

## OBSOLETE: FOR INFORMATION PURPOSES ONLY Contact Linear Technology for Potential Replacement

#### **FEATURES**

- Guaranteed 1% Initial Tolerance
- Guaranteed 0.3% Load Regulation
- Guaranteed 5A Output Current
- 100% Thermal Limit Burn-In
- 12A Transient Output Current

#### **APPLICATIONS**

- High Power Linear Regulator
- Battery Chargers
- Power Driver
- Constant-Current Regulator

# 5A Positive Adjustable Voltage Regulator

#### DESCRIPTION

The LT®138A series of adjustable regulators provide 5A output current over an output voltage range of 1.2V to 32V. The internal voltage reference is trimmed to less than 1%, enabling a very tight output voltage. In addition to excellent line and load regulation, with full overload protection, the LT138A incorporates new current limiting circuitry allowing large transient load currents to be handled for short periods. Transient load currents of up to 12A can be supplied without limiting, eliminating the need for a large output capacitor.

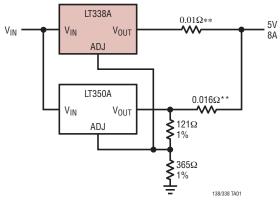
The LT138A is an improved version of the popular LM138 with improved circuit design and advanced process techniques to provide superior performance and reliability.

The graph below shows the significant improvement in output voltage tolerance achieved by using the LT138A or LT338A.

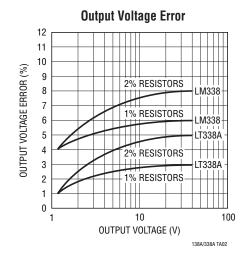
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### TYPICAL APPLICATION

#### Parallel Regulators for Higher Current\*



\*\*THIS CIRCUIT WILL NOT WORK WITH LM VERSION DEVICES
\*\*CURRENT SHARING RESISTORS DEGRADE REGULATION TO 1%



### **ABSOLUTE MAXIMUM RATINGS** (Note 1)

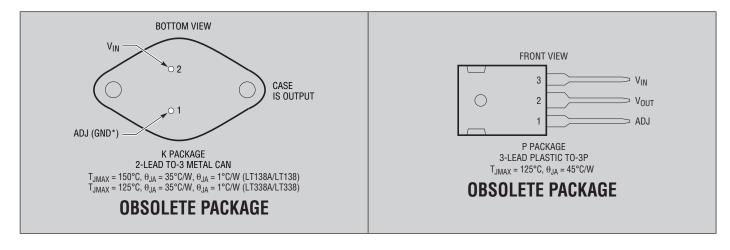
Power Dissipation	Internally Limited
Input-to-Output Voltage Differential	35V
Operating Junction Temperature Range	е
LT138A/LM138	55°C to 150°C
LT338A/LM338	0°C to 125°C

Storage Temperature Range65°C to	150°C
Lead Temperature (Soldering, 10 sec)	300°C

#### **PRECONDITIONING**

100% Thermal Limit Burn-In

#### PIN CONFIGURATION



## **ELECTRICAL CHARACTERISTICS** The $\bullet$ denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25\,^{\circ}\text{C}$ . (Note 2)

SYMBOL	PARAMETER	CONDITIONS		MIN	LT138A TYP	MAX	MIN	LM138 TYP	MAX	UNITS
							IVIIIV	Ш	IVIAA	_
$V_{REF}$	Reference Voltage	I <sub>OUT</sub> = 10mA, T <sub>J</sub> = 25°C		1.238	1.250	1.262	1 1 1 0		4.00	V
		$ 3V \le (V_{IN} - V_{OUT}) \le 35V, $ $ 10mA \le I_{OUT} \le 5A, P \le 50W $	•	1.225	1.250	1.270	1.119	1.24	1.29	V
$\Delta V_{OUT}$ $\Delta V_{IN}$	Line Regulation	$3V \le (V_{IN} - V_{OUT}) \le 35V$ , (Note 3)	•		0.005 0.02	0.01 0.04		0.005 0.02	0.01 0.04	%/V %/V
$\Delta V_{OUT}$ $\Delta V_{OUT}$	Load Regulation	$ \begin{array}{l} 10\text{mA} \leq I_{OUT} \leq 5\text{A, (Note 3)} \\ V_{OUT} \leq 5\text{V} \\ V_{OUT} \geq 5\text{V} \end{array} $			5 0.1	15 0.3		5 0.1	15 0.3	mV %
		$V_{OUT} \le 5V$ $V_{OUT} \ge 5V$	•		20 0.3	30 0.6		20 0.3	30 0.6	mV %
	Thermal Regulation	20ms Pulse			0.002	0.01		0.002	0.01	%/W
	Ripple Rejection	$V_{OUT}$ = 10V, f = 120Hz $C_{ADJ}$ = 0 $\mu$ F $C_{ADJ}$ = 10 $\mu$ F	•	60	60 75		60	60 75		dB dB
$I_{ADJ}$	Adjust Pin Current		•		45	100		45	100	μА
$\Delta I_{ADJ}$	Adjust Pin Current Change	$\begin{array}{l} 10mA \leq I_{OUT} \leq 5A, \\ 3V \leq (V_{IN} - V_{OUT}) \leq 35V \end{array}$	•		0.2	5		0.2	5	μА
	Minimum Load Current	$(V_{IN} - V_{OUT}) = 35V$	•		3.5	5		3.5	5	mA
I <sub>SC</sub>	Current Limit	$(V_{IN} - V_{OUT}) \le 10V$ DC 0.5ms Peak $(V_{IN} - V_{OUT}) = 30V, T_J = 25^{\circ}C$	•	5 6	8 12 1	2	5 6	8 12 1		A A A
$\Delta V_{OUT}$ $\Delta Temp$	Temperature Stability		•		1	2		1		%
ΔV <sub>OUT</sub> ΔTime	Long-Term Stability	T <sub>A</sub> = 125°C, 1000 Hours			0.3	1		0.3	1	%
en	RMS Output Noise (% of V <sub>OUT</sub> )	$10HZ \le f \le 10kHz$			0.001			0.003		%
$\theta_{\sf JC}$	Thermal Resistance Junction-to-Case	K Package				1			1	°C/W

## The ullet denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^{\circ}C$ .

					LT138A			LM138		
SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	MIN	TYP	MAX	UNITS
$V_{REF}$	Reference Voltage	I <sub>OUT</sub> = 10mA		1.238	1.250	1.262				V
		$3V \le (V_{IN} - V_{OUT}) \le 35V,$ $10\text{mA} \le I_{OUT} \le 5A, P \le 50W$	•	1.225	1.250	1.270	1.19	1.24	1.29	V
$\Delta V_{OUT}$ $\Delta V_{IN}$	Line Regulation	$3V \le (V_{IN} - V_{OUT}) \le 35V$ , (Note 3)	•		0.005 0.02	0.01 0.04		0.005 0.02	0.03 0.06	%/V %/V
$\Delta V_{OUT}$ $\Delta V_{OUT}$	Load Regulation	$ \begin{array}{c} 10mA \leq I_{OUT} \leq 5A, \text{ (Note 3)} \\ V_{OUT} \leq 5V \\ V_{OUT} \geq 5V \end{array} $			5 0.1	15 0.3		5 0.1	25 0.5	mV %
		$V_{OUT} \le 5V$ $V_{OUT} \ge 5V$	•		20 0.3	30 0.6		20 0.3	50 1	mV %
	Thermal Regulation	20ms Pulse			0.002	0.02		0.002	0.02	%/W



## **ELECTRICAL CHARACTERISTICS** The $\bullet$ denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25 \, ^{\circ}C$ .

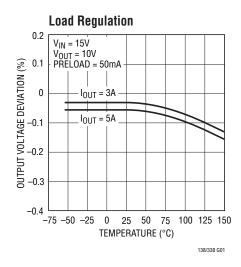
				LT138A		LM138				
SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	MIN	TYP	MAX	UNITS
	Ripple Rejection	$V_{OUT}$ = 10V, f = 120Hz $C_{ADJ}$ = 0 $\mu$ F $C_{ADJ}$ = 10 $\mu$ F	•	60	60 75		60	60 75		dB dB
I <sub>ADJ</sub>	Adjust Pin Current		•		45	100		45	100	μА
$\Delta I_{ADJ}$	Adjust Pin Current Change	$\begin{array}{l} 10mA \leq I_{OUT} \leq 5A, \\ 3V \leq (V_{IN} - V_{OUT}) \leq 35V \end{array}$	•		0.2	5		0.2	5	μА
	Minimum Load Current	$(V_{IN} - V_{OUT}) = 35V$	•		3.5	10		3.5	10	mA
I <sub>SC</sub>	Current Limit	$(V_{IN} - V_{OUT}) \le 10V$ DC 0.5ms Peak	•	5 6	8 12		5 6	8 12		A A
		$(V_{IN} - V_{OUT}) = 30V, T_J = 25^{\circ}C$			1	2		1		А
$\Delta V_{OUT}$ $\Delta Temp$	Temperature Stability		•		1	2		1		%
$\Delta V_{OUT}$ $\Delta Time$	Long-Term Stability	T <sub>A</sub> = 125°C, 1000 Hours			0.3	1		0.3	1	%
e <sub>n</sub>	RMS Output Noise (% of V <sub>OUT</sub> )	10Hz ≤ f ≤ 10kHz			0.001			0.003		%
$\theta_{\sf JC}$	Thermal Resistance Junction-to-Case	K Package				1			1	°C/W

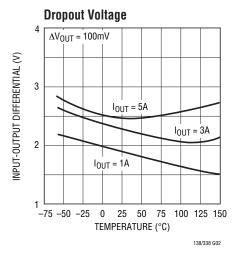
**Note 1:** Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. Exposure to any Absolute Maximum Rating condition for extended periods may affect device reliability and lifetime.

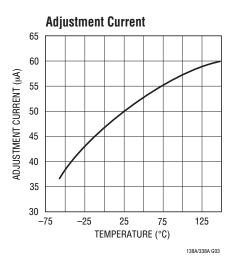
**Note 2:** Unless otherwise specified, these specifications apply:  $V_{IN} - V_{OUT} = 5V$  and  $I_{OUT} = 2.5A$ . These specifications are applicable for power dissipations up to 50W.

**Note 3:** See thermal regulation specifications for changes in output voltage due to heating effects. Load and line regulation are measured at a constant junction temperature by low duty cycle pulse testing.

#### TYPICAL PERFORMANCE CHARACTERISTICS



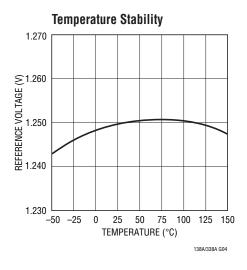


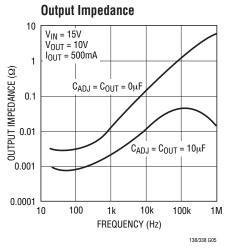


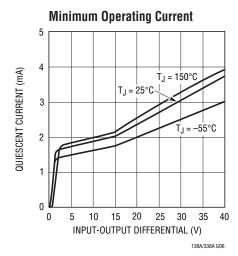
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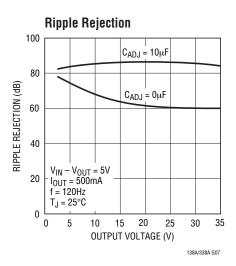


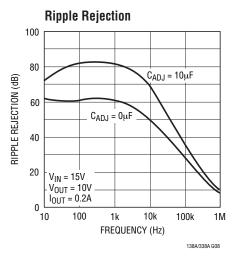
#### TYPICAL PERFORMANCE CHARACTERISTICS

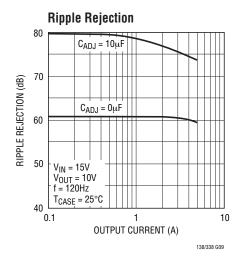


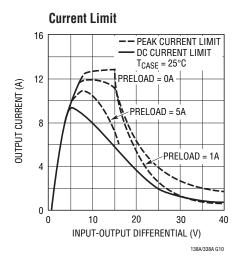


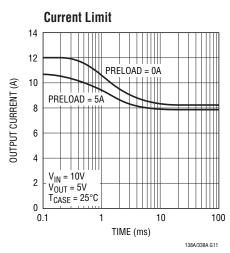


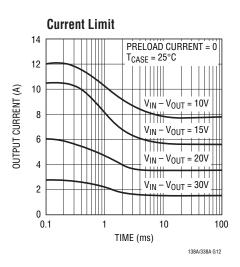






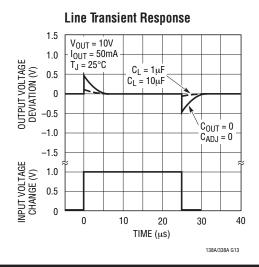


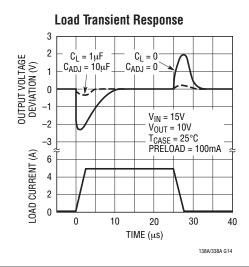




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#### TYPICAL PERFORMANCE CHARACTERISTICS





#### APPLICATIONS INFORMATION

#### General

The LT138A develops a 1.25V reference voltage between the output and the adjustable terminal (see Figure 1). By placing a resistor, R1, between these two terminals, a constant current is caused to flow through R1 and down through R2 to set the overall output voltage. Normally this current is the specified minimum load current of 5mA or 10mA. Because IAD, is very small and constant when compared with the current through R1, it represents a small error and can usually be ignored. It is easily seen from the output voltage equation, that even if the resistors were of exact value, the accuracy of the output is limited by the accuracy of V<sub>RFF</sub>. Earlier adjustable regulators had a reference tolerance of ±4% which is dangerously close to the ±5% supply tolerance required in many logic and analog systems. Further, even 1% resistors can drift 0.01%/°C. adding additional error to the output voltage tolerance.

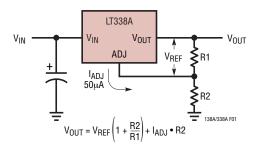


Figure 1. Basic Adjustable Regulator

For example, using 2% resistors and  $\pm 4\%$  tolerance for  $V_{REF}$ , calculations will show that the expected range of a 5V regulator design would be  $4.66V \le V_{OUT} \le 5.36V$  or approximately  $\pm 7\%$ . If the same example were used for a 15V regulator, the expected tolerance would be  $\pm 8\%$ . With these results most applications required some method of trimming, usually a trim pot. This solution is both expensive and not conductive to volume production.

One of the enhancements of Linear Technology's adjustable regulators over existing devices is the tightened initial tolerance of  $V_{REF}$ . This allows relatively inexpensive 1% or 2% film resistors to be used for R1 and R2 to set the output voltage within an acceptable tolerance.

With a guaranteed 1% reference, a 5V power supply design, using  $\pm 2\%$  resistors, would have a worst-case manufacturing tolerance of  $\pm 4\%$ . If 1% resistors are used, the tolerance will drop to  $\pm 2.5\%$ . A plot of the worst-case output voltage tolerance as a function of resistor tolerance is shown on the front page of this data sheet.

For convenience, a table of standard 1% resistor values is shown in Table 1.

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#### APPLICATIONS INFORMATION

Table 1. 0.5% and 1% Standard Resistance Values

1.00	1.47	2.15	3.16	4.64	6.81
1.02	1.50	2.21	3.24	4.75	6.98
1.05	1.54	2.26	3.32	4.87	7.15
1.07	1.58	2.32	3.40	4.99	7.32
1.10	1.62	2.37	3.48	5.11	7.50
1.13	1.65	2.43	3.57	5.23	7.68
1.15	1.69	2.49	3.65	5.36	7.87
1.18	1.74	2.55	3.74	5.49	8.06
1.21	1.78	2.61	3.83	5.62	8.25
1.24	1.82	2.67	3.92	5.76	8.45
1.27	1.87	2.74	4.02	5.90	8.66
1.30	1.91	2.80	4.12	6.04	8.87
1.33	1.96	2.87	4.22	6.19	9.09
1.37	2.00	2.94	4.32	6.34	9.31
1.40	2.05	3.01	4.42	6.49	9.53
1.43	2.10	3.09	4.53	6.65	9.76

Standard resistance values are obtained from the Decade Table by multiplying by multiples of 10. As an example, 1.21 can represent 1.21 $\Omega$ , 121 $\Omega$ , 1.21k etc.

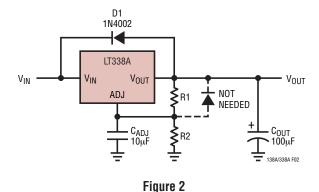
#### **Bypass Capacitors**

Input bypassing using a 1µF tantalum or 25µF electrolytic is recommended when the input filter capacitors are more than 5 inches from the device. Improved ripple rejection (80dB) can be accomplished by adding a 10µF capacitor from the ADJ pin to ground. Increasing the size of the capacitor to 20µF will help ripple rejection at low output voltage since the reactance of this capacitor should be small compared to the voltage setting resistor, R2. For improved AC transient response and to prevent the possibility of oscillation due to unknown reactive load, a 1µF capacitor is also recommended at the output. Because of their low impedance at high frequencies, the best type of capacitor to use is solid tantalum.

#### **Protection Diodes**

The LT138A/LT338A do not require a protection diode from the adjustment terminal to the output (see Figure 2). Improved internal circuitry eliminates the need for this diode when the adjustment pin is bypassed with a capacitor to improve ripple rejection.

If a very large output capacitor is used, such as a  $100\mu F$  shown in Figure 2, the regulator could be damaged or destroyed if the input is accidentally shorted to ground or crowbarred, due to the output capacitor discharging into the output terminal of the regulator. To prevent this, a diode D1 as shown, is recommended to safely discharge the capacitor.



#### **Load Regulation**

Because the LT138A is a three-terminal device, it is not possible to provide true remote load sensing. Load regulation will be limited by the resistance of the wire connecting the regulator to the load. The data sheet specification for load regulation is measured at the bottom of the package. Negative side sensing is a true Kelvin connection, with the bottom of the output divider returned to the negative side of the load. Although it may not be immediately obvious, best load regulation is obtained when the top of the resistor divider, R1, is connected *directly* to the case *not to the load*. This is illustrated in *Figure 3*. If R1 were connected to the load, the effective resistance between the regulator and the load would be:

$$R_P\left(\frac{R2+R1}{R1}\right)$$
,  $R_P$  = Parasitic Line Resistance

Connected as shown,  $R_P$  is not multiplied by the divider ratio.  $R_P$  is about  $0.004\Omega$  per foot using 16 gauge wire. This translates to 4mV/ft at 1A load current, so it is important to keep the positive lead between regulator and load as short as possible, and use large wire or PC board traces.



#### APPLICATIONS INFORMATION

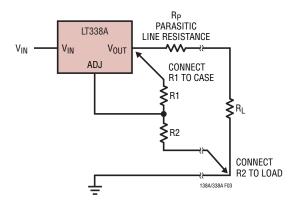
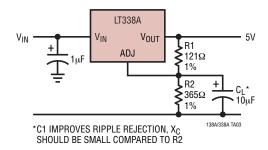


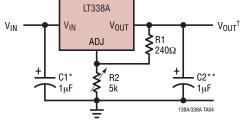
Figure 3. Connections for Best Load Regulation

#### TYPICAL APPLICATIONS

#### Improving Ripple Rejection



#### 1.2V to 25V Adjustable Regulator



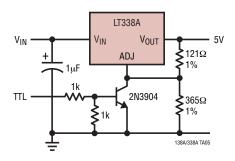
\*NEEDED IF DEVICE IS FAR FROM FILTER CAPACITORS \*\*OPTIONAL, IMPROVES TRANSIENT RESPONSE

$$^{\dagger}V_{OUT} = 1.25V \left(1 + \frac{R2}{R1}\right)$$

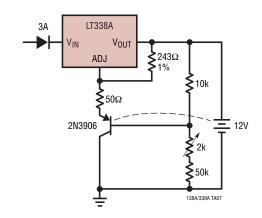


## TYPICAL APPLICATIONS

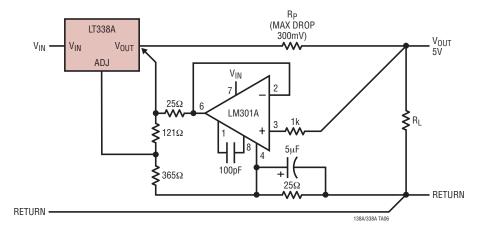
#### **5V Regulator with Shutdown**



#### **Temperature Compensated Lead Acid Battery Charger**

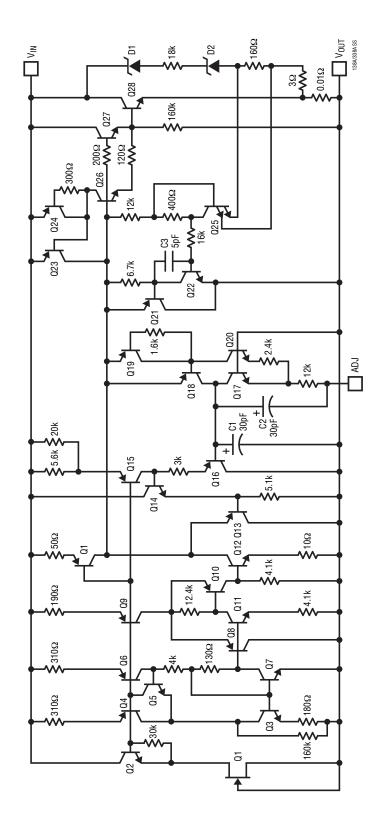


#### **Remote Sensing**



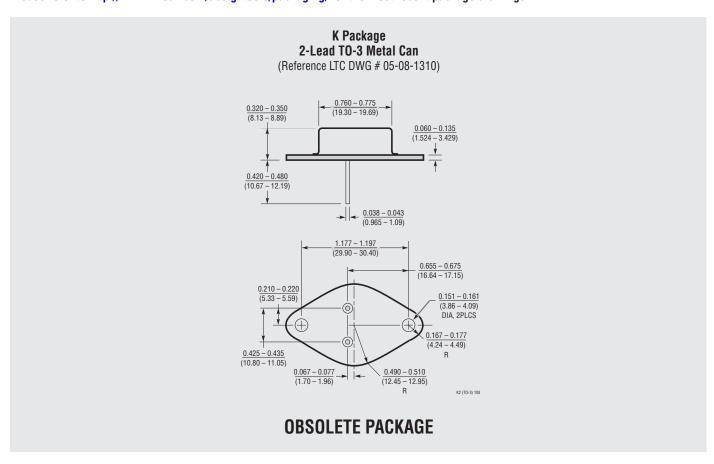


### SCHEMATIC DIAGRAM LT138A/LT338A



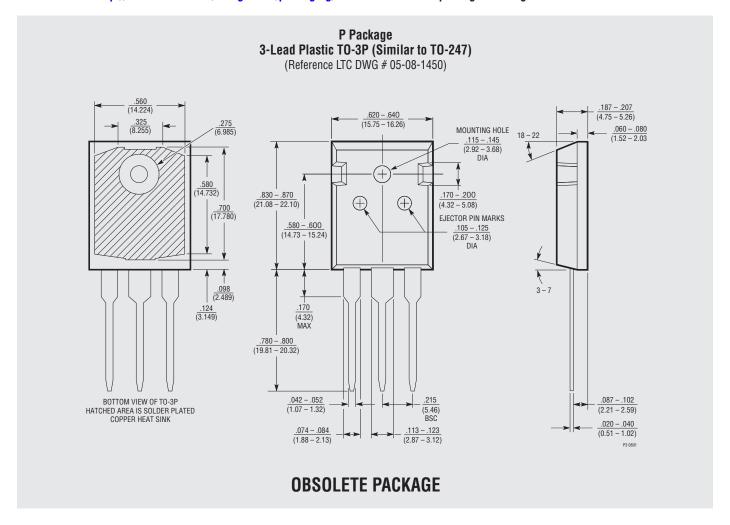
#### PACKAGE DESCRIPTION

Please refer to http://www.linear.com/designtools/packaging/ for the most recent package drawings.



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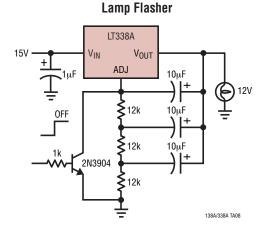


## **REVISION HISTORY** (Revision history begins at Rev D)

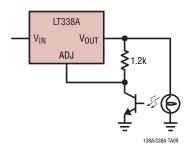
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D	05/15	Obsolete packaged parts.	1, 2, 12



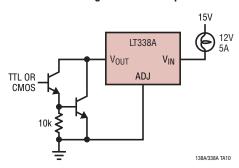
#### TYPICAL APPLICATION



#### **Automatic Light Control**



#### **Protected High Current Lamp Driver**



### **RELATED PARTS**

PART NUMBER	DESCRIPTION	COMMENTS
LT1083/LT1084/ LT1085	3A/5A/7.5A Low Dropout Regulators	Fixed Outputs, V <sub>IN</sub> Up to 30V
LT1580	7A Fast Transient Response Regulator with 0.7V Dropout	For 3.3V to 2.xxV Applications
LT1581	10A Fast Transient Response Regulator	For 3.3V to 2.xxV Applications
LT1584/LT1585/ LT1587	7A/4.6A/3A Low Dropout Fast Transient Response Regulator	For 1.2V to 3.3V Outputs from 5V
LT1764	3A Fast Transient Response Regulator	Dropout Voltage 340mV, Low Noise: 40µV <sub>RMS</sub>