

PULSE-WIDTH-MODULATION CONTROL CIRCUITS

- Complete PWM Power Control Circuitry
- Uncommitted Outputs for 200-mA Sink or Source Current
- Output Control Selects Single-Ended or Push-Pull Operation
- Internal Circuitry Prohibits Double Pulse at Either Output
- Variable Dead Time Provides Control Over Total Range
- Internal Regulator Provides a Stable 5-V Reference Supply With 5% Tolerance
- Circuit Architecture Allows Easy Synchronization

Description

The PL494 incorporates on a single monolithic chip all the functions required in the construction of a pulse-width-modulation control circuit.

Designed primarily for power supply control, this device offers the systems engineer the flexibility to tailor the power supply control circuitry to a specific application.

The PL494 contains two error amplifiers, an on-chip adjustable oscillator, a dead-time control (DTC) comparator, a pulse-steering control flip-flop, a 5-V, 5%-precision regulator, and output-control circuits.

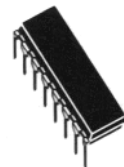
The error amplifiers exhibit a common-mode voltage range from $-0.3V$ to $V_{cc} - 2V$.

The dead-time control comparator has a fixed offset that provides approximately 5% dead time. The on-chip oscillator may be bypassed by terminating RT to the reference output and providing a sawtooth input to CT, or it may drive the common circuits in synchronous multiple-rail power supplies.

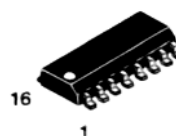
FUNCTION TABLE

INPUT TO OUTPUT CTRL	OUTPUT FUNCTION
$V_i = GND$	Single-ended or parallel output
$V_i = V_{ref}$	Normal push-pull operation

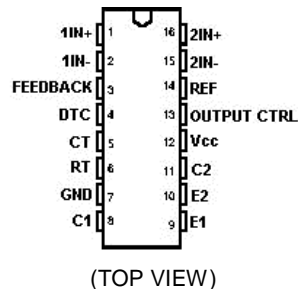
PLASTIC PACKAGE DIP-16



PLASTIC PACKAGE SO-16



PIN CONNECTIONS

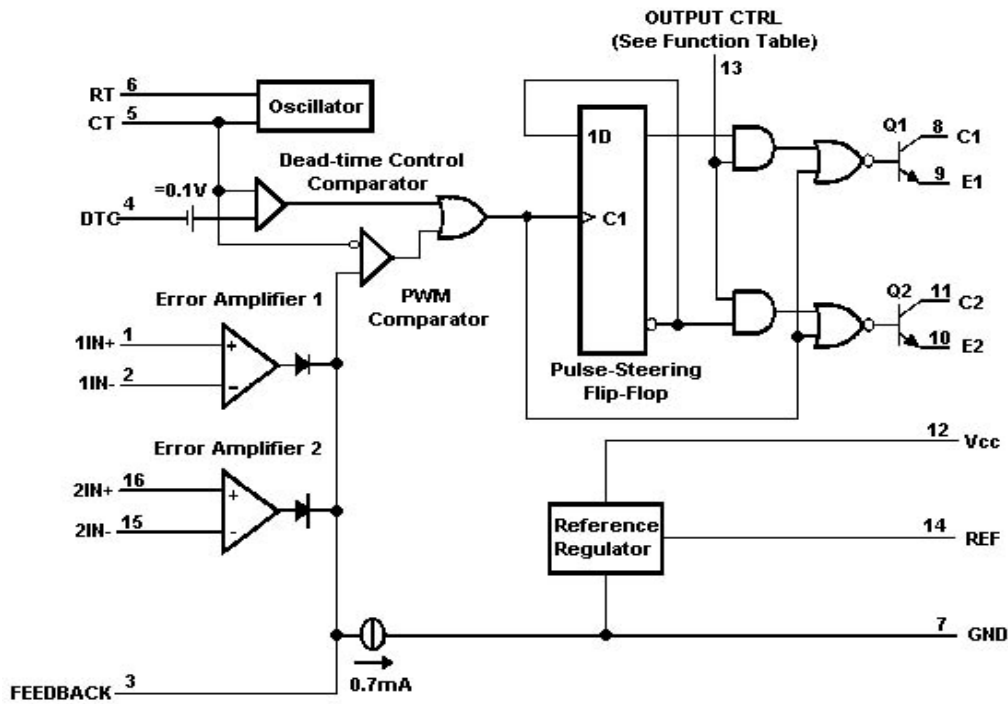


Description (continued)

The uncommitted output transistors provide either common-emitter or emitter-follower output capability. The PL494 provides for push-pull or single-ended output operation, which may be selected through the output-control function. The architecture of this device prohibits the possibility of either output being pulsed twice during push-pull operation.

The PL494CN is characterized for operation from 0°C to 70°C.

Functional block diagram



Absolute maximum ratings over operating free-air temperature range (unless otherwise noted)*

	PL494CN	UNIT
Supply voltage, V_{CC} (see Note 1)	41	V
Amplifier input voltage, V_I	- 0.3 to + 42	V
Collector output voltage, V_O	41	V
Collector output current, I_O	250	mA
Power Dissipation T_A 45	1	W
Operating free-air temperature range, T_A	0 to 70	
Storage temperature range, T_{stg}	- 55 to + 125	

*Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

Note 1: All voltage values, except differential voltages, are with respect to the network ground terminal.

Recommended operating conditions

	PL494CN		UNIT
	MIN	MAX	
Supply voltage, V_{CC}	7	40	V
Amplifier input voltage, V_I	-0.3	$V_{CC} - 2.0$	V
Collector output voltage, V_O		40	V
Collector output current, (each transistor)		200	mA
Current into feedback terminal		0.3	mA
Oscillator frequency, f_{OSC}	1	300	kHz
Timing capacitor, C_T	0.47	10000	nF
Timing resistor, R_T	1.8	500	k
Operating free-air temperature, T_A	0	70	°C

Electrical characteristics over recommended operating free-air temperature range, $V_{CC}=15V$, $f=10kHz$ (unless otherwise noted)

Reference section

PARAMETER	TEST CONDITIONS*	PL494CN			UNIT
		MIN	TVP**	MAX	
Output voltage (REF)	$I_O=1mA$	4.75	5	5.25	V
Input regulation	$V_{CC}=7V$ to 40V		2	25	mV
Output regulation	$I_O=1mA$ to 10mA		1	15	mV
Output voltage change with temperature	$T_A=MIN$ to MAX		2	10	mV/V
Short-circuit output current***	REF= 0V		25		mA

*For conditions shown as MIN or MAX, use the appropriate value specified under recommended operating conditions.

**All typical values except for parameter changes with temperature are at $T_A=25^\circ C$.

***Duration of the short circuit should not exceed one second.

Oscillator section, $C_T=0.01$ mkF, $R_T=12k$

PARAMETER	TEST CONDITIONS*	PL494CN		UNIT
		MIN	TVP**MAX	
Frequency			10	kHz
Standard deviation of frequency	All values of V_{CC} , C_T , R_T , and T_A constant		100	Hz/kHz
Frequency change with voltage	$V_{CC}=7V$ to 40V, $T_A=25^\circ C$		1	Hz/kHz
Frequency change with temperature***	$T_A=MIN$ to MAX		10	Hz/kHz

*For conditions shown as MIN or MAX, use the appropriate value specified under recommended operating conditions.

**All typical values except for parameter changes with temperature are at $T_A=25^\circ C$.

***Temperature coefficient of timing capacitor and timing resistor not taken into account.

Error amplifier section

PARAMETER	TEST CONDITIONS	PL494CN		UNIT
		MIN	TYP* MAX	
Input offset voltage	V_O (FEEDBACK) = 2.5V		2 10	mV
Input offset current	V_O (FEEDBACK) = 2.5V	250	25	nA
Input bias current	V_O (FEEDBACK) = 2.5V	1	0.2	mkA
Common-mode input voltage range	$V_{CC} = 7V$ to 40V		-0.3 to $V_{CC}-2$	V
Open-loop voltage amplification	$V_O=3V$, $R_L=2k$, $V_O=0.5V$ to 3.5V	70	95	dB
Unity-gain bandwidth	$V_O=0.5V$ to 3.5V, $R_L=2k$		800	kHz
Common-mode rejection ratio	$V_O = 40V$, $T_A=25^\circ C$	65	80	dB
Output sink current (FEEDBACK)	$V_{ID}=-15mV$ to $-5V$, $V(FEEDBACK)=0.7V$	0.3	0.7	mA
Output source current (FEEDBACK)	$V_{ID}=15mV$ to 5V, $V(FEEDBACK)=3.5V$	-2		mA

*All typical values except for temperature coefficient are at $T_A=25^\circ C$

electrical characteristics over recommended operating free-air temperature range, $V_{CC}=15V$, $f = 10kHz$ (unless otherwise noted)

Output section

PARAMETER		TEST CONDITIONS	PL494CN		UNIT
			MIN	TYP* MAX	
Corrector off-state current		$V_{CE}=40V, V_{CC}=40V$	2	100	mA
Emitter off-state current		$V_{CC}=V_C=40V, V_E = 0$	-100		mA
Collector emitter saturation voltage	Common emitter	$V_E=0, I_C=200mA$	1.1	1.3	V
	Emitter follower	$V_{O(C1orC2)}=15V, I_E=-200mA$	1.5	2.5	
Output control input current		$V_I=V_{ref}$	3.5		mA

*All typical values except for temperature coefficient are at $T_A=25^\circ C$

dead-time control section

PARAMETER		TEST CONDITIONS	PL494CN		UNIT
			MIN	TYP* MAX	
Input bias current (DEAD-TIME CTRL)		$V_I=0$ to 5.25V	-2	-10	nA
Maximum duty cycle, each output		V_I (DEAD-TIME CTRL) = 0, $C_T=0.1mkF, R_T=12k$	45%		
Input threshold voltage (DEAD-TIME CTRL)	Zero duty cycle		3	3.3	V
	Maximum duty cycle		0°		

On products compliant to MIL-STD-883, Class B, this parameter is not production tested.

*All typical values except for temperature coefficient are at $T_A=25^\circ C$.

PWM comparator section

PARAMETER		TEST CONDITIONS	PL494CN		UNIT
			MIN	TYP* MAX	
Input threshold voltage (FEEDBACK)		Zero duty cycle	4	4.5	V
Input sink current (FEEDBACK)		V (FEEDBACK) = 0.7V	0.3	0.7	mA

*All typical values except for temperature coefficient are at $T_A=25^\circ C$.

Total Device

PARAMETER		TEST CONDITIONS	PL494CN		UNIT
			MIN	TYP* MAX	
Standby supply current	$R_T = V_{ref}$ All other inputs and outputs open	$V_{CC} = 15V$	6	10	mA
		$V_{CC} = 40V$	9	15	
Average supply current		V_I (DEAD-TIME CTRL) = 2V	7.5		mA

*All typical values except for temperature coefficient are at $T_A=25^\circ C$.

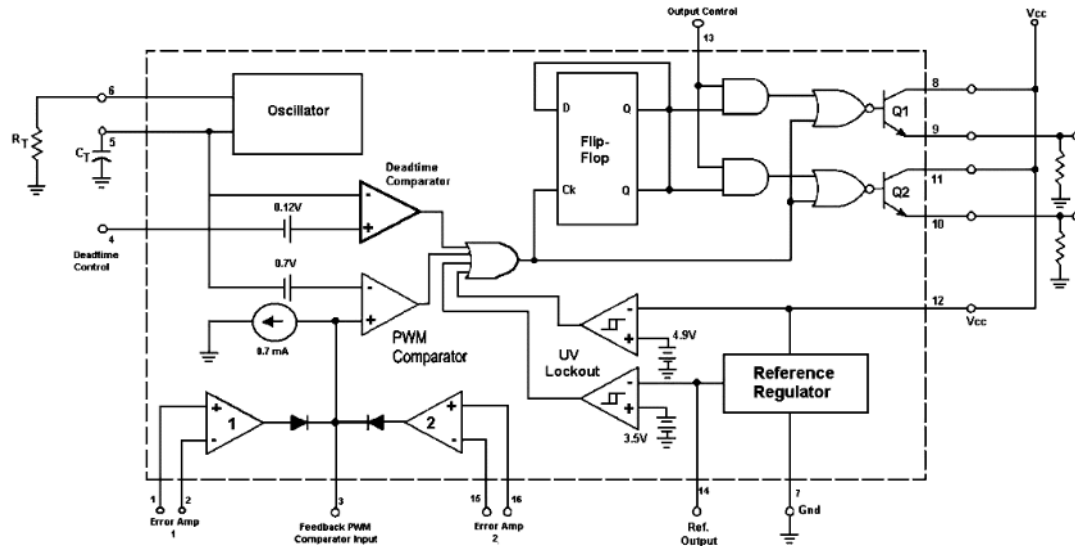
electrical characteristics over recommended operating free-air temperature range, $V_{CC}=15V$, $f=10kHz$ (unless otherwise noted) (continued)

Switching characteristics, $T_A=25^\circ C$

PARAMETER	PL494CN		UNIT
	MIN	TYP* MAX	
Rise time	100	200	ns
Fall time	25	100	ns
Rise time	100	200	ns
Fall time	40	100	ns

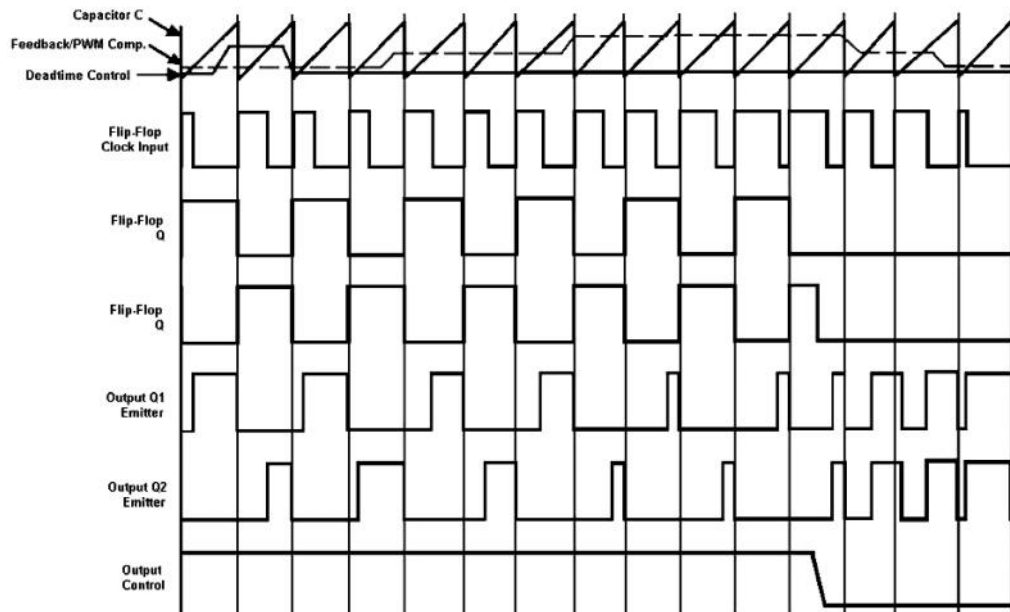
*All typical values except for temperature coefficient are at $T_A=25^\circ C$.

Figure 1. Representative Block Diagram



This device Contains 46 active transistors

Figure 2. Timing Diagram



APPLICATIONS INFORMATION

Description

The PL494CN is a fixed-frequency pulse width modulation control circuit, incorporating the primary building blocks required for the control of a switching power supply. (See Figure1.) An internal-linear sawtooth oscillator is frequency-programmable by two external components, R_T and C_T . The Approximate oscillator frequency is determined by:

$$f_{osc} \approx \frac{1.1}{R_T \cdot C_T}$$

For more information refer to Figure 3.

Output pulse width modulation is accomplished by comparison of the positive sawtooth waveform across capacitor C_T to either of two control signals. The NOR gates, which drive output transistors Q1 and Q2, are enabled only when the flip-flop clock-input line is in its low state. This happens only during that portion of time when the sawtooth voltage is greater than the control signals. Therefore, an increase in control-signal amplitude causes a corresponding linear decrease of output pulse width. (Refer to the Timing Diagram shown in Figure2.)

The Control signals are external inputs that can be fed into the deadtime control, the error amplifier inputs, or the feedback input. The deadtime control comparator has an effective 120 mV input offset which limits the minimum output deadtime to approximately the first 4% of the sawtooth-cycle time. This would result in a maximum duty cycle on a given output of 96% with the output control grounded, and 48% with it connected to the reference line. Additional deadtime may be imposed on the output by setting the deadtime-control input to a fixed voltage, ranging between 0 V to 3.3 V.

Functional Table

Input/Output Controls	Output Function	$\frac{f_{out}}{f_{osc}} =$
Grounded	Single-ended PWM @ Q1 and Q2	1.0
@ Vref	Push-pull Operation	0.5

The pulse width modulator comparator provides a means for the error amplifiers to

adjust the output pulse width from the maximum percent on-time, established by the deadtime control input, down to zero, as the voltage at the feedback pin varies from 0.5 V to 3.5 V. Both error amplifiers have a common mode input range from -0.3 V to ($V_{cc} - 2V$), and may be used to sense power-supply output voltage and current. The error-amplifier outputs are active high and are ORed together at the noninverting input of the pulse-width modulator comparator. With this configuration, the amplifier that demands minimum output on time, dominates control of the loop.

When capacitor C_T is discharged, a positive pulse is generated on the output of the deadtime comparator, which clocks the pulse-steering flip-flop and inhibits the output transistor, Q1 and Q2. With the output-control connected to the reference line, the pulse-steering flip-flop directs the modulated pulses to each of the two output transistors equal to half that of the oscillator. Output drive can also be taken from Q1 or Q2, when single-ended operation with a maximum on-time of less than 50% is required. This is desirable when the output transformer has a ringback winding with a catch diode used for snubbing. When higher operation, Q1 and Q2 may be connected in parallel, and the output-mode pin must be tied to ground to disable the flip-flop. The output frequency will now be equal to that of the oscillator.

The PL494CN has an internal 5.0 V reference capable of sourcing up to 10 mA of load current for external bias circuits. The reference has an internal accuracy of $\pm 5.0\%$ with a typical thermal drift of less than 50 mV over an operating temperature range of 0° to 70°.

Figure 3. Oscillator Frequency versus Timing Resistance

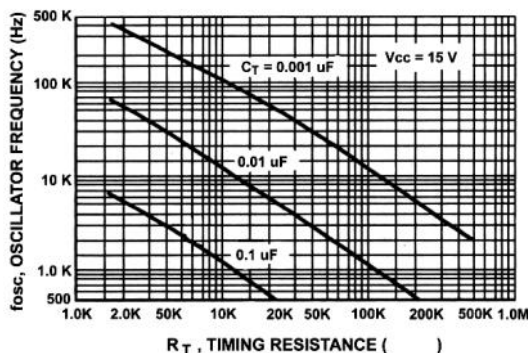


Figure 4. Open Loop Voltage Gain and Phase versus Frequency

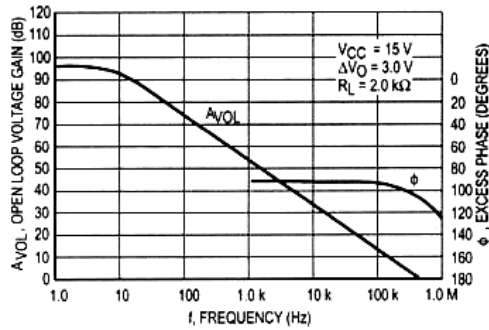


Figure 5. Percent Deadtime versus Oscillator Frequency

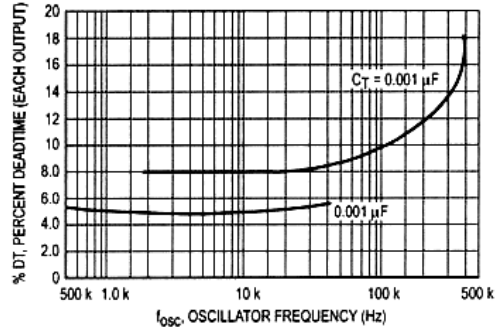


Figure 6. Percent Duty Cycle versus Deadtime Control Voltage

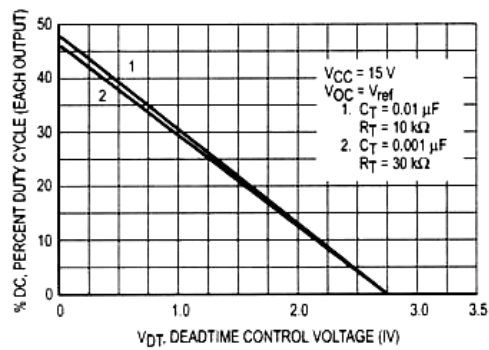


Figure 7. Emitter-Follower Configuration Output Saturation Voltage versus Emitter Current

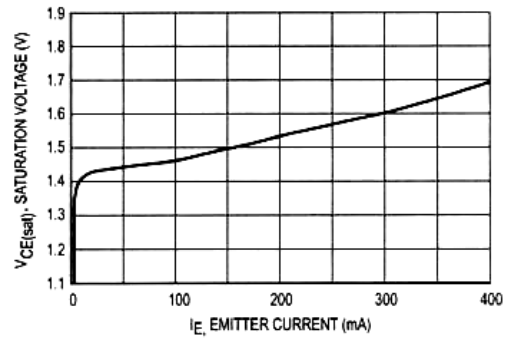


Figure 8. Common-Emitter Configuration Output Saturation Voltage versus Collector Current

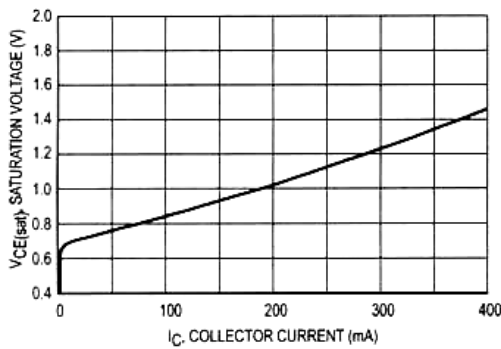


Figure 9. Standby Supply Current versus Supply Voltage

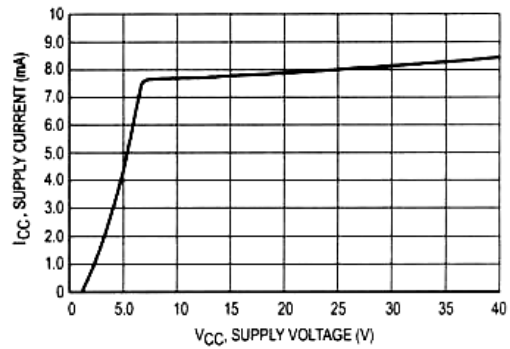


Figure 10. Error-Amplifier Characteristics

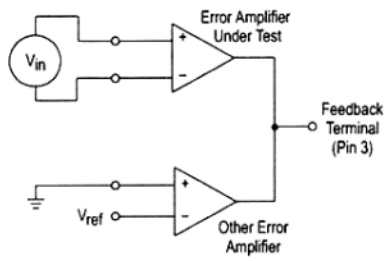


Figure 11. Deadtime and Feedback Control Circuit

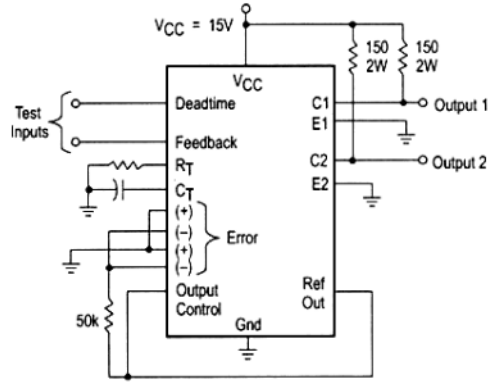


Figure 12. Common-Emitter Configuration Test Circuit and Waveform

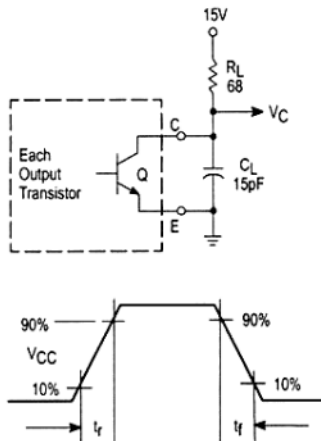


Figure 13. Emitter-Follower Configuration Test Circuit and Waveform

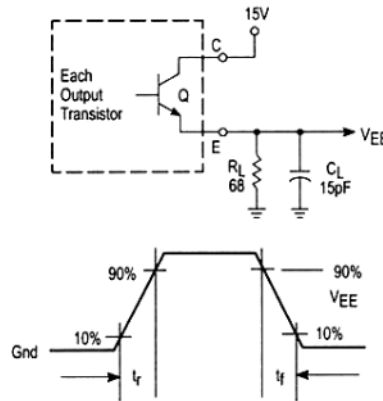


Figure 14. Error-Amplifier Sensing Techniques

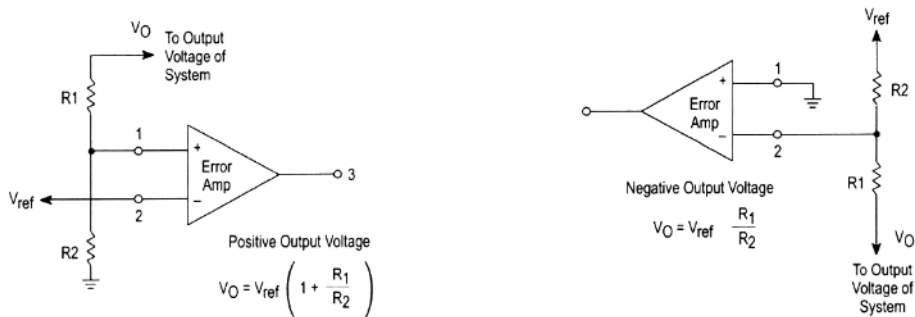


Figure 15. Deadtime Control Circuit

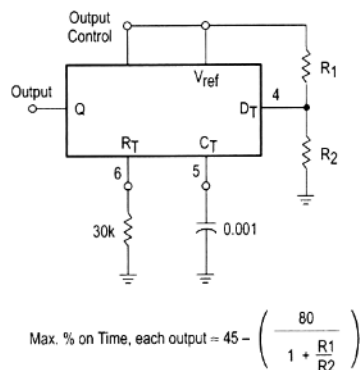


Figure 16. Soft-Start Circuit

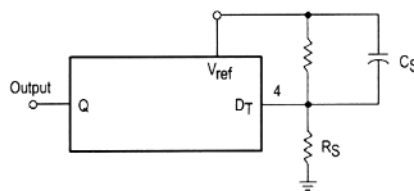


Figure 17. Output Connections for Single-Ended and Push-Pull Configurations

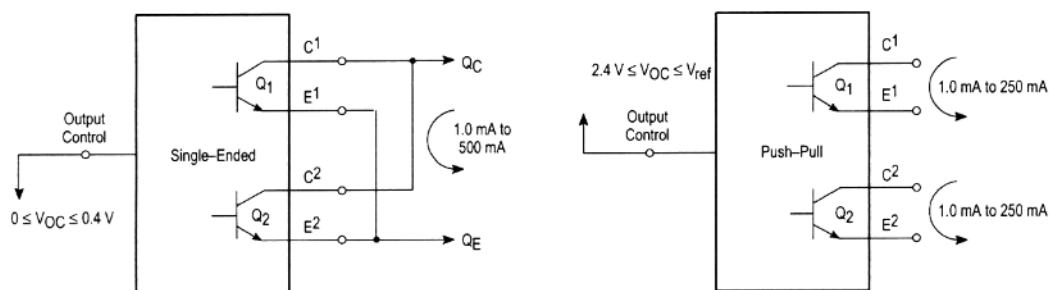


Figure 18. Slaving Two or More Control Circuits

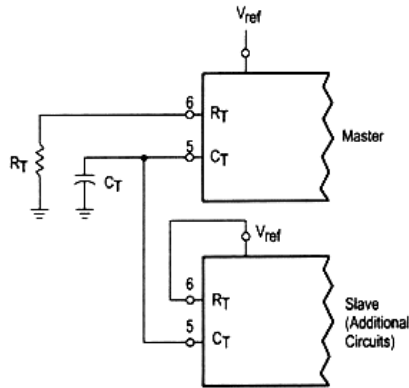


Figure 19. Operation with $V_{in} > 40V$ Using External Zener

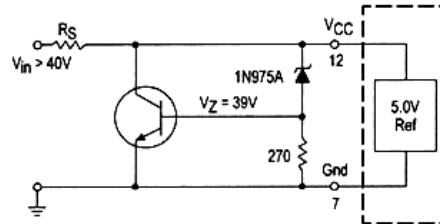
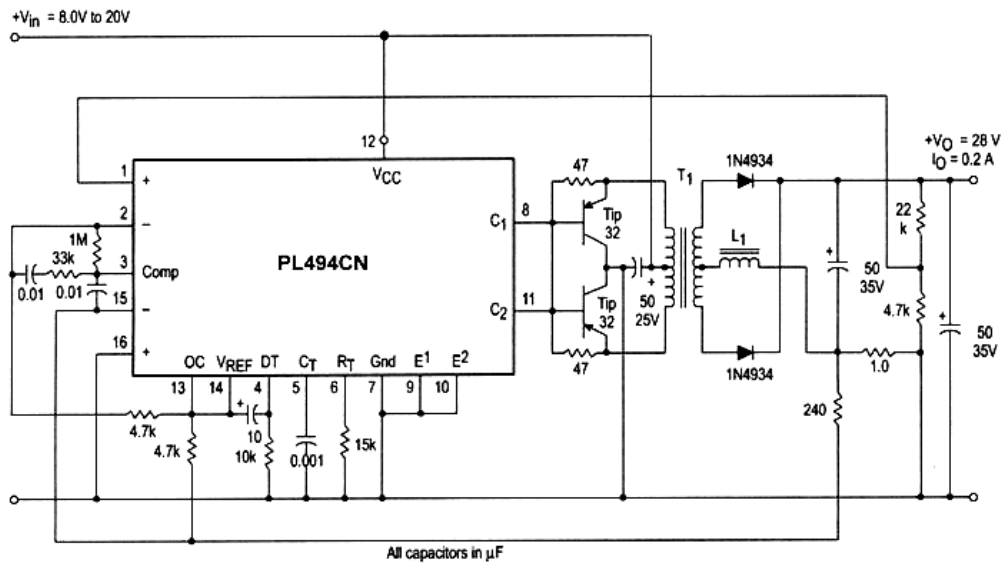


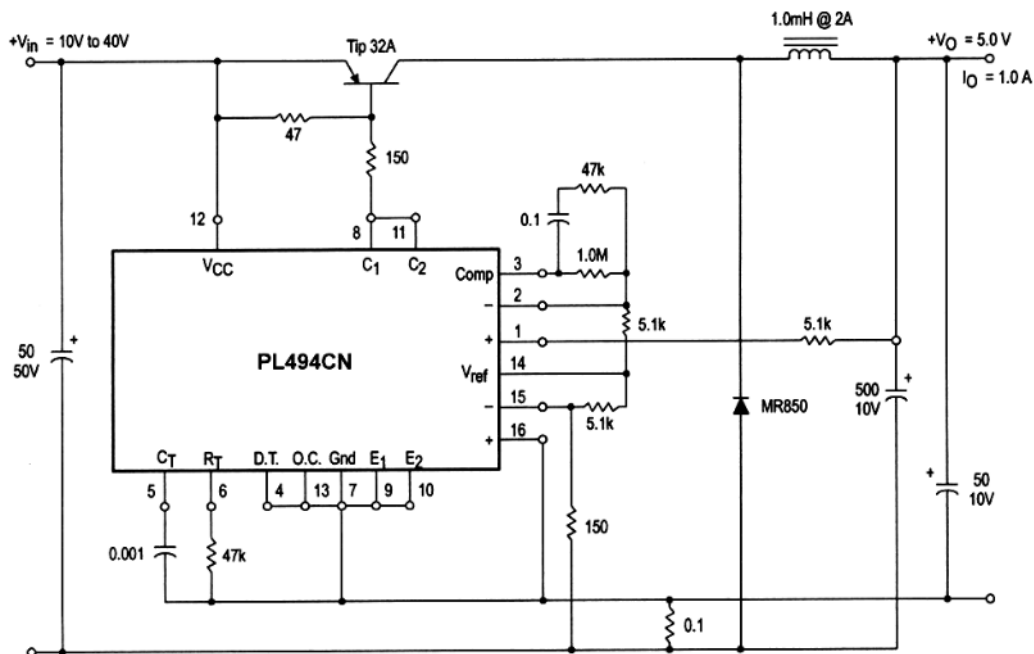
Figure 20. Pulse Width Modulated Push-Pull Converter



Test	Conditions	Results
Line Regulation	$V_{in} = 10V$ to $40V$	14 mV 0.28%
Load Regulation	$V_{in} = 28V$, $I_O = 1.0mA$ to $1.0A$	3.0 mV 0.06%
Output Ripple	$V_{in} = 28V$, $I_O = 1.0A$	65 mV pp P.A.R.D.
Short Circuit Current	$V_{in} = 28V$, $R_L = 0.1\Omega$	1.6 A
Efficiency	$V_{in} = 28V$, $I_O = 1.0A$	71%

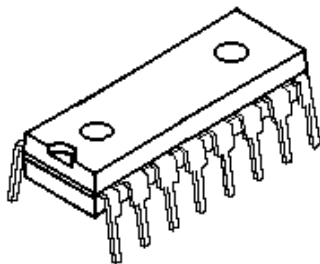
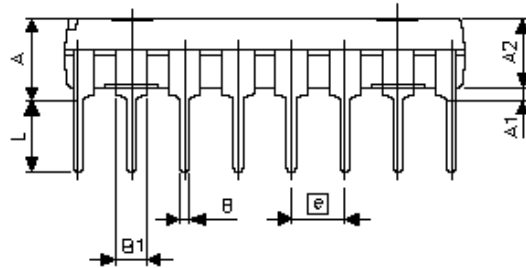
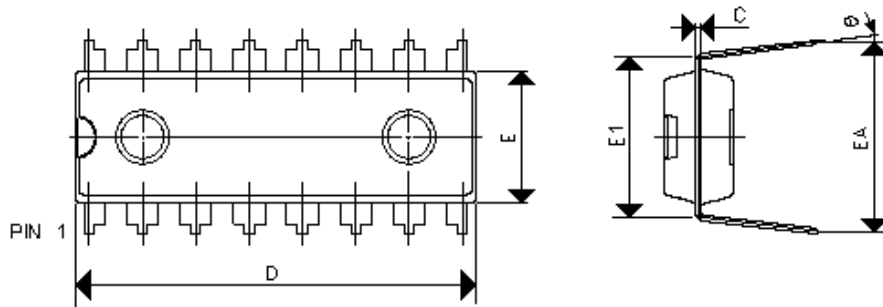
L1 - 3.5 mH @ 0.3 A
T1 - Primary: 20T C.T. #28 AWG
Secondary: 120T C.T. #36 AWG
Core: Ferroxcube 1408P-L00-3CB

Figure 21. Pulse Width Modulated Step-Down Converter



Test	Conditions	Results
Line Regulation	$V_{in} = 8.0V$ to $40V$	$3.0mV$ 0.01%
Load Regulation	$V_{in} = 12.6V$, $I_o = 0.2$ mA to 200 mA	$5.0mV$ 0.02%
Output Ripple	$V_{in} = 12.6V$, $I_o = 200$ mA	$40mV_{pp}$ P.A.R.D.
Short Circuit Current	$V_{in} = 12.6V$, $R_L = 0.1$	250mA
Efficiency	$V_{in} = 12.6V$, $I_o = 200$ mA	72%

DIP-16P Dimension

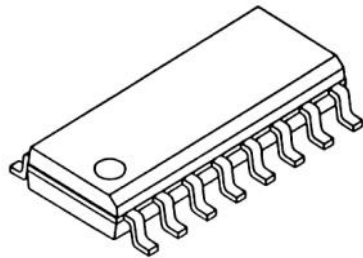
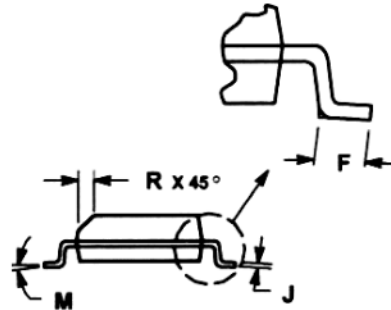
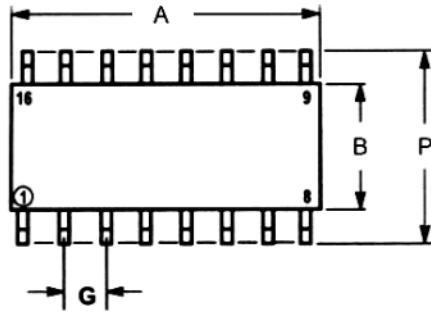


SYMBOL	DIMENSION IN INCH	DIMENSION IN MM
A	0.170 MAX.	4.318 MAX.
A1	0.015 MIN	0.381 MIN.
A2	0.130±0.005	3.302±0.127
B	0.018 TYP.	0.457 TYP.
B1	0.060 TYP.	1.524 TYP.
C	0.010 NOM.	0.254 NOM.
D	0.752±0.005	19.101±0.127
E	0.252±0.005	6.401±0.127
E1	0.300±0.010	7.62±0.254
EA	0.355±0.020	9.017±0.508
e	0.100 TYP.	2.540 TYP.
L	0.130±0.010	3.302±0.254
	0°~15°	0°~15°

NOTE :

1. DIMENSION D & E DOES NOT INCLUDE FLASH.

SOP-16 Dimension



DIM	MILLIMETERS
A	9.90 ± 0.10
B	3.90 ± 0.09
C	1.60 ± 0.13
D	0.400 (REF.)
F	0.700 ± 0.20
G	1.27 (BSC.)
J	0.250
K	0.160 ± 0.08
M	0° ~ 5°
P	5.99 ± 0.20
R	0.375