

**MJE13009****NPN SILICON TRANSISTOR**

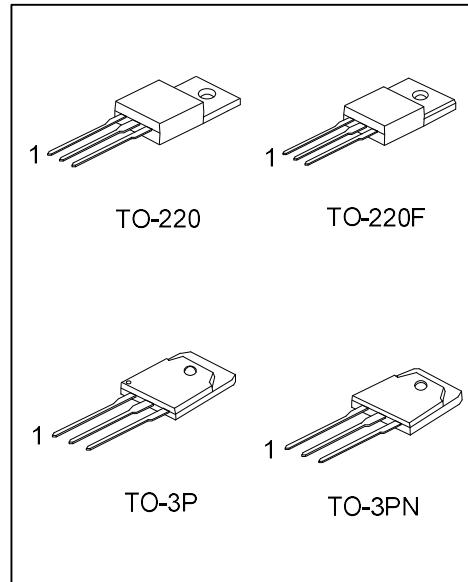
# **SWITCHMODE SERIES NPN SILICON POWER TRANSISTORS**

## ■ DESCRIPTION

The **MJE13009** is designed for high-voltage, high-speed power switching inductive circuits where fall time is critical. They are particularly suited for 115 and 220V switch mode applications such as Switching Regulators, Inverters, Motor Controls, Solenoid/Relay drivers and Deflection circuits.

## ■ FEATURES

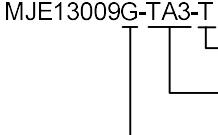
- \* V<sub>CEO</sub> 400V and 300V
- \* Reverse Bias SOA with Inductive Loads @ T<sub>c</sub> = 100°C
- \* Inductive Switching Matrix 3 ~ 12 Amp, 25 and 100°C  
t<sub>c</sub> @ 8A, 100°C is 120 ns (Typ.).
- \* 700 V Blocking Capability
- \* SOA and Switching Applications Information.



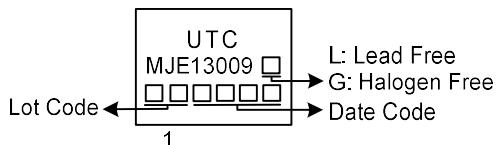
## ■ ORDERING INFORMATION

Ordering Number		Package	Pin Assignment			Packing
Lead Free	Halogen Free		1	2	3	
MJE13009L-TA3-T	MJE13009G-TA3-T	TO-220	B	C	E	Tube
MJE13009L-TF3-T	MJE13009G-TF3-T	TO-220F	B	C	E	Tube
MJE13009L-T3P-T	MJE13009G-T3P-T	TO-3P	B	C	E	Tube
MJE13009L-T3N-T	MJE13009G-T3N-T	TO-3PN	B	C	E	Tube

Note: Pin Assignment: B: Base    C: Collector    E: Emitter

 (1)Packing Type (2)Package Type (3)Green Package	(1) T: Tube (2) TA3: TO-220, TF3: TO-220F, T3P: TO-3P T3N: TO-3PN (3) G: Halogen Free and Lead Free, L: Lead Free	

## ■ MARKING



■ ABSOLUTE MAXIMUM RATINGS ( $T_A=25^\circ\text{C}$ , unless otherwise specified)

PARAMETER		SYMBOL	RATINGS	UNIT
Collector-Emitter Voltage		$V_{CEO}$	400	V
Collector-Emitter Voltage ( $V_{BE}=-1.5\text{V}$ )		$V_{CEV}$	700	V
Emitter Base Voltage		$V_{EBO}$	9	V
Collector Current	Continuous	$I_C$	12	A
	Peak (Note 3)	$I_{CM}$	24	A
Base Current	Continuous	$I_B$	6	A
	Peak (Note 3)	$I_{BM}$	12	A
Emitter Current	Continuous	$I_E$	18	A
	Peak (Note 3)	$I_{EM}$	36	A
Power Dissipation	TO-220	$P_D$	2	W
	TO-220F		2	W
	TO-3P		5.8	W
Derate above $25^\circ\text{C}$	TO-220/TO-220F		16	$\text{mW}/^\circ\text{C}$
	TO-3P		47	$\text{mW}/^\circ\text{C}$
Junction Temperature	$T_J$		+150	$^\circ\text{C}$
Storage Temperature	$T_{STG}$		-40 ~ +150	$^\circ\text{C}$

Notes: 1. Absolute maximum ratings are those values beyond which the device could be permanently damaged.

Absolute maximum ratings are stress ratings only and functional device operation is not implied.

2. Pulse Test: Pulse Width = 5ms, Duty Cycle  $\leq 10\%$ .

3. Pulse Test: Pulse Width = 300 $\mu\text{s}$ , Duty Cycle = 2%.

■ THERMAL DATA

PARAMETER		SYMBOL	RATINGS	UNIT
Junction to Ambient	TO-220/TO-220F	$\theta_{JA}$	62.5	$^\circ\text{C}/\text{W}$
	TO-3P		21	$^\circ\text{C}/\text{W}$
Junction to Case	TO-220	$\theta_{JC}$	1.56	$^\circ\text{C}/\text{W}$
	TO-220F		3.13	$^\circ\text{C}/\text{W}$
	TO-3P		0.6	$^\circ\text{C}/\text{W}$

■ ELECTRICAL CHARACTERISTICS ( $T_C=25^\circ\text{C}$ , unless otherwise specified)

PARAMETER	SYMBOL	TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>OFF CHARACTERISTICS (Note)</b>						
Collector-Emitter Sustaining Voltage	$V_{CEO}$	$I_C = 10\text{mA}, I_B = 0$	400			V
Collector Cutoff Current $V_{CBO}=\text{Rated Value}$	$I_{CEV}$	$V_{BE(OFF)} = 1.5\text{V}_{DC}$ $V_{BE(OFF)} = 1.5\text{V}_{DC}, T_C = 100^\circ\text{C}$			1 5	mA
Emitter Cutoff Current	$I_{EBO}$	$V_{EB} = 9\text{V}_{DC}, I_C = 0$			1	mA
<b>ON CHARACTERISTICS (Note)</b>						
DC Current Gain	$h_{FE1}$	$I_C = 5\text{A}, V_{CE} = 5\text{V}$			40	
	$h_{FE2}$	$I_C = 8\text{A}, V_{CE} = 5\text{V}$			30	
Current-Emitter Saturation Voltage	$V_{CE(SAT)}$	$I_C = 5\text{A}, I_B = 1\text{A}$			1	V
		$I_C = 8\text{A}, I_B = 1.6\text{A}$			1.5	V
		$I_C = 12\text{A}, I_B = 3\text{A}$			3	V
		$I_C = 8\text{A}, I_B = 1.6\text{A}, T_C = 100^\circ\text{C}$			2	V
Base-Emitter Saturation Voltage	$V_{BE(SAT)}$	$I_C = 5\text{A}, I_B = 1\text{A}$			1.2	V
		$I_C = 8\text{A}, I_B = 1.6\text{A}$			1.6	V
		$I_C = 8\text{A}, I_B = 1.6\text{A}, T_C = 100^\circ\text{C}$			1.5	V
<b>DYNAMIC CHARACTERISTICS</b>						
Transition frequency	$f_T$	$I_C = 500\text{mA}, V_{CE} = 10\text{V}, f = 1\text{MHz}$	4			MHz
Output Capacitance	$C_{OB}$	$V_{CB} = 10\text{V}, I_E = 0, f = 0.1\text{MHz}$		180		pF
<b>SWITCHING CHARACTERISTICS (Resistive Load, Table 1)</b>						
Delay Time	$t_{DLY}$	$V_{CC} = 125\text{V}_{DC}, I_C = 8\text{A}$ $I_{B1} = I_{B2} = 1.6\text{A}, t_P = 25\mu\text{s}$ Duty Cycle $\leq 1\%$		0.06	0.1	$\mu\text{s}$
Rise Time	$t_R$			0.45	1	$\mu\text{s}$
Storage Time	$t_S$			1.3	3	$\mu\text{s}$
Fall Time	$t_F$			0.2	0.7	$\mu\text{s}$
<b>Inductive Load, Clamped (Table 1, Fig. 13)</b>						
Voltage Storage Time	$t_S$	$I_C = 8\text{A}, V_{CLAMP} = 300\text{V}, I_{B1} = 1.6\text{A}$		0.92	2.3	$\mu\text{s}$
Crossover Time	$t_C$	$V_{BE(OFF)} = 5\text{V}, T_C = 100^\circ\text{C}$		0.12	0.7	$\mu\text{s}$

Note: Pulse Test: Pulse Width = 300 $\mu\text{s}$ , Duty Cycle = 2%.

■ TABLE 1. TEST CONDITIONS FOR DYNAMIC PERFORMANCE

REVERSE BIAS SAFE OPERATING AREA AND INDUCTIVE SWITCHING		RESISTIVE SWITCHING
TEST CIRCUITS		
<p>Note: P<sub>w</sub> and V<sub>CC</sub> Adjusted for Desired I<sub>c</sub> R<sub>B</sub> Adjusted for Desired I<sub>b1</sub></p> <p>DUTY CYCLE ≤ 10% t<sub>R</sub>, t<sub>F</sub> ≤ 10 ns</p>	<p>*SELECTED FOR .1 KV</p>	
CIRCUIT VALUES	<p>Coil Data: Ferroxcube Core #6656 Full Bobbin (~16 Turns) #16</p> <p>GAP for 200μH/20A L<sub>COIL</sub> = 200μH</p>	<p>V<sub>CC</sub> = 20V V<sub>CLAMP</sub> = 300V<sub>DC</sub></p> <p>V<sub>CC</sub> = 125V R<sub>C</sub> = 15Ω D1 = 1N5820 or Equiv. R<sub>B</sub> = Ω</p>
TEST WAVEFORMS	<p>OUTPUT WAVEFORMS</p> <p>t1 ADJUSTED TO OBTAIN IC</p> $t_1 \approx \frac{L_{COIL} (I_{CM})}{V_{CC}}$ $t_2 \approx \frac{L_{COIL} (I_{CM})}{V_{CLAMP}}$	<p>t<sub>R</sub>, t<sub>F</sub> &lt; 10 ns Duty Cycle = 1.0% R<sub>B</sub> and R<sub>C</sub> adjusted for desired I<sub>B</sub> and I<sub>C</sub></p>

■ TABLE 2. APPLICATIONS EXAMPLES OF SWITCHING CIRCUITS

CIRCUIT	LOAD LINE DIAGRAMS	TIME DIAGRAMS
SERIES SWITCHING REGULATOR	<p>TURN-ON (FORWARD BIAS) SOA  <math>t_{ON} \leq 10\text{ ms}</math>  DUTY CYCLE <math>\leq 10\%</math>  <math>P_D = 4000\text{ W}</math> ②</p> <p><math>T_C = 100^\circ\text{C}</math></p> <p>TURN-ON (REVERSE BIAS) SOA  <math>1.5\text{ V} \leq V_{BE(OFF)} \leq 9.0\text{ V}</math>  DUTY CYCLE <math>\leq 10\%</math></p> <p>Collector Current: 24A, 12A  Collector Voltage: 350V, 400V ①, 700V ①</p>	<p><math>I_C</math> vs TIME</p> <p><math>V_{CE}</math> vs TIME</p>
RINGING CHOKE INVERTER	<p>TURN-ON (FORWARD BIAS) SOA  <math>t_{ON} \leq 10\text{ ms}</math>  DUTY CYCLE <math>\leq 10\%</math></p> <p><math>T_C = 100^\circ\text{C}</math></p> <p>TURN-OFF (REVERSE BIAS) SOA  <math>1.5\text{ V} \leq V_{BE(OFF)} \leq 9.0\text{ V}</math>  DUTY CYCLE <math>\leq 10\%</math></p> <p>Collector Current: 24A, 12A  Collector Voltage: 350V, 400V ①, 700V ①</p>	<p><math>I_C</math> vs TIME</p> <p><math>V_{CE}</math> vs TIME</p> <p>LEAKAGE SPIKE</p>
PUSH-PULL INVERTER/CONVERTER	<p>TURN-ON (FORWARD BIAS) SOA  <math>t_{ON} \leq 10\text{ ms}</math>  DUTY CYCLE <math>\leq 10\%</math></p> <p><math>T_C = 100^\circ\text{C}</math></p> <p>TURN-OFF (REVERSE BIAS) SOA  <math>1.5\text{ V} \leq V_{BE(OFF)} \leq 9.0\text{ V}</math>  DUTY CYCLE <math>\leq 10\%</math></p> <p>Collector Current: 24A, 12A  Collector Voltage: 350V, 400V ①, 700V ①, 2V<sub>CC</sub></p>	<p><math>I_C</math> vs TIME</p> <p><math>V_{CE}</math> vs TIME</p>
SOLENOID DRIVER	<p>TURN-ON (FORWARD BIAS) SOA  <math>t_{ON} \leq 10\text{ ms}</math>  DUTY CYCLE <math>\leq 10\%</math></p> <p><math>T_C = 100^\circ\text{C}</math></p> <p>TURN-OFF (REVERSE BIAS) SOA  <math>1.5\text{ V} \leq V_{BE(OFF)} \leq 9.0\text{ V}</math>  DUTY CYCLE <math>\leq 10\%</math></p> <p>Collector Current: 24A, 12A  Collector Voltage: 350V, 400V ①, 700V ①, 2V<sub>CC</sub></p>	<p><math>I_C</math> vs TIME</p> <p><math>V_{CE}</math> vs TIME</p>

**■ TABLE 3. TYPICAL INDUCTIVE SWITCHING PERFORMANCE**

$I_C(A)$	$T_C(^{\circ}\text{C})$	$t_{SV}(\text{ns})$	$t_{RV}(\text{ns})$	$t_{FI}(\text{ns})$	$t_{TI}(\text{ns})$	$t_C(\text{ns})$
3	25	770	100	150	200	240
	100	1000	230	160	200	320
5	25	630	72	26	10	100
	100	820	100	55	30	180
8	25	720	55	27	2	77
	100	920	70	50	8	120
12	25	640	20	17	2	41
	100	800	32	24	4	54

**■ SWITCHING TIME NOTES**

In resistive switching circuits, rise, fall, and storage times have been defined and apply to both current and voltage waveforms since they are in phase. However, for inductive loads which are common to SWITCHMODE power supplies and hammer drivers, current and voltage waveforms are not in phase. Therefore, separate measurements must be made on each waveform to determine the total switching time. For this reason, the following new terms have been defined.

$t_{SV}$  = Voltage Storage Time, 90%  $I_{B1}$  to 10%  $V_{CEM}$

$t_{RV}$  = Voltage Rise Time, 10–90%  $V_{CEM}$

$t_{FI}$  = Current Fall Time, 90–10%  $I_{CM}$

$t_{TI}$  = Current Tail, 10–2%  $I_{CM}$

$t_C$  = Crossover Time, 10%  $V_{CEM}$  to 10%  $I_{CM}$

An enlarged portion of the turn-off waveforms is shown in Fig. 13 to aid in the visual identity of these terms.

For the designer, there is minimal switching loss during storage time and the predominant switching power losses occur during the crossover interval and can be obtained using the standard equation from AN-222:

$$P_{SWT} = 1/2 V_{CC} I_C(t_C) f$$

Typical inductive switching waveforms are shown in Fig. 14. In general,  $t_{RV} + t_{FI} \approx t_C$ . However, at lower test currents this relationship may not be valid.

As is common with most switching transistors, resistive switching is specified at 25°C and has become a benchmark for designers. However, for designers of high frequency converter circuits, the user oriented specifications which make this a "SWITCHMODE" transistor are the inductive switching speeds ( $t_C$  and  $t_{SV}$ ) which are guaranteed at 100°C.

## ■ TYPICAL CHARACTERISTICS

Fig. 1 Forward Bias Safe Operating Area

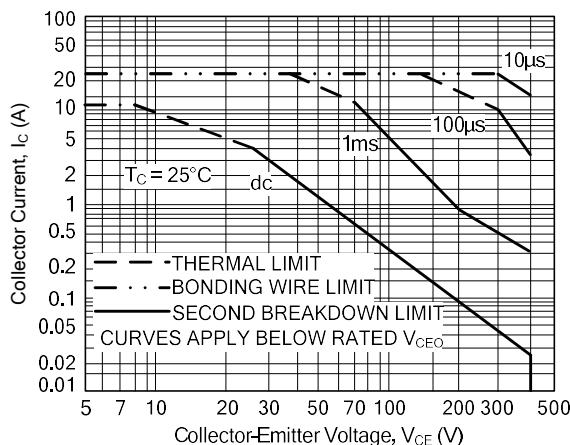


Fig. 2 Reverse Bias Switching Safe Operating Area

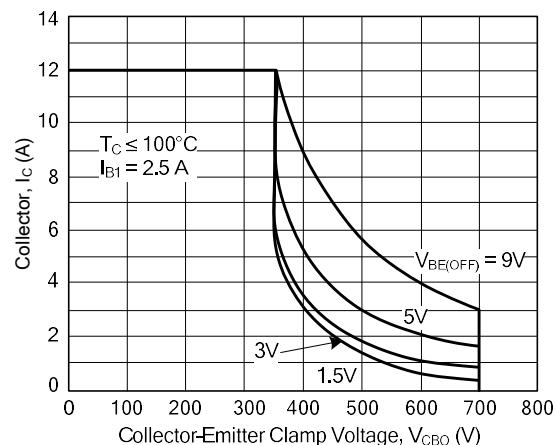
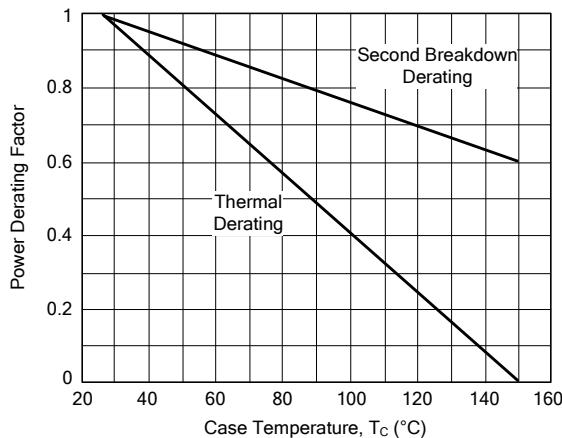


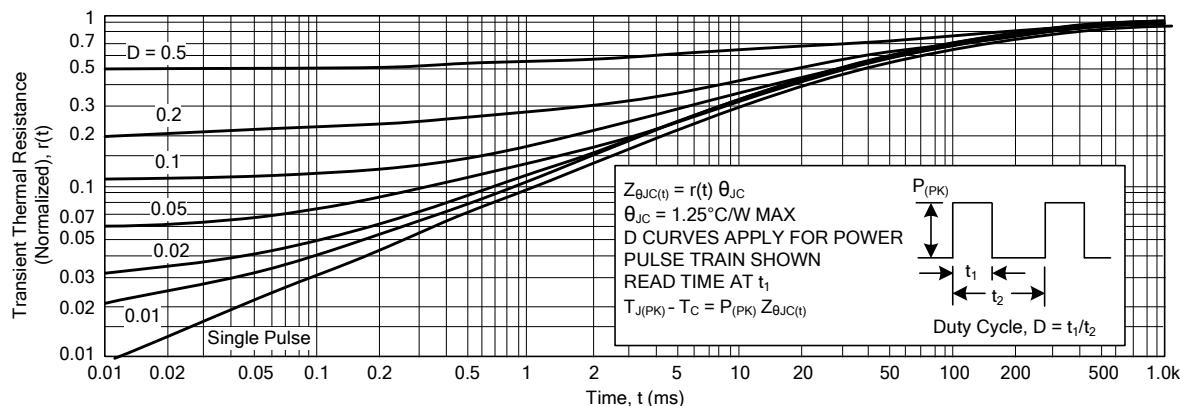
Fig. 3 Forward Bias Power Derating



There are two limitations on the power handling ability of a transistor: average junction temperature and second breakdown. Safe operating area curves indicate  $I_c$ - $V_{CE}$  limits of the transistor that must be observed for reliable operation; i.e., the transistor must not be subjected to greater dissipation than the curves indicate.

The data of Fig. 1 is based on  $T_c=25^\circ\text{C}$ ;  $T_{J(PK)}$  is variable depending on power level. Second breakdown pulse limits are valid for duty cycles to 10% but must be derated when  $T_c \geq 25^\circ\text{C}$ . Second breakdown limitations do not derate the same as thermal limitations. Allowable current at the voltages shown on Fig. 1 may be found at any case temperature by using the appropriate curve on Fig. 3.

$T_{J(PK)}$  may be calculated from the data in Fig. 4. At high case temperatures, thermal limitations will reduce the power that can be handled to values less than the limitations imposed by second breakdown. Use of reverse biased safe operating area data (Fig. 2) is discussed in the applications information section.

Fig. 4 Typical Thermal Response [ $Z_{\theta JC}(t)$ ]

### ■ TYPICAL CHARACTERISTICS (Cont.)

Fig. 5 DC Current Gain

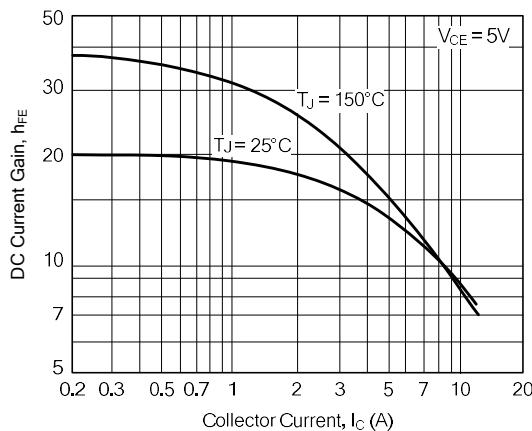


Fig. 6 Collector Saturation Region

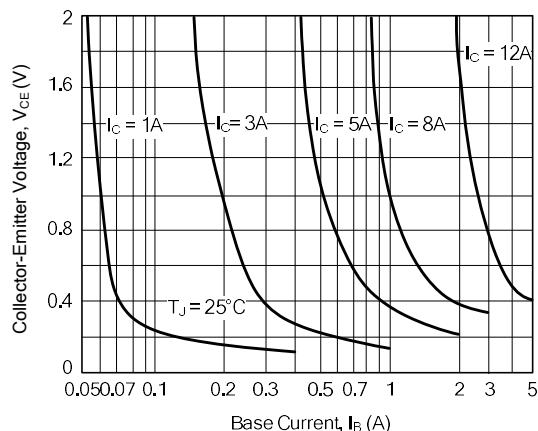


Fig. 7 Base-Emitter Saturation Voltage

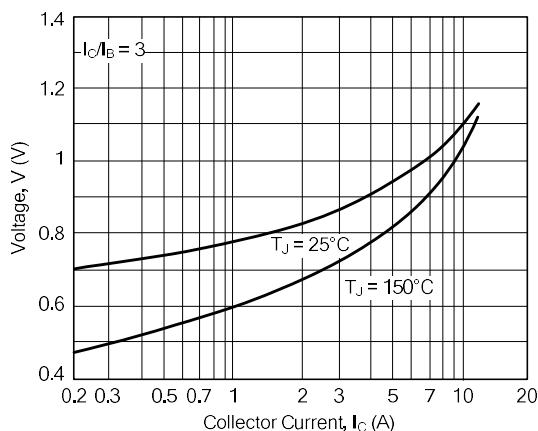


Fig. 8 Collector-Emitter Saturation Voltage

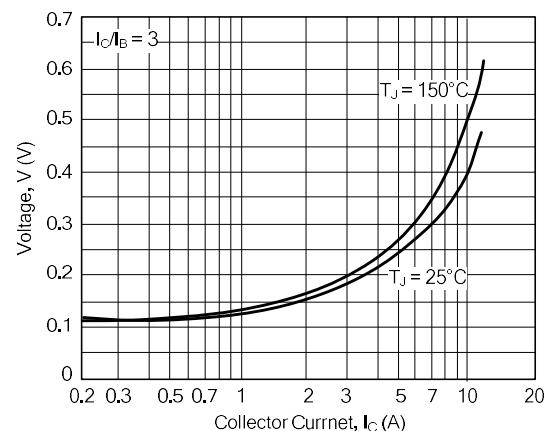


Fig. 9 Collector Cutoff Region

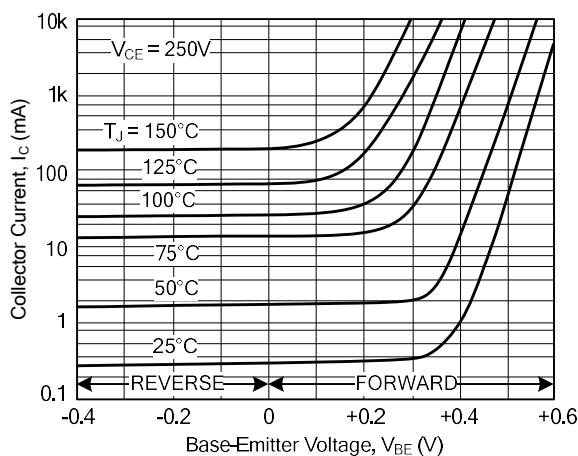
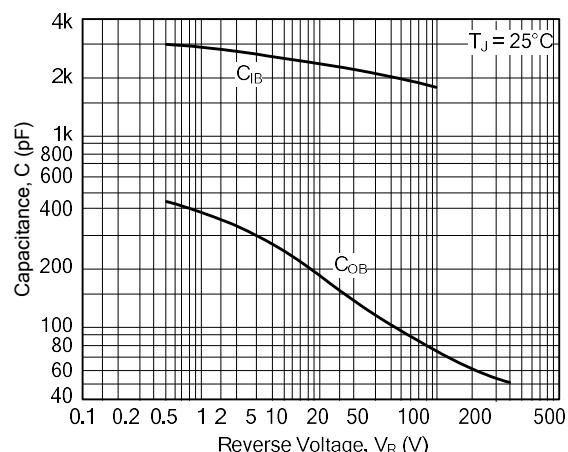


Fig. 10 Capacitance



### ■ RESISTIVE SWITCHING PERFORMANCE

Fig. 11. Turn-On Time

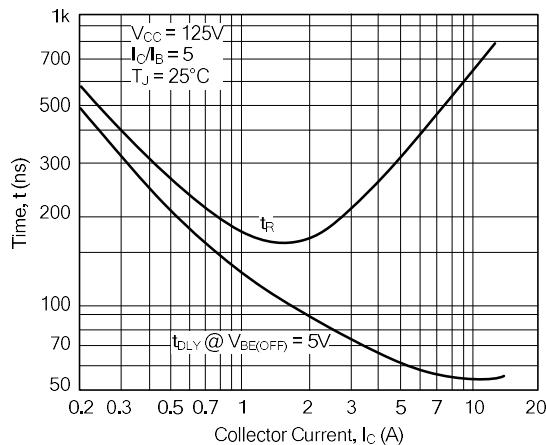


Fig. 12 Turn-Off Time

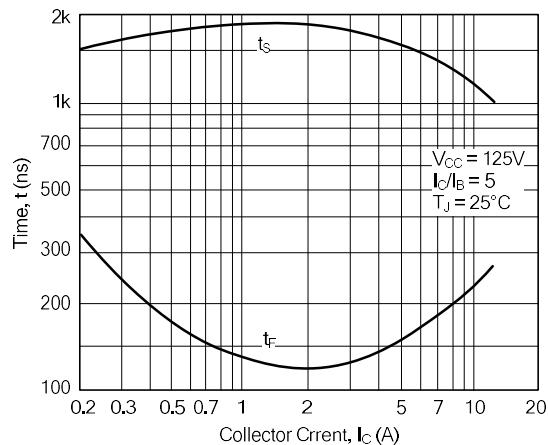
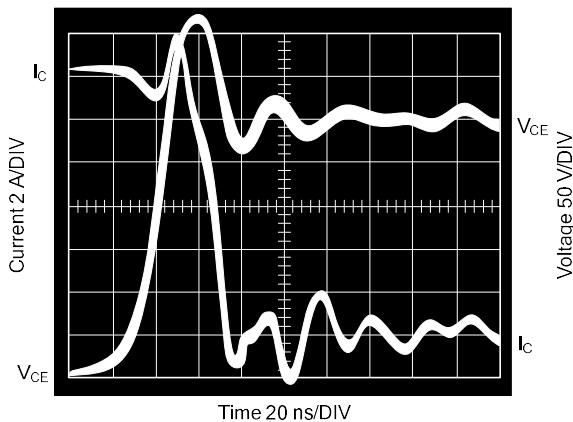


Fig. 13 Typical Inductive Switching Waveforms  
(at 300V and 12A with  $I_{B1} = 2.4A$  and  $V_{BE(off)} = 5V$ )



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