

Single/Dual/Quad 220MHz, 1500V/µs Operational Amplifiers with Programmable Supply Current

FEATURES

- 220MHz Gain-Bandwidth Product
- 1500V/us Slew Rate
- 6.5mA Supply Current per Amplifier
- Space Saving MSOP and SSOP Packages
- Ultra Small SOT-23 and Leadless DFN Packages
- Programmable Current Option
- 6nV/√Hz Input Noise Voltage
- Unity-Gain Stable
- 1.5mV Maximum Input Offset Voltage
- 8µA Maximum Input Bias Current
- 800nA Maximum Input Offset Current
- 50mA Minimum Output Current, V_{OUT} = ±3V
- ± 3.5 V Minimum Input CMR, $V_S = \pm 5$ V
- Specified at ±5V, Single 5V Supplies
- Operating Temperature Range: -40°C to 85°C

APPLICATIONS

- Wideband Amplifiers
- Buffers
- Active Filters
- Video and RF Amplification
- Communication Receivers
- Cable Drivers
- Data Acquisition Systems

DESCRIPTION

The LT®1815/LT1816/LT1817 are low power, high speed, very high slew rate operational amplifiers with excellent DC performance. The LT1815/LT1816/LT1817 feature higher bandwidth and slew rate, much lower input offset voltage and lower noise and distortion than other devices with comparable supply current. A programmable current option (LT1815 and LT1816A) allows power savings and flexibility by operating at reduced supply current and speed. The circuit topology is a voltage feedback amplifier with the slewing characteristics of a current feedback amplifier.

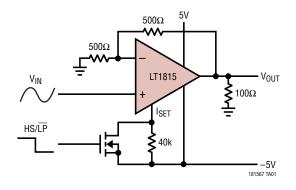
The output drives a 100Ω load to $\pm 3.8 V$ with $\pm 5 V$ supplies. On a single 5V supply, the output swings from 1V to 4V with a 100Ω load connected to 2.5V. Harmonic distortion is -70 dB for a 5 MHz, $2 V_{P-P}$ output driving a 100Ω load in a gain of -1.

The LT1815/LT1816/LT1817 are manufactured on Linear Technology's advanced low voltage complementary bipolar process and are available in a variety of SOT-23, SO, MSOP, SSOP and leadless DFN packages.

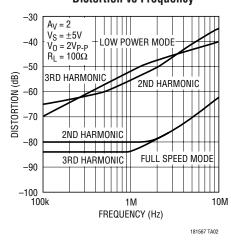
17, LTC and LT are registered trademarks of Linear Technology Corporation.

TYPICAL APPLICATION

Programmable Current Amplifier Switches from Low Power Mode to Full Speed Mode



Distortion vs Frequency

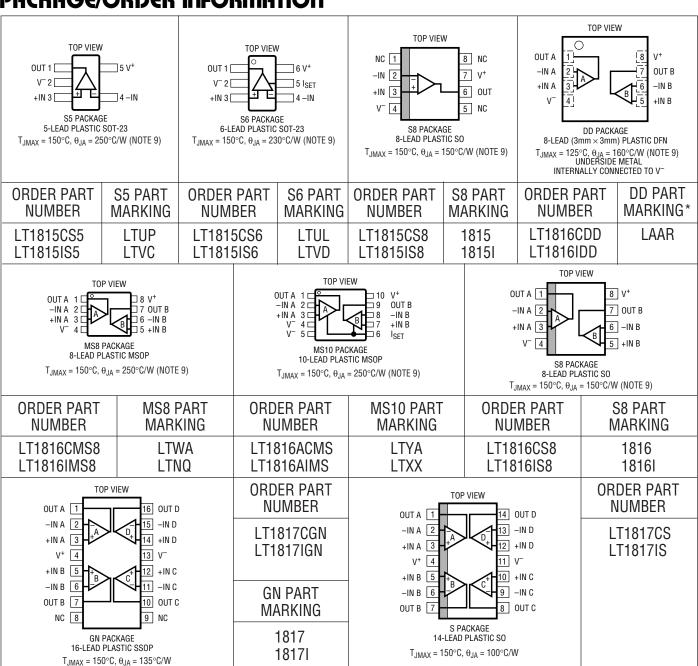


ABSOLUTE MAXIMUM RATINGS (Note 1)

Total Supply Voltage (V + to V -)	12.6V
Differential Input Voltage	
(Transient Only, Note 2)	±6V
Input Voltage	
Output Short-Circuit Duration (Note 3) Inc	definite
Operating Temperature Range40°C to	o 85°C

Specified Temperature Range (Note 8)40°C to 85°C
Maximum Junction Temperature 150°C
(DD Package)125°C
Storage Temperature Range65°C to 150°C
(DD Package)65°C to 125°C
Lead Temperature (Soldering, 10 sec)300°C

PACKAGE/ORDER INFORMATION



Consult LTC Marketing for parts specified with wider operating temperature ranges.

www.DataSheet4U.com

^{*}The temperature grades are identified by a label on the shipping container.

ELECTRICAL CHARACTERISTICS The ullet denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^{\circ}C$ (Note 8). $V_S = \pm 5V$, $V_{CM} = 0V$ unless otherwise noted. For the programmable current option (LT1815S6 or LT1816A), the I_{SET} pin must be connected to V^- through 75Ω or less, unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
V _{OS}	Input Offset Voltage	(Note 4) $T_A = 0$ °C to 70°C $T_A = -40$ °C to 85°C	•		0.2	1.5 2.0 3.0	mV mV mV
	Input Offset Voltage (Low Power Mode) (Note 10)	LT1815S6/LT1816A, $40k\Omega$ Between I_{SET} and $V^ T_A$ = 0°C to 70°C T_A = -40 °C to 85°C	•		2	7 9 10	mV mV mV
$\frac{\Delta V_{0S}}{\Delta T}$	Input Offset Voltage Drift	$T_A = 0^{\circ}\text{C to } 70^{\circ}\text{C (Note 7)}$ $T_A = -40^{\circ}\text{C to } 85^{\circ}\text{C (Note 7)}$	•		10 10	15 30	μV/°C μV/°C
I _{0S}	Input Offset Current	T _A = 0°C to 70°C T _A = -40°C to 85°C	•		60	800 1000 1200	nA nA nA
I _B	Input Bias Current	T _A = 0°C to 70°C T _A = -40°C to 85°C	•		-2	±8 ±10 ±12	μΑ μΑ μΑ
e _n	Input Noise Voltage Density	f = 10kHz			6		nV/√Hz
i _n	Input Noise Current Density	f = 10kHz			1.3		pA/√Hz
R _{IN}	Input Resistance	$V_{CM} = \pm 3.5V$ Differential		1.5	5 750		MΩ kΩ
C _{IN}	Input Capacitance				2		pF
V _{CM}	Input Voltage Range	Guaranteed by CMRR $T_A = -40^{\circ}\text{C}$ to 85°C	•	±3.5 ±3.5	±4.2		V
CMRR	Common Mode Rejection Ratio	$V_{CM} = \pm 3.5V$ $T_A = 0^{\circ}C$ to $70^{\circ}C$ $T_A = -40^{\circ}C$ to $85^{\circ}C$	• •	75 73 72	85		dB dB dB
	Minimum Supply Voltage	Guaranteed by PSRR T _A = -40°C to 85°C	•		±1.25	±2 ±2	V
PSRR	Power Supply Rejection Ratio	$V_S = \pm 2V \text{ to } \pm 5.5V$ $T_A = 0^{\circ}\text{C to } 70^{\circ}\text{C}$ $T_A = -40^{\circ}\text{C to } 85^{\circ}\text{C}$	•	78 76 75	97		dB dB dB
	Channel Separation	V_{OUT} = $\pm 3V$, R_L = 100Ω , LT1816/LT1817 T_A = 0° C to 70° C T_A = -40° C to 85° C	•	82 81 80	100		dB dB dB
A _{VOL}	Large-Signal Voltage Gain	$\begin{split} V_{OUT} &= \pm 3 \text{V, R}_L = 500 \Omega \\ T_A &= 0^{\circ} \text{C to } 70^{\circ} \text{C} \\ T_A &= -40^{\circ} \text{C to } 85^{\circ} \text{C} \end{split}$	•	1.5 1.0 0.8	3		V/mV V/mV V/mV
		V_{OUT} = ±3V, R _L = 100 Ω T_A = 0°C to 70°C T_A = -40°C to 85°C	•	0.7 0.5 0.4	2.5		V/mV V/mV V/mV
V _{OUT}	Maximum Output Swing	$R_L = 500\Omega$, 30mV Overdrive $T_A = 0^{\circ}\text{C}$ to 70°C $T_A = -40^{\circ}\text{C}$ to 85°C	•	±3.8 ±3.7 ±3.6	±4.1		V V V
		$R_L = 100\Omega$, 30mV Overdrive $T_A = 0^{\circ}\text{C}$ to 70°C $T_A = -40^{\circ}\text{C}$ to 85°C	•	±3.50 ±3.25 ±3.15	±3.8		V V V



ELECTRICAL CHARACTERISTICS The • denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^{\circ}C$ (Note 8). $V_S = \pm 5V$, $V_{CM} = 0V$ unless otherwise noted. For the programmable current option (LT1815S6 or LT1816A), the I_{SET} pin must be connected to V^- through 75Ω or less, unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
I _{OUT}	Maximum Output Current	$V_{OUT} = \pm 3V$, 30mV Overdrive $T_A = 0^{\circ}C$ to $70^{\circ}C$ $T_A = -40^{\circ}C$ to $85^{\circ}C$	•	±50 ±45 ±40	±80		mA mA mA
	Maximum Output Current (Low Power Mode) (Note 10)	LT1815S6/LT1816A; $40k\Omega$ Between I_{SET} and V^- ; $V_{OUT}=\pm 3V$, $30mV$ Overdrive $T_A=0^{\circ}C$ to $70^{\circ}C$ $T_A=-40^{\circ}C$ to $85^{\circ}C$	•	±50 ±40 ±30	±75		mA mA mA
I _{SC}	Output Short-Circuit Current	$V_{OUT} = 0V$, 1V Overdrive (Note 3) $T_A = 0$ °C to 70°C $T_A = -40$ °C to 85°C	•	±100 ±90 ±70	±200		mA mA mA
SR	Slew Rate	$A_V = -1$ (Note 5) $T_A = 0^{\circ}C$ to $70^{\circ}C$ $T_A = -40^{\circ}C$ to $85^{\circ}C$	•	900 750 600	1500		V/µs V/µs V/µs
FPBW	Full-Power Bandwidth	6V _{P-P} (Note 6)			80		MHz
GBW	Gain-Bandwidth Product	$ f = 200 \text{kHz}, \ R_L = 500 \Omega, \ LT1815 $ $ T_A = 0^\circ \text{C to } 70^\circ \text{C} $ $ T_A = -40^\circ \text{C to } 85^\circ \text{C} $	•	150 140 130	220		MHz MHz MHz
		$f = 200 \text{kHz}$, $R_L = 500 \Omega$, LT1816/LT1817 $T_A = 0^{\circ} \text{C}$ to 70°C $T_A = -40^{\circ} \text{C}$ to 85°C	•	140 130 120	220		MHz MHz MHz
	Gain-Bandwidth Product (Low Power Mode) (Note 10)	LT1815S6/LT1816A; $40k\Omega$ Between I _{SET} and V ⁻ ; f = $200kHz$, R _L = 500Ω T _A = 0° C to 70° C T _A = -40° C to 85° C	•	35 30 25	55		MHz MHz MHz
–3dB BW	-3dB Bandwidth	$A_V = 1$, $R_L = 500\Omega$			350		MHz
t_r , t_f	Rise Time, Fall Time	$A_V = 1$, 10% to 90%, 0.1V, $R_L = 100\Omega$			1		ns
t _{PD}	Propagation Delay	$A_V = 1,50\%$ to 50%, 0.1V, $R_L = 100\Omega$			1.4		ns
OS	Overshoot	$A_V = 1, 0.1V; R_L = 100\Omega$			25		%
ts	Settling Time	$A_V = -1, 0.1\%, 5V$			15		ns
THD	Total Harmonic Distortion	$A_V = 2$, $f = 5MHz$, $V_{OUT} = 2V_{P-P}$, $R_L = 500\Omega$			-70		dB
dG	Differential Gain	$A_V = 2$, $V_{OUT} = 2V_{P-P}$, $R_L = 150\Omega$			0.08		%
dP	Differential Phase	$A_V = 2$, $V_{OUT} = 2V_{P-P}$, $R_L = 150\Omega$			0.04		Deg
R _{OUT}	Output Resistance	$A_V = 1$, $f = 1MHz$			0.20		Ω
I _S	Supply Current	LT1815 $T_A = 0^{\circ}C$ to $70^{\circ}C$ $T_A = -40^{\circ}C$ to $85^{\circ}C$	•		6.5	7 9 10	mA mA mA
		LT1816/LT1817, per Amplifier $T_A = 0^{\circ}C \text{ to } 70^{\circ}C$ $T_A = -40^{\circ}C \text{ to } 85^{\circ}C$	•		6.5	7.8 10.5 11.5	mA mA mA
	Supply Current (Low Power Mode) (Note 10)	LT1815S6/LT1816A, $40k\Omega$ Between I_{SET} and V^- , per Amplifier $T_A=0^{\circ}C$ to $70^{\circ}C$ $T_A=-40^{\circ}C$ to $85^{\circ}C$	•		1	1.5 1.8 2.0	mA mA mA
I _{SET}	I _{SET} Pin Current (Note 10)	LT1815S6/LT1816A T _A = 0°C to 70°C T _A = -40°C to 85°C	•	-150 -175 -200	-100		μΑ μΑ μΑ

ELECTRICAL CHARACTERISTICS The • denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^{\circ}C$ (Note 8). $V_S = 5V$, 0V; $V_{CM} = 2.5V$, R_L to 2.5V unless otherwise noted. For the programmable current option (LT1815S6 or LT1816A), the I_{SET} pin must be connected to V^- through 75Ω or less, unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
V _{OS}	Input Offset Voltage	(Note 4) $T_A = 0^{\circ}C$ to $70^{\circ}C$ $T_A = -40^{\circ}C$ to $85^{\circ}C$	•		0.4	2.0 2.5 3.5	mV mV mV
	Input Offset Voltage (Low Power Mode) (Note 10)	LT1815S6/LT1816A, $40k\Omega$ Between I _{SET} and V ⁻ $T_A = 0$ °C to 70 °C $T_A = -40$ °C to 85 °C	•		2	7 9 10	mV mV mV
$\frac{\Delta V_{0S}}{\Delta T}$	Input Offset Voltage Drift	$T_A = 0$ °C to 70°C (Note 7) $T_A = -40$ °C to 85°C (Note 7)	•		10 10	15 30	μV/°C μV/°C
I _{OS}	Input Offset Current	T _A = 0°C to 70°C T _A = -40°C to 85°C	•		60	800 1000 1200	nA nA nA
I _B	Input Bias Current	T _A = 0°C to 70°C T _A = -40°C to 85°C	•		-2.4	±8 ±10 ±12	μΑ μΑ μΑ
e _n	Input Noise Voltage Density	f = 10kHz			6		nV/√Hz
i _n	Input Noise Current Density	f = 10kHz			1.3		pA/√Hz
R _{IN}	Input Resistance	V _{CM} = 1.5V to 3.5V Differential		1.5	5 750		MΩ kΩ
C _{IN}	Input Capacitance				2		pF
V _{CM}	Input Voltage Range (High)	Guaranteed by CMRR $T_A = -40$ °C to 85°C	•	3.5 3.5	4.1		V
	Input Voltage Range (Low)	Guaranteed by CMRR $T_A = -40^{\circ}C$ to 85°C	•		0.9	1.5 1.5	V
CMRR	Common Mode Rejection Ratio	$V_{CM} = 1.5V \text{ to } 3.5V$ $T_A = 0^{\circ}\text{C to } 70^{\circ}\text{C}$ $T_A = -40^{\circ}\text{C to } 85^{\circ}\text{C}$	•	73 71 70	82		dB dB dB
	Channel Separation	V_{OUT} = 1.5V to 3.5V, R_L = 100 Ω , LT1816/LT1817 T_A = 0°C to 70°C T_A = -40°C to 85°C	•	81 80 79	100		dB dB dB
	Minimum Supply Voltage	Guaranteed by PSRR T _A = -40°C to 85°C	•		2.5	4 4	V
A _{VOL}	Large-Signal Voltage Gain	V_{OUT} = 1.5V to 3.5V, R_L = 500 Ω T_A = 0°C to 70°C T_A = -40°C to 85°C	•	1.0 0.7 0.6	2		V/mV V/mV V/mV
		V_{OUT} = 1.5V to 3.5V, R_L = 100 Ω T_A = 0°C to 70°C T_A = -40°C to 85°C	•	0.7 0.5 0.4	1.5		V/mV V/mV V/mV
V _{OUT}	Maximum Output Swing (High)	$R_L = 500\Omega$, 30mV Overdrive $T_A = 0^{\circ}C$ to 70°C $T_A = -40^{\circ}C$ to 85°C	•	3.9 3.8 3.7	4.2		V V V
		$R_L = 100\Omega$, 30mV Overdrive $T_A = 0^{\circ}C$ to 70°C $T_A = -40^{\circ}C$ to 85°C	•	3.7 3.6 3.5	4		V V V



ELECTRICAL CHARACTERISTICS The • denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^{\circ}C$ (Note 8). $V_S = 5V$, 0V; $V_{CM} = 2.5V$, R_L to 2.5V unless otherwise noted. For the programmable current option (LT1815S6 or LT1816A), the I_{SET} pin must be connected to V^- through 75Ω or less, unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
V _{OUT}	Maximum Output Swing (Low)	$R_L = 500\Omega$, 30mV Overdrive $T_A = 0^{\circ}C$ to $70^{\circ}C$	•		0.8	1.1 1.2	V
		$T_A = -40^{\circ}$ C to 85°C $R_L = 100\Omega$, 30mV Overdrive $T_A = 0^{\circ}$ C to 70°C $T_A = -40^{\circ}$ C to 85°C	•		1	1.3 1.3 1.4 1.5	V V V
I _{OUT}	Maximum Output Current	V_{OUT} = 1.5V or 3.5V, 30mV Overdrive T_A = 0°C to 70°C T_A = -40°C to 85°C	•	±30 ±25 ±20	±50		mA mA mA
	Maximum Output Current (Low Power Mode) (Note 10)	LT1815S6/LT1816A; $40k\Omega$ Between I _{SET} and V ⁻ ; V_{OUT} = 1.5V or 3.5V, 30mV Overdrive T_A = 0°C to 70°C T_A = -40°C to 85°C	•	±30 ±25 ±20	±50		mA mA mA
I _{SC}	Output Short-Circuit Current	V_{OUT} = 2.5V, 1V Overdrive (Note 3) T_A = 0°C to 70°C T_A = -40°C to 85°C	•	±80 ±70 ±50	±140		mA mA mA
SR	Slew Rate	$A_V = -1$ (Note 5) $T_A = 0$ °C to 70°C $T_A = -40$ °C to 85°C	•	450 375 300	750		V/µs V/µs V/µs
FPBW	Full-Power Bandwidth	2V _{P-P} (Note 6)			120		MHz
GBW	Gain-Bandwidth Product	$f = 200kHz$, $R_L = 500Ω$, LT1815 $T_A = 0°C$ to $70°C$ $T_A = -40°C$ to $85°C$	•	140 130 120	200		MHz MHz MHz
		$f = 200kHz$, $R_L = 500Ω$, LT1816/LT1817 $T_A = 0$ °C to 70°C $T_A = -40$ °C to 85°C	•	130 110 100	200		MHz MHz MHz
	Gain-Bandwidth Product (Low Power Mode) (Note 10)	LT1815S6/LT1816A; $40k\Omega$ Between I_{SET} and V^- ; $f=200kHz$, $R_L=500\Omega$ $T_A=0^{\circ}C$ to $70^{\circ}C$ $T_A=-40^{\circ}C$ to $85^{\circ}C$	•	30 25 20	50		MHz MHz MHz
-3dB BW	-3dB Bandwidth	$A_V = 1, R_L = 500\Omega$			300		MHz
t _r , t _f	Rise Time, Fall Time	$A_V = 1$, 10% to 90%, 0.1V, $R_L = 100\Omega$			1.2		ns
t _{PD}	Propagation Delay	$A_V = 1,50\%$ to 50%, 0.1V, $R_L = 100\Omega$			1.5		ns
OS	Overshoot	$A_V = 1, 0.1V; R_L = 100\Omega$			25		%
t_S	Settling Time	$A_V = -1, 0.1\%, 2V$			15		ns
THD	Total Harmonic Distortion	$A_V = 2$, $f = 5MHz$, $V_{OUT} = 2V_{P-P}$, $R_L = 500\Omega$			-65		dB
dG	Differential Gain	$A_V = 2$, $V_{OUT} = 2V_{P-P}$, $R_L = 150\Omega$			0.08		%
dP	Differential Phase	$A_V = 2$, $V_{OUT} = 2V_{P-P}$, $R_L = 150\Omega$			0.13		Deg
R _{OUT}	Output Resistance	$A_V = 1$, $f = 1MHz$			0.24		Ω
I _S	Supply Current	LT1815 $T_A = 0^{\circ}\text{C to } 70^{\circ}\text{C}$ $T_A = -40^{\circ}\text{C to } 85^{\circ}\text{C}$	•		6.3	8 10 11	mA mA mA
		LT1816/LT1817, per Amplifier $T_A = 0^{\circ}C$ to $70^{\circ}C$ $T_A = -40^{\circ}C$ to $85^{\circ}C$	•		6.3	9 12 13	mA mA mA
	Supply Current (Low Power Mode) (Note 10)	LT1815S6/LT1816A, $40k\Omega$ Between I_{SET} and V^- , per Amplifier $T_A = 0^{\circ}C$ to $70^{\circ}C$ $T_A = -40^{\circ}C$ to $85^{\circ}C$	•		0.9	1.5 1.8 2.0	mA mA mA

www.DataSheet4U.com

ELECTRICAL CHARACTERISTICS The • denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^{\circ}C$ (Note 8). $V_S = 5V$, OV; $V_{CM} = 2.5V$, R_L to 2.5V unless otherwise noted.

For the programmable current option (LT1815S6 or LT1816A), the I_{SET} pin must be connected to V^- through 75 Ω or less, unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
I _{SET}	I _{SET} Pin Current (Note 10)	LT1815S6/LT1816A T _A = 0°C to 70°C T _A = -40°C to 85°C	• •	-150 -175 -200	-100		μΑ μΑ μΑ

Note 1: Absolute Maximum Ratings are those values beyond which the life of a device may be impaired.

Note 2: Differential inputs of $\pm 6V$ are appropriate for transient operation only, such as during slewing. Large sustained differential inputs can cause excessive power dissipation and may damage the part.

Note 3: A heat sink may be required to keep the junction temperature below absolute maximum when the output is shorted indefinitely.

Note 4: Input offset voltage is pulse tested and is exclusive of warm-up

Note 5: Slew rate is measured between $\pm 2V$ at the output with $\pm 3V$ input for $\pm 5V$ supplies and $2V_{P-P}$ at the output with a $3V_{P-P}$ input for single 5V

Note 6: Full-power bandwidth is calculated from the slew rate: FPBW = $SR/2\pi V_P$.

Note 7: This parameter is not 100% tested.

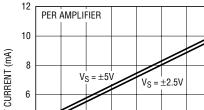
Note 8: The LT1815C/LT1816C/LT1817C are guaranteed to meet specified performance from 0°C to 70°C and are designed, characterized and expected to meet the extended temperature limits, but are not tested at -40°C and 85°C. The LT1815I/LT1816I/LT1817I are guaranteed to meet the extended temperature limits.

Note 9: Thermal resistance (θ_{JA}) varies with the amount of PC board metal connected to the package. The specified values are for short traces connected to the leads. If desired, the thermal resistance can be substantially reduced by connecting Pin 2 of the SOT-23, Pin 4 of the SO-8 and MS8, Pin 5 of the MS10 or the underside metal of the DD package to a large metal area.

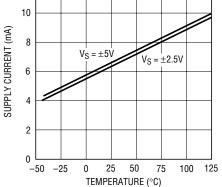
Note 10: A resistor of 40k or less is required between the I_{SET} and V⁻ pins of the LT1815S6 and the LT1816AMS. See the applications section for information on selecting a suitable resistor.

TYPICAL PERFORMANCE CHARACTERISTICS

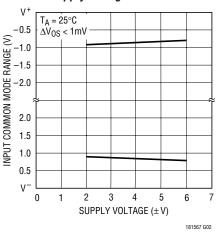
181567 G01



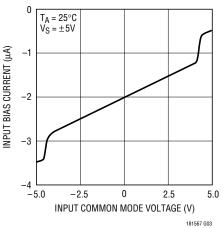
Supply Current vs Temperature

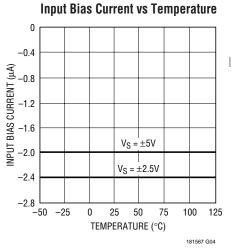


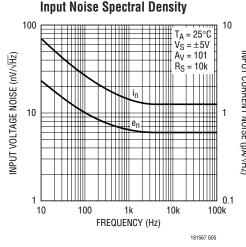
Input Common Mode Range vs Supply Voltage

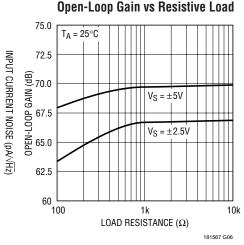


Input Bias Current vs Common Mode Voltage

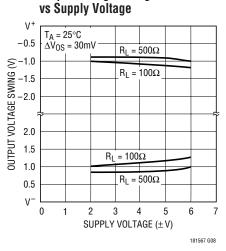




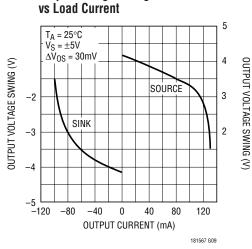




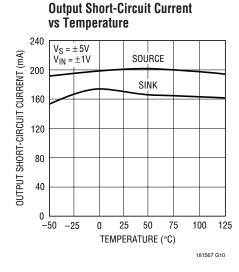
Open-Loop Gain vs Temperature 75.0 $V_S = \pm 5V$ $V_0 = \pm 3V$ 72.5 OPEN-LOOP GAIN (dB) $R_L = 500\Omega$ 70.0 67.5 $R_L = 100\Omega$ 65.0 62.5 60.0 -25 125 -50 0 25 50 75 100 TEMPERATURE (°C) 181567 G07

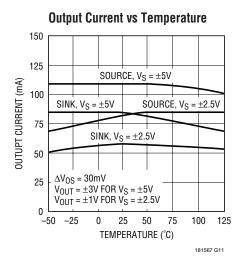


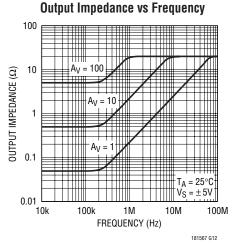
Output Voltage Swing



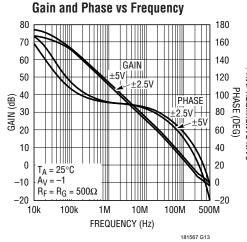
Output Voltage Swing

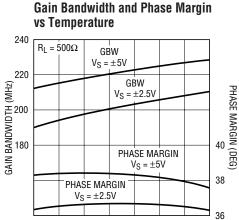












25 50

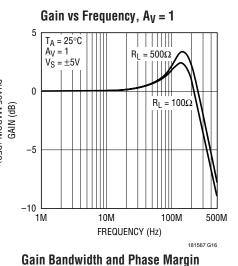
TEMPERATURE (°C)

0

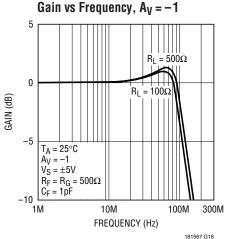
100 125

181567 G15

75

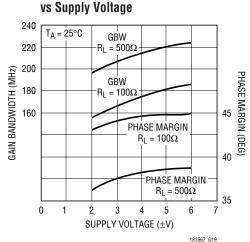


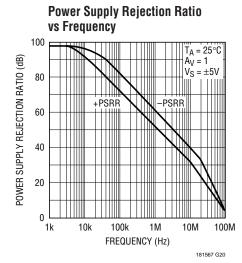
Gain vs Frequency, $A_V = 2$ 10 $R_L = 500\Omega$ 5 $R_L = 100\Omega$ GAIN (dB) $T_A = 25^{\circ}C$ -5 $A_V = 2$ $V_S = \pm 5V$ $R_F = R_G = 500\Omega$ $C_F = 1pF$ -10 1M 10M 100M 300M FREQUENCY (Hz) 181567 G17

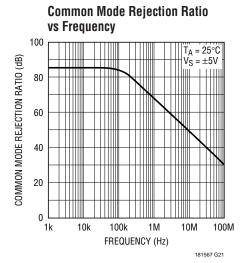


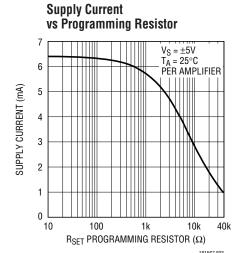
-25

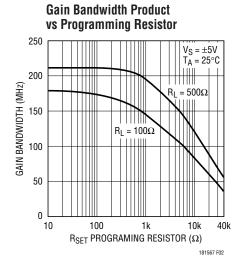
-50

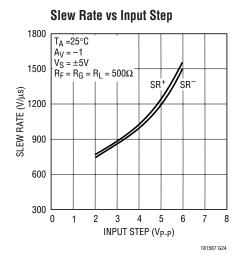


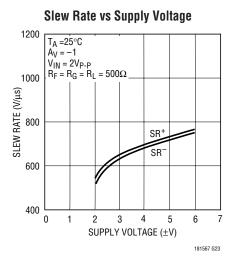


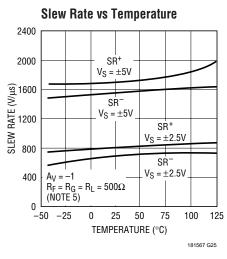


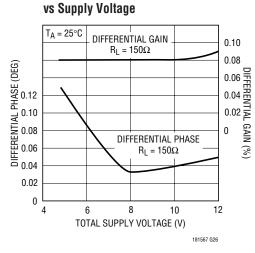




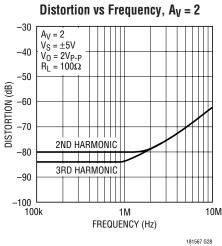


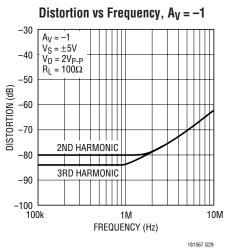


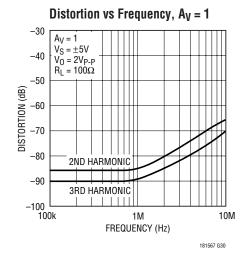




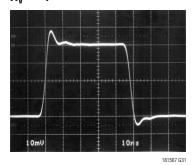
Differential Gain and Phase



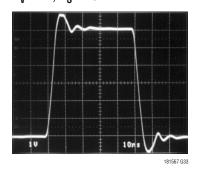




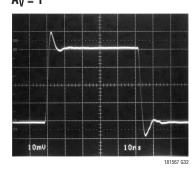
Small-Signal Transient, $A_V = -1$



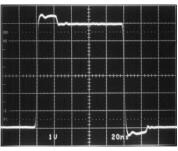
Large-Signal Transient, $A_V = -1$, $V_S = \pm 5V$



Small-Signal Transient, $A_V = 1$



Large-Signal Transient, $A_V = 1$, $V_S = \pm 5V$



APPLICATIONS INFORMATION

Layout and Passive Components

As with all high speed amplifiers, the LT1815/LT1816/LT1817 require some attention to board layout. A ground plane is recommended and trace lengths should be minimized, especially on the negative input lead.

Low ESL/ESR bypass capacitors should be placed directly at the positive and negative supply $(0.01\mu F)$ ceramics are recommended). For high drive current applications, additional $1\mu F$ to $10\mu F$ tantalums should be added.

The parallel combination of the feedback resistor and gain setting resistor on the inverting input combine with the input capacitance to form a pole that can cause peaking or even oscillations. If feedback resistors greater than 1k are used, a parallel capacitor of value:

$$C_F > R_G \cdot C_{IN}/R_F$$

should be used to cancel the input pole and optimize dynamic performance. For applications where the DC noise gain is 1 and a large feedback resistor is used, C_F should be greater than or equal to C_{IN} . An example would be an I-to-V converter.

Input Considerations

The inputs of the LT1815/LT1816/LT1817 amplifiers are connected to the base of an NPN and PNP bipolar transistor in parallel. The base currents are of opposite polarity and provide first-order bias current cancellation. Due to variation in the matching of NPN and PNP beta, the polarity of the input bias current can be positive or negative. The offset current, however, does not depend on beta matching and is tightly controlled. Therefore, the use of balanced source resistance at each input is recommended for applications where DC accuracy must be maximized. For example, with a 100Ω source resistance at each input, the 800nA maximum offset current results in only $80\mu V$ of extra offset, while without balance the $8\mu A$ maximum input bias current could result in a 0.8mV offset contribution.

The inputs can withstand differential input voltages of up to 6V without damage and without needing clamping or series resistance for protection. This differential input voltage generates a large internal current (up to 80mA),

which results in the high slew rate. In normal transient closed-loop operation, this does not increase power dissipation significantly because of the low duty cycle of the transient inputs. Sustained differential inputs, however, will result in excessive power dissipation and therefore this device should not be used as a comparator.

Capacitive Loading

The LT1815/LT1816/LT1817 are optimized for high bandwidth and low distortion applications. They can drive a capacitive load of 10pF in a unity-gain configuration and more with higher gain. When driving a larger capacitive load, a resistor of 10Ω to 50Ω should be connected between the output and the capacitive load to avoid ringing or oscillation. The feedback should still be taken from the output so that the resistor will isolate the capacitive load to ensure stability.

Slew Rate

The slew rate of the LT1815/LT1816/LT1817 is proportional to the differential input voltage. Therefore, highest slew rates are seen in the lowest gain configurations. For example, a 5V output step in a gain of 10 has a 0.5V input step, whereas in unity gain there is a 5V input step. The LT1815/LT1816/LT1817 are tested for a slew rate in a gain of -1. Lower slew rates occur in higher gain configurations.

Programmable Supply Current (LT1815/LT1816A)

In order to operate the LT1815S6 or LT1816A at full speed (and full supply current), connect the I_{SET} pin to the negative supply through a resistance of 75Ω or less.

To adjust or program the supply current and speed of the LT1815S6 or LT1816A, connect an external resistor (R_{SET}) between the I_{SET} pin and the negative supply as shown in Figure 1. The amplifiers are fully functional with $0 \leq R_{SET} \leq 40k$. Figures 2 and 3 show how the gain bandwidth and supply current vary with the value of the programming resistor R_{SET} . In addition, the Electrical Characteristics section of the data sheet specifies maximum supply current and offset voltage, as well as minimum gain bandwidth and output current at the maximum R_{SET} value of 40k.

www.DataSheet4U.com

APPLICATIONS INFORMATION

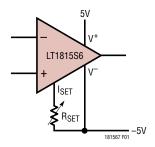


Figure 1. Programming Resistor Between I_{SET} and V⁻

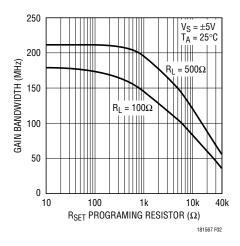


Figure 2. Gain Bandwidth Product vs R_{SET} Programming Resistor

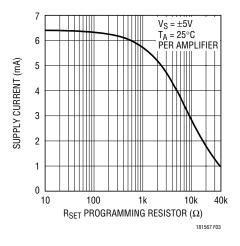


Figure 3. Supply Current vs R_{SET} Programming Resistor

Power Dissipation

The LT1815/LT1816/LT1817 combine high speed and large output drive in small packages. It is possible to exceed the maximum junction temperature specification (150°C) under certain conditions. Maximum junction temperature (T_J) is calculated from the ambient temperature (T_A), power dissipation per amplifier (P_D) and number of amplifiers (n) as follows:

$$T_J = T_A + (n \cdot P_D \cdot \theta_{JA})$$

Power dissipation is composed of two parts. The first is due to the quiescent supply current and the second is due to on-chip dissipation caused by the load current. The worst-case load induced power occurs when the output voltage is at 1/2 of either supply voltage (or the maximum swing if less than 1/2 the supply voltage). Therefore P_{DMAX} is:

$$\begin{split} P_{DMAX} &= (V^{+} - V^{-}) \bullet (I_{SMAX}) + (V^{+}/2)^{2}/R_{L} \text{ or } \\ P_{DMAX} &= (V^{+} - V^{-}) \bullet (I_{SMAX}) + (V^{+} - V_{OMAX}) \bullet \\ & (V_{OMAX}/R_{L}) \end{split}$$

Example: LT1816IS8 at 85°C, $V_S = \pm 5V$, $R_L = 100\Omega$

$$P_{DMAX} = (10V) \cdot (11.5mA) + (2.5V)^2 / 100\Omega = 178mW$$

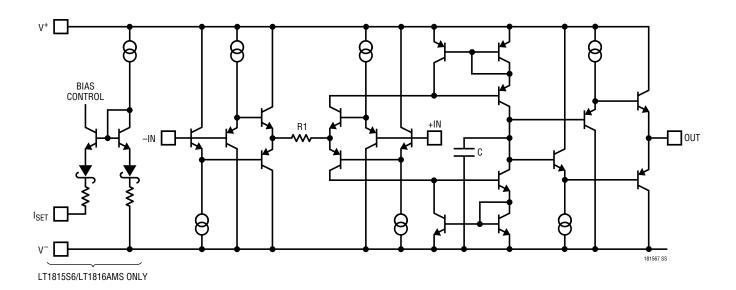
$$T_{JMAX} = 85^{\circ}C + (2 \cdot 178 \text{mW}) \cdot (150^{\circ}C/W) = 138^{\circ}C$$

Circuit Operation

The LT1815/LT1816/LT1817 circuit topology is a true voltage feedback amplifier that has the slewing behavior of a current feedback amplifier. The operation of the circuit can be understood by referring to the Simplified Schematic. Complementary NPN and PNP emitter followers buffer the inputs and drive an internal resistor. The input voltage appears across the resistor, generating current that is mirrored into the high impedance node.

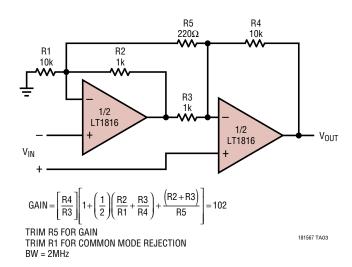
Complementary followers form an output stage that buffers the gain node from the load. The input resistor, input stage transconductance and the capacitor on the high impedance node determine the bandwidth. The slew rate is determined by the current available to charge the gain node capacitance. This current is the differential input voltage divided by R1, so the slew rate is proportional to the input step. Highest slew rates are therefore seen in the lowest gain configurations.

SIMPLIFIED SCHEMATIC (one amplifier)



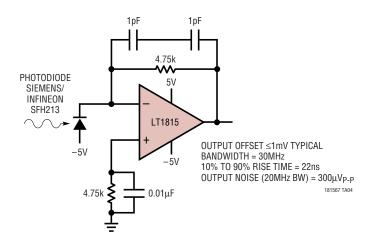
TYPICAL APPLICATIONS

Two Op Amp Instrumentation Amplifier

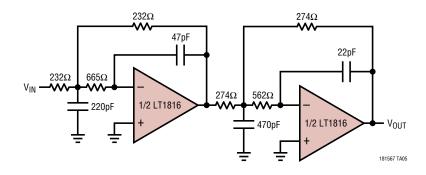


TYPICAL APPLICATIONS

Photodiode Transimpedance Amplifier

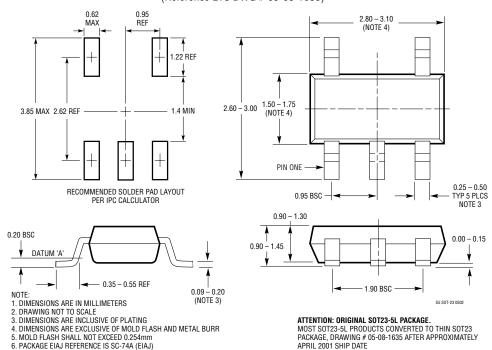


4MHz, 4th Order Butterworth Filter



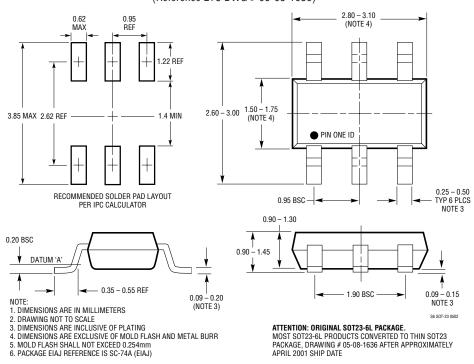
S5 Package 5-Lead Plastic SOT-23

(Reference LTC DWG # 05-08-1633) (Reference LTC DWG # 05-08-1635)



S6 Package 6-Lead Plastic SOT-23

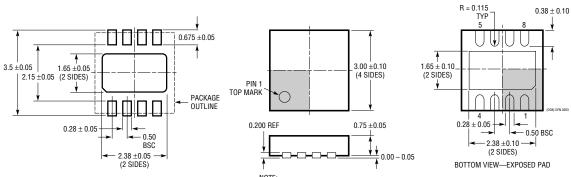
(Reference LTC DWG # 05-08-1634) (Reference LTC DWG # 05-08-1636)



www.DataSheet4U.com

DD Package 8-Lead Plastic DFN (3mm × 3mm)

(Reference LTC DWG # 05-08-1698)



RECOMMENDED SOLDER PAD PITCH AND DIMENSIONS

- NOTE:

 1. DRAWING TO BE MADE A JEDEC PACKAGE OUTLINE M0-229 VARIATION OF (WEED-1)

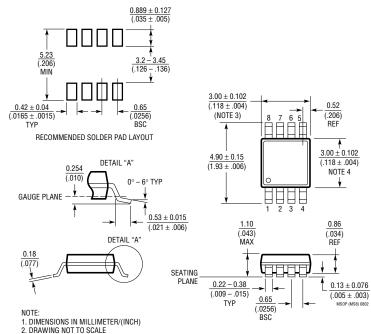
 2. ALL DIMENSIONS ARE IN MILLIMETERS

 3. DIMENSIONS OF EXPOSED PAD ON BOTTOM OF PACKAGE DO NOT INCLUDE MOLD FLASH. MOLD FLASH, IF PRESENT, SHALL NOT EXCEED 0.15mm ON ANY SIDE

 4. EXPOSED PAD SHALL BE SOLDER PLATED

MS8 Package 8-Lead Plastic MSOP

(Reference LTC DWG # 05-08-1660)

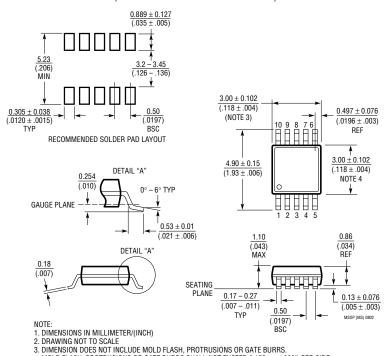


- 2. DRAWING NOT 10 SCALE
 3. DIMENSION DOES NOT INCLUDE MOLD FLASH, PROTRUSIONS OR GATE BURRS.
 MOLD FLASH, PROTRUSIONS OR GATE BURRS SHALL NOT EXCEED 0.152mm (.006') PER SIDE
 4. DIMENSION DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSIONS.
 INTERLEAD FLASH OR PROTRUSIONS SHALL NOT EXCEED 0.152mm (.006') PER SIDE
 5. LEAD COPLANARITY (BOTTOM OF LEADS AFTER FORMING) SHALL BE 0.102mm (.004') MAX



MS10 Package 10-Lead Plastic MSOP

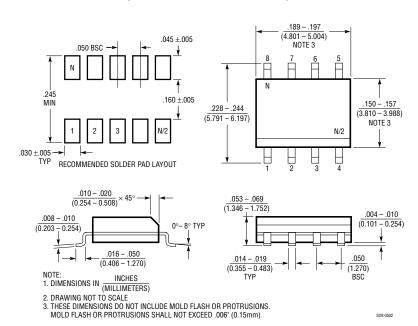
(Reference LTC DWG # 05-08-1661)



S8 Package 8-Lead Plastic Small Outline (Narrow .150 Inch)

MOLD FLASH, PROTRUSIONS OR GATE BURRS SHALL NOT EXCEED 0.152mm (.006') PER SIDE 4. DIMENSION DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSIONS. INTERLEAD FLASH OR PROTRUSIONS SHALL NOT EXCEED 0.152mm (.006°) PER SIDE 5. LEAD COPLANARITY (BOTTOM OF LEADS AFTER FORMING) SHALL BE 0.102mm (.004°) MAX

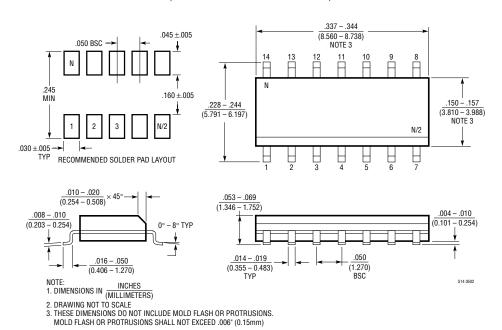
(Reference LTC DWG # 05-08-1610)



www.DataSheet4U.com

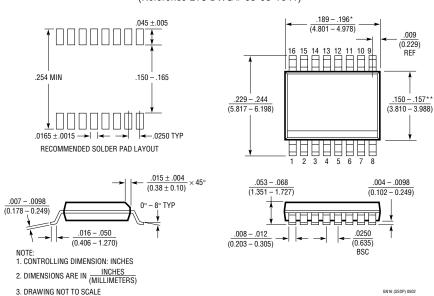
S Package 14-Lead Plastic Small Outline (Narrow .150 Inch)

(Reference LTC DWG # 05-08-1610)



GN Package 16-Lead Plastic SSOP (Narrow .150 Inch)

(Reference LTC DWG # 05-08-1641)



181567fa

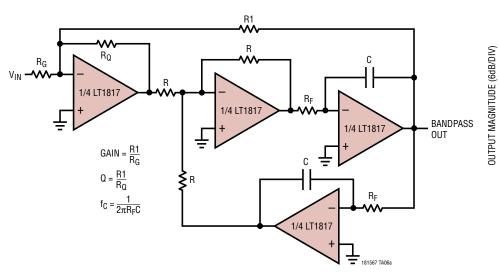


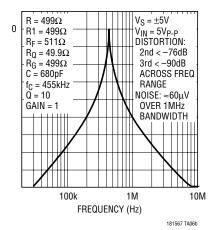
*DIMENSION DOES NOT INCLUDE MOLD FLASH. MOLD FLASH SHALL NOT EXCEED 0.006" (0.152mm) PER SIDE **DIMENSION DOES NOT INCLUDE INTERLEAD FLASH. INTERLEAD FLASH SHALL NOT EXCEED 0.010" (0.254mm) PER SIDE

TYPICAL APPLICATIONS

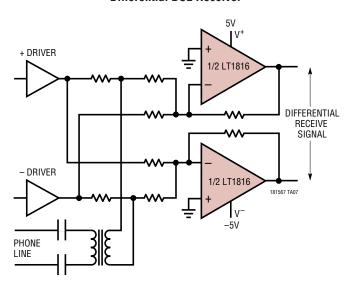
Bandpass Filter with Independently Settable Gain, Q and fc

455kHz Filter Frequency Response





Differential DSL Receiver



RELATED PARTS

PART NUMBER	DESCRIPTION	COMMENTS
LT1363/LT1364/LT1365	Single/Dual/Quad 70MHz, 1V/ns, C-Load™ Op Amp	Wide Supply Range: ±2.5V to ±15V
LT1395/LT1396/LT1397	Single/Dual/Quad 400MHz Current Feedback Amplifier	4.6mA Supply Current, 800V/µs, 80mA Output Current
LT1806/LT1807	Single/Dual 325MHz, 140V/µs Rail-to-Rail I/O Op Amp	Low Noise: 3.5nV/√Hz
LT1809/LT1810	Single/Dual 180MHz, 350V/µs Rail-to-Rail I/O Op Amp	Low Distortion: 90dBc at 5MHz
LT1812/LT1813/LT1814	Single/Dual/Quad 3mA, 100MHz, 750V/µs Op Amp	Low Power: 3.6mA Max at ±5V

C-Load is a trademark of Linear Technology Corporation.

LT/TP 0303 TWEET ALEST TEET ALEST OM LINEAR TECHNOLOGY CORPORATION 2001