

# 3.5A Synchronous Buck Li-ion Charger With Intelligent Path Management, Adapter Adaptive and OTG

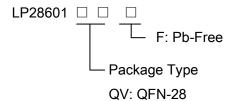
# **General Description**

The LP28601 is a 3.5A Li-lon battery charger intended for 4.4V~14V wall adapters. It utilizes a high efficiency synchronous buck converter topology to reduce power dissipation during charging. And a synchronous step up function could supply 5V/2A from Li-lon battery which can be used as the OTG. The LP28601 includes complete charge termination circuitry, automatic recharge and a ±1% 4.2V/4.3V /4.35V float voltage.

The LP28601 could manage the power supply for system intelligently. The adapter would satisfy the system's demand firstly, then charge battery with extra current output capacity from adapter. If the current capacity of adapter could not meet the system demand, the adapter and battery would supply power for system together. When the adapter's current capacity is low than the set charge current, the LP28601 would decrease the charge current automatically to keep the output of adapter would not be pull down by the chip.

Additional features include shorted cell detection; temperature qualified charging and overvoltage protection. The LP28601 is available in a low profile QFN-28 package.

# **Order Information**



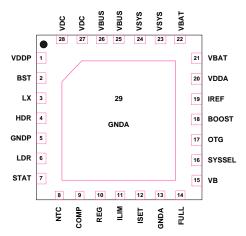
# **Applications**

- ♦ Quick charge 2.0/3.0 (QC2.0 / QC3.0)
- ♦ Portable Media Players
- ♦ Cellular and Smart mobile phone
- ♦ PDA/DSC
- ♦ Handheld Battery-Powered Devices
- Handheld Computers
- ♦ Charging Docks and Cradles

## **Features**

- Power Path Management Intelligently
- Adapter Adaptive
- ◆ CHARGE:
  - Very Low Power Dissipation
  - 3.5A Maximum Charge Current
  - Efficiency up to 90%
  - Input voltage: 4.4V~14V
  - Programmable charge complete voltage: 4.2V /4.3V /4.35V
  - Operation with Thermal Regulation to Maximize Charge Rate Without Risk of Overheating
  - Charges Single Cell Li-Ion Batteries Directly from USB Port
- OTG:
  - 5V /2A @ V<sub>BAT</sub>=3V Synchronous step up
  - Efficiency up to 92%
- Available in QFN28(4\*4mm) Package
- ◆ RoHS Compliant and 100% Lead (Pb)-Free

# **Functional Pin Description**



# **Marking Information**

Device	Marking	Package	Shipping	
LP28601QVF	LPS	QFN-28	3K/REEL	
LP28601				
	YWX			
Y: Year code. W: Week code. X: Batch numbers.				

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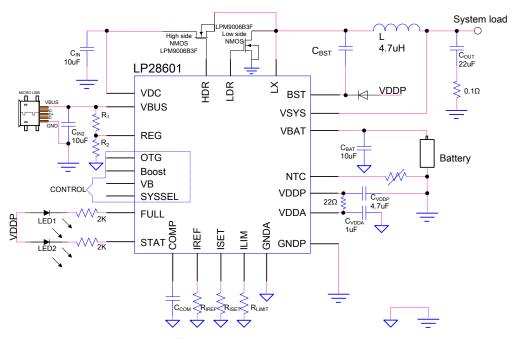


# **Pin Description**

Pin	Name	Description				
1	VDDP	Internal LDO output. Connect a decoupling 4.7uF capacitor to GNDP.				
2	BST	Positive supply for the high side driver. A 0.22µF capacitor should be placed between BST and LX.				
3	LX	Switching Node Connection.				
4	HDR	High side drive gate.				
5	GNDP	Ground for Power section.				
6	LDR	Low side drive gate.				
7	STAT	Indicates charge status. Active low when charging is on. STAT will blink with timeout, V <sub>SYS-OVP</sub> , NTC fault.				
8	NTC	Connect a 10K NTC resistor to GNDA, 100uA(constant current source) current output from NTC pin.				
9	COMP	Compensation pin, a 2.2nF ceramic capacitor is needed from COMP to GNDA.				
		Input voltage feedback for the input voltage regulation loop. Connect to tap of an external resistor divider				
10	REG	from VBUS to GNDA to program the input voltage regulation. Once the voltage at REG pin drops to the				
		inner threshold, the charge current is reduced to maintain the input voltage at the regulation value.				
		Adapter current limit setting pin. A resistor R <sub>ILIM</sub> is needed from ILIM to GNDA. Adapter current is				
11	ILIM	programmed by: I <sub>LIM</sub> =550×2V/R <sub>ILIM</sub> . This pin could not be floating.				
		Recommend: 40K> R <sub>ILIM</sub> >0.37K. The chip would set the max input current limit when 0.37K> R <sub>ILIM</sub> .				
		Charging current setting pin, a resistor RISET is needed from ISET to GNDA. CC current is programmed				
12	ISET	by I <sub>CHG</sub> =1500V/R <sub>ISET</sub> . The internal reference for ISET comparator is 1.5V when V <sub>BAT</sub> >V <sub>TRIKL</sub> .				
	l l	Recommend: 6.8K> R <sub>ISET</sub> >0.43K.				
13,29	GNDA	Ground for the analog circuits.				
14	FULL	Battery ful <mark>l</mark> indication pin, active low.				
15	VB	Programmable battery-full voltage. Connect to GND for 4.35V, leave floating to 4.2V, and connect to				
13	V D	VDDA for 4.3V.				
16	SYSSEL	Programmable VSYS minimum voltage. Connect to GND for 3V, leave floating to 3.3V, and connect to				
	010022	VDDA for3.5V.				
17	OTG	USB On-The-Go function enable pin, active low. With $1M\Omega$ internal resistor pulls up. When OTG active,				
.,		the charge function is disable. This pin must be high voltage to charge the battery.				
18	Boost	Boost function enable pin, active high. With $1M\Omega$ internal resistor pulls up.				
19	IREF	Current reference generator. A 100k resistor connect to GNDA, internal voltage reference is 1V.				
20	VDDA	Power supply for the internal analog circuit.				
21,22	VBAT	Battery charger output and battery voltage sense pin. Connect to battery cell.				
23,24	VSYS	System voltage output.				
25,26	VBUS	USB or adapter input.				
27,28	VDC	A capacitor is needed from this pin to GNDP.				

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# **Typical Application Circuit**



# Absolute Maximum Ratings Note 1

$\diamond$	Input and VOUT to GND(VDC,VBUS)	0.3V to 18V
$\diamondsuit$	Other Pin to GND	0.3V to 6.5V
$\diamond$	LX voltage to GND	0.3V to 18V
$\diamondsuit$	HDR,BST voltage to GND	
$\diamondsuit$	BST referred to LX	0.3V to 6.5V
$\diamond$	BAT Short-circuit Duration	Continuous
$\diamondsuit$	Maximum Junction Temperature	150°C
$\diamondsuit$	Storage Temperature	45°C to 165°C
$\diamond$	Operating Ambient Temperature Range (T <sub>A</sub> )	40°C to 85°C
<b></b>	Maximum Soldering Temperature (at leads, 10 sec)	260°C

**Note 1.** Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## **Thermal Information**

- ♦ Maximum Power Dissipation (QFN-28, P<sub>D</sub>, T<sub>A</sub>=25°C) ------ 2W
- ♦ Thermal Resistance (QFN-28, θ<sub>JA</sub>) ------ 50°C/W

# **ESD Susceptibility**

- ♦ HBM(Human Body Mode) ------ 2KV
- ♦ MM(Machine Mode) ------ 200V

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# **Electrical Characteristics**

(The specifications which apply over the full operating temperature range, T<sub>A</sub>=25°C, V<sub>BUS</sub>=5V, unless otherwise noted.)

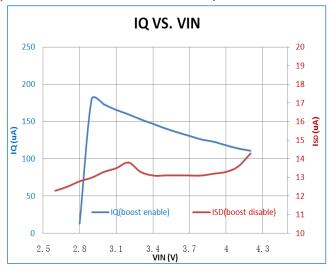
Parameter	Test Conditions	Measured	Min	Тур	Max	Units
	VBUS, VSYS,VDD					
Input voltage		V <sub>BUS</sub>	4.4		14	V
VBUS port protection threshold	V <sub>BUS</sub> Rising	UVLO	4.15	4.25	4.35	V
	V <sub>BUS</sub> Falling		3.95	4.05	4.15	V
Input voltage regulation reference		$V_{REG}$	2.3	2.4	2.5	V
	SYSSEL=float, V <sub>BAT</sub> =2V			3.3		V
Minimum output voltage set for VSYS	SYSSEL=high, V <sub>BAT</sub> =2V			3.5		V
VDDDA/DDA	SYSSEL=low, V <sub>BAT</sub> =2V			3.0		V
VDDP/VDDA	Dower Doth Management			5		V
VDC Dower by USP/Adenter MOSEET Days	Power Path Management  1A Current Load	D	I	50	I	
VDC Power by USB/Adapter MOSFET R <sub>DSON</sub> Switch between VSYS and VBAT	V <sub>BAT</sub> =4.2V, VBUS Absent, I <sub>BAT</sub> =3A	R(VBUS,VDC) R(VSYS, VBAT)		50 40		mΩ mΩ
Switch between VS13 and VBA1	Quiescent Currents	N(VSYS, VBAT)		40		11122
	OTG=float	1		750		μA
VBUS Current	OTG=low	I <sub>BUS</sub>		20		μΑ
VDOS Current	V <sub>BUS</sub> =4V	IBUS		80		μΑ
Battery Discharge Current in Standby mode	V <sub>BUS</sub> -4-V V <sub>BAT</sub> =4.2V, VBOOST Floating VBUS Absent	I <sub>BAT</sub>		15		μA
Battery Discharge Current in boost mode	V <sub>BAT</sub> =4.2V, V <sub>BUS</sub> =5V, I <sub>OUT</sub> =0A	IBOOST		110		μΑ
Dattery Disoriarge Ourient in boost mode	Charger Controller	180081		1 110		_ μΛ
	V <sub>BAT</sub> <1.4V	I	I	0.04	I	l v
$V_{\text{ISET}}$ Voltage in TRICK Mode and Low $V_{\text{BAT}}$	1.4V< VBAT < VTRICKL	V <sub>ISET</sub>		0.04		V
	VB=float, R <sub>ISET</sub> =1K, I <sub>BAT</sub> =100mA		4.158	4.2	4.242	V
VBAT voltage	VB=high, R <sub>ISET</sub> =1K, I <sub>BAT</sub> =100mA		4.100	4.3	7.272	V
VD/(I Vollage	VB=low, R <sub>ISET</sub> =1K, I <sub>BAT</sub> =100mA			4.35		V
	R <sub>ISET</sub> =1K,V <sub>BAT</sub> =3.6V			1500		mA
Charge Current in CC Mode	RISET=0.5K,VBAT=3.6V	I <sub>BAT</sub>		3000		mA
V <sub>ISET</sub> Voltage in CC Mode	TUSET-0.5IX, V BAT-0.0 V	V <sub>ISET</sub>	1.45	1.5	1.55	V
Trickle Charge Voltage Threshold	V <sub>BAT</sub> rising	VISET	2.8	2.9	3.0	V
Trickle Charge Voltage Threshold Hysteresis	VBAITISING	VTRIKL_HYS	2.0	200	0.0	mV
Thomas charge vertage threathers thystoreass	V <sub>BAT</sub> < 1.4V	V TRIRE_TITO		15		mA
Charge Current in Trickle Charge Condition	1.4V < V <sub>BAT</sub> < V <sub>TRIKL</sub> , R <sub>ISET</sub> =1K	I <sub>TRIKL</sub>	= $H$	200		mA
charge carrone in Thomas charge condition	1.4V < V <sub>BAT</sub> < V <sub>TRIKL</sub> , RISET=0.5K	TITAL		400		mA
End of Charger Current	T. TV - V BAT - V TRIRE, TRISET - 0.0TC			13.3		%I <sub>BAT</sub>
Adapter Current Limit		V <sub>ILIM</sub>	1.9	2	2.1	V
·	R <sub>ILIM</sub> =1K	- ILIW		1.1		A
Current Limit through VBUS	R <sub>ILIM</sub> =0.5K	I <sub>LIMIT</sub>		2.2		A
Switch Frequency	V <sub>BUS</sub> =5V,V <sub>BAT</sub> =3.6V,R <sub>ISET</sub> =0.5K			1.2		MHz
Trickle Charge Timer	Default register, wake-up mode			90		min
Charge Timer	Default register, CC+CV mode			10		hr
Recharge Threshold	V <sub>BAT</sub> falling			150		mV
STAT Low Level	Open drain pulled up with 5mA	STAT	-	-	0.2	V
Leakage Current to STAT	V <sub>BAT</sub> =4.3V, I <sub>BAT</sub> =0			0.6	1	μA
FULL Low Level	Open drain pulled up with 5mA	FULL			0.2	V
Leakage Current to FULL	V <sub>BAT</sub> =4.3V I <sub>BAT</sub> =0			0.4	1	μA
Charging Enable Time	OTG: L→H	T <sub>C_DELAY</sub>		6	-	ms
Ondrighing Endoire Time	Temperature sense comparators	10_020				16
High Voltage Threshold	Temp fault at V <sub>NTC</sub> >V <sub>LTF</sub>	V <sub>LTF</sub>	2.45	2.50	2.55	V
Low Voltage Threshold	Temp fault at V <sub>NTC</sub> <v<sub>HTF</v<sub>	VHTF	0.48	0.50	0.52	V
Temperature Sense Current Sense	R <sub>NTC</sub> =10k	I <sub>NTC</sub>	94	100	106	μA
Charging and OTG Boost Temperature	Temperature rising			145		°C
Shutdown	Hysteresis falling	1		25		°C
	Boost Controller Section				l	
Boost Output Voltage	V <sub>BAT</sub> =4.2V	V <sub>OUT</sub>	4.9V	5.05	5.2V	ΙV
	V <sub>BAT</sub> =3V,I <sub>OUT</sub> =1A	- 501		1.4	J.2.	MHz
DOOSE SWILCH Frequency		1				
Boost Switch Frequency	V <sub>BAT</sub> rising			2.9		V
Boost UVLO	V <sub>BAT</sub> rising V <sub>BAT</sub> falling	V <sub>BAT</sub>		2.9 2.6		V

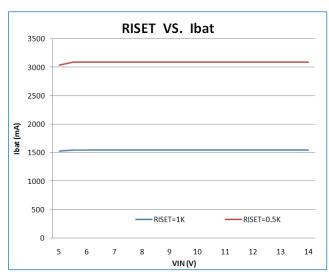
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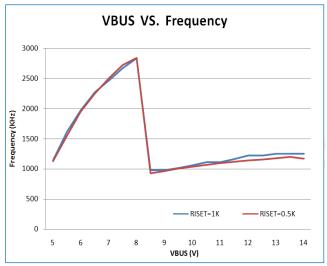


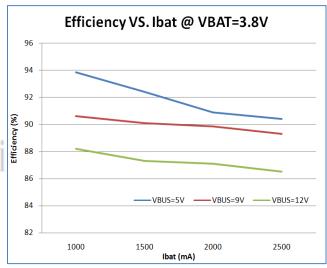
# **Typical Operating Characteristics for Charger**

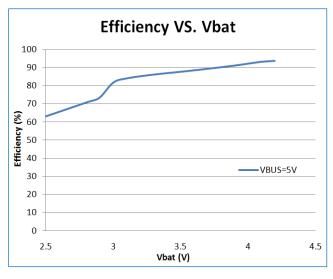
## (T<sub>A</sub>=25°C, unless otherwise noted)

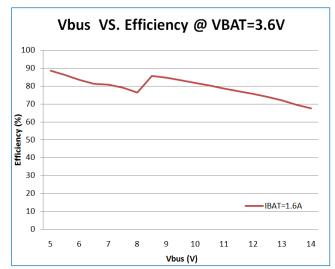




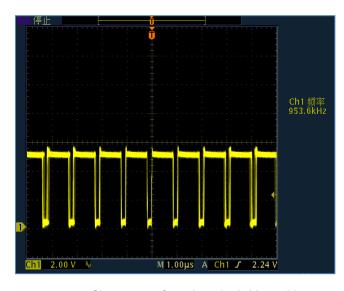




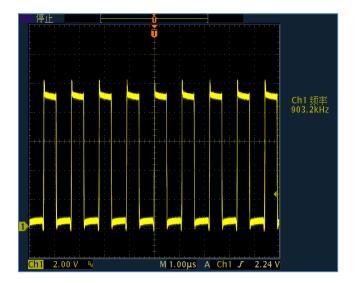




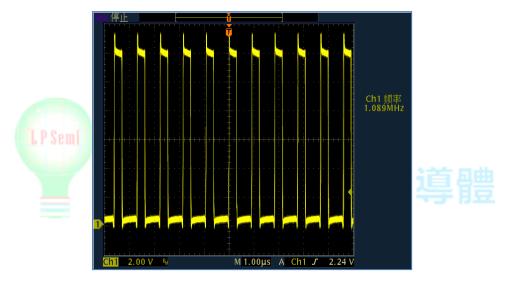
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Charge waveform: I<sub>BAT</sub>=2.5A, V<sub>BUS</sub>=5V



Charge waveform: IBAT=2.5A, VBUS=9V



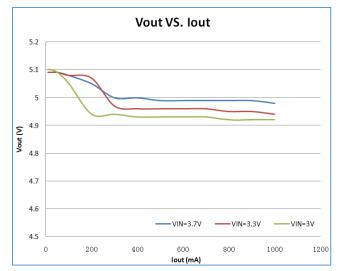
Charge waveform: IBAT=2.5A, VBUS=12V

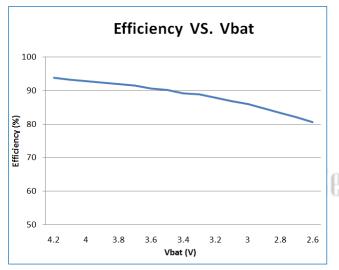
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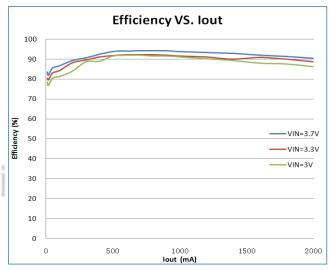
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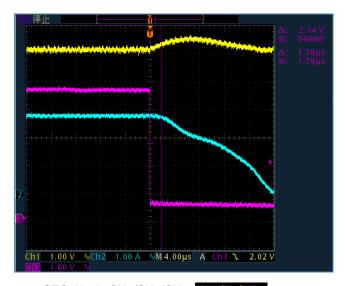
Email: marketing@lowpowersemi.com

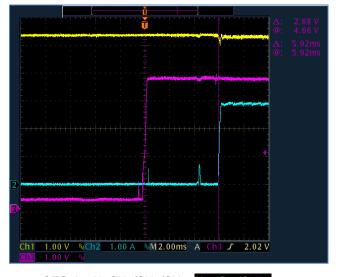
# **Typical Operating Characteristics for Boost(OTG)**







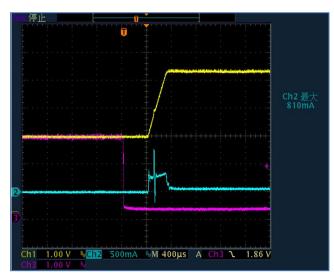




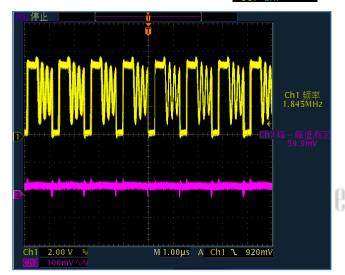
OTG: H→L; CH1/CH2/CH3=V<sub>BUS</sub>/I<sub>BA</sub>/OTG

OTG: L→H; CH1/CH2/CH3=V<sub>BUS</sub>/I<sub>BAT</sub>/OTG

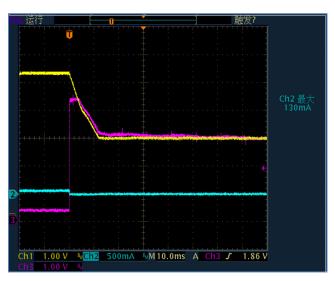
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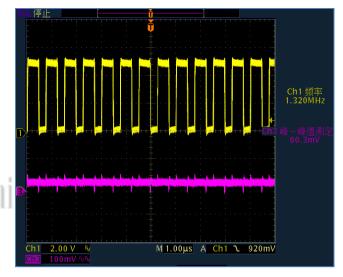
BOOST ON waveform: CH1/CH2/CH3=V<sub>OUT</sub>/I<sub>BAT</sub>/Boost



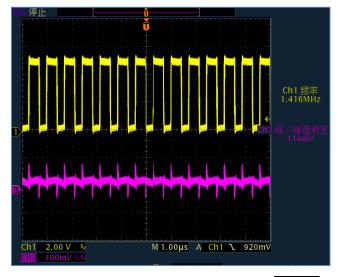
Boost waveform: V<sub>BAT</sub>=3V, I<sub>OUT</sub>=100mA; CH1/CH3=LX/Vp-p



BOOST OFF waveform: CH1/CH2/CH3=V<sub>OUT</sub>/I<sub>BAT</sub>/Boost



Boost waveform: V<sub>BAT</sub>=3V, I<sub>OUT</sub>=200mA; CH1/CH3=LX/Vp-p



Boost waveform: V<sub>BAT</sub>=3V, I<sub>OUT</sub>=1A; CH1/CH3=LX/Vp-p

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# **Application Information**

The LP28601 is an easy controlled power path management device and a single cell Li-lon battery charger. It integrates the input reverse-blocking FET, high-side switching FET, low-side switching FET, and BATFET between system and battery. The device also integrates the bootstrap diode for the high-side gate drive.

## Device Power Up from Battery without DC Source

If only battery is present and the voltage is above depletion threshold, the BATFET turns on and connects battery to system. The REG LDO stays off to minimize the quiescent current. The low R<sub>DSON</sub> in BATFET and the low quiescent current on BAT minimize the conduction loss and maximize the battery run time.

## Power Up from DC Source

When the DC source plugs in, the LP28601 checks the input source voltage to turn on REG、 LDO and all the bias circuits. It also checks the input current limit before starts the buck converter.

#### **Input Source Qualification**

After REG LDO powers up, the LP28601 checks the current capability of the input source. The input source has to meet the following requirements to start the buck converter.

- 1. VBUS voltage below 14V
- 2. VREG voltage above 2.43V

Once the input source passes all the conditions above, the permit signal is asserted to the chip.

#### **Input Current Limit Detection**

The USB ports on personal computers are convenient charging source for portable devices (PDs). If the portable device is attached to a USB host, the USB specification requires the portable device to draw limited current. If the portable device is attached to a charging port, it is allowed to draw up to the maximum current form the USB host by two parts limit:

- 1. VREG voltage above 2.43V
- 2. The maximum input current < I<sub>LIM</sub>

## **Boost Mode Operation from Battery**

The LP28601 supports boost converter operation to deliver power from the battery to other portable devices through USB port. The boost mode output current rating meets the 2A charging requirements for smart phone and tablet. The boost operation can be enabled after the OTG and BOOST is pulled low. In battery boost mode, the LP28601 employs a synchronous step-up switching regulator. During boost mode, the output voltage is fixed at 5.1V and the chip could supply 5V /2A from V<sub>BAT</sub>=3V with high efficiency on VDC. It is recommended to use the minimum CIN2 cap value 20uF for boost current.

#### **VSYS and Narrow VDC Architecture**

The device deploys Narrow VDC architecture (NVDC) with BATFET separating system from battery. The minimum system voltage is set by battery voltage. With a not fully depleted battery, the system is regulated (R<sub>BATFET</sub>×I<sub>(BAT,SYS)</sub>)V less than the battery voltage. And a selectable VSYS could be set by SYSSEL when battery is fully depleted and adapter is applied to VBUS.

Charge state indication

As showed below, the STAT and FULL LED respond to this six STATES.

STATE	STAT	FULL	
Without Battery	Flicker	Light On	
Charging	Light On	Light Off	
Charge Complete	Light Off	Light On	
Battery Overheat	Flicker	Light Off	
Time Out	Flicker	Light Off	
$V_{REG} < V_{REG(th)}$	Light On	Light Off	

## **Dynamic Power Management**

To meet maximum current limit in USB spec and avoid over loading the adapter, the LP28601 features Dynamic Power Management (DPM), which continuously monitors the input current and input voltage. When input source is over-loaded because the charge current is too large, either the current exceeds the input current limit or the voltage falls below the input voltage limit by detection from REG. The device then reduces the charge current until the REG voltage rises above the threshold voltage and the input current is less than the current limit.

#### **Power Path Management**

The LP28601 accommodates a wide range of input sources from USB to wall adapter. The device provides automatic power path selection to supply the system (SYS) from input source (VBUS), battery (BAT), or both.

If the system current and the input current limit is large than the adapter current limit, the adapter voltage would be pulled down by this large system current.

## **Battery Charging Management**

The LP28601 charges 1-cell Li-lon battery with up to 3.5A charge current for high capacity tablet battery. The low dissipation BATFET improves charging efficiency and minimizes the voltage drop during discharging.

## **Autonomous Charging Cycle**

With battery charging enabled, the LP28601 can complete a charging cycle. The charger device automatically terminates the charging cycle when the charging current is below termination threshold and charge voltage is above recharge threshold. When a full battery voltage is discharged below recharge threshold 0.15V, the LP28601 automatically starts another charging cycle. The STAT output indicates the charging status of charging (LOW), charging complete or charge disable (HIGH) or charging fault (Blinking). The three states indicate the different charging phases: low-charging, high-charge complete, blink-charge fault. Another charge down indication is FULL(low when charge complete or without battery).

## **Battery Charging Profile**

The device charges the battery in three phases: preconditioning, constant current and constant voltage. At the beginning of a charging cycle, the device checks the battery voltage and applies current.

If the charger device is in DPM regulation or thermal regulation during charging, the actual charging current will be less than the programmed value. In this case, termination is temporarily disabled and the charging safety timer is counted.

# **Preliminary Datasheet**

LP28601

## **Battery Temperature Detection**

The LP28601 continuously monitors battery temperature by measuring the voltage between the NTC pins and ground, typically determined by a negative temperature coefficient thermistor and an external voltage divider. The device compares this voltage against its internal thresholds to determine if charging is allowed. To initiate a charge cycle, the battery temperature must be within the V<sub>LTF</sub> to V<sub>HTF</sub> thresholds. There is a constant current source in NTC which is 100uA(I<sub>NTC</sub>)flowing out form this pin. So V<sub>NTC</sub> is I<sub>NTC</sub>×R<sub>NTC</sub>.

When the NTC fault occurs, the STAT pin will blink to indicate the fault.

## Input Current LIMIT on ILIM and ISET

For safe operation, the LP28601 has an additional hardware pin on ILIM to limit maximum input current on ILIM pin. The input maximum current is set by a resistor from ILIM pin to ground as:

$$I_{LIM}(A) = 550 \times \frac{2V}{R_{ILIM}(\Omega)}$$

## **ISET ramming Charge Current**

The charge current is  $R_{\mathsf{ISET}}$  rammed using a single resistor from the ISET pin to ground. The battery charge current is 1000 times the current out of the ISET pin. The ISET ram resistor and the charge current are calculated using the following equations:

$$I_{BAT}(A) = 1000 \times \frac{1.5V}{R_{ISET}(\Omega)}$$

Note: V<sub>ISET</sub> is 1.5Volts when V<sub>BAT</sub>>V<sub>TRIKL</sub>.



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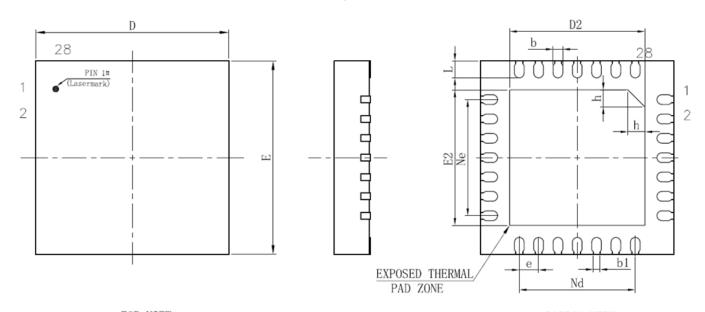
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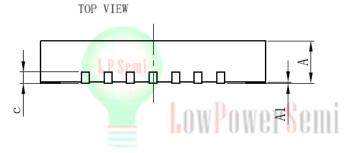
Mar.-2017

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# **Packaging Information**

## **QFN-28**





BO	TΤ	'OM	VI	EW

SYMBOL	MILLIMETER			
STMBOL	MIN	NOM	MAX	
A	0.70	0.75	0.80	
A1	0	0.02	0.05	
b	0. 15	0.20	0. 25	
b1	0. 14REF			
с	0. 18	0.20	0. 25	
D	3. 90	4.00	4. 10	
<b>D</b> 2	2. 70	2.80	2. 90	
e	0. 40BSC			
Ne	2. 40BSC			
Nd	2. 40BSC			
Е	3. 90	4.00	4. 10	
E2	2. 70	2.80	2. 90	
L	0.30	0. 35	0.40	
h	0.30	0. 35	0.40	

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