

# **QUAD N-CHANNEL** MOSFET POWER MODULE 301

M.S.KENNEDY CORP.

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(315) 701-6751

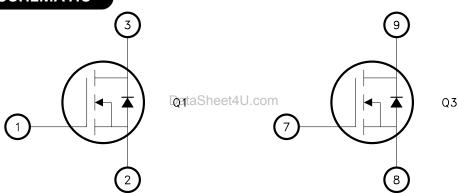
#### **FEATURES:**

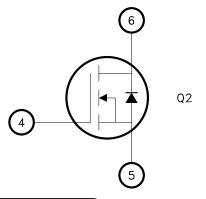
- Pin Compatible with MPM3013
- QUAD Independent N Channel MOSFETS
- Isolated Package for Direct Heat Sinking, Excellent Thermal Conductivity
- Avalanche Rated Devices
- 55 Volt, 25 Amp Rated
- Low RDS (ON) 0.022Ω For Each Die

#### **DESCRIPTION:**

The MSK 3013 is a QUAD N-Channel power circuit packaged in a space efficient isolated ceramic tab power SIP package. The MSK 3013 consists of four totally isolated N-Channel MOSFETs. The MSK 3013 uses M.S.Kennedy's proven power hybrid technology to bring a cost effective high performance circuit for use in today's sophisticated servo motor and disk drive systems. The MSK 3013 is a replacement for the MPM3013 with only minor differences in specifications.

#### **EQUIVALENT SCHEMATIC**





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#### TYPICAL APPLICATIONS

- Stepper Motor Servo Control
- Disk Drive Head Control
- X-Y Table Control
- DataSheetAz-EPAntenna Control
  - Various Switching Applications

#### **PIN-OUT INFORMATION**

Q1 Gate 7 Q3 Gate Q1 Source 8 Q3 Source 3 Q1 Drain 9 Q3 Drain

Q4 Gate Q2 Gate 10 5 Q2 Source Q4 Drain

Q2 Drain 12 Q4 Source

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## **ABSOLUTE MAXIMUM RATINGS**

VDSS	Drain to Source Voltage55V MAX	TJ	Junction Temperature + 175°C MAX
VDGDR	Drain to Gate Voltage	Tst	Storage Temperature
	$(RGS = 1M\Omega)$	Tc	Case Operating Temperature Range55°C TO 125°C
Vgs	Gate to Source Voltage	TLD	Lead Temperature Range
	(Continuous) ± 20V MAX		(10 Seconds)
ID	Continuous Current 25A MAX		
IDM	Pulsed Current 49A MAX		
RTH-JC	Thermal Resistance		
	(Junction to Case)0.3°C/W		

#### **ELECTRICAL SPECIFICATIONS**

Dayamata:	Tank Cam distant		MSK 3013			7
Parameter	Test Conditions (4)	Min.	Тур.	Max.	Units	
Drain-Source Breakdown Voltage	VGS = 0 ID = 0.25 mA	55	-	-	V	
Drain-Source Leakage Current	VDS = 55V VGS = 0V	-	-	25	μA	
Gate-Source Leakage Current	VGS = ±20V VDS = 0	-	-	±100	nA	
Gate-Source Threshold Voltage	VDS = VGS	2	-	4	V	
Drain-Source on Resistance 2	VGS = 10V ID = 25A	-	0.033	0.040	Ω	
Drain-Source on Resistance 3	VGS = 10V ID = 25A	-	-	0.022	Ω	
Forward Transconductance 1	VDS = 25V ID = 25A	17	-	-	S	DataShe
Total Gate Charge 1	Data-25A Data-25Aet4U.com	-	-	65	nC	
Gate-Source Charge 1	VDS = 28V	-	-	12	nC	
Gate-Drain Charge 1	VGS = 10V	_	-	27	nC	
Turn-On Delay Time 1	VDD = 28V	_	7.3	-	nS	
Rise Time 1	ID = 25A	_	69	-	nS	
Turn-Off Delay Time 1	$RG = 12^{\Omega}$	-	47	-	nS	
Fall Time 1	RD = 1.1		60	-	nS	
nput Capacitance 1 VGS = 0V		-	1300	-	pF	
Output Capacitance 1 VDS = 25V		-	410	-	pF	
Reverse Transfer Capacitance 1	f = 1 MHz	-	150	-	pF	
BODY DIODE						
Forward on Voltage 1	IS = 25A VGS = 0V	-	1.3	1.75	V	
Reverse Recovery Time 1	IS = 25A di/dt = 100A/μS	-	65	98	nS	
Reverse Recovery Charge 1	IS = 25A di/dt = 100A/μS	-	160	240	μC	
LOTES.						

**NOTES:** 

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device performance but are for reference only.

DataShe 2 Resistance as seen at package pins.

3 Resistance for die only; use for thermal calculations.

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<sup>(1)</sup> This parameter is guaranteed by design but need not be tested. Typical parameters are representative of

<sup>(4)</sup> TA = 25 °C unless otherwise specified. Parameters apply to each transistor in the module.

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#### **APPLICATION NOTES**

#### N-CHANNEL GATES

For driving the N-Channel gates, it is important to keep in mind that it is essentially like driving a capacitance to a sufficient voltage to get the channel fully on. Driving the gates to +15 volts with respect to their sources assures that the transistors are on. This will keep the dissipation down to a minimum level [RDS(ON) specified in the data sheet]. How quickly the gate gets turned ON and OFF will determine the dissipation of the transistor while it is transitioning from OFF to ON, and vice-versa. Turning the gate ON and OFF too slow will cause excessive dissipation, while turning it ON and OFF too fast will cause excessive switching noise in the system. It is important to have as low a driving impedance as practical for the size of the transistor. Many motor drive IC's have sufficient gate drive capability for the MSK 3013. If not, paralleled CMOS standard gates will usually be sufficient. A series resistor in the gate circuit slows it down, but also suppresses any ringing caused by stray inductances in the MOSFET circuit. The selection of the resistor is determined by how fast the MOSFET wants to be switched. See Figure 1 for circuit details.

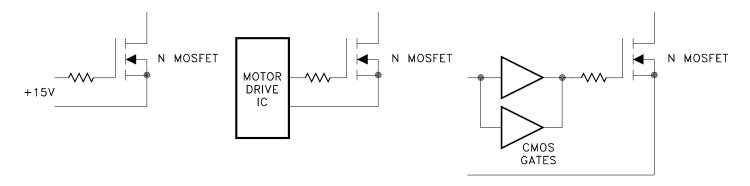


Figure 1

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#### BRIDGE DRIVE CONSIDERATIONS

It is important that the logic used to turn ON and OFF the various transistors allow sufficient "dead time" between a high side transistor and its low side transistor to make sure that at no time are they both ON. When they are, this is called "shoot-through", and it places a momentary short across the power supply. This overly stresses the transistors and causes excessive noise as well. See Figure 3.

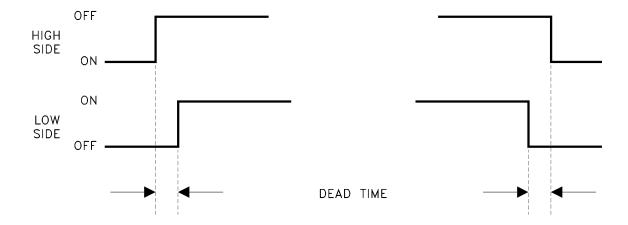


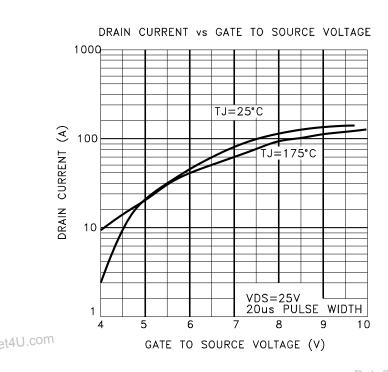
Figure 2

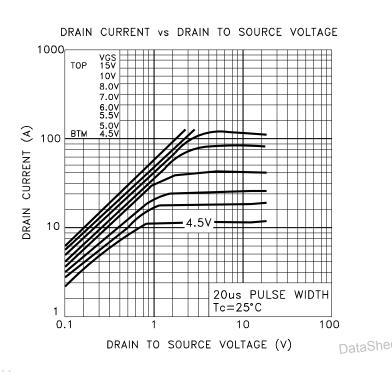
DataSheetThis deadtime should allow for the turn on and turn off time of the transistors, especially when slowing them who with gate U.com resistors. This situation will be present when switching motor direction, or when sophisticated timing schemes are used for servo systems such as locked antiphase PWM'ing for high bandwidth operation.

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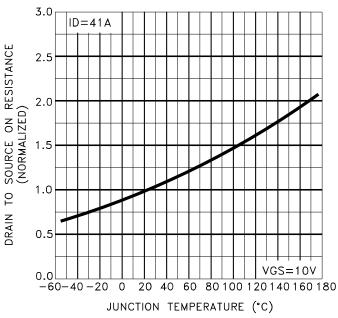
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### **TYPICAL PERFORMANCE CURVES**





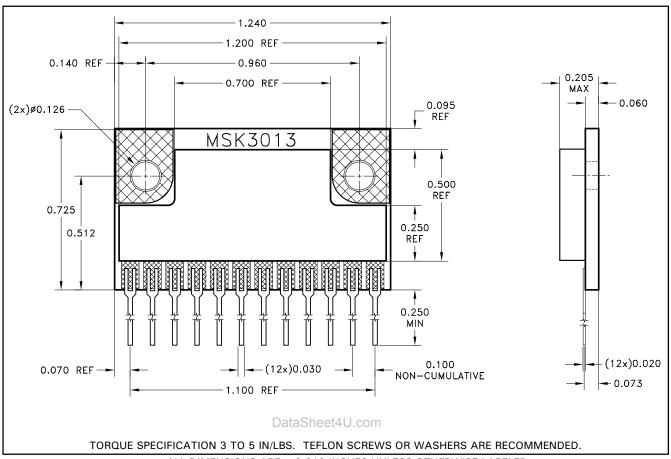
DRAIN TO SOURCE ON RESISTANCE VS JUNCTION TEMP.



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#### **MECHANICAL SPECIFICATIONS**



ALL DIMENSIONS ARE  $\pm 0.010$  INCHES UNLESS OTHERWISE LABELED.

# ORDERING INFORMATION

PART NUMBER	SCREENING LEVEL		
MSK 3013	Industrial		

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