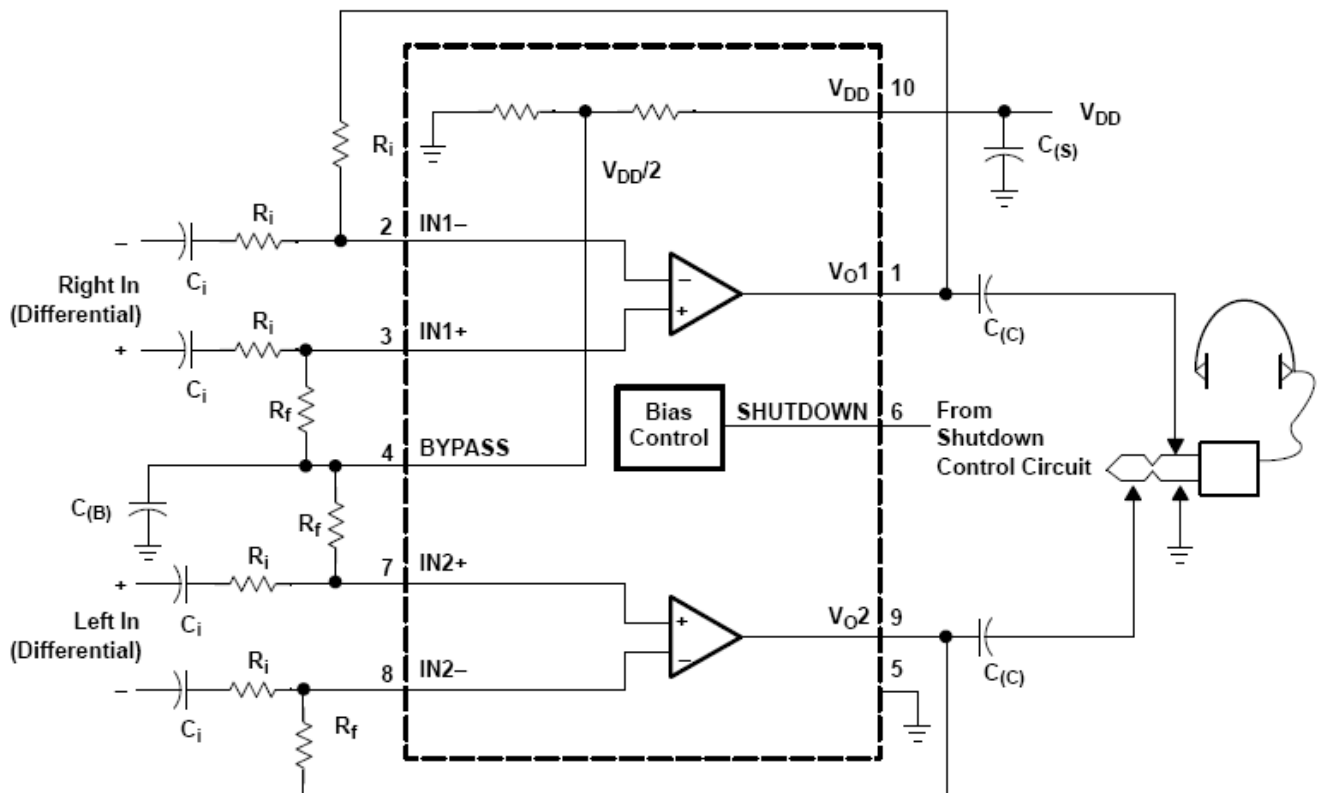
## Marking Rule

- MSOP-10



The second behavior wafer version  
The third production information behavior

## Typical Application Circuit



## Absolute Maximum Ratings

Parameter	Symbol	Value	Unit
Supply Voltage	VDD	-0.3—5.0	V
Input Voltage	VIN	-0.3—VDD+0.3	V
Operating Free-air Temperature	Topr	-40—85	°C
Storage Temperature Range	Tstg	-65—150	°C
ESD (HBM)	-	4000	V

## Electrical Characteristics

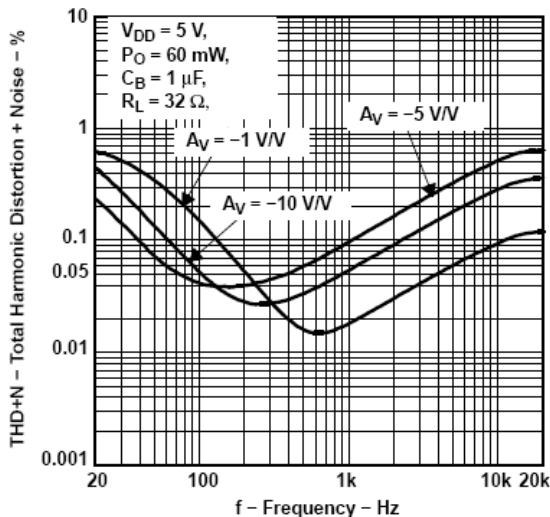
VDD=5V

(Unless otherwise specified. Limits apply for TA = 25°C.)

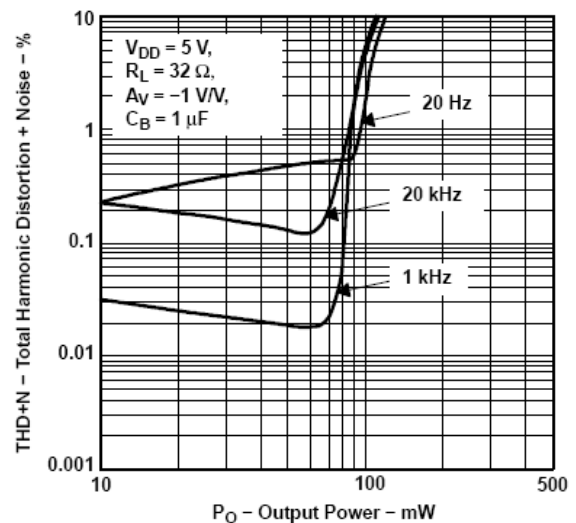
Parameter	Symbol	Condition	Min	Typ	Max	Unit
I <sub>DD</sub>	Supply Current		—	1.5	3	mA
I <sub>SD</sub>	Supply Current in SHUTDOWN Mode	Vshutdown=5V	—		0.1	μA
V <sub>SDIH</sub>	Shutdown Voltage Input High		1.2	—	—	V
V <sub>SDIL</sub>	Shutdown Voltage Input Low		—	—	0.4	V
V <sub>OS</sub>	Output Offset Voltage	A <sub>V</sub> =2V/V	—		15	mV
P <sub>O</sub>	Output Power (each channel)	THD = 1% (max); f = 1 kHz 16Ω Load		150		mW
T <sub>WU</sub>	Wake-up time		—	170	220	ms
T <sub>SD</sub>	Thermal Shutdown Temperature		150	170	190	°C
THD+N	Total harmonic distortion + noise	P <sub>O</sub> = 100mWrms; 20-20KHz	—	0.06	—	%
B <sub>om</sub>	Maximum Output Power BW	G=10, THD<5%	20			KHz
PSRR	Supply Current	V <sub>ripple</sub> = 200mV <sub>sine p-p</sub> f=1kHz	55	60	—	dB
PCS	Channel Isolation	1KHz		90		dB
T <sub>SdT</sub>	Shut Down Time	16Ω Load	—	1.0	—	ms

## Typical Performance Characteristics

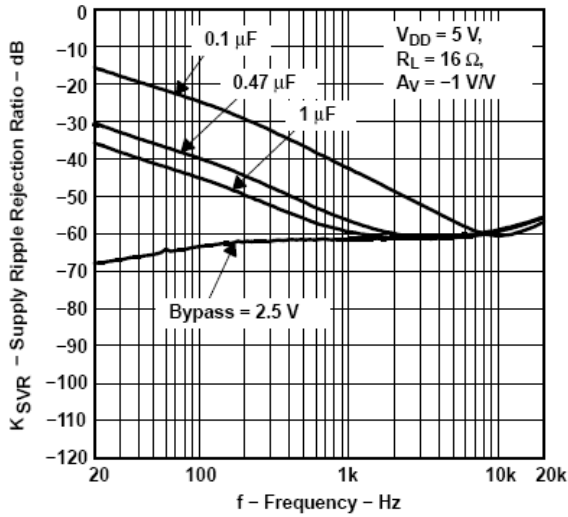
Total Harmonic Distortion + Noise VS Frequency



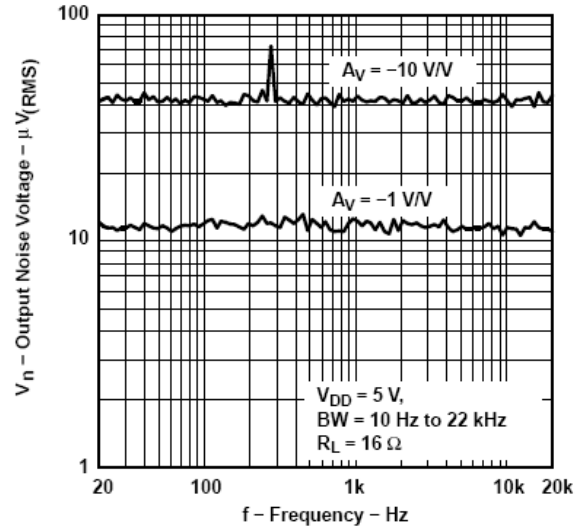
Total Harmonic Distortion + Noise VS Output Power



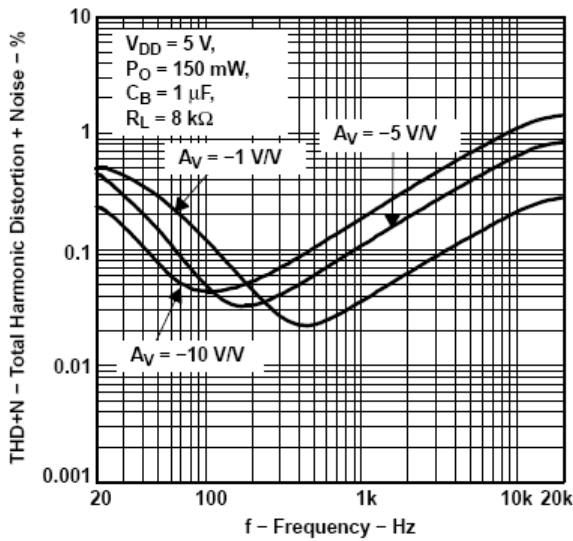
Supply Ripple Rejection Ratio VS Frequency



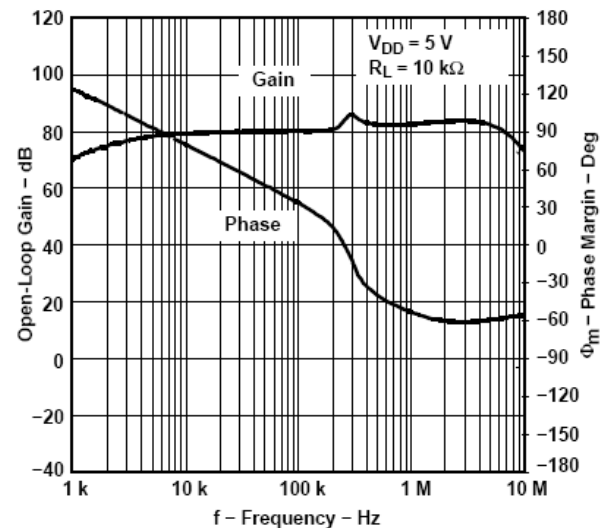
Output Noise Voltage VS Frequency



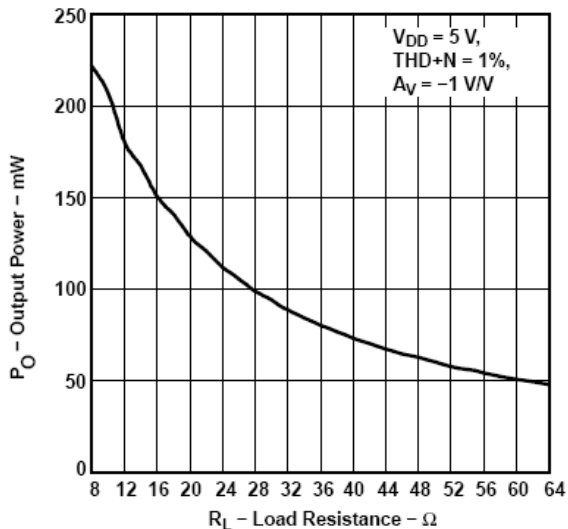
Crosstalk VS Frequency



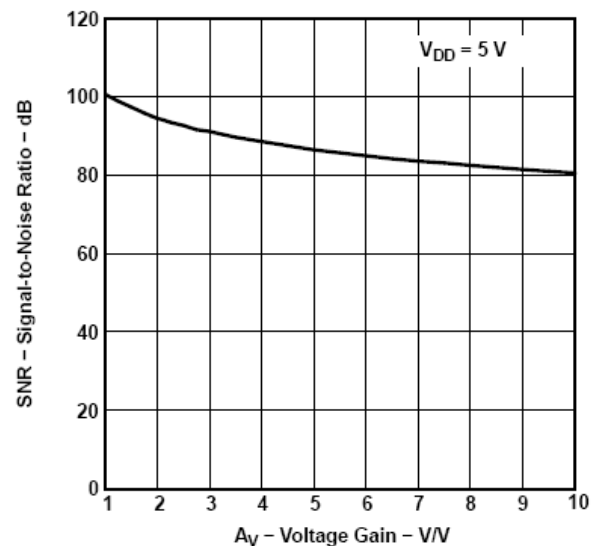
Open-loop Gain and Phase Margin VS Frequency



Output Power VS Load Resistance



Signal-to-Noise Ratio VS Voltage Gain



## ■ Application Information

### ● Gain Setting Resistors

The gain for the LN4812 is set by resistors  $R_f$  and  $R_i$  according to  $GAIN = - (R_f/R_i)$ , and  $R_i$  is usually selected for resistance 5K-20K.

When  $R_f > 50K$ , recommended to use metal film resistors, so that we can get better performance. Meanwhile, in order to prevent system instability, it is recommended  $R_f$  in parallel with a capacitor  $C_f$ , to forming a low-pass filter network together with  $R_f$ , and the cutoff frequency of the low pass filter is  $f_c = 1 / (2\pi R_f C_f)$ .

For example, if  $R_f$  is 100 k $\Omega$  and  $C_f$  is 5 pF then  $f_c(\text{lowpass})$  is 318kHz, which is well outside the audio range.

### ● Input Capacitor, $C_i$

Input resistance  $R_i$  and the input capacitance  $C_i$  form a high-pass filter.  $F_c(\text{highpass}) = f_c = 1/(2\pi R_i C_i)$ . The value of  $C_i$  directly affects the bass (low frequency) performance of the circuit. Consider the example where  $R_i$  is 10k $\Omega$  and the specification calls for a flat bass response down to 20Hz, and  $C_i$  is 1 $\mu$ F.

ESR additional parasitic capacitance of the resistor will affect the audio signal, so we recommend using a low ESR ceramic capacitor.

### ● Power Coupling Capacitor $C(s)$

For higher frequency transients, spikes, or digital hash on the line, a good low equivalent-series-resistance (ESR) ceramic capacitor, typically 0.1  $\mu$ F, placed as close as possible to the device VDD lead, works best. For filtering lower-frequency noise signals, a larger aluminum

### ● BYPASS Capacitor $C_{(B)}$

BYPASS circuit is a voltage divider network through an internal resistor to achieve, and the internal resistor is designed to in series 100K. To get a good job is usually the characteristics required to meet the  $C_B * 100K > R_i C_i$ .

For example,  $R_i = 10K$ ,  $C_i = 1\mu F$ , the  $C_B$  value preferably greater than 0.1 $\mu F$ , 1 $\mu F$  select more suggestions

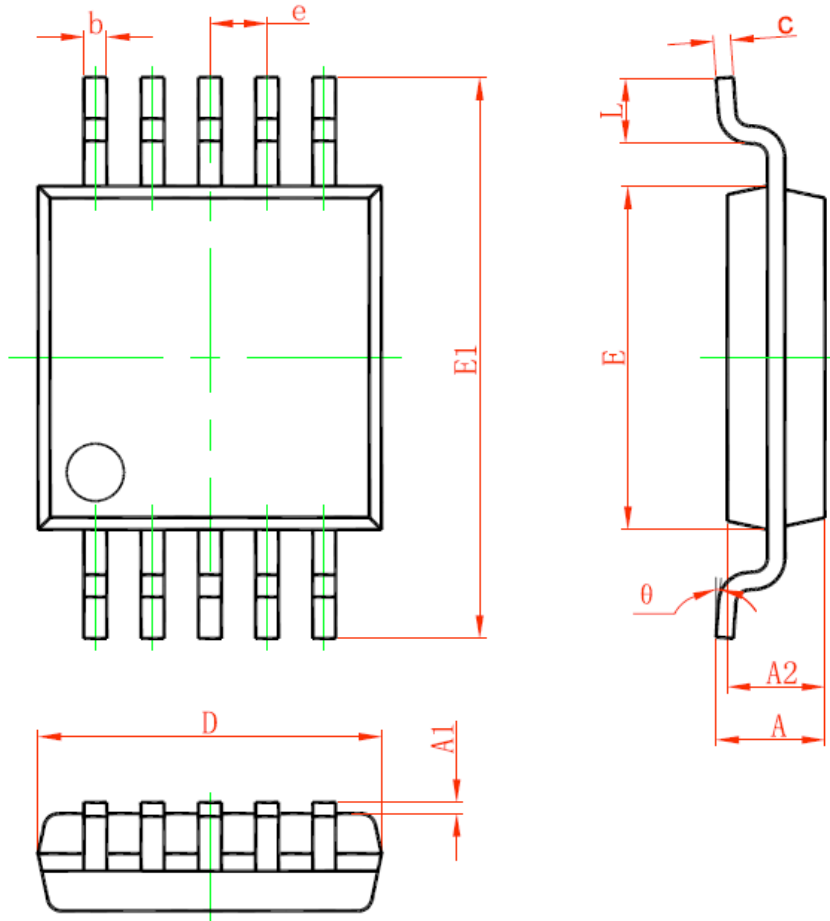
### ● Output Coupling Capacitors $C_{(C)}$

In atypical single-supply, single-ended(SE) configuration, an output coupling capacitor ( $C_{(C)}$ ) is required to block the dc bias at the output of the amplifier, thus preventing dc currents in the load. As with the input coupling capacitor, the output coupling capacitor and impedance of the load form a high-pass filter  $F_c = 1/(2\pi R_L C_c)$ .

The main disadvantage, from a performance stand-point, is that the typically-small load impedance drives the low-frequency corner higher. Large values of  $C_{(C)}$  are required to pass low frequencies into the load. The output coupling capacitor required in single-supply SE mode also places additional constraints on the selection of other components in the amplifier circuit. With the rules described earlier still valid, add the following relationship:  $C_B * 100K > R_i C_i \gg R_L C_c$ .

■ Package Information

● MSOP-10



Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min	Max	Min	Max
A	0.820	1.100	0.032	0.043
A1	0.020	0.150	0.001	0.006
A2	0.750	0.950	0.030	0.037
b	0.180	0.280	0.007	0.011
c	0.090	0.230	0.004	0.009
D	2.900	3.100	0.114	0.122
e	0.50(BSC)		0.020(BSC)	
E	2.900	3.100	0.114	0.122
E1	4.750	5.050	0.187	0.199
L	0.400	0.800	0.016	0.031
$\theta$	0°	6°	0°	6°