

NEGATIVE OUTPUT WIDE BANDWIDTH HIGH VOLTAGE AMPLIFIER

604

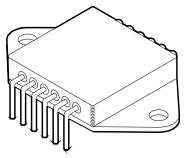
4707 Dey Road Liverpool, N.Y. 13088

(315) 701-6751

FEATURES:

- · Negative Going Output Voltage
- Ultra Low Quiescent Current ± 10mA for High Voltage
- · 95V Peak to Peak Output Voltage Swing
- Slew Rate 3000V/µS Typical
- · Full Power Output Frequency 2 MHz Typical
- · Output Current 250mA Peak
- · Adjustable VHV Power Supply Minimizes Power Dissipation
- Compact Package Offers Superior Power Dissipation.

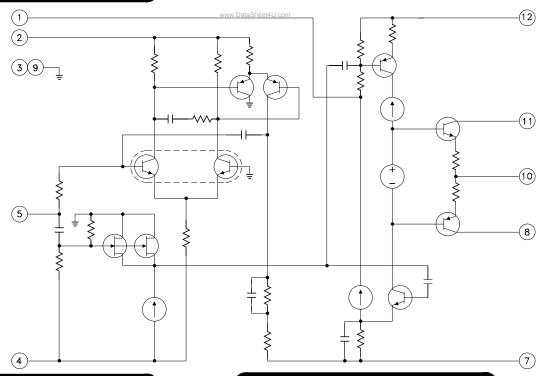
MIL-PRF-38534 QUALIFIED



DESCRIPTION:

The MSK 604(B) is a high voltage wideband amplifier designed to provide large negative voltage swings at high slew rates in wideband systems. The true inverting op-amp topology employed in the MSK 604 provides excellent D.C. specifications such as input offset voltage and input bias current. These attributes are important in amplifiers that will be used in high gain configurations since the input error voltages will be multiplied by the system gain. The MSK 604 achieves impressive slew rate specifications by employing a feed forward A.C. path through the amplifier; however, the device is internally configured in inverting mode to utilize this benefit. Internal compensation for gains of -5V/V or greater keeps the MSK 604 stable in this range. The MSK 604 is packaged in a space efficient, hermetically sealed, 12 pin power dual in line package that has a high thermal conductivity for efficient device cooling.

EQUIVALENT SCHEMATIC



TYPICAL APPLICATIONS

- · Wideband High Voltage Amplifier
- · High Resolution CRT Monitor Grid Drive
- · Ultra High Performance Video Processing
- · CRT Beam Intensity Control
- · Varactor Tuned VCO Driver
- · Automatic Test Equipment

PIN-OUT INFORMATION

- 1 COMP
- 2 + VCC
- 3 GROUND
- 4 -VCC

1

- 5 -INPUT
- 6 NO CONNECTION
- 12 + VHV
- 12 + V \(\text{V}\) \(\text{V}\)
- 10 OUTPUT
- 9 CASE/GROUND
- 8 -VSC
- 7 -VHV

Rev. B 8/00

ABSOLUTE MAXIMUM RATINGS

± V н∨	Supply Voltage + 20,-120VDC	Тѕт	Storage Temperature Range -65°C to +150°C
\pmV_{IN}	Input Voltage Range ± Vcc	T_LD	Lead Temperature Range 300°C
$\pm V$ cc	Supply Voltage (Input Stage) ± 18VDC		(10 Seconds)
hetaJC	Thermal Resistance	Tc	Case Operating Temperature
	(Output Devices)		MSK60440°C to +85°C
			MSK604B55°C to +125°C
		ТJ	Junction Temperature

ELECTRICAL SPECIFICATIONS

Parameter	Took Conditions	Group A	MSK 604B 4		MSK 604 ③				
Farameter	Test Conditions	Subgroup	Min.	Тур.	Max.	Min.	Тур.	Max.	Units
STATIC									
	VIN=0 @ +Vcc	1,2,3	-	1.5	2.0	-	1.8	2.5	mA
	VIN = 0 @ -Vcc	1,2,3	-	15	25	-	20	30	mA
Quiescent Current	VIN = 0 @ + VHV	1	-	10	15	-	10	20	mA
		2,3	-	14	17	-	-	-	mA
	VIN = 0 @ -VHV	1	-	10	15	-	10	20	mA
	VIIV — O @ -V HV	2,3	-	14	17	-	-	-	mA
Input Offset Voltage	VIN = 0	1	-	±1.0	±5.0	-	±1.0	±10	mV
input Offset Voltage	VIIV — O	2,3	-	±2.0	±10.0	-	±2.0	-	mV
Input Bias Current	Input Bias Current			50	250	-	50	500	nA
		2,3	-	100	350	-	100	-	nA
Input Offset Voltage Drift ②	VIN = 0	2,3	-	±10	±50	-	± 10	-	μV/°C
	±Vcc	-	±12	±15	± 18	±12	± 15	±18	V
Power Supply Range	+VHV	-	0	+ 15	+ 20	0	+ 15	+20	>
	-VHV	-	-50	-100	-120	-50	-100	-120	٧
DYNAMIC CHARACTERISTICS									
Output Voltage Swing ②	f = 1 KHz	4	-90/+5	-95/8	-	-90/+5	-95/8	-	٧
Peak Output Current ②	f = 1 KHz	-	±100	± 250	-	±100	±250	-	mA
Full Power Output ②	Vo = 70V	-	1	2	-	1	2	-	MHz
Unity Gain Bandwidth	Vo = 1.0V	-	50	100	-	50	100	-	MHz
Slew Rate ②	Vo=80V	4	2000	3000	-	2000	3000	-	V/μS
Voltage Gain ②	f = 1 KHz	-	94	100	-	90	100	-	dB
Settling Time to 1% ②	Av = -10V/V $Vo = 50V$	-	-	200	-	-	200	-	nS
Settling Time to 0.1%	Av = -10V/V Vo = 50V	-	-	1000	-	-	1000	-	nS

NOTES:

2

Rev. B 8/00

① Unless otherwise specified, $\pm V_{CC} = \pm 15V_{DC}$, $+V_{HV} = +15V_{DC}$, $-V_{HV} = -100V$ CL = 8pF (probe capacitance) and Av = -10V/V.
② This parameter is guaranteed by design but not tested. Typical parameters are representative of actual device performance but are for reference only.
③ Industrial grade devices shall be tested to subgroups 1 and 4 unless otherwise specified.
④ Military grade devices ('B' suffix) shall be 100% tested to subgroups 1,2,3 and 4.

⁽⁵⁾ Subgroup 1,4 TA = Tc = +25°C Subgroup 2,5 TA = Tc = +125°C Subgroup 3,6 $T_A = T_C = -55$ °C

APPLICATION NOTES

FEED FORWARD TOPOLOGY

The MSK 604 employs a circuit topology known as "feed forward". This inverting configuration allows the user to realize the excellent D.C. input characteristics of a differential amplifier without losing system bandwidth. The incoming signal is split at the input into it's A.C. and D.C. components. The D.C. component is allowed to run through the differential amplifier where any common mode noise is rejected. The A.C. component is "fed forward" to the output section through a very high speed linear amplifier where it is mixed back together with the D.C. component. The result is an amplifier with most of the benefits of a differential amplifier without the loss in system bandwidth.

INTERNAL COMPENSATION

Since the MSK 604 is a high voltage amplifier, it is commonly used in circuits employing large gains. Therefore, the internal compensation was chosen for gains of -5V/V or greater. In circuits running at gains of less than -5V/V, the user can further compensate the device by adding compensation networks at the input or feedback node. Pin 1 (comp) should be bypassed with a 0.1uF ceramic capacitor to +VHV for all applications.

HIGH VOLTAGE SUPPLIES

The negative high voltage supply on the MSK 604 can be adjusted to reduce power dissipation. The output of the MSK 604 will typically swing to within 5V of the -VHV power supply rail. Therefore, if the system in question only needs the output of the amplifier to swing -40V peak, the power supply rails could be set to -50V safely. For best performance, the minimum value of -VHV should be -50VDC. The high voltage and low voltage power supplies should be decoupled as shown in Figure 1. The +VHV supply can range from ground to +20VDC depending on application requirements.

TRANSITION TIMES

Transition time optimization of the MSK 604 follows the same basic rules as most any other amplifier. Best transition times will be realized with minumum load capacitance, minimum external feedback resistance and lowest circuit gain. Transition times will degrade if the output is driven too close to either supply rail. Feedback and input resistor values will affect transition time as well. See Figure 1 and Table 1 for recommended component values.

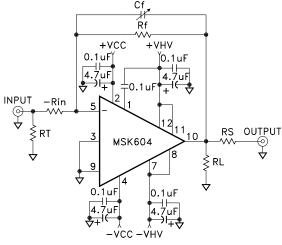


Figure 1

VOLTAGE GAIN	-Rin	RF	Сғ		
-10V/V	1ΚΩ	10ΚΩ	0.5-5pF		
-20V/V	499Ω	10ΚΩ	N/A		
-50V/V	402Ω	20ΚΩ	N/A		

Table 1

CURRENT LIMIT

Figure 2 is the recommended active short circuit protection scheme for the MSK 604. The following formula may be used for setting current limit:

RBASE must be selected based on the value of $\pm\,\text{VHV}$ as follows:

RBASE =
$$((+VHV - (-VHV)) - 1.2V) / 4mA$$

This formula guarantees that Q2 and Q4 will always have sufficient base current to be in operation. This circuit can be made tolerant of high frequency output current spikes with the addition of Csc. The corresponding time constant would be:

$$T = (Rsc) (Csc)$$

A common value for Csc is approximately 1000pF. If current limit is unnecessary, short pin 7 to pin 8 and pin 11 to pin 12 as shown in Figure 1.

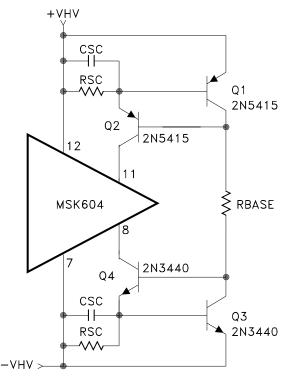
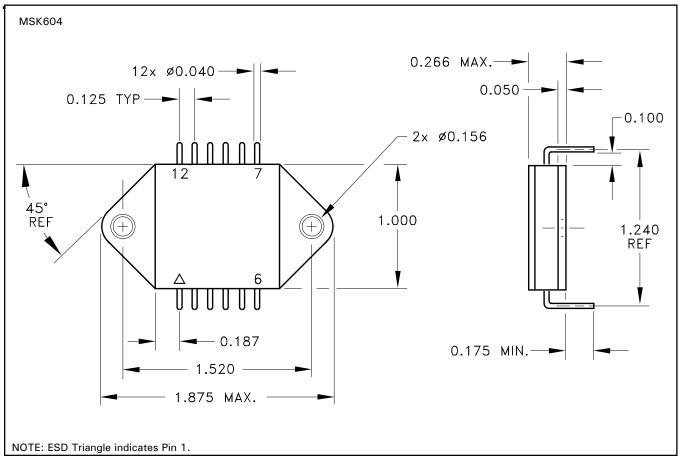


Figure 2



ALL DIMENSIONS ARE ± 0.010 INCHES UNLESS OTHERWISE LABELED

ORDERING INFORMATION

Part Number	Screening Level		
MSK604	Industrial		
MSK604B	Military-Mil-PRF-38534		

M.S. Kennedy Corp.
4707 Dey Road, Liverpool, New York 13088
Phone (315) 701-6751
FAX (315) 701-6752
www.mskennedy.com

The information contained herein is believed to be accurate at the time of printing. MSK reserves the right to make changes to its products or specifications without notice, however, and assumes no liability for the use of its products.

4 Rev. B 8/00