Ordering number : ENA1999

SANYO

SANYO Semiconductors
DATA SHEET

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LV5028TT — Bi-CMOS IC LED Driver IC

Overview

LV5028TT is a High voltage LED drive controller which drives LED current up to 3A with external MOSFET. LV5028TT is realized very simple LED circuits with a few external parts. It corresponds to various wide dimming controls including the TRIAC dimming control.

Note) This LV5028TT is designed or developed for general use or consumer appliance. Therefore, it is NOT permitted to use for automotive, communication, office equipment, industrial equipment.

Functions

- High voltage LED controller
- Switching frequency: 50kHz
- Soft Start function
- Built-in TRIAC stabilized function
- Built-in circuit of detection of overvoltage of CS pin.
- Corresponds to TRIAC stabilized
- Selectable reference Voltage
  -Internal 0.605V & External Input Voltage
- Low noise switching system/skip frequency function
  - 5 stages skip mode Frequency
  - Soft driving

Specifications

Maximum Ratings at Ta = 25°C

Parameter	Symbol	Conditions	Ratings	Unit
Maximum input voltage	V_{IN} max (Note1)		-0.3 to 42	V
REF_OUT, REF_IN, CS, PWM_D, ACS			-0.3 to 7	V
OUT1 pin	V_{OUT}_abs		-0.3 to 42	V
OUT2 pin	V_{OUT}2_abs		-0.3 to 42	V
Allowable power dissipation	Pd max	With specified board*	1.0	W

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Parameter	Symbol	Conditions	Ratings	Unit
Junction temperature	Tj		150	°C
Operating junction temperature	Topj (Note2)		-30 to +125	°C
Storage temperature	Tstg		-40 to +150	°C

*1 Specified board: 58.0mm × 54.0mm × 1.6mm (glass epoxy board)

Note1) Absolute maximum ratings represent the values which cannot be exceeded for any length of time.

Note2) Even when the device is used within the range of absolute maximum ratings, as a result of continuous usage under high temperature, high current, high voltage, or drastic temperature change, the reliability of the IC may be degraded. Please contact us for the further details.

Recommended Operating Conditions at Ta = 25°C

Parameter	Symbol	Conditions	Ratings	Unit
Input voltage	V_{IN}		8.5 to 24	V

* Note : supply the stabilized voltage.

Electrical Characteristics at Ta = 25°C, V_{CC} = 5.0V

Parameter	Symbol	Conditions	Ratings min	Ratings typ	Ratings max	Unit
Reference voltage block						
Built-in reference voltage	VREF		0.585	0.605	0.625	V
VREF V_{IN} line regulation	VREF_LN	V_{IN} = 8.5 to 24V		±0.5		%
Reference output voltage	REFOUT	I_{REFOUT} = 0.5mA		3.0		V
- Maximum load	REFOUT_MAX		0.5			mA
- equivalent output impedance	REFOUT_RO			10		Ω
Under voltage lockout						
Operation start Input voltage	UVLOON		8	9	10	V
Operation stop input voltage	UVLOOFF		6.3	7.3	8.3	V
Hysteresis voltage	UVLOH			1.7		V
Oscillation						
Frequency	FOSC		40	50	60	kHz
Maximum ON duty	MAXDuty			93		%
Comparator						
Input offset voltage (Between CS and VREF)	V_{IO}_VR			1	10	mV
Input offset voltage (Between CS and REFOUT)	V_{IO}_RI			1	10	mV
Input current	I_{IO}SC			160		nA
	I_{IO}REF			80		nA
CS pin max voltage	VOM				1	V
malfunction prevention mask time	TMSK			150		ns
Thermal protection circuit						
Thermal shutdown temperature	TSD	*Design guarantee		165		°C
Thermal shutdown hysteresis	ΔTSD	*Design guarantee		30		°C
Drive Circuit						
OUT sink current	I_OI		500	1000		mA
OUT source current	I_OO			120		mA
Minimum On time	TMIN			200	300	ns
TRIAC Stabilization circuit						
Threshold of OUT2	V_{ACS}	OUT2 = High [less than right record]	2.8	3.0	3.2	V
OUT2 sink current	I_O2I	V_{IN} = 12V, OUT2 = 6V		0.6		mA
OUT2 source current	I_O2O	V_{IN} = 12V, OUT2 = 6V		0.6		mA
V_{CC} current						
UVLO mode V_{IN} current	I_{CC}OFF	V_{IN} < UVLOON		80	120	µA
Normal mode V_{IN} current	I_{CC}ON	V_{IN} > UVLOON, OUT = OPEN		0.8		mA

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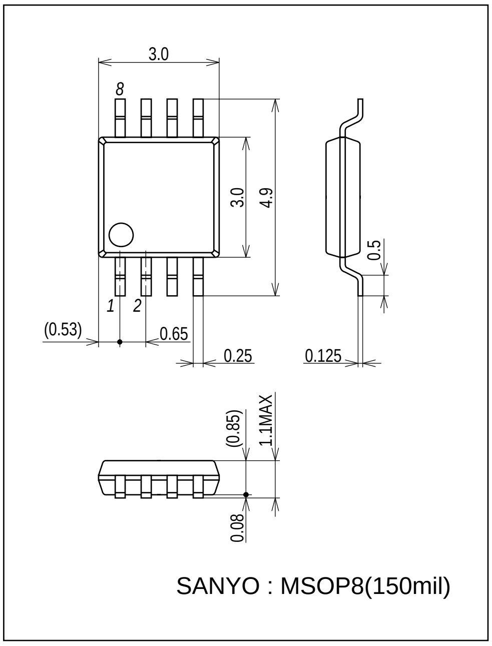
Parameter	Symbol	Conditions	Ratings			Unit
			min	typ	max	
V_{IN} over voltage protection circuit						
V_{IN} over voltage protection voltage	V_{IN}OVP		24	27	30	V
V_{IN} current at OVP	I_{IN}OVP	V_{IN} = 30V	0.7	1.0	1.5	mA
CS terminal abnormal sensing circuit						
Abnormal sensing voltage	CSOCP			1.9		V

*: Design guarantee (value guaranteed by design and not tested before shipment)

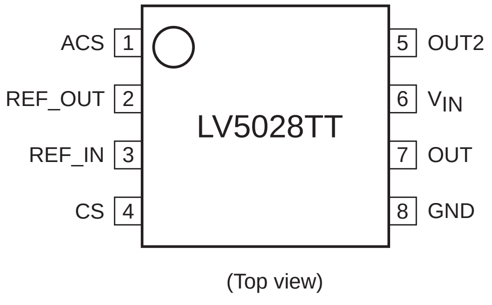
Package Dimensions

unit : mm (typ)

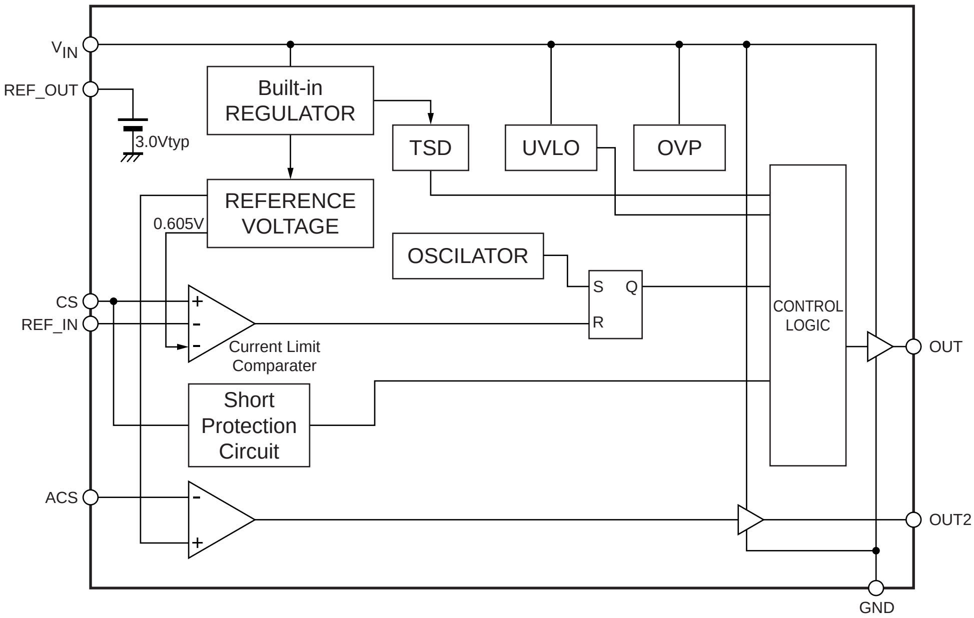
3245B



Pin Assignment

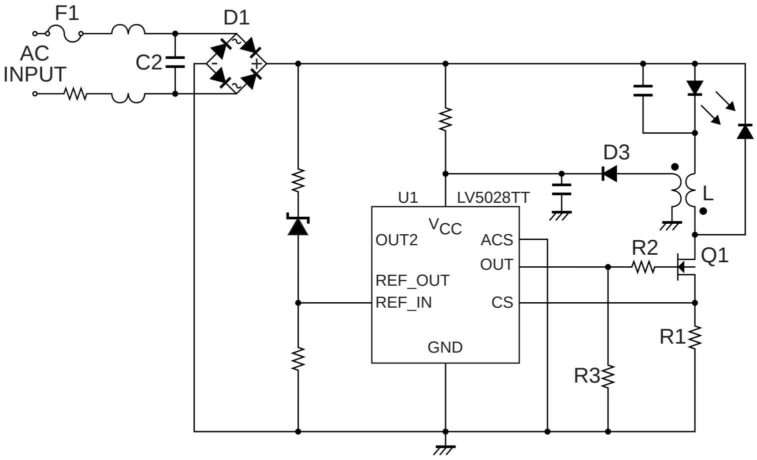


Block Diagram

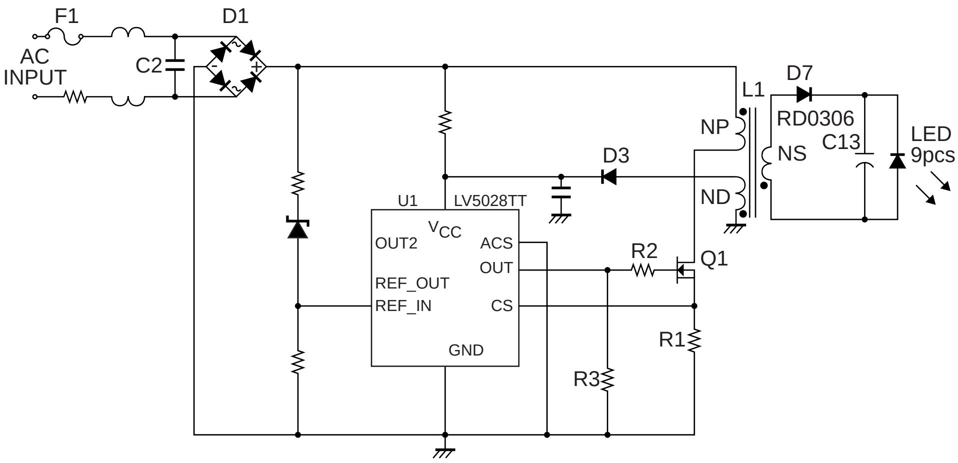


Sample Application Circuit

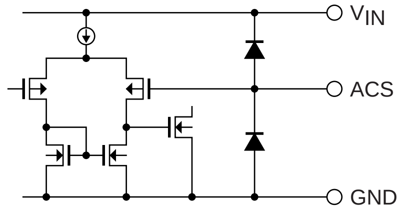
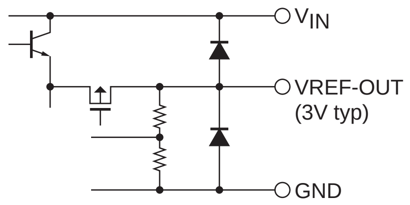
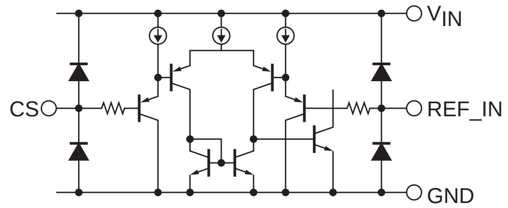
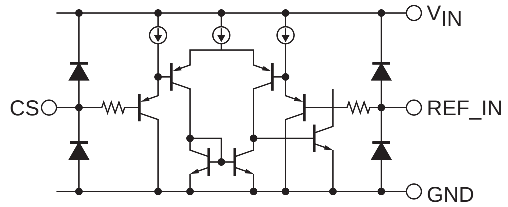
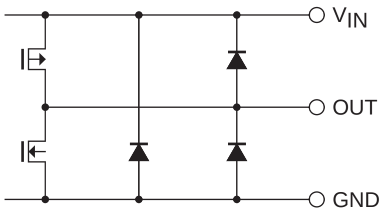
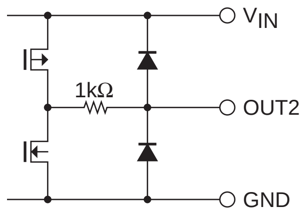
Non isolation



Isolation



Pin Functions

Pin No.	Pin name	Pin function	Equivalent circuit
1	ACS	ACS pin senses AC Voltage. If this function isn't used, please connect GND.	
2	REF_OUT	Built-in 3V Regulate out Pin. If this function isn't used, please connect to nothing.	
3	REF_IN	External LED current Limit Setting pin. If less than VREF (0.61V) voltage is input, Peak current value is used at the input voltage. If more than REF_IN voltage is input, it is done at VREF voltage. If this function isn't used, please connect nothing.	
4	CS	LED current sensing in. If this terminal voltage exceeds VREF (Or REF_IN), external FET is OFF. And if the voltage of the terminal exceeds 1.9V, LV5028TT turns to latch-off mod	
5	GND	GND pin.	
6	OUT	Driving the external FET Gate Pin.	
7	V_{IN}	Power supply pin. Operation : V_{IN} > UVLOON Stop: V_{IN} < UVLOOFF Switching Stop : V_{IN} > V_{IN}OVP	
8	OUT2	This pin drive the FET which is stabilized the TRIAC dimming application. If ACS is less than 3V, OUT2 turn High voltage. If this function isn't used, please connect nothing.	

LED current and inductande setting

• Relation ship beween REF_IN and CS pin voltage(Power Factor Crrection(PFC))

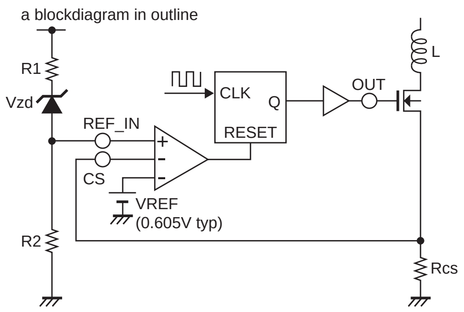
The output current value is the average of the current value that flows during one cycle. The current value that flows into coil is a triangular wave shown in the figure below. Make sure to set Ipk so that (average of current value at one cycle) is equal to (LED current value).Ipk is set by the relationship between REF_IN voltage and Rcs voltage.

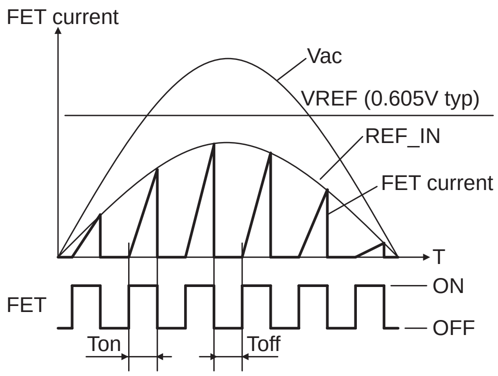
This relationship make Power Factor Correction (PFC).Therefore, it is available to make LED current a sine curve.

• Setting Zener voltage

Vzd depend on LED voltage (VF). Choose Zener diode around Vf (LED voltage).When VAC voltage is lower than Vf, LED operation is not normal. Using Zener diode prevents incorrect operating during VAC voltage lower than Vf. In detail, refer to [LED current and inductance setting]

In case of REF_IN pin open, this error amplifier negative input(-) is under control of internal VREF voltage (0.605Vtyp).





$$Ipk = \frac{(Vac\text{-}Vzd) \times \dfrac{R2}{R1+R2}}{Rcs}$$

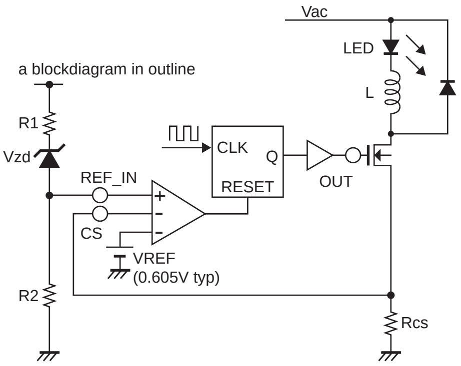
Ipk: peak inductor current Vf: LED forward voltage drop Vac: effective value, R.M.S value VREF: Built-in reference voltage (0.605V) VREF_IN: REF_IN voltage (6 pin) Rs: External sense resistor Vzd: Zener diode voltage (REF_IN pin)

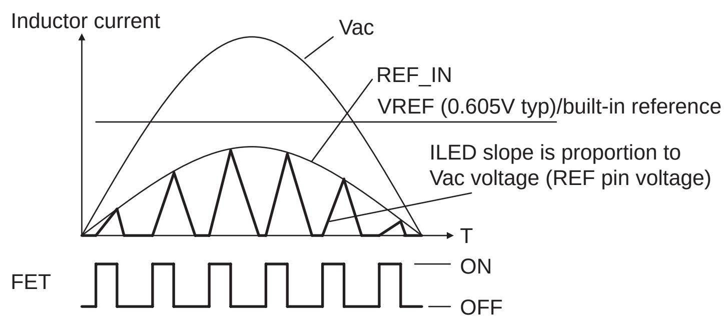
LED current and inductance setting

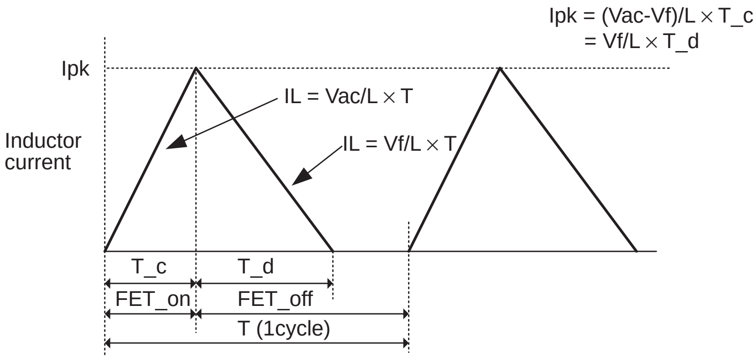
It is available to use both no-isolation and isolation applications.

(For non-isolation application)

The output current value is the average of the current value that flows during one cycle. The current value that flows into coil is a triangular wave shown in the figure below. Make sure to set IL_PK so that (average of current value at one cycle) is equal to (LED current value).







Given that the period when current flows into coil is

$$\text{DutyI} = \frac{T\_c+T\_d}{T}$$

$$Ipk \times \frac{1}{2} \times (Duty \times T)/T = ILED$$

$$Ipk \times \frac{2 \times ILED}{DutyI} \quad \text{(1) since} \quad Ipk \times \frac{VREF\_IN}{Rcs}$$

$$Rcs \times \frac{VFEF\_IN}{Ipk} = \frac{DutyI \times VFEF\_IN}{2ILED} \quad (2)$$

Ipk: peak inductor current
Vf: LED forward voltage drop
Vac: effective value(R.M.S value)
VREF: Built-in reference voltage (0.605V)
VREF_IN: REF_IN voltage (6 pin)
Rs: External sense resistor
Vzd: Zener diode voltage (REF_IN pin)

Since formula for LED current is different between on period and off period as shown above,

$$Ipk \times \frac{Vac\text{-}Vf}{L} \times T\_c = \frac{Vf}{L} \times T\_d \quad (3)$$

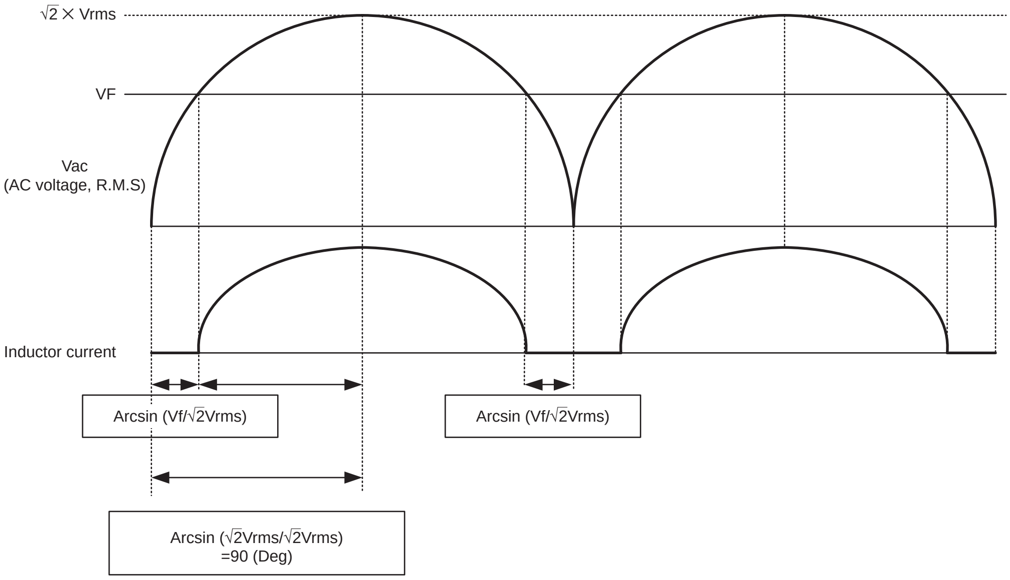
Since $T\_c + T\_d = DutyI \times T,\ T\_c = DutyI \times T - T\_d$ (4)

Based on the result of (3) and (4), $T\_d = DutyI \times T \times \frac{Vac\text{-}Vf}{Vac}$ (5)

To obtain L from the equation (1), (3), (5),

$$L \times \frac{Vf \times DutyI}{2 \times ILED} \times DutyI \times T = \frac{Vac - Vf}{Vac} = \frac{Vf}{2 \times ILED} \times \frac{1}{fosc} \times \frac{Vac - Vf}{Vac} \times (DutyI)^2 \quad (6)$$

Since LED and inductor are connected in serial in non-isolation mode, LED current flows only when AC voltage exceed VF.



Given that the ratio of inductor current to AC input is DutyAC.

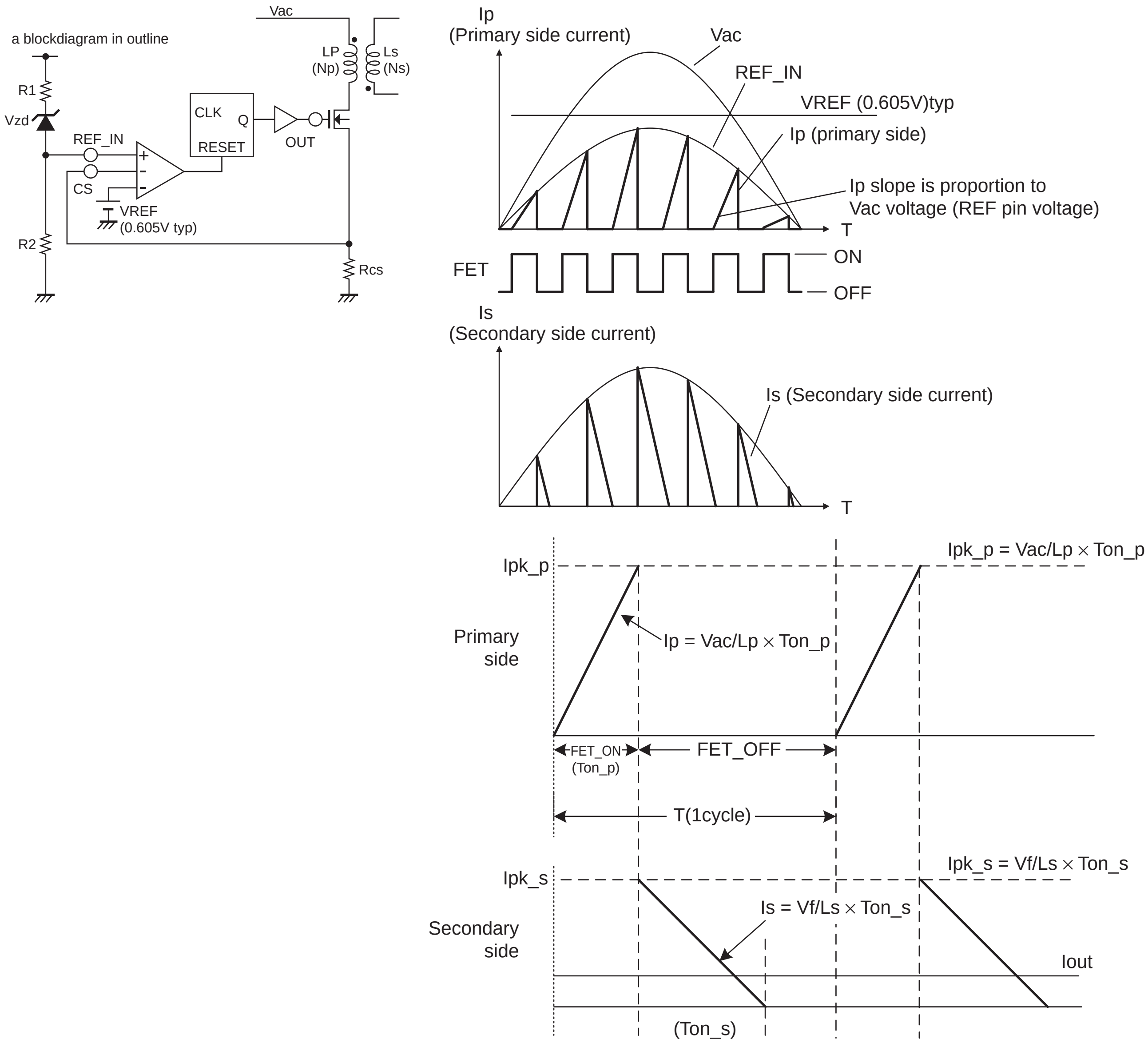
$$DutyAC = \frac{90 - \arcsin\left(\frac{Vf}{\sqrt{2}Vrms}\right)}{90}$$

Since the period when the inductor current flows are limited by DutyAC, the formula (6) is represented as follows:

$$L = \frac{Vf}{2 \times ILED} \times \frac{1}{fosc} \times \frac{Vac - Vf}{V_{IN}} \times (DutyI)^2 \times \left(\frac{90 - \arcsin\left(\frac{Vf}{\sqrt{2}Vrms}\right)}{90}\right)^2 \qquad (7)$$

(for Isolation circuit)

Using the circuit diagram below, the wave form of the current that flows to Np and Ns is as follows.
Current waveform flows to primary side and secondary.



[Inductance Lp of primary side and sense resistor Rs]

If a peak current flow to transformer is represented as Ipk_p, the power (Pin) charged to the transformer on primary side can be represented as:

$$Pin = \frac{1}{2} \times Lp \times (Ipk\_p)^2 \times fosc \quad (11)$$

$$Ipk\_p = \frac{Vac}{Lp} \times Ton\_p \quad (12)$$

$$Lp = \frac{Vac^2 \times Ton\_p^2 \times fosc}{2 \times Pin} = \frac{Vac^2 \times Don\_p^2}{2 \times Pin \times fosc} \quad (13)$$

$$(Don\_p = \frac{Ton\_p}{T} = Ton\_p \times fosc),$$

To substitute the following to the formula below,

$$\because \eta = \frac{Pout}{Pin} \quad (14)$$

$$\therefore Lp = \frac{Vac^2 \times Ton\_p^2 \times fosc \times \eta}{2 \times Pout} = \frac{Vac^2 \times Don^2 \times \eta}{2 \times Pout \times fosc} \quad (15)$$

Sense resistor is obtained as follows.

$$Rs = \frac{VREF\_IN}{Ipk\_p} = \frac{VREF\_IN \times Lp}{Vac \times Ton\_p} = \frac{VREF\_IN \times Lp}{Vac \times Don\_p \times T} \quad (16)$$

[Inductance Ls of secondary side]
Since output current Iout is the average value of current flows to transformer of secondary side

$$Iout = Ipk\_s \times \frac{Ton\_s}{T} \times \frac{1}{2} = \frac{Ipk\_s \times Don\_s}{2} \left(Don\_s = \frac{Ton\_s}{T} = Ton\_s \times fosc\right) \quad (17)$$

$$Ipk\_s = \frac{Vout}{Ls} \times Ton\_s = \frac{Vout}{Ls} = \frac{Don\_s}{fosc} \quad (18)$$

$$Ls = \frac{Vout \times T \times Don\_s^2}{2 \times Iout} = \frac{Vout \times Don\_s^2}{2 \times Iout \times fosc} = \frac{Vout^2 \times Don\_s^2}{2 \times Pout \times fosc} \quad (19)$$

Calculation of the ratio of transformer coil on primary side and secondary side
Since ratio and inductance of transformer coil is

$$\frac{Ns}{Np} = \frac{\sqrt{Ls}}{\sqrt{Lp}} \quad (20)$$

substituted equations (15), (19) for (20)

$$\therefore \frac{Np}{Ns} = \frac{Vac}{Vout} \times \sqrt{\eta} \times \frac{Don\_p}{Don\_s} \quad (21)$$

Calculation of transformer coil on primary side and secondary side

$$N = \frac{Vac \times 10^8}{2 \times \Delta B \times Ae \times fosc} \quad (22)$$

ΔB: variation range of core flux density [Gauss]
Ae: core section area [cm^2]

To use Al (L value at 100T),

$$N = \sqrt{\frac{L}{Al}} \times 10^2 \quad (23)$$

L: inductance [μH]
Al: L value at 100T [uH/N^2]

lg (Air gap) is obtained as follows:

$$lg = \frac{\mu_r \, \mu_0 \, N^2 \, A_e \, 10^2}{L} \quad (24)$$

μr: relative magnetic permeability, μr = 1
μ0: vacuum magnetic permeability $\mu 0 = 4\pi * 10^{-7}$
N: turn count [T]
Ae: core section area [m^2]
L: inductance [H]

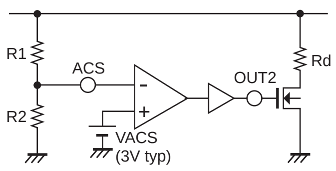
Bleeder current cuircuit for TRIAC dimmer

1. Operating voltage setting
   ACS pin voltage set operating voltage at OUT2. ACS pin threshold volage is 3V typ.
   OUT2 operating voltage is set by R1 and R2. R1 and R2 is determined below.

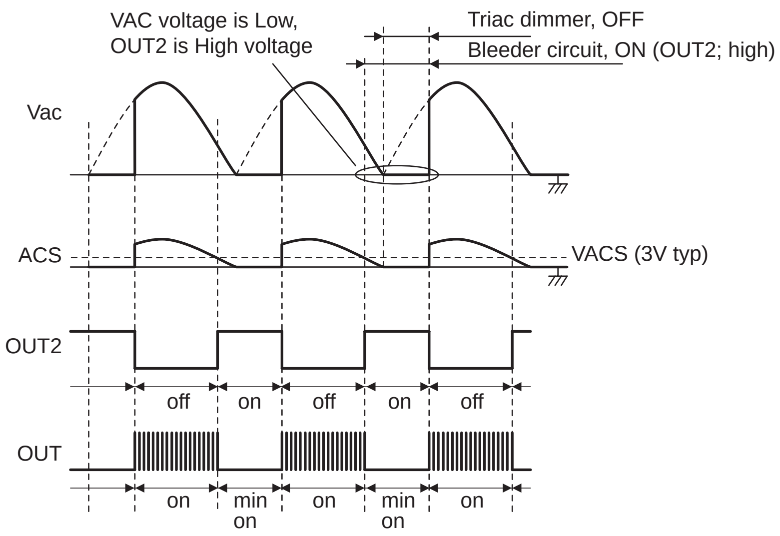
$$ACS = Vac \times \frac{R2}{R1+R2}$$

2. Bleeder current setting
   Rd set hold current at Triac dimmer.
   Bleeder current is set at Rd depending on Triac dimmer.

a blockdiagram in outline



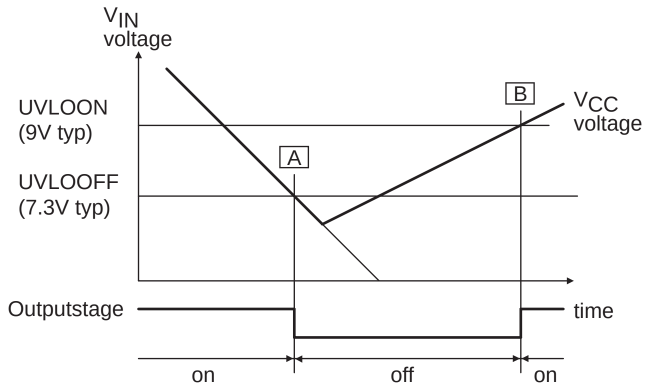
a blockdiagram in outline



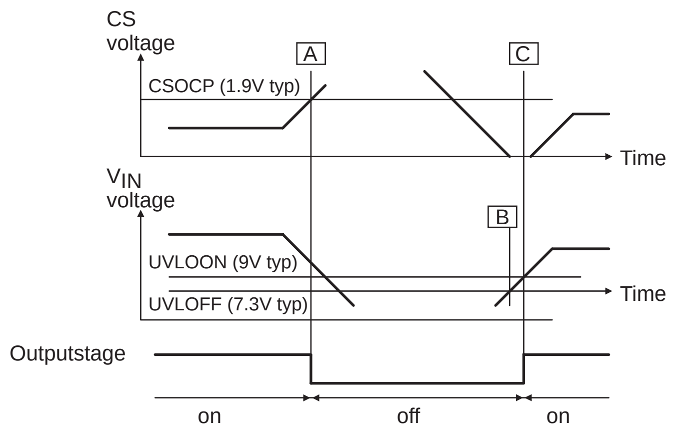
Description of operation
protection function

	tilte	outline	monitor point	note
1	UVLO	Under voltage lock out	V_{CC} voltage	
2	OCP	Over current protection	CS voltage	available FET current
3	OVP	Over voltage protection	V_{CC} voltage	
4	OTP (TSD)	Over Temperature Protection (Thermal Shut Down)	PN Junction temperature	

1. UVLO (Under voltage lock out)
If V_{IN} voltage is 7.3V or lower, then UVLO operates and the IC stops. When UVLO operates, the power supply current of the IC is about 80μA or lower. If V_{IN} voltage is 9V or higher, then the IC starts switching operation.



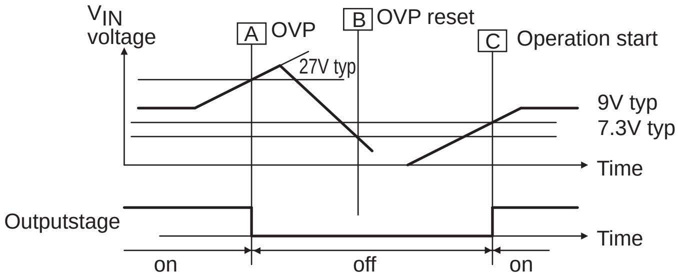
2. UVLO (Under voltage lock out)
The CS pin sense the current through the MOS FET switch and the primary side of the transformer. This provides an additional level of protection in the event of a fault. If the voltage of the CS pin exceeds VCSOCP (1.9V typ) (A), the iternal comparator will detect the event and turn off the MOSFET. The peak switch current is calculated
Io (peak) [A] = VSOCP [V]/Rsense [Ω]
The V_{CC} pin is pulled down to fixed level, keeping the controller lached off.The lach reset occurs when the user disconnects LED from VAC and lets the V_{CC} falls below the V_{CC} reset voltage, UVLOOFF (7.3V typ)(B). Then V_{CC} rise UVLOON (9V typ) (C), restart the switching.



3. OVP (Over voltage protection)

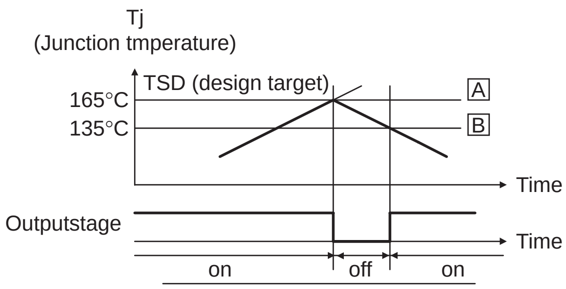
If the voltage of V_{IN} pin is higher than the internal reference voltage V_{IN}OVP (27V typ), switching operation is stopped.

The stopping operation is kept until the voltage of V_{IN} is lower than 7.3V. If the voltage of V_{IN} pin is higher than 9V, the switching operation is restated.



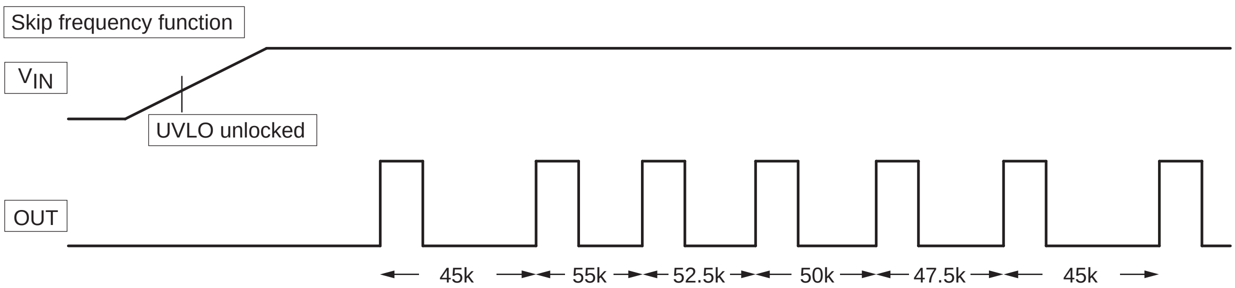
4. TSD (Thermal shut down protection)

The thermal shutdown function works when the junction temperature of IC is 165°C (typ) (A), and the IC switching stops. The IC starts switching operation again when the junction temperature is 135°C typ (B) or lower.



Skip frequency function

LV5028TT contains the skip frequency function for reduction of the peak value of conduction noise. This function changes the frequency as follows.

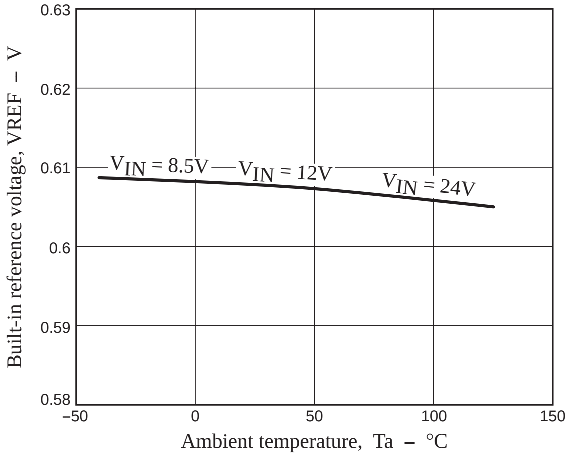


Switching frequency is changed as follows.

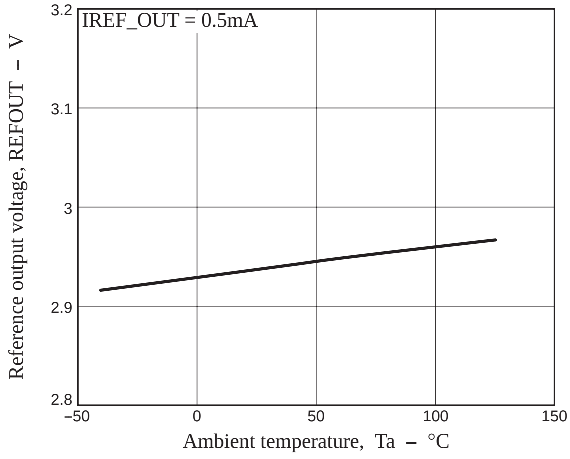
…×0.9 →	×1.1 →	×1.05 →	×1 →	×0.95 →	×0.9 →	×1.1 …
(45kHz)	(55kHz)	(52.5kHz)	(50kHz)	(47.5kHz)	(45kHz)	(55kHz)

It's repeated by this loop.

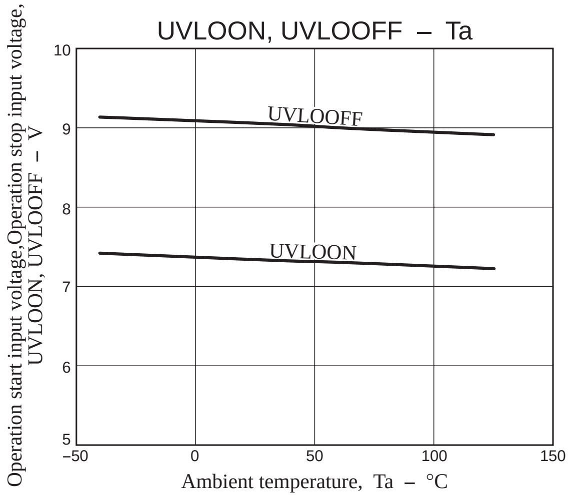
VREF – Ta



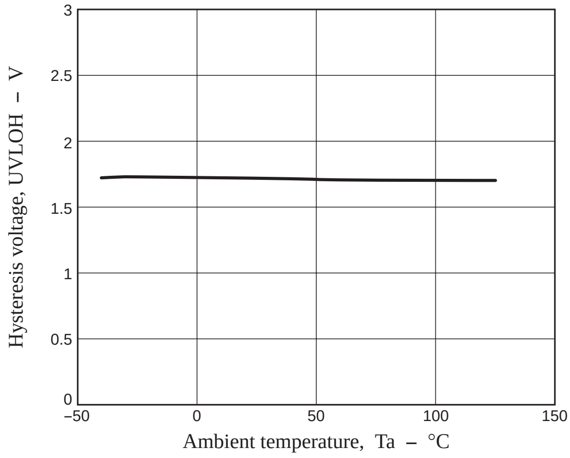
REFOUT – Ta



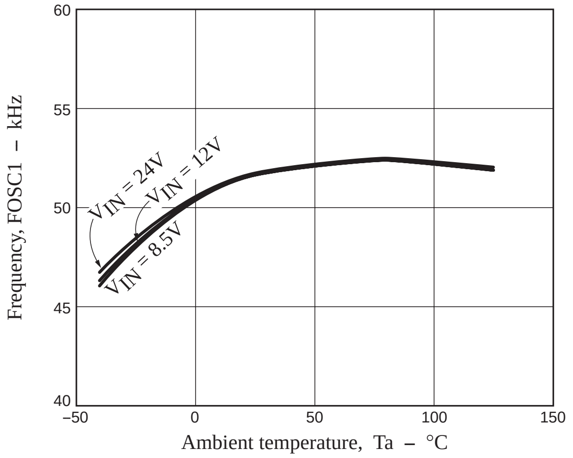
UVLOON, UVLOOFF – Ta



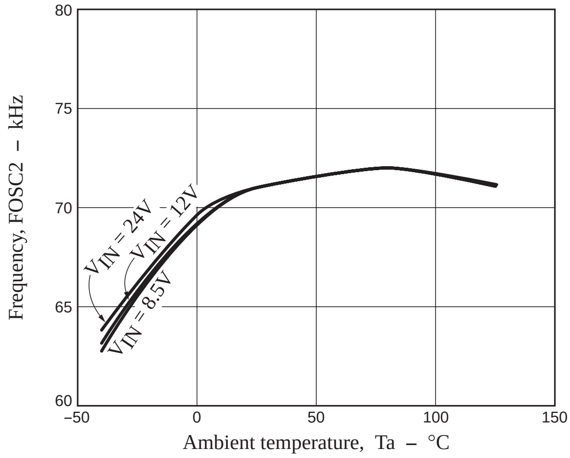
UVLOH – Ta



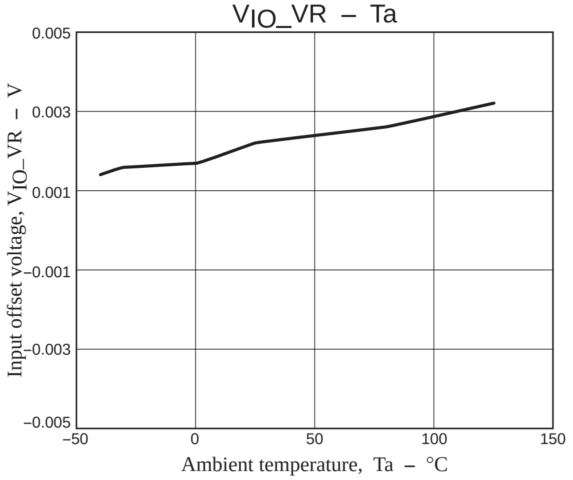
FOSC1 – Ta



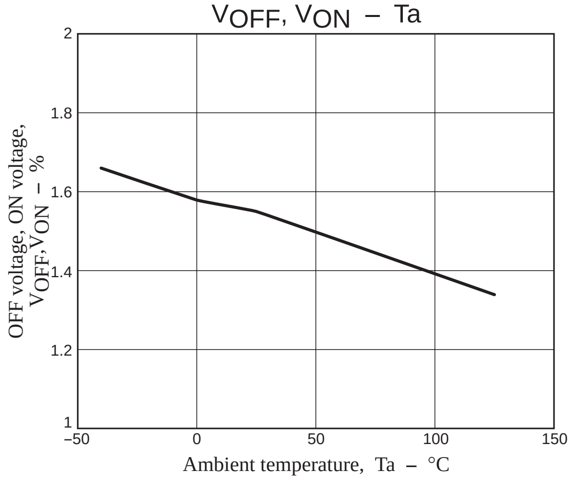
FOSC2 – Ta

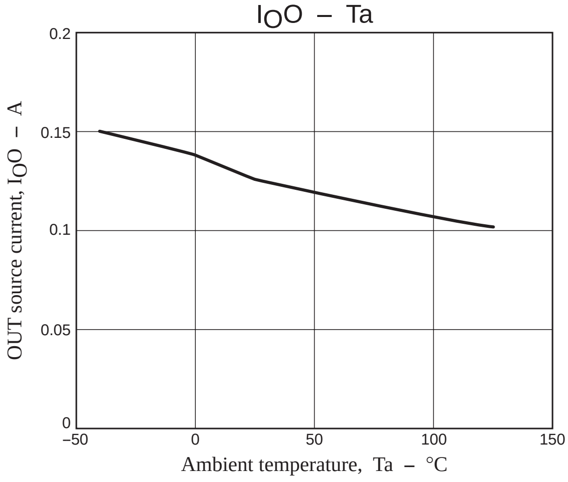


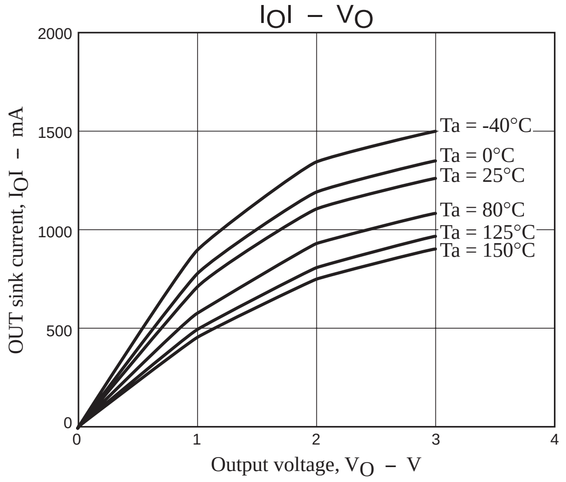
V_{IO}_VR – Ta

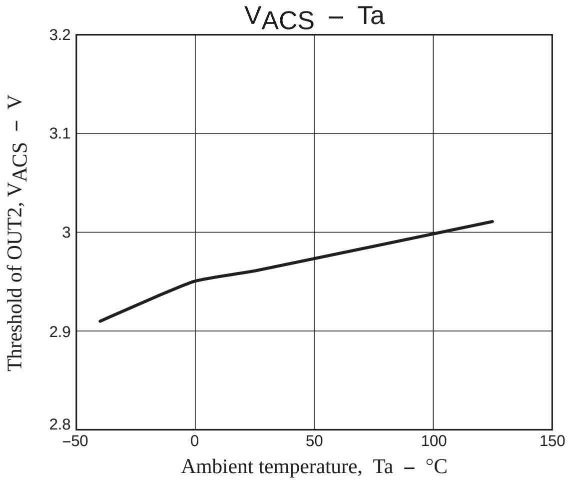


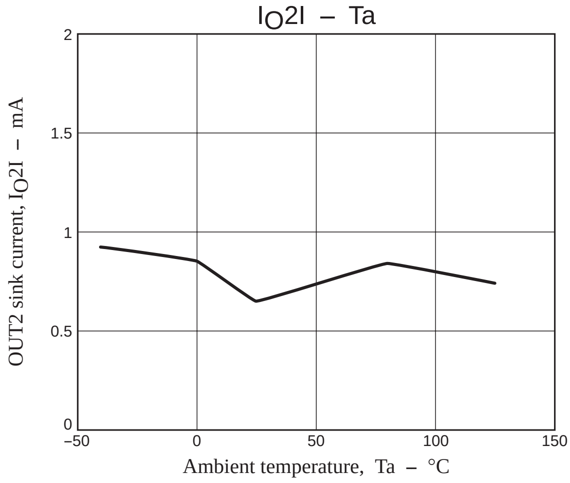
V_{OFF}, V_{ON} – Ta

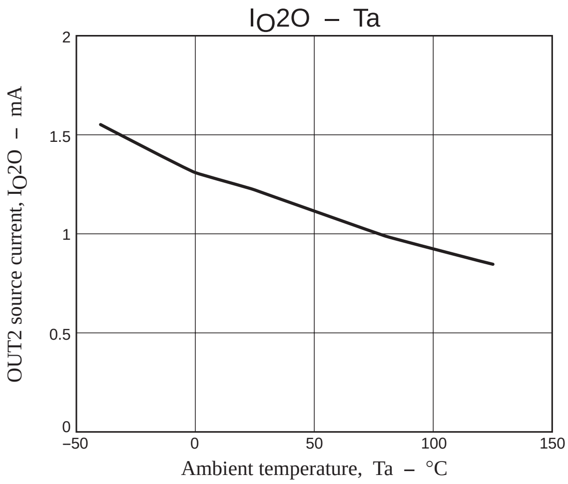


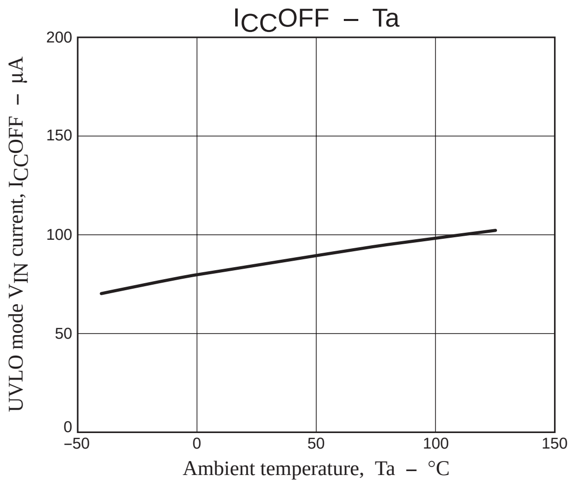

IOO – Ta
OUT source current, IOO – A
0.2
0.15
0.1
0.05
0
–50
0
50
100
150
Ambient temperature, Ta – °C

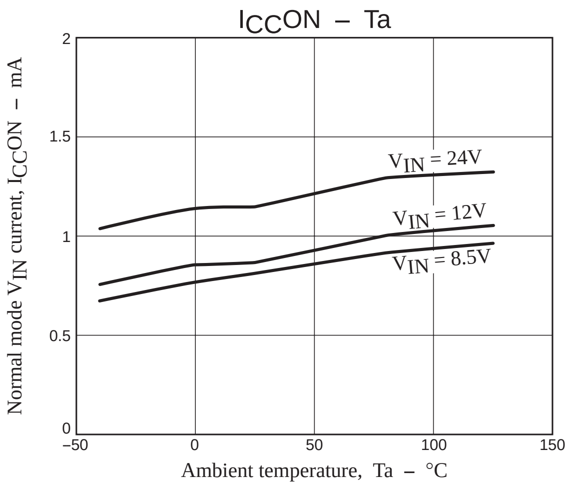

IOI – VO
OUT sink current, IOI – mA
2000
1500
1000
500
0
0
1
2
3
4
Output voltage, VO – V
Ta = -40°C
Ta = 0°C
Ta = 25°C
Ta = 80°C
Ta = 125°C
Ta = 150°C

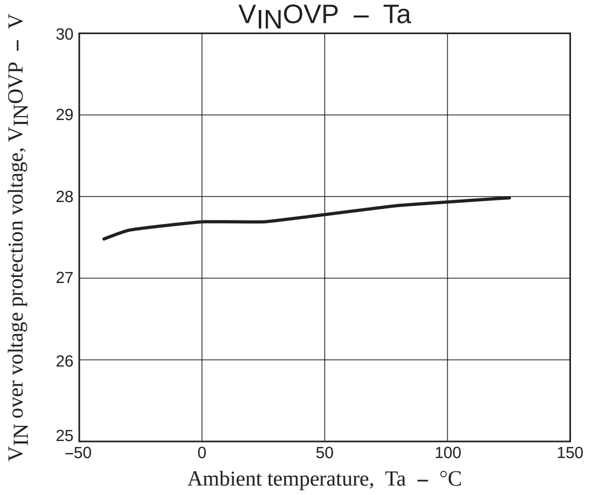

VACS – Ta
Threshold of OUT2, VACS – V
3.2
3.1
3
2.9
2.8
–50
0
50
100
150
Ambient temperature, Ta – °C

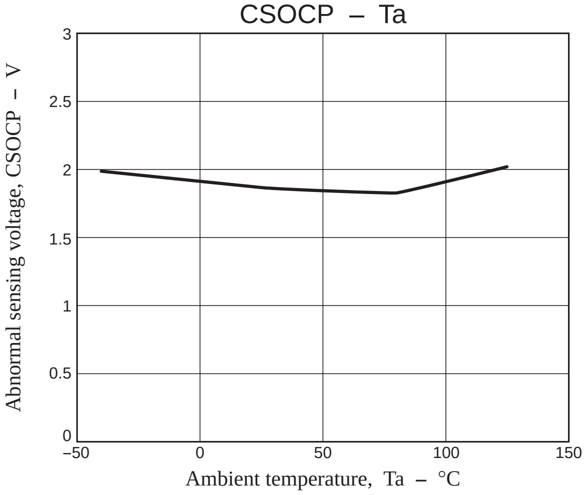

IO2I – Ta
OUT2 sink current, IO2I – mA
2
1.5
1
0.5
0
–50
0
50
100
150
Ambient temperature, Ta – °C


IO2O – Ta
OUT2 source current, IO2O – mA
2
1.5
1
0.5
0
–50
0
50
100
150
Ambient temperature, Ta – °C


ICCOFF – Ta
UVLO mode VIN current, ICCOFF – μA
200
150
100
50
0
–50
0
50
100
150
Ambient temperature, Ta – °C


ICCON – Ta
Normal mode VIN current, ICCON – mA
2
1.5
1
0.5
0
–50
0
50
100
150
Ambient temperature, Ta – °C
VIN = 24V
VIN = 12V
VIN = 8.5V





This catalog provides information as of December, 2011. Specifications and information herein are subject to change without notice.