

Rev. 1.7

2.5 W/CH Stereo Class D Audio power Amplifier (WCSP/ TSSOP)

FEATURES

- 2.5 W/CH Into 4Ω from 5V power supply at THD = 10% (Typ.).
- 2.5V~5.5V Power supply.
- Low shutdown Current.
- Low Quiescent Current.
- Minimum external components.
- No output filter required for inductive loads.
- Output Pin Short-Circuit Protection (Short to Output Pin, Short to GND, Short to VCC)
- Low noise during turn-on and turn-off transitions.
- Lead free and green package available. (RoHS Compliant)
- Wafer Chip Scale (WCSP) and TSSOP 20 pin packaging available.

GENERAL DESCRIPTION

The LY8210 is a high efficiency, 2.5 W stereo class D audio power amplifier. It is a low noise, filterless PWM architecture eliminates the output filter, reducing external component count, system cost, and simplify design.

The LY8210 is designed to meet of Multimedia application includes mobile phones and other portable electronic devices. The LY8210 is a single 5V supply, it is capable of driving 4Ω speaker load at a continuous average output of 2.5 W/CH with 10% THD+N.

In cellular handsets, the earpiece, speaker phone, and melody ringer can each be driven by the LY8210. Output pin short circuit (short to output pin, short to ground and short to VDD) protection prevent the device from damage during fault conditions.

APPLICATION

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- Portable electronic devices
- Mobile Phones, PDAs
- DVD/CD Players, TFT LCD TVs/Monitors
- USB Audio, 2.1 / 5.1 CH Audio System

BLOCK DIAGRAM

V_{DD}(L) AV_{DD} (L) 150KΩ -IN (L) $V_{O+}(L)$ PWM H-Differential Bridge Input Modulator + IN (L) $V_{O-}(L)$ 150KΩ Internal Oscillator 150KΩ -IN (R) $V_{O+}(R)$ PWM H_ Differential Bridge Modulator Input + IN (R) V_{O_}(R) 150KΩ SHUTDOWN (L) Bias VSS Circuit SHUTDOWN (R) $AV_{DD}(R)$ $V_{DD}(R)$

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PIN CONFIGURATION

LY8210 WCSP 18 Ball Pin Configuration (Top View)

_	1	2	3	4	5	6
A	VOPL	SDL	INNL	INPR	VSS	VONR
В	VSS	VDD	AVDDL	AVDDR	VDD	VSS
С	VONL	VSS	INPL	INNR	SDR	VOPR

LY8210 TSSOP20 Pin Configuration (Top View)

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			1
VONL	1	20	VSS
NC	2	19	VOPL
VDDL	3	18	VSS
SDL	4	17	INNL
INPL	5	16	AVDDR
AVDDL	6	15	INNR
INPR	7	14	SDR
VSS	8	13	VDDR
NC	9	12	VOPR
VONR	10	11	VSS
			J

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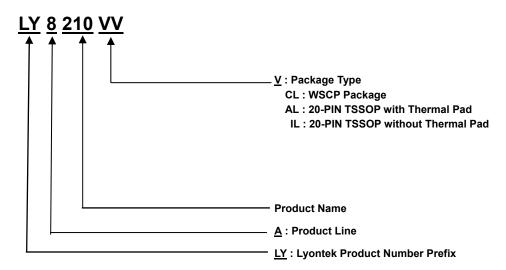
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PIN DESCRIPTION

SYMBOL	Pin No.		DESCRIPTION
31WIDOL	WCSP	TSSOP	DESCRIPTION
SDL	A2	4	Shutdown pin of left channel. (when low level is active the pin).
INPL(+)	C3	5	Positive(+) input of left channel.
INNL(-)	A3	17	Negative(-) input of left channel.
VOPL(+)	A1	19	Positive(+) BTL output of left channel.
VONL(-)	C1	1	Negative(-) BTL output of left channel.
AV _{DD}	B3/B4	6/16	Analog Power supply of left and right channel.
Vdd	B2/B5	3/13	Power supply of left and right channel.
Vss	C2/A5 B1/B6	8/11/18/20	Ground
SDR	C5	14	Shutdown pin of right channel. (when low level is active the pin).
INPR(+)	A4	7	Positive(+) input of right channel.
INNR(-)	C4	15	Negative(-) input of right channel.
VOPR(+)	C6	12	Positive(+) BTL output of right channel.
VONR(-)	A6	10	Negative(-) BTL output of right channel.
NC		2/9	No Connection.

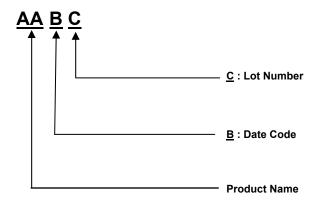
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PACKAGE MARKING FORMAT (For WCSP Package only)



TYPICAL APPLICATION CIRCUIT

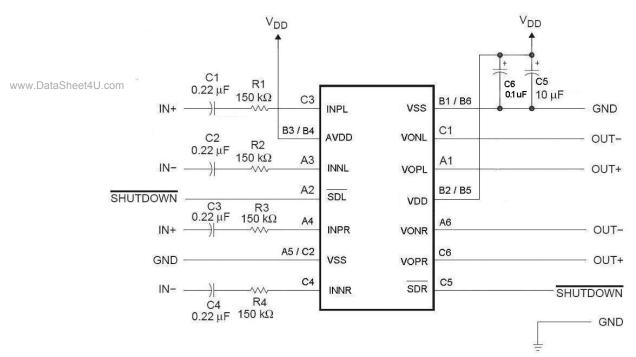


Figure 1. Typical Application Schematic With Differential Input and AVDD Connect to VDD Configuration (WCSP package)

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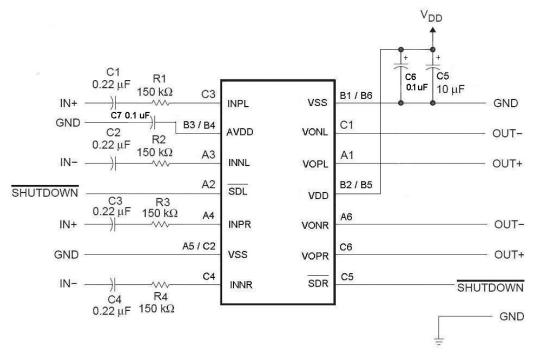


Figure 2. Typical Application Schematic With Differential Input and AVDD Increase Capacitors Configuration(WCSP package).

Note: In case of AVDD and VDD, the use of AVDD Increase Capacitors will optimize the THD+N.

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ABSOLUTE MAXIMUN RATINGS*

PARAMETER	SYMBOL	RATING	UNIT
Supply Voltage	V _{DD}	6.0	V
Operating Temperature	TA	-40 to 85 (I grade)	$^{\circ}\!\mathbb{C}$
Input Voltage	Vı	-0.3V to V _{DD} +0.3V	V
Storage Temperature	Тѕтс	-65 to 150	$^{\circ}\!\mathbb{C}$
Power Dissipation	PD	Internally Limited	W
ESD Susceptibility	Vesd	2000	V
Junction Temperature	Тумах	150	$^{\circ}\!\mathbb{C}$
Soldering Temperature (under 10 sec)	Tsolder	260	$^{\circ}\!\mathbb{C}$

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ELECTRICAL CHARACTERISTICS (TA = 25°C)

PARAMETER	SYMBOL	TEST CONDITION	MIN.	TYP.	MAX.	UNIT
Output offset voltage (measured differentially)		V _I = 0 V, A _V = 2 V/V, V _{DD} = 2.5 V to 5.5 V	-	-	50	mV
High-level input current	liH	V _{DD} = 5.5 V, V _I = 5.8 V	-	-	200	uA
Low-level input current	lıL	V _{DD} = 5.5 V, V _I = 0.3 V	-	-	10	uA
		V _{DD} = 5.5V, No Load	-	6.8	9.0	
Quiescent Current	lq	V _{DD} = 3.6V, No Load	-	5.6	-	mA
		V _{DD} = 2.5V, No Load	-	4.4	6.4	
Shutdown Current	l len	Vshutdow _N = 0.8V, Vdd = 2.5V to 5.5V	-	0.6	4	μA

■ OPERATING CHARACTERISTICS (1)(TA = 25°C, Gain = 2 V/V)

	PARAMETER	SYMBOL	TEST CONDITION	1	MIN.	TYP.	MAX.	UNIT
Ī			THD+N= 10%, f = 1 kHz,	VDD=5.0V	-	1.5	-	
			$R_L = 8\Omega$	V _{DD} =3.6V	-	0.7	-	
			112 032	VDD=2.5V	-	0.35	-	
			THD+N= 1%, f = 1 kHz,	VDD=5.0V	-	1.2	-	
			$R_L = 8\Omega$	V _{DD} =3.6V	-	0.6	-	
	Out Power / Channel	Po	1	VDD=2.5V	-	0.3	-	w
	Out i owei / Chaimei	10	THD+N= 10%, f = 1 kHz,	VDD=5.0V	-	2.5	-	V V
			$RL=4\Omega$	VDD=3.6V	-	1.2	-	
			122	VDD=2.5V	-	0.5	-	
			THD+N= 1%, f = 1 kHz,	VDD=5.0V	-	2.0	-	
			$R_L = 4\Omega$	VDD=3.6V	-	0.9	-	
			112	VDD=2.5V	-	0.4	-	
www	w.DataSheet4U.com		Po = 1.0 W, f = 1 kHz, R _L = 8 Ω	VDD=5.0V -	-	0.03	-	- %
		THD+N	Po = 2.0 W, f = 1 kHz, RL = 4 Ω	VDD-5.0V	-	0.06	-	
	Total harmonic distortion +		Po = 0.5 W, f = 1 kHz, R _L = 8 Ω	VDD=3.6V -	-	0.03	-	
	noise		Po = 0.8 W, f = 1 kHz, R _L = 4 Ω		-	0.04	-	
			$P_0 = 0.2 \text{ W, f} = 1 \text{ kHz,}$ R _L = 8 Ω		-	0.04	-	
			Po = 0.2 W, f = 1 kHz, RL = 4Ω	V _{DD} =2.5V -	-	0.04	-	
l	Power supply rejection ratio	PSRR	V _{DD} = 2.5 V to 5.5 V		-	-75	-	dB
l	Common mode rejection ratio	CMRR	Vic = 1 Vpp , f = 217 Hz	VDD=3.6V	-	-63	-	dB
ĺ	Signal-to-noise ratio	SNR	PO= 1.0 W, RL = 8 Ω	VDD=5.0V	-	91	-	dB
	Output voltage noise	Vn	V _{DD} = 3.6 V, f = 20 Hz to 20 kHz,	No weighting	-	50	-	uV _{RMS}
	Calpat Voltage Holse	VII	Inputs ac-grounded with Ci = 2 μF	A weighting	-	36	-	G V RIVIS
	Closed-loop Gain	Gv	V _{DD} = 2.5V to 5.5V		-	150kΩ/Ri	-	V/V
	Start-up time from shutdown	Zı	$V_{DD} = 2.5V \text{ to } 5.5V$		-	1	-	ms

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TYPICAL PERFORMANCE CHARACTERISTICS

Figure 3
Total Harmonic Distortion + Noise vs Output Power (RL = 4Ω , Stereo)

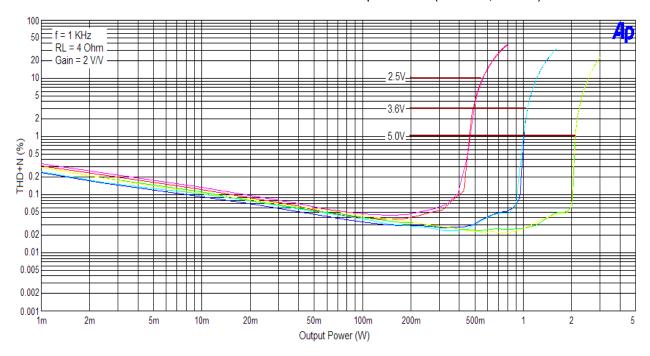
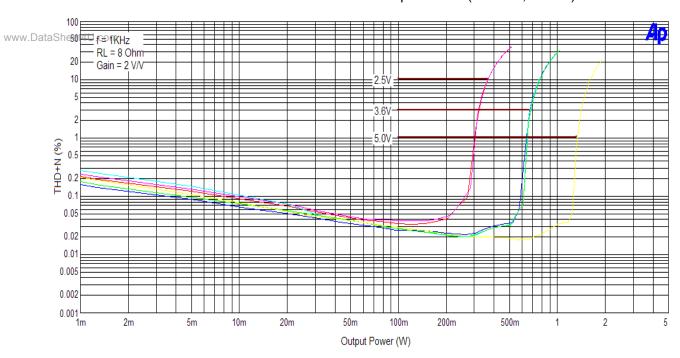


Figure 4
Total Harmonic Distortion + Noise vs Output Power ($RL = 8\Omega$, Stereo)



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Figure 5
Total Harmonic Distortion + Noise vs Frequency (VDD=5.0V, 4Ω)

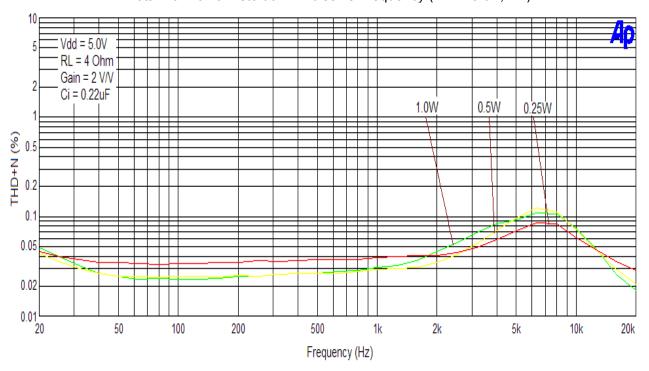
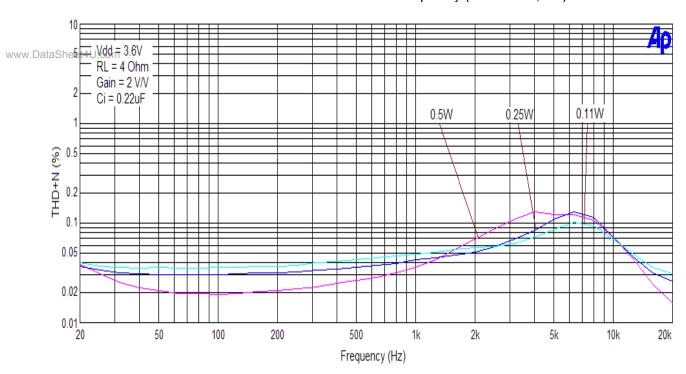


Figure 6 Total Harmonic Distortion + Noise vs Frequency (VDD=3.6V, 4Ω)



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Figure 7
Total Harmonic Distortion + Noise vs Frequency (VDD=2.5V, 4Ω)

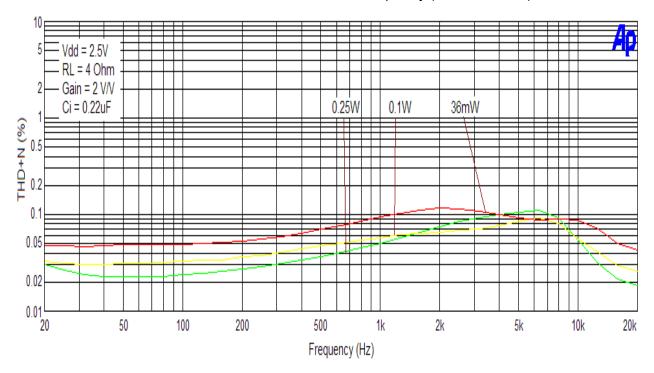
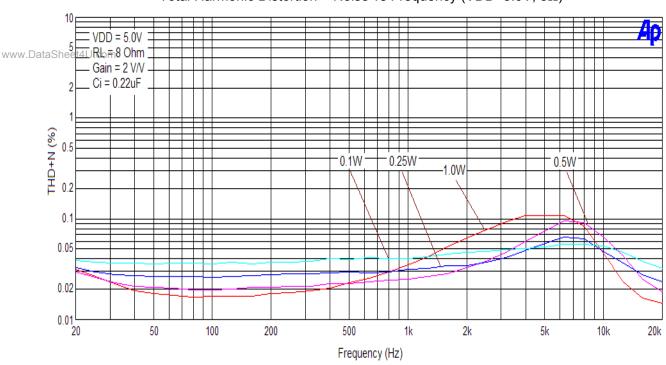


Figure 8
Total Harmonic Distortion + Noise vs Frequency (VDD=5.0V, 8Ω)



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Figure 9
Total Harmonic Distortion + Noise vs Frequency (VDD=3.6V, 8Ω)

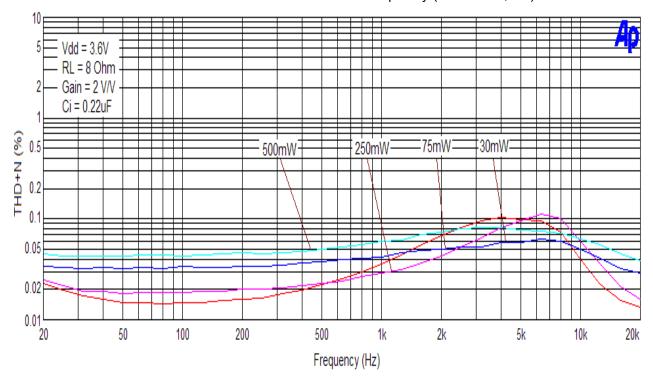
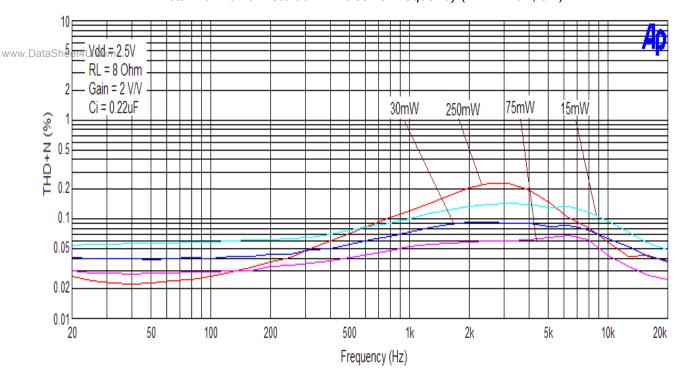


Figure 10

Total Harmonic Distortion + Noise vs Frequency (VDD=2.5V, 8Ω)



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Figure 11 Efficiency vs Output power (RL= 4Ω , 33μ H)

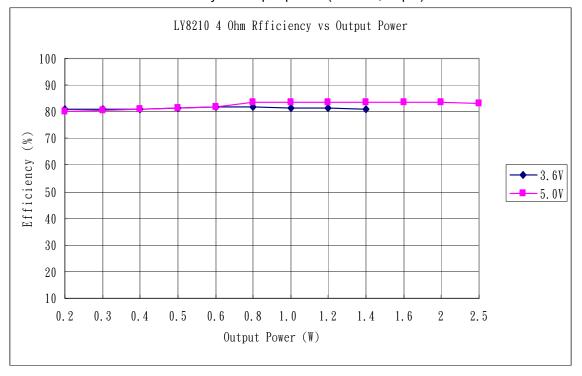
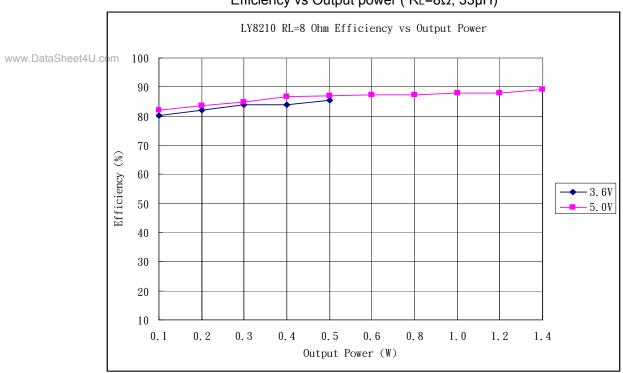


Figure 12 Efficiency vs Output power (RL=8 Ω , 33 μ H)



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APPLICATION INFORMATION

Fully Differential Amplifier

The LY8210 is a fully differential amplifier with differential inputs and outputs. The fully differential amplifier consists of a differential amplifier and a common-mode amplifier. The differential amplifier ensures that the amplifier outputs a differential voltage on the output that is equal to the differential input times the gain. The common-mode feedback ensures that the common-mode voltage at the output is biased around VDD/2 regardless of the common-mode voltage at the input. The fully differential LY8210 can still be used with a single-ended input; however, the LY8210 should be used with differential inputs when in a noisy environment, like a wireless handset, to ensure maximum noise rejection.

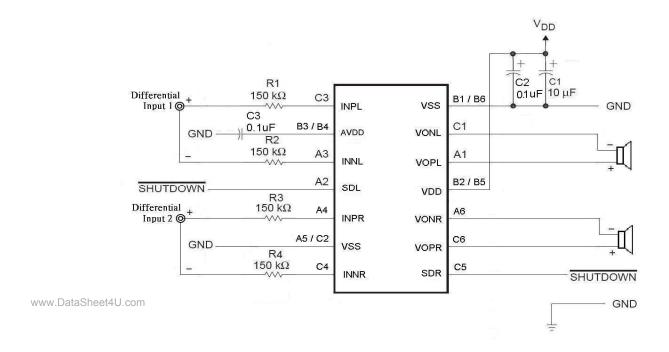


Figure 13. Application Schematic With Differential Input Configuration(WCSP Package)

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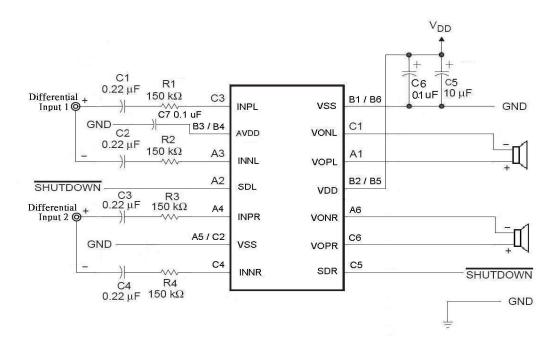


Figure 14. Application Schematic With Differential and Capacitors Input Configuration(WCSP Package)

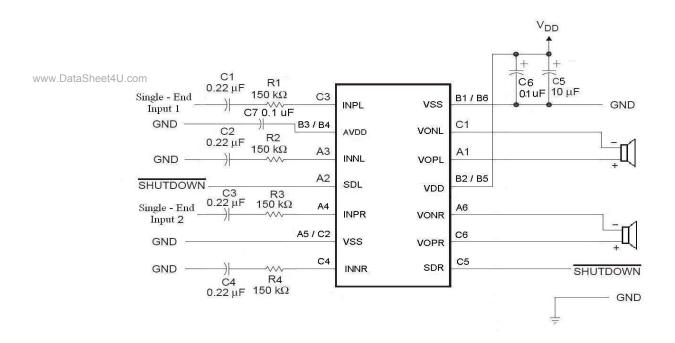


Figure 15. Application Schematic With Single-Ended Input Configuration(WCSP Package)

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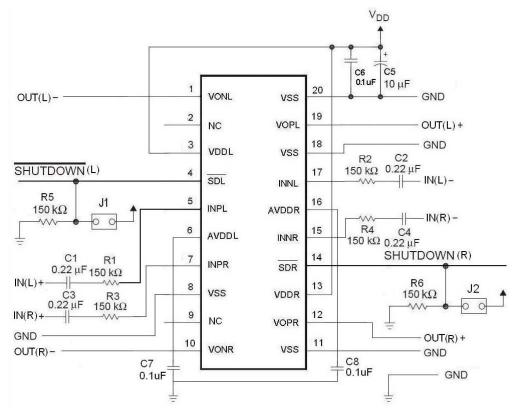


Figure 16. Application Schematic With Differential and Capacitors Input Configuration(TSSOP Package)

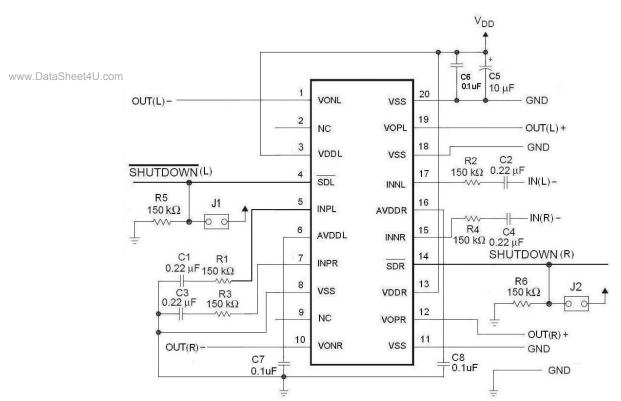


Figure 17. Application Schematic With Single-Ended Input Configuration(TSSOP Package)

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Input Resistors (Ri) and Gain

The LY8210 has two internal amplifier stages. The pre-amplifier gain is externally configurable, while the total gain is internally fixed. The closed-loop gain of the pre-amplifier gain is set by selecting the Rf to Ri while the total gain is fixed at 2x. So the input resistors (Ri) set the gain of the amplifier according to the equation.

Pre-Amplifier Gain = Rf / Ri

Total Gain = $(Rf/Ri) \times 2$

 $A_{VD} = 20 \times \log [2 \times (Rf/Ri)]$

The resistor matching is very important in the amplifiers. Balance of the output on the reference voltage depends on matched ratio of the resistors. CMRR, PSRR, and cancellation of the second harmonic distortion if resistor mismatch occurs. Therefore, it is recommended to use 1% tolerance resistors or better to keep the performance optimized. Matching is more important than overall tolerance.

Resistor arrays with 1% matching can be used with a tolerance greater than 1%. Place the input resistors very close to the LY8210 to limit noise injection on the high-impedance nodes. For optimal performance the gain should be set to 2 V/V or lower. Lower gain allows the LY8210 to operate at its best,

For example

Table 1. Typical Total Gain and AvD Values

Rf (KΩ)	150	150	150	150	150	150
Ri (KΩ)	150	75	50	37.5	25	18.75
Pre AMP. Gain	1	2	3	4	6	8
Total Gain	2	4	6	8	12	16
Avd (db)	6.02	12.04	15.56	18.06	21.58	24.08

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Input Capacitors (Ci)

The LY8210 using fully differential source, So the input coupling capacitors are required. The input capacitors and input resistors form a high-pass filter with the corner frequency(fc), determined in the equation.

 $fc = 1 / (2\pi Ri Ci)$

The value of the input capacitor is important to consider as it directly affects the bass (low frequency) performance of the circuit. Speakers in wireless phones cannot usually respond well to low frequencies, so the corner frequency can be set to block low frequencies in this application. Equation is reconfigured to solve for the input coupling capacitance.

 $Ci = 1 / (2\pi Ri fc)$

If the corner frequency is within the audio band, the capacitors should have a tolerance of $\pm 10\%$ or better, because any mismatch in capacitance causes an impedance mismatch at the corner frequency and below.

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For example

In the table 2 shows the external components. Rin in connect with Cin to create a high-pass filter.

Table 2. Typical Component Values

	71	
Reference	Description	Note
Ri	150ΚΩ	1% tolerance resistors
Ci	0.22uF	80%/–20%

 $Ci = 1 / (2\pi Ri fc)$

Ci = 1 / ($2\pi \times 150$ K $\Omega \times 4.8$ Hz)=0.221uF , Use 0.22uF

Decoupling Capacitor (Cs)

The LY8210 is a high-performance class-D audio amplifier that requires adequate power supply decoupling to ensure the efficiency is high and total harmonic distortion (THD) is low. For higher frequency transients, spikes, or digital hash on the line, a good low equivalent-series-resistance (ESR) ceramic capacitor, typically 1.0uF, placed as close as possible to the device VDD lead works best. Placing this decoupling capacitor close to the LY8210 is very important for the efficiency of the class-D amplifier, because any resistance or inductance in the trace between the device and the capacitor can cause a loss in efficiency. For filtering lower-frequency noise signals, a 10.0uF or greater capacitor placed near the audio power amplifier would also help, but it is not required in most applications because of the high PSRR of this device.

PCB Layout

www.Dallthe external components must place very close to the LY8210. The input resistors need to be very close to the LY8210 input pins so noise does not couple on the high impedance nodes between the input resistors and the input amplifier of the LY8210. Then place the decoupling capacitor Cs, close to the LY8210 is important for the efficiency of the class-D amplifier. Any resistance or inductance in the trace between the device and the capacitor can cause a loss in efficiency.

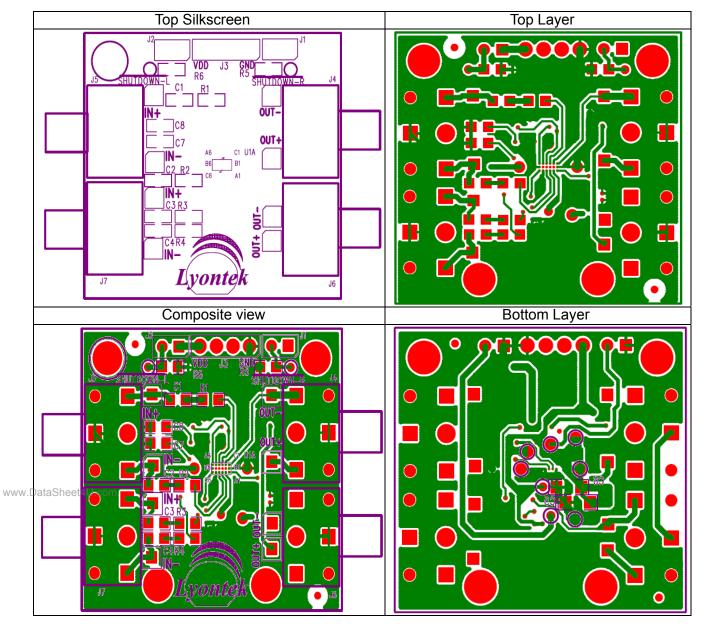
Making the high current traces going to VDD, GND, Vo+ and Vo- pins of the LY8210 should be as wide as possible to minimize trace resistance. If these traces are too thin, the LY8210's performance and output power will decrease. The input traces do not need to be wide, but do need to run side-by-side to enable common-mode noise cancellation.



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LY8210CL Demo Board Artwork



LY8210 V1.1 BOM List

No.	Description	Reference	Note
1	Resistor, 150K Ω	R1,R2,R3,R4,R5,R6	
2	Capacitor, 0.1uF	C6,C7,C8	
3	Capacitor, 0.22uF	C1,C2,C3,C4	
4	Capacitor, 10.0uF	C5	
5	LY8210CL	U1	
6	1*2 Pin Header	J1,J2	Close → Active

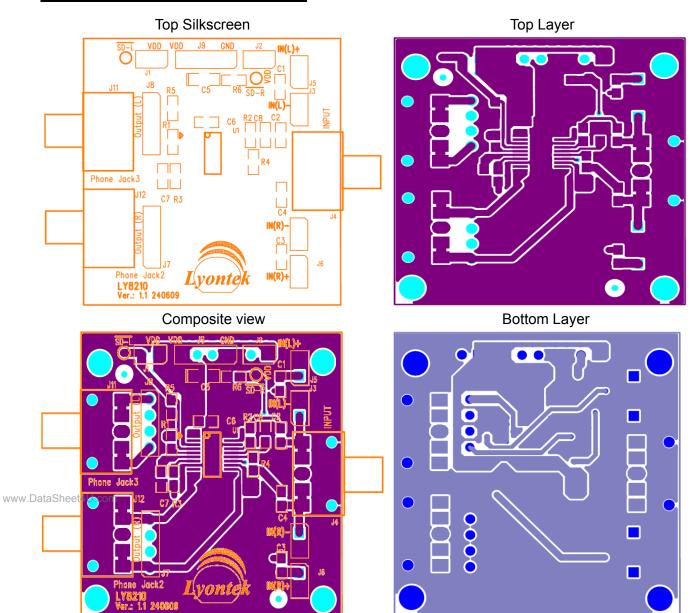
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LY8210AL/IL Demo Board Artwork



LY8210 V1.1 BOM List

No.	Description	Reference	Note
1	Resistor, 150KΩ	R1,R2,R3,R4,R5,R6	
2	Capacitor, 0.1uF	C6,C7,C8	
3	Capacitor, 0.22uF	C1,C2,C3,C4	
4	Capacitor, 10.0uF	C5	
5	LY8210AL/IL	U1	
6	1*2 Pin Header	J1,J2	Close → Active

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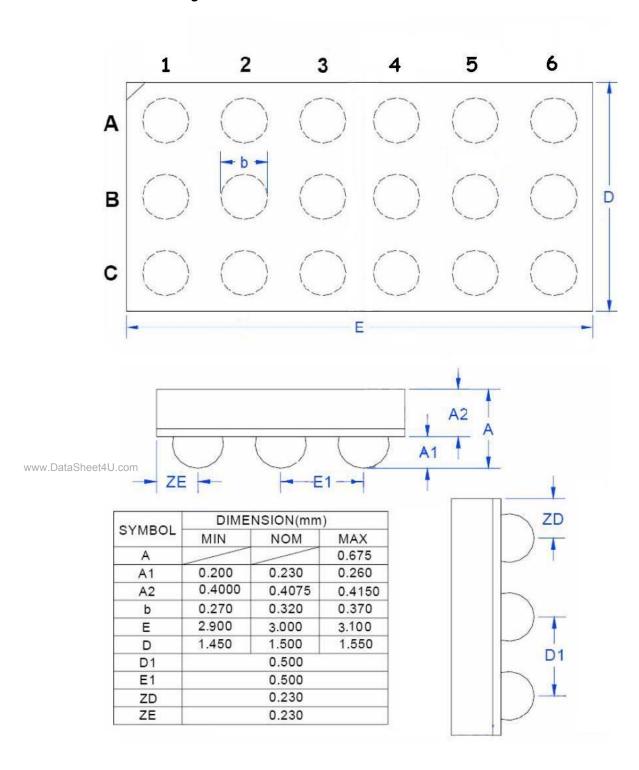
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PACKAGE OUTLINE DIMENSION

18 Ball WCSP Package Outline Dimension

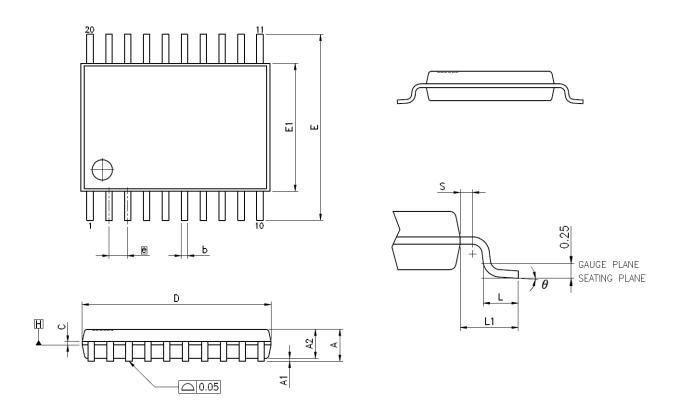


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TSSOP 20 Pin Package Outline Dimension



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SYMBOLS	MIN.	NOM.	MAX.		
Α	_	_	1.20		
A1	0.05	_	0.15		
A2	0.80	0.90	1.05		
b	0.19	_	0.30		
С	0.09	_	0.20		
D	6.40	6.50	6.60		
E1	4.30	4.40	4.50		
E		6.40 BSC			
е		0.65 BSC			
L1	1.00 REF				
L	0.50	0.60	0.75		
S	0.20	_	_		
θ	0*	_	8*		

VARIATIONS (ALL DIMENSIONS SHOWN IN MM)