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# 3850 Group (Spec. H)

User's Manual

RENESAS 8-BIT CISC SINGLE-CHIP MICROCOMPUTER  
740 FAMILY / 38000 SERIES

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REVISION HISTORY

3850 Group (Spec. H) User's Manual

Rev.	Date	Description	
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1.0	Aug. 30, 2001	–	First edition issued
1.1	Sep. 10, 2001	3-5	Limits and test conditions into Table 3.1.5 are partly added.
1.02	Aug. 29, 2003	1-6 1-7 1-7 1-37 1-38 2-103 2-104 2-105 2-106  2-107 3-4 3-20 3-22 3-24	Fig. 4 is partly revised. Table 2 is partly added. Note of Table 3 is added. Fig. 42 is partly revised. Fig. 43 is partly revised. Clause name of "2.11 Flash memory mode" is revised. Notes of Fig. 2.11.3 are partly revised. Table 2.11.2 is partly revised. Explanations of "[Beginning procedure]" of "2.11.6 CPU rewrite mode" are partly added.  Explanations of this page are added. Parameter of Table 3.1.4 is partly revised. Fig. 3.2.16 is partly revised. Fig. 3.2.20 is partly revised. Fig. 3.2.24 is partly revised.
1.03	Sep. 18, 2003	1-51	Fig. 52 is partly revised.



## Preface

This user's manual describes Renesas's CMOS 8-bit microcomputers 3850 Group (Spec. H).

After reading this manual, the user should have a thorough knowledge of the functions and features of the 3850 Group (Spec. H), and should be able to fully utilize the product. The manual starts with specifications and ends with application examples.

For details of software, refer to the "740 Family Software Manual".

The user who is using the 3850 Group (standard) needs to refer to not this manual but "3850/3851 Group User's Manual".

# BEFORE USING THIS MANUAL

This user's manual consists of the following three chapters. Refer to the chapter appropriate to your conditions, such as hardware design or software development. Chapter 3 also includes necessary information for systems development. You must refer to that chapter.

## 1. Organization

### ● CHAPTER 1 HARDWARE

This chapter describes features of the microcomputer and operation of each peripheral function.

### ● CHAPTER 2 APPLICATION

This chapter describes usage and application examples of peripheral functions, based mainly on setting examples of relevant registers.

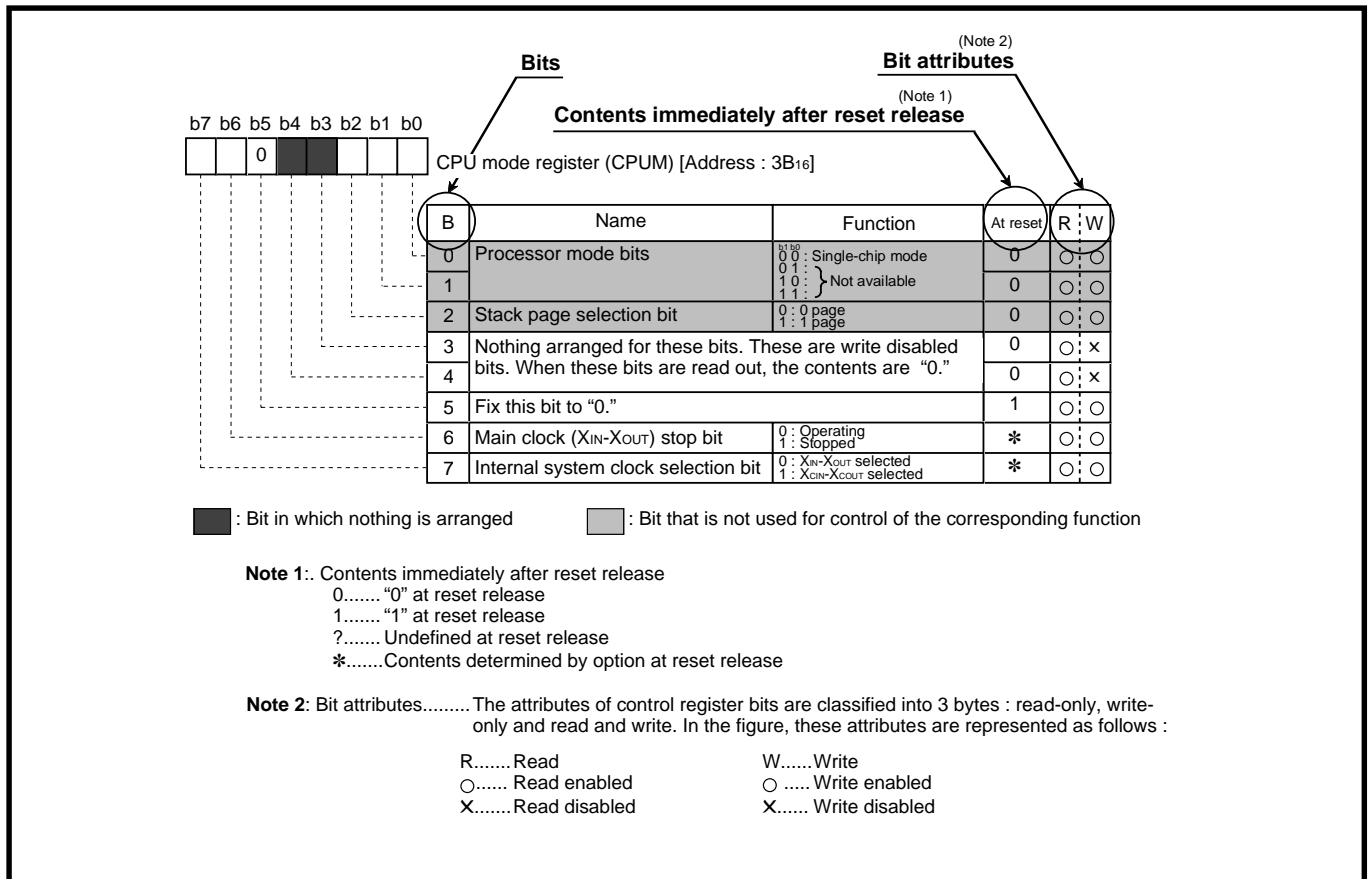
### ● CHAPTER 3 APPENDIX

This chapter includes necessary information for systems development using the microcomputer, such as the electrical characteristics, the notes, and the list of registers.

\*For the mask ROM confirmation form, the ROM programming confirmation form, and the mark specifications, refer to the "Renesas Technology Corp." Homepage (<http://www.renesas.com/en/rom>).

## 2. Structure of register

The figure of each register structure describes its functions, contents at reset, and attributes as follows :



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# CHAPTER 1

## **HARDWARE**

DESCRIPTION

FEATURES

APPLICATION

PIN CONFIGURATION

FUNCTIONAL BLOCK

PIN DESCRIPTION

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FUNCTIONAL DESCRIPTION

NOTES ON PROGRAMMING

NOTES ON USAGE

DATA REQUIRED FOR MASK ORDERS

DATA REQUIRED FOR One Time

PROM PROGRAMMING ORDERS

ROM PROGRAMMING METHOD

FUNCTIONAL DESCRIPTION

SUPPLEMENT

# HARDWARE

## DESCRIPTION/FEATURES/APPLICATION/PIN CONFIGURATION

### DESCRIPTION

The 3850 group (spec. H) is the 8-bit microcomputer based on the 740 family core technology.

The 3850 group (spec. H) is designed for the household products and office automation equipment and includes serial I/O functions, 8-bit timer, and A-D converter.

### FEATURES

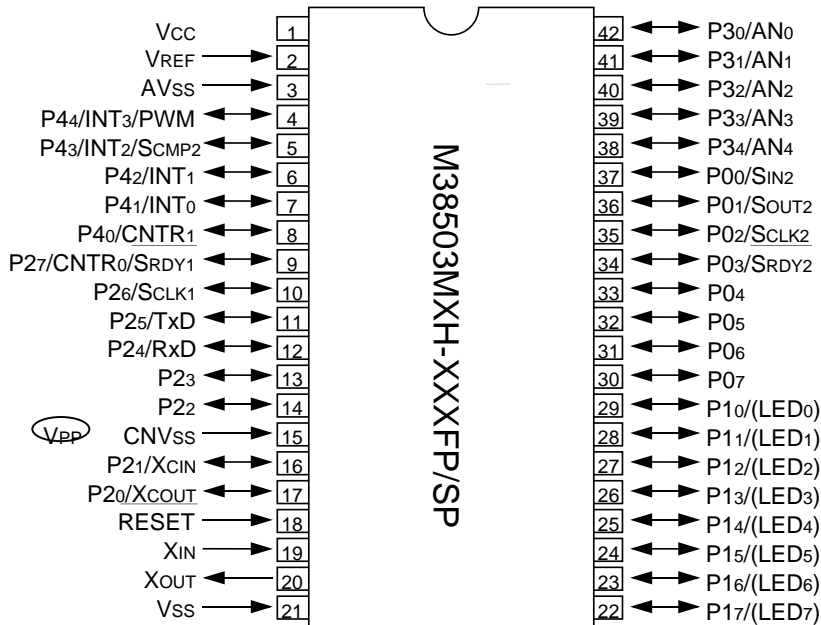
- Basic machine-language instructions ..... 71
- Minimum instruction execution time ..... 0.5  $\mu$ s  
(at 8 MHz oscillation frequency)
- Memory size
  - ROM ..... 8K to 32K bytes
  - RAM ..... 512 to 1024 bytes
- Programmable input/output ports ..... 34
- Interrupts ..... 15 sources, 14 vectors
- Timers ..... 8-bit X 4
- Serial I/O1 ..... 8-bit X 1 (UART or Clock-synchronized)
- Serial I/O2 ..... 8-bit X 1 (Clock-synchronized)
- PWM ..... 8-bit X 1
- A-D converter ..... 10-bit X 5 channels
- Watchdog timer ..... 16-bit X 1
- Clock generating circuit ..... Built-in 2 circuits  
(connect to external ceramic resonator or quartz-crystal oscillator)

- Power source voltage
  - In high-speed mode ..... 4.0 to 5.5 V  
(at 8 MHz oscillation frequency)
  - In middle-speed mode ..... 2.7 to 5.5 V  
(at 8 MHz oscillation frequency)
  - In low-speed mode ..... 2.7 to 5.5 V  
(at 32 kHz oscillation frequency)
- Power dissipation
  - In high-speed mode ..... 34 mW  
(at 8 MHz oscillation frequency, at 5 V power source voltage)
  - In low-speed mode
    - Except M38507F8FP/SP ..... 60  $\mu$ W
    - M38507F8FP/SP ..... 450  $\mu$ W  
(at 32 kHz oscillation frequency, at 3 V power source voltage)
- Operating temperature range ..... -20 to 85°C

### APPLICATION

Office automation equipment, FA equipment, Household products, Consumer electronics, etc.

### PIN CONFIGURATION (TOP VIEW)



○ : Flash memory version

Package type : FP ..... 42P2R-A/E (42-pin plastic-molded SSOP)

Package type : SP ..... 42P4B (42-pin plastic-molded SDIP)

Fig. 1 M38503MXH-XXXFP/SP pin configuration

### FUNCTIONAL BLOCK

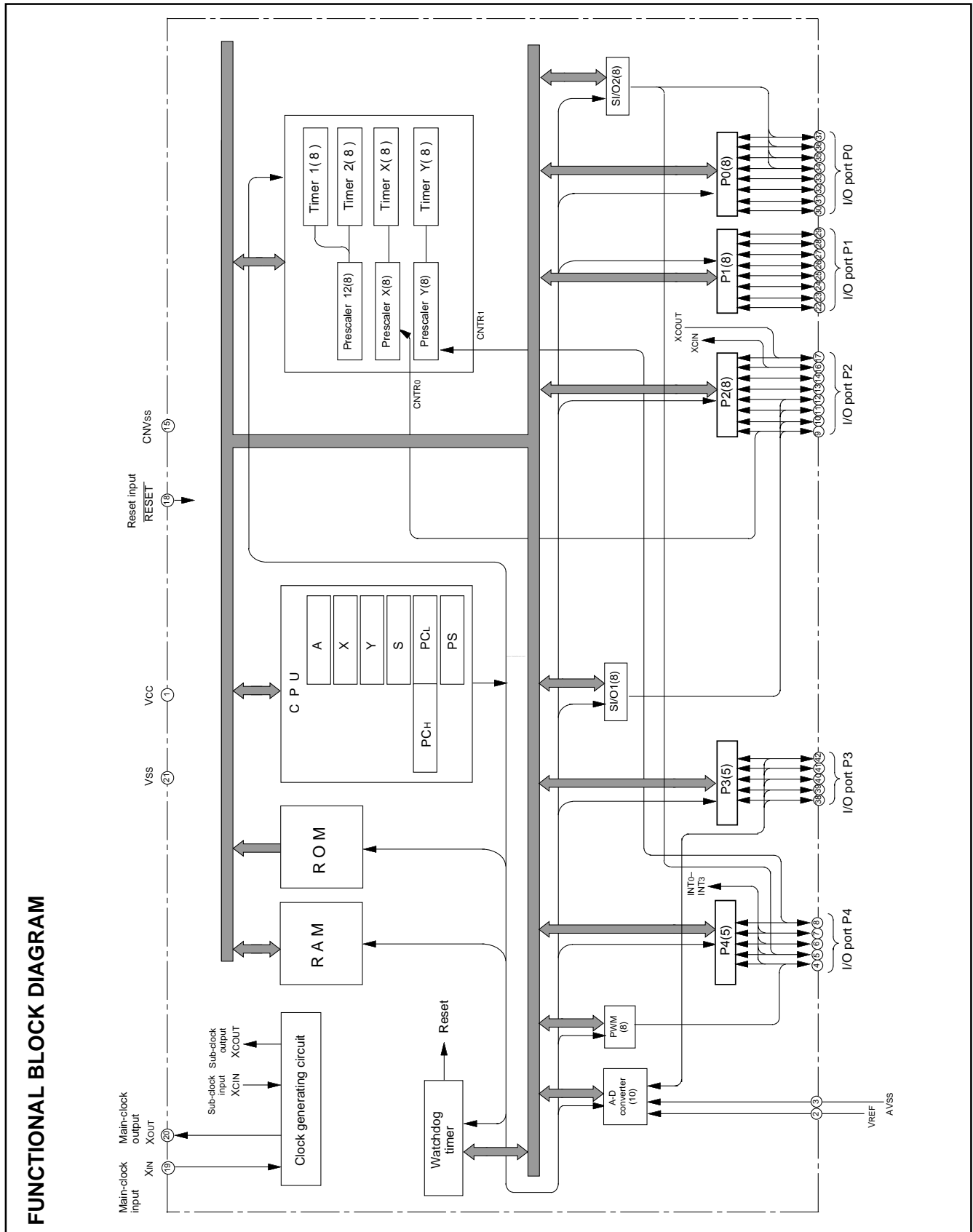


Fig. 2 Functional block diagram

# HARDWARE

## PIN DESCRIPTION

### PIN DESCRIPTION

Table 1 Pin description

Pin	Name	Functions	
			Function except a port function
VCC, VSS	Power source	•Apply voltage of 2.7 V – 5.5 V to Vcc, and 0 V to Vss.	
CNVss	CNVss input	•This pin controls the operation mode of the chip. •Normally connected to Vss.	
$\overline{\text{RESET}}$	Reset input	•Reset input pin for active “L”.	
XIN	Clock input	•Input and output pins for the clock generating circuit. •Connect a ceramic resonator or quartz-crystal oscillator between the XIN and XOUT pins to set the oscillation frequency.	
XOUT	Clock output	•When an external clock is used, connect the clock source to the XIN pin and leave the XOUT pin open.	
P00/SIN2 P01/SOUT2 P02/SCLK2 P03/SRDY2 P04–P07	I/O port P0	<ul style="list-style-type: none"> <li>•8-bit CMOS I/O port.</li> <li>•I/O direction register allows each pin to be individually programmed as either input or output.</li> <li>•CMOS compatible input level.</li> <li>•CMOS 3-state output structure.</li> </ul>	• Serial I/O2 function pin
P10–P17	I/O port P1	•P10 to P17 (8 bits) are enabled to output large current for LED drive.	
P20/XCOUT P21/XCIN P22 P23	I/O port P2	<ul style="list-style-type: none"> <li>•8-bit CMOS I/O port.</li> <li>•I/O direction register allows each pin to be individually programmed as either input or output.</li> <li>•CMOS compatible input level.</li> </ul>	• Sub-clock generating circuit I/O pins (connect a resonator)
P24/RxD P25/TxD		<ul style="list-style-type: none"> <li>•P20, P21, P24 to P27: CMOS3-state output structure.</li> <li>•P22, P23: N-channel open-drain structure.</li> </ul>	• Serial I/O1 function pin
P26/SCLK1 P27/CNTR0/ SRDY1			• Serial I/O1 function pin/ Timer X function pin
P30/AN0– P34/AN4		I/O port P3	<ul style="list-style-type: none"> <li>•8-bit CMOS I/O port with the same function as port P0.</li> <li>•CMOS compatible input level.</li> <li>•CMOS 3-state output structure.</li> </ul>
P40/CNTR1 P41/INT0 P42/INT1 P43/INT2/SCMP2 P44/INT3/PWM	I/O port P4	<ul style="list-style-type: none"> <li>•8-bit CMOS I/O port with the same function as port P0.</li> <li>•CMOS compatible input level.</li> <li>•CMOS 3-state output structure.</li> </ul>	• Timer Y function pin • Interrupt input pins
		• Interrupt input pin • SCMP2 output pin	
		• Interrupt input pin • PWM output pin	

### PART NUMBERING

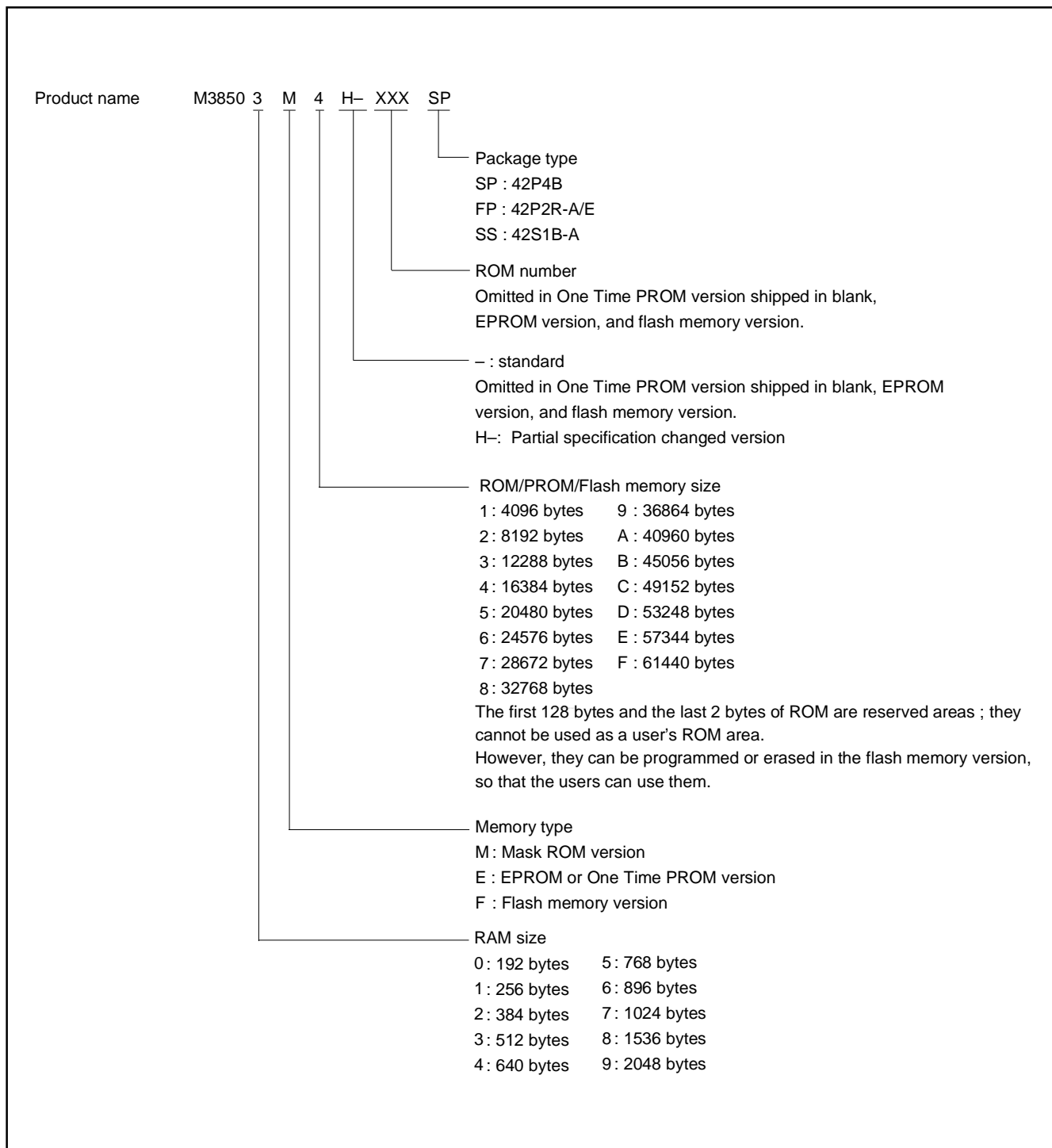


Fig. 3 Part numbering

# HARDWARE

## GROUP EXPANSION

### GROUP EXPANSION

Renesas Technology plans to expand the 3850 group (spec. H) as follows.

### Memory Type

Support for mask ROM, One Time PROM, and flash memory versions.

### Memory Size

Flash memory size ..... 32 K bytes

One Time PROM size ..... 24 K bytes

Mask ROM size ..... 8 K to 32 K bytes

RAM size ..... 512 to 1 K bytes

### Packages

42P4B ..... 42-pin shrink plastic-molded DIP

42P2R-A/E ..... 42-pin plastic-molded SSOP

42S1B-A ..... 42-pin shrink ceramic DIP (EPROM version)

### Memory Expansion Plan

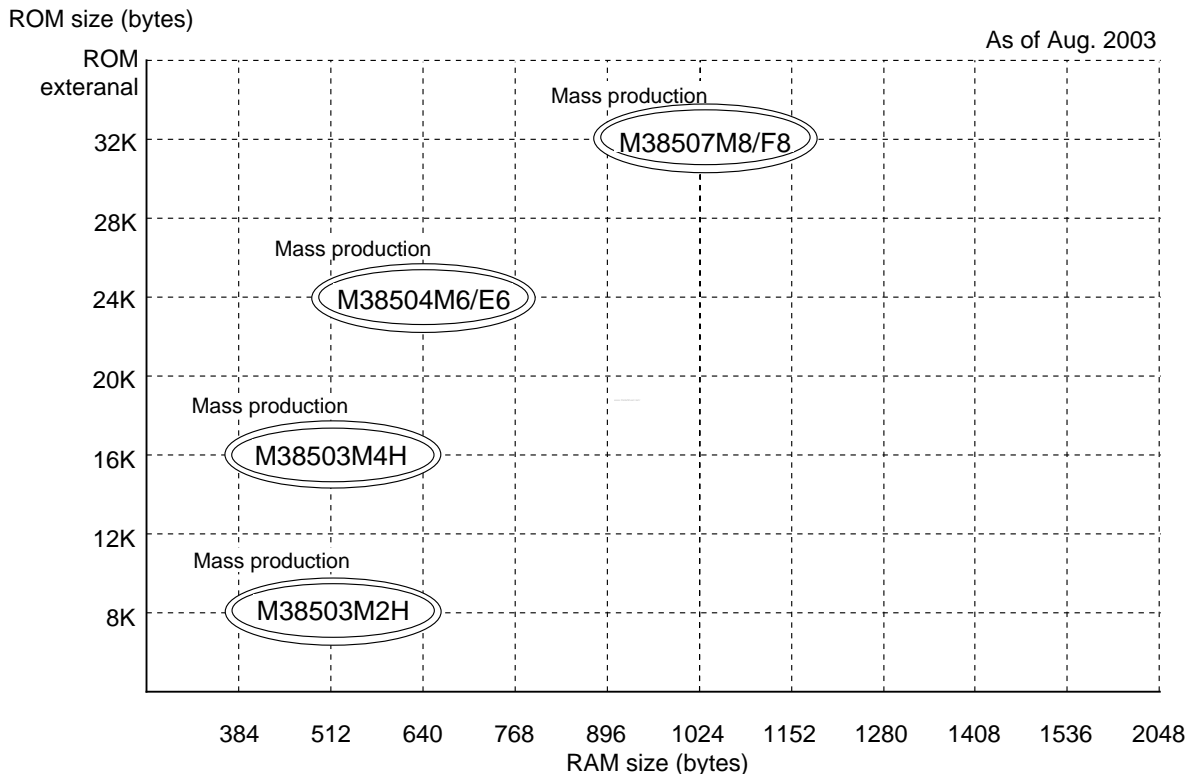


Fig. 4 Memory expansion plan



Currently support products are listed below.

**Table 2 Support products**

As of Aug. 2003

Product name	ROM size (bytes) ROM size for User in ( )	RAM size (bytes)	Package	Remarks
M38503M2H-XXXSP	8192 (8062)	512	42P4B	Mask ROM version
M38503M2H-XXXFP			42P2R-A/E	Mask ROM version
M38503M4H-XXXSP	16384 (16254)	512	424P4B	Mask ROM version
M38503M4H-XXXFP			42P2R-A/E	Mask ROM version
M38504M6-XXXSP	24576 (24446)	640	424P4B	Mask ROM version
M38504E6-XXXSP				One Time PROM version
M38504E6SP				One Time PROM version (blank)
M38504E6SS			42S1B-A	EPROM version
M38504M6-XXXFP				Mask ROM version
M38504E6-XXXFP			42P2R-A/E	One Time PROM version
M38504E6FP				One Time PROM version (blank)
M38507M8-XXXSP				42P4B
M38507M8-XXXFP			42P2R-A/E	
M38507F8SP			32768 (32638)	1024
M38507F8FP	42P2R-A/E			

**Table 3 3850 group (standard) and 3850 group (spec. H) corresponding products**

3850 group (standard) (Note)	3850 group (spec. H)
M38503M2-XXXFP/SP	M38503M2H-XXXFP/SP
M38503M4-XXXFP/SP	M38503M4H-XXXFP/SP
M38503E4-XXXFP/SP	M38504M6-XXXFP/SP
M38503E4FP/SP	M38504E6-XXXFP/SP
M38503E4SS	M38504E6FP/SP
	M38504E6SS
	M38507M8-XXXFP/SP
	M38507F8FP/SP

**Note:** The user who is using the 3850 Group (standard) needs to refer to not this manual but "3850/3851 Group User's Manual".

**Table 4 Differences between 3850 group (standard) and 3850 group (spec. H)**

	3850 group (standard)	3850 group (spec. H)
Serial I/O	1: Serial I/O (UART or Clock-synchronized)	2: Serial I/O1 (UART or Clock-synchronized) Serial I/O2 (Clock-synchronized)
A-D converter	Unserviceable in low-speed mode	Serviceable in low-speed mode
Large current port	5: P13–P17	8: P10–P17

### Notes on differences between 3850 group (standard) and 3850 group (spec. H)

- (1) The absolute maximum ratings of 3850 group (spec. H) is smaller than that of 3850 group (standard).
  - Power source voltage  $V_{cc} = -0.3$  to  $6.5$  V
  - CNVss input voltage  $V_I = -0.3$  to  $V_{cc} + 0.3$  V
- (2) The oscillation circuit constants of X<sub>IN</sub>-X<sub>OUT</sub>, X<sub>CIN</sub>-X<sub>COUT</sub> may be some differences between 3850 group (standard) and 3850 group (spec. H).
- (3) Do not write any data to the reserved area and the reserved bit. (Do not change the contents after reset.)
- (4) Fix bit 3 of the CPU mode register to "1".
- (5) Be sure to perform the termination of unused pins.

# HARDWARE

## FUNCTIONAL DESCRIPTION

### FUNCTIONAL DESCRIPTION CENTRAL PROCESSING UNIT (CPU)

The 3850 group (spec. H) uses the standard 740 Family instruction set. Refer to the table of 740 Family addressing modes and machine instructions or the 740 Family Software Manual for details on the instruction set.

Machine-resident 740 Family instructions are as follows:

- The FST and SLW instructions cannot be used.
- The STP, WIT, MUL, and DIV instructions can be used.

#### [Accumulator (A)]

The accumulator is an 8-bit register. Data operations such as data transfer, etc., are executed mainly through the accumulator.

#### [Index Register X (X)]

The index register X is an 8-bit register. In the index addressing modes, the value of the OPERAND is added to the contents of register X and specifies the real address.

#### [Index Register Y (Y)]

The index register Y is an 8-bit register. In partial instruction, the value of the OPERAND is added to the contents of register Y and specifies the real address.

#### [Stack Pointer (S)]

The stack pointer is an 8-bit register used during subroutine calls and interrupts. This register indicates start address of stored area (stack) for storing registers during subroutine calls and interrupts. The low-order 8 bits of the stack address are determined by the contents of the stack pointer. The high-order 8 bits of the stack address are determined by the stack page selection bit. If the stack page selection bit is "0", the high-order 8 bits becomes "0016". If the stack page selection bit is "1", the high-order 8 bits becomes "0116".

The operations of pushing register contents onto the stack and popping them from the stack are shown in Figure 6.

Store registers other than those described in Figure 6 with program when the user needs them during interrupts or subroutine calls.

#### [Program Counter (PC)]

The program counter is a 16-bit counter consisting of two 8-bit registers PCH and PCL. It is used to indicate the address of the next instruction to be executed.

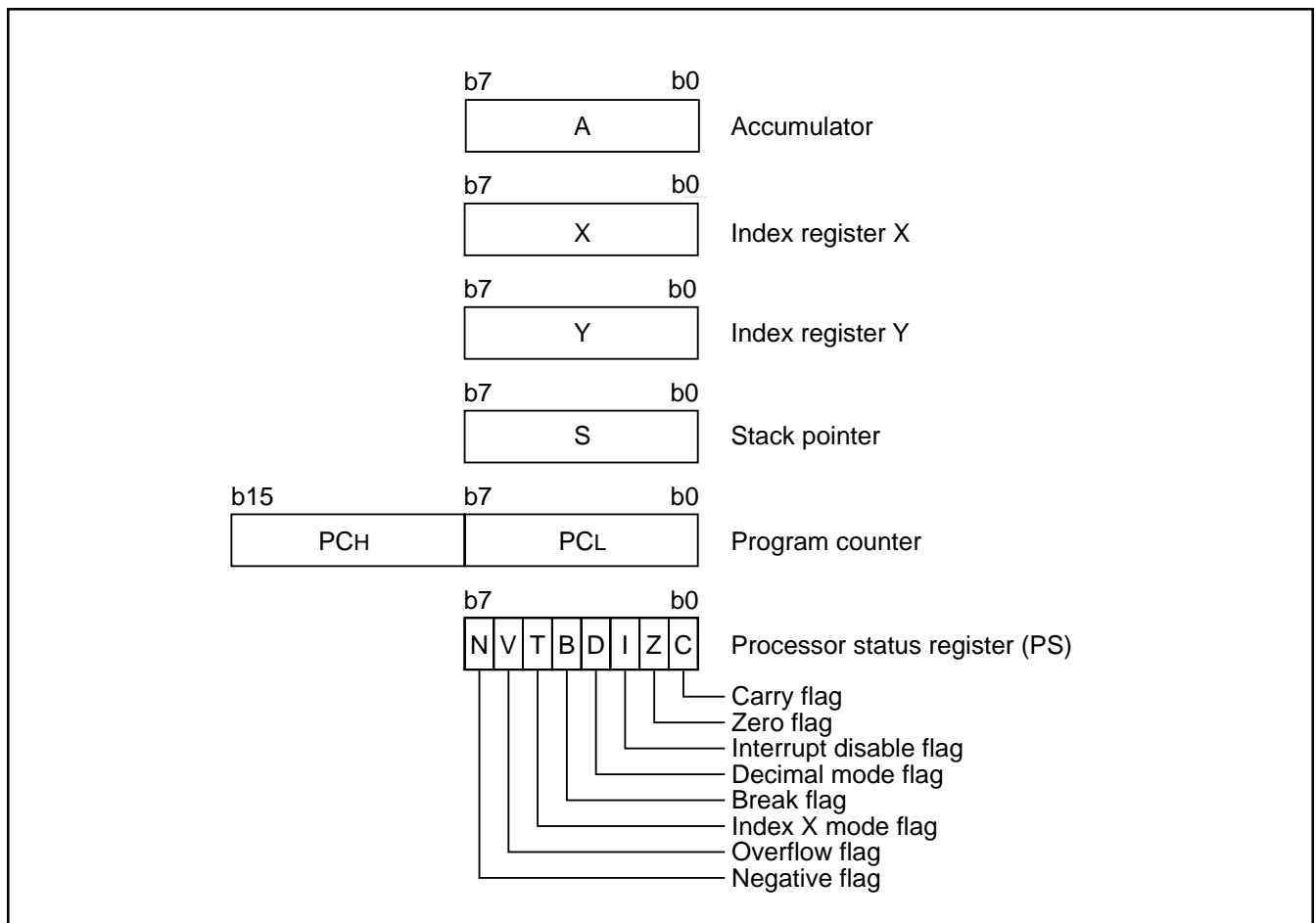


Fig. 5 740 Family CPU register structure

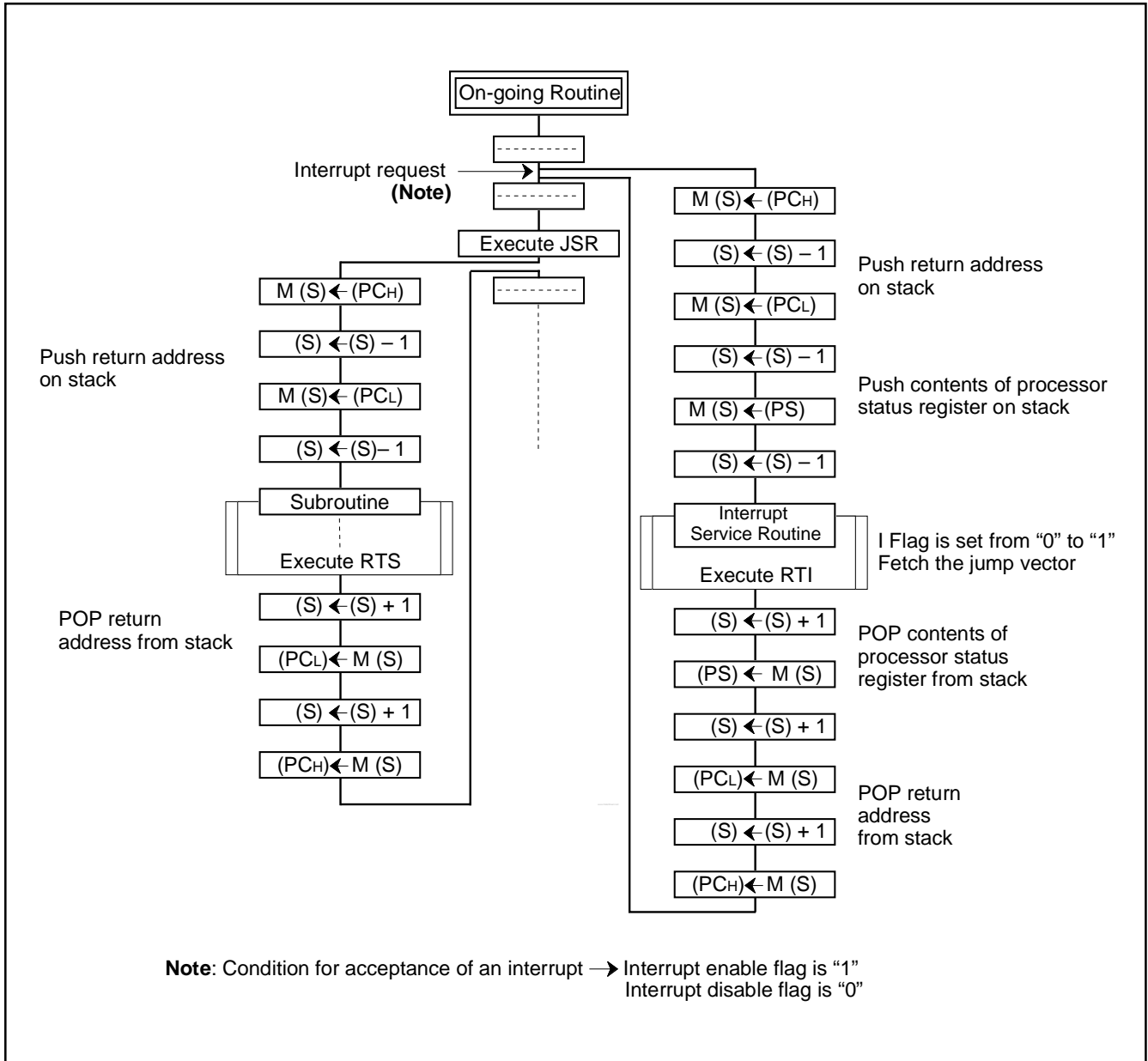


Fig. 6 Register push and pop at interrupt generation and subroutine call

Table 5 Push and pop instructions of accumulator or processor status register

	Push instruction to stack	Pop instruction from stack
Accumulator	PHA	PLA
Processor status register	PHP	PLP

# HARDWARE

## FUNCTIONAL DESCRIPTION

### [Processor status register (PS)]

The processor status register is an 8-bit register consisting of 5 flags which indicate the status of the processor after an arithmetic operation and 3 flags which decide MCU operation. Branch operations can be performed by testing the Carry (C) flag, Zero (Z) flag, Overflow (V) flag, or the Negative (N) flag. In decimal mode, the Z, V, N flags are not valid.

•Bit 0: Carry flag (C)

The C flag contains a carry or borrow generated by the arithmetic logic unit (ALU) immediately after an arithmetic operation. It can also be changed by a shift or rotate instruction.

•Bit 1: Zero flag (Z)

The Z flag is set if the result of an immediate arithmetic operation or a data transfer is "0", and cleared if the result is anything other than "0".

•Bit 2: Interrupt disable flag (I)

The I flag disables all interrupts except for the interrupt generated by the BRK instruction.

Interrupts are disabled when the I flag is "1".

•Bit 3: Decimal mode flag (D)

The D flag determines whether additions and subtractions are executed in binary or decimal. Binary arithmetic is executed when this flag is "0"; decimal arithmetic is executed when it is "1". Decimal correction is automatic in decimal mode. Only the ADC

•Bit 4: Break flag (B)

The B flag is used to indicate that the current interrupt was generated by the BRK instruction. The BRK flag in the processor status register is always "0". When the BRK instruction is used to generate an interrupt, the processor status register is pushed onto the stack with the break flag set to "1".

•Bit 5: Index X mode flag (T)

When the T flag is "0", arithmetic operations are performed between accumulator and memory. When the T flag is "1", direct arithmetic operations and direct data transfers are enabled between memory locations.

•Bit 6: Overflow flag (V)

The V flag is used during the addition or subtraction of one byte of signed data. It is set if the result exceeds +127 to -128. When the BIT instruction is executed, bit 6 of the memory location operated on by the BIT instruction is stored in the overflow flag.

•Bit 7: Negative flag (N)

The N flag is set if the result of an arithmetic operation or data transfer is negative. When the BIT instruction is executed, bit 7 of the memory location operated on by the BIT instruction is stored in the negative flag.

Table 6 Set and clear instructions of each bit of processor status register

	C flag	Z flag	I flag	D flag	B flag	T flag	V flag	N flag
Set instruction	SEC	–	SEI	SED	–	SET	–	–
Clear instruction	CLC	–	CLI	CLD	–	CLT	CLV	–

### [CPU Mode Register (CPUM)] 003B16

The CPU mode register contains the stack page selection bit, etc.

The CPU mode register is allocated at address 003B16.

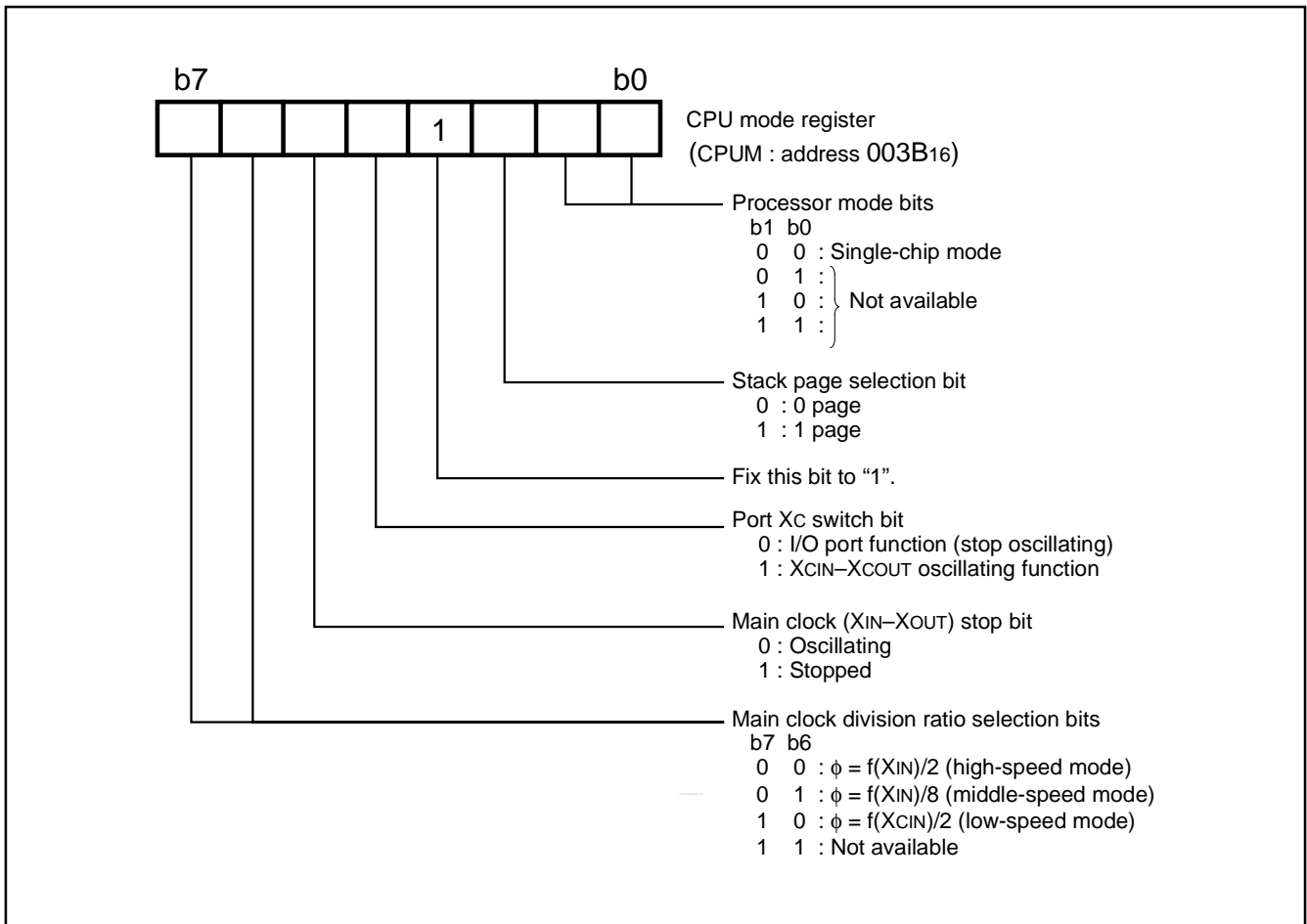


Fig. 7 Structure of CPU mode register

# HARDWARE

## FUNCTIONAL DESCRIPTION

### MEMORY

#### Special Function Register (SFR) Area

The Special Function Register area in the zero page contains control registers such as I/O ports and timers.

#### RAM

RAM is used for data storage and for stack area of subroutine calls and interrupts.

#### ROM

The first 128 bytes and the last 2 bytes of ROM are reserved for device testing and the rest is user area for storing programs.

#### Interrupt Vector Area

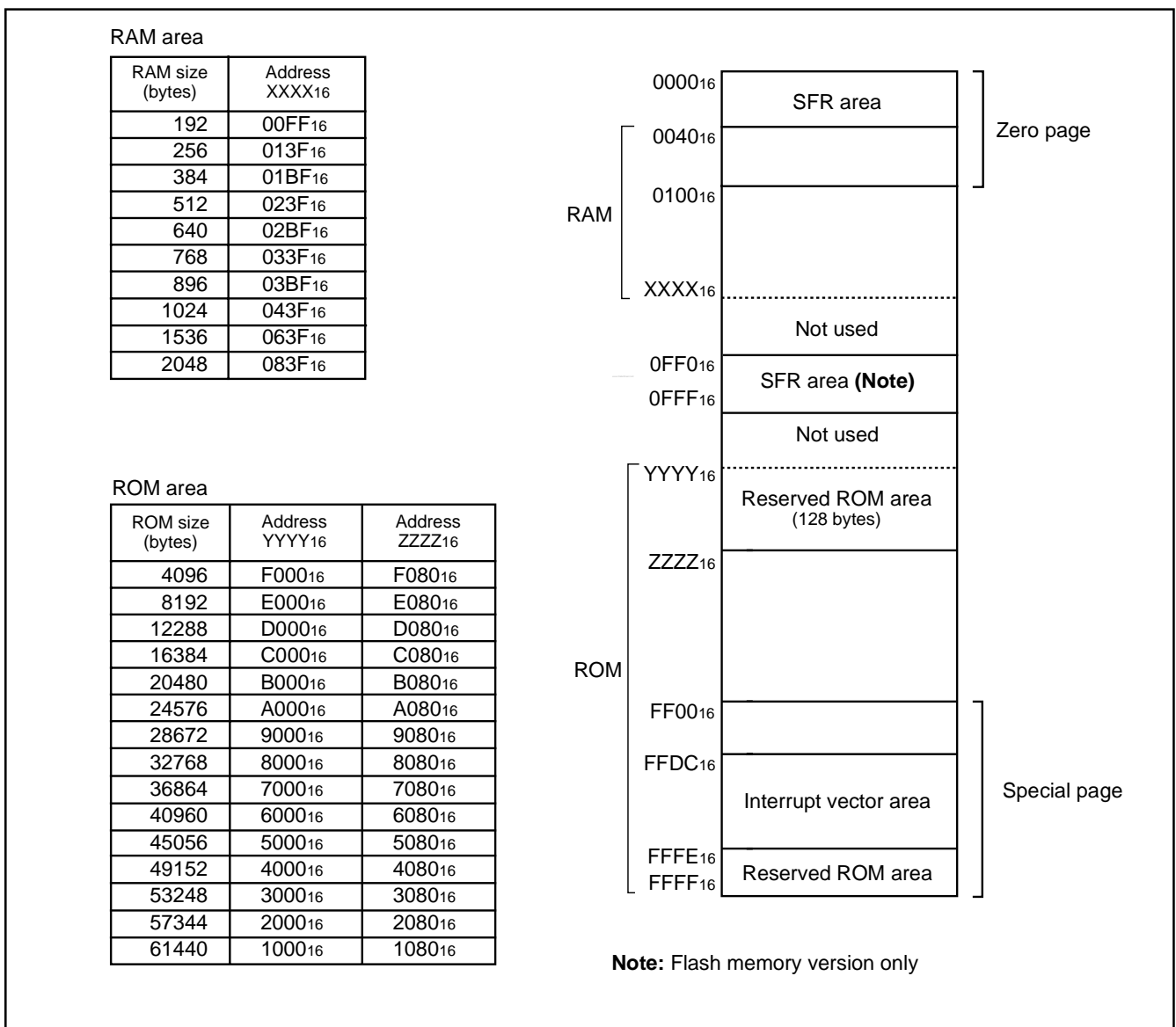
The interrupt vector area contains reset and interrupt vectors.

#### Zero Page

Access to this area with only 2 bytes is possible in the zero page addressing mode.

#### Special Page

Access to this area with only 2 bytes is possible in the special page addressing mode.



**Note:** Flash memory version only

Fig. 8 Memory map diagram

0000 <sub>16</sub>	Port P0 (P0)	0020 <sub>16</sub>	Prescaler 12 (PRE12)
0001 <sub>16</sub>	Port P0 direction register (P0D)	0021 <sub>16</sub>	Timer 1 (T1)
0002 <sub>16</sub>	Port P1 (P1)	0022 <sub>16</sub>	Timer 2 (T2)
0003 <sub>16</sub>	Port P1 direction register (P1D)	0023 <sub>16</sub>	Timer XY mode register (TM)
0004 <sub>16</sub>	Port P2 (P2)	0024 <sub>16</sub>	Prescaler X (PREX)
0005 <sub>16</sub>	Port P2 direction register (P2D)	0025 <sub>16</sub>	Timer X (TX)
0006 <sub>16</sub>	Port P3 (P3)	0026 <sub>16</sub>	Prescaler Y (PREY)
0007 <sub>16</sub>	Port P3 direction register (P3D)	0027 <sub>16</sub>	Timer Y (TY)
0008 <sub>16</sub>	Port P4 (P4)	0028 <sub>16</sub>	Timer count source selection register (TCSS)
0009 <sub>16</sub>	Port P4 direction register (P4D)	0029 <sub>16</sub>	
000A <sub>16</sub>		002A <sub>16</sub>	
000B <sub>16</sub>		002B <sub>16</sub>	Reserved *
000C <sub>16</sub>		002C <sub>16</sub>	Reserved *
000D <sub>16</sub>		002D <sub>16</sub>	Reserved *
000E <sub>16</sub>		002E <sub>16</sub>	Reserved *
000F <sub>16</sub>		002F <sub>16</sub>	Reserved *
0010 <sub>16</sub>		0030 <sub>16</sub>	Reserved *
0011 <sub>16</sub>		0031 <sub>16</sub>	Reserved *
0012 <sub>16</sub>	Reserved *	0032 <sub>16</sub>	
0013 <sub>16</sub>	Reserved *	0033 <sub>16</sub>	
0014 <sub>16</sub>	Reserved *	0034 <sub>16</sub>	A-D control register (ADCON)
0015 <sub>16</sub>	Serial I/O2 control register 1 (SIO2CON1)	0035 <sub>16</sub>	A-D conversion low-order register (ADL)
0016 <sub>16</sub>	Serial I/O2 control register 2 (SIO2CON2)	0036 <sub>16</sub>	A-D conversion high-order register (ADH)
0017 <sub>16</sub>	Serial I/O2 register (SIO2)	0037 <sub>16</sub>	Reserved *
0018 <sub>16</sub>	Transmit/Receive buffer register (TB/RB)	0038 <sub>16</sub>	MISRG
0019 <sub>16</sub>	Serial I/O1 status register (SIOSTS)	0039 <sub>16</sub>	Watchdog timer control register (WDTCON)
001A <sub>16</sub>	Serial I/O1 control register (SIOCON)	003A <sub>16</sub>	Interrupt edge selection register (INTEDGE)
001B <sub>16</sub>	UART control register (UARTCON)	003B <sub>16</sub>	CPU mode register (CPUM)
001C <sub>16</sub>	Baud rate generator (BRG)	003C <sub>16</sub>	Interrupt request register 1 (IREQ1)
001D <sub>16</sub>	PWM control register (PWMCON)	003D <sub>16</sub>	Interrupt request register 2 (IREQ2)
001E <sub>16</sub>	PWM prescaler (PREPWM)	003E <sub>16</sub>	Interrupt control register 1 (ICON1)
001F <sub>16</sub>	PWM register (PWM)	003F <sub>16</sub>	Interrupt control register 2 (ICON2)
		00FE <sub>16</sub>	Flash memory control register (FMCR)

\* Reserved : Do not write any data to this addresses, because these areas are reserved.

Fig. 9 Memory map of special function register (SFR)

# HARDWARE

## FUNCTIONAL DESCRIPTION

### I/O PORTS

The I/O ports have direction registers which determine the input/output direction of each individual pin. Each bit in a direction register corresponds to one pin, and each pin can be set to be input port or output port.

When “0” is written to the bit corresponding to a pin, that pin becomes an input pin. When “1” is written to that bit, that pin becomes an output pin.

If data is read from a pin which is set to output, the value of the port output latch is read, not the value of the pin itself. Pins set to input are floating. If a pin set to input is written to, only the port output latch is written to and the pin remains floating.

**Table 7 I/O port function**

Pin	Name	Input/Output	I/O Structure	Non-Port Function	Related SFRs	Ref.No.			
P00/SIN2	Port P0	Input/output, individual bits	CMOS compatible input level CMOS 3-state output	Serial I/O2 function I/O	Serial I/O2 control register	(1)			
P01/SOUT2						(2)			
P02/SCLK2						(3)			
P03/SRDY2						(4)			
P04–P07	Port P1						(5)		
P10–P17									
P20/XCOUT	Port P2				Sub-clock generating circuit	CPU mode register	(6)		
P21/XCIN							(7)		
P22	Port P2			CMOS compatible input level N-channel open-drain output			(8)		
P23									
P24/RxD							Serial I/O1 function I/O	Serial I/O1 control register	(9)
P25/TxD									(10)
P26/SCLK1	(11)								
P27/CNTR0/SRDY1	Port P3			CMOS compatible input level CMOS 3-state output	Serial I/O1 function I/O Timer X function I/O	Serial I/O1 control register Timer XY mode register	(12)		
P30/AN0– P34/AN4							A-D conversion input	A-D control register	(13)
P40/CNTR1						Timer Y function I/O	Timer XY mode register	(14)	
P41/INT0						External interrupt input	Interrupt edge selection register	(15)	
P42/INT1	External interrupt input SCMP2 output	Interrupt edge selection register Serial I/O2 control register	(16)						
P43/INT2/SCMP2			External interrupt input PWM output	Interrupt edge selection register PWM control register	(17)				
P44/INT3/PWM									

**Note:** When bits 5 to 7 of Ports P3 and P4 are read out, the contents are undefined.



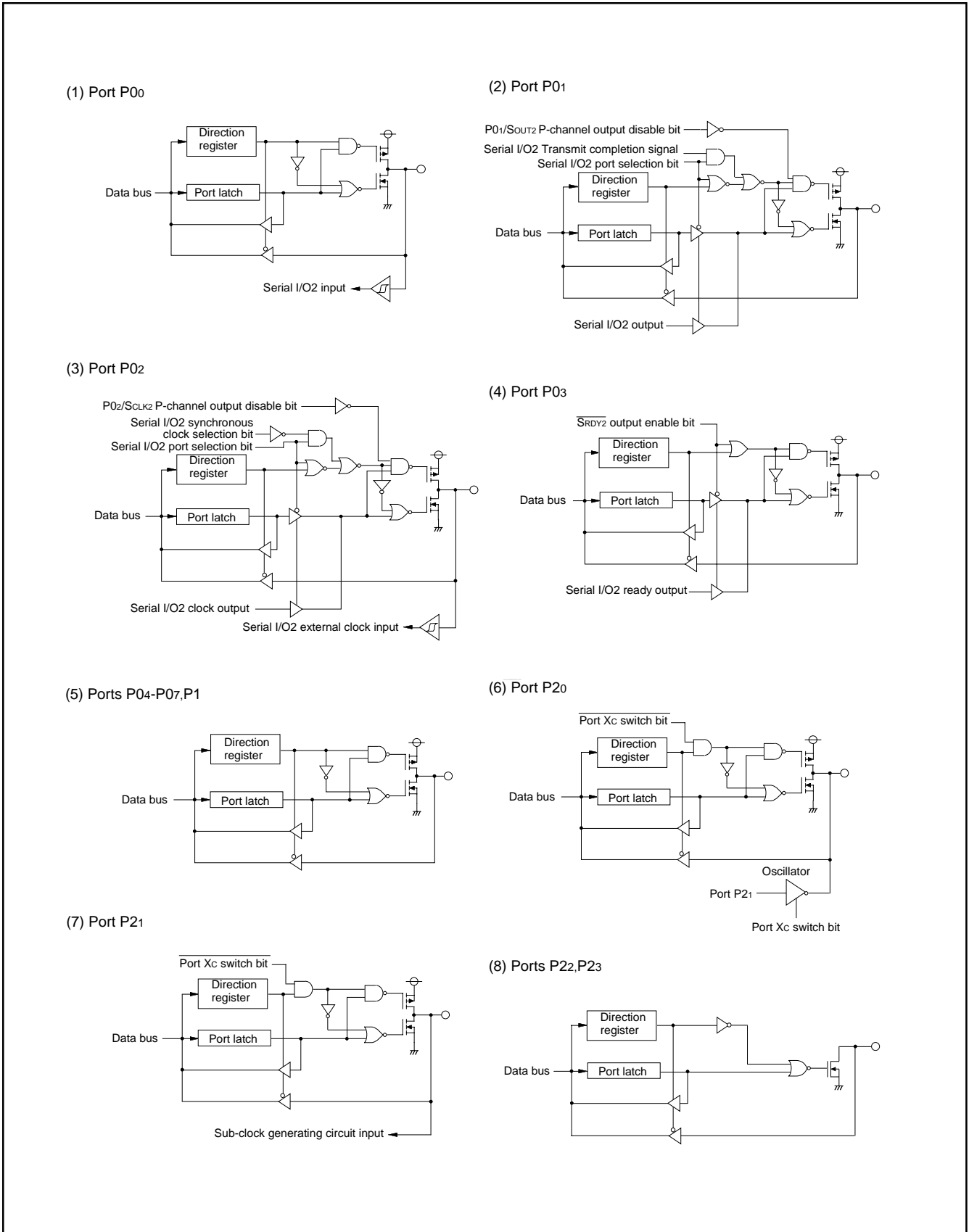


Fig. 10 Port block diagram (1)

# HARDWARE

## FUNCTIONAL DESCRIPTION

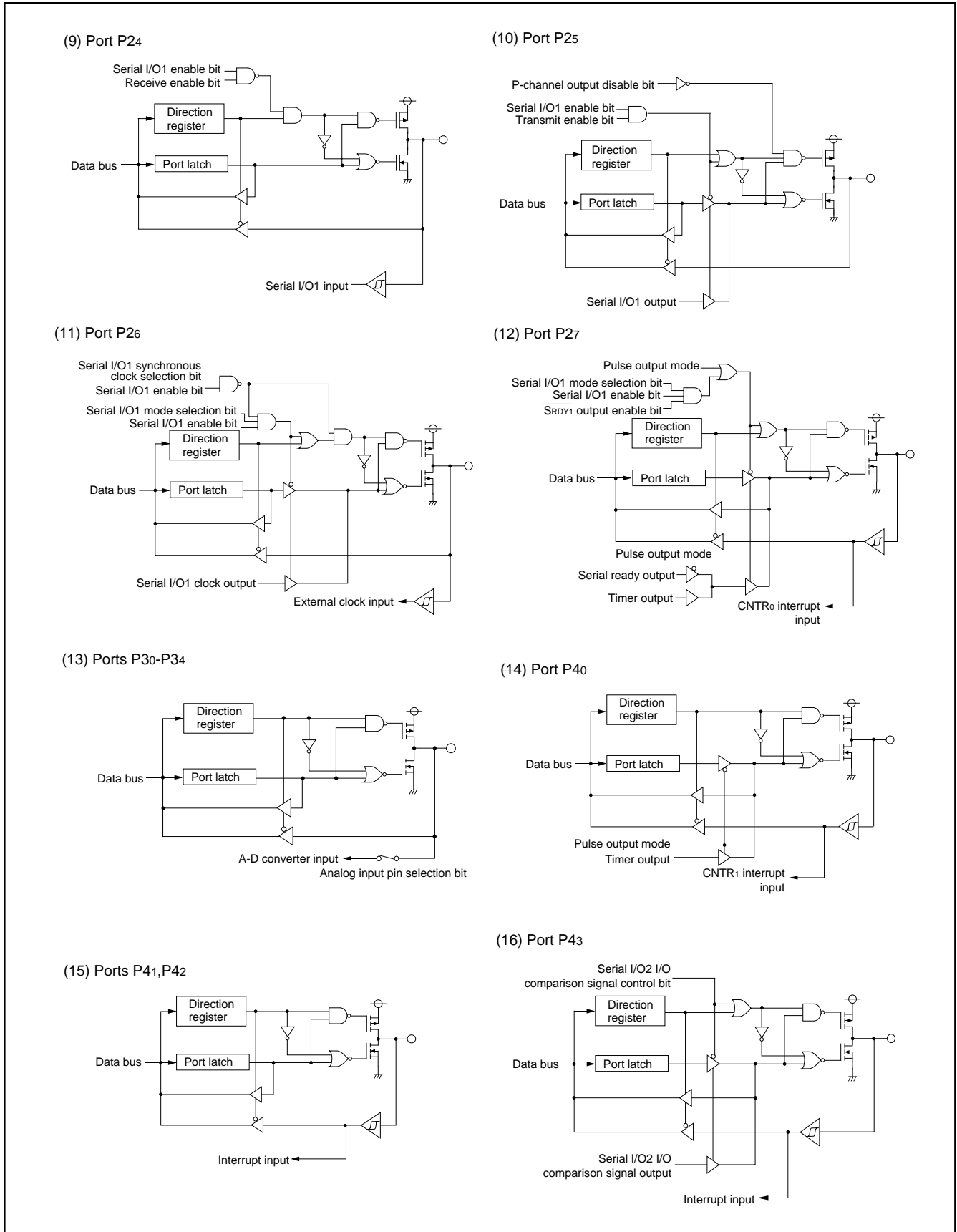


Fig. 11 Port block diagram (2)

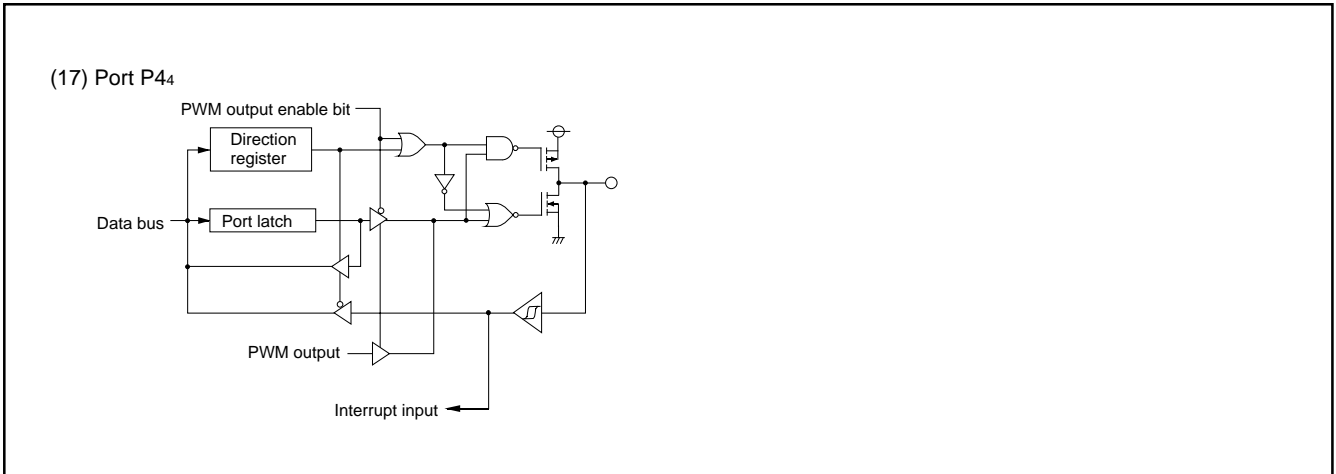


Fig. 12 Port block diagram (3)

# HARDWARE

## FUNCTIONAL DESCRIPTION

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### INTERRUPTS

Interrupts occur by 15 sources among 15 sources: six external, eight internal, and one software.

#### Interrupt Control

Each interrupt is controlled by an interrupt request bit, an interrupt enable bit, and the interrupt disable flag except for the software interrupt set by the BRK instruction. An interrupt occurs if the corresponding interrupt request and enable bits are "1" and the interrupt disable flag is "0".

Interrupt enable bits can be set or cleared by software.

Interrupt request bits can be cleared by software, but cannot be set by software.

The BRK instruction cannot be disabled with any flag or bit. The I (interrupt disable) flag disables all interrupts except the BRK instruction interrupt.

When several interrupts occur at the same time, the interrupts are received according to priority.

#### Interrupt Operation

By acceptance of an interrupt, the following operations are automatically performed:

1. The contents of the program counter and the processor status register are automatically pushed onto the stack.
2. The interrupt disable flag is set and the corresponding interrupt request bit is cleared.
3. The interrupt jump destination address is read from the vector table into the program counter.

### ■Notes

When setting the followings, the interrupt request bit may be set to "1".

- When setting external interrupt active edge

Related register: Interrupt edge selection register (address 3A16)  
Timer XY mode register (address 2316)

- When switching interrupt sources of an interrupt vector address where two or more interrupt sources are allocated

Related register: Interrupt edge selection register (address 3A16)  
When not requiring for the interrupt occurrence synchronized with these setting, take the following sequence.

- ①Set the corresponding interrupt enable bit to "0" (disabled).
- ②Set the interrupt edge select bit or the interrupt source select bit to "1".
- ③Set the corresponding interrupt request bit to "0" after 1 or more instructions have been executed.
- ④Set the corresponding interrupt enable bit to "1" (enabled).

**Table 8 Interrupt vector addresses and priority**

Interrupt Source	Priority	Vector Addresses (Note 1)		Interrupt Request Generating Conditions	Remarks
		High	Low		
Reset (Note 2)	1	FFFD <sub>16</sub>	FFFC <sub>16</sub>	At reset	Non-maskable
INT <sub>0</sub>	2	FFFB <sub>16</sub>	FFFA <sub>16</sub>	At detection of either rising or falling edge of INT <sub>0</sub> input	External interrupt (active edge selectable)
Reserved	3	FFF9 <sub>16</sub>	FFF8 <sub>16</sub>	Reserved	
INT <sub>1</sub>	4	FFF7 <sub>16</sub>	FFF6 <sub>16</sub>	At detection of either rising or falling edge of INT <sub>1</sub> input	External interrupt (active edge selectable)
INT <sub>2</sub>	5	FFF5 <sub>16</sub>	FFF4 <sub>16</sub>	At detection of either rising or falling edge of INT <sub>2</sub> input	External interrupt (active edge selectable)
INT <sub>3</sub> / Serial I/O <sub>2</sub>	6	FFF3 <sub>16</sub>	FFF2 <sub>16</sub>	At detection of either rising or falling edge of INT <sub>3</sub> input/ At completion of serial I/O <sub>2</sub> data reception/transmission	External interrupt (active edge selectable) Switch by Serial I/O <sub>2</sub> /INT <sub>3</sub> interrupt source bit
Reserved	7	FFF1 <sub>16</sub>	FFF0 <sub>16</sub>	Reserved	
Timer X	8	FFEF <sub>16</sub>	FFEE <sub>16</sub>	At timer X underflow	
Timer Y	9	FFED <sub>16</sub>	FFEC <sub>16</sub>	At timer Y underflow	
Timer 1	10	FFEB <sub>16</sub>	FFEA <sub>16</sub>	At timer 1 underflow	STP release timer underflow
Timer 2	11	FFE9 <sub>16</sub>	FFE8 <sub>16</sub>	At timer 2 underflow	
Serial I/O <sub>1</sub> reception	12	FFE7 <sub>16</sub>	FFE6 <sub>16</sub>	At completion of serial I/O <sub>1</sub> data reception	Valid when serial I/O <sub>1</sub> is selected
Serial I/O <sub>1</sub> transmission	13	FFE5 <sub>16</sub>	FFE4 <sub>16</sub>	At completion of serial I/O <sub>1</sub> transfer shift or when transmission buffer is empty	Valid when serial I/O <sub>1</sub> is selected
CNTR <sub>0</sub>	14	FFE3 <sub>16</sub>	FFE2 <sub>16</sub>	At detection of either rising or falling edge of CNTR <sub>0</sub> input	External interrupt (active edge selectable)
CNTR <sub>1</sub>	15	FFE1 <sub>16</sub>	FFE0 <sub>16</sub>	At detection of either rising or falling edge of CNTR <sub>1</sub> input	External interrupt (active edge selectable)
A-D converter	16	FFDF <sub>16</sub>	FFDE <sub>16</sub>	At completion of A-D conversion	
BRK instruction	17	FFDD <sub>16</sub>	FFDC <sub>16</sub>	At BRK instruction execution	Non-maskable software interrupt

**Notes 1:** Vector addresses contain interrupt jump destination addresses.

**2:** Reset function in the same way as an interrupt with the highest priority.

# HARDWARE

## FUNCTIONAL DESCRIPTION

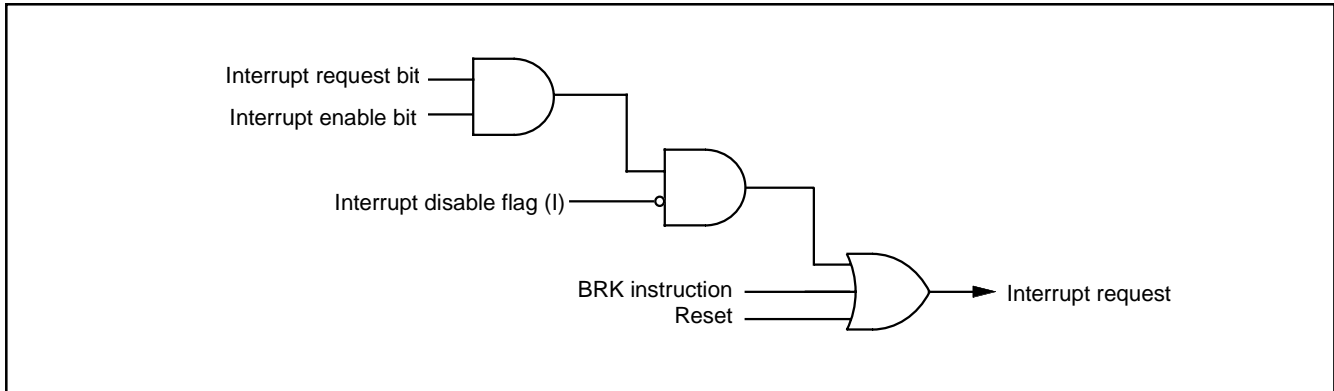


Fig. 13 Interrupt control

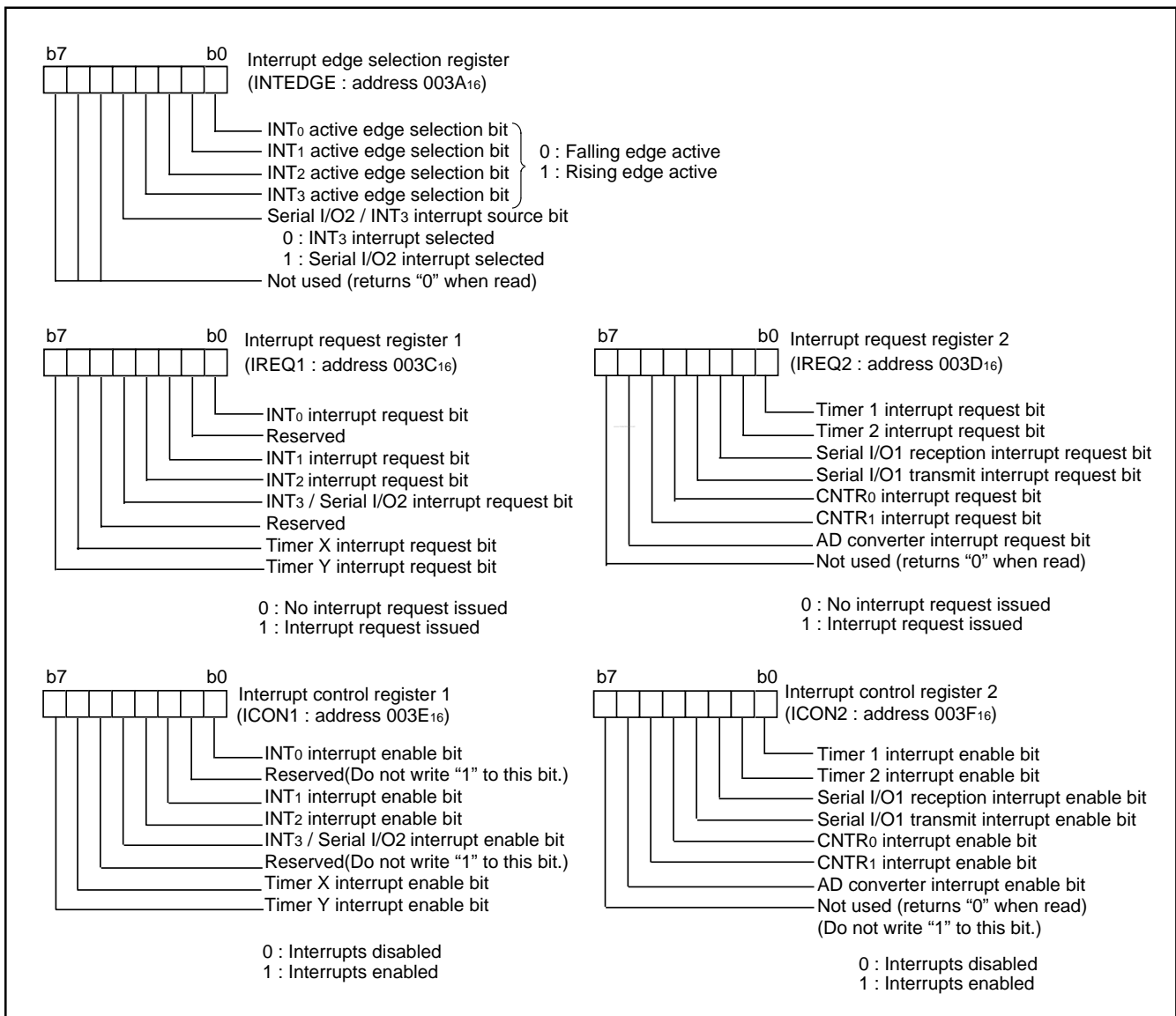


Fig. 14 Structure of interrupt-related registers

### TIMERS

The 3850 group (spec. H) has four timers: timer X, timer Y, timer 1, and timer 2.

The division ratio of each timer or prescaler is given by  $1/(n + 1)$ , where  $n$  is the value in the corresponding timer or prescaler latch. All timers are count down. When the timer reaches "0016", an underflow occurs at the next count pulse and the corresponding timer latch is reloaded into the timer and the count is continued. When a timer underflows, the interrupt request bit corresponding to that timer is set to "1".

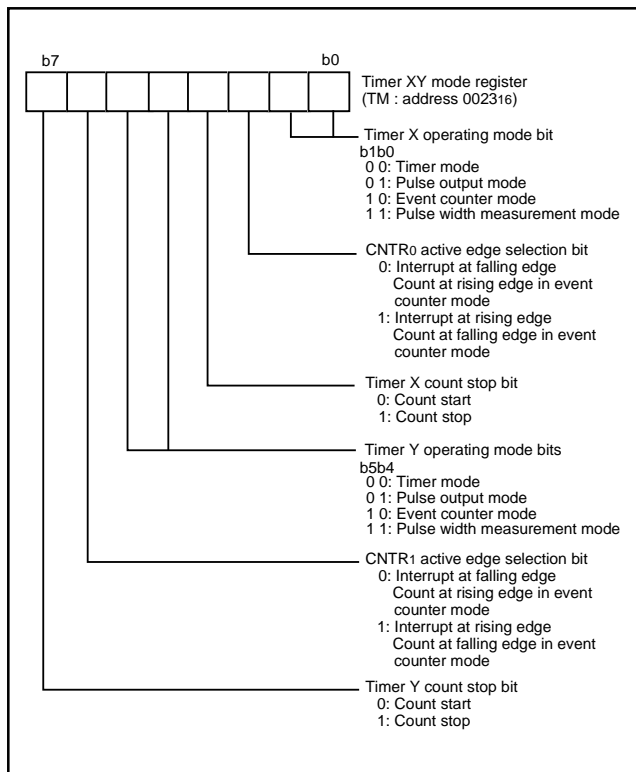


Fig. 15 Structure of timer XY mode register

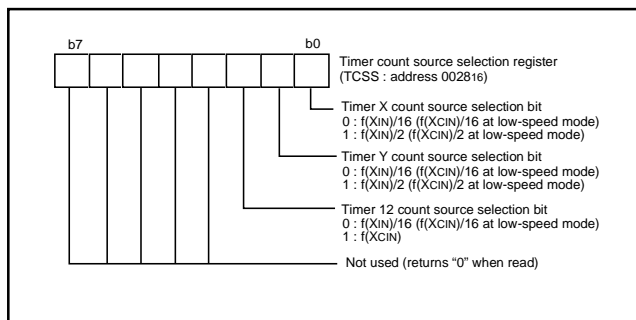


Fig. 16 Structure of timer count source selection register

### Timer 1 and Timer 2

The count source of prescaler 12 is the oscillation frequency which is selected by timer 12 count source selection bit. The output of prescaler 12 is counted by timer 1 and timer 2, and a timer underflow sets the interrupt request bit.

### Timer X and Timer Y

Timer X and Timer Y can each select in one of four operating modes by setting the timer XY mode register.

#### (1) Timer Mode

The timer counts the count source selected by Timer count source selection bit.

#### (2) Pulse Output Mode

The timer counts the count source selected by Timer count source selection bit. Whenever the contents of the timer reach "0016", the signal output from the CNTR0 (or CNTR1) pin is inverted. If the CNTR0 (or CNTR1) active edge selection bit is "0", output begins at "H".

If it is "1", output starts at "L". When using a timer in this mode, set the corresponding port P27 (or port P40) direction register to output mode.

#### (3) Event Counter Mode

Operation in event counter mode is the same as in timer mode, except that the timer counts signals input through the CNTR0 or CNTR1 pin.

When the CNTR0 (or CNTR1) active edge selection bit is "0", the rising edge of the CNTR0 (or CNTR1) pin is counted.

When the CNTR0 (or CNTR1) active edge selection bit is "1", the falling edge of the CNTR0 (or CNTR1) pin is counted.

#### (4) Pulse Width Measurement Mode

If the CNTR0 (or CNTR1) active edge selection bit is "0", the timer counts the selected signals by the count source selection bit while the CNTR0 (or CNTR1) pin is at "H". If the CNTR0 (or CNTR1) active edge selection bit is "1", the timer counts it while the CNTR0 (or CNTR1) pin is at "L".

The count can be stopped by setting "1" to the timer X (or timer Y) count stop bit in any mode. The corresponding interrupt request bit is set each time a timer underflows.

#### ■Note

When switching the count source by the timer 12, X and Y count source bit, the value of timer count is altered in unconsiderable amount owing to generating of a thin pulses in the count input signals.

Therefore, select the timer count source before set the value to the prescaler and the timer.

# HARDWARE

## FUNCTIONAL DESCRIPTION

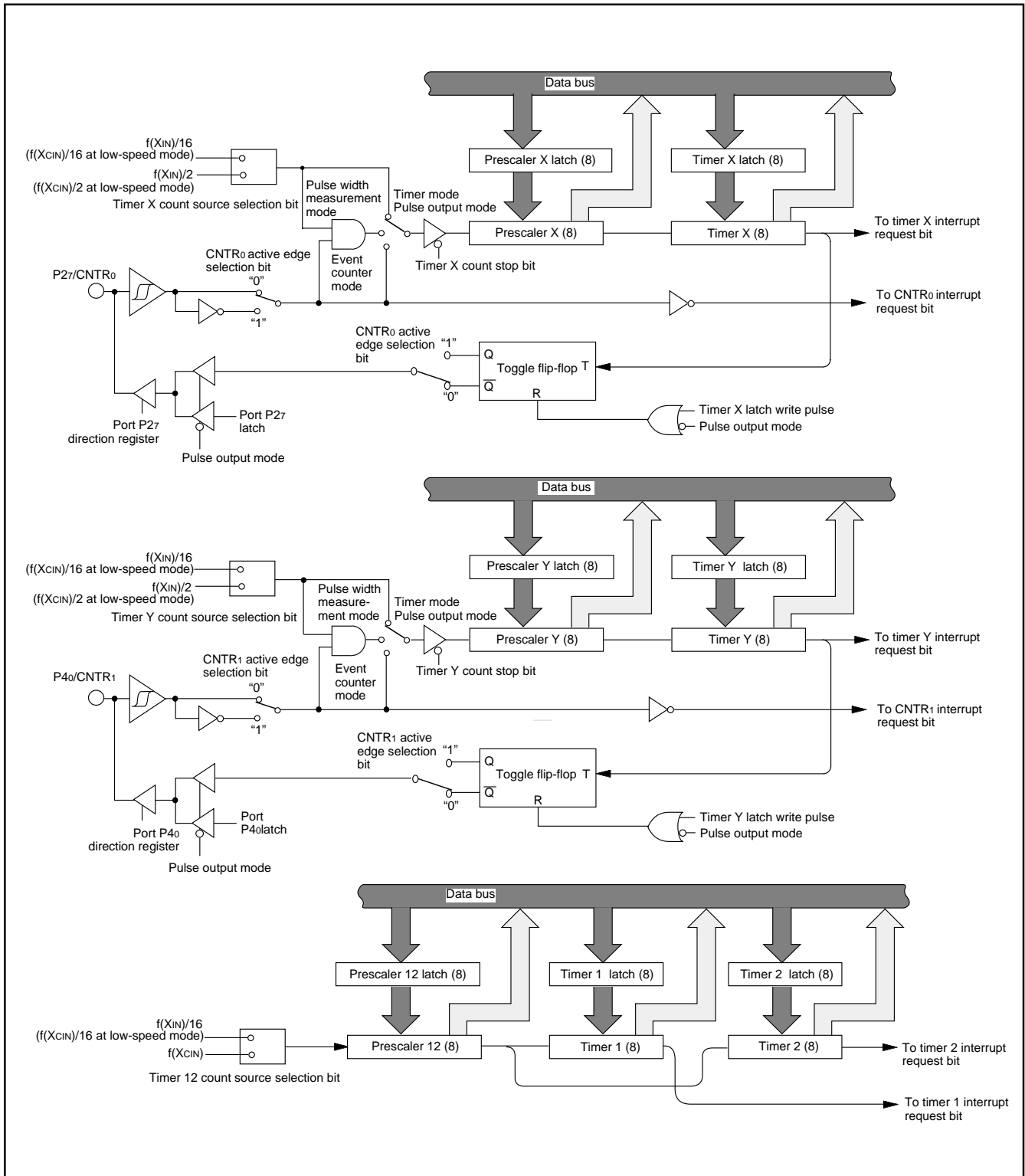


Fig. 17 Block diagram of timer X, timer Y, timer 1, and timer 2



### SERIAL I/O

#### ●SERIAL I/O1

Serial I/O1 can be used as either clock synchronous or asynchronous (UART) serial I/O. A dedicated timer is also provided for baud rate generation.

#### (1) Clock Synchronous Serial I/O Mode

Clock synchronous serial I/O mode can be selected by setting the serial I/O1 mode selection bit of the serial I/O1 control register (bit 6 of address 001A16) to "1".

For clock synchronous serial I/O, the transmitter and the receiver must use the same clock. If an internal clock is used, transfer is started by a write signal to the TB/RB.

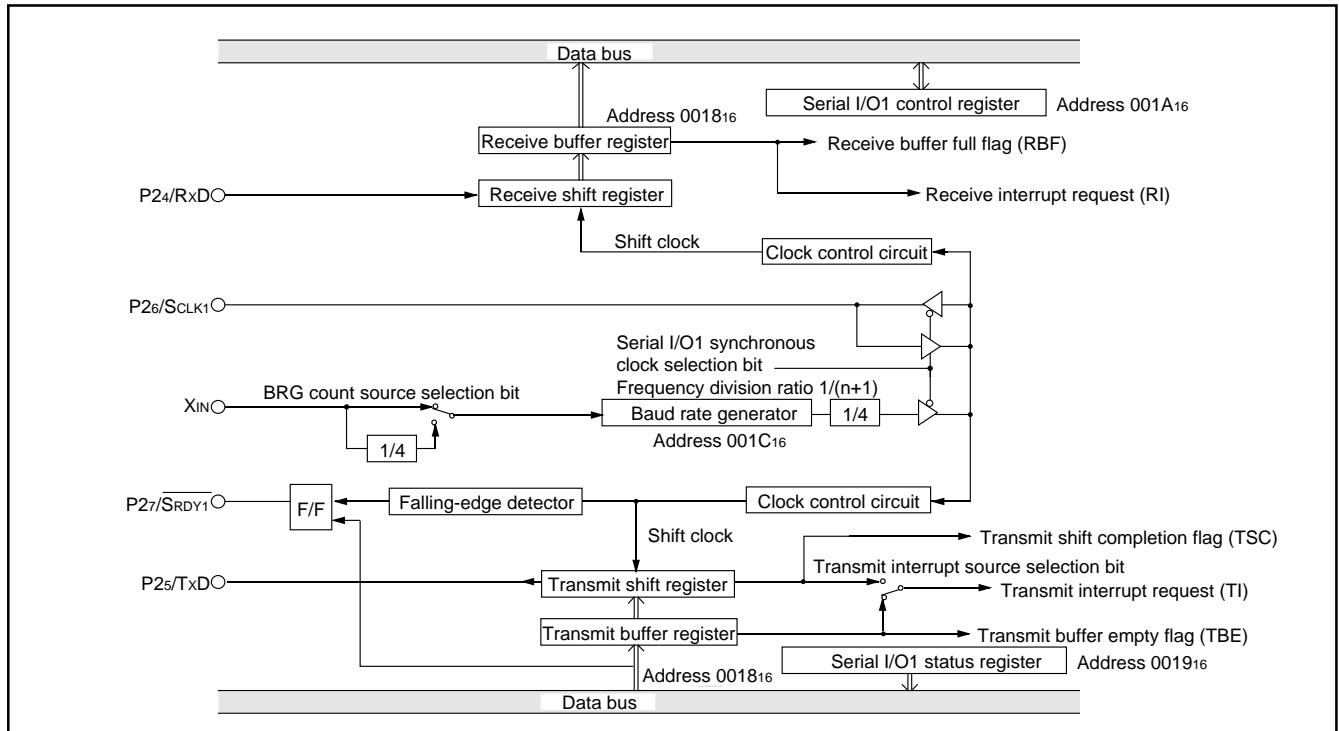


Fig. 18 Block diagram of clock synchronous serial I/O1

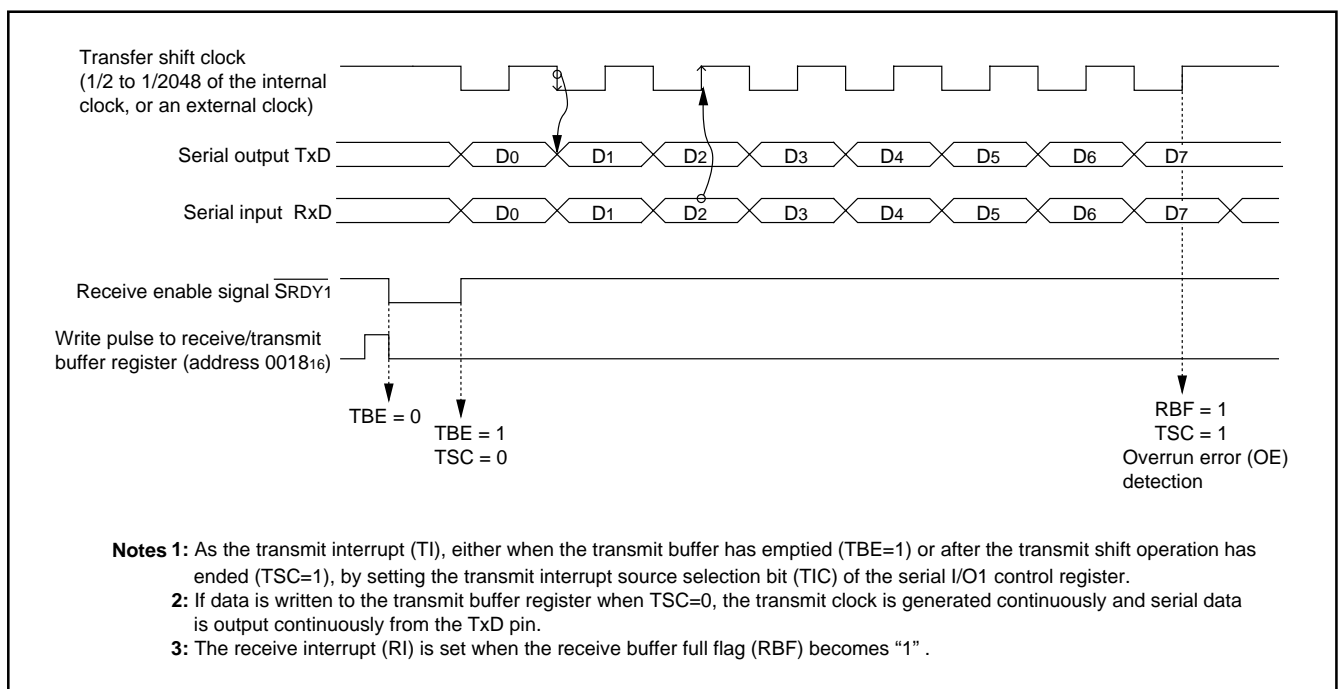


Fig. 19 Operation of clock synchronous serial I/O1 function

# HARDWARE

## FUNCTIONAL DESCRIPTION

### (2) Asynchronous Serial I/O (UART) Mode

Clock asynchronous serial I/O mode (UART) can be selected by clearing the serial I/O mode selection bit (b6) of the serial I/O1 control register to "0".

Eight serial data transfer formats can be selected, and the transfer formats used by a transmitter and receiver must be identical.

The transmit and receive shift registers each have a buffer, but the

two buffers have the same address in memory. Since the shift register cannot be written to or read from directly, transmit data is written to the transmit buffer register, and receive data is read from the receive buffer register.

The transmit buffer register can also hold the next data to be transmitted, and the receive buffer register can hold a character while the next character is being received.

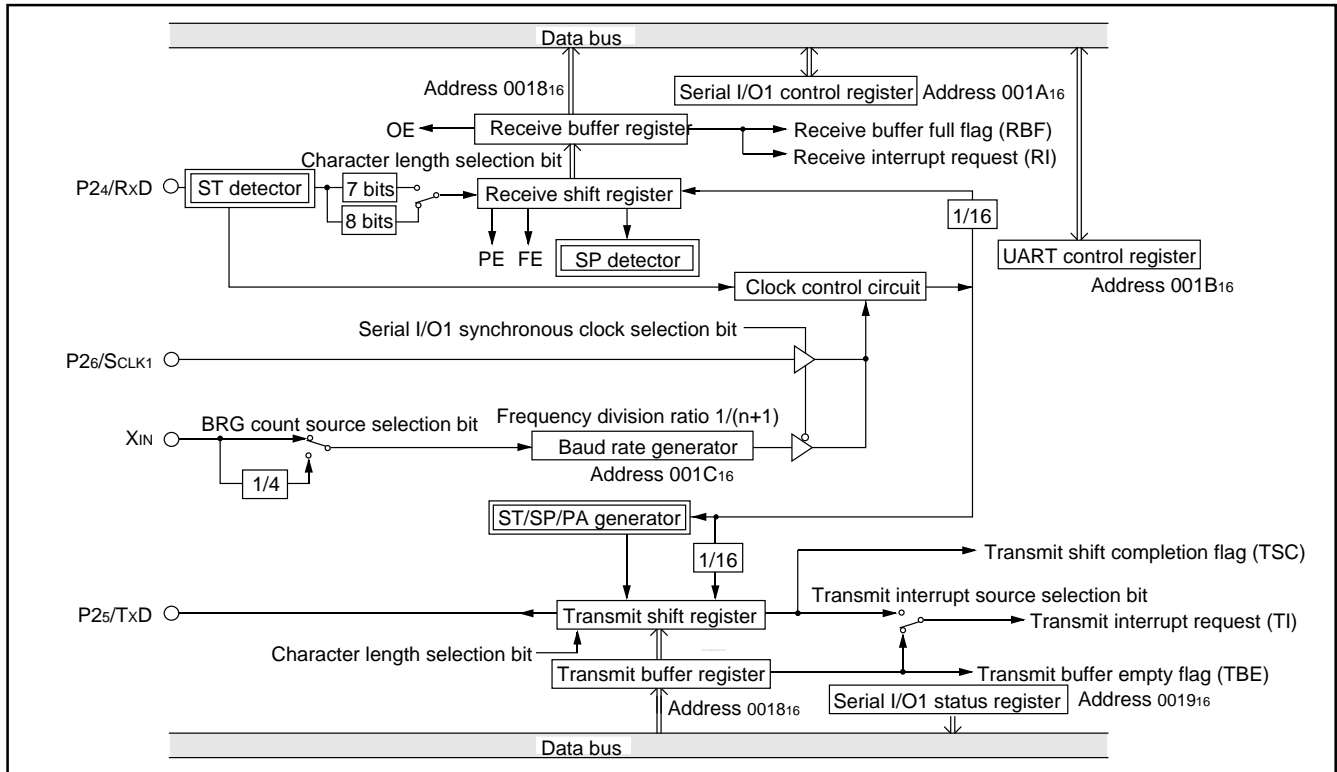
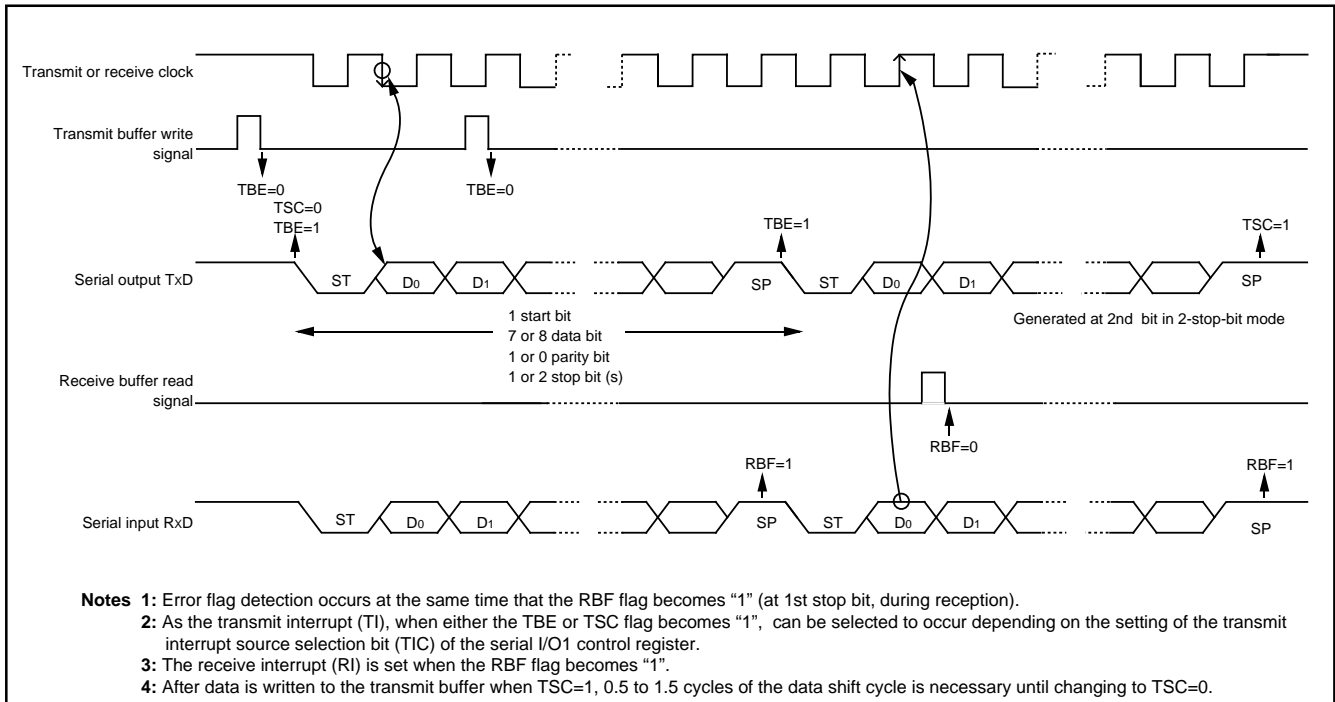


Fig. 20 Block diagram of UART serial I/O1



**Fig. 21 Operation of UART serial I/O1 function**

### [Transmit Buffer Register/Receive Buffer Register (TB/RB)] 001816

The transmit buffer register and the receive buffer register are located at the same address. The transmit buffer is write-only and the receive buffer is read-only. If a character bit length is 7 bits, the MSB of data stored in the receive buffer is "0".

### [Serial I/O1 Status Register (SIOSTS)] 001916

The read-only serial I/O1 status register consists of seven flags (bits 0 to 6) which indicate the operating status of the serial I/O1 function and various errors.

Three of the flags (bits 4 to 6) are valid only in UART mode.

The receive buffer full flag (bit 1) is cleared to "0" when the receive buffer register is read.

If there is an error, it is detected at the same time that data is transferred from the receive shift register to the receive buffer register, and the receive buffer full flag is set. A write to the serial I/O1 status register clears all the error flags OE, PE, FE, and SE (bit 3 to bit 6, respectively). Writing "0" to the serial I/O1 enable bit SIOE (bit 7 of the serial I/O1 control register) also clears all the status flags, including the error flags.

Bits 0 to 6 of the serial I/O1 status register are initialized to "0" at reset, but if the transmit enable bit (bit 4) of the serial I/O1 control register has been set to "1", the transmit shift completion flag (bit 2) and the transmit buffer empty flag (bit 0) become "1".

### [Serial I/O1 Control Register (SIOCON)] 001A16

The serial I/O1 control register consists of eight control bits for the serial I/O1 function.

### [UART Control Register (UARTCON)] 001B16

The UART control register consists of four control bits (bits 0 to 3) which are valid when asynchronous serial I/O is selected and set the data format of a data transfer and one bit (bit 4) which is always valid and sets the output structure of the P25/TxD pin.

### [Baud Rate Generator (BRG)] 001C16

The baud rate generator determines the baud rate for serial transfer.

The baud rate generator divides the frequency of the count source by  $1/(n + 1)$ , where  $n$  is the value written to the baud rate generator.

# HARDWARE

## FUNCTIONAL DESCRIPTION

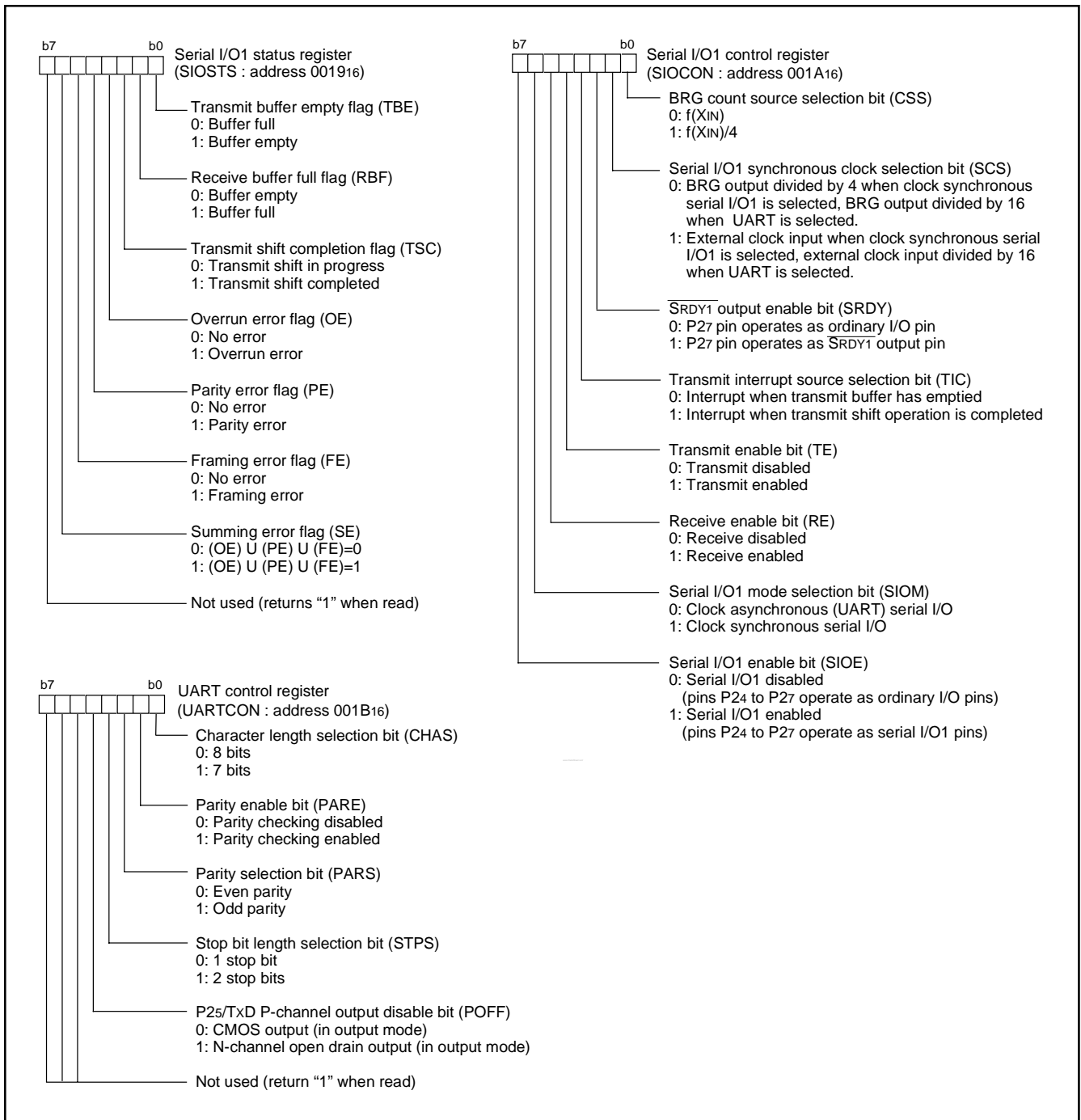


Fig. 22 Structure of serial I/O1 control registers

### ■Notes on serial I/O

When setting the transmit enable bit of serial I/O1 to "1", the serial I/O1 transmit interrupt request bit is automatically set to "1". When not requiring the interrupt occurrence synchronized with the transmission enabled, take the following sequence.

- ① Set the serial I/O1 transmit interrupt enable bit to "0" (disabled).
- ② Set the transmit enable bit to "1".
- ③ Set the serial I/O1 transmit interrupt request bit to "0" after 1 or more instructions have been executed.
- ④ Set the serial I/O1 transmit interrupt enable bit to "1" (enabled).

### ● SERIAL I/O2

The serial I/O2 can be operated only as the clock synchronous type. As a synchronous clock for serial transfer, either internal clock or external clock can be selected by the serial I/O2 synchronous clock selection bit (b6) of serial I/O2 control register 1.

The internal clock incorporates a dedicated divider and permits selecting 6 types of clock by the internal synchronous clock selection bits (b2, b1, b0) of serial I/O2 control register 1.

Regarding SOUT2 and SCLK2 being output pins, either CMOS output format or N-channel open-drain output format can be selected by the P01/SOUT2, P02/SCLK2 P-channel output disable bit (b7) of serial I/O2 control register 1.

When the internal clock has been selected, a transfer starts by a write signal to the serial I/O2 register (address 001716). After completion of data transfer, the level of the SOUT2 pin goes to high impedance automatically but bit 7 of the serial I/O2 control register 2 is not set to "1" automatically.

When the external clock has been selected, the contents of the serial I/O2 register is continuously sifted while transfer clocks are input. Accordingly, control the clock externally. Note that the SOUT2 pin does not go to high impedance after completion of data transfer.

To cause the SOUT2 pin to go to high impedance in the case where the external clock is selected, set bit 7 of the serial I/O2 control register 2 to "1" when SCLK2 is "H" after completion of data transfer. After the next data transfer is started (the transfer clock falls), bit 7 of the serial I/O2 control register 2 is set to "0" and the SOUT2 pin is put into the active state.

Regardless of the internal clock to external clock, the interrupt request bit is set after the number of bits (1 to 8 bits) selected by the optional transfer bit is transferred. In case of a fractional number of bits less than 8 bits as the last data, the received data to be stored in the serial I/O2 register becomes a fractional number of bits close to MSB if the transfer direction selection bit of serial I/O2 control register 1 is LSB first, or a fractional number of bits close to LSB if the said bit is MSB first. For the remaining bits, the previously received data is shifted.

At transmit operation using the clock synchronous serial I/O, the SCMP2 signal can be output by comparing the state of the transmit pin SOUT2 with the state of the receive pin SIN2 in synchronization with a rise of the transfer clock. If the output level of the SOUT2 pin is equal to the input level to the SIN2 pin, "L" is output from the SCMP2 pin. If not, "H" is output. At this time, an INT2 interrupt request can also be generated. Select a valid edge by bit 2 of the interrupt edge selection register (address 003A16).

### [Serial I/O2 Control Registers 1, 2 (SIO2CON1 / SIO2CON2)] 001516, 001616

The serial I/O2 control registers 1 and 2 are containing various selection bits for serial I/O2 control as shown in Figure 23.

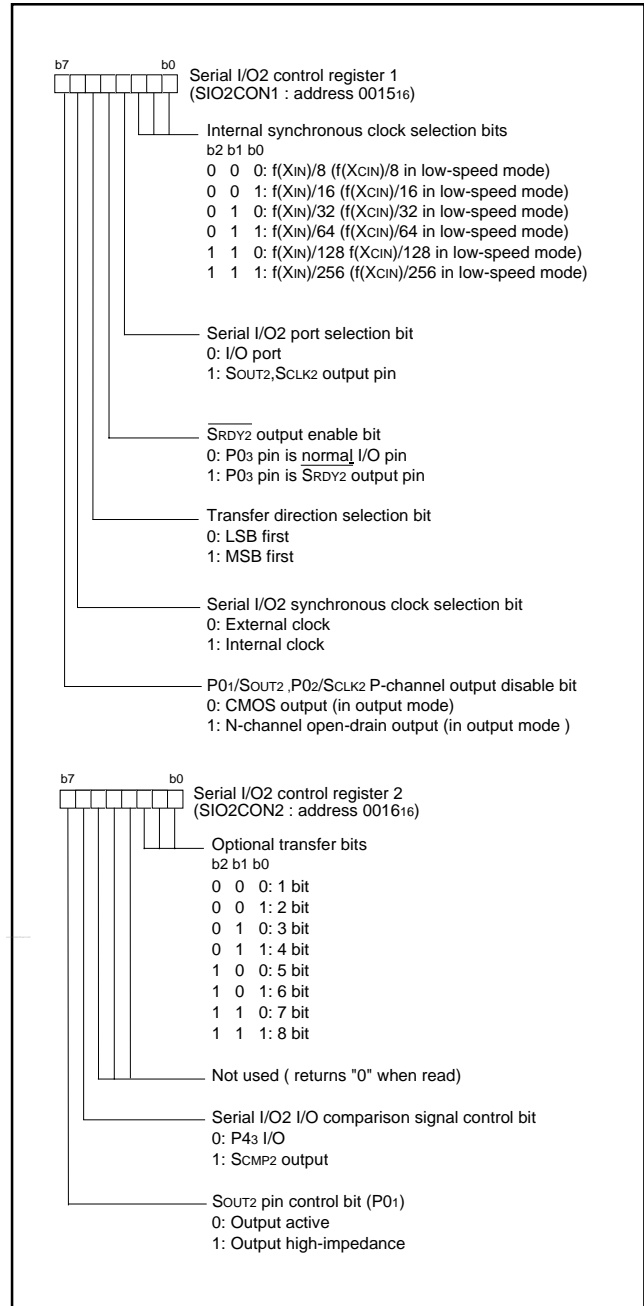


Fig. 23 Structure of Serial I/O2 control registers 1, 2

# HARDWARE

## FUNCTIONAL DESCRIPTION

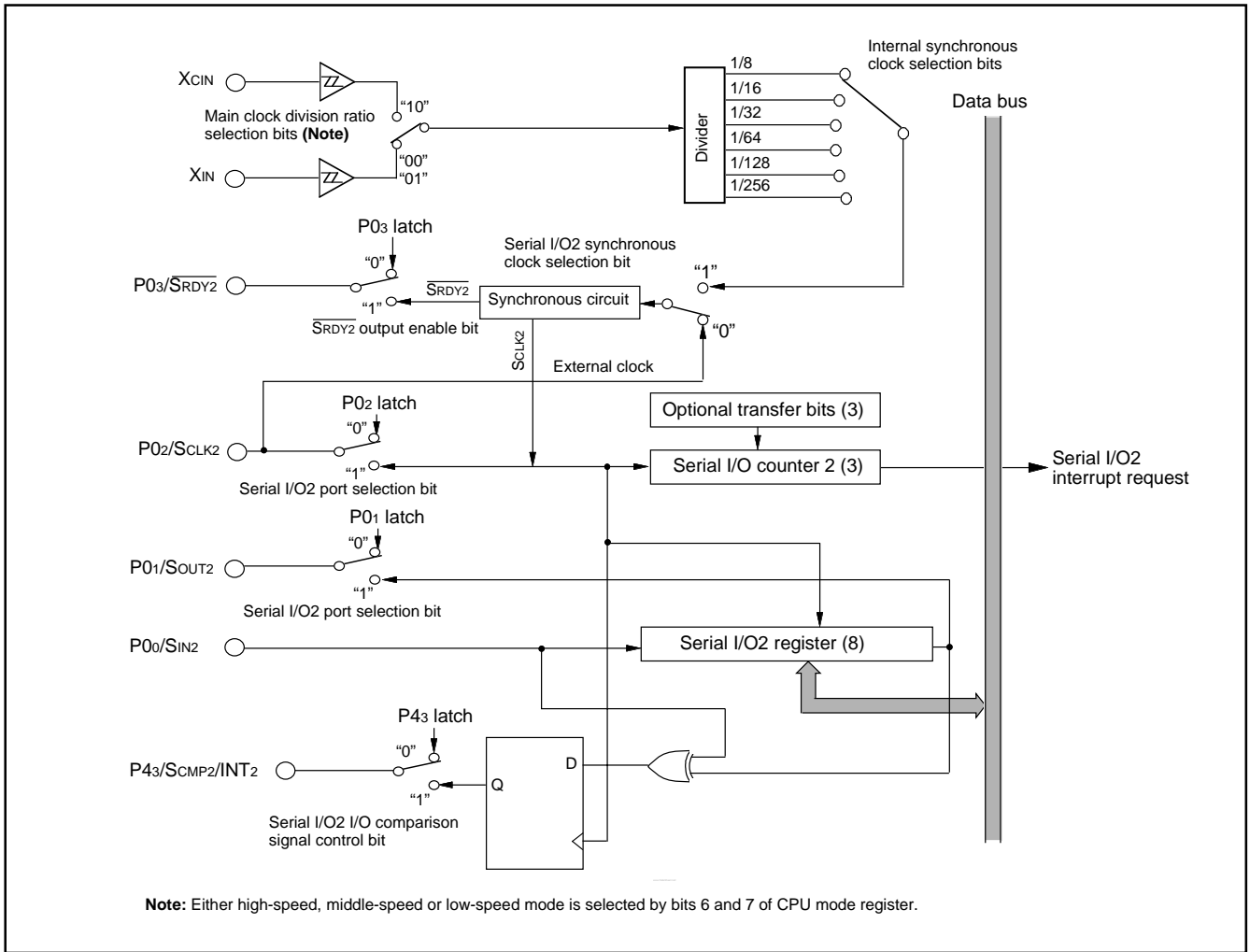


Fig. 24 Block diagram of Serial I/O2

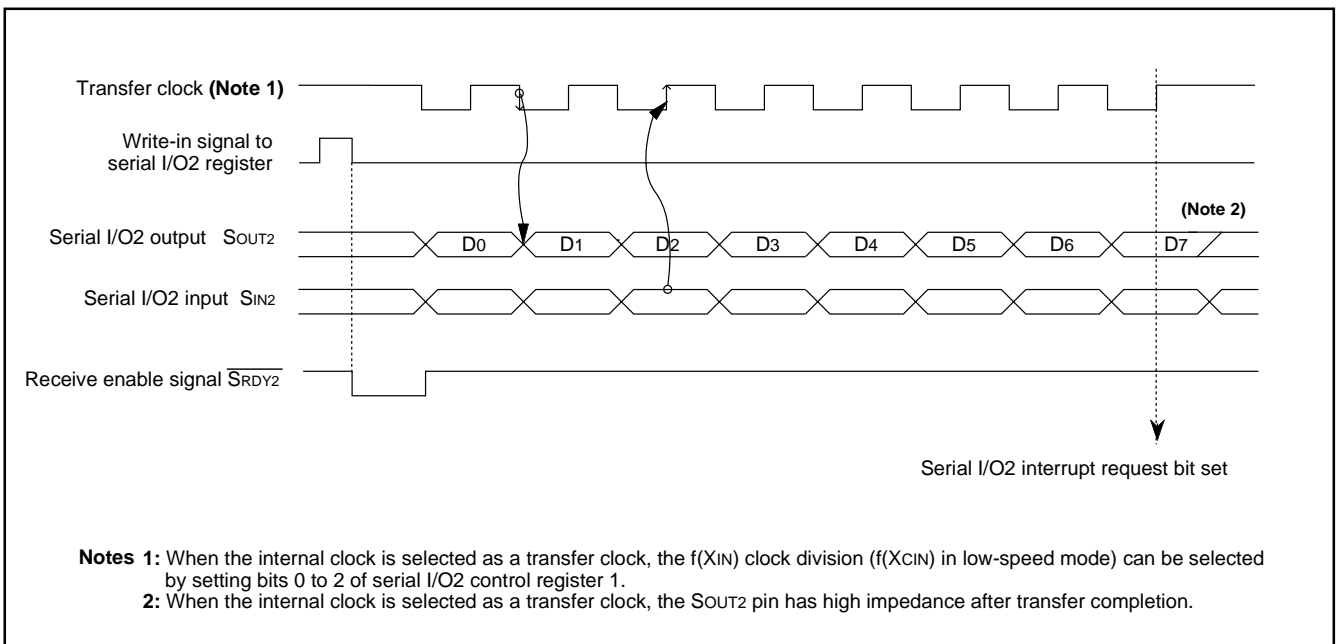


Fig. 25 Timing chart of Serial I/O2

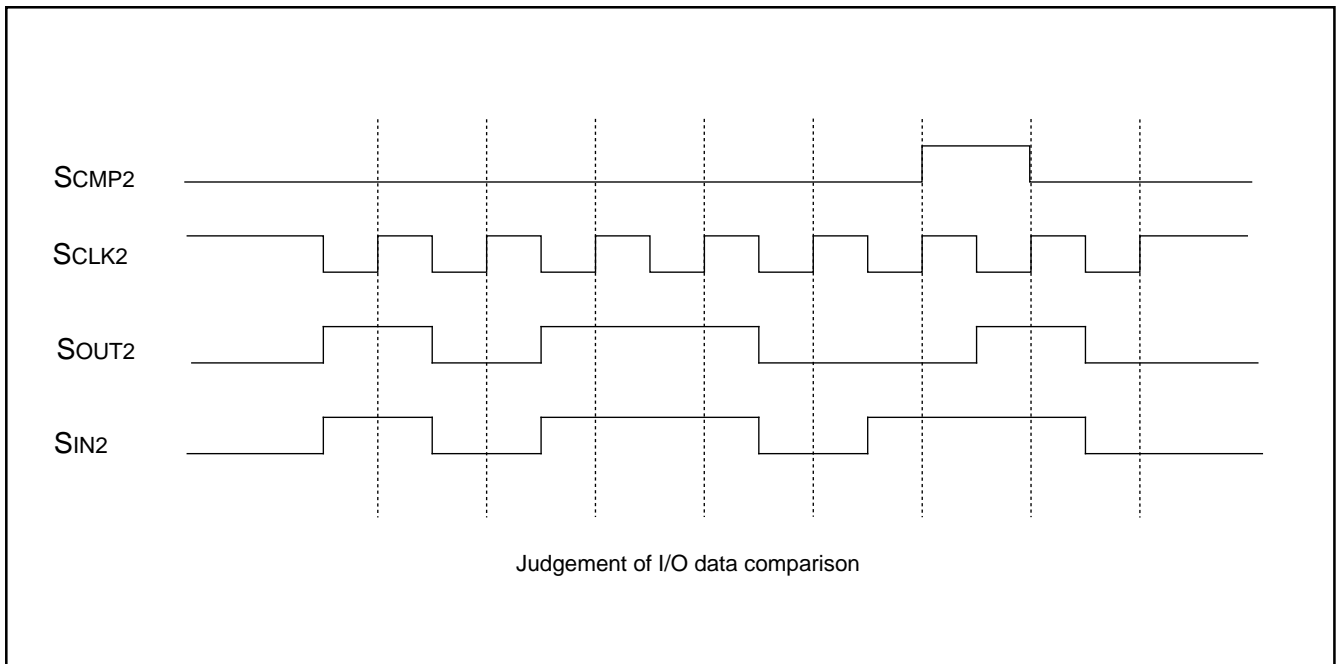


Fig. 26 SCMP2 output operation

# HARDWARE

## FUNCTIONAL DESCRIPTION

### PULSE WIDTH MODULATION (PWM)

The 3850 group (spec. H) has a PWM function with an 8-bit resolution, based on a signal that is the clock input X<sub>IN</sub> or that clock input divided by 2.

### Data Setting

The PWM output pin also functions as port P44. Set the PWM period by the PWM prescaler, and set the "H" term of output pulse by the PWM register.

If the value in the PWM prescaler is n and the value in the PWM register is m (where n = 0 to 255 and m = 0 to 255) :

$$\text{PWM period} = 255 \times (n+1) / f(X_{IN})$$

$$= 31.875 \times (n+1) \mu\text{s}$$

(when f(X<sub>IN</sub>) = 8 MHz, count source selection bit = "0")

$$\text{Output pulse "H" term} = \text{PWM period} \times m / 255$$

$$= 0.125 \times (n+1) \times m \mu\text{s}$$

(when f(X<sub>IN</sub>) = 8 MHz, count source selection bit = "0")

### PWM Operation

When bit 0 (PWM enable bit) of the PWM control register is set to "1", operation starts by initializing the PWM output circuit, and pulses are output starting at an "H".

If the PWM register or PWM prescaler is updated during PWM output, the pulses will change in the cycle after the one in which the change was made.

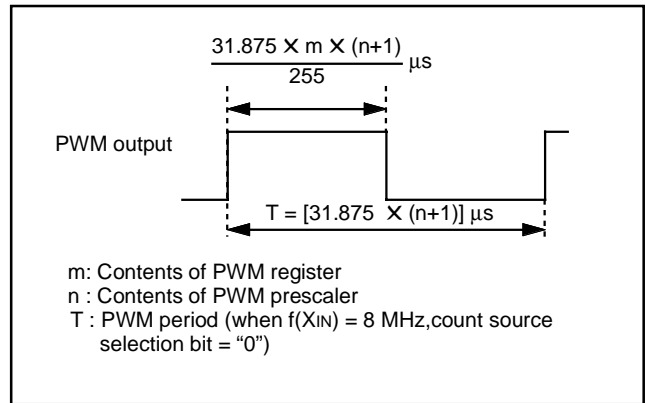


Fig. 27 Timing of PWM period

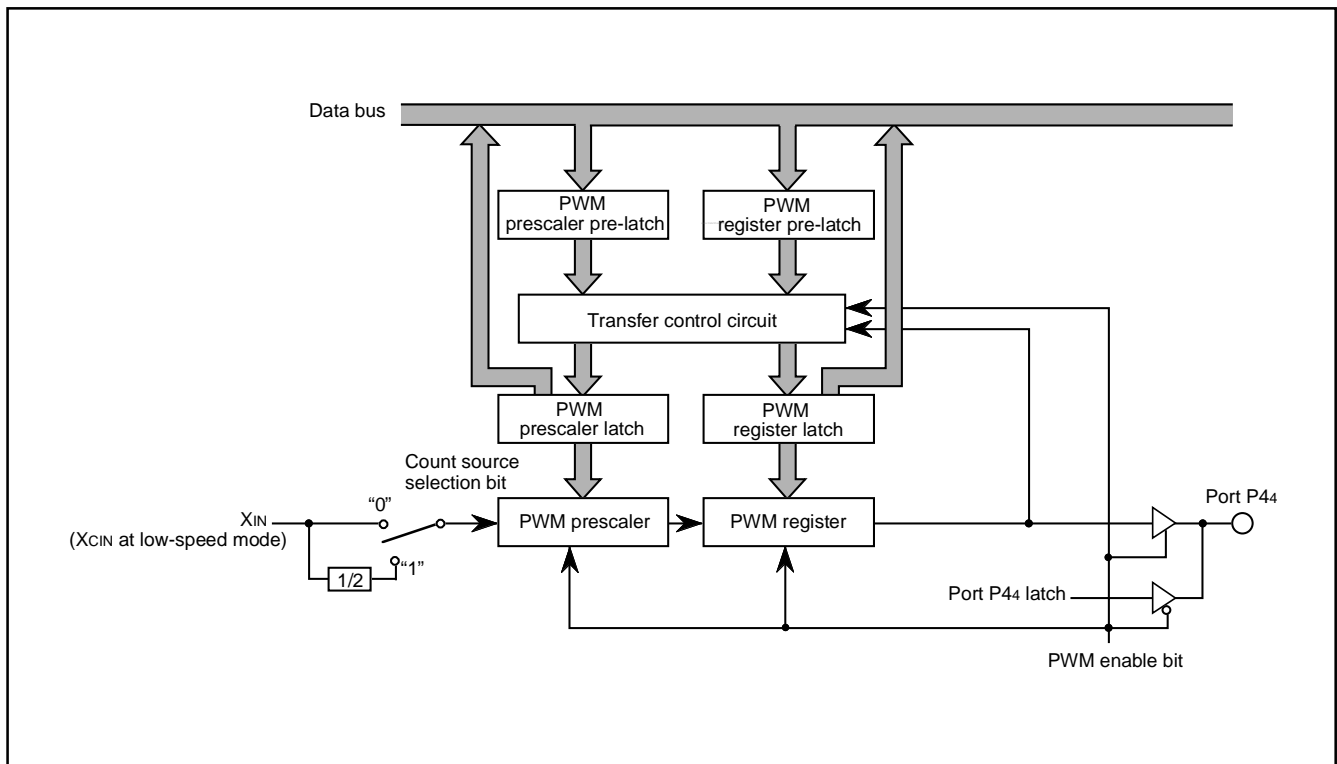


Fig. 28 Block diagram of PWM function



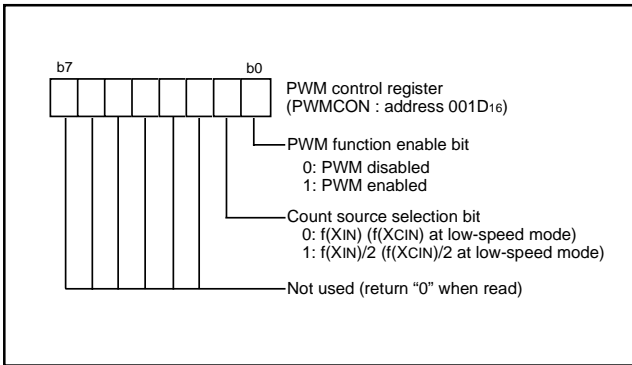


Fig. 29 Structure of PWM control register

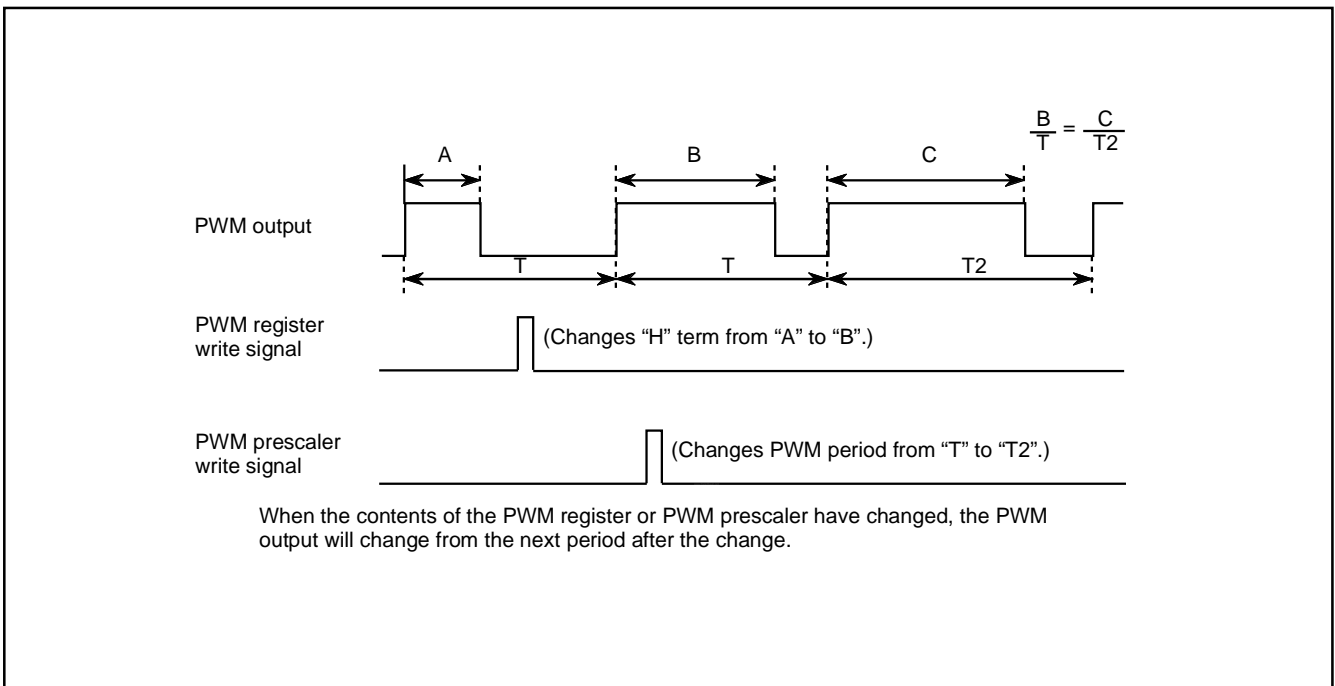


Fig. 30 PWM output timing when PWM register or PWM prescaler is changed

### ■Note

The PWM starts after the PWM function enable bit is set to enable and "L" level is output from the PWM pin. The length of this "L" level output is as follows:

$$\frac{n+1}{2 \cdot f(X_{IN})} \text{ sec} \quad (\text{Count source selection bit} = 0, \text{ where } n \text{ is the value set in the prescaler})$$

$$\frac{n+1}{f(X_{IN})} \text{ sec} \quad (\text{Count source selection bit} = 1, \text{ where } n \text{ is the value set in the prescaler})$$

# HARDWARE

## FUNCTIONAL DESCRIPTION

### A-D CONVERTER

#### [A-D Conversion Registers (ADL, ADH)] 003516, 003616

The A-D conversion registers are read-only registers that store the result of an A-D conversion. Do not read these registers during an A-D conversion.

#### [AD Control Register (ADCON)] 003416

The AD control register controls the A-D conversion process. Bits 0 to 2 select a specific analog input pin. Bit 4 indicates the completion of an A-D conversion. The value of this bit remains at "0" during an A-D conversion and changes to "1" when an A-D conversion ends. Writing "0" to this bit starts the A-D conversion.

#### Comparison Voltage Generator

The comparison voltage generator divides the voltage between AVSS and VREF into 1024 and outputs the divided voltages.

#### Channel Selector

The channel selector selects one of ports P30/AN0 to P34/AN4 and inputs the voltage to the comparator.

#### Comparator and Control Circuit

The comparator and control circuit compare an analog input voltage with the comparison voltage, and the result is stored in the A-D conversion registers. When an A-D conversion is completed, the control circuit sets the A-D conversion completion bit and the A-D interrupt request bit to "1".

Note that because the comparator consists of a capacitor coupling, set  $f(X_{IN})$  to 500 kHz or more during an A-D conversion. When the A-D converter is operated at low-speed mode,  $f(X_{IN})$  and  $f(X_{CIN})$  do not have the lower limit of frequency, because of the A-D converter has a built-in self-oscillation circuit.

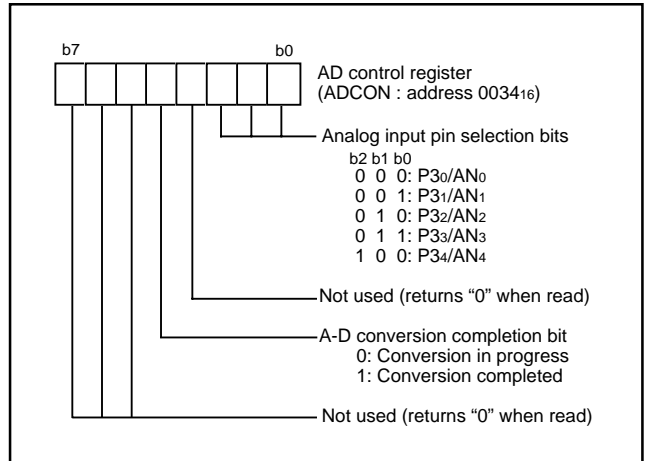


Fig. 31 Structure of AD control register

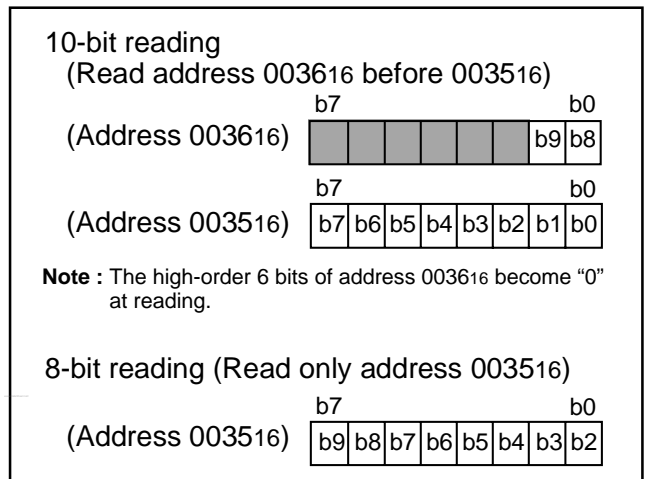


Fig. 32 Structure of A-D conversion registers

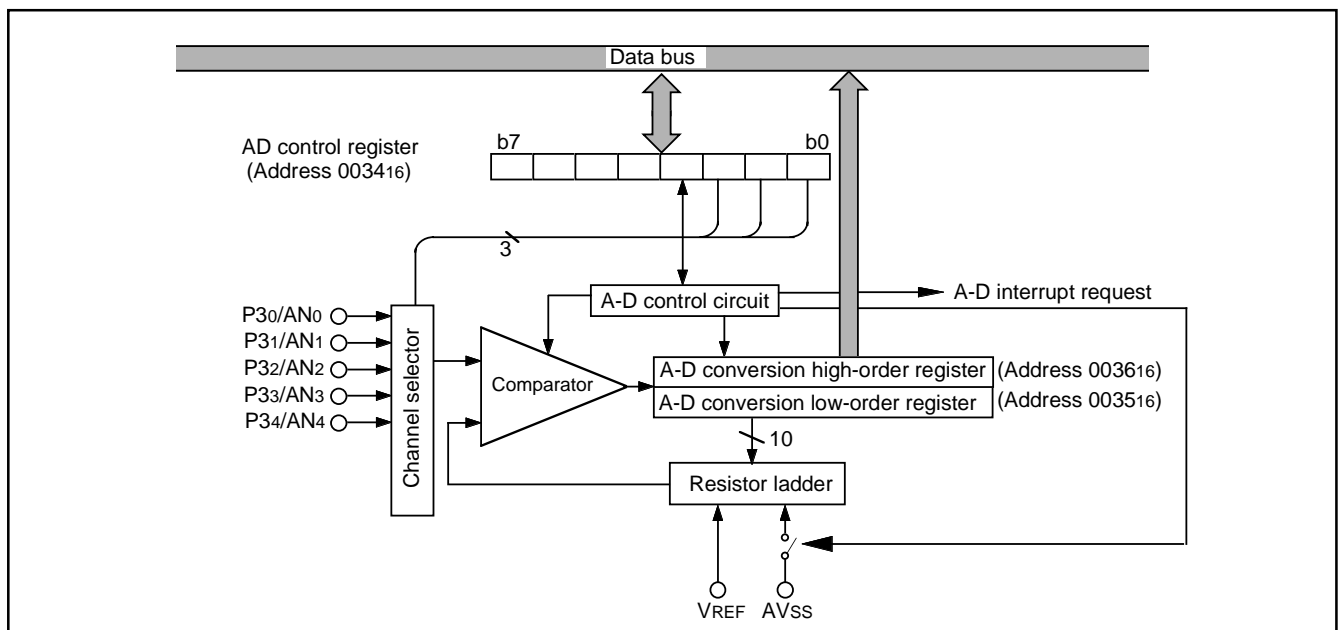


Fig. 33 Block diagram of A-D converter

### WATCHDOG TIMER

The watchdog timer gives a mean of returning to the reset status when a program cannot run on a normal loop (for example, because of a software run-away). The watchdog timer consists of an 8-bit watchdog timer L and an 8-bit watchdog timer H.

#### Standard Operation of Watchdog Timer

When any data is not written into the watchdog timer control register (address 0039<sub>16</sub>) after reset, the watchdog timer is in the stop state. The watchdog timer starts to count down by writing an optional value into the watchdog timer control register (address 0039<sub>16</sub>) and an internal reset occurs at an underflow of the watchdog timer H.

Accordingly, programming is usually performed so that writing to the watchdog timer control register (address 0039<sub>16</sub>) may be started before an underflow. When the watchdog timer control register (address 0039<sub>16</sub>) is read, the values of the high-order 6 bits of the watchdog timer H, STP instruction disable bit, and watchdog timer H count source selection bit are read.

#### Initial value of watchdog timer

At reset or writing to the watchdog timer control register (address 0039<sub>16</sub>), each watchdog timer H and L is set to "FF<sub>16</sub>".

#### Watchdog timer H count source selection bit operation

Bit 7 of the watchdog timer control register (address 0039<sub>16</sub>) permits selecting a watchdog timer H count source. When this bit is set to "0", the count source becomes the underflow signal of watchdog timer L. The detection time is set to 131.072 ms at  $f(XIN) = 8$  MHz frequency and 32.768 s at  $f(XCIN) = 32$  kHz frequency. When this bit is set to "1", the count source becomes the signal divided by 16 for  $f(XIN)$  (or  $f(XCIN)$ ). The detection time in this case is set to 512  $\mu$ s at  $f(XIN) = 8$  MHz frequency and 128 ms at  $f(XCIN) = 32$  kHz frequency. This bit is cleared to "0" after reset.

#### Operation of STP instruction disable bit

Bit 6 of the watchdog timer control register (address 0039<sub>16</sub>) permits disabling the STP instruction when the watchdog timer is in operation.

When this bit is "0", the STP instruction is enabled.

When this bit is "1", the STP instruction is disabled, once the STP instruction is executed, an internal reset occurs. When this bit is set to "1", it cannot be rewritten to "0" by program. This bit is cleared to "0" after reset.

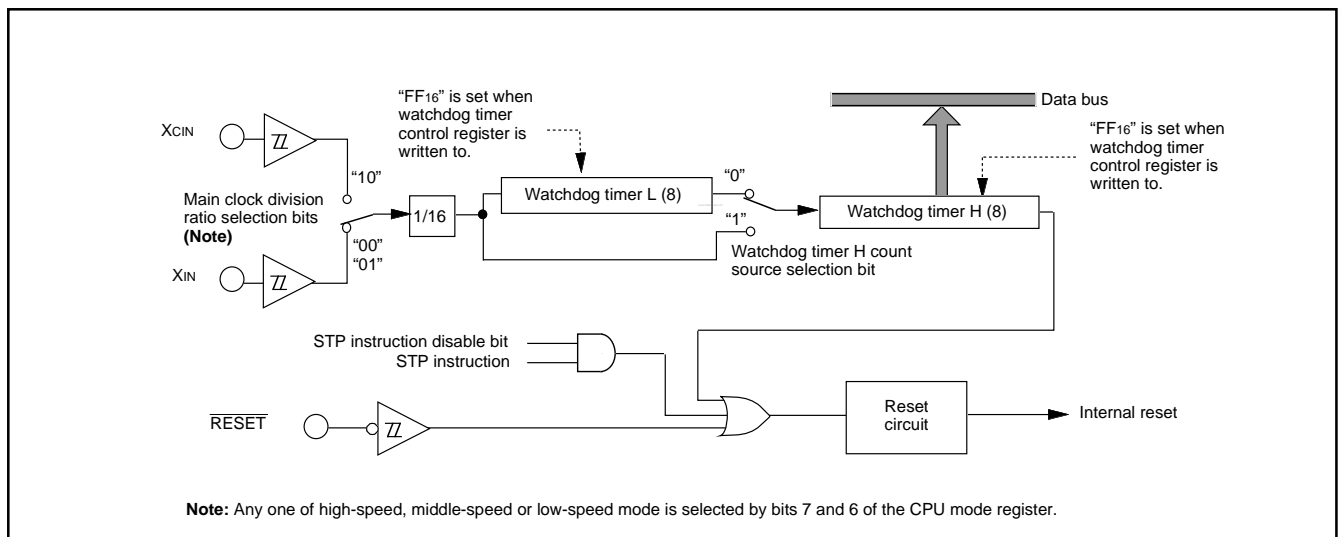


Fig. 34 Block diagram of Watchdog timer

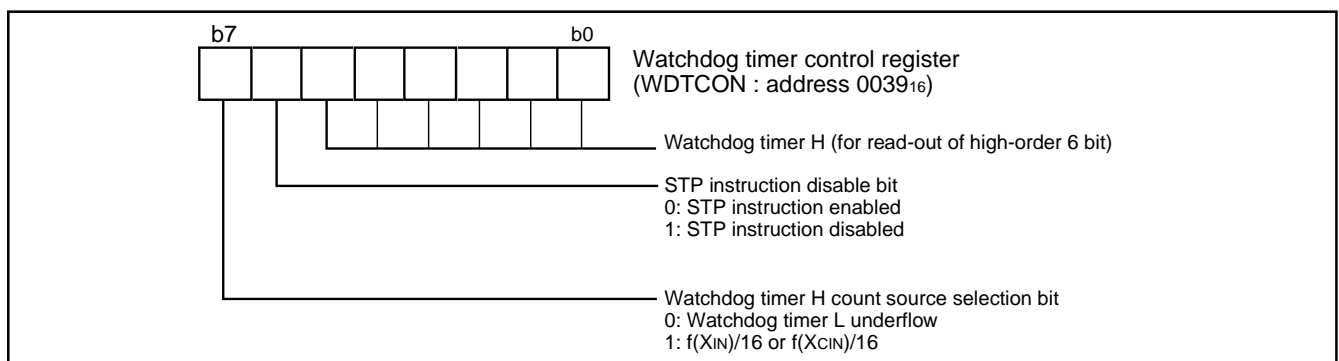


Fig. 35 Structure of Watchdog timer control register

# HARDWARE

## FUNCTIONAL DESCRIPTION

### RESET CIRCUIT

To reset the microcomputer,  $\overline{\text{RESET}}$  pin must be held at an "L" level for 20 cycles or more of  $X_{IN}$ . Then the  $\overline{\text{RESET}}$  pin is returned to an "H" level (the power source voltage must be between 2.7 V and 5.5 V, and the oscillation must be stable), reset is released. After the reset is completed, the program starts from the address contained in address  $\text{FFFD}_{16}$  (high-order byte) and address  $\text{FFFC}_{16}$  (low-order byte). Make sure that the reset input voltage is less than 0.54 V for  $V_{CC}$  of 2.7 V.

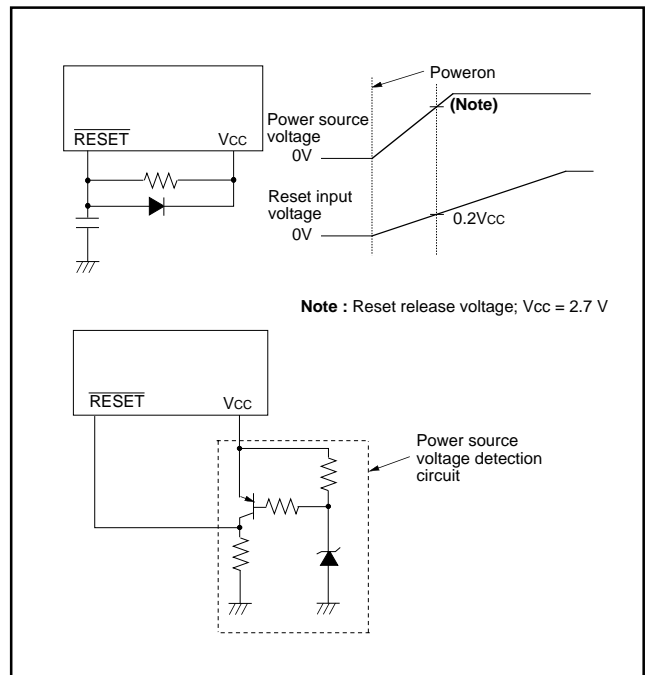


Fig. 36 Reset circuit example

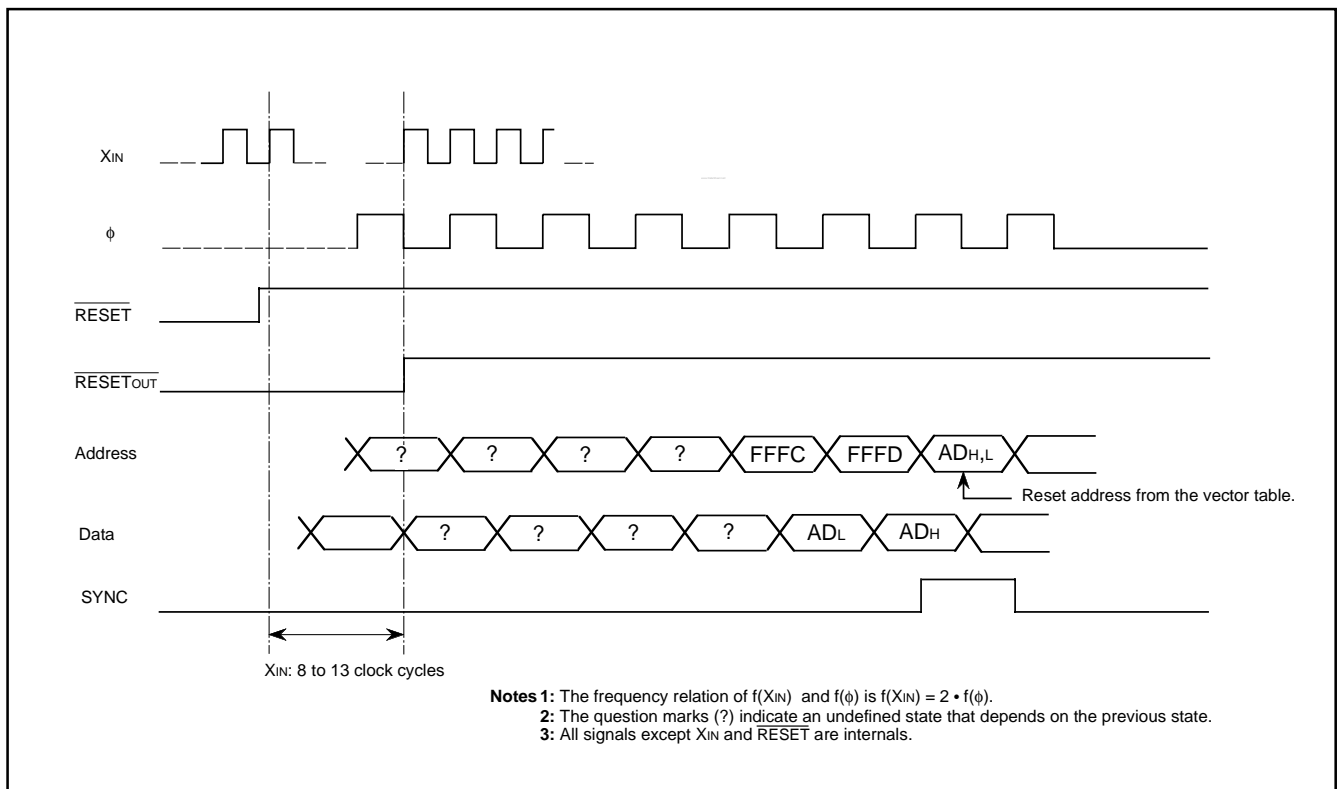


Fig. 37 Reset sequence

	Address	Register contents		Address	Register contents
(1) Port P0 (P0)	0000 <sub>16</sub>	00 <sub>16</sub>	(34) MISRG	0038 <sub>16</sub>	00 <sub>16</sub>
(2) Port P0 direction register (P0D)	0001 <sub>16</sub>	00 <sub>16</sub>	(35) Watchdog timer control register (WDTCON)	0039 <sub>16</sub>	00111111
(3) Port P1 (P1)	0002 <sub>16</sub>	00 <sub>16</sub>	(36) Interrupt edge selection register (INTEDGE)	003A <sub>16</sub>	00 <sub>16</sub>
(4) Port P1 direction register (P1D)	0003 <sub>16</sub>	00 <sub>16</sub>	(37) CPU mode register (CPUM)	003B <sub>16</sub>	01001000
(5) Port P2 (P2)	0004 <sub>16</sub>	00 <sub>16</sub>	(38) Interrupt request register 1 (IREQ1)	003C <sub>16</sub>	00 <sub>16</sub>
(6) Port P2 direction register (P2D)	0005 <sub>16</sub>	00 <sub>16</sub>	(39) Interrupt request register 2 (IREQ2)	003D <sub>16</sub>	00 <sub>16</sub>
(7) Port P3 (P3)	0006 <sub>16</sub>	00 <sub>16</sub>	(40) Interrupt control register 1 (ICON1)	003E <sub>16</sub>	00 <sub>16</sub>
(8) Port P3 direction register (P3D)	0007 <sub>16</sub>	00 <sub>16</sub>	(41) Interrupt control register 2 (ICON2)	003F <sub>16</sub>	00 <sub>16</sub>
(9) Port P4 (P4)	0008 <sub>16</sub>	00 <sub>16</sub>	(42) Processor status register	(PS)	XXXXXXXX1XX
(10) Port P4 direction register (P4D)	0009 <sub>16</sub>	00 <sub>16</sub>	(43) Program counter	(PCH)	FFFD <sub>16</sub> contents
(11) Serial I/O2 control register 1 (SIO2CON1)	0015 <sub>16</sub>	00 <sub>16</sub>		(PCL)	FFFC <sub>16</sub> contents
(12) Serial I/O2 control register 2 (SIO2CON2)	0016 <sub>16</sub>	00000111			
(13) Serial I/O2 register (SIO2)	0017 <sub>16</sub>	XXXXXXXX			
(14) Transmit/Receive buffer register (TB/RB)	0018 <sub>16</sub>	XXXXXXXX			
(15) Serial I/O1 status register (SIOSTS)	0019 <sub>16</sub>	10000000			
(16) Serial I/O1 control register (SIOCON)	001A <sub>16</sub>	00 <sub>16</sub>			
(17) UART control register (UARTCON)	001B <sub>16</sub>	11100000			
(18) Baud rate generator (BRG)	001C <sub>16</sub>	XXXXXXXX			
(19) PWM control register (PWMCON)	001D <sub>16</sub>	00 <sub>16</sub>			
(20) PWM prescaler (PREPWM)	001E <sub>16</sub>	XXXXXXXX			
(21) PWM register (PWM)	001F <sub>16</sub>	XXXXXXXX			
(22) Prescaler 12 (PRE12)	0020 <sub>16</sub>	FF <sub>16</sub>			
(23) Timer 1 (T1)	0021 <sub>16</sub>	01 <sub>16</sub>			
(24) Timer 2 (T2)	0022 <sub>16</sub>	00 <sub>16</sub>			
(25) Timer XY mode register (TM)	0023 <sub>16</sub>	00 <sub>16</sub>			
(26) Prescaler X (PREX)	0024 <sub>16</sub>	FF <sub>16</sub>			
(27) Timer X (TX)	0025 <sub>16</sub>	FF <sub>16</sub>			
(28) Prescaler Y (PREY)	0026 <sub>16</sub>	FF <sub>16</sub>			
(29) Timer Y (TY)	0027 <sub>16</sub>	FF <sub>16</sub>			
(30) Timer count source selection register (TCSS)	0028 <sub>16</sub>	00 <sub>16</sub>			
(31) A-D control register (ADCON)	0034 <sub>16</sub>	00010000			
(32) A-D conversion low-order register (ADL)	0035 <sub>16</sub>	XXXXXXXX			
(33) A-D conversion high-order register (ADH)	0036 <sub>16</sub>	000000XX			

**Note** : X : Not fixed  
 Since the initial values for other than above mentioned registers and RAM contents are indefinite at reset, they must be set.

**Fig. 38 Internal status at reset**

# HARDWARE

## FUNCTIONAL DESCRIPTION

### CLOCK GENERATING CIRCUIT

The 3850 group (spec. H) has two built-in oscillation circuits. An oscillation circuit can be formed by connecting a resonator between  $X_{IN}$  and  $X_{OUT}$  ( $X_{CIN}$  and  $X_{COUT}$ ). Use the circuit constants in accordance with the resonator manufacturer's recommended values. No external resistor is needed between  $X_{IN}$  and  $X_{OUT}$  since a feed-back resistor exists on-chip. However, an external feed-back resistor is needed between  $X_{CIN}$  and  $X_{COUT}$ . Immediately after power on, only the  $X_{IN}$  oscillation circuit starts oscillating, and  $X_{CIN}$  and  $X_{COUT}$  pins function as I/O ports.

### Frequency Control

#### (1) Middle-speed mode

The internal clock  $\phi$  is the frequency of  $X_{IN}$  divided by 8. After reset is released, this mode is selected.

#### (2) High-speed mode

The internal clock  $\phi$  is half the frequency of  $X_{IN}$ .

#### (3) Low-speed mode

The internal clock  $\phi$  is half the frequency of  $X_{CIN}$ .

#### ■Note

If you switch the mode between middle/high-speed and low-speed, stabilize both  $X_{IN}$  and  $X_{CIN}$  oscillations. The sufficient time is required for the sub-clock to stabilize, especially immediately after power on and at returning from the stop mode. When switching the mode between middle/high-speed and low-speed, set the frequency on condition that  $f(X_{IN}) > 3 \cdot f(X_{CIN})$ .

#### (4) Low power dissipation mode

The low power consumption operation can be realized by stopping the main clock  $X_{IN}$  in low-speed mode. To stop the main clock, set bit 5 of the CPU mode register to "1". When the main clock  $X_{IN}$  is restarted (by setting the main clock stop bit to "0"), set sufficient time for oscillation to stabilize.

The sub-clock  $X_{CIN}$ - $X_{COUT}$  oscillating circuit can not directly input clocks that are generated externally. Accordingly, make sure to cause an external resonator to oscillate.

### Oscillation Control

#### (1) Stop mode

If the STP instruction is executed, the internal clock  $\phi$  stops at an "H" level, and  $X_{IN}$  and  $X_{CIN}$  oscillation stops. When the oscillation stabilizing time set after STP instruction released bit is "0", the prescaler 12 is set to "FF16" and timer 1 is set to "0116". When the oscillation stabilizing time set after STP instruction released bit is "1", set the sufficient time for oscillation of used oscillator to stabilize since nothing is set to the prescaler 12 and timer 1.

Either  $X_{IN}$  or  $X_{CIN}$  divided by 16 is input to the prescaler 12 as count source. Oscillator restarts when an external interrupt is received, but the internal clock  $\phi$  is not supplied to the CPU (remains at "H") until timer 1 underflows. The internal clock  $\phi$  is supplied for the first time, when timer 1 underflows. This ensures time for the clock oscillation using the ceramic resonators to be stabilized. When the oscillator is restarted by reset, apply "L" level to the  $\overline{RESET}$  pin until the oscillation is stable since a wait time will not be generated.

#### (2) Wait mode

If the WIT instruction is executed, the internal clock  $\phi$  stops at an "H" level, but the oscillator does not stop. The internal clock  $\phi$  restarts at reset or when an interrupt is received. Since the oscillator does not stop, normal operation can be started immediately after the clock is restarted.

To ensure that the interrupts will be received to release the STP or WIT state, their interrupt enable bits must be set to "1" before executing of the STP or WIT instruction.

When releasing the STP state, the prescaler 12 and timer 1 will start counting the clock  $X_{IN}$  divided by 16. Accordingly, set the timer 1 interrupt enable bit to "0" before executing the STP instruction.

#### ■Note

When using the oscillation stabilizing time set after STP instruction released bit set to "1", evaluate time to stabilize oscillation of the used oscillator and set the value to the timer 1 and prescaler 12.

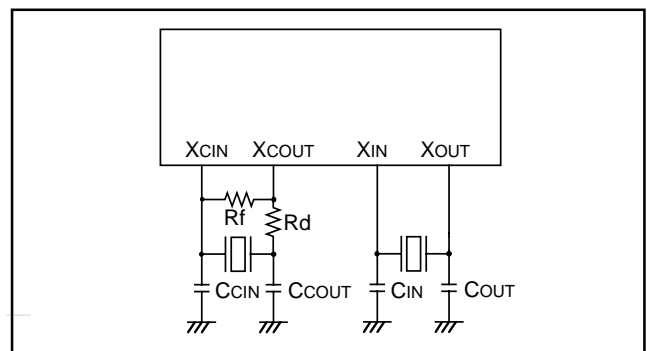


Fig. 39 Ceramic resonator circuit

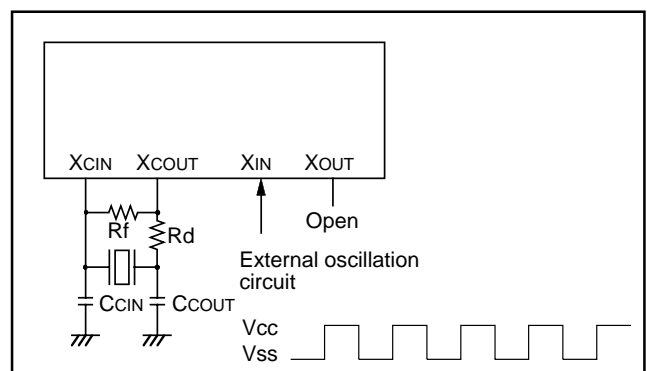


Fig. 40 External clock input circuit

### [MISRG (MISRG)] 003816

MISRG consists of three control bits (bits 1 to 3) for middle-speed mode automatic switch and one control bit (bit 0) for oscillation stabilizing time set after STP instruction released.

By setting the middle-speed mode automatic switch start bit to "1" while operating in the low-speed mode and setting the middle-speed mode automatic switch set bit to "1", XIN oscillation automatically starts and the mode is automatically switched to the middle-speed mode.

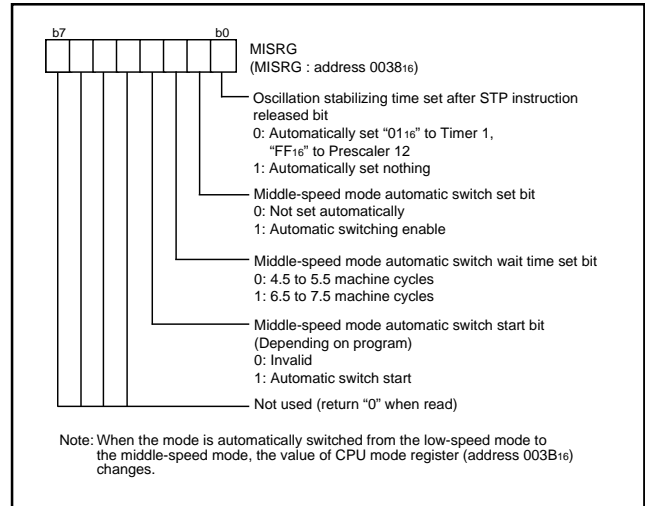


Fig. 41 Structure of MISRG

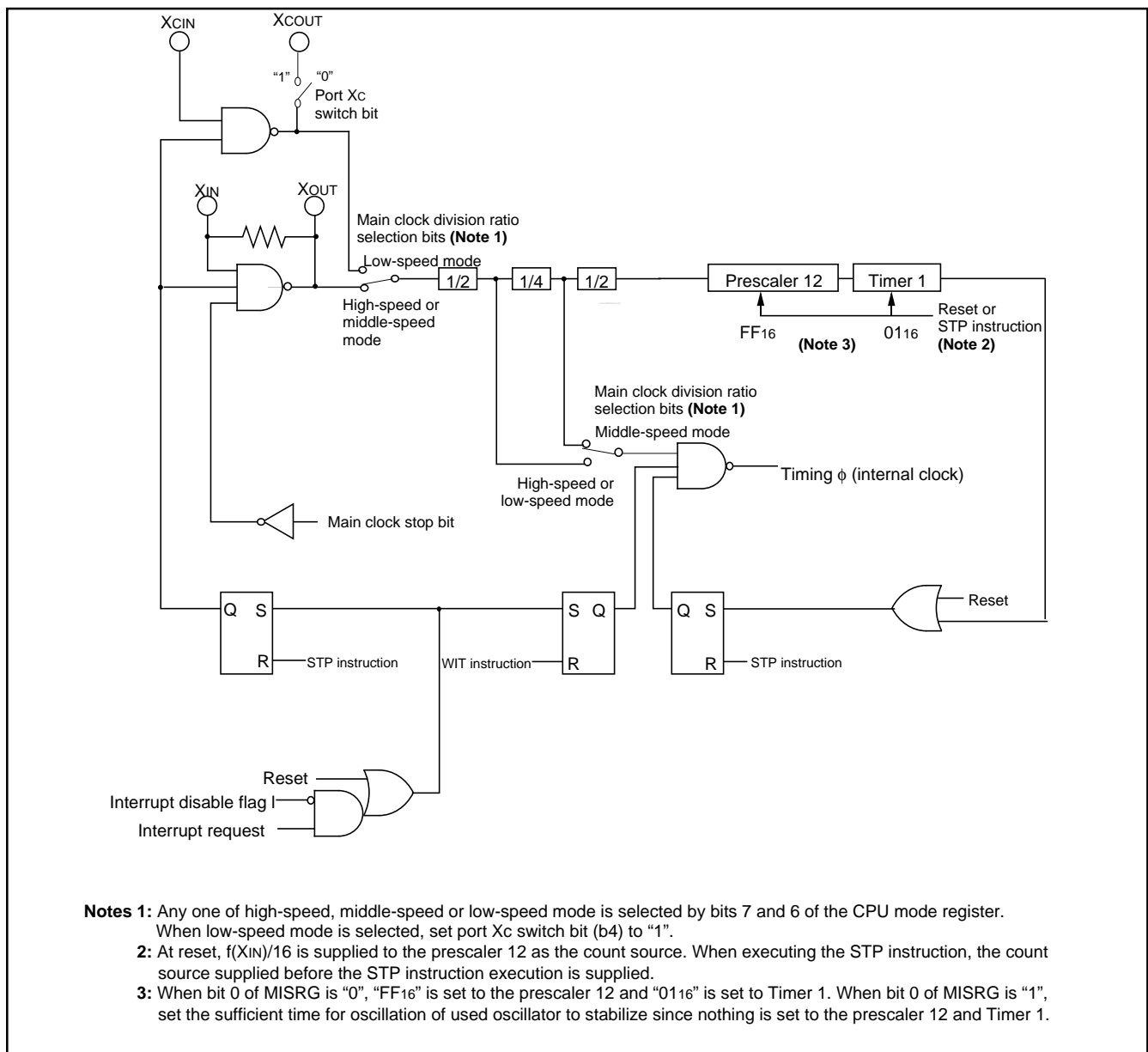
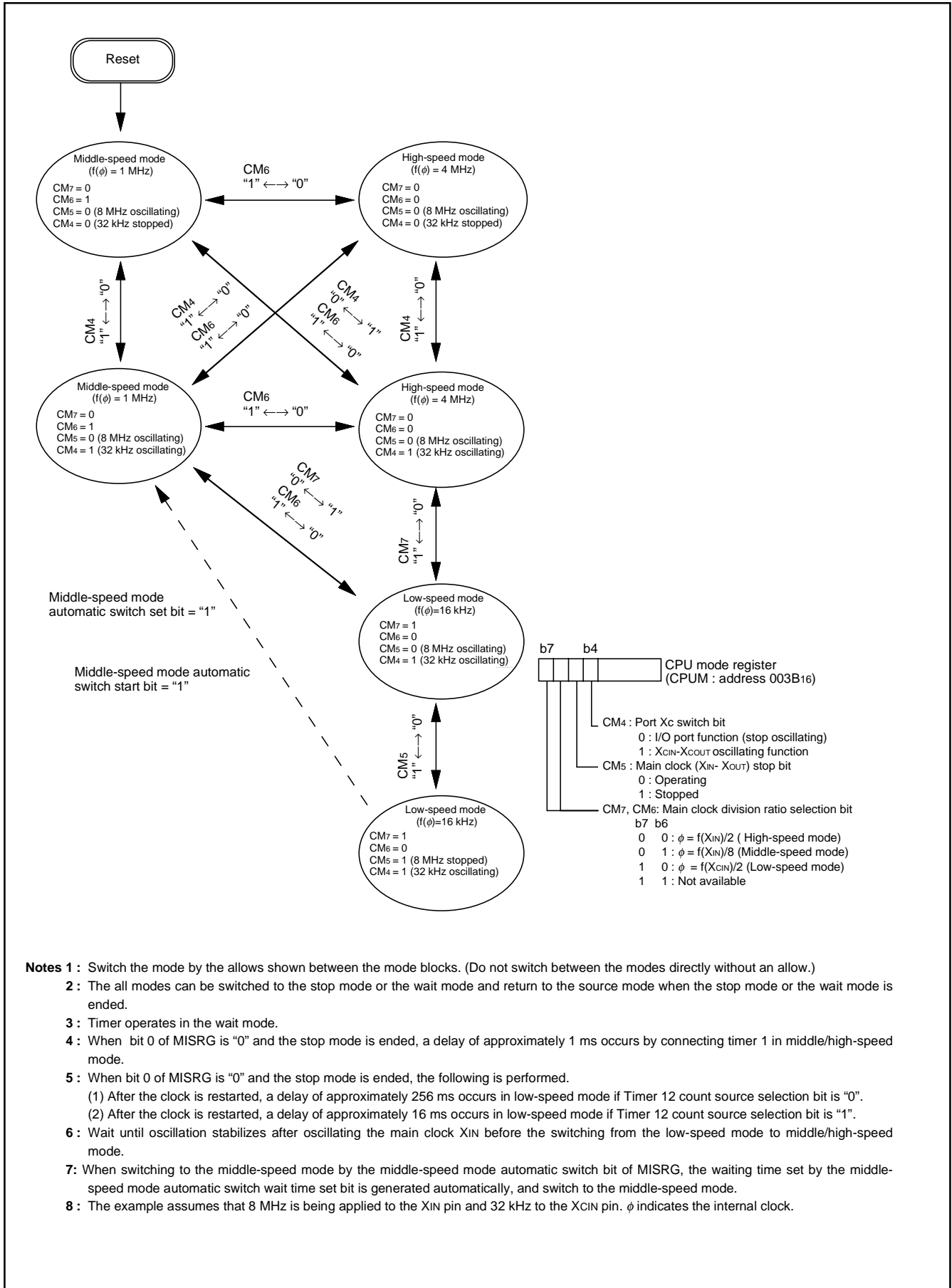


Fig. 42 System clock generating circuit block diagram (Single-chip mode)

# HARDWARE

## FUNCTIONAL DESCRIPTION



- Notes 1 :** Switch the mode by the allows shown between the mode blocks. (Do not switch between the modes directly without an allow.)
- 2 :** The all modes can be switched to the stop mode or the wait mode and return to the source mode when the stop mode or the wait mode is ended.
- 3 :** Timer operates in the wait mode.
- 4 :** When bit 0 of MISRG is "0" and the stop mode is ended, a delay of approximately 1 ms occurs by connecting timer 1 in middle/high-speed mode.
- 5 :** When bit 0 of MISRG is "0" and the stop mode is ended, the following is performed.  
 (1) After the clock is restarted, a delay of approximately 256 ms occurs in low-speed mode if Timer 12 count source selection bit is "0".  
 (2) After the clock is restarted, a delay of approximately 16 ms occurs in low-speed mode if Timer 12 count source selection bit is "1".
- 6 :** Wait until oscillation stabilizes after oscillating the main clock XIN before the switching from the low-speed mode to middle/high-speed mode.
- 7 :** When switching to the middle-speed mode by the middle-speed mode automatic switch bit of MISRG, the waiting time set by the middle-speed mode automatic switch wait time set bit is generated automatically, and switch to the middle-speed mode.
- 8 :** The example assumes that 8 MHz is being applied to the XIN pin and 32 kHz to the XCIN pin.  $\phi$  indicates the internal clock.

Fig. 43 State transitions of system clock



### FLASH MEMORY VERSION

#### Summary

Table 9 shows the summary of the M38507F8 (flash memory version).

**Table 9 Summary of M38507F8 (flash memory version)**

Item		Specification
Power source voltage		V <sub>CC</sub> = 2.7–5.5 V ( <b>Note 1</b> ) V <sub>CC</sub> = 2.7–3.6 V ( <b>Note 2</b> )
Program/Erase V <sub>PP</sub> voltage		4.5–5.5 V, f(X <sub>IN</sub> ) = 8 MHz
Flash memory mode		3 modes (Parallel I/O mode, Standard serial I/O mode, CPU rewrite mode)
Erase block division	User ROM area	1 block (32 Kbytes)
	Boot ROM area	1 block (4 Kbytes) ( <b>Note 3</b> )
Program method		Byte program
Erase method		Batch erasing
Program/Erase control method		Program/Erase control by software command
Number of commands		6 commands
Number of program/Erase times		100 times
ROM code protection		Available in parallel I/O mode, and standard serial I/O mode

**Notes 1:** The power source voltage must be V<sub>CC</sub> = 4.5–5.5 V at program and erase operation.

**2:** The power source voltage can be V<sub>CC</sub> = 3.0–3.6 V also at program and erase operation.

**3:** The Boot ROM area has had a standard serial I/O mode control program stored in it when shipped from the factory. This Boot ROM area can be rewritten in only parallel I/O mode.

# HARDWARE

## FUNCTIONAL DESCRIPTION

### Flash Memory Mode

The M38507F8 (flash memory version) has an internal new DINOR (Divided bit line NOR) flash memory that can be rewritten with a single power source when VCC is 5 V, and 2 power sources when VCC is 3.3-5.0 V.

For this flash memory, three flash memory modes are available in which to read, program, and erase: parallel I/O and standard serial I/O modes in which the flash memory can be manipulated using a programmer and a CPU rewrite mode in which the flash memory can be manipulated by the Central Processing Unit (CPU). Each mode is detailed in the pages to follow.

The flash memory of the M38507F8 is divided into User ROM area and Boot ROM area as shown in Figure 44.

In addition to the ordinary user ROM area to store a microcomputer operation control program, the flash memory has a Boot ROM area that is used to store a program to control rewriting in CPU rewrite and standard serial I/O modes. This Boot ROM area has had a standard serial I/O mode control program stored in it when shipped from the factory. However, the user can write a rewrite control program in this area that suits the user's application system. This Boot ROM area can be rewritten in only parallel I/O mode.

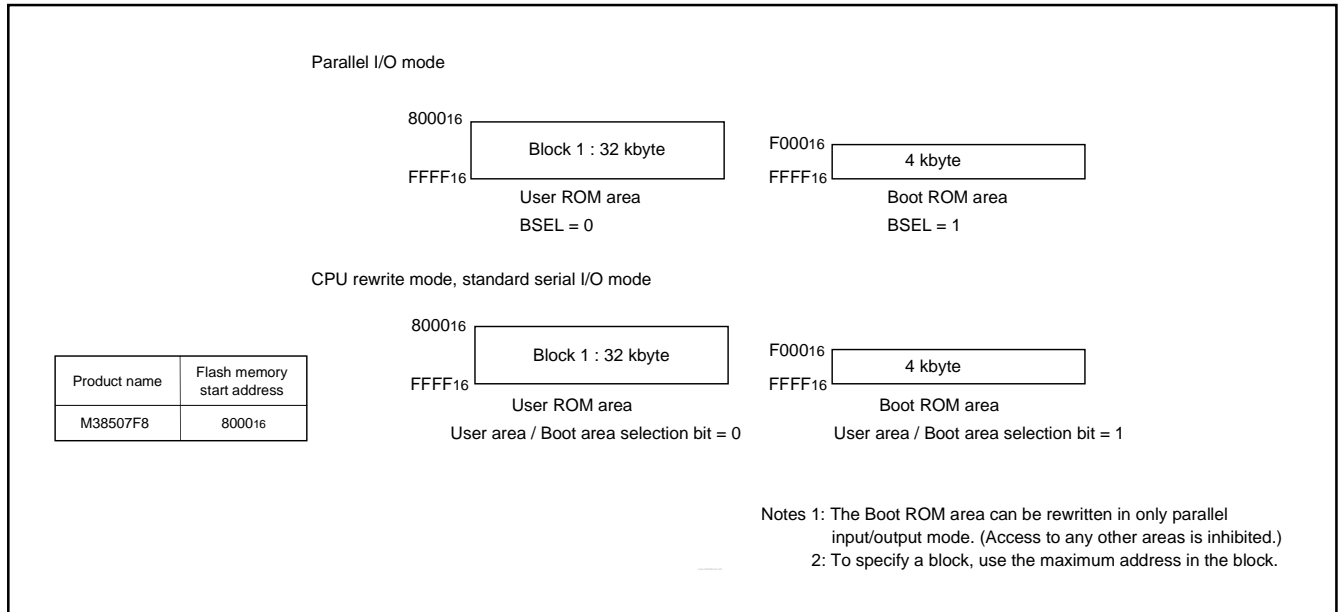


Fig. 44 Block diagram of flash memory version

### CPU Rewrite Mode

In CPU rewrite mode, the on-chip flash memory can be operated on (read, program, or erase) under control of the Central Processing Unit (CPU).

In CPU rewrite mode, only the user ROM area shown in Figure 44 can be rewritten; the Boot ROM area cannot be rewritten. Make sure the program and block erase commands are issued for only the user ROM area and each block area.

The control program for CPU rewrite mode can be stored in either user ROM or Boot ROM area. In the CPU rewrite mode, because the flash memory cannot be read from the CPU, the rewrite control program must be transferred to internal RAM area before it can be executed.

### Microcomputer Mode and Boot Mode

The control program for CPU rewrite mode must be written into the user ROM or Boot ROM area in parallel I/O mode beforehand. (If the control program is written into the Boot ROM area, the standard serial I/O mode becomes unusable.)

See Figure 44 for details about the Boot ROM area.

Normal microcomputer mode is entered when the microcomputer is reset with pulling CNVSS pin low. In this case, the CPU starts operating using the control program in the user ROM area.

When the microcomputer is reset by pulling the P41/INT0 pin high, the CNVSS pin high, the CPU starts operating using the control program in the Boot ROM area (program start address is FFFC16, FFFD16 fixation). This mode is called the "boot" mode.

### Block Address

Block addresses refer to the maximum address of each block. These addresses are used in the block erase command. In case of the M38507F8, it has only one block.

### Outline Performance (CPU Rewrite Mode)

In the CPU rewrite mode, the CPU erases, programs and reads the internal flash memory as instructed by software commands. This rewrite control program must be transferred to internal RAM before it can be executed.

The CPU rewrite mode is accessed by applying  $5V \pm 10\%$  to the CNVSS pin and writing "1" for the CPU rewrite mode select bit (bit 1 in address 0FFE16). Software commands are accepted once the mode is accessed.

Use software commands to control program and erase operations. Whether a program or erase operation has terminated normally or in error can be verified by reading the status register.

Figure 45 shows the flash memory control register.

Bit 0 is the RY/B $\bar{Y}$  status flag used exclusively to read the operating status of the flash memory. During programming and erase operations, it is "0". Otherwise, it is "1".

Bit 1 is the CPU rewrite mode select bit. When this bit is set to "1" and  $5V \pm 10\%$  are applied to the CNVSS pin, the M38507F8 accesses the CPU rewrite mode. Software commands are accepted once the mode is accessed. In CPU rewrite mode, the CPU becomes unable to access the internal flash memory directly. Therefore, use the control program in RAM for write to bit 1. To set this bit to "1", it is necessary to write "0" and then write "1" in succession. The bit can be set to "0" by only writing a "0".

Bit 2 is the CPU rewrite mode entry flag. This bit can be read to check whether the CPU rewrite mode has been entered or not.

Bit 3 is the flash memory reset bit used to reset the control circuit of the internal flash memory. This bit is used when exiting CPU rewrite mode and when flash memory access has failed. When the CPU rewrite mode select bit is "1", writing "1" for this bit resets the control circuit. To release the reset, it is necessary to set this bit to "0".

Bit 4 is the User area/Boot area selection bit. When this bit is set to "1", Boot ROM area is accessed, and CPU rewrite mode in Boot ROM area is available. In boot mode, this bit is set "1" automatically. Operation of this bit must be in RAM area.

Figure 46 shows a flowchart for setting/releasing the CPU rewrite mode.

### Precautions on CPU Rewrite Mode

Described below are the precautions to be observed when rewriting the flash memory in CPU rewrite mode.

#### (1) Operation speed

During CPU rewrite mode, set the internal clock frequency 4MHz or less using the main clock division ratio selection bits (bit 6, 7 at 003B16).

#### (2) Instructions inhibited against use

The instructions which refer to the internal data of the flash memory cannot be used during CPU rewrite mode.

#### (3) Interrupts inhibited against use

The interrupts cannot be used during CPU rewrite mode because they refer to the internal data of the flash memory.

#### (4) Watchdog timer

In case of the watchdog timer has been running already, the internal reset generated by watchdog timer underflow does not happen, because of watchdog timer is always clearing during program or erase operation.

#### (5) Reset

Reset is always valid. In case of CNVSS = H when reset is released, boot mode is active. So the program starts from the address contained in address FFFC16 and FFFD16 in boot ROM area.

# HARDWARE

## FUNCTIONAL DESCRIPTION

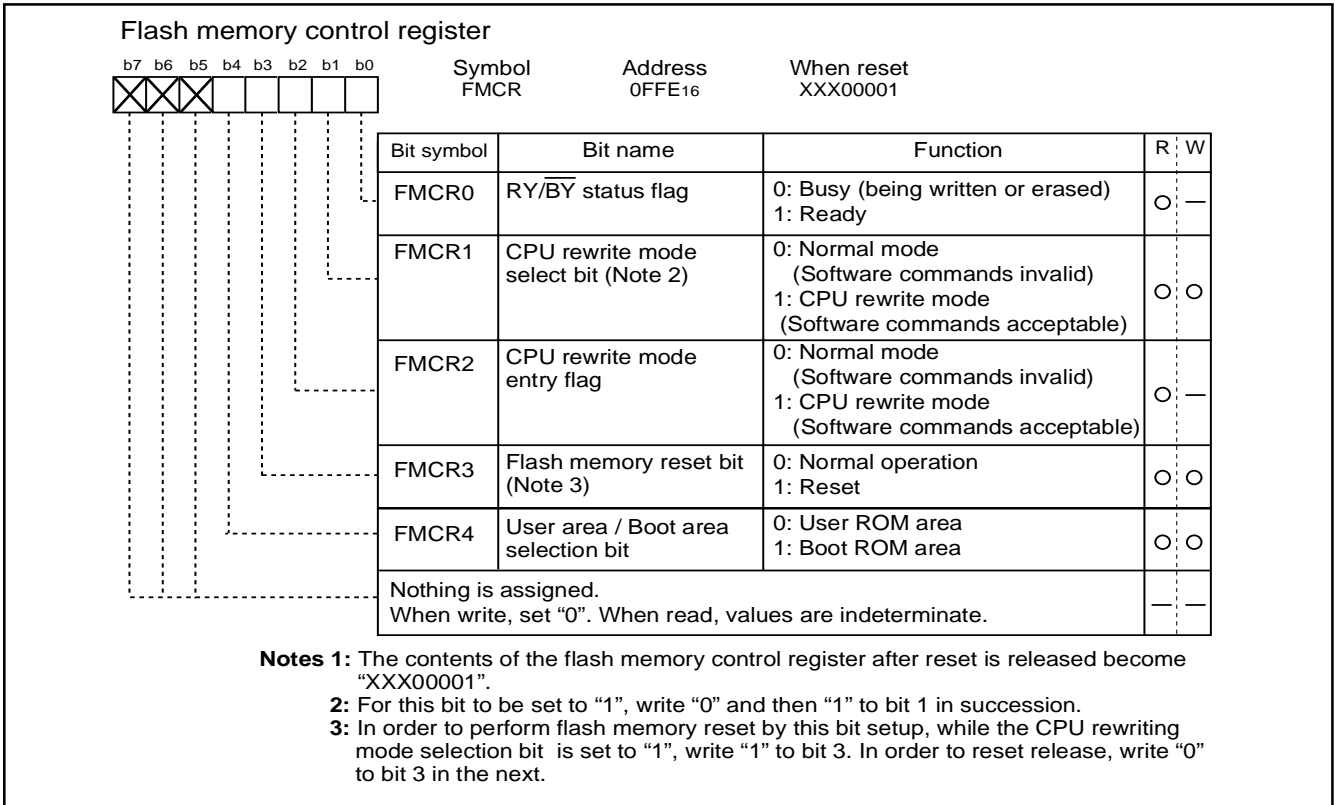


Fig. 45 Flash memory control registers

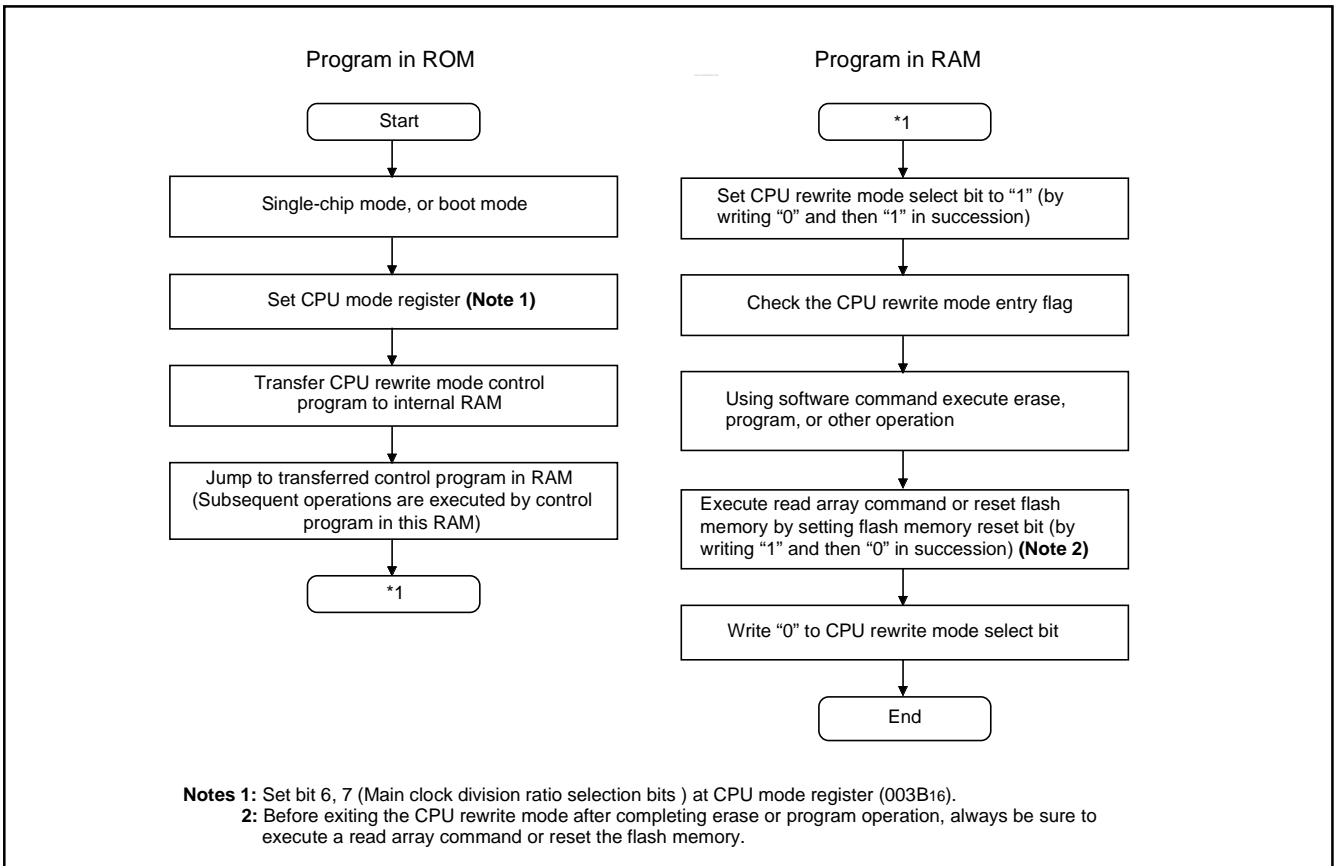


Fig. 46 CPU rewrite mode set/reset flowchart

### Software Commands

Table 10 lists the software commands.

After setting the CPU rewrite mode select bit to "1", write a software command to specify an erase or program operation.

The content of each software command is explained below.

#### Read Array Command (FF<sub>16</sub>)

The read array mode is entered by writing the command code "FF<sub>16</sub>" in the first bus cycle. When an address to be read is input in one of the bus cycles that follow, the content of the specified address is read out at the data bus (D<sub>0</sub>–D<sub>7</sub>).

The read array mode is retained intact until another command is written. And after power on and after recover from deep power down mode, this mode is selected also.

#### Read Status Register Command (70<sub>16</sub>)

When the command code "70<sub>16</sub>" is written in the first bus cycle, the content of the status register is read out at the data bus (D<sub>0</sub>–D<sub>7</sub>) by a read in the second bus cycle.

The status register is explained in the next section.

#### Clear Status Register Command (50<sub>16</sub>)

This command is used to clear the bits SR1,SR4 and SR5 of the status register after they have been set. These bits indicate that operation has ended in an error. To use this command, write the command code "50<sub>16</sub>" in the first bus cycle.

**Table 10 List of software commands (CPU rewrite mode)**

Command	Cycle number	First bus cycle			Second bus cycle		
		Mode	Address	Data (D <sub>0</sub> to D <sub>7</sub> )	Mode	Address	Data (D <sub>0</sub> to D <sub>7</sub> )
Read array	1	Write	X (Note 4)	FF <sub>16</sub>			
Read status register	2	Write	X	70 <sub>16</sub>	Read	X	SRD (Note 1)
Clear status register	1	Write	X	50 <sub>16</sub>			
Program	2	Write	X	40 <sub>16</sub>	Write	WA (Note 2)	WD (Note 2)
Erase all block	2	Write	X	20 <sub>16</sub>	Write	X	20 <sub>16</sub>
Block erase	2	Write	X	20 <sub>16</sub>	Write	BA (Note 3)	D0 <sub>16</sub>

**Notes 1:** SRD = Status Register Data

**2:** WA = Write Address, WD = Write Data

**3:** BA = Block Address (Enter the maximum address of each block.)

**4:** X denotes a given address in the user ROM area .

# HARDWARE

## FUNCTIONAL DESCRIPTION

### Program Command (40<sub>16</sub>)

Program operation starts when the command code “40<sub>16</sub>” is written in the first bus cycle. Then, if the address and data to program are written in the 2nd bus cycle, program operation (data programming and verification) will start.

Whether the write operation is completed can be confirmed by reading the status register or the RY/ $\overline{\text{BY}}$  status flag. When the program starts, the read status register mode is accessed automatically and the content of the status register is read into the data bus (D0–D7). The status register bit 7 (SR7) is set to “0” at the same time the write operation starts and is returned to “1” upon completion of the write operation. In this case, the read status register mode remains active until the read array command (FF<sub>16</sub>) is written.

The RY/ $\overline{\text{BY}}$  status flag is “0” during write operation and “1” when the write operation is completed as is the status register bit 7.

At program end, program results can be checked by reading the status register.

### Erase All Blocks Command (20<sub>16</sub>/20<sub>16</sub>)

By writing the command code “20<sub>16</sub>” in the first bus cycle and the confirmation command code “20<sub>16</sub>” in the second bus cycle that follows, the system starts erase all blocks( erase and erase verify).

Whether the erase all blocks command is terminated can be confirmed by reading the status register or the RY/ $\overline{\text{BY}}$  status flag. When the erase all blocks operation starts, the read status register mode is accessed automatically and the content of the status register can be read out. The status register bit 7 (SR7) is set to “0” at the same time the erase operation starts and is returned to “1” upon completion of the erase operation. In this case, the read status register mode remains active until the read array command (FF<sub>16</sub>) is written.

The RY/ $\overline{\text{BY}}$  status flag is “0” during erase operation and “1” when the erase operation is completed as is the status register bit 7.

At erase all blocks end, erase results can be checked by reading the status register. For details, refer to the section where the status register is detailed.

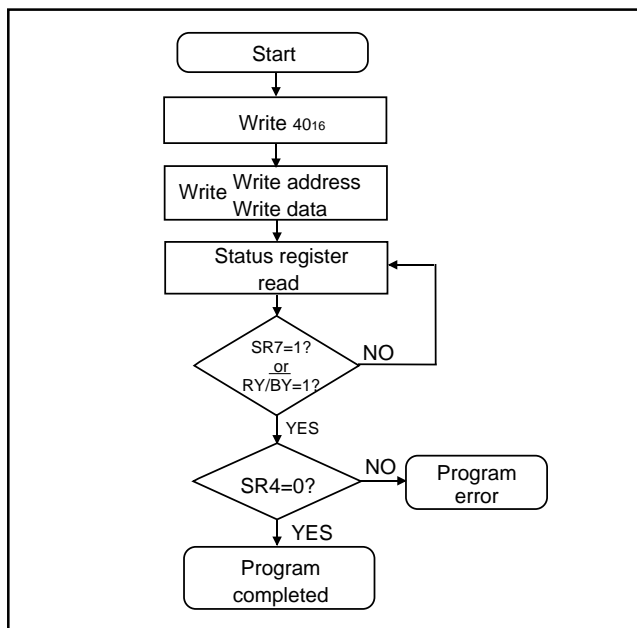


Fig. 47 Program flowchart

### Block Erase Command (20<sub>16</sub>/D0<sub>16</sub>)

By writing the command code “20<sub>16</sub>” in the first bus cycle and the confirmation command code “D0<sub>16</sub>” in the second bus cycle that follows to the block address of a flash memory block, the system initiates a block erase (erase and erase verify) operation.

Whether the block erase operation is completed can be confirmed by reading the status register or the RY/ $\overline{\text{BY}}$  status flag. At the same time the block erase operation starts, the read status register mode is automatically entered, so the content of the status register can be read out. The status register bit 7 (SR7) is set to “0” at the same time the block erase operation starts and is returned to “1” upon completion of the block erase operation. In this case, the read status register mode remains active until the read array command (FF<sub>16</sub>) is written.

The RY/ $\overline{\text{BY}}$  status flag is “0” during block erase operation and “1” when the block erase operation is completed as is the status register bit 7.

After the block erase operation is completed, the status register can be read out to know the result of the block erase operation. For de-

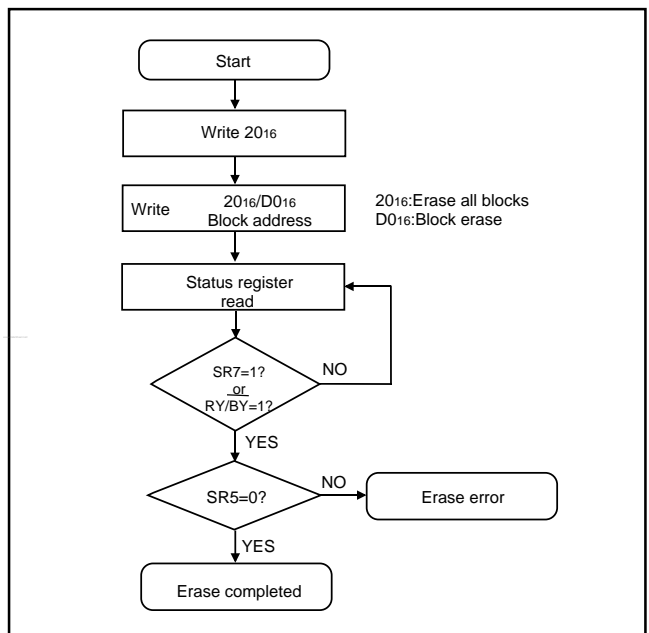


Fig. 48 Erase flowchart

### Status Register

The status register shows the operating state of the flash memory and whether erase operations and programs ended successfully or in error. It can be read in the following ways.

- (1) By reading an arbitrary address from the user ROM area after writing the read status register command (70<sub>16</sub>)
- (2) By reading an arbitrary address from the user ROM area in the period from when the program starts or erase operation starts to when the read array command (FF<sub>16</sub>) is input

Table 11 shows the status register.

Also, the status register can be cleared in the following way.

- (1) By writing the clear status register command (50<sub>16</sub>)
- (2) In the deep power down mode
- (3) In the power supply off state

After a reset, the status register is set to "80<sub>16</sub>".

Each bit in this register is explained below.

### Sequencer status (SR7)

After power-on, and after recover from deep power down mode, the sequencer status is set to "1"(ready).

The sequencer status indicates the operating status of the device. This status bit is set to "0" (busy) during write or erase operation and is set to "1" upon completion of these operations.

### Erase status (SR5)

The erase status informs the operating status of erase operation to the CPU. When an erase error occurs, it is set to "1".

The erase status is reset to "0" when cleared.

### Program status (SR4)

The program status informs the operating status of write operation to the CPU. When a write error occurs, it is set to "1".

The program status is reset to "0" when cleared.

If "1" is written for any of the SR5 or SR4 bits, the program, erase all blocks, and block erase commands are not accepted. Before executing these commands, execute the clear status register command (50<sub>16</sub>) and clear the status register.

Also, any commands are not correct, both SR5 and SR4 are set to "1".

**Table 11 Definition of each bit in status register**

Each bit of SRD0 bits	Status name	Definition	
		"1"	"0"
SR7 (bit7)	Sequencer status	Ready	Busy
SR6 (bit6)	Reserved	-	-
SR5 (bit5)	Erase status	Terminated in error	Terminated normally
SR4 (bit4)	Program status	Terminated in error	Terminated normally
SR3 (bit3)	Reserved	-	-
SR2 (bit2)	Reserved	-	-
SR1 (bit1)	Reserved	-	-
SR0 (bit0)	Reserved	-	-

# HARDWARE

## FUNCTIONAL DESCRIPTION

### Full Status Check

By performing full status check, it is possible to know the execution results of erase and program operations. Figure 49 shows a full sta-

tus check flowchart and the action to be taken when each error occurs.

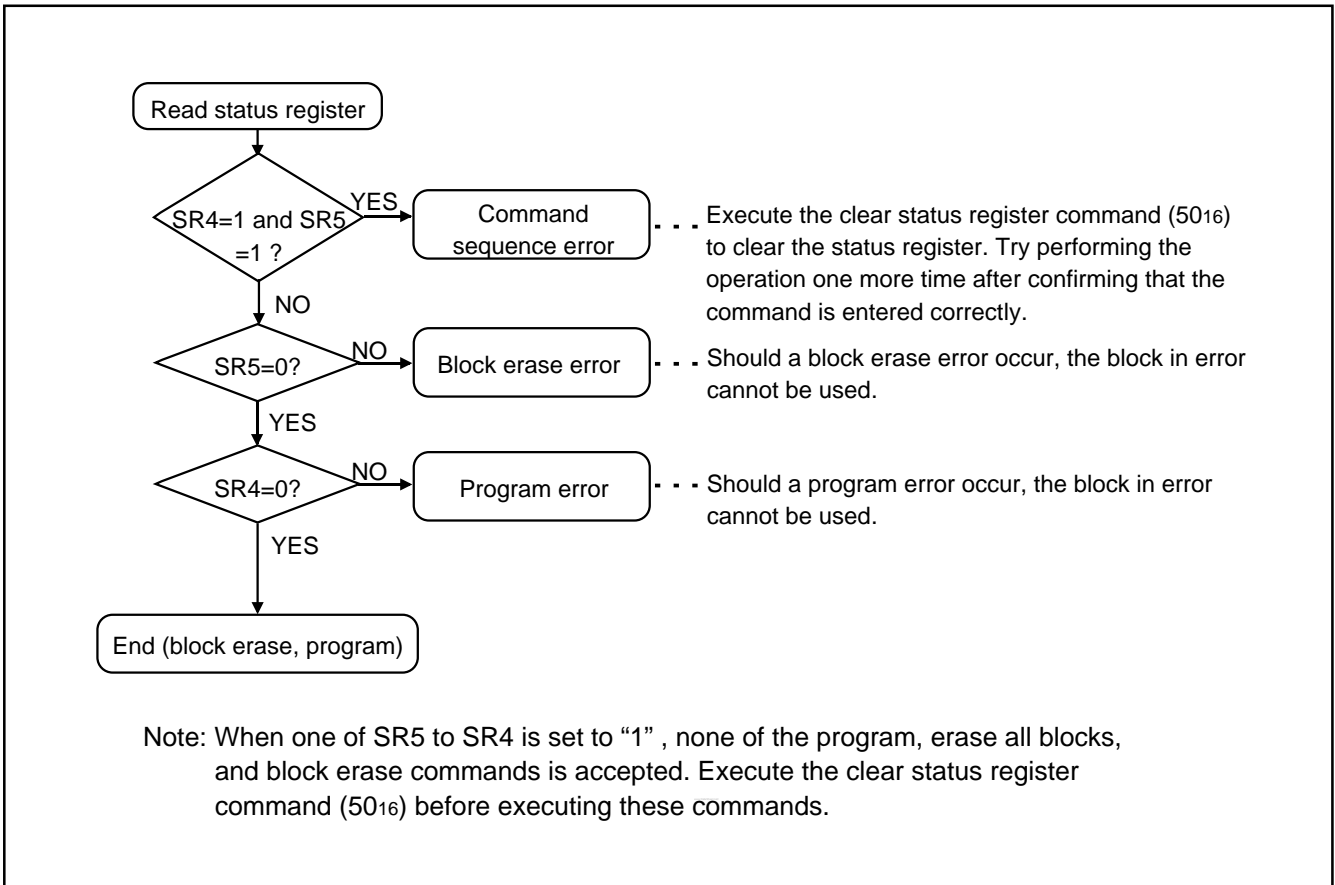


Fig. 49 Full status check flowchart and remedial procedure for errors



### Functions To Inhibit Rewriting Flash Memory Version

To prevent the contents of the flash memory version from being read out or rewritten easily, the device incorporates a ROM code protect function for use in parallel I/O mode and an ID code check function for use in standard serial I/O mode.

#### ROM code protect function

The ROM code protect function is the function inhibit reading out or modifying the contents of the flash memory version by using the ROM code protect control address (FFDB<sub>16</sub>) during parallel I/O mode. Figure 50 shows the ROM code protect control address (FFDB<sub>16</sub>). (This address exists in the user ROM area.)

If one of the pair of ROM code protect bits is set to “0”, ROM code protect is turned on, so that the contents of the flash memory version are protected against readout and modification. ROM code protect is implemented in two levels. If level 2 is selected, the flash memory is protected even against readout by a shipment inspection LSI tester, etc. When an attempt is made to select both level 1 and level 2, level 2 is selected by default.

If both of the two ROM code protect reset bits are set to “00”, ROM code protect is turned off, so that the contents of the flash memory version can be read out or modified. Once ROM code protect is turned on, the contents of the ROM code protect reset bits cannot be modified in parallel I/O mode. Use the serial I/O or some other mode to rewrite the contents of the ROM code protect reset bits.

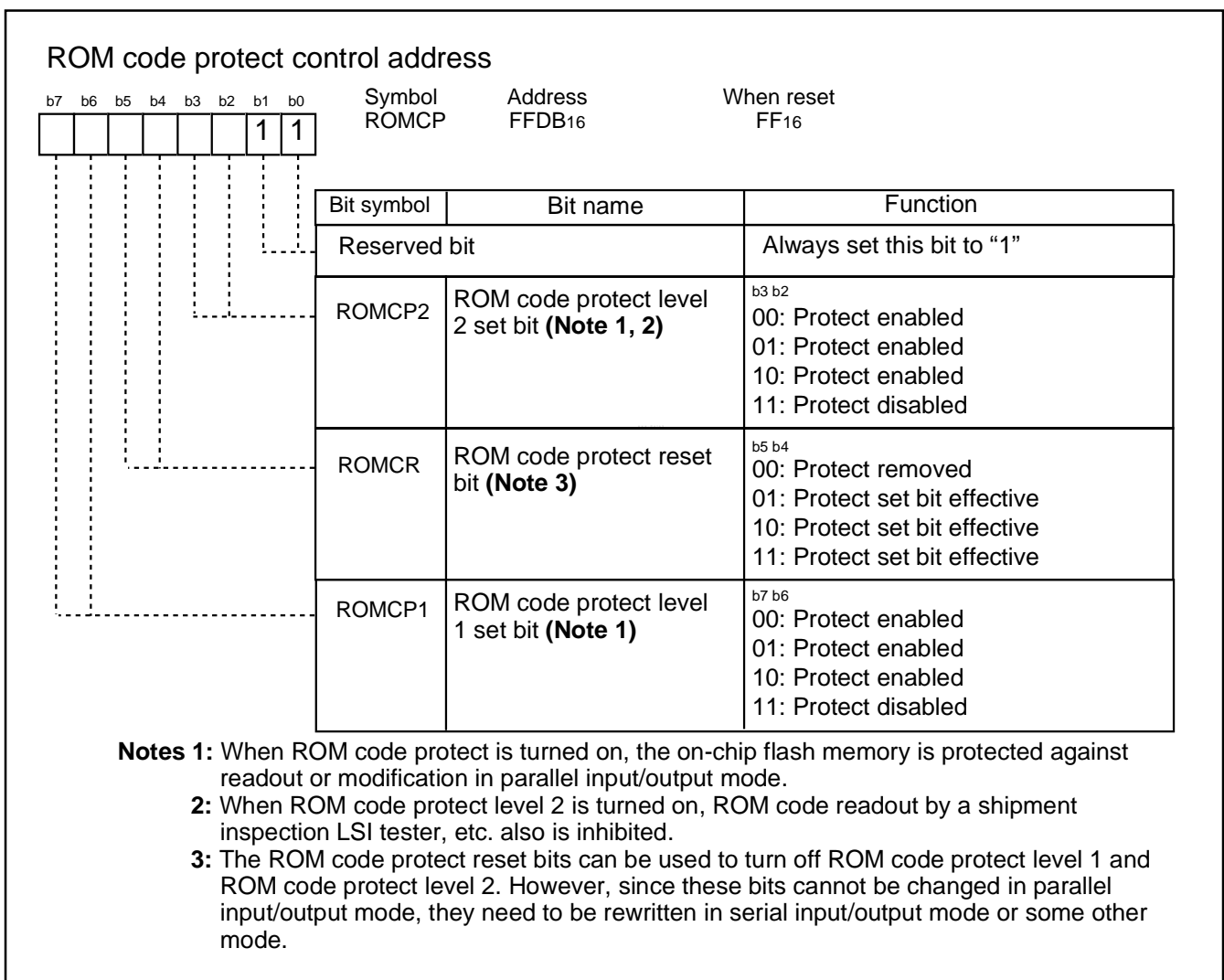


Fig. 50 ROM code protect control address

# HARDWARE

## FUNCTIONAL DESCRIPTION

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### ID Code Check Function

Use this function in standard serial I/O mode. When the contents of the flash memory are not blank, the ID code sent from the peripheral unit is compared with the ID code written in the flash memory to see if they match. If the ID codes do not match, the commands sent from the peripheral unit are not accepted. The ID code consists of 8-bit data, the areas of which are FFD4<sub>16</sub> to FFDA<sub>16</sub>. Write a program which has had the ID code preset at these addresses to the flash memory.

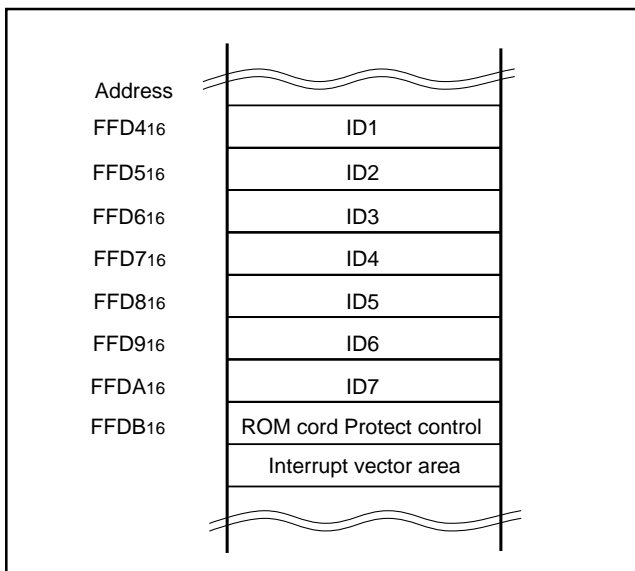


Fig. 51 ID code store addresses

### Parallel I/O Mode

The parallel I/O mode is entered by making connections shown in Figure 52 and then turning the Vcc power supply on.

### Address

The user ROM is only one block as shown in Figure 44. The block address referred to in this data sheet is the maximum address value of each block.

### User ROM and Boot ROM Areas

In parallel I/O mode, the user ROM and boot ROM areas shown in Figure 44 can be rewritten. The BSEL pin is used to choose between these two areas. The user ROM area is selected by pulling the BSEL input low; the boot ROM area is selected by driving the BSEL input high. Both areas of flash memory can be operated on in the same way.

Program and block erase operations can be performed in the user ROM area. The user ROM area and its block is shown in Figure 44.

The user ROM area is 32 Kbytes in size. In parallel I/O mode, it is located at addresses 8000<sub>16</sub> through FFFF<sub>16</sub>. The boot ROM area is 4 Kbytes in size. In parallel I/O mode, it is located at addresses F000<sub>16</sub> through FFFF<sub>16</sub>. Make sure program and block erase operations are always performed within this address range. (Access to any location outside this address range is prohibited.)

In the Boot ROM area, an erase block operation is applied to only one 4 Kbyte block. The boot ROM area has had a standard serial I/O mode control program stored in it when shipped from the Renesas factory. Therefore, using the device in standard serial input/output mode, you do not need to write to the boot ROM area.

### Functional Outline (Parallel I/O Mode)

In parallel I/O mode, bus operation modes—Read, Output Disable, Standby, Write, and Deep Power Down—are selected by the status of the CE, OE, WE, and RP input pins.

The contents of erase, program, and other operations are selected by writing a software command. The data, status register, etc. in memory can only be read out by a read after software command input.

Program and erase operations are controlled using software commands.

The following explains about bus operation modes, software commands, and status register.

### Bus Operation Modes

#### Read

The Read mode is entered by pulling the OE pin low when the CE pin is low and the WE and RP pins are high. There are two read modes: array, and status register, which are selected by software command input. In read mode, the data corresponding to each software command entered is output from the data I/O pins D0–D7. The read array mode is automatically selected when the device is powered on or after it exits deep power down mode.

#### Output Disable

The output disable mode is entered by pulling the CE pin low and the WE, OE, and RP pins high. Also, the data I/O pins are placed in the high-impedance state.

#### Standby

The standby mode is entered by driving the CE pin high when the RP pin is high. Also, the data I/O pins are placed in the high-impedance state. However, if the CE pin is set high during erase or program operation, the internal control circuit does not halt immediately and normal power consumption is required until the operation under way is completed.

#### Write

The write mode is entered by pulling the WE pin low when the CE pin is low and the OE and RP pins are high. In this mode, the device accepts the software commands or write data entered from the data I/O pins. A program, erase, or some other operation is initiated depending on the content of the software command entered here. The input data such as address and software command is latched at the rising edge of WE or CE whichever occurs earlier.

#### Deep Power Down

The deep power down is entered by pulling the RP pin low. Also, the data I/O pins are placed in the high-impedance state. When the device is freed from deep power down mode, the read array mode is selected and the content of the status register is set to “80<sub>16</sub>”. If the RP pin is pulled low during erase or program operation, the operation under way is canceled and the data in the relevant block becomes invalid.

**Table 12 Relationship between control signals and bus operation modes**

Mode		Pin name	CE	OE	WE	RP	D0 to D7
Read	Array		VIL	VIL	VIH	VIH	Data output
	Status register		VIL	VIL	VIH	VIH	Status register data output
Output disabled			VIL	VIH	VIH	VIH	High impedance
Stand by			VIH	X	X	VIH	High impedance
Write	Program		VIL	VIH	VIL	VIH	Command/data input
	Erase all blocks		VIL	VIH	VIL	VIH	Command input
	Block erase		VIL	VIH	VIL	VIH	Command input
Deep power down			X	X	X	VIL	High impedance

Note : X can be VIL or VIH.

# HARDWARE

## FUNCTIONAL DESCRIPTION

**Table 13 Description of Pin Function (Flash Memory Parallel I/O Mode)**

Pin name	Signal name	I/O	Function
Vcc,Vss	Power supply input		Apply $5.0 \pm 0.5$ V to the Vcc pin and 0 V to the Vss pin.
CNVss	CNVss	I	Connect this pin to Vcc.
$\overline{\text{RESET}}$	Reset input	I	Reset input pin. When reset is held low, more than 20 cycles of clock are required at the XIN pin.
XIN	Clock input	I	Connect a ceramic or crystal resonator between the XIN and XOUT pins. When entering an externally driven clock, enter it from XIN and leave XOUT open.
XOUT	Clock output	O	
AVss	Analog power supply input		Connect AVss to Vss.
VREF	Reference voltage input	I	Input AD reference voltage or keep open.
P00 to P07	Data I/O D0 to D7	I/O	These are data D0–D7 input/output pins.
P10 to P17	Address input A8 to A15	I	These are address A8–A15 input pins.
P20 to P27	Address input A0 to A7	I	These are address A0–A7 input pins.
P30	BSEL input	I	This is a BSEL input pin.
P31	$\overline{\text{RP}}$ input	I	This is a $\overline{\text{RP}}$ input pin.
P32	$\overline{\text{WE}}$ input	I	This is a $\overline{\text{WE}}$ input pin.
P33	$\overline{\text{CE}}$ input	I	This is a $\overline{\text{CE}}$ input pin.
P34	$\overline{\text{OE}}$ input	I	This is a $\overline{\text{OE}}$ input pin.
P40	RY/ $\overline{\text{BY}}$ output	O	This is a RY/ $\overline{\text{BY}}$ output pin.
P41	Input P41	I	Enter low signals to this pin.
P42 to P44	Input P4	I	Input “H” or “L” or keep open.

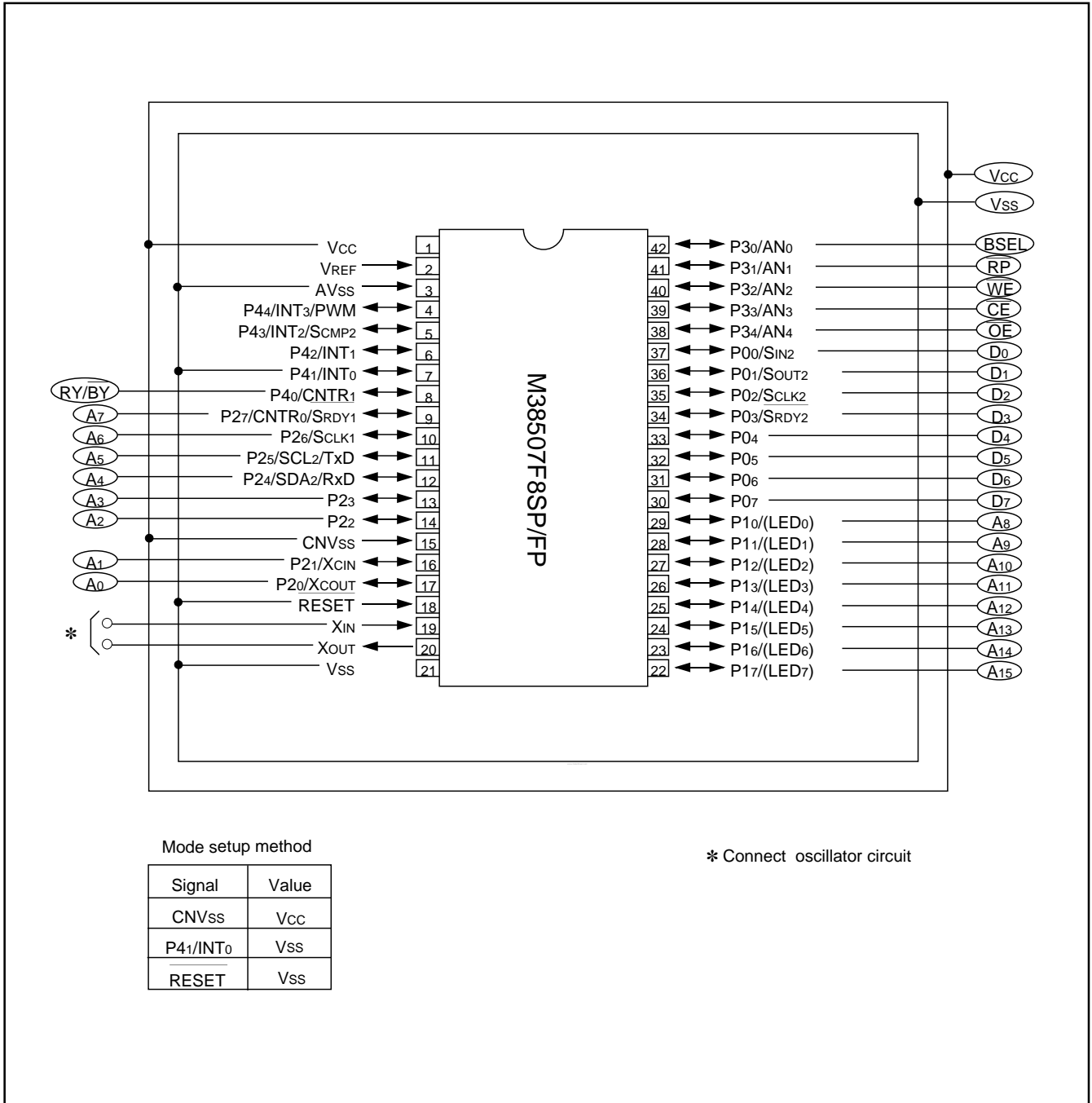


Fig. 52 Pin connection diagram in parallel I/O mode

# HARDWARE

## FUNCTIONAL DESCRIPTION

### Software Commands

Table 14 lists the software commands. By entering a software command from the data I/O pins (D0–D7) in Write mode, specify the content of the operation, such as erase or program operation, to be performed.

The following explains the content of each software command.

#### Read Array Command (FF16)

The read array mode is entered by writing the command code “FF16” in the first bus cycle. When an address to be read is input in one of the bus cycles that follow, the content of the specified address is output from the data I/O pins (D0–D7).

The read array mode is retained intact until another command is written.

The read array mode is also selected automatically when the device is powered on and after it exits deep power down mode.

#### Read Status Register Command (7016)

When the command code “7016” is written in the first bus cycle, the content of the status register is output from the data I/O pins (D0–D7) by a read in the second bus cycle. Since the content of the status register is updated at the falling edge of OE or CE, the OE or CE signal must be asserted each time the status is read. The status register is explained in the next section.

#### Clear Status Register Command (5016)

This command is used to clear the bits SR4,SR5 of the status register after they have been set. These bits indicate that operation has ended in an error. To use this command, write the command code “5016” in the first bus cycle.

**Table 14 Software command list (parallel I/O mode)**

Command	Cycle number	First bus cycle			Second bus cycle		
		Mode	Address	Data (D0 to D7)	Mode	Address	Data (D0 to D7)
Read array	1	Write	X(Note 4)	FF16			
Read status register	2	Write	X	7016	Read	X	SRD(Note 1)
Clear status register	1	Write	X	5016			
Program	2	Write	X	4016	Write	WA(Note 2)	WD(Note 2)
All block erase	2	Write	X	2016	Write	X	2016
Block erase	2	Write	X	2016	Write	BA(Note 3)	D016

**Notes 1:** SRD = Status Register Data

**2:** WA = Write Address, WD = Write Data

**3:** BA = Block Address (Enter the maximum address of each block)

**4:** X denotes a given address in the user ROM area or boot ROM area.

### Program Command (40<sub>16</sub>)

The program operation starts when the command code “40<sub>16</sub>” is written in the first bus cycle. Then, if the address and data to program are written in the 2nd bus cycle, program operation (data programming and verification) will start.

Whether the write operation is completed can be confirmed by reading the status register or the RY/B $\bar{Y}$  signal status. When the program starts, the read status register mode is accessed automatically and the content of the status register can be read out from the data bus (D<sub>0</sub>–D<sub>7</sub>). The status register bit 7 (SR7) is set to “0” at the same time the write operation starts and is returned to “1” upon completion of the write operation. In this case, the read status register mode remains active until the read array command (FF<sub>16</sub>) is written.

The RY/B $\bar{Y}$  pin is “L” during write operation and “H” when the write operation is completed as is the status register bit 7.

At program end, program results can be checked by reading the status register.

### Erase All Blocks Command (20<sub>16</sub>/20<sub>16</sub>)

By writing the command code “20<sub>16</sub>” in the first bus cycle and the confirmation command code “20<sub>16</sub>” in the second bus cycle that follows, the system starts erase all blocks( erase and erase verify).

Whether the erase all blocks command is terminated can be confirmed by reading the status register or the RY/B $\bar{Y}$  signal status . When the erase all blocks operation starts, the read status register mode is accessed automatically and the content of the status register can be read out. The status register bit 7 (SR7) is set to “0” at the same time the erase operation starts and is returned to “1” upon completion of the erase operation. In this case, the read status register mode remains active until the read array command (FF<sub>16</sub>) is written.

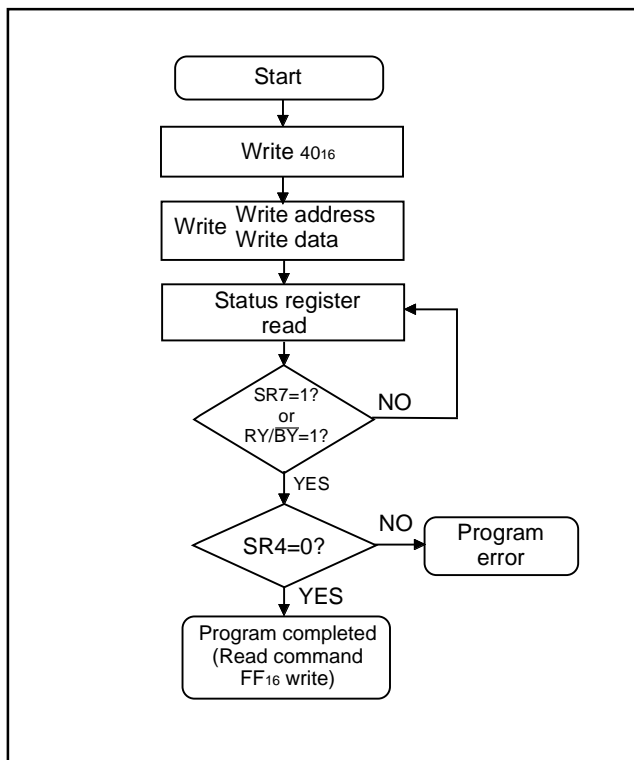


Fig. 53 Page program flowchart

The RY/B $\bar{Y}$  pin is “L” during erase operation and “H” when the erase operation is completed as is the status register bit 7.

At erase all blocks end, erase results can be checked by reading the status register. For details, refer to the section where the status register is detailed.

### Block Erase Command (20<sub>16</sub>/D0<sub>16</sub>)

By writing the command code “20<sub>16</sub>” in the first bus cycle and the confirmation command code “D0<sub>16</sub>” in the second bus cycle that follows to the block address of a flash memory block, the system initiates a block erase (erase and erase verify) operation.

Whether the block erase operation is completed can be confirmed by reading the status register or the RY/B $\bar{Y}$  signal. At the same time the block erase operation starts, the read status register mode is automatically entered, so the content of the status register can be read out. The status register bit 7 (SR7) is set to “0” at the same time the block erase operation starts and is returned to “1” upon completion of the block erase operation. In this case, the read status register mode remains active until the read array command (FF<sub>16</sub>) is written.

The RY/B $\bar{Y}$  pin is “L” during block erase operation and “H” when the block erase operation is completed as is the status register bit 7.

After the block erase operation is completed, the status register can be read out to know the result of the block erase operation. For details, refer to the section where the status register is detailed.

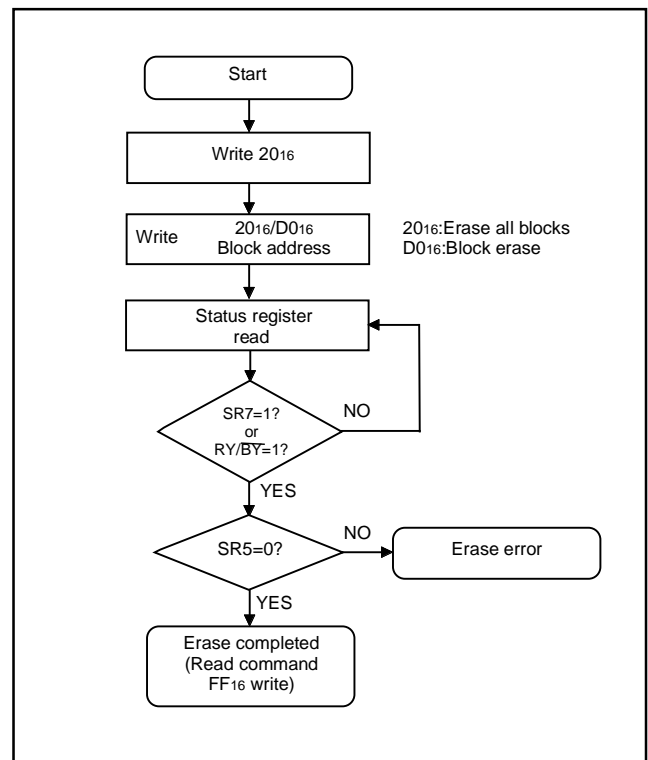


Fig. 54 Block erase flowchart

# HARDWARE

## FUNCTIONAL DESCRIPTION

### Status Register

The status register indicates status such as whether an erase operation or a program ended successfully or in error. It can be read under the following conditions.

- (1) In the read array mode when the read status register command (70<sub>16</sub>) is written and the block address is subsequently read.
- (2) In the period from when the program write or auto erase starts to when the read array command (FF<sub>16</sub>)

The status register is cleared in the following situations.

- (1) By writing the clear status register command (50<sub>16</sub>)
- (2) In the deep power down mode
- (3) In the power supply off state

Table 15 gives the definition of each status register bit. When power is turned on or returning from the deep power down mode, the status register outputs "80<sub>16</sub>".

### Sequencer status (SR7)

The sequencer status indicates the operating status of the flash memory. When power is turned on or returning from the deep power down mode, "1" is set for it. This bit is "0" (busy) during the write or erase operations and becomes "1" when these operations ends.

### Erase Status (SR5)

The erase status reports the operating status of the erase operation. If an erase error occurs, it is set to "1". When the erase status is cleared, it is set to "0".

### Program Status (SR4)

The program status reports the operating status of the write operation. If a write error occurs, it is set to "1". When the program status is cleared, it is set to "0".

If "1" is written for any of the SR5, SR4 bits, the program erase all blocks, block erase, commands are not accepted. Before executing these commands, execute the clear status register command (50<sub>16</sub>) and clear the status register.

Also, any commands are not correct, both SR5 and SR4 are set to "1".

### Full Status Check

Results from executed erase and program operations can be known by running a full status check. Figure 55 shows a flowchart of the full status check and explains how to remedy errors which occur.

### Ready/Busy (RY/ $\overline{\text{BY}}$ ) pin

The RY/ $\overline{\text{BY}}$  pin is an output pin (N-channel open drain output) which, like the sequencer status (SR7), indicates the operating status of the flash memory. It is "L" level during auto program or auto erase operations and becomes to the high impedance state (ready state) when these operations end. The RY/ $\overline{\text{BY}}$  pin requires an external pull-up.

Table 15 Status register

Each bit of SRD0 bits	Status name	Definition	
		"1"	"0"
SR7 (D7)	Sequencer status	Ready	Busy
SR6 (D6)	Reserved	-	-
SR5 (D5)	Erase status	Ended in error	Ended successfully
SR4 (D4)	Program status	Ended in error	Ended successfully
SR3 (D3)	Reserved	-	-
SR2 (D2)	Reserved	-	-
SR1 (D1)	Reserved	-	-
SR0 (D0)	Reserved	-	-



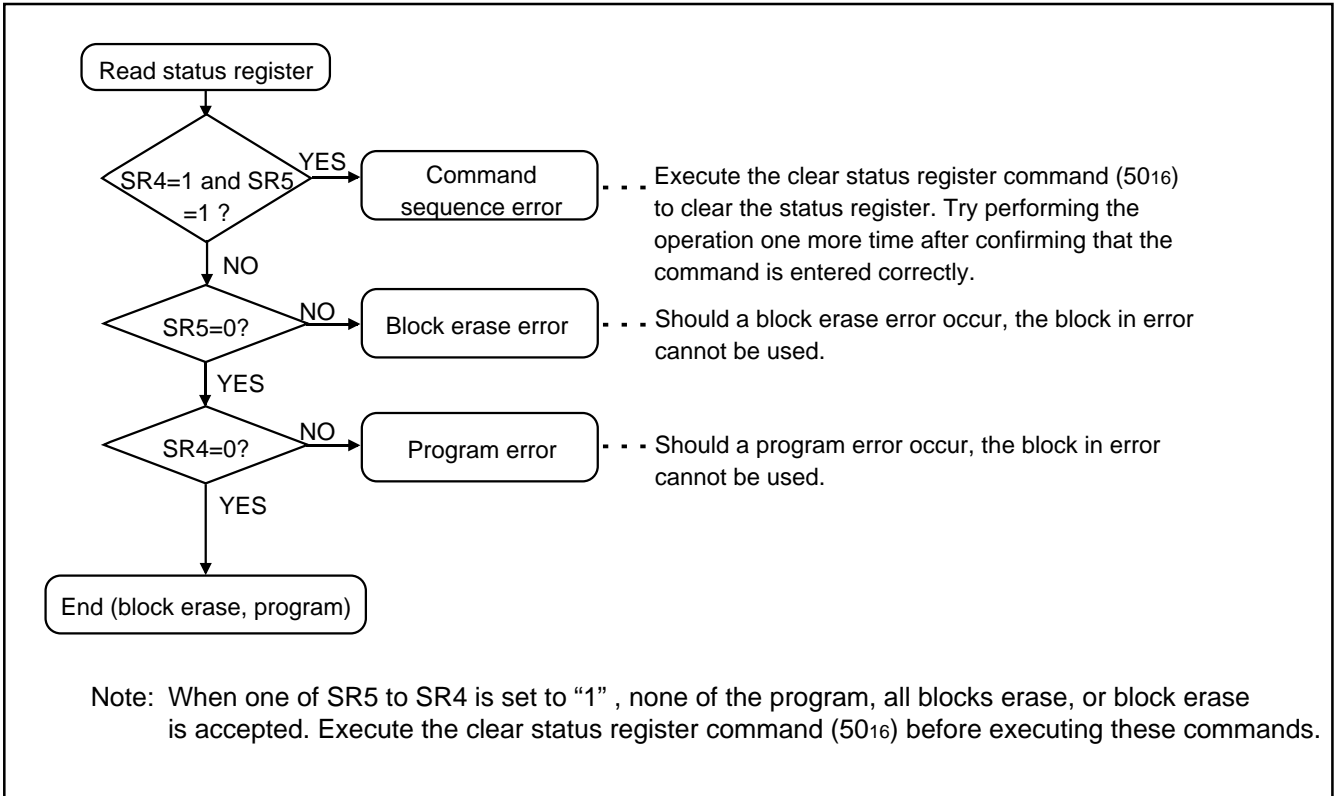


Fig. 55 Full status check flowchart and remedial procedure for errors

# HARDWARE

## FUNCTIONAL DESCRIPTION

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### Standard serial I/O mode

The standard serial I/O mode inputs and outputs the software commands, addresses and data needed to operate (read, program, erase, etc.) the internal flash memory. This I/O is clock synchronized serial. This modes require a purpose-specific peripheral unit.

The standard serial I/O mode is different from the parallel I/O mode in that the CPU controls flash memory rewrite (uses the CPU's rewrite mode), rewrite data input and so forth. The standard serial I/O mode is started by connecting "H" to the P26 (SCLK1) pin and the P41(INT0) pin and "H" to the CNVSS pin (when VCC = 4.5 V to 5.5 V, connect to VCC; when VCC = 2.7 V to 4.5 V, supply 4.5 V to 5.5 V to Vpp from an external source), and releasing the reset operation. (In the ordinary command mode, set CNVSS pin to "L" level.)

This control program is written in the boot ROM area when the product is shipped from Renesas Technology Corp. Accordingly, make note of the fact that the standard serial I/O mode cannot be used if the boot ROM area is rewritten in the parallel I/O mode. Figure 56 shows the pin connections for the standard serial I/O mode. Serial data I/O uses SI/O1 data serially in 8-bit units.

To use standard serial I/O mode. The operation uses the four SI/O1 pins SCLK1, RxD, TxD and  $\overline{\text{SRDY1}}$  (BUSY). The SCLK1 pin is the transfer clock input pin through which an external transfer clock is input. The TxD pin is for CMOS output. The  $\overline{\text{SRDY1}}$  (BUSY) pin outputs an "L" level when ready for reception and an "H" level when reception starts.

In the standard serial I/O mode, only the user ROM area indicated in Figure 44 can be rewritten. The boot ROM cannot.

In the standard serial I/O mode, a 7-byte ID code is used. When there is data in the flash memory, commands sent from the peripheral unit (programmer) are not accepted unless the ID code matches.

### Overview of standard serial I/O mode

In standard serial I/O mode, software commands, addresses and data are input and output between the MCU and peripheral units (serial programmer, etc.) using 4-wire clock-synchronized serial I/O (SI/O1).

In reception, software commands, addresses and program data are synchronized with the rise of the transfer clock that is input to the SCLK1 pin, and are then input to the MCU via the RxD pin. In transmission, the read data and status are synchronized with the fall of the transfer clock, and output from the TxD pin.

The TxD pin is for CMOS output. Transfer is in 8-bit units with LSB first.

When busy, such as during transmission, reception, erasing or program execution, the  $\overline{\text{SRDY1}}$  (BUSY) pin is "H" level. Accordingly, always start the next transfer after the  $\overline{\text{SRDY1}}$  (BUSY) pin is "L" level. Also, data and status registers in memory can be read after inputting software commands. Status, such as the operating state of the flash memory or whether a program or erase operation ended successfully or not, can be checked by reading the status register. Here following are explained software commands, status registers, etc.

**Table 16 Pin functions (Flash memory standard serial I/O mode)**

Pin	Name	I/O	Description
Vcc,Vss	Power input		Apply program/erase protection voltage to Vcc pin and 0 V to Vss pin.
CNVss	CNVss	I	Connect to Vcc when Vcc = 4.5 V to 5.5 V. Connect to Vpp (=4.5 V to 5.5 V) when Vcc = 2.7 V to 4.5 V.
RESET	Reset input	I	Reset input pin. While reset is "L" level, a 20 cycle or longer clock must be input to XIN pin.
XIN	Clock input	I	Connect a ceramic resonator or crystal oscillator between XIN and XOUT pins. To input an externally generated clock, input it to XIN pin and open XOUT pin.
XOUT	Clock output	O	
AVss	Analog power supply input		Connect AVss to Vss .
VREF	Reference voltage input	I	Enter the reference voltage for AD from this pin.
P00 to P07	Input port P0	I	Input "H" or "L" level signal or open.
P10 to P17	Input port P1	I	Input "H" or "L" level signal or open.
P20 to P23	Input port P2	I	Input "H" or "L" level signal or open.
P24	RxD input	I	Serial data input pin
P25	TxD output	O	Serial data output pin
P26	SCLK1 input	I	Serial clock input pin
P27	BUSY output	O	BUSY signal output pin
P30 to P34	Input port P3	I	Input "H" or "L" level signal or open.
P40, P42 to P44	Input port P4	I	Input "H" or "L" level signal or open.
P41	Input P41	I	Input "H" level signal, when reset is released.

# HARDWARE

## FUNCTIONAL DESCRIPTION

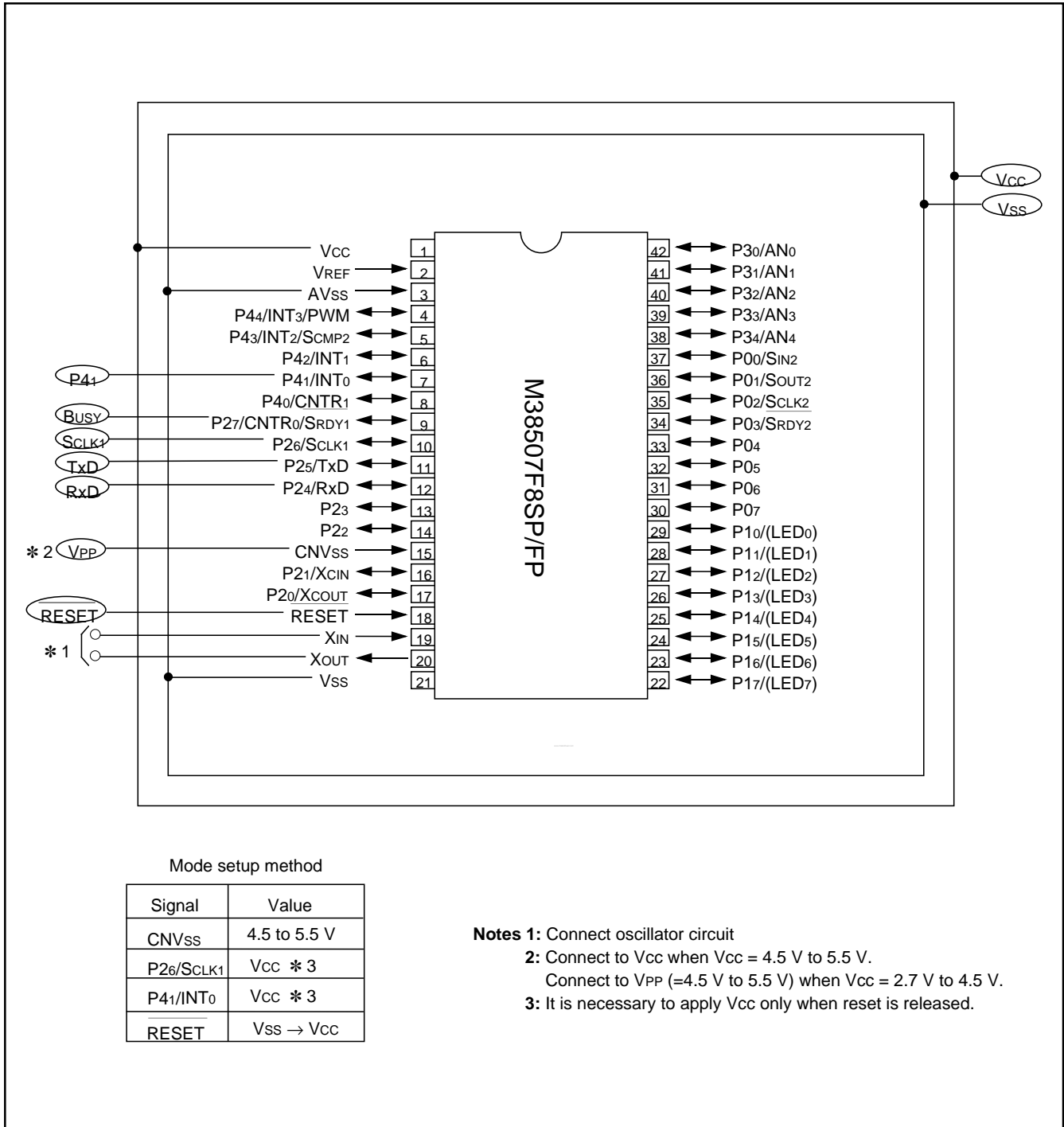


Fig. 56 Connection for serial I/O mode

### Software Commands

Table 17 lists software commands. In the standard serial I/O mode, erase operations, programs and reading are controlled by transferring software commands via the Rx/D pin. Software commands are

explained here below. Basically, the software commands of the standard serial I/O mode is as same as that of the parallel I/O mode, but it is excluded 1 command of block erase, and it is added 3 command of ID check, download function, version data output function.

**Table 17 Software commands (Standard serial I/O mode 1)**

	Control command	1st byte transfer	2nd byte	3rd byte	4th byte	5th byte	6th byte		When ID is not verified
1	Page read	FF <sub>16</sub>	Address (middle)	Address (high)	Data output	Data output	Data output	Data output to 259th byte	Not acceptable
2	Page program	41 <sub>16</sub>	Address (middle)	Address (high)	Data input	Data input	Data input	Data input to 259th byte	Not acceptable
3	Erase all blocks	A7 <sub>16</sub>	D0 <sub>16</sub>						Not acceptable
4	Read status register	70 <sub>16</sub>	SRD output	SRD1 output					Acceptable
5	Clear status register	50 <sub>16</sub>							Not acceptable
6	ID check	F5 <sub>16</sub>	Address (low)	Address (middle)	Address (high)	ID size	ID1	To ID7	Acceptable
7	Download function	FA <sub>16</sub>	Size (low)	Size (high)	Check-sum	Data input	To required number of times		Not acceptable
8	Version data output function	FB <sub>16</sub>	Version data output	Version data output	Version data output	Version data output	Version data output	Version data output to 9th byte	Acceptable

**Notes 1:** Shading indicates transfer from flash memory microcomputer to peripheral unit. All other data is transferred from the peripheral unit to the flash memory microcomputer.

**2:** SRD refers to status register data. SRD1 refers to status register 1 data.

**3:** All commands can be accepted when the flash memory is totally blank.

**4:** Address high (A<sub>16</sub> to A<sub>23</sub>) must be "0016".

# HARDWARE

## FUNCTIONAL DESCRIPTION

### Page Read Command

This command reads the specified page (256 bytes) in the flash memory sequentially one byte at a time. Execute the page read command as explained here following.

(1) Transfer the "FF<sub>16</sub>" command code with the 1st byte.

(2) Transfer addresses A<sub>8</sub> to A<sub>15</sub> and A<sub>16</sub> to A<sub>23</sub> ("00<sub>16</sub>") with the 2nd and 3rd bytes respectively.

(3) From the 4th byte onward, data (D<sub>0</sub>–D<sub>7</sub>) for the page (256 bytes) specified with addresses A<sub>8</sub> to A<sub>23</sub> will be output sequentially from the smallest address first in sync with the fall of the clock.

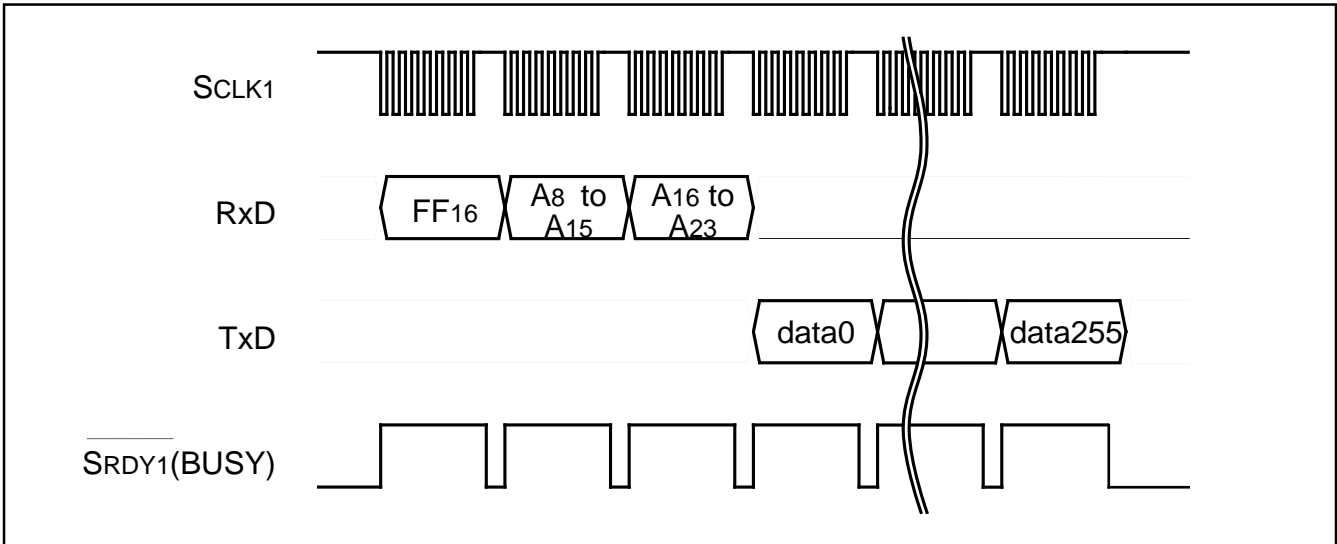


Fig. 57 Timing for page read

### Read Status Register Command

This command reads status information. When the "70<sub>16</sub>" command code is sent with the 1st byte, the contents of the status register (SRD) specified with the 2nd byte and the contents of status register 1 (SRD1) specified with the 3rd byte are read.

### Clear Status Register Command

This command clears the bits (SR4–SR5) which are set when the status register operation ends in error. When the "50<sub>16</sub>" command code is sent with the 1st byte, the aforementioned bits are cleared. When the clear status register operation ends, the SRDY1 (BUSY) signal changes from the "H" to the "L" level.

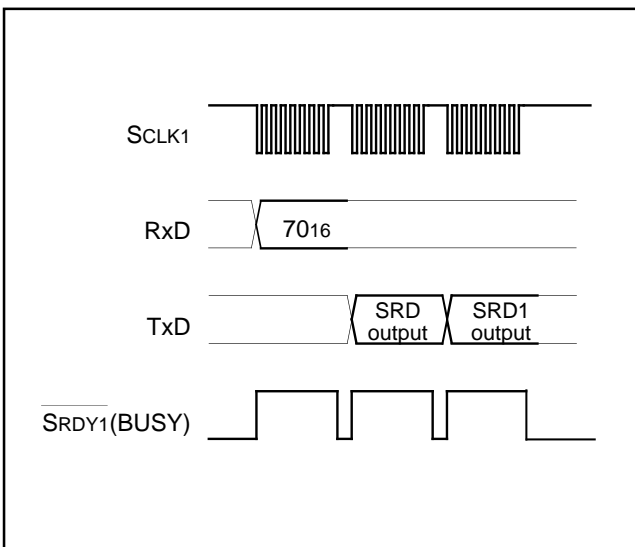


Fig. 58 Timing for reading the status register

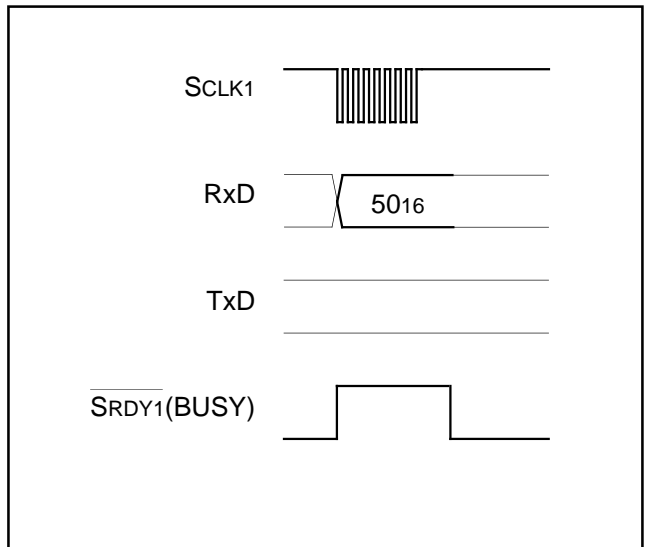


Fig. 59 Timing for clearing the status register

### Page Program Command

This command writes the specified page (256 bytes) in the flash memory sequentially one byte at a time. Execute the page program command as explained here following.

- (1) Transfer the "4116" command code with the 1st byte.
- (2) Transfer addresses A8 to A15 and A16 to A23 ("0016") with the 2nd and 3rd bytes respectively.

- (3) From the 4th byte onward, as write data (D0–D7) for the page (256 bytes) specified with addresses A8 to A23 is input sequentially from the smallest address first, that page is automatically written.

When reception setup for the next 256 bytes ends, the  $\overline{\text{SRDY1}}$  (BUSY) signal changes from the "H" to the "L" level. The result of the page program can be known by reading the status register. For more information, see the section on the status register.

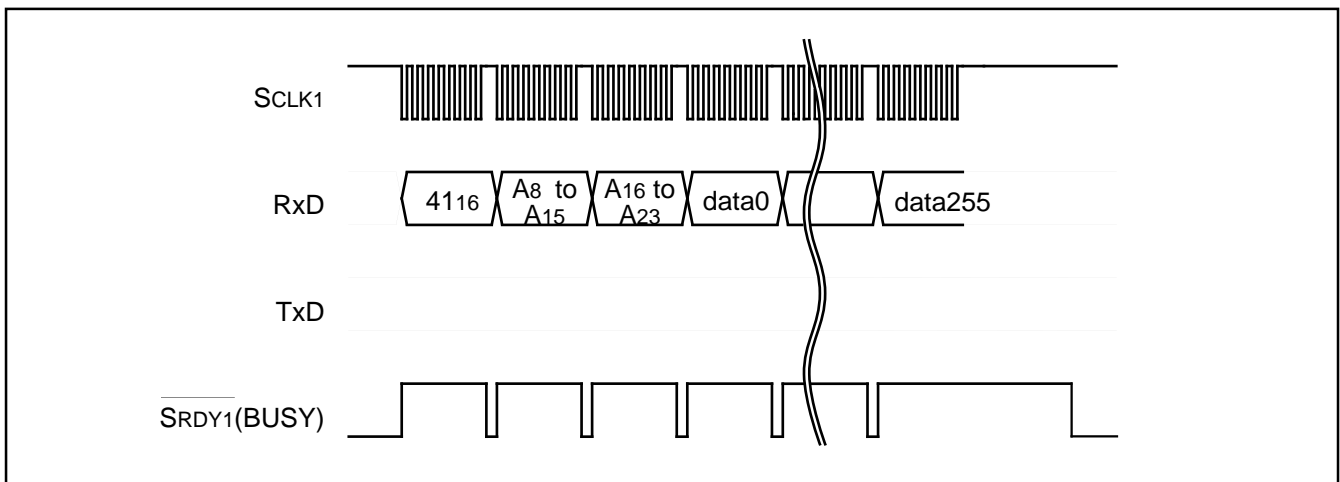


Fig. 60 Timing for the page program

### Erase All Blocks Command

This command erases the content of all blocks. Execute the erase all blocks command as explained here following.

- (1) Transfer the "A716" command code with the 1st byte.

- (2) Transfer the verify command code "D016" with the 2nd byte. With the verify command code, the erase operation will start and continue for all blocks in the flash memory.

When block erasing ends, the  $\overline{\text{SRDY1}}$  (BUSY) signal changes from the "H" to the "L" level. The result of the erase operation can be known by reading the status register.

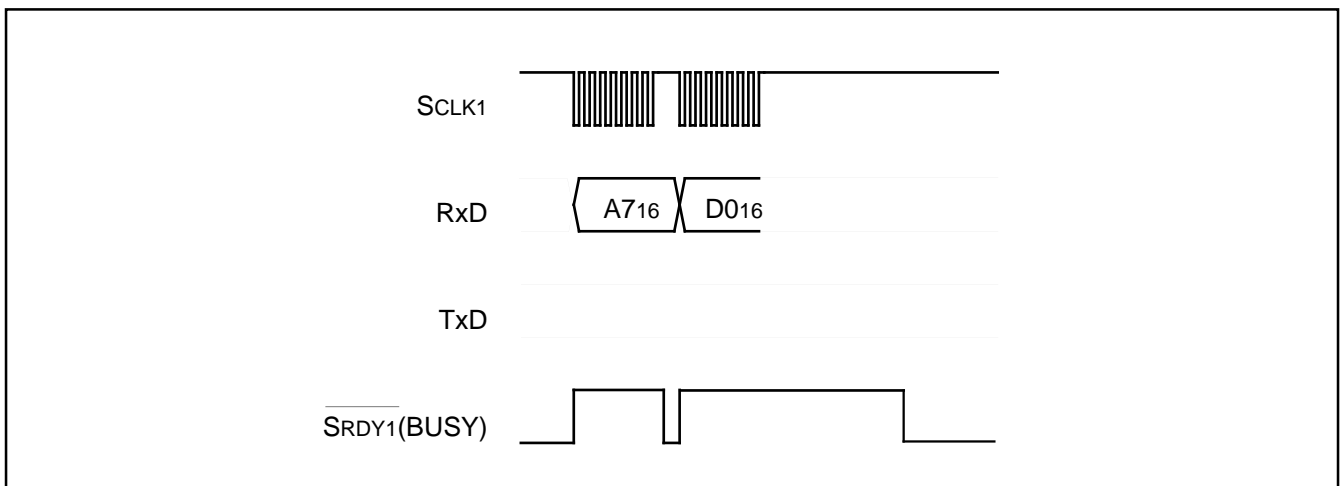


Fig. 61 Timing for erasing all blocks

# HARDWARE

## FUNCTIONAL DESCRIPTION

### Download Command

This command downloads a program to the RAM for execution. Execute the download command as explained here following.

- (1) Transfer the "FA16" command code with the 1st byte.
- (2) Transfer the program size with the 2nd and 3rd bytes.
- (3) Transfer the check sum with the 4th byte. The check sum is added to all data sent with the 5th byte onward.

- (4) The program to execute is sent with the 5th byte onward.

When all data has been transmitted, if the check sum matches, the downloaded program is executed. The size of the program will vary according to the internal RAM.

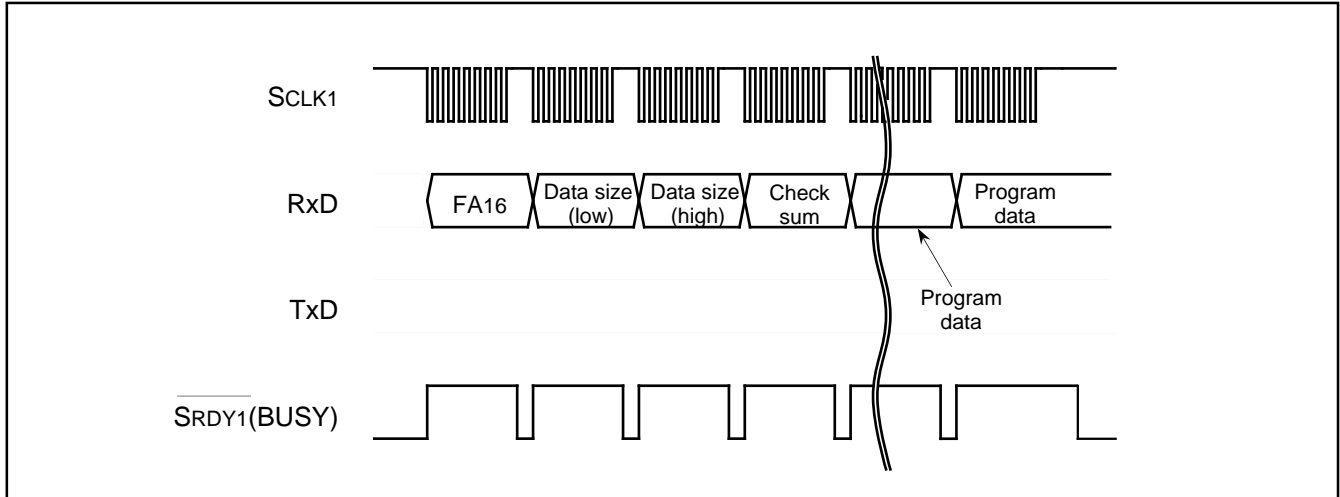


Fig. 62 Timing for download

### Version Information Output Command

This command outputs the version information of the control program stored in the boot area. Execute the version information output command as explained here following.

- (1) Transfer the "FB16" command code with the 1st byte.
- (2) The version information will be output from the 2nd byte onward. This data is composed of 8 ASCII code characters.

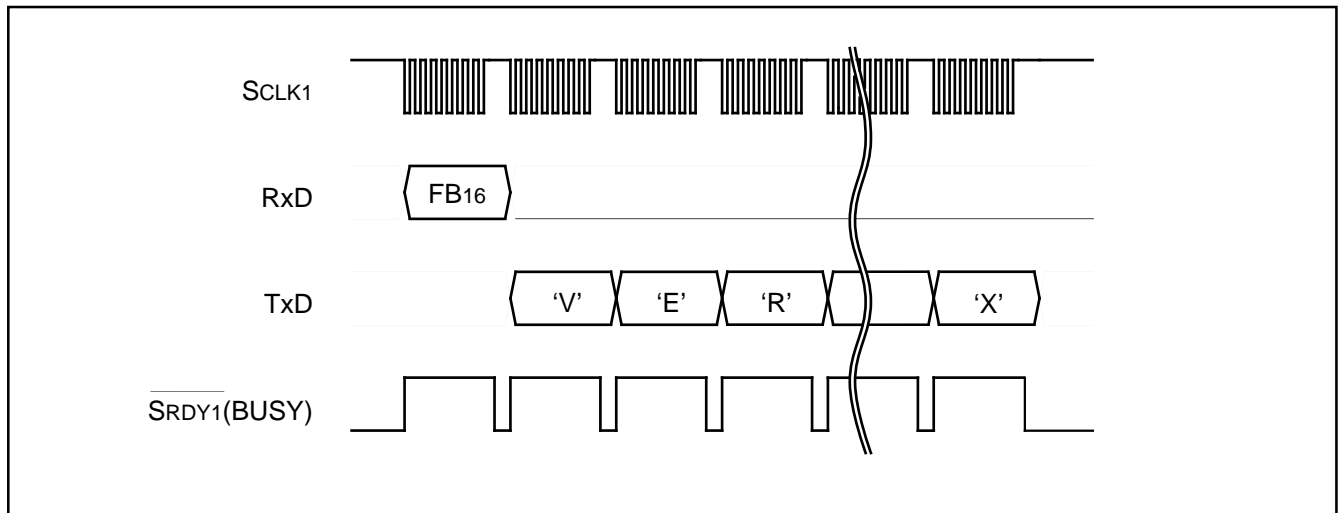


Fig. 63 Timing for version information output



### ID Check

This command checks the ID code. Execute the boot ID check command as explained here following.

- (1) Transfer the "F5<sub>16</sub>" command code with the 1st byte.
- (2) Transfer addresses A0 to A7, A8 to A15 and A16 to A23 ("00<sub>16</sub>") of

the 1st byte of the ID code with the 2nd, 3rd and 4th bytes respectively.

- (3) Transfer the number of data sets of the ID code with the 5th byte.
- (4) The ID code is sent with the 6th byte onward, starting with the 1st byte of the code.

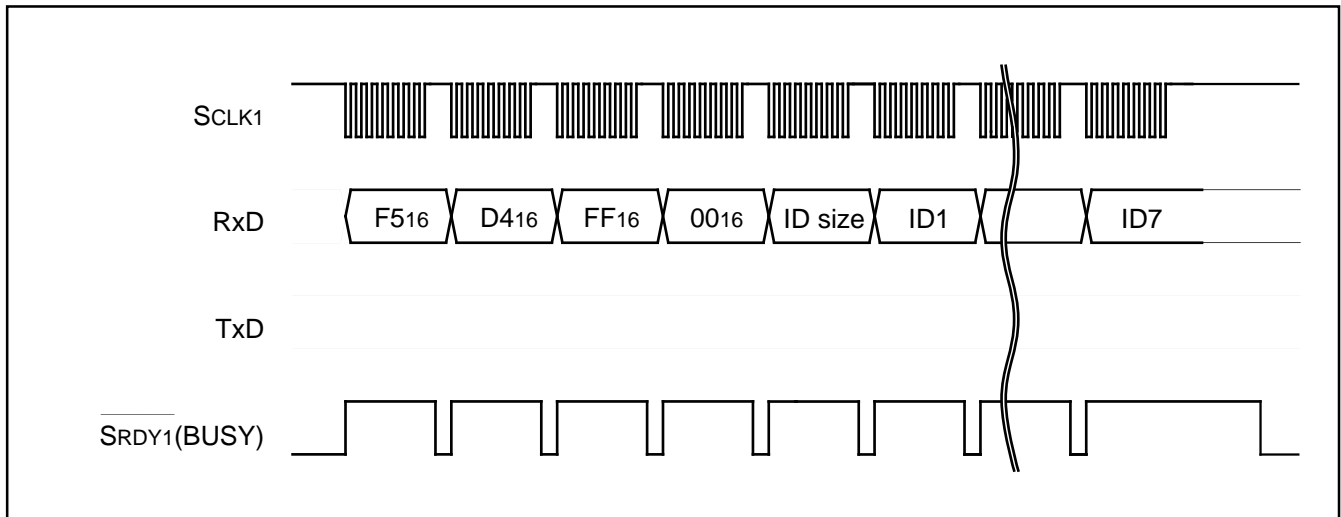


Fig. 64 Timing for the ID check

### ID Code

When the flash memory is not blank, the ID code sent from the peripheral units and the ID code written in the flash memory are compared to see if they match. If the codes do not match, the command

sent from the peripheral units is not accepted. An ID code contains 8 bits of data. Area is, from the 1st byte, addresses FFD4<sub>16</sub> to FFDA<sub>16</sub>. Write a program into the flash memory, which already has the ID code set for these addresses.

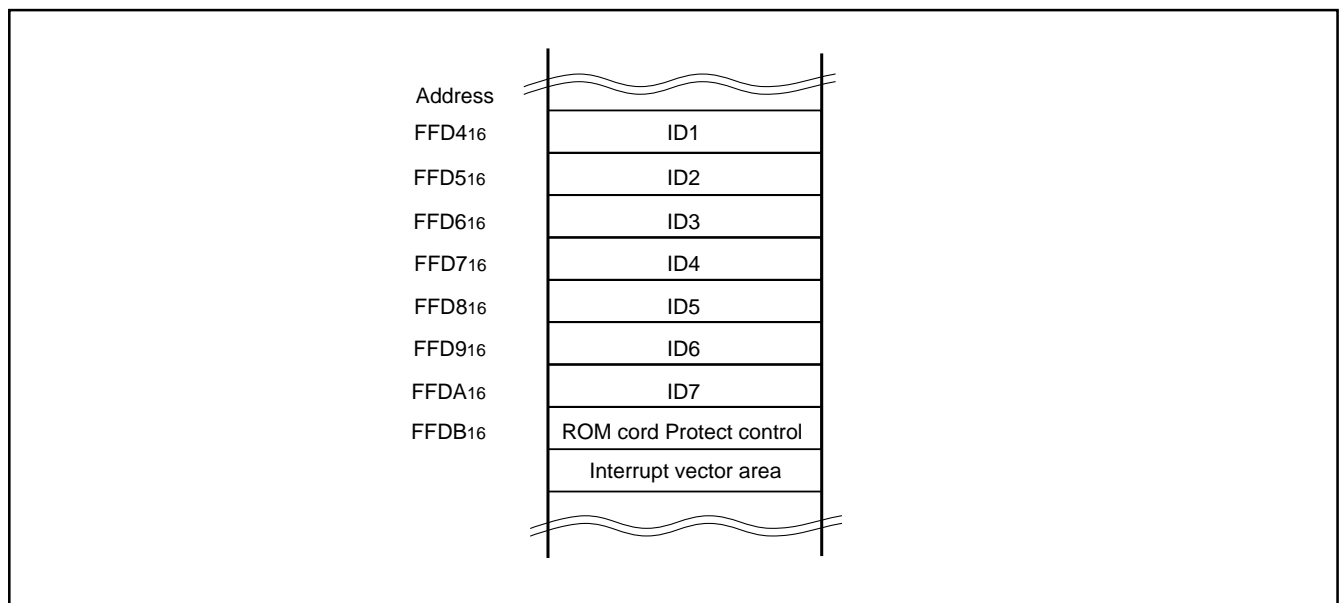


Fig. 65 ID code storage addresses

# HARDWARE

## FUNCTIONAL DESCRIPTION

### Status Register (SRD)

The status register indicates operating status of the flash memory and status such as whether an erase operation or a program ended successfully or in error. It can be read by writing the read status register command (7016). Also, the status register is cleared by writing the clear status register command (5016).

Table 18 gives the definition of each status register bit. After clearing the reset, the status register outputs "8016".

### Sequencer status (SR7)

After power-on and recover from deep power down mode, the sequencer status is set to "1"(ready).

**Table 18 Status register (SRD)**

SRD0 bits	Status name	Definition	
		"1"	"0"
SR7 (bit7)	Sequencer status	Ready	Busy
SR6 (bit6)	Reserved	-	-
SR5 (bit5)	Erase status	Terminated in error	Terminated normally
SR4 (bit4)	Program status	Terminated in error	Terminated normally
SR3 (bit3)	Reserved	-	-
SR2 (bit2)	Reserved	-	-
SR1 (bit1)	Reserved	-	-
SR0 (bit0)	Reserved	-	-

### Status Register 1 (SRD1)

Status register 1 indicates the status of serial communications, results from ID checks and results from check sum comparisons. It can be read after the SRD by writing the read status register command (7016). Also, status register 1 is cleared by writing the clear status register command (5016).

Table 19 gives the definition of each status register bit. "0016" is output when power is turned on and the flag status is maintained even after the reset.

### Boot Update Completed Bit (SR15)

This flag indicates whether the control program was downloaded to the RAM or not, using the download function.

**Table 19 Status register 1 (SRD1)**

SRD1 bits	Status name	Definition	
		"1"	"0"
SR15 (bit7)	Boot update completed bit	Update completed	Not Update
SR14 (bit6)	Reserved	-	-
SR13 (bit5)	Reserved	-	-
SR12 (bit4)	Checksum match bit	Match	Mismatch
SR11 (bit3) SR10 (bit2)	ID check completed bits	00 01 10 11	Not verified Verification mismatch Reserved Verified
SR9 (bit1)	Data reception time out	Time out	Normal operation
SR8 (bit0)	Reserved	-	-

The sequencer status indicates the operating status of the device. This status bit is set to "0" (busy) during write or erase operation and is set to "1" upon completion of these operations.

### Erase Status (SR5)

The erase status reports the operating status of the auto erase operation. If an erase error occurs, it is set to "1". When the erase status is cleared, it is set to "0".

### Program Status (SR4)

The program status reports the operating status of the auto write operation. If a write error occurs, it is set to "1". When the program status is cleared, it is set to "0".

### Check Sum Consistency Bit (SR12)

This flag indicates whether the check sum matches or not when a program is downloaded for execution using the download function.

### ID Check Completed Bits (SR11 and SR10)

These flags indicate the result of ID checks. Some commands cannot be accepted without an ID check.

### Data Reception Time Out (SR9)

This flag indicates when a time out error is generated during data reception. If this flag is attached during data reception, the received data is discarded and the microcomputer returns to the command wait state.

### Full Status Check

Results from executed erase and program operations can be known

by running a full status check. Figure 66 shows a flowchart of the full status check and explains how to remedy errors which occur.

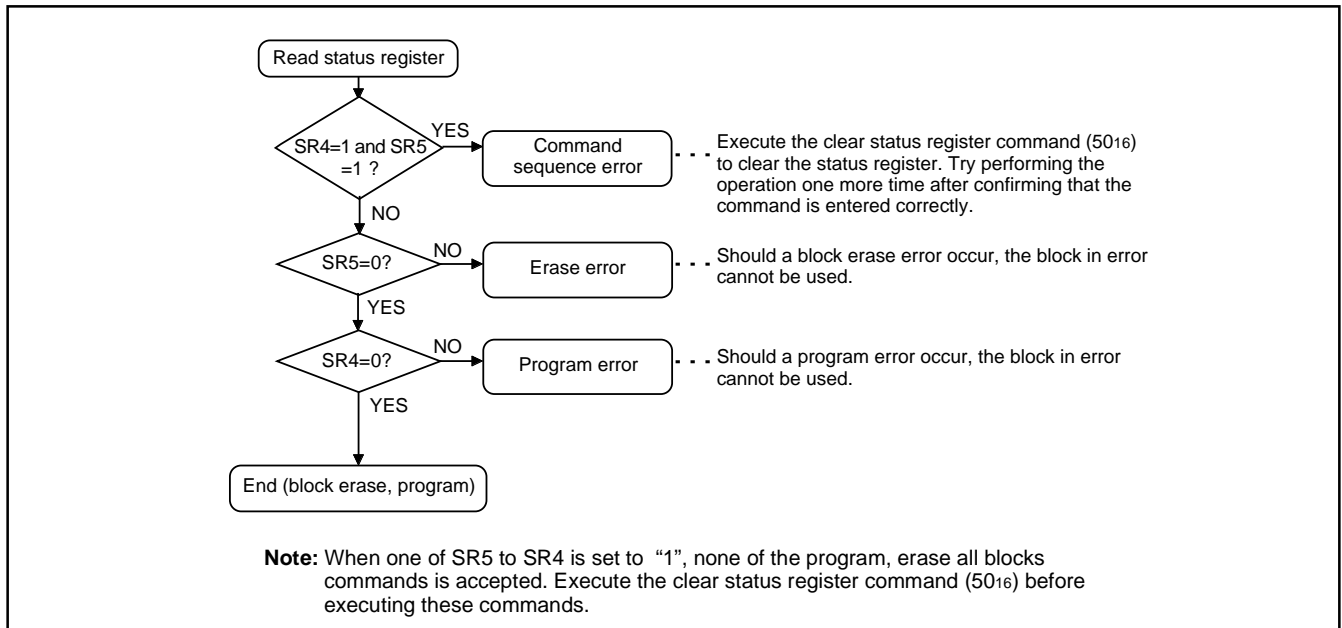


Fig. 66 Full status check flowchart and remedial procedure for errors

### Example Circuit Application for The Standard Serial I/O Mode

Figure 67 shows a circuit application for the standard serial I/O mode. Control pins will vary according to programmer, therefore see the peripheral unit manual for more information.

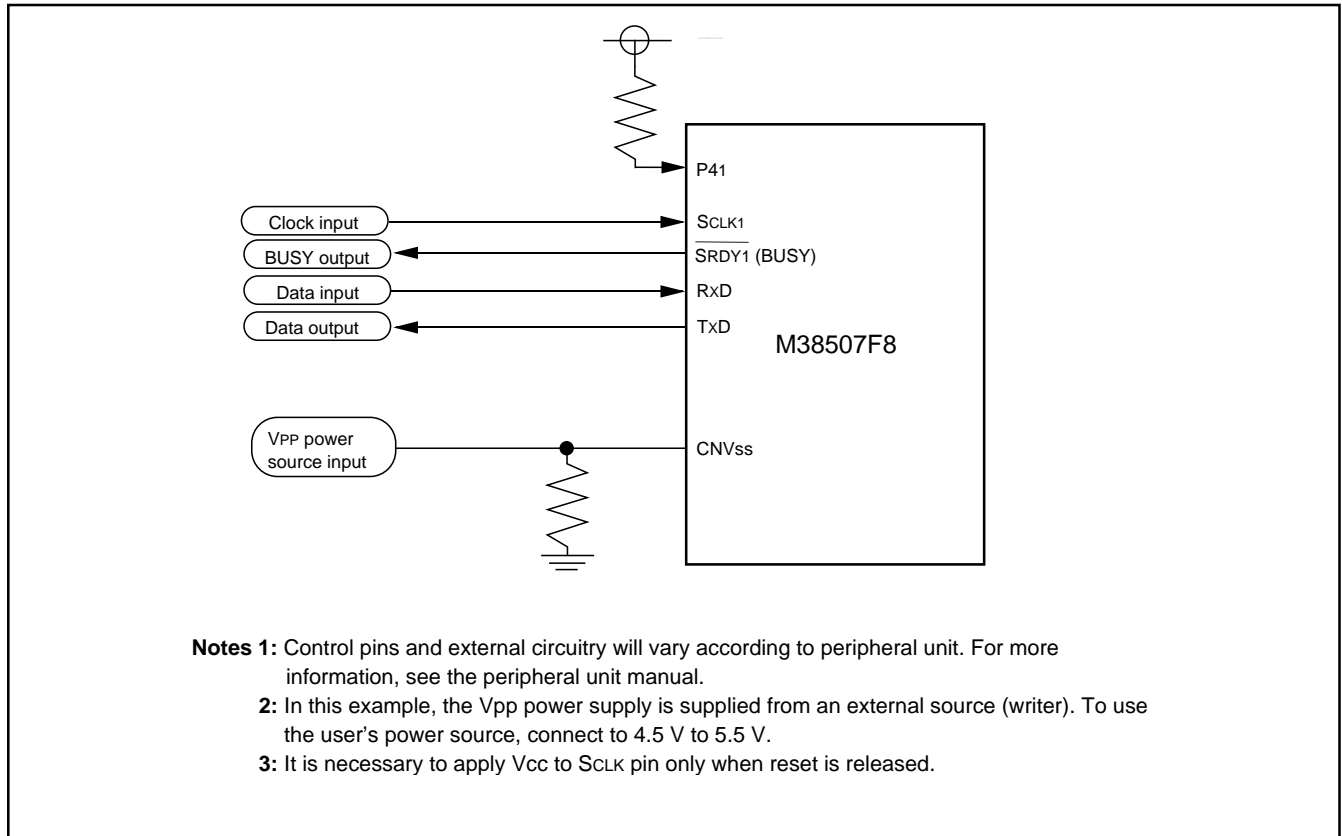


Fig. 67 Example circuit application for the standard serial I/O mode

# HARDWARE

## FUNCTIONAL DESCRIPTION

### Flash memory Electrical characteristics

**Table 20 Absolute maximum ratings**

Symbol	Parameter	Conditions	Ratings	Unit
VCC	Power source voltage		-0.3 to 6.5	V
VI	Input voltage P00–P07, P10–P17, P20, P21, P24–P27, P30–P34, P40–P44, VREF	All voltages are based on Vss. Output transistors are cut off.	-0.3 to VCC +0.3	V
VI	Input voltage P22, P23		-0.3 to 5.8	V
VI	Input voltage RESET, XIN		-0.3 to VCC +0.3	V
VI	Input voltage CNVSS		-0.3 to VCC +0.3	V
VO	Output voltage P00–P07, P10–P17, P20, P21, P24–P27, P30–P34, P40–P44, XOUT		-0.3 to VCC +0.3	V
VO	Output voltage P22, P23		-0.3 to 5.8	V
Pd	Power dissipation	Ta = 25 °C	1000 (Note)	mW
Topr	Operating temperature		25±5	°C
Tstg	Storage temperature		-40 to 125	°C

**Note:** The rating becomes 300 mW at the 42P2R-A/E package.

**Table 21 Flash memory mode Electrical characteristics**  
(Ta = 25°C, VCC = 4.5 to 5.5V unless otherwise noted)

Symbol	Parameter	Conditions	Limits			Unit
			Min.	Typ.	Max.	
I <sub>PP1</sub>	V <sub>PP</sub> power source current (read)	V <sub>PP</sub> = V <sub>CC</sub>			100	μA
I <sub>PP2</sub>	V <sub>PP</sub> power source current (program)	V <sub>PP</sub> = V <sub>CC</sub>			60	mA
I <sub>PP3</sub>	V <sub>PP</sub> power source current (erase)	V <sub>PP</sub> = V <sub>CC</sub>			30	mA
V <sub>IL</sub>	“L” input voltage (Note)		0		0.8	V
V <sub>IH</sub>	“H” input voltage (Note)		2.0		V <sub>CC</sub>	V
V <sub>PP</sub>	V <sub>PP</sub> power source voltage		4.5		5.5	V
V <sub>CC</sub>	V <sub>CC</sub> power source voltage	Microcomputer mode operation at V <sub>CC</sub> = 2.7 to 5.5V	4.5		5.5	V
		Microcomputer mode operation at V <sub>CC</sub> = 2.7 to 3.6V	3.0		3.6	V

**Note:** Input pins for parallel I/O mode.

### AC Electrical characteristics

(Ta = 25°C, Vcc = 4.5 to 5.5V unless otherwise noted)

**Table 22 Read-only mode**

Symbol	Parameter	Limits			Unit
		Min.	Typ.	Max.	
tRC	Read cycle time	200			ns
ta (AD)	Address access time			100	ns
ta (CE)	$\overline{CE}$ access time			100	ns
ta (OE)	$\overline{OE}$ access time			80	ns
tCLZ	Output enable time (after $\overline{CE}$ )	0			ns
tDF(CE)	Output floating time (after $\overline{CE}$ )			25	ns
tOLZ	Output enable time (after $\overline{OE}$ )	0			ns
tDF(OE)	Output floating time (after $\overline{OE}$ )			25	ns
tPHZ	Output floating time (after $\overline{PR}$ )			300	ns
tOH	Output valid time (after $\overline{CE}$ , $\overline{OE}$ , address)	0			ns
tOEH	Write recovery time (before read)	200			ns
tPS	$\overline{RP}$ recovery time	10			$\mu$ s

**Note :** Timing measurement condition is showed in Figure 68.

**Table 23 Read / Write mode (WE control)**

Symbol	Parameter	Limits			Unit
		Min.	Typ.	Max.	
tWC	Write cycle time	200			ns
tAS	Address set up time	100			ns
tAH	Address hold time	25			ns
tDS	Data set up time	100			ns
tDH	Data hold time	25			ns
tCS	$\overline{CE}$ set up time	0			ns
tCH	$\overline{CE}$ hold time	0			ns
tWP	$\overline{WE}$ pulse width	100			ns
tWPH	"H" write pulse width	50			ns
tDAP	Program time		25		$\mu$ s
tDAE	Erase all blocks time		1.5		s
tWHRL	RY/BY delay time			200	ns
tPS	$\overline{RP}$ recovery time	10			$\mu$ s

**Note :** The read timing parameter in the command write operation mode is same as that of the read-only mode.  
Typical value is at Vcc = 5.0 V, Ta = 25 °C condition.

# HARDWARE

## FUNCTIONAL DESCRIPTION

Flash memory mode Electrical characteristics  
(Ta = 25°C, Vcc = 4.5 to 5.5V unless otherwise noted)

Table 24 Read / Write mode (CE control)

Symbol	Parameter	Limits			Unit
		Min.	Typ.	Max.	
tWC	Write cycle time	200			ns
tAS	Address set up time	100			ns
tAH	Address hold time	25			ns
tDS	Data set up time	100			ns
tDH	Data hold time	25			ns
tWS	WE set up time	0			ns
tWH	WE hold time	0			ns
tCEP	CE pulse width	100			ns
tCEPH	"H" CE pulse width	50			ns
tDAP	Program time		25		μs
tDAE	Erase all blocks time		1.5		s
tEHL	RY/BY delay time			200	ns
tPS	RP recovery time	10			μs

**Note :** The read timing parameter in the command write operation mode is same as that of the read-only mode.  
Typical value is at Vcc = 5.0 V, Ta = 25 °C condition.

Table 25 Erase and program operation

Parameter	Min.	Typ.	Max.	Unit
Erase all blocks time		1.5		s
Block erase time		1.0		s
Program time (1byte)		25		μs

Table 26 Vcc power up / power down timing

Symbol	Parameter	Min.	Typ.	Max.	Unit
tvCS	RP = VIH set up time (after rised Vcc = Vcc min.)	10			μs

**Note :** Miserase or miswrite may happen, in case of noise pulse due to the power supply on or off is input to the control pins. Therefore disabling the write mode is need for prevent from memory data break at the power supply on or off. 10μs (min.) waiting time is need to initiate read or write operation after Vcc rises to Vcc min. at power supply on. The memory data is protected owing to keep the RP pin VIL level at power supply off. The RP pin must be kept VIL level for 10μs (min.) after Vcc rises to Vcc min. at the power supply on. The RP pin must be kept VIL level until the Vcc falls to the GND level at power supply off. RP pin doesn't have latch mode, so RP pin must be kept VIH level during read, erase and program operation.

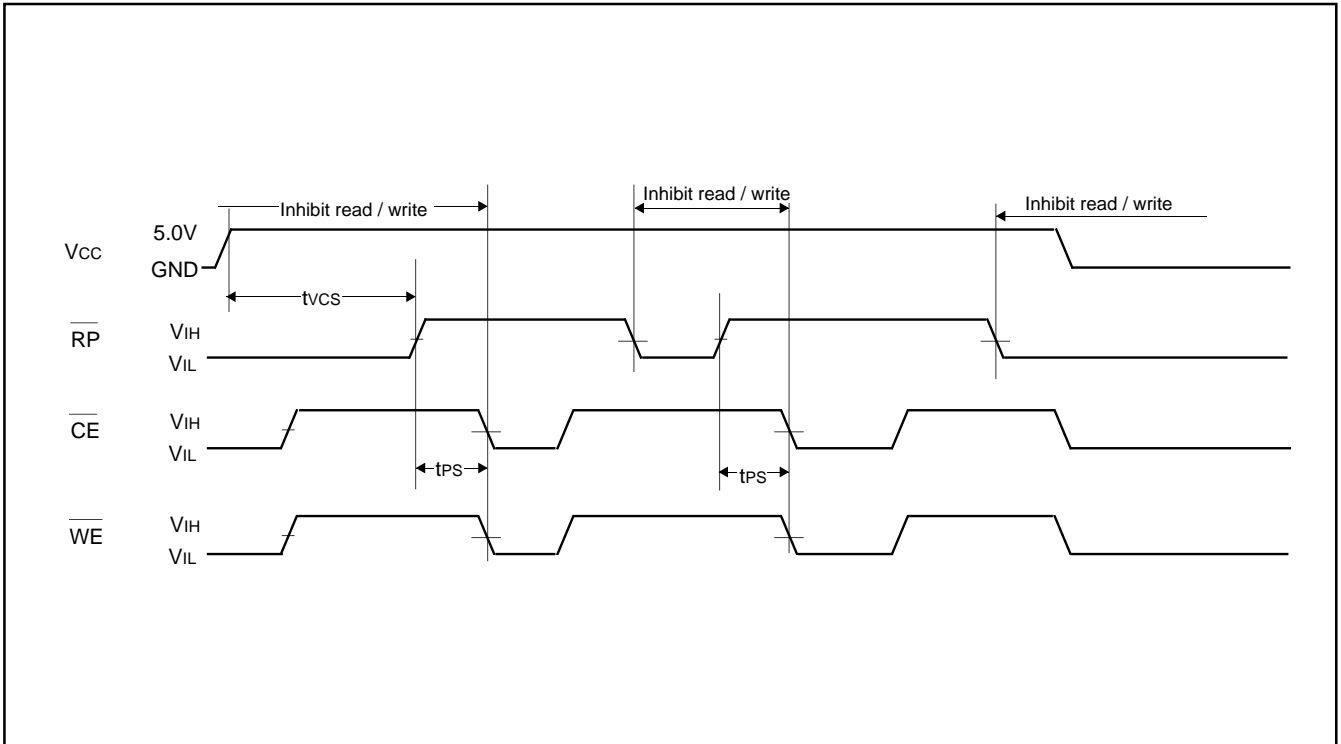


Fig. 68 Vcc power up / power down timing

# HARDWARE

## FUNCTIONAL DESCRIPTION

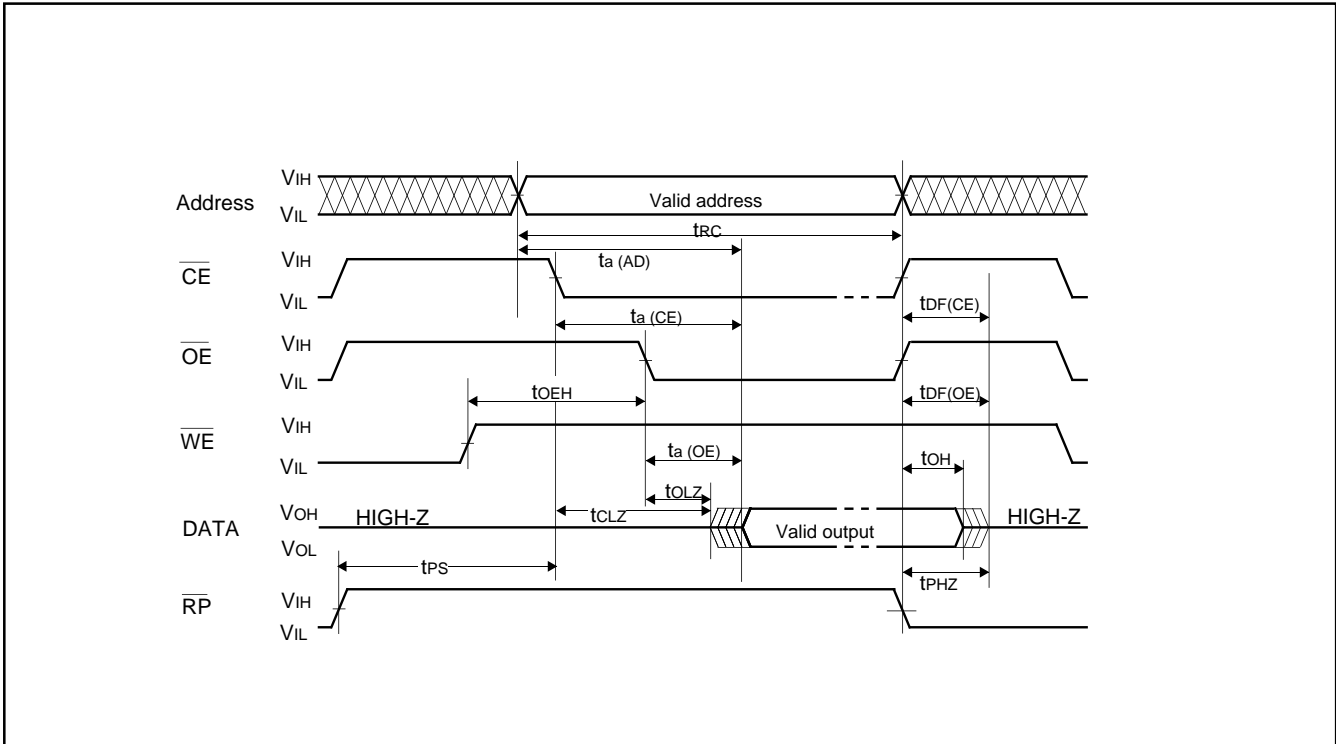


Fig. 69 AC wave for read operation

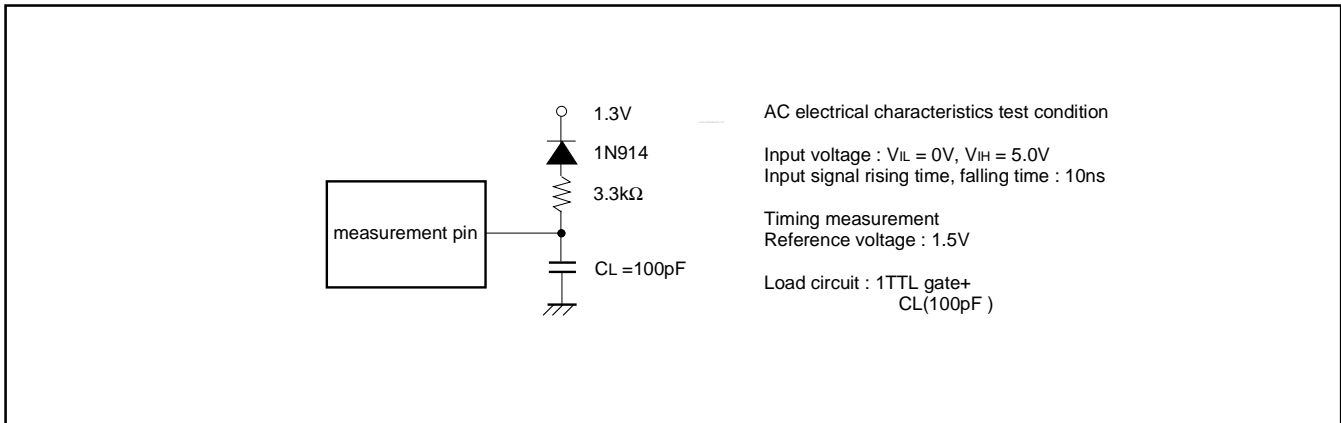


Fig. 70 AC electrical characteristics test condition for read operation



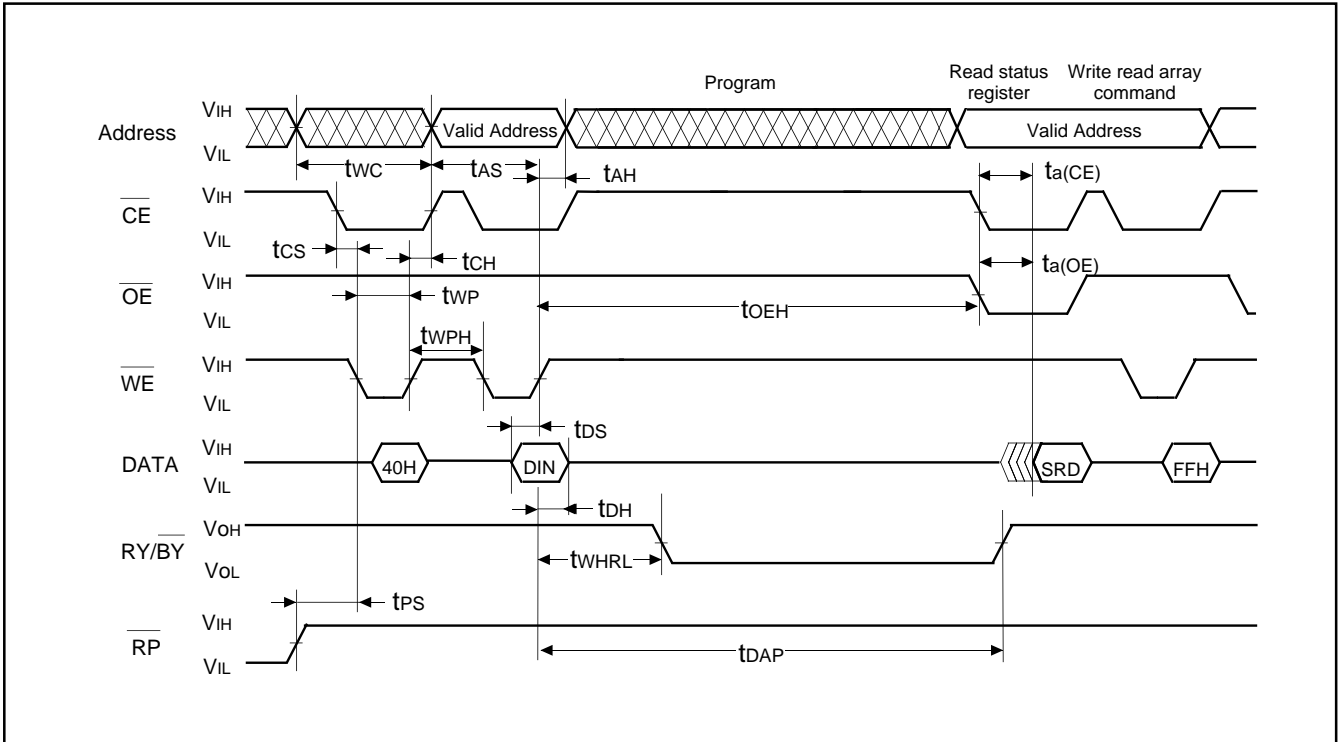


Fig. 71 AC wave for program operation (WE control)

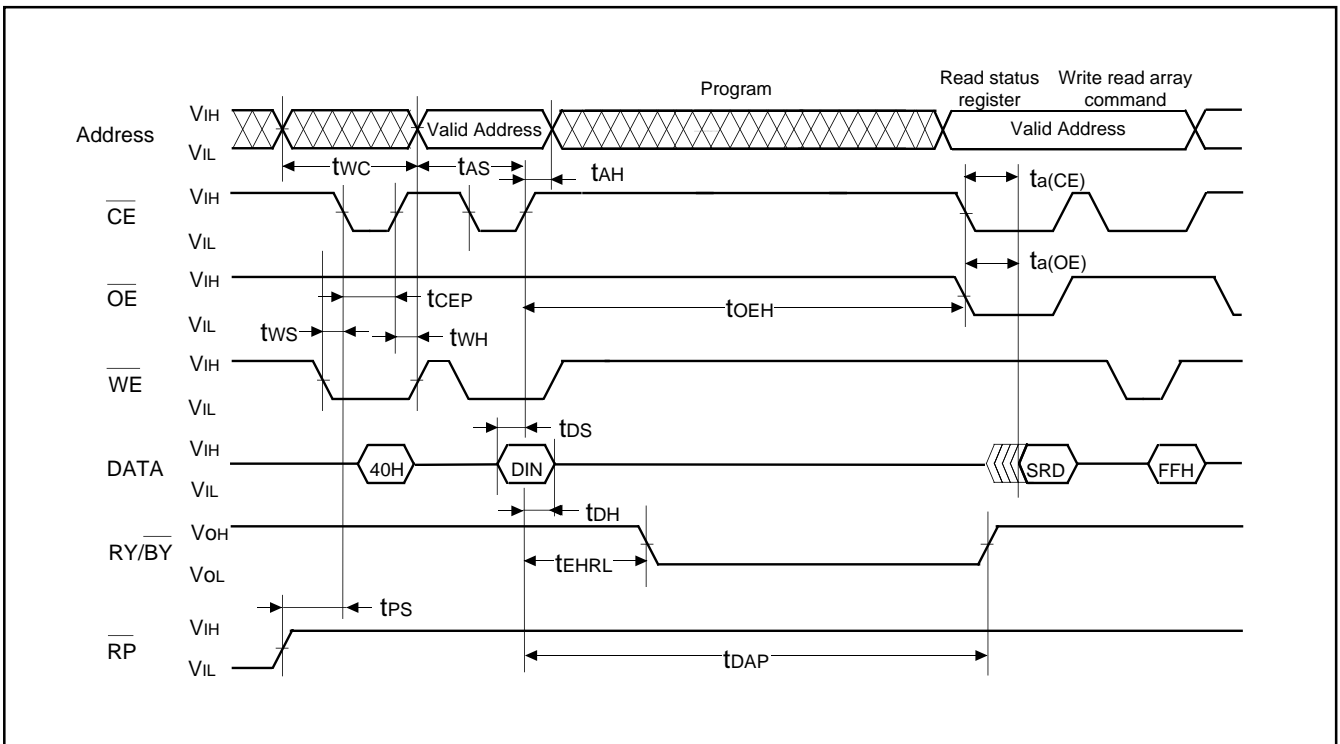


Fig. 72 AC wave for program operation (CE control)

# HARDWARE

## FUNCTIONAL DESCRIPTION

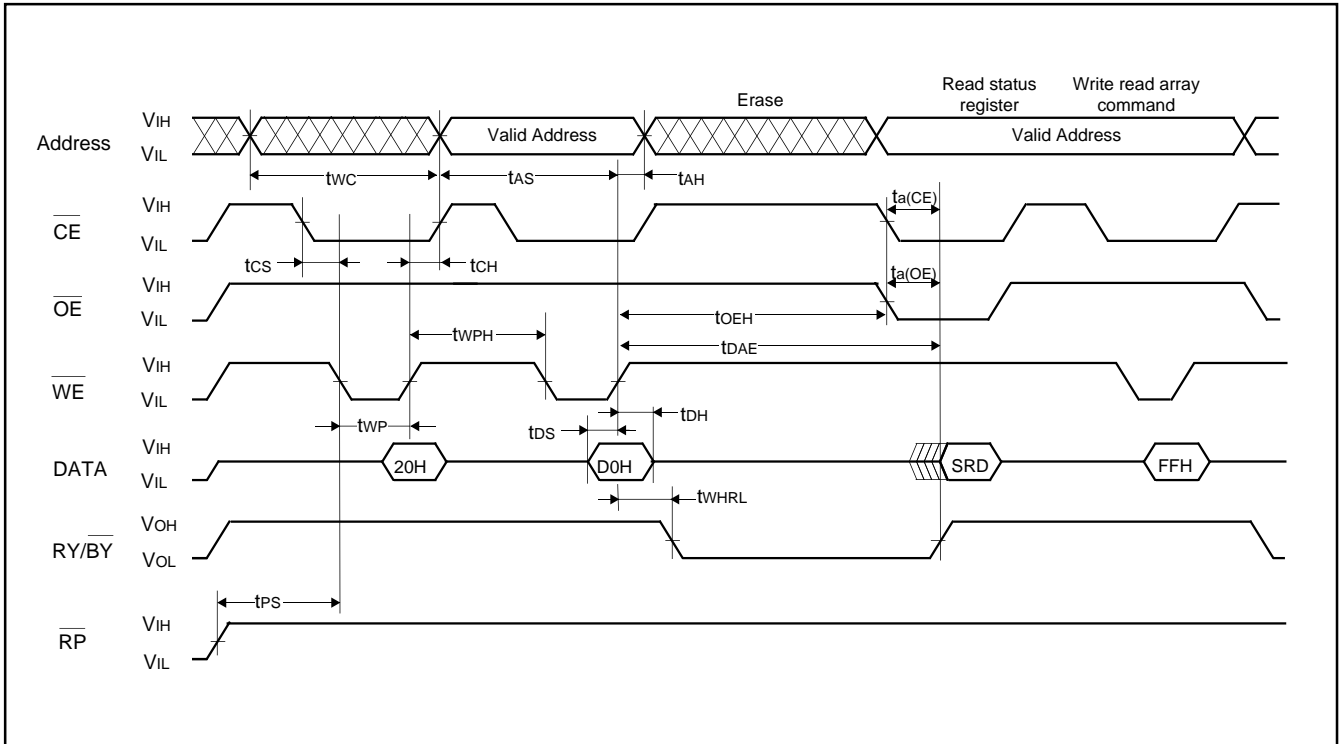


Fig. 73 AC wave for erase operation ( $\overline{WE}$  control)

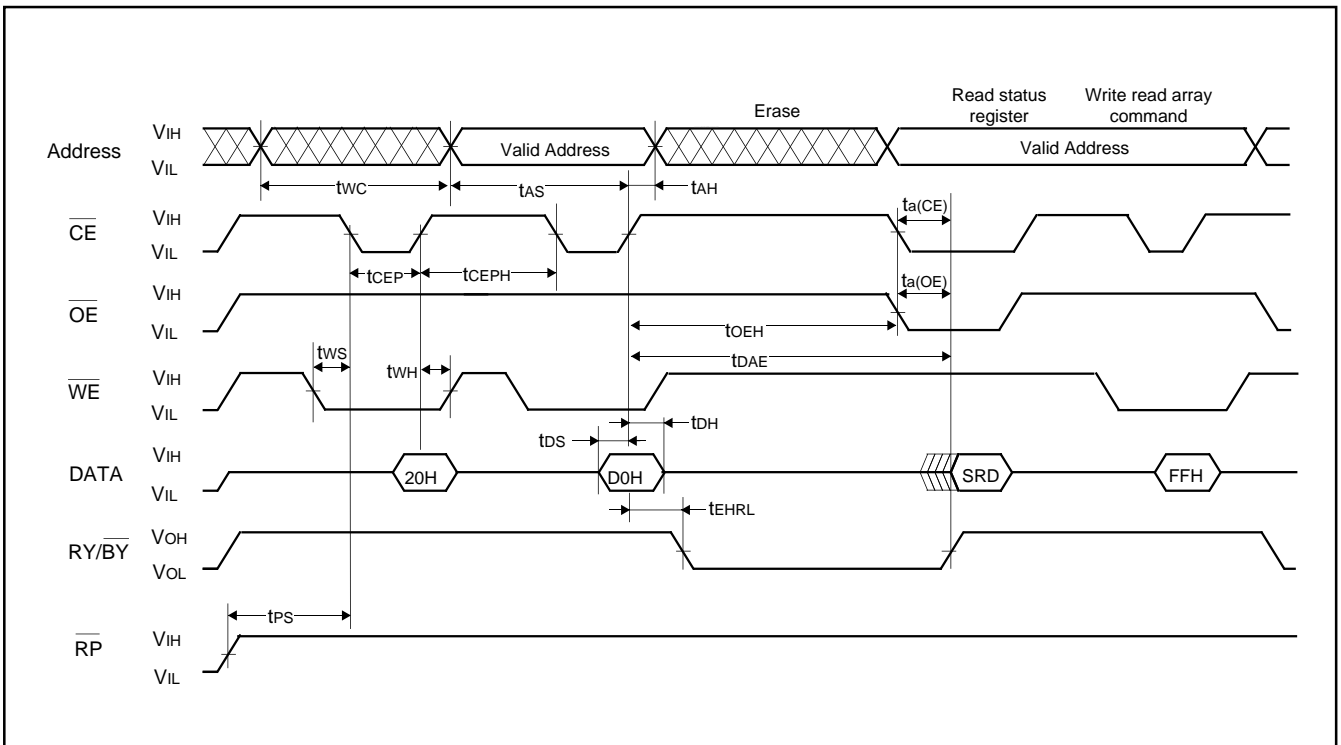


Fig. 74 AC wave for erase operation ( $\overline{CE}$  control)

### NOTES ON PROGRAMMING Processor Status Register

The contents of the processor status register (PS) after a reset are undefined, except for the interrupt disable flag (I) which is "1". After a reset, initialize flags which affect program execution. In particular, it is essential to initialize the index X mode (T) and the decimal mode (D) flags because of their effect on calculations.

### Interrupts

The contents of the interrupt request bits do not change immediately after they have been written. After writing to an interrupt request register, execute at least one instruction before performing a BBC or BBS instruction.

### Decimal Calculations

- To calculate in decimal notation, set the decimal mode flag (D) to "1", then execute an ADC or SBC instruction. After executing an ADC or SBC instruction, execute at least one instruction before executing a SEC, CLC, or CLD instruction.
- In decimal mode, the values of the negative (N), overflow (V), and zero (Z) flags are invalid.

### Timers

If a value n (between 0 and 255) is written to a timer latch, the frequency division ratio is  $1/(n+1)$ .

### Multiplication and Division Instructions

- The index X mode (T) and the decimal mode (D) flags do not affect the MUL and DIV instruction.
- The execution of these instructions does not change the contents of the processor status register.

### Ports

The contents of the port direction registers cannot be read. The following cannot be used:

- The data transfer instruction (LDA, etc.)
- The operation instruction when the index X mode flag (T) is "1"
- The addressing mode which uses the value of a direction register as an index
- The bit-test instruction (BBC or BBS, etc.) to a direction register
- The read-modify-write instructions (ROR, CLB, or SEB, etc.) to a direction register.

Use instructions such as LDM and STA, etc., to set the port direction registers.

### Serial I/O

In clock synchronous serial I/O, if the receive side is using an external clock and it is to output the  $\overline{\text{SRDY}}_1$  signal, set the transmit enable bit, the receive enable bit, and the  $\overline{\text{SRDY}}_1$  output enable bit to "1".

Serial I/O1 continues to output the final bit from the TXD pin after transmission is completed.

SOUT2 pin for serial I/O2 goes to high impedance after transmission is completed.

When an external clock is used as synchronous clock in serial I/O1 or serial I/O2, write transmission data to the transmit buffer register or serial I/O2 register while the transfer clock is "H".

### A-D Converter

The comparator uses capacitive coupling amplifier whose charge will be lost if the clock frequency is too low.

Therefore, make sure that  $f(\text{XIN})$  in the middle/high-speed mode is at least on 500 kHz during an A-D conversion.

Do not execute the STP instruction during an A-D conversion.

### Instruction Execution Time

The instruction execution time is obtained by multiplying the frequency of the internal clock  $\phi$  by the number of cycles needed to execute an instruction.

The number of cycles required to execute an instruction is shown in the list of machine instructions.

The frequency of the internal clock  $\phi$  is half of the XIN frequency in high-speed mode.

### NOTES ON USAGE

#### Differences between 3850 group (standard) and 3850 group (spec. H)

- (1) The absolute maximum ratings of 3850 group (spec. H) is smaller than that of 3850 group (standard).
  - Power source voltage  $V_{cc} = -0.3$  to  $6.5$  V
  - CNVss input voltage  $V_i = -0.3$  to  $V_{cc} + 0.3$  V
- (2) The oscillation circuit constants of XIN-XOUT, XCIN-XCOUT may be some differences between 3850 group (standard) and 3850 group (spec. H).
- (3) Do not write any data to the reserved area and the reserved bit. (Do not change the contents after rest.)
- (4) Fix bit 3 of the CPU mode register to "1".
- (5) Be sure to perform the termination of unused pins.

### Handling of Source Pins

In order to avoid a latch-up occurrence, connect a capacitor suitable for high frequencies as bypass capacitor between power source pin ( $V_{cc}$  pin) and GND pin ( $V_{ss}$  pin) and between power source pin ( $V_{cc}$  pin) and analog power source input pin ( $AV_{ss}$  pin). Besides, connect the capacitor to as close as possible. For bypass capacitor which should not be located too far from the pins to be connected, a ceramic capacitor of  $0.01 \mu\text{F}$ – $0.1 \mu\text{F}$  is recommended.

### EPROM Version/One Time PROM Version/Flash Memory Version

The CNVss pin is connected to the internal memory circuit block by a low-ohmic resistance, since it has the multiplexed function to be a programmable power source pin ( $V_{PP}$  pin) as well.

To improve the noise reduction, connect a track between CNVss pin and  $V_{ss}$  pin or  $V_{cc}$  pin with 1 to  $10 \text{ k}\Omega$  resistance.

The mask ROM version track of CNVss pin has no operational interference even if it is connected to  $V_{ss}$  pin or  $V_{cc}$  pin via a resistor.

# HARDWARE

## DATA REQUIRED FOR MASK ORDERS/DATA REQUIRED FOR One Time PROM PROGRAMMING ORDERS/ROM PROGRAMMING METHOD

### DATA REQUIRED FOR MASK ORDERS

The following are necessary when ordering a mask ROM production:

1. Mask ROM Order Confirmation Form\*
2. Mark Specification Form\*
3. Data to be written to ROM, in EPROM form (three identical copies) or one floppy disk.

### DATA REQUIRED FOR One Time PROM PROGRAMMING ORDERS

The following are necessary when ordering a PROM programming service:

1. ROM Programming Confirmation Form\*
2. Mark Specification Form\* (only special mark with customer's trade mark logo)
3. Data to be programmed to PROM, in EPROM form (three identical copies) or one floppy disk.

\*For the mask ROM confirmation, the ROM programming confirmation form, and the mark specifications, refer to the "Renesas Technology Corp." Homepage (<http://www.renesas.com/en/rom>).

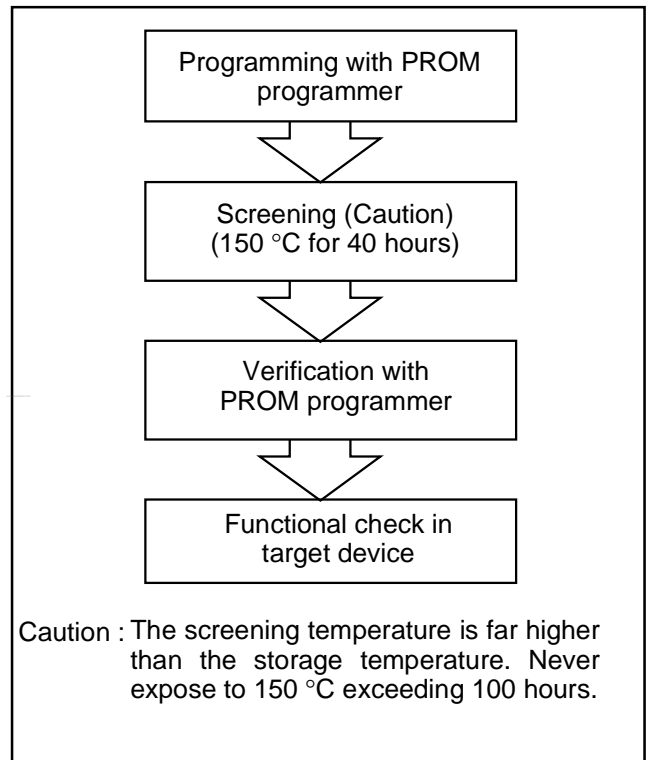
### ROM PROGRAMMING METHOD

The built-in PROM of the blank One Time PROM version and built-in EPROM version can be read or programmed with a general-purpose PROM programmer using a special programming adapter. Set the address of PROM programmer in the user ROM area.

**Table 27 Programming adapter**

Package	Name of Programming Adapter
42P4B, 42S1B	PCA4738S-42A
42P2R-A/E	PCA4738F-42A

The PROM of the blank One Time PROM version is not tested or screened in the assembly process and following processes. To ensure proper operation after programming, the procedure shown in Figure 75 is recommended to verify programming.



**Fig. 75 Programming and testing of One Time PROM version**

### FUNCTIONAL DESCRIPTION SUPPLEMENT A-D Converter

A-D conversion is started by setting AD conversion completion bit to "0". During A-D conversion, internal operations are performed as follows.

1. After the start of A-D conversion, A-D conversion register goes to "0016".
2. The highest-order bit of A-D conversion register is set to "1", and the comparison voltage  $V_{ref}$  is input to the comparator. Then,  $V_{ref}$  is compared with analog input voltage  $V_{IN}$ .
3. As a result of comparison, when  $V_{ref} < V_{IN}$ , the highest-order bit of A-D conversion register becomes "1". When  $V_{ref} > V_{IN}$ , the highest-order bit becomes "0".

By repeating the above operations up to the lowest-order bit of the A-D conversion register, an analog value converts into a digital value.

A-D conversion completes at 61 clock cycles (15.25  $\mu$ s at  $f(XIN) = 8$  MHz) after it is started, and the result of the conversion is stored into the A-D conversion register.

Concurrently with the completion of A-D conversion, A-D conversion interrupt request occurs, so that the AD conversion interrupt request bit is set to "1".

**Table 28 Relative formula for a reference voltage  $V_{REF}$  of A-D converter and  $V_{ref}$**

When $n = 0$	$V_{ref} = 0$
When $n = 1$ to 1023	$V_{ref} = \frac{V_{REF}}{1024} \times n$

n: Value of A-D converter (decimal numeral)

**Table 29 Change of A-D conversion register during A-D conversion**

	Change of A-D conversion register	Value of comparison voltage ( $V_{ref}$ )
<b>At start of conversion</b>	0 0 0 0 0 0 0 0 0 0 0	0
<b>First comparison</b>	1 0 0 0 0 0 0 0 0 0 0	$\frac{V_{REF}}{2}$
<b>Second comparison</b>	*1 1 0 0 0 0 0 0 0 0 0	$\frac{V_{REF}}{2} \pm \frac{V_{REF}}{4}$
<b>Third comparison</b>	*1 *2 1 0 0 0 0 0 0 0 0	$\frac{V_{REF}}{2} \pm \frac{V_{REF}}{4} \pm \frac{V_{REF}}{8}$
≈		
<b>After completion of tenth comparison</b>	A result of A-D conversion *1 *2 *3 *4 *5 *6 *7 *8 *9 *10	$\frac{V_{REF}}{2} \pm \frac{V_{REF}}{4} \pm \dots \pm \frac{V_{REF}}{1024}$

\*1~\*10: A result of the first comparison to the tenth comparison

# HARDWARE

## FUNCTIONAL DESCRIPTION SUPPLEMENT

Figure 76 shows the A-D conversion equivalent circuit, and Figure 77 shows the A-D conversion timing chart.

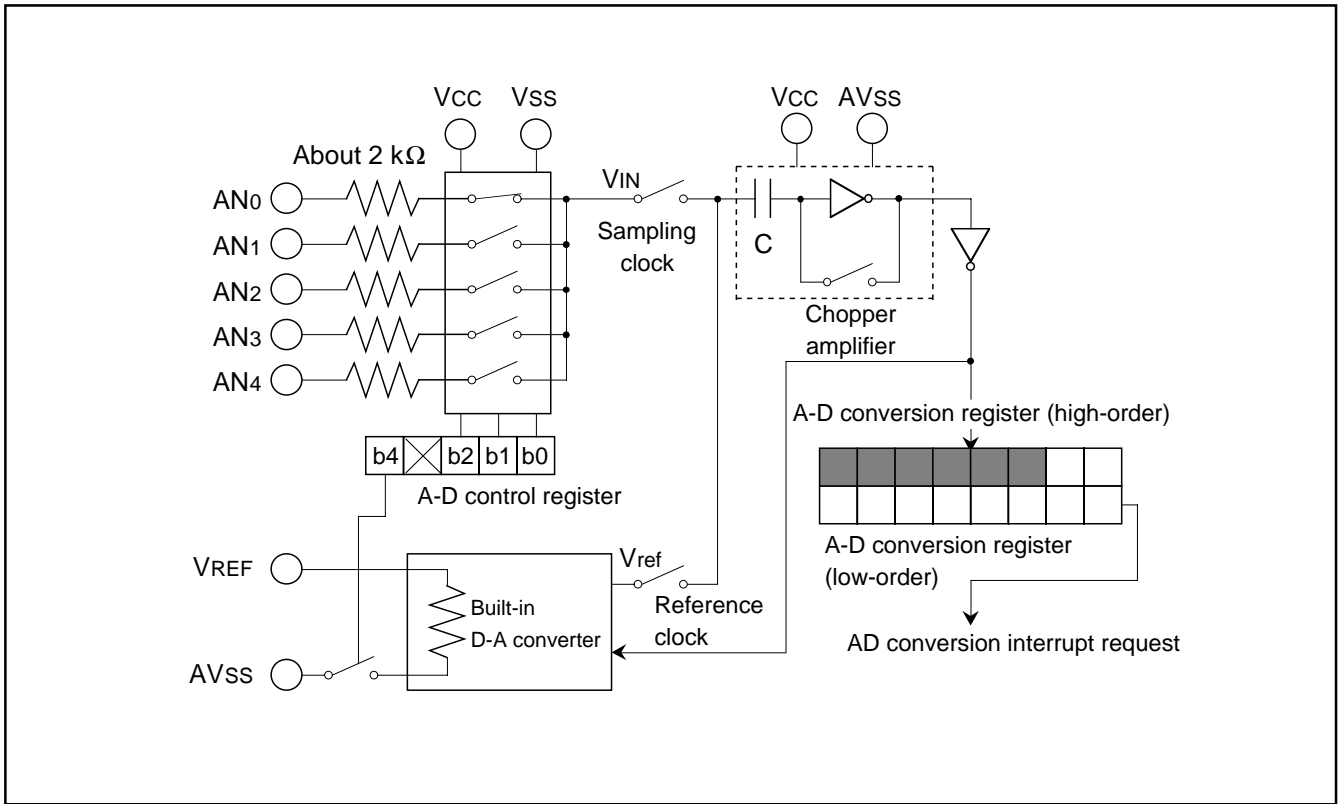


Fig. 76 A-D conversion equivalent circuit

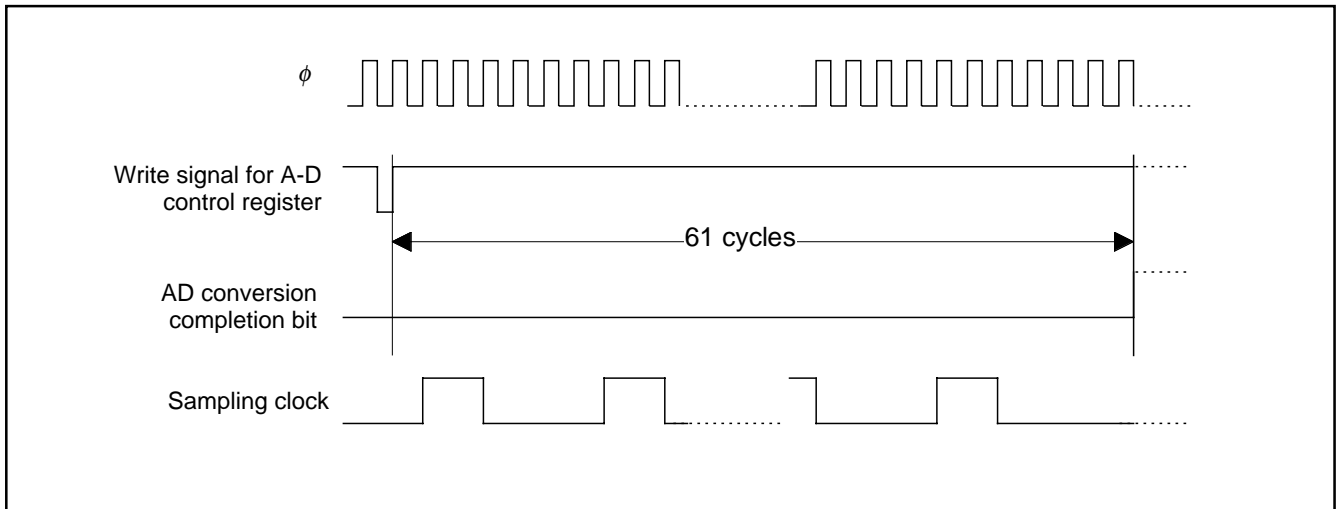


Fig. 77 A-D conversion timing chart



# CHAPTER 2

## **APPLICATION**

- 2.1 I/O port
- 2.2 Interrupt
- 2.3 Timer
- 2.4 Serial I/O
- 2.5 PWM
- 2.6 A-D converter
- 2.7 Watchdog timer
- 2.8 Reset
- 2.9 Clock generating circuit
- 2.10 Standby function
- 2.11 Flash memory mode

# APPLICATION

## 2.1 I/O port

### 2.1 I/O port

This paragraph describes the setting method of I/O port relevant registers, notes etc.

#### 2.1.1 Memory map

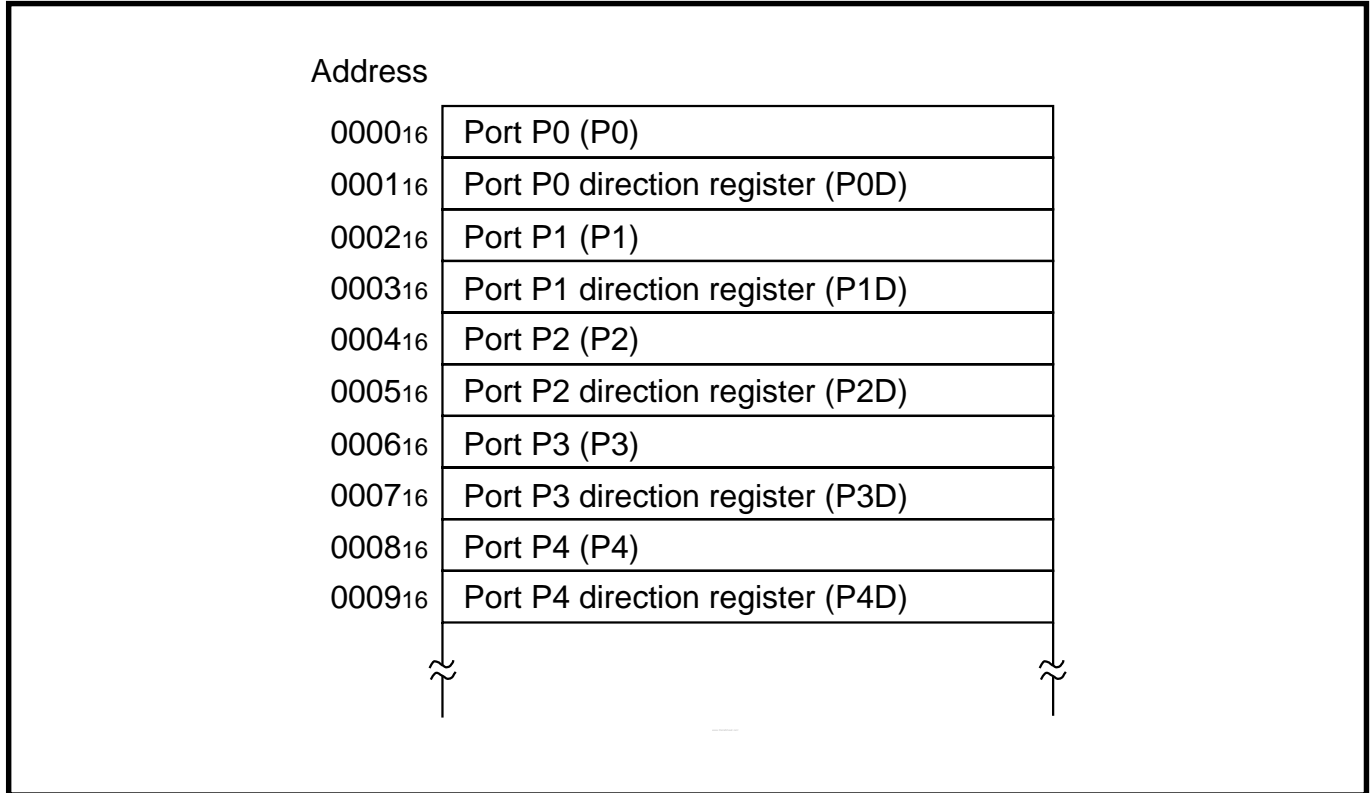


Fig. 2.1.1 Memory map of I/O port relevant registers

#### 2.1.2 Relevant registers

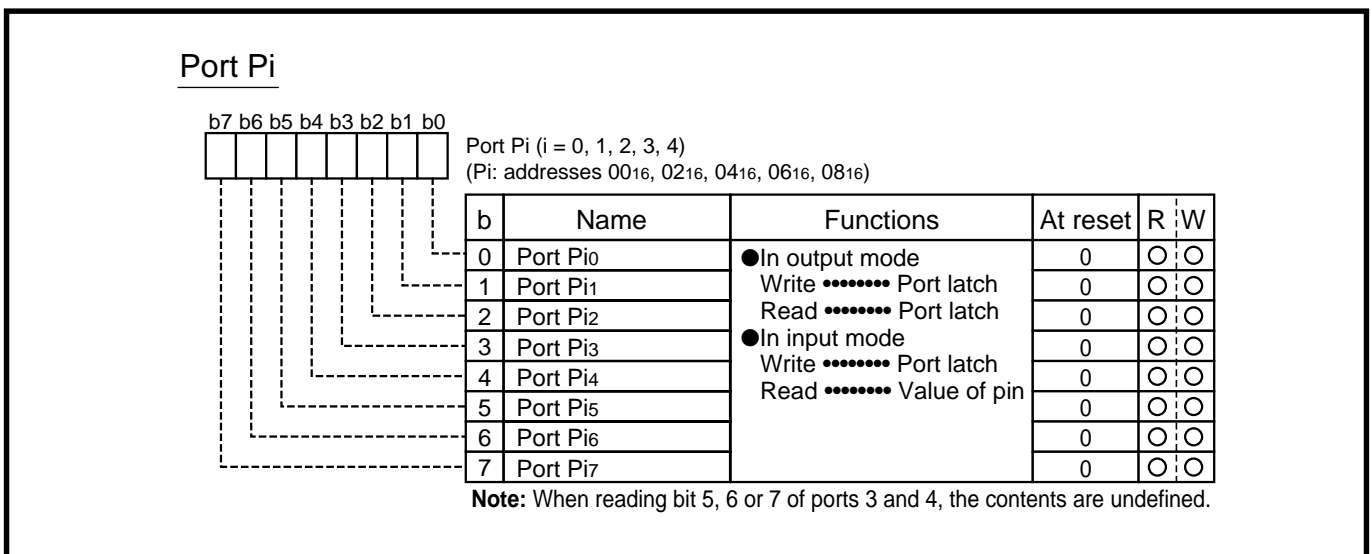


Fig. 2.1.2 Structure of Port Pi (i = 0, 1, 2, 3, 4)



### Port Pi direction register

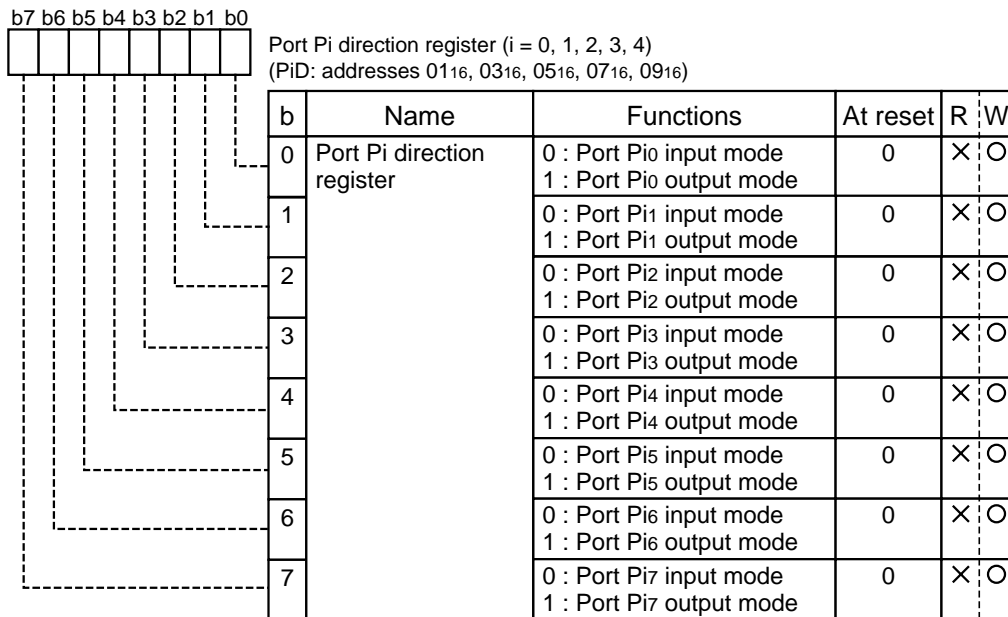


Fig. 2.1.3 Structure of Port Pi direction register ( $i = 0, 1, 2, 3, 4$ )

### 2.1.3 Terminate unused pins

Table 2.1.1 Termination of unused pins

Pins/Ports name	Termination
P0, P1, P2, P3, P4	<ul style="list-style-type: none"> <li>Set to the input mode and connect each to <math>V_{CC}</math> or <math>V_{SS}</math> through a resistor of 1 k<math>\Omega</math> to 10 k<math>\Omega</math>.</li> <li>Set to the output mode and open at "L" or "H" output state.</li> </ul>
$V_{REF}$	Connect to $V_{SS}$ (GND).
$AV_{SS}$	Connect to $V_{SS}$ (GND).
$X_{OUT}$	Open (only when using external clock)

# APPLICATION

## 2.1 I/O port

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### 2.1.4 Notes on I/O port

#### (1) Notes in standby state

In standby state\*<sup>1</sup>, do not make input levels of an I/O port “undefined”, especially for I/O ports of the N-channel open-drain. When setting the N-channel open-drain port as an output, do not make input levels of an I/O port “undefined”, too.

Pull-up (connect the port to VCC) or pull-down (connect the port to VSS) these ports through a resistor.

When determining a resistance value, note the following points:

- External circuit
- Variation of output levels during the ordinary operation

#### ● Reason

When setting as an input port with its direction register, the transistor becomes the OFF state, which causes the ports to be the high-impedance state.

Accordingly, the potential which is input to the input buffer in a microcomputer is unstable in the state that input levels of an I/O port are “undefined”. This may cause power source current.

In I/O ports of N-channel open-drain, when the contents of the port latch are “1”, even if it is set as an output port with its direction register, it becomes the same phenomenon as the case of an input port.

\*1 standby state: stop mode by executing **STP** instruction  
wait mode by executing **WIT** instruction

#### (2) Modifying output data with bit managing instruction

When the port latch of an I/O port is modified with the bit managing instruction\*<sup>2</sup>, the value of the unspecified bit may be changed.

#### ● Reason

The bit managing instructions are read-modify-write form instructions for reading and writing data by a byte unit. Accordingly, when these instructions are executed on a bit of the port latch of an I/O port, the following is executed to all bits of the port latch.

- As for bit which is set for input port:

The pin state is read in the CPU, and is written to this bit after bit managing.

- As for bit which is set for output port:

The bit value is read in the CPU, and is written to this bit after bit managing.

Note the following:

- Even when a port which is set as an output port is changed for an input port, its port latch holds the output data.
- As for a bit of which is set for an input port, its value may be changed even when not specified with a bit managing instruction in case where the pin state differs from its port latch contents.

\*2 Bit managing instructions: **SEB** and **CLB** instructions

---

### 2.1.5 Termination of unused pins

#### (1) Terminate unused pins

① Output ports : Open

② Input ports :

Connect each pin to  $V_{CC}$  or  $V_{SS}$  through each resistor of 1 k $\Omega$  to 10 k $\Omega$ .

A for pins whose potential affects to operation modes such as pins CNV<sub>SS</sub>, INT or others, select the  $V_{CC}$  pin or the  $V_{SS}$  pin according to their operation mode.

③ I/O ports :

- Set the I/O ports for the input mode and connect them to  $V_{CC}$  or  $V_{SS}$  through each resistor of 1 k $\Omega$  to 10 k $\Omega$ .

Set the I/O ports for the output mode and open them at "L" or "H".

- When opening them in the output mode, the input mode of the initial status remains until the mode of the ports is switched over to the output mode by the program after reset. Thus, the potential at these pins is undefined and the power source current may increase in the input mode. With regard to an effects on the system, thoroughly perform system evaluation on the user side.

- Since the direction register setup may be changed because of a program runaway or noise, set direction registers by program periodically to increase the reliability of program.

④ The AV<sub>SS</sub> pin when not using the A-D converter :

- When not using the A-D converter, handle a power source pin for the A-D converter, AV<sub>SS</sub> pin as follows:

AV<sub>SS</sub>: Connect to the  $V_{SS}$  pin.

#### (2) Termination remarks

① Input ports and I/O ports :

Do not open in the input mode.

● Reason

- The power source current may increase depending on the first-stage circuit.
- An effect due to noise may be easily produced as compared with proper termination ② and ③ shown on the above.

② I/O ports :

When setting for the input mode, do not connect to  $V_{CC}$  or  $V_{SS}$  directly.

● Reason

If the direction register setup changes for the output mode because of a program runaway or noise, a short circuit may occur between a port and  $V_{CC}$  (or  $V_{SS}$ ).

③ I/O ports :

When setting for the input mode, do not connect multiple ports in a lump to  $V_{CC}$  or  $V_{SS}$  through a resistor.

● Reason

If the direction register setup changes for the output mode because of a program runaway or noise, a short circuit may occur between ports.

- At the termination of unused pins, perform wiring at the shortest possible distance (20 mm or less) from microcomputer pins.

# APPLICATION

## 2.2 Interrupt

---

### 2.2 Interrupt

This paragraph explains the registers setting method and the notes relevant to the interrupt.

#### 2.2.1 Memory map

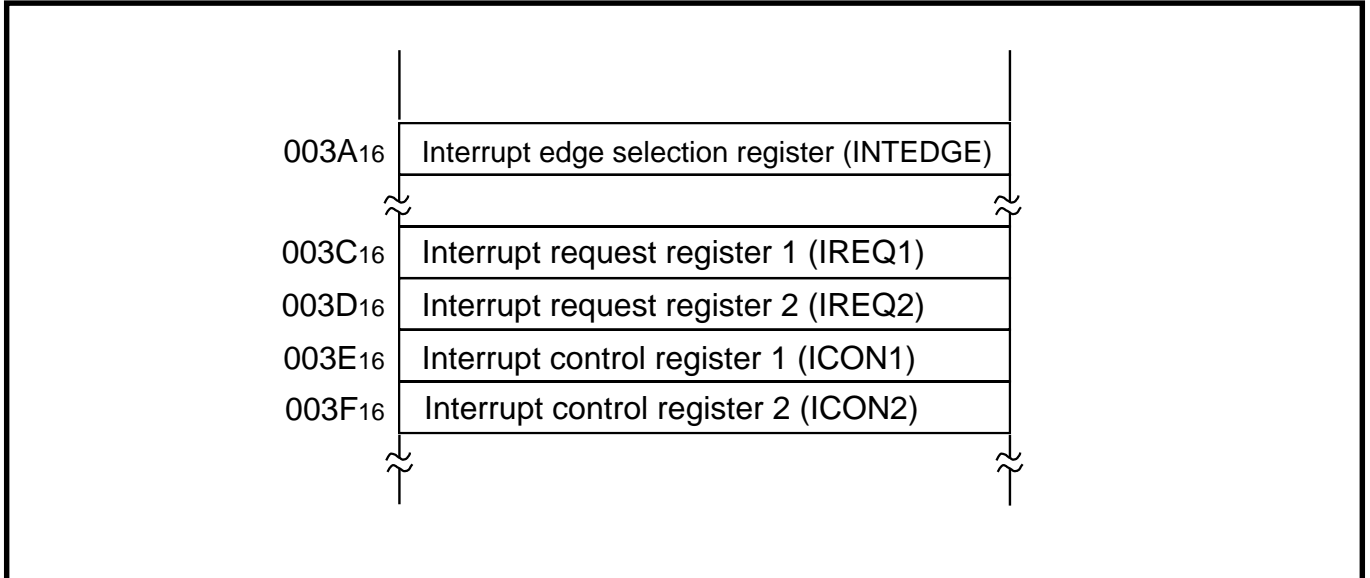


Fig. 2.2.1 Memory map of registers relevant to interrupt

## 2.2.2 Relevant registers

## Interrupt edge selection register

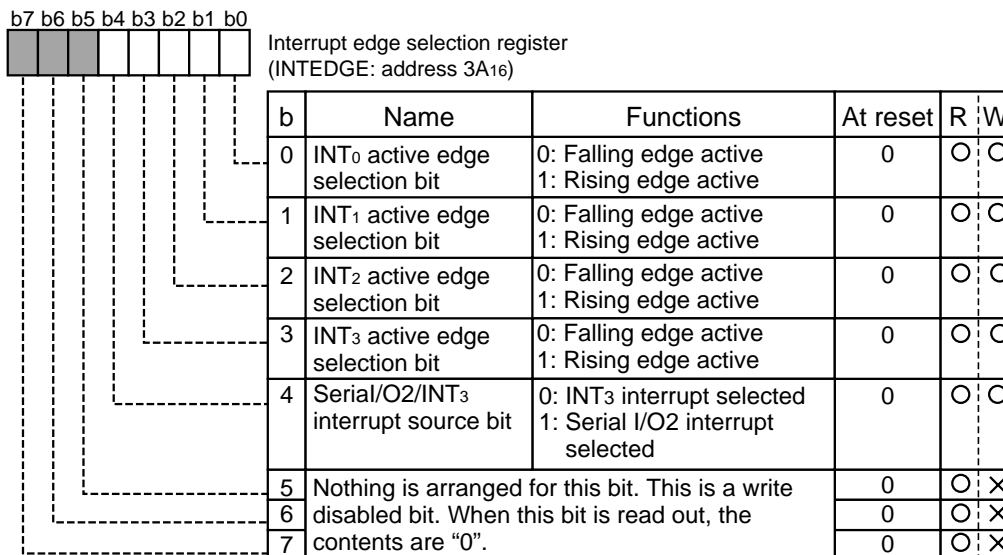


Fig. 2.2.2 Structure of Interrupt edge selection register

# APPLICATION

## 2.2 Interrupt

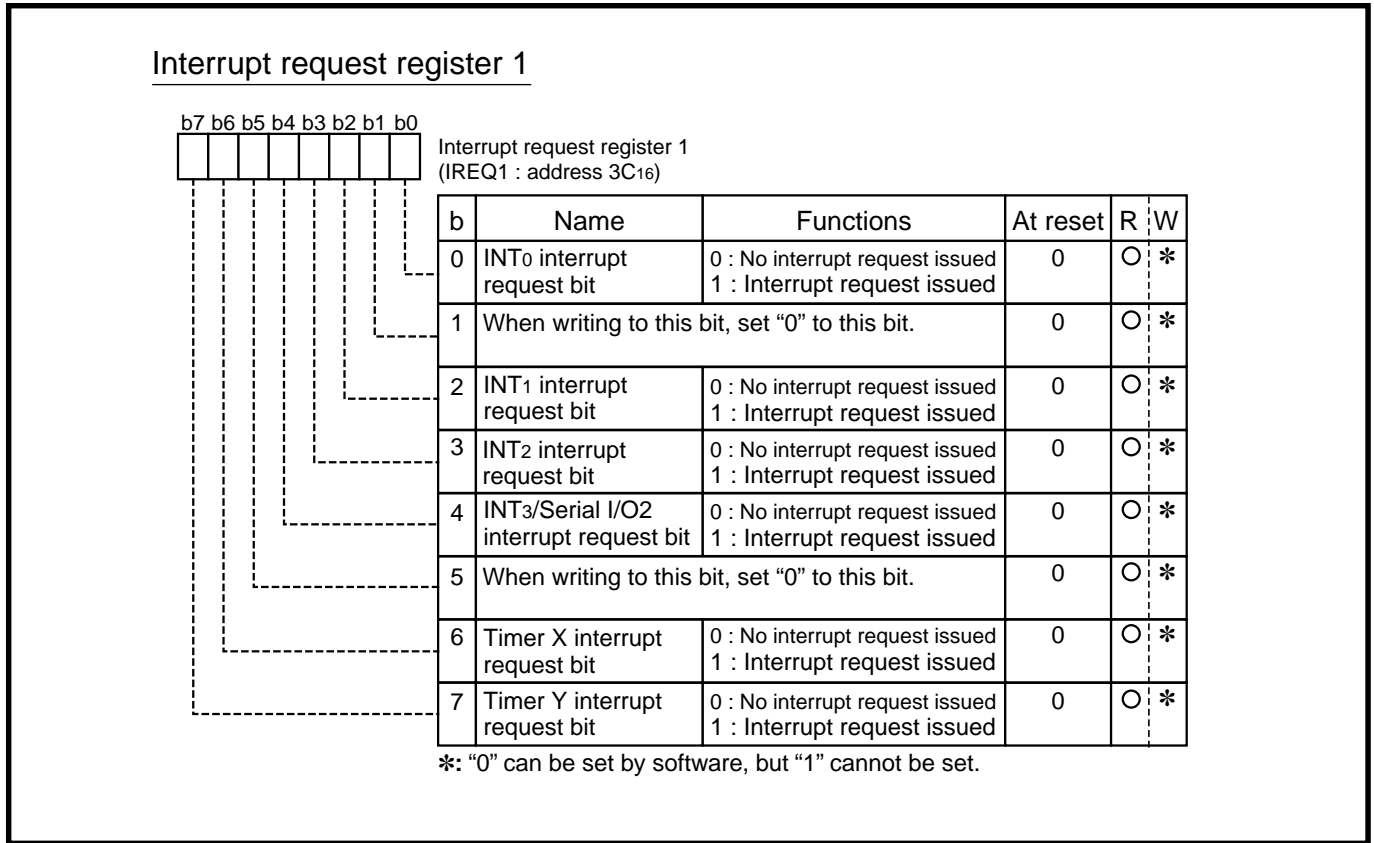


Fig. 2.2.3 Structure of Interrupt request register 1

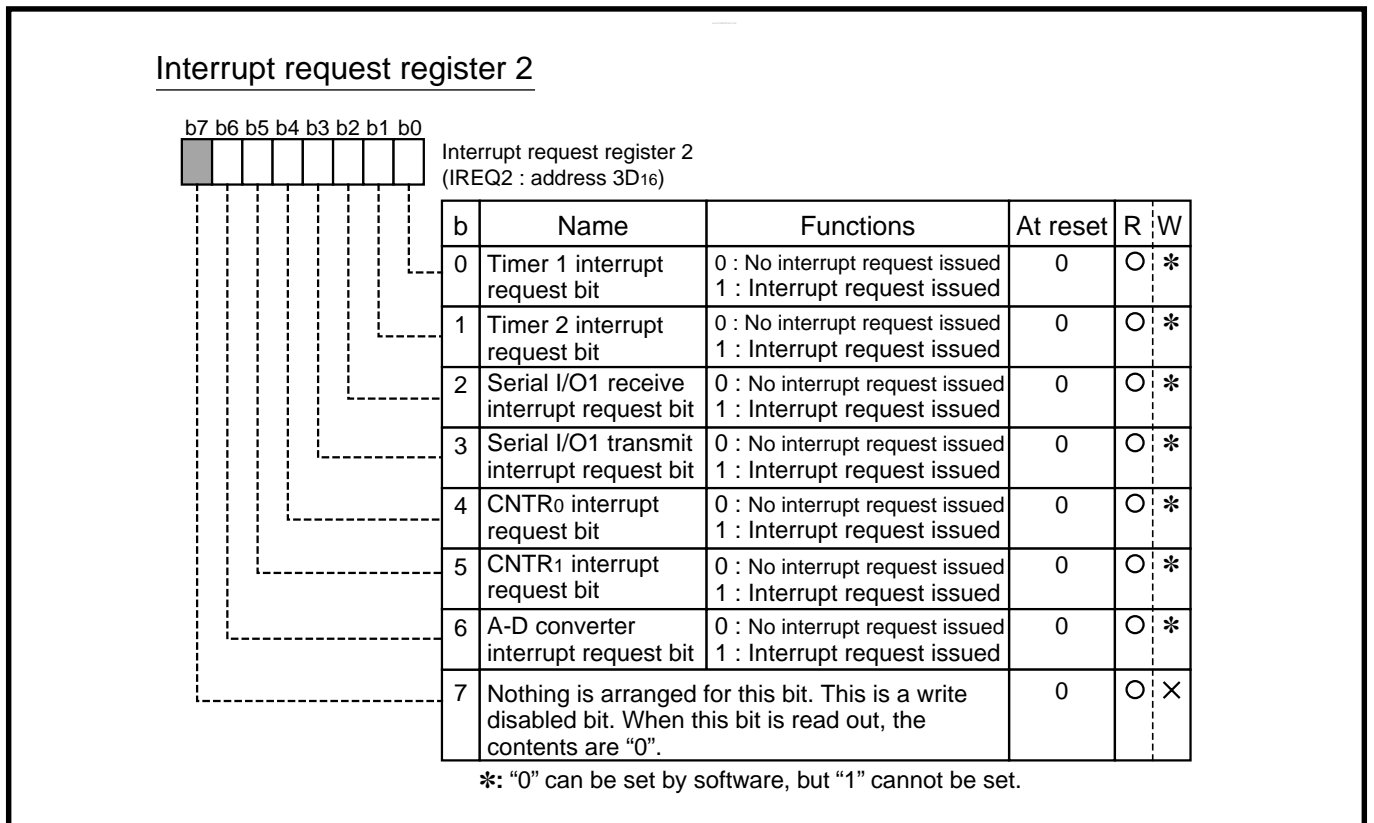


Fig. 2.2.4 Structure of Interrupt request register 2

### Interrupt control register 1

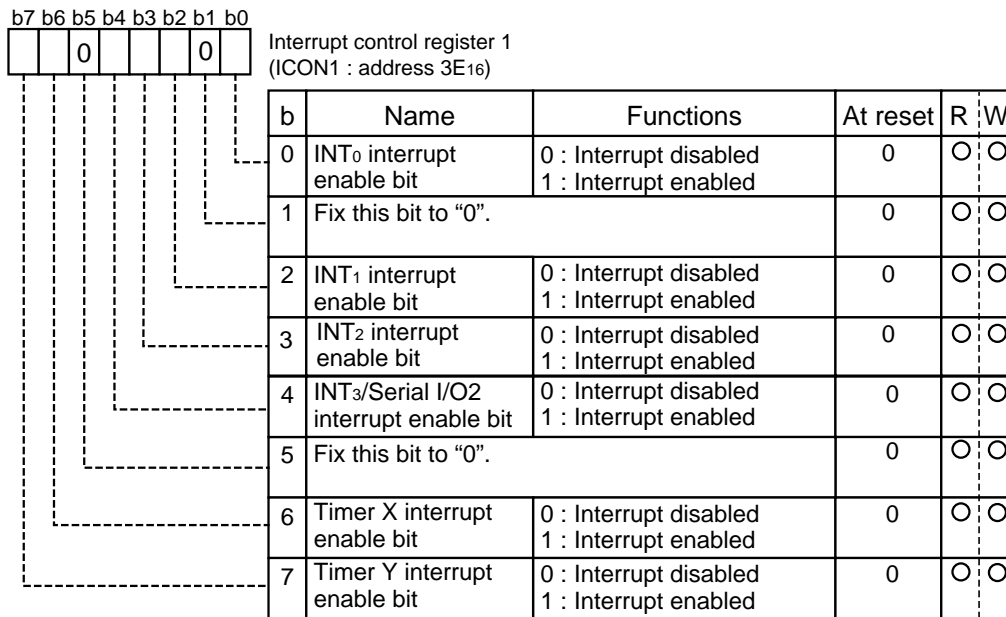


Fig. 2.2.5 Structure of Interrupt control register 1

### Interrupt control register 2

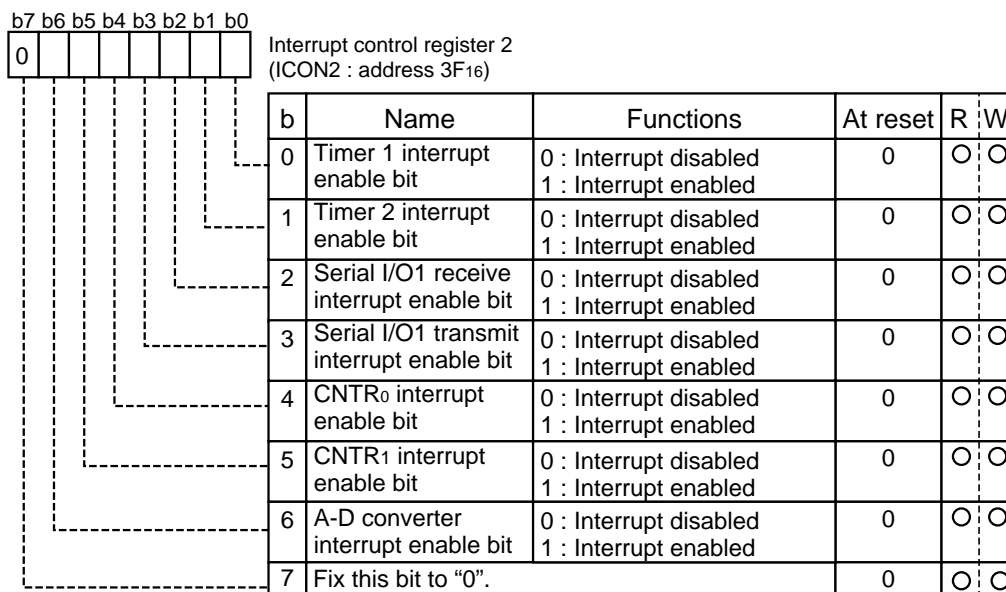


Fig. 2.2.6 Structure of Interrupt control register 2

# APPLICATION

## 2.2 Interrupt

### 2.2.3 Interrupt source

The 3850 group permits interrupts of 15 sources. These are vector interrupts with a fixed priority system. Accordingly, when two or more interrupt requests occur during the same sampling, the higher-priority interrupt is accepted first. This priority is determined by hardware, but a variety of priority processing can be performed by software, using an interrupt enable bit and an interrupt disable flag.

For interrupt sources, vector addresses and interrupt priority, refer to Table 2.2.1.

**Table 2.2.1 Interrupt sources, vector addresses and priority of 3850 group**

Interrupt Source	Priority	Vector Addresses (Note 1)		Interrupt Request Generating Conditions	Remarks
		High	Low		
Reset (Note 2)	1	FFFD <sub>16</sub>	FFFC <sub>16</sub>	At reset	Non-maskable
INT <sub>0</sub>	2	FFFB <sub>16</sub>	FFFA <sub>16</sub>	At detection of either rising or falling edge of INT <sub>0</sub> input	External interrupt (active edge selectable)
Reserved	3	FFF9 <sub>16</sub>	FFF8 <sub>16</sub>	Reserved	
INT <sub>1</sub>	4	FFF7 <sub>16</sub>	FFF6 <sub>16</sub>	At detection of either rising or falling edge of INT <sub>1</sub> input	External interrupt (active edge selectable)
INT <sub>2</sub>	5	FFF5 <sub>16</sub>	FFF4 <sub>16</sub>	At detection of either rising or falling edge of INT <sub>2</sub> input	External interrupt (active edge selectable)
INT <sub>3</sub>	6	FFF3 <sub>16</sub>	FFF2 <sub>16</sub>	At detection of either rising or falling edge of INT <sub>3</sub> input	External interrupt (active edge selectable)
Serial I/O <sub>2</sub>				At completion of serial I/O <sub>2</sub> data transfer	Switch by Serial I/O <sub>2</sub> /INT <sub>3</sub> interrupt source bit
Reserved	7	FFF1 <sub>16</sub>	FFF0 <sub>16</sub>	Reserved	
Timer X	8	FFEF <sub>16</sub>	FFEE <sub>16</sub>	At timer X underflow	
Timer Y	9	FFED <sub>16</sub>	FFEC <sub>16</sub>	At timer Y underflow	
Timer 1	10	FFEB <sub>16</sub>	FFEA <sub>16</sub>	At timer 1 underflow	STP release timer underflow
Timer 2	11	FFE9 <sub>16</sub>	FFE8 <sub>16</sub>	At timer 2 underflow	
Serial I/O <sub>1</sub> received	12	FFE7 <sub>16</sub>	FFE6 <sub>16</sub>	At completion of serial I/O <sub>1</sub> data reception	Valid when serial I/O <sub>1</sub> is selected
Serial I/O <sub>1</sub> transmit	13	FFE5 <sub>16</sub>	FFE4 <sub>16</sub>	At completion of serial I/O <sub>1</sub> transfer shift or when transmission buffer is empty	Valid when serial I/O <sub>1</sub> is selected
CNTR <sub>0</sub>	14	FFE3 <sub>16</sub>	FFE2 <sub>16</sub>	At detection of either rising or falling edge of CNTR <sub>0</sub> input	External interrupt (active edge selectable)
CNTR <sub>1</sub>	15	FFE1 <sub>16</sub>	FFE0 <sub>16</sub>	At detection of either rising or falling edge of CNTR <sub>1</sub> input	External interrupt (active edge selectable)
A-D converter	16	FFDF <sub>16</sub>	FFDE <sub>16</sub>	At completion of A-D conversion	
BRK instruction	17	FFDD <sub>16</sub>	FFDC <sub>16</sub>	At BRK instruction execution	Non-maskable software interrupt

**Notes 1:** Vector addresses contain interrupt jump destination addresses.

**2:** Reset function in the same way as an interrupt with the highest priority.



### 2.2.4 Interrupt operation

When an interrupt request is accepted, the contents of the following registers just before acceptance of the interrupt requests are automatically pushed onto the stack area in the order of ①, ② and ③.

- ① High-order contents of program counter (PC<sub>H</sub>)
- ② Low-order contents of program counter (PC<sub>L</sub>)
- ③ Contents of processor status register (PS)

After the contents of the above registers are pushed onto the stack area, the accepted interrupt vector address enters the program counter and consequently the interrupt processing routine is executed.

When the RTI instruction is executed at the end of the interrupt processing routine, the contents of the above registers pushed onto the stack area are restored to the respective registers in the order of ③, ② and ①; and the microcomputer resumes the processing executed just before acceptance of the interrupts. Figure 2.2.7 shows an interrupt operation diagram.

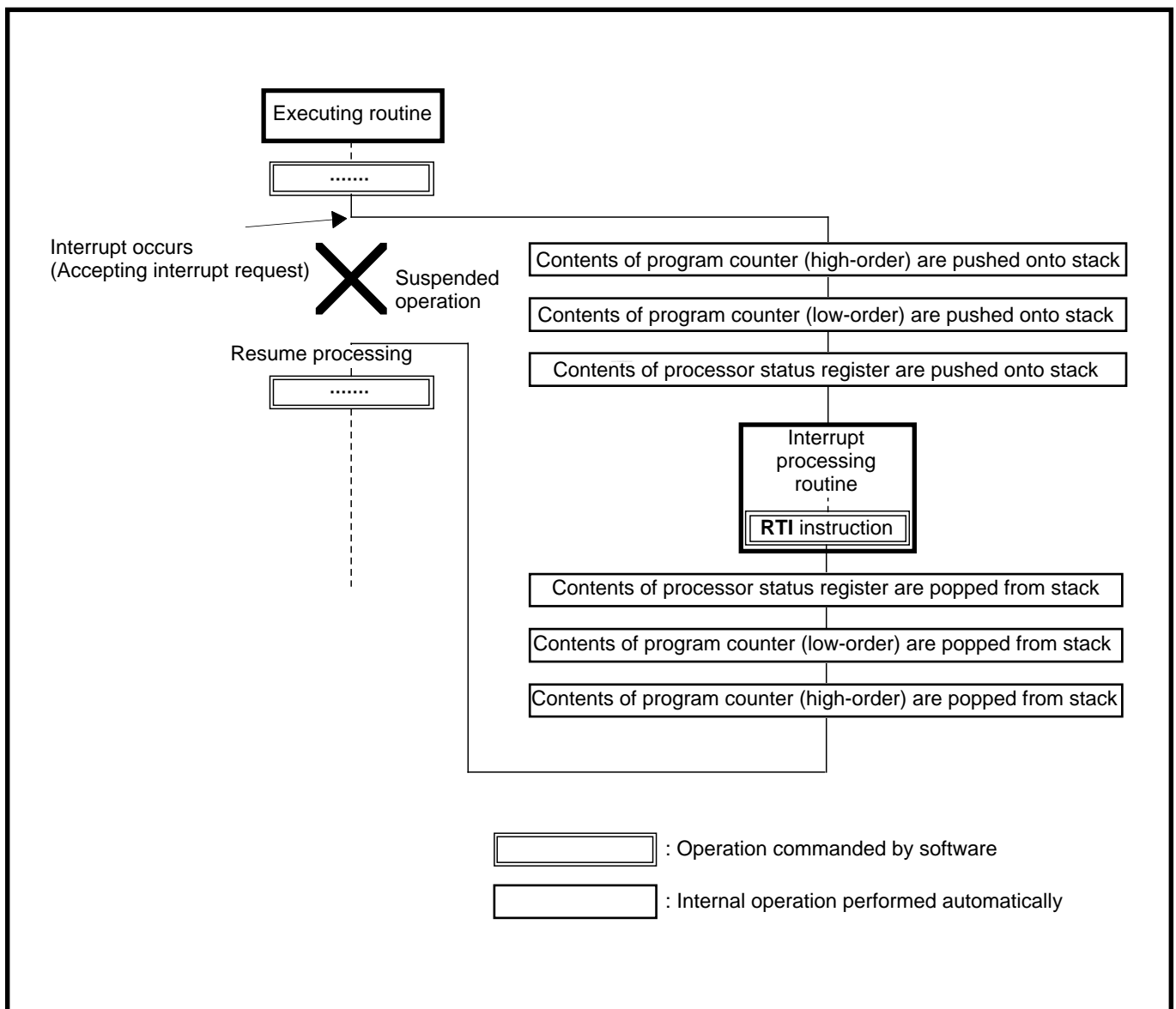


Fig. 2.2.7 Interrupt operation diagram

# APPLICATION

## 2.2 Interrupt

### (1) Processing upon acceptance of interrupt request

Upon acceptance of an interrupt request, the following operations are automatically performed.

- ① The processing being executed is stopped.
- ② The contents of the program counter and the processor status register are pushed onto the stack area. Figure 2.2.8 shows the changes of the stack pointer and the program counter upon acceptance of an interrupt request.
- ③ Concurrently with the push operation, the jump destination address (the beginning address of the interrupt processing routine) of the occurring interrupt stored in the vector address is set in the program counter, then the interrupt processing routine is executed.
- ④ After the interrupt processing routine is started, the corresponding interrupt request bit is automatically cleared to "0". The interrupt disable flag is set to "1" so that multiple interrupts are disabled.

Accordingly, for executing the interrupt processing routine, it is necessary to set the jump destination address in the vector area corresponding to each interrupt.

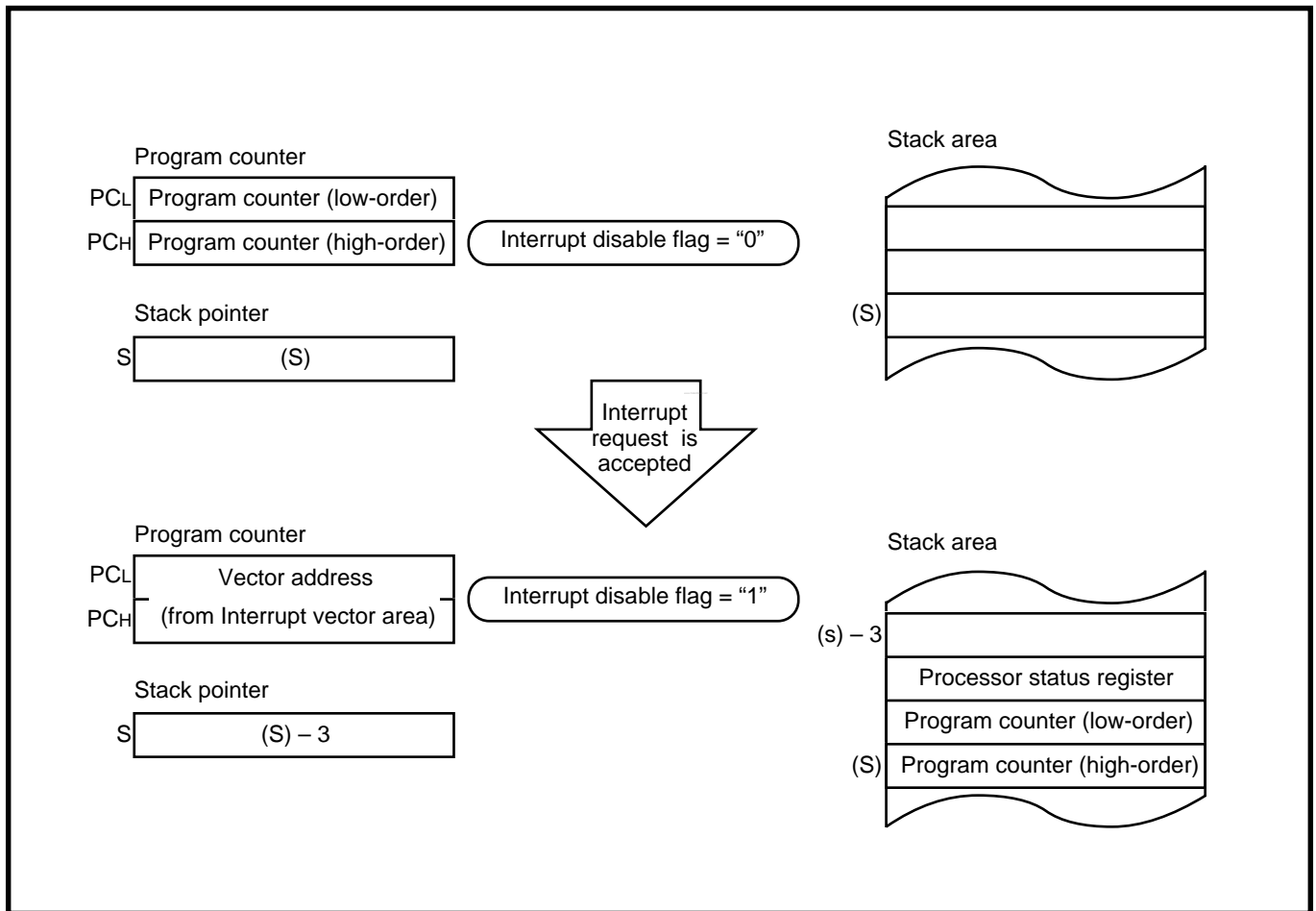


Fig. 2.2.8 Changes of stack pointer and program counter upon acceptance of interrupt request

### (2) Timing after acceptance of interrupt request

The interrupt processing routine begins with the machine cycle following the completion of the instruction that is currently being executed.

Figure 2.2.9 shows the time up to execution of interrupt processing routine and Figure 2.2.10 shows the timing chart after acceptance of interrupt request.

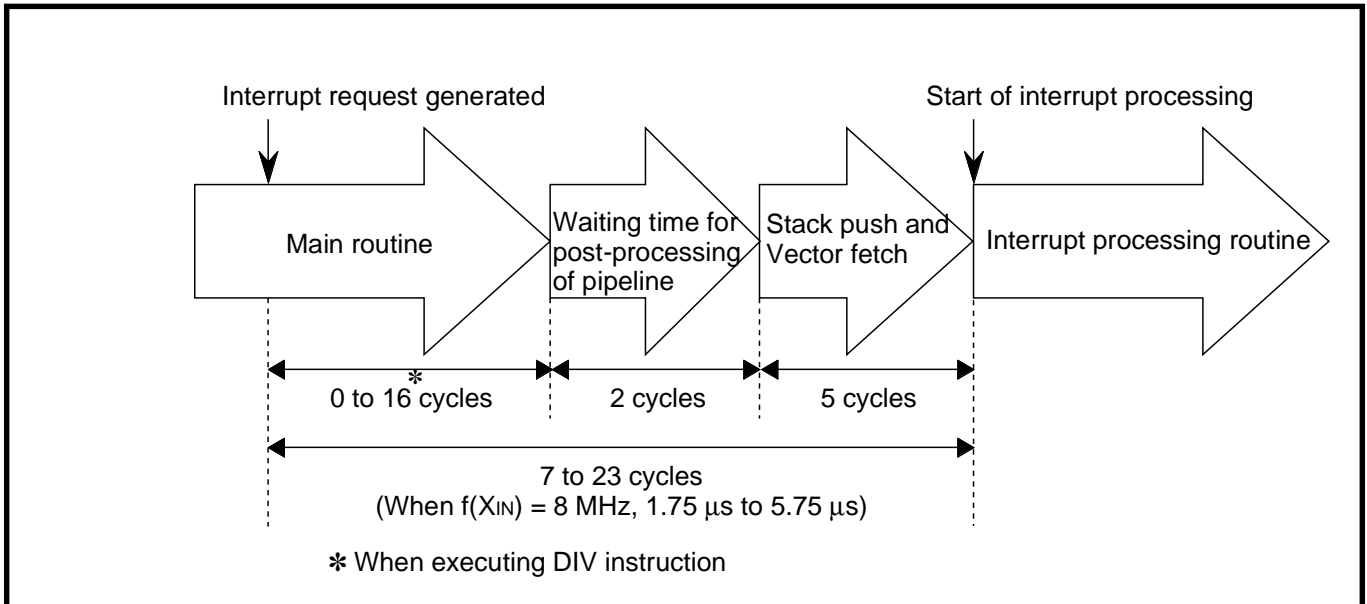


Fig. 2.2.9 Time up to execution of interrupt processing routine

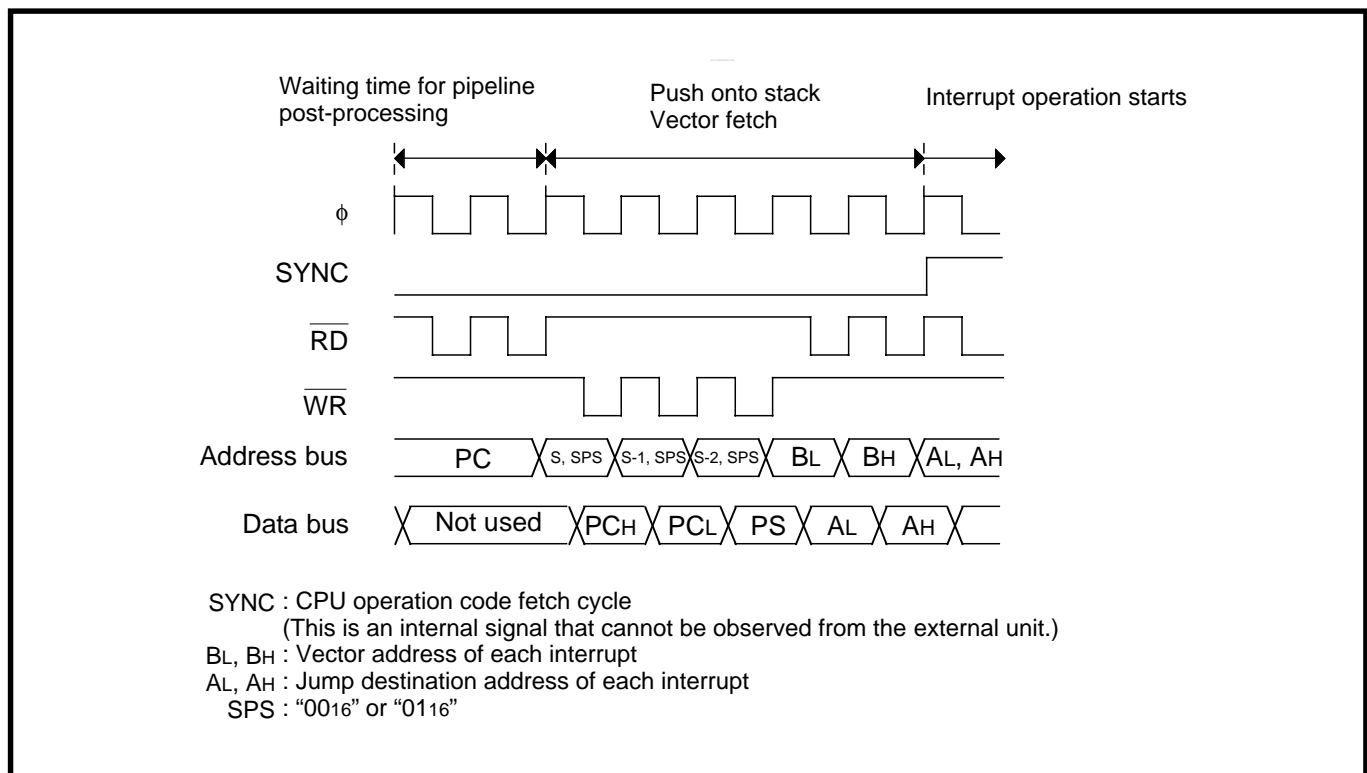


Fig. 2.2.10 Timing chart after acceptance of interrupt request

# APPLICATION

## 2.2 Interrupt

### 2.2.5 Interrupt control

The acceptance of all interrupts, excluding the BRK instruction interrupt, can be controlled by the interrupt request bit, interrupt enable bit, and an interrupt disable flag, as described in detail below. Figure 2.2.11 shows an interrupt control diagram.

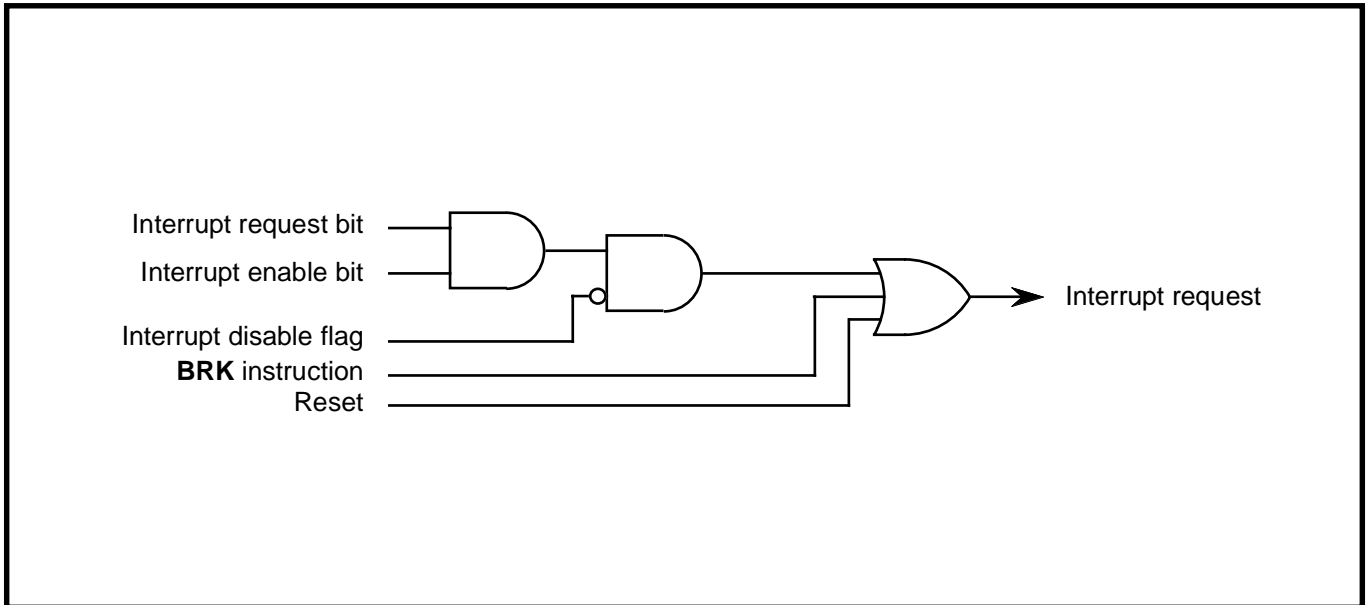


Fig. 2.2.11 Interrupt control diagram

The interrupt request bit, interrupt enable bit and interrupt disable flag function independently and do not affect each other. An interrupt is accepted when all the following conditions are satisfied.

- Interrupt request bit ..... "1"
- Interrupt enable bit ..... "1"
- Interrupt disable flag ..... "0"

Though the interrupt priority is determined by hardware, a variety of priority processing can be performed by software using the above bits and flag. Table 2.2.2 shows a list of interrupt control bits according to the interrupt source.

#### (1) Interrupt request bits

The interrupt request bits are allocated to the interrupt request register 1 (address 3C<sub>16</sub>) and interrupt request register 2 (address 3D<sub>16</sub>).

The occurrence of an interrupt request causes the corresponding interrupt request bit to be set to "1". The interrupt request bit is held in the "1" state until the interrupt is accepted. When the interrupt is accepted, this bit is automatically cleared to "0".

Each interrupt request bit can be set to "0", but cannot be set to "1", by software.

#### (2) Interrupt enable bits

The interrupt enable bits are allocated to the interrupt control register 1 (address 003E<sub>16</sub>) and the interrupt control register 2 (address 3F<sub>16</sub>).

The interrupt enable bits control the acceptance of the corresponding interrupt request.

When an interrupt enable bit is "0", the corresponding interrupt request is disabled. If an interrupt request occurs when this bit is "0", the corresponding interrupt request bit is set to "1" but the interrupt is not accepted. In this case, unless the interrupt request bit is set to "0" by software, the interrupt request bit remains in the "1" state.

When an interrupt enable bit is "1", the corresponding interrupt is enabled. If an interrupt request occurs when this bit is "1", the interrupt is accepted (when interrupt disable flag = "0").

Each interrupt enable bit can be set to "0" or "1" by software.

### (3) Interrupt disable flag

The interrupt disable flag is allocated to bit 2 of the processor status register. The interrupt disable flag controls the acceptance of interrupt request except BRK instruction.

When this flag is “1”, the acceptance of interrupt requests is disabled. When the flag is “0”, the acceptance of interrupt requests is enabled. This flag is set to “1” with the SEI instruction and is set to “0” with the CLI instruction.

When a main routine branches to an interrupt processing routine, this flag is automatically set to “1”, so that multiple interrupts are disabled. To use multiple interrupts, set this flag to “0” with the CLI instruction within the interrupt processing routine. Figure 2.2.12 shows an example of multiple interrupts.

**Table 2.2.2 List of interrupt bits according to interrupt source**

Interrupt source	Interrupt enable bit		Interrupt request bit	
	Address	Bit	Address	Bit
INT <sub>0</sub>	003E <sub>16</sub>	b0	003C <sub>16</sub>	b0
INT <sub>1</sub>	003E <sub>16</sub>	b2	003C <sub>16</sub>	b2
INT <sub>2</sub>	003E <sub>16</sub>	b3	003C <sub>16</sub>	b3
INT <sub>3</sub> /Serial I/O2	003E <sub>16</sub>	b4	003C <sub>16</sub>	b4
Timer X	003E <sub>16</sub>	b6	003C <sub>16</sub>	b6
Timer Y	003E <sub>16</sub>	b7	003C <sub>16</sub>	b7
Timer 1	003F <sub>16</sub>	b0	003D <sub>16</sub>	b0
Timer 2	003F <sub>16</sub>	b1	003D <sub>16</sub>	b1
Serial I/O1 reception	003F <sub>16</sub>	b2	003D <sub>16</sub>	b2
Serial I/O1 transmission	003F <sub>16</sub>	b3	003D <sub>16</sub>	b3
CNTR <sub>0</sub>	003F <sub>16</sub>	b4	003D <sub>16</sub>	b4
CNTR <sub>1</sub>	003F <sub>16</sub>	b5	003D <sub>16</sub>	b5
A-D converter	003F <sub>16</sub>	b6	003D <sub>16</sub>	b6

# APPLICATION

## 2.2 Interrupt

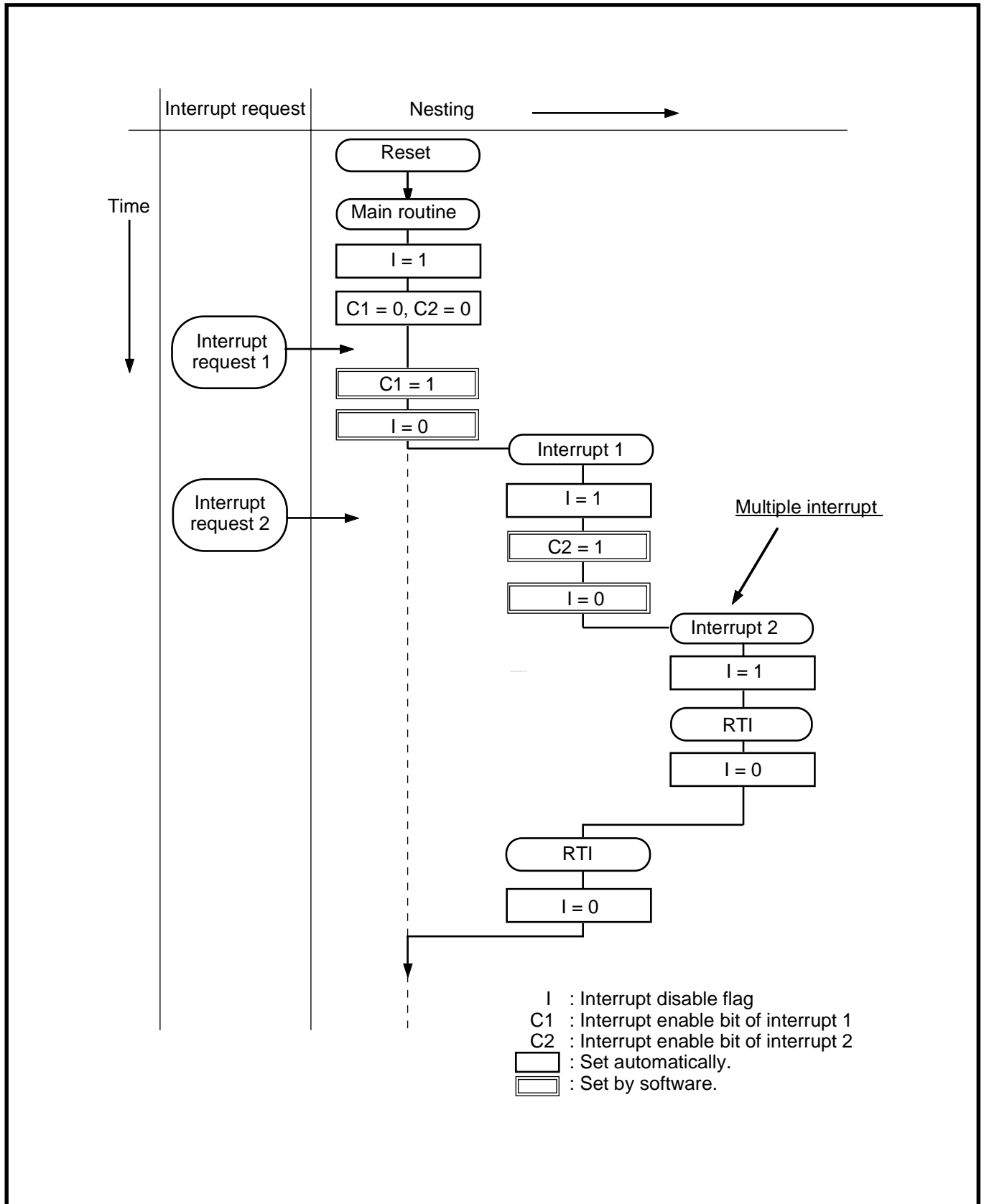


Fig. 2.2.12 Example of multiple interrupts

### 2.2.6 INT interrupt

The INT interrupt requests is generated when the microcomputer detects a level change of each INT pin (INT<sub>0</sub>–INT<sub>3</sub>).

**(1) Active edge selection**

INT<sub>0</sub>–INT<sub>3</sub> can be selected from either a falling edge or rising edge detection as an active edge by the interrupt edge selection register. In the “0” state, the falling edge of the corresponding pin is detected. In the “1” state, the rising edge of the corresponding pin is detected.

**(2) INT<sub>3</sub> interrupt source selection**

Which of interrupt source of the serial I/O2/INT<sub>3</sub> interrupt source can be selected by the serial I/O2/INT<sub>3</sub> interrupt source bit (bit 4 of address 3A<sub>16</sub>). (Set this bit to “0” when using INT<sub>3</sub>.)

# APPLICATION

## 2.2 Interrupt

### 2.2.7 Notes on interrupts

#### (1) Change of relevant register settings

When the setting of the following registers or bits is changed, the interrupt request bit may be set to "1". When not requiring the interrupt occurrence synchronized with these setting, take the following sequence.

- Interrupt edge selection register (address 3A<sub>16</sub>)
- Timer XY mode register (address 23<sub>16</sub>)

Set the above listed registers or bits as the following sequence.

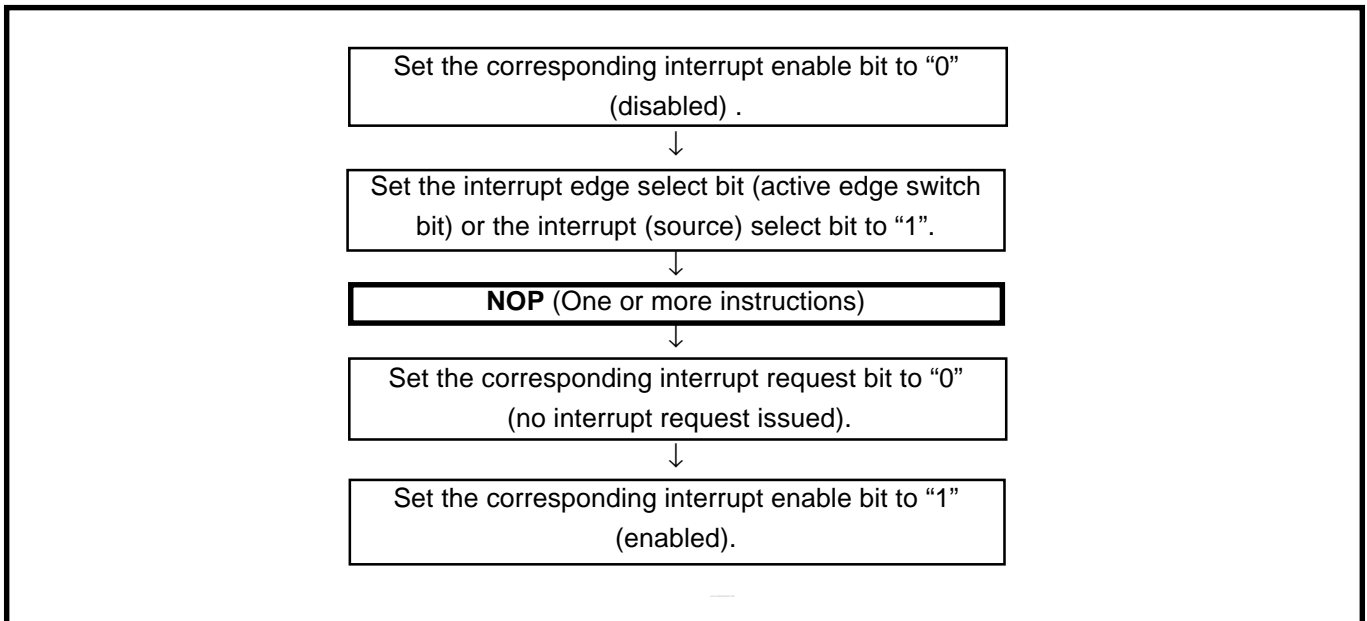


Fig. 2.2.13 Sequence of changing relevant register

#### ■ Reason

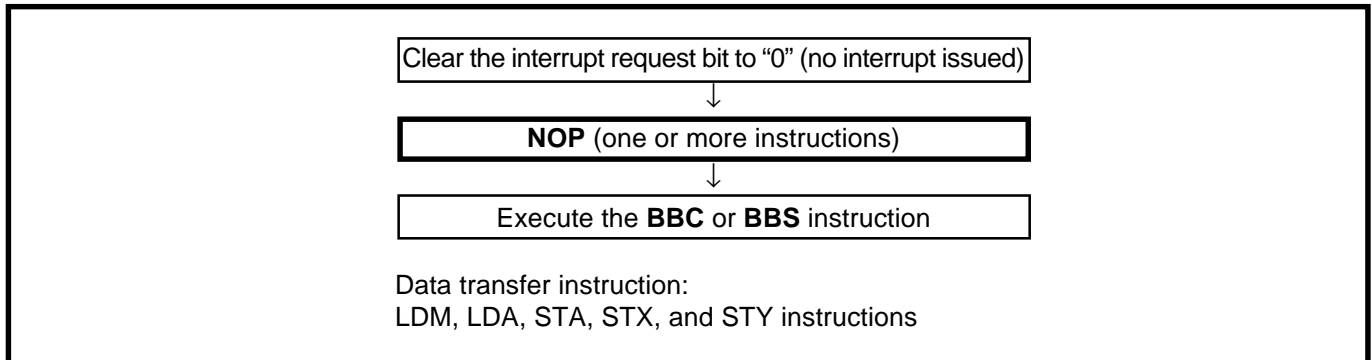
When setting the followings, the interrupt request bit may be set to "1".

- When setting external interrupt active edge  
Concerned register: Interrupt edge selection register (address 3A<sub>16</sub>)  
Timer XY mode register (address 23<sub>16</sub>)
- When switching interrupt sources of an interrupt vector address where two or more interrupt sources are allocated.  
Concerned register: Interrupt edge selection register (address 3A<sub>16</sub>)



**(2) Check of interrupt request bit**

- When executing the **BBC** or **BBS** instruction to an interrupt request bit of an interrupt request register immediately after this bit is set to "0" by using a data transfer instruction, execute one or more instructions before executing the **BBC** or **BBS** instruction.

**Fig. 2.2.14 Sequence of check of interrupt request bit****■ Reason**

If the **BBC** or **BBS** instruction is executed immediately after an interrupt request bit of an interrupt request register is cleared to "0", the value of the interrupt request bit before being cleared to "0" is read.

# APPLICATION

## 2.3 Timer

### 2.3 Timer

This paragraph explains the registers setting method and the notes relevant to the timers.

#### 2.3.1 Memory map

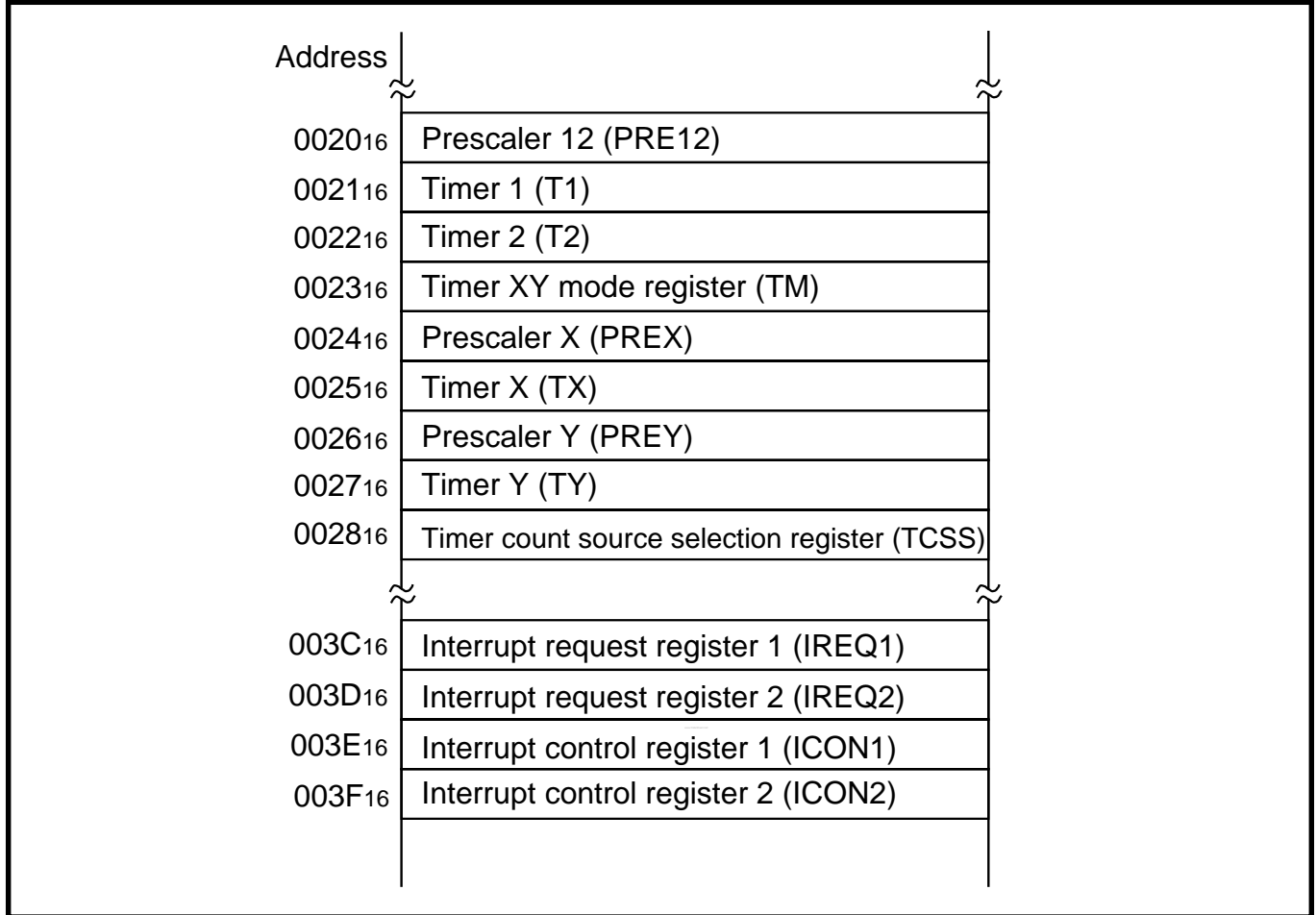


Fig. 2.3.1 Memory map of registers relevant to timers

#### 2.3.2 Relevant registers

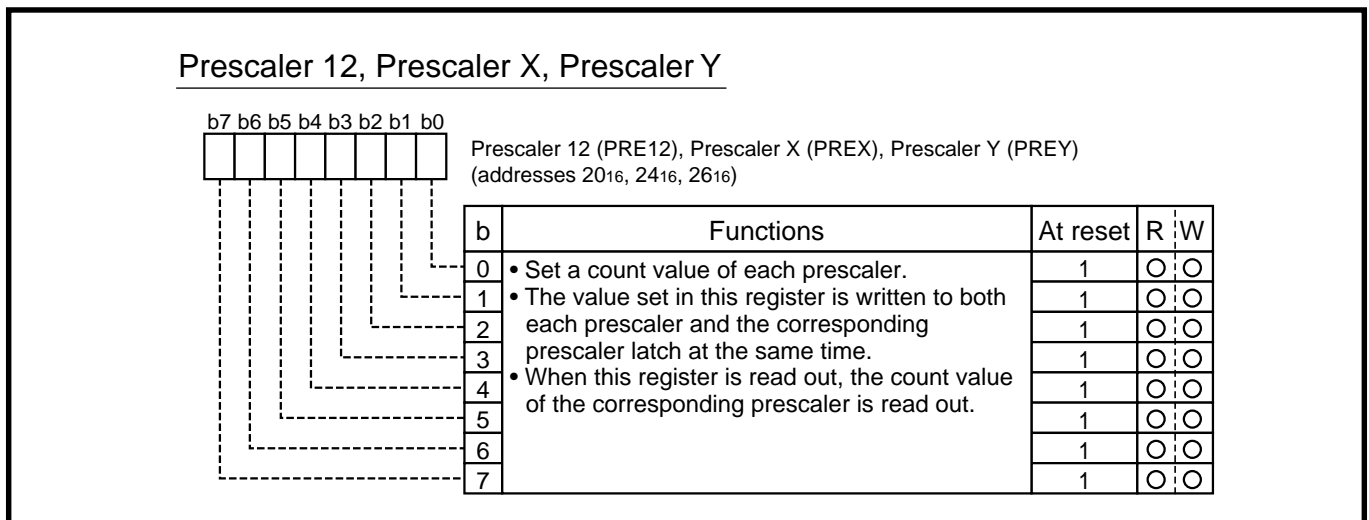


Fig. 2.3.2 Structure of Prescaler 12, Prescaler X, Prescaler Y

### Timer 1

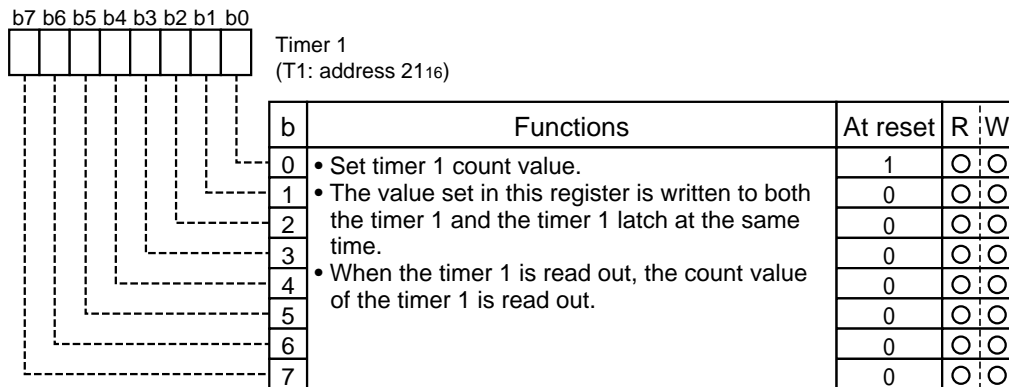


Fig. 2.3.3 Structure of Timer 1

### Timer 2

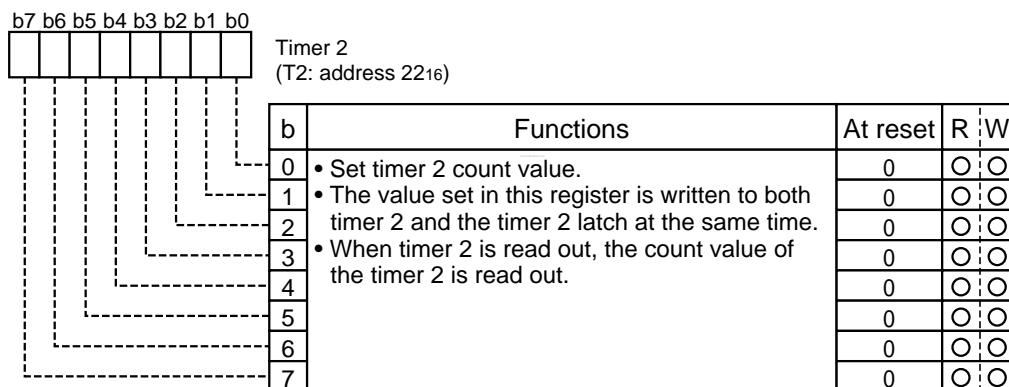


Fig. 2.3.4 Structure of Timer 2

# APPLICATION

## 2.3 Timer

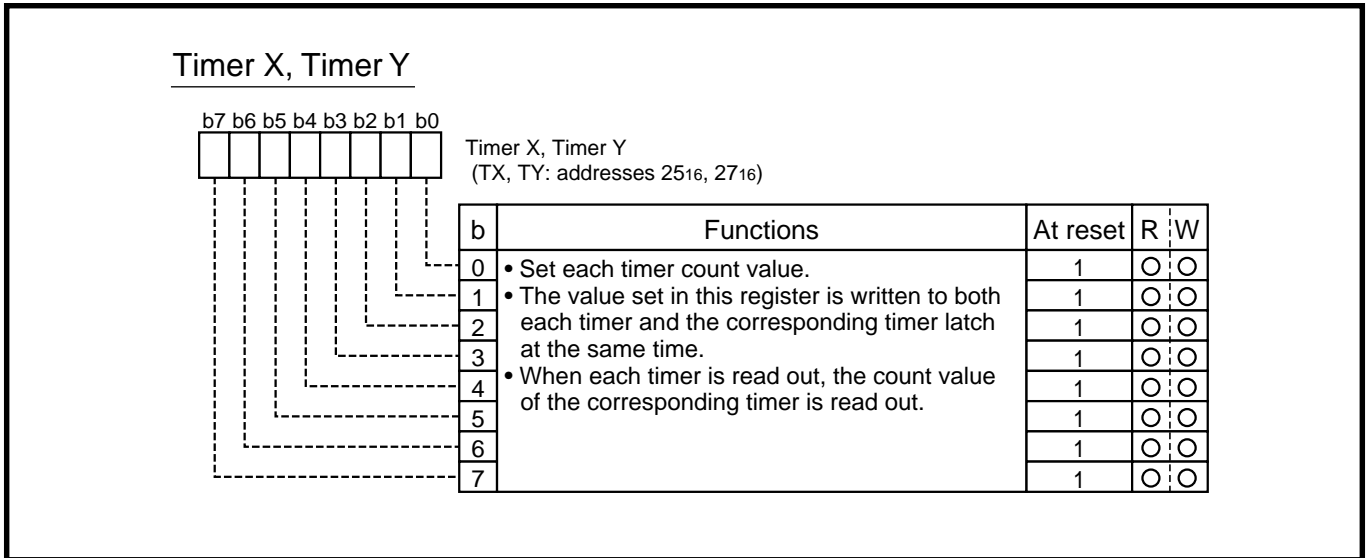


Fig. 2.3.5 Structure of Timer X, Timer Y

### Timer XY mode register

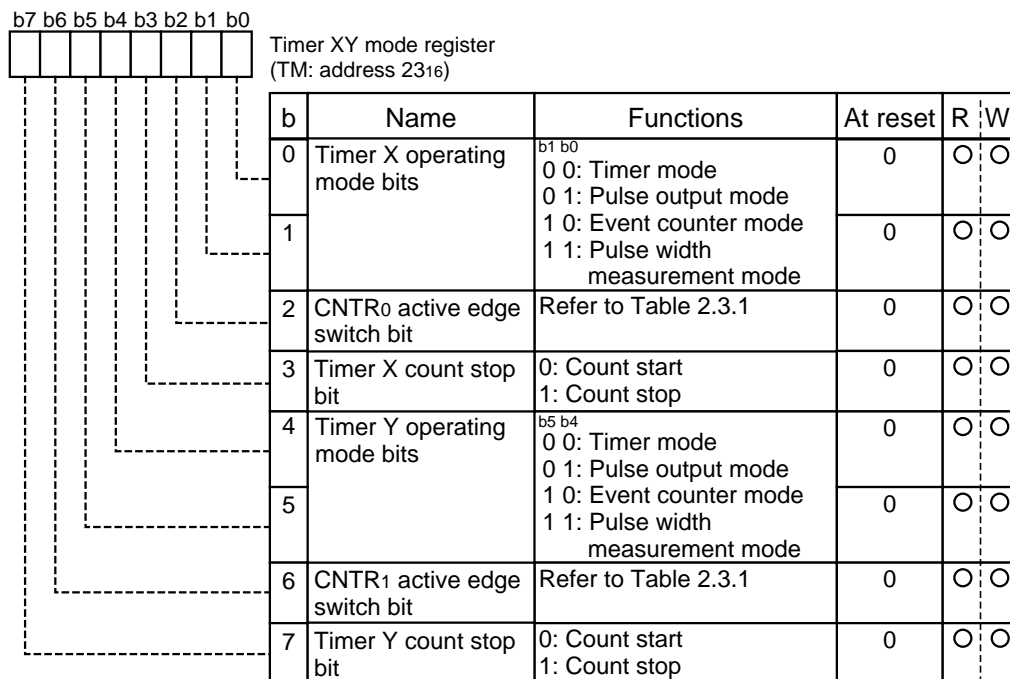


Fig. 2.3.6 Structure of Timer XY mode register

Table 2.3.1 CNTR0/CNTR1 active edge switch bit function

Timer X /Timer Y operation modes	Set value	Timer function	CNTR0 / CNTR1 interrupt request occurrence source
Timer mode	"0"	No influence to timer count	CNTR0/CNTR1 input signal falling edge
	"1"	No influence to timer count	CNTR0/CNTR1 input signal rising edge
Pulse output mode	"0"	Pulse output start: Beginning at "H" level	Output signal falling edge count
	"1"	Pulse output start: Beginning at "L" level	Output signal rising edge count
Event counter mode	"0"	Rising edge count	Input signal falling edge count
	"1"	Falling edge count	Input signal rising edge count
Pulse width measurement mode	"0"	"H" level width measurement	Input signal falling edge count
	"1"	"L" level width measurement	Input signal rising edge count

# APPLICATION

## 2.3 Timer

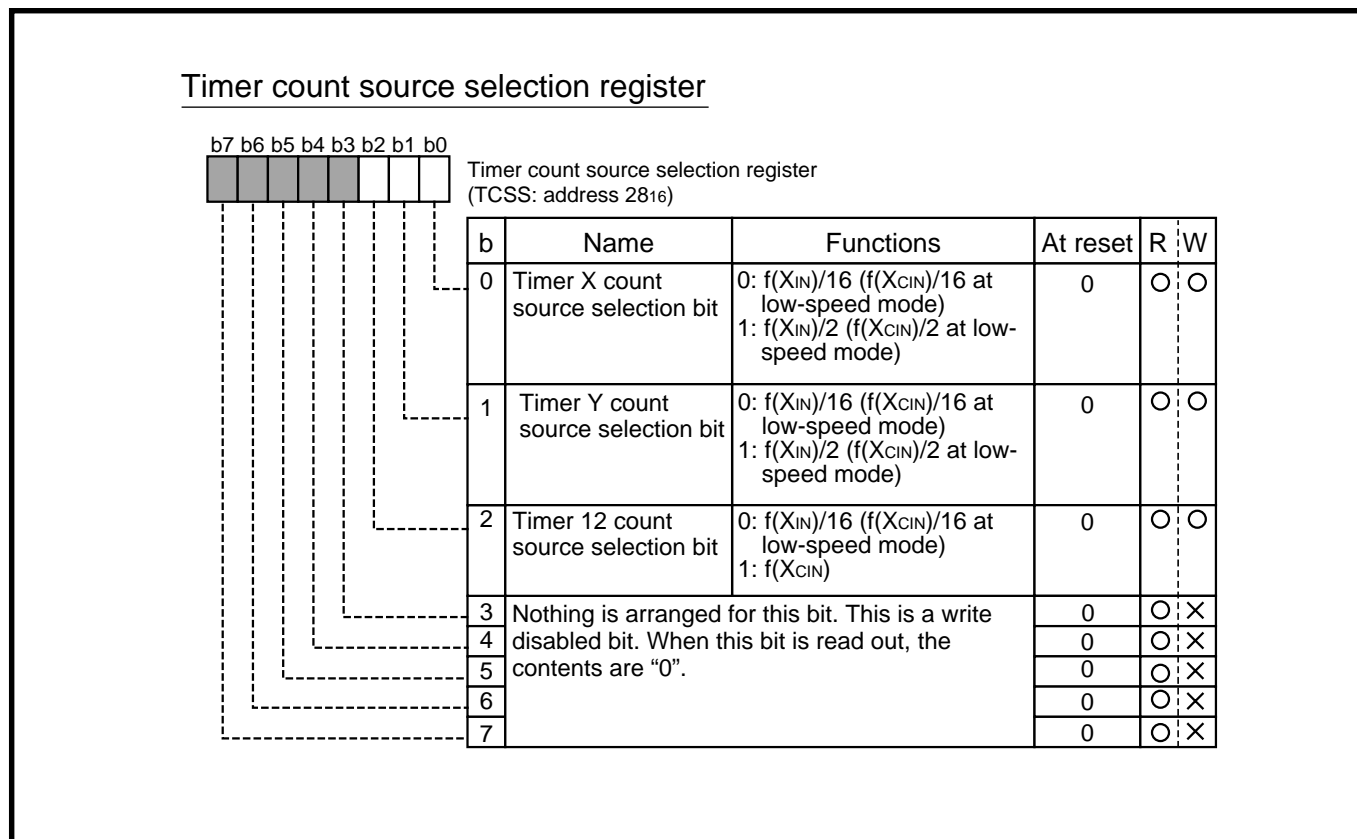


Fig. 2.3.7 Structure of Timer count source selection register

### Interrupt request register 1

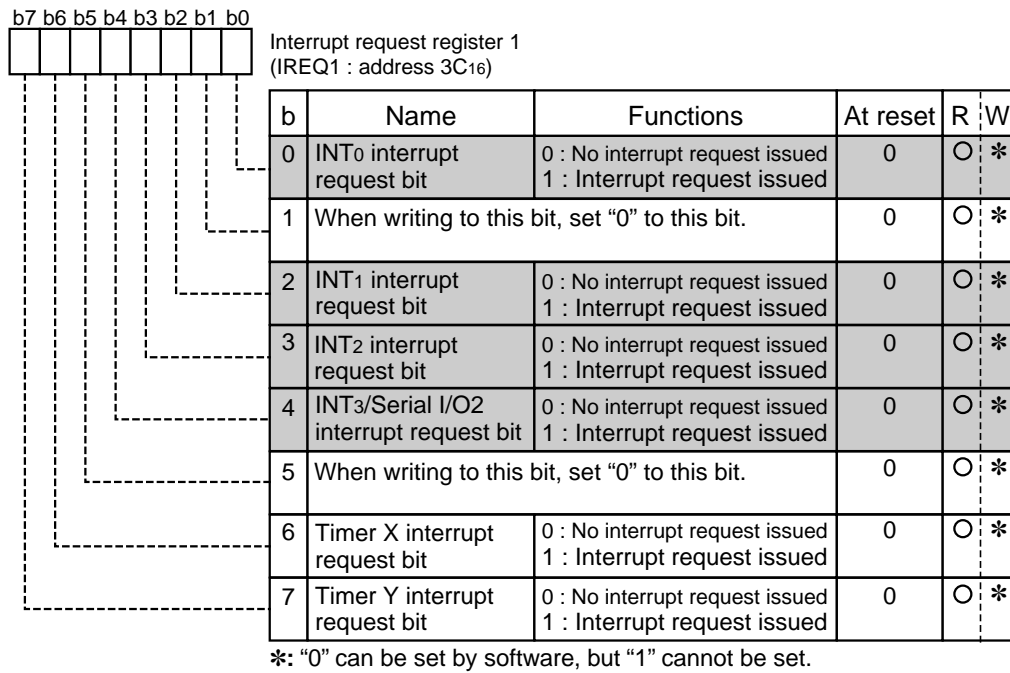


Fig. 2.3.8 Structure of Interrupt request register 1

### Interrupt request register 2

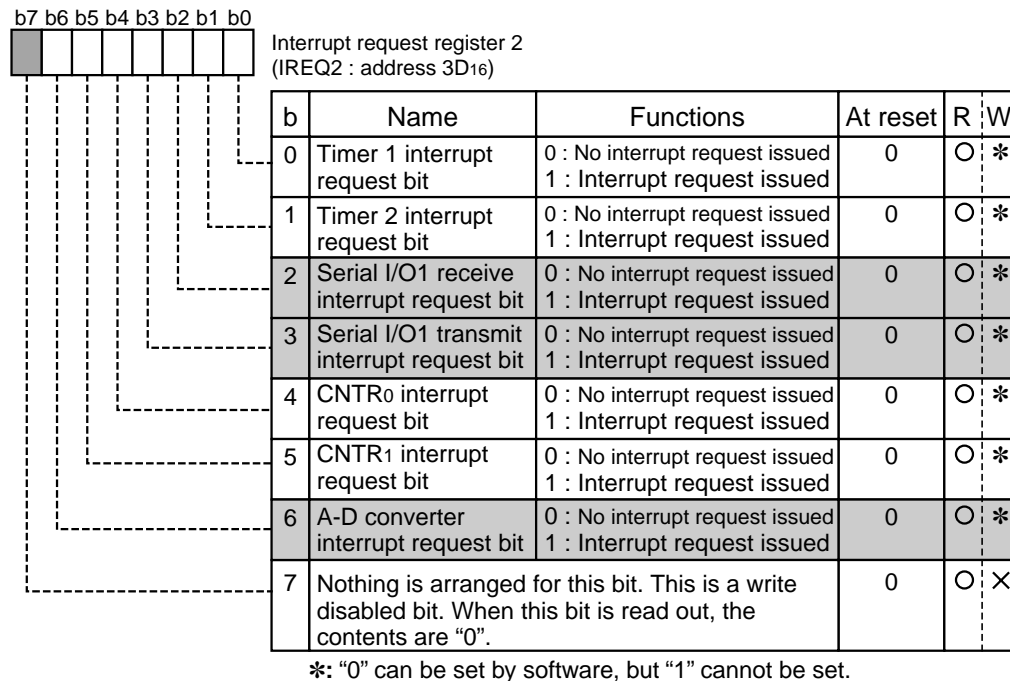


Fig. 2.3.9 Structure of Interrupt request register 2

# APPLICATION

## 2.3 Timer

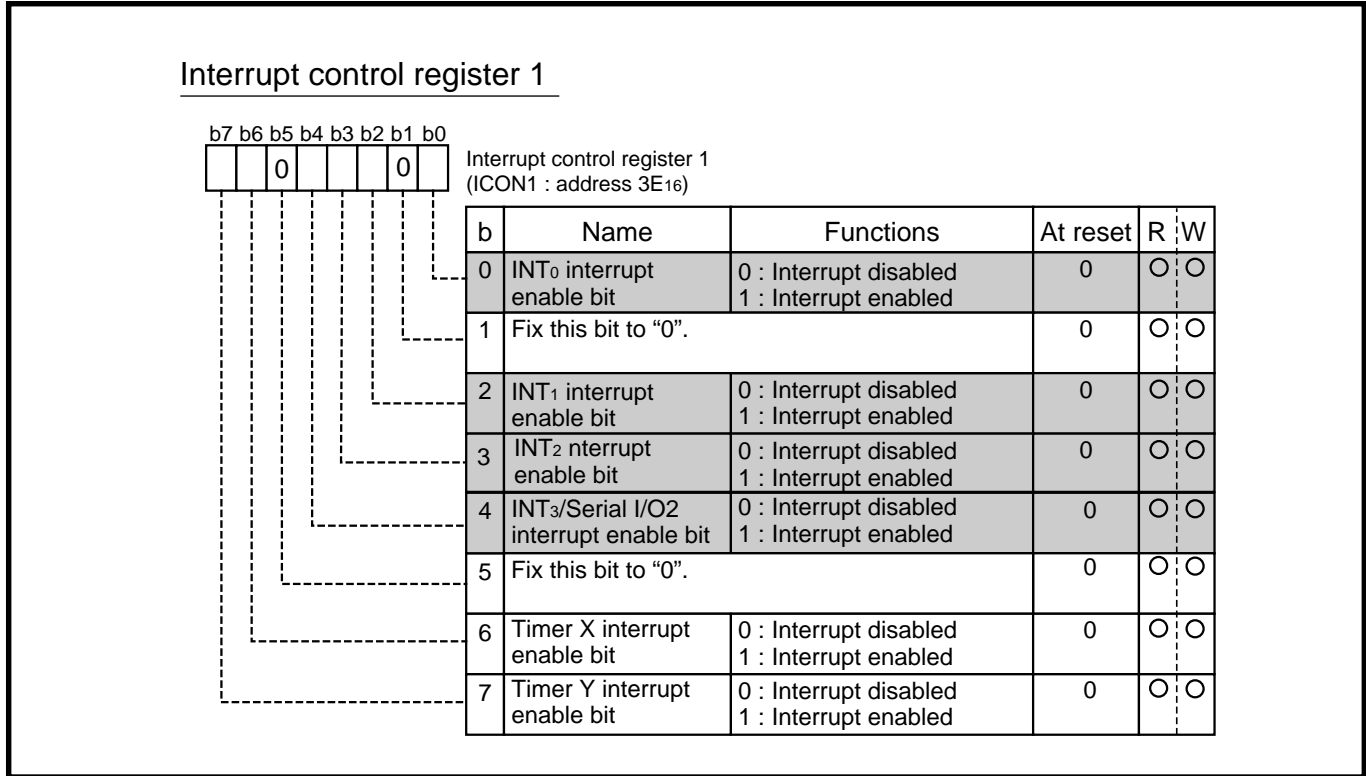


Fig. 2.3.10 Structure of Interrupt control register 1

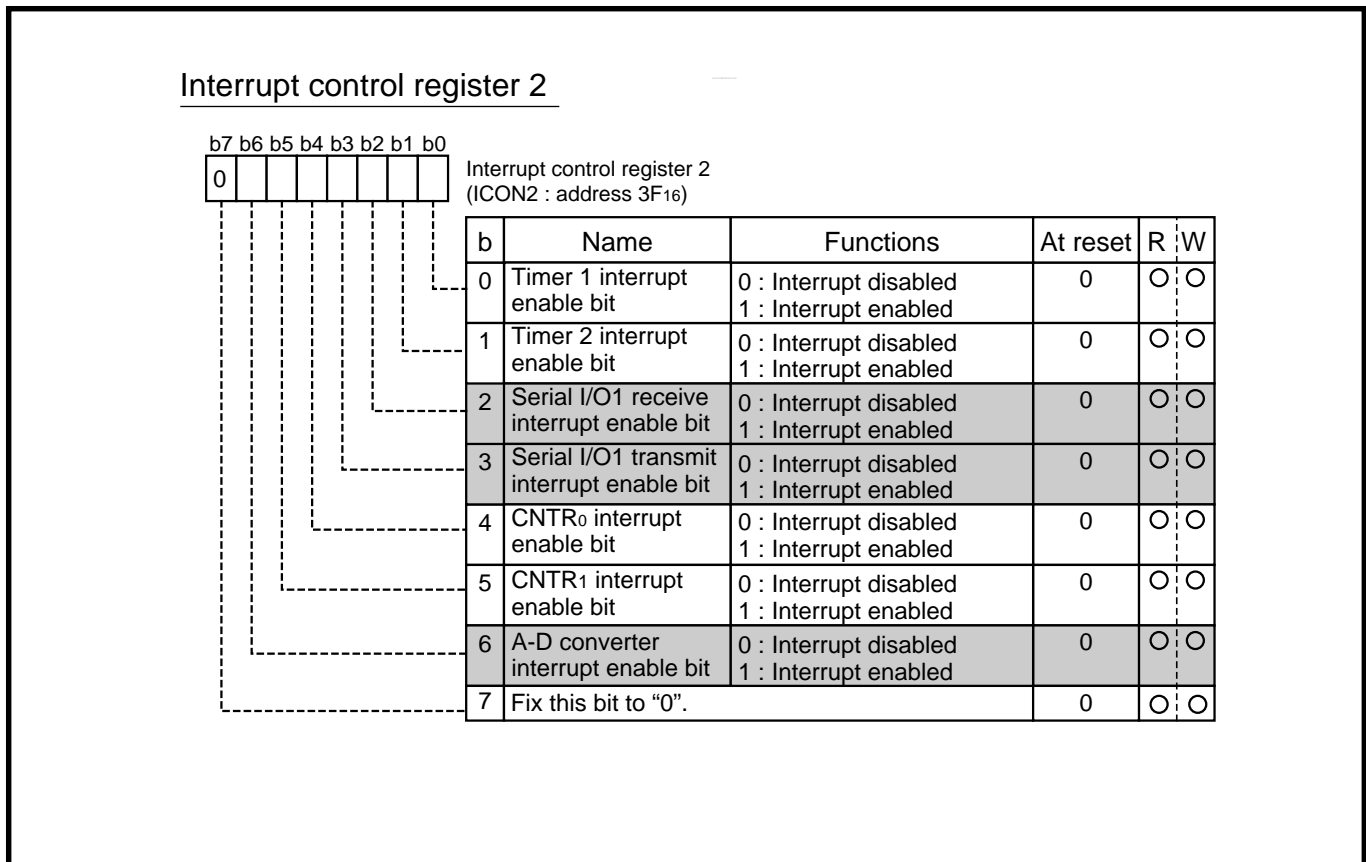


Fig. 2.3.11 Structure of Interrupt control register 2



### 2.3.3 Timer application examples

#### (1) Basic functions and uses

**[Function 1] Control of Event interval (Timer X, Timer Y, Timer 1, Timer 2)**

When a certain time, by setting a count value to each timer, has passed, the timer interrupt request occurs.

<Use>

- Generation of an output signal timing
- Generation of a wait time

**[Function 2] Control of Cyclic operation (Timer X, Timer Y, Timer 1, Timer 2)**

The value of the timer latch is automatically written to the corresponding timer each time the timer underflows, and each timer interrupt request occurs in cycles.

<Use>

- Generation of cyclic interrupts
- Clock function (measurement of 250 ms); see Application example 1
- Control of a main routine cycle

**[Function 3] Output of Rectangular waveform (Timer X, Timer Y)**

The output level of the CNTR pin is inverted each time the timer underflows (in the pulse output mode).

<Use>

- Piezoelectric buzzer output; see Application example 2
- Generation of the remote control carrier waveforms

**[Function 4] Count of External pulses (Timer X, Timer Y)**

External pulses input to the CNTR pin are counted as the timer count source (in the event counter mode).

<Use>

- Frequency measurement; see Application example 3
- Division of external pulses
- Generation of interrupts due to a cycle using external pulses as the count source; count of a reel pulse

**[Function 5] Measurement of External pulse width (Timer X, Timer Y)**

The "H" or "L" level width of external pulses input to CNTR pin is measured (in the pulse width measurement mode).

<Use>

- Measurement of external pulse frequency (measurement of pulse width of FG pulse\* for a motor); see Application example 4
- Measurement of external pulse duty (when the frequency is fixed)

FG pulse\*: Pulse used for detecting the motor speed to control the motor speed.

# APPLICATION

## 2.3 Timer

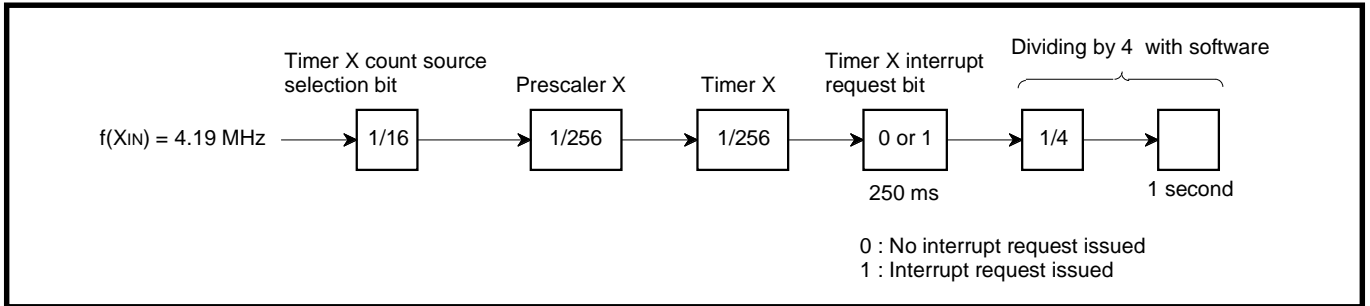
### (2) Timer application example 1: Clock function (measurement of 250 ms)

**Outline:** The input clock is divided by the timer so that the clock can count up at 250 ms intervals.

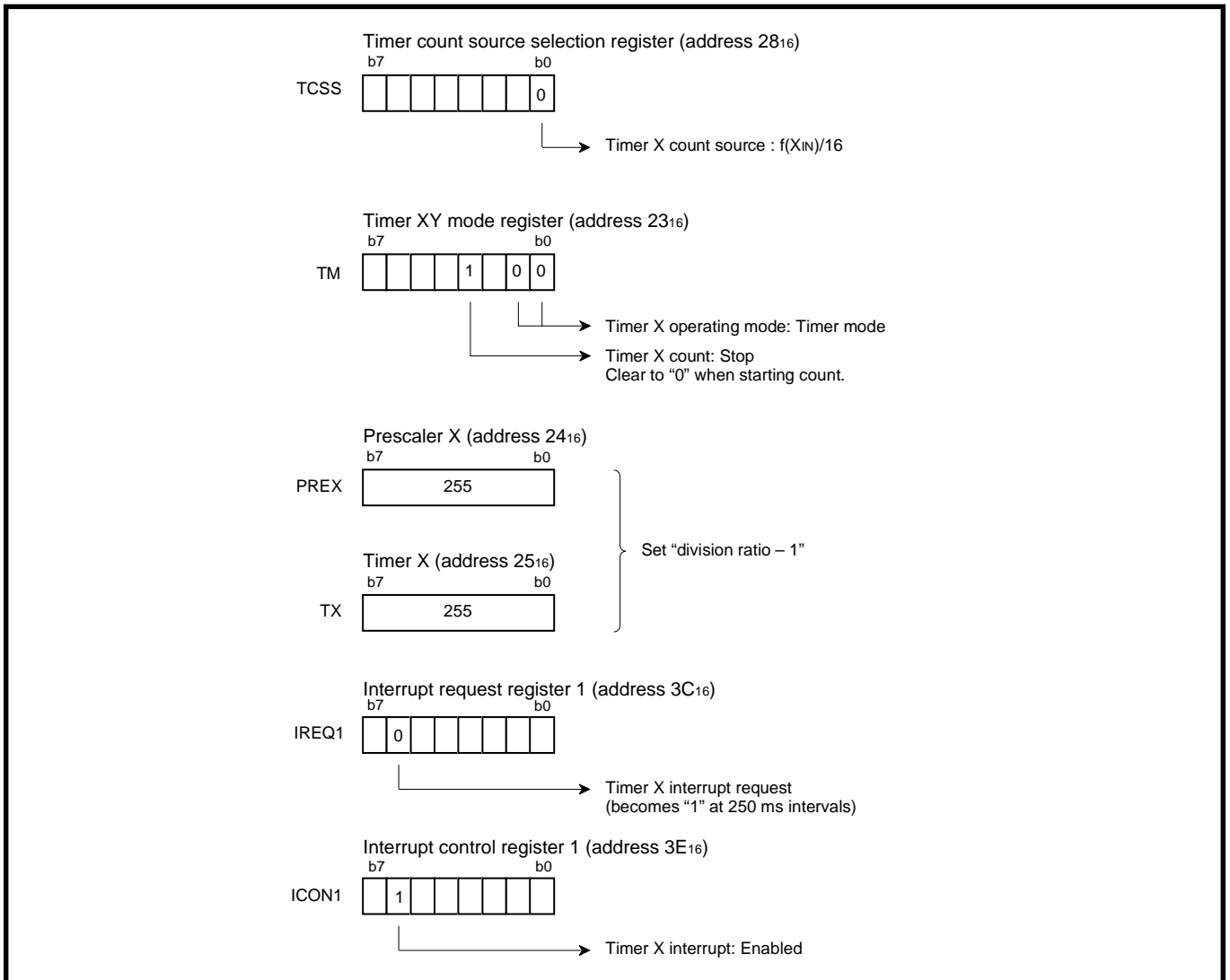
**Specifications:** •The clock  $f(X_{IN}) = 4.19 \text{ MHz}$  ( $2^{22} \text{ Hz}$ ) is divided by the timer X.

- The clock is counted up in the process routine of the timer X interrupt which occurs at 250 ms intervals.

Figure 2.3.12 shows the timers connection and setting of division ratios; Figure 2.3.13 shows the relevant registers setting; Figure 2.3.14 shows the control procedure.



**Fig. 2.3.12 Timers connection and setting of division ratios**



**Fig. 2.3.13 Relevant registers setting**

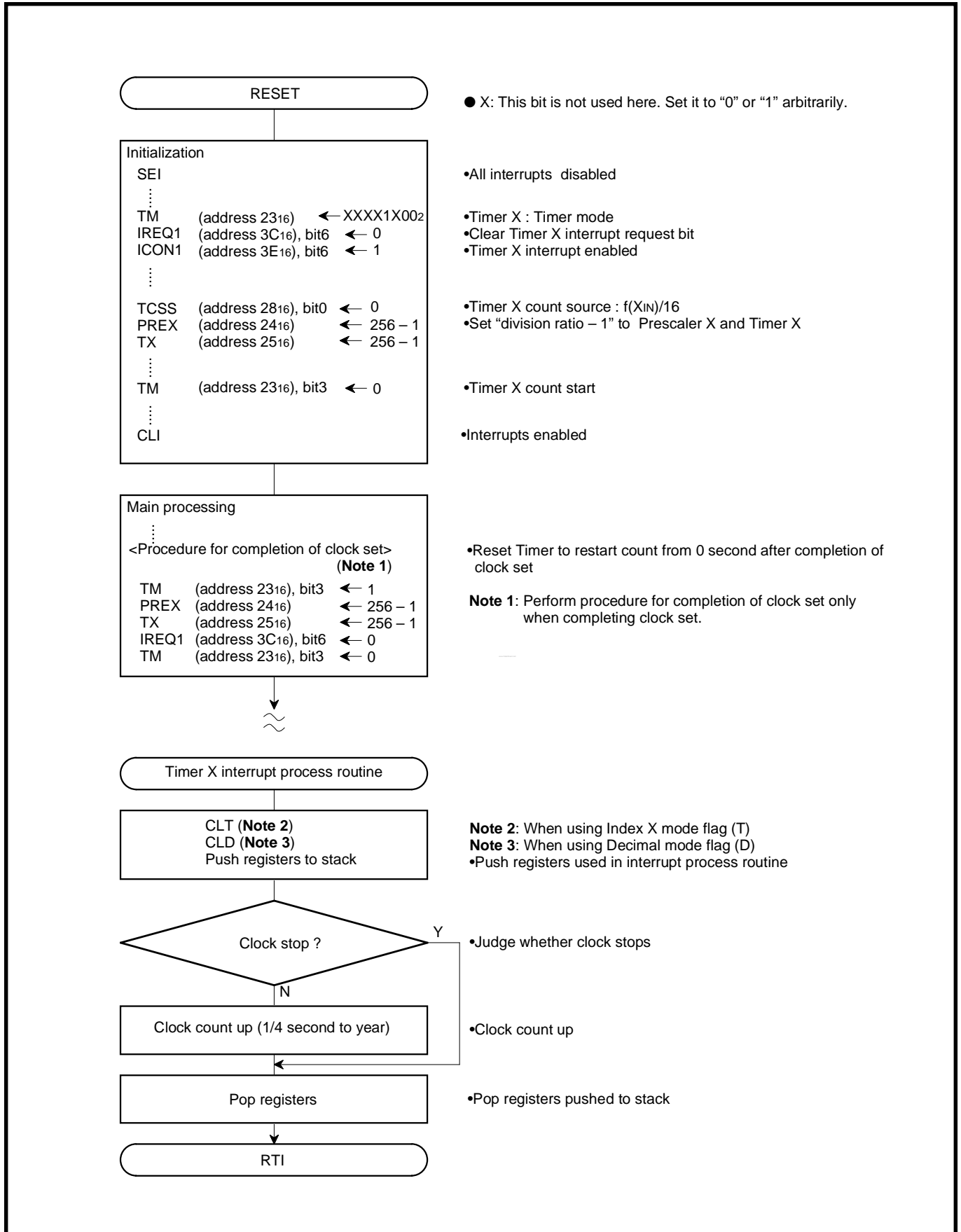


Fig. 2.3.14 Control procedure

# APPLICATION

## 2.3 Timer

### (3) Timer application example 2: Piezoelectric buzzer output

**Outline:** The rectangular waveform output function of the timer is applied for a piezoelectric buzzer output.

- Specifications:**
- The rectangular waveform, dividing the clock  $f(X_{IN}) = 4.19 \text{ MHz}$  ( $2^{22} \text{ Hz}$ ) into about 2 kHz (2048 Hz), is output from the P27/CNTR<sub>0</sub> pin.
  - The level of the P27/CNTR<sub>0</sub> pin is fixed to “H” while a piezoelectric buzzer output stops.

Figure 2.3.15 shows a peripheral circuit example, and Figure 2.3.16 shows the timers connection and setting of division ratios. Figure 2.3.17 shows the relevant registers setting, and Figure 2.3.18 shows the control procedure.

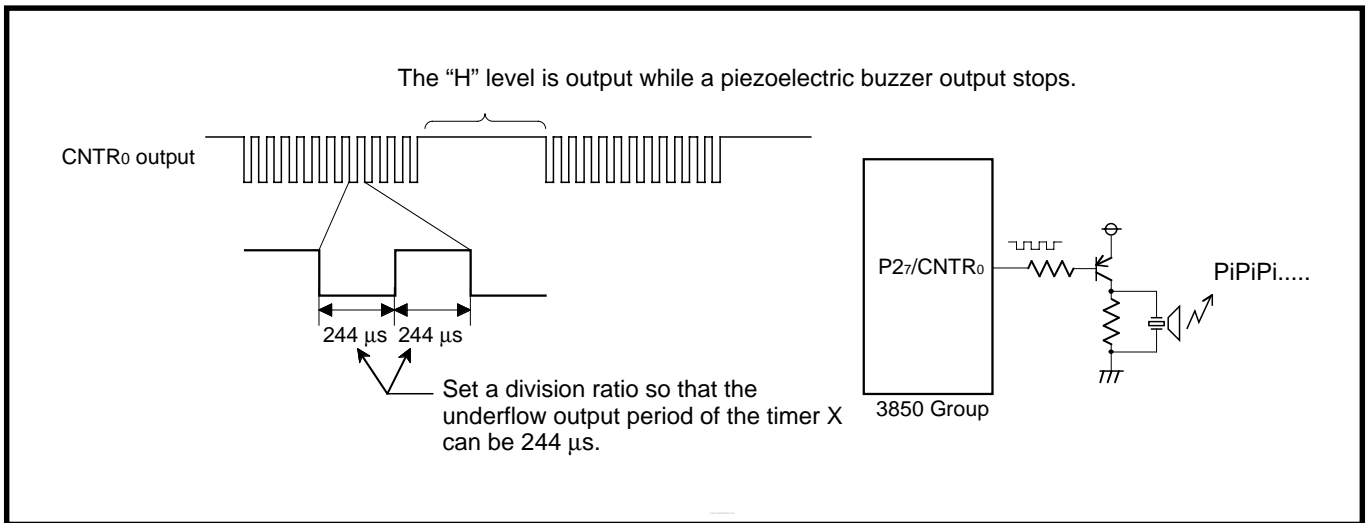


Fig. 2.3.15 Peripheral circuit example

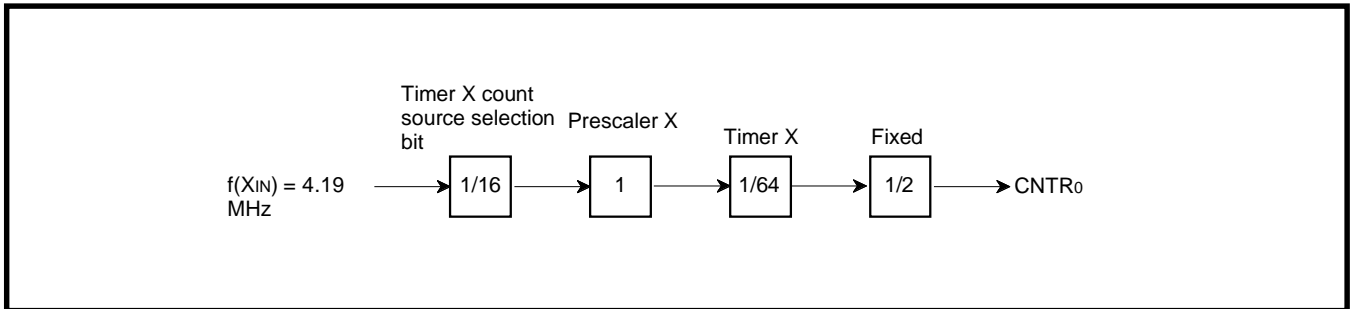
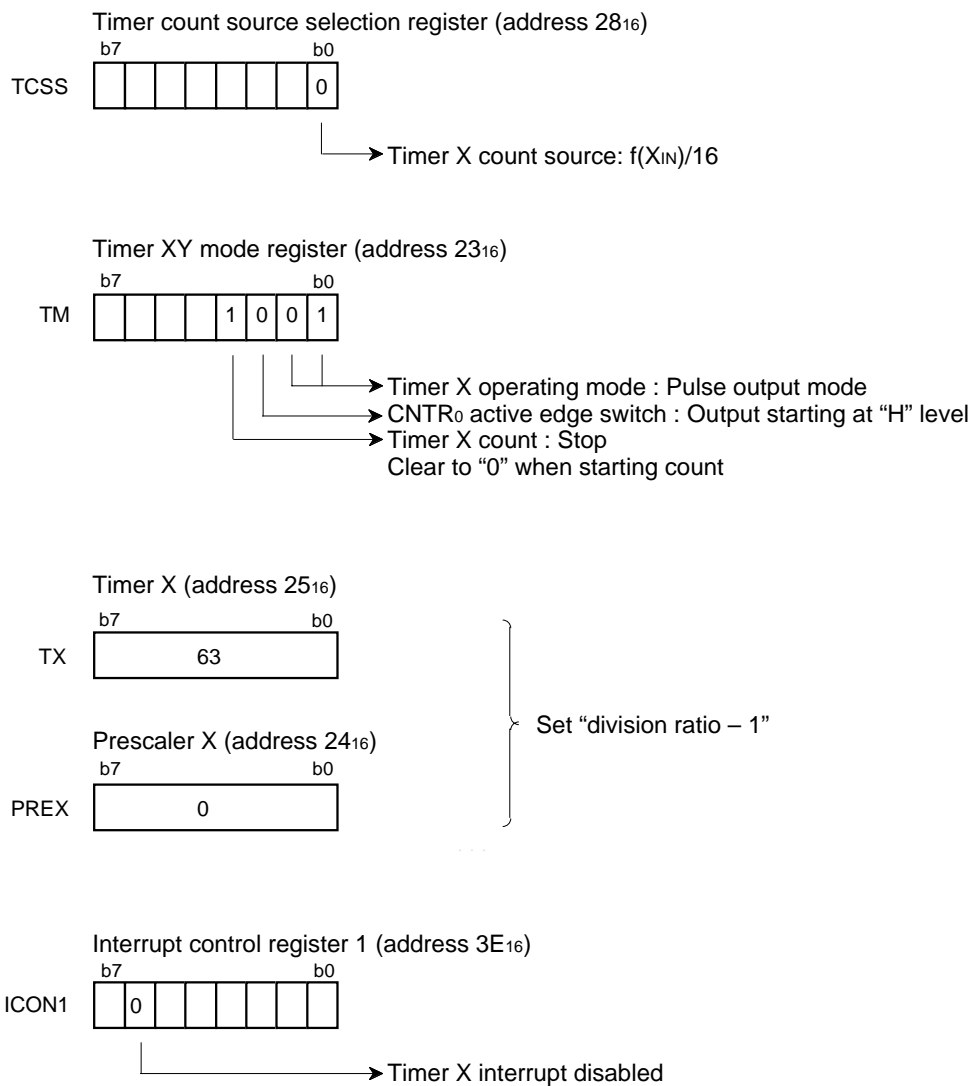


Fig. 2.3.16 Timers connection and setting of division ratios



**Fig. 2.3.17 Relevant registers setting**

# APPLICATION

## 2.3 Timer

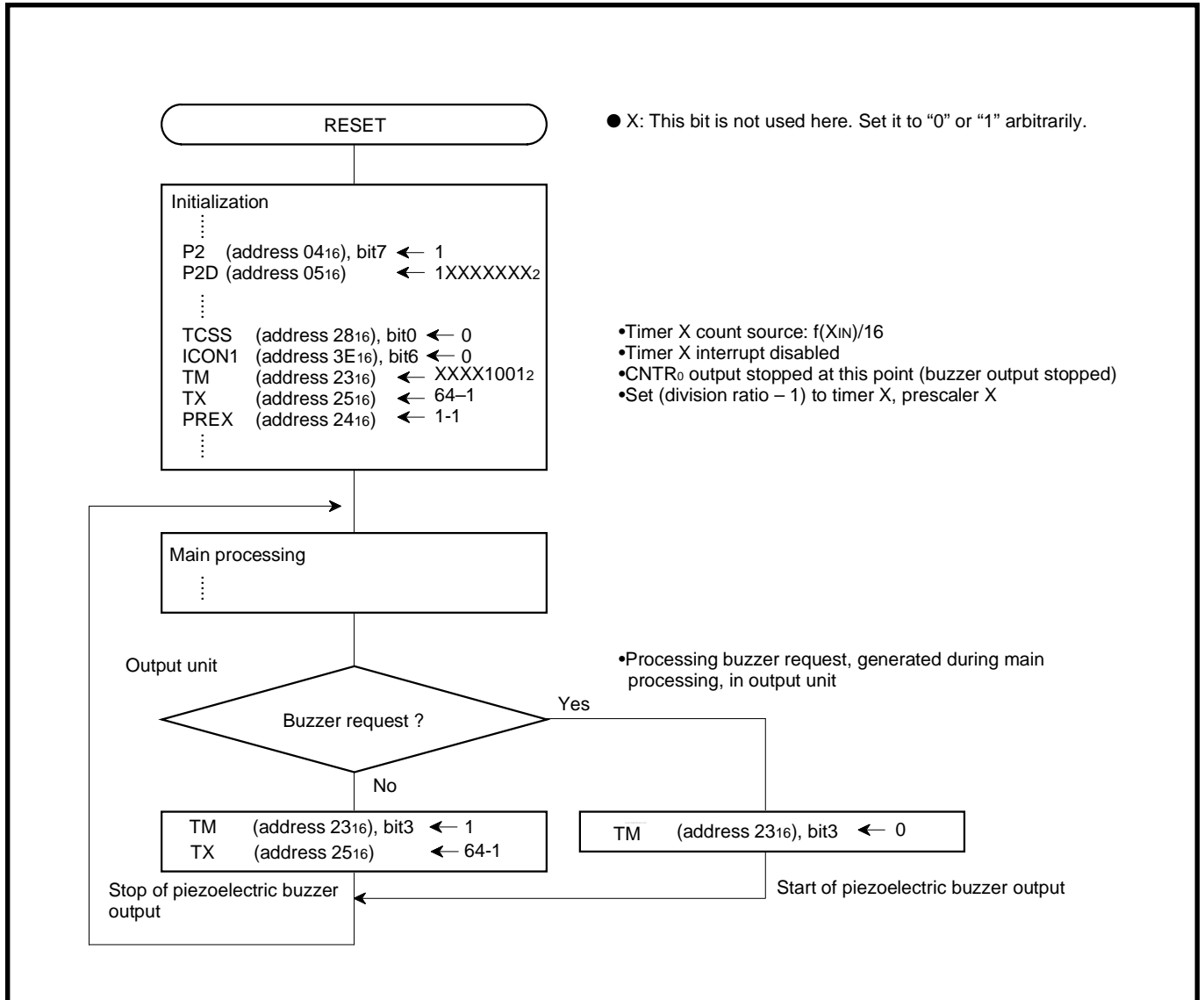


Fig. 2.3.18 Control procedure

**(4) Timer application example 3: Frequency measurement**

**Outline:** The following two values are compared to judge whether the frequency is within a valid range.

- A value by counting pulses input to P4<sub>0</sub>/CNTR<sub>1</sub> pin with the timer.
- A reference value

**Specifications:** •Clock  $f(X_{IN}) = 4.19 \text{ MHz}$  ( $2^{22} \text{ Hz}$ )

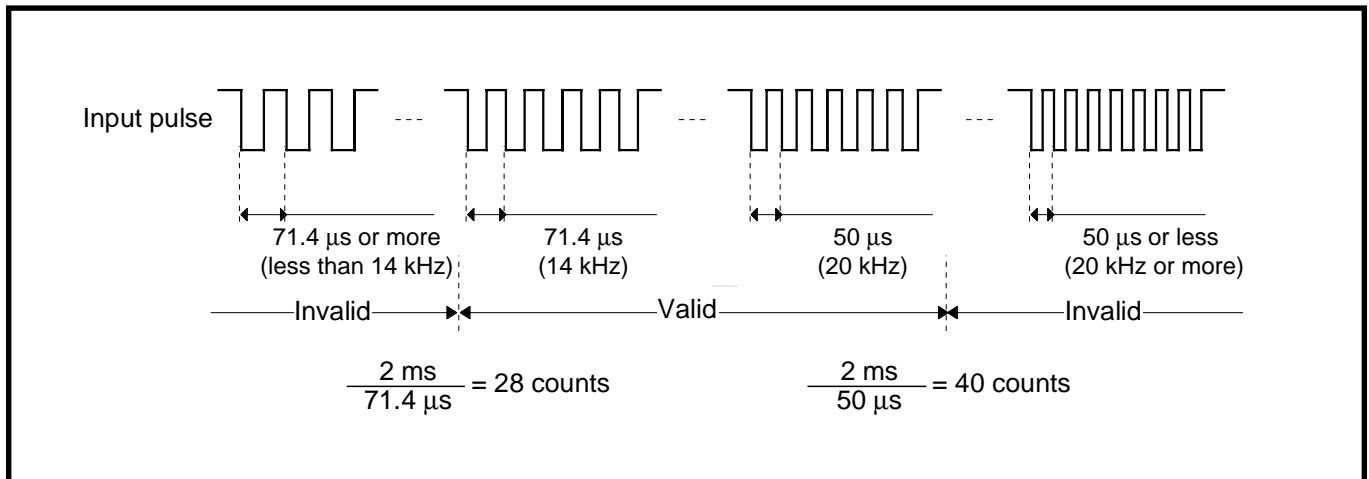
•The pulse is input to the P4<sub>0</sub>/CNTR<sub>1</sub> pin and counted by the timer Y.

•A count value is read out at about 2 ms intervals, the timer 1 interrupt interval. When the count value is 28 to 40, it is judged that the input pulse is valid.

•Because the timer is a down-counter, the count value is compared with 227 to 215 **(Note)**.

**Note:** 227 to 215 = {255 (initial value of counter) – 28} to {255 – 40}; 28 to 40 means the number of valid value.

Figure 2.3.19 shows the judgment method of valid/invalid of input pulses; Figure 2.3.20 shows the relevant registers setting; Figure 2.3.21 shows the control procedure.



**Fig. 2.3.19 Judgment method of valid/invalid of input pulses**

# APPLICATION

## 2.3 Timer

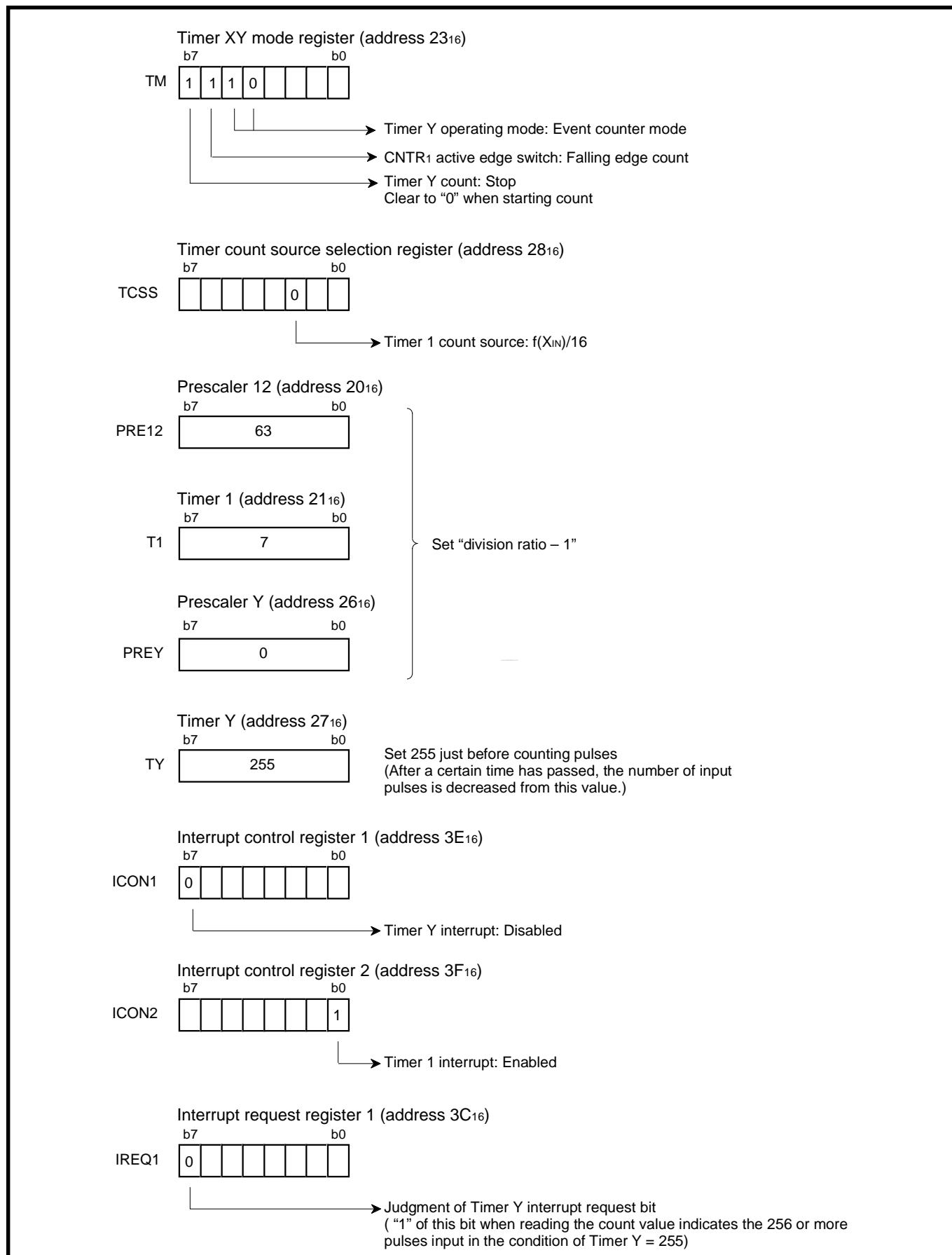


Fig. 2.3.20 Relevant registers setting



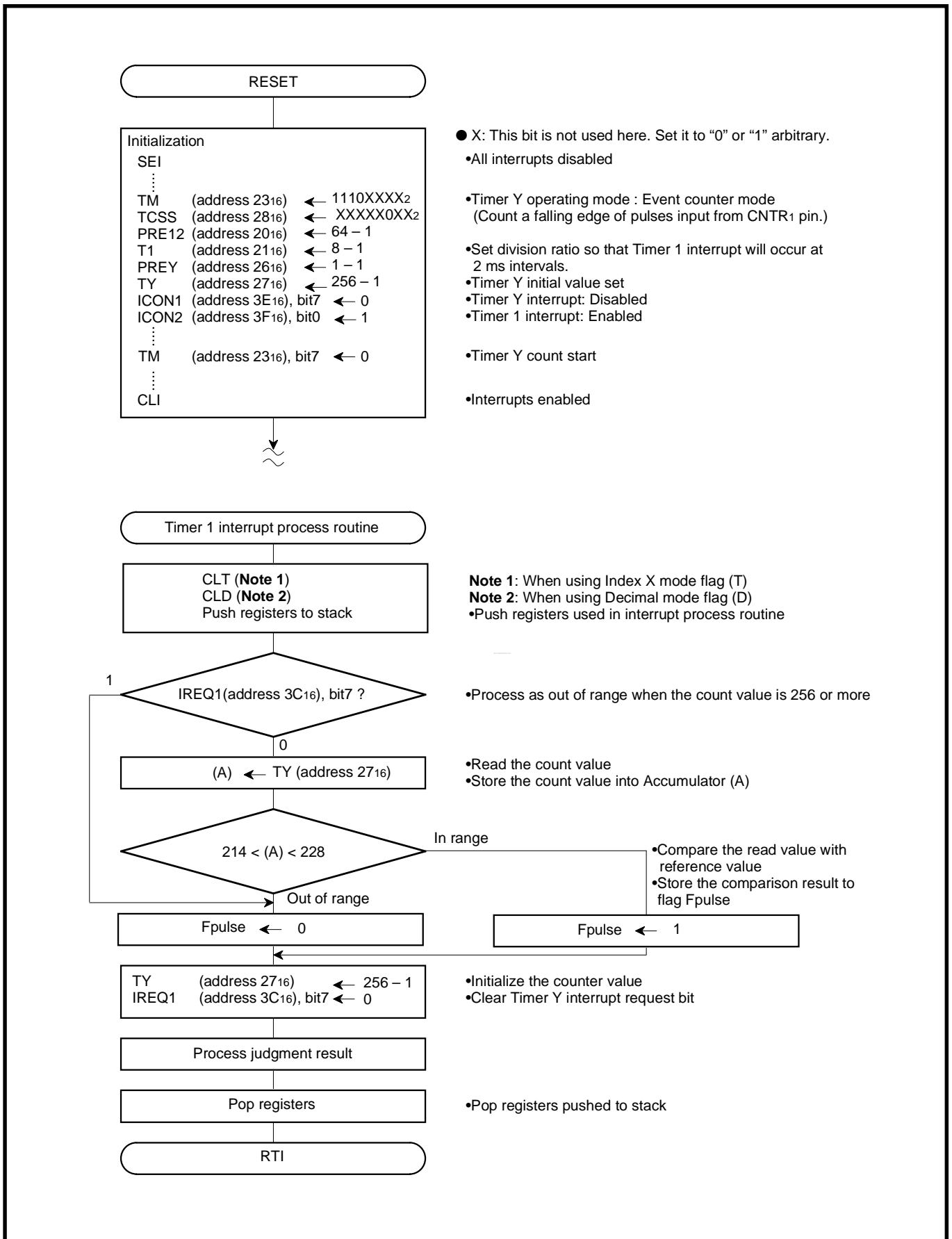


Fig. 2.3.21 Control procedure

# APPLICATION

## 2.3 Timer

### (5) Timer application example 4: Measurement of FG pulse width for motor

**Outline:** The timer X counts the “H” level width of the pulses input to the CNTR<sub>0</sub> pin. An underflow is detected by the timer X interrupt and an end of the input pulse “H” level is detected by the CNTR<sub>0</sub> interrupt.

**Specifications:** •The timer X counts the “H” level width of the FG pulse input to the CNTR<sub>0</sub> pin.

<Example>

When the clock frequency is 4.19 MHz, the count source is 3.8 μs, which is obtained by dividing the clock frequency by 16. Measurement can be made up to 250 ms in the range of FFFF<sub>16</sub> to 0000<sub>16</sub>.

Figure 2.3.22 shows the timers connection and setting of division ratio; Figure 2.3.23 shows the relevant registers setting; Figure 2.3.24 shows the control procedure.

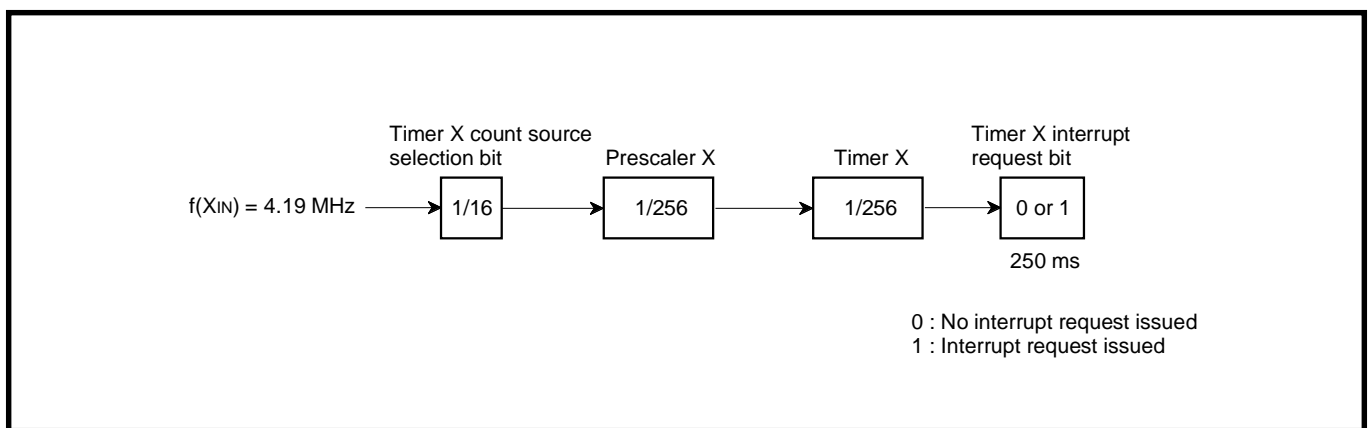
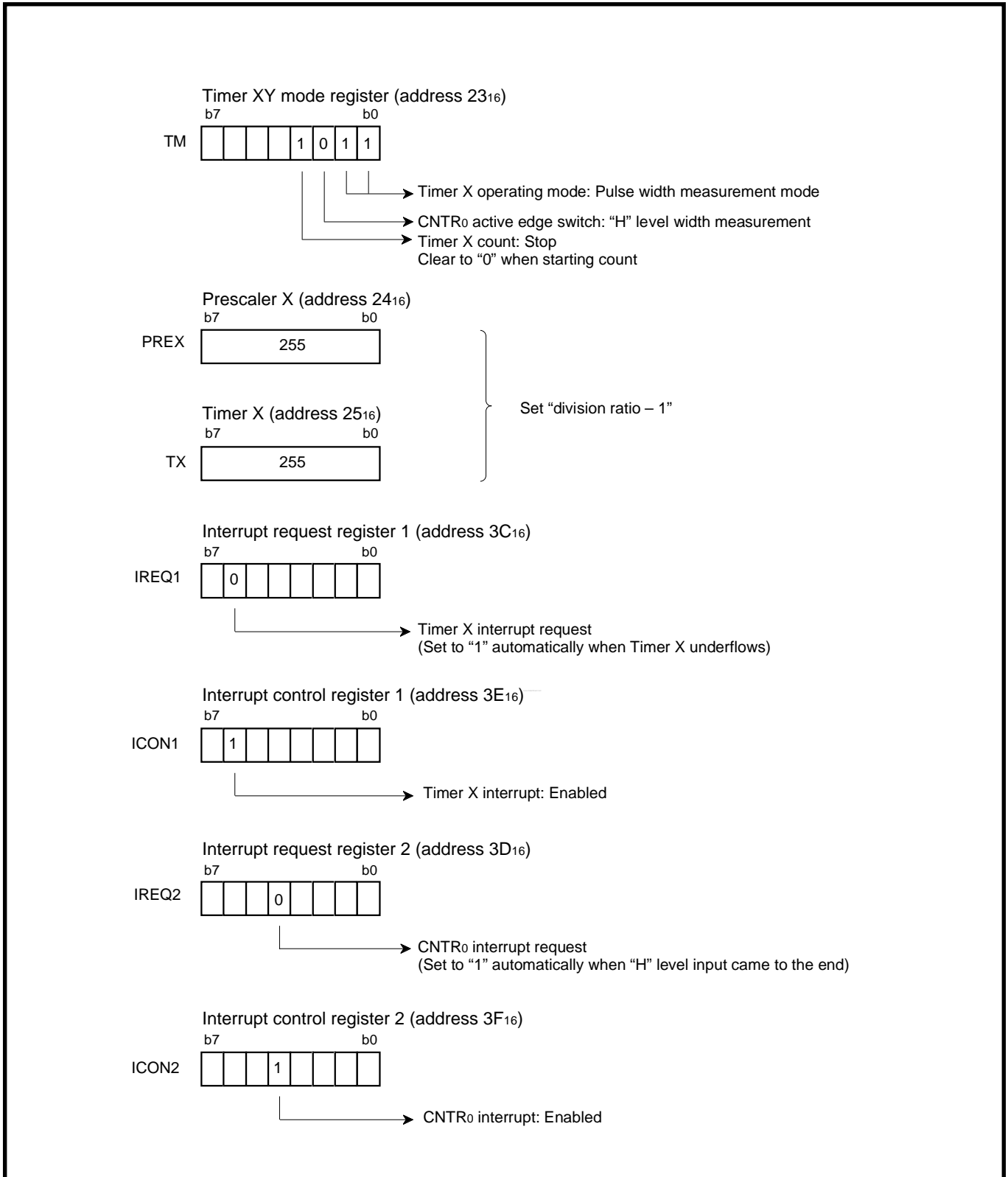


Fig. 2.3.22 Timers connection and setting of division ratios



**Fig. 2.3.23 Relevant registers setting**

# APPLICATION

## 2.3 Timer

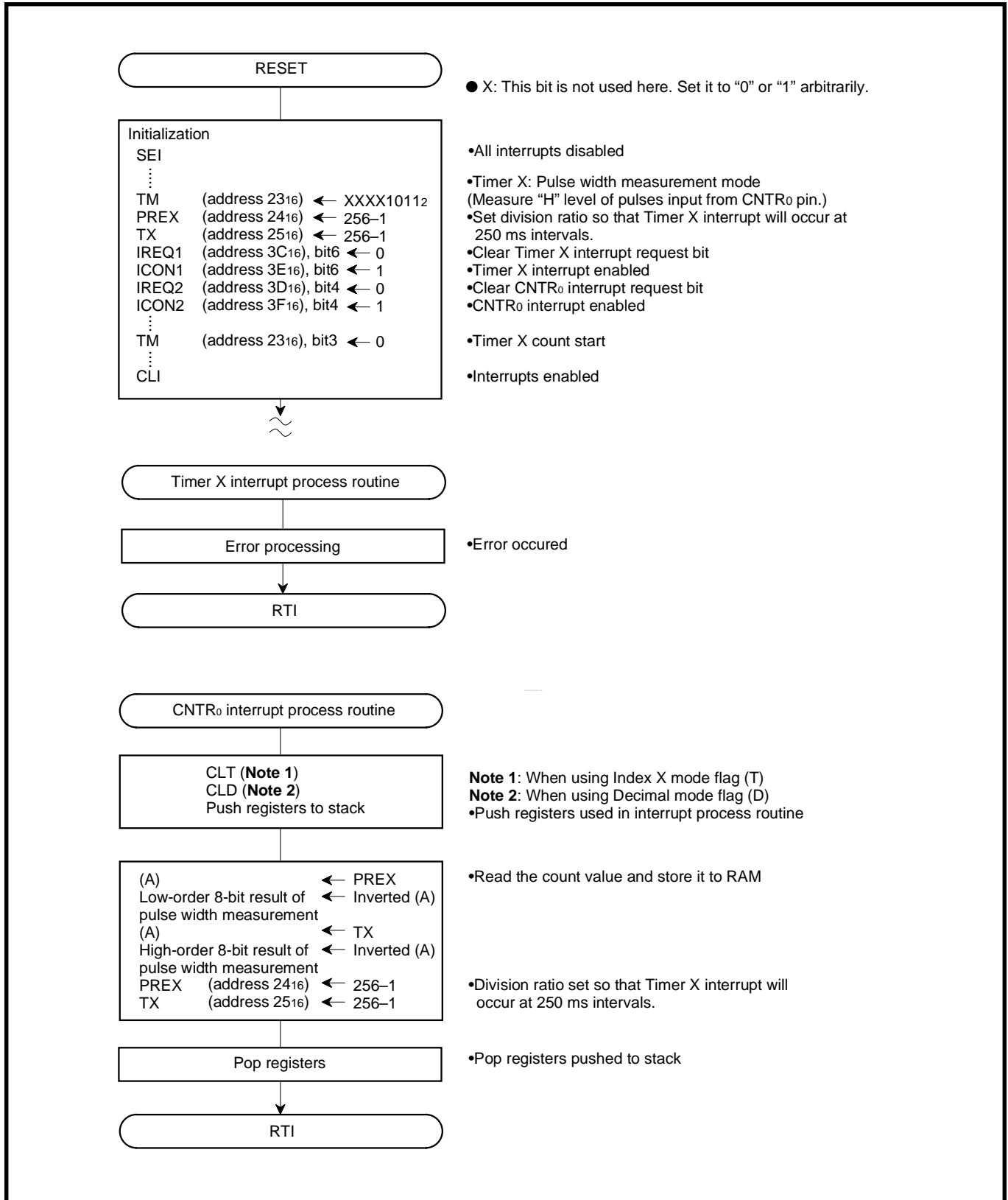


Fig. 2.3.24 Control procedure

### 2.3.4 Notes on timer

- If a value  $n$  (between 0 and 255) is written to a timer latch, the frequency division ratio is  $1/(n+1)$ .
- When switching the count source by the timer 12, X and Y count source selection bits, the value of timer count is altered in unconsiderable amount owing to generating of thin pulses in the count input signals.

Therefore, select the timer count source before set the value to the prescaler and the timer.

# APPLICATION

## 2.4 Serial I/O

### 2.4 Serial I/O

This paragraph explains the registers setting method and the notes relevant to the Serial I/O.

#### 2.4.1 Memory map

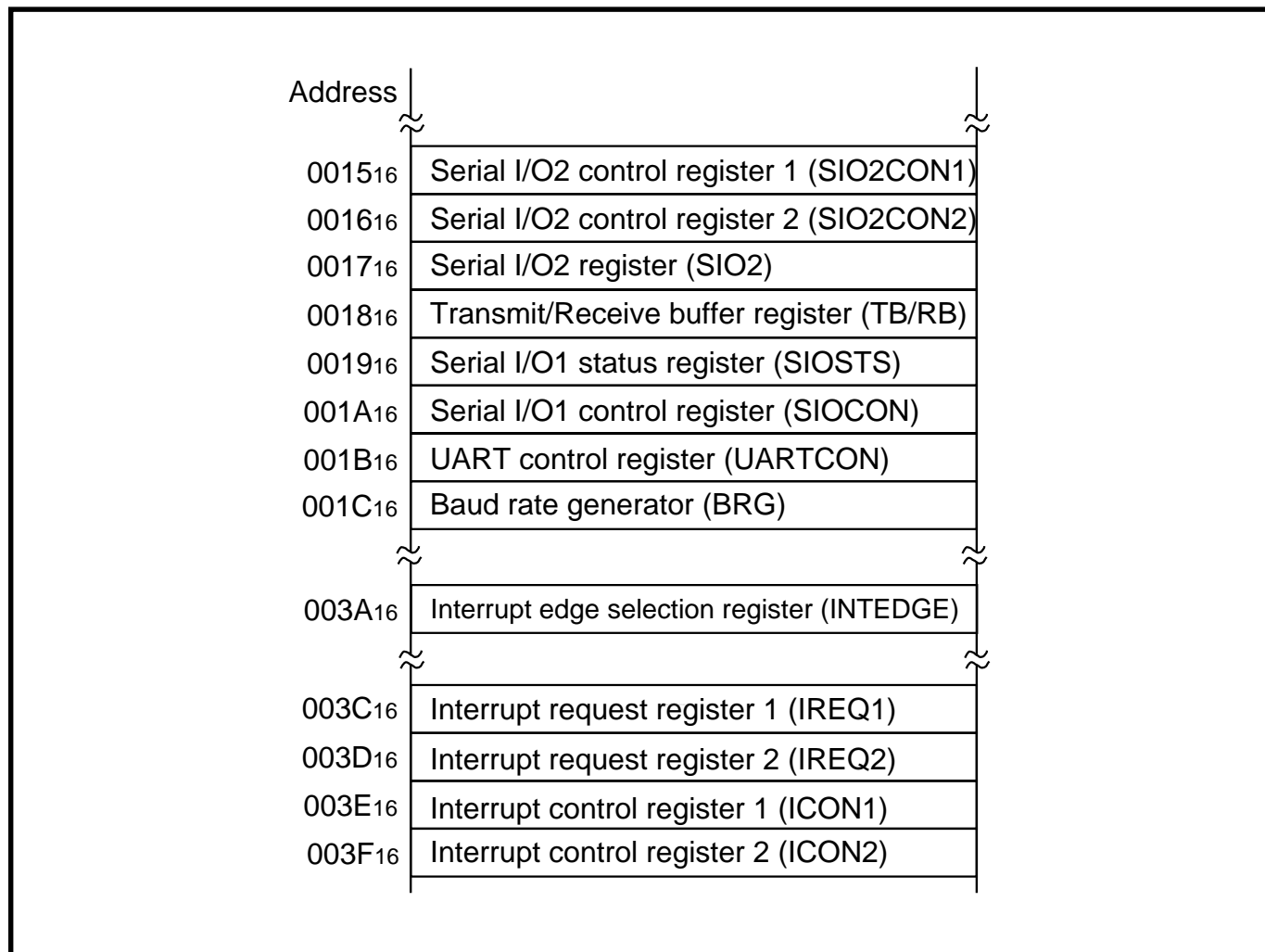


Fig. 2.4.1 Memory map of registers relevant to Serial I/O

### 2.4.2 Relevant registers

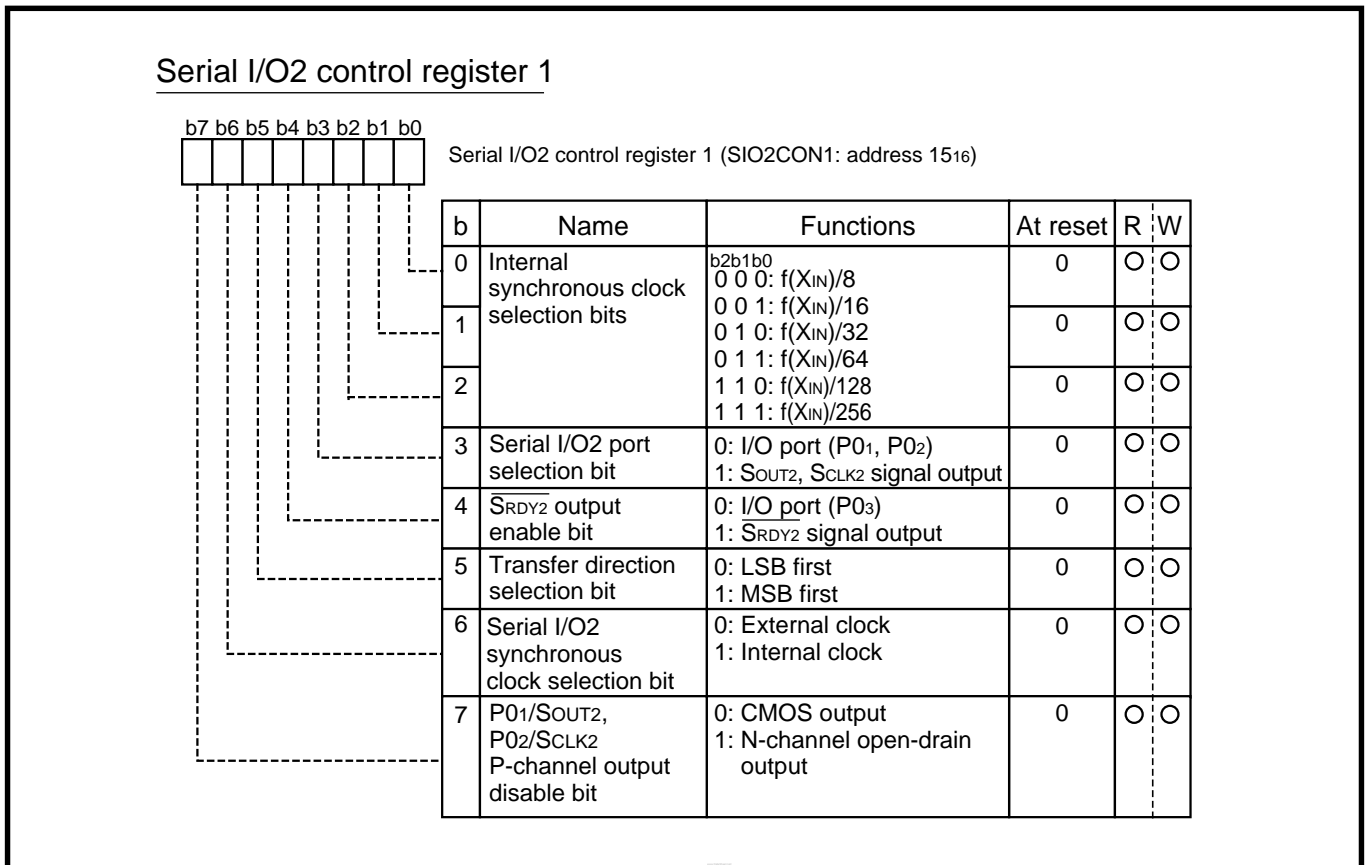


Fig. 2.4.2 Structure of Serial I/O2 control register 1

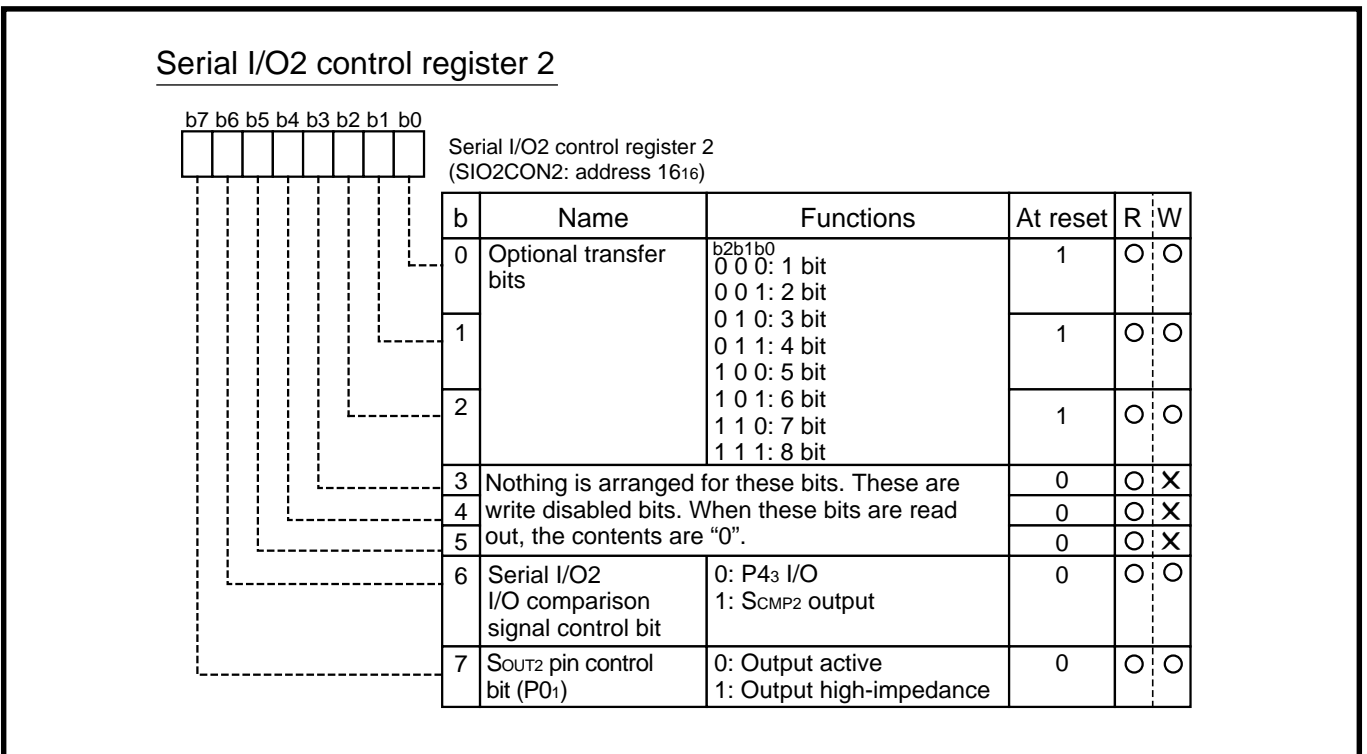


Fig. 2.4.3 Structure of Serial I/O2 control register 2

# APPLICATION

## 2.4 Serial I/O

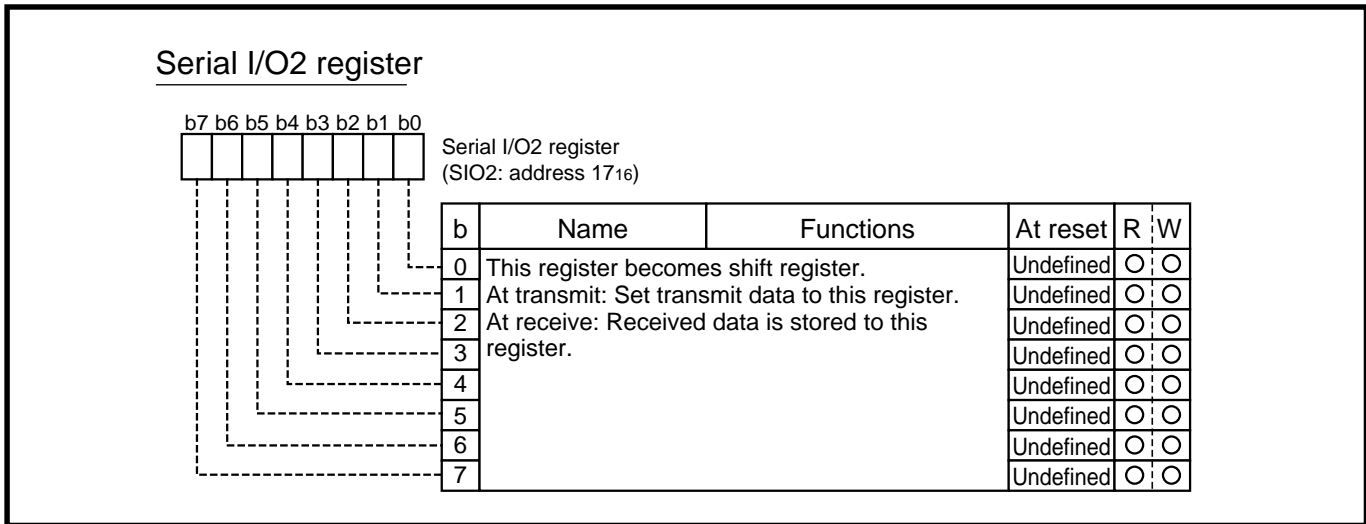


Fig. 2.4.4 Structure of Serial I/O2 register

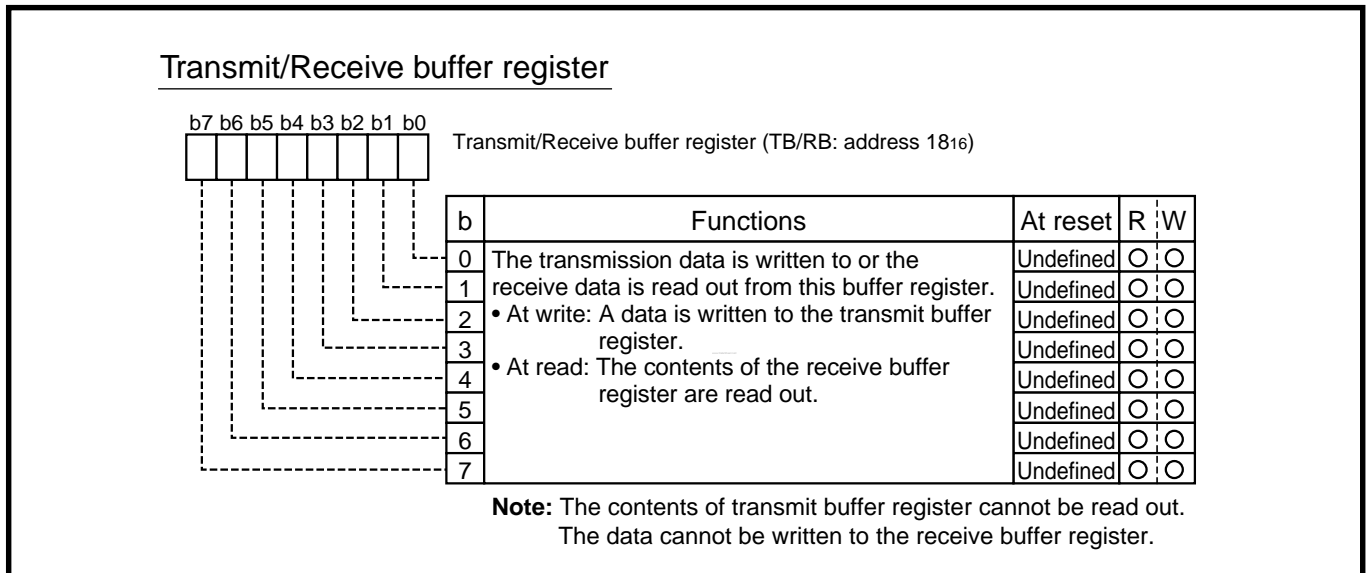


Fig. 2.4.5 Structure of Transmit/Receive buffer register



### Serial I/O1 status register

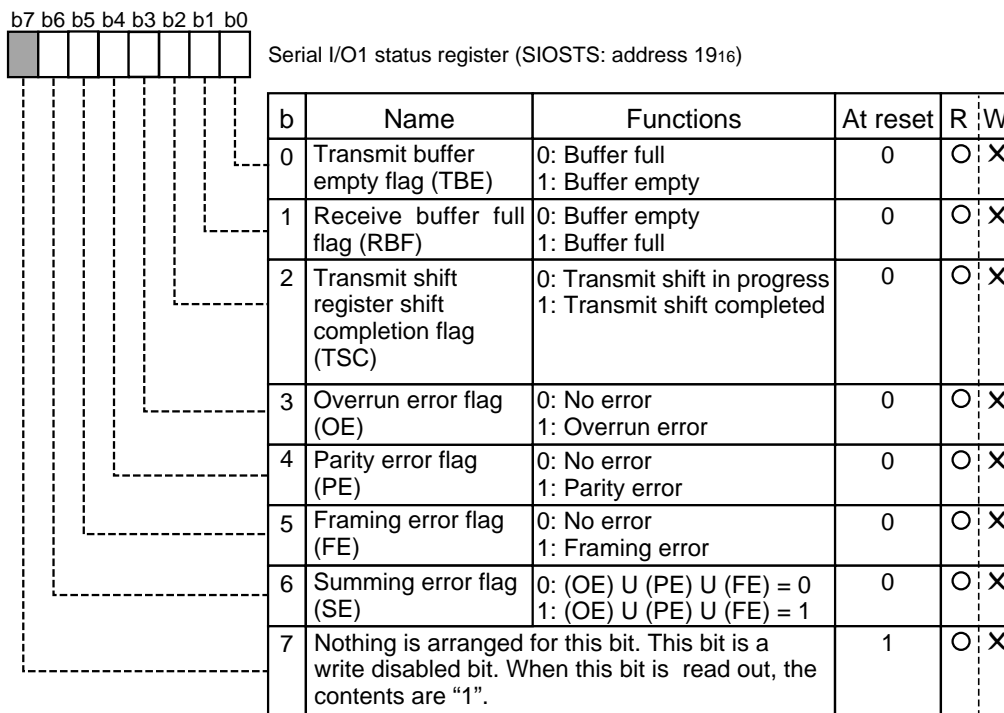


Fig. 2.4.6 Structure of Serial I/O1 status register

# APPLICATION

## 2.4 Serial I/O

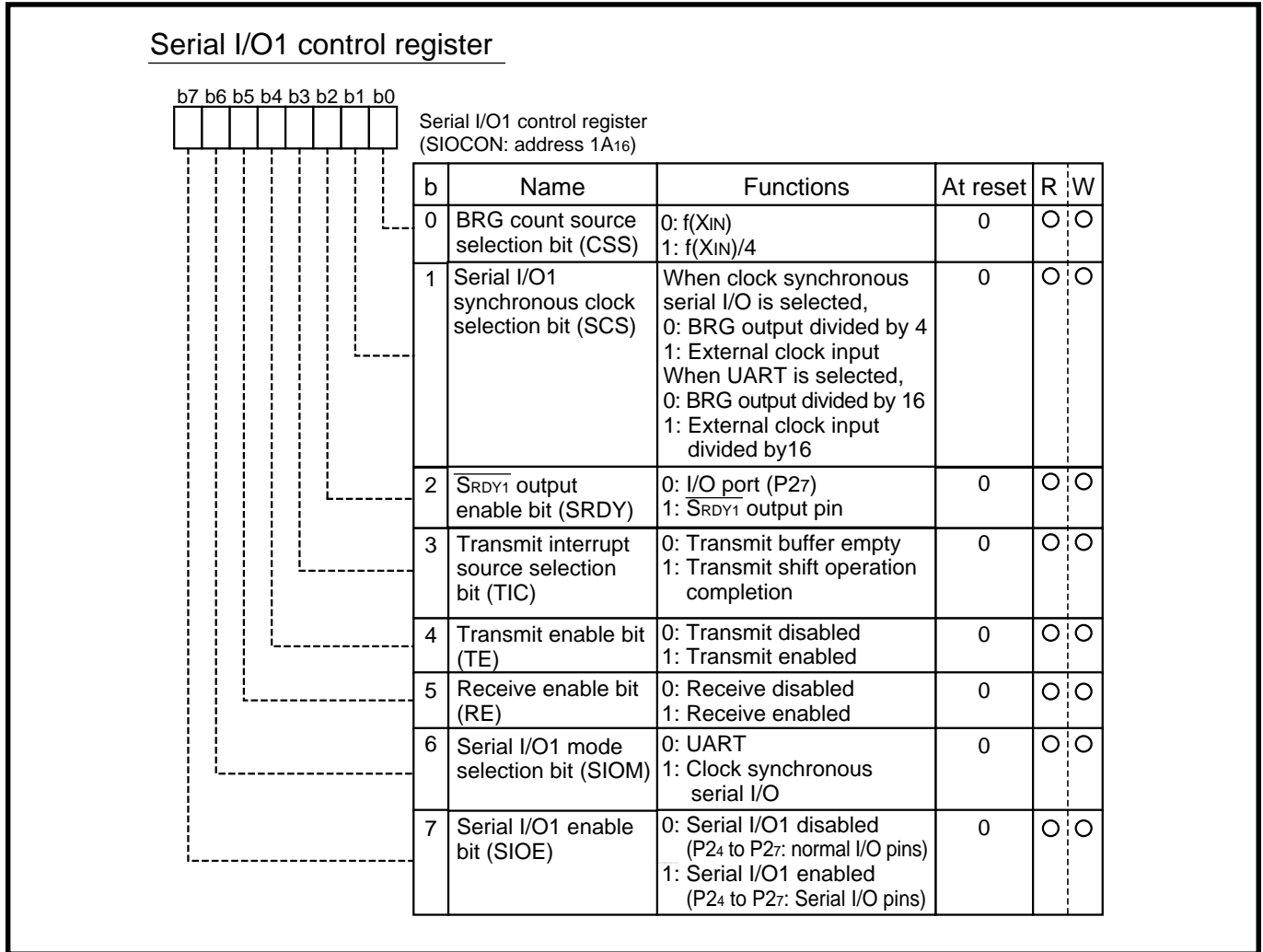


Fig. 2.4.7 Structure of Serial I/O1 control register

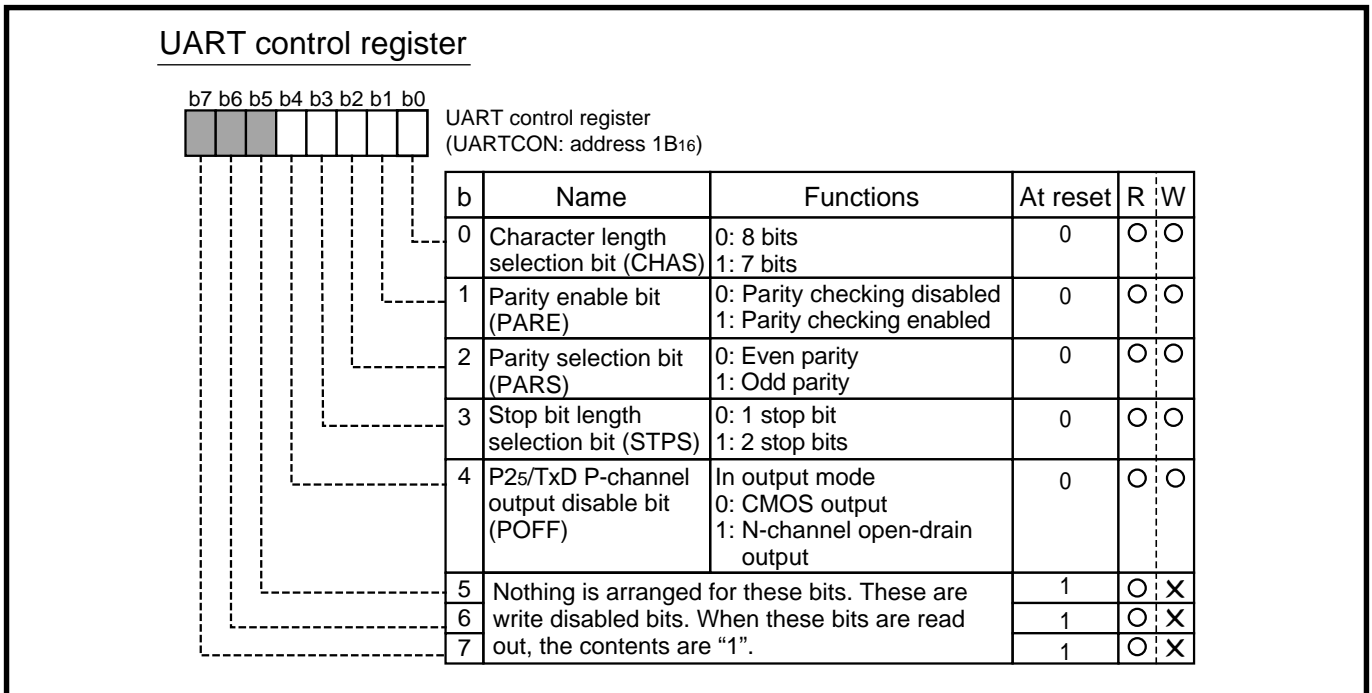


Fig. 2.4.8 Structure of UART control register

### Baud rate generator

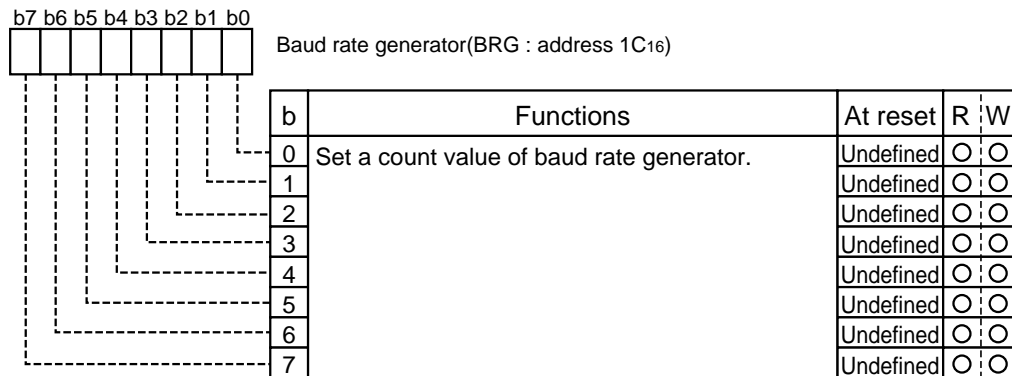


Fig. 2.4.9 Structure of Baud rate generator

### Interrupt edge selection register

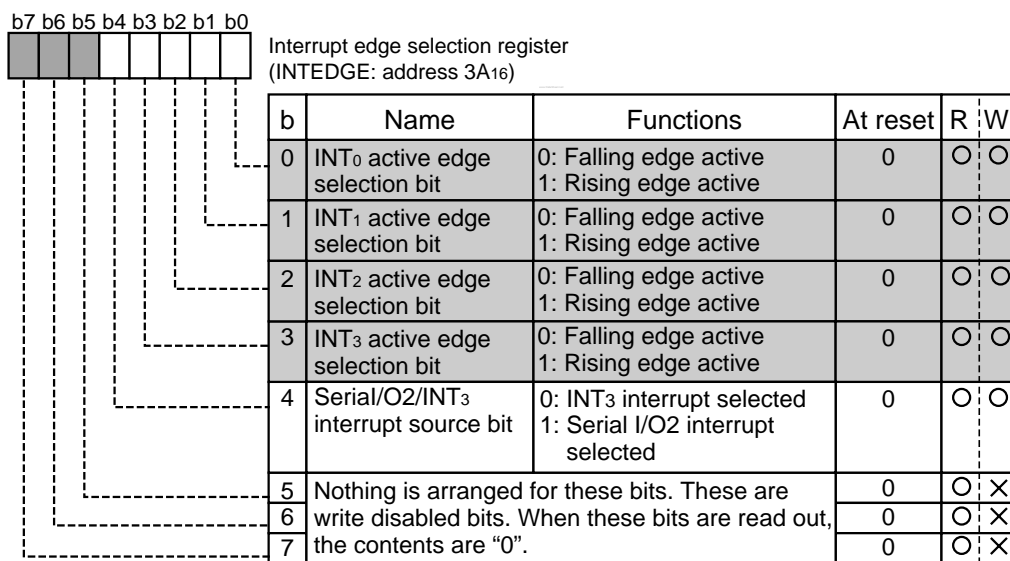


Fig. 2.4.10 Structure of Interrupt edge selection register

# APPLICATION

## 2.4 Serial I/O

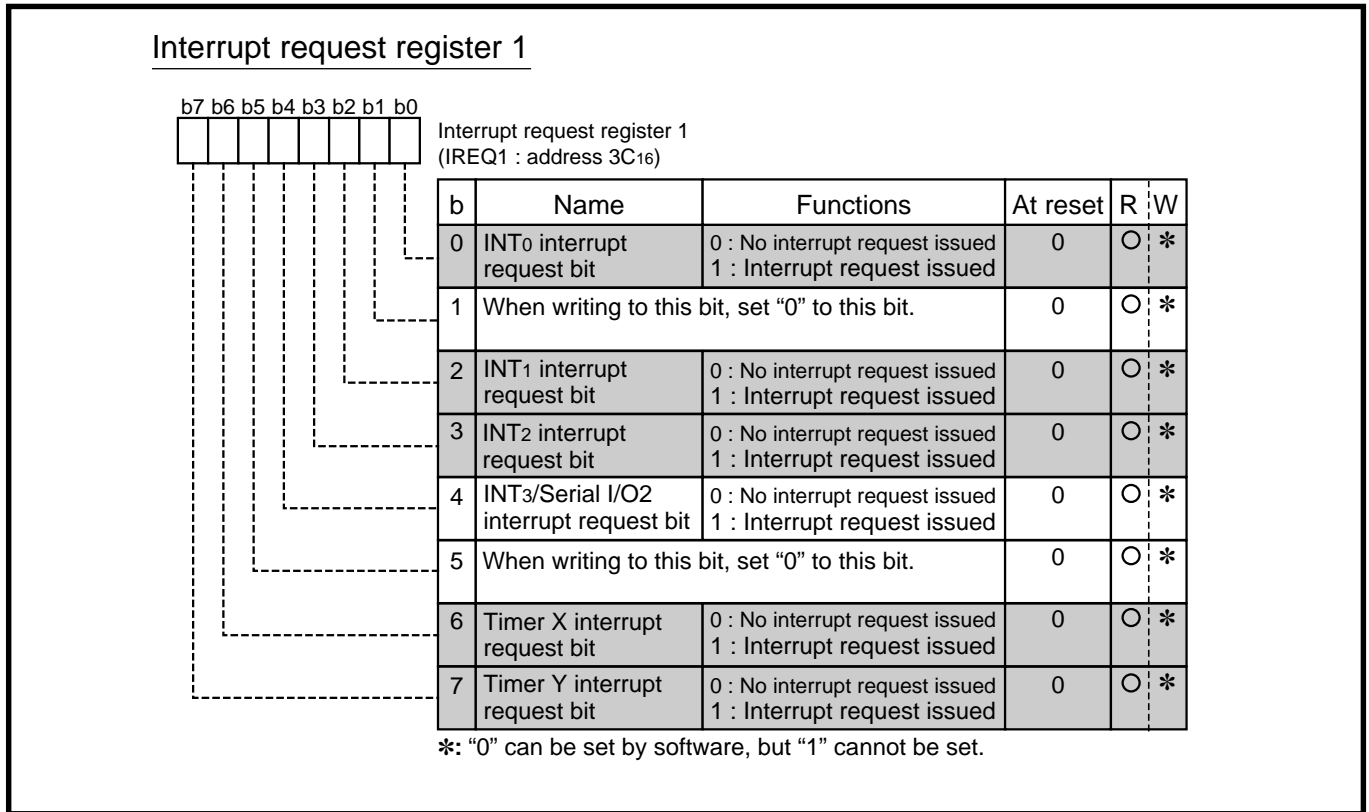


Fig. 2.4.11 Structure of Interrupt request register 1

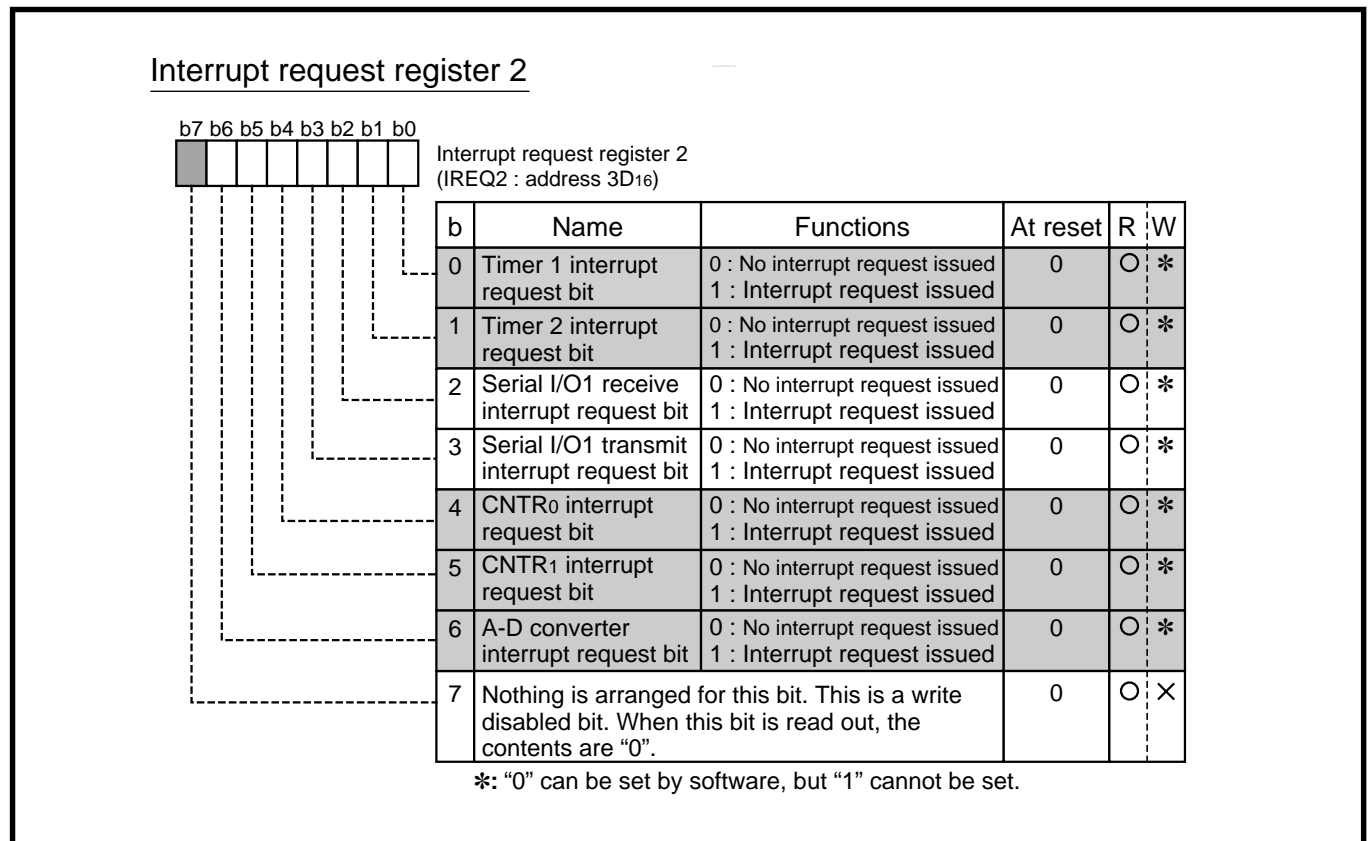


Fig. 2.4.12 Structure of Interrupt request register 2

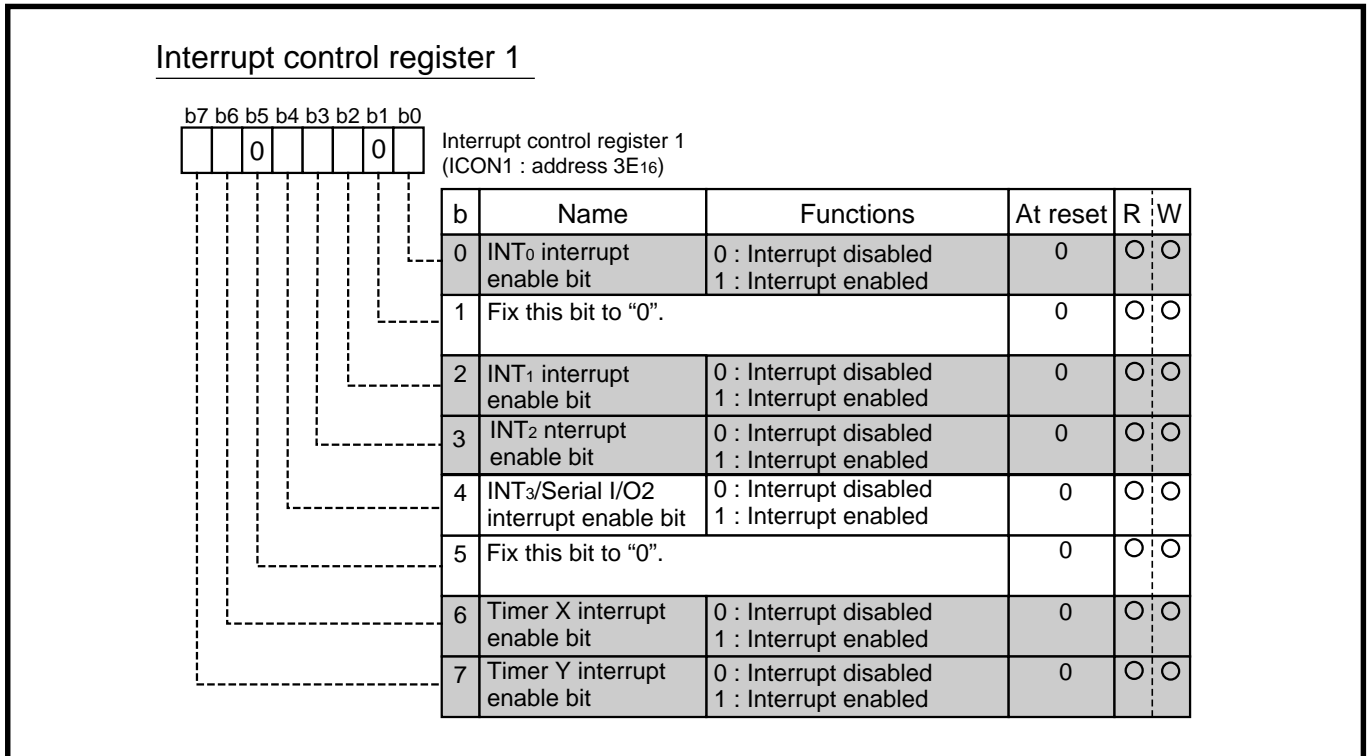


Fig. 2.4.13 Structure of Interrupt control register 1

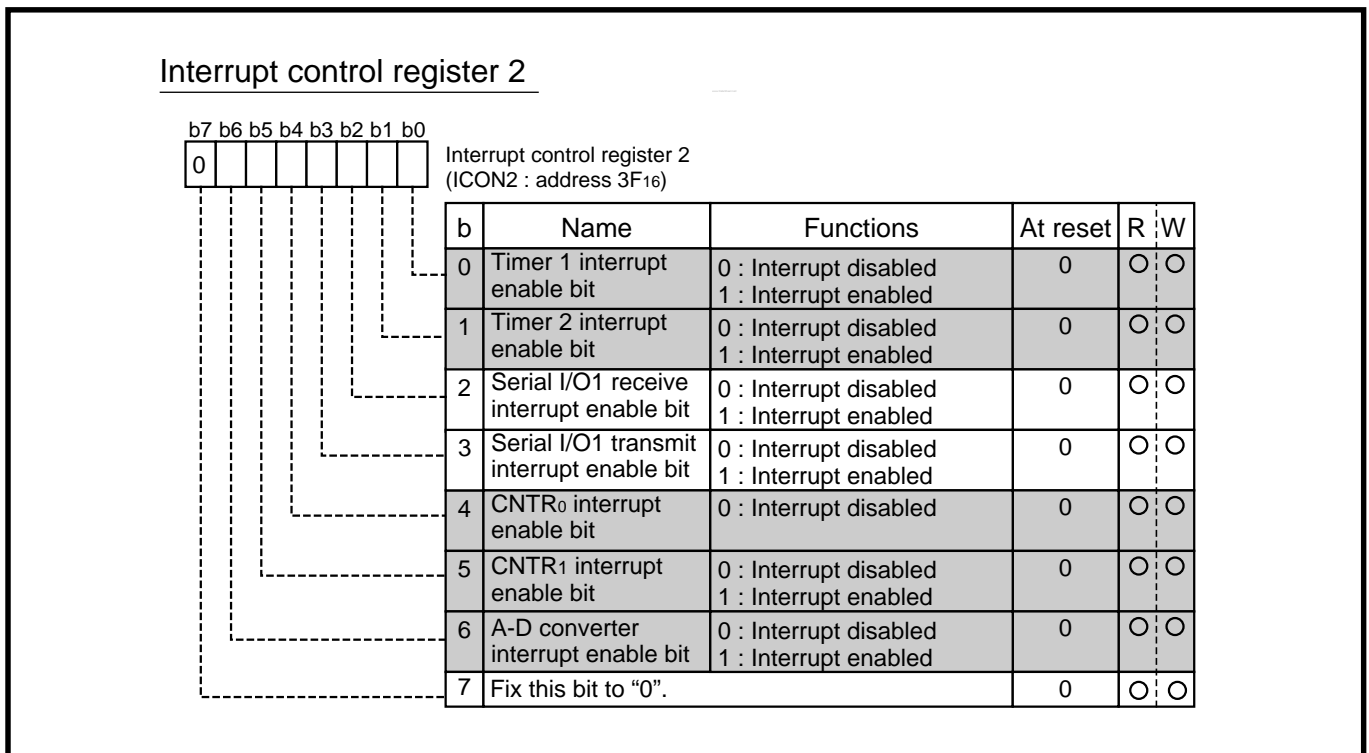


Fig. 2.4.14 Structure of Interrupt control register 2

# APPLICATION

## 2.4 Serial I/O

### 2.4.3 Serial I/O connection examples

#### (1) Control of peripheral IC equipped with CS pin

Figure 2.4.15 shows connection examples of a peripheral IC equipped with the CS pin. There are connection examples using a clock synchronous serial I/O mode.

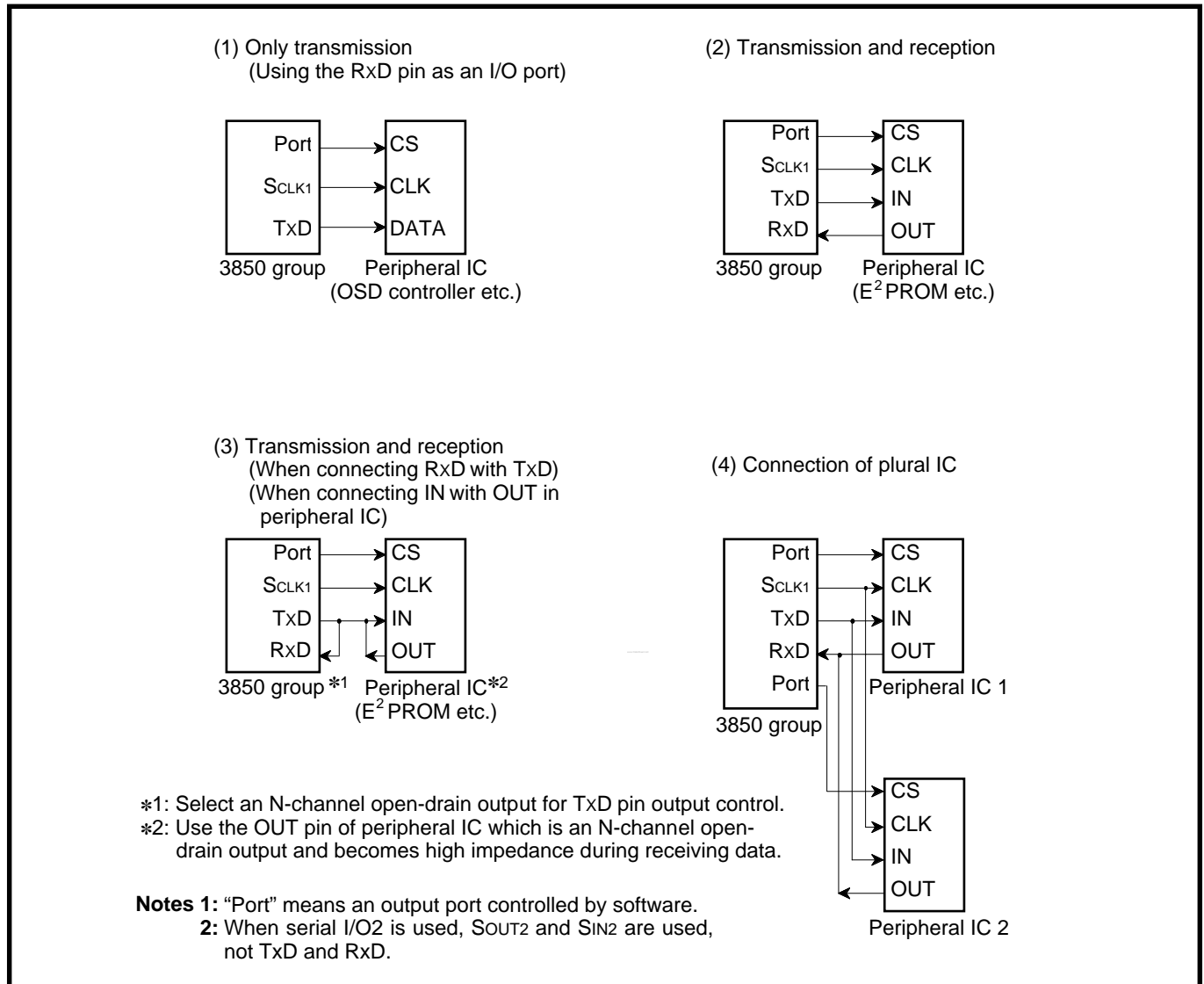


Fig. 2.4.15 Serial I/O connection examples (1)

### (2) Connection with microcomputer

Figure 2.4.16 shows connection examples with another microcomputer.

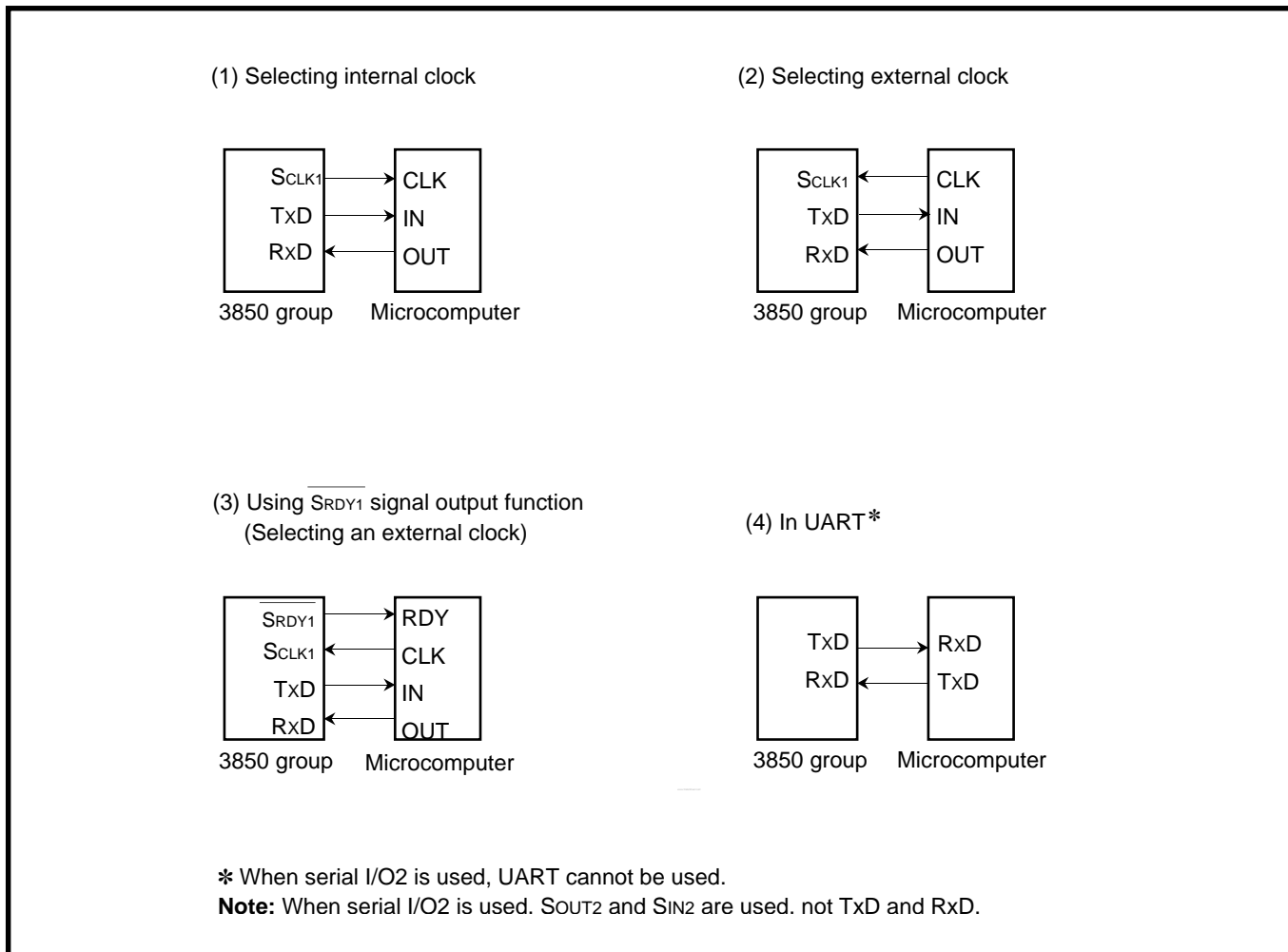


Fig. 2.4.16 Serial I/O connection examples (2)

# APPLICATION

## 2.4 Serial I/O

### 2.4.4 Setting of serial I/O transfer data format

A clock synchronous or clock asynchronous (UART) can be selected as a data format of Serial I/O1. A clock synchronous is used as a data format of Serial I/O2.

Figure 2.4.17 shows the serial I/O transfer data format.

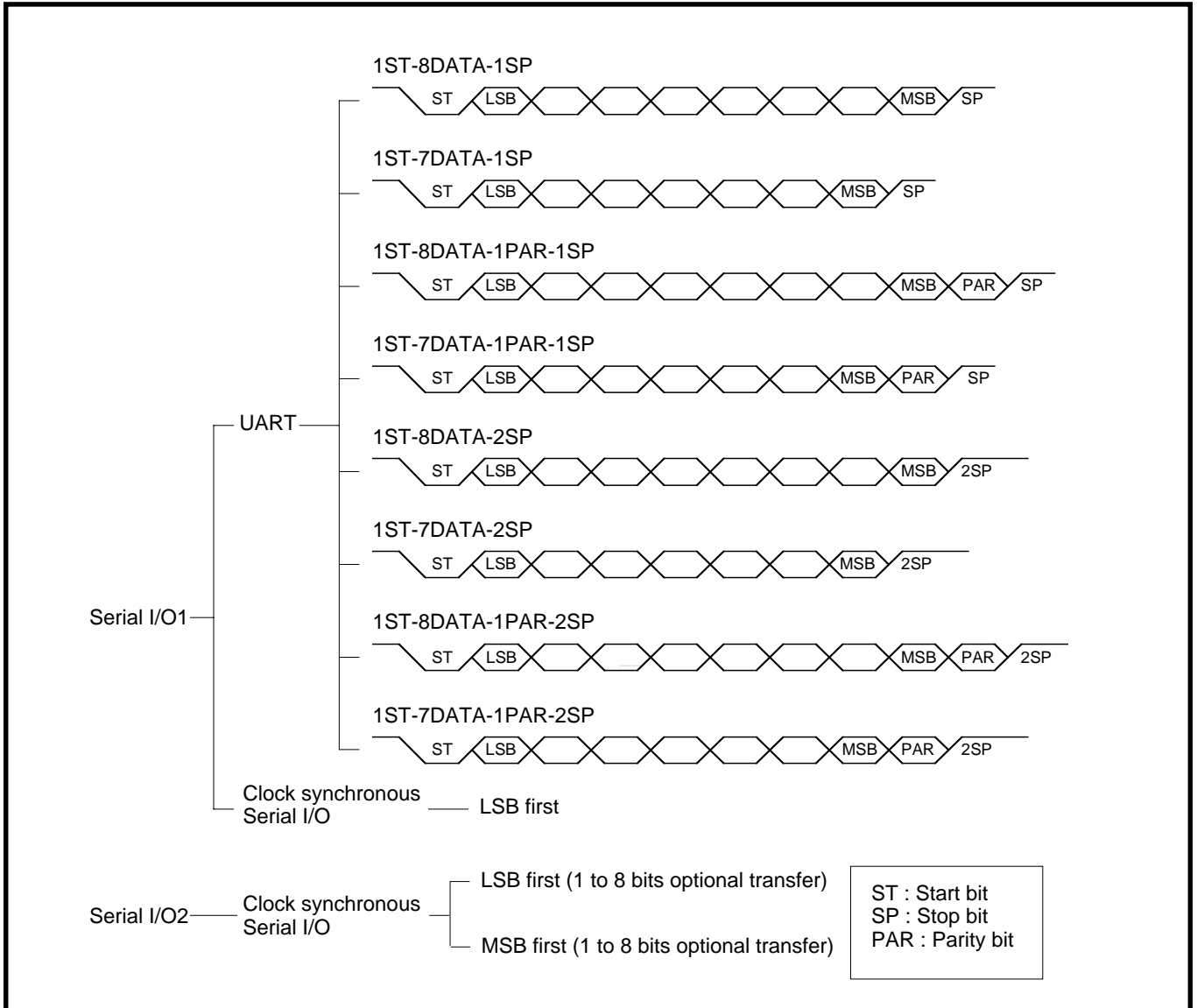


Fig. 2.4.17 Serial I/O transfer data format



### 2.4.5 Serial I/O application examples

#### (1) Communication using clock synchronous serial I/O (transmit/receive)

**Outline :** 2-byte data is transmitted and received, using the clock synchronous serial I/O.  
The  $\overline{\text{SRDY1}}$  signal is used for communication control.

Figure 2.4.18 shows a connection diagram, and Figure 2.4.19 shows a timing chart.  
Figure 2.4.20 shows a registers setting relevant to the transmitting side, and Figure 2.4.21 shows registers setting relevant to the receiving side.

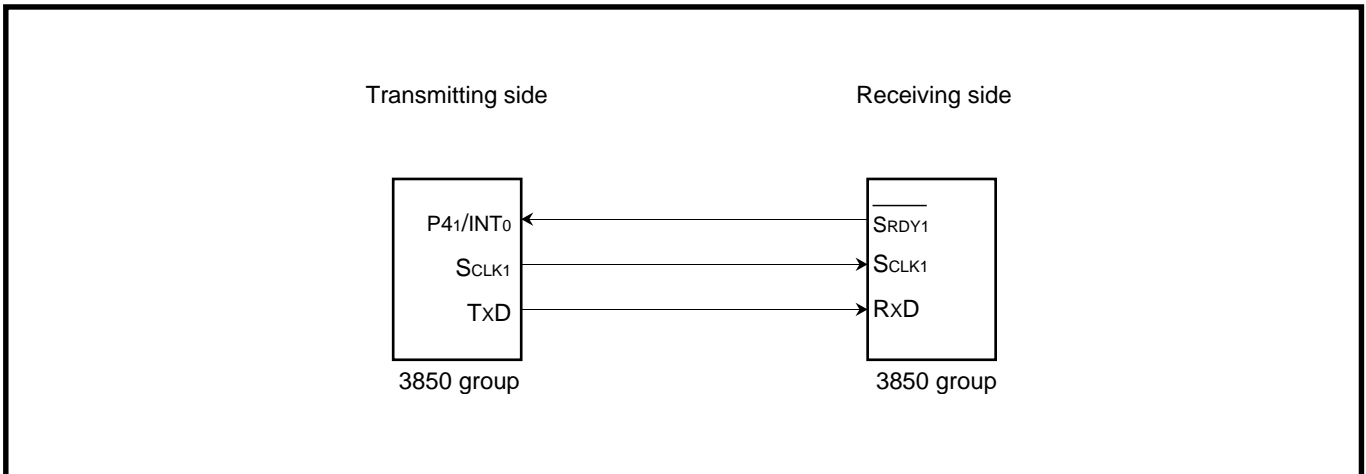


Fig. 2.4.18 Connection diagram

- Specifications :**
- The Serial I/O is used (clock synchronous serial I/O is selected.)
  - Synchronous clock frequency : 125 kHz ( $f(\text{X}_{\text{IN}}) = 4 \text{ MHz}$  is divided by 32)
  - The  $\overline{\text{SRDY1}}$  (receivable signal) is used.
  - The receiving side outputs the  $\overline{\text{SRDY1}}$  signal at intervals of 2 ms (generated by timer), and 2-byte data is transferred from the transmitting side to the receiving side.

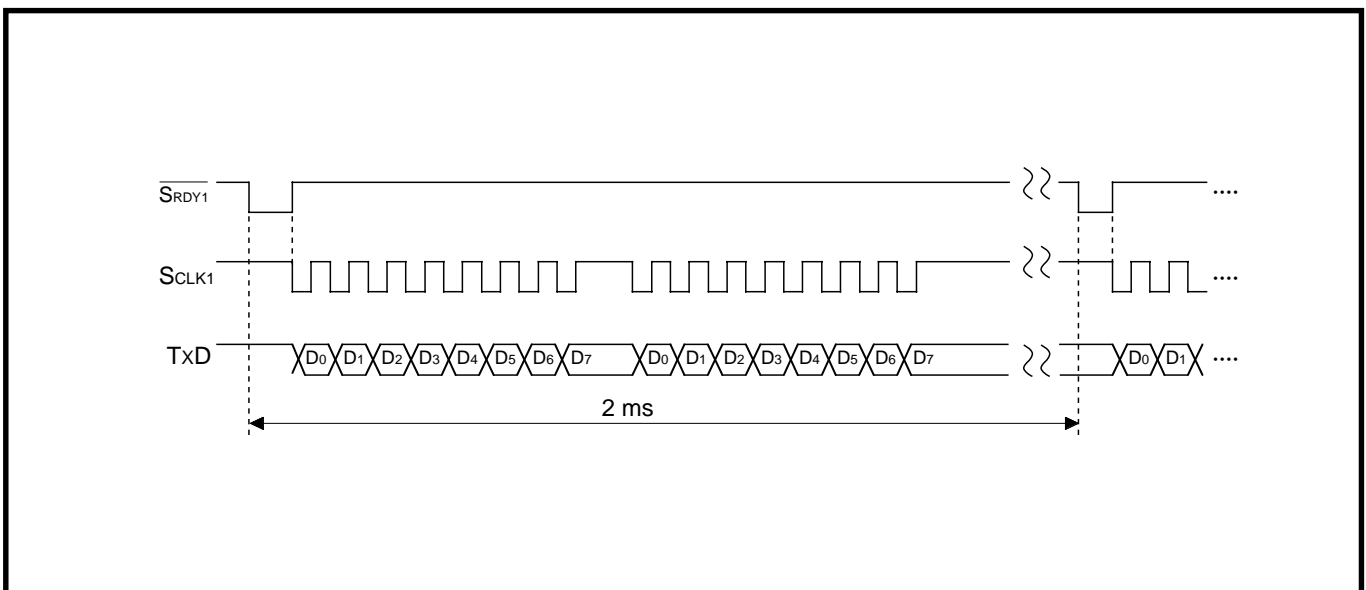


Fig. 2.4.19 Timing chart

# APPLICATION

## 2.4 Serial I/O

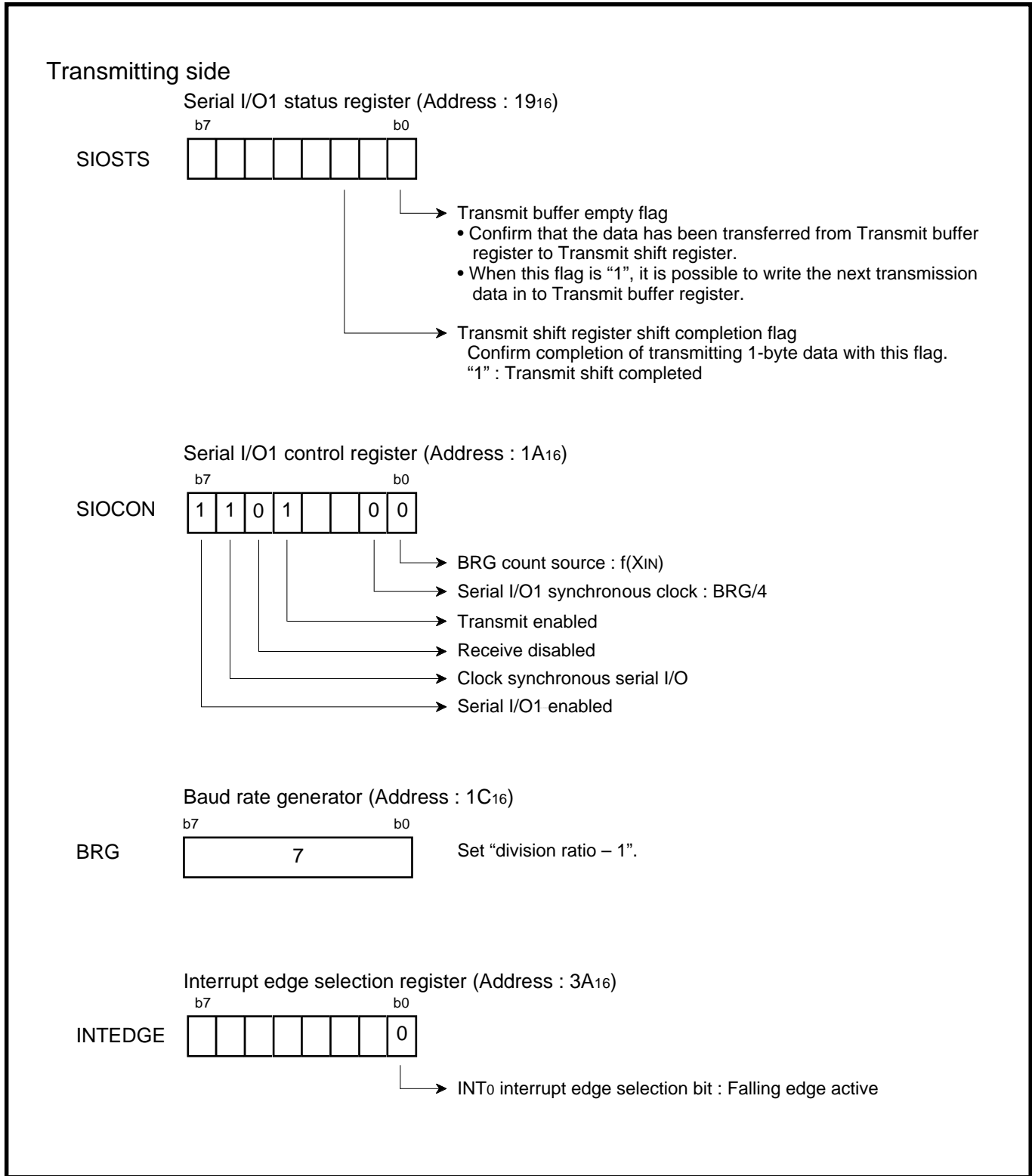


Fig. 2.4.20 Registers setting relevant to transmitting side

Receiving side

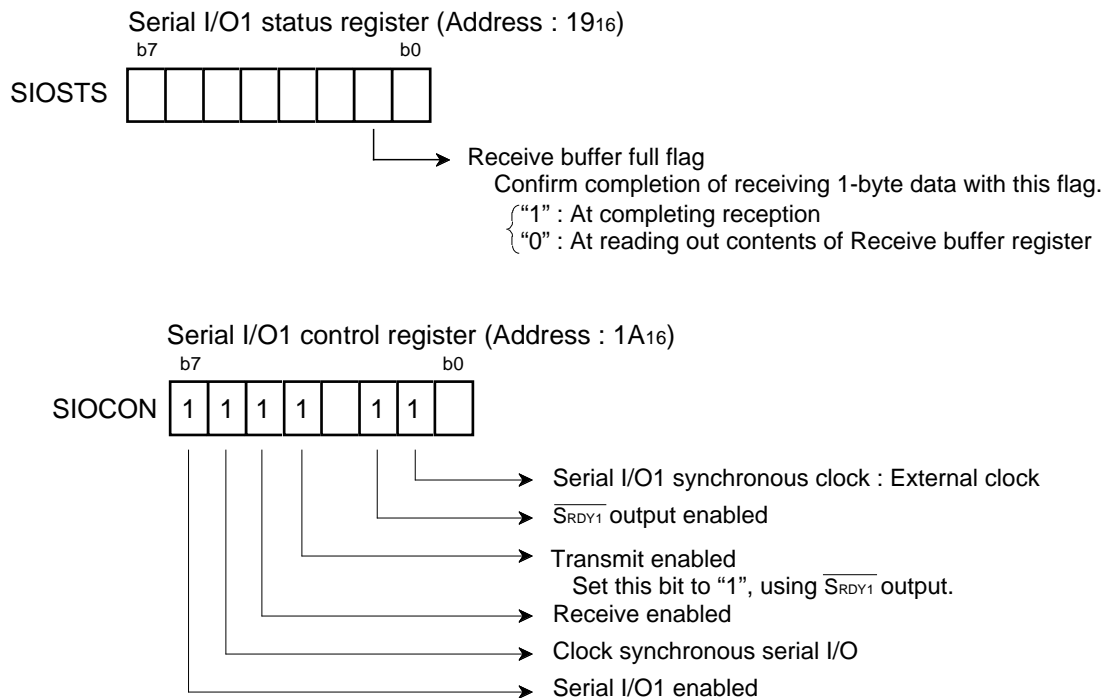


Fig. 2.4.21 Registers setting relevant to receiving side

# APPLICATION

## 2.4 Serial I/O

Figure 2.4.22 shows a control procedure of the transmitting side, and Figure 2.4.23 shows a control procedure of the receiving side.

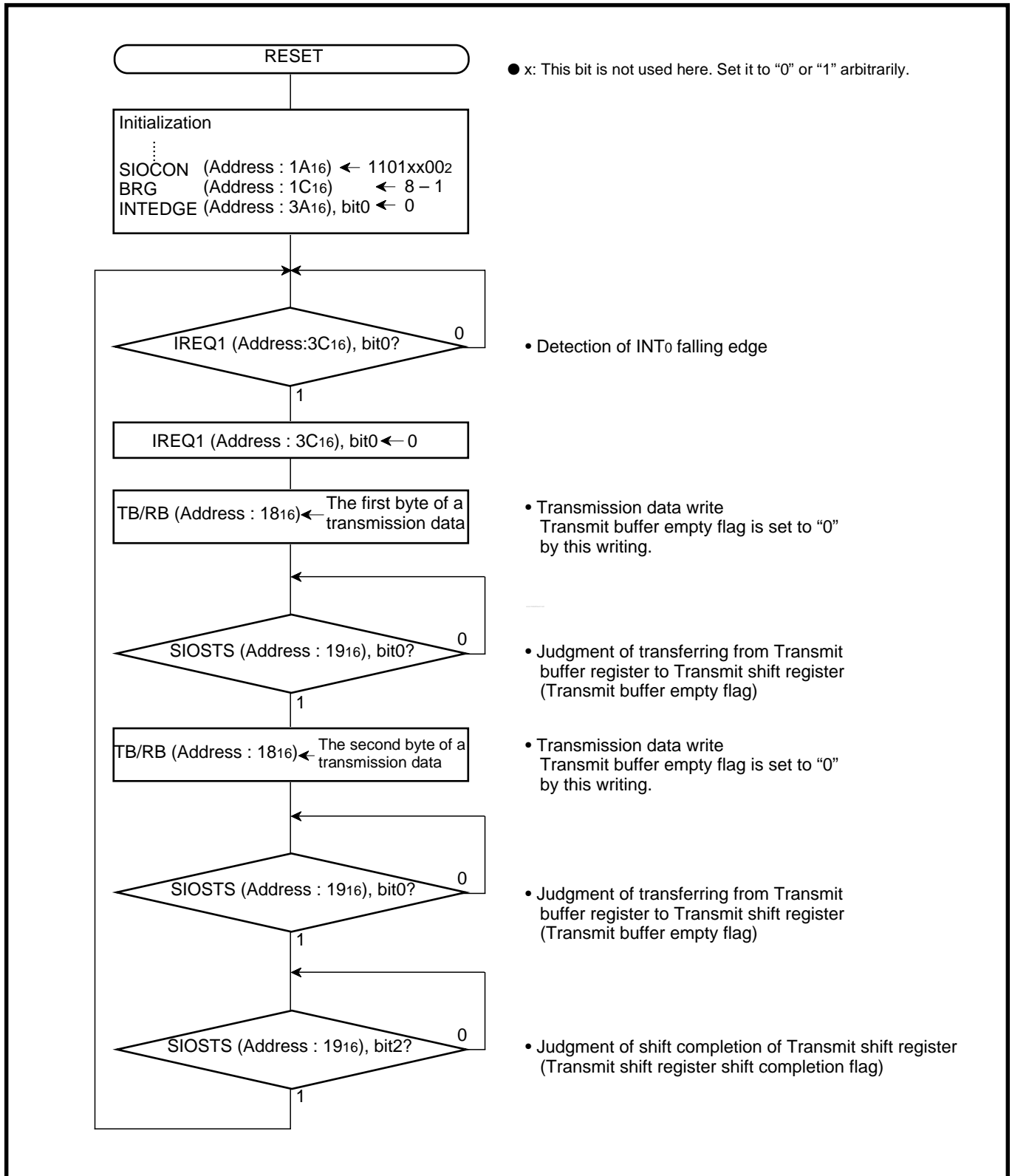


Fig. 2.4.22 Control procedure of transmitting side

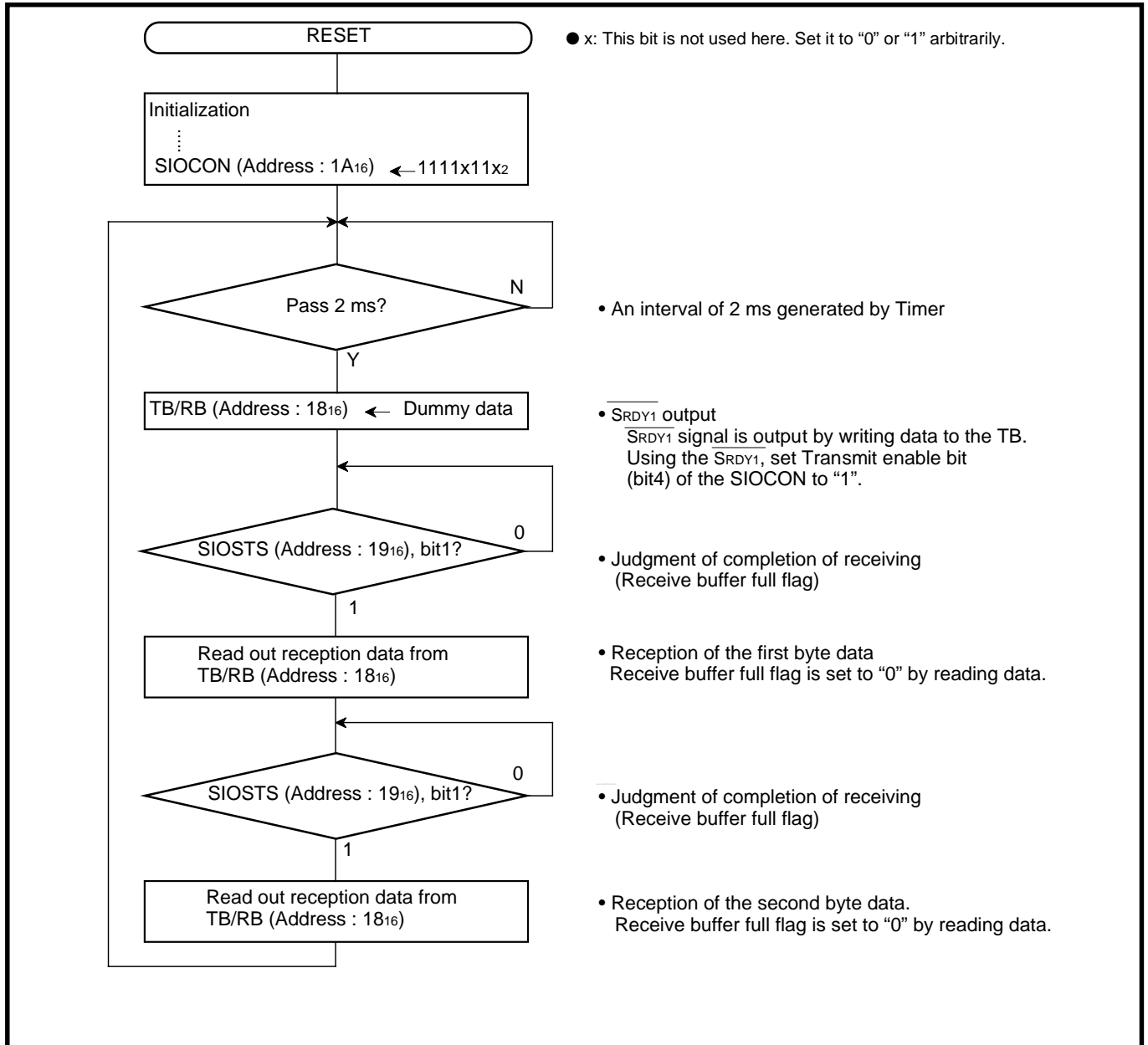


Fig. 2.4.23 Control procedure of receiving side

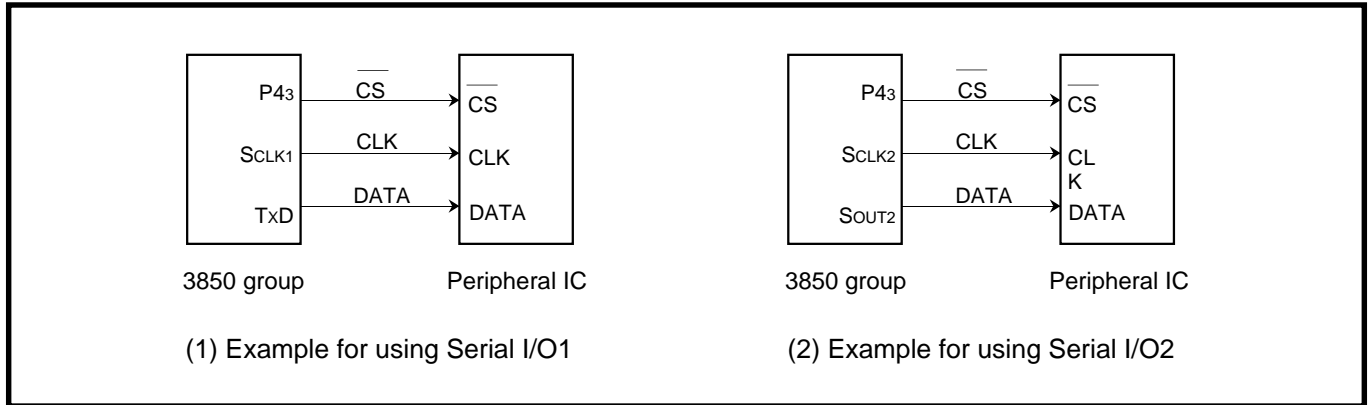
# APPLICATION

## 2.4 Serial I/O

### (2) Output of serial data (control of peripheral IC)

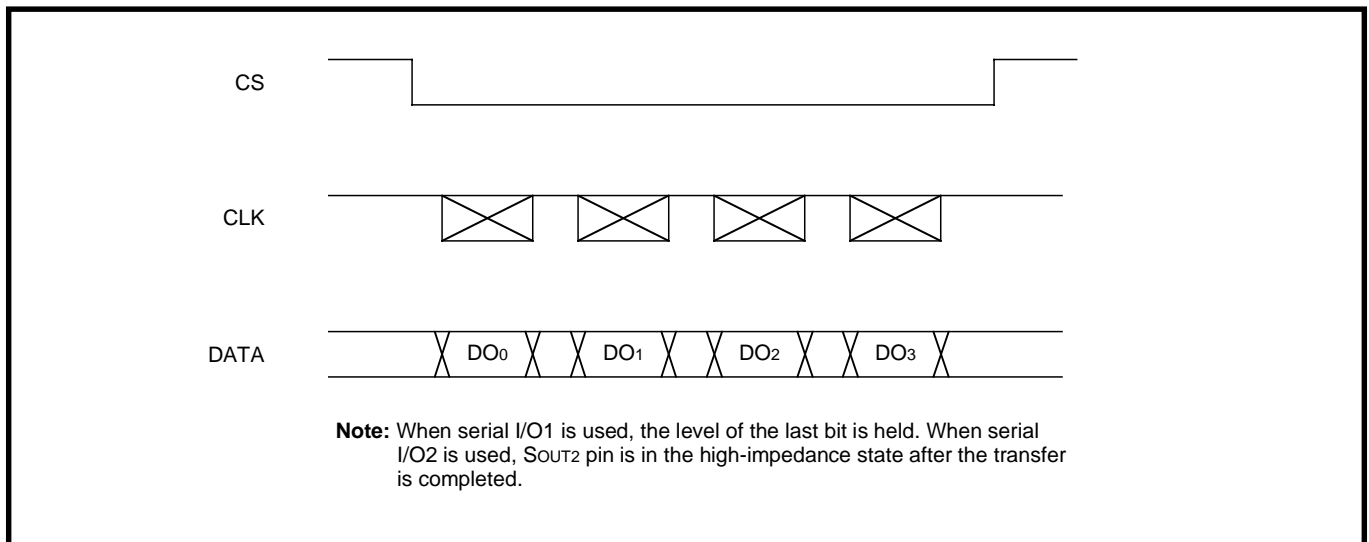
**Outline :** 4-byte data is transmitted and received, using the clock synchronous serial I/O.  
The CS signal is output to a peripheral IC through port P4<sub>3</sub>.

Figure 2.4.24 shows a connection diagram, and Figure 2.4.25 shows a timing chart.



**Fig. 2.4.24 Connection diagram**

- Specifications :**
- The Serial I/O is used (clock synchronous serial I/O is selected.)
  - Synchronous clock frequency : 125 kHz ( $f(X_{IN}) = 4 \text{ MHz}$  is divided by 32)
  - Transfer direction : LSB first
  - The Serial I/O interrupt is not used.
  - Port P4<sub>3</sub> is connected to the CS pin ("L" active) of the peripheral IC for transmission control; the output level of port P4<sub>3</sub> is controlled by software.



**Fig. 2.4.25 Timing chart (Serial I/O1)**

Figure 2.4.26 shows registers setting relevant to Serial I/O1, and Figure 2.4.27 shows a setting of serial I/O1 transmission data.

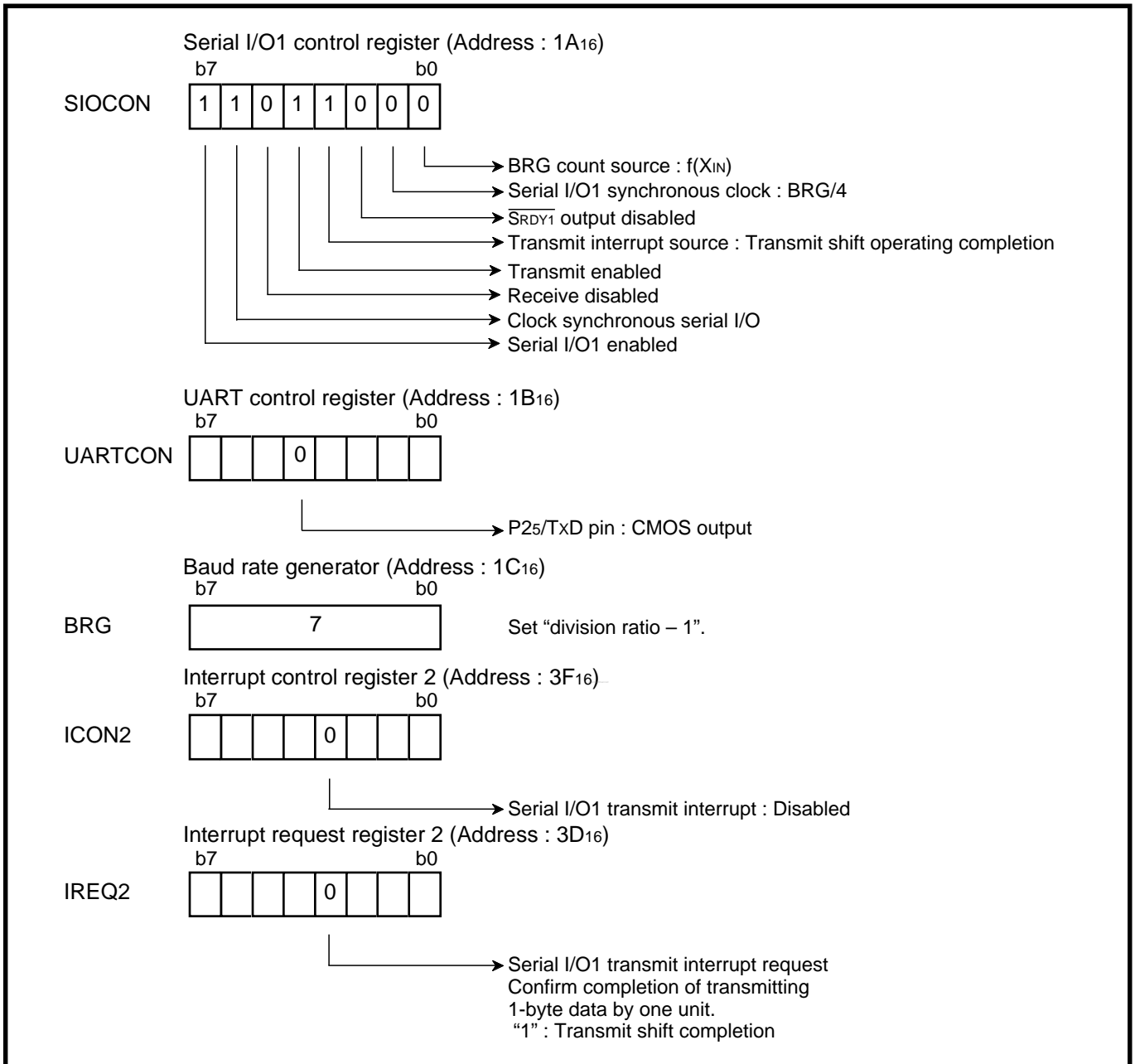


Fig. 2.4.26 Registers setting relevant to Serial I/O1

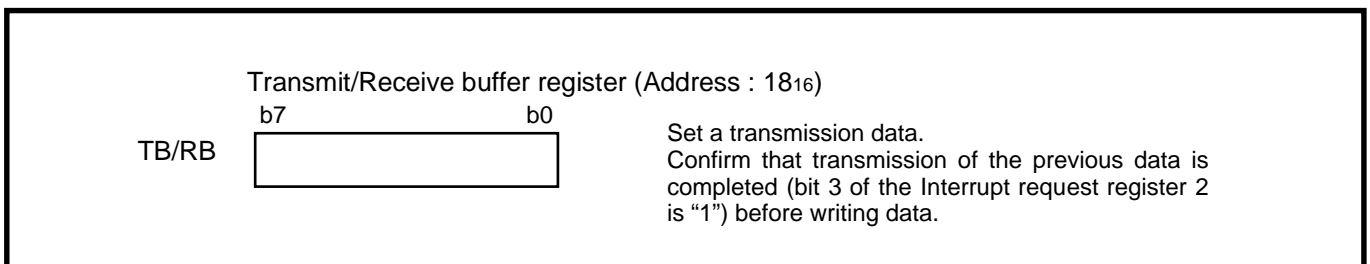


Fig. 2.4.27 Setting of serial I/O1 transmission data

# APPLICATION

## 2.4 Serial I/O

### Example for using Serial I/O1

When the registers are set as shown in Figure 2.4.26, the Serial I/O1 can transmit 1-byte data by writing data to the transmit buffer register.

Thus, after setting the  $\overline{CS}$  signal to "L", write the transmission data to the transmit buffer register by each 1 byte, and return the  $\overline{CS}$  signal to "H" when 4-byte data has been transmitted.

Figure 2.4.28 shows a control procedure of Serial I/O1.

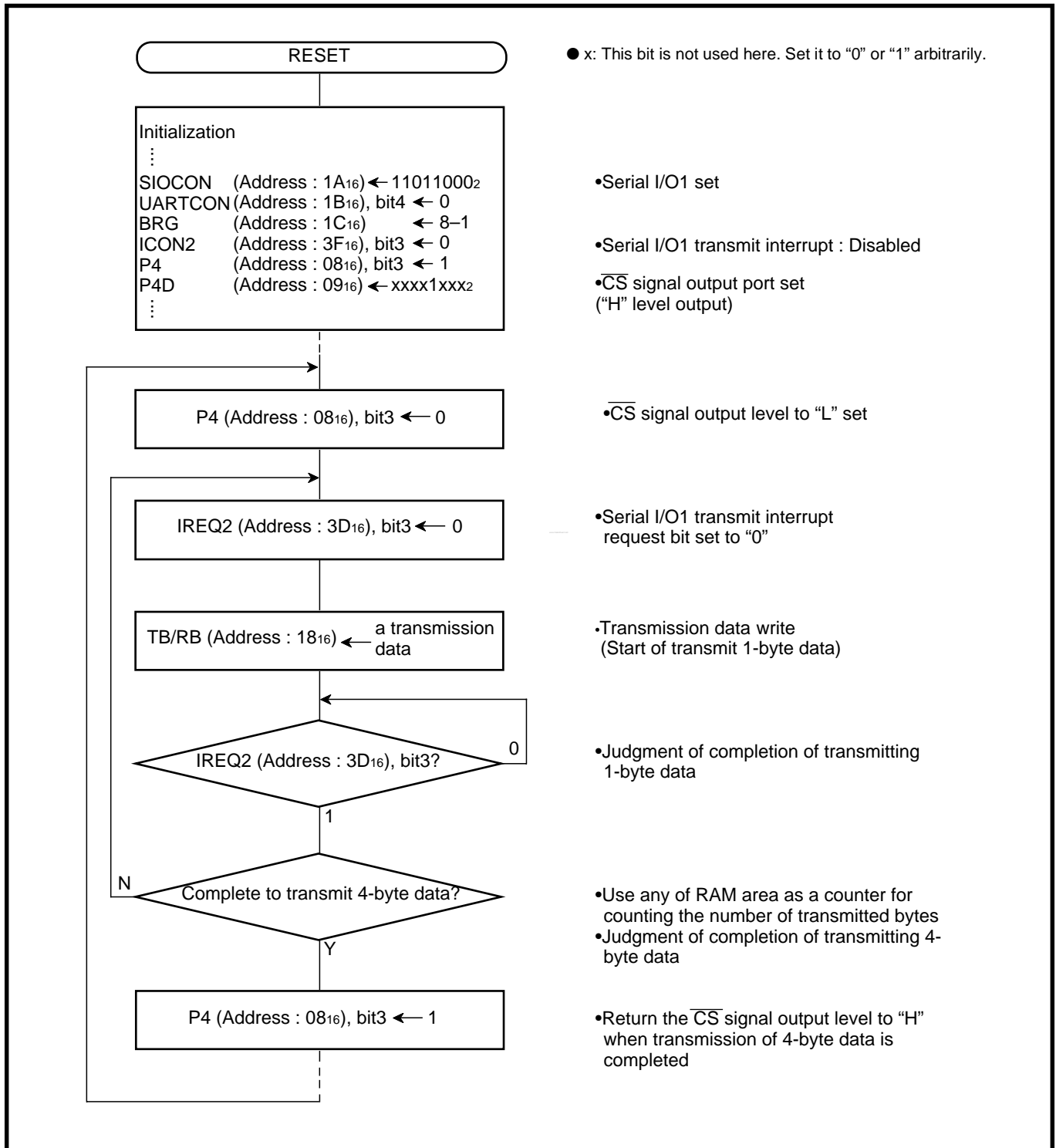
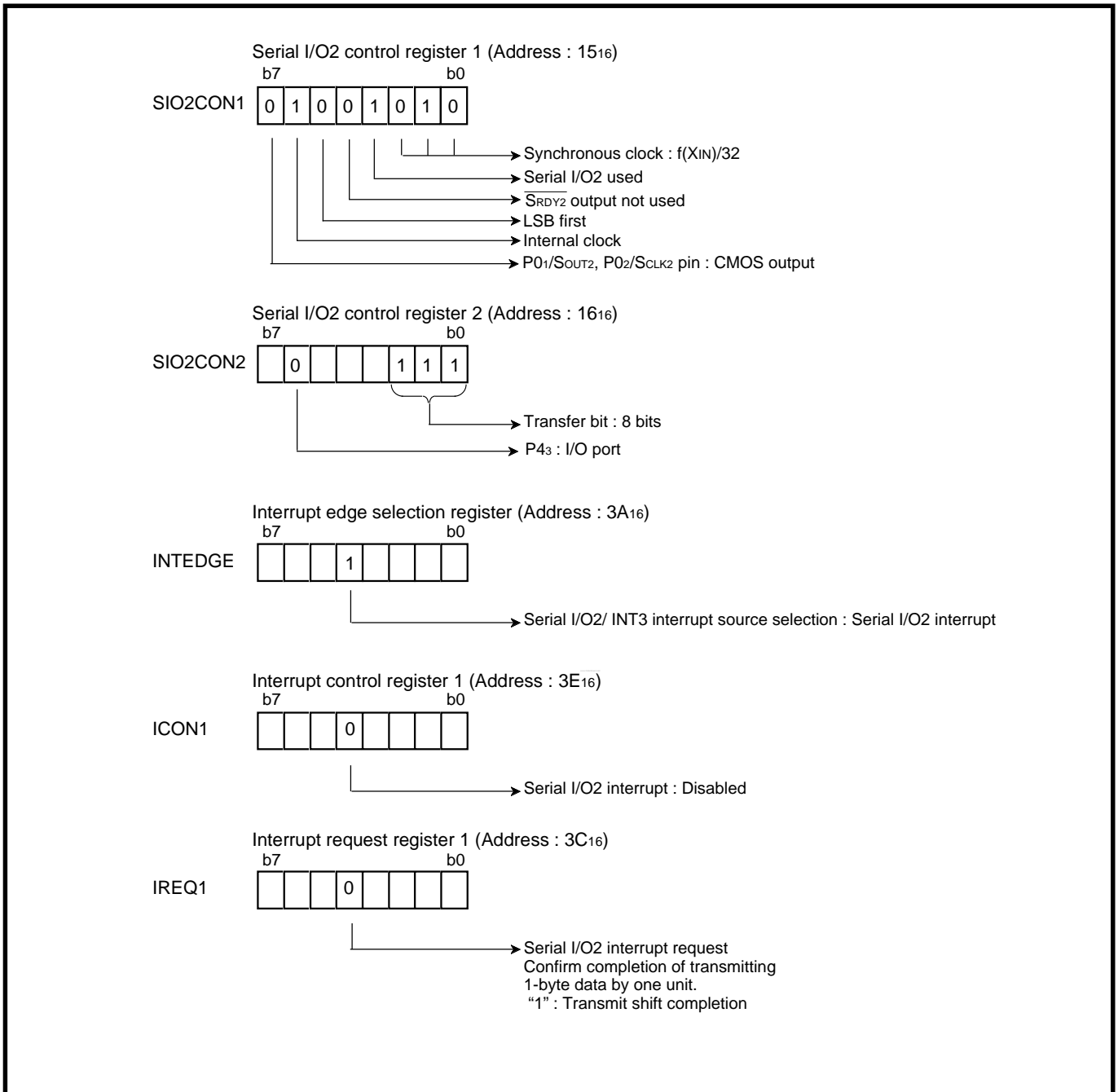


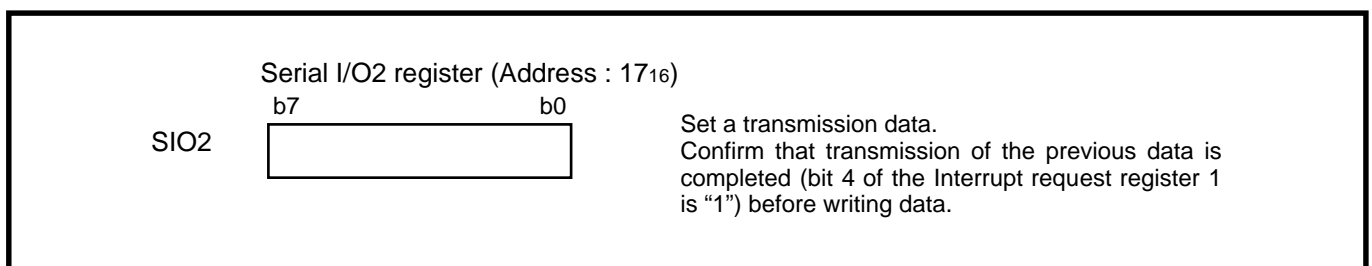
Fig. 2.4.28 Control procedure of Serial I/O1



Figure 2.4.29 shows registers setting relevant to Serial I/O2, and Figure 2.4.30 shows a setting of serial I/O2 transmission data.



**Fig. 2.4.29 Registers setting relevant to Serial I/O2**



**Fig. 2.4.30 Setting of serial I/O2 transmission data**

# APPLICATION

## 2.4 Serial I/O

### Example for using Serial I/O2

When the registers are set as shown in Fig. 2.4.29, the Serial I/O2 can transmit 1-byte data by writing data to the serial I/O2 register.

Thus, after setting the  $\overline{CS}$  signal to "L", write the transmission data to Serial I/O2 by each 1 byte, and return the  $\overline{CS}$  signal to "H" when 4-byte data has been transmitted.

Figure 2.4.31 shows a control procedure of Serial I/O2.

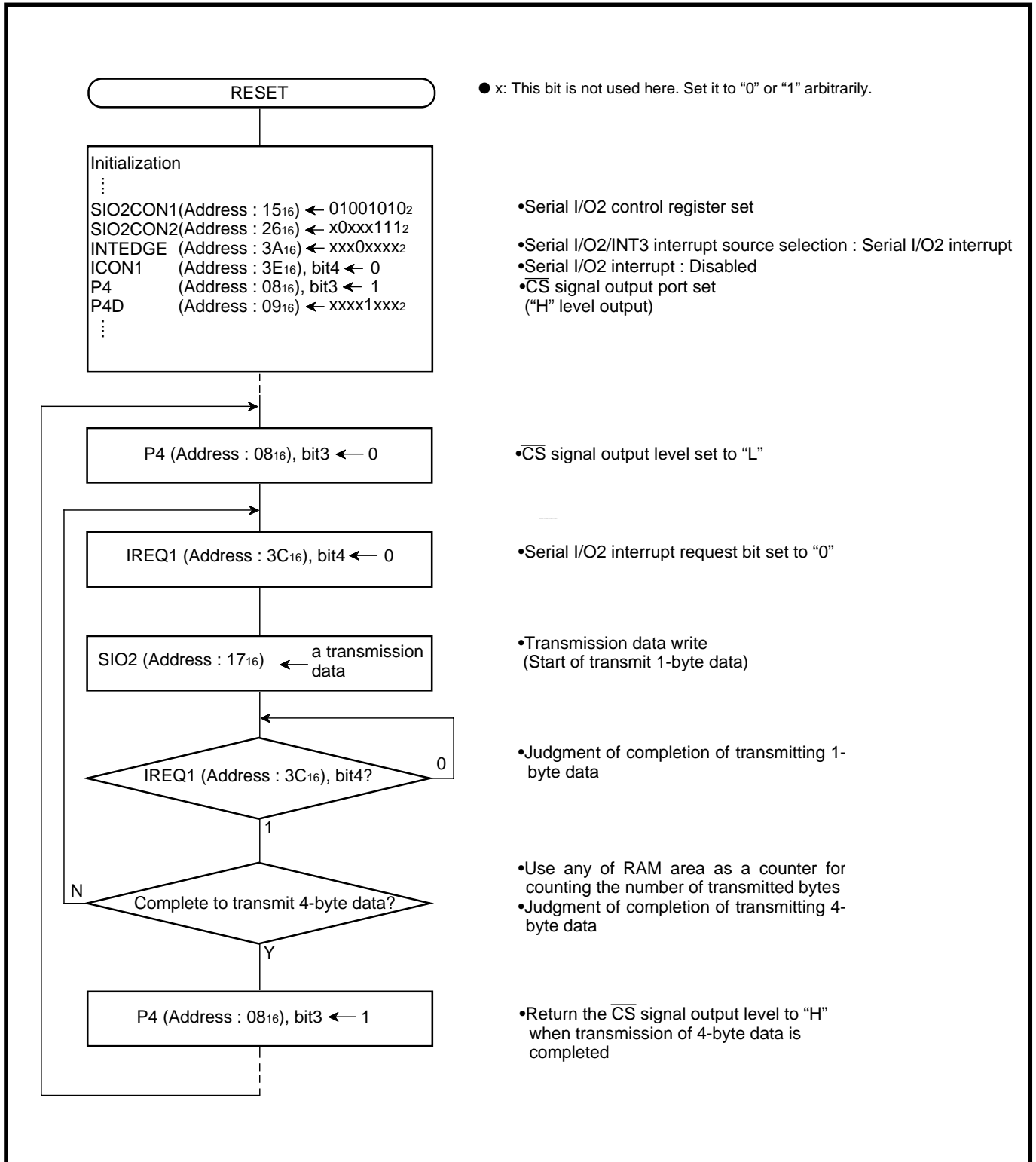


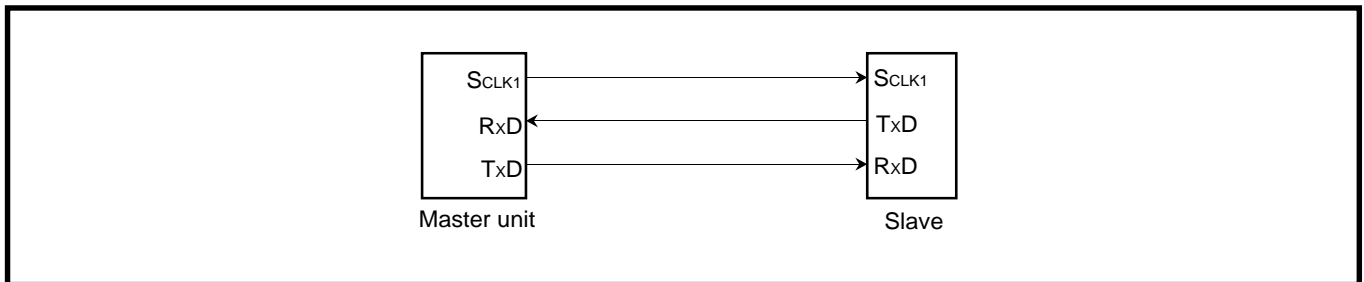
Fig. 2.4.31 Control procedure of Serial I/O2

**(3) Cyclic transmission or reception of block data (data of specified number of bytes) between two microcomputers**

**Outline :** When the clock synchronous serial I/O is used for communication, synchronization of the clock and the data between the transmitting and receiving sides may be lost because of noise included in the synchronous clock. It is necessary to correct that constantly, using "heading adjustment".

This "heading adjustment" is carried out by using the interval between blocks in this example.

Figure 2.4.32 shows a connection diagram.



**Fig. 2.4.32 Connection diagram**

**Specifications :**

- The serial I/O is used (clock synchronous serial I/O is selected).
- Synchronous clock frequency : 131 kHz ( $f(X_{IN}) = 4.19 \text{ MHz}$  is divided by 32)
- Byte cycle: 488  $\mu\text{s}$
- Number of bytes for transmission or reception : 8 byte/block
- Block transfer cycle : 16 ms
- Block transfer term : 3.5 ms
- Interval between blocks : 12.5 ms
- Heading adjustment time : 8 ms

**Limitations of specifications :**

- Reading of the reception data and setting of the next transmission data must be completed within the time obtained from "byte cycle – time for transferring 1-byte data" (in this example, the time taken from generating of the serial I/O1 receive interrupt request to input of the next synchronous clock is 431  $\mu\text{s}$ ).
- "Heading adjustment time < interval between blocks" must be satisfied.

# APPLICATION

## 2.4 Serial I/O

The communication is performed according to the timing shown in Figure 2.4.33. In the slave unit, when a synchronous clock is not input within a certain time (heading adjustment time), the next clock input is processed as the beginning (heading) of a block.

When a clock is input again after one block (8 byte) is received, the clock is ignored.

Figure 2.4.34 shows relevant registers setting.

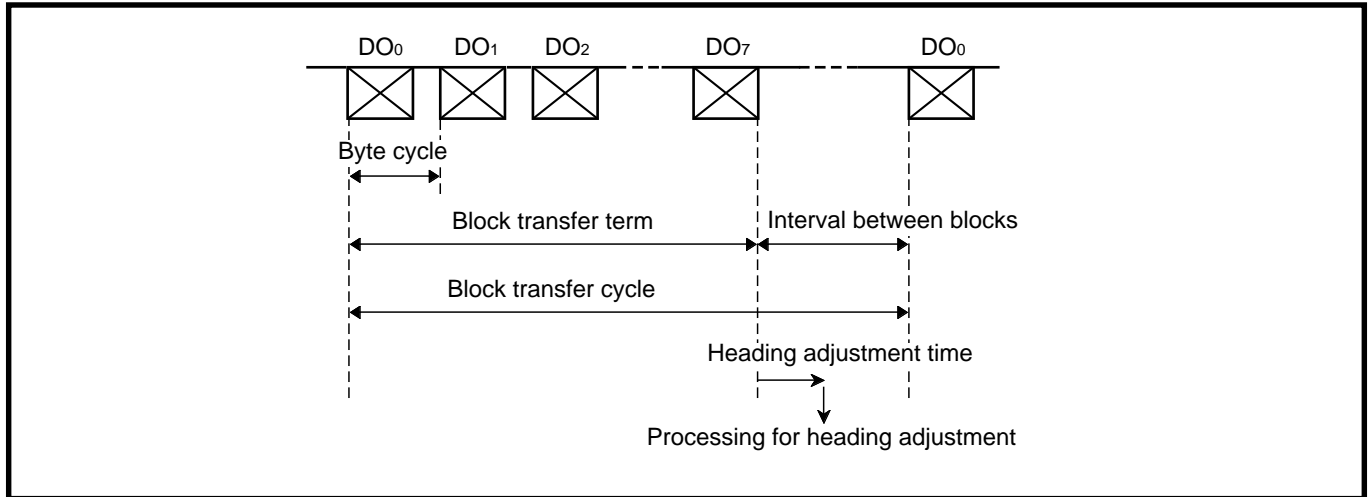


Fig. 2.4.33 Timing chart

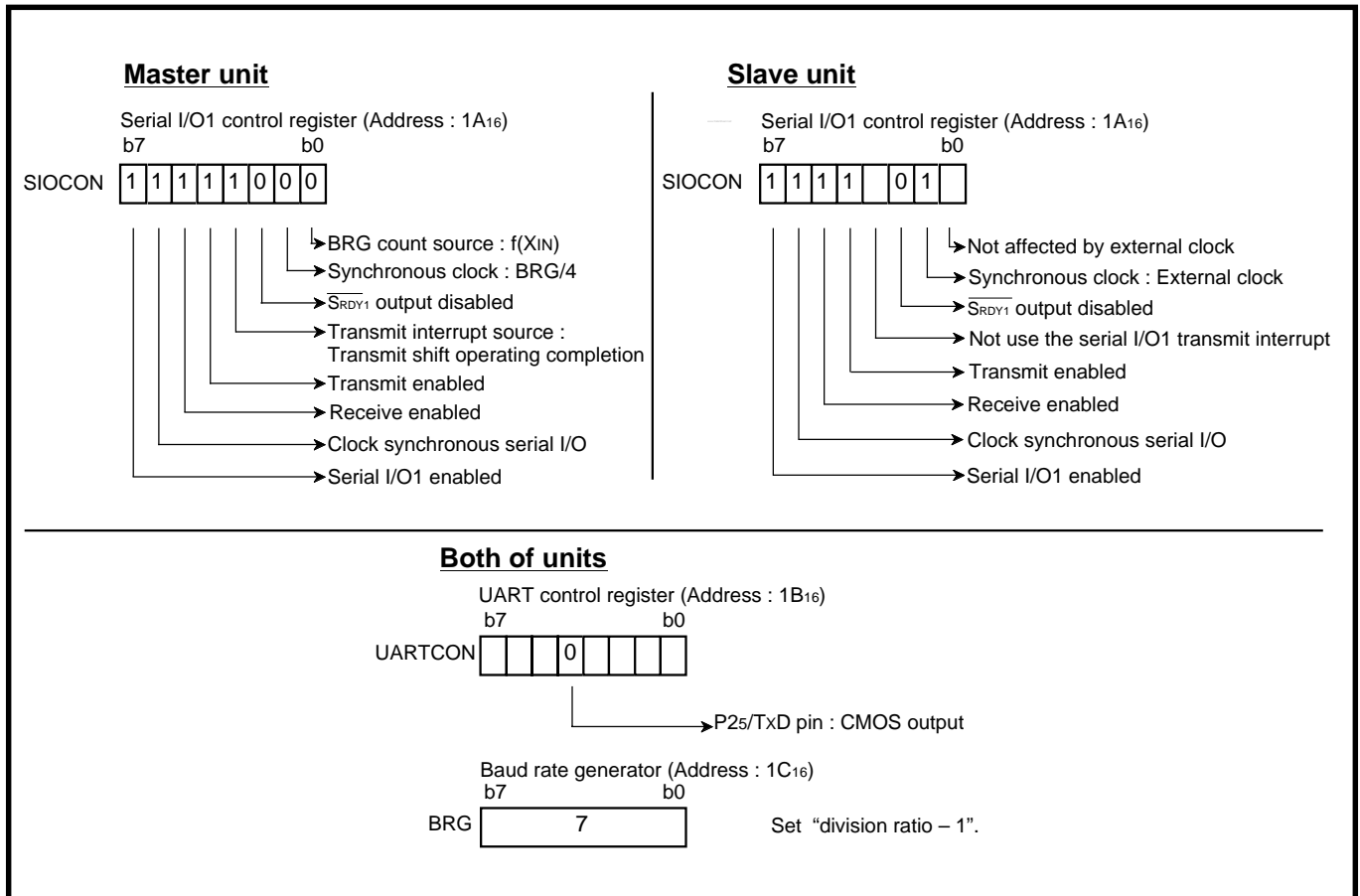


Fig. 2.4.34 Relevant registers setting

### Control procedure :

- Control in the master unit

After setting the relevant registers shown in Figure 2.4.34, the master unit starts transmission or reception of 1-byte data by writing transmission data to the transmit buffer register.

To perform the communication in the timing shown in Figure 2.4.33, take the timing into account and write transmission data. Additionally, read out the reception data when the serial I/O1 transmit interrupt request bit is set to "1", or before the next transmission data is written to the transmit buffer register.

Figure 2.4.35 shows a control procedure of the master unit using timer interrupts.

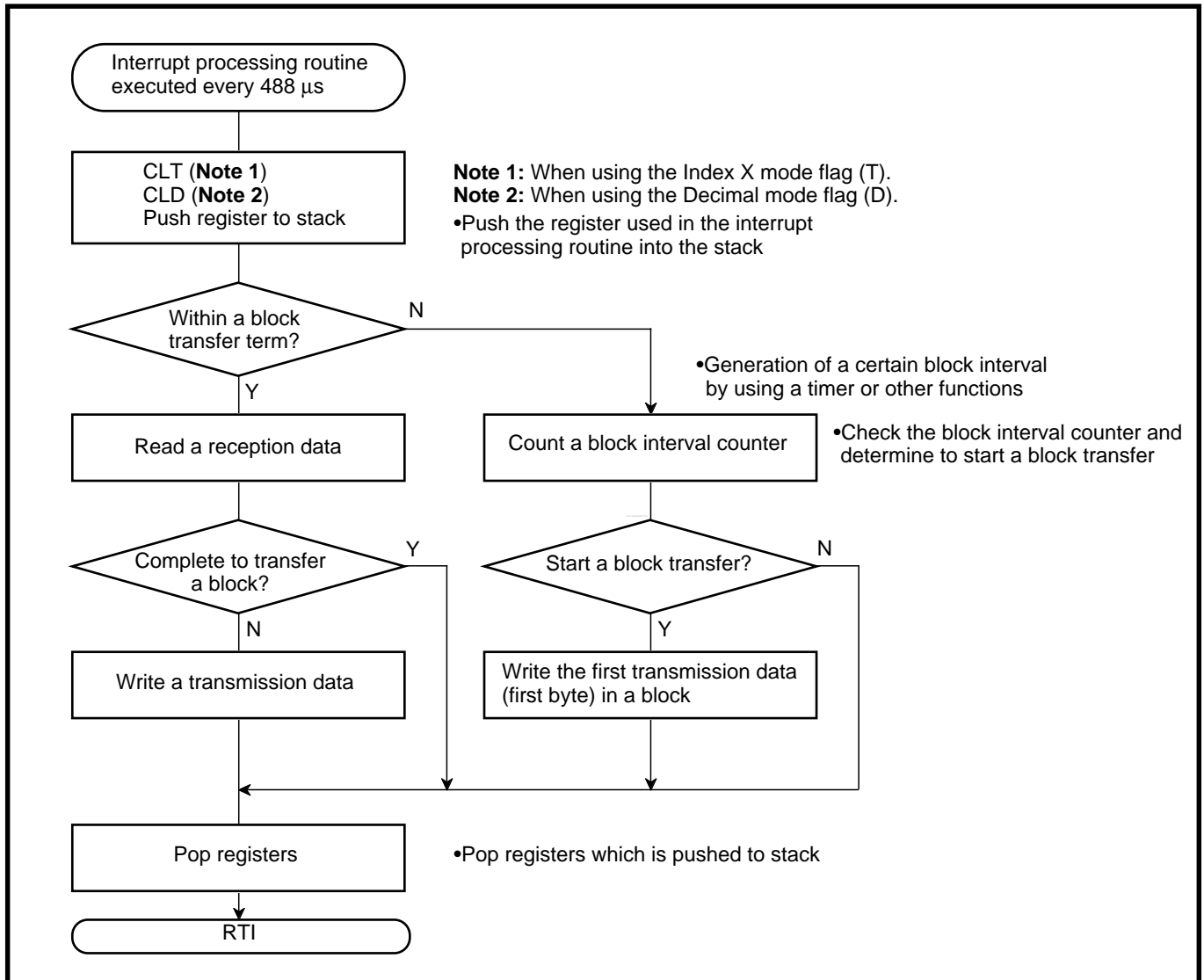


Fig. 2.4.35 Control procedure of master unit

# APPLICATION

## 2.4 Serial I/O

● Control in the slave unit

After setting the relevant registers as shown in Figure 2.4.34, the slave unit becomes the state where a synchronous clock can be received at any time, and the serial I/O receive interrupt request bit is set to “1” each time an 8-bit synchronous clock is received.

In the serial I/O receive interrupt processing routine, the data to be transmitted next is written to the transmit buffer register after the received data is read out.

However, if no serial I/O receive interrupt occurs for a certain time (heading adjustment time or more), the following processing will be performed.

1. The first 1-byte data of the transmission data in the block is written into the transmit buffer register.
  2. The data to be received next is processed as the first 1 byte of the received data in the block.
- Figure 2.4.36 shows a control procedure of the slave unit using the serial I/O receive interrupt and any timer interrupt (for heading adjustment).

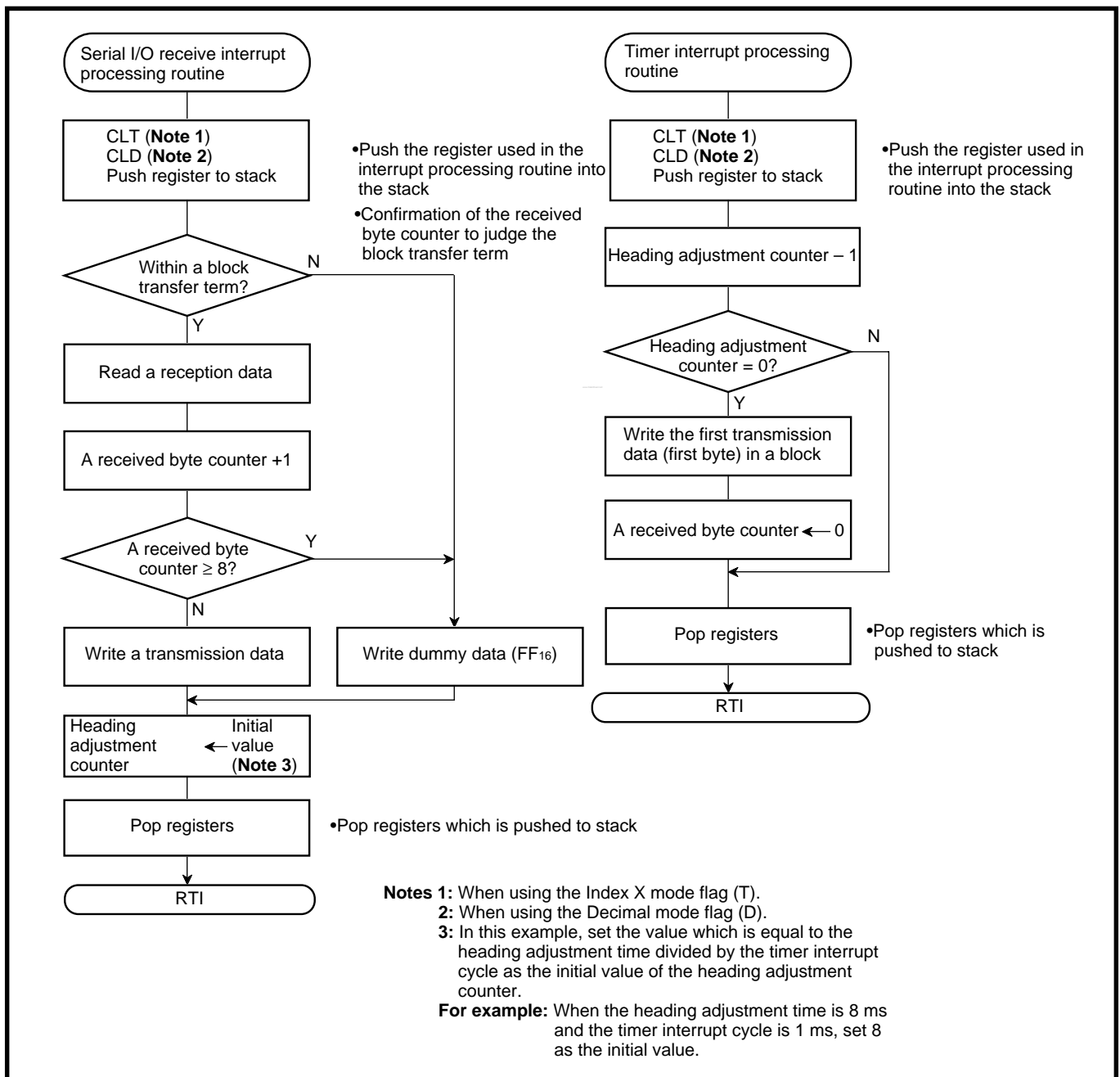
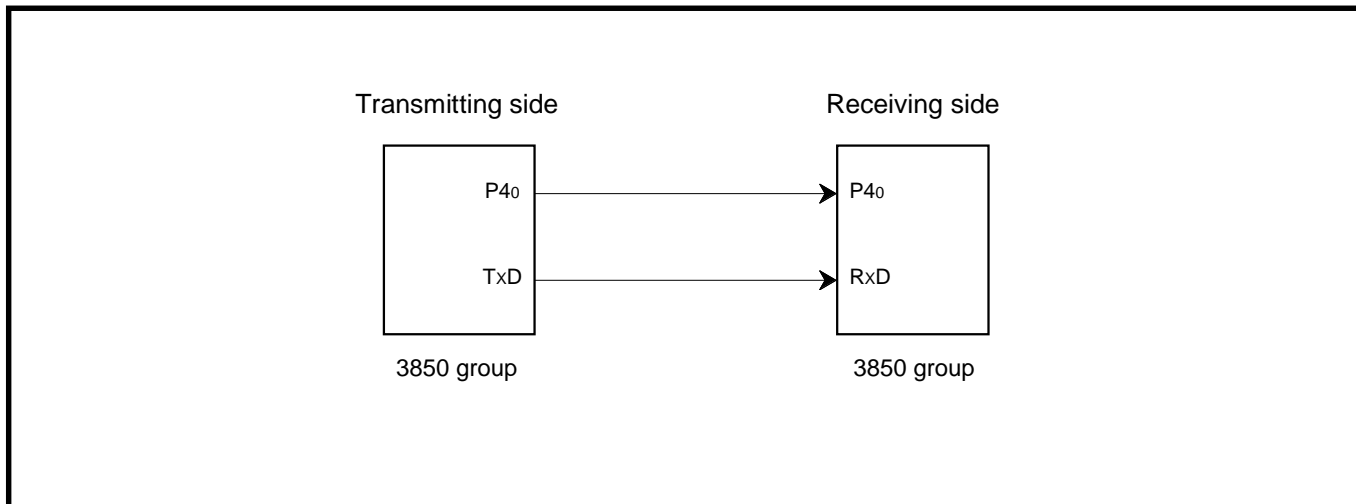


Fig. 2.4.36 Control procedure of slave unit

### (4) Communication (transmit/receive) using asynchronous serial I/O (UART)

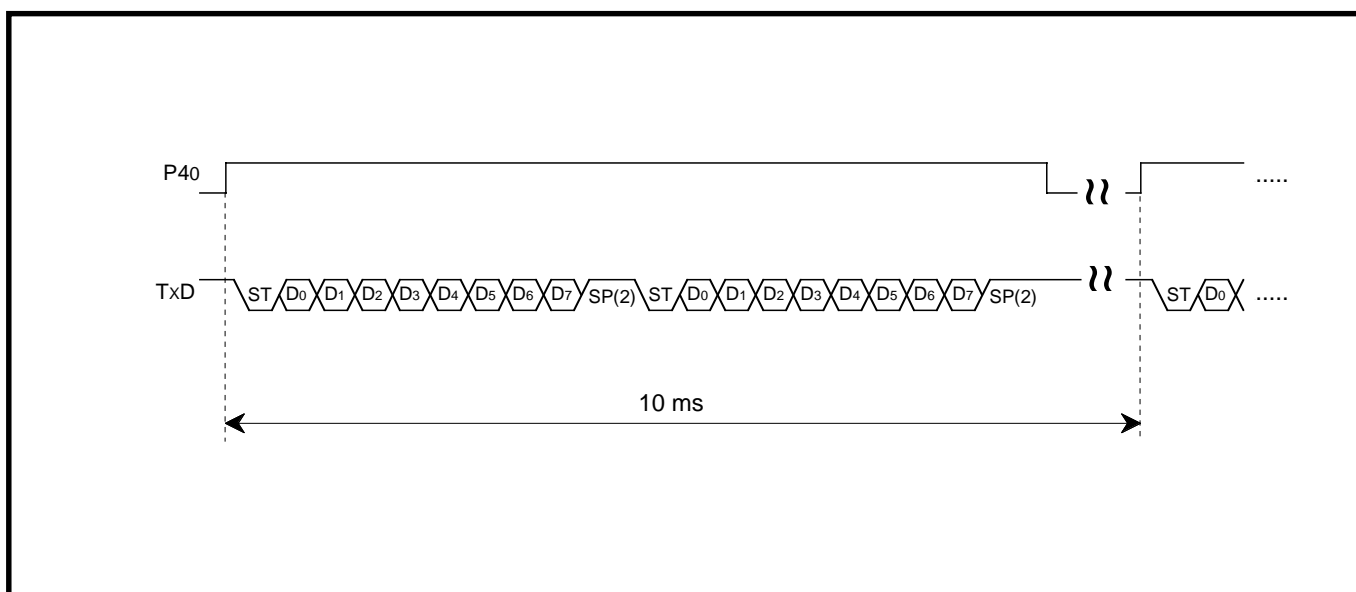
**Outline :** 2-byte data is transmitted and received, using the asynchronous serial I/O.  
Port P4<sub>0</sub> is used for communication control.

Figure 2.4.37 shows a connection diagram, and Figure 2.4.38 shows a timing chart.



**Fig. 2.4.37 Connection diagram**

- Specifications :**
- The Serial I/O1 is used (UART is selected).
  - Transfer bit rate : 9600 bps ( $f(X_{IN}) = 4.9152 \text{ MHz}$  is divided by 512)
  - Communication control using port P4<sub>0</sub>  
(The output level of port P4<sub>0</sub> is controlled by software.)
  - 2-byte data is transferred from the transmitting side to the receiving side at intervals of 10 ms generated by the timer.



**Fig. 2.4.38 Timing chart (using UART)**

# APPLICATION

## 2.4 Serial I/O

Table 2.4.1 and Table 2.4.2 show setting examples of the baud rate generator (BRG) values and transfer bit rate values; Figure 2.4.39 shows registers setting relevant to the transmitting side; Figure 2.4.40 shows registers setting relevant to the receiving side.

**Table 2.4.1 Setting examples of Baud rate generator values and transfer bit rate values (1)**

BRG count source (Note 1)	BRG setting value	Transfer bit rate (bps) (Note 2)	
		at f(XIN) = 4.9152 MHz	at f(XIN) = 8 MHz
f(XIN)/4	255(FF <sub>16</sub> )	300	488.28125
f(XIN)/4	127(7F <sub>16</sub> )	600	976.5625
f(XIN)/4	63(3F <sub>16</sub> )	1200	1953.125
f(XIN)/4	31(1F <sub>16</sub> )	2400	3906.25
f(XIN)/4	15(0F <sub>16</sub> )	4800	7812.5
f(XIN)/4	7(07 <sub>16</sub> )	9600	15625
f(XIN)/4	3(03 <sub>16</sub> )	19200	31250
f(XIN)/4	1(01 <sub>16</sub> )	38400	62500
f(XIN)	3(03 <sub>16</sub> )	76800	125000
f(XIN)	1(01 <sub>16</sub> )	153600	250000
f(XIN)	0(00 <sub>16</sub> )	307200	500000

**Table 2.4.2 Setting examples of Baud rate generator values and transfer bit rate values (2)**

BRG count source (Note 1)	BRG setting value	Transfer bit rate (bps) (Note 2)
		at f(XIN) = 7.9872 MHz
f(XIN)/4	207(CF <sub>16</sub> )	600
f(XIN)/4	103(67 <sub>16</sub> )	1200
f(XIN)/4	51(33 <sub>16</sub> )	2400
f(XIN)/4	25(19 <sub>16</sub> )	4800
f(XIN)/4	12(0C <sub>16</sub> )	9600
f(XIN)	25(19 <sub>16</sub> )	19200
f(XIN)	12(0C <sub>16</sub> )	38400

**Notes 1:** Select the BRG count source with bit 0 of the serial I/O1 control register (Address : 1A<sub>16</sub>).

**2:** Equation of transfer bit rate:

$$\text{Transfer bit rate (bps)} = \frac{f(\text{XIN})}{(\text{BRG setting value} + 1) \times 16 \times m^*}$$

\*m: When bit 0 of the serial I/O1 control register (Address : 1A<sub>16</sub>) is set to "0", a value of m is 1.

When bit 0 of the serial I/O1 control register (Address : 1A<sub>16</sub>) is set to "1", a value of m is 4.



### Transmitting side

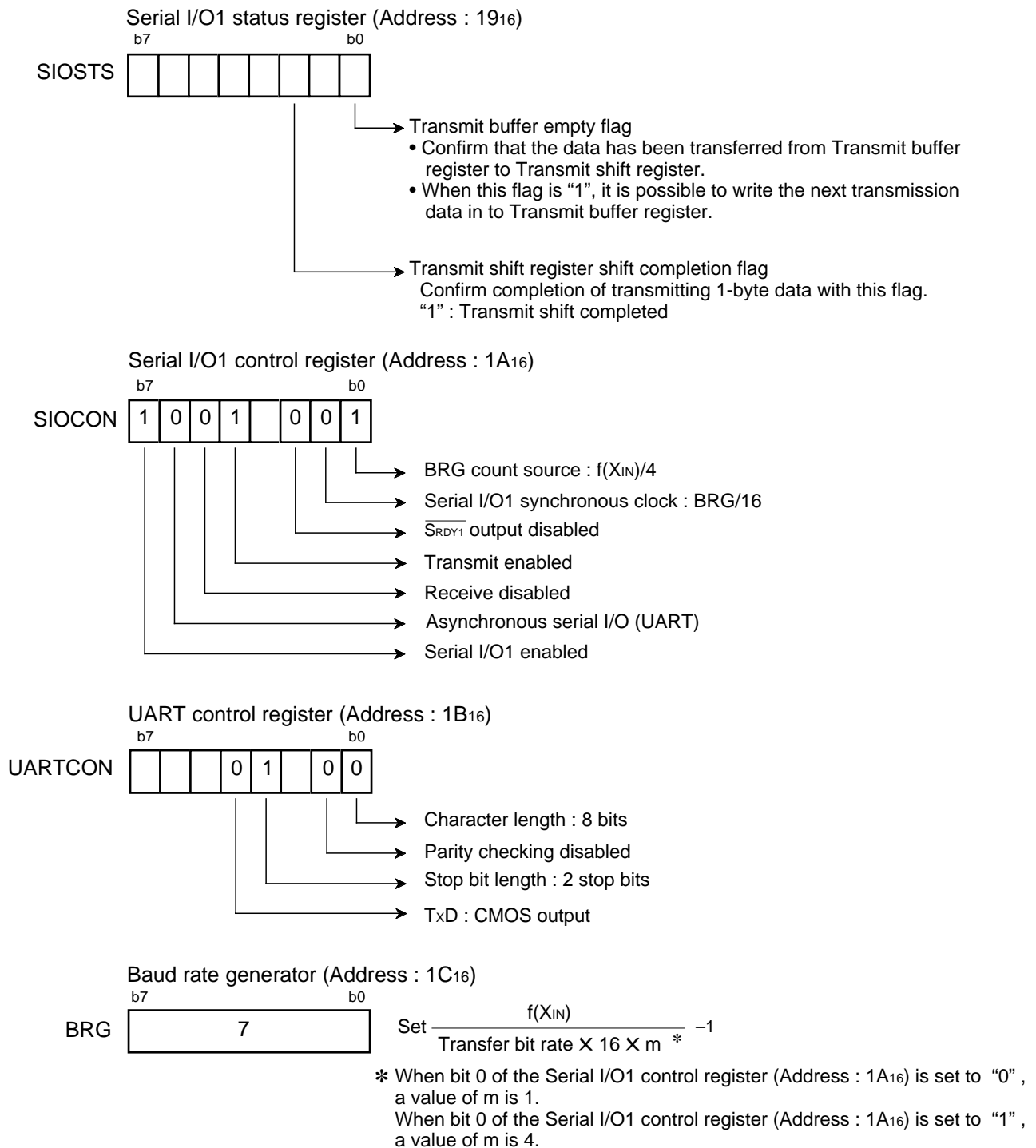


Fig. 2.4.39 Registers setting relevant to transmitting side

# APPLICATION

## 2.4 Serial I/O

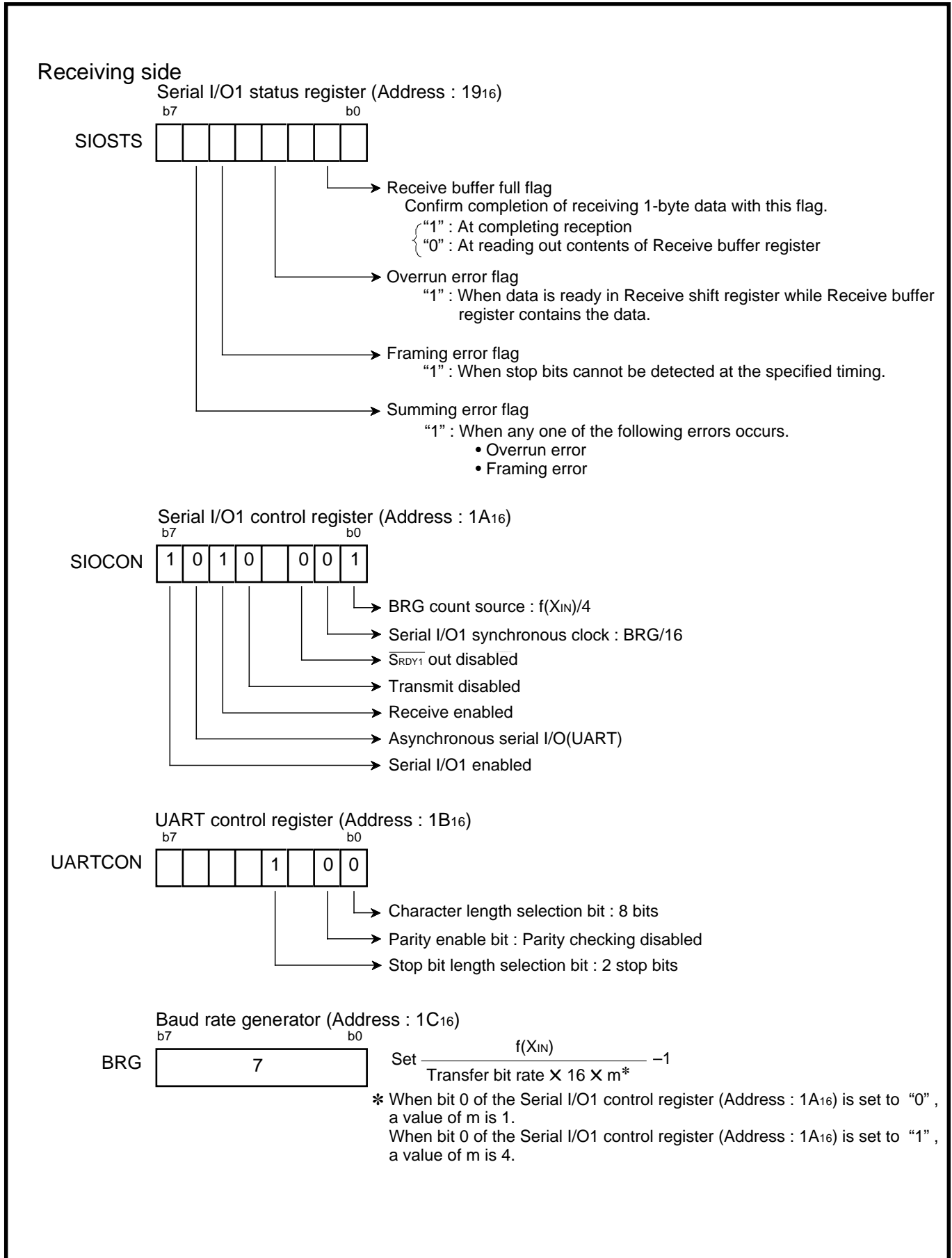


Fig. 2.4.40 Registers setting relevant to receiving side

Figure 2.4.41 shows a control procedure of the transmitting side, and Figure 2.4.42 shows a control procedure of the receiving side.

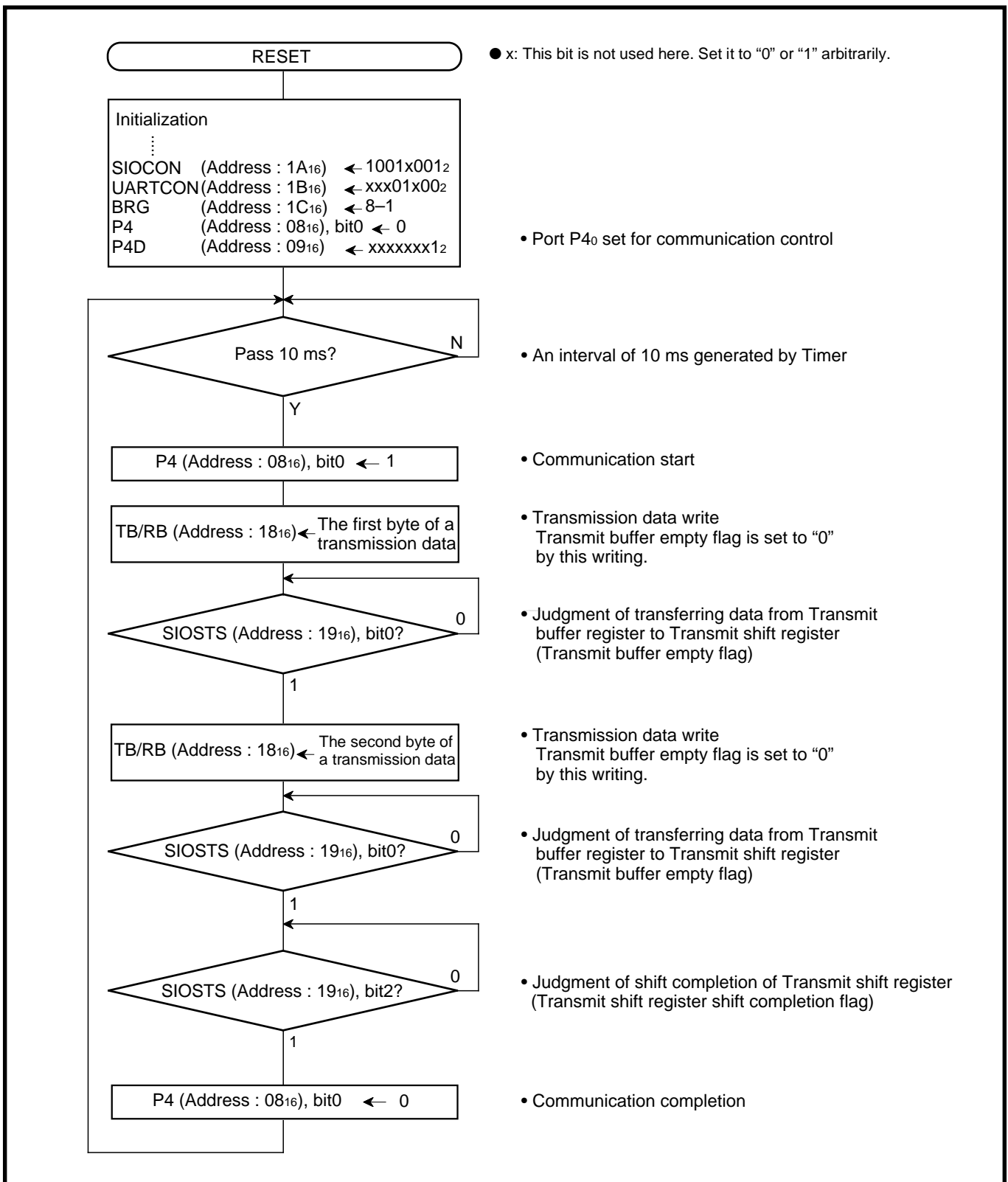


Fig. 2.4.41 Control procedure of transmitting side

# APPLICATION

## 2.4 Serial I/O

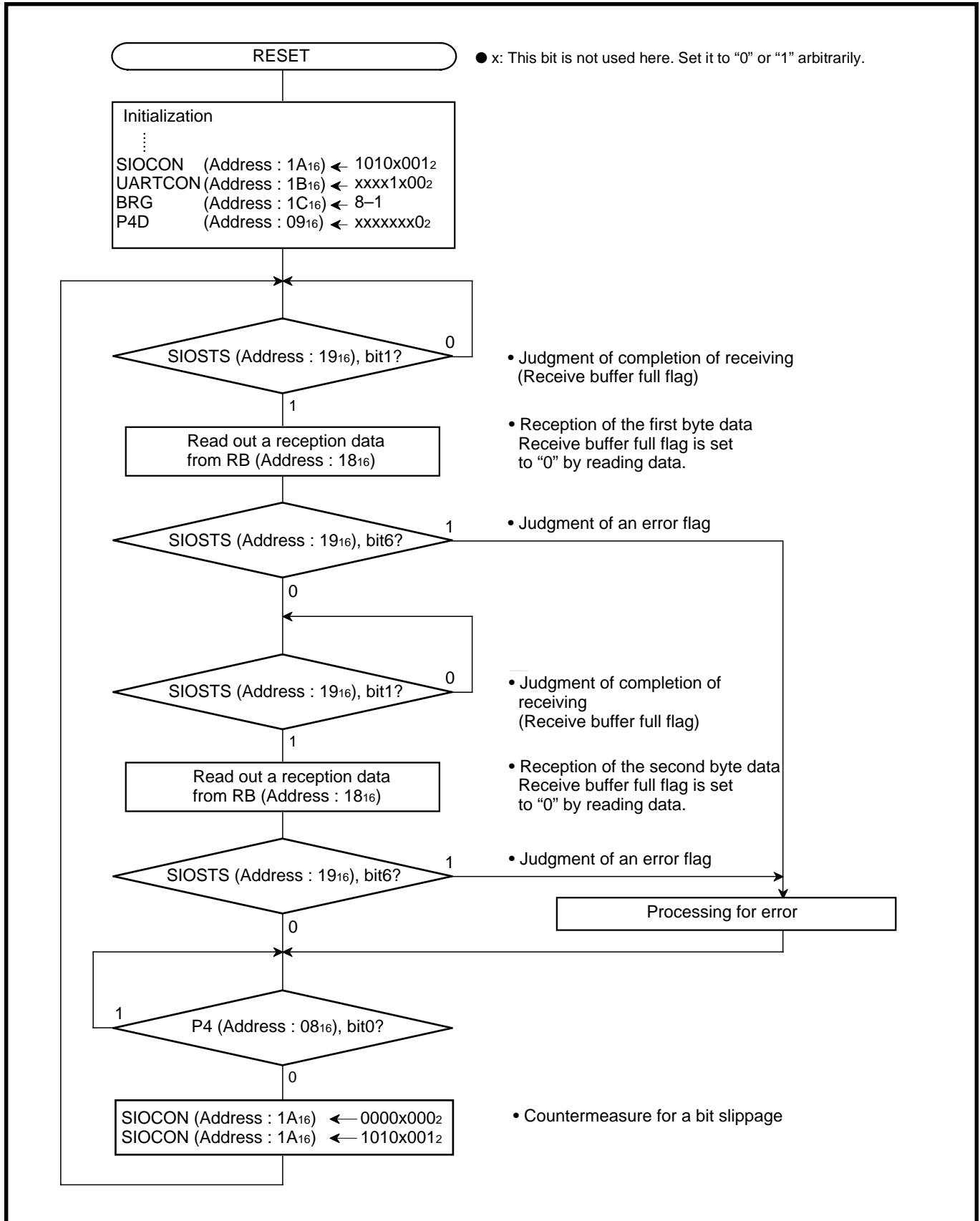


Fig. 2.4.42 Control procedure of receiving side

### 2.4.6 Notes on serial I/O

#### (1) Notes when selecting clock synchronous serial I/O (Serial I/O1)

##### ① Stop of transmission operation

Clear the serial I/O1 enable bit and the transmit enable bit to "0" (Serial I/O1 and transmit disabled).

##### ● Reason

Since transmission is not stopped and the transmission circuit is not initialized even if only the serial I/O1 enable bit is cleared to "0" (Serial I/O1 disabled), the internal transmission is running (in this case, since pins TxD, RxD, SCLK1, and  $\overline{\text{SRDY1}}$  function as I/O ports, the transmission data is not output). When data is written to the transmit buffer register in this state, data starts to be shifted to the transmit shift register. When the serial I/O1 enable bit is set to "1" at this time, the data during internally shifting is output to the TxD pin and an operation failure occurs.

##### ② Stop of receive operation

Clear the receive enable bit to "0" (receive disabled), or clear the serial I/O1 enable bit to "0" (Serial I/O1 disabled).

##### ③ Stop of transmit/receive operation

Clear the transmit enable bit and receive enable bit to "0" simultaneously (transmit and receive disabled).

(when data is transmitted and received in the clock synchronous serial I/O mode, any one of data transmission and reception cannot be stopped.)

##### ● Reason

In the clock synchronous serial I/O mode, the same clock is used for transmission and reception. If any one of transmission and reception is disabled, a bit error occurs because transmission and reception cannot be synchronized.

In this mode, the clock circuit of the transmission circuit also operates for data reception. Accordingly, the transmission circuit does not stop by clearing only the transmit enable bit to "0" (transmit disabled). Also, the transmission circuit is not initialized by clearing the serial I/O1 enable bit to "0" (Serial I/O1 disabled) (refer to (1) ①).

# APPLICATION

## 2.4 Serial I/O

### (2) Notes when selecting clock asynchronous serial I/O (Serial I/O1)

① **Stop of transmission operation**

Clear the transmit enable bit to "0" (transmit disabled).

● **Reason**

Since transmission is not stopped and the transmission circuit is not initialized even if only the serial I/O1 enable bit is cleared to "0" (Serial I/O1 disabled), the internal transmission is running (in this case, since pins TxD, RxD, SCLK1, and SRDY1 function as I/O ports, the transmission data is not output). When data is written to the transmit buffer register in this state, data starts to be shifted to the transmit shift register. When the serial I/O1 enable bit is set to "1" at this time, the data during internally shifting is output to the TxD pin and an operation failure occurs.

② **Stop of receive operation**

Clear the receive enable bit to "0" (receive disabled).

③ **Stop of transmit/receive operation**

**Only transmission operation is stopped.**

Clear the transmit enable bit to "0" (transmit disabled).

● **Reason**

Since transmission is not stopped and the transmission circuit is not initialized even if only the serial I/O1 enable bit is cleared to "0" (Serial I/O1 disabled), the internal transmission is running (in this case, since pins TxD, RxD, SCLK1, and SRDY1 function as I/O ports, the transmission data is not output). When data is written to the transmit buffer register in this state, data starts to be shifted to the transmit shift register. When the serial I/O1 enable bit is set to "1" at this time, the data during internally shifting is output to the TxD pin and an operation failure occurs.

**Only receive operation is stopped.**

Clear the receive enable bit to "0" (receive disabled).

### (3) SRDY1 output of reception side

When signals are output from the SRDY1 pin on the reception side by using an external clock in the clock synchronous serial I/O mode, set all of the receive enable bit, the SRDY1 output enable bit, and the transmit enable bit to "1" (transmit enabled).

### (4) Setting serial I/O1 control register again (Serial I/O1)

Set the serial I/O1 control register again after the transmission and the reception circuits are reset by clearing both the transmit enable bit and the receive enable bit to "0".

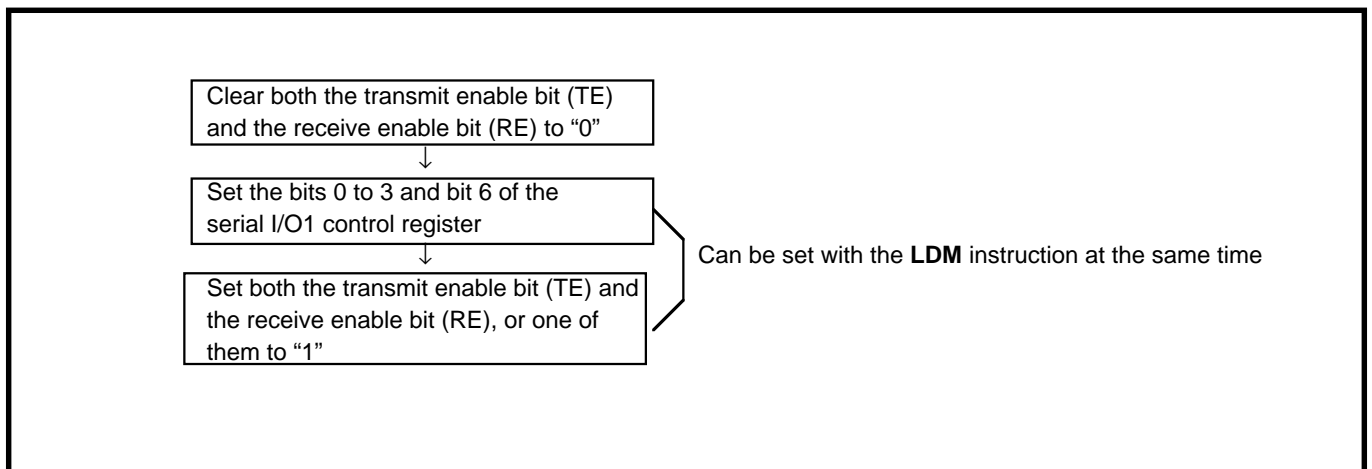


Fig. 2.4.43 Sequence of setting serial I/O1 control register again

**(5) Data transmission control with referring to transmit shift register completion flag (Serial I/O1)**  
The transmit shift register completion flag changes from “1” to “0” with a delay of 0.5 to 1.5 shift clocks. When data transmission is controlled with referring to the flag after writing the data to the transmit buffer register, note the delay.

**(6) Transmission control when external clock is selected (Serial I/O1)**  
When an external clock is used as the synchronous clock for data transmission, set the transmit enable bit to “1” at “H” of the SCLK1 input level. Also, write the transmit data to the transmit buffer register at “H” of the SCLK1 input level.

**(7) Transmit interrupt request when transmit enable bit is set (Serial I/O1)**  
When the transmit interrupt is used, set the transmit interrupt enable bit to transmit enabled as shown in the following sequence.

- ① Set the interrupt enable bit to “0” (disabled) with CLB instruction.
- ② Prepare serial I/O for transmission/reception.
- ③ Set the interrupt request bit to “0” with CLB instruction after 1 or more instruction has been executed.
- ④ Set the interrupt enable bit to “1” (enabled).

● **Reason**

When the transmission enable bit is set to “1”, the transmit buffer empty flag and transmit shift register completion flag are set to “1”. The interrupt request is generated and the transmission interrupt bit is set regardless of which of the two timings listed below is selected as the timing for the transmission interrupt to be generated.

- Transmit buffer empty flag is set to “1”
- Transmit shift register completion flag is set to “1”

**(8) Transmit data writing (Serial I/O2)**  
In the clock synchronous serial I/O, when selecting an external clock as synchronous clock, write the transmit data to the serial I/O2 register (serial I/O shift register) at “H” of the transfer clock input level.

# APPLICATION

## 2.5 PWM

### 2.5 PWM

This paragraph explains the registers setting method and the notes relevant to the PWM.

#### 2.5.1 Memory map

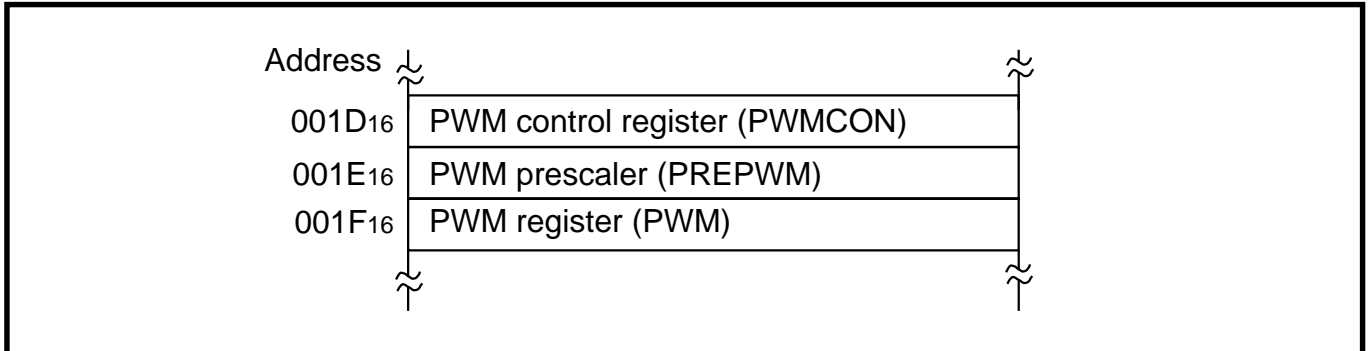


Fig. 2.5.1 Memory map of registers relevant to PWM

#### 2.5.2 Relevant registers

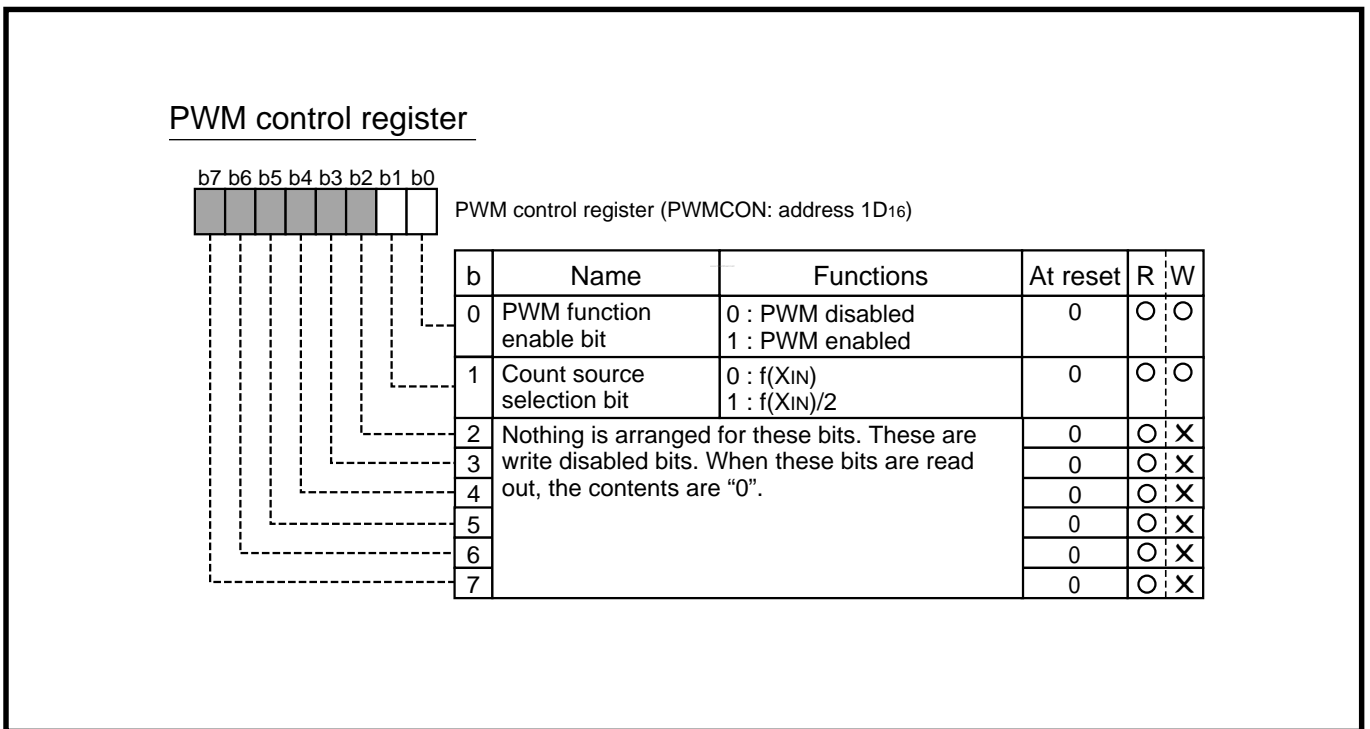


Fig. 2.5.2 Structure of PWM control register



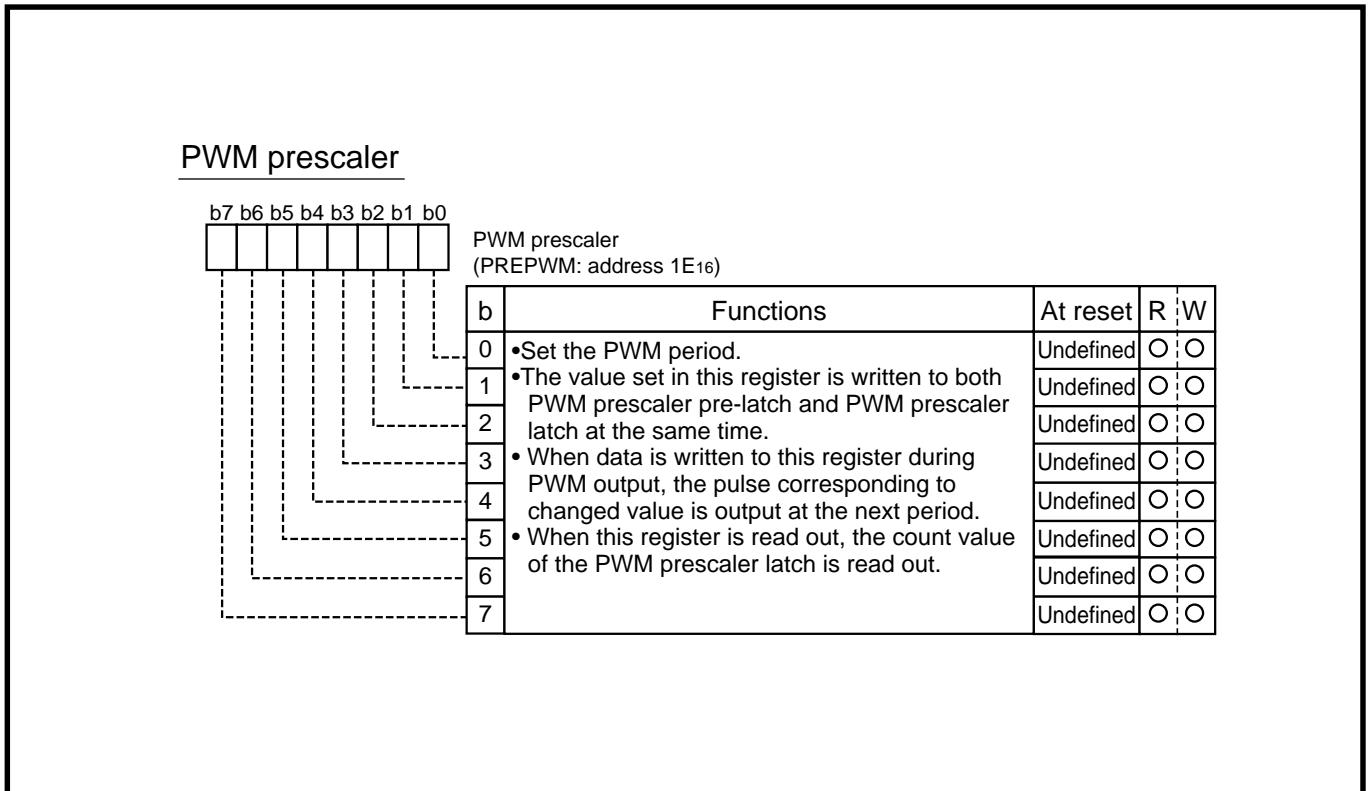


Fig. 2.5.3 Structure of PWM prescaler

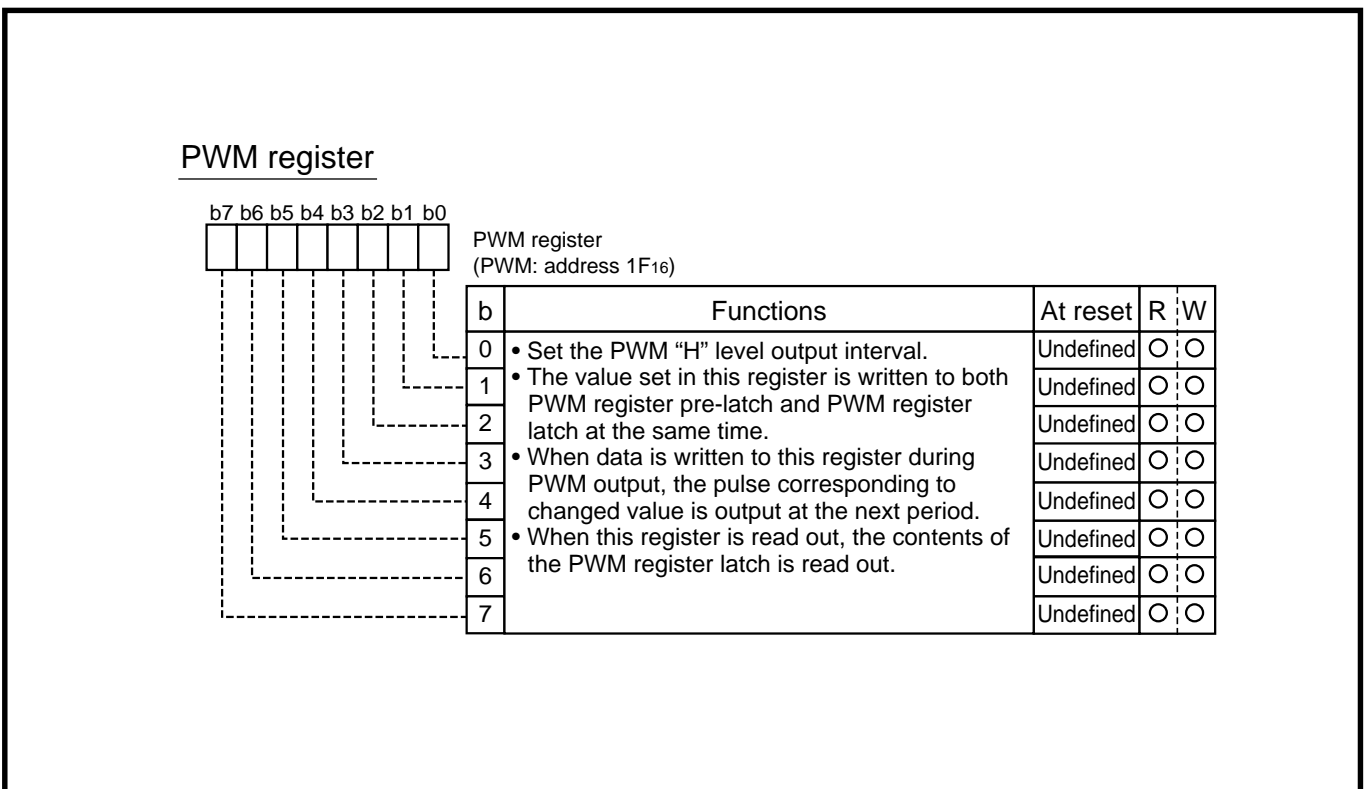


Fig. 2.5.4 Structure of PWM register

# APPLICATION

## 2.5 PWM

### 2.5.3 PWM output circuit application example

#### <Motor control>

**Outline :** The rotation speed of the motor is controlled by using PWM (pulse width modulation) output.

Figure 2.5.5 shows a connection diagram ; Figures 2.5.6 shows PWM output timing, and Figure 2.5.7 shows a setting of the related registers.

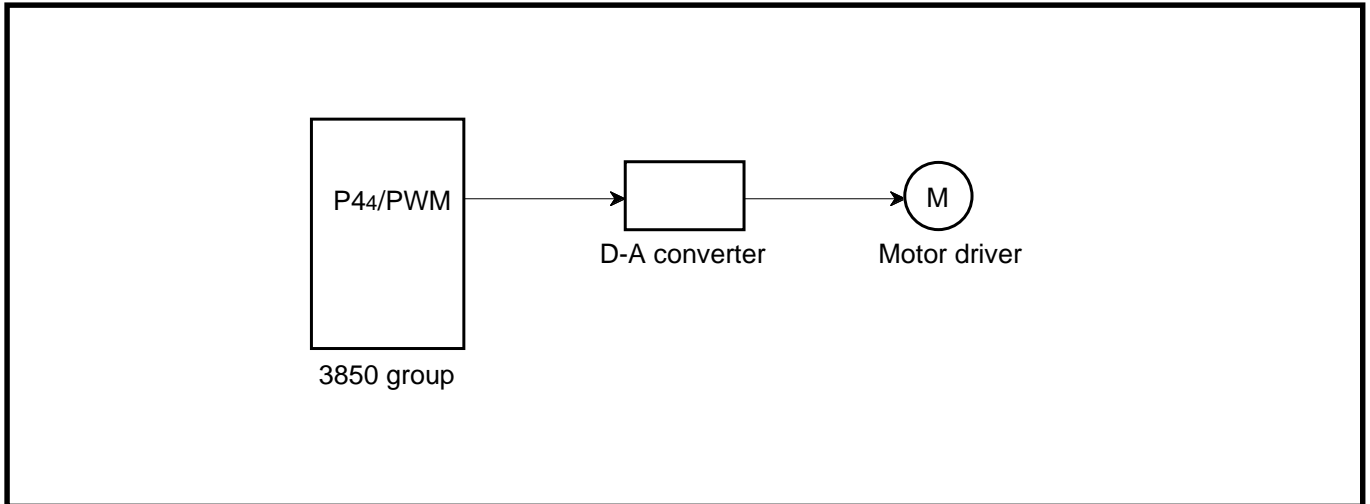


Fig. 2.5.5 Connection diagram

- Specifications :**
- Motor is controlled by using the PWM output function of 8-bit resolution.
  - Clock  $f(X_{IN}) = 5.0 \text{ MHz}$
  - "T", PWM cycle :  $102 \mu\text{s}$
  - "t", "H" level width of output pulse :  $40 \mu\text{s}$  (Fixed speed)
- \* A motor speed can be changed by modifying the "H" level width of output pulse.

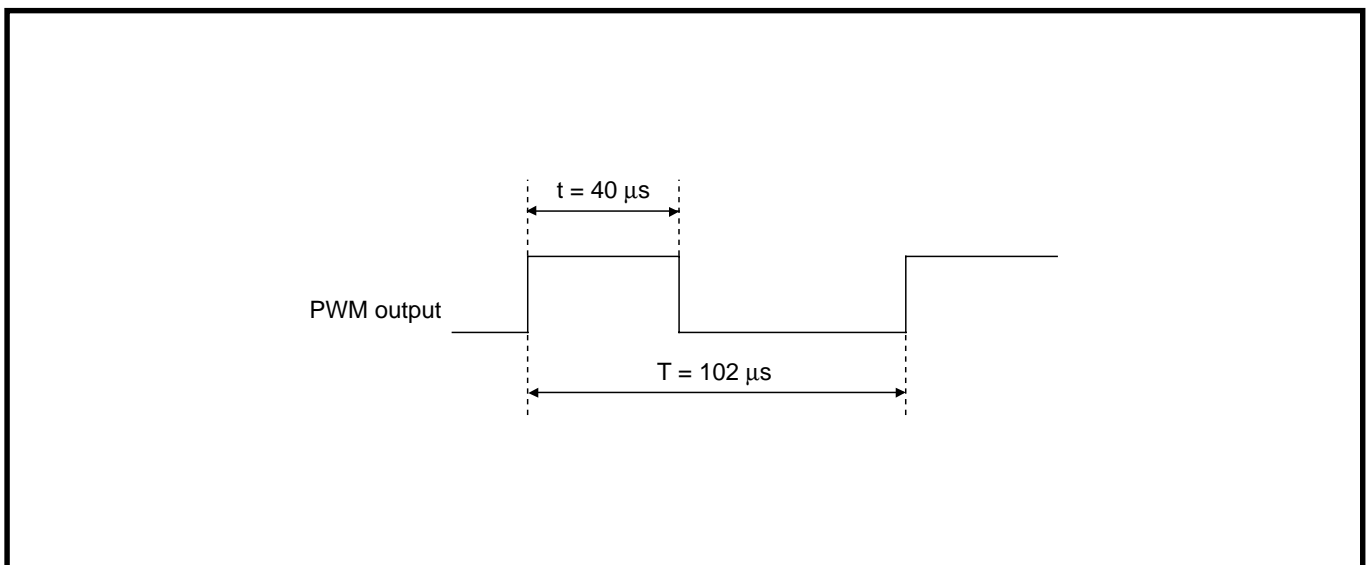
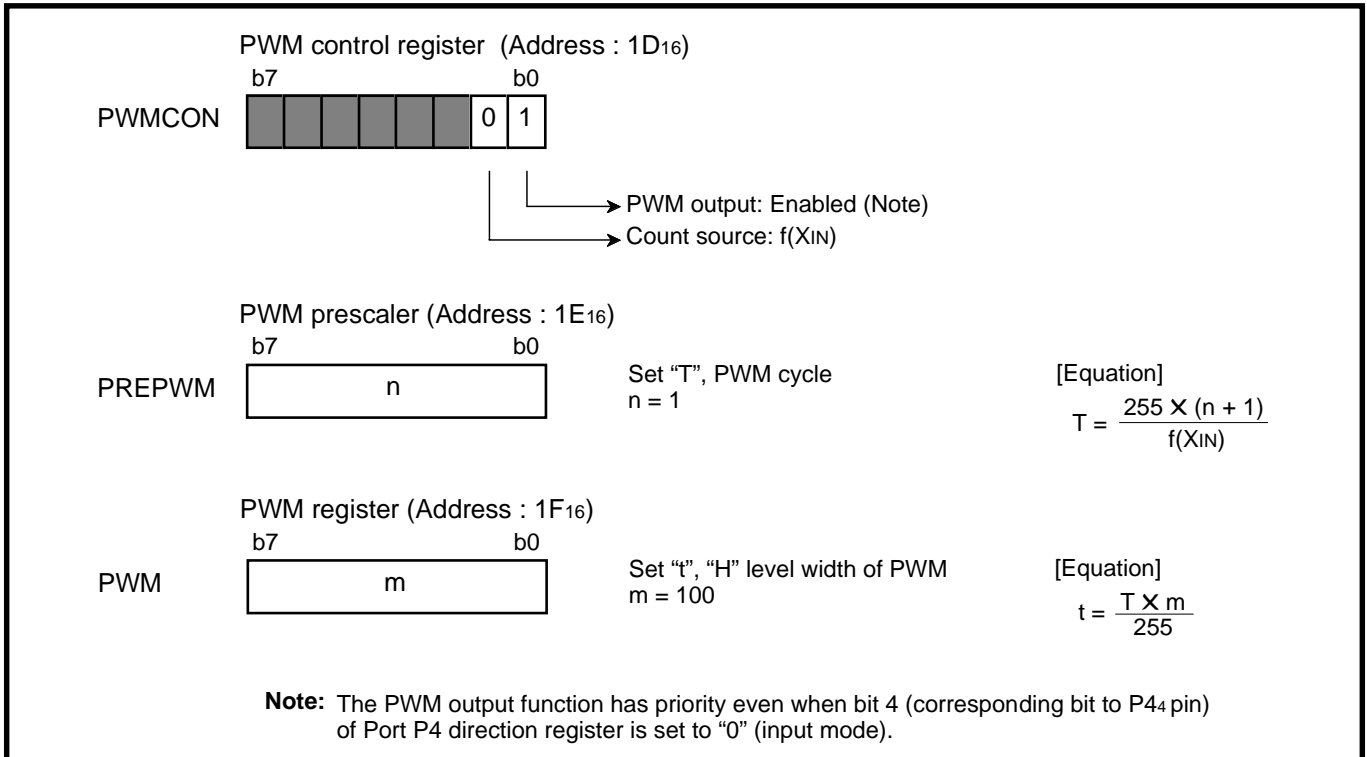


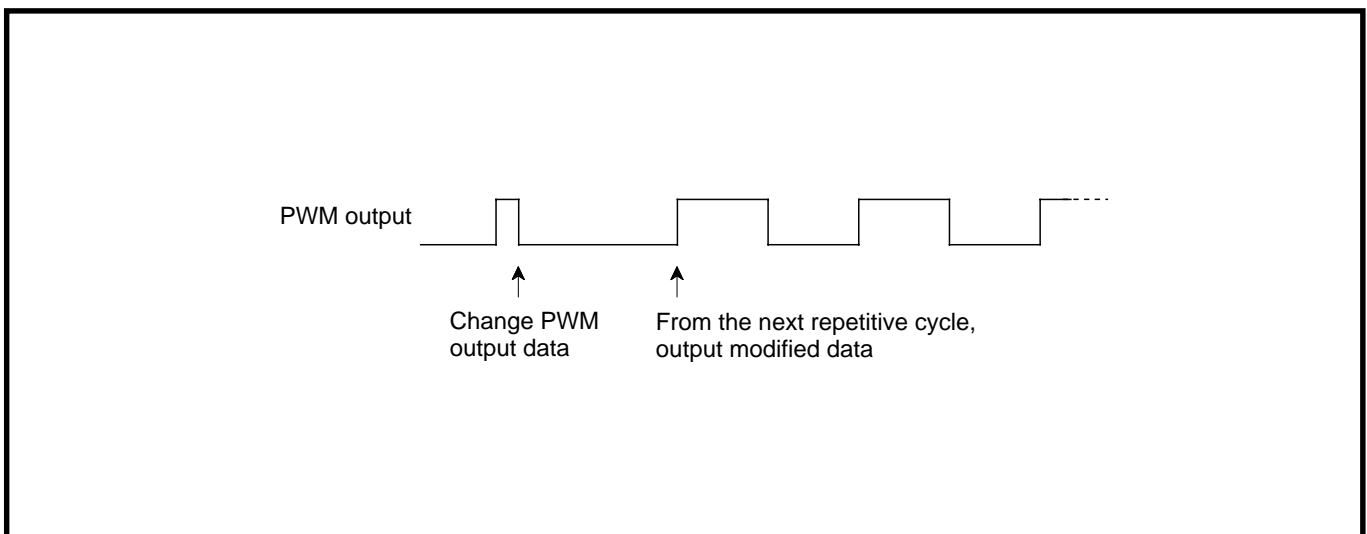
Fig. 2.5.6 PWM output timing



**Fig. 2.5.7 Setting of relevant registers**

**<About PWM output>**

1. Set the PWM function enable bit to "1" : The P4<sub>4</sub>/PWM pin is used as the PWM pin. The pulse beginning with "H" level pulse is output.
2. Set the PWM function enable bit to "0" : The P4<sub>4</sub>/PWM pin is used as the port P4<sub>4</sub>. Thus, when fixing the output level, take the following procedure:
  - (1) Write an output value to bit 6 of the port P4 register.
  - (2) Write "00010000<sub>2</sub>" to the port P4 direction register.
3. After data is set to the PWM prescaler and the PWM register, the PWM waveforms corresponding to updated data will be output from the next repetitive cycle.



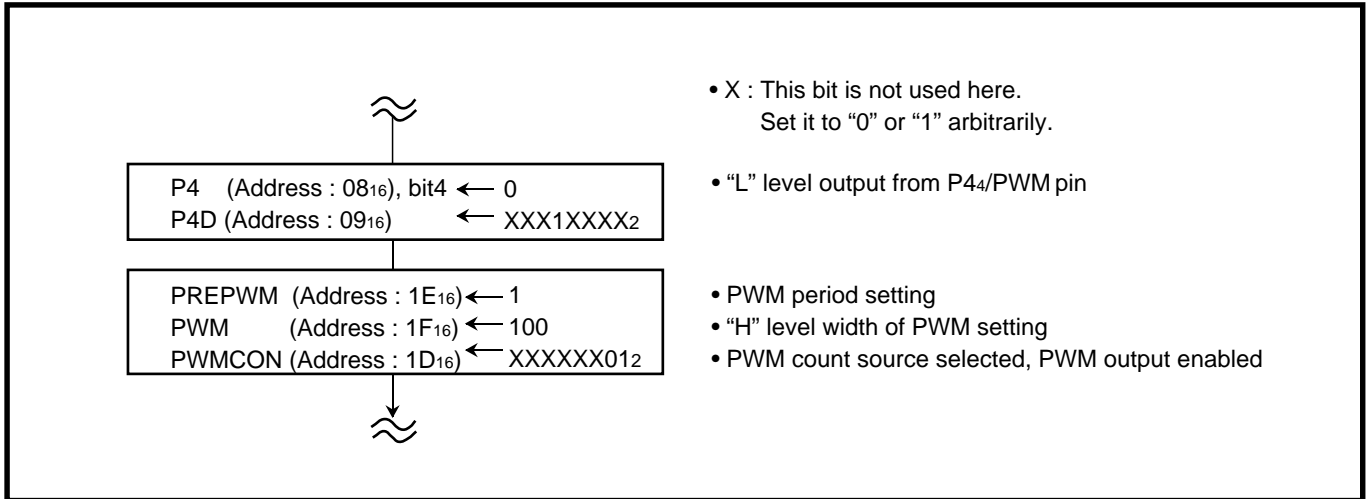
**Fig. 2.5.8 PWM output**

# APPLICATION

## 2.5 PWM

**Control procedure :** By setting the related registers as shown by Figure 2.5.7, PWM waveforms are output to the externals. This PWM output is integrated through the low pass filter, and that converted into DC signals is used for control of the motor.

Figure 2.5.9 shows control procedure.



**Fig. 2.5.9 Control procedure**

### 2.5.4 Notes on PWM

The PWM starts after the PWM enable bit is set to enable and "L" level is output from the PWM pin. The length of this "L" level output is as follows:

$$\frac{n + 1}{2 \cdot f(X_{IN})} \text{ sec. (Count source selection bit = 0, where n is the value set in the prescaler)}$$

$$\frac{n + 1}{f(X_{IN})} \text{ sec. (Count source selection bit = 1, where n is the value set in the prescaler)}$$

### 2.6 A-D converter

This paragraph explains the registers setting method and the notes relevant to the A-D converter.

#### 2.6.1 Memory map

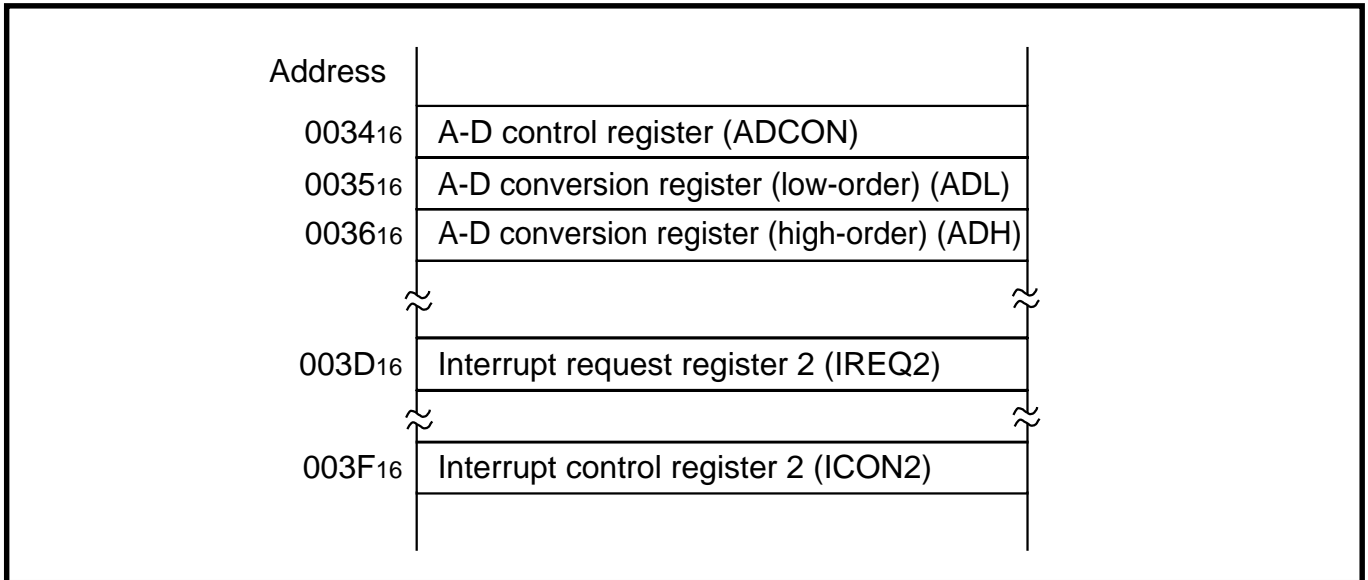


Fig. 2.6.1 Memory map of registers relevant to A-D converter

#### 2.6.2 Relevant registers

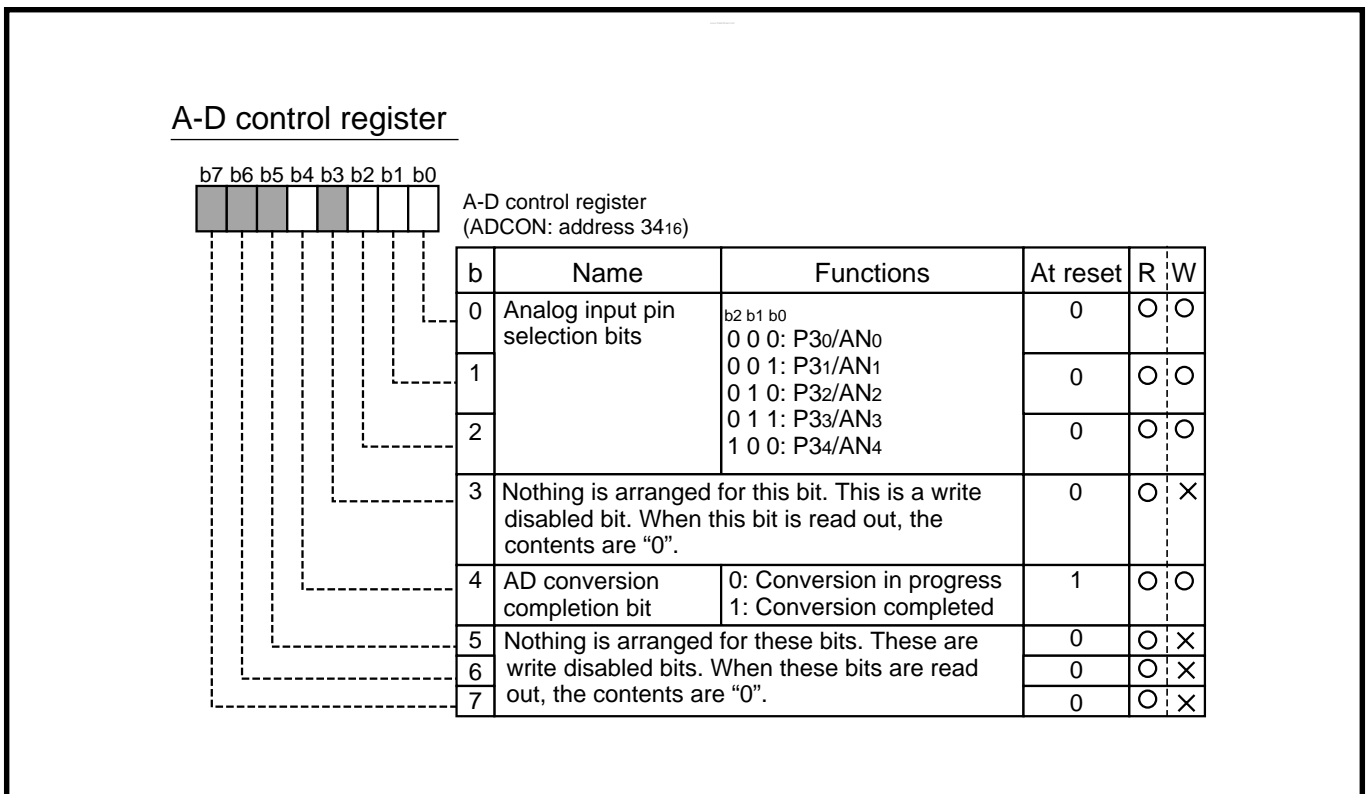


Fig. 2.6.2 Structure of A-D control register

# APPLICATION

## 2.6 A-D converter

### A-D conversion register (high-order)

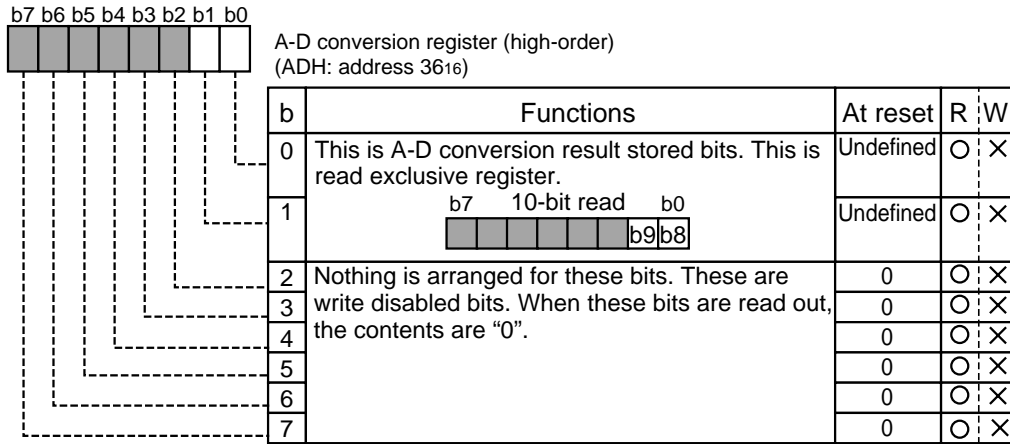


Fig. 2.6.3 Structure of A-D conversion register (high-order)

### A-D conversion register (low-order)

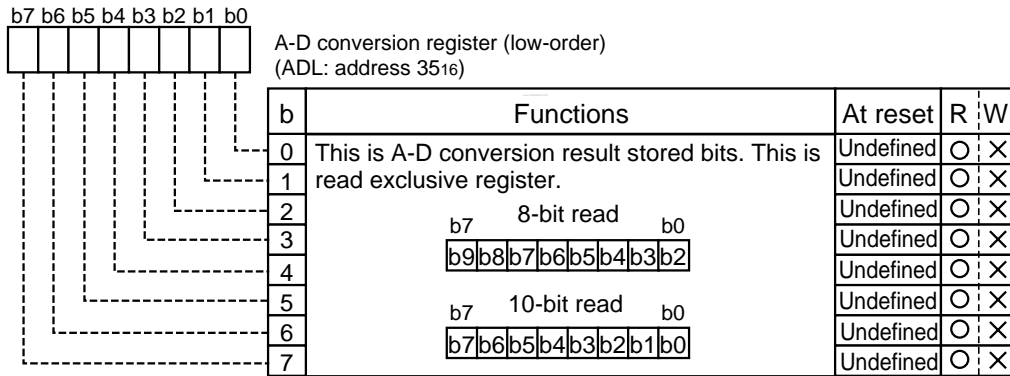


Fig. 2.6.4 Structure of A-D conversion register (low-order)

### Interrupt request register 2

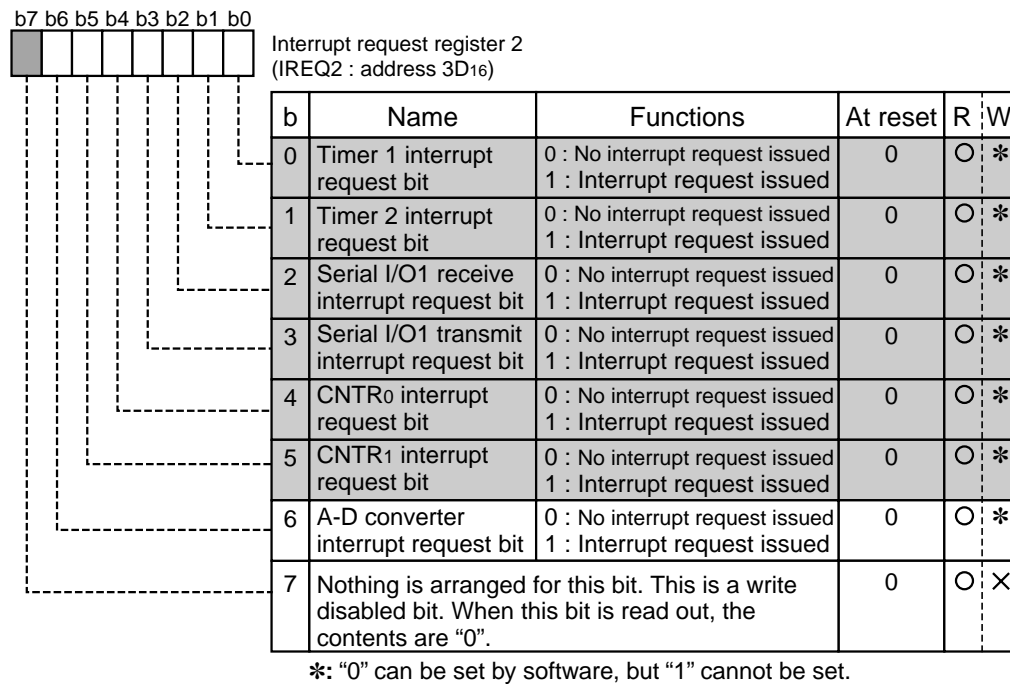


Fig. 2.6.5 Structure of Interrupt request register 2

### Interrupt control register 2

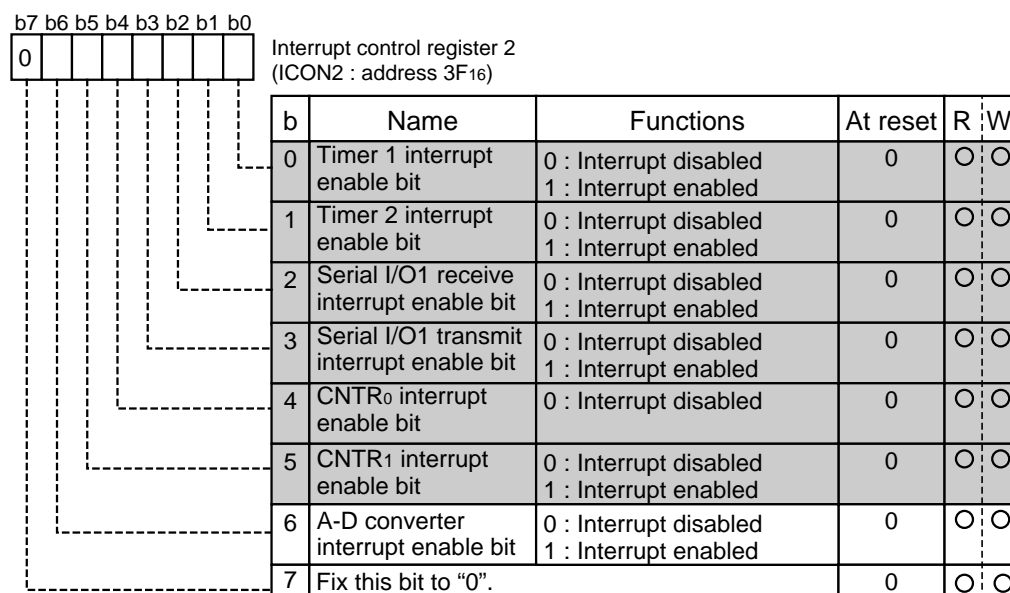


Fig. 2.6.6 Structure of Interrupt control register 2

# APPLICATION

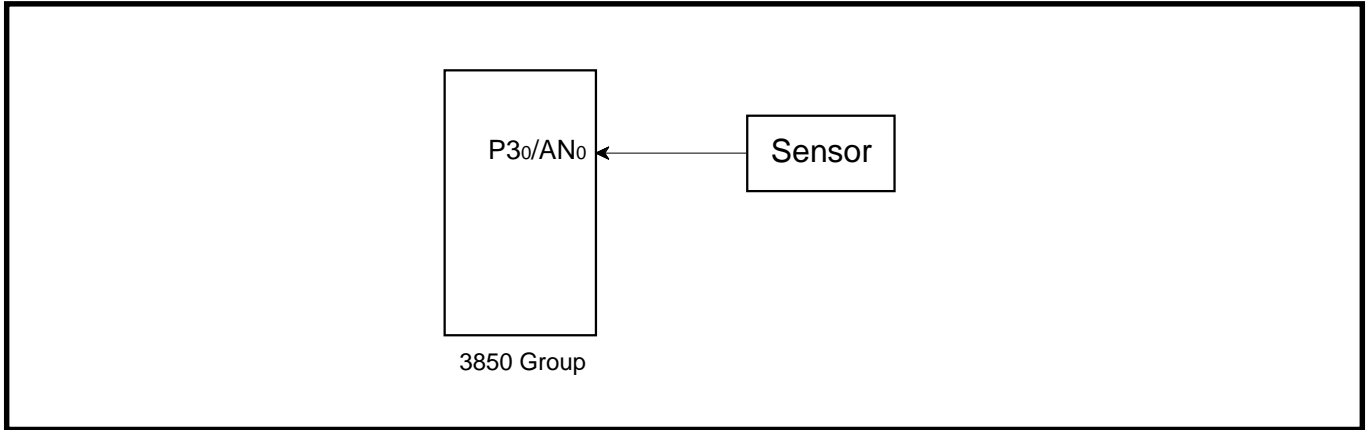
## 2.6 A-D converter

### 2.6.3 A-D converter application examples

#### (1) Conversion of analog input voltage

**Outline :** The analog input voltage input from a sensor is converted to digital values.

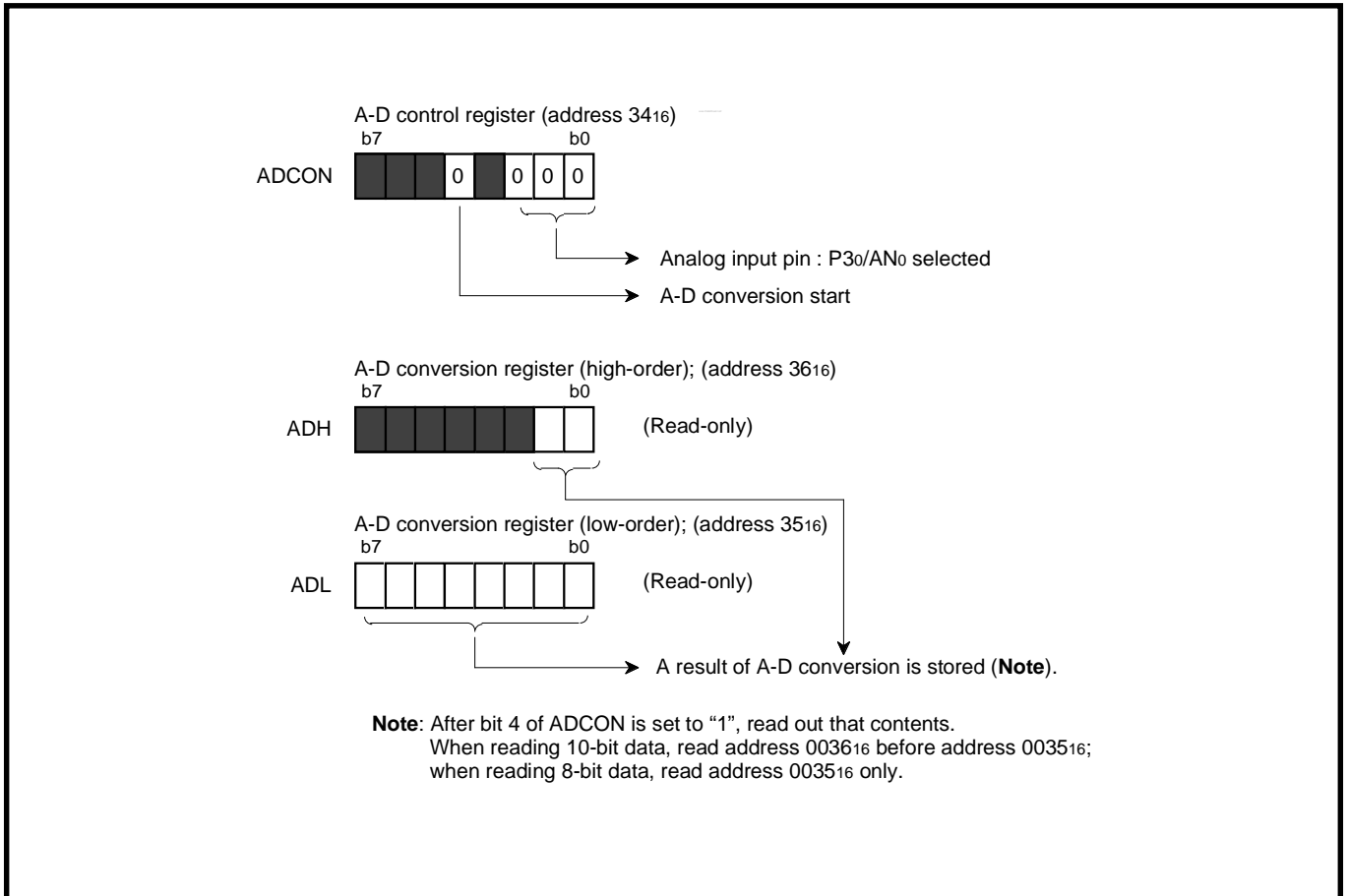
Figure 2.6.7 shows a connection diagram, and Figure 2.6.8 shows the relevant registers setting.



**Fig. 2.6.7 Connection diagram**

**Specifications :**

- The analog input voltage input from a sensor is converted to digital values.
- P30/AN0 pin is used as an analog input pin.



**Fig. 2.6.8 Relevant registers setting**



An analog input signal from a sensor is converted to the digital value according to the relevant registers setting shown by Figure 2.6.8. Figure 2.6.9 shows the control procedure for 8-bit read, and Figure 2.6.10 shows the control procedure for 10-bit read.

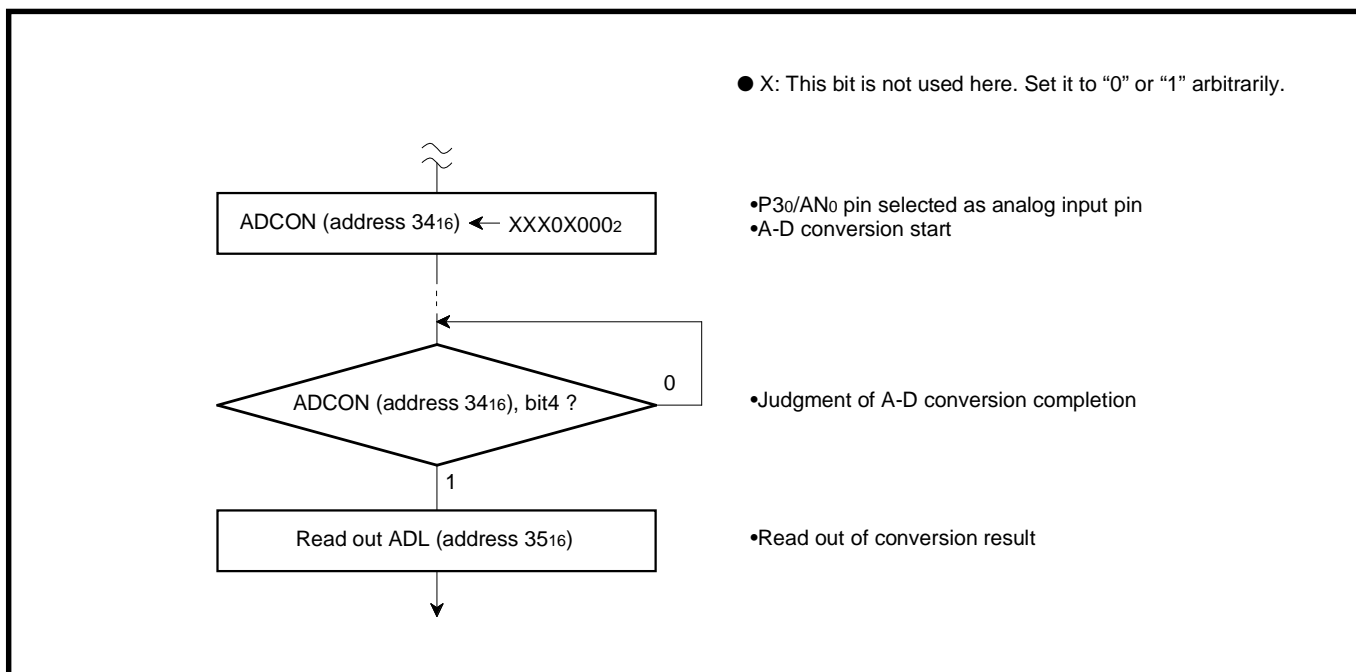


Fig. 2.6.9 Control procedure for 8-bit read

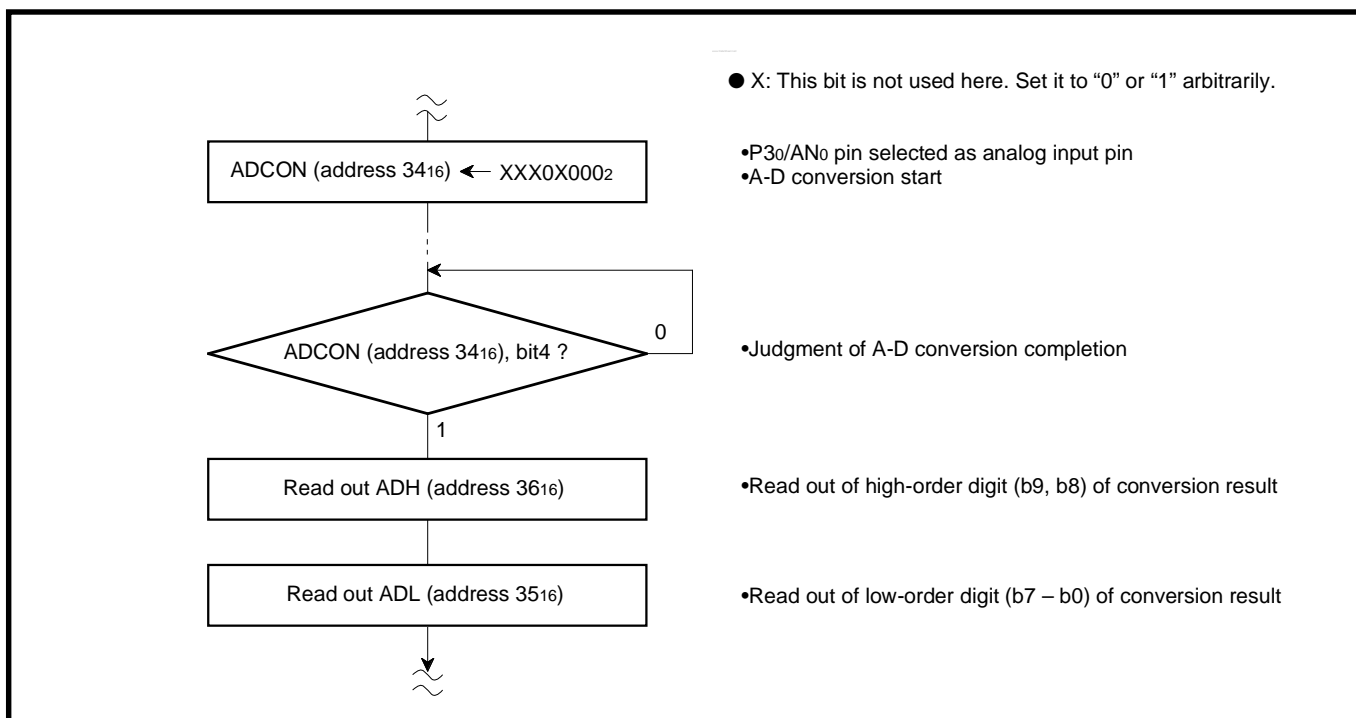


Fig. 2.6.10 Control procedure for 10-bit read

# APPLICATION

## 2.6 A-D converter

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### 2.6.4 Notes on A-D converter

#### (1) Analog input pin

Make the signal source impedance for analog input low, or equip an analog input pin with an external capacitor of 0.01  $\mu$ F to 1  $\mu$ F. Further, be sure to verify the operation of application products on the user side.

##### ● Reason

An analog input pin includes the capacitor for analog voltage comparison. Accordingly, when signals from signal source with high impedance are input to an analog input pin, charge and discharge noise generates. This may cause the A-D conversion precision to be worse.

#### (2) A-D converter power source pin

The AVss pin is A-D converter power source pin. Regardless of using the A-D conversion function or not, connect it as following :

- AVss : Connect to the Vss line

##### ● Reason

If the AVss pin is opened, the microcomputer may have a failure because of noise or others.

#### (3) Clock frequency during A-D conversion

The comparator consists of a capacity coupling, and a charge of the capacity will be lost if the clock frequency is too low. Thus, make sure the following during an A-D conversion.

- $f(X_{IN})$  is 500 kHz or more in middle-/high-speed mode.
- Do not execute the **STP** instruction.
- When the A-D converter is operated at low-speed mode,  $f(X_{IN})$  do not have the lower limit of frequency, because of the A-D converter has a built-in self-oscillation circuit.

### 2.7 Watchdog timer

This paragraph explains the registers setting method and the notes relevant to the watchdog timer.

#### 2.7.1 Memory map

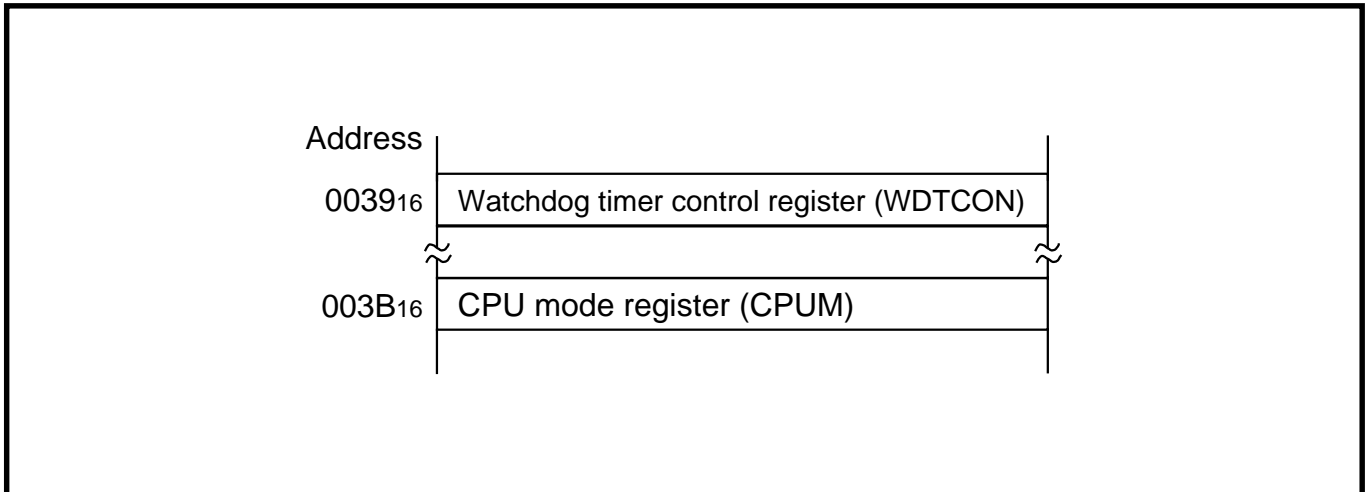


Fig. 2.7.1 Memory map of registers relevant to watchdog timer

#### 2.7.2 Relevant registers

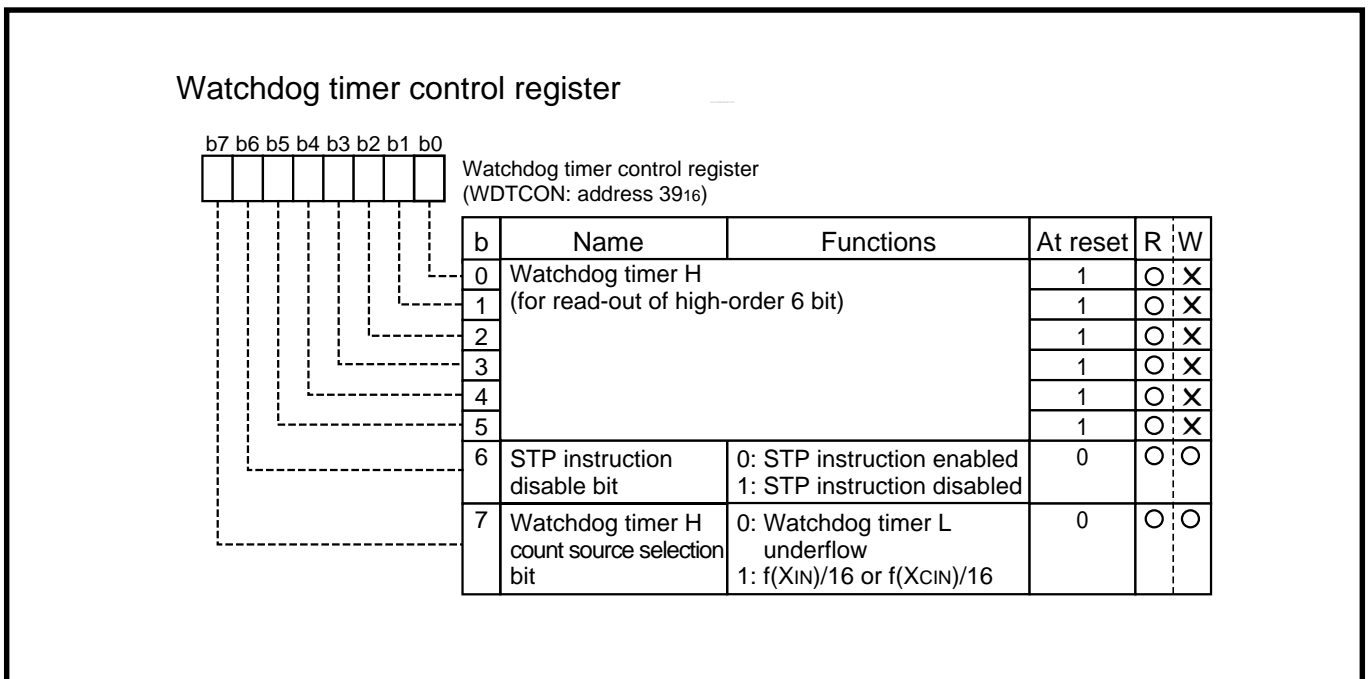


Fig. 2.7.2 Structure of Watchdog timer control register

# APPLICATION

## 2.7 Watchdog timer

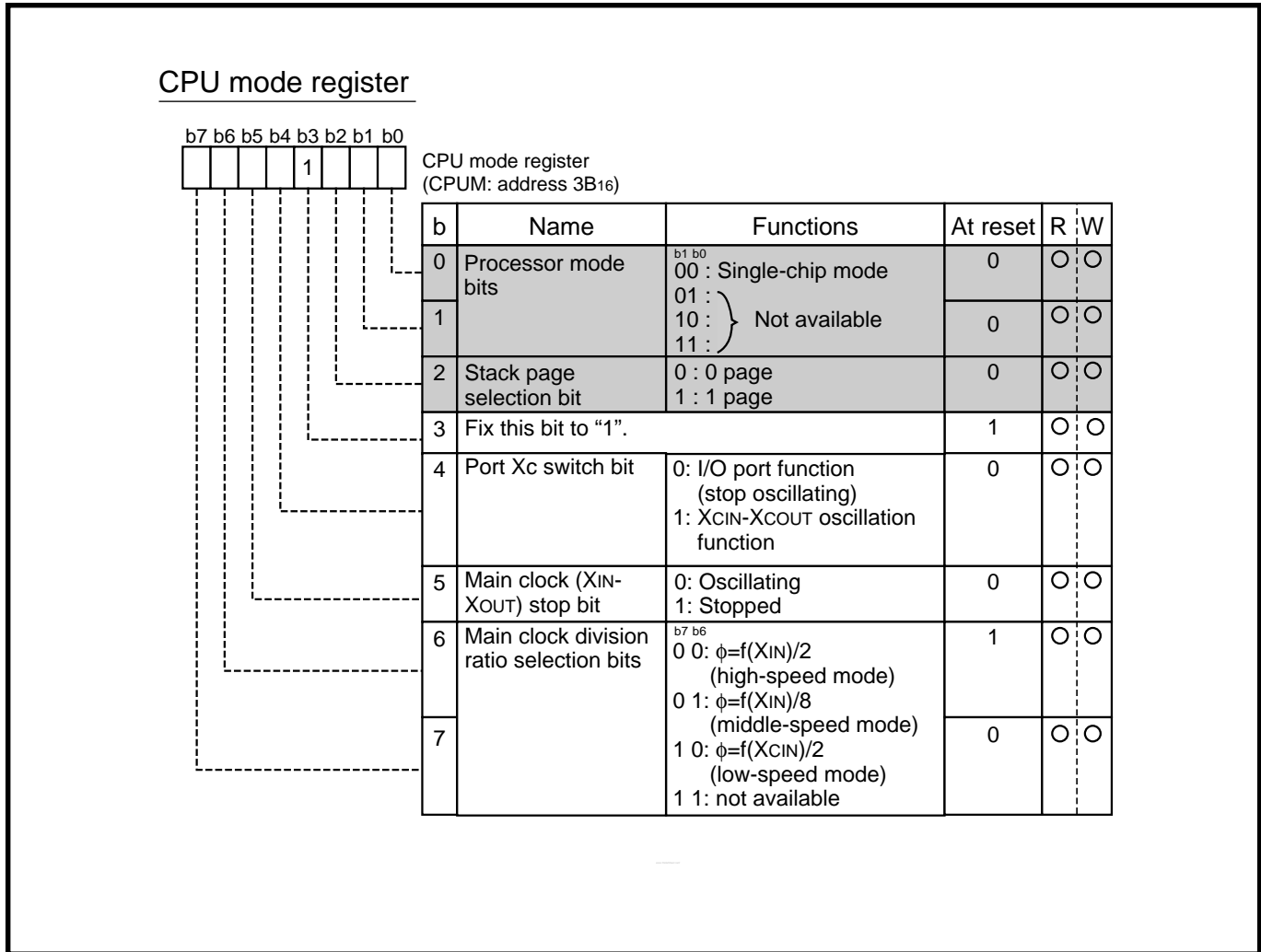


Fig. 2.7.3 Structure of CPU mode register

## 2.7.3 Watchdog timer application examples

## (1) Detection of program runaway

**Outline:** If program runaway occurs, let the microcomputer reset, using the internal timer for detection of program runaway.

- Specifications:**
- An underflow of watchdog timer H is judged to be program runaway, and the microcomputer is returned to the reset status.
  - Before the watchdog timer underflows, "0" is set into bits 6 and 7 of the watchdog timer control register at every cycle in a main routine.
  - High-speed mode is used as a main clock division ratio.
  - An underflow signal of the watchdog timer L is supplied as the count source of watchdog timer H.

Figure 2.7.4 shows a watchdog timer connection and division ratio setting; Figure 2.7.5 shows the relevant registers setting; Figure 2.7.6 shows the control procedure.

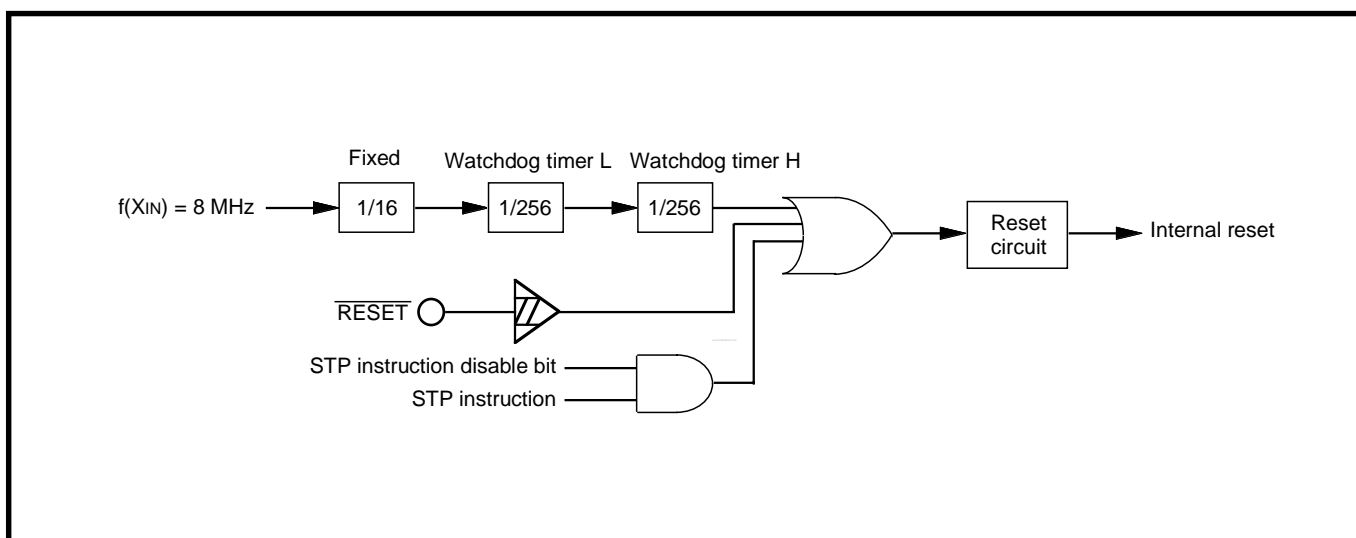


Fig. 2.7.4 Watchdog timer connection and division ratio setting

# APPLICATION

## 2.7 Watchdog timer

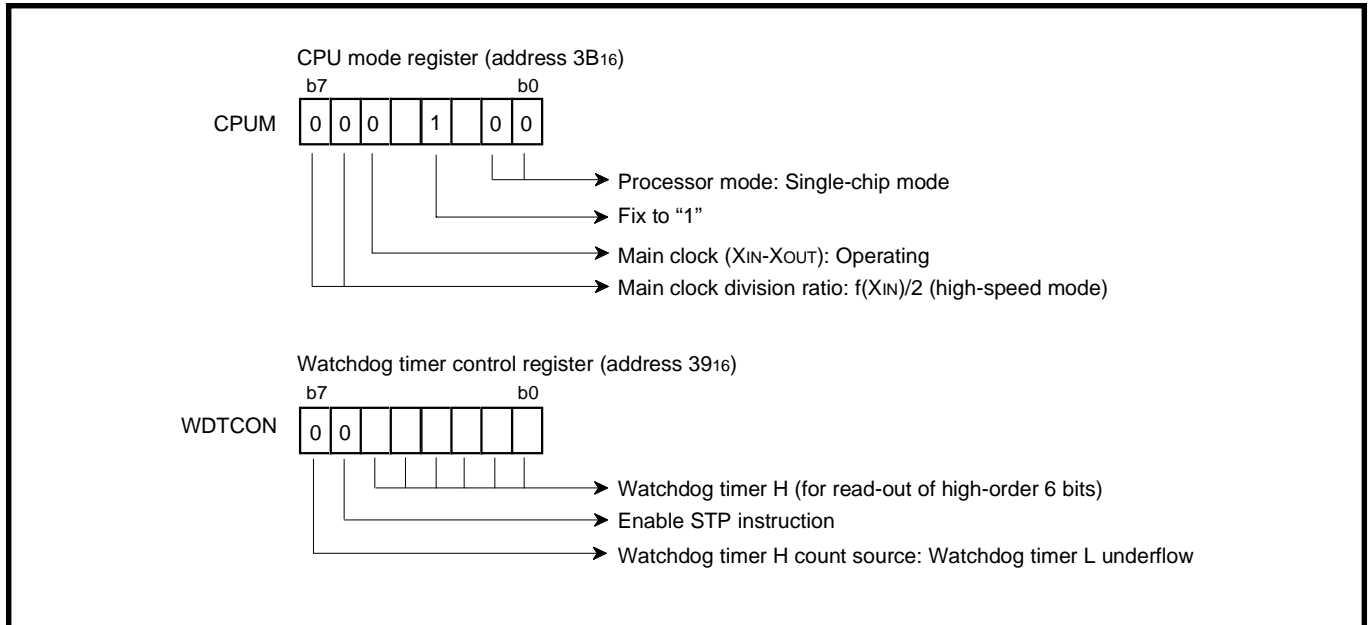


Fig. 2.7.5 Relevant registers setting

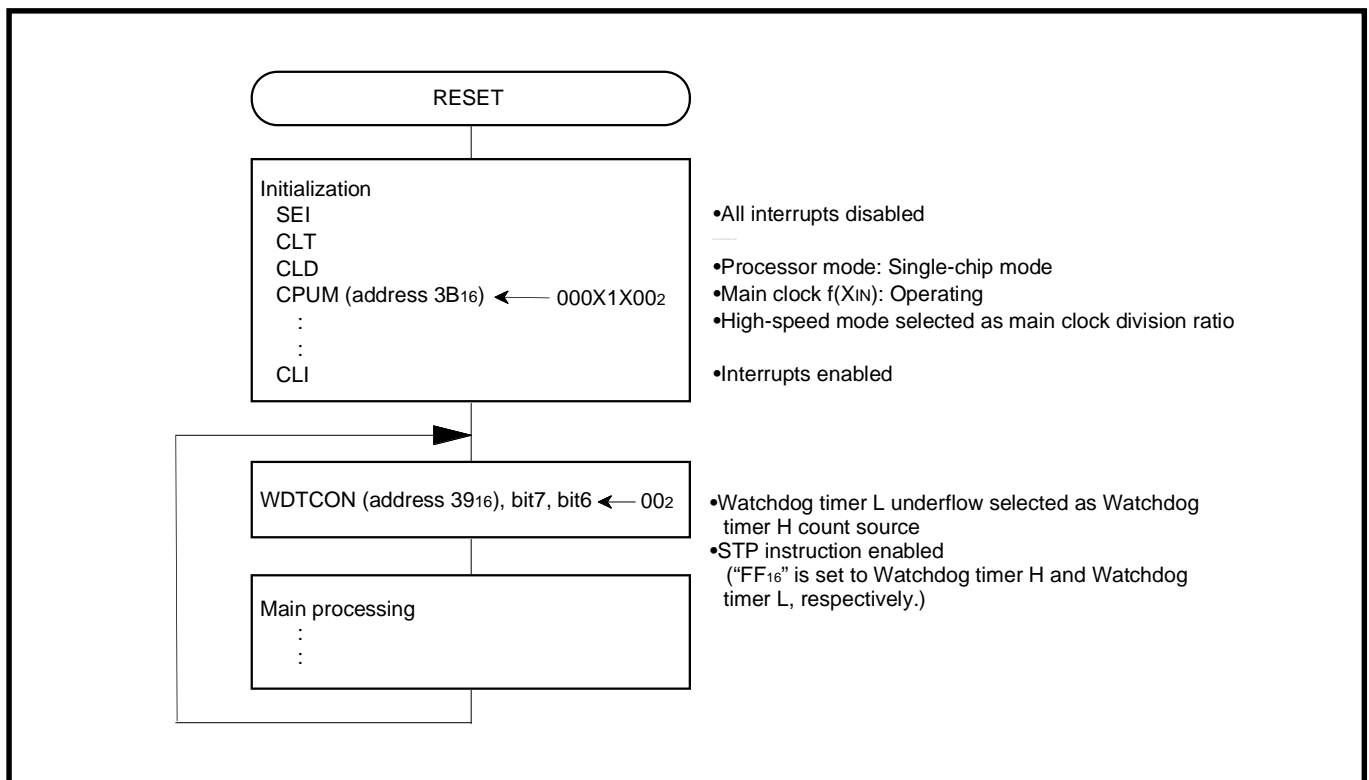


Fig. 2.7.6 Control procedure

### 2.7.4 Notes on watchdog timer

- Make sure that the watchdog timer does not underflow while waiting Stop release, because the watchdog timer keeps counting during that term.
- When the STP instruction disable bit has been set to "1", it is impossible to switch it to "0" by a program.

### 2.8 Reset

#### 2.8.1 Connection example of reset IC

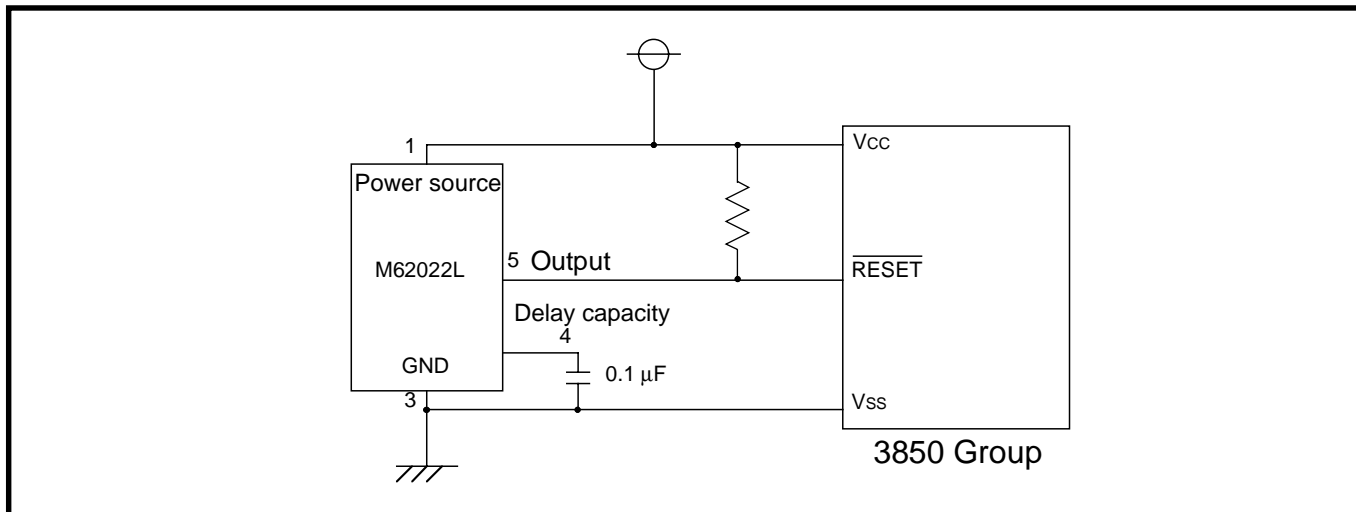


Fig. 2.8.1 Example of poweron reset circuit

Figure 2.8.2 shows the system example which switches to the RAM backup mode by detecting a drop of the system power source voltage with the INT interrupt.

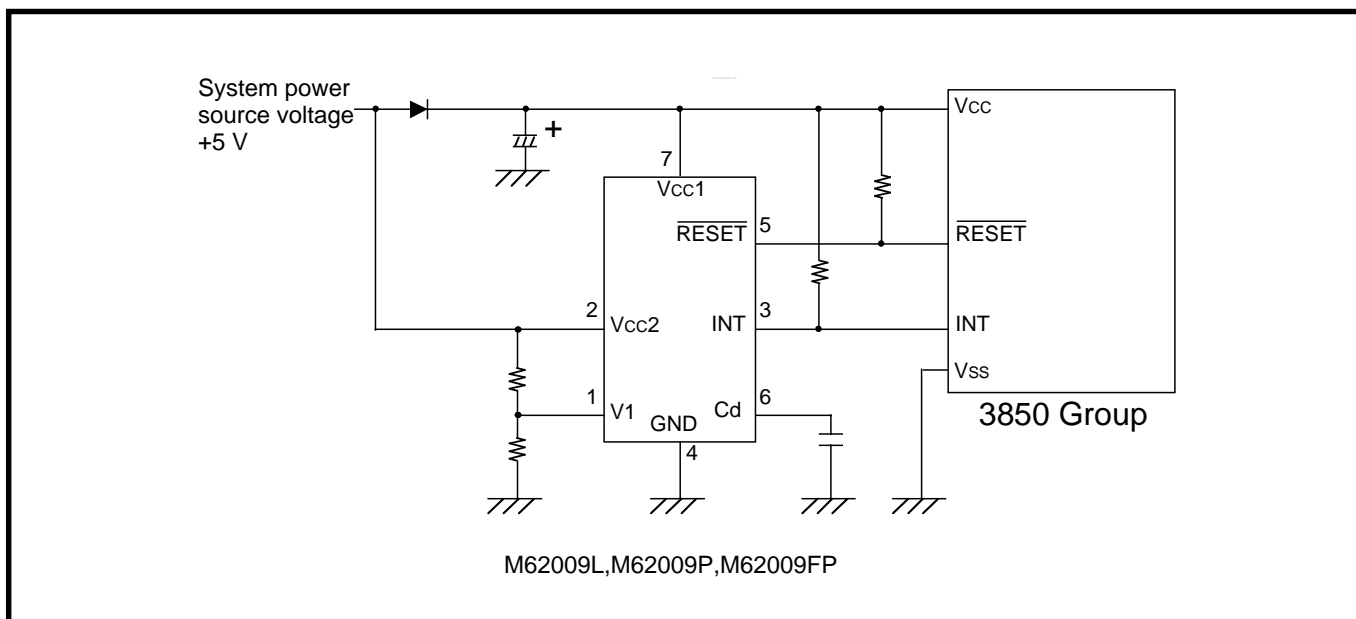


Fig. 2.8.2 RAM backup system

# APPLICATION

## 2.8 Reset

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### 2.8.2 Notes on $\overline{\text{RESET}}$ pin

#### (1) Connecting capacitor

In case where the  $\overline{\text{RESET}}$  signal rise time is long, connect a ceramic capacitor or others across the  $\overline{\text{RESET}}$  pin and the  $V_{SS}$  pin. Use a 1000 pF or more capacitor for high frequency use. When connecting the capacitor, note the following :

- Make the length of the wiring which is connected to a capacitor as short as possible.
- Be sure to verify the operation of application products on the user side.

#### ● Reason

If the several nanosecond or several ten nanosecond impulse noise enters the  $\overline{\text{RESET}}$  pin, it may cause a microcomputer failure.

#### (2) Reset release after power on

When releasing the reset after power on, such as power-on reset, release reset after  $X_{IN}$  passes more than 20 cycles in the state where the power supply voltage is 2.7 V or more and the  $X_{IN}$  oscillation is stable.

#### ● Reason

To release reset, the  $\overline{\text{RESET}}$  pin must be held at an "L" level for 20 cycles or more of  $X_{IN}$  in the state where the power source voltage is between 2.7 V and 5.5 V, and  $X_{IN}$  oscillation is stable.



### 2.9 Clock generating circuit

This paragraph explains how to set the registers relevant to the clock generating circuit and describes an application example.

#### 2.9.1 Relevant registers

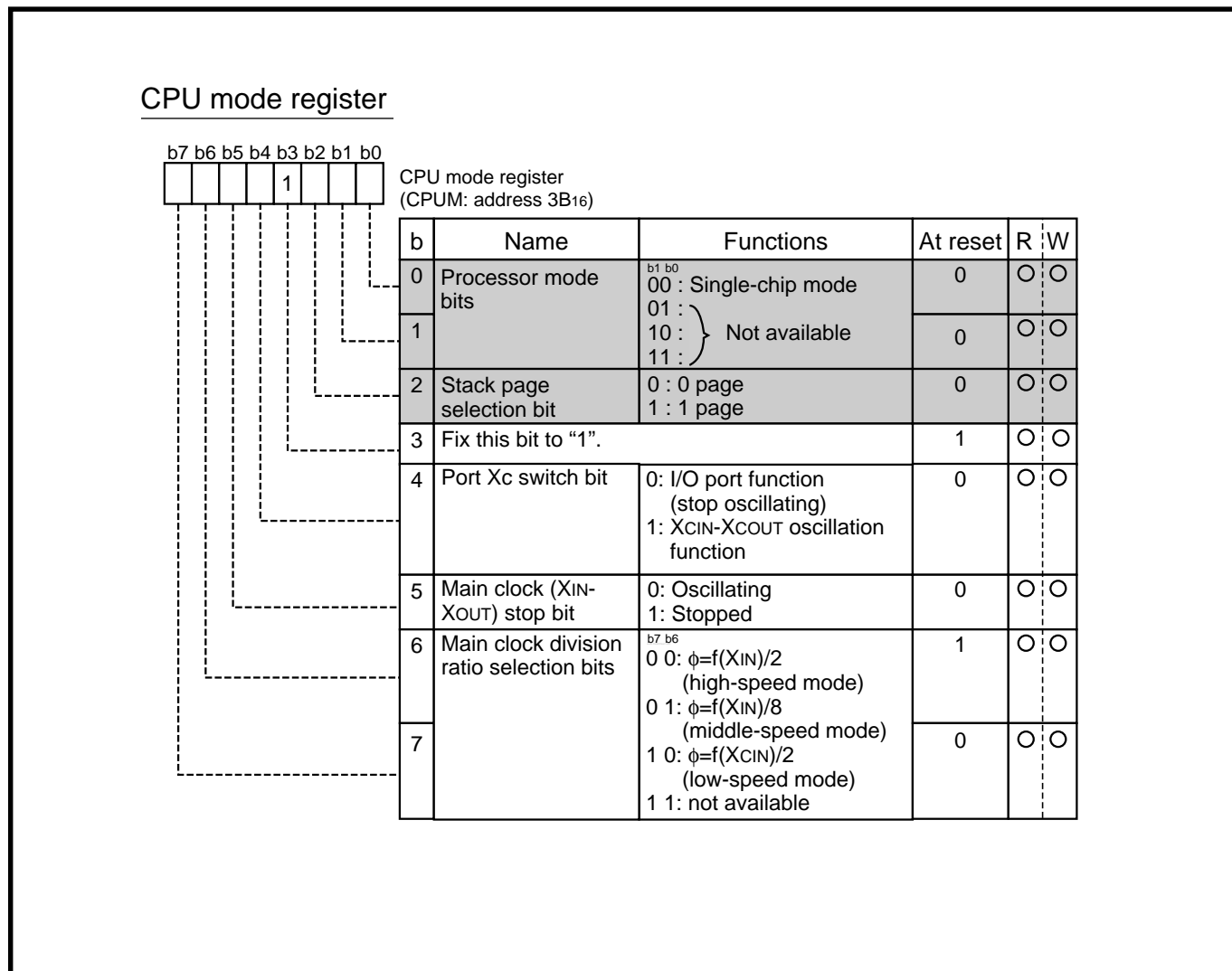


Fig. 2.9.1 Structure of CPU mode register

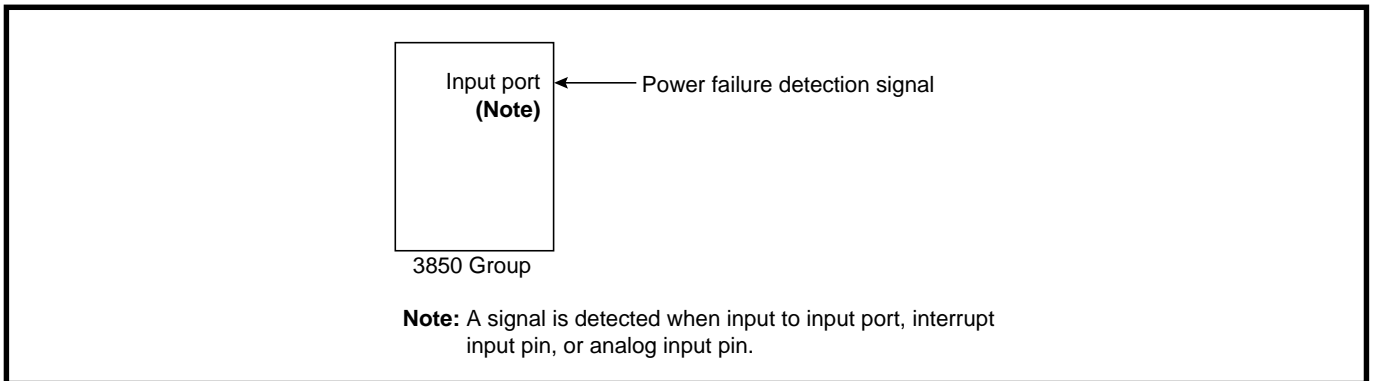
# APPLICATION

## 2.9 Clock generating circuit

### 2.9.2 Clock generating circuit application example

#### (1) Status transition during power failure

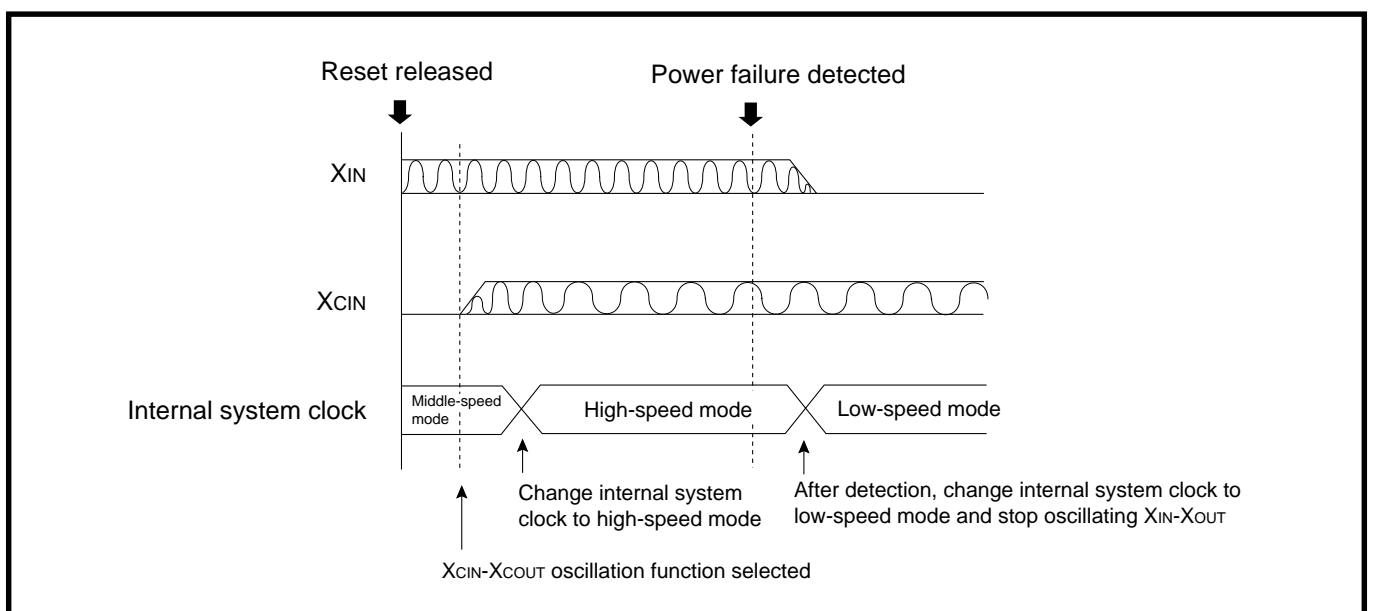
**Outline:** The clock counts up every second by using the timer interrupt during a power failure.



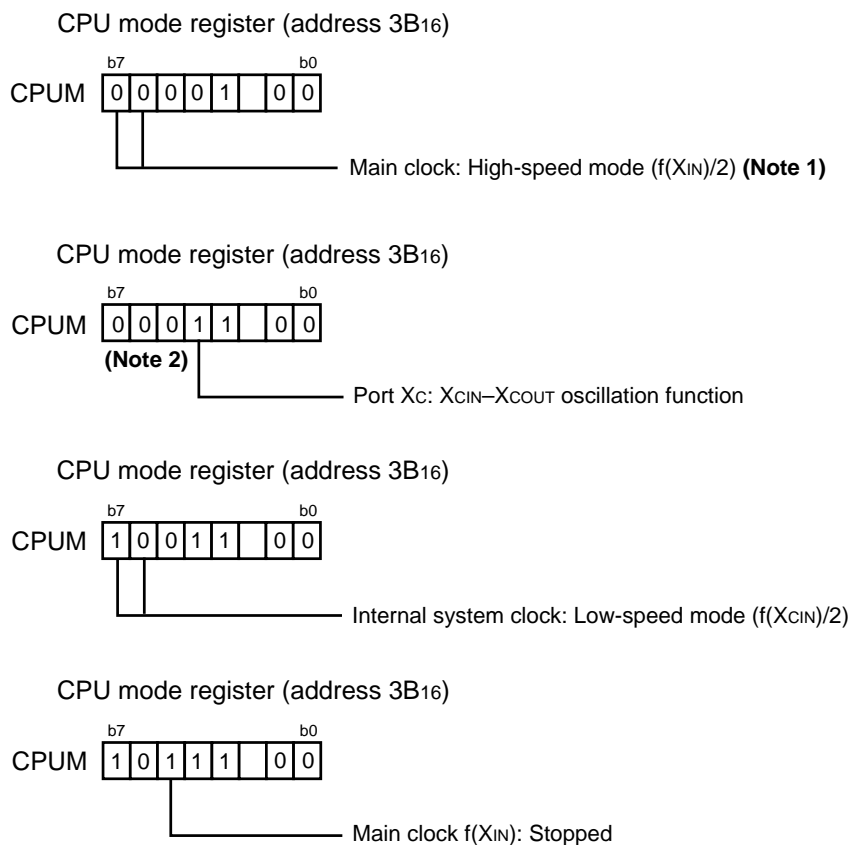
**Fig. 2.9.2 Connection diagram**

- Specifications:**
- Reducing power dissipation as low as possible while maintaining clock function
  - Clock:  $f(X_{IN}) = 8 \text{ MHz}$ ,  $f(X_{CIN}) = 32.768 \text{ kHz}$
  - Port processing
    - Input port: Fixed to “H” or “L” level externally.
    - Output port: Fixed to output level that does not cause current flow to the external.  
(Example) Fix to “H” for an LED circuit that turns on at “L” output level.
    - I/O port: Input port → Fixed to “H” or “L” level externally.  
Output port → Output of data that does not consume current
    - $V_{REF}$  pin: Stop  $V_{REF}$  current dissipation by terminating A-D conversion operation.

Figure 2.9.3 shows the status transition diagram during power failure and Figure 2.9.4 shows the setting of relevant registers.



**Fig. 2.9.3 Status transition diagram during power failure**



**Notes 1:** This setting is necessary only when selecting the high-speed mode.

**2:** When selecting the middle-speed mode, bit 6 is "1".

**Fig. 2.9.4** Setting of relevant registers

# APPLICATION

## 2.9 Clock generating circuit

**Control procedure:** To prepare for a power failure, set the relevant registers in the order shown below.

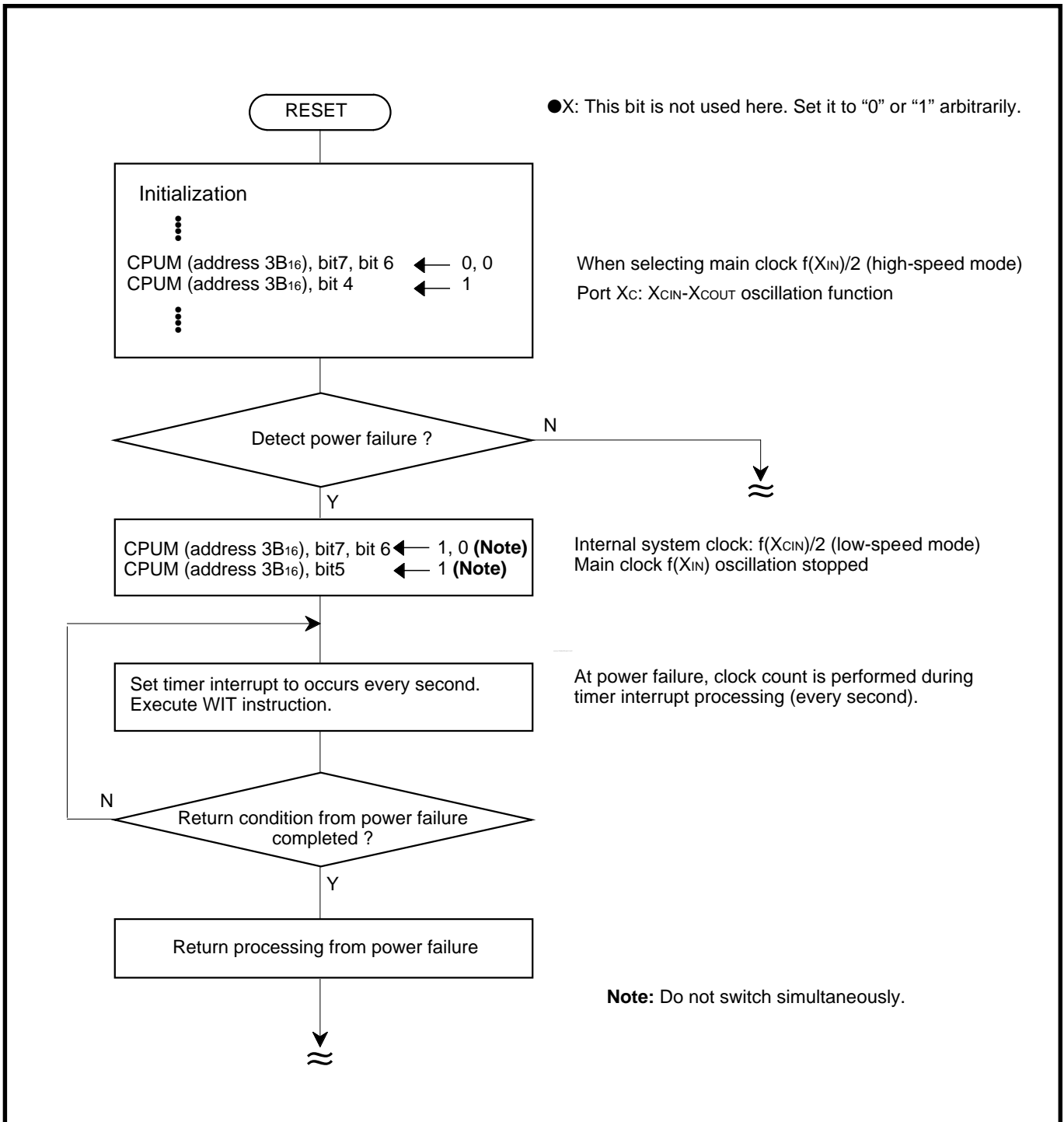


Fig. 2.9.5 Control procedure

## 2.10 Standby function

The 3850 group is provided with standby functions to stop the CPU by software and put the CPU into the low-power operation.

The following two types of standby functions are available.

- Stop mode using STP instruction
- Wait mode using WIT instruction

### 2.10.1 Relevant registers

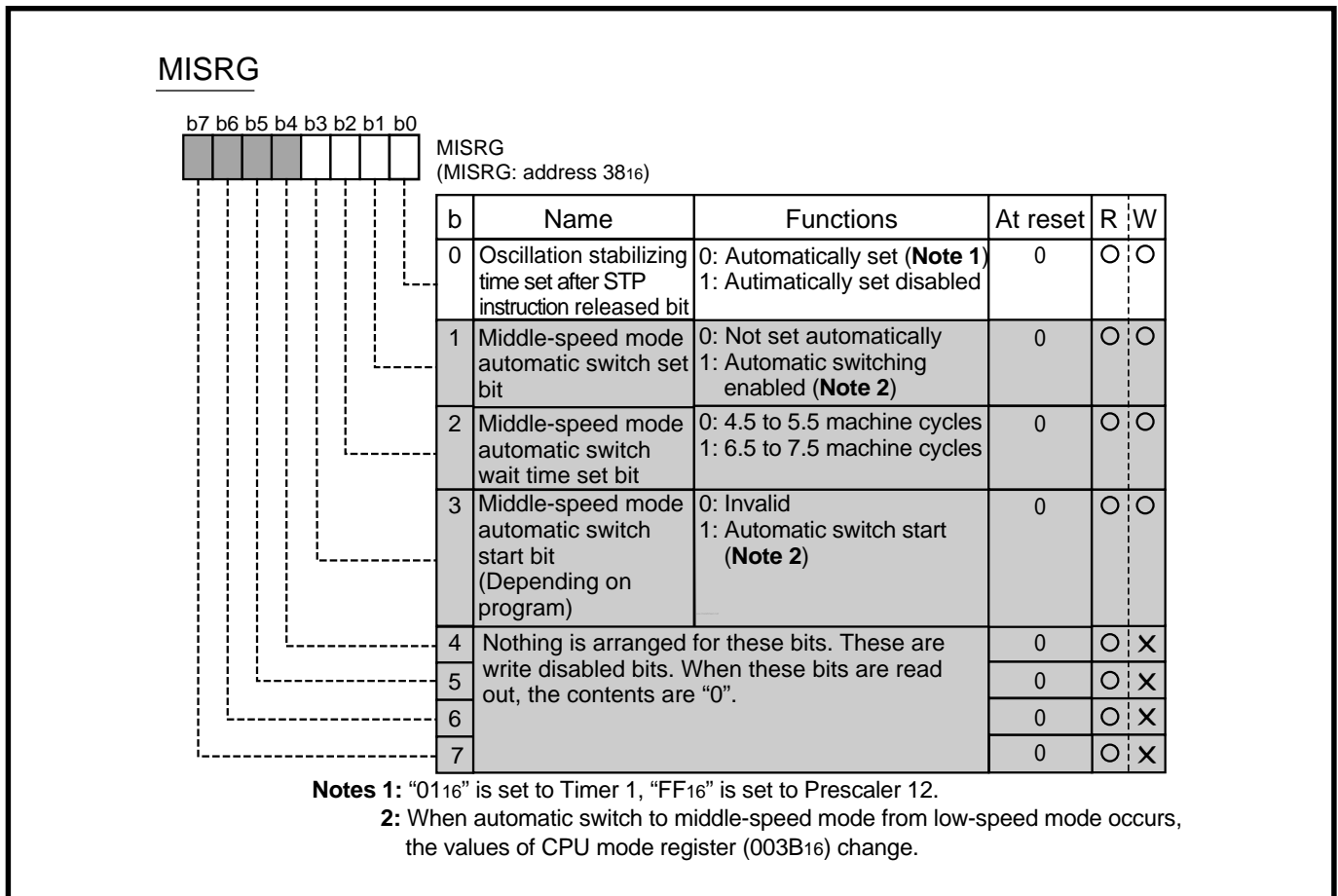


Fig. 2.10.1 Structure of MISRG

# APPLICATION

## 2.10 Standby function

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### 2.10.2 Stop mode

The stop mode is set by executing the STP instruction. In the stop mode, the oscillation of both clocks ( $X_{IN}$ – $X_{OUT}$ ,  $X_{CIN}$ – $X_{COUT}$ ) stop and the internal clock  $\phi$  stops at the “H” level. The CPU stops and peripheral units stop operating. As a result, power dissipation is reduced.

#### (1) State in stop mode

Table 2.10.1 shows the state in the stop mode.

**Table 2.10.1 State in stop mode**

Item	State in stop mode
Oscillation	Stopped.
CPU	Stopped.
Internal clock $\phi$	Stopped at “H” level.
I/O ports P0–P4	Retains the state at the STP instruction execution.
Timer	Stopped. (Timers 1, 2, X, Y) However, Timers X and Y can be operated in the event counter mode.
PWM	Stopped.
Watchdog timer	Stopped.
Serial I/O1, Serial I/O2	Stopped. However, these can be operated only when an external clock is selected.
A-D converter	Stopped.

**(2) Release of stop mode**

The stop mode is released by a reset input or by the occurrence of an interrupt request. Note the differences in the restoration process according to reset input or interrupt request, as described below.

**■ Restoration by reset input**

The stop mode is released by holding the  $\overline{\text{RESET}}$  pin to the “L” input level during the stop mode. Oscillation is started when all ports are in the input state and the stop mode of the main clock ( $X_{\text{IN}}$ - $X_{\text{OUT}}$ ) is released.

Oscillation is unstable when restarted. For this reason, time for stabilizing of oscillation (oscillation stabilizing time) (**Note**) is required. The input of the  $\overline{\text{RESET}}$  pin should be held at the “L” level until oscillation stabilizes.

When the  $\overline{\text{RESET}}$  pin is held at the “L” level for 20 cycles or more of  $X_{\text{IN}}$  after the oscillation has stabilized, the microcomputer will go to the reset state. After the input level of the  $\overline{\text{RESET}}$  pin is returned to “H”, the reset state is released in approximately 10.5 to 18.5 cycles of the  $X_{\text{IN}}$  input. Figure 2.10.2 shows the oscillation stabilizing time at restoration by reset input.

At release of the stop mode by reset input, the internal RAM retains its contents previous to the reset. However, the previous contents of the CPU register and SFR are not retained. For more details concerning reset, refer to “2.8 Reset”.

**Note:** For the setting of oscillation stabilizing time, refer to MISRG (address 0038<sub>16</sub>).

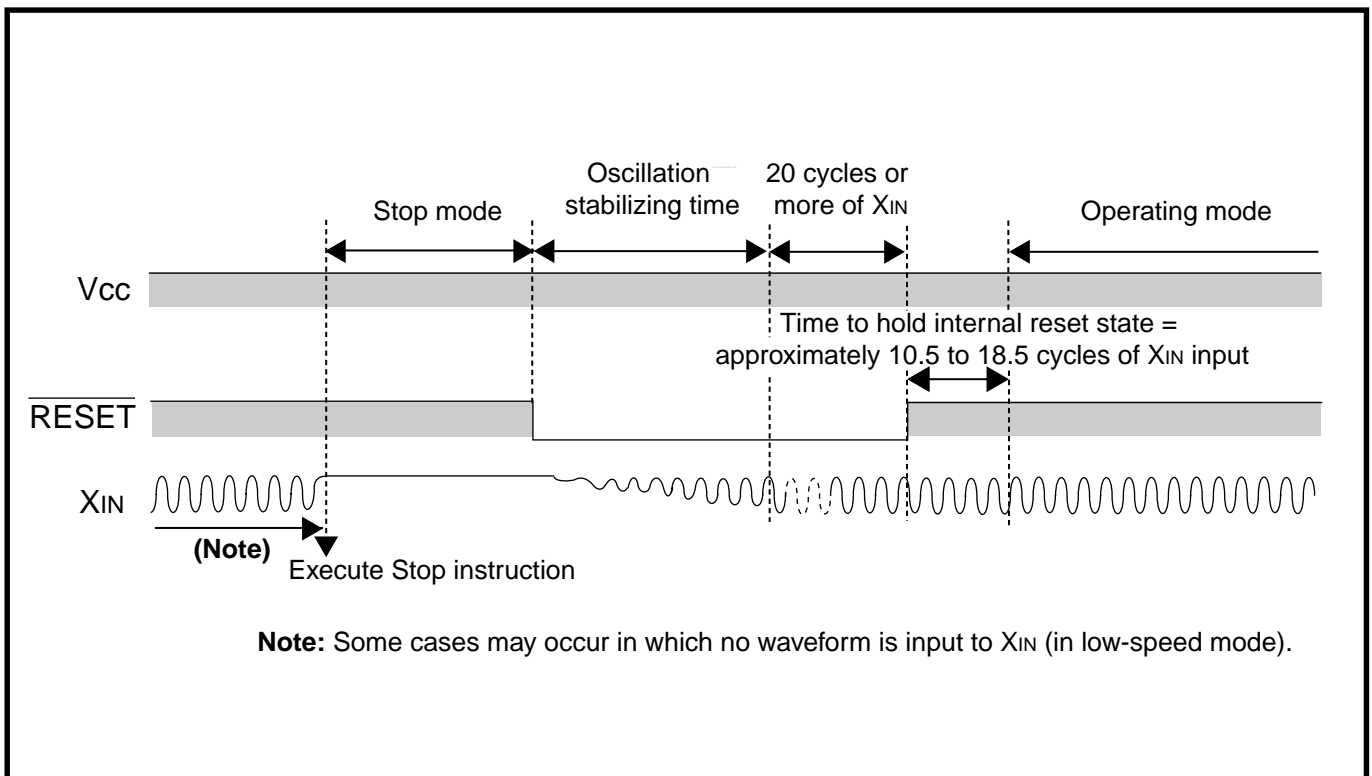


Fig. 2.10.2 Oscillation stabilizing time at restoration by reset input

# APPLICATION

## 2.10 Standby function

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### ■ Restoration by interrupt request

The occurrence of an interrupt request in the stop mode releases the stop mode. As a result, oscillation is resumed. The interrupts available for restoration are:

- INT<sub>0</sub>–INT<sub>3</sub>
- CNTR<sub>0</sub>, CNTR<sub>1</sub>
- Serial I/O (1, 2) using an external clock
- Timer X, Y using an external event count

However, when using any of these interrupt requests for restoration from the stop mode, in order to enable the selected interrupt, you must execute the STP instruction after setting the following conditions.

[Necessary register setting]

- ① Interrupt disable flag I = “0” (interrupt enabled)
- ② Timer 1 interrupt enable bit = “0” (interrupt disabled)
- ③ Interrupt request bit of interrupt source to be used for restoration = “0” (no interrupt request issued)
- ④ Interrupt enable bit of interrupt source to be used for restoration = “1” (interrupts enabled)

For more details concerning interrupts, refer to “2.2 Interrupts”.

Oscillation is unstable when restarted. For this reason, time for stabilizing of oscillation (oscillation stabilizing time) is required. For restoration by an interrupt request, waiting time prior to supplying internal clock  $\phi$  to the CPU is automatically generated\*2 by Prescaler 12 and Timer 1\*1. This waiting time is reserved as the oscillation stabilizing time on the system clock side. The supply of internal clock  $\phi$  to the CPU is started at the Timer 1 underflow.

Figure 2.10.3 shows an execution sequence example at restoration by the occurrence of an INT<sub>0</sub> interrupt request.

\*1: If the STP instruction is executed when the oscillation stabilizing time set after STP instruction released bit is “0”, “FF<sub>16</sub>” and “01<sub>16</sub>” are automatically set in the Prescaler 12 counter/latch and Timer 1 counter/latch, respectively. When the oscillation stabilizing time set after STP instruction released bit is “1”, nothing is automatically set to either Prescaler 12 or Timer 1. For this reason, any suitable value can be set to Prescaler 12 and Timer 1 for the oscillation stabilizing time.

\*2: Immediately after the oscillation is started, the count source is supplied to the prescaler 12 so that a count operation is started.



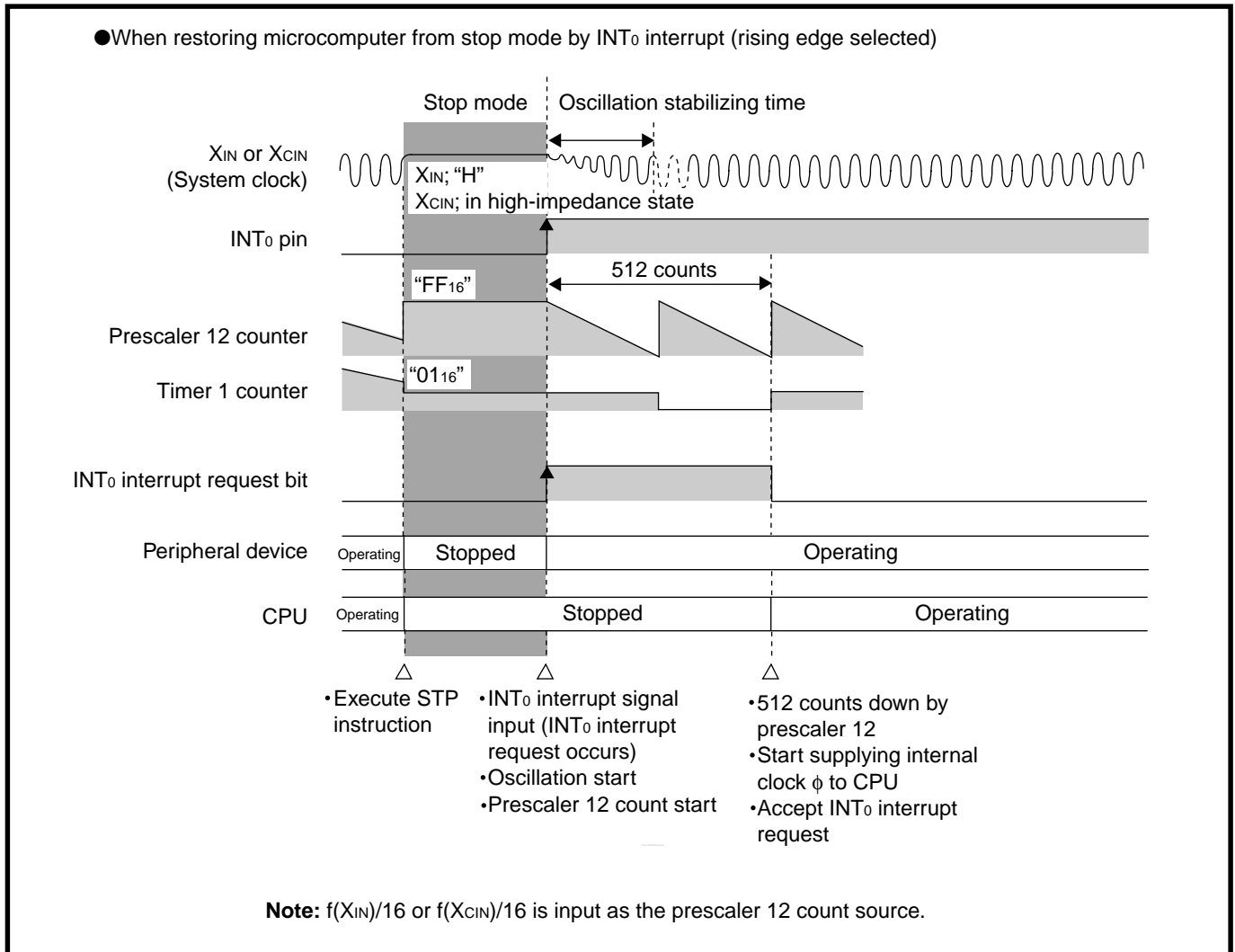


Fig. 2.10.3 Execution sequence example at restoration by occurrence of INT<sub>0</sub> interrupt request

### (3) Notes on using stop mode

#### ■Register setting

Since values of the prescaler 12 and Timer 1 are automatically reloaded when returning from the stop mode, set them again, respectively. (When the oscillation stabilizing time set after STP instruction released bit is "0")

#### ■Clock restoration

After restoration from the stop mode to the normal mode by an interrupt request, the contents of the CPU mode register previous to the STP instruction execution are retained. Accordingly, if both main clock and sub clock were oscillating before execution of the STP instruction, the oscillation of both clocks is resumed at restoration.

In the above case, when the main clock side is set as a system clock, the oscillation stabilizing time for approximately 8,000 cycles of the X<sub>IN</sub> input is reserved at restoration from the stop mode. At this time, note that the oscillation on the sub clock side may not be stabilized even after the lapse of the oscillation stabilizing time of the main clock side.

# APPLICATION

## 2.10 Standby function

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### 2.10.3 Wait mode

The wait mode is set by execution of the WIT instruction. In the wait mode, oscillation continues, but the internal clock  $\phi$  stops at the "H" level.

The CPU stops, but most of the peripheral units continue operating.

#### (1) State in wait mode

The continuation of oscillation permits clock supply to the peripheral units. Table 2.10.2 shows the state in the wait mode.

**Table 2.10.2 State in wait mode**

Item	State in wait mode
Oscillation	Operating.
CPU	Stopped.
Internal clock $\phi$	Stopped at "H" level.
I/O ports P0–P4	Retains the state at the WIT instruction execution.
Timer	Operating.
PWM	Operating.
Watchdog timer	Operating.
Serial I/O1, Serial I/O2	Operating.
A-D converter	Operating.

**(2) Release of wait mode**

The wait mode is released by reset input or by the occurrence of an interrupt request. Note the differences in the restoration process according to reset input or interrupt request, as described below.

In the wait mode, oscillation is continued, so an instruction can be executed immediately after the wait mode is released.

**■Restoration by reset input**

The wait mode is released by holding the input level of the  $\overline{\text{RESET}}$  pin at "L" in the wait mode. Upon release of the wait mode, all ports are in the input state, and supply of the internal clock  $\phi$  to the CPU is started. To reset the microcomputer, the  $\overline{\text{RESET}}$  pin should be held at an "L" level for 20 cycles or more of  $X_{\text{IN}}$ . The reset state is released in approximately 10.5 cycles to 18.5 cycles of the  $X_{\text{IN}}$  input after the input of the  $\overline{\text{RESET}}$  pin is returned to the "H" level.

At release of wait mode, the internal RAM retains its contents previous to the reset. However, the previous contents of the CPU register and SFR are not retained.

Figure 2.10.4 shows the reset input time.

For more details concerning reset, refer to "2.8 Reset".

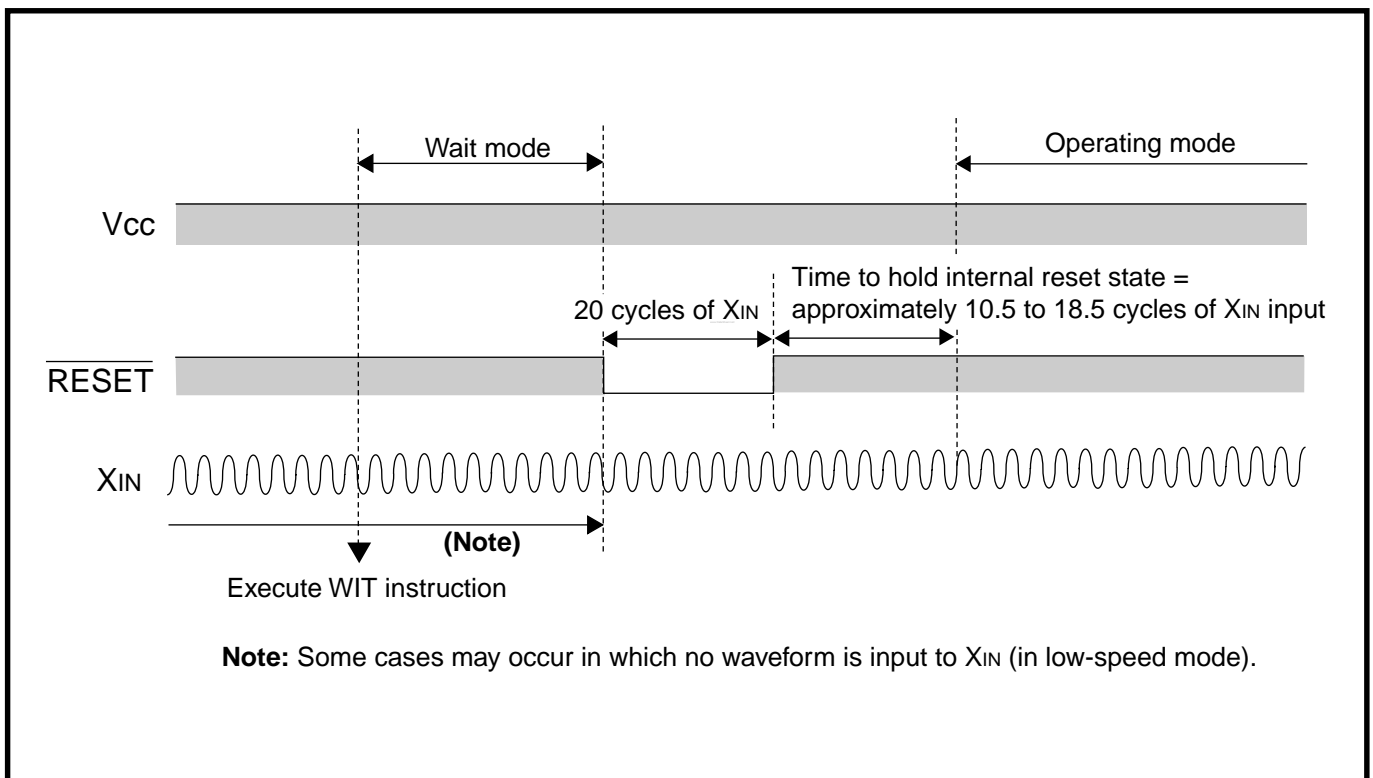


Fig. 2.10.4 Reset input time

# APPLICATION

## 2.10 Standby function

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### ■Restoration by interrupt request

In the wait mode, the occurrence of an interrupt request releases the wait mode and supply of the internal clock  $\phi$  to the CPU is started. At the same time, the interrupt request used for restoration is accepted, so the interrupt processing routine is executed.

However, when using an interrupt request for restoration from the wait mode, in order to enable the selected interrupt, you must execute the WIT instruction after setting the following conditions.

[Necessary register setting]

- ① Interrupt disable flag I = "0" (interrupt enabled)
- ② Interrupt request bit of interrupt source to be used for restoration = "0" (no interrupt request issued)
- ③ Interrupt enable bit of interrupt source to be used for restoration = "1" (interrupts enabled)

For more details concerning interrupts, refer to "2.2 Interrupts".

### (3) Notes on wait mode

#### ■Clock restoration

If the wait mode is released by a reset when  $X_{CIN}$  is set as the system clock and  $X_{IN}$  oscillation is stopped during execution of the WIT instruction,  $X_{CIN}$  oscillation stops,  $X_{IN}$  oscillations starts, and  $X_{IN}$  is set as the system clock.

In the above case, the  $\overline{RESET}$  pin should be held at "L" until the oscillation is stabilized.

### 2.11 Flash memory mode

This paragraph explains the registers setting method and the notes relevant to the flash memory version.

#### 2.11.1 Overview

The functions of the flash memory version are similar to those of the mask ROM version except that the flash memory is built-in and some of the SFR area differ from that of the mask ROM version (refer to “2.11.2 Memory map”).

In the flash memory version, the built-in flash memory can be programmed or erased by using the following three modes.

- CPU rewrite mode
- Parallel I/O mode
- Standard serial I/O mode

#### 2.11.2 Memory map

M38507F8FP/SP have 32 Kbytes of built-in flash memory.

Figure 2.11.1 shows the memory map of the flash memory version.

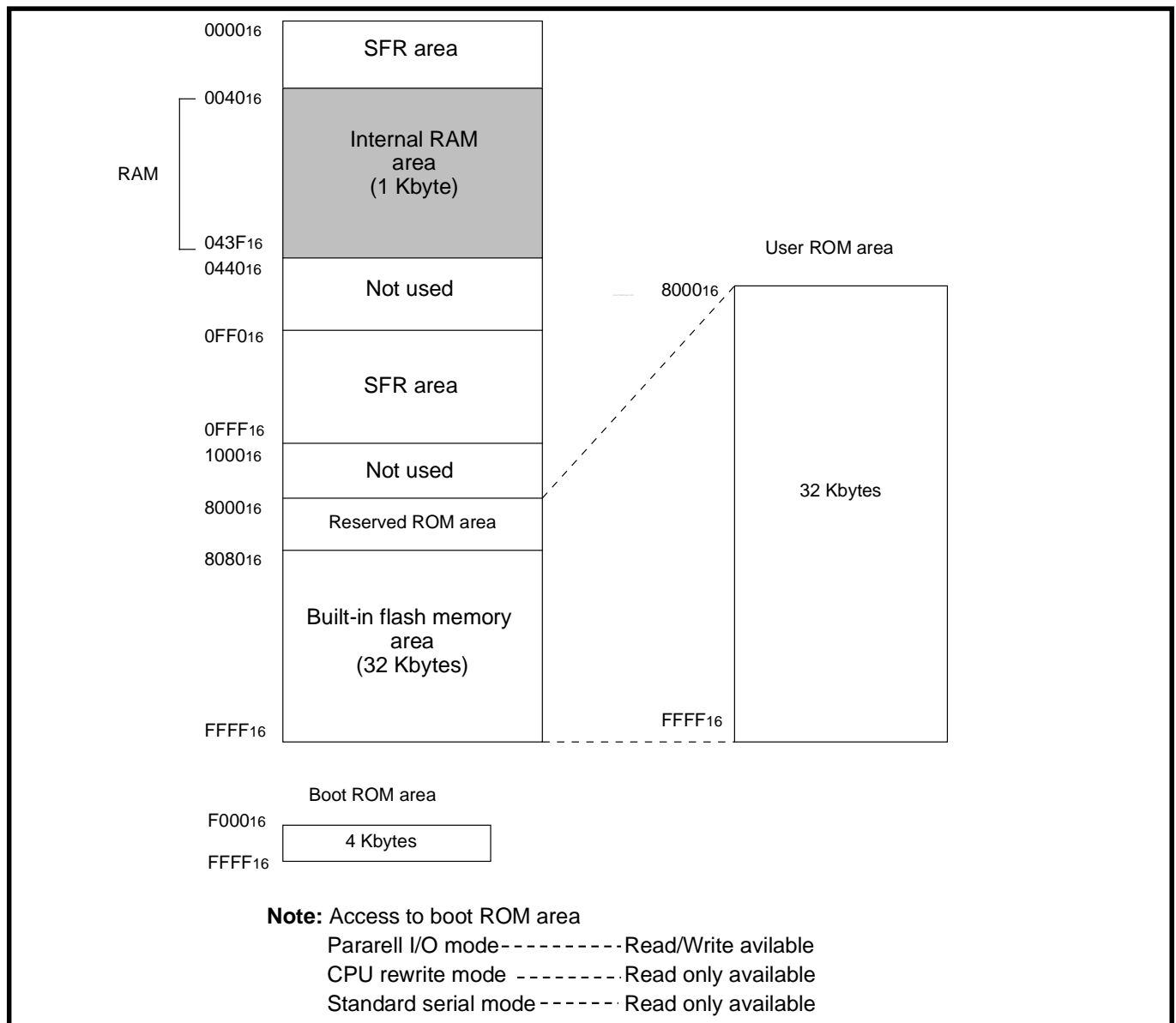


Fig. 2.11.1 Memory map of flash memory version for 3850 Group

# APPLICATION

## 2.11 Flash memory mode

### 2.11.3 Relevant registers

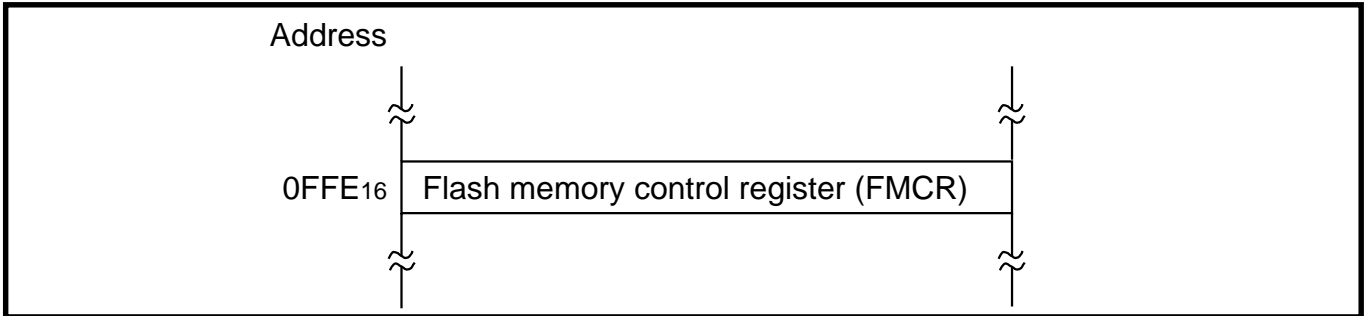


Fig. 2.11.2 Memory map of registers relevant to flash memory

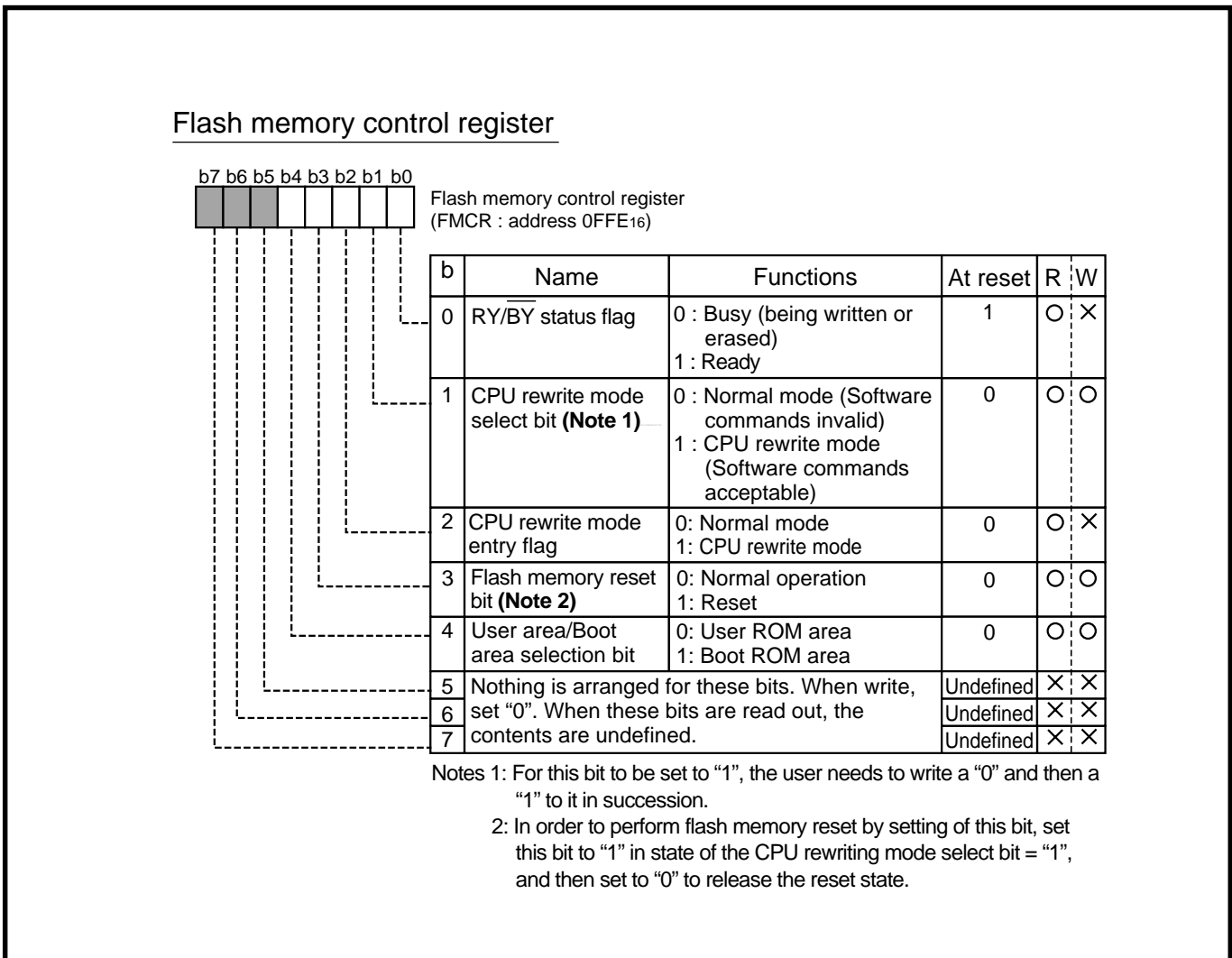


Fig. 2.11.3 Structure of Flash memory control register

### 2.11.4 Parallel I/O mode

In the parallel I/O mode, program/erase to the built-in flash memory can be performed by a EPROM programmer (EFP-I).

The memory area of program/erase is from 0F000<sub>16</sub> to 0FFFF<sub>16</sub> (boot ROM area) or from 08000<sub>16</sub> to 0FFFF<sub>16</sub> (user ROM area). Be especially careful when erasing; if the memory area is not set correctly, the products will be damaged eternally.

Table 2.11.1 shows the setting of programmers when programming in the parallel I/O mode.

•EFP-I provided by Suisai Electronics System Co., Ltd. ([http://www.suisai.co.jp/index\\_e.htm](http://www.suisai.co.jp/index_e.htm))  
(product available in Asia and Oceania only)

**Table 2.11.1 Setting of programmers when parallel programming**

Products	Parallel unit	Boot ROM area	User ROM area
M38507F8FP	EF3850F-42E	0F000 <sub>16</sub> to 0FFFF <sub>16</sub>	08000 <sub>16</sub> to 0FFFF <sub>16</sub>
M38507F8SP	EF3850F-42S		

### 2.11.5 Standard serial I/O mode

Table 2.11.2 shows a pin connection example (4 wires) between the programmer (EFP-I; Serial unit EF1SRP-01U is required additionally) and the microcomputer when programming in the serial I/O mode.

•EFP-I provided by Suisai Electronics System Co., Ltd. ([http://www.suisai.co.jp/index\\_e.htm](http://www.suisai.co.jp/index_e.htm))  
(product available in Asia and Oceania only)

**Table 2.11.2 Connection example to programmer when serial programming (4 wires)**

Function	EFP-I (EF1SRP-01U)		3850 Group flash memory version	
	Signal name	EF1SRP-01U side connector Line number	Pin name	Pin number
Transfer clock input	T_SCLK1	9	P2 <sub>6</sub> /SCLK <sub>1</sub>	10
Serial data input	T_RXD	11	P2 <sub>5</sub> /TxD	11
Serial data output	T_TXD	10	P2 <sub>4</sub> /RxD	12
Transmit/Receive enable output	T_BUSY	12	P2 <sub>7</sub> /CNTR <sub>0</sub> /S <sub>RDY1</sub>	9
5 V input	T_VPP	3	CNV <sub>SS</sub>	15
Reset input	T_RESET	14	RESET (Note 1)	18
Target board power source monitor input	T_VDD (Note 2)	4	V <sub>CC</sub> (Note 2)	1
GND	GND (Note 3)	1, 2, 15, 16	V <sub>SS</sub> , AV <sub>SS</sub> (Note 3)	21, 3

**Notes 1:** Since reset release after write verification is not performed, when operating MCU after writing, separate a target connection cable.

**2:** Supply V<sub>CC</sub> of EFP-I side from user side so that the power supply voltage of the output buffer used by the EFP-I side becomes the same as user side power supply voltage (V<sub>CC</sub>).

**3:** Four pins (No. 1, 2, 15, and 16) of the EF1SRP-01U side connector are prepared for GND signal. When connecting with a target board, although connection of only one pin does not have a problem, we recommend connecting with two or more pins.

# APPLICATION

## 2.11 Flash memory mode

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### 2.11.6 CPU rewrite mode

In the CPU rewrite mode, issuing software commands through the Central Processing Unit (CPU) can rewrite the built-in flash memory. Accordingly, the contents of the built-in flash memory can be rewritten with the microcomputer itself mounted on board, without using the programmer.

Store the rewrite control program to the built-in flash memory in advance. The built-in flash memory cannot be read in the CPU rewrite mode. Accordingly, after transferring the rewrite control program to the internal RAM, execute it on the RAM.

The following commands can be used in the CPU rewrite mode: read array, read status register, clear status register, program, erase all block, and block erase. For details concerning each command, refer to "CHAPTER 1 Flash memory mode (CPU rewrite mode)".

#### (1) CPU rewrite mode beginning/release procedures

Operation procedure in the CPU rewrite mode for the built-in flash memory is described below. As for the control example, refer to "2.11.7 (2) Control example in the CPU rewrite mode".

##### [Beginning procedure]

- ① Apply 5  $V_{\pm 10}$  % to the CNV<sub>SS</sub>/V<sub>PP</sub> pin (at selecting boot ROM area).
- ② Release reset.
- ③ Set bits 6 and 7 (main clock division ratio selection bits) of the CPU mode register.
- ④ After CPU rewrite mode control program is transferred to internal RAM, jump to this control program on RAM. (The following operations are controlled by this control program).
- ⑤ Apply 5  $V_{\pm 10}$  % to the CNV<sub>SS</sub>/V<sub>PP</sub> pin (in single-chip mode).
- ⑥ Set "1" to the CPU rewrite mode select bit (bit 1 of address 0FFE<sub>16</sub>).  
For this bit to be set to "1", the user needs to write "0" and then "1" to it in succession.
- ⑦ Read the CPU rewrite mode entry flag (bit 2 of address 0FFE<sub>16</sub>) to confirm that the CPU rewrite mode is set to "1".
- ⑧ Flash memory operations are executed by using software commands.

**Note:** The following procedures are also necessary.

- Control for data which is input from the external (serial I/O etc.) and to be programmed to the flash memory.
- Initial setting for ports, etc.
- Writing to the watchdog timer

##### [Release procedure]

- ① Execute the read command or set the flash memory reset bit (bit 3 of address 0FFE<sub>16</sub>).
- ② Set the CPU rewrite mode select bit (bit 0 of address 0FFE<sub>16</sub>) to "0".



Also, execute the following processing before the CPU reprogramming mode is selected so that interrupts will not occur during the CPU reprogramming mode.

- Set the interrupt disable flag (I) to "1"

When the watchdog timer has already started, write to the watchdog timer control register (address 1E16) periodically during the CPU reprogramming mode in order not to generate the reset by the underflow of the watchdog timer H.

During the program or erase execution, watchdog timer is automatically cleared. Accordingly, the internal reset by underflow does not occur.

When the interrupt request or reset occurs in the CPU reprogramming mode, the microcomputer enters the following state;

- Interrupt occurs

This may cause a program runaway because the read from the flash memory which has the interrupt vector area cannot be performed.

- Underflow of watchdog timer H, reset

This may cause a microcomputer reset; the built-in flash memory control circuit and the flash memory control register are reset. When reset state is released with CNVss = "H", CPU starts in the boot mode.

Also, when the above interrupt and reset occur during program/erase, error data may still exist after reset release because the reprogramming of the flash memory is not completed, so that reprogramming of the flash memory in the parallel I/O mode or serial I/O mode is required.

# APPLICATION

## 2.11 Flash memory mode

### 2.11.7 Flash memory mode application examples

The control pin processing example on the system board in the serial I/O mode and the control example in the CPU rewrite mode are described below.

#### (1) Control pin connection example on the system board in serial I/O mode

As shown in Figure 2.11.4, in the serial I/O mode, the built-in flash memory can be rewritten with the microcomputer mounted on board. Connection examples of control pins (P2<sub>4</sub>/RxD, P2<sub>5</sub>/TxD, P2<sub>6</sub>/SCLK1, P2<sub>7</sub>/S $\overline$ RDY1, P4<sub>1</sub>, CNV<sub>SS</sub>, and RESET pin) in the serial I/O mode are described below.

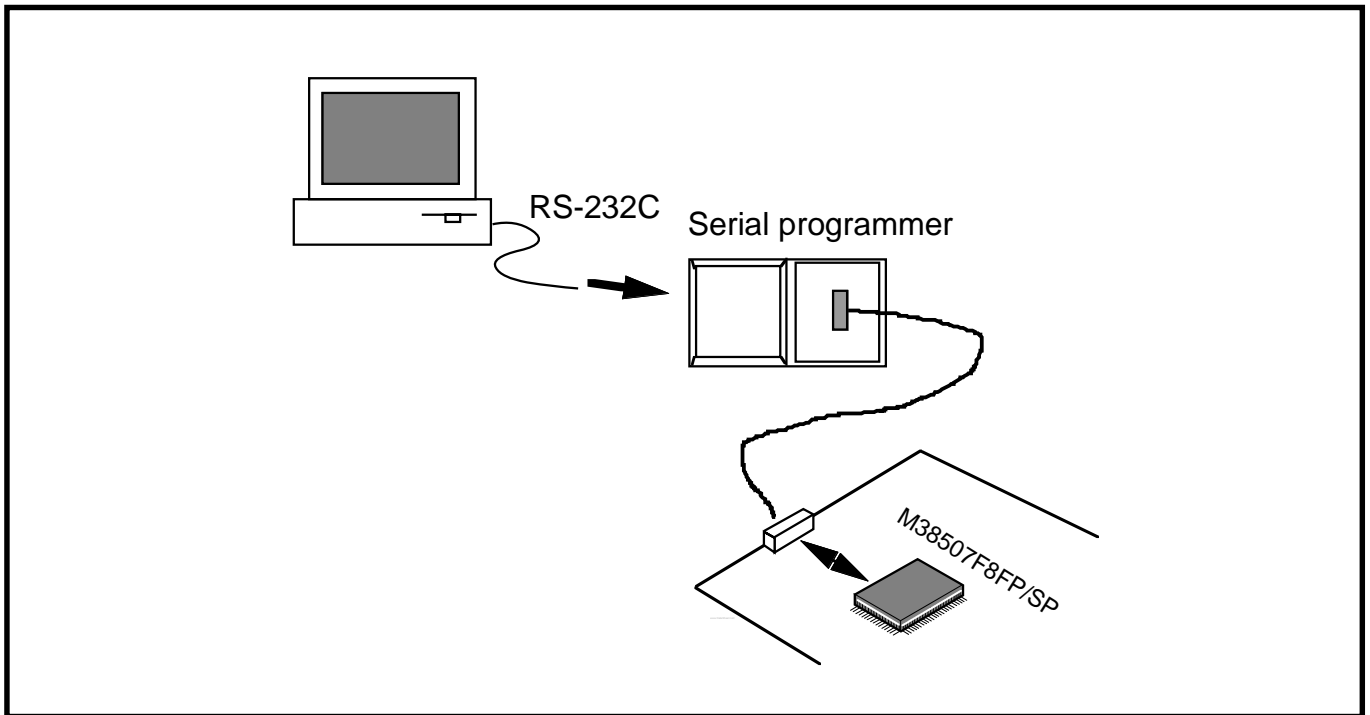


Fig. 2.11.4 Rewrite example of built-in flash memory in serial I/O mode

### ① When control signals are not affected to user system circuit

When the control signals in the serial I/O mode are not used or not affected to the user system circuit, they can be connected as shown in Figure 2.11.5.

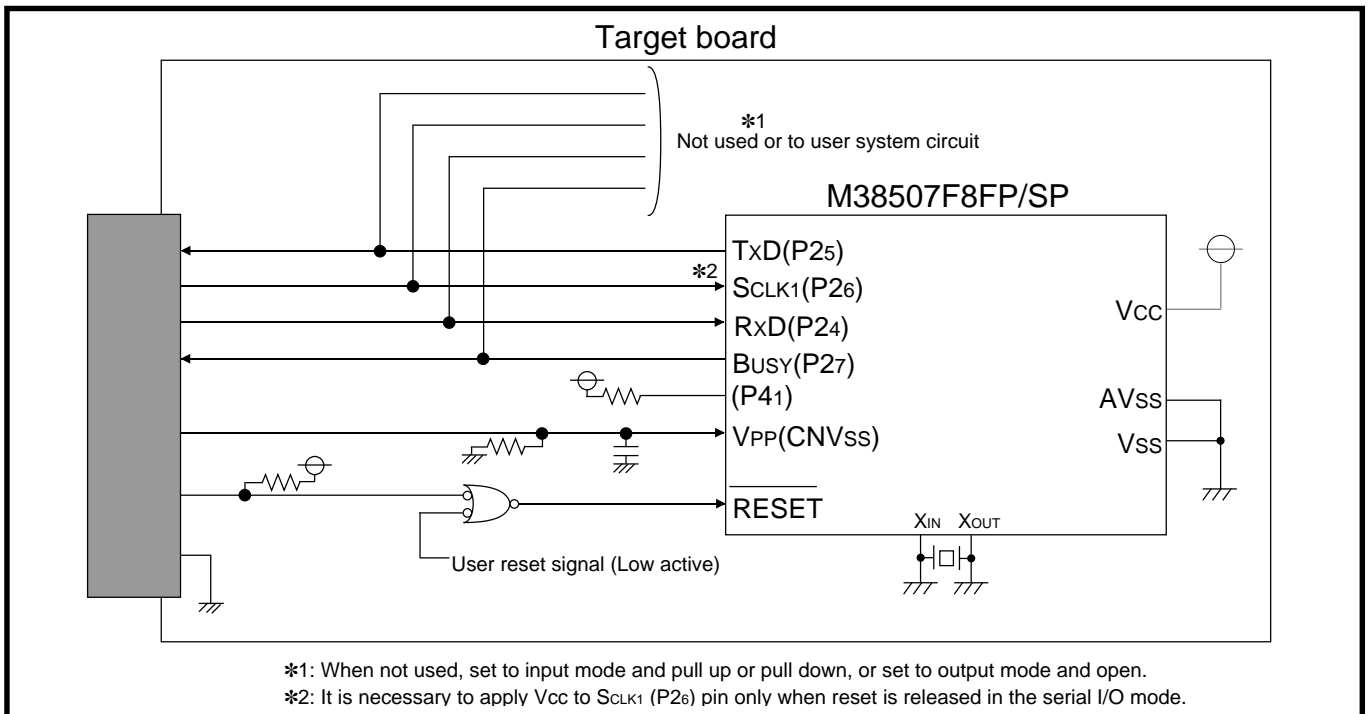


Fig. 2.11.5 Connection example in serial I/O mode (1)

### ② When control signals are affected to user system circuit-1

Figure 2.11.6 shows an example that the jumper switch cut-off the control signals not to supply to the user system circuit in the serial I/O mode.

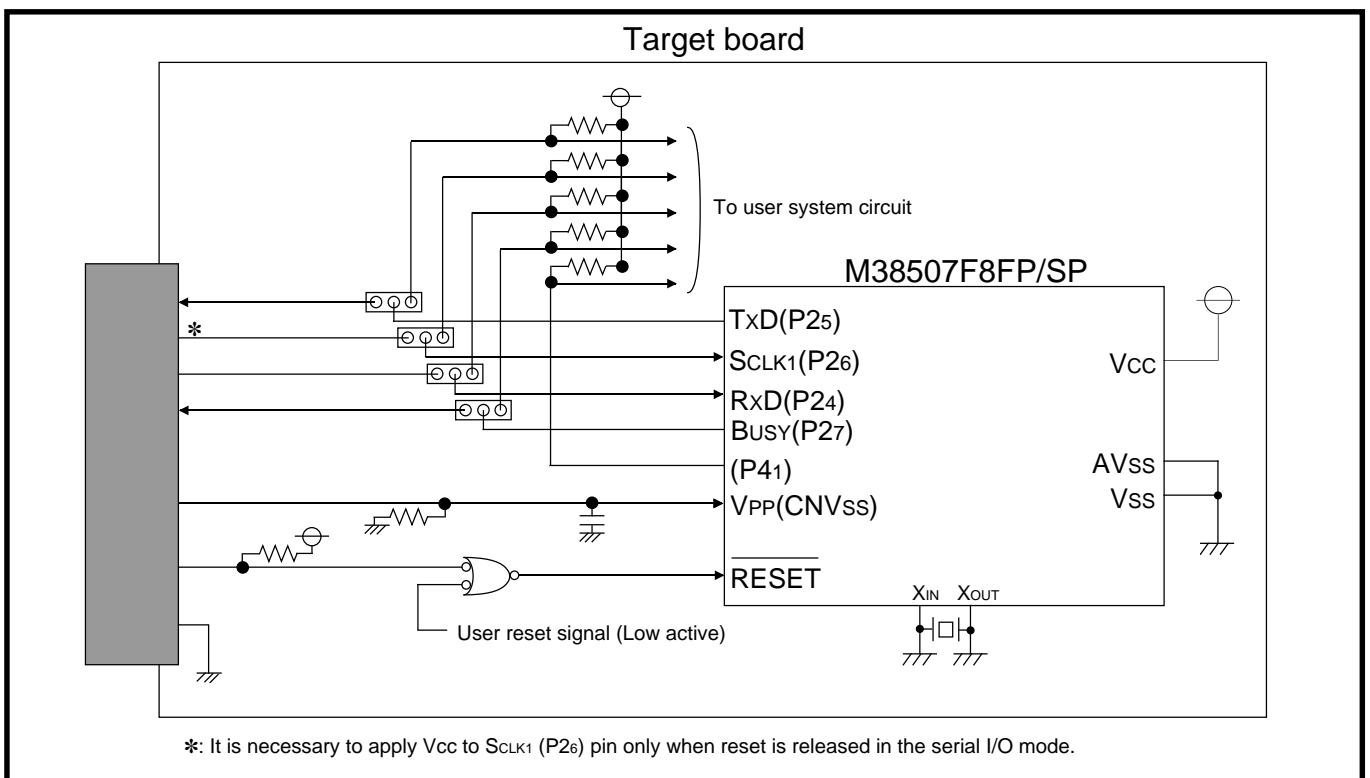


Fig. 2.11.6 Connection example in serial I/O mode (2)

# APPLICATION

## 2.11 Flash memory mode

### ③ When control signals are affected to user system circuit-2

Figure 2.11.7 shows an example that the analog switch (74HC4066) cut-off the control signals not to supply to the user system circuit in the serial I/O mode.

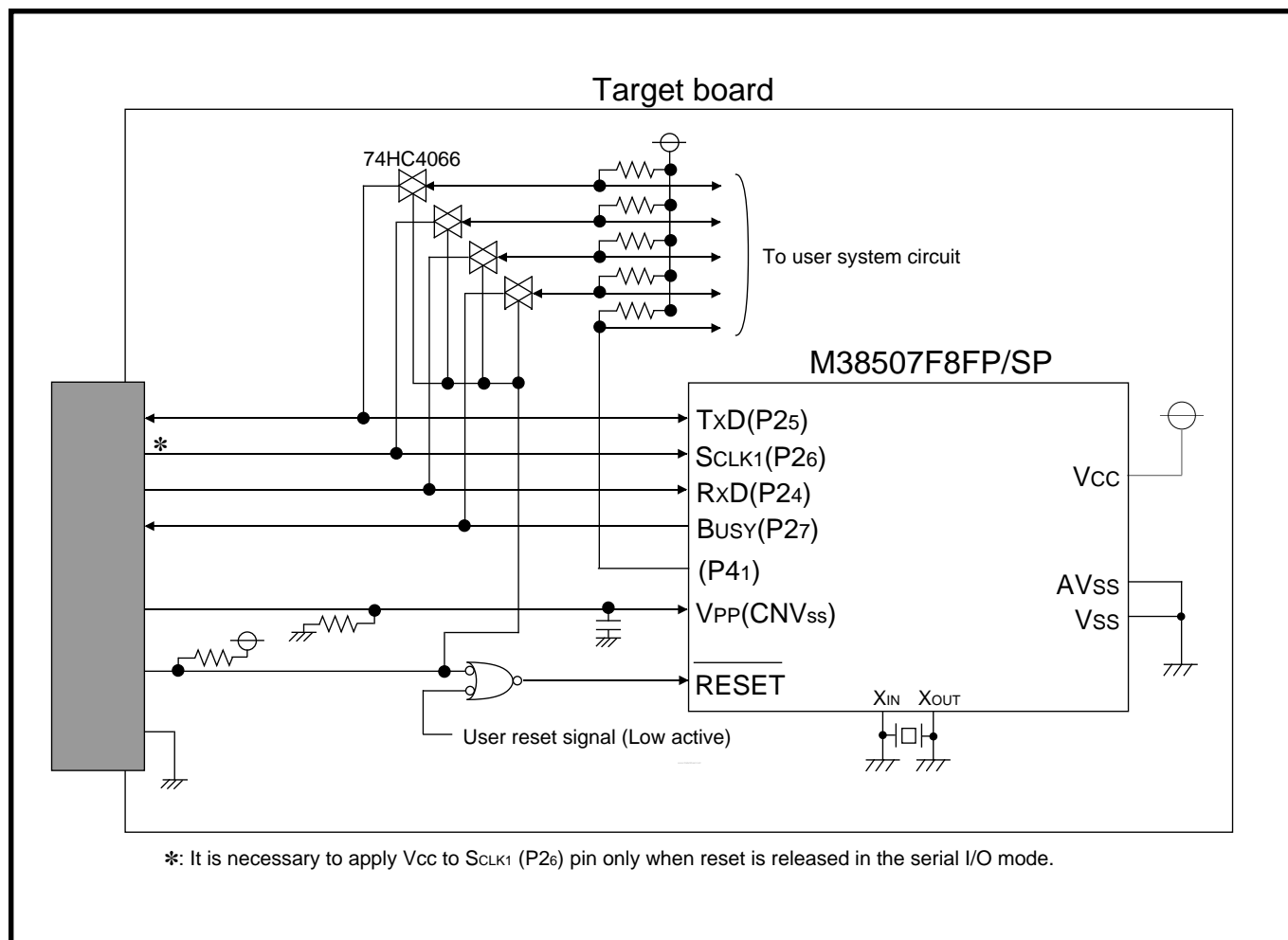


Fig. 2.11.7 Connection example in serial I/O mode (3)

### (2) Control example in CPU rewrite mode

In this example, data is received by using serial I/O, and the data is programmed to the built-in flash memory in the CPU rewrite mode.

Figure 2.11.8 shows an example of the reprogramming system for the built-in flash memory in the CPU rewrite mode. Figure 2.11.9 shows the CPU rewrite mode beginning/release flowchart.

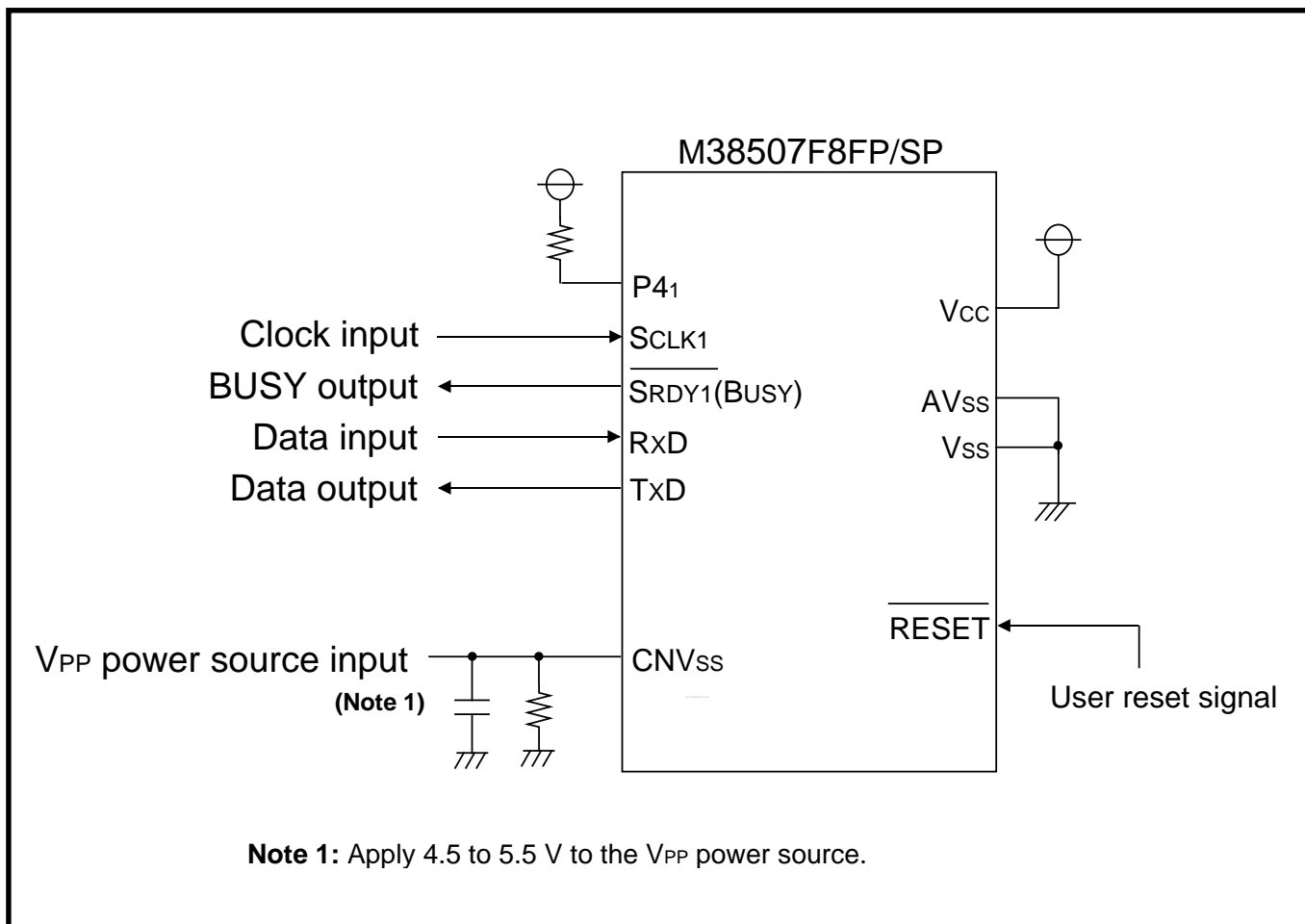
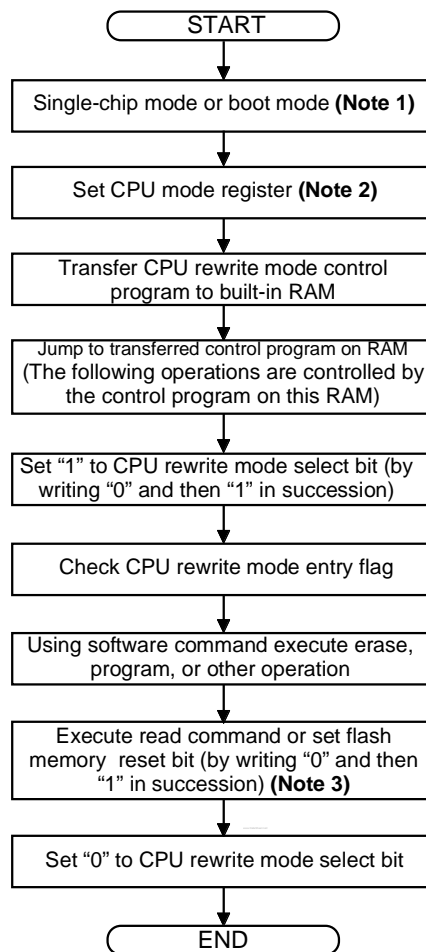


Fig. 2.11.8 Example of rewrite system for built-in flash memory in CPU rewrite mode

# APPLICATION

## 2.11 Flash memory mode



- Notes**
- 1:** When MCU starts in the single-chip mode, it is necessary to apply  $5\text{ V} \pm 10\%$  to the CNVss pin until confirming the CPU rewrite mode entry flag.
  - 2:** Set bits 6 and 7 (main clock division ratio selection bits) of the CPU mode register (address 003B<sub>16</sub>).
  - 3:** Before releasing the CPU rewrite mode after completing erase or program operation, always be sure to execute a read command or reset the flash memory.

Fig. 2.11.9 CPU rewrite mode beginning/release flowchart

### 2.11.8 Notes on CPU rewrite mode

**(1) Operation speed**

During CPU rewrite mode, set the internal clock  $\phi$  4 MHz or less using the main clock division ratio selection bits (bits 6 and 7 of address 003B<sub>16</sub>).

**(2) Instructions inhibited against use**

The instructions which refer to the internal data of the flash memory cannot be used during the CPU rewrite mode.

**(3) Interrupts inhibited against use**

The interrupts cannot be used during the CPU rewrite mode because they refer to the internal data of the flash memory.

**(4) Watchdog timer**

In case of the watchdog timer has been running already, the internal reset generated by watchdog timer underflow does not happen, because of watchdog timer is always clearing during program or erase operation.

**(5) Reset**

Reset is always valid. In case of CNV<sub>SS</sub> = "H" when reset is released, boot mode is active. So the program starts from the address contained in address FFFC<sub>16</sub> and FFFD<sub>16</sub> in boot ROM area.

# APPLICATION

## 2.11 Flash memory mode

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### *MEMORANDUM*





# CHAPTER 3

## **APPENDIX**

- 3.1 Electrical characteristics
- 3.2 Standard characteristics
- 3.3 Notes on use
- 3.4 Countermeasures against noise
- 3.5 List of registers
- 3.6 Package outline
- 3.7 Machine instructions
- 3.8 List of instruction code
- 3.9 SFR memory map
- 3.10 Pin configurations

# APPENDIX

## 3.1 Electrical characteristics

### 3.1 Electrical characteristics

#### 3.1.1 Absolute maximum ratings

Table 3.1.1 Absolute maximum ratings

Symbol	Parameter	Conditions	Ratings	Unit
V <sub>CC</sub>	Power source voltage		-0.3 to 6.5	V
V <sub>I</sub>	Input voltage P00–P07, P10–P17, P20, P21, P24–P27, P30–P34, P40–P44, V <sub>REF</sub>	All voltages are based on V <sub>SS</sub> . Output transistors are cut off.	-0.3 to V <sub>CC</sub> +0.3	V
V <sub>I</sub>	Input voltage P22, P23		-0.3 to 5.8	V
V <sub>I</sub>	Input voltage RESET, X <sub>IN</sub>		-0.3 to V <sub>CC</sub> +0.3	V
V <sub>I</sub>	Input voltage CNV <sub>SS</sub>		-0.3 to V <sub>CC</sub> +0.3	V
V <sub>O</sub>	Output voltage P00–P07, P10–P17, P20, P21, P24–P27, P30–P34, P40–P44, X <sub>OUT</sub>		-0.3 to V <sub>CC</sub> +0.3	V
V <sub>O</sub>	Output voltage P22, P23		-0.3 to 5.8	V
P <sub>d</sub>	Power dissipation	T <sub>a</sub> = 25 °C	1000 ( <b>Note</b> )	mW
T <sub>opr</sub>	Operating temperature		-20 to 85	°C
T <sub>stg</sub>	Storage temperature		-40 to 125	°C

**Note :** The rating becomes 300mW at the 42P2R-A/E package.

### 3.1.2 Recommended operating conditions

**Table 3.1.2 Recommended operating conditions (1)**

(VCC = 2.7 to 5.5 V, Ta = -20 to 85 °C, unless otherwise noted)

Symbol	Parameter		Limits			Unit
			Min.	Typ.	Max.	
VCC	Power source voltage	8 MHz (high-speed mode)	4.0	5.0	5.5	V
		8 MHz (middle-speed mode), 4 MHz (high-speed mode)	2.7	5.0	5.5	
VSS	Power source voltage			0		V
VREF	A-D convert reference voltage		2.0		VCC	V
AVSS	Analog power source voltage			0		V
VIA	Analog input voltage	AN0-AN4	AVSS		VCC	V
VIH	"H" input voltage	P00-P07, P10-P17, P20-P27, P30-P34, P40-P44	0.8VCC		VCC	V
VIH	"H" input voltage	RESET, XIN, CNVSS	0.8VCC		VCC	V
VIL	"L" input voltage	P00-P07, P10-P17, P20-P27, P30-P34, P40-P44	0		0.2VCC	V
VIL	"L" input voltage	RESET, CNVSS	0		0.2VCC	V
VIL	"L" input voltage	XIN	0		0.16VCC	V
ΣIOH(peak)	"H" total peak output current (Note)	P00-P07, P10-P17, P30-P34			-80	mA
ΣIOH(peak)	"H" total peak output current (Note)	P20, P21, P24-P27, P40-P44			-80	mA
ΣIOL(peak)	"L" total peak output current (Note)	P00-P07, P30-P34			80	mA
ΣIOL(peak)	"L" total peak output current (Note)	P10-P17			120	mA
ΣIOL(peak)	"L" total peak output current (Note)	P20-P27, P40-P44			80	mA
ΣIOH(avg)	"H" total average output current (Note)	P00-P07, P10-P17, P30-P34			-40	mA
ΣIOH(avg)	"H" total average output current (Note)	P20, P21, P24-P27, P40-P44			-40	mA
ΣIOL(avg)	"L" total average output current (Note)	P00-P07, P30-P34			40	mA
ΣIOL(avg)	"L" total average output current (Note)	P10-P17			60	mA
ΣIOL(avg)	"L" total average output current (Note)	P20-P27, P40-P44			40	mA

**Note :** The total output current is the sum of all the currents flowing through all the applicable ports. The total average current is an average value measured over 100 ms. The total peak current is the peak value of all the currents.

# APPENDIX

## 3.1 Electrical characteristics

**Table 3.1.3 Recommended operating conditions (2)**

(VCC = 2.7 to 5.5 V, Ta = -20 to 85 °C, unless otherwise noted)

Symbol	Parameter	Limits			Unit	
		Min.	Typ.	Max.		
IOH(peak)	"H" peak output current (Note 1)	P00–P07, P10–P17, P20, P21, P24–P27, P30–P34, P40–P44			-10	mA
IOL(peak)	"L" peak output current (Note 1)	P00–P07, P20–P27, P30–P34, P40–P44			10	mA
IOL(peak)	"L" peak output current (Note 1)	P10–P17			20	mA
IOH(avg)	"H" average output current (Note 2)	P00–P07, P10–P17, P20, P21, P24–P27, P30–P34, P40–P44			-5	mA
IOL(avg)	"L" average output current (Note 2)	P00–P07, P20–P27, P30–P34, P40–P44			5	mA
IOL(avg)	"L" average output current (Note 2)	P10–P17			15	mA
f(XIN)	Internal clock oscillation frequency (VCC = 4.0 to 5.5V) (Note 3)				8	MHz
f(XIN)	Internal clock oscillation frequency (VCC = 2.7 to 5.5V) (Note 3)				4	MHz

**Notes** 1: The peak output current is the peak current flowing in each port.

2: The average output current IOL(avg), IOH(avg) are average value measured over 100 ms.

3: When the oscillation frequency has a duty cycle of 50%.

### 3.1.3 Electrical characteristics

**Table 3.1.4 Electrical characteristics (1)**

(VCC = 2.7 to 5.5 V, VSS = 0 V, Ta = -20 to 85 °C, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min.	Typ.	Max.	
VOH	"H" output voltage P00–P07, P10–P17, P20, P21, P24–P27, P30–P34, P40–P44 (Note)	IOH = -10 mA VCC = 4.0–5.5 V	VCC-2.0			V
		IOH = -1.0 mA VCC = 2.7–5.5 V	VCC-1.0			V
VOL	"L" output voltage P00–P07, P20–P27, P30–P34, P40–P44	IOH = 10 mA VCC = 4.0–5.5 V			2.0	V
		IOH = 1.0 mA VCC = 2.7–5.5 V			1.0	V
VOL	"L" output voltage P10–P17	IOH = 20 mA VCC = 4.0–5.5 V			2.0	V
		IOH = 10 mA VCC = 2.7–5.5 V			1.0	V
VT+–VT-	Hysteresis CNTR0, CNTR1, INT0–INT3			0.4		V
VT+–VT-	Hysteresis RxD, SCLK1, SCLK2, SIN2			0.5		V
VT+–VT-	Hysteresis RESET			0.5		V
IiH	"H" input current P00–P07, P10–P17, P20, P21, P24–P27, P30–P34, P40–P44	Vi = VCC			5.0	μA
IiH	"H" input current RESET, CNVSS	Vi = VCC			5.0	μA
IiH	"H" input current XIN	Vi = VCC		4		μA
IiL	"L" input current P00–P07, P10–P17, P20–P27, P30–P34, P40–P44	Vi = VSS			-5.0	μA
IiL	"L" input current RESET, CNVSS	Vi = VSS			-5.0	μA
IiL	"L" input current XIN	Vi = VSS		-4		μA
VRAM	RAM hold voltage	When clock stopped	2.0		5.5	V

**Note:** P25 is measured when the P25/TxD P-channel output disable bit of the UART control register (bit 4 of address 001B16) is "0".

**Table 3.1.5 Electrical characteristics (2)**

(V<sub>CC</sub> = 2.7 to 5.5 V, V<sub>SS</sub> = 0 V, T<sub>a</sub> = -20 to 85 °C, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min.	Typ.	Max.	
I <sub>CC</sub>	Power source current	High-speed mode f(X <sub>IN</sub> ) = 8 MHz f(XC <sub>IN</sub> ) = 32.768 kHz Output transistors "off"		6.8	13	mA
		High-speed mode f(X <sub>IN</sub> ) = 8 MHz (in WIT state) f(XC <sub>IN</sub> ) = 32.768 kHz Output transistors "off"		1.6		mA
		Low-speed mode f(X <sub>IN</sub> ) = stopped f(XC <sub>IN</sub> ) = 32.768 kHz Output transistors "off"	Except M38507F8FP/SP	60	200	μA
			M38507F8FP/SP	250		μA
		Low-speed mode f(X <sub>IN</sub> ) = stopped f(XC <sub>IN</sub> ) = 32.768 kHz (in WIT state) Output transistors "off"	Except M38507F8FP/SP	20	40	μA
			M38507F8FP/SP	70		μA
		Low-speed mode (V <sub>CC</sub> = 3 V) f(X <sub>IN</sub> ) = stopped f(XC <sub>IN</sub> ) = 32.768 kHz Output transistors "off"	Except M38507F8FP/SP	20	55	μA
			M38507F8FP/SP	150		μA
		Low-speed mode (V <sub>CC</sub> = 3 V) f(X <sub>IN</sub> ) = stopped f(XC <sub>IN</sub> ) = 32.768 kHz (in WIT state) Output transistors "off"	Except M38507F8FP/SP	5.0	10.0	μA
			M38507F8FP/SP	20		μA
		Middle-speed mode f(X <sub>IN</sub> ) = 8 MHz f(XC <sub>IN</sub> ) = stopped Output transistors "off"		4.0	7.0	mA
		Middle-speed mode f(X <sub>IN</sub> ) = 8 MHz (in WIT state) f(XC <sub>IN</sub> ) = stopped Output transistors "off"	—	1.5		mA
Increment when A-D conversion is executed f(X <sub>IN</sub> ) = 8 MHz		800		μA		
All oscillation stopped (in STP state) Output transistors "off"	T <sub>a</sub> = 25 °C	0.1	1.0	μA		
	T <sub>a</sub> = 85 °C		10	μA		

# APPENDIX

## 3.1 Electrical characteristics

### 3.1.4 A-D converter characteristics

Table 3.1.6 A-D converter characteristics

(V<sub>CC</sub> = 2.7 to 5.5 V, V<sub>SS</sub> = AV<sub>SS</sub> = 0 V, T<sub>a</sub> = -20 to 85 °C, f(X<sub>IN</sub>) = 8 MHz, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit	
			Min.	Typ.	Max.		
–	Resolution				10	bit	
–	Absolute accuracy (excluding quantization error)				±4	LSB	
t <sub>CONV</sub>	Conversion time	High-speed mode, Middle-speed mode			61	2t <sub>c</sub> (X <sub>IN</sub> )	
		Low-speed mode		40		μs	
RLADDER	Ladder resistor			35		kΩ	
I <sub>VREF</sub>	Reference power source input current	V <sub>REF</sub> "on"	V <sub>REF</sub> = 5.0 V	50	150	200	μA
		V <sub>REF</sub> "off"				5.0	
I <sub>I(AD)</sub>	A-D port input current			0.5	5.0	μA	

### 3.1.5 Timing requirements and switching characteristics

**Table 3.1.7 Timing requirements (1)**

(V<sub>CC</sub> = 4.0 to 5.5 V, V<sub>SS</sub> = 0 V, T<sub>a</sub> = -20 to 85 °C, unless otherwise noted)

Symbol	Parameter	Limits			Unit
		Min.	Typ.	Max.	
t <sub>w</sub> (RESET)	Reset input "L" pulse width	20			X <sub>IN</sub> cycle
t <sub>c</sub> (X <sub>IN</sub> )	External clock input cycle time	125			ns
t <sub>WH</sub> (X <sub>IN</sub> )	External clock input "H" pulse width	50			ns
t <sub>WL</sub> (X <sub>IN</sub> )	External clock input "L" pulse width	50			ns
t <sub>c</sub> (CNTR)	CNTR <sub>0</sub> , CNTR <sub>1</sub> input cycle time	200			ns
t <sub>WH</sub> (CNTR)	CNTR <sub>0</sub> , CNTR <sub>1</sub> input "H" pulse width	80			ns
t <sub>WL</sub> (CNTR)	CNTR <sub>0</sub> , CNTR <sub>1</sub> input "L" pulse width	80			ns
t <sub>WH</sub> (INT)	INT <sub>0</sub> to INT <sub>3</sub> input "H" pulse width	80			ns
t <sub>WL</sub> (INT)	INT <sub>0</sub> to INT <sub>3</sub> input "L" pulse width	80			ns
t <sub>c</sub> (SCLK1)	Serial I/O1 clock input cycle time <b>(Note)</b>	800			ns
t <sub>WH</sub> (SCLK1)	Serial I/O1 clock input "H" pulse width <b>(Note)</b>	370			ns
t <sub>WL</sub> (SCLK1)	Serial I/O1 clock input "L" pulse width <b>(Note)</b>	370			ns
t <sub>su</sub> (RxD-SCLK1)	Serial I/O1 input setup time	220			ns
t <sub>h</sub> (SCLK1-RxD)	Serial I/O1 input hold time	100			ns
t <sub>c</sub> (SCLK2)	Serial I/O2 clock input cycle time	1000			ns
t <sub>WH</sub> (SCLK2)	Serial I/O2 clock input "H" pulse width	400			ns
t <sub>WL</sub> (SCLK2)	Serial I/O2 clock input "L" pulse width	400			ns
t <sub>su</sub> (SIN2-SCLK2)	Serial I/O2 clock input setup time	200			ns
t <sub>h</sub> (SCLK2-SIN2)	Serial I/O2 clock input hold time	200			ns

**Note :** When f(X<sub>IN</sub>) = 8 MHz and bit 6 of address 001A16 is "1" (clock synchronous).  
Divide this value by four when f(X<sub>IN</sub>) = 8 MHz and bit 6 of address 001A16 is "0" (UART).

**Table 3.1.8 Timing requirements (2)**

(V<sub>CC</sub> = 2.7 to 5.5 V, V<sub>SS</sub> = 0 V, T<sub>a</sub> = -20 to 85 °C, unless otherwise noted)

Symbol	Parameter	Limits			Unit
		Min.	Typ.	Max.	
t <sub>w</sub> (RESET)	Reset input "L" pulse width	20			X <sub>IN</sub> cycle
t <sub>c</sub> (X <sub>IN</sub> )	External clock input cycle time	250			ns
t <sub>WH</sub> (X <sub>IN</sub> )	External clock input "H" pulse width	100			ns
t <sub>WL</sub> (X <sub>IN</sub> )	External clock input "L" pulse width	100			ns
t <sub>c</sub> (CNTR)	CNTR <sub>0</sub> , CNTR <sub>1</sub> input cycle time	500			ns
t <sub>WH</sub> (CNTR)	CNTR <sub>0</sub> , CNTR <sub>1</sub> input "H" pulse width	230			ns
t <sub>WL</sub> (CNTR)	CNTR <sub>0</sub> , CNTR <sub>1</sub> input "L" pulse width	230			ns
t <sub>WH</sub> (INT)	INT <sub>0</sub> to INT <sub>3</sub> input "H" pulse width	230			ns
t <sub>WL</sub> (INT)	INT <sub>0</sub> to INT <sub>3</sub> input "L" pulse width	230			ns
t <sub>c</sub> (SCLK1)	Serial I/O1 clock input cycle time <b>(Note)</b>	2000			ns
t <sub>WH</sub> (SCLK1)	Serial I/O1 clock input "H" pulse width <b>(Note)</b>	950			ns
t <sub>WL</sub> (SCLK1)	Serial I/O1 clock input "L" pulse width <b>(Note)</b>	950			ns
t <sub>su</sub> (RxD-SCLK1)	Serial I/O1 input setup time	400			ns
t <sub>h</sub> (SCLK1-RxD)	Serial I/O1 input hold time	200			ns
t <sub>c</sub> (SCLK2)	Serial I/O2 clock input cycle time	2000			ns
t <sub>WH</sub> (SCLK2)	Serial I/O2 clock input "H" pulse width	950			ns
t <sub>WL</sub> (SCLK2)	Serial I/O2 clock input "L" pulse width	950			ns
t <sub>su</sub> (SIN2-SCLK2)	Serial I/O2 clock input setup time	400			ns
t <sub>h</sub> (SCLK2-SIN2)	Serial I/O2 clock input hold time	300			ns

**Note :** When f(X<sub>IN</sub>) = 4 MHz and bit 6 of address 001A16 is "1" (clock synchronous).  
Divide this value by four when f(X<sub>IN</sub>) = 4 MHz and bit 6 of address 001A16 is "0" (UART).

# APPENDIX

## 3.1 Electrical characteristics

**Table 3.1.9 Switching characteristics (1)**

(V<sub>CC</sub> = 4.0 to 5.5 V, V<sub>SS</sub> = 0 V, T<sub>a</sub> = -20 to 85 °C, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min.	Typ.	Max.	
t <sub>WH</sub> (SCLK1)	Serial I/O1 clock output "H" pulse width	Fig.3.1.1	tc(SCLK1)/2-30			ns
t <sub>WL</sub> (SCLK1)	Serial I/O1 clock output "L" pulse width		tc(SCLK1)/2-30			ns
t <sub>d</sub> (SCLK1-TxD)	Serial I/O1 output delay time ( <b>Note 1</b> )				140	ns
t <sub>v</sub> (SCLK1-TxD)	Serial I/O1 output valid time ( <b>Note 1</b> )		-30			ns
t <sub>r</sub> (SCLK1)	Serial I/O1 clock output rising time				30	ns
t <sub>f</sub> (SCLK1)	Serial I/O1 clock output falling time				30	ns
t <sub>WH</sub> (SCLK2)	Serial I/O2 clock output "H" pulse width		tc(SCLK2)/2-160			ns
t <sub>WL</sub> (SCLK2)	Serial I/O2 clock output "L" pulse width		tc(SCLK2)/2-160			ns
t <sub>d</sub> (SCLK2-SOUT2)	Serial I/O2 output delay time ( <b>Note 2</b> )				200	ns
t <sub>v</sub> (SCLK2-SOUT2)	Serial I/O2 output valid time ( <b>Note 2</b> )		0			ns
t <sub>f</sub> (SCLK2)	Serial I/O2 clock output falling time				30	ns
t <sub>r</sub> (CMOS)	CMOS output rising time ( <b>Note 3</b> )			10	30	ns
t <sub>f</sub> (CMOS)	CMOS output falling time ( <b>Note 3</b> )			10	30	ns

**Notes** 1: When the P25/TxD P-channel output disable bit of the UART control register (bit 4 of address 001B16) is "0".

2: When the P01/SOUT2 and P02/SCLK2 P-channel output disable bit of the Serial I/O2 control register 1 (bit 7 of address 001516) is "0".

3: The XOUT pin is excluded.

**Table 3.1.10 Switching characteristics (2)**

(V<sub>CC</sub> = 2.7 to 5.5 V, V<sub>SS</sub> = 0 V, T<sub>a</sub> = -20 to 85 °C, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min.	Typ.	Max.	
t <sub>WH</sub> (SCLK1)	Serial I/O1 clock output "H" pulse width	Fig.3.1.1	tc(SCLK1)/2-50			ns
t <sub>WL</sub> (SCLK1)	Serial I/O1 clock output "L" pulse width		tc(SCLK1)/2-50			ns
t <sub>d</sub> (SCLK1-TxD)	Serial I/O1 output delay time ( <b>Note 1</b> )				350	ns
t <sub>v</sub> (SCLK1-TxD)	Serial I/O1 output valid time ( <b>Note 1</b> )		-30			ns
t <sub>r</sub> (SCLK1)	Serial I/O1 clock output rising time				50	ns
t <sub>f</sub> (SCLK1)	Serial I/O1 clock output falling time				50	ns
t <sub>WH</sub> (SCLK2)	Serial I/O2 clock output "H" pulse width		tc(SCLK2)/2-240			ns
t <sub>WL</sub> (SCLK2)	Serial I/O2 clock output "L" pulse width		tc(SCLK2)/2-240			ns
t <sub>d</sub> (SCLK2-SOUT2)	Serial I/O2 output delay time ( <b>Note 2</b> )				400	ns
t <sub>v</sub> (SCLK2-SOUT2)	Serial I/O2 output valid time ( <b>Note 2</b> )		0			ns
t <sub>f</sub> (SCLK2)	Serial I/O2 clock output falling time				50	ns
t <sub>r</sub> (CMOS)	CMOS output rising time ( <b>Note 3</b> )			20	50	ns
t <sub>f</sub> (CMOS)	CMOS output falling time ( <b>Note 3</b> )			20	50	ns

**Notes** 1: When the P25/TxD P-channel output disable bit of the UART control register (bit 4 of address 001B16) is "0".

2: When the P01/SOUT2 and P02/SCLK2 P-channel output disable bit of the Serial I/O2 control register 1 (bit 7 of address 001516) is "0".

3: The XOUT pin is excluded.



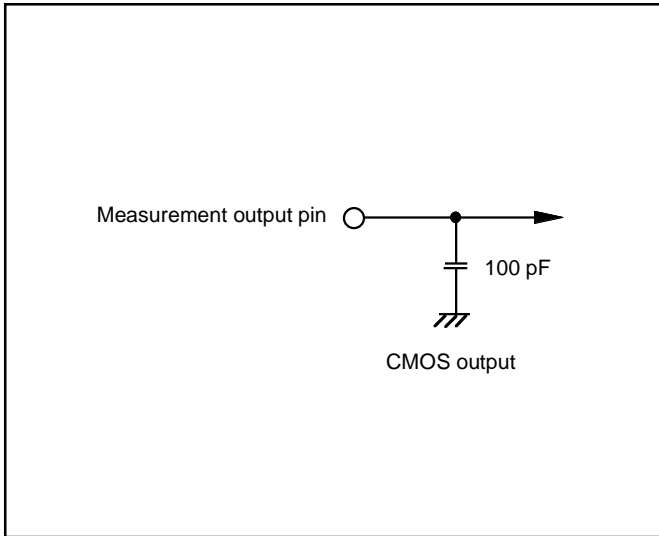


Fig. 3.1.1 Circuit for measuring output switching characteristics

# APPENDIX

## 3.1 Electrical characteristics

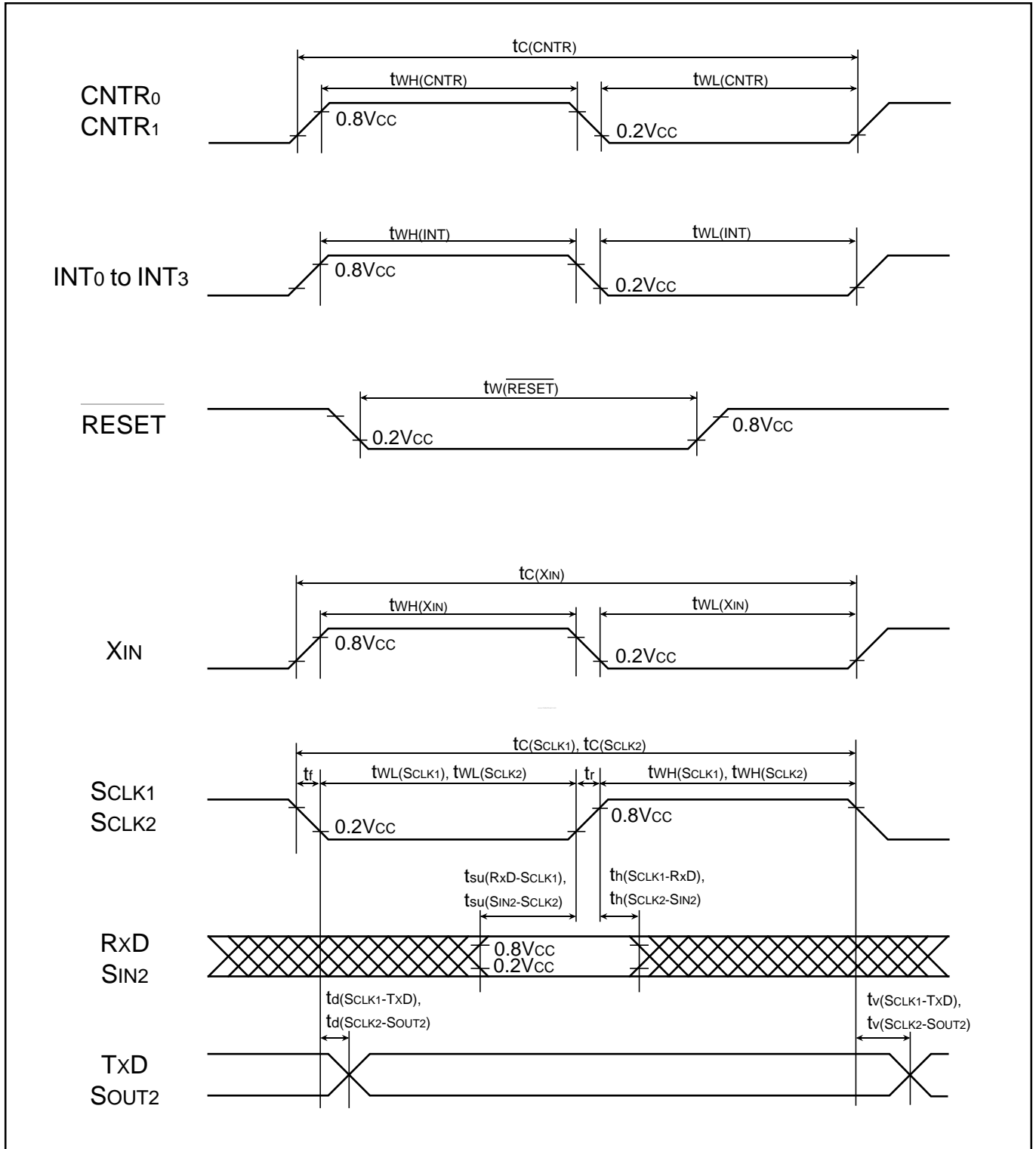


Fig. 3.1.2 Timing diagram

### 3.2 Standard characteristics

Standard characteristics described below are just examples of the 3850 Group (spec. H)'s characteristics and are not guaranteed. For rated values, refer to “3.1 Electrical characteristics”.

#### 3.2.1 Flash memory version power source current standard characteristics

Figure 3.2.1, Figure 3.2.2, Figure 3.2.3, Figure 3.2.4, and Figure 3.2.5 show flash memory version (M38507F8) power source current standard characteristics.

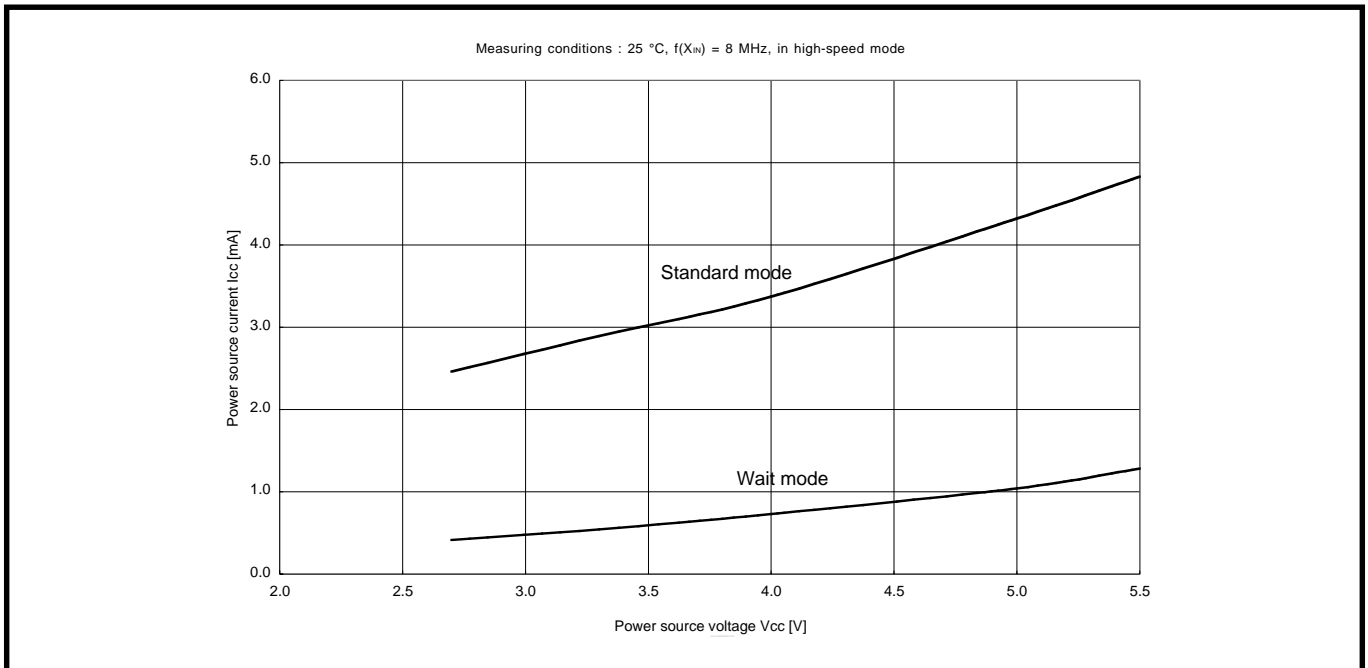


Fig. 3.2.1 Flash memory version power source current standard characteristics (in high-speed mode,  $f(X_{IN}) = 8 \text{ MHz}$ )

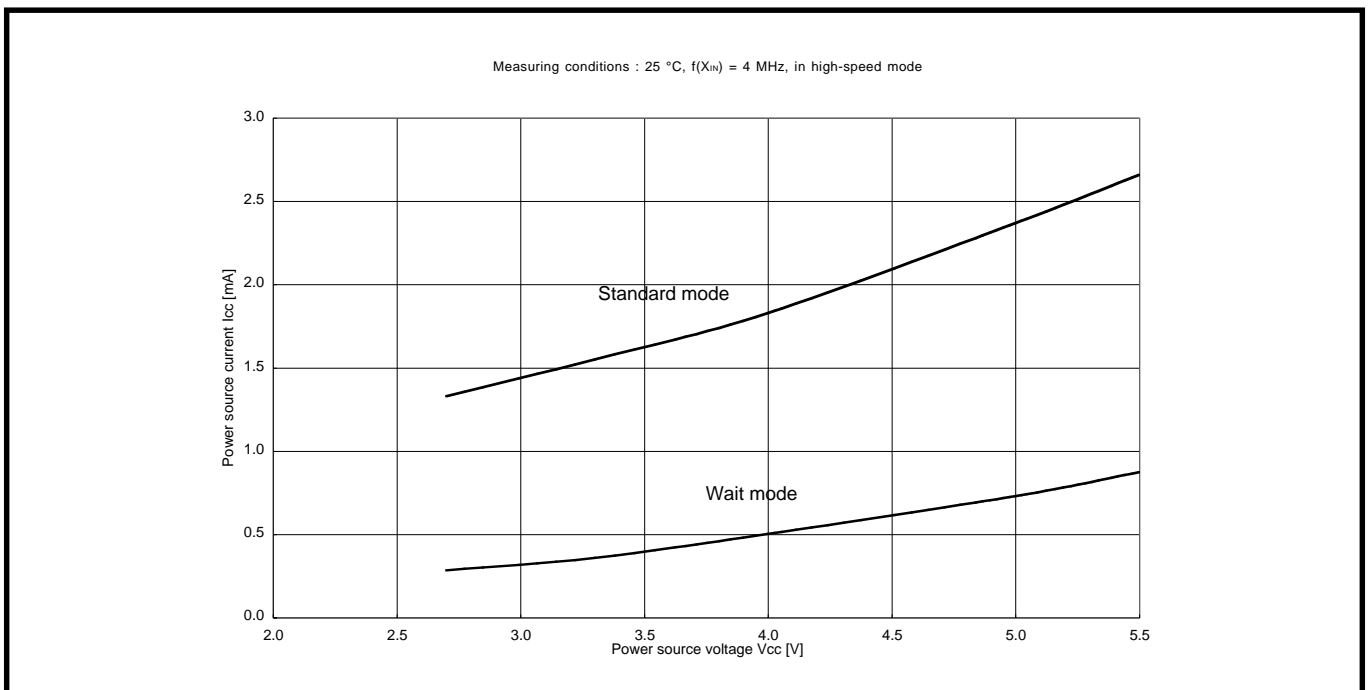


Fig. 3.2.2 Flash memory version power source current standard characteristics (in high-speed mode,  $f(X_{IN}) = 4 \text{ MHz}$ )

# APPENDIX

## 3.2 Standard characteristics

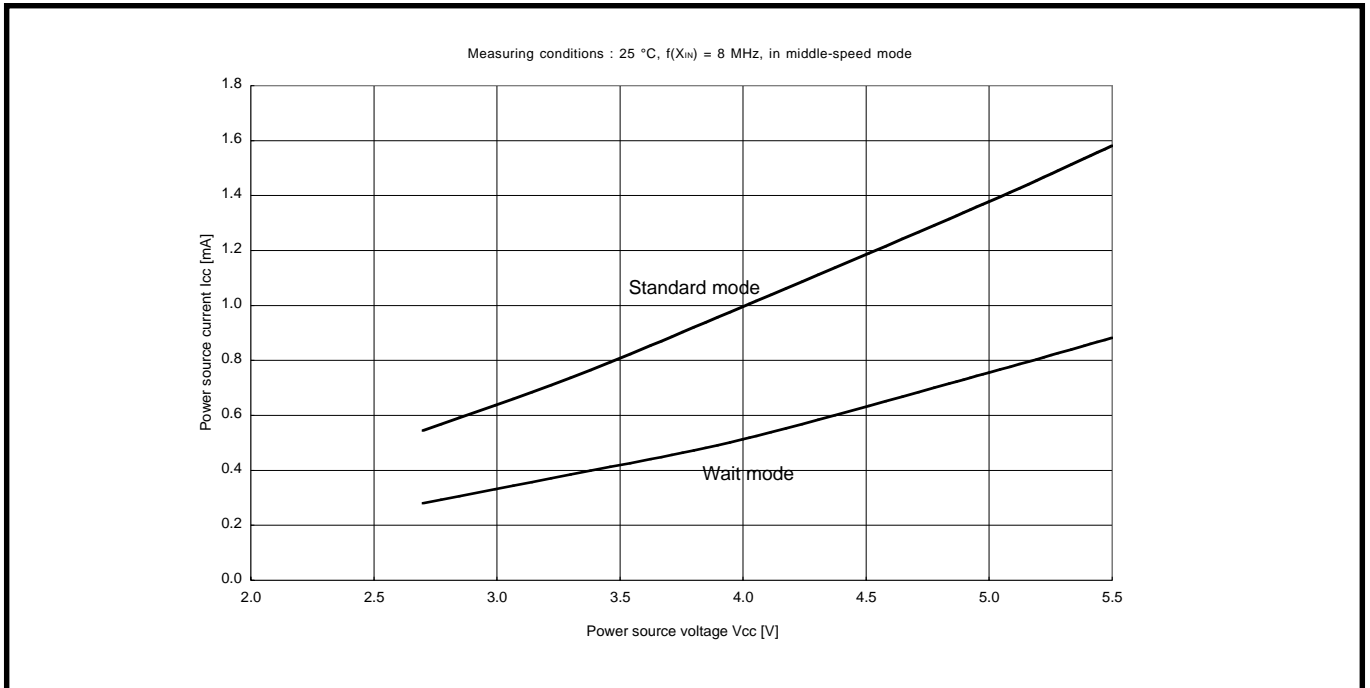


Fig. 3.2.3 Flash memory version power source current standard characteristics (in middle-speed mode,  $f(X_{IN}) = 8 \text{ MHz}$ )

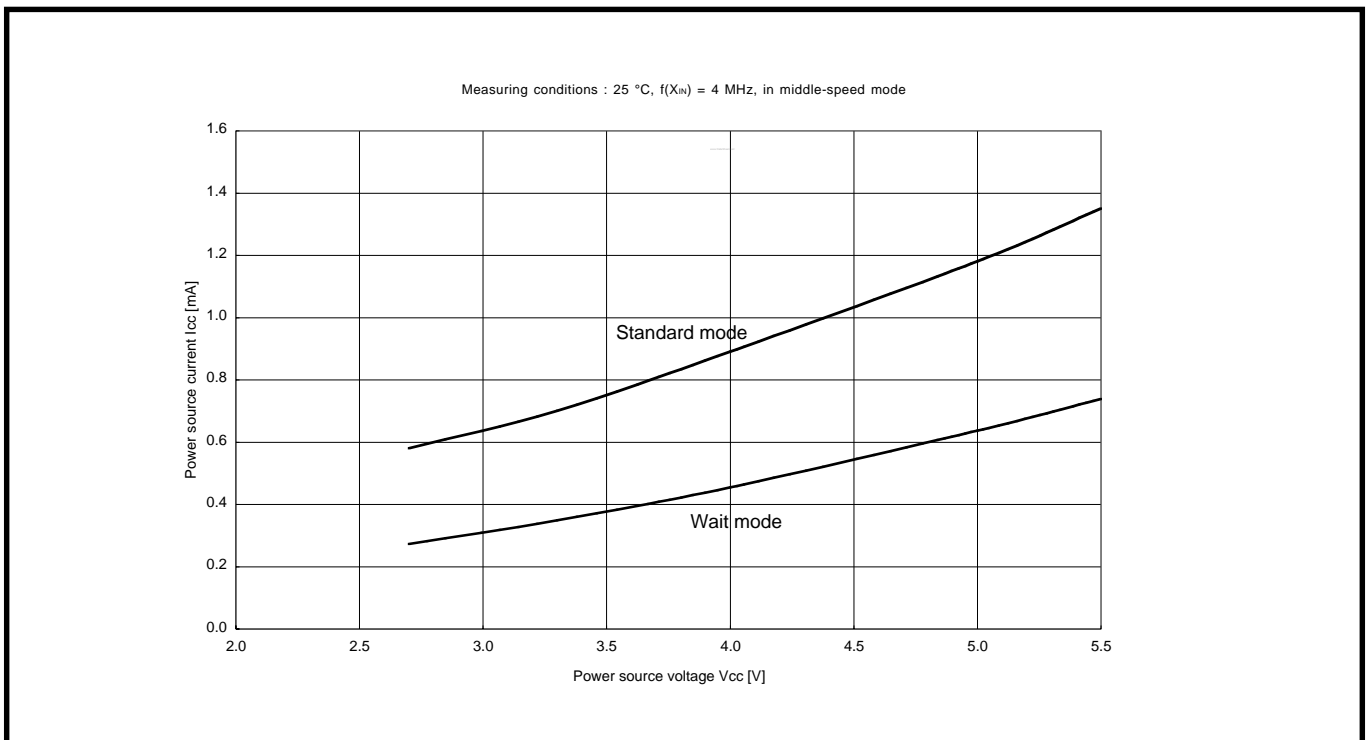


Fig. 3.2.4 Flash memory version power source current standard characteristics (in middle-speed mode,  $f(X_{IN}) = 4 \text{ MHz}$ )

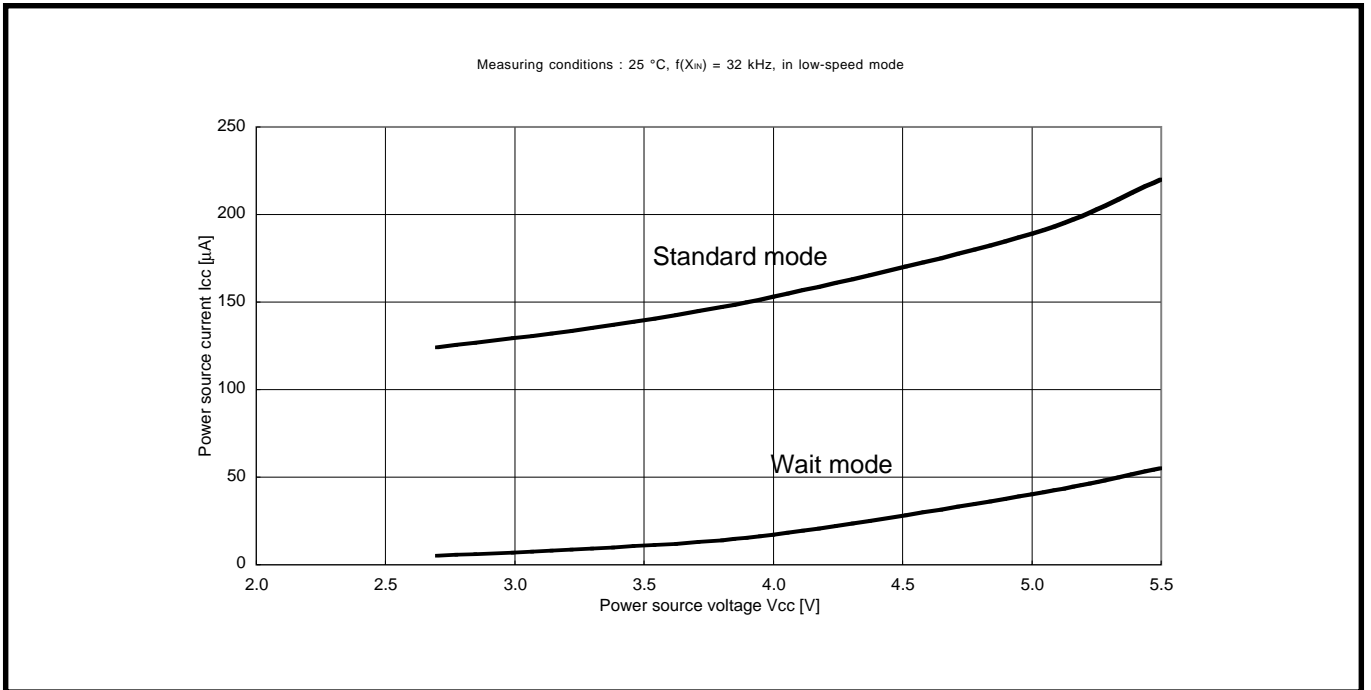


Fig. 3.2.5 Flash memory version power source current standard characteristics (in low-speed mode)

# APPENDIX

## 3.2 Standard characteristics

### 3.2.2 Mask ROM version power source current standard characteristics

Figure 3.2.6, Figure 3.2.7, Figure 3.2.8, Figure 3.2.9 and Figure 3.2.10 show mask ROM version (M38503M2H, M38503M4H, M38504M6, M38507M8) power source current standard characteristics.

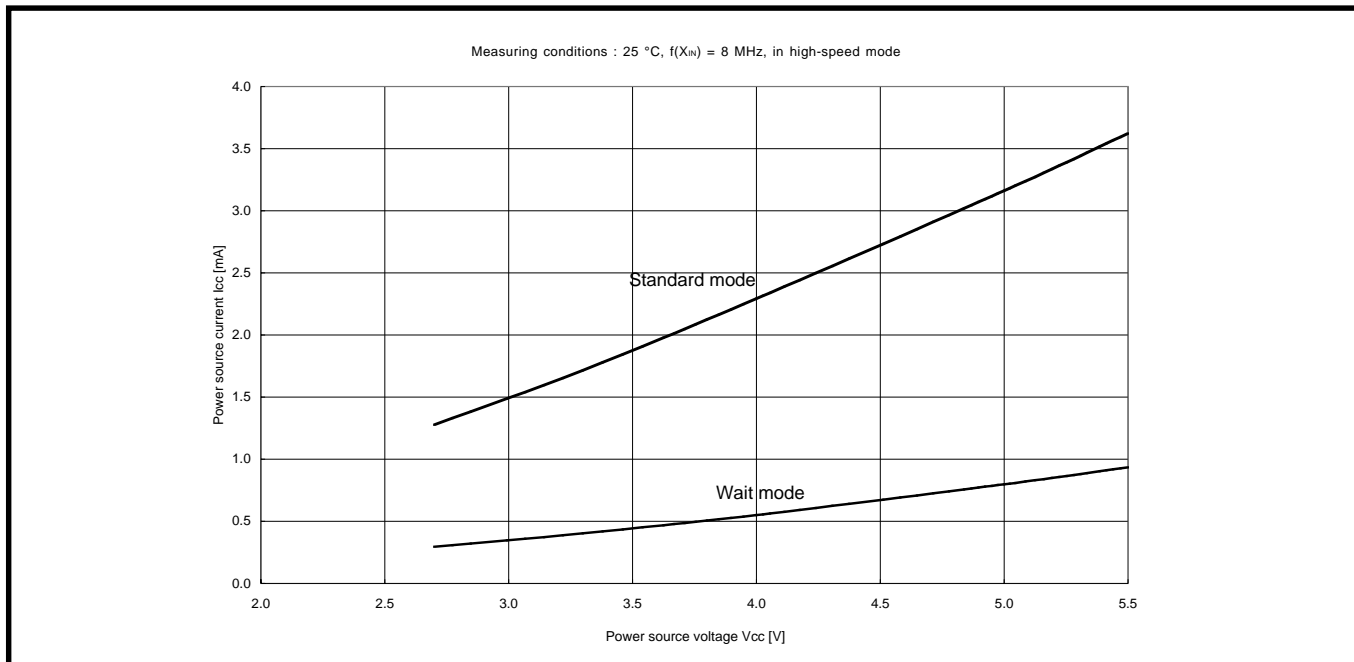


Fig. 3.2.6 Mask ROM version power source current standard characteristics (in high-speed mode,  $f(X_{IN}) = 8$  MHz)

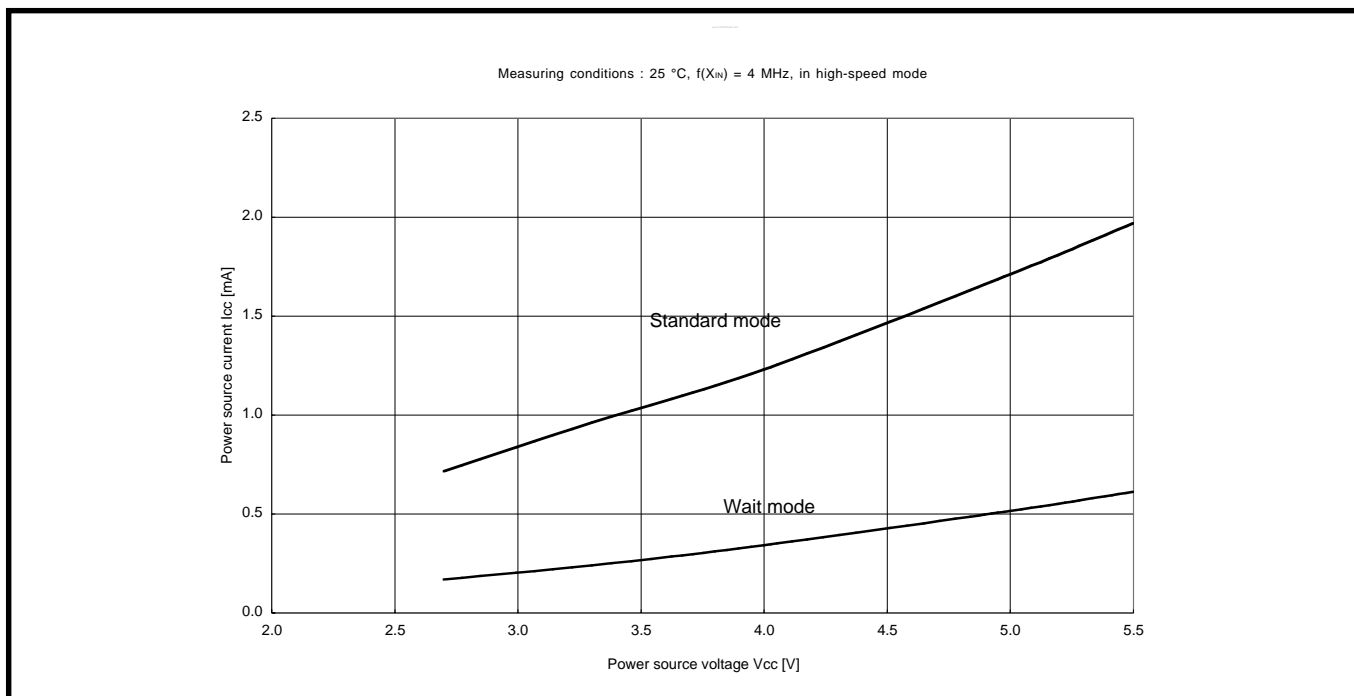
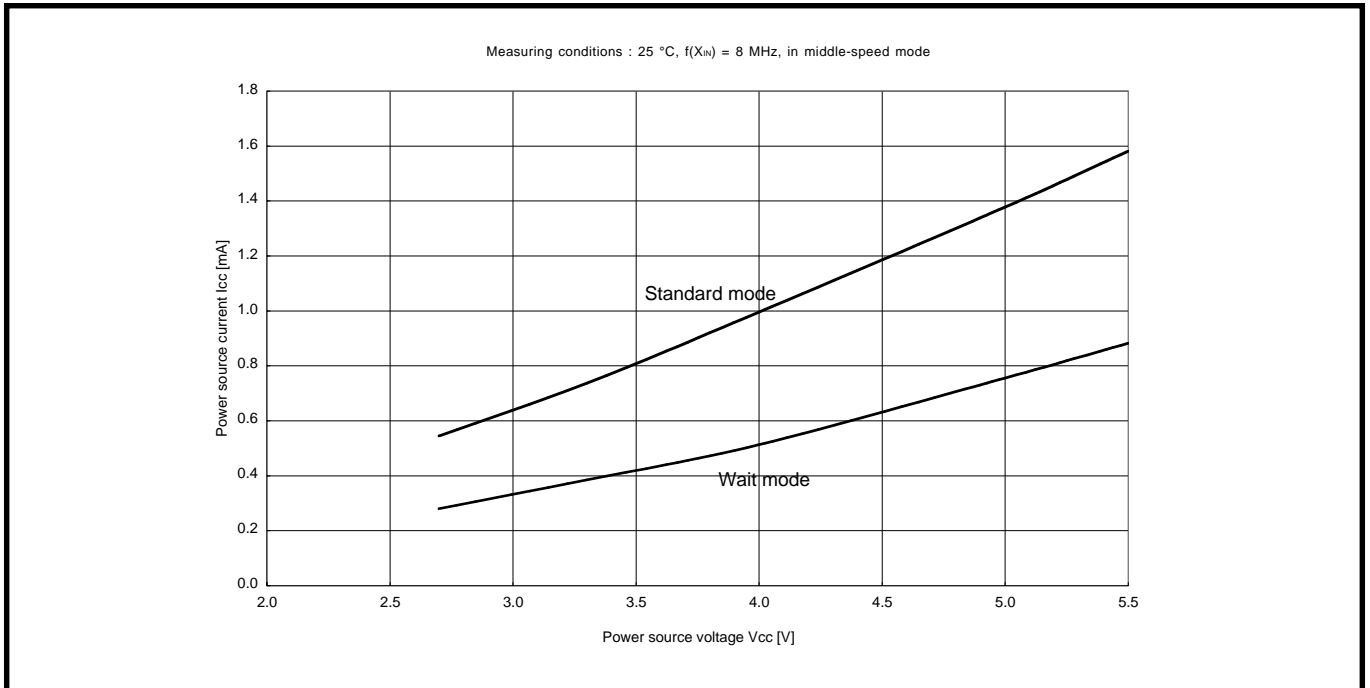
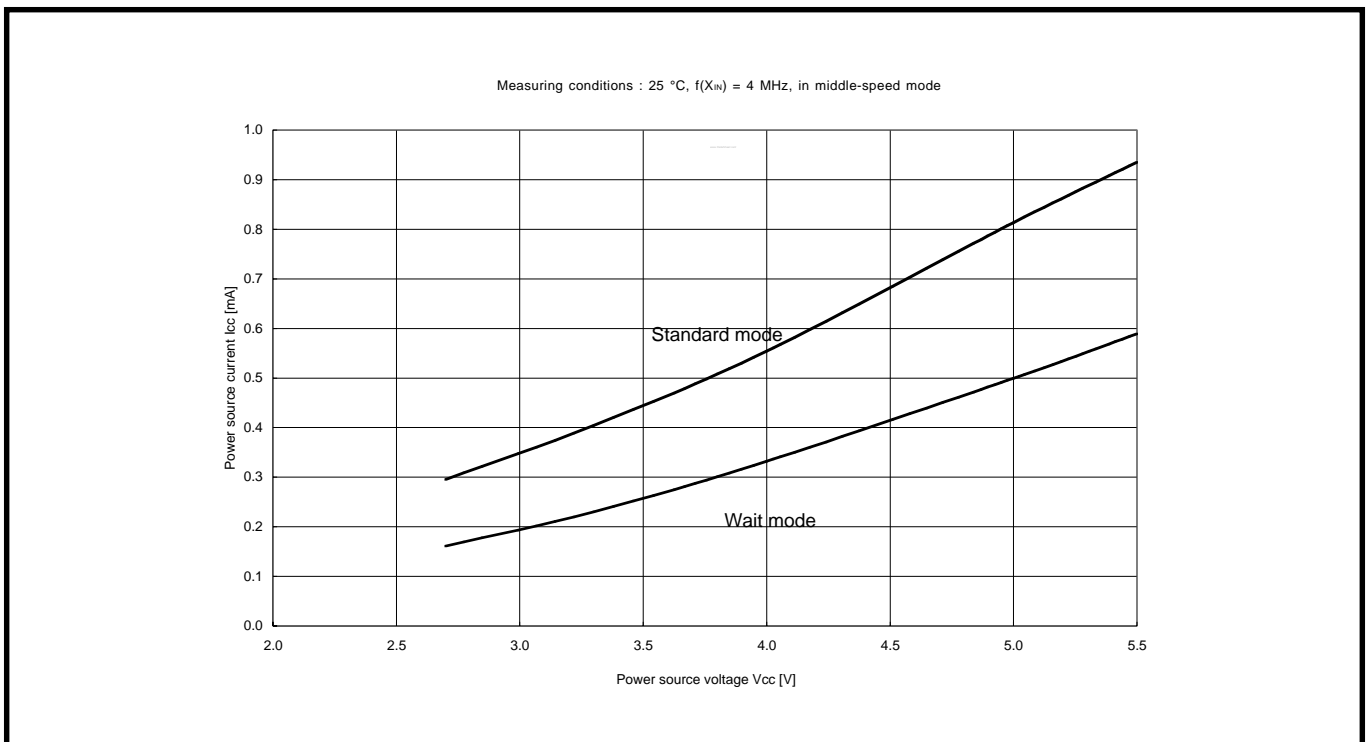


Fig. 3.2.7 Mask ROM version power source current standard characteristics (in high-speed mode,  $f(X_{IN}) = 4$  MHz)



**Fig. 3.2.8 Mask ROM version power source current standard characteristics (in middle-speed mode,  $f(X_{IN}) = 8$  MHz)**



**Fig. 3.2.9 Mask ROM version power source current standard characteristics (in middle-speed mode,  $f(X_{IN}) = 4$  MHz)**

# APPENDIX

## 3.2 Standard characteristics

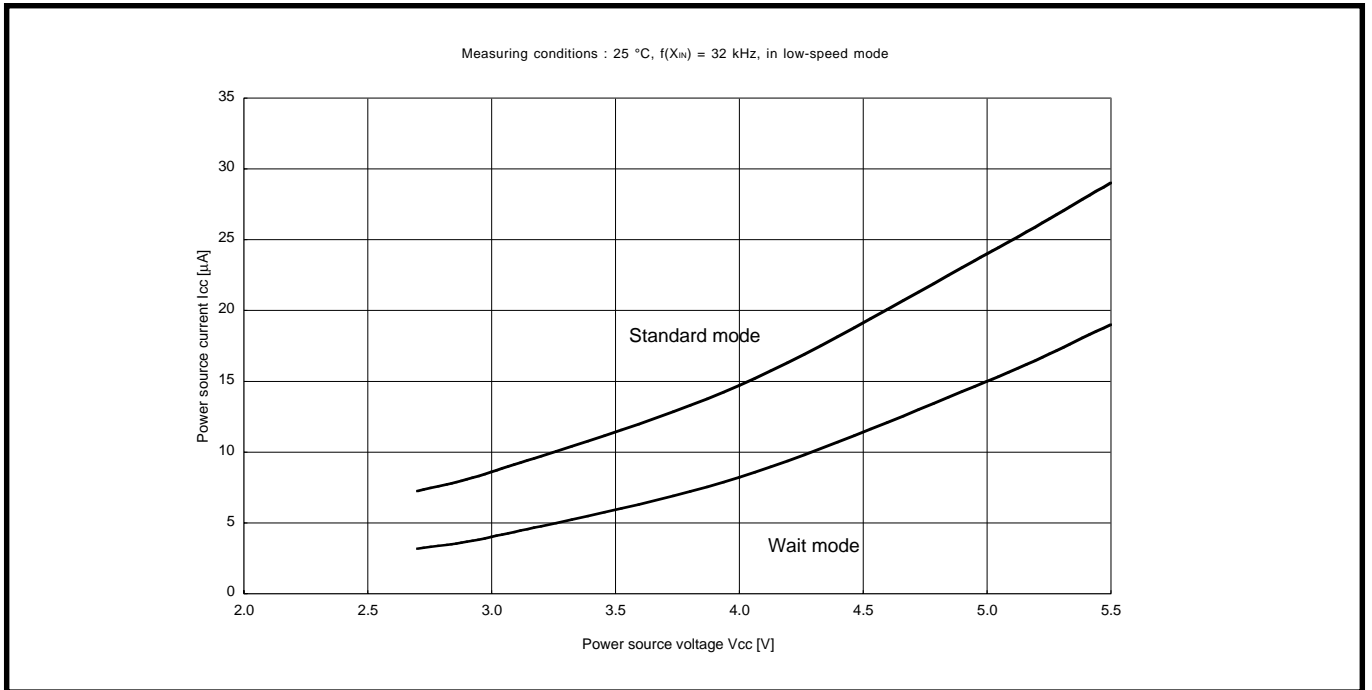


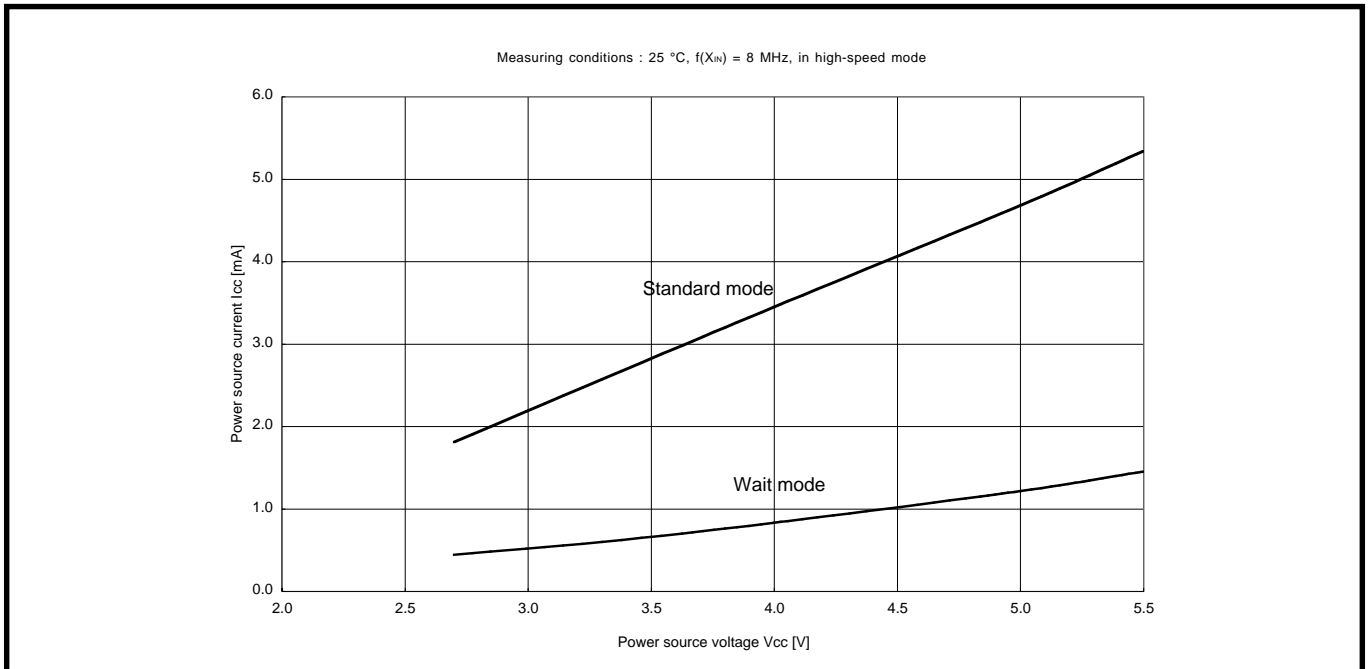
Fig. 3.2.10 Mask ROM version power source current standard characteristics (in low-speed mode)



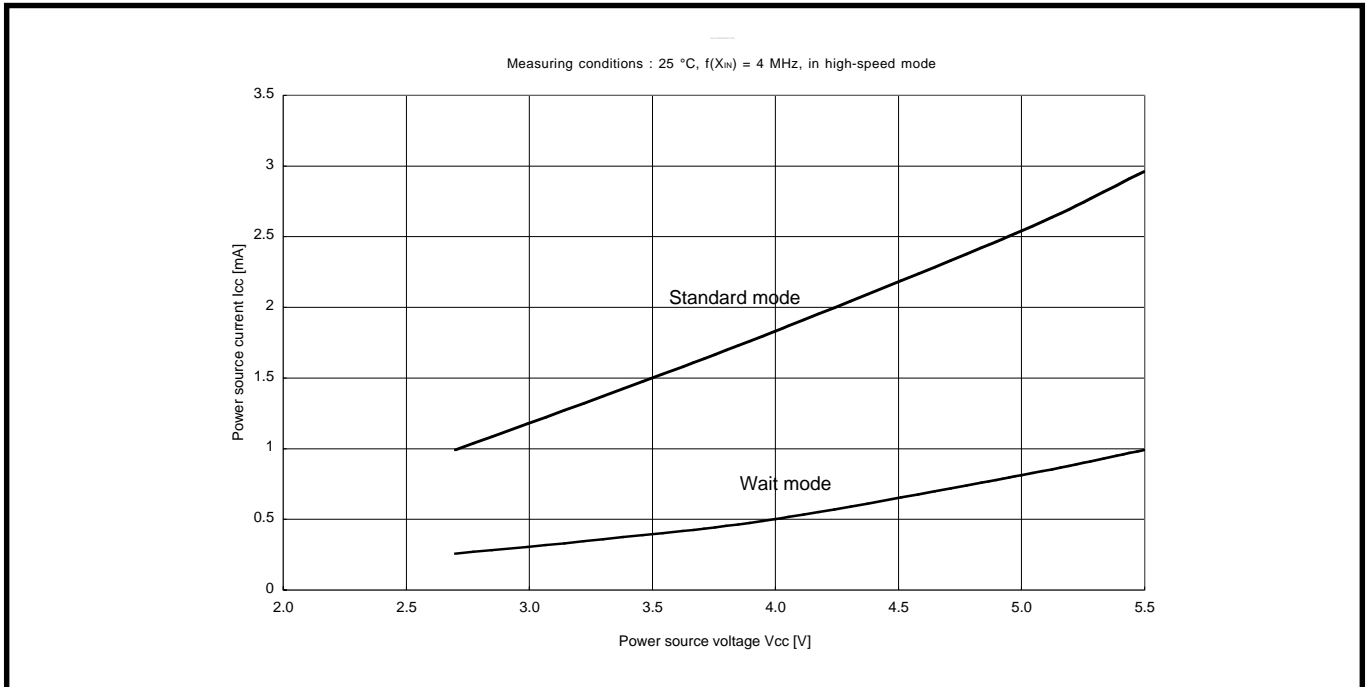
## 3.2 Standard characteristics

### 3.2.3 PROM version power source current standard characteristics

Figure 3.2.11, Figure 3.2.12, Figure 3.2.13, Figure 3.2.14, and Figure 3.2.15 show flash memory version (M38504E6) power source current standard characteristics.



**Fig. 3.2.11 PROM version power source current standard characteristics (in high-speed mode,  $f(X_{IN}) = 8$  MHz)**



**Fig. 3.2.12 PROM version power source current standard characteristics (in high-speed mode,  $f(X_{IN}) = 4$  MHz)**

# APPENDIX

## 3.2 Standard characteristics

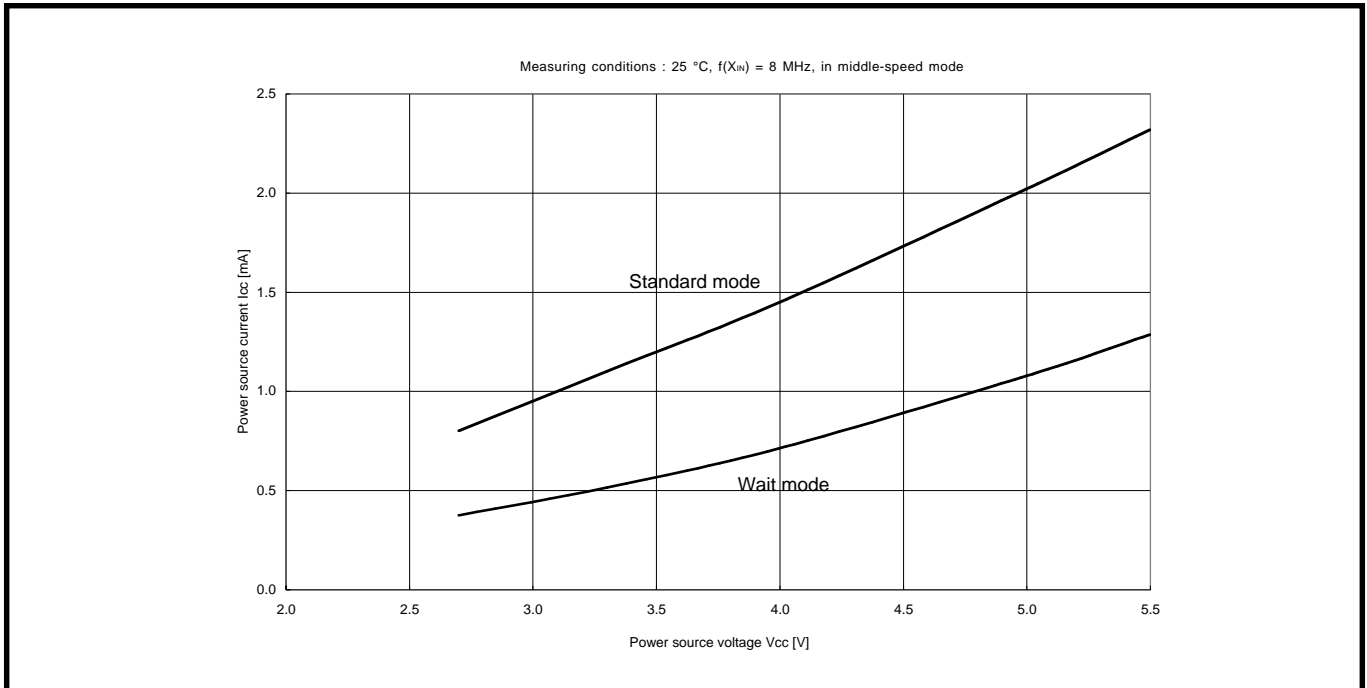


Fig. 3.2.13 PROM version power source current standard characteristics (in middle-speed mode,  $f(X_{IN}) = 8$  MHz)

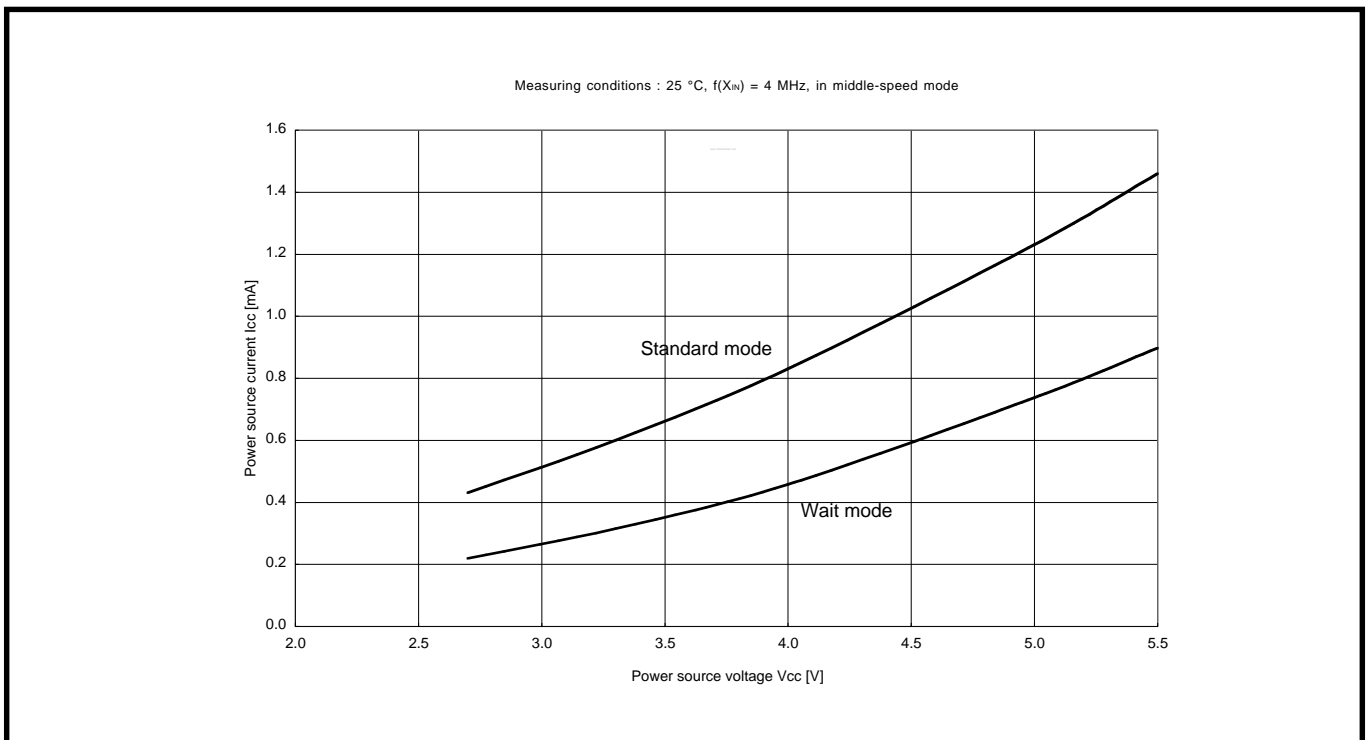


Fig. 3.2.14 PROM version power source current standard characteristics (in middle-speed mode,  $f(X_{IN}) = 4$  MHz)

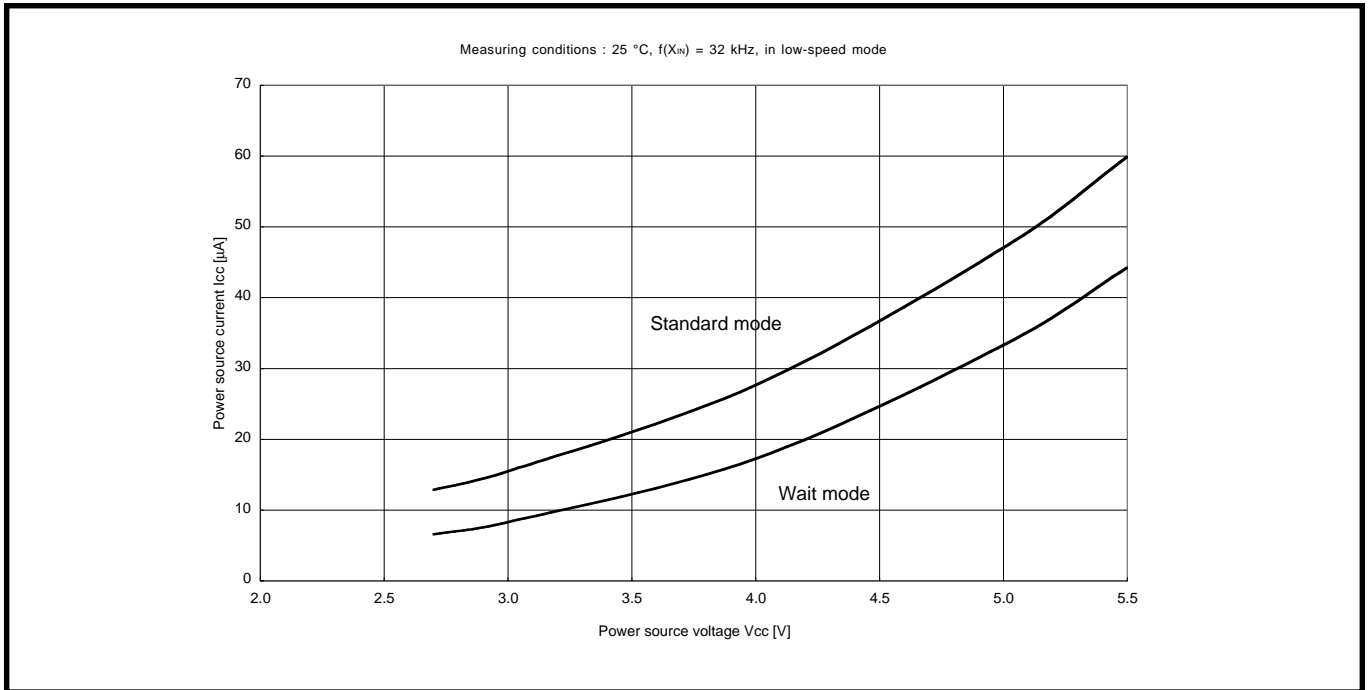


Fig. 3.2.15 PROM version power source current standard characteristics (in low-speed mode)

# APPENDIX

## 3.2 Standard characteristics

### 3.2.4 Flash memory version port standard characteristics

Figure 3.2.16, Figure 3.2.17, Figure 3.2.18 and Figure 3.2.19 show flash memory version (M38507F8) port standard characteristics.

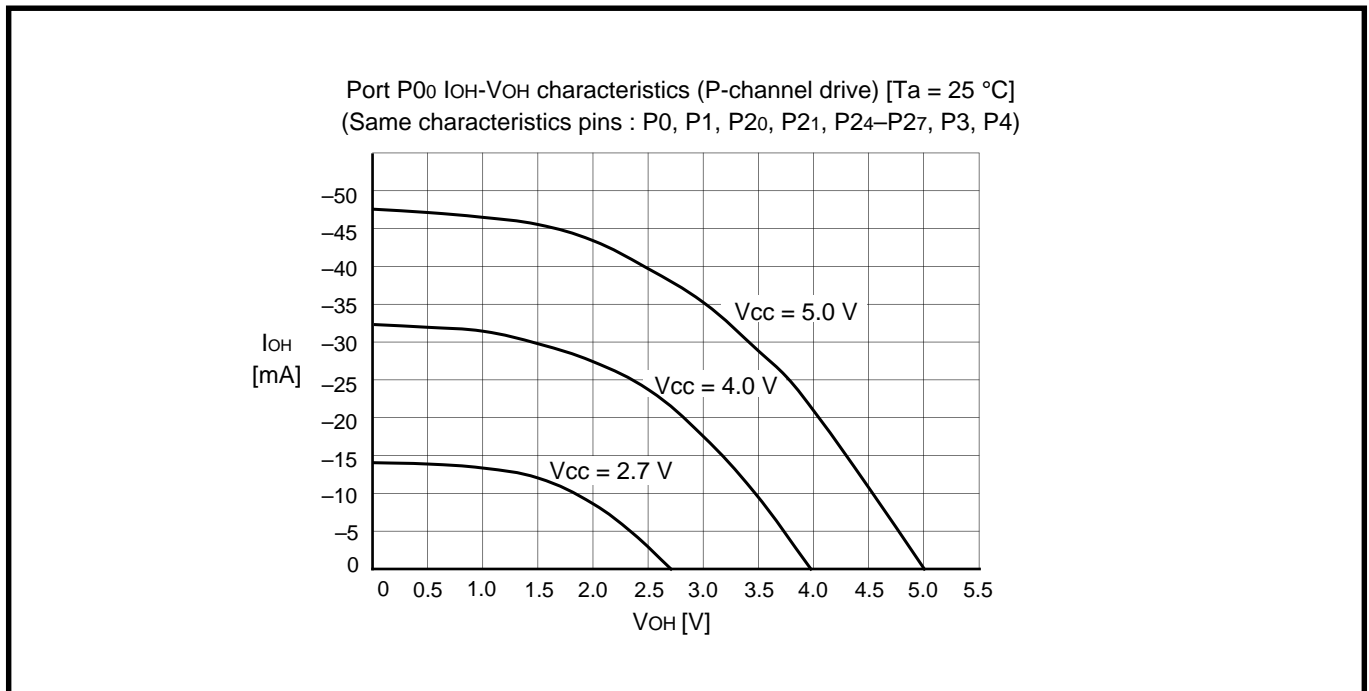


Fig. 3.2.16 CMOS output port P-channel side characteristics (T<sub>a</sub> = 25 °C)

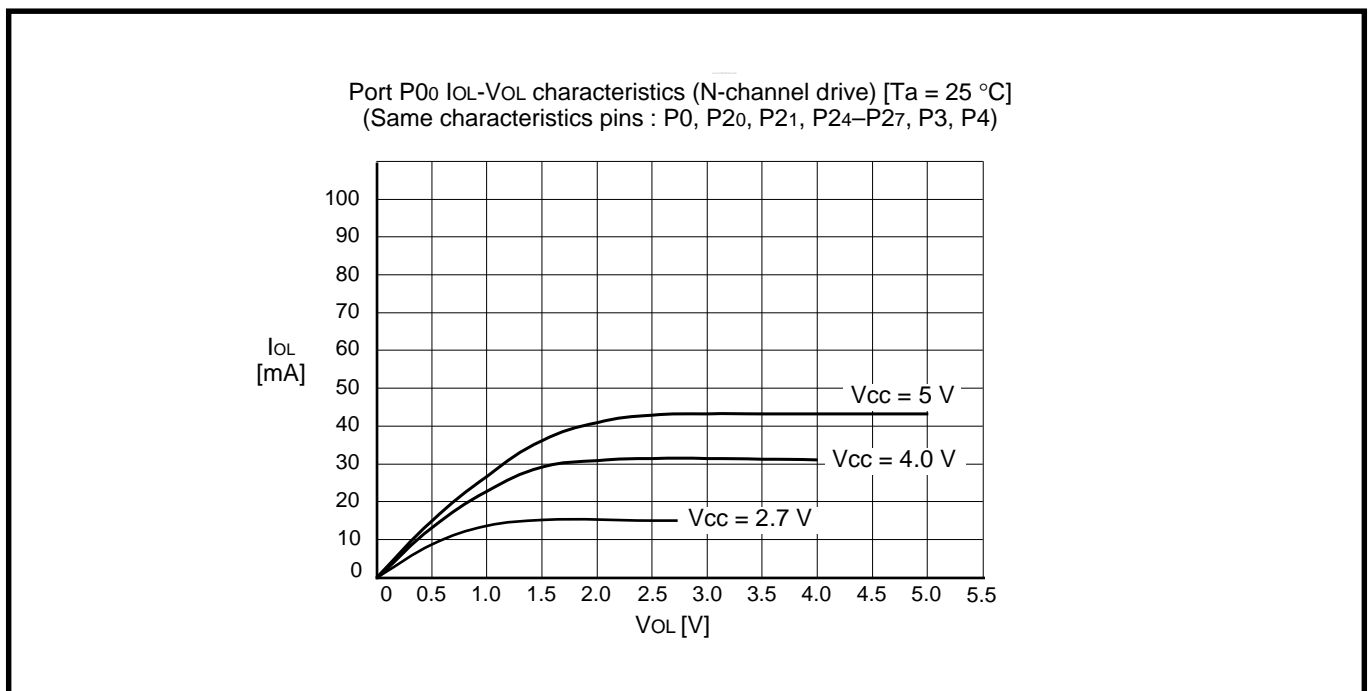
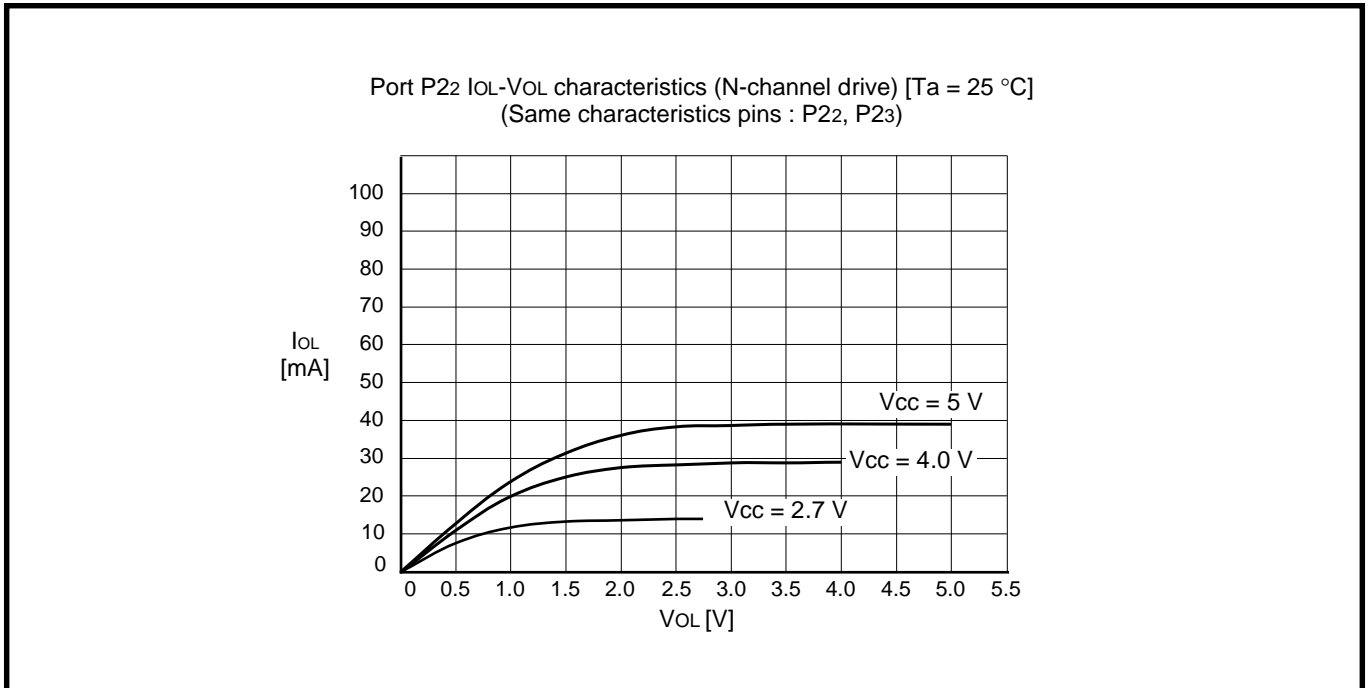
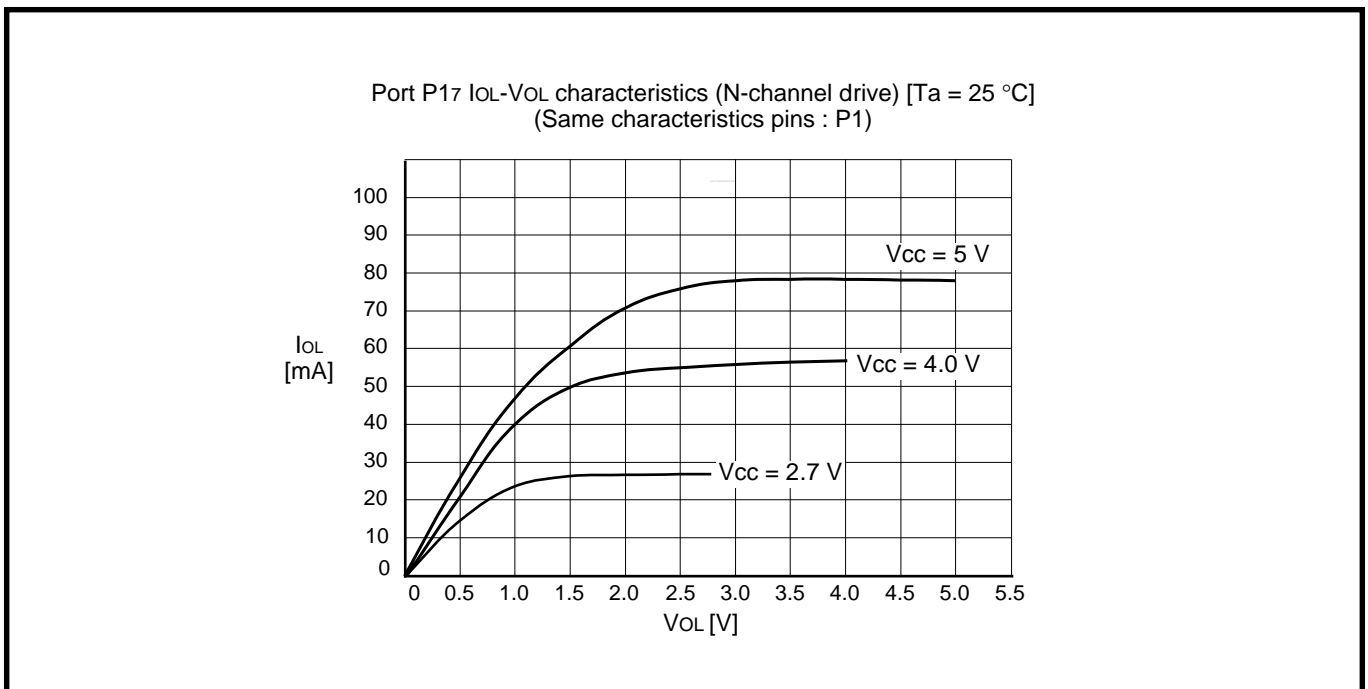


Fig. 3.2.17 CMOS output port N-channel side characteristics (T<sub>a</sub> = 25 °C)



**Fig. 3.2.18 N-channel open-drain output port N-channel side characteristics (Ta = 25 °C)**



**Fig. 3.2.19 CMOS large current output port N-channel side characteristics (Ta = 25 °C)**

# APPENDIX

## 3.2 Standard characteristics

### 3.2.5 Mask ROM version port standard characteristics

Figure 3.2.20, Figure 3.2.21, Figure 3.2.22 and Figure 3.2.23 show mask ROM version (M38503M2H, M38503M4H, M38504M6, M38507M8) port standard characteristics.

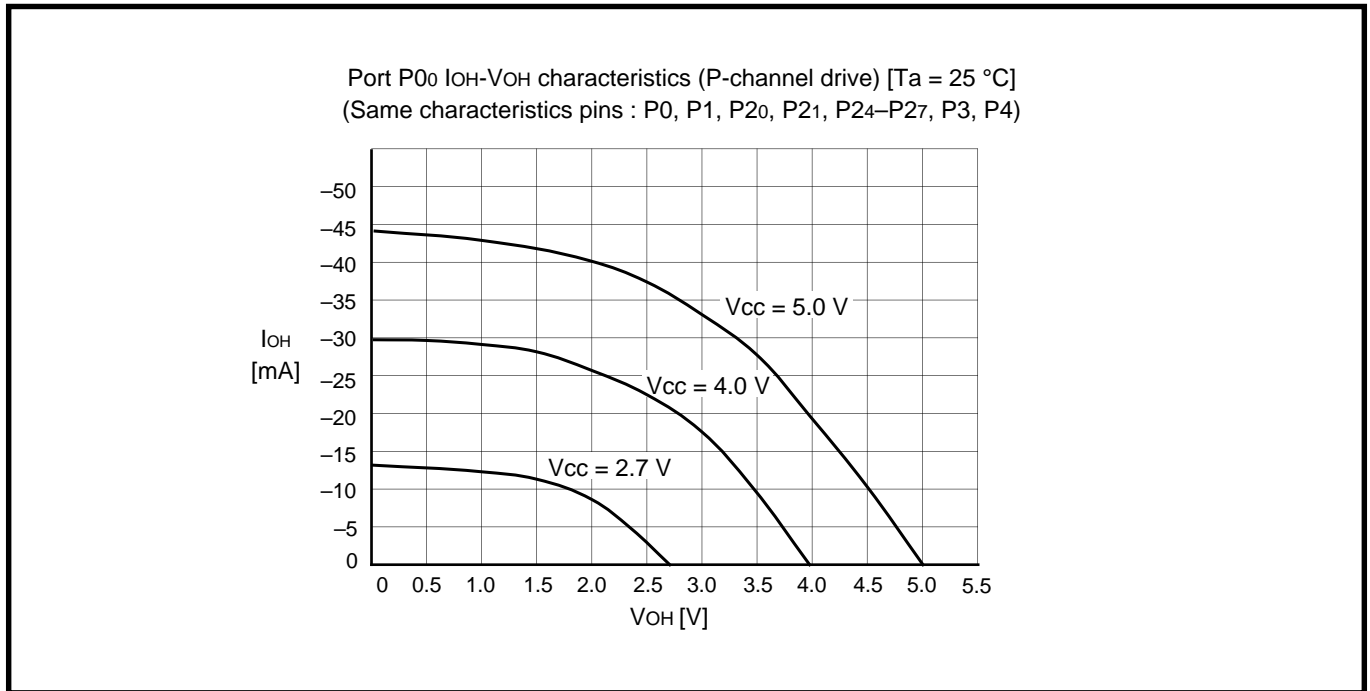


Fig. 3.2.20 CMOS output port P-channel side characteristics (Ta = 25 °C)

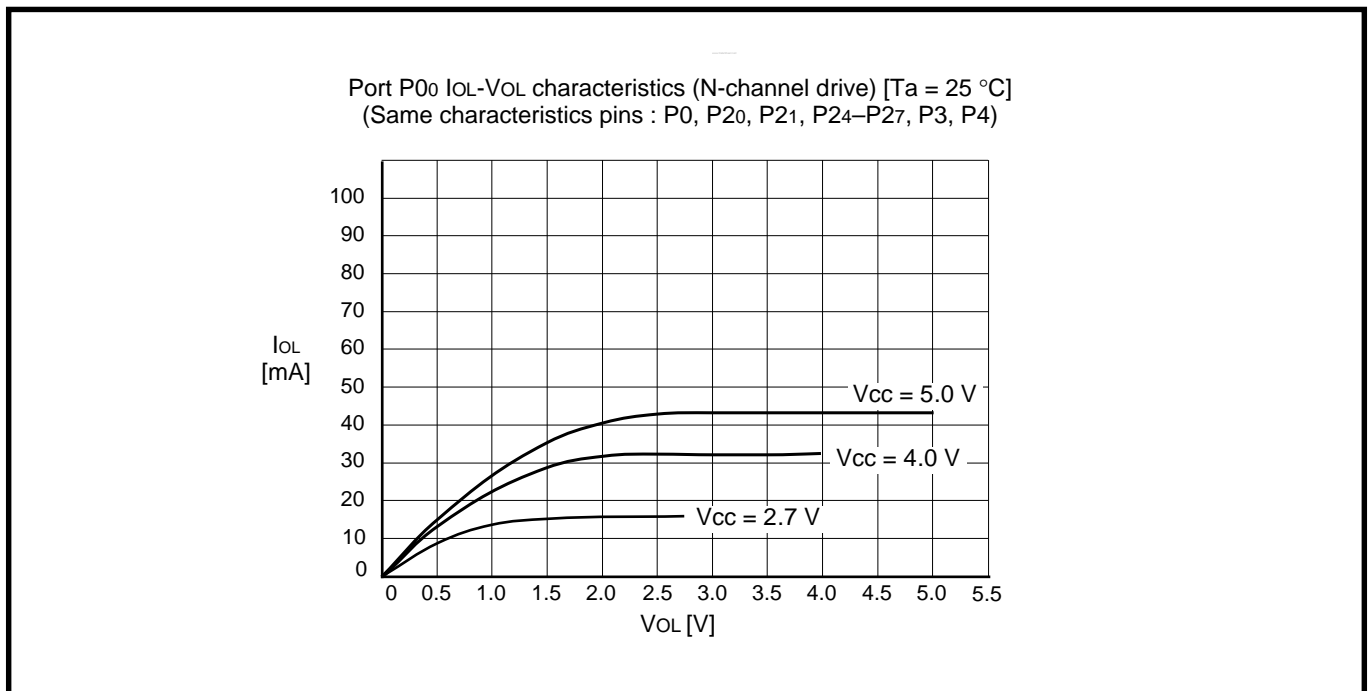


Fig. 3.2.21 CMOS output port N-channel side characteristics (Ta = 25 °C)

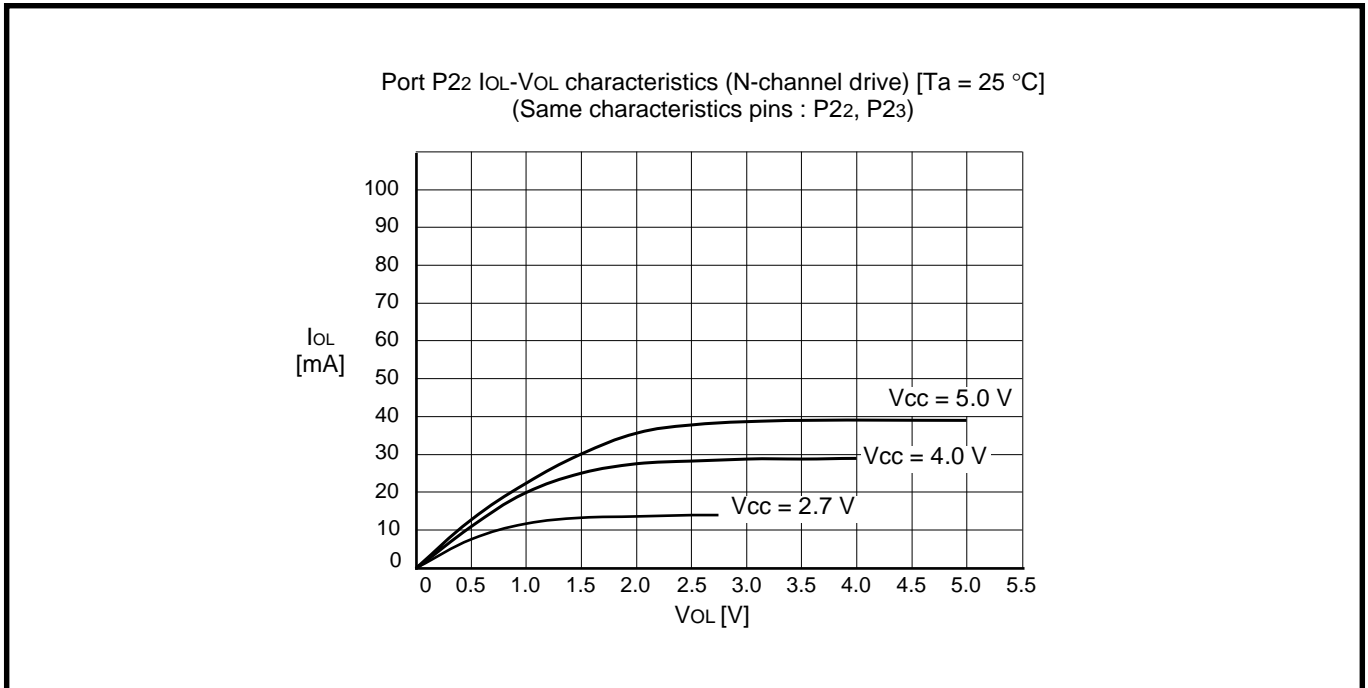


Fig. 3.2.22 N-channel open-drain output port N-channel side characteristics ( $T_a = 25\text{ }^\circ\text{C}$ )

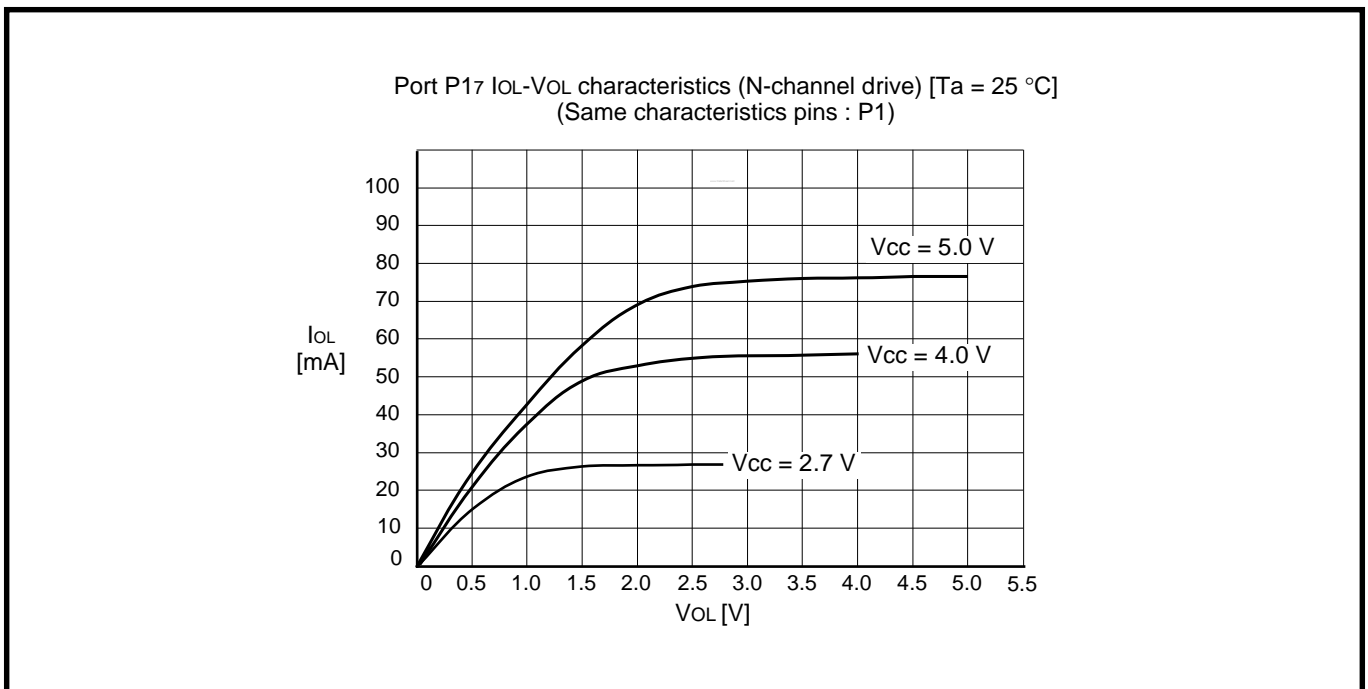


Fig. 3.2.23 CMOS large current output port N-channel side characteristics ( $T_a = 25\text{ }^\circ\text{C}$ )

# APPENDIX

## 3.2 Standard characteristics

### 3.2.6 PROM version port standard characteristics

Figure 3.2.24, Figure 3.2.25, Figure 3.2.26 and Figure 3.2.27 show PROM version (M38504E6) port standard characteristics.

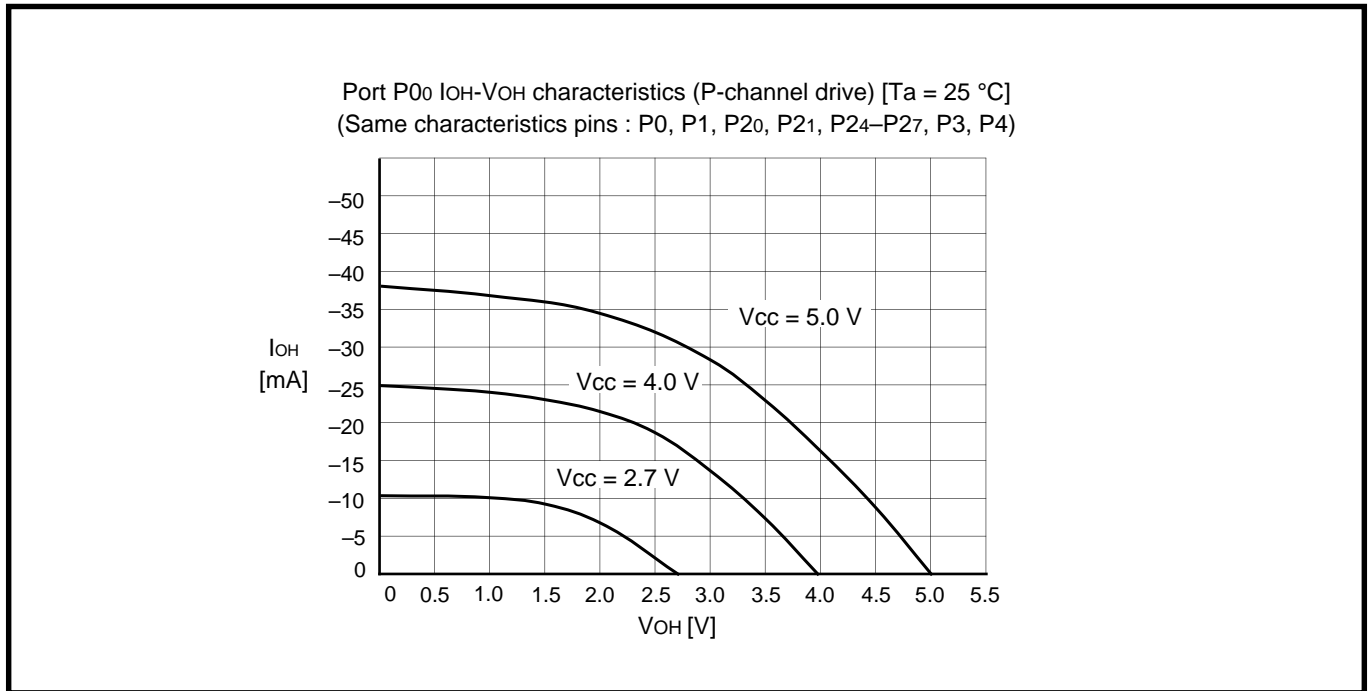


Fig. 3.2.24 CMOS output port P-channel side characteristics (T<sub>a</sub> = 25 °C)

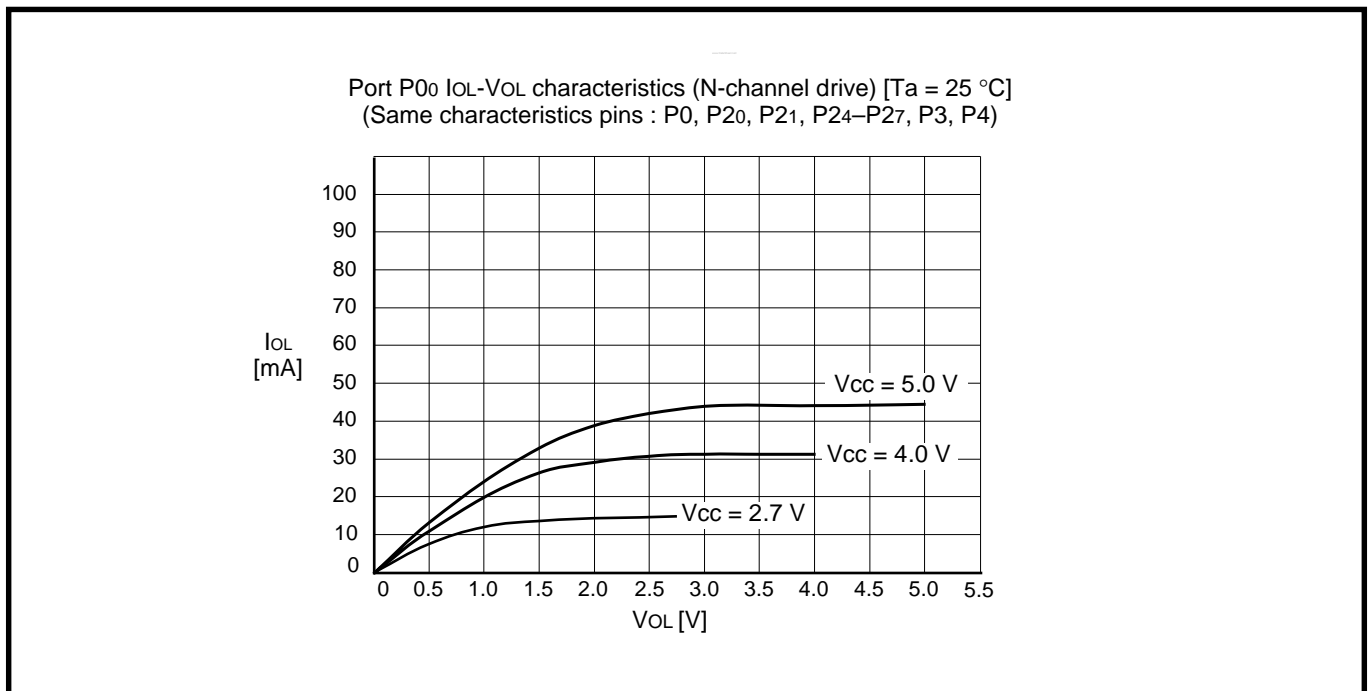
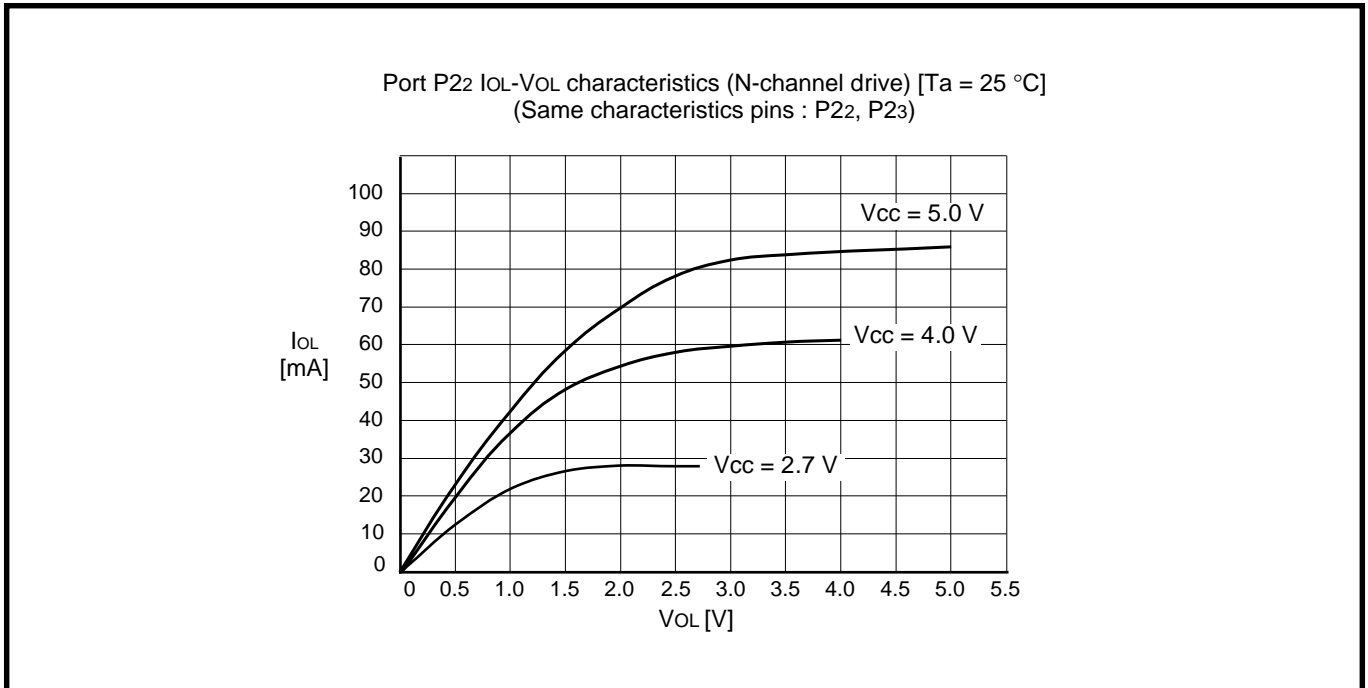
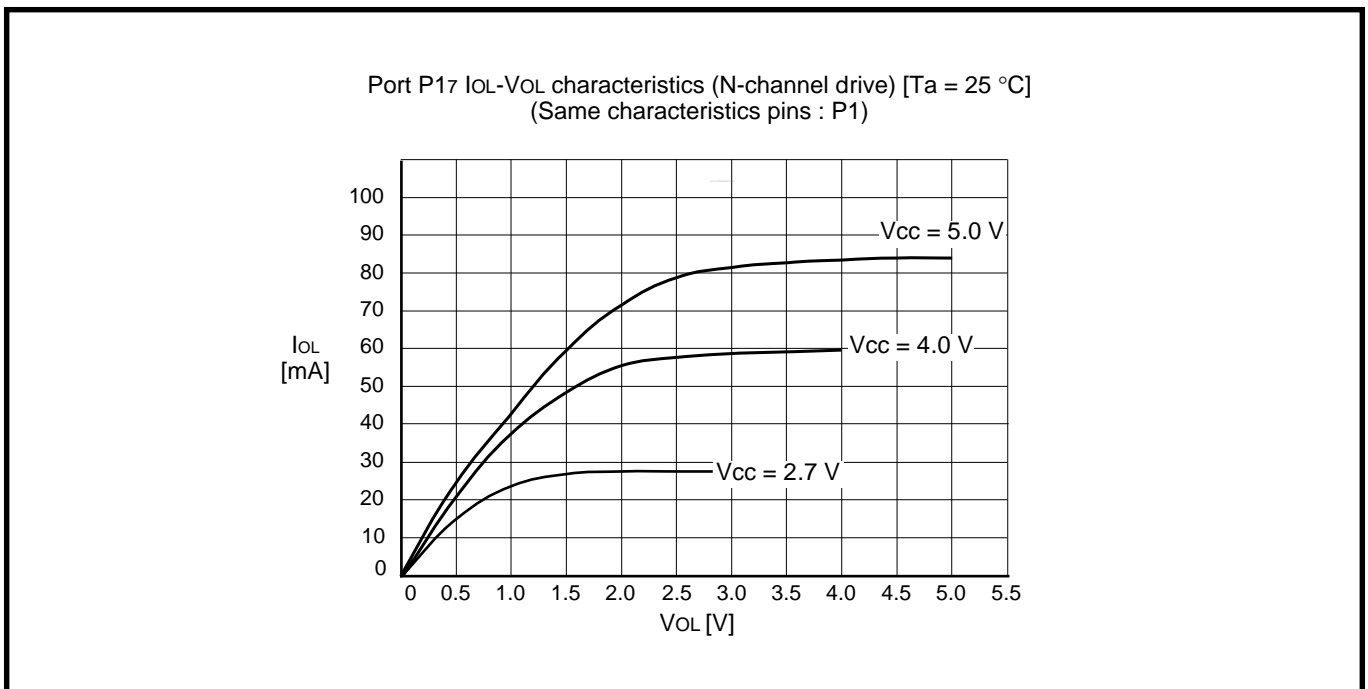


Fig. 3.2.25 CMOS output port N-channel side characteristics (T<sub>a</sub> = 25 °C)





**Fig. 3.2.26 N-channel open-drain output port N-channel side characteristics (Ta = 25 °C)**



**Fig. 3.2.27 CMOS large current output port N-channel side characteristics (Ta = 25 °C)**

# APPENDIX

## 3.2 Standard characteristics

### 3.2.7 A-D conversion standard characteristics

#### (1) Definition of A-D conversion accuracy

The A-D conversion accuracy is defined below.

##### ●Relative accuracy

###### ① Zero transition voltage ( $V_{0T}$ )

This means an analog input voltage when the actual A-D conversion output data changes from “0” to “1”.

###### ② Full-scale transition voltage ( $V_{FST}$ )

This means an analog input voltage when the actual A-D conversion output data changes from “1023” to “1022”.

###### ③ Linearity error

This means a deviation from the lone between  $V_{0T}$  and  $V_{FST}$  of a converted value between  $V_{0T}$  and  $V_{FST}$ .

###### ④ Differential non-linearity error

This means a deviation from the input potential difference required to change a converted value between  $V_{0T}$  and  $V_{FST}$  by 1 LSB of the 1 LSB at the relative accuracy.

##### ●Absolute accuracy

This means a deviation from the ideal characteristics between 0 to  $V_{REF}$  of actual A-D conversion characteristics.

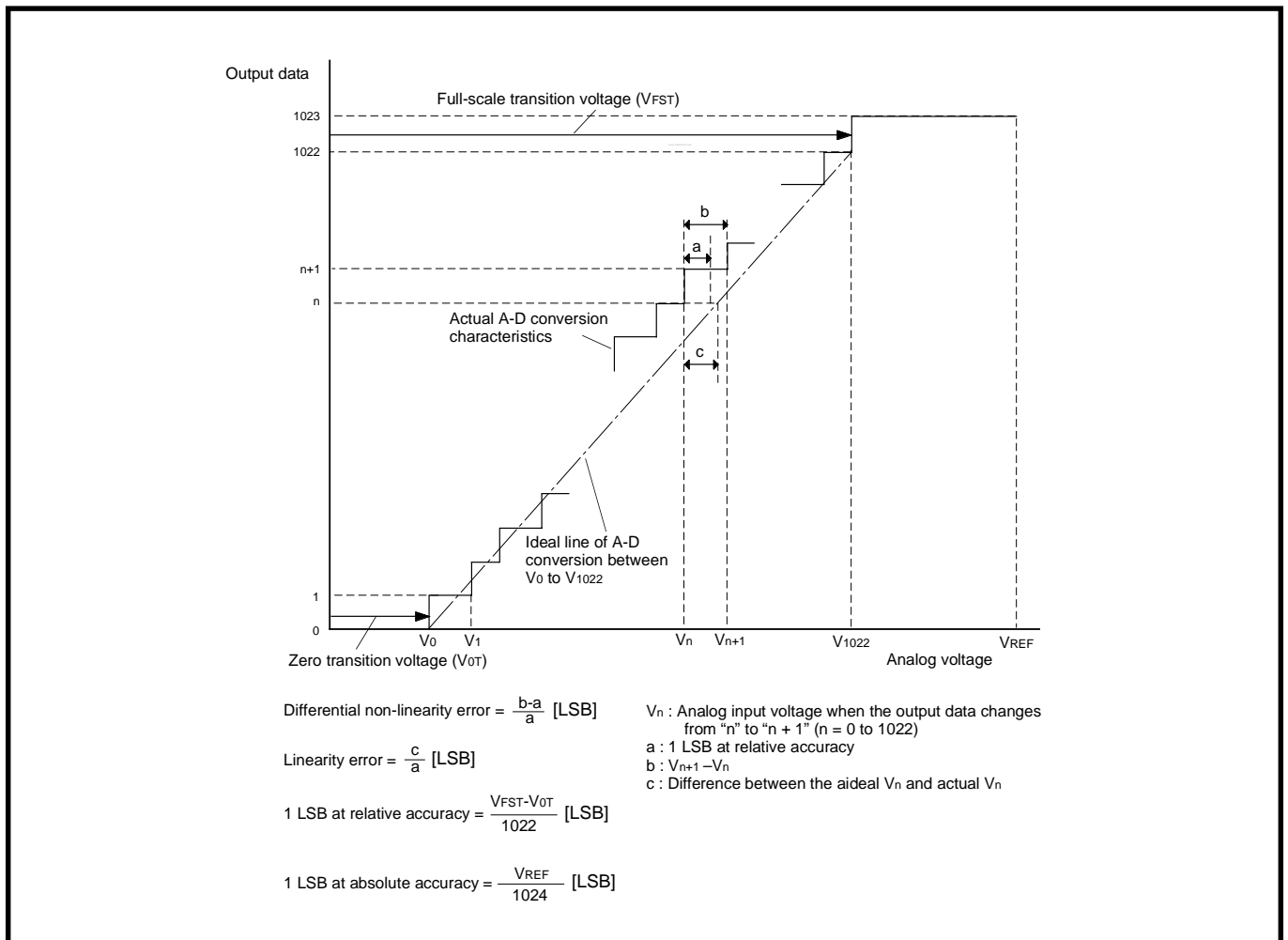


Fig. 3.2.28 Definition of A-D conversion accuracy

### (2) A-D conversion standard characteristics

Figure 3.2.29, Figure 3.2.30, and Figure 3.2.31 show the A-D conversion standard characteristics of flash memory version, mask ROM version, and PROM version, respectively.

The thick lines of the graph indicate the absolute precision errors. These are expressed as the deviation from the ideal value when the output code changes. For example, the change in output code from 256 to 257 should occur at 1280 mV, but the measured value is 2.5 mV. Accordingly, the measured point of change is  $1280 + 2.5 = 1282.5$  mV.

The thin lines of the graph indicate the input voltage width for which the output code is constant. For example, the measured input voltage width for which the output code is 256 is 5.0 mV, so that the differential non-linear error is  $5.0 - 5.0 = 0$  mV (0 LSB).

# APPENDIX

## 3.2 Standard characteristics

### M38507F8 A-D CONVERTER ERROR & STEP WIDTH MEASUREMENT

VCC = 5.12 [V], VREF = 5.12 [V]  
 XIN = 8 [MHz], Ta = 25 [deg.]

Zero transition voltage	:	10.625 [mV]	
Full-scale transition voltage	:	5122.812 [mV]	
Differential non-linearity error	:	1.719 [mV]	0.344 [LSB]
Linearity error	:	-5.659 [mV]	-1.131 [LSB]
Absolute accuracy	:	8.906 [mV]	1.781 [LSB]

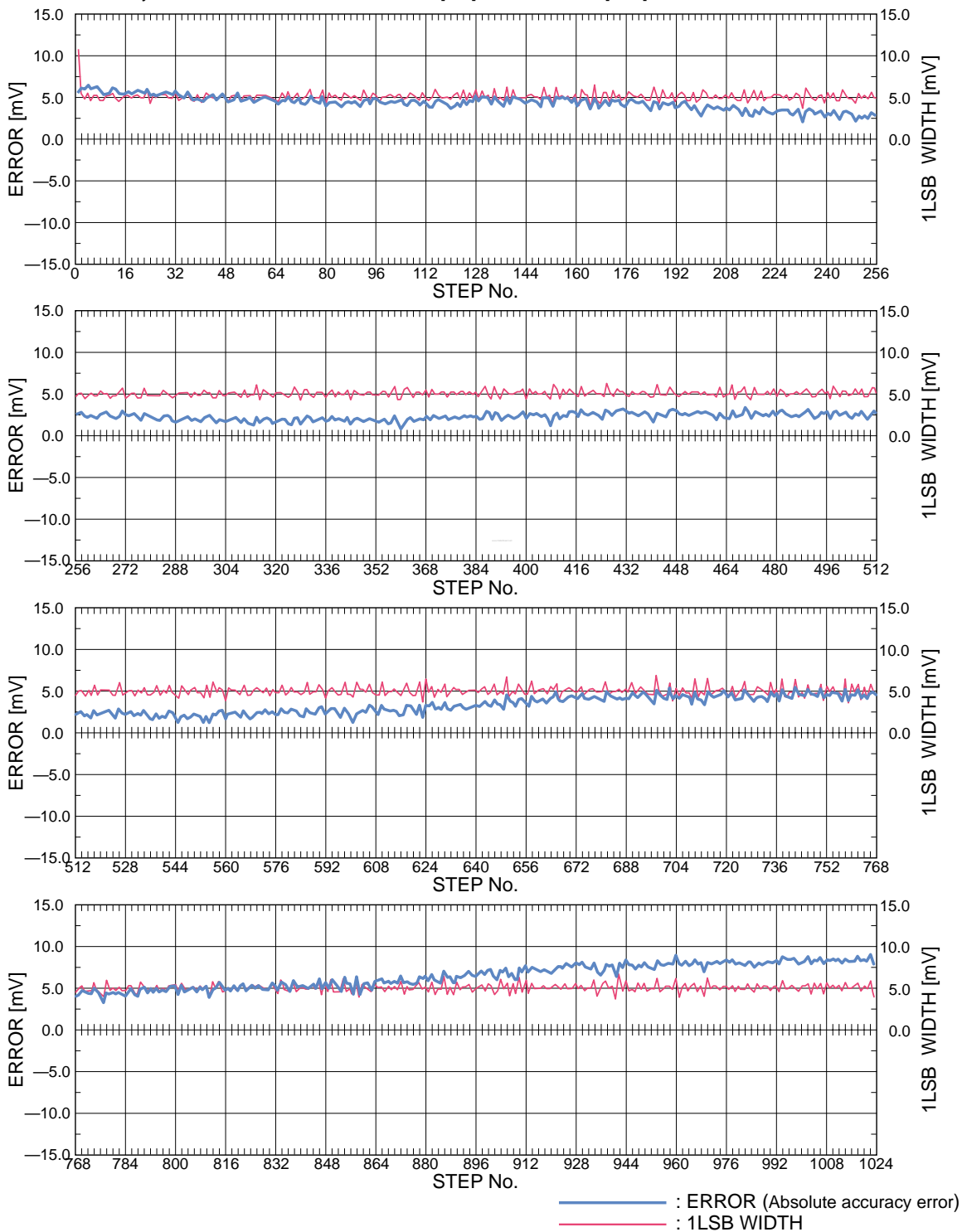


Fig. 3.2.29 Flash memory version (M38507F8) A-D conversion standard characteristics

### M38503M2H, M38503M4H, M38504M6, M38507M8 A-D CONVERTER ERROR & STEP WIDTH MEASUREMENT

VCC = 5.12 [V], VREF = 5.12 [V]  
 XIN = 8 [MHz], Ta = 25 [deg.]

Zero transition voltage	:	10.31 [mV]	
Full-scale transition voltage	:	5118.12 [mV]	
Differential non-linearity error	:	-4.41 [mV]	-0.28 [LSB]
Linearity error	:	-4.72 [mV]	-0.94 [LSB]
Absolute accuracy	:	6.25 [mV]	1.25 [LSB]

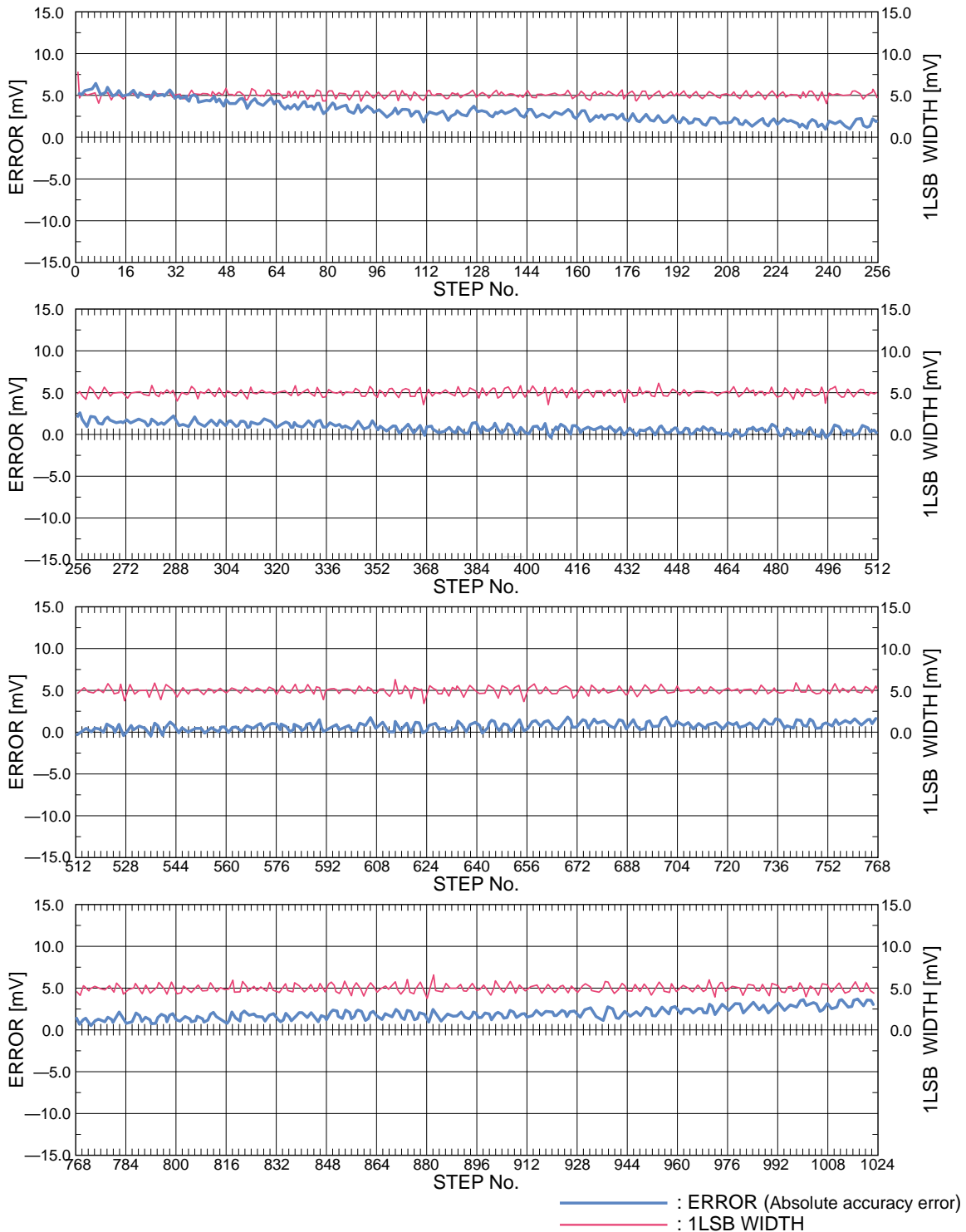


Fig. 3.2.30 Mask ROM version (M38503M2H, M38503M4H, M38504M6, M38507M8) A-D conversion standard characteristics

# APPENDIX

## 3.2 Standard characteristics

### M38504E6 A-D CONVERTER ERROR & STEP WIDTH MEASUREMENT

VCC = 5.12 [V], VREF = 5.12 [V]  
 XIN = 8 [MHz], Ta = 25 [deg.]

Zero transition voltage	:	-0.62 [mV]	
Full-scale transition voltage	:	5112.19 [mV]	
Differential non-linearity error	:	2.97 [mV]	0.59 [LSB]
Linearity error	:	-3.41 [mV]	-0.68 [LSB]
Absolute accuracy	:	-7.03 [mV]	-1.41 [LSB]

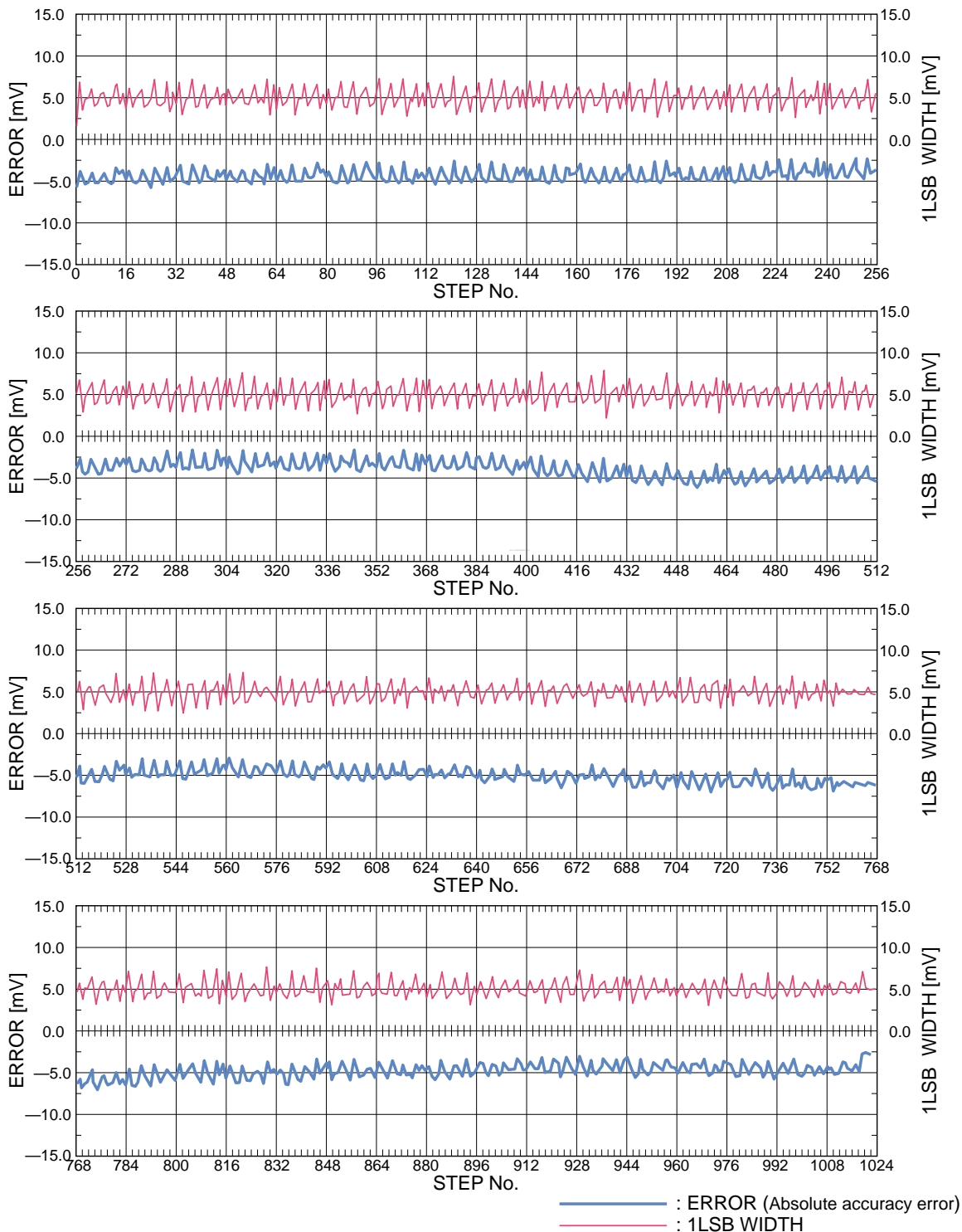


Fig. 3.2.31 PROM version (M38504E6) A-D conversion standard characteristics

## 3.3 Notes on use

### 3.3.1 Notes on input and output ports

#### (1) Notes in standby state

In standby state\*<sup>1</sup>, do not make input levels of an I/O port “undefined”, especially for I/O ports of the N-channel open-drain. When setting the N-channel open-drain port as an output, do not make input levels of an I/O port “undefined”, too.

Pull-up (connect the port to VCC) or pull-down (connect the port to VSS) these ports through a resistor.

When determining a resistance value, note the following points:

- External circuit
- Variation of output levels during the ordinary operation

#### ● Reason

When setting as an input port with its direction register, the transistor becomes the OFF state, which causes the ports to be the high-impedance state.

Accordingly, the potential which is input to the input buffer in a microcomputer is unstable in the state that input levels of an I/O port are “undefined”. This may cause power source current.

In I/O ports of N-channel open-drain, when the contents of the port latch are “1”, even if it is set as an output port with its direction register, it becomes the same phenomenon as the case of an input port.

\*1 standby state: stop mode by executing **STP** instruction  
wait mode by executing **WIT** instruction

#### (2) Modifying output data with bit managing instruction

When the port latch of an I/O port is modified with the bit managing instruction\*<sup>2</sup>, the value of the unspecified bit may be changed.

#### ● Reason

The bit managing instructions are read-modify-write form instructions for reading and writing data by a byte unit. Accordingly, when these instructions are executed on a bit of the port latch of an I/O port, the following is executed to all bits of the port latch.

- As for bit which is set for input port:

The pin state is read in the CPU, and is written to this bit after bit managing.

- As for bit which is set for output port:

The bit value is read in the CPU, and is written to this bit after bit managing.

Note the following:

- Even when a port which is set as an output port is changed for an input port, its port latch holds the output data.
- As for a bit of which is set for an input port, its value may be changed even when not specified with a bit managing instruction in case where the pin state differs from its port latch contents.

\*2 Bit managing instructions: **SEB** and **CLB** instructions

# APPENDIX

## 3.3 Notes on use

---

### 3.3.2 Termination of unused pins

#### (1) Terminate unused pins

① Output ports : Open

② Input ports :

Connect each pin to  $V_{CC}$  or  $V_{SS}$  through each resistor of 1 k $\Omega$  to 10 k $\Omega$ .

As for pins whose potential affects to operation modes such as pins  $CNV_{SS}$ , INT or others, select the  $V_{CC}$  pin or the  $V_{SS}$  pin according to their operation mode.

③ I/O ports :

- Set the I/O ports for the input mode and connect them to  $V_{CC}$  or  $V_{SS}$  through each resistor of 1 k $\Omega$  to 10 k $\Omega$ .

Set the I/O ports for the output mode and open them at "L" or "H".

- When opening them in the output mode, the input mode of the initial status remains until the mode of the ports is switched over to the output mode by the program after reset. Thus, the potential at these pins is undefined and the power source current may increase in the input mode. With regard to an effects on the system, thoroughly perform system evaluation on the user side.
- Since the direction register setup may be changed because of a program runaway or noise, set direction registers by program periodically to increase the reliability of program.

④ The AVss pin when not using the A-D converter :

- When not using the A-D converter, handle a power source pin for the A-D converter, AVss pin as follows:

AVss: Connect to the Vss pin.

#### (2) Termination remarks

① Input ports and I/O ports :

Do not open in the input mode.

● Reason

- The power source current may increase depending on the first-stage circuit.
- An effect due to noise may be easily produced as compared with proper termination ② and ③ shown on the above.

② I/O ports :

When setting for the input mode, do not connect to  $V_{CC}$  or  $V_{SS}$  directly.

● Reason

If the direction register setup changes for the output mode because of a program runaway or noise, a short circuit may occur between a port and  $V_{CC}$  (or  $V_{SS}$ ).

③ I/O ports :

When setting for the input mode, do not connect multiple ports in a lump to  $V_{CC}$  or  $V_{SS}$  through a resistor.

● Reason

If the direction register setup changes for the output mode because of a program runaway or noise, a short circuit may occur between ports.

- At the termination of unused pins, perform wiring at the shortest possible distance (20 mm or less) from microcomputer pins.



## 3.3.3 Notes on interrupts

## (1) Change of relevant register settings

When the setting of the following registers or bits is changed, the interrupt request bit may be set to "1". When not requiring the interrupt occurrence synchronized with these setting, take the following sequence.

- Interrupt edge selection register (address  $3A_{16}$ )
- Timer XY mode register (address  $23_{16}$ )

Set the above listed registers or bits as the following sequence.

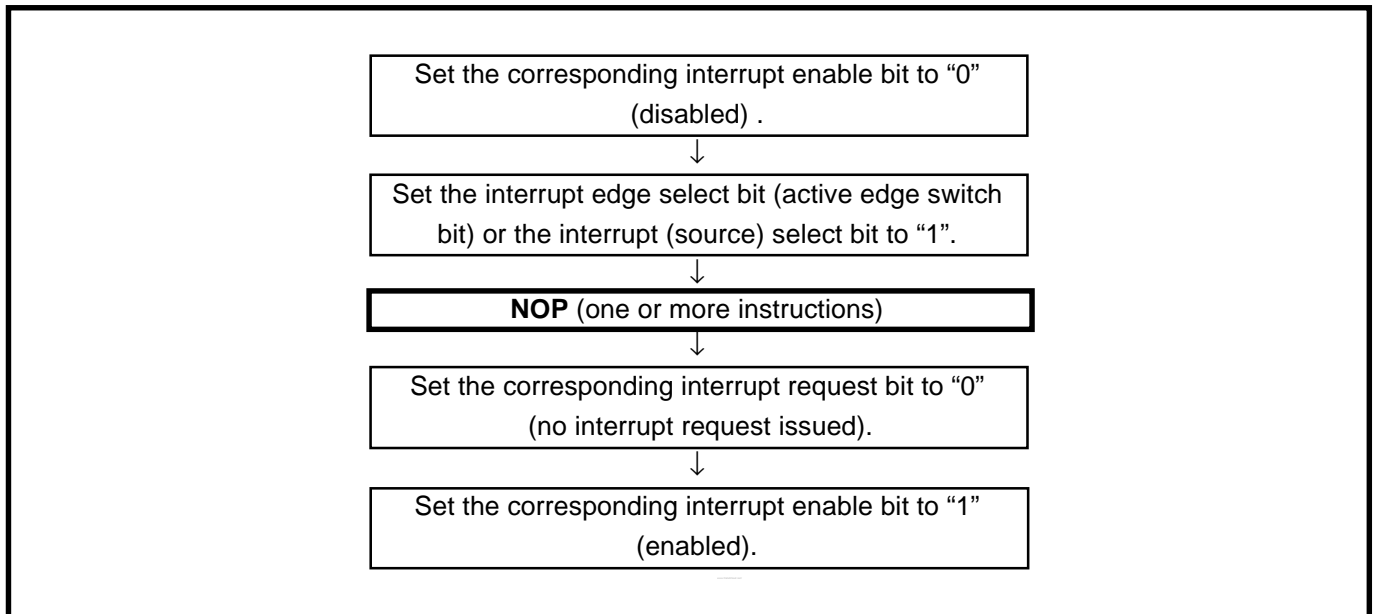


Fig. 3.3.1 Sequence of changing relevant register

■ Reason

When setting the followings, the interrupt request bit may be set to "1".

- When setting external interrupt active edge  
Concerned register: Interrupt edge selection register (address  $3A_{16}$ )  
Timer XY mode register (address  $23_{16}$ )
- When switching interrupt sources of an interrupt vector address where two or more interrupt sources are allocated.  
Concerned register: Interrupt edge selection register (address  $3A_{16}$ )

# APPENDIX

## 3.3 Notes on use

---

### (2) Check of interrupt request bit

- When executing the **BBC** or **BBS** instruction to an interrupt request bit of an interrupt request register immediately after this bit is set to "0" by using a data transfer instruction, execute one or more instructions before executing the **BBC** or **BBS** instruction.

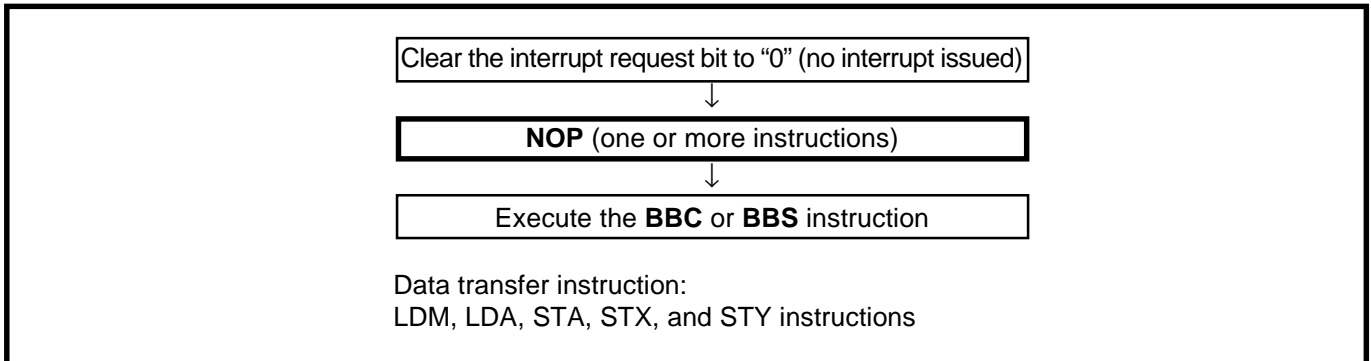


Fig. 3.3.2 Sequence of check of interrupt request bit

#### ■ Reason

If the BBC or BBS instruction is executed immediately after an interrupt request bit of an interrupt request register is cleared to "0", the value of the interrupt request bit before being cleared to "0" is read.

### 3.3.4 Notes on timer

- If a value  $n$  (between 0 and 255) is written to a timer latch, the frequency division ratio is  $1/(n+1)$ .
- When switching the count source by the timer 12, X and Y count source selection bits, the value of timer count is altered in unconsiderable amount owing to generating of thin pulses in the count input signals.  
Therefore, select the timer count source before set the value to the prescaler and the timer.

### 3.3.5 Notes on serial I/O

#### (1) Notes when selecting clock synchronous serial I/O (Serial I/O1)

##### ① Stop of transmission operation

Clear the serial I/O1 enable bit and the transmit enable bit to "0" (Serial I/O1 and transmit disabled).

##### ● Reason

Since transmission is not stopped and the transmission circuit is not initialized even if only the serial I/O1 enable bit is cleared to "0" (Serial I/O1 disabled), the internal transmission is running (in this case, since pins TxD, RxD, SCLK1, and  $\overline{\text{SRDY}}_1$  function as I/O ports, the transmission data is not output). When data is written to the transmit buffer register in this state, data starts to be shifted to the transmit shift register. When the serial I/O1 enable bit is set to "1" at this time, the data during internally shifting is output to the TxD pin and an operation failure occurs.

##### ② Stop of receive operation

Clear the receive enable bit to "0" (receive disabled), or clear the serial I/O1 enable bit to "0" (Serial I/O1 disabled).

### ③ Stop of transmit/receive operation

Clear the transmit enable bit and receive enable bit to “0” simultaneously (transmit and receive disabled).

(when data is transmitted and received in the clock synchronous serial I/O mode, any one of data transmission and reception cannot be stopped.)

#### ● Reason

In the clock synchronous serial I/O mode, the same clock is used for transmission and reception. If any one of transmission and reception is disabled, a bit error occurs because transmission and reception cannot be synchronized.

In this mode, the clock circuit of the transmission circuit also operates for data reception. Accordingly, the transmission circuit does not stop by clearing only the transmit enable bit to “0” (transmit disabled). Also, the transmission circuit is not initialized by clearing the serial I/O1 enable bit to “0” (Serial I/O1 disabled) (refer to (1) ①).

### ④ $\overline{\text{SRDY1}}$ output of reception side (Serial I/O1)

When signals are output from the  $\overline{\text{SRDY1}}$  pin on the reception side by using an external clock in the clock synchronous serial I/O mode, set all of the receive enable bit, the  $\overline{\text{SRDY1}}$  output enable bit, and the transmit enable bit to “1” (transmit enabled).

## (2) Notes when selecting clock asynchronous serial I/O (Serial I/O1)

### ① Stop of transmission operation

Clear the transmit enable bit to “0” (transmit disabled).

#### ● Reason

Since transmission is not stopped and the transmission circuit is not initialized even if only the serial I/O1 enable bit is cleared to “0” (Serial I/O1 disabled), the internal transmission is running (in this case, since pins TxD, RxD, SCLK1, and  $\overline{\text{SRDY1}}$  function as I/O ports, the transmission data is not output). When data is written to the transmit buffer register in this state, data starts to be shifted to the transmit shift register. When the serial I/O1 enable bit is set to “1” at this time, the data during internally shifting is output to the TxD pin and an operation failure occurs.

### ② Stop of receive operation

Clear the receive enable bit to “0” (receive disabled).

### ③ Stop of transmit/receive operation

#### Only transmission operation is stopped.

Clear the transmit enable bit to “0” (transmit disabled).

#### ● Reason

Since transmission is not stopped and the transmission circuit is not initialized even if only the serial I/O1 enable bit is cleared to “0” (Serial I/O1 disabled), the internal transmission is running (in this case, since pins TxD, RxD, SCLK1, and  $\overline{\text{SRDY1}}$  function as I/O ports, the transmission data is not output). When data is written to the transmit buffer register in this state, data starts to be shifted to the transmit shift register. When the serial I/O1 enable bit is set to “1” at this time, the data during internally shifting is output to the TxD pin and an operation failure occurs.

#### Only receive operation is stopped.

Clear the receive enable bit to “0” (receive disabled).

# APPENDIX

## 3.3 Notes on use

### (3) Setting serial I/O1 control register again (Serial I/O1)

Set the serial I/O1 control register again after the transmission and the reception circuits are reset by clearing both the transmit enable bit and the receive enable bit to "0".

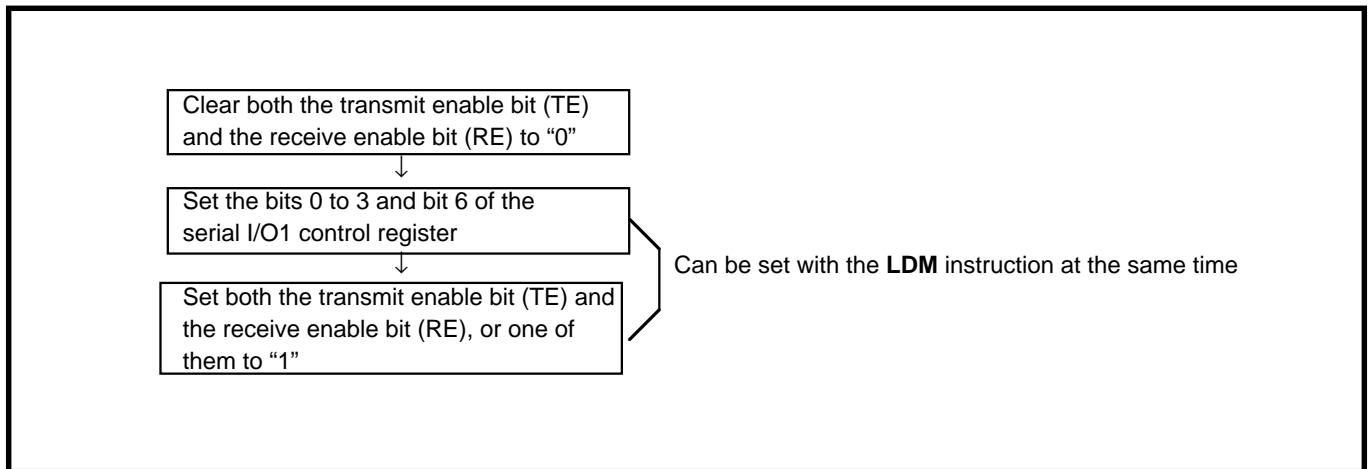


Fig. 3.3.3 Sequence of setting serial I/O1 control register again

### (4) Data transmission control with referring to transmit shift register completion flag (Serial I/O1)

The transmit shift register completion flag changes from "1" to "0" with a delay of 0.5 to 1.5 shift clocks. When data transmission is controlled with referring to the flag after writing the data to the transmit buffer register, note the delay.

### (5) Transmit interrupt request when transmit enable bit is set (Serial I/O1)

When the transmit interrupt is used, set the transmit interrupt enable bit to transmit enabled as shown in the following sequence.

- ① Set the interrupt enable bit to "0" (disabled) with CLB instruction.
- ② Prepare serial I/O for transmission/reception.
- ③ Set the interrupt request bit to "0" with CLB instruction after 1 or more instruction has been executed.
- ④ Set the interrupt enable bit to "1" (enabled).

#### ● Reason

When the transmission enable bit is set to "1", the transmit buffer empty flag and transmit shift register completion flag are set to "1". The interrupt request is generated and the transmission interrupt request bit is set regardless of which of the two timings listed below is selected as the timing for the transmission interrupt to be generated.

- Transmit buffer empty flag is set to "1"
- Transmit shift register completion flag is set to "1"

### (6) Transmission control when external clock is selected (Serial I/O1 clock synchronous mode)

When an external clock is used as the synchronous clock for data transmission, set the transmit enable bit to "1" at "H" of the SCLK1 input level. Also, write the transmit data to the transmit buffer register (serial I/O shift register) at "H" of the SCLK1 input level.

### (7) Transmit data writing (Serial I/O2)

In the clock synchronous serial I/O, when selecting an external clock as synchronous clock, write the transmit data to the serial I/O2 register (serial I/O shift register) at "H" of the transfer clock input level.

**3.3.6 Notes on PWM**

The PWM starts after the PWM enable bit is set to enable and “L” level is output from the PWM pin. The length of this “L” level output is as follows:

$$\frac{n + 1}{2 \cdot f(X_{IN})} \text{ sec. (Count source selection bit = "0", where n is the value set in the prescaler)}$$

$$\frac{n + 1}{f(X_{IN})} \text{ sec. (Count source selection bit = "1", where n is the value set in the prescaler)}$$

**3.3.7 Notes on A-D converter****(1) Analog input pin**

Make the signal source impedance for analog input low, or equip an analog input pin with an external capacitor of 0.01  $\mu$ F to 1  $\mu$ F. Further, be sure to verify the operation of application products on the user side.

**● Reason**

An analog input pin includes the capacitor for analog voltage comparison. Accordingly, when signals from signal source with high impedance are input to an analog input pin, charge and discharge noise generates. This may cause the A-D conversion precision to be worse.

**(2) A-D converter power source pin**

The AVSS pin is A-D converter power source pin. Regardless of using the A-D conversion function or not, connect it as following :

- AVSS : Connect to the VSS line

**● Reason**

If the AVSS pin is opened, the microcomputer may have a failure because of noise or others.

**(3) Clock frequency during A-D conversion**

The comparator consists of a capacity coupling, and a charge of the capacity will be lost if the clock frequency is too low. Thus, make sure the following during an A-D conversion.

- $f(X_{IN})$  is 500 kHz or more in middle-/high-speed mode.
- Do not execute the **STP** instruction.
- When the A-D converter is operated at low-speed mode,  $f(X_{IN})$  do not have the lower limit of frequency, because of the A-D converter has a built-in self-oscillation circuit.

**3.3.8 Notes on watchdog timer**

- Make sure that the watchdog timer does not underflow while waiting Stop release, because the watchdog timer keeps counting during that term.
- When the STP instruction disable bit has been set to “1”, it is impossible to switch it to “0” by a program.

# APPENDIX

## 3.3 Notes on use

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### 3.3.9 Notes on $\overline{\text{RESET}}$ pin

#### (1) Connecting capacitor

In case where the  $\overline{\text{RESET}}$  signal rise time is long, connect a ceramic capacitor or others across the  $\overline{\text{RESET}}$  pin and the VSS pin. Use a 1000 pF or more capacitor for high frequency use. When connecting the capacitor, note the following :

- Make the length of the wiring which is connected to a capacitor as short as possible.
- Be sure to verify the operation of application products on the user side.

#### ● Reason

If the several nanosecond or several ten nanosecond impulse noise enters the  $\overline{\text{RESET}}$  pin, it may cause a microcomputer failure.

#### (2) Reset release after power on

When releasing the reset after power on, such as power-on reset, release reset after  $X_{\text{IN}}$  passes more than 20 cycles in the state where the power supply voltage is 2.7 V or more and the  $X_{\text{IN}}$  oscillation is stable.

#### ● Reason

To release reset, the  $\overline{\text{RESET}}$  pin must be held at an “L” level for 20 cycles or more of  $X_{\text{IN}}$  in the state where the power source voltage is between 2.7 V and 5.5 V, and  $X_{\text{IN}}$  oscillation is stable.

### 3.3.10 Notes on using stop mode

#### ■ Register setting

Since values of the prescaler 12 and Timer 1 are automatically reloaded when returning from the stop mode, set them again, respectively. (When the oscillation stabilizing time set after STP instruction released bit is “0”)

When using the oscillation stabilizing time set after STP instruction released bit set to “1”, evaluate time to stabilize oscillation of the used oscillator and set the value to the timer 1 and prescaler 12.

#### ■ Clock restoration

After restoration from the stop mode to the normal mode by an interrupt request, the contents of the CPU mode register previous to the STP instruction execution are retained. Accordingly, if both main clock and sub clock were oscillating before execution of the STP instruction, the oscillation of both clocks is resumed at restoration.

In the above case, when the main clock side is set as a system clock, the oscillation stabilizing time for approximately 8,000 cycles of the  $X_{\text{IN}}$  input is reserved at restoration from the stop mode. At this time, note that the oscillation on the sub clock side may not be stabilized even after the lapse of the oscillation stabilizing time of the main clock side.

### 3.3.11 Notes on wait mode

#### ■ Clock restoration

If the wait mode is released by a reset when  $X_{\text{CIN}}$  is set as the system clock and  $X_{\text{IN}}$  oscillation is stopped during execution of the WIT instruction,  $X_{\text{CIN}}$  oscillation stops,  $X_{\text{IN}}$  oscillations starts, and  $X_{\text{IN}}$  is set as the system clock.

In the above case, the  $\overline{\text{RESET}}$  pin should be held at “L” until the oscillation is stabilized.

### 3.3.12 Notes on CPU rewrite mode of flash memory version

**(1) Operation speed**

During CPU rewrite mode, set the internal clock frequency 4MHz or less by using the main clock division ratio selection bits (bits 6, 7 at address 003B<sub>16</sub>).

**(2) Instructions inhibited against use**

The instructions which refer to the internal data of the flash memory cannot be used during CPU rewrite mode .

**(3) Interrupts inhibited against use**

The interrupts cannot be used during CPU rewrite mode because they refer to the internal data of the flash memory.

**(4) Watchdog timer**

In case of the watchdog timer has been running already, the internal reset generated by watchdog timer underflow does not happen, because of watchdog timer is always clearing during program or erase operation.

**(5) Reset**

Reset is always valid. In case of CNV<sub>SS</sub> = "H" when reset is released, boot mode is active. So the program starts from the address contained in addresses FFFC<sub>16</sub> and FFFD<sub>16</sub> in boot ROM area.

### 3.3.13 Notes on restarting oscillation

■ **Restarting oscillation**

Usually, when the MCU stops the clock oscillation by STP instruction and the STP instruction has been released by an external interrupt source, the fixed values of Timer 1 and Prescaler 12 (Timer 1 = "01<sub>16</sub>", Prescaler 12 = "FF<sub>16</sub>") are automatically reloaded in order for the oscillation to stabilize. The user can inhibit the automatic setting by writing "1" to bit 0 of MISRG (address 0038<sub>16</sub>).

However, by setting this bit to "1", the previous values, set just before the STP instruction was executed, will remain in Timer 1 and Prescaler 12. Therefore, you will need to set an appropriate value to each register, in accordance with the oscillation stabilizing time, before executing the STP instruction.

● **Reason**

Oscillation will restart when an external interrupt is received. However, internal clock  $\phi$  is supplied to the CPU only when Timer 1 starts to underflow. This ensures time for the clock oscillation using the ceramic resonators to be stabilized.

# APPENDIX

## 3.3 Notes on use

### 3.3.14 Notes on programming

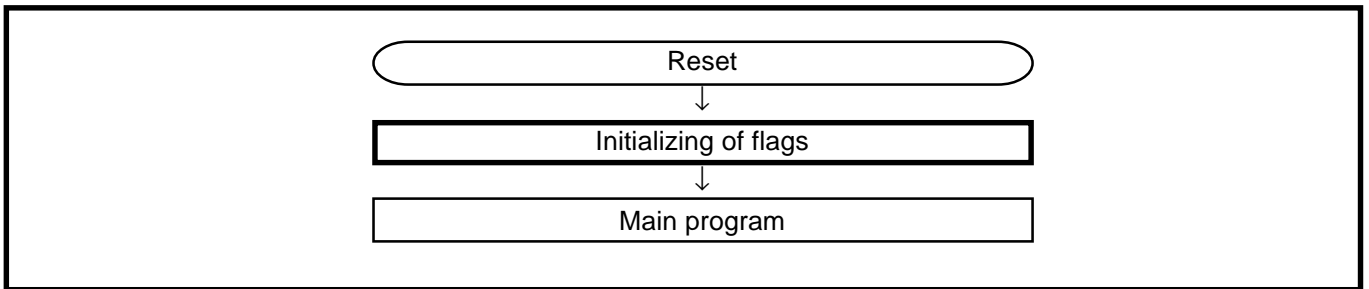
#### (1) Processor status register

① **Initializing of processor status register**

Flags which affect program execution must be initialized after a reset. In particular, it is essential to initialize the T and D flags because they have an important effect on calculations.

● **Reason**

After a reset, the contents of the processor status register (PS) are undefined except for the I flag which is "1".

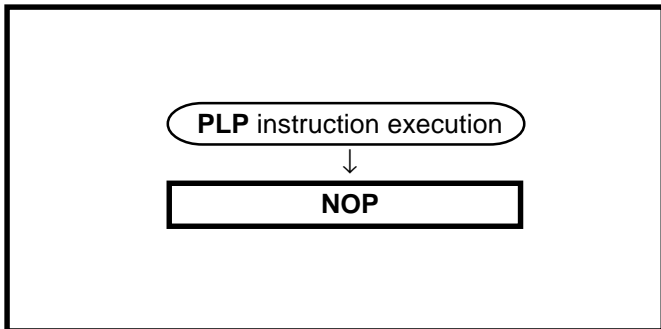


**Fig. 3.3.4 Initialization of processor status register**

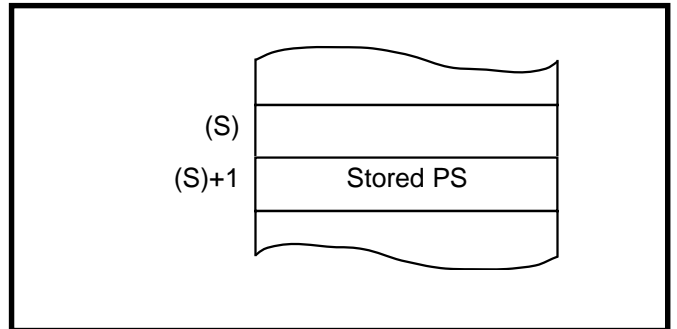
② **How to reference the processor status register**

To reference the contents of the processor status register (PS), execute the **PHP** instruction once then read the contents of (S+1). If necessary, execute the **PLP** instruction to return the PS to its original status.

A **NOP** instruction should be executed after every **PLP** instruction.



**Fig. 3.3.5 Sequence of PLP instruction execution**



**Fig. 3.3.6 Stack memory contents after PHP instruction execution**

#### (2) BRK instruction

① **Interrupt priority level**

When the BRK instruction is executed with the following conditions satisfied, the interrupt execution is started from the address of interrupt vector which has the highest priority.

- Interrupt request bit and interrupt enable bit are set to "1".
- Interrupt disable flag (I) is set to "1" to disable interrupt.



**(3) Decimal calculations****① Execution of decimal calculations**

The **ADC** and **SBC** are the only instructions which will yield proper decimal notation, set the decimal mode flag (D) to “1” with the **SED** instruction. After executing the **ADC** or **SBC** instruction, execute another instruction before executing the **SEC**, **CLC**, or **CLD** instruction.

**② Notes on status flag in decimal mode**

When decimal mode is selected, the values of three of the flags in the status register (the N, V, and Z flags) are invalid after a **ADC** or **SBC** instruction is executed.

The carry flag (C) is set to “1” if a carry is generated as a result of the calculation, or is cleared to “0” if a borrow is generated. To determine whether a calculation has generated a carry, the C flag must be initialized to “0” before each calculation. To check for a borrow, the C flag must be initialized to “1” before each calculation.

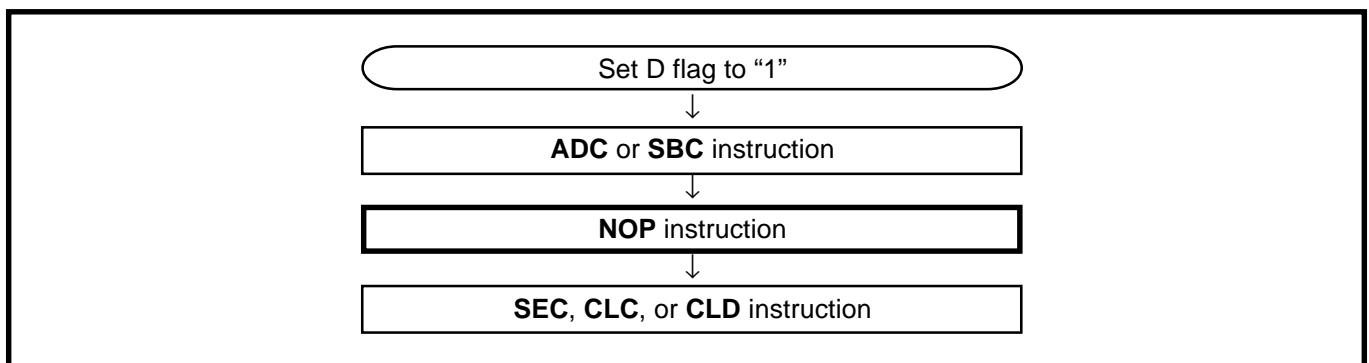


Fig. 3.3.7 Status flag at decimal calculations

**(4) JMP instruction**

When using the **JMP** instruction in indirect addressing mode, do not specify the last address on a page as an indirect address.

**(5) Multiplication and Division Instructions**

- The index X mode (T) and the decimal mode (D) flags do not affect the **MUL** and **DIV** instruction.
- The execution of these instructions does not change the contents of the processor status register.

**(6) Ports**

The contents of the port direction registers cannot be read. The following cannot be used:

- The data transfer instruction (**LDA**, etc.)
- The operation instruction when the index X mode flag (T) is “1”
- The addressing mode which uses the value of a direction register as an index
- The bit-test instruction (**BBC** or **BBS**, etc.) to a direction register
- The read-modify-write instructions (**ROR**, **CLB**, or **SEB**, etc.) to a direction register.

Use instructions such as **LDM** and **STA**, etc., to set the port direction registers.

**(7) Instruction Execution Time**

The instruction execution time is obtained by multiplying the frequency of the internal clock  $\phi$  by the number of cycles needed to execute an instruction.

The number of cycles required to execute an instruction is shown in the list of machine instructions.

The frequency of the internal clock  $\phi$  is half of the  $X_{IN}$  frequency in high-speed mode.

# APPENDIX

## 3.3 Notes on use

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### 3.3.15 EPROM Version/One Time PROM Version/Flash Memory Version

The CNVss pin is connected to the internal memory circuit block by a low-ohmic resistance, since it has the multiplexed function to be a programmable power source pin ( $V_{PP}$  pin) as well.

To improve the noise reduction, connect a track between CNVss pin and Vss pin or Vcc pin with 1 to 10 k $\Omega$  resistance.

The mask ROM version track of CNVss pin has no operational interference even if it is connected to Vss pin or Vcc pin via a resistor.

### 3.3.16 Handling of Source Pins

In order to avoid a latch-up occurrence, connect a capacitor suitable for high frequencies as bypass capacitor between power source pin ( $V_{CC}$  pin) and GND pin ( $V_{SS}$  pin) and between power source pin ( $V_{CC}$  pin) and analog power source input pin ( $AV_{SS}$  pin). Besides, connect the capacitor to as close as possible. For bypass capacitor which should not be located too far from the pins to be connected, a ceramic capacitor of 0.01  $\mu$ F–0.1 $\mu$ F is recommended.

### 3.3.17 Differences between 3850 group (standard) and 3850 group (spec. H)

- (1) The absolute maximum ratings of 3850 group (spec. H) is smaller than that of 3850 group (standard).
  - Power source voltage  $V_{CC} = 0.3$  to 6.5 V
  - CNVss input voltage  $V_i = -0.3$  to  $V_{CC} + 0.3$  V
- (2) The oscillation circuit constants of  $X_{IN}$ - $X_{OUT}$ ,  $X_{CIN}$ - $X_{COUT}$  may be some differences between 3850 group (standard) and 3850 group (spec. H).
- (3) Do not write any data to the reserved area and the reserved bit. (Do not change the contents after reset.)
- (4) Fix bit 3 of the CPU mode register to "1".
- (5) Be sure to perform the termination of unused pins.

### 3.4 Countermeasures against noise

Countermeasures against noise are described below. The following countermeasures are effective against noise in theory, however, it is necessary not only to take measures as follows but to evaluate before actual use.

#### 3.4.1 Shortest wiring length

The wiring on a printed circuit board can function as an antenna which feeds noise into the microcomputer. The shorter the total wiring length (by mm unit), the less the possibility of noise insertion into a microcomputer.

##### (1) Package

Select the smallest possible package to make the total wiring length short.

##### ● Reason

The wiring length depends on a microcomputer package. Use of a small package, for example QFP and not DIP, makes the total wiring length short to reduce influence of noise.

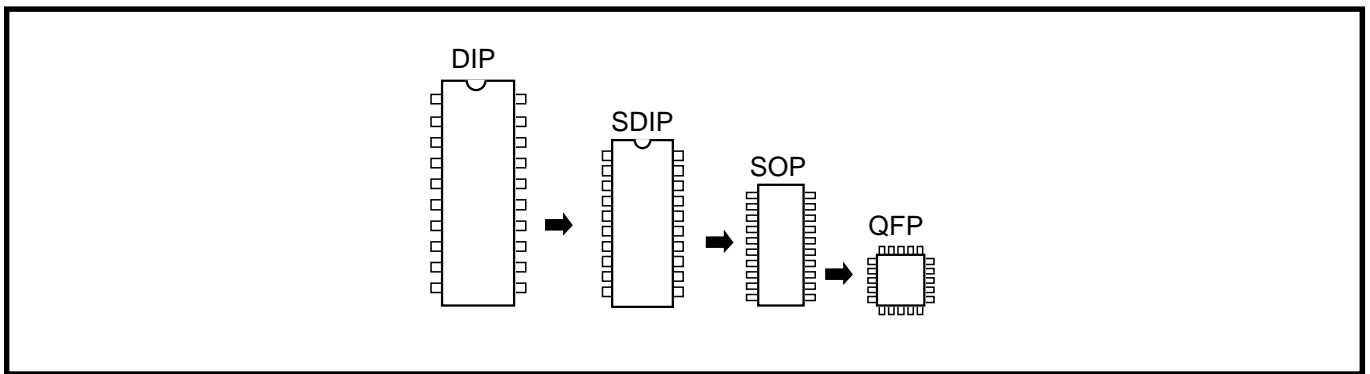


Fig. 3.4.1 Selection of packages

##### (2) Wiring for $\overline{\text{RESET}}$ pin

Make the length of wiring which is connected to the  $\overline{\text{RESET}}$  pin as short as possible. Especially, connect a capacitor across the  $\overline{\text{RESET}}$  pin and the  $V_{\text{ss}}$  pin with the shortest possible wiring (within 20mm).

##### ● Reason

The width of a pulse input into the  $\overline{\text{RESET}}$  pin is determined by the timing necessary conditions. If noise having a shorter pulse width than the standard is input to the  $\overline{\text{RESET}}$  pin, the reset is released before the internal state of the microcomputer is completely initialized. This may cause a program runaway.

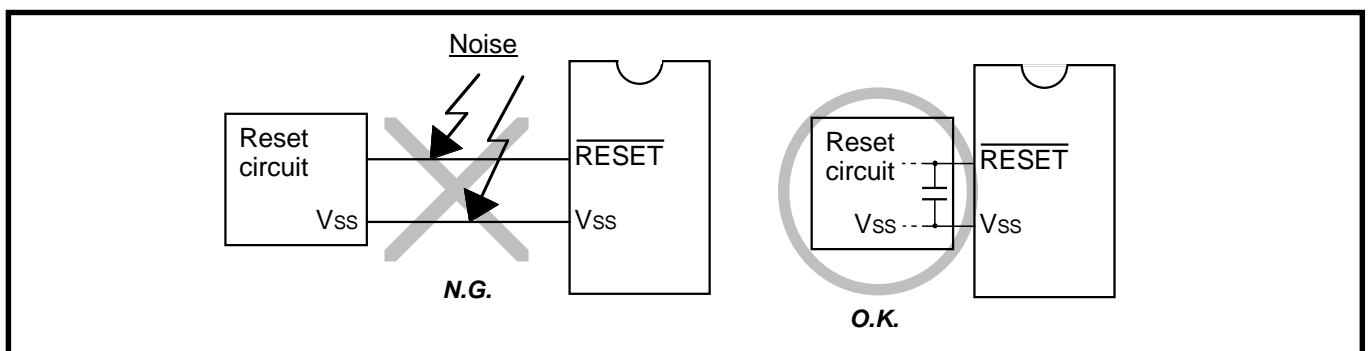


Fig. 3.4.2 Wiring for the  $\overline{\text{RESET}}$  pin

# APPENDIX

## 3.4 Countermeasures against noise

### (3) Wiring for clock input/output pins

- Make the length of wiring which is connected to clock I/O pins as short as possible.
- Make the length of wiring (within 20 mm) across the grounding lead of a capacitor which is connected to an oscillator and the VSS pin of a microcomputer as short as possible.
- Separate the VSS pattern only for oscillation from other VSS patterns.

● Reason

If noise enters clock I/O pins, clock waveforms may be deformed. This may cause a program failure or program runaway. Also, if a potential difference is caused by the noise between the VSS level of a microcomputer and the VSS level of an oscillator, the correct clock will not be input in the microcomputer.

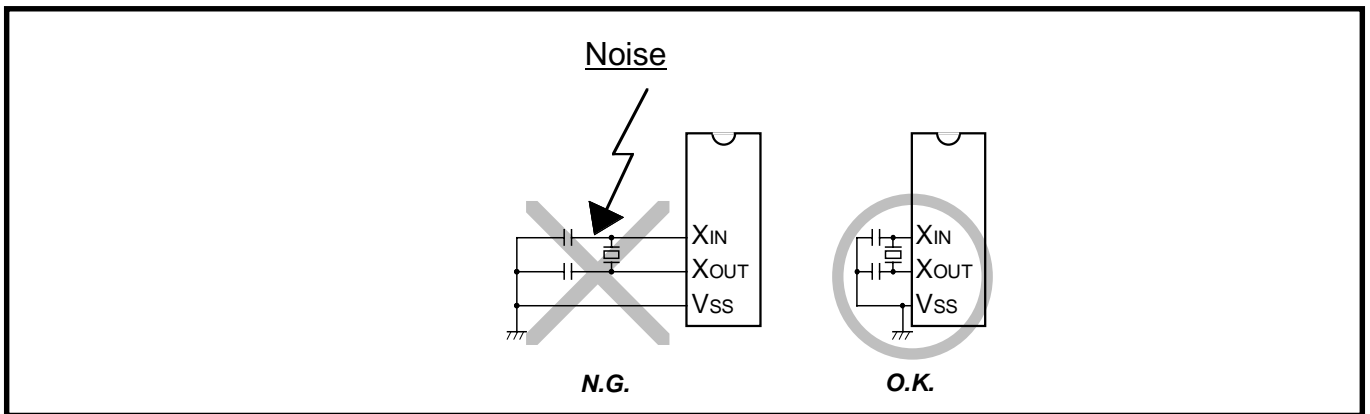


Fig. 3.4.3 Wiring for clock I/O pins

### (4) Wiring to CNVss pin

Connect the CNVss pin to the VSS pin with the shortest possible wiring.

● Reason

The processor mode of a microcomputer is influenced by a potential at the CNVss pin. If a potential difference is caused by the noise between pins CNVss and VSS, the processor mode may become unstable. This may cause a microcomputer malfunction or a program runaway.

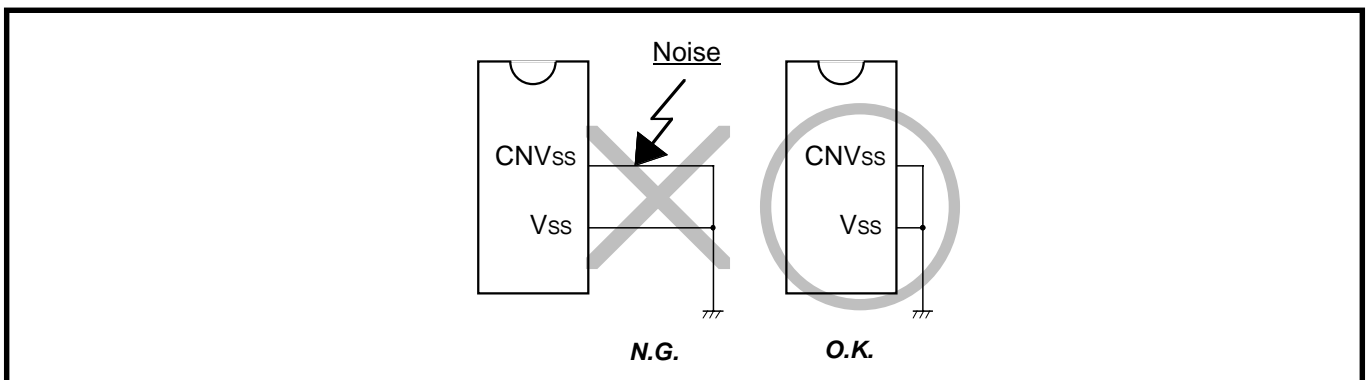


Fig. 3.4.4 Wiring for CNVss pin

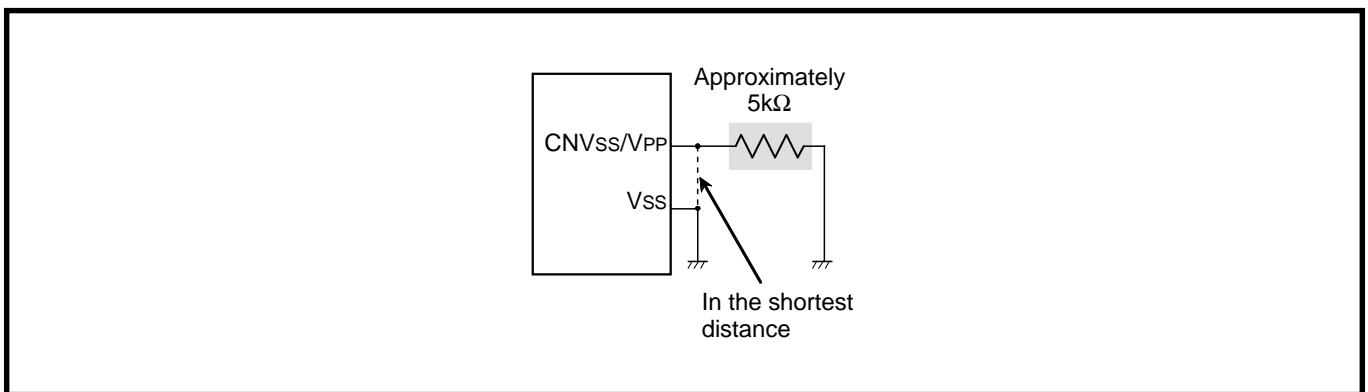
## 3.4 Countermeasures against noise

- (5) **Wiring to VPP pin of One Time PROM version, EPROM version, and Flash memory version**  
 Connect an approximately 5 kΩ resistor to the VPP pin the shortest possible in series and also to the VSS pin. When not connecting the resistor, make the length of wiring between the VPP pin and the VSS pin the shortest possible.

**Note:** Even when a circuit which included an approximately 5 kΩ resistor is used in the Mask ROM version, the microcomputer operates correctly.

● **Reason**

The VPP pin of the One Time PROM, the EPROM version, and the flash memory version is the power source input pin for the built-in PROM. When programming in the built-in PROM, the impedance of the VPP pin is low to allow the electric current for writing flow into the PROM. Because of this, noise can enter easily. If noise enters the VPP pin, abnormal instruction codes or data are read from the built-in PROM, which may cause a program runaway.

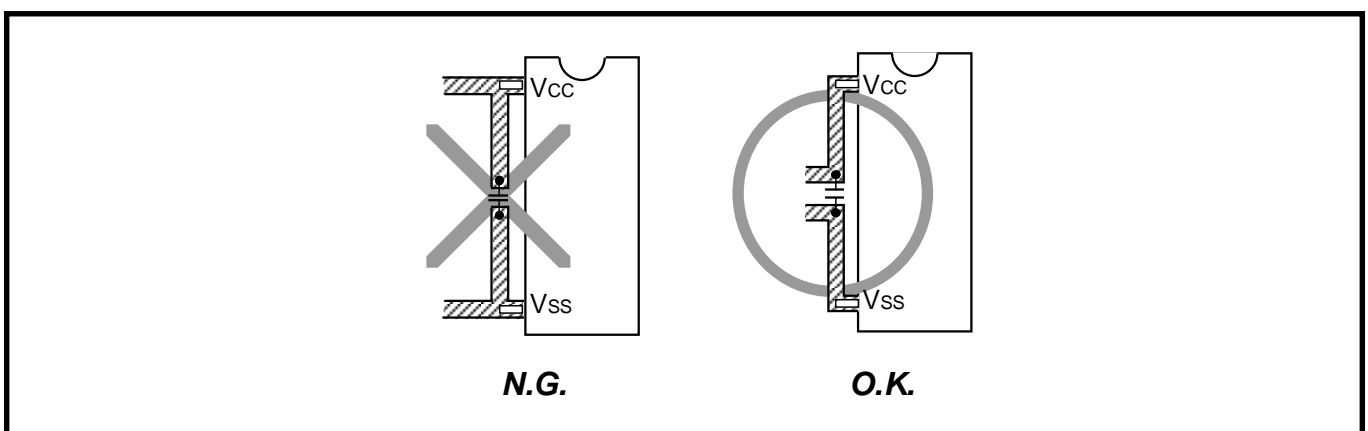


**Fig. 3.4.5 Wiring for the VPP pin of the One Time PROM version, the EPROM version, and the flash memory version**

### 3.4.2 Connection of bypass capacitor across Vss line and Vcc line

Connect an approximately 0.1 μF bypass capacitor across the VSS line and the VCC line as follows:

- Connect a bypass capacitor across the VSS pin and the VCC pin at equal length.
- Connect a bypass capacitor across the VSS pin and the VCC pin with the shortest possible wiring.
- Use lines with a larger diameter than other signal lines for VSS line and VCC line.
- Connect the power source wiring via a bypass capacitor to the VSS pin and the VCC pin.



**Fig. 3.4.6 Bypass capacitor across the Vss line and the Vcc line**

# APPENDIX

## 3.4 Countermeasures against noise

### 3.4.3 Wiring to analog input pins

- Connect an approximately 100  $\Omega$  to 1 k $\Omega$  resistor to an analog signal line which is connected to an analog input pin in series. Besides, connect the resistor to the microcomputer as close as possible.
- Connect an approximately 1000 pF capacitor across the Vss pin and the analog input pin. Besides, connect the capacitor to the Vss pin as close as possible. Also, connect the capacitor across the analog input pin and the Vss pin at equal length.

#### ● Reason

Signals which is input in an analog input pin (such as an A-D converter/comparator input pin) are usually output signals from sensor. The sensor which detects a change of event is installed far from the printed circuit board with a microcomputer, the wiring to an analog input pin is longer necessarily. This long wiring functions as an antenna which feeds noise into the microcomputer, which causes noise to an analog input pin.

If a capacitor between an analog input pin and the Vss pin is grounded at a position far away from the Vss pin, noise on the GND line may enter a microcomputer through the capacitor.

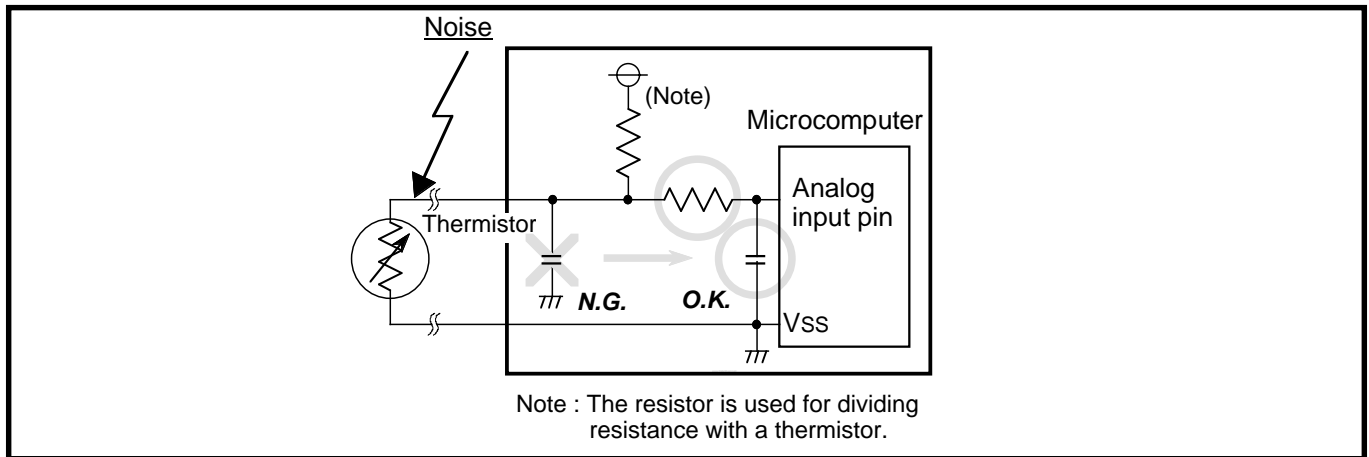


Fig. 3.4.7 Analog signal line and a resistor and a capacitor

### 3.4.4 Oscillator concerns

Take care to prevent an oscillator that generates clocks for a microcomputer operation from being affected by other signals.

#### (1) Keeping oscillator away from large current signal lines

Install a microcomputer (and especially an oscillator) as far as possible from signal lines where a current larger than the tolerance of current value flows.

##### ● Reason

In the system using a microcomputer, there are signal lines for controlling motors, LEDs, and thermal heads or others. When a large current flows through those signal lines, strong noise occurs because of mutual inductance.

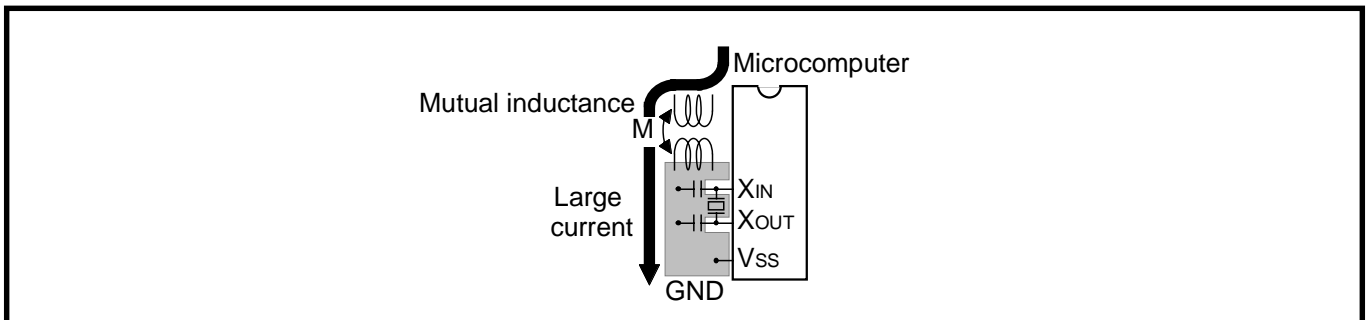


Fig. 3.4.8 Wiring for a large current signal line

#### (2) Installing oscillator away from signal lines where potential levels change frequently

Install an oscillator and a connecting pattern of an oscillator away from signal lines where potential levels change frequently. Also, do not cross such signal lines over the clock lines or the signal lines which are sensitive to noise.

##### ● Reason

Signal lines where potential levels change frequently (such as the CNTR pin signal line) may affect other lines at signal rising edge or falling edge. If such lines cross over a clock line, clock waveforms may be deformed, which causes a microcomputer failure or a program runaway.

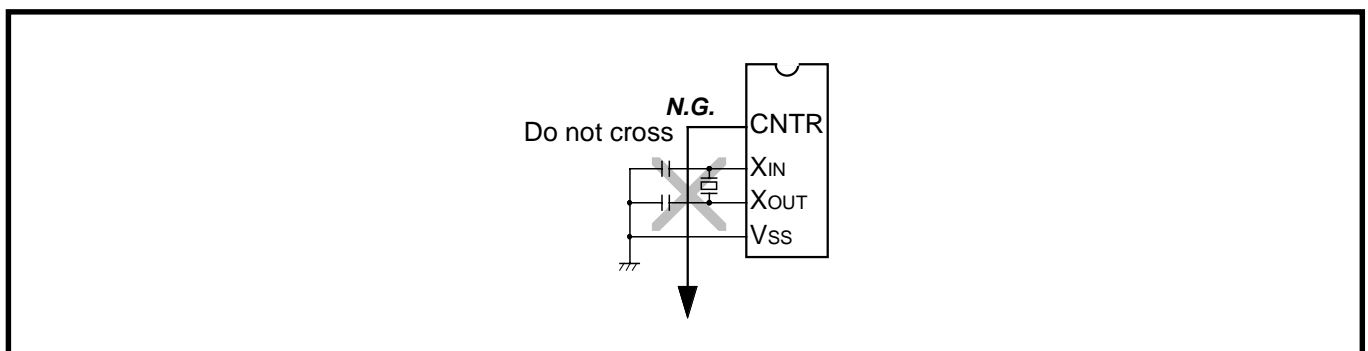


Fig. 3.4.9 Wiring of  $\overline{\text{RESET}}$  pin

# APPENDIX

## 3.4 Countermeasures against noise

### (3) Oscillator protection using Vss pattern

As for a two-sided printed circuit board, print a Vss pattern on the underside (soldering side) of the position (on the component side) where an oscillator is mounted. Connect the Vss pattern to the microcomputer Vss pin with the shortest possible wiring. Besides, separate this Vss pattern from other Vss patterns.

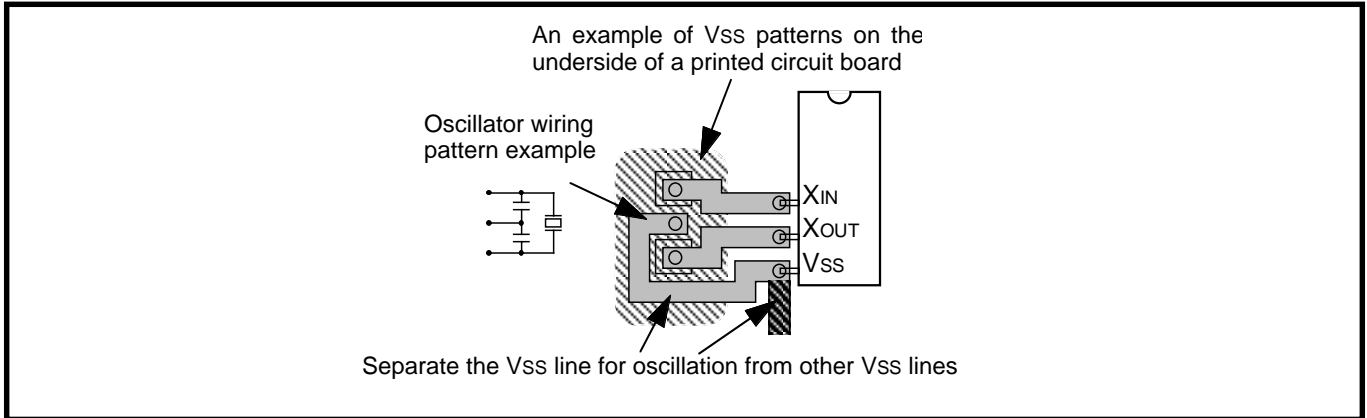


Fig. 3.4.10 Vss pattern on the underside of an oscillator

### 3.4.5 Setup for I/O ports

Setup I/O ports using hardware and software as follows:

#### <Hardware>

- Connect a resistor of 100  $\Omega$  or more to an I/O port in series.

#### <Software>

- As for an input port, read data several times by a program for checking whether input levels are equal or not.
- As for an output port, since the output data may reverse because of noise, rewrite data to its port latch at fixed periods.
- Rewrite data to direction registers at fixed periods.

**Note:** When a direction register is set for input port again at fixed periods, a several-nanosecond short pulse may be output from this port. If this is undesirable, connect a capacitor to this port to remove the noise pulse.

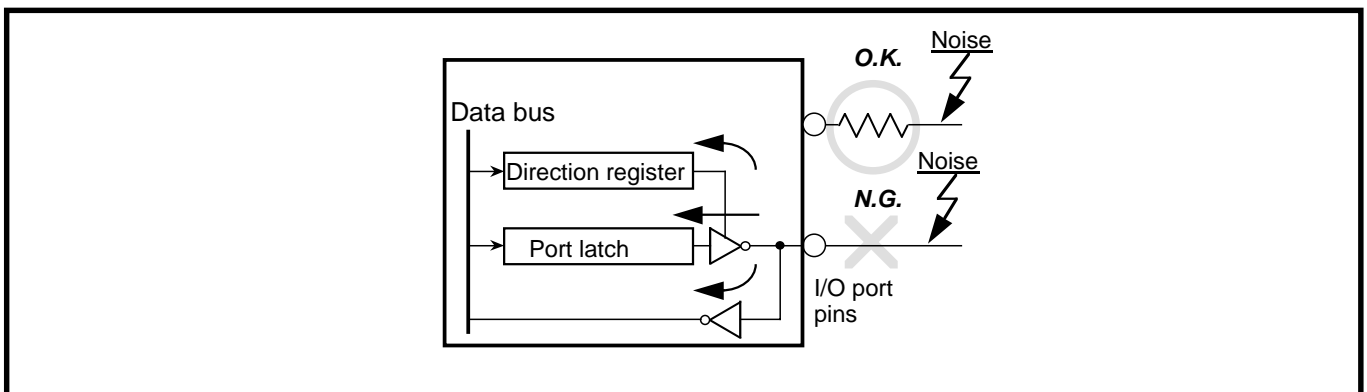


Fig. 3.4.11 Setup for I/O ports



### 3.4.6 Providing of watchdog timer function by software

If a microcomputer runs away because of noise or others, it can be detected by a software watchdog timer and the microcomputer can be reset to normal operation. This is equal to or more effective than program runaway detection by a hardware watchdog timer. The following shows an example of a watchdog timer provided by software.

In the following example, to reset a microcomputer to normal operation, the main routine detects errors of the interrupt processing routine and the interrupt processing routine detects errors of the main routine. This example assumes that interrupt processing is repeated multiple times in a single main routine processing.

#### <The main routine>

- Assigns a single byte of RAM to a software watchdog timer (SWDT) and writes the initial value N in the SWDT once at each execution of the main routine. The initial value N should satisfy the following condition:  

$$N+1 \geq (\text{Counts of interrupt processing executed in each main routine})$$
 As the main routine execution cycle may change because of an interrupt processing or others, the initial value N should have a margin.
- Watches the operation of the interrupt processing routine by comparing the SWDT contents with counts of interrupt processing after the initial value N has been set.
- Detects that the interrupt processing routine has failed and determines to branch to the program initialization routine for recovery processing in the following case:  
 If the SWDT contents do not change after interrupt processing.

#### <The interrupt processing routine>

- Decrements the SWDT contents by 1 at each interrupt processing.
- Determines that the main routine operates normally when the SWDT contents are reset to the initial value N at almost fixed cycles (at the fixed interrupt processing count).
- Detects that the main routine has failed and determines to branch to the program initialization routine for recovery processing in the following case:  
 If the SWDT contents are not initialized to the initial value N but continued to decrement and if they reach 0 or less.

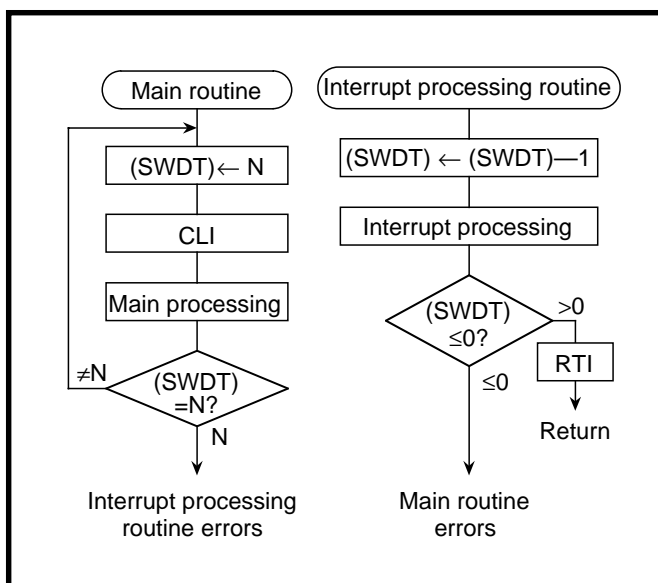


Fig. 3.4.12 Watchdog timer by software

# APPENDIX

## 3.5 List of registers

### 3.5 List of registers

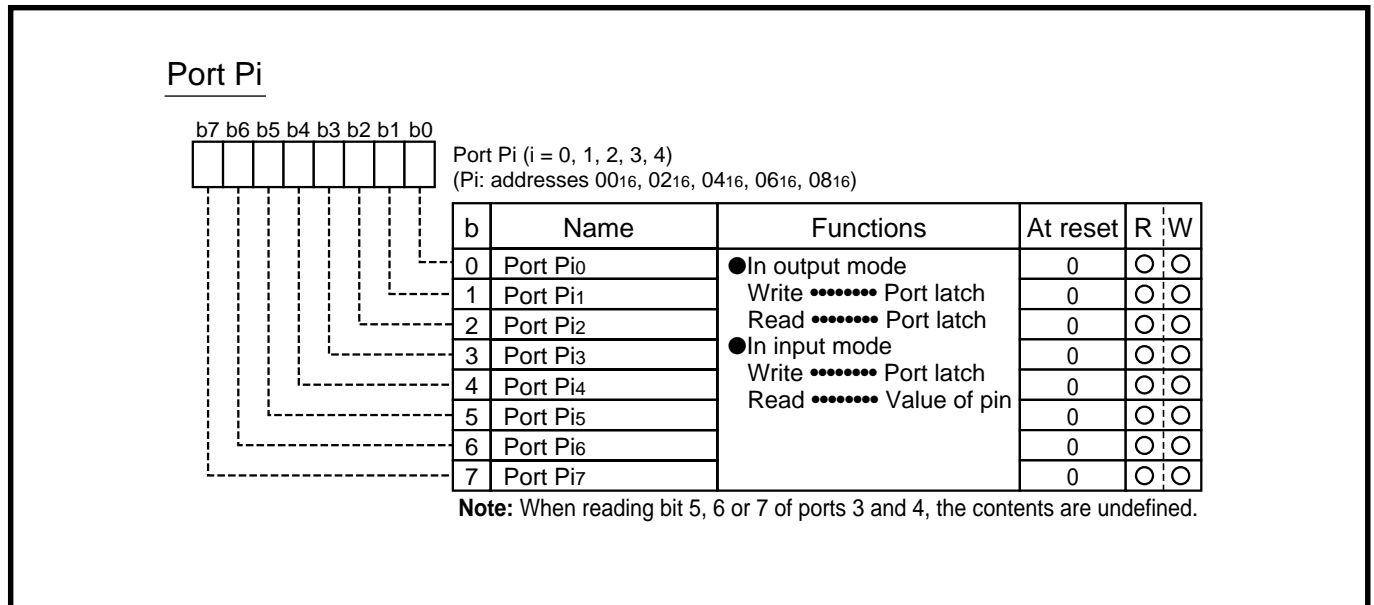


Fig. 3.5.1 Structure of Port Pi

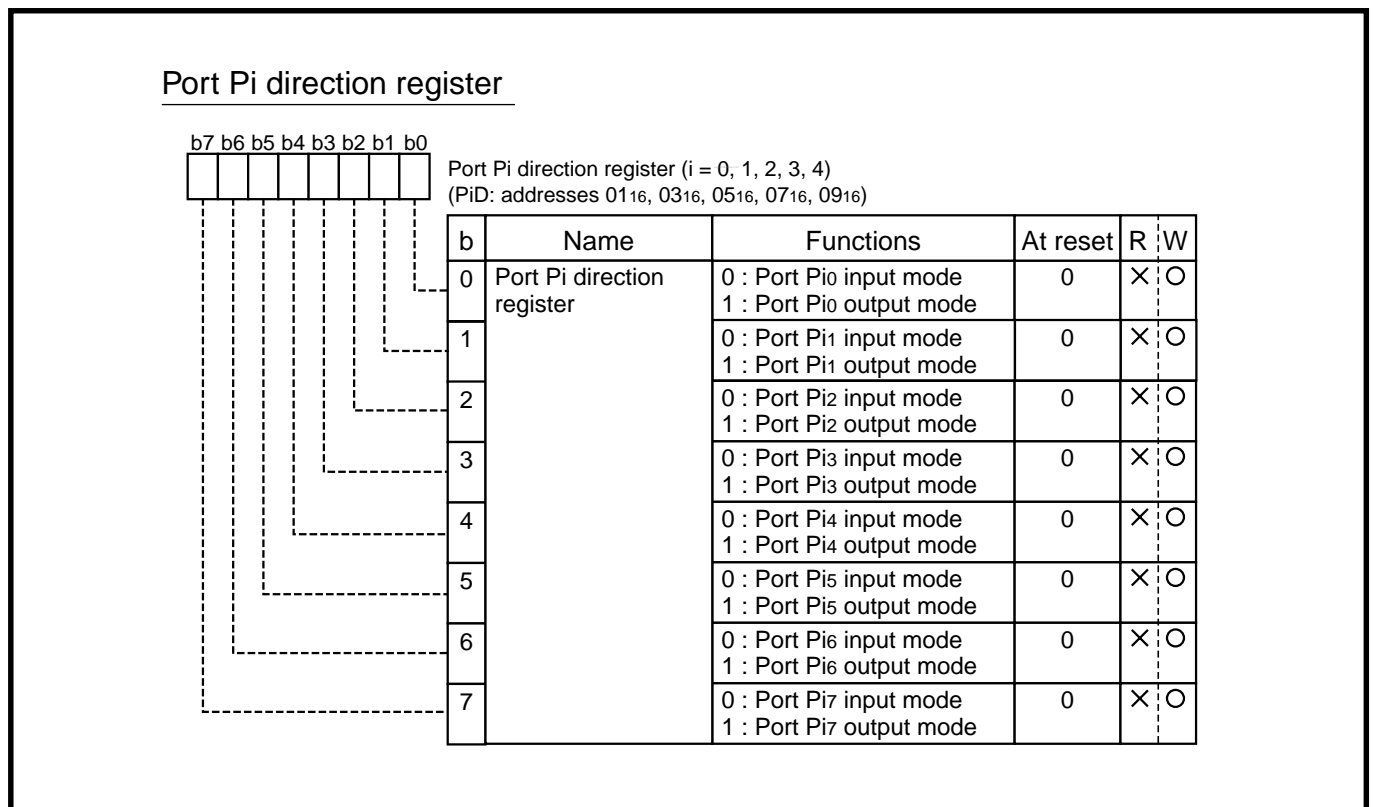


Fig. 3.5.2 Structure of Port Pi direction register

### Serial I/O2 control register 1

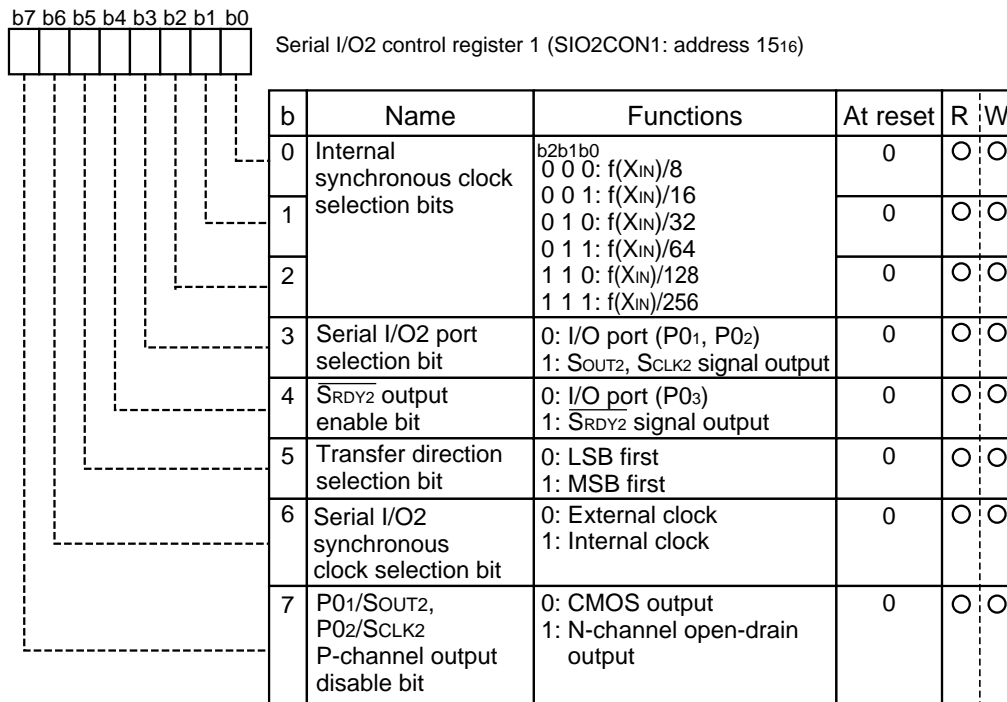


Fig. 3.5.3 Structure of Serial I/O2 control register 1

### Serial I/O2 control register 2

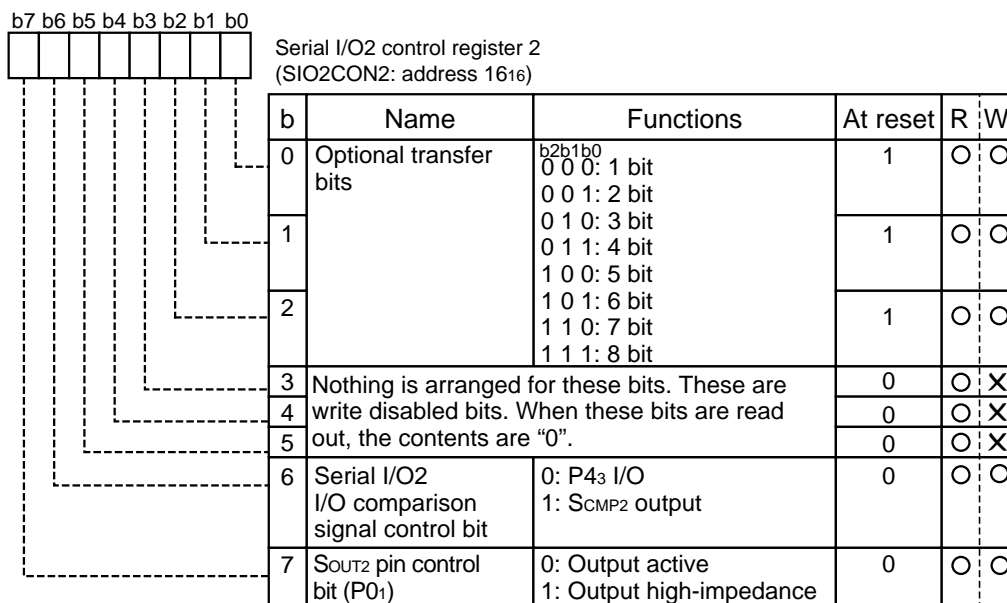


Fig. 3.5.4 Structure of Serial I/O2 control register 2

# APPENDIX

## 3.5 List of registers

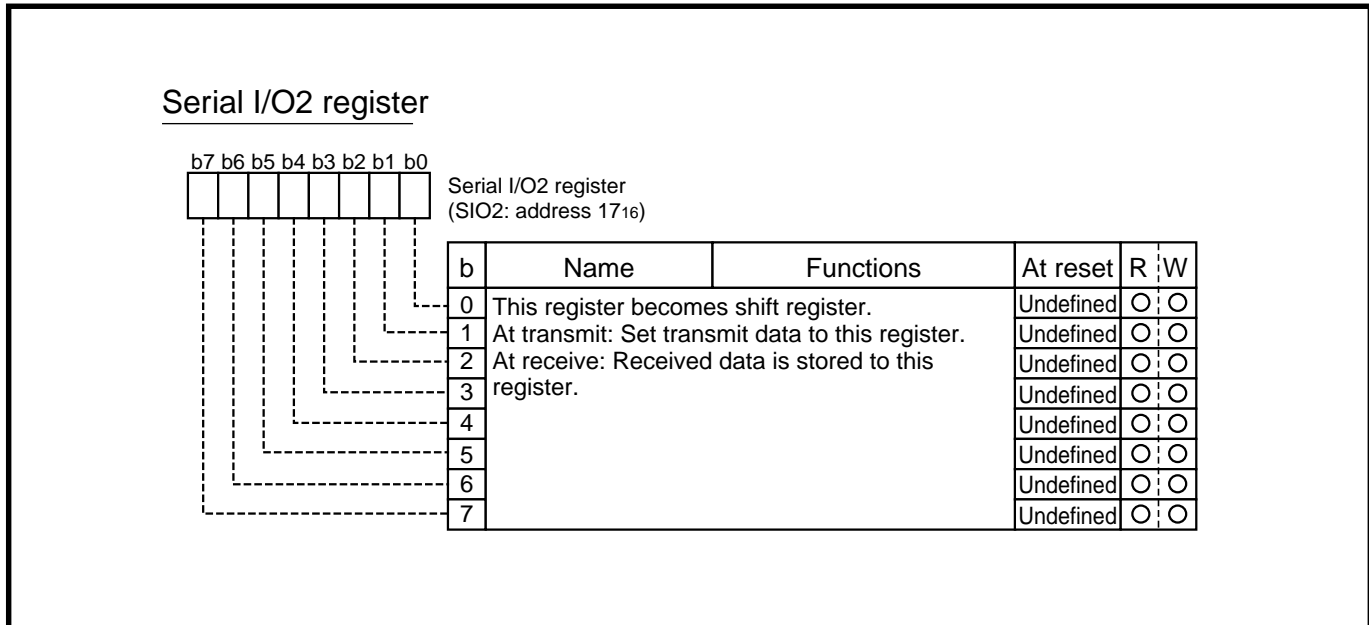


Fig. 3.5.5 Structure of Serial I/O2 register

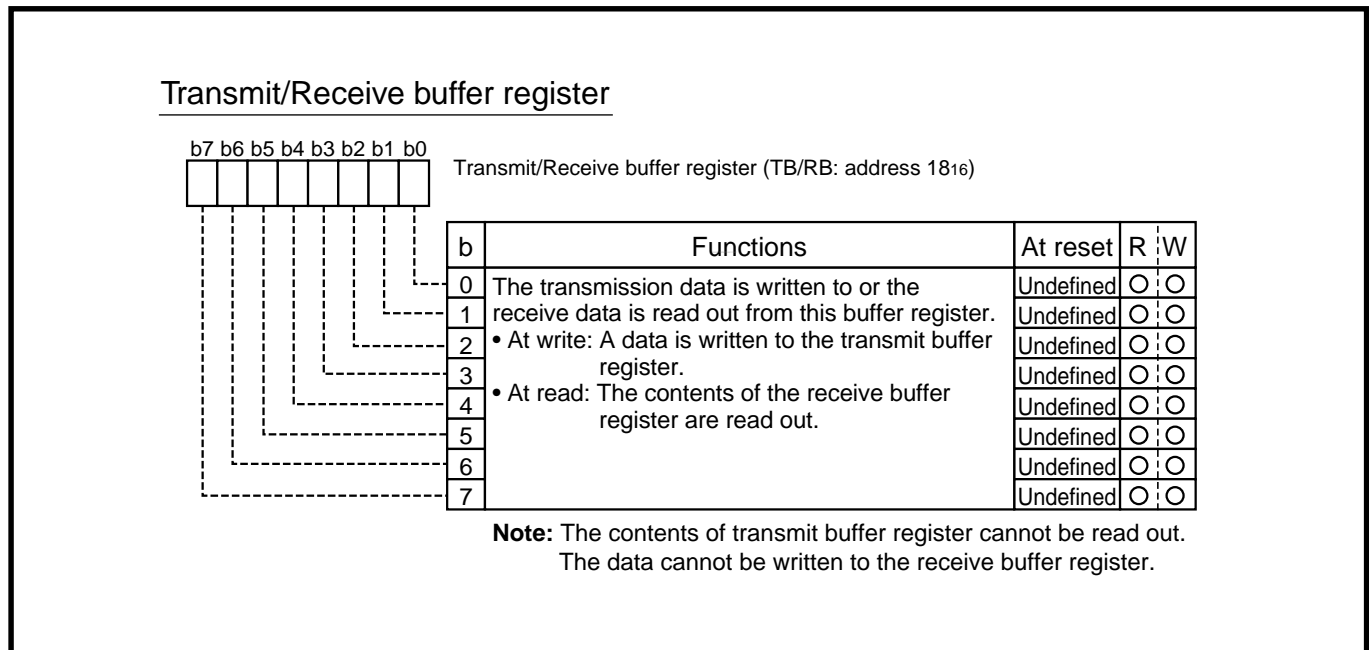


Fig. 3.5.6 Structure of Transmit/Receive buffer register

### Serial I/O1 status register

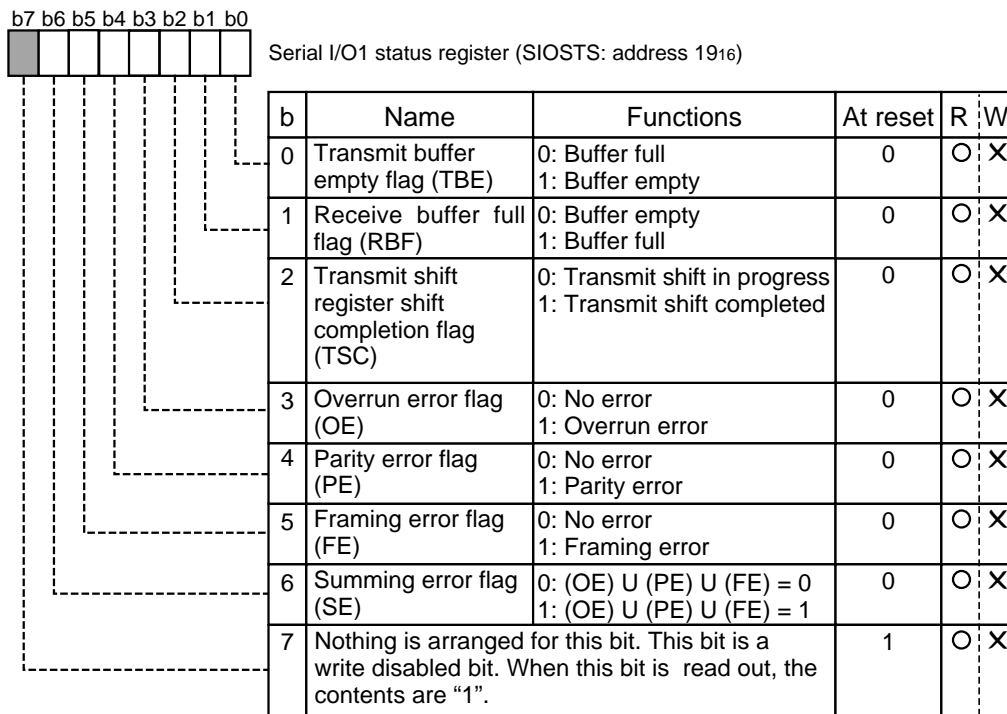


Fig. 3.5.7 Structure of Serial I/O1 status register

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## 3.5 List of registers

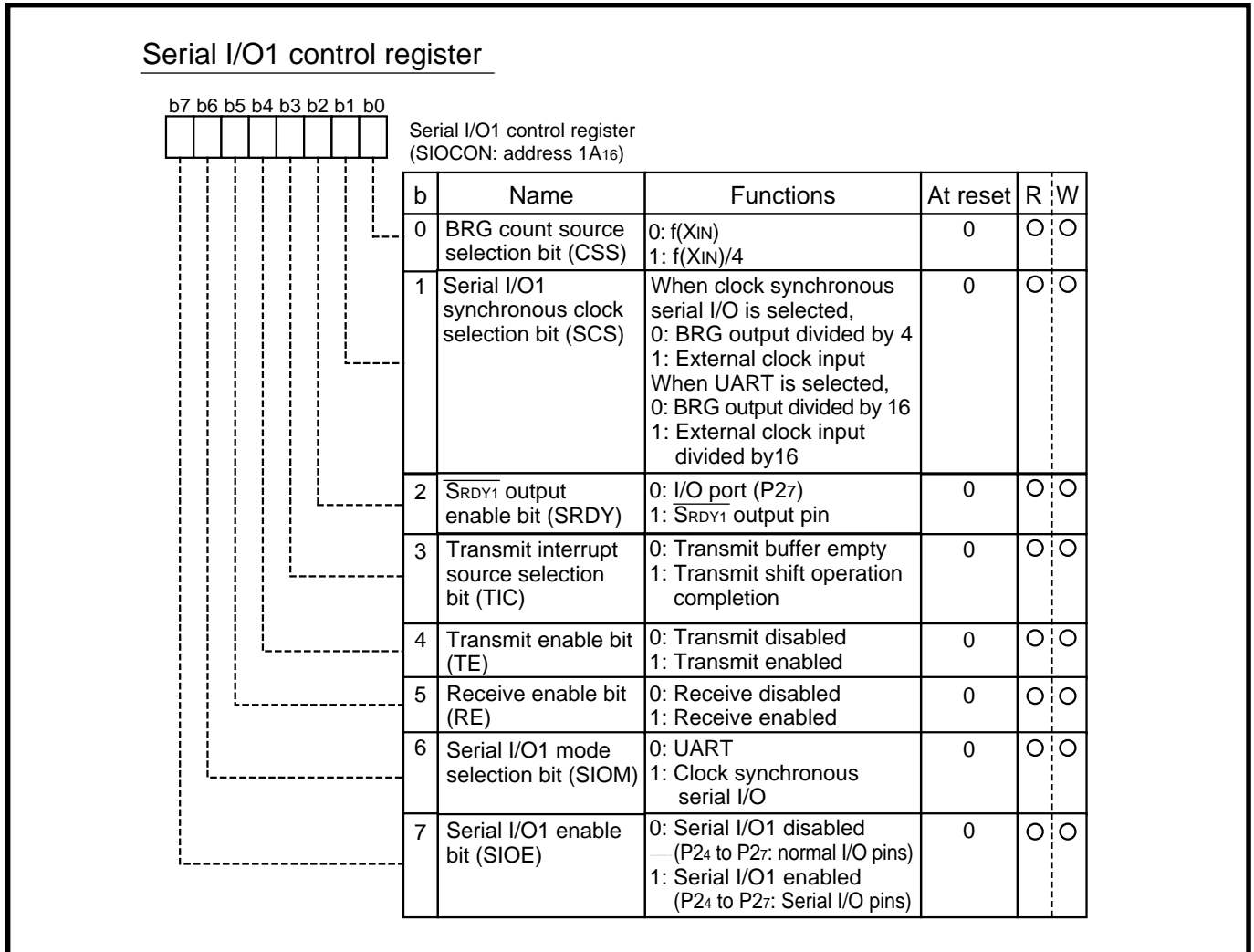


Fig. 3.5.8 Structure of Serial I/O1 control register

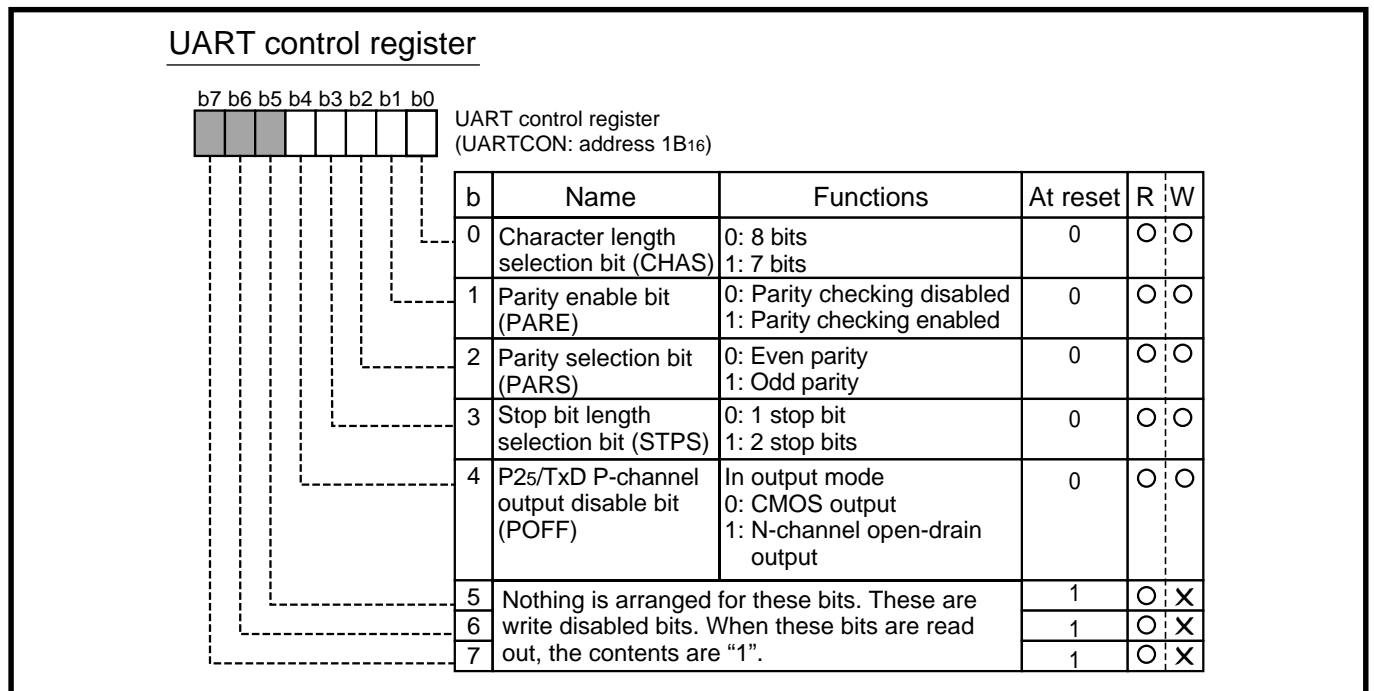


Fig. 3.5.9 Structure of UART control register

### Baud rate generator

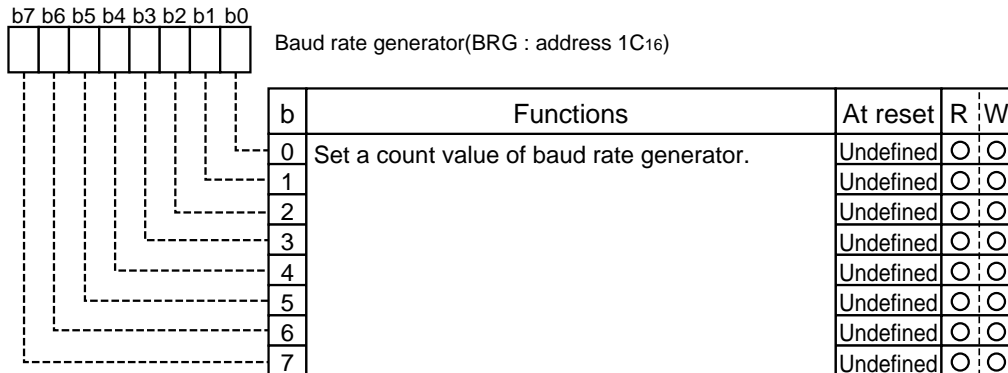


Fig. 3.5.10 Structure of Baud rate generator

### PWM control register

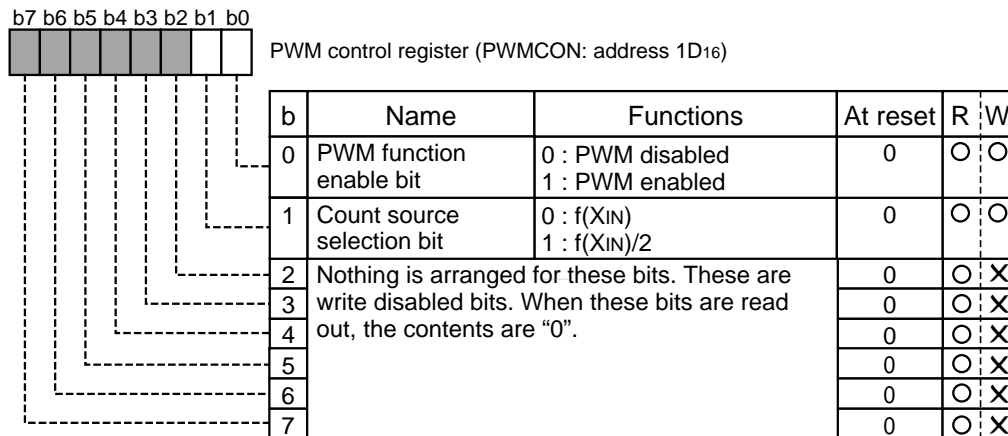


Fig. 3.5.11 Structure of PWM control register

### PWM prescaler

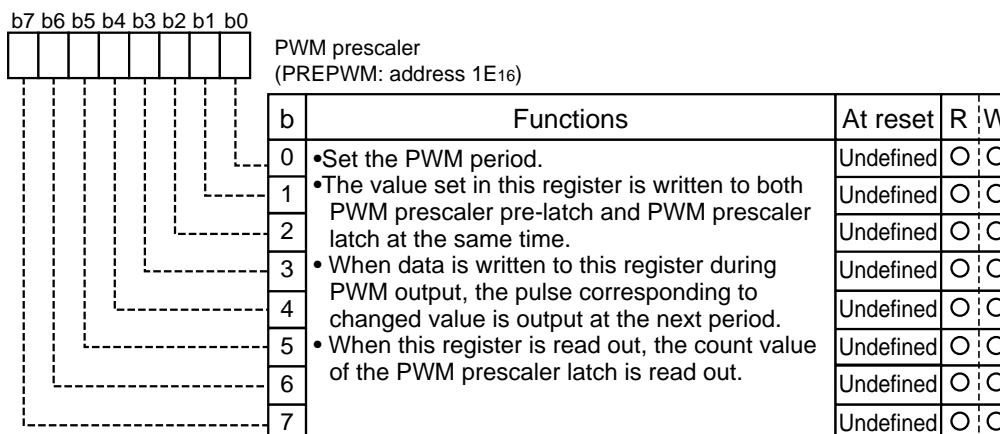


Fig. 3.5.12 Structure of PWM prescaler

# APPENDIX

## 3.5 List of registers

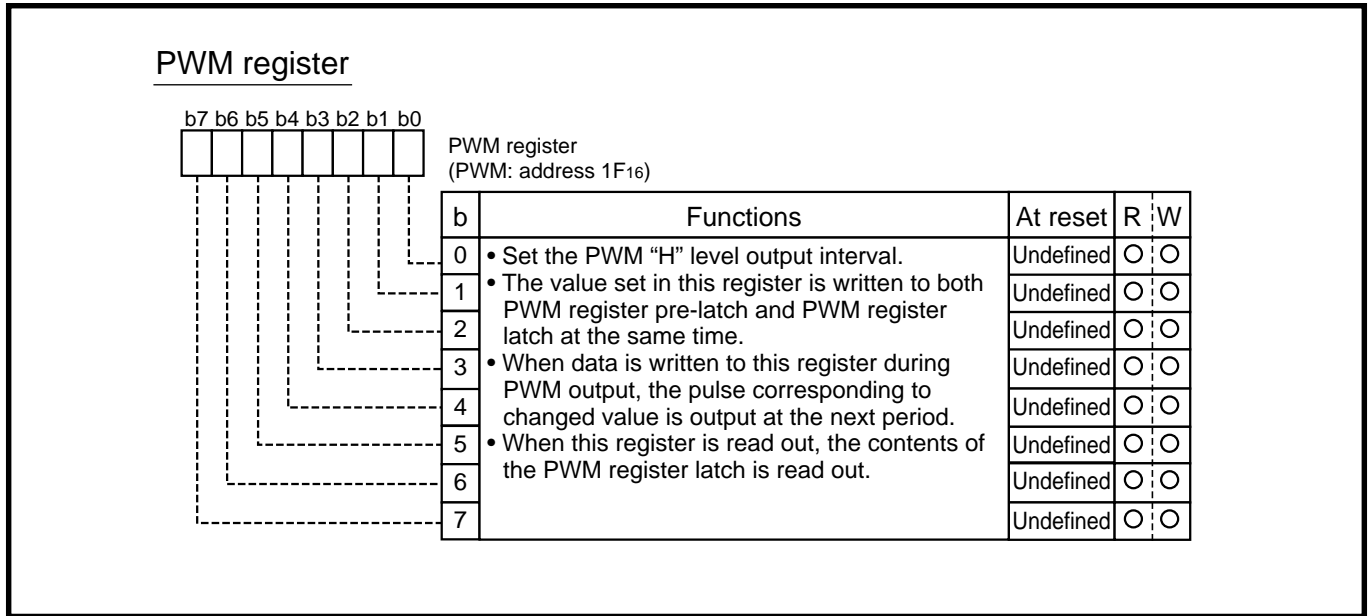


Fig. 3.5.13 Structure of PWM register

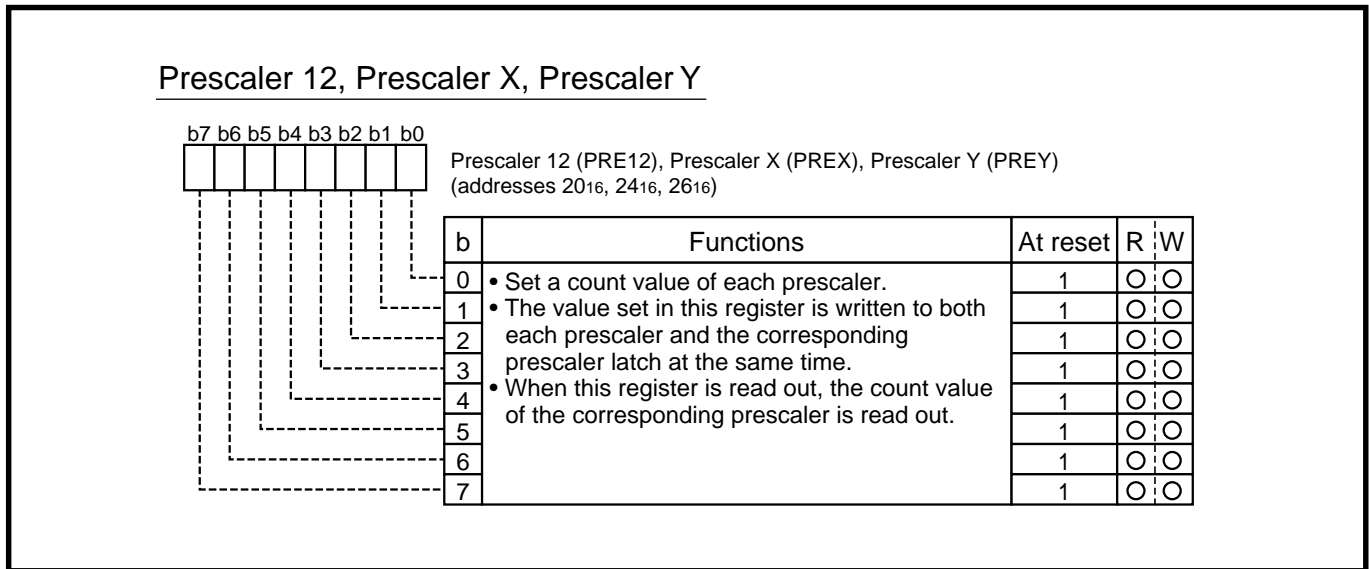


Fig. 3.5.14 Structure of Prescaler 12, Prescaler X, Prescaler Y



### Timer 1

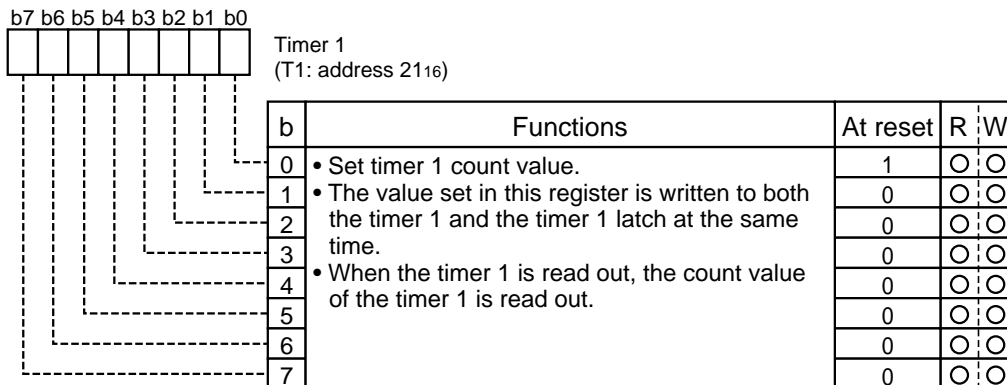


Fig. 3.5.15 Structure of Timer 1

### Timer 2

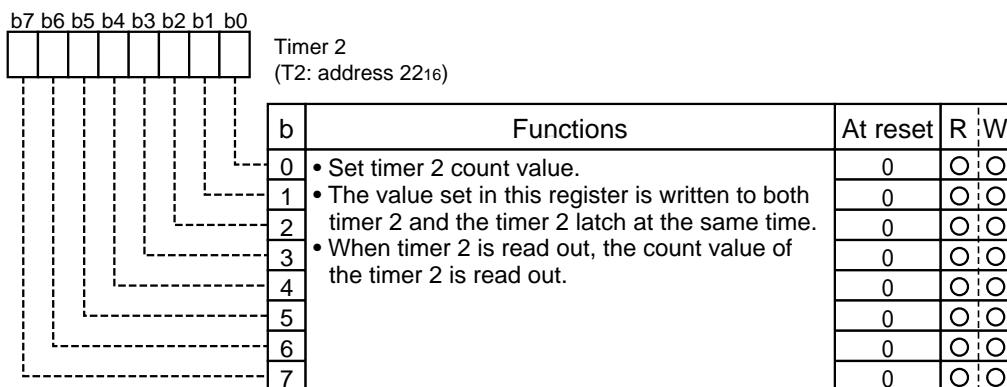


Fig. 3.5.16 Structure of Timer 2

# APPENDIX

## 3.5 List of registers

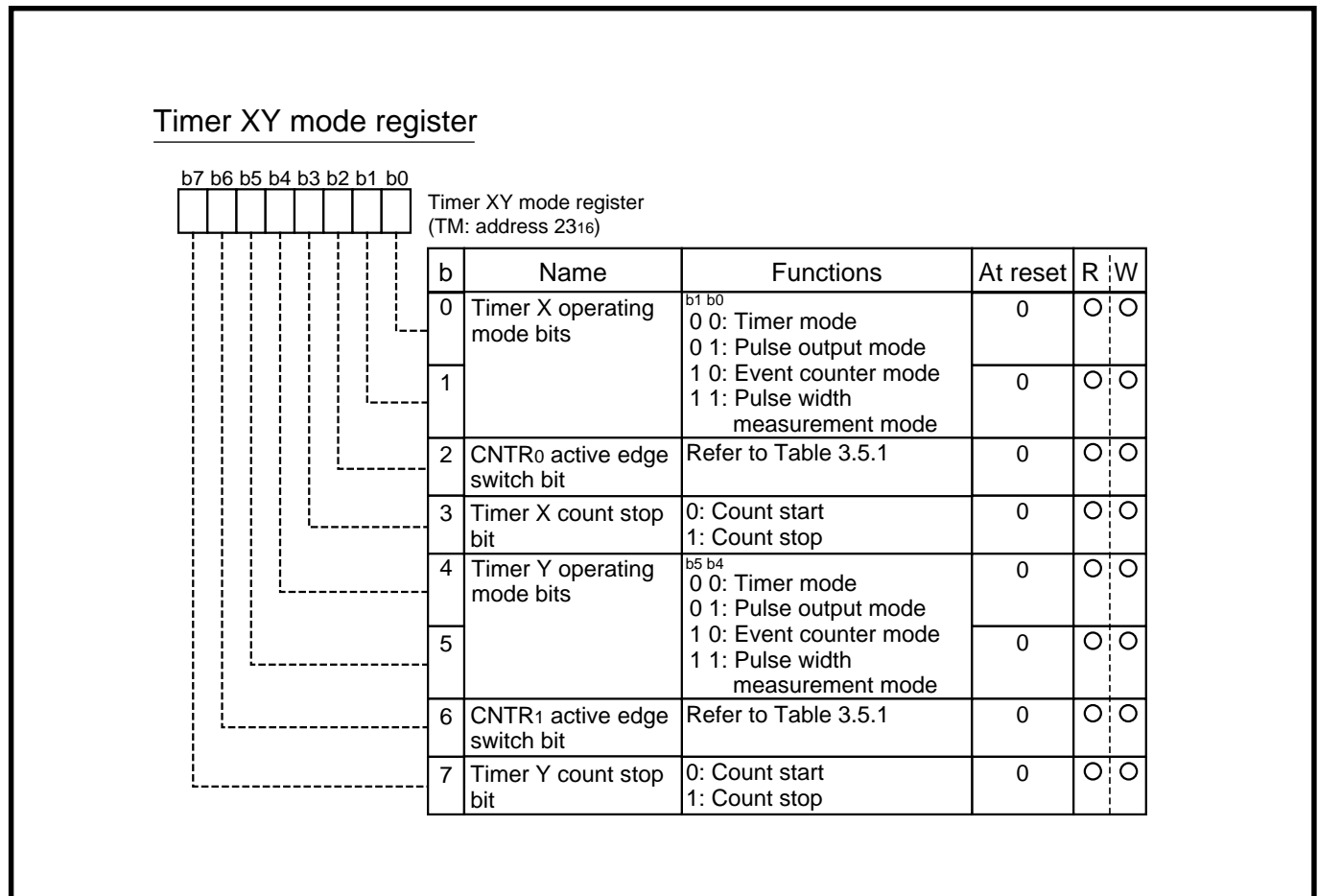


Fig. 3.5.17 Structure of Timer XY mode register

Table 3.5.1 CNTR<sub>0</sub>/CNTR<sub>1</sub> active edge switch bit function

Timer X /Timer Y operation modes	Set value	Timer function	CNTR <sub>0</sub> / CNTR <sub>1</sub> interrupt request occurrence source
Timer mode	“0”	No influence to timer count	CNTR <sub>0</sub> /CNTR <sub>1</sub> input signal falling edge
	“1”	No influence to timer count	CNTR <sub>0</sub> /CNTR <sub>1</sub> input signal rising edge
Pulse output mode	“0”	Pulse output start: Beginning at “H” level	Output signal falling edge count
	“1”	Pulse output start: Beginning at “L” level	Output signal rising edge count
Event counter mode	“0”	Rising edge count	Input signal falling edge count
	“1”	Falling edge count	Input signal rising edge count
Pulse width measurement mode	“0”	“H” level width measurement	Input signal falling edge count
	“1”	“L” level width measurement	Input signal rising edge count

### Timer X, Timer Y

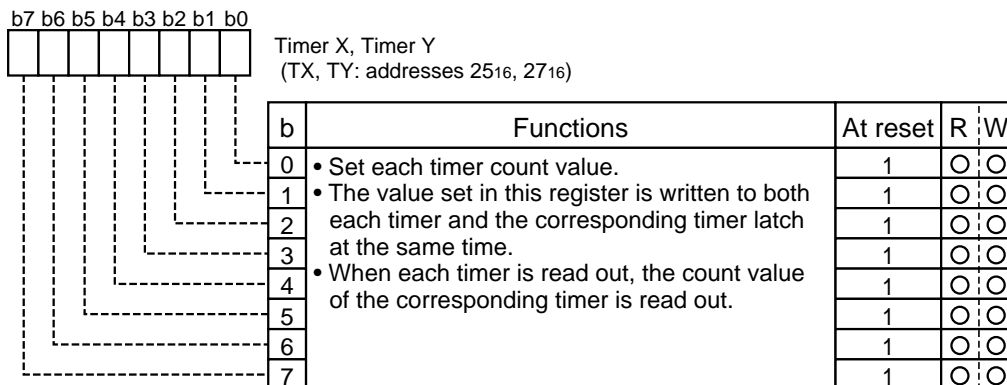


Fig. 3.5.18 Structure of Timer X, Timer Y

### Timer count source selection register

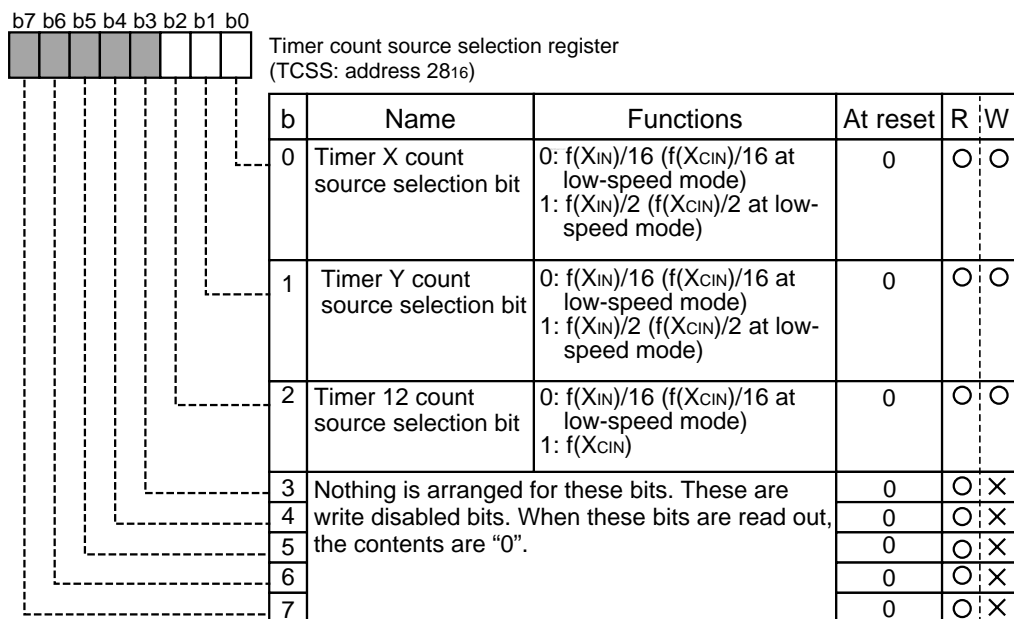


Fig. 3.5.19 Structure of Timer count source selection register

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## 3.5 List of registers

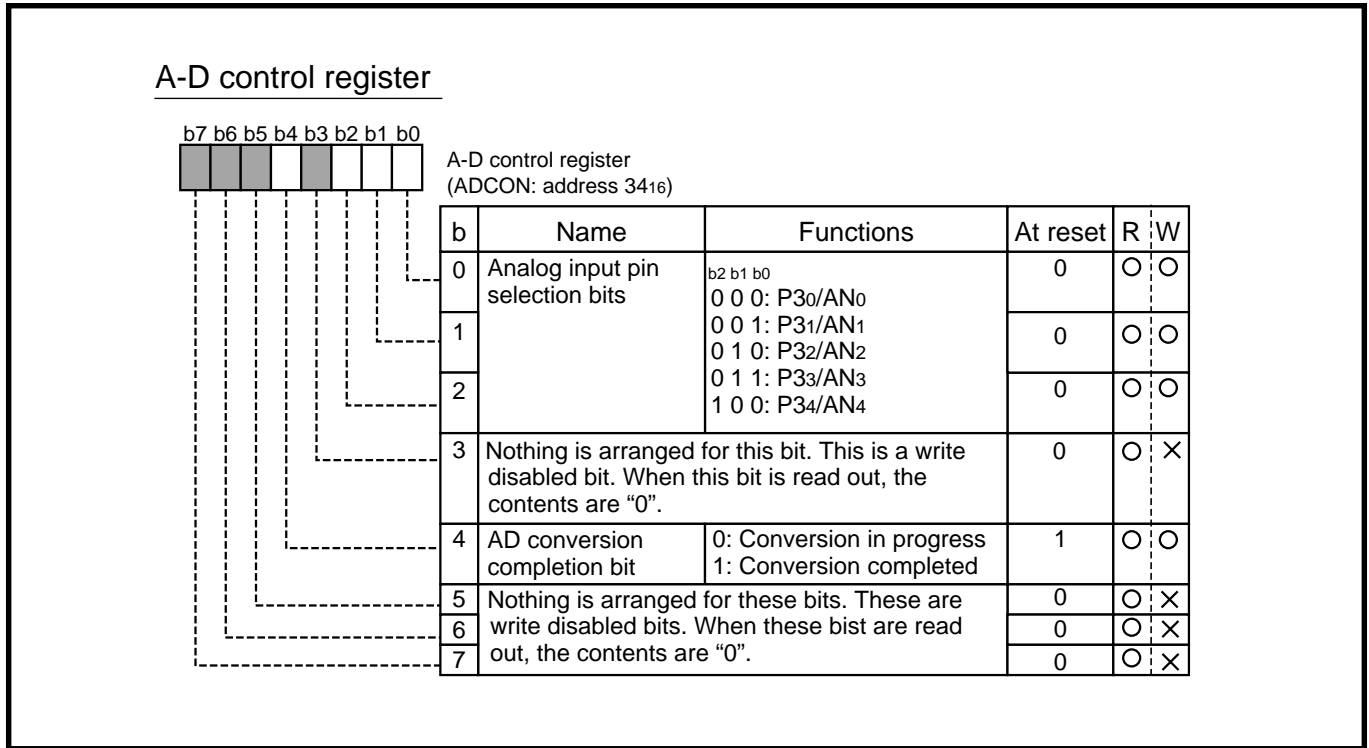


Fig. 3.5.20 Structure of A-D control register

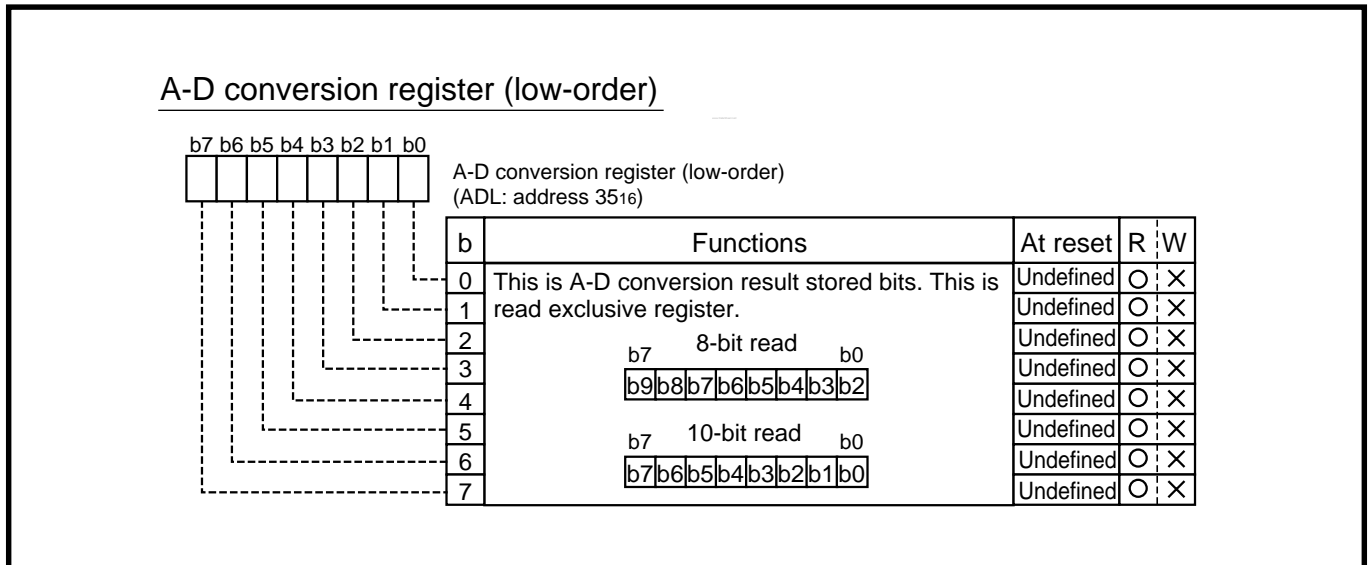


Fig. 3.5.21 Structure of A-D conversion low-order register

### A-D conversion register (high-order)

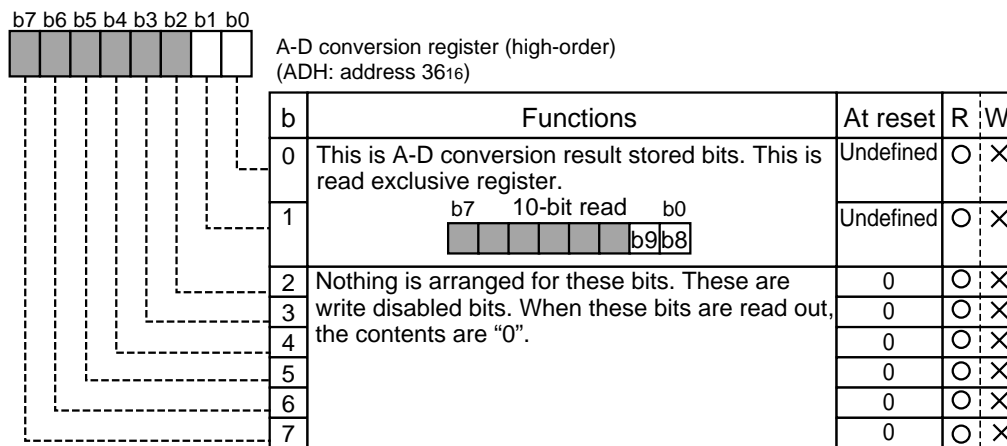
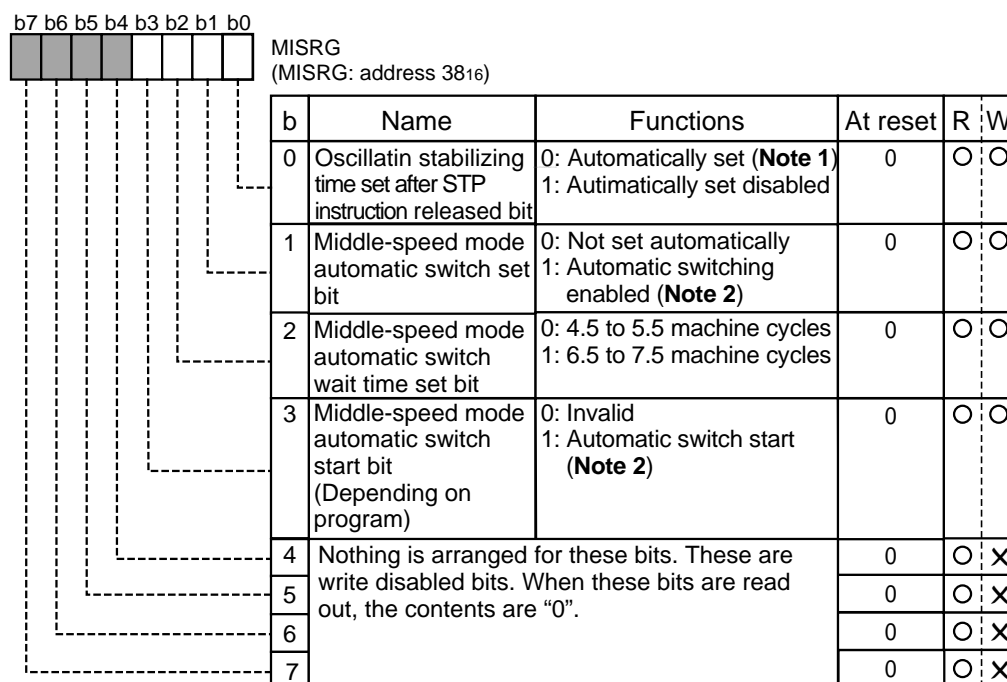


Fig. 3.5.22 Structure of A-D conversion high-order register

### MISRG



**Notes 1:** "01<sub>16</sub>" is set to Timer 1, "FF<sub>16</sub>" is set to Prescaler 12.

**2:** When automatic switch to middle-speed mode from low-speed mode occurs, the values of CPU mode register (3B<sub>16</sub>) change.

Fig. 3.5.23 Structure of MISRG

# APPENDIX

## 3.5 List of registers

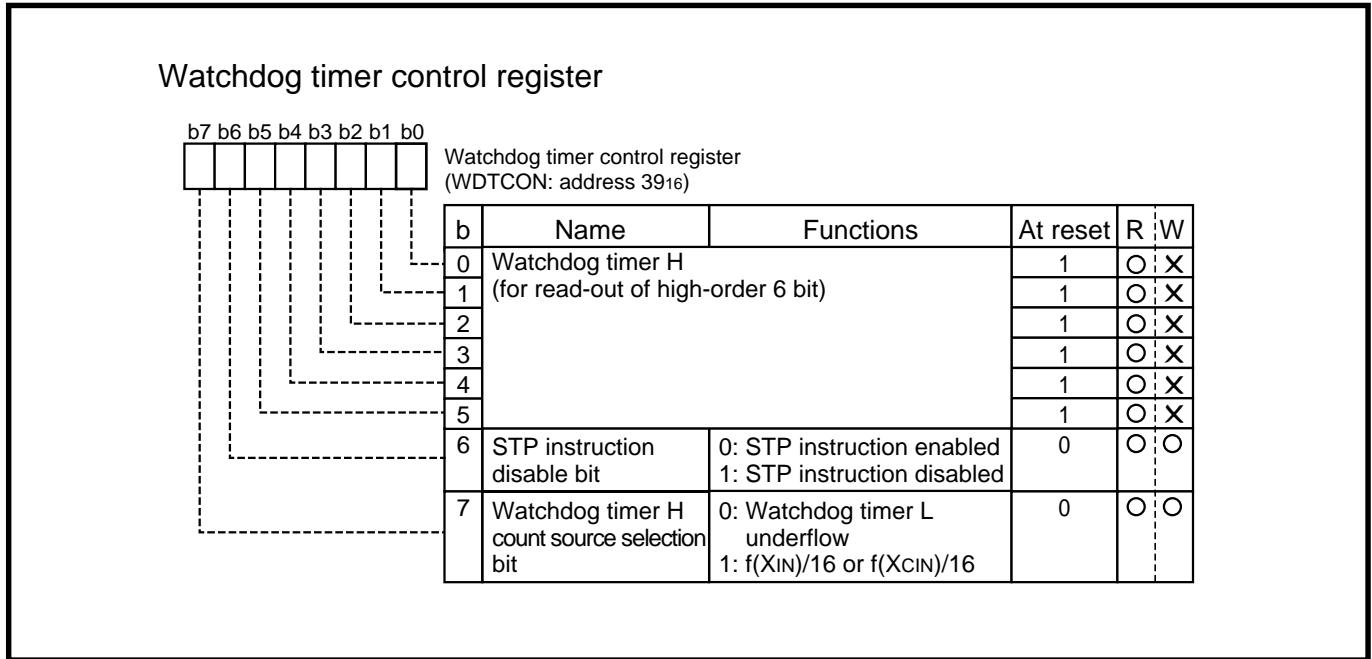


Fig. 3.5.24 Structure of Watchdog timer control register

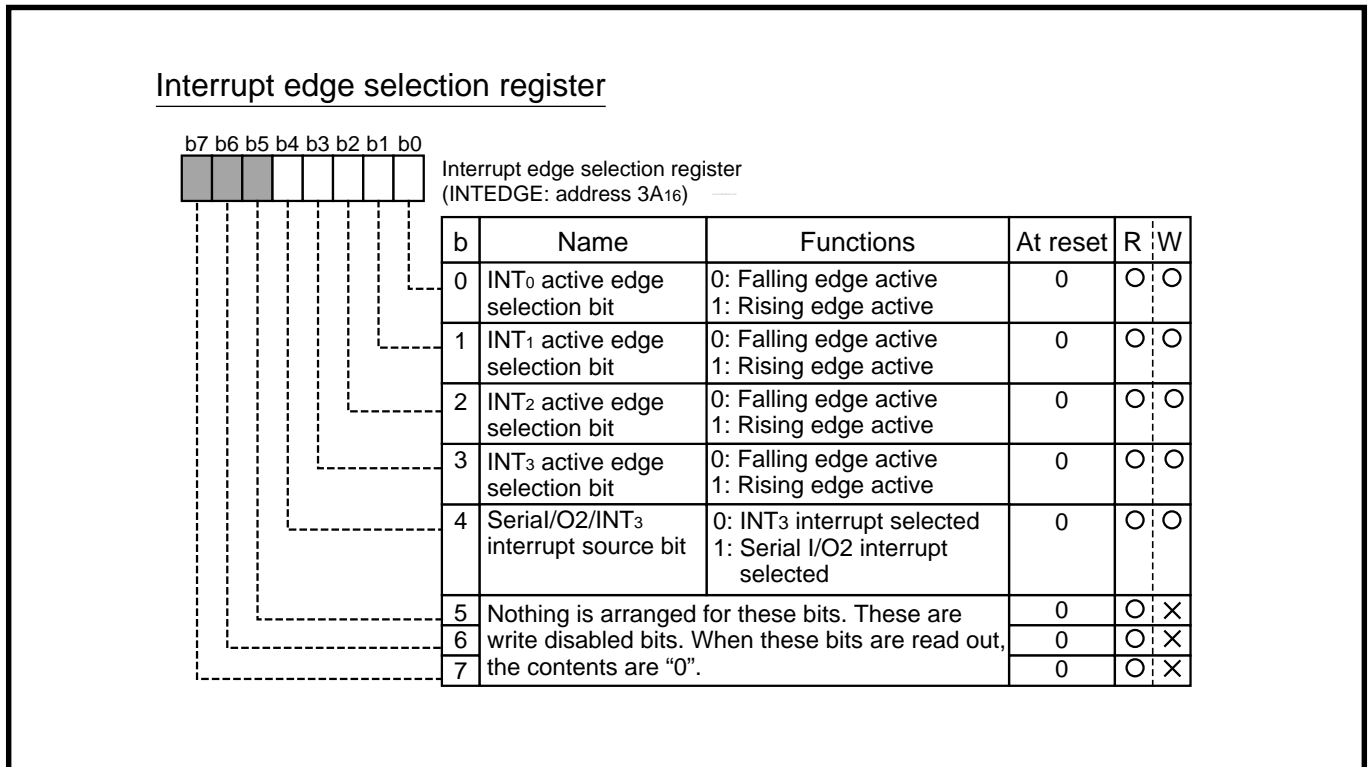


Fig. 3.5.25 Structure of Interrupt edge selection register

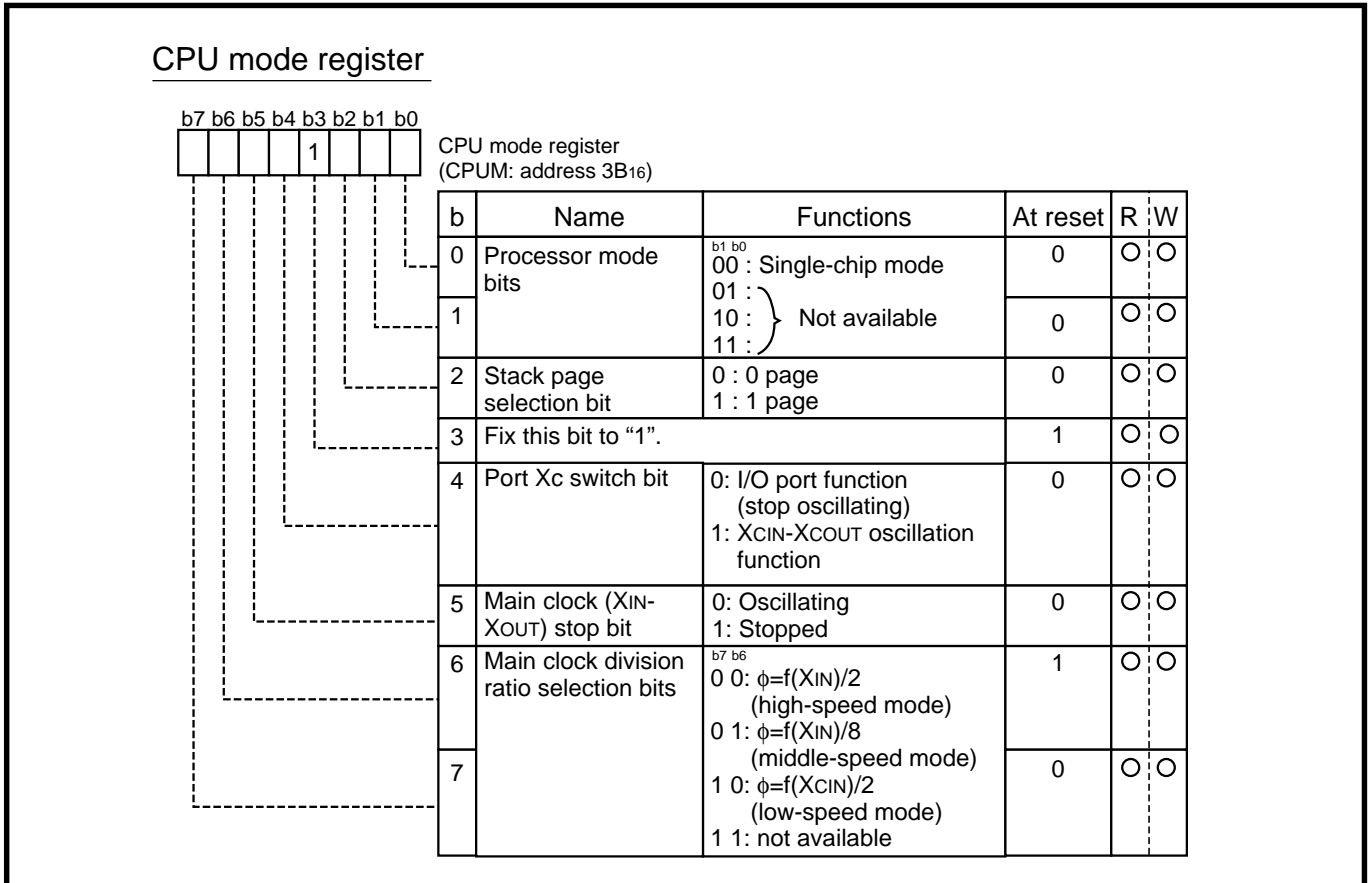


Fig. 3.5.26 Structure of CPU mode register

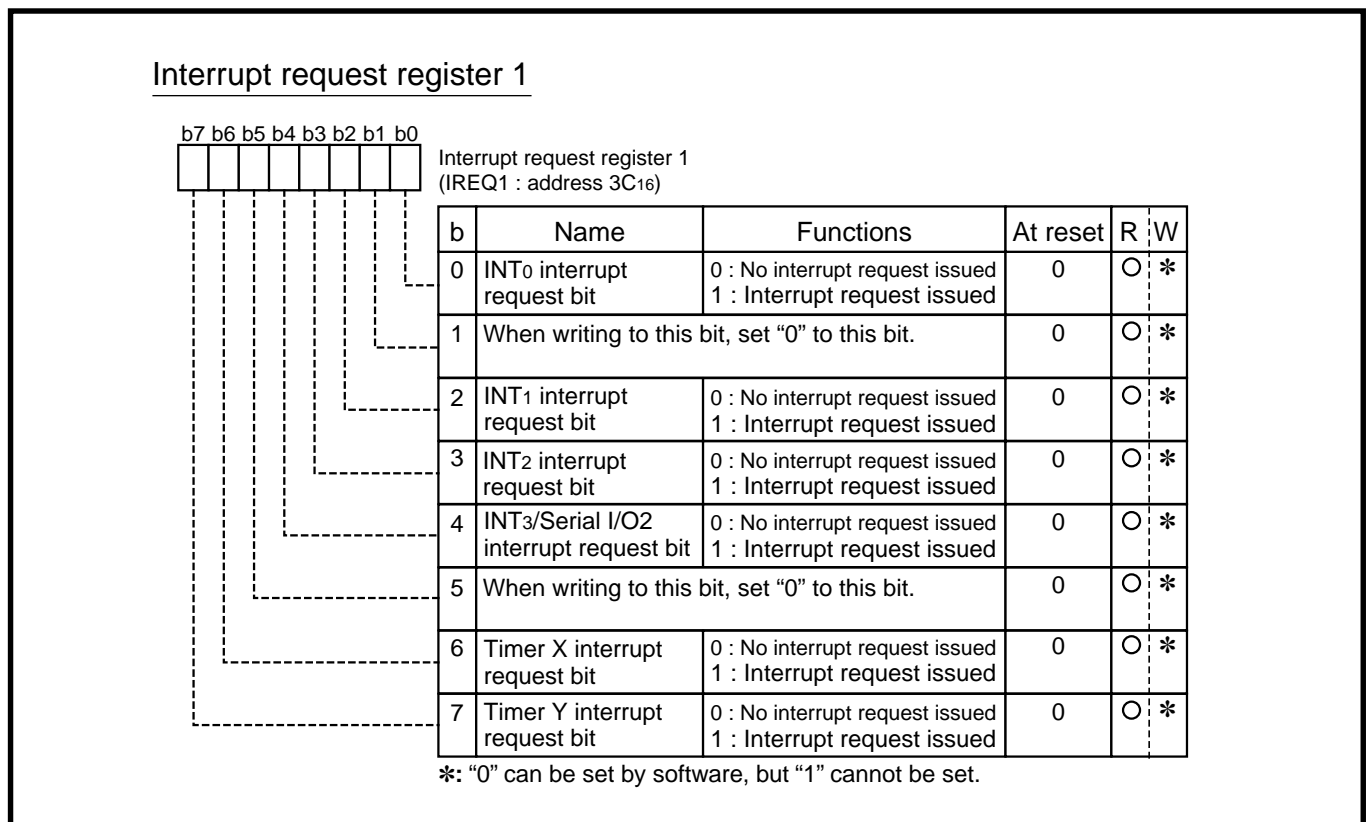


Fig. 3.5.27 Structure of Interrupt request register 1

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## 3.5 List of registers

### Interrupt request register 2

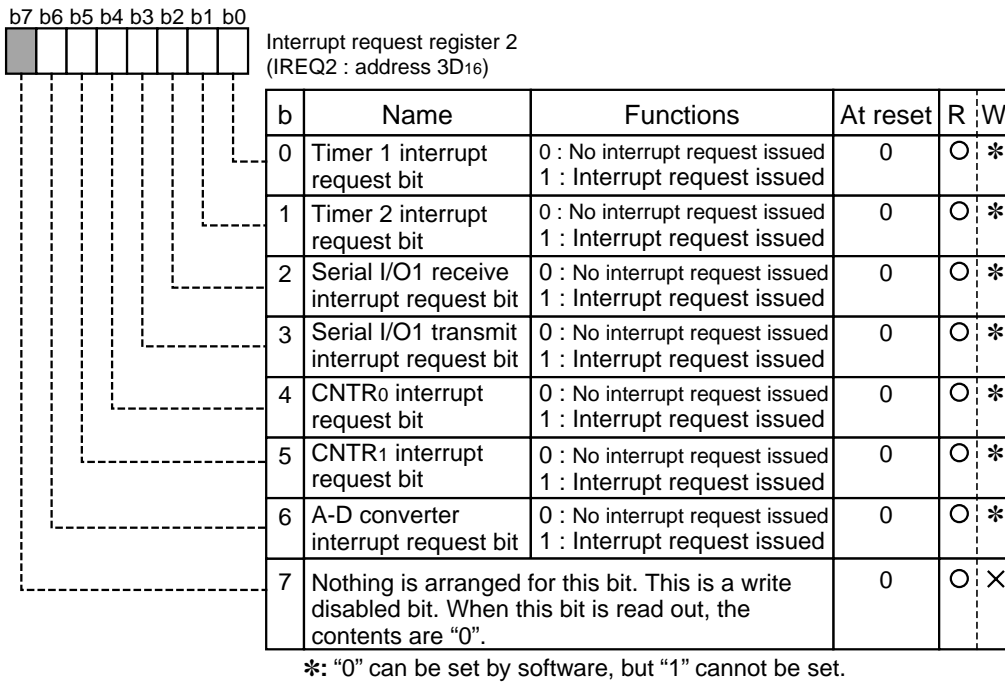


Fig. 3.5.28 Structure of Interrupt request register 2

### Interrupt control register 1

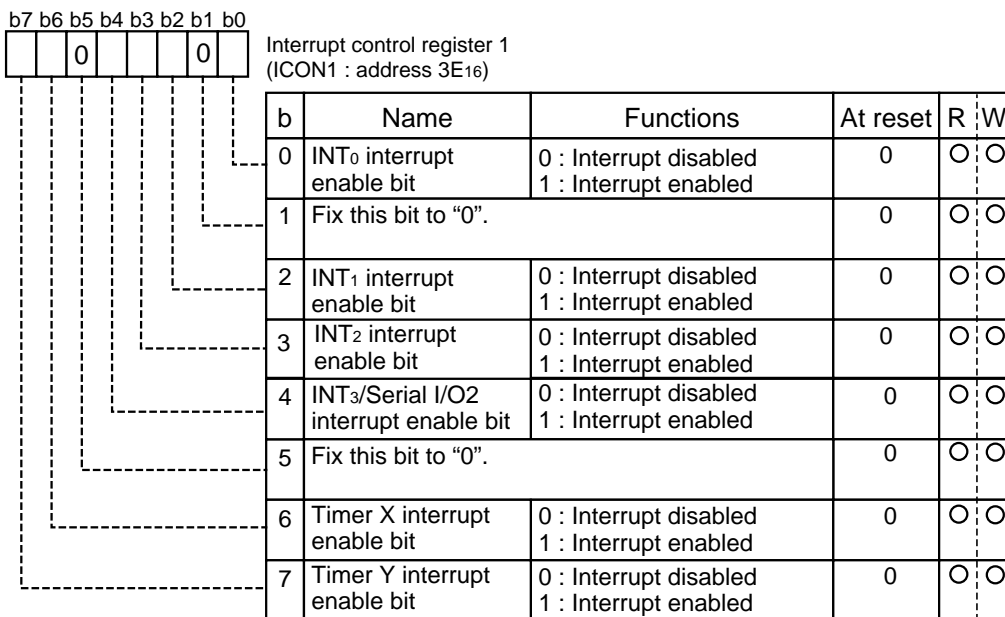


Fig. 3.5.29 Structure of Interrupt control register 1



### Interrupt control register 2

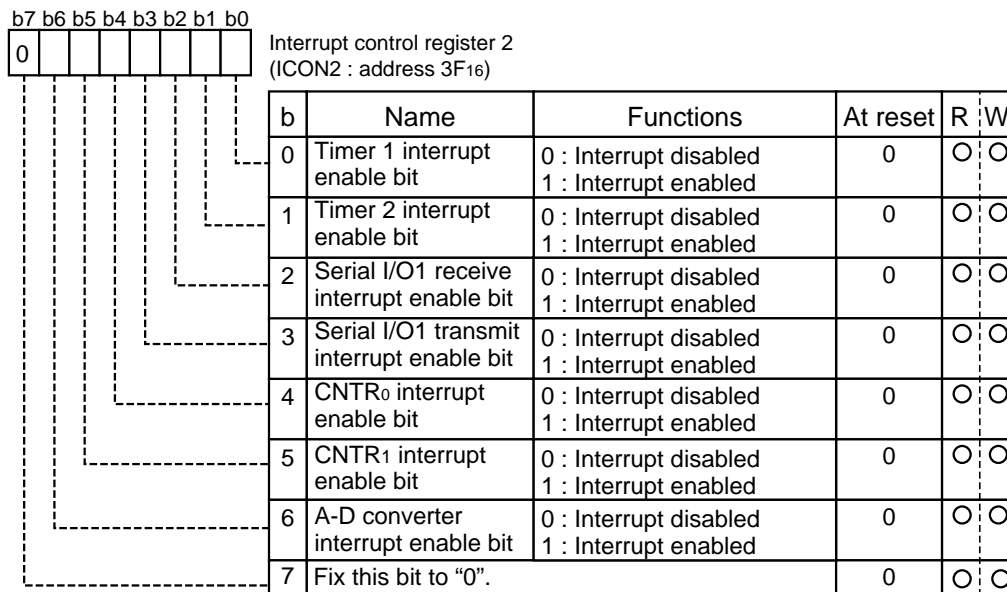
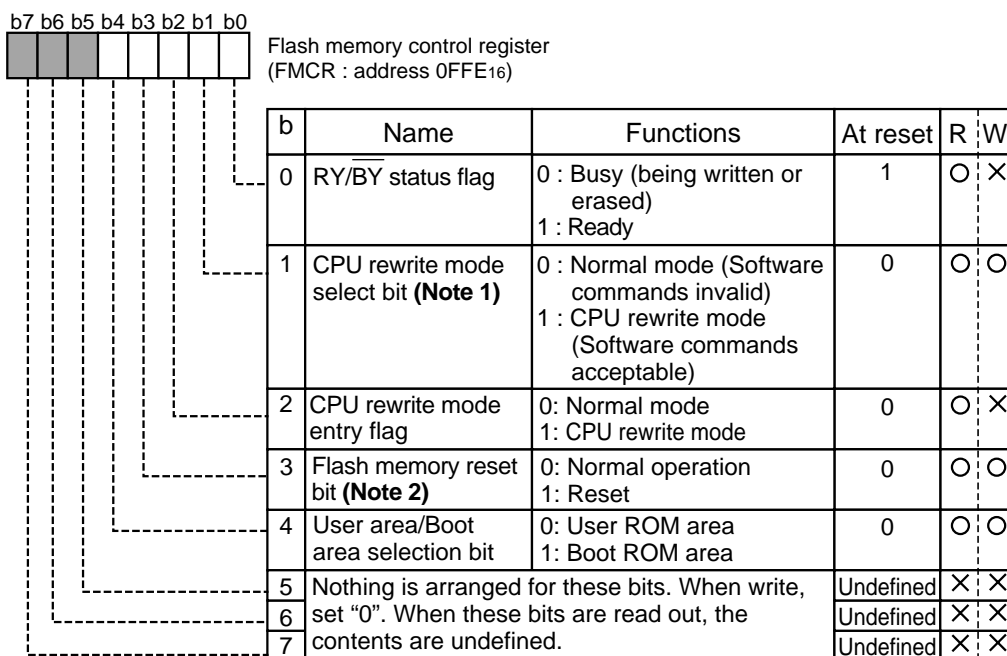


Fig. 3.5.30 Structure of Interrupt control register 2

### Flash memory control register



**Notes 1:** For this bit to be set to "1", the user needs to write a "0" and then a "1" to it in succession.

**2:** Effective only when the CPU rewrite mode select bit = "1". Set this bit to "0" subsequently after setting it to "1" (reset).

Fig. 3.5.31 Structure of Flash memory control register

# APPENDIX

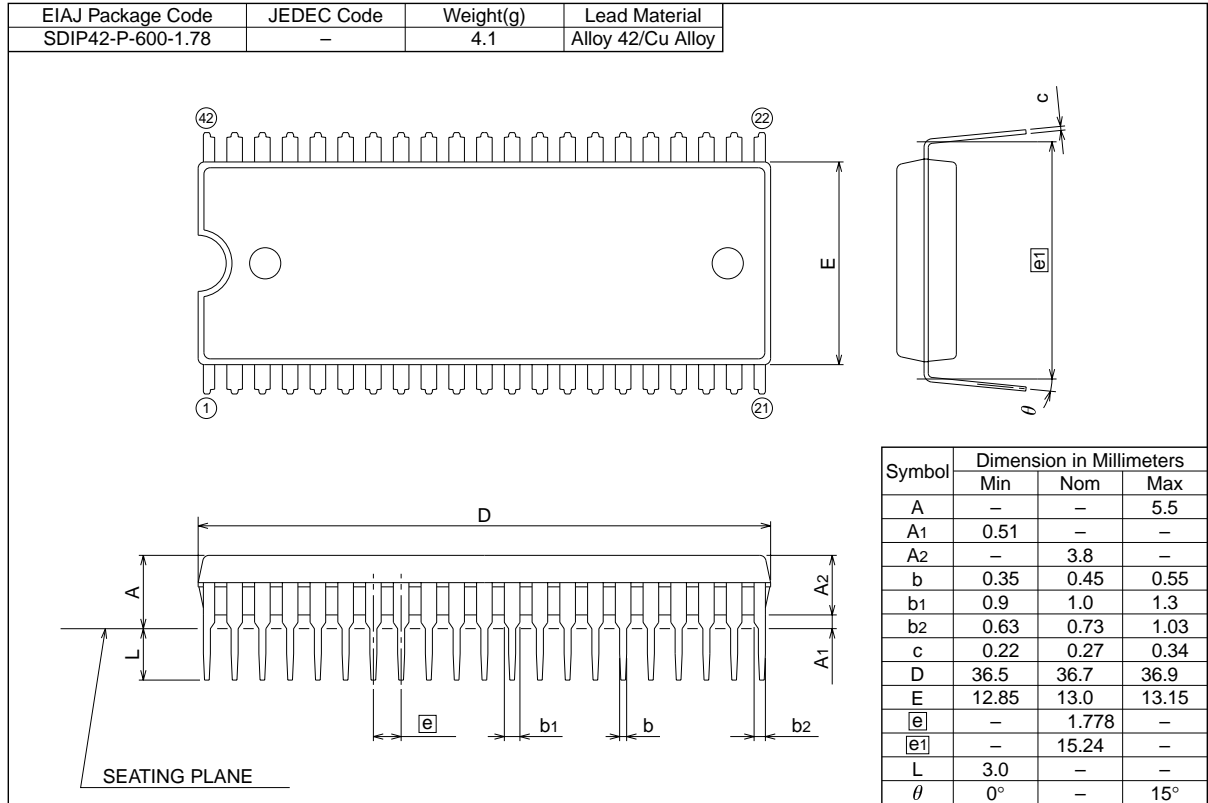
## 3.6 Package outline

### 3.6 Package outline

**42P4B**

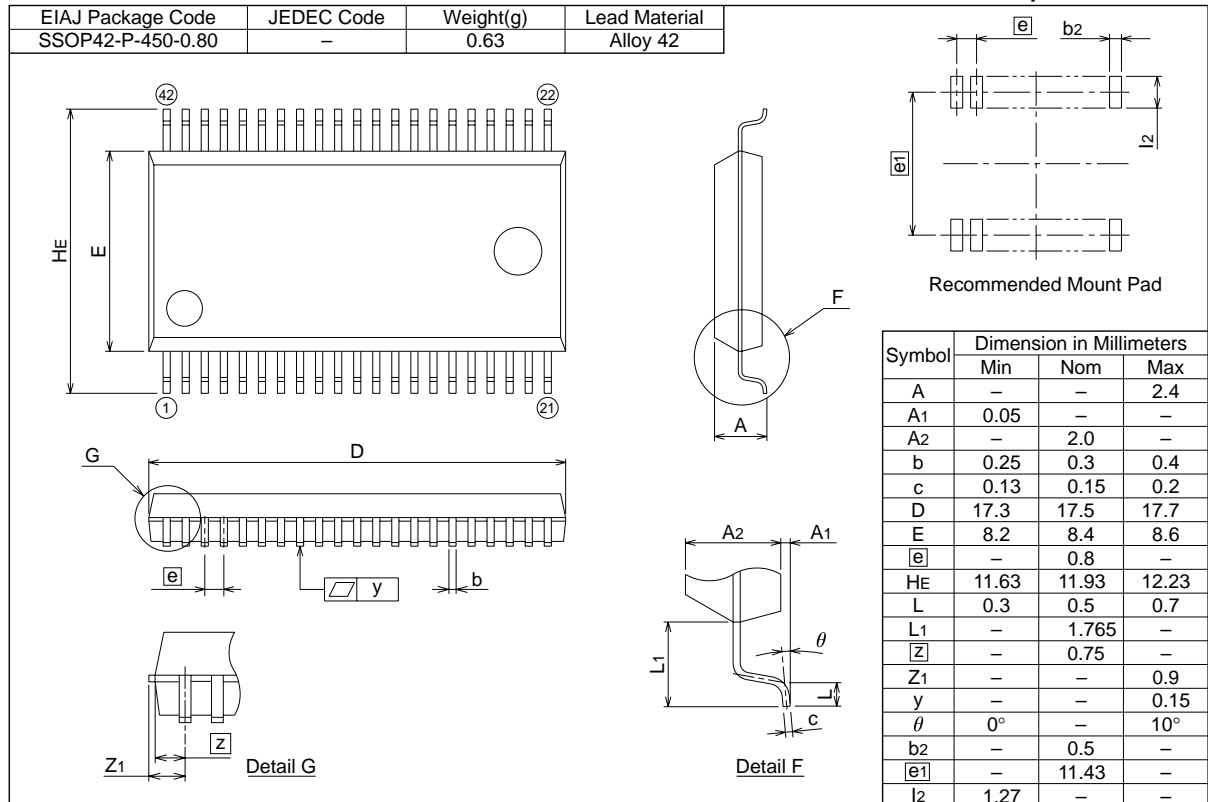
(MMP)

Plastic 42pin 600mil SDIP



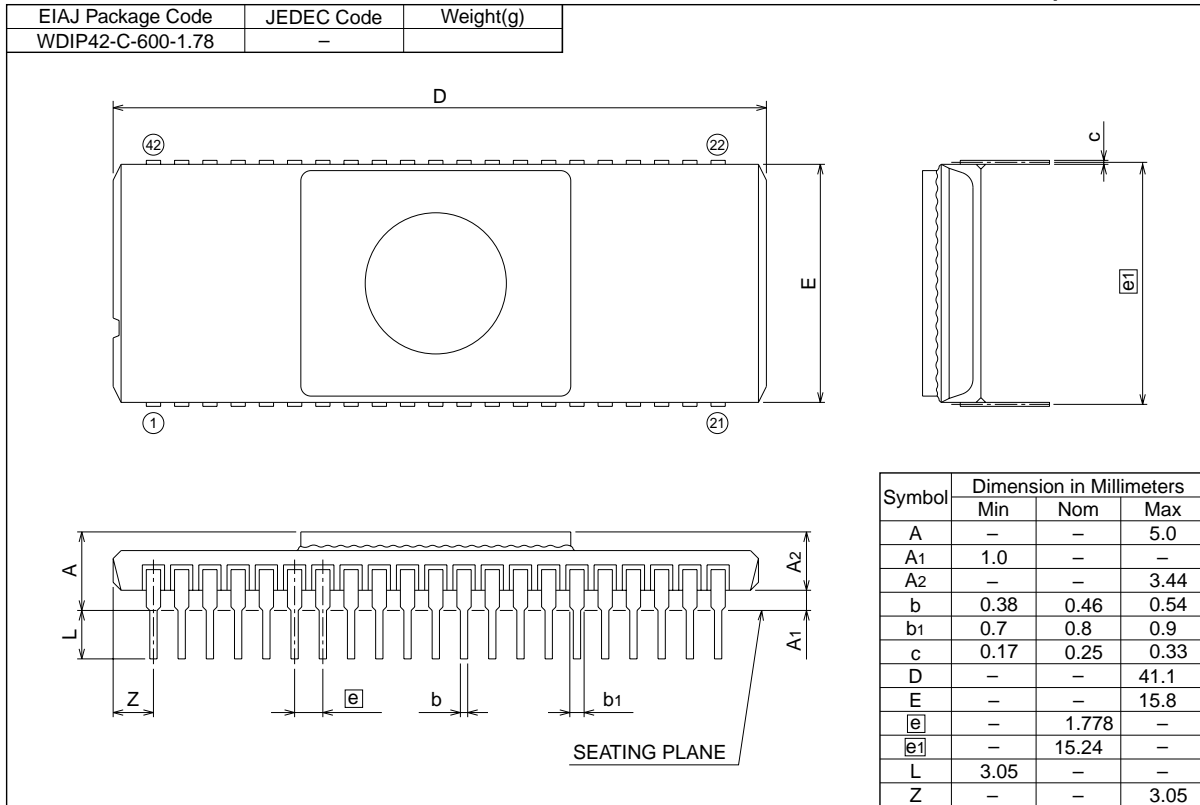
**42P2R-A/E**

Plastic 42pin 450mil SSOP

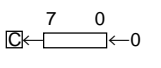


### 42S1B-A

Metal seal 42pin 600mil DIP



3.7 Machine instructions

Symbol	Function	Details	Addressing mode																				
			IMP			IMM			A			BIT, A, R			ZP			BIT, ZP, R					
			OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#			
ADC (Note 1) (Note 5)	When T = 0 $A \leftarrow A + M + C$  When T = 1 $M(X) \leftarrow M(X) + M + C$	When T = 0, this instruction adds the contents M, C, and A; and stores the results in A and C. When T = 1, this instruction adds the contents of M(X), M and C; and stores the results in M(X) and C. When T=1, the contents of A remain unchanged, but the contents of status flags are changed. M(X) represents the contents of memory where is indicated by X.				69	2	2										65	3	2			
AND (Note 1)	When T = 0 $A \leftarrow A \wedge M$  When T = 1 $M(X) \leftarrow M(X) \wedge M$	When T = 0, this instruction transfers the contents of A and M to the ALU which performs a bit-wise AND operation and stores the result back in A. When T = 1, this instruction transfers the contents M(X) and M to the ALU which performs a bit-wise AND operation and stores the results back in M(X). When T = 1 the contents of A remain unchanged, but status flags are changed. M(X) represents the contents of memory where is indicated by X.				29	2	2										25	3	2			
ASL		This instruction shifts the content of A or M by one bit to the left, with bit 0 always being set to 0 and bit 7 of A or M always being contained in C.							0A	2	1							06	5	2			
BBC (Note 4)	Ai or Mi = 0?	This instruction tests the designated bit i of M or A and takes a branch if the bit is 0. The branch address is specified by a relative address. If the bit is 1, next instruction is executed.										13 20i	4	2							17 20i	5	3
BBS (Note 4)	Ai or Mi = 1?	This instruction tests the designated bit i of the M or A and takes a branch if the bit is 1. The branch address is specified by a relative address. If the bit is 0, next instruction is executed.										03 20i	4	2							07 20i	5	3
BCC (Note 4)	C = 0?	This instruction takes a branch to the appointed address if C is 0. The branch address is specified by a relative address. If C is 1, the next instruction is executed.																			90	2	2
BCS (Note 4)	C = 1?	This instruction takes a branch to the appointed address if C is 1. The branch address is specified by a relative address. If C is 0, the next instruction is executed.																			B0	2	2
BEQ (Note 4)	Z = 1?	This instruction takes a branch to the appointed address when Z is 1. The branch address is specified by a relative address. If Z is 0, the next instruction is executed.																			F0	2	2
BIT	$A \wedge M$	This instruction takes a bit-wise logical AND of A and M contents; however, the contents of A and M are not modified. The contents of N, V, Z are changed, but the contents of A, M remain unchanged.																24	3	2			
BMI (Note 4)	N = 1?	This instruction takes a branch to the appointed address when N is 1. The branch address is specified by a relative address. If N is 0, the next instruction is executed.																			30	2	2
BNE (Note 4)	Z = 0?	This instruction takes a branch to the appointed address if Z is 0. The branch address is specified by a relative address. If Z is 1, the next instruction is executed.																			D0	2	2

Addressing mode															Processor status register																															
ZP, X			ZP, Y			ABS			ABS, X			ABS, Y			IND			ZP, IND			IND, X			IND, Y			REL		SP		7	6	5	4	3	2	1	0								
OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	N	V	T	B	D	I	Z	C
75	4	2				6D	4	3	7D	5	3	79	5	3										61	6	2	71	6	2										N	V	.	.	.	.	Z	C
35	4	2				2D	4	3	3D	5	3	39	5	3										21	6	2	31	6	2							N	.	.	.	.	.	Z	.			
16	6	2				0E	6	3	1E	7	3																						N	.	.	.	.	.	Z	C						
																																	.	.	.	.	.	.	.	.						
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																																	.	.	.	.	.	.	.	.						
																																	.	.	.	.	.	.	.	.						
																																	M7	M6	.	.	.	.	Z	.						
																																	.	.	.	.	.	.	.	.						
																																	.	.	.	.	.	.	.	.						



Symbol	Function	Details	Addressing mode																				
			IMP			IMM			A			BIT, A			ZP			BIT, ZP					
			OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#			
DEX	$X \leftarrow X - 1$	This instruction subtracts one from the current contents of X.	CA	2	1																		
DEY	$Y \leftarrow Y - 1$	This instruction subtracts one from the current contents of Y.	88	2	1																		
DIV	$A \leftarrow (M(zz + X + 1), M(zz + X)) / A$ $M(S) \leftarrow \text{one's complement of Remainder}$ $S \leftarrow S - 1$	Divides the 16-bit data in M(zz+(X)) (low-order byte) and M(zz+(X)+1) (high-order byte) by the contents of A. The quotient is stored in A and the one's complement of the remainder is pushed onto the stack.																					
EOR (Note 1)	When T = 0 $A \leftarrow A \vee M$  When T = 1 $M(X) \leftarrow M(X) \vee M$	When T = 0, this instruction transfers the contents of the M and A to the ALU which performs a bit-wise Exclusive OR, and stores the result in A. When T = 1, the contents of M(X) and M are transferred to the ALU, which performs a bit-wise Exclusive OR and stores the results in M(X). The contents of A remain unchanged, but status flags are changed. M(X) represents the contents of memory where is indicated by X.				49	2	2							45	3	2						
INC	$A \leftarrow A + 1$ or $M \leftarrow M + 1$	This instruction adds one to the contents of A or M.							3A	2	1				E6	5	2						
INX	$X \leftarrow X + 1$	This instruction adds one to the contents of X.	E8	2	1																		
INY	$Y \leftarrow Y + 1$	This instruction adds one to the contents of Y.	C8	2	1																		
JMP	If addressing mode is ABS $PCL \leftarrow ADL$ $PCH \leftarrow ADH$ If addressing mode is IND $PCL \leftarrow M(ADH, ADL)$ $PCH \leftarrow M(ADH, ADL + 1)$ If addressing mode is ZP, IND $PCL \leftarrow M(00, ADL)$ $PCH \leftarrow M(00, ADL + 1)$	This instruction jumps to the address designated by the following three addressing modes: Absolute Indirect Absolute Zero Page Indirect Absolute																					
JSR	$M(S) \leftarrow PCH$ $S \leftarrow S - 1$ $M(S) \leftarrow PCL$ $S \leftarrow S - 1$ After executing the above, if addressing mode is ABS, $PCL \leftarrow ADL$ $PCH \leftarrow ADH$ if addressing mode is SP, $PCL \leftarrow ADL$ $PCH \leftarrow FF$ If addressing mode is ZP, IND, $PCL \leftarrow M(00, ADL)$ $PCH \leftarrow M(00, ADL + 1)$	This instruction stores the contents of the PC in the stack, then jumps to the address designated by the following addressing modes: Absolute Special Page Zero Page Indirect Absolute																			22	5	2
LDA (Note 2)	When T = 0 $A \leftarrow M$ When T = 1 $M(X) \leftarrow M$	When T = 0, this instruction transfers the contents of M to A. When T = 1, this instruction transfers the contents of M to M(X). The contents of A remain unchanged, but status flags are changed. M(X) represents the contents of memory where is indicated by X.				A9	2	2							A5	3	2						
LDM	$M \leftarrow nn$	This instruction loads the immediate value in M.													3C	4	3						
LDX	$X \leftarrow M$	This instruction loads the contents of M in X.				A2	2	2							A6	3	2						
LDY	$Y \leftarrow M$	This instruction loads the contents of M in Y.				A0	2	2							A4	3	2						

Addressing mode															Processor status register																															
ZP, X			ZP, Y			ABS			ABS, X			ABS, Y			IND			ZP, IND			IND, X			IND, Y			REL			SP			7	6	5	4	3	2	1	0						
OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	N	V	T	B	D	I	Z	C
E2	16	2																																												
55	4	2				4D	4	3	5D	5	3	59	5	3							41	6	2	51	6	2										N	.	.	.	.	.	Z	.			
F6	6	2				EE	6	3	FE	7	3																									N	.	.	.	.	.	Z	.			
															6C	5	3	B2	4	2																.	.	.	.	.	.	.	.			
B5	4	2							AD	4	3	BD	5	3	B9	5	3				A1	6	2	B1	6	2										N	.	.	.	.	.	Z	.			
																		B6	4	2	AE	4	3				BE	5	3							N	.	.	.	.	.	Z	.			
B4	4	2																			AC	4	3	BC	5	3										N	.	.	.	.	.	Z	.			

Symbol	Function	Details	Addressing mode																	
			IMP			IMM			A			BIT, A			ZP			BIT, ZP		
			OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#
LSR		This instruction shifts either A or M one bit to the right such that bit 7 of the result always is set to 0, and the bit 0 is stored in C.							4A	2	1				46	5	2			
MUL	$M(S) \cdot A \leftarrow A * M(zz + X)$ $S \leftarrow S - 1$	Multiplies Accumulator with the memory specified by the Zero Page X address mode and stores the high-order byte of the result on the Stack and the low-order byte in A.																		
NOP	$PC \leftarrow PC + 1$	This instruction adds one to the PC but does no other operation.	EA	2	1															
ORA (Note 1)	When T = 0 $A \leftarrow A \vee M$  When T = 1 $M(X) \leftarrow M(X) \vee M$	When T = 0, this instruction transfers the contents of A and M to the ALU which performs a bit-wise "OR", and stores the result in A. When T = 1, this instruction transfers the contents of M(X) and the M to the ALU which performs a bit-wise OR, and stores the result in M(X). The contents of A remain unchanged, but status flags are changed. M(X) represents the contents of memory where is indicated by X.				09	2	2							05	3	2			
PHA	$S \leftarrow S - 1$	This instruction pushes the contents of A to the memory location designated by S, and decrements the contents of S by one.	48	3	1															
PHP	$M(S) \leftarrow PS$ $S \leftarrow S - 1$	This instruction pushes the contents of PS to the memory location designated by S and decrements the contents of S by one.	08	3	1															
PLA	$S \leftarrow S + 1$ $A \leftarrow M(S)$	This instruction increments S by one and stores the contents of the memory designated by S in A.	68	4	1															
PLP	$S \leftarrow S + 1$ $PS \leftarrow M(S)$	This instruction increments S by one and stores the contents of the memory location designated by S in PS.	28	4	1															
ROL		This instruction shifts either A or M one bit left through C. C is stored in bit 0 and bit 7 is stored in C.							2A	2	1				26	5	2			
ROR		This instruction shifts either A or M one bit right through C. C is stored in bit 7 and bit 0 is stored in C.							6A	2	1				66	5	2			
RRF		This instruction rotates 4 bits of the M content to the right.													82	8	2			
RTI	$S \leftarrow S + 1$ $PS \leftarrow M(S)$ $S \leftarrow S + 1$ $PCL \leftarrow M(S)$ $S \leftarrow S + 1$ $PCH \leftarrow M(S)$	This instruction increments S by one, and stores the contents of the memory location designated by S in PS. S is again incremented by one and stores the contents of the memory location designated by S in PCL. S is again incremented by one and stores the contents of memory location designated by S in PCH.	40	6	1															
RTS	$S \leftarrow S + 1$ $PCL \leftarrow M(S)$ $S \leftarrow S + 1$ $PCH \leftarrow M(S)$ $(PC) \leftarrow (PC) + 1$	This instruction increments S by one and stores the contents of the memory location designated by S in PCL. S is again incremented by one and the contents of the memory location is stored in PCH. PC is incremented by 1.	60	6	1															

Addressing mode															Processor status register																																		
ZP, X			ZP, Y			ABS			ABS, X			ABS, Y			IND			ZP, IND			IND, X			IND, Y			REL			SP			7	6	5	4	3	2	1	0									
OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	N	V	T	B	D	I	Z	C
56	6	2				4E	6	3	5E	7	3																												0	.	.	.	.	.	Z	C			
62	15	2																																		.	.	.	.	.	.	.	.						
																																				.	.	.	.	.	.	.	.						
15	4	2				0D	4	3	1D	5	3	19	5	3							01	6	2	11	6	2										N	.	.	.	.	.	Z	.						
																																				.	.	.	.	.	.	.	.						
																																				N	.	.	.	.	.	Z	.						
																																										(Value saved in stack)							
36	6	2				2E	6	3	3E	7	3																									N	.	.	.	.	.	Z	C						
76	6	2				6E	6	3	7E	7	3																									N	.	.	.	.	.	Z	C						
																																				.	.	.	.	.	.	.	.						
																																										(Value saved in stack)							
																																				.	.	.	.	.	.	.	.						

Symbol	Function	Details	Addressing mode																				
			IMP			IMM			A			BIT, A			ZP			BIT, ZP					
			OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#			
SBC (Note 1) (Note 5)	When T = 0 $A \leftarrow A - M - C$  When T = 1 $M(X) \leftarrow M(X) - M - C$	When T = 0, this instruction subtracts the value of M and the complement of C from A, and stores the results in A and C. When T = 1, the instruction subtracts the contents of M and the complement of C from the contents of M(X), and stores the results in M(X) and C. A remain unchanged, but status flag are changed. M(X) represents the contents of memory where is indicated by X.				E9	2	2										E5	3	2			
SEB	$A_i$ or $M_i \leftarrow 1$	This instruction sets the designated bit i of A or M.										0B	2	1							0F	5	2
SEC	$C \leftarrow 1$	This instruction sets C.	38	2	1																		
SED	$D \leftarrow 1$	This instruction set D.	F8	2	1																		
SEI	$I \leftarrow 1$	This instruction set I.	78	2	1																		
SET	$T \leftarrow 1$	This instruction set T.	32	2	1																		
STA	$M \leftarrow A$	This instruction stores the contents of A in M. The contents of A does not change.																85	4	2			
STP		This instruction resets the oscillation control F/F and the oscillation stops. Reset or interrupt input is needed to wake up from this mode.	42	2	1																		
STX	$M \leftarrow X$	This instruction stores the contents of X in M. The contents of X does not change.																86	4	2			
STY	$M \leftarrow Y$	This instruction stores the contents of Y in M. The contents of Y does not change.																84	4	2			
TAX	$X \leftarrow A$	This instruction stores the contents of A in X. The contents of A does not change.	AA	2	1																		
TAY	$Y \leftarrow A$	This instruction stores the contents of A in Y. The contents of A does not change.	A8	2	1																		
TST	$M = 0?$	This instruction tests whether the contents of M are "0" or not and modifies the N and Z.																64	3	2			
TSX	$X \leftarrow S$	This instruction transfers the contents of S in X.	BA	2	1																		
TXA	$A \leftarrow X$	This instruction stores the contents of X in A.	8A	2	1																		
TXS	$S \leftarrow X$	This instruction stores the contents of X in S.	9A	2	1																		
TYA	$A \leftarrow Y$	This instruction stores the contents of Y in A.	98	2	1																		
WIT		The WIT instruction stops the internal clock but not the oscillation of the oscillation circuit is not stopped. CPU starts its function after the Timer X over flows (comes to the terminal count). All registers or internal memory contents except Timer X will not change during this mode. (Of course needs VDD).	C2	2	1																		

Notes 1 : The number of cycles "n" is increased by 3 when T is 1.  
 2 : The number of cycles "n" is increased by 2 when T is 1.  
 3 : The number of cycles "n" is increased by 1 when T is 1.  
 4 : The number of cycles "n" is increased by 2 when branching has occurred.  
 5 : N, V, and Z flags are invalid in decimal operation mode.

Addressing mode															Processor status register																															
ZP, X			ZP, Y			ABS			ABS, X			ABS, Y			IND			ZP, IND			IND, X			IND, Y			REL			SP			7	6	5	4	3	2	1	0						
OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	OP	n	#	N	V	T	B	D	I	Z	C
F5	4	2				ED	4	3	FD	5	3	F9	5	3										E1	6	2	F1	6	2										N	V	.	.	.	.	Z	C
																																				.	.	.	.	.	.	.	.			
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95	5	2				8D	5	3	9D	6	3	99	6	3										81	7	2	91	7	2							.	.	.	.	.	.	.	.			
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# APPENDIX


## 3.7 Machine instructions


Symbol	Contents	Symbol	Contents
IMP	Implied addressing mode	+	Addition
IMM	Immediate addressing mode	-	Subtraction
A	Accumulator or Accumulator addressing mode	*	Multiplication
BIT, A	Accumulator bit addressing mode	/	Division
BIT, A, R	Accumulator bit relative addressing mode	∧	Logical OR
ZP	Zero page addressing mode	∨	Logical AND
BIT, ZP	Zero page bit addressing mode	⊕	Logical exclusive OR
BIT, ZP, R	Zero page bit relative addressing mode	—	Negation
ZP, X	Zero page X addressing mode	←	Shows direction of data flow
ZP, Y	Zero page Y addressing mode	X	Index register X
ABS	Absolute addressing mode	Y	Index register Y
ABS, X	Absolute X addressing mode	S	Stack pointer
ABS, Y	Absolute Y addressing mode	PC	Program counter
IND	Indirect absolute addressing mode	PS	Processor status register
ZP, IND	Zero page indirect absolute addressing mode	PCH	8 high-order bits of program counter
IND, X	Indirect X addressing mode	PCL	8 low-order bits of program counter
IND, Y	Indirect Y addressing mode	ADH	8 high-order bits of address
REL	Relative addressing mode	ADL	8 low-order bits of address
SP	Special page addressing mode	FF	FF in Hexadecimal notation
C	Carry flag	nn	Immediate value
Z	Zero flag	zz	Zero page address
I	Interrupt disable flag	M	Memory specified by address designation of any addressing mode
D	Decimal mode flag	M(X)	Memory of address indicated by contents of index register X
B	Break flag	M(S)	Memory of address indicated by contents of stack pointer
T	X-modified arithmetic mode flag	M(ADH, ADL)	Contents of memory at address indicated by ADH and ADL, in ADH is 8 high-order bits and ADL is 8 low-order bits.
V	Overflow flag	M(00, ADL)	Contents of address indicated by zero page ADL
N	Negative flag	Ai	Bit i (i = 0 to 7) of accumulator
		Mi	Bit i (i = 0 to 7) of memory
		OP	Opcode
		n	Number of cycles
		#	Number of bytes

### 3.8 List of instruction code

D3 – D0 D7 – D4		Hexadecimal notation															
		0000	0001	0010	0011	0100	0101	0110	0111	1000	1001	1010	1011	1100	1101	1110	1111
		0	1	2	3	4	5	6	7	8	9	A	B	C	D	E	F
0000	0	BRK	ORA IND, X	JSR ZP, IND	BBS 0, A	—	ORA ZP	ASL ZP	BBS 0, ZP	PHP	ORA IMM	ASL A	SEB 0, A	—	ORA ABS	ASL ABS	SEB 0, ZP
0001	1	BPL	ORA IND, Y	CLT	BBC 0, A	—	ORA ZP, X	ASL ZP, X	BBC 0, ZP	CLC	ORA ABS, Y	DEC A	CLB 0, A	—	ORA ABS, X	ASL ABS, X	CLB 0, ZP
0010	2	JSR ABS	AND IND, X	JSR SP	BBS 1, A	BIT ZP	AND ZP	ROL ZP	BBS 1, ZP	PLP	AND IMM	ROL A	SEB 1, A	BIT ABS	AND ABS	ROL ABS	SEB 1, ZP
0011	3	BMI	AND IND, Y	SET	BBC 1, A	—	AND ZP, X	ROL ZP, X	BBC 1, ZP	SEC	AND ABS, Y	INC A	CLB 1, A	LDM ZP	AND ABS, X	ROL ABS, X	CLB 1, ZP
0100	4	RTI	EOR IND, X	STP	BBS 2, A	COM ZP	EOR ZP	LSR ZP	BBS 2, ZP	PHA	EOR IMM	LSR A	SEB 2, A	JMP ABS	EOR ABS	LSR ABS	SEB 2, ZP
0101	5	BVC	EOR IND, Y	—	BBC 2, A	—	EOR ZP, X	LSR ZP, X	BBC 2, ZP	CLI	EOR ABS, Y	—	CLB 2, A	—	EOR ABS, X	LSR ABS, X	CLB 2, ZP
0110	6	RTS	ADC IND, X	MUL ZP, X	BBS 3, A	TST ZP	ADC ZP	ROR ZP	BBS 3, ZP	PLA	ADC IMM	ROR A	SEB 3, A	JMP IND	ADC ABS	ROR ABS	SEB 3, ZP
0111	7	BVS	ADC IND, Y	—	BBC 3, A	—	ADC ZP, X	ROR ZP, X	BBC 3, ZP	SEI	ADC ABS, Y	—	CLB 3, A	—	ADC ABS, X	ROR ABS, X	CLB 3, ZP
1000	8	BRA	STA IND, X	RRF ZP	BBS 4, A	STY ZP	STA ZP	STX ZP	BBS 4, ZP	DEY	—	TXA	SEB 4, A	STY ABS	STA ABS	STX ABS	SEB 4, ZP
1001	9	BCC	STA IND, Y	—	BBC 4, A	STY ZP, X	STA ZP, X	STX ZP, Y	BBC 4, ZP	TYA	STA ABS, Y	TXS	CLB 4, A	—	STA ABS, X	—	CLB 4, ZP
1010	A	LDY IMM	LDA IND, X	LDX IMM	BBS 5, A	LDY ZP	LDA ZP	LDX ZP	BBS 5, ZP	TAY	LDA IMM	TAX	SEB 5, A	LDY ABS	LDA ABS	LDX ABS	SEB 5, ZP
1011	B	BCS	LDA IND, Y	JMP ZP, IND	BBC 5, A	LDY ZP, X	LDA ZP, X	LDX ZP, Y	BBC 5, ZP	CLV	LDA ABS, Y	TSX	CLB 5, A	LDY ABS, X	LDA ABS, X	LDX ABS, Y	CLB 5, ZP
1100	C	CPY IMM	CMP IND, X	WIT	BBS 6, A	CPY ZP	CMP ZP	DEC ZP	BBS 6, ZP	INY	CMP IMM	DEX	SEB 6, A	CPY ABS	CMP ABS	DEC ABS	SEB 6, ZP
1101	D	BNE	CMP IND, Y	—	BBC 6, A	—	CMP ZP, X	DEC ZP, X	BBC 6, ZP	CLD	CMP ABS, Y	—	CLB 6, A	—	CMP ABS, X	DEC ABS, X	CLB 6, ZP
1110	E	CPX IMM	SBC IND, X	DIV ZP, X	BBS 7, A	CPX ZP	SBC ZP	INC ZP	BBS 7, ZP	INX	SBC IMM	NOP	SEB 7, A	CPX ABS	SBC ABS	INC ABS	SEB 7, ZP
1111	F	BEQ	SBC IND, Y	—	BBC 7, A	—	SBC ZP, X	INC ZP, X	BBC 7, ZP	SED	SBC ABS, Y	—	CLB 7, A	—	SBC ABS, X	INC ABS, X	CLB 7, ZP

 : 3-byte instruction

 : 2-byte instruction

 : 1-byte instruction

# APPENDIX

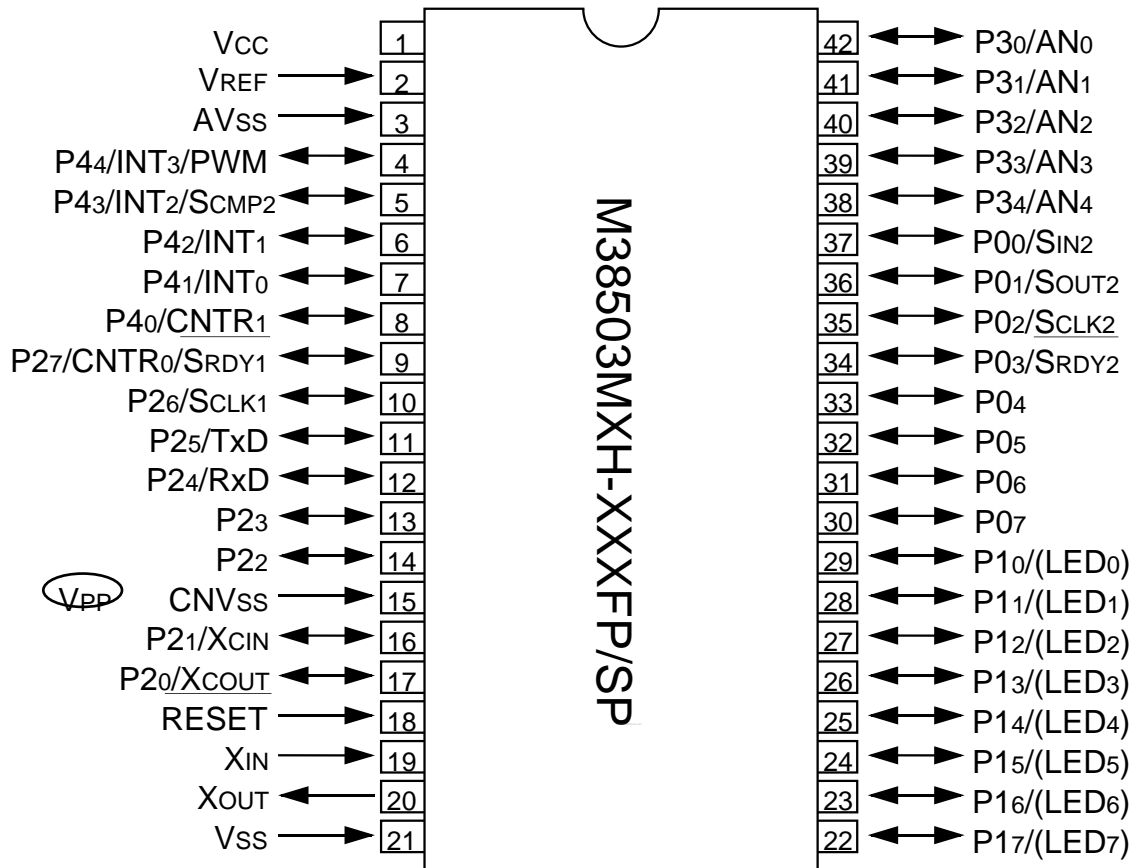
## 3.9 SFR memory map

### 3.9 SFR memory map

0000 <sub>16</sub>	Port P0 (P0)	0020 <sub>16</sub>	Prescaler 12 (PRE12)
0001 <sub>16</sub>	Port P0 direction register (P0D)	0021 <sub>16</sub>	Timer 1 (T1)
0002 <sub>16</sub>	Port P1 (P1)	0022 <sub>16</sub>	Timer 2 (T2)
0003 <sub>16</sub>	Port P1 direction register (P1D)	0023 <sub>16</sub>	Timer XY mode register (TM)
0004 <sub>16</sub>	Port P2 (P2)	0024 <sub>16</sub>	Prescaler X (PREX)
0005 <sub>16</sub>	Port P2 direction register (P2D)	0025 <sub>16</sub>	Timer X (TX)
0006 <sub>16</sub>	Port P3 (P3)	0026 <sub>16</sub>	Prescaler Y (PREY)
0007 <sub>16</sub>	Port P3 direction register (P3D)	0027 <sub>16</sub>	Timer Y (TY)
0008 <sub>16</sub>	Port P4 (P4)	0028 <sub>16</sub>	Timer count source selection register (TCSS)
0009 <sub>16</sub>	Port P4 direction register (P4D)	0029 <sub>16</sub>	
000A <sub>16</sub>		002A <sub>16</sub>	
000B <sub>16</sub>		002B <sub>16</sub>	Reserved *
000C <sub>16</sub>		002C <sub>16</sub>	Reserved *
000D <sub>16</sub>		002D <sub>16</sub>	Reserved *
000E <sub>16</sub>		002E <sub>16</sub>	Reserved *
000F <sub>16</sub>		002F <sub>16</sub>	Reserved *
0010 <sub>16</sub>		0030 <sub>16</sub>	Reserved *
0011 <sub>16</sub>		0031 <sub>16</sub>	Reserved *
0012 <sub>16</sub>	Reserved *	0032 <sub>16</sub>	
0013 <sub>16</sub>	Reserved *	0033 <sub>16</sub>	
0014 <sub>16</sub>	Reserved *	0034 <sub>16</sub>	A-D control register (ADCON)
0015 <sub>16</sub>	Serial I/O2 control register 1 (SIO2CON1)	0035 <sub>16</sub>	A-D conversion low-order register (ADL)
0016 <sub>16</sub>	Serial I/O2 control register 2 (SIO2CON2)	0036 <sub>16</sub>	A-D conversion high-order register (ADH)
0017 <sub>16</sub>	Serial I/O2 register (SIO2)	0037 <sub>16</sub>	Reserved *
0018 <sub>16</sub>	Transmit/Receive buffer register (TB/RB)	0038 <sub>16</sub>	MISRG
0019 <sub>16</sub>	Serial I/O1 status register (SIOSTS)	0039 <sub>16</sub>	Watchdog timer control register (WDTCN)
001A <sub>16</sub>	Serial I/O1 control register (SIOCON)	003A <sub>16</sub>	Interrupt edge selection register (INTEDGE)
001B <sub>16</sub>	UART control register (UARTCON)	003B <sub>16</sub>	CPU mode register (CPUM)
001C <sub>16</sub>	Baud rate generator (BRG)	003C <sub>16</sub>	Interrupt request register 1 (IREQ1)
001D <sub>16</sub>	PWM control register (PWMCON)	003D <sub>16</sub>	Interrupt request register 2 (IREQ2)
001E <sub>16</sub>	PWM prescaler (PREPWM)	003E <sub>16</sub>	Interrupt control register 1 (ICON1)
001F <sub>16</sub>	PWM register (PWM)	003F <sub>16</sub>	Interrupt control register 2 (ICON2)
		00FE <sub>16</sub>	Flash memory control register (FMCR)

\* Reserved : Do not write any data to this addresses, because these areas are reserved.

### 3.10 Pin configurations

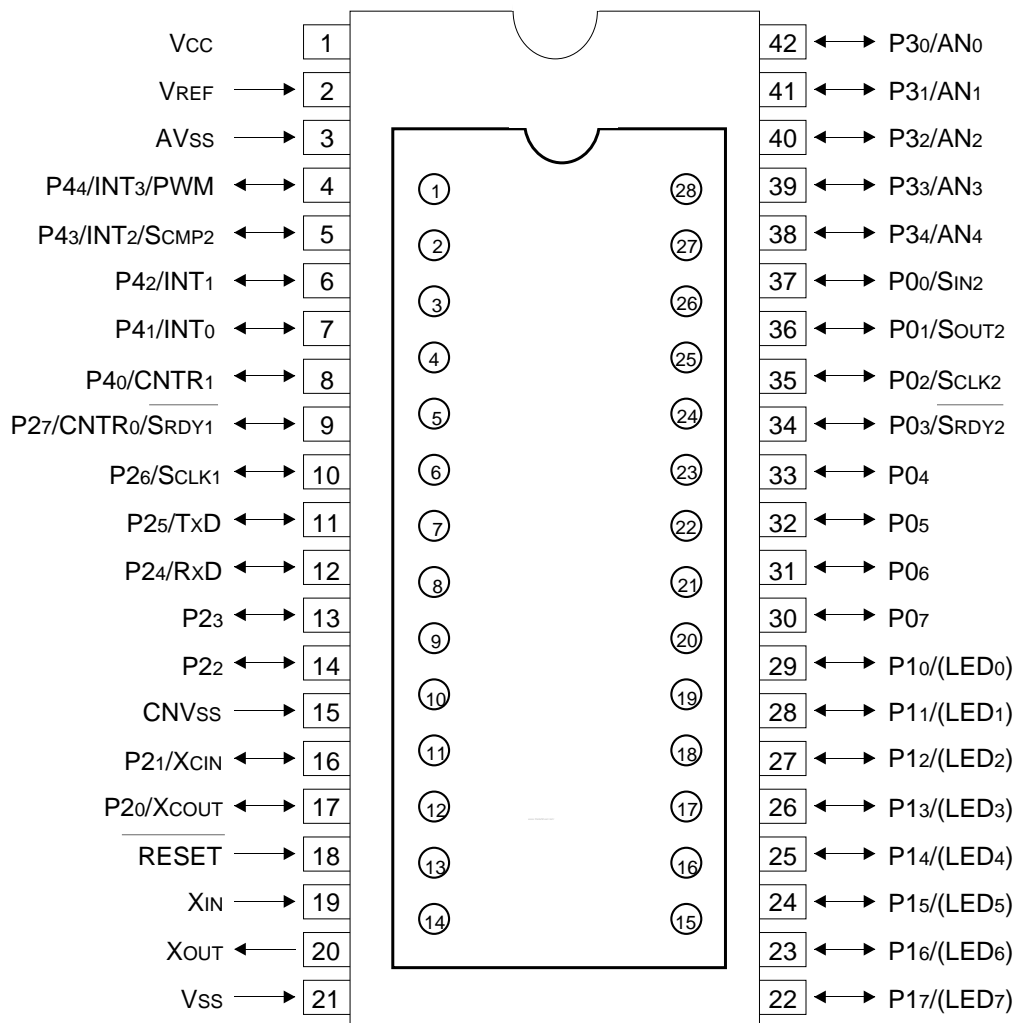


○ : Flash memory version

# APPENDIX

## 3.10 Pin configurations

### M38517RSS PIN CONFIGURATION (TOP VIEW)



Outline : 42S1M

**RENESAS 8-BIT CISC SINGLE-CHIP MICROCOMPUTER  
USER'S MANUAL  
3850 Group (Spec. H) Rev.1.03**

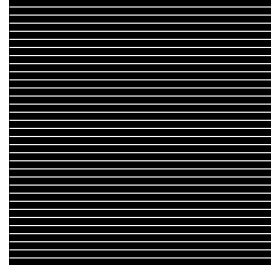
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3850 Group (Spec. H)  
User's Manual



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