

SANYO Semiconductors DATA SHEET

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PWM Constant-Current Control Stepping Motor Driver

Overview

The LV8772 is a stepping motor driver, which is capable of micro-step drive and supports 4W 1-2 phase excitation. It is stepping motors used in office equipment and amusement applications.

Features

- Low on resistance (upper side : 0.3Ω ; lower side : 0.25Ω ; total of upper and lower : 0.55Ω ; Ta = 25°C, IO = 2.5A)
- Excitation mode can be set to 2-phase, 1-2 phase, W1-2 phase , or 4W1-2 phase
- BiCDMOS process IC
- Motor current selectable in four steps
- Unusual condition warning output pins
- Excitation step proceeds only by step signal input
- Output short-circuit protection circuit incorporated
- No control power supply required

Specifications

Absolute Maximum Ratings at Ta = 25°C

Parameter	Symbol	Conditions	Ratings	Unit
Supply voltage	VM max		36	V
Output peak current	I _O peak	tw ≤ 10ms, duty 20%	3.0	А
Output current	I _O max		2.5	А
Logic input voltage	V _{IN}		-0.3 to +6	V
MONI/EMO input voltage	Vmoni/Vemo		-0.3 to +6	V
Allowable power dissipation	Pd max1	1 unit	3.0	W
	Pd max2	*	5.4	W
Operating temperature	Topr		-20 to +85	°C
Storage temperature	Tstg		-55 to +150	°C

* Specified circuit board : 90.0mm×90.0mm×1.6mm, glass epoxy 2-layer board, with backside mounting.

Caution 1) Absolute maximum ratings represent the value which cannot be exceeded for any length of time.

Caution 2) Even when the device is used within the range of absolute maximum ratings, as a result of continuous usage under high temperature, high current, high voltage, or drastic temperature change, the reliability of the IC may be degraded. Please contact us for the further details.

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Allowable Operating Ratings at $Ta = 25^{\circ}C$

Parameter	Symbol	Conditions	Ratings	Unit
Supply voltage range	VM		9 to 32	V
Logic input voltage	V _{IN}		0 to 5.5	V
VREF input voltage range	VREF		0 to 3	V

Electrical Characteristics at $Ta = 25^{\circ}C$, VM = 24V, VREF = 1.5V

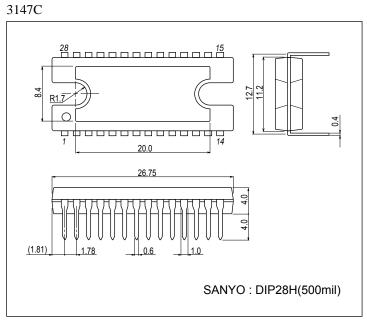
Dara	meter	Symbol	Conditions	ļ	Ratings		Unit
Para	melei	Symbol	Conditions	min	typ	max	Unit
Standby mode cu	irrent drain	IMst	ST = "L"		100	400	μA
Current drain		IM	ST = "H", with no load		3.2	5	mA
VREG5 output vo	ltage	Vreg5	I _O = -1mA	4.5	5	5.5	V
Thermal shutdow	n temperature	TSD	Design guarantee	150	180	200	°C
Thermal hysteres	is width	∆TSD	Design guarantee		40		°C
Motor driver							
Output on resista	nce	Ronu	I_{O} = 2.5A, Upper-side on resistance		0.3	0.4	Ω
		Rond	I_{O} = 2.5A, Lower-side on resistance		0.25	0.33	Ω
Output leakage c	urrent	l _O leak				50	μA
Diode forward vo	Itage	VD	ID = -2.5A		1.2	1.4	V
Logic pin input cu	ırrent	I _{IN} L	V _{IN} = 0.8V	4	8	12	μA
		I _{IN} H	V _{IN} = 5V	30	50	70	μA
Logic high-level in	nput voltage	V _{IN} H		2.0			V
Logic low-level in	put voltage	V _{IN} L				0.8	V
Current setting comparator	4W1-2-phase drive	Vtdac0_4W	Step 0 (When initialized : channel 1 comparator level)	0.291	0.3	0.309	V
threshold		Vtdac1_4W	Step 1 (Initial state+1)	0.291	0.3	0.309	V
voltage		Vtdac2_4W	Step 2 (Initial state+2)	0.285	0.294	0.303	V
(current step switching)		Vtdac3_4W	Step 3 (Initial state+3)	0.279	0.288	0.297	V
		Vtdac4_4W	Step 4 (Initial state+4)	0.267	0.276	0.285	V
		Vtdac5_4W	Step 5 (Initial state+5)	0.255	0.264	0.273	V
		Vtdac6_4W	Step 6 (Initial state+6)	0.240	0.249	0.258	V
		Vtdac7_4W	Step 7 (Initial state+7)	0.222	0.231	0.240	V
		Vtdac8_4W	Step 8 (Initial state+8)	0.201	0.210	0.219	V
		Vtdac9_4W	Step 9 (Initial state+9)	0.180	0.189	0.198	V
		Vtdac10_4W	Step 10 (Initial state+10)	0.157	0.165	0.173	V
		Vtdac11_4W	Step 11 (Initial state+11)	0.134	0.141	0.148	V
		Vtdac12_4W	Step 12 (Initial state+12)	0.107	0.114	0.121	V
		Vtdac13_4W	Step 13 (Initial state+13)	0.080	0.087	0.094	V
		Vtdac14_4W	Step 14 (Initial state+14)	0.053	0.060	0.067	V
		Vtdac15_4W	Step 15 (Initial state+15)	0.023	0.030	0.037	V
	W1-2-phase drive	Vtdac0_W	Step 0 (When initialized : channel 1 comparator level)	0.291	0.3	0.309	V
		Vtdac4_W	Step 4 (Initial state+1)	0.267	0.276	0.285	V
		Vtdac8_W	Step 8 (Initial state+2)	0.201	0.21	0.219	V
		Vtdac12_W	Step 12 (Initial state+3)	0.107	0.114	0.121	V
	1-2 phase drive	Vtdac0_H	Step 0 (When initialized : channel 1 comparator level)	0.291	0.3	0.309	V
		Vtdac8_H	Step 8 (Initial state+1)	0.201	0.21	0.219	V
	2 phase drive	Vtdac8_F	Step 8' (When initialized : channel 1 comparator level)	0.291	0.3	0.309	V
Current setting co	omparator	Vtatt00	ATT1 = L, ATT2 = L	0.291	0.3	0.309	V
threshold voltage		Vtatt01	ATT1 = H, ATT2 = L	0.232	0.24	0.248	V
(current attenuation	on rate switching)	Vtatt10	ATT1 = L, ATT2 = H	0.143	0.15	0.157	V
		Vtatt11	ATT1 = H, ATT2 = H	0.053	0.06	0.067	V
Chopping frequer	псу	Fchop	Cchop = 180pF	45	55	65	kHz
	/discharge current	Ichop		7	10	13	μA

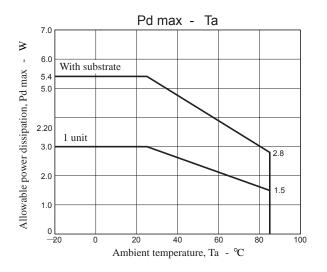
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Deventer	0 set et	mbal		Ratings		
Parameter	Symbol	Symbol Conditions		typ	max	Unit
Chopping oscillation circuit	Vtup		0.8	1	1.2	V
threshold voltage	Vtdown		0.4	0.5	0.6	V
VREF pin input current	Iref	VREF = 1.5V	-0.5			μA
MONI pin saturation voltage	Vsatmon	Imoni = 1mA			400	mV
Charge pump						
VG output voltage	VG		28	28.7	29.8	V
Rise time	tONG	VG = 0.1μ F, CP1-CP2 = 0.1μ F, ST = "H" \rightarrow VG = VM+4V		200	500	μS
Oscillator frequency	Fosc		90	125	150	kHz
Output short-circuit protection		•				
EMO pin saturation voltage	Vsatemo	lemo = 1mA			400	mV

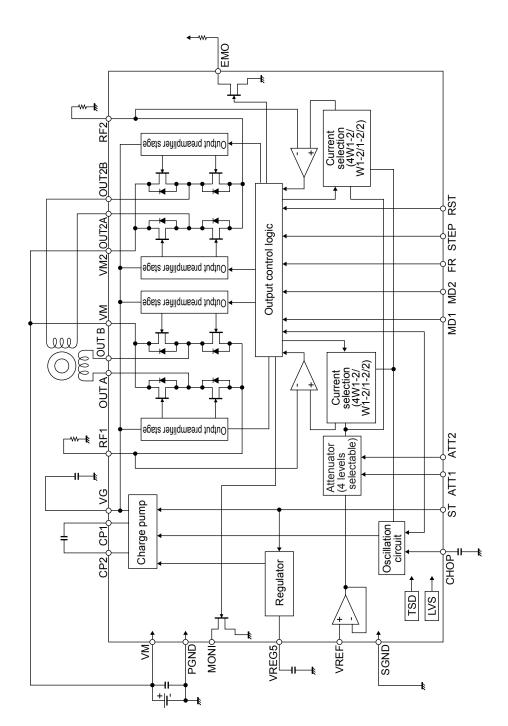
Package Dimensions

unit : mm (typ)

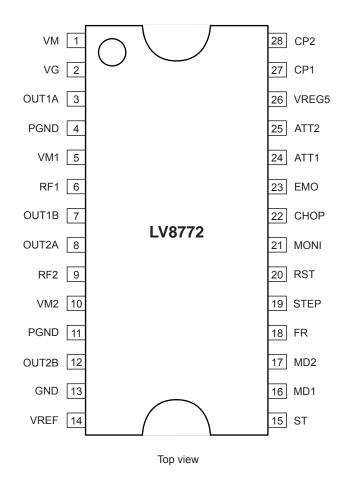




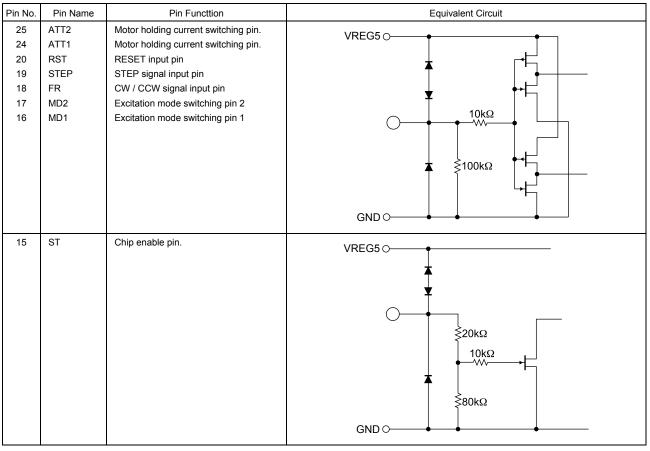
Block Diagram



Pin Assignment



Pin Functions

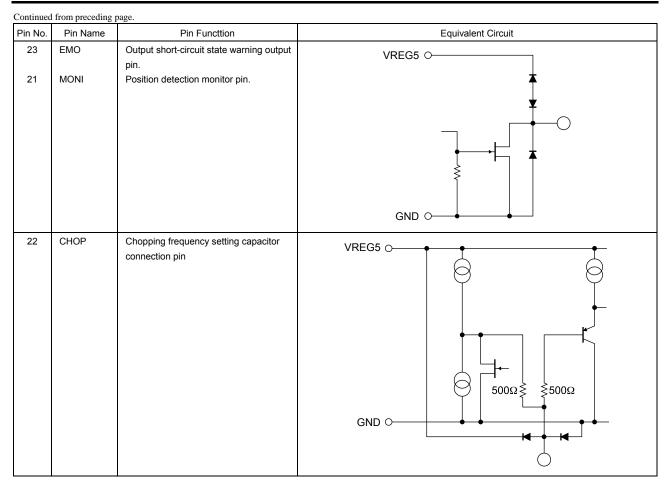


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Pin No.	Pin Name	Pin Functtion	Equivalent Circuit
12	OUT2B	Channel 2 OUTB output pin.	
4/11	PGND	Power system ground.	(5)
10	VM2	Channel 2 motor power supply	
		connection pin.	
9	RF2	Channel 2 current-sense resistor	
		connection pin.	│
8	OUT2A	Channel 2 OUTA output pin.	
7	OUT1B	Channel 1 OUTB output pin.	
6	RF1	Channel 1 current-sense resistor	
		connection pin.	
5	VM1	Channel 1 motor power supply pin.	│
3	OUT1A	Channel 1 OUTA output pin	
			9
			GND O
2	VG	Charge pump capacitor connection pin.	
2	VG VM	Charge pump capacitor connection pin. Motor power supply connection pin.	
28	CP2	Charge pump capacitor connection pin.	
27	CP1	Charge pump capacitor connection pin	
		5 - 5 - F - F - F F	
			₹100Ω
14	VREF	Constant current control reference	VREG5 0
		voltage input pin	
			★ ↓ ↓ ↓
			¥ '
			▲
26	VREG5	Internal power supply capacitor	VM 0
		connection pin	
			\forall \downarrow
			78kΩ
			ξ 26kΩ
			GND O

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Description of operation

Input Pin Function

(1) Chip enable function

This IC is switched between standby and operating mode by setting the ST pin. In standby mode, the IC is set to power-save mode and all logic is reset. In addition, the internal regulator circuit and charge pump circuit do not operate in standby mode.

ST	Mode	Internal regulator	Charge pump
Low or Open	Standby mode	Standby	Standby
High	Operating mode	Operating	Operating

Stepping mode drive method

(1) STEP pin function

Inj	out	Operating mode		
ST	STP			
Low	*	Standby mode		
High		Excitation step proceeds		
High		Excitation step is kept		

(2) Excitation mode setting function

MD1	MD2	Excitation mode	Initial position	
			Channel 1 Channel 2	
Low	Low	2 phase excitation	100%	-100%
High	Low	1-2 phase excitation	100%	0%
Low	High	W1-2 phase excitation	100%	0%
High	High	4W1-2 phase excitation	100% 0%	

This is the initial position of each excitation mode in the initial state after power-on and when the counter is reset.

(3) Position detection monitoring function

The MONI position detection monitoring pin is of an open drian type. When the excitation position is in the initial position, the MONI output is placed in the ON state. (Refer to "Examples of current waveforms in each of the excitation modes.")

(4) Setting constant-current control reference current

This IC is designed to automatically exercise PWM constant-current chopping control for the motor current by setting the output current. Based on the voltage input to the VREF pin and the resistance connected between RF and GND, the output current that is subject to the constant-current control is set using the calculation formula below :

$I_{OUT} = (VREF/5)/RF$ resistance

* The above setting is the output current at 100% of each excitation mode.

The voltage input to the VREF pin can be switched to four-step settings depending on the statuses of the two inputs, ATT1 and ATT2. This is effective for reducing power consumption when motor holding current is supplied.

Attenuation function for	VREF input voltage
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ATT1	ATT2	Current setting reference voltage attenuation ratio
Low	Low	100%
High	Low	80%
Low	High	50%
High	High	20%

The formula used to calculate the output current when using the function for attenuating the VREF input voltage is given below.

 $I_{OUT} = (VREF/5) \times (attenuation ratio)/RF$ resistance

Example : At VREF of 1.5V, a reference voltage setting of 100% [(ATT1, ATT2) = (L, L)] and an RF resistance of 0.22Ω , the output current is set as shown below.

- $I_{OUT} = 1.5V/5 \times 100\%/0.22\Omega = 1.36A$
- If, in this state, (ATT1, ATT2) is set to (H, H), IOUT will be as follows : $I_{OUT} = 1.36A \times 20\% = 272mA$

In this way, the output current is attenuated when the motor holding current is supplied so that power can be conserved.

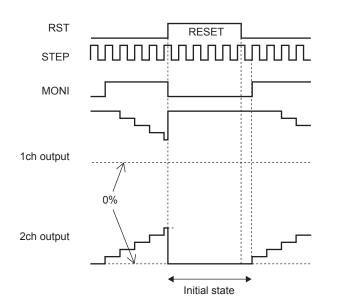
(5) Blanking period

If, when exercising PWM constant-current chopping control over the motor current, the mode is switched from decay to charge, the recovery current of the parasitic diode may flow to the current sensing resistance, causing noise to be carried on the current sensing resistance pin, and this may result in erroneous detection. To prevent this erroneous detection, a blanking period is provided to prevent the noise occurring during mode switching from being received. During this period, the mode is not switched from charge to decay even if noise is carried on the current sensing resistance pin.

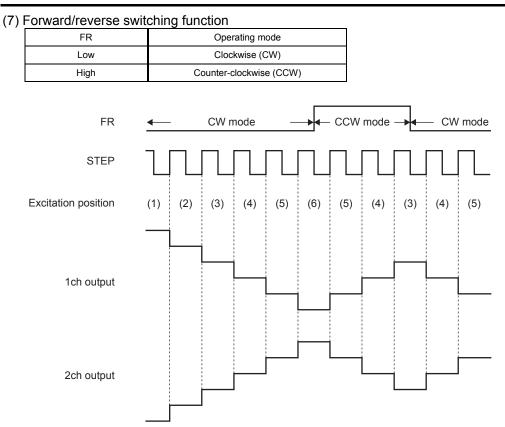
This IC is the blanking time is fixed at approximately 1µs.

(6) Reset function

RST	Operating mode
Low	Normal operation
High	Reset state



When the RST pin is set to High, the excitation position of the output is forcibly set to the initial state, and the MONI output is placed in the ON state. When RST is then set to Low, the excitation position is advanced by the next STEP input.



The internal D/A converter proceeds by one bit at the rising edge of the input STEP pulse. In addition, CW and CCW mode are switched by setting the FR pin. In CW mode, the channel 2 current phase is delayed by 90° relative to the channel 1 current. In CCW mode, the channel 2 current phase is advanced by 90° relative to the channel 1 current.

(8) Chopping frequency setting

For constant-current control, this IC performs chopping operations at the frequency determined by the capacitor (Cchop) connected between the CHOP pin and GND.

The chopping frequency is set as shown below by the capacitor (Cchop) connected between the CHOP pin and GND.

Fchop = Ichop/ (Cchop \times Vtchop \times 2) (Hz)

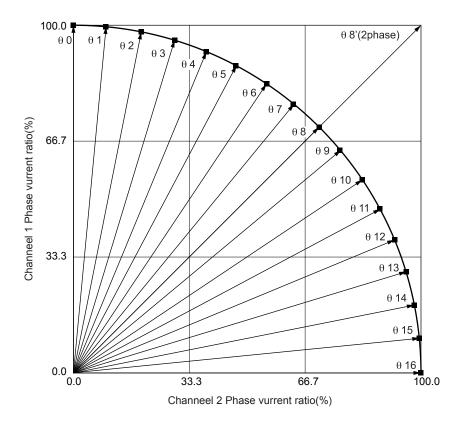
Ichop : Capacitor charge/discharge current, typ 10µA

Vtchop : Charge/discharge hysteresis voltage (Vtup-Vtdown), typ 0.5V

For instance, when Cchop is 180pF, the chopping frequency will be as follows :

 $Fchop = 10\mu A / (180pF \times 0.5V \times 2) = 55 kHz$

(9) Output current vector locus (one step is normalized to 90 degrees)

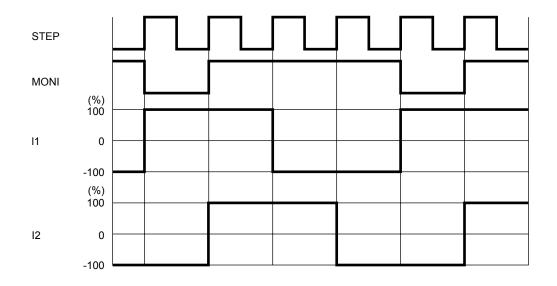


Setting current ration in each excitation mode

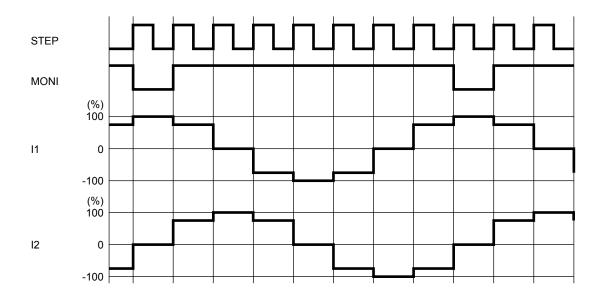
STEP	4W1-2 pl	hase (%)	W1-2 ph	nase (%)	1-2 pha	ise (%)	2-phas	e (%)
	Channel 1	Channel 2						
θ 0	100	0	100	0	100	0		
θ1	100	10						
θ2	98	20						
03	96	29						
04	92	38	92	38				
θ5	88	47						
06	83	55						
θ7	77	63						
08	70	70	70	70	70	70	100	100
θ9	63	77						
θ 10	55	83						
θ11	47	88						
θ 12	38	92	38	92				
θ 13	29	96						
014	20	98						
θ15	10	100						
θ16	0	100	0	100	0	100		



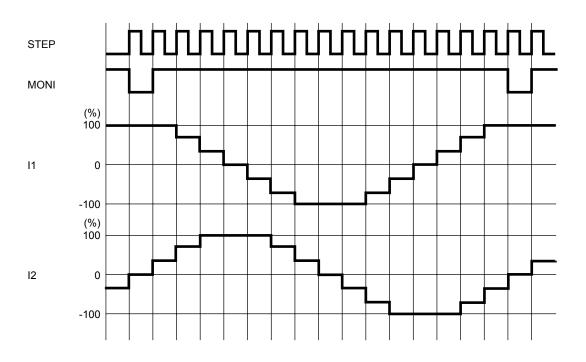
2-phase excitation (CW mode)



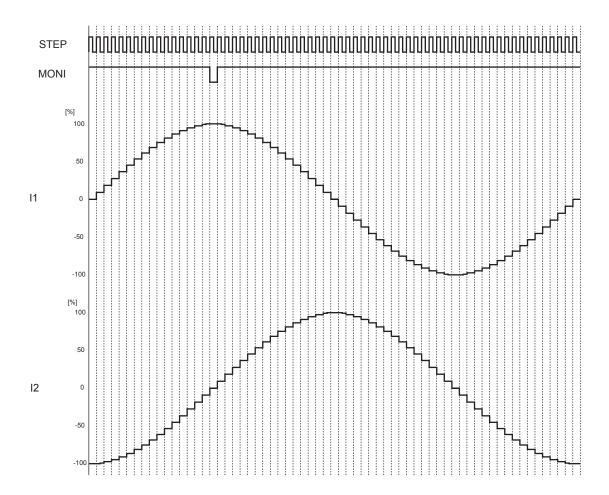
1-2 phase excitation (CW mode)



W1-2 phase excitation (CW mode)

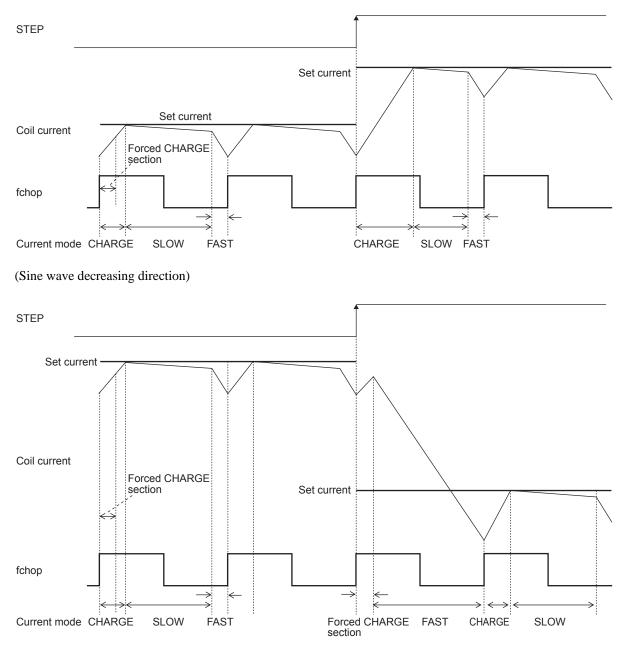


4W1-2 phase excitation (CW mode)



(11) Current control operation specification

(Sine wave increasing direction)



In each current mode, the operation sequence is as described below :

- At rise of chopping frequency, the CHARGE mode begins. (In the time defined as the "blanking time," the CHARGE mode is forced regardless of the magnitude of the coil current (ICOIL) and set current (IREF).)
- The coil current (ICOIL) and set current (IREF) are compared in this blanking time.
 - When (ICOIL < IREF) state exists ;

The CHARGE mode up to ICOIL \geq IREF, then followed by changeover to the SLOW DECAY mode, and finally by the FAST DECAY mode for approximately 1µs.

When (ICOIL < IREF) state does not exist;

The FAST DECAY mode begins. The coil current is attenuated in the FAST DECAY mode till one cycle of chopping is over.

Above operations are repeated. Normally, the SLOW (+FAST) DECAY mode continues in the sine wave increasing direction, then entering the FAST DECAY mode till the current is attenuated to the set level and followed by the SLOW DECAY mode.

Output short-circuit protection function

This IC incorporates an output short-circuit protection circuit that, when the output has been shorted by an event such as shorting to power or shorting to ground, sets the output to the standby mode and turns on the warning output in order to prevent the IC from being damaged.

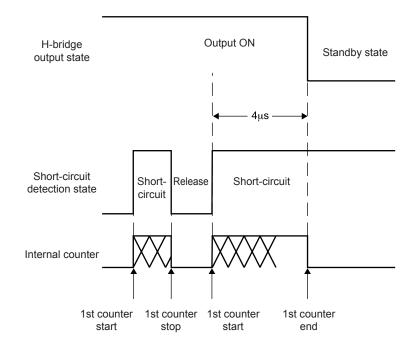
This function sets the output to the standby mode for both channels by detecting the short-circuiting in one of the channels.

(1) Output short-circuit protection method

The output short-circuit protection method of LV8772 is a latch method to turn off the output when the output current exceeds the detection current, and to maintain the state.

The detection of the output short-circuited state by the IC causes the output short-circuit protection circuit to be activated.

All the outputs of correspondence ch side where the short-circuit was first detected are switched to the standby mode when the short-circuit is the consecutive between internal timers (approximately 4μ s), and the state is maintained. This state is released by setting ST to low.



(2) Unusual condition warning output pins (EMO)

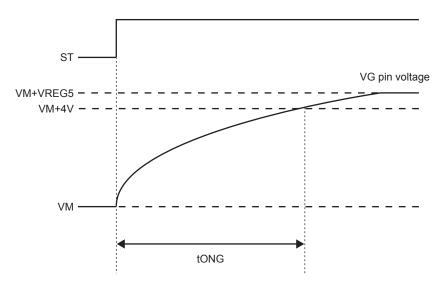
The LV8772 is provided with the EMO pin which notifies the CPU of an unusual condition if the protection circuit operates by detecting an unusual condition of the IC. This pin is of the open-drain output type and when an unusual condition is detected, the EMO output is placed in the ON (EMO = Low) state.

Furthermore, the EMO pin is placed in the ON state when one of the following conditions occurs.

- 1. Shorting-to-power, shorting-to-ground, or shorting-to-load occurs at the output pin and the output short-circuit protection circuit is activated.
- 2. The IC junction temperature rises and the thermal protection circuit is activated.

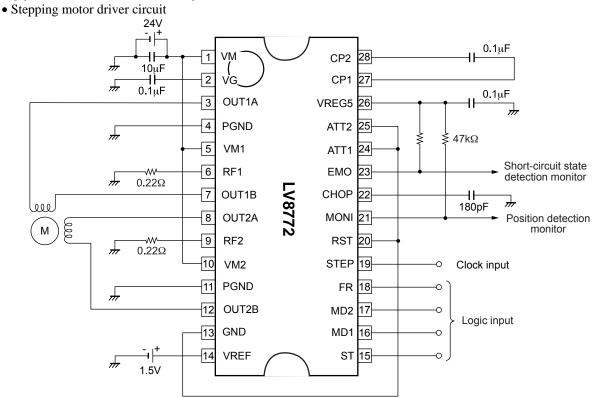
Charge Pump Circuit

When the ST pin is set High, the charge pump circuit operates and the VG pin voltage is boosted from the VM voltage to the VM + VREG5 voltage. If the VG pin voltage is not boosted sufficiently, the output cannot be controlled, so be sure to provide a wait time of tONG or more after setting the ST pin High before starting to drive the motor.





Application Circuit Example



The formulae for setting the constants in the examples of the application circuits above are as follows :

Constant current (100%) setting

When VREF = 1.5V

 $I_{OUT} = VREF/5/RF$ resistance

 $= 1.5 V/5/0.22 \Omega = 1.36 A$

Chopping frequency setting

Fchop = Ichop/ (Cchop × Vtchop × 2) = 10μ A/ (180pF × 0.5V × 2) = 55kHz

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