

GENERAL DESCRIPTION

OB2530T is a high performance offline PSR power switch for low power AC/DC charger and adapter applications. It operates in primary-side sensing and regulation. Consequently, opto-coupler and TL431 could be eliminated. Proprietary Constant Voltage (CV) and Constant Current (CC) control is integrated as shown in the figure below.

In CC control, the current and output power setting can be adjusted externally by the sense resistor Rs at CS pin. In CV control, multi-mode operations are utilized to achieve high performance and high efficiency. In addition, good load regulation is achieved by the built-in cable drop compensation. Device operates in PFM in CC mode at large load condition and it operates in PWM with frequency reduction at light/medium load. The chip consumes very low operation current. It achieves less than 75mW standby power to meet strict standby power standard.

OB2530T offers comprehensive protection coverage with auto-recovery feature including Cycle-by-Cycle current limiting, VDD over voltage protection, feedback loop open protection, short circuit protection, built-in leading edge blanking, VDD under voltage lockout (UVLO), OTP etc. OB2530T is offered in DIP8/SOP8 packages.

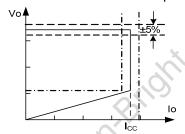


Figure.1. Typical CC/CV Curve

FEATURES

- Primary-side sensing and regulation without TL431 and opto-coupler
- High precision constant voltage and current regulation at universal AC input
- Multi-mode PWM/PFM operation for efficiency improving
- Integrated Rdson 2Ω MOSFET
- Good dynamic response
- Programmable CV and CC regulation
- Built-in primary winding inductance compensation
- Programmable cable drop compensation
- No need for control loop compensation
- Audio noise free operation
- Built-in leading edge blanking (LEB)
- Ultra low start-up current and low operating current
- Comprehensive protection coverage with auto-recovery
 - VDD over voltage protection
 - VDD under voltage lockout with hysteresis (UVLO)
 - Cycle-by-Cycle current limiting
 - Feedback loop open protection
 - Output short circuit protection
 - On-chip OTP

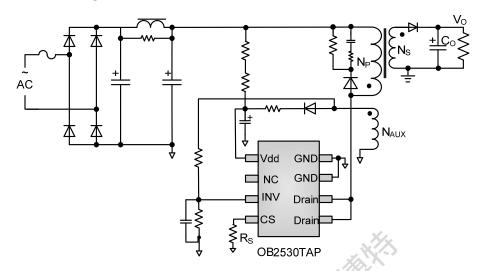
APPLICATIONS

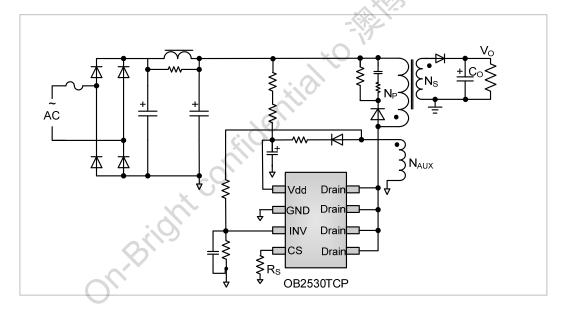
Low Power AC/DC offline SMPS for

- Cell Phone Charger
- Digital Cameras Charger
- Small Power Adapter
- Auxiliary Power for PC, TV etc.
- Linear Regulator/RCC Replacement



TYPICAL APPLICATION



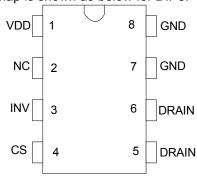




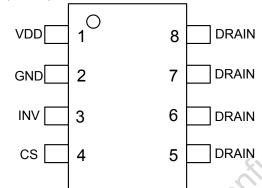
GENERAL INFORMATION

Pin Configuration

The pin map is shown as below for DIP8.



The pin map is shown as below for SOP8.



Ordering Information

Part Number	Description
OB2530TAP	DIP8, Pb-free, Tube
OB2530TCP-H	SOP8, Pb-free, Tube
OB2530TCPA-H	SOP8, Pb-free, T&R

Recommended Operating Condition

- tto common a parating committee				
Symbol	Parameter	Range		
VDD	VDD Supply Voltage	8 to 25V		

Package Dissipation Rating

Package	RθJA (℃/W)
DIP8	75
SOP8	90

Absolute Maximum Ratings

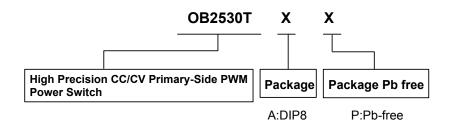
Absolute Maximum Natings				
Parameter	Value			
Drain Voltage (off state)	-0.3V to Bvdss			
VDD Voltage	-0.3 to 28V			
INV Input Voltage	-0.3 to 7V			
CS Input Voltage	-0.3 to 7V			
Min/Max Operating Junction	-40 to 150 ℃			
Temperature T _J	-40 to 150 C			
Operating Ambient	-20 to 85 ℃			
Temperature T _A	-20 to 65 C			
Min/Max Storage	-55 to 150 ℃			
Temperature T _{stg}	-55 10 150 0			
Lead Temperature	260 ℃			
(Soldering, 10secs)				
N. 4 Or 1 1 1 1 1 1 1 1	. "			

Note: Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute maximum-rated conditions for extended periods may affect device reliability.

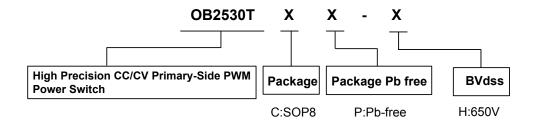
Output Power Table

Part Number		90Vac~264Vac
	OB2530TAP	15W
	OB2530TCP-H	15W

Note: Maximum continuous power in a typical non-ventilated enclosed adapter with adequate PCB heat sinking, measured at $+45^{\circ}$ C







Marking Information

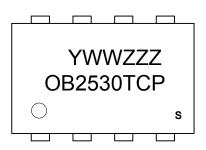


Y:Year Code WW:Week Code(01-52) ZZZ:Lot Code A:DIP8 Package P:Pb-free Package S:Internal Code(Optional)

TERMINAL ASSIGNMENTS for OB2530TAP

Pin Num.	Pin Name	I/O	Description	
1	VDD	P	Power Supply	
2	NC			
3	INV	I	The voltage feedback from auxiliary winding. Connected to resistor divider from auxiliary winding reflecting output voltage.	
4	cs	I	Power MOSFET source	
5/6	Drain	0	Drain of internal power MOSFET	
7/8	GND	Р	Ground	





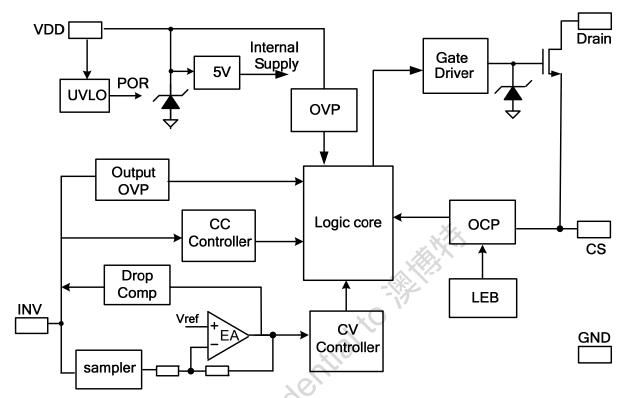
Y:Year Code WW:Week Code(01-52) ZZZ:Lot Code C:SOP8 Package P:Pb-free Package S:Internal Code(Optional)

TERMINAL ASSIGNMENTS for OB2530TCP-H

Pin Num.	Pin Name	I/O	Description
1	VDD	Р	Power Supply
2	GND	Р	Ground
3	INV	ı	The voltage feedback from auxiliary winding. Connected to resistor divider from auxiliary winding reflecting output voltage.
4	CS	I	Power MOSFET source
5/6/7/8	Drain	0	Drain of internal power MOSFET



BLOCK DIAGRAM





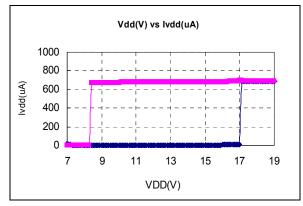
ELECTRICAL CHARACTERISTICS

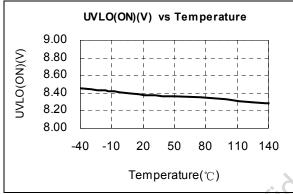
(TA = 25°C, VDD=15V, if not otherwise noted)

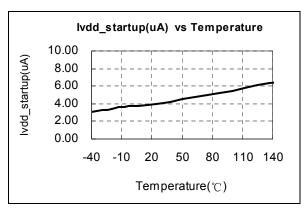
Symbol	Parameter	Test Conditions	Min	Тур.	Max	Unit	
Supply Voltage (VDD) Section							
I start-up	Start up current	VDD=UVLO_OFF-1V		5		uA	
I static	Static current			0.65	1.0	mA	
UVLO(OFF)	VDD under voltage lockout exit		15.5	16.5	17.5	V	
UVLO(ON)	VDD under voltage lockout enter		7.3	8.3	9.3	V	
VDD_OVP	VDD over voltage protection		26.0	27.0	28.5	V	
Max. Operating Voltage					25	٧	
Current Sense	Input Section	. /	,				
TLEB	LEB time		Š	300		ns	
Vth_ocp_min	Minimum over current threshold		485	500	515	mV	
Vth_ocp_max	Maximum over current threshold			590		mV	
Td_oc	OCP Propagation delay			100		ns	
INV Input Section	on	×O '					
Vref_INV	Reference voltage for feedback threshold		2.475	2.500	2.525	V	
Tpause_min	Minimum Toff			2.0		us	
F_min	Minimum frequency	<i>ن</i>	400	445	490	Hz	
F_max	Maximum frequency		72			KHz	
lcomp_cable	Maximum cable compensation current		40	45	50	uA	
Output Over Vo	Itage Protection						
V_OVP	Output Over voltage threshold		2.85	3.0	3.15	V	
MOSFET Section							
BVdss	Mosfet Drain-Source Breakdown	OB2530TAP	620			V	
D V 033	Voltage	OB2530TCP	650			V	
Rdson	Static drain to source on	OB2530TAP		2		Ω	
1 (3001)	resistance	OB2530TCP		2.1		Ω	

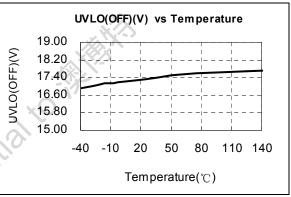


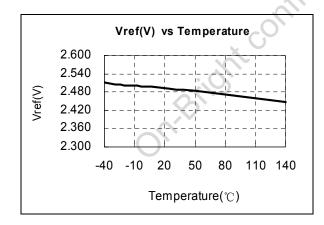
CHARACTERIZATION PLOTS

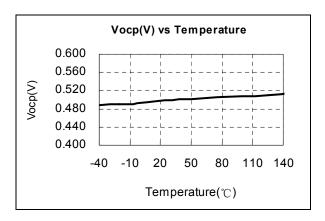


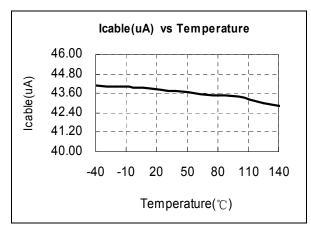


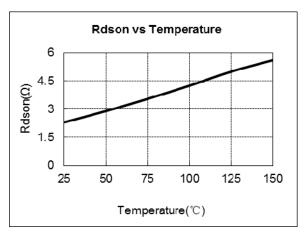














OPERATION DESCRIPTION

OB2530T is a cost effective PSR power switch optimized for off-line low power AC/DC applications including battery chargers. It operates in primary side sensing and regulation, thus opto-coupler and TL431 are not required. Proprietary built-in CV and CC control can achieve high precision CC/CV control meeting most charger application requirements.

Startup Current and Start up Control

Startup current of OB2530T is designed to be very low so that VDD could be charged up above UVLO threshold level and device starts up quickly. A large value startup resistor can therefore be used to minimize the power loss yet achieve a reliable startup in application.

Operating Current

The operating current of OB2530T is as low as 650uA (typical). Good efficiency and les than 75mW standby power is achieved with the low operating current.

CC/CV Operation

OB2530T is designed to produce good CC/CV control characteristic as shown in the Figure. 1. In charger applications, a discharged battery charging starts in the CC portion of the curve until it is nearly full charged and smoothly switches to operate in CV portion of the curve. The CC portion provides output current limiting. In CV operation, the output voltage is regulated through the primary side control. In CC operation mode, OB2530P will regulate the output current constant regardless of the output voltage drop.

Principle of Operation

To support OB2530T series proprietary CC/CV control, system needs to be designed in DCM mode for flyback system (Refer to Typical Application Diagram in page1).

In the DCM flyback converter, the output voltage can be sensed via the auxiliary winding. During MOSFET turn-on time, the load current is supplied from the output filter capacitor, Co, and the current in the primary winding ramps up. When MOSFET turns off, the energy stored in the primary winding is transferred to the secondary side and the current in the secondary winding is

$$I_S = \frac{N_P}{N_S} \cdot I_P \tag{1}$$

The auxiliary voltage reflects the output voltage as shown in Figure 2 and it is given by

$$V_{AUX} = \frac{N_{AUX}}{N_{s}} \cdot (V_O + \Delta V)$$
 (2)

Where ΔV indicates the voltage drop of the

output Diode.

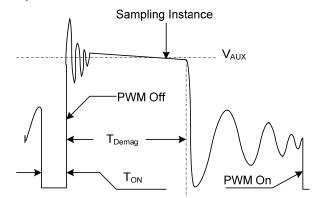


Figure.2. Auxiliary voltage waveform

Via a resistor divider connected between the auxiliary winding and INV, the auxiliary voltage is sampled at the middle of the de-magnetization and it is hold until the next sampling. The sampled voltage is compared with reference voltage Vref (typical 2.5V) and the difference is amplified. The error amplifier output reflects the load condition and controls the switching off time to regulate the output voltage, thus constant output voltage can be achieved.

When the sampled voltage is below Vref and the error amplifier output reaches its minimum, the switching frequency is controlled by the sampled voltage to regulate the output current, thus the constant output current can be achieved.

Adjustable CC Point and Output Power

In OB2530T series, the CC point and maximum output power can be externally adjusted by external current sense resistor Rs at CS pin as illustrated in the typical application diagram. The larger Rs, the smaller CC point is, and the smaller output power becomes, and vice versa as shown in Figure 3.

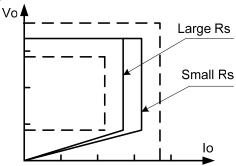


Figure.3. Adjustable output power by changing Rs

Operation Switching Frequency

The switching frequency of OB2530T is adaptively controlled according to the load conditions and the operation modes.

For flyback operating in DCM, the maximum output power is given by



$$Po_{MAX} = \frac{1}{2} L_p F_{SW} I_p^2$$
 (3)

Where Lp indicate the inductance of primary winding and Ip is the peak current of primary winding.

Refer to the equation 3, the change of the primary winding inductance results in the change of the maximum output power and the constant output current in CC mode. To compensate the change from variations of primary winding inductance, the switching frequency is locked by an internal loop and the switching frequency is

$$F_{SW} = \frac{1}{2T_{Demag}} \tag{4}$$

Since T_{Demag} is inversely proportional to the inductance, as a result, the product Lp and fsw is constant, thus the maximum output power and constant current in CC mode will not change as primary winding inductance changes. Up to $\pm 7\%$ variation of the primary winding inductance can be compensated.

On Time OCP Compensation

The variation of max output current in CC mode can be rather large if no compensation is provided. The OCP threshold value is self adjusted higher at higher AC voltage. This OCP threshold slope adjustment helps to compensate the increased output current limit at higher AC voltage. In OB2530T, a proprietary OCP compensation block is integrated and no external components are needed. The OCP threshold in OB2530T series is a function of the switching ON time. For the ON time between 1.6us to 6.4us (typical), the OCP threshold changes linearly from 500mV to 590mV. For the ON time larger than 6.4us, the OCP threshold is clamped to 590mV, as shown in Figure 4.

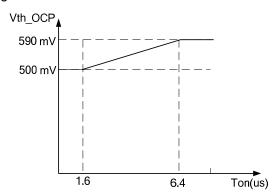


Figure.4. On time OCP compensation

Programmable Cable Drop Compensation
In OB2530T, cable drop compensation is implemented to achieve good load regulation. An offset voltage is generated at INV pin by an

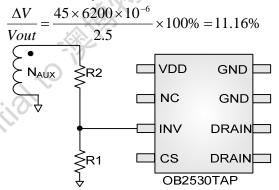
internal current flowing into the resister divider. The current is proportional to the switching off time, as a result, it is inversely proportional to the output load current, and the drop due to the cable loss can be compensated. As the load current decreases from full-load to no-load, the offset voltage at INV will increase. It can also be programmed by adjusting the resistance of the divider to compensate the drop for various cable lines used.

The percentage of maximum compensation is

$$\frac{\Delta V}{Vout} = \frac{Icomp_cable \times (R1//R2) \times 10^{-6}}{2.5} \times 100\%$$

 ΔV is load compensation voltage and Vout is output voltage;

For example: R1//R2=6.2Kohm, the percentage of maximum compensation is



Current Sensing and Leading Edge Blanking

Cycle-by-Cycle current limiting is offered in OB2530T. The switch current is detected by a sense resistor into the CS pin. An internal leading edge blanking circuit chops off the sensed voltage spike at initial power MOSFET on state so that the external RC filtering on sense input is no longer needed.

Gate Driver

The GATE pin is connected to the gate of an external power switch. An internal 11V (typical) clamp is added for MOSFET gate protection at high VCC voltage. When VCC voltage drops below UVLO(ON), the GATE pin is internally pull low to maintain the off state.

Protection Control

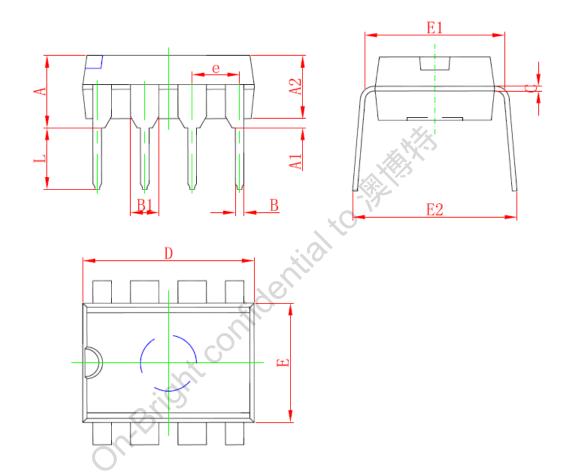
Good power supply system reliability is achieved with its rich protection features including precise on-chip OTP, cycle-by-cycle current limiting (OCP), output over voltage protection, VDD over voltage protection, short circuit protection, under voltage lockout on VDD (UVLO).

VDD is supplied by transformer auxiliary winding output. The output of OB2530T is shut down when VDD drops below UVLO (ON) and the power converter enters power on start-up sequence thereafter.



PACKAGE MECHANICAL DATA

DIP8 PACKAGE OUTLINE DIMENSIONS

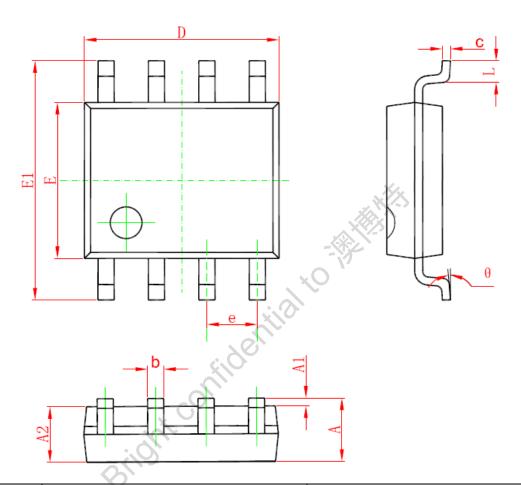


Symbol	Dimensions I	n Millimeters	Dimensions In Inches	
Symbol	Min	Max	Min	Max
Α	3.710	5.334	0.146	0.210
A1	0.381		0.015	
A2	2.921	4.953	0.115	0.195
В	0.350	0.650	0.014	0.026
B1	1.524 (BSC)		0.06 (BSC)	
С	0.200	0.360	0.008	0.014
D	9.000	10.160	0.354	0.400
E	6.096	7.112	0.240	0.280
E1	7.320	8.255	0.288	0.325
е	2.540 (BSC)		0.1 (BSC)	
L	2.921	3.810	0.115	0.150
E2	7.620	10.920	0.300	0.430



PACKAGE MECHANICAL DATA

SOP8 PACKAGE OUTLINE DIMENSIONS



Symbol	Dimensions I	n Millimeters	Dimensions In Inches		
Symbol	Min	Max	Min	Max	
Α	1.350	1.750	0.053	0.069	
A1	0.050	0.250	0.002	0.010	
A2	1.250	1.650	0.049	0.065	
b	0.310	0.510	0.012	0.020	
С	0.100	0.250	0.004	0.010	
D	4.700	5.150	0.185	0.203	
Е	3.800	4.000	0.150	0.157	
E1	5.800	6.200	0.228	0.244	
е	1.270	(BSC)	0.050	(BSC)	
L	0.400	1.270	0.016	0.050	
θ	0°	8°	0°	8°	



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