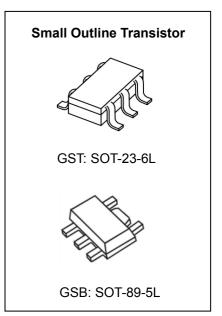


Preliminary Datasheet

Step-Down, 1A Dimmable LED Driver

Features

- 6~40V wide input voltage range
- Maximum 1A constant output current (1.2A for MBI6657GSB)
- PWM/DC input for dimming control
- Hysteretic PFM operation eliminates external compensation design
- Integrated power switch with 0.25-ohm low R_{ds(on)}
- Full protections: Thermal Foldback/ Thermal Shutdown/ Schottky Diode Short/ LED Open/Short-Circuit
- Only 5 external components required
- Package MSL Level: 3



Product Description

MBI6657 is a step-down constant-current high-brightness LED driver to provide a cost-effective design solution for interior/exterior illumination applications. It is designed to deliver constant current to light up high power LED with only 5 external components. With hysteretic PFM control scheme, MBI6657 eliminates external compensation design and makes the design simple.

The output current of MBI6657 can be programmed by an external resistor and dimmed via pulse width modulation (PWM) or DC voltage through DIM pin to achieve higher efficiency linear current modulation.

MBI6657 features completed protection design to handle faulty situations. The start-up function limits the inrush current while the power is switched on. Thermal shutdown, Thermal Foldback and Schottky diode short (SDS) guard the system to be robust and keep the driver away from being damaged which results from LED open-circuited, short-circuited and other abnormal events. Besides SOT-236 small package, MBI6657 provides thermal-enhanced SOT-89 packages as well to handle power dissipation more efficiently.

Applications

- Automotive Lighting
- Signage and Decorative LED Lighting
- High Power LED Lighting
- Constant Current Source

Typical Application Circuit

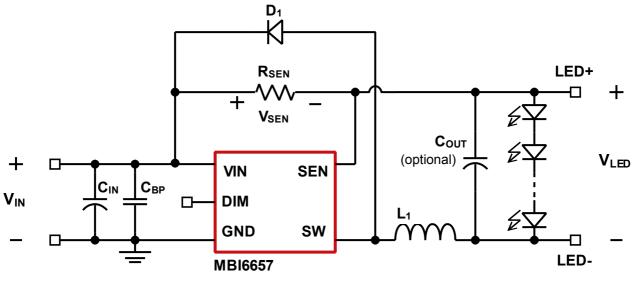
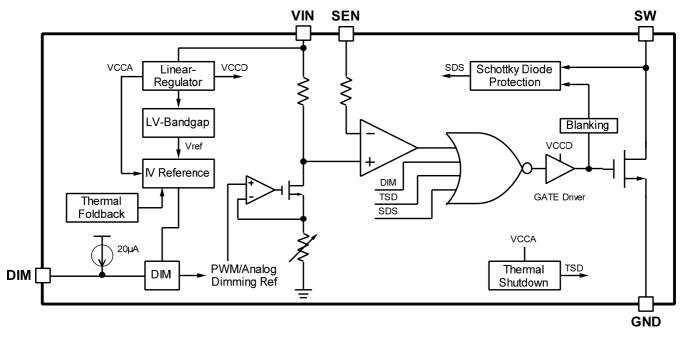
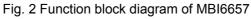


Fig. 1 Application circuit of MBI6657

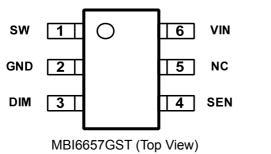
Functional Diagram

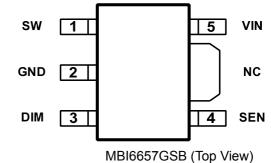




Step-Down, 1A Dimmable LED Driver

Pin Configuration





Pin Description

| Pin Name | Function |
|----------|--|
| GND | Ground terminal for control logic and current sink |
| SW | Switch output terminal |
| DIM | Digital/Analog dimming control terminal. PWM or DC voltage signal is applied into the terminal for brightness control. |
| SEN | Output current sense terminal |
| VIN | Supply voltage terminal |
| NC | No function pin* |

*To improve the noise immunity, this NC pin is suggested to connect to GND on PCB. In addition for SOT89, when a heat-conducting copper foil on PCB is soldered with NC pin, the desired thermal conductivity will be improved.

Step-Down, 1A Dimmable LED Driver

Maximum Ratings

Operation above the maximum ratings may cause device failure. Operation at the extended periods of the maximum ratings may reduce the device reliability.

| Characteri | Symbol | Rating | Unit | |
|---|--|-----------------|--------------------|------|
| Supply Voltage | | V _{IN} | 0~45 | V |
| Sustaining Voltage at DIM pin | V _{PWMD} | -0.3~45 | V | |
| Sustaining Voltage at SW pin | | V _{SW} | -0.3~45 | V |
| Sustaining Voltage at SEN pin | V_{SEN} | -0.3~45 | V | |
| Power Dissipation (On 4-Layer PCB, Ta=25°C) | 0.07.7 | P _D | 0.5 | W |
| Thermal Resistance (By simulation, on 4-Layer PCB) | GST Type | $R_{th(j-a)}$ | 244.0 | °C/W |
| Power Dissipation (On 4-Layer PCB, Ta=25°C) | GSB Type | P _D | 0.75 | W |
| Empirical Thermal Resistance (On PCB, Ta=25°C)* | GSB Type | $R_{th(j-a)}$ | 121.5 | °C/W |
| Junction Temperature | $T_{j,max}$ | 150** | °C | |
| Storage Temperature | T _{stg} | -55~+150 | °C | |
| ESD Rating | Human Body Mode (MIL-STD-883G Method 3015.8) | HBM | Class 3A (5KV) | - |
| | Machine Mode (ANSI/ ESD S5.2-2009) | MM | Class M4 (400V) | - |

*The PCB size is 76.2mm*114.3mm in simulation. Please refer to JEDEC JESD51-7 thermal measurement standard.

**Operation at the maximum rating for extended periods may reduce the device reliability; therefore, the suggested junction temperature of the device is under 125°C.

Note: The performance of thermal dissipation is strongly related to the size of thermal pad, thickness and layer numbers of the PCB. The empirical thermal resistance may be different from simulative value. Users should plan for expected thermal dissipation performance by selecting package and arranging layout of the PCB to maximize the capability.

Recommended Operating Conditions

| Characteristic | Symbol | Rating | Unit |
|--------------------------------------|-----------------|----------|------|
| Supply Voltage | V _{IN} | 6~40 | V |
| Operating Junction Temperature Range | T _i | -40~+125 | °C |

Step-Down, 1A Dimmable LED Driver

Electrical Characteristics

 $V_{\text{IN}}=12V, V_{\text{OUT}}=3.6V, L_{1}=68\mu\text{H}, C_{\text{IN}}=C_{\text{OUT}}=10\mu\text{F}, \ T_{\text{A}}=25^{\circ}\text{C}; \text{ unless otherwise specified}.$

| Characte | | Symbol | Condition | Min. | Тур. | Max. | Unit |
|---------------------------------------|------------------------|-------------------------------|--|------|------|-----------------|--------|
| INPUT AND C | | | | | | | |
| Supply Voltage | e | V _{IN} | - | 6 | - | 40 | V |
| Supply Curren | t | I _{IN} | V _{IN} =6V~40V | - | 1.2 | 2 | mA |
| Outrast Ourraget | + | I _{OUT} | For GST Type | - | - | 1 | Α |
| Output Curren | ι | | For GSB Type | - | - | 1.2 | Α |
| HYSTERESIS | CONTROL | | | · | | | |
| Mean Sense \ | | V _{SENSE} | - | 95 | 100 | 105 | mV |
| Sense Voltage hysteresis | e threshold | $V_{\text{SENSE},\text{HYS}}$ | - | - | 15 | - | % |
| Internal Propa Delay Time | gation | T _{PD} | - | - | 250 | 400 | ns |
| MOS SWITCH | | | | | | | 1 |
| Switch ON Re | sistance | R _{ds(on)} | V _{IN} =12V | - | 0.25 | - | Ω |
| Minimum Swit Time* | ch ON | T _{ON,min} | - | - | 300 | - | ns |
| Minimum Swit Time* | ch OFF | $T_{OFF,min}$ | - | - | 300 | - | ns |
| Recommende Cycle Range d | | D _{sw} | - | 20 | - | 80 | % |
| Maximum Ope frequency | | Freq _{Max} | - | 40 | - | 1000 | kHz |
| THERMAL O | /ERLOAD | | | | | | |
| Thermal Shuto Threshold* | down | T _{SD} | - | - | 155 | - | °C |
| Thermal Shuto Hystersis* | lown | T _{SD-HYS} | - | - | 30 | - | °C |
| Thermal Foldb Threshold* | back | T _{FB} | - | | 135 | - | °C |
| Current Decre | ase Rate* | $\Delta I_{OUT} / \Delta T$ | T _{FB} <t<sub>i < T_{SD}</t<sub> | - | -3 | _ | %/°C |
| PWM DIMMIN | | 001/ | | | Ū | I | , o. C |
| Input | "H" level | V _{IH} | - | 2.5 | - | V _{IN} | V |
| voltage of PWMD | "L" level | V _{IL} | - | - | - | 0.3 | V |
| Duty Cycle PWM Signal DIM pin | Range of Applied to | Duty _{PWM} | PWM Frequency: 1kHz | 0 | - | 100 | % |
| ANALOG DIM | IMING | | | | | | |
| Analog Dimming Input Clamp Voltage | | V _{DIM.CLAMP} | - | - | 2.5 | - | V |
| Analog Dimmi Voltage turn of | ng Input | $V_{\text{DIM. SWOFF}}$ | - | - | 0.3 | - | V |
| Source Currer | | I _{DIM} | V _{IN} =12V, V _{DIM} =0V | | 20 | - | μA |

*Parameters are not tested at production. Parameters are guaranteed by design.

Typical Performance Characteristic

1. Efficiency vs. Input Voltage at Various LED Cascaded Numbers

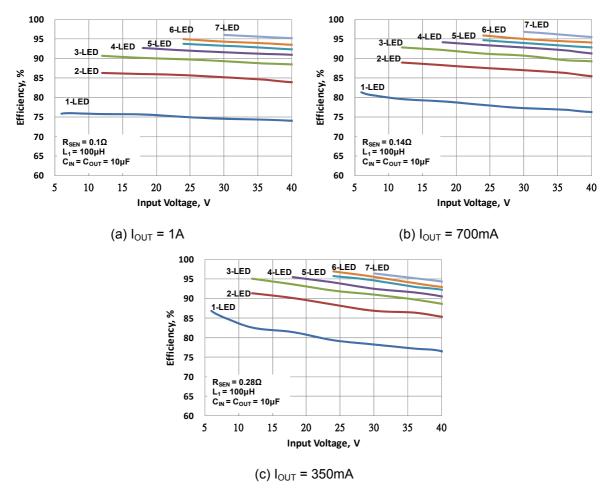


Fig. 3 Efficiency vs. input voltage at I_{OUT} = 1A, 700mA and 350mA

2. Output Current vs. Ambient Temperature

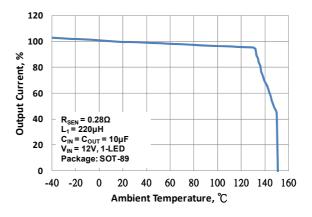


Fig. 4 Output Current vs. Ambient Temperature at V_{IN} = 12V/ I_{OUT} = 350mA

5. $R_{ds(on)}$ vs. Ambient Temperature at V_{IN} = 12V

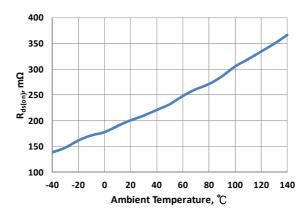


Fig. 5 $R_{ds(on)}$ vs. Temperature at V_{IN} = 12V

Step-Down, 1A Dimmable LED Driver

Application Information

MBI6657 is a simple and high efficient buck converter with capability to drive up to 1A of loading (1.2A for MBI6657GSB). The device adopts hysteretic PFM control scheme to regulate loading and input voltage variations. The hysteretic PFM control requires no loop compensation bringing very fast load transient response and simplicity of the design.

The device is well suited for applications requiring a wide input voltage range. The high-side current sensing and an integrated current-setting circuitry minimize the number of external components while delivering an average output current with $\pm 5\%$ accuracy. Featured by PWM dimming and analog dimming capability, MBI6657 offers flexible ways to meet LED dimming related applications.

Setting Average Output Current

The average output current (I_{OUT}) is set by an external resistor, R_{SEN} . The relationship between I_{OUT} and R_{SEN} is as below:

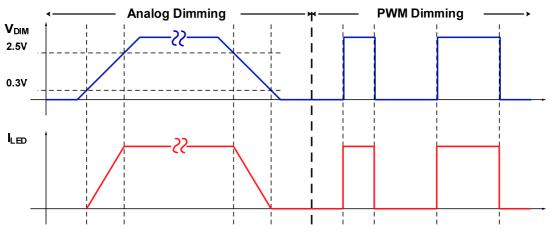
 R_{SEN} =(V_{SEN}/I_{OUT})=(0.1V/I_{OUT}); V_{SEN}=0.1V;

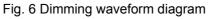
$I_{OUT}=(V_{SEN}/R_{SEN})=(0.1V/R_{SEN})$

where R_{SEN} is the resistance of the external resistor connecting to SEN pin, and V_{SEN} is the voltage of external resistor. The magnitude of current (as a function of R_{SEN}) is around 1000mA at 0.1 Ω .

Dimming Functions

Dimming is achieved by applying either a PWM signal or a DC voltage at the DIM pin. For analog dimming application, the LED current increases linearly with the rising V_{DIM} , and the dimming range of VDIM is from 0.3V to 2.5V. For PWM dimming application, the LED current can be adjusted digitally, by applying a PWM logic signal to the DIM pin to turn the device on and off, as shown on Fig. 14.





A. PWM dimming

The dimming of LEDs can be performed by applying PWM signals of DIM pin. A logic low (below 0.3V) at DIM disables the internal MOSFET and shuts off the current flow to the LED array. An internal pull-up circuit ensures that the MBI6657 is ON when DIM pin is unconnected. Therefore, the need for an external pull-up resistor will be eliminated. The following Fig. 13 shows good linearity in dimming control.

Step-Down, 1A Dimmable LED Driver

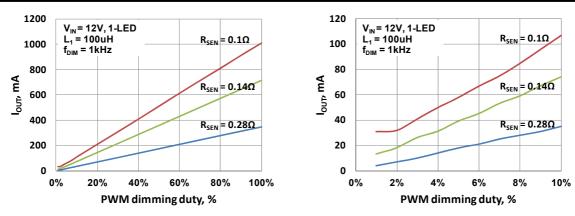


Fig. 7 PWM dimming curve

B. Analog dimming

Users can apply DC voltage directly to DIM for modulating LED current. The result is shown in the following figures.

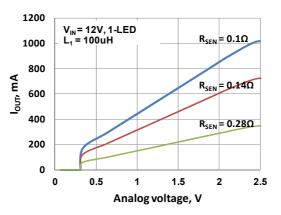


Fig. 8 Analog dimming curve

Additionally, users can also connect a resistor between DIM and GND to set $V_{\text{DIM}},$ that is determined by

$$V_{\text{DIM}} = 20 \mu A \times R_{\text{DIM}}$$

Step-Down, 1A Dimmable LED Driver

Component Selection

A. Inductor Selection

The inductance is determined by two factors: the switching frequency and the inductor ripple current. The calculation of the inductance, L_1 , can be described as

$$L_{1} = \frac{(V_{\text{IN}} - V_{\text{LED}})\frac{V_{\text{LED}}}{V_{\text{IN}}}}{0.3 \times f_{\text{S}} \times I_{\text{LED}}}$$

When selecting an inductor, not only the inductance but also the saturation current that should be considered as the factors to affect the performance of module. In general, it is recommended to choose an inductor with 1.5 times of LED current as the saturation current. Also, the larger inductance gains the better line/load regulation. However, the inductance and saturation current become a trade-off at the same inductor size. An inductor with shield is recommended to reduce the EMI interference. However, this is another trade-off with heat dissipation.

B. Schottky Diode Selection

The MBI6657 needs a flywheel diode, D_1 , to carry the inductor current when the MOSFET is off. The recommended flywheel diode is schottky diode with low forward voltage for better efficiency. Two factors determine the selection of schottky diode. One is the maximum reverse voltage. The recommended rated voltage of the reverse voltage is at least 1.5 times of input voltage. The other is the maximum forward current, which works when the MOSFET is off. And the recommended forward current is 1.5 times of output current. Users should carefully choose an appropriate schottky diode which can perform low leakage current at high temperature.

C. Input Capacitor Selection

The input capacitor, C_{IN} , can supply pulses of current for the MBI6657 when the MOSFET is ON. And C_{IN} is charged by input voltage when the MOSFET is OFF. As the input voltage is lower than the tolerable input voltage, the internal MOSFET of the MBI6657 remains constantly ON, and the LED current is limited to 1.15 times of normal current. The recommended value of input capacitor is 10uF to stabilize the lighting system. The rated voltage of the input capacitor should be at least 1.5 times of the input voltage.

For system stability, it is recommended to place the C_{IN} to the VIN pin of MBI6657 as close as possible. However, the PCB size might limit this requirement. Therefore, to avoid the noise interference, a bypass capacitor, whose capacitance range is from 0.1uF to 1uF and the material is ceramic, parallels with the VIN and GND pins of MBI6657 is recommended. The rated voltage of the bypass capacitor should be at least 1.5times of the input voltage.

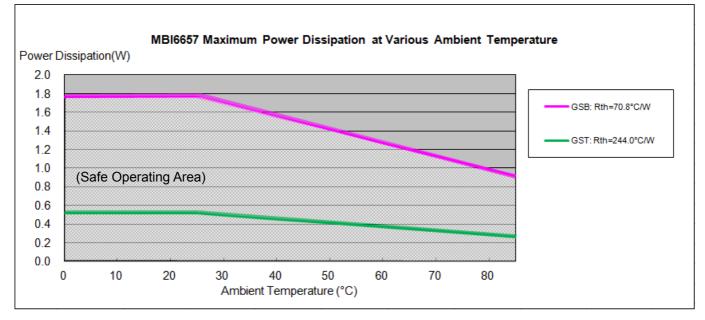
The rated voltage and capacitance are not the only concerns when selecting capacitors, but also the maximum ripple current. If the actual ripple current is larger than the specified maximum ripple current, the capacitor and the IC might be damaged. In general, the ripple current is related to the inductor ripple current. The specification of maximum ripple current of capacitor should be larger than 1.3 times of the inductor ripple current.

D. Output Capacitor Selection (Optional)

A capacitor paralleled with cascaded LED can reduce the LED ripple current and allow smaller inductance.

Package Power Dissipation (P_D)

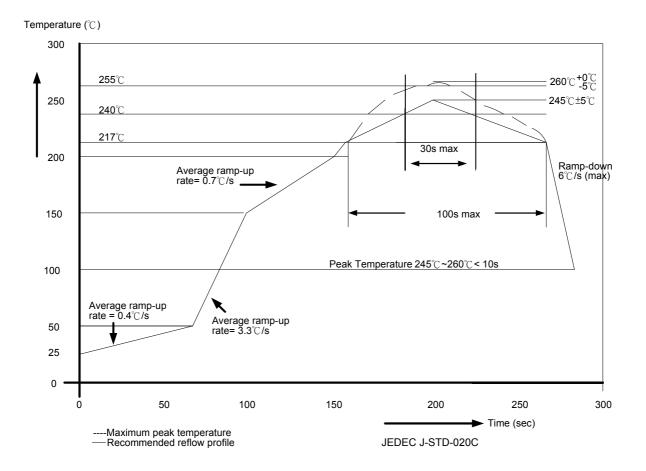
The maximum power dissipation. $P_{D}(max)=(Ti-Ta)/R_{th/i-a}$, decreases as the ambient temperature increases.



MBI6657 Step-Down, 1A Dimmable LED Driver Soldering Process of "Pb-free" Package Plating*

Macroblock has defined "Pb-Free & Green" to mean semiconductor products that are compatible with the current RoHS requirements and selected 100% pure tin (Sn) to provide forward and backward compatibility with both the current industry-standard SnPb-based soldering processes and higher-temperature Pb-free processes. Pure tin is widely accepted by customers and suppliers of electronic devices in Europe, Asia and the US as the lead-free surface finish of choice to replace tin-lead. Also, it adopts tin/lead (SnPb) solder paste, and please refer to the JEDEC J-STD-020C for the temperature of solder bath. However, in the whole Pb-free soldering processes and materials, 100% pure tin (Sn) will all require from 245 °C to 260°C for proper soldering on boards, referring to JEDEC J-STD-020C as shown below.

For managing MSL3 Package, it should refer to JEDEC J-STD-020C about floor life management & refer to JEDEC J-STD-033C about re-bake condition while IC's floor life exceeds MSL3 limitation.

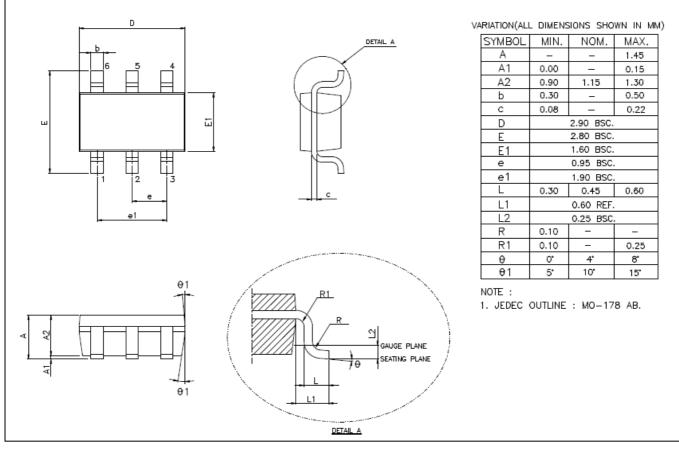


| Package Thickness | Volume mm ³ <350 | Volume mm ³ 350-2000 | Volume mm ³ \geq 2000 |
|-------------------|--------------------------------|------------------------------------|------------------------------------|
| <1.6mm | 260 + 0 °C | 260 + 0 °C | 260 + 0 °C |
| 1.6mm – 2.5mm | 260 + 0 °C | 250 + 0 °C | 245 + 0 °C |
| ≧2.5mm | 250 + 0 °C | 245 + 0 °C | 245 + 0 °C |

* For details, please refer to Macroblock's "Policy on Pb-free & Green Package".

Step-Down, 1A Dimmable LED Driver

Outline Drawing

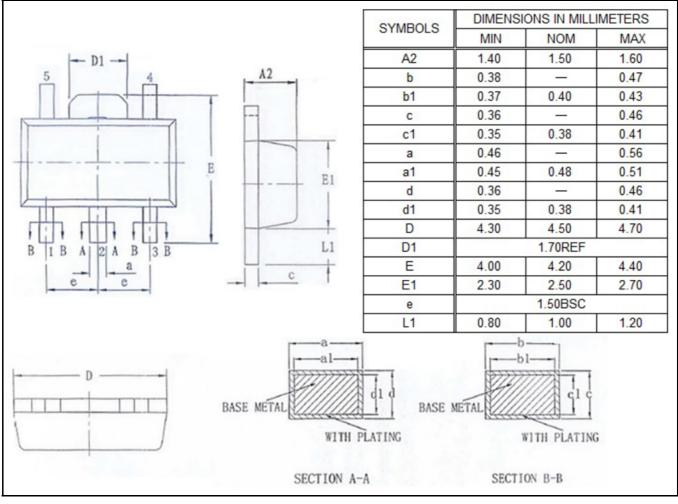


MBI6657GST Outline Drawing

http://YesLED.com / Info@YesLED.com

MBI6657

Step-Down, 1A Dimmable LED Driver

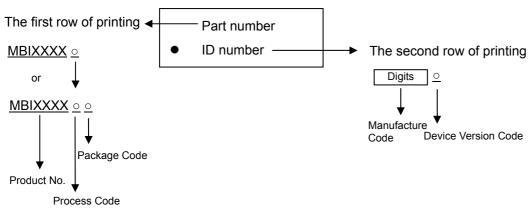


MBI6657GSB Outline Drawing

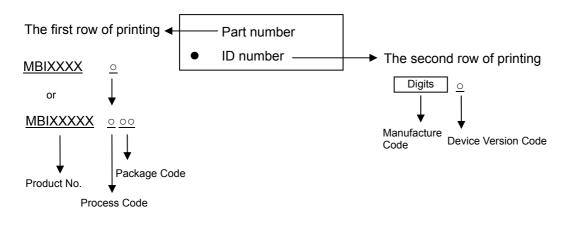
Note: Please use the maximum dimensions for the thermal pad layout. To avoid the short circuit risk, the vias or circuit traces shall not pass through the maximum area of thermal pad.

Product Top Mark Information

GST(SOT-23-6L)



GSB(SOT-89-5L)



Product Revision History

| Datasheet Version | Device Version Code |
|-------------------|---------------------|
| V1.00 | A |

Product Ordering Information

| Product Ordering Number* | RoHS Compliant Package Type | Weight (g) |
|-----------------------------|--------------------------------|------------|
| MBI6657GST-A | SOT-23-6L | 0.016 |
| MBI6657GSB-A | SOT-89-5L | 0.080 |

*Please place your order with the "product ordering number" information on your purchase order (PO).

Step-Down, 1A Dimmable LED Driver

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