# Freescale Semiconductor

Technical Data

# **RF Power Field Effect Transistors**

# N-Channel Enhancement-Mode Lateral MOSFETs

Designed for broadband commercial and industrial applications with frequencies to 175 MHz. The high gain and broadband performance of these devices make them ideal for large-signal, common source amplifier applications in 12.5 volt mobile FM equipment.

• Specified Performance @ 175 MHz, 12.5 Volts Output Power — 50 Watts Power Gain — 12 dB Efficiency — 50%

• Capable of Handling 20:1 VSWR, @ 15.6 Vdc, 175 MHz, 2 dB Overdrive

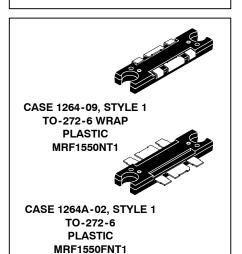
#### **Features**

- · Excellent Thermal Stability
- Characterized with Series Equivalent Large-Signal Impedance Parameters
- Broadband Full Power Across the Band: 135-175 MHz
- Broadband Demonstration Amplifier Information Available **Upon Request**
- 200°C Capable Plastic Package
- N Suffix Indicates Lead-Free Terminations. RoHS Compliant.
- In Tape and Reel. T1 Suffix = 500 Units per 44 mm, 13 inch Reel.

# Document Number: MRF1550N Rev. 11, 9/2006

# MRF1550NT1 **MRF1550FNT1**

175 MHz, 50 W, 12.5 V **LATERAL N-CHANNEL BROADBAND RF POWER MOSFETs** 



### **Table 1. Maximum Ratings**

Rating	Symbol	Value	Unit
Drain-Source Voltage	V <sub>DSS</sub>	-0.5, +40	Vdc
Gate-Source Voltage	$V_{GS}$	±20	Vdc
Drain Current — Continuous	I <sub>D</sub>	12	Adc
Total Device Dissipation @ T <sub>C</sub> = 25°C (1) Derate above 25°C	P <sub>D</sub>	165 0.50	W W/°C
Storage Temperature Range	T <sub>stg</sub>	- 65 to +150	°C
Operating Junction Temperature	TJ	200	°C

### **Table 2. Thermal Characteristics**

Characteristic	Symbol	Value <sup>(2)</sup>	Unit
Thermal Resistance, Junction to Case	$R_{ heta JC}$	0.75	°C/W

## **Table 3. Moisture Sensitivity Level**

Test Methodology	Rating	Package Peak Temperature	Unit
Per JESD 22-A113, IPC/JEDEC J-STD-020	1	260	°C

1. Calculated based on the formula P<sub>D</sub> =

2. MTTF calculator available at http://www.freescale.com/rf. Select Tools/Software/Application Software/Calculators to access the MTTF calculators by product.

NOTE - CAUTION - MOS devices are susceptible to damage from electrostatic charge. Reasonable precautions in handling and packaging MOS devices should be observed.



Table 4. Electrical Characteristics ( $T_C = 25^{\circ}C$  unless otherwise noted)

Characteristic	Symbol	Min	Тур	Max	Unit
Off Characteristics	•				
Zero Gate Voltage Drain Current $(V_{DS} = 60 \text{ Vdc}, V_{GS} = 0 \text{ Vdc})$	I <sub>DSS</sub>	_	_	1	μAdc
Gate-Source Leakage Current (V <sub>GS</sub> = 10 Vdc, V <sub>DS</sub> = 0 Vdc)	I <sub>GSS</sub>	=	_	0.5	μAdc
On Characteristics		•	•		
Gate Threshold Voltage ( $V_{DS}$ = 12.5 Vdc, $I_D$ = 800 $\mu$ A)	V <sub>GS(th)</sub>	1	_	3	Vdc
Drain-Source On-Voltage $(V_{GS} = 5 \text{ Vdc}, I_D = 1.2 \text{ A})$	R <sub>DS(on)</sub>	_	_	0.5	Ω
Drain-Source On-Voltage (V <sub>GS</sub> = 10 Vdc, I <sub>D</sub> = 4.0 Adc)	V <sub>DS(on)</sub>	=	_	1	Vdc
Dynamic Characteristics	•				
Input Capacitance (Includes Input Matching Capacitance) (V <sub>DS</sub> = 12.5 Vdc, V <sub>GS</sub> = 0 V, f = 1 MHz)	C <sub>iss</sub>	_	_	500	pF
Output Capacitance (V <sub>DS</sub> = 12.5 Vdc, V <sub>GS</sub> = 0 V, f = 1 MHz)	C <sub>oss</sub>	_	_	250	pF
Reverse Transfer Capacitance (V <sub>DS</sub> = 12.5 Vdc, V <sub>GS</sub> = 0 V, f = 1 MHz)	C <sub>rss</sub>	_	_	35	pF
RF Characteristics (In Freescale Test Fixture)		•			
Common-Source Amplifier Power Gain (V <sub>DD</sub> = 12.5 Vdc, P <sub>out</sub> = 50 Watts, I <sub>DQ</sub> = 500 mA) f = 175 MHz	G <sub>ps</sub>	_	14.5	_	dB
Drain Efficiency $(V_{DD} = 12.5 \text{ Vdc}, P_{out} = 50 \text{ Watts}, I_{DQ} = 500 \text{ mA})$ f = 175 MHz	η	_	55	_	%

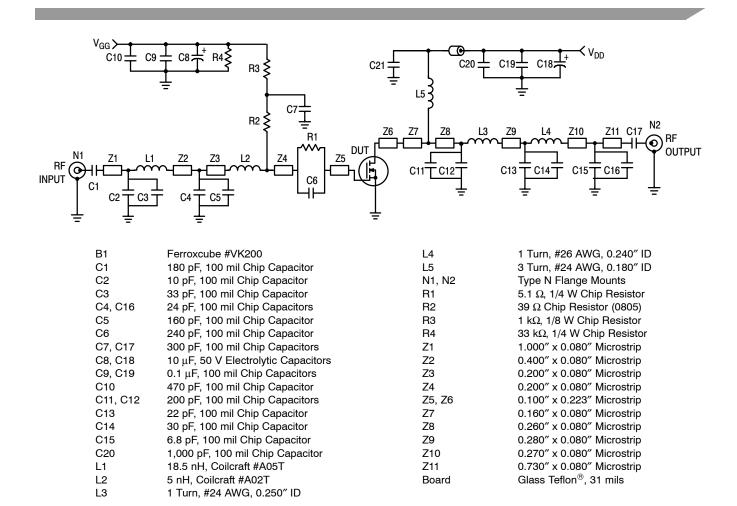


Figure 1. 135 - 175 MHz Broadband Test Circuit

#### TYPICAL CHARACTERISTICS

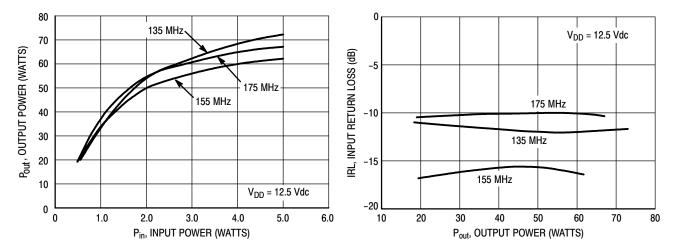


Figure 2. Output Power versus Input Power

Figure 3. Input Return Loss versus Output Power

#### TYPICAL CHARACTERISTICS

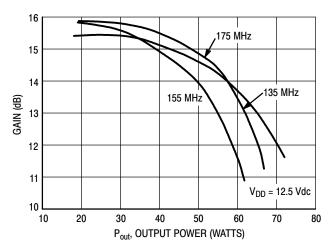


Figure 4. Gain versus Output Power

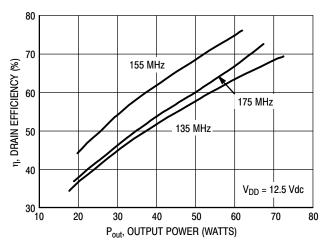


Figure 5. Drain Efficiency versus Output Power

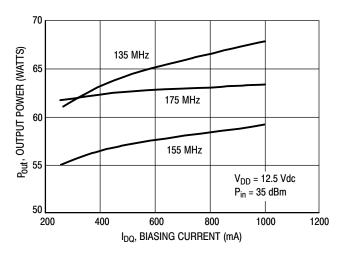


Figure 6. Output Power versus Biasing Current

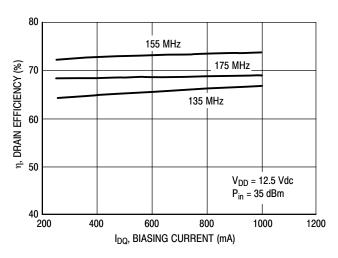


Figure 7. Drain Efficiency versus
Biasing Current

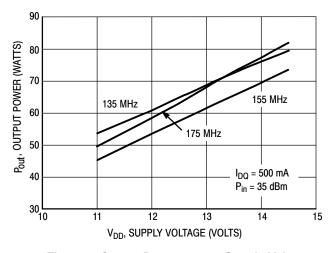


Figure 8. Output Power versus Supply Voltage

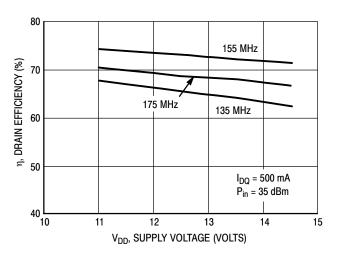
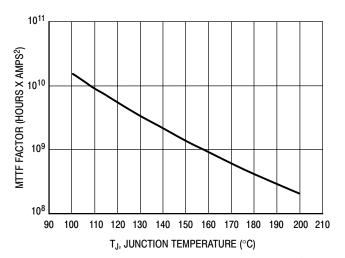


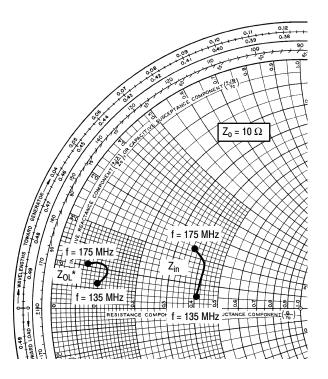
Figure 9. Drain Efficiency versus Supply Voltage

# **TYPICAL CHARACTERISTICS**



This above graph displays calculated MTTF in hours x ampere<sup>2</sup> drain current. Life tests at elevated temperatures have correlated to better than  $\pm 10\%$  of the theoretical prediction for metal failure. Divide MTTF factor by  $I_D{}^2$  for MTTF in a particular application.

Figure 10. MTTF Factor versus Junction Temperature



 $V_{DD}$  = 12.5 V,  $I_{DQ}$  = 500 mA,  $P_{out}$  = 50 W

f MHz	<b>Z</b> <sub>in</sub> Ω	<b>Z<sub>OL</sub>*</b> Ω
135	4.1 + j0.5	1.0 + j0.6
155	4.2 + j1.7	1.2 + j.09
175	3.7 + j2.3	0.7 + j1.1

Z<sub>in</sub> = Complex conjugate of source impedance.

$$\begin{split} Z_{OL}{}^{\star} &= & \text{Complex conjugate of the load} \\ &\text{impedance at given output power,} \\ &\text{voltage, frequency, and } \eta_D > 50 \text{ \%}. \end{split}$$

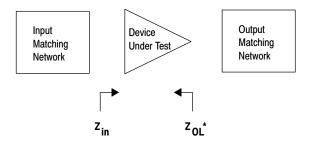


Figure 11. Series Equivalent Input and Output Impedance

Table 5. Common Source Scattering Parameters ( $V_{DD}$  = 12.5 Vdc)

# $I_{DQ} = 500 \text{ mA}$

f	S <sub>11</sub> S <sub>21</sub>		S <sub>12</sub>		S	22		
MHz	S <sub>11</sub>	∠ φ	S <sub>21</sub>	∠ φ	S <sub>12</sub>	∠ ф	S <sub>22</sub>	∠ ¢
50	0.93	-178	4.817	80	0.009	-39	0.86	-176
100	0.94	-178	2.212	69	0.009	-3	0.88	-175
150	0.95	-178	1.349	61	0.008	-8	0.90	-174
200	0.95	-178	0.892	54	0.006	-13	0.92	-174
250	0.96	-178	0.648	51	0.005	-7	0.93	-174
300	0.97	-178	0.481	47	0.004	-8	0.95	-174
350	0.97	-178	0.370	46	0.005	4	0.95	-174
400	0.98	-178	0.304	43	0.001	15	0.97	-174
450	0.98	-178	0.245	43	0.005	81	0.97	-174
500	0.98	-178	0.209	43	0.003	84	0.97	-174
550	0.99	-177	0.178	41	0.007	70	0.98	-175
600	0.98	-178	0.149	41	0.010	106	0.96	-175

# I<sub>DQ</sub> = 2.0 mA

f	S <sub>11</sub>		S	21	S	S <sub>12</sub>		22
MHz	S <sub>11</sub>	∠ ф	S <sub>21</sub>	∠ φ	S <sub>12</sub>	∠ ф	S <sub>22</sub>	∠ ф
50	0.93	-177	4.81	80	0.003	-119	0.93	-178
100	0.94	-178	2.20	69	0.006	4	0.93	-178
150	0.95	-178	1.35	61	0.003	-1	0.93	-177
200	0.95	-178	0.89	54	0.004	18	0.93	-176
250	0.96	-178	0.65	51	0.001	28	0.94	-176
300	0.97	-178	0.48	47	0.004	77	0.94	-175
350	0.97	-178	0.37	46	0.006	85	0.95	-175
400	0.98	-178	0.30	43	0.007	53	0.96	-174
450	0.98	-178	0.25	43	0.006	74	0.97	-174
500	0.98	-177	0.21	44	0.006	84	0.97	-174
550	0.99	-177	0.18	41	0.002	106	0.97	-175
600	0.98	-178	0.15	41	0.004	116	0.96	-174

# $I_{DQ} = 4.0 \text{ mA}$

f	S <sub>11</sub>		S	S <sub>21</sub>		S <sub>12</sub>		S <sub>22</sub>	
MHz	S <sub>11</sub>	∠ ф	S <sub>21</sub>	∠ ф	S <sub>12</sub>	∠ ф	S <sub>22</sub>	∠ ф	
50	0.97	-179	5.04	87	0.002	-116	0.94	-179	
100	0.96	-179	2.43	82	0.006	42	0.94	-178	
150	0.96	-179	1.60	77	0.004	13	0.94	-177	
200	0.96	-179	1.14	74	0.003	43	0.95	-176	
250	0.97	-179	0.89	71	0.004	65	0.95	-175	
300	0.97	-179	0.71	68	0.006	68	0.95	-175	
350	0.97	-179	0.57	67	0.006	74	0.97	-174	

Table 5. Common Source Scattering Parameters ( $V_{DD} = 12.5 \text{ Vdc}$ ) (continued)

# I<sub>DQ</sub> = 4.0 mA (continued)

4	S <sub>11</sub>		S <sub>21</sub>		S <sub>12</sub>		S <sub>22</sub>	
MHz	S <sub>11</sub>	∠ ф	S <sub>21</sub>	∠ ф	S <sub>12</sub>	∠ ф	S <sub>22</sub>	∠ φ
400	0.97	-179	0.49	63	0.005	58	0.97	-173
450	0.98	-178	0.41	63	0.005	73	0.98	-173
500	0.98	-178	0.36	62	0.003	128	0.98	-173
550	0.98	-178	0.32	58	0.004	57	0.99	-174
600	0.98	-178	0.27	58	0.009	83	0.98	-174

#### **APPLICATIONS INFORMATION**

#### **DESIGN CONSIDERATIONS**

This device is a common-source, RF power, N-Channel enhancement mode, Lateral Metal-Oxide Semiconductor Eield-Effect Transistor (MOSFET). Freescale Application Note AN211A, "FETs in Theory and Practice", is suggested reading for those not familiar with the construction and characteristics of FETs.

This surface mount packaged device was designed primarily for VHF and UHF mobile power amplifier applications. Manufacturability is improved by utilizing the tape and reel capability for fully automated pick and placement of parts. However, care should be taken in the design process to insure proper heat sinking of the device.

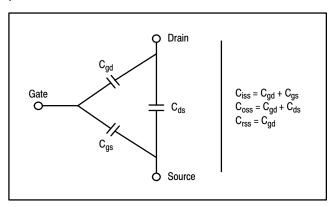
The major advantages of Lateral RF power MOSFETs include high gain, simple bias systems, relative immunity from thermal runaway, and the ability to withstand severely mismatched loads without suffering damage.

#### **MOSFET CAPACITANCES**

The physical structure of a MOSFET results in capacitors between all three terminals. The metal oxide gate structure determines the capacitors from gate-to-drain  $(C_{\rm gd})$ , and gate-to-source  $(C_{\rm gs})$ . The PN junction formed during fabrication of the RF MOSFET results in a junction capacitance from drain-to-source  $(C_{\rm ds})$ . These capacitances are characterized as input  $(C_{\rm iss})$ , output  $(C_{\rm oss})$  and reverse transfer  $(C_{\rm rss})$  capacitances on data sheets. The relationships between the inter-terminal capacitances and those given on data sheets are shown below. The  $C_{\rm iss}$  can be specified in two ways:

- 1. Drain shorted to source and positive voltage at the gate.
- Positive voltage of the drain in respect to source and zero volts at the gate.

In the latter case, the numbers are lower. However, neither method represents the actual operating conditions in RF applications.



#### **DRAIN CHARACTERISTICS**

One critical figure of merit for a FET is its static resistance in the full-on condition. This on-resistance,  $R_{DS(on)}$ , occurs in the linear region of the output characteristic and is specified at a specific gate-source voltage and drain current. The

drain-source voltage under these conditions is termed  $V_{DS(on)}$ . For MOSFETs,  $V_{DS(on)}$  has a positive temperature coefficient at high temperatures because it contributes to the power dissipation within the device.

 $\rm BV_{DSS}$  values for this device are higher than normally required for typical applications. Measurement of  $\rm BV_{DSS}$  is not recommended and may result in possible damage to the device.

## **GATE CHARACTERISTICS**

The gate of the RF MOSFET is a polysilicon material, and is electrically isolated from the source by a layer of oxide. The DC input resistance is very high - on the order of  $10^9~\Omega$  — resulting in a leakage current of a few nanoamperes.

Gate control is achieved by applying a positive voltage to the gate greater than the gate-to-source threshold voltage,  $V_{\rm GS(th)}$ .

**Gate Voltage Rating** — Never exceed the gate voltage rating. Exceeding the rated V<sub>GS</sub> can result in permanent damage to the oxide layer in the gate region.

**Gate Termination** — The gates of these devices are essentially capacitors. Circuits that leave the gate open-circuited or floating should be avoided. These conditions can result in turn-on of the devices due to voltage build-up on the input capacitor due to leakage currents or pickup.

Gate Protection — These devices do not have an internal monolithic zener diode from gate-to-source. If gate protection is required, an external zener diode is recommended. Using a resistor to keep the gate-to-source impedance low also helps dampen transients and serves another important function. Voltage transients on the drain can be coupled to the gate through the parasitic gate-drain capacitance. If the gate-to-source impedance and the rate of voltage change on the drain are both high, then the signal coupled to the gate may be large enough to exceed the gate-threshold voltage and turn the device on.

#### DC BIAS

Since this device is an enhancement mode FET, drain current flows only when the gate is at a higher potential than the source. RF power FETs operate optimally with a quiescent drain current ( $I_{DQ}$ ), whose value is application dependent. This device was characterized at  $I_{DQ} = 150$  mA, which is the suggested value of bias current for typical applications. For special applications such as linear amplification,  $I_{DQ}$  may have to be selected to optimize the critical parameters.

The gate is a dc open circuit and draws no current. Therefore, the gate bias circuit may generally be just a simple resistive divider network. Some special applications may require a more elaborate bias system.

#### **GAIN CONTROL**

Power output of this device may be controlled to some degree with a low power dc control signal applied to the gate, thus facilitating applications such as manual gain control, ALC/AGC and modulation systems. This characteristic is very dependent on frequency and load line.

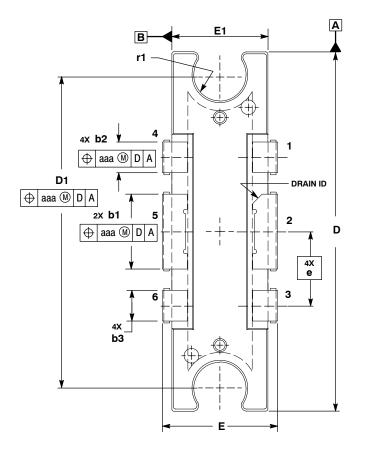
#### **AMPLIFIER DESIGN**

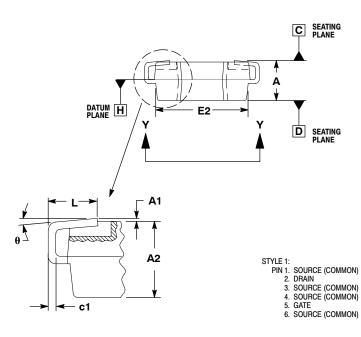
Impedance matching networks similar to those used with bipolar transistors are suitable for this device. For examples see Freescale Application Note AN721, "Impedance Matching Networks Applied to RF Power Transistors." Large-signal impedances are provided, and will yield a good first pass approximation.

Since RF power MOSFETs are triode devices, they are not unilateral. This coupled with the very high gain of this device yields a device capable of self oscillation. Stability may be achieved by techniques such as drain loading, input shunt resistive loading, or output to input feedback. The RF test fixture implements a parallel resistor and capacitor in series with the gate, and has a load line selected for a higher efficiency, lower gain, and more stable operating region.

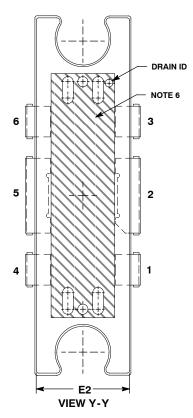
Two-port stability analysis with this device's S-parameters provides a useful tool for selection of loading or feedback circuitry to assure stable operation. See Freescale Application Note AN215A, "RF Small-Signal Design Using Two-Port Parameters" for a discussion of two port network theory and stability.

#### **PACKAGE DIMENSIONS**





**CASE 1264-09 ISSUE K TO-272-6 WRAP PLASTIC** MRF1550NT1



#### NOTES:

- NOTES:

  1. CONTROLLING DIMENSION: INCH.
  2. INTERPRET DIMENSIONS AND TOLERANCES
  PER ASME Y14.5M, 1994.
  3. DATUM PLANE -H- IS LOCATED AT TOP OF LEAD
  AND IS COINCIDENT WITH THE LEAD WHERE
  THE LEAD EXITS THE PLASTIC BODY AT THE
  TOP OF THE PARTING LINE.

  DIMENSION DAND E JOD NOT INCLINIO
- TOP OF THE PARTING LINE.

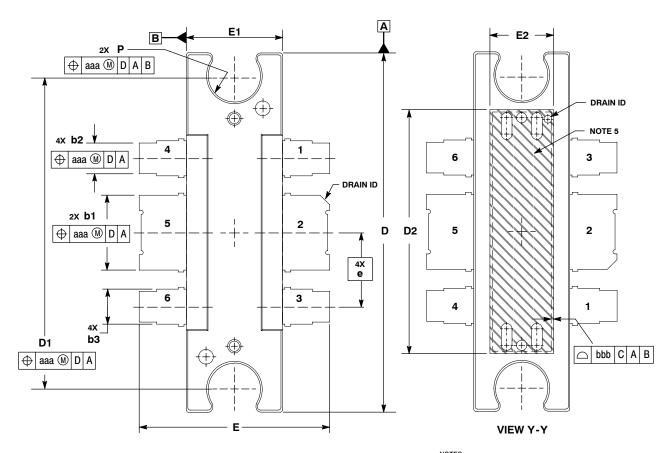
  A DIMENSION D AND E I DO NOT INCLUDE MOLD
  PROTRUSION. ALLOWABLE PROTRUSION IS

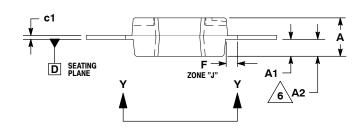
  0.006 PER SIDE. DIMENSION D AND E I DO
  INCLUDE MOLD MISMATCH AND ARE
  DETERMINED AT DATUM PLANE -H-.

  5. DIMENSIONS DI AND B3 DO NOT INCLUDE
  DAMBAR PROTRUSION. ALLOWABLE DAMBAR
  PROTRUSION SHALL BE 0.005 TOTAL IN EXCESS
  OF THE AND BE DIMENSIONS AT MAYMAIN.
- OF THE b1 AND b2 DIMENSIONS AT MAXIMUM MATERIAL CONDITION.

  6. CROSSHATCHING REPRESENTS THE EXPOSED
- AREA OF THE HEAT SLUG.

	INC	HES	MILLIN	IETERS
DIM	MIN	MAX	MIN	MAX
Α	0.098	0.108	2.49	2.74
A1	0.000	0.004	0.00	0.10
A2	0.100	0.104	2.54	2.64
D	0.928	0.932	23.57	23.67
D1	0.806	0.814	20.47	20.68
Е	0.296	0.304	7.52	7.72
E1	0.248	0.252	6.30	6.40
E2	0.241	0.245	6.12	6.22
L	0.060	0.070	1.52	1.78
b1	0.193	0.199	4.90	5.05
b2	0.078	0.084	1.98	2.13
b3	0.088	0.094	2.24	2.39
c1	0.007	0.011	0.18	0.28
е	0.193 BSC		4.90	BSC
r1	0.063	0.068	1.60	1.73
θ	0°	6°	0 °	6 °
aaa	0.004		0.	10





- STYLE 1: PIN 1. SOURCE (COMMON)
  - 2. DRAIN
    3. SOURCE (COMMON)
  - 4. SOURCE (COMMON) 5. GATE

  - 6. SOURCE (COMMON)

**CASE 1264A-02 ISSUE C** TO-272-6 **PLASTIC** MRF1550FNT1

- NOTES:
  1. CONTROLLING DIMENSION: INCH.
  2. INTERPRET DIMENSIONS AND TOLERANCES
  PER ASME Y14.5M, 1994.
  3. DIMENSIONS D AND E1 DO NOT INCLUDE MOLD
  PROTRUSION. ALLOWABLE PROTRUSION IS
  0.006 PER SIDE. DIMENSIONS D AND E1 DO
  INCLUDE MOLD MISMATCH AND ARE
  DETERMINED AT DATUM PLANE -H-.
  4. DIMENSIONS b1 AND RS DO NOT INCLUDE
- DETERMINED AT DATUM PLANE -HF.

  A DIMENSIONS b1 AND b3 DO NOT INCLUDE

  DAMBAR PROTRUSION. ALLOWABLE DAMBAR
  PROTRUSION SHALL BE 0.005 TOTAL IN EXCESS
  OF THE b1 AND b2 DIMENSIONS AT MAXIMUM
  MATERIAL CONDITION.

  5. CROSSHATCHING REPRESENTS THE EXPOSED
- AREA OF THE HEAT SLUG.

  6. DIMENSION A2 APPLIES WITHIN ZONE J ONLY.

INC	HES	MILLIN	ETERS	
MIN	MAX	MIN	MAX	
0.098	0.106	2.49	2.69	
0.038	0.044	0.96	1.12	
0.040	0.042	1.02	1.07	
0.926	0.934	23.52	23.72	
0.810	BSC	20.57	BSC	
0.608 BSC		15.44	BSC	
0.492	0.500	12.50	12.70	
0.246	0.254	6.25	6.45	
0.170	BSC	4.32 BSC		
0.025	BSC	0.64 BSC		
0.126	0.134	3.20	3.40	
0.193	0.199	4.90	5.05	
0.078	0.084	1.98	2.13	
0.088	0.094	2.24	2.39	
0.007	0.011	0.178	0.279	
0.193	BSC	4.90	BSC	
0.0	004	0.	10	
0.008 0.20			20	
	MIN 0.098 0.038 0.040 0.926 0.810 0.608 0.492 0.246 0.1726 0.126 0.193 0.078 0.088 0.007 0.193 0.008	0.098 0.106 0.038 0.044 0.040 0.042 0.926 0.934 0.810 BSC 0.608 BSC 0.492 0.500 0.246 0.254 0.170 BSC 0.126 0.134 0.193 0.199 0.078 0.084 0.088 0.094 0.008 0.094 0.009 0.004	MIN         MAX         MIN           0.098         0.106         2.49           0.038         0.044         0.96           0.040         0.042         1.02           0.926         0.934         23.52           0.810 BSC         20.57           0.608 BSC         15.44           0.492         0.500         12.50           0.246         0.254         6.25           0.170 BSC         4.32           0.025 BSC         0.64           0.126         0.134         3.20           0.193         0.199         4.90           0.078         0.084         1.98           0.088         0.094         2.24           0.093         0.938 BSC         4.90           0.094         0.094         0.094           0.098         0.094         0.094           0.098         0.094         0.094           0.094         0.094         0.094           0.094         0.094         0.094           0.095         0.094         0.094           0.096         0.094         0.094           0.097         0.011         0.191	

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