LANSDALE
Semiconductor, Inc.

ML1490
RF/IF/Audio Amplifier
Wideband Amplifier With AGC

Legacy Device: *Motorola MC1490*
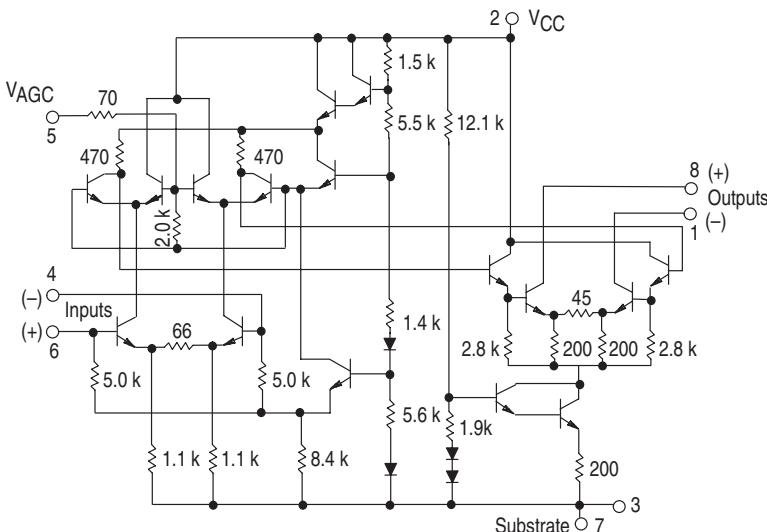
The ML1490 is an integrated circuit featuring wide–range AGC for use in RF/IF amplifiers and audio amplifiers.

- High Power Gain:
    - 50 dB Typ at 10 MHz
    - 45 dB Typ at 60 MHz
    - 35 dB Typ at 100 MHz
- Wide Range AGC: 60 dB Min, DC to 60 MHz
- 6.0 V to 15 V Operation, Single Polarity Supply
- Operating Temperature Range T_A = –40° to +85°C
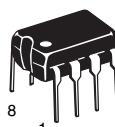
Note: See Similar ML1350 For Possible Option

MAXIMUM RATINGS (T_A = +25°C, unless otherwise noted.)

Rating	Symbol	Value	Unit
Power Supply Voltage	V_{CC}	+18	Vdc
AGC Supply	V_{AGC}	V_{CC}	Vdc
Input Differential Voltage	V_{ID}	5.0	Vdc
Operating Temperature Range	T_A	–40 to +85	°C
Storage Temperature Range	T_{stg}	–65 to +150	°C
Junction Temperature	T_J	+150	°C

Representative Schematic Diagram

Pins 3 and 7 should both be connected to circuit ground.
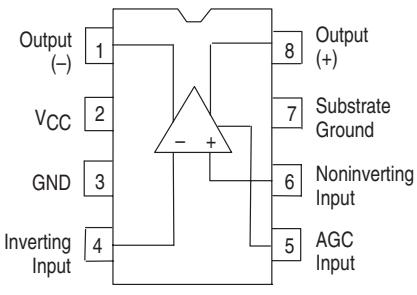
**P DIP 8 = PP**
PLASTIC PACKAGE
CASE 626

CROSS REFERENCE/ORDERING INFORMATION

PACKAGE	MOTOROLA	LANSDALE
P DIP 8	MC1490P	ML1490PP

Note: Lansdale lead free (**Pb**) product, as it becomes available, will be identified by a part number prefix change from **ML** to **MLE**.

PIN CONNECTIONS

Output (–)	1	8 Output (+)
V_{CC}	2	7 Substrate Ground
GND	3	6 Noninverting Input
Inverting Input	4	5 AGC Input
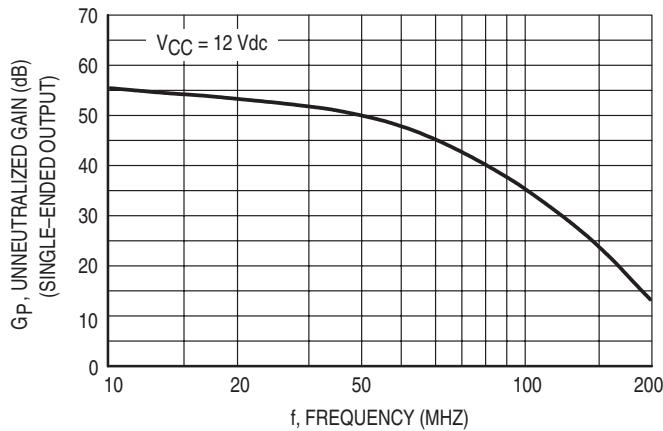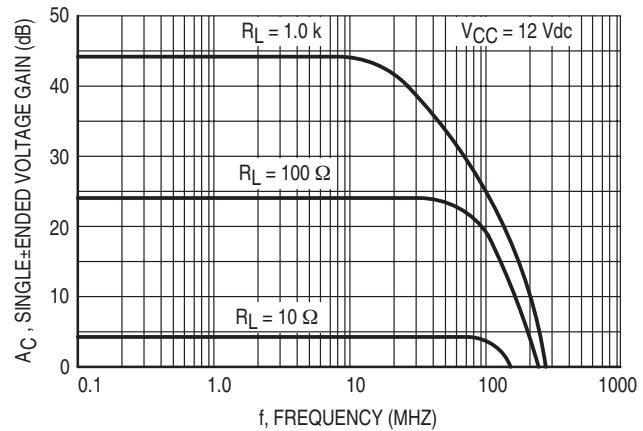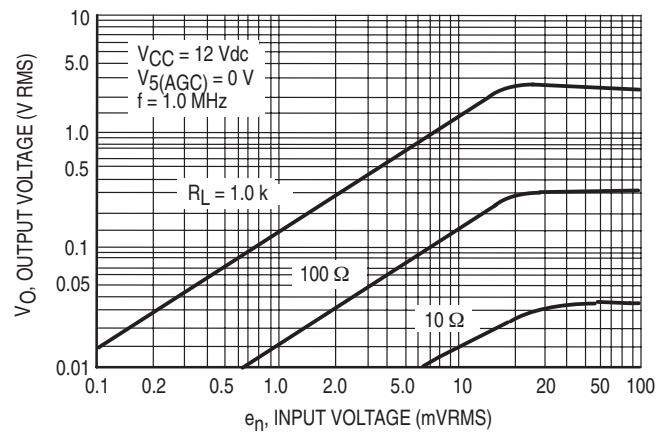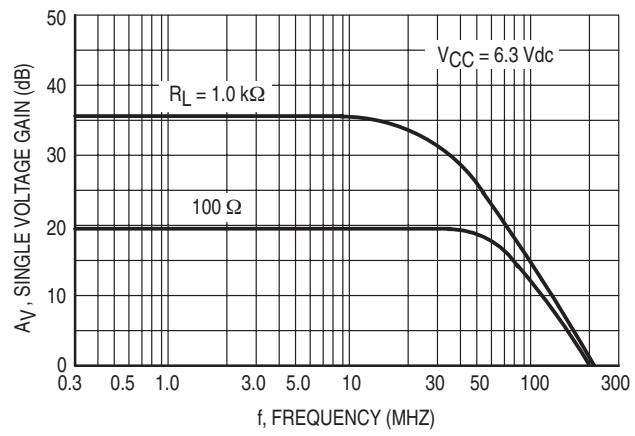
(Top View)

SCATTERING PARAMETERS
(V_{CC} = +12 Vdc, T_A = +25°C, Z_O = 50 Ω)

Parameter	Symbol	f = MHz Typ 30	60	Unit
Input Reflection Coefficient	$|S_{11}|$ $\theta 11$	0.95 −7.3	0.93 −16	– deg
Output Reflection Coefficient	$|S_{22}|$ $\theta 22$	0.99 −3.0	0.98 −5.5	– deg
Forward Transmission Coefficient	$|S_{21}|$ $\theta 21$	16.8 128	14.7 64.3	– deg
Reverse Transmission Coefficient	S_{12} $\theta 12$	0.00048 84.9	0.00092 79.2	– deg

ELECTRICAL CHARACTERISTICS (V_{CC} = 12 Vdc, f = 60 MHz, BW = 1.0 MHz, T_A = 25°C)

| Characteristic	Figure	Symbol	Min	Typ	Max	Unit
Power Supply Current Drain	–	I_{CC}	–	–	17	mA
AGC Range (AGC) 5.0 V Min to 7.0 V Max	19	M_{AGC}	–60	–	–	dB
Output Stage Current (Sum of Pins 1 and 8)	–	I_O	4.0	–	7.5	mA
Single–Ended Power Gain R_S = R_L = 50 Ω	19	G_P	40	–	–	dB
Noise Figure R_S = 50 Ohms	19	NF	–	6.0	–	dB
Power Dissipation	–	P_D	–	168	204	mW

Figure 1. Unneutralized Power Gain versus Frequency (Tuned Amplifier, See Figure 19)


Figure 2. Voltage Gain versus Frequency (Video Amplifier, See Figure 20)

Figure 3. Dynamic Range: Output Voltage versus Input Voltage (Video Amplifier, See Figure 20)


Figure 4. Voltage Gain versus Frequency (Video Amplifier, See Figure 20)

Legacy Applications Information

**Figure 5. Voltage Gain and Supply Current versus
Supply Voltage (Video Amplifier, See Figure 20)**
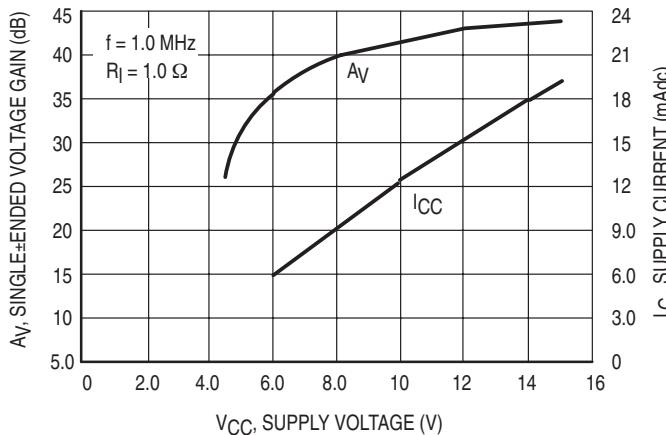

**Figure 6. Typical Gain Reduction
versus AGC Voltage**
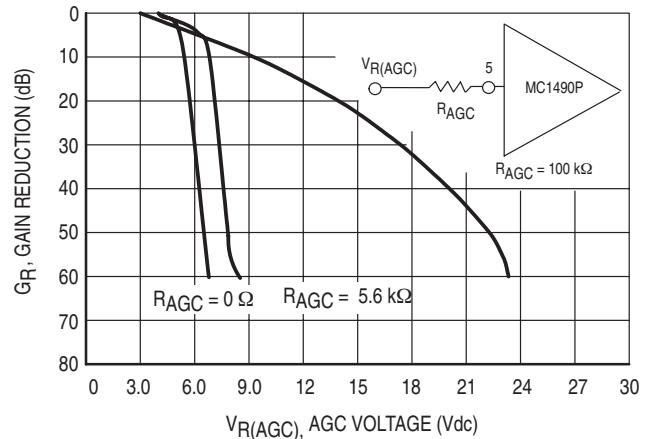
**Figure 7. Typical Gain Reduction
versus AGC Current**
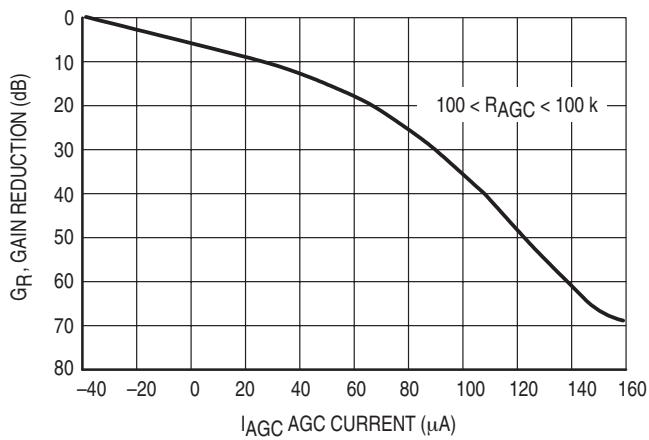
**Figure 8. Fixed Tuned Power Gain Reduction versus
Temperature (See Test Circuit, Figure 19)**
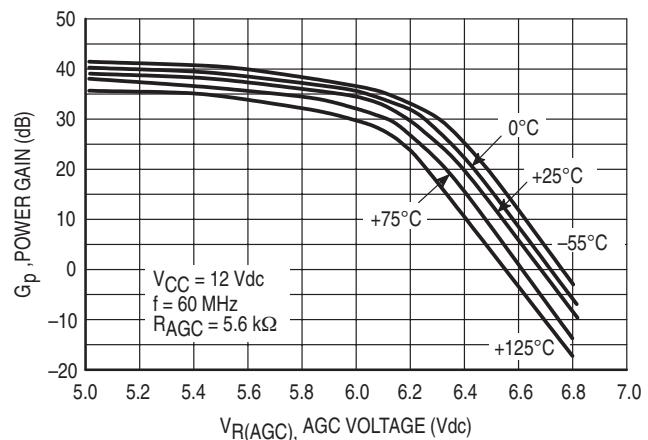
**Figure 9. Power Gain versus Supply Voltage
(See Test Circuit, Figure 19)**
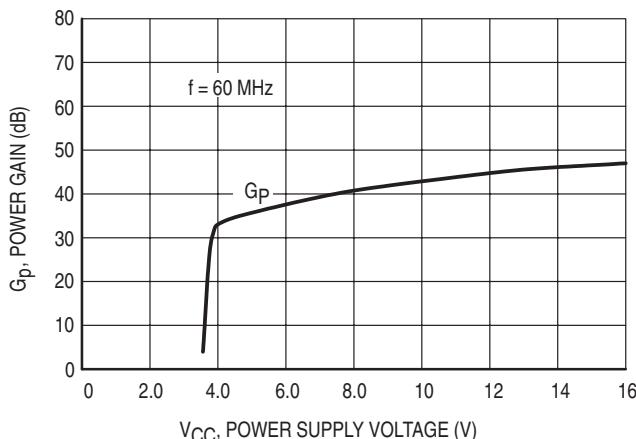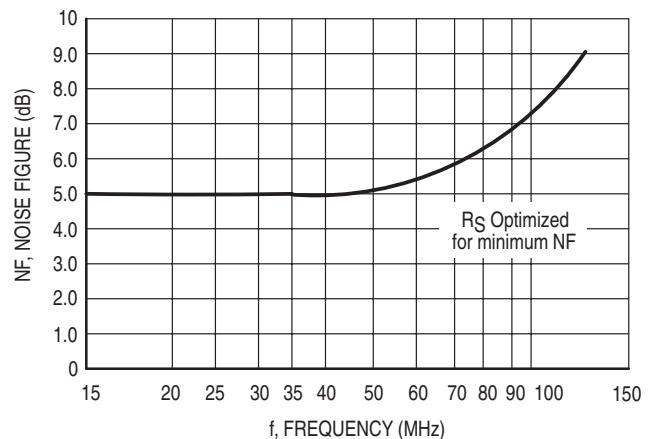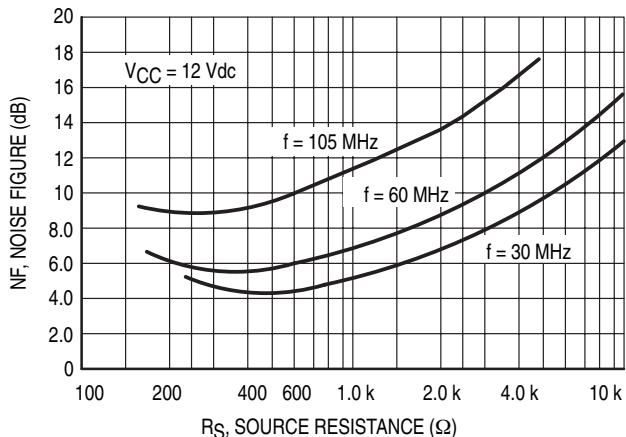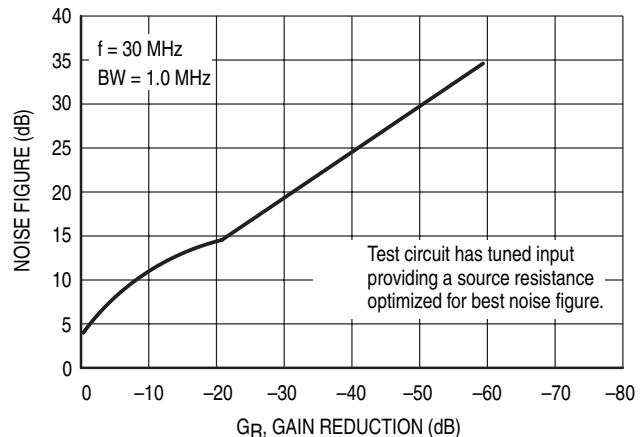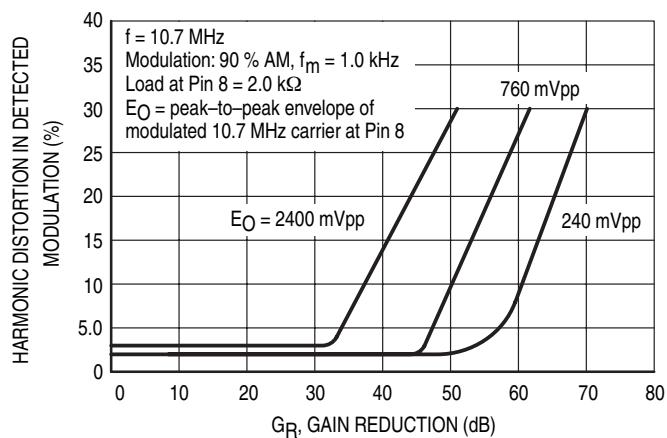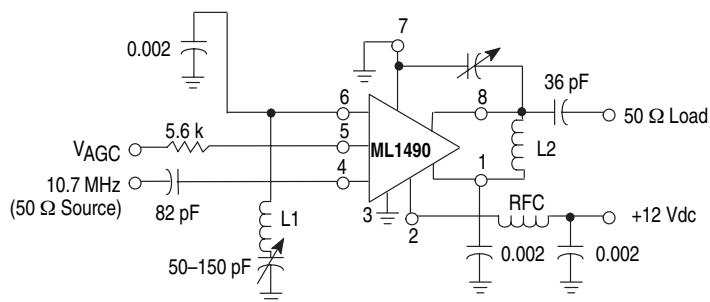
**Figure 10. Noise Figure versus Frequency**

Legacy Applications Information

**Figure 11. Noise Figure versus
Source Resistance**

V_{CC} = 12 Vdc

f = 105 MHz

f = 60 MHz

f = 30 MHz

NF, NOISE FIGURE (dB)

R_S, SOURCE RESISTANCE (Ω)

**Figure 12. Noise Figure versus
AGC Gain Reduction**

f = 30 MHz
BW = 1.0 MHz

NOISE FIGURE (dB)

Test circuit has tuned input
providing a source resistance
optimized for best noise figure.

G_R, GAIN REDUCTION (dB)

**Figure 13. Harmonic Distortion versus AGC Gain
Reduction for AM Carrier (For Test Circuit, See Figure 14)**

f = 10.7 MHz
Modulation: 90 % AM, f_m = 1.0 kHz
Load at Pin 8 = 2.0 kΩ
E_O = peak–to–peak envelope of
modulated 10.7 MHz carrier at Pin 8

760 mVpp

E_O = 2400 mVpp

240 mVpp

HARMONIC DISTORTION IN DETECTED MODULATION (%)

G_R, GAIN REDUCTION (dB)

Figure 14. 10.7 MHz Amplifier Gain   55 dB, BW   100 kHz

0.002

7

6
5
4

8

36 pF

50 Ω Load

V_{AGC}

5.6 k

ML1490

L2

10.7 MHz
(50 Ω Source)

82 pF

L1

1

RFC

+12 Vdc

3

2

0.002   0.002

50–150 pF

L1 = 24 turns, #22 AWG wire
on a T12–44 micro metal
Toroid core (–124 pF)

L2 = 20 turns, #22 AWG wire
on a T12–44 micro metal
Toroid core (–100 pF)

Legacy Applications Information

Figure 15. S$_{11}$ and S$_{22}$, Input and Output Reflection Coefficient
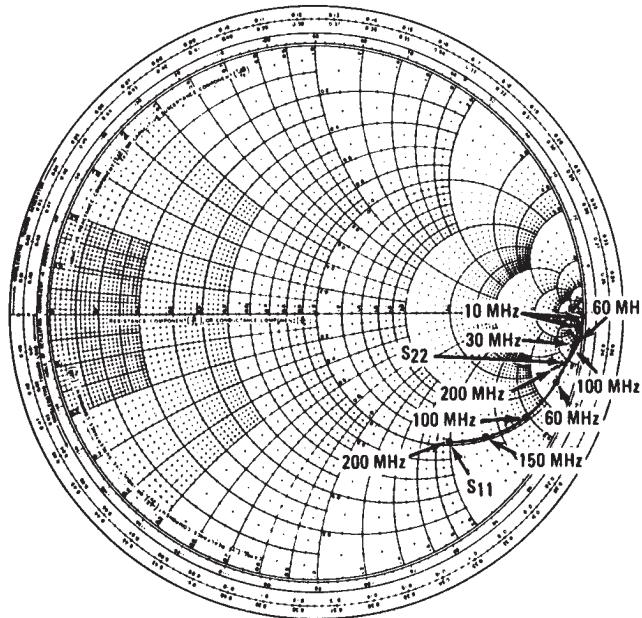
Figure 16. S$_{11}$ and S$_{22}$, Input and Output Reflection Coefficient
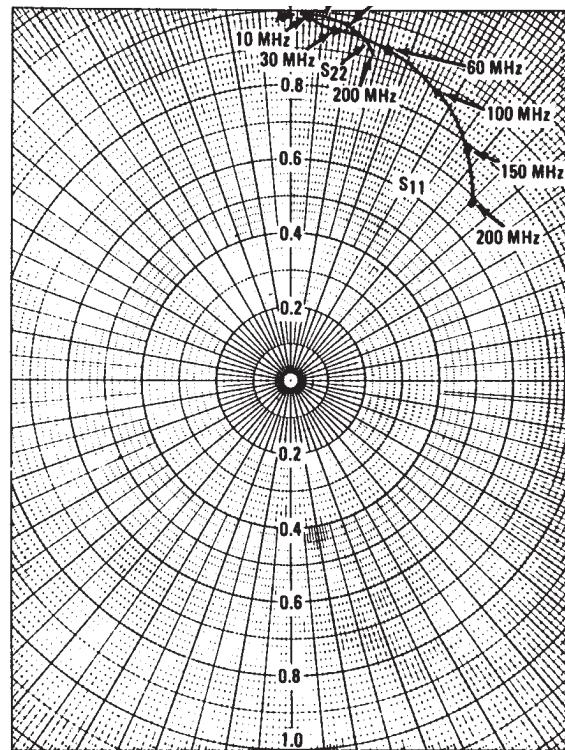

Figure 17. S$_{21}$, Forward Transmission Coefficient (Gain)
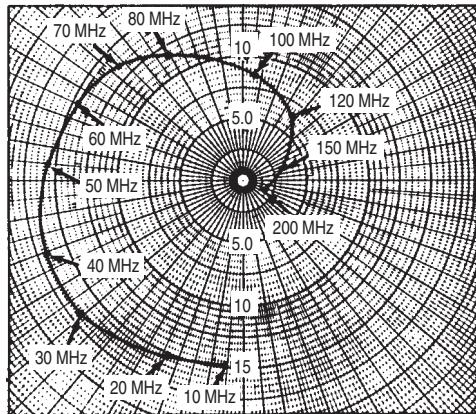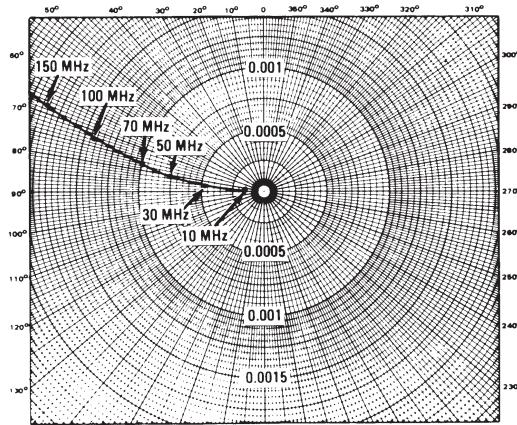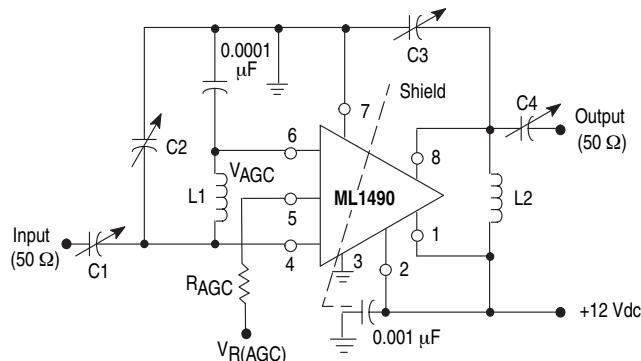

Figure 18. S$_{12}$, Reverse Transmission Coefficient (Feedback)
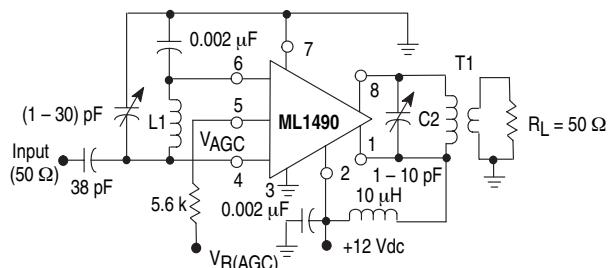
Legacy Applications Information

Figure 19. 60 MHz Power Gain Test Circuit

L1 = 7 turns, #20 AWG wire, 5/16" Dia.,5/8" long
L2 = 6 turns, #14 AWG wire, 9/16" Dia.,3/4" long
C1,C2,C3 = (1–30) pF
C4 = (1–10) pF

Figure 20. Video Amplifier

Figure 21. 30 MHz Amplifier
(Power Gain = 50 dB, BW   1.0 MHz)

L1 = 12 turns, #22 AWG wire on a Toroid core,
(T37–6 micro metal or equiv).
T1: Primary = 17 turns, #20 AWG wire on a Toroid core, (T44–6).
Secondary = 2 turns, #20 AWG wire.
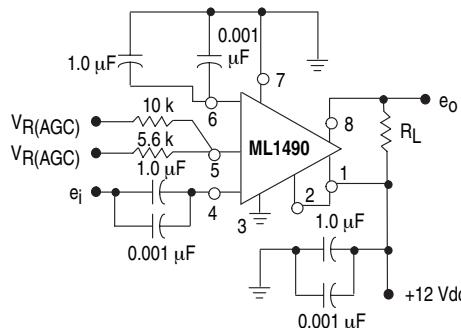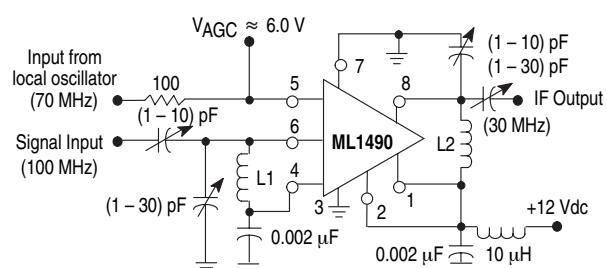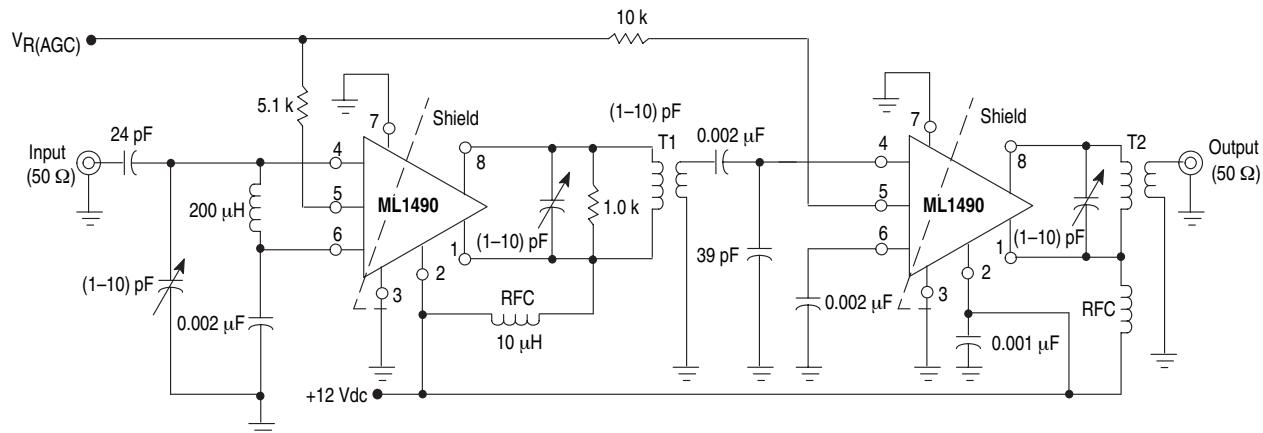
Figure 22. 100 MHz Mixer

L1 = 5 turns, #16 AWG wire, 1/4", ID Dia., 5/8" long
L2 = 16 turns, #20 AWG wire on a Toroid core, (T44–6).

Figure 23. Two–Stage 60 MHz IF Amplifier (Power Gain 80 dB, BW 1.5 MHz)

T1: Primary Winding = 15 turns, #22 AWG wire, 1/4" ID Air Core
     Secondary Winding = 4 turns, #22 AWG wire,
     Coefficient of Coupling ≈ 1.0

T2: Primary Winding = 10 turns, #22 AWG wire, 1/4" ID Air Core
     Secondary Winding = 2 turns, #22 AWG wire,
     Coefficient of Coupling ≈ 1.0

DESCRIPTION OF SPEECH COMPRESSOR

The amplifier drives the base of a PNP transistor operating common–emitter with a voltage gain of approximately 20. The control R1 varies the quiescent Q point of this transistor so that varying amounts of signal exceed the level V_r. Diode D1 rectifies the positive peaks of Q1's output only when these peaks are greater than $V_r \approx 7.0$ V. The resulting output is filtered by C_X, R_X.

R_X controls the charging time constant or attack time. C_X is involved in both charge and discharge. R2 (the 150 kΩ and input resistance of the emitter–follower Q2) controls the decay time. Making the decay long and attack short is accomplished by making R_X small and R2 large. (A Darlington emitter–follower may be needed if extremely slow decay times are required.)

The emitter–follower Q2 drives the AGC Pin 5 of the ML1490PP and reduces the gain. R3 controls the slope of signal compression.
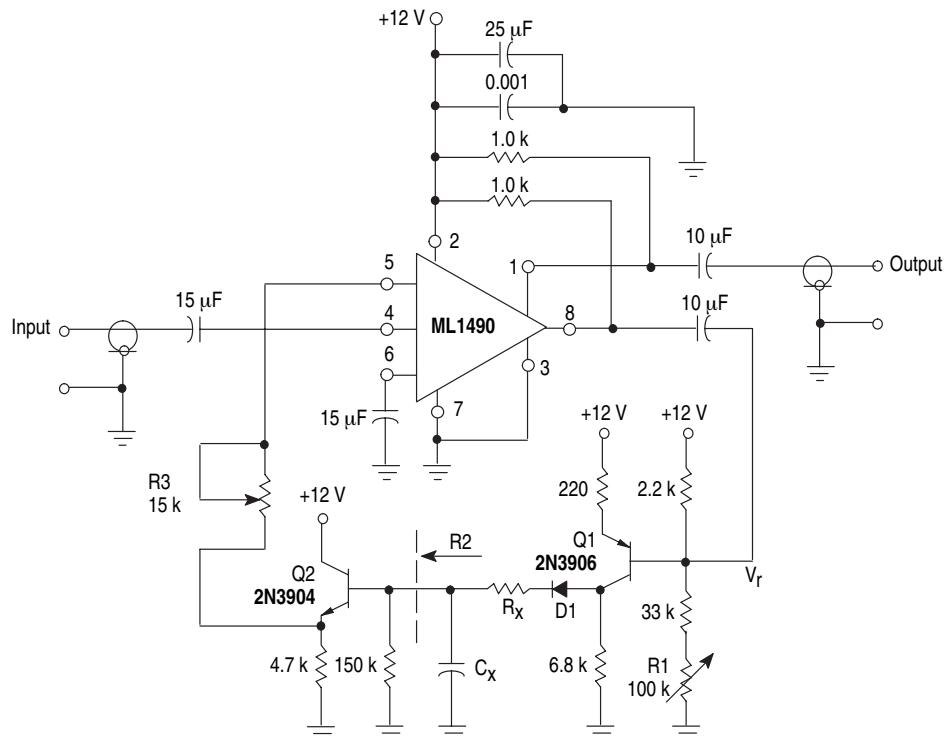
Table 1. Distortion versus Frequency

Frequency	Distortion		Distortion	
	10 mV e_i	100 mV e_i	10 mV e_i	100 mV e_i
100 Hz	3.5%	12%	15%	27%
300 Hz	2%	10%	6%	20%
1.0 kHz	1.5%	8%	3%	9%
10 kHz	1.5%	8%	1%	3%
100 kHz	1.5%	8%	1%	3%
	Notes 1 and 2		Notes 3 and 4	
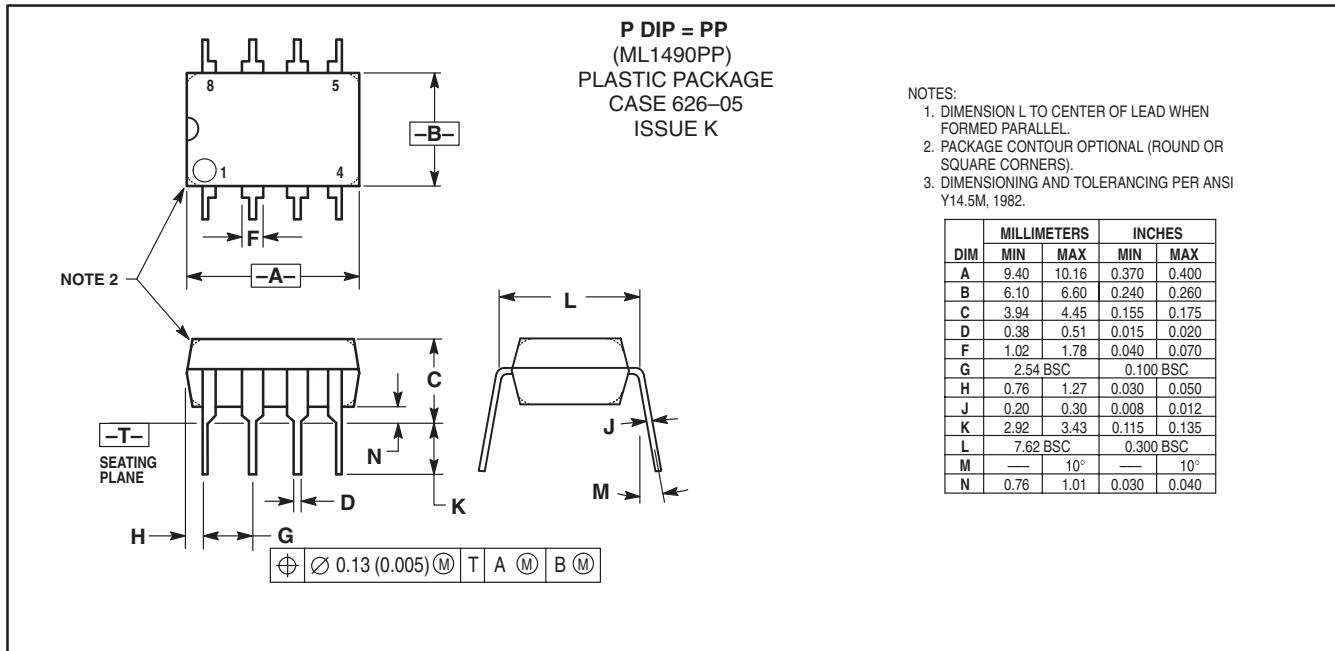
Notes:
(1) Decay = 300 ms
    Attack = 20 ms
(2) C_X = 7.5 μF
    R_X = 0 (Short)
(3) Decay = 20 ms
    Attack = 3.0 ms
(4) C_X = 0.68 μF
    R_X = 1.5 kΩ

Figure 24. Speech Compressor

OUTLINE DIMENSIONS

P DIP = PP
(ML1490PP)
PLASTIC PACKAGE
CASE 626–05
ISSUE K

8 5

1 4

—B—

NOTE 2

—F—

—A—

L

C

—T—
SEATING
PLANE

N

J

D

M

K

H

G

⊕ | ⌀ 0.13 (0.005) Ⓜ | T | A Ⓜ | B Ⓜ

NOTES:
1. DIMENSION L TO CENTER OF LEAD WHEN FORMED PARALLEL.
2. PACKAGE CONTOUR OPTIONAL (ROUND OR SQUARE CORNERS).
3. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.

DIM	MILLIMETERS		INCHES	
	MIN	MAX	MIN	MAX
A	9.40	10.16	0.370	0.400
B	6.10	6.60	0.240	0.260
C	3.94	4.45	0.155	0.175
D	0.38	0.51	0.015	0.020
F	1.02	1.78	0.040	0.070
G	2.54 BSC		0.100 BSC	
H	0.76	1.27	0.030	0.050
J	0.20	0.30	0.008	0.012
K	2.92	3.43	0.115	0.135
L	7.62 BSC		0.300 BSC	
M	—	10°	—	10°
N	0.76	1.01	0.030	0.040