

## **High Performance Digital FM Transmitter for Portable Devices**

General Description
---------------------

The QN8007 is a high performance, low power, full-featured single-chip stereo FM transmitter designed for portable audio/video players, automotive accessories, cell phones, and GPS personal navigation devices. It integrates transmitting functions, clear channel scan, and antenna tuning to ease matching in real applications. Advanced digital architecture enables variable input gain programming; selectable pre-emphasis, precision low-spur MPX stereo encoding and pilot tone generation, low-noise PLL-based modulation, and an on-chip power amplifier with variable output level and RF band-pass filtering to ensure optimum transmit spectrum purity.

With its small footprint, minimal external component count and multiple clock frequency support, the QN8007 is easy to integrate into a variety of small form-factor low power portable applications. An integrated voltage regulator enables direct connection to a battery and provides high PSRR for superior noise suppression. A low-power Standby mode extends battery life. ESD protection is on all pins. The QN8007 is fabricated in highly reliable CMOS technology.

### Key Features

#### • Worldwide FM Band Transmit

- 76 MHz to 108 MHz full band tuning in 50/100/200 kHz step sizes
- 50/75 μs pre-emphasis

### • High Performance FM Transmitter (FMT)

- 66dB Stereo SNR, 0.03% THD
- Maximum 121dBµV RF output level with 42dB adjustable range
- · Integrated Clear Channel Scan

#### • RDS/RBDS Transmit

• Supports US and European data service, including TMC (Traffic Messaging Channel) (not available in QN8007L)

#### • Flexible Audio Interfaces

- Digital audio interface supports PS and a variety of PCM data formats with 4 different data rates
- · Programmable analog audio input/output
- · Integrated audio AGC and soft clipping

### Very Low Power Consumption

- 9.2 mA
- Integrated voltage regulator, direct connect to battery
- Power saving idle and standby modes
- Low shutdown leakage current

### • Ease of Integration

- Small footprint, 4 x 4 x 0.85mm QFN24
- Only 2 external passive components required
- Adaptive antenna tuning
- Low cellular and GPS band spurs
- High Immunity to TDMA (GSM/GPRS) burst noise
- Multiple crystal frequencies supported
- 2-wire and 3-wire control interfaces

#### • Robust Operation

- $-25^{\circ}C$  to  $+85^{\circ}C$  operation
- · ESD protection on all input and output pads

### Typical Applications \_

- Cell Phones / PDAs / Smart Phones
- Portable Audio & Media Players

- GPS Personal Navigation Devices
- Automotive and Accessories



### **CONTENTS**

1	Functional Block Diagram	4
2	Pin Assignments	5
3	Electrical Specifications	7
	3.1 I <sup>2</sup> S Interface Timing	15
4	Functional Description	17
	4.1 Transmit Mode	17
	4.2 Idle and Standby Mode	17
	4.3 Audio Interface	17
	4.4 Audio Processing	
	4.5 Channel Setting	18
	4.6 RDS/RBDS	19
5	Control Interface Protocol	20
	5.1 2-wire Serial Control Interface	20
		21
6		22
		22
	6.1 Introduction	
	6.3 I <sup>2</sup> S Interface Timing Description	
	6.3.1 MSB-Justified (Format 0)	
	0.5.1 Wisb-sustified (Format 0)	23
	6.3.2 I <sup>2</sup> S (Format 1)	
	6.3.3 DSP1 (Format 2)	23
	6.3.4 DSP2 (Format 3)	24
	6.3.5 LSB-Justified (Format 4)	24
7	User Control Registers	25
8	Package Description	44
9	Solder Reflow Profile	
,	9.1 Package Peak Reflow Temperature	
	9.2 Classification Reflow Profiles	
	9.3 Maximum Reflow Times	
	/.J 1710/MILIUM INCHOW 1 IMPO	<del>-</del> /



### **REVISION HISTORY**

REVISION	CHANGE DESCRIPTION	DATE
1.0	Datasheet Release	06/12/08
2.0	B1 Register set incorporated, INT pin added to Figure 1.	8/26/08
2.01	<ol> <li>Table 3 Vcc range:2.7~5.0 v, TYP:3.3 v;</li> <li>In Section 4.1: Modify 124 to 121dBμVp.</li> <li>In Section 4.4: Delete" When there is no audio signal for a pre-determined period, AGC will power down the transmitter."</li> </ol>	03/17/09
2.02	Modify Figure 1, Figure 2 and the pictures in Section 6.3: Rin20, should be DIN not DIN/INT	05/12/09
2.03	Add the Register STATUS2(RO) Address 1Bh	06/02/09
2.04	Update the data in Table 4 and Table 6.	06/19/09
2.05	Modify the register PWROUT_CAL PAC_TARGET and the description	07/30/09
2.06	Update the registers in Section 7.	08/05/09
2.07	Modify t <sub>dtHD</sub> MIN 100 ns 20 ns in table 8	09/09/09
2.08	Modify the description of register PAC_TARGET	10/12/09
2.09	<ol> <li>Modify I<sup>2</sup>C to 2-wire;</li> <li>Add/Reg 59h, Reg 5Ah.</li> <li>Update the grammar and syntax.</li> <li>Modify V<sub>ain</sub> Max value: 1400 mV→2000 mV</li> </ol>	11/19/09

### **STATEMENT:**

Users are responsible for compliance with local regulatory requirements for low power unlicensed FM broadcast operation. Quintic is not responsible for any violations resulting from user's intentional or unintentional breach of regulatory requirements in personal or commercial use.



## 1 FUNCTIONAL BLOCK DIAGRAM

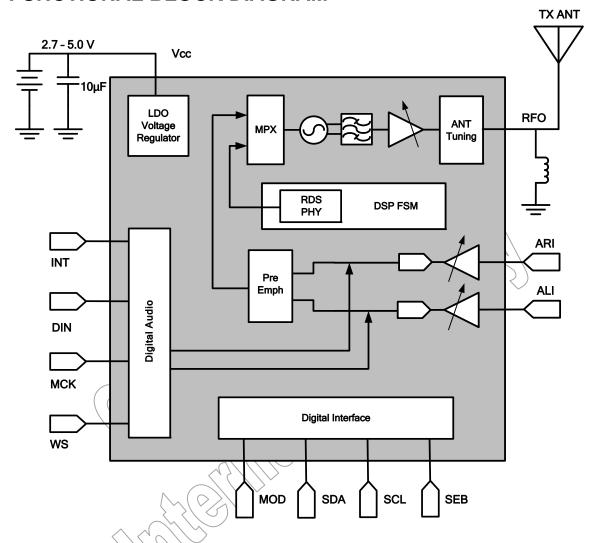
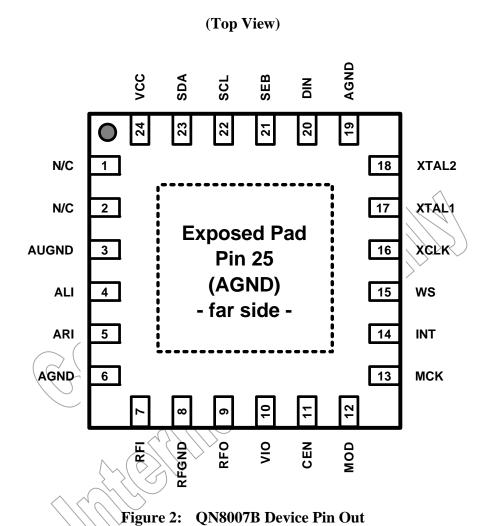


Figure 1: QN8007 Functional Blocks



## **2 PIN ASSIGNMENTS**





## **Table 1: Pin Descriptions**

PINS	NAME	DESCRIPTION
1	N/C	No connect
2	N/C	No connect
3	AUGND	Audio ground
4	ALI	Analog audio input – left channel
5	ARI	Analog audio input – right channel
6	AGND	Analog ground
7	RFI	Receiver RF input
8	RFGND	RF ground
9	RFO	Transmitter RF output – connect to matched antenna.
10	VIO	IO voltage - specifies voltage limit for all digital pins.
11	CEN	Chip enable: Chip power down if less than 0.6V, (see also MOD pin below) power up if voltage applied >min (0.7*VIQ, 1.8V).
12	MOD	Bus mode: HIGH \$\infty\$3-wire serial operation. LOW \$=\; 2-wire serial operation  Note: Both MOD=0 and CEN=0 to disable chip.
13	MCK	Master clock - for digital audio interface.
14	INT	Interrupt
15	WS	Word select (I <sup>2</sup> S mode only)
16	XCLK	External clock input (register 49h, bit 4 must be HIGH)
17	XTALL	On-chip crystal driver port 1. If using an external clock source, connect this pin to ground.
18	XTAL2	On-chip crystal driver port 2  If using an external clock source, connect this pin to ground.
19	AGND	Analog ground
20	DIN	Data in (12S mode only)
21	SEB	Serves as the bus enable pin in 3-wire serial mode; serves as the address select pin in 2-wire serial mode, SEB = Low for default address, SEB = High for register controlled address.
22	SCL	Clock for 2-wire or 3-wire serial bus.
23	SDA	Bi-directional data line for 2- or 3-wire serial bus.
24	VCC	Voltage supply
25	PAD	Exposed pad, must be soldered to the ground on the PCB.



## **3 ELECTRICAL SPECIFICATIONS**

**Table 2: Absolute Maximum Ratings** 

SYMBOL	PARAMETER	CONDITIONS	MIN	MAX	UNIT
$V_{bat}$	Supply voltage	VCC to GND	-0.3	5	V
V <sub>IO</sub>	Logic signals	CEN, SEB, SCL, SDA, MOD, to GND	-0.3	3.6	V
$T_{s}$	Storage temperature		-55	+150	°C

**Table 3: Recommended Operating Conditions** 

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
Vcc	Supply voltage	VCC to GND	2.7	3.3	5.0	V
$T_{A}$	Operating temperature		-25		+85	°C
V <sub>ain</sub>	L/R channel input signal level	Single ended peak to peak voltage		1000	2000	mV
V <sub>io</sub>	Digital I/O voltage		1.6		3.6	V



### **Table 4: DC Characteristics**

 $(Vcc = 2.7 \sim 5.0 \text{ V}, T_A = -25 \sim 85 \text{ }^{\circ}\text{C}, \text{ unless otherwise noticed.}$  Typical values are at Vcc = 3.3 V, f carrier=88 MHz and  $T_A = 25 \text{ }^{\circ}\text{C})$ .

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT		
$I_{TX}$	Transmit mode supply current <sup>1</sup>	digital audio interface		6.8		mA		
1TX	Transmit mode supply current	analog audio interface	9.2		12.9	mA		
$I_{IDLE}$	Idle mode supply current	Idle mode		3.5		mA		
$I_{STBY}$	Standby mode supply current	Standby mode		350		μΑ		
$I_{PDN}$	Power down leakage current	power down		5	15	μΑ		
Interface								
$V_{\mathrm{OH}}$	High level output voltage		0.9*V <sub>IO</sub>			V		
V <sub>OL</sub>	Low level output voltage				0.1*V <sub>IO</sub>	V		
$V_{\mathrm{IH}}$	High level input voltage		1.7 or 0.7*V <sub>10</sub>			V		
V <sub>IL</sub>	Low level input voltage				0.6	V		
Notes: 1. Max: F	RFO output level is 121dBuVp, Mi	n: RFO=82 dBµVp.						

### **Table 5: AC Characteristics**

(Vcc =  $2.7 \sim 5.0$  V,  $T_A = -25 \sim 85$  °C, unless otherwise noticed. Typical values are at Vcc = 3.3V and  $T_A = 25$ °C).

SYMBOL	PARAMETERS	CONDITIONS	MIN	TYP	MAX	UNIT			
$F_{xtal}$	Crystal or Clock frequency	Real-Time Clock		7.6 - 38.41		MHz			
$F_{xtal\_err}$	Crystal frequency accuracy	Over temperature, and aging	-20		20	ppm			
Notes: 1. See	Notes: 1. See also XSEL [3:0] (register 03h, bits 3:0).								



### **Table 6: Transmitter Characteristics**

 $(Vcc = 2.7 \sim 5.0 \text{ V}, T_A = -25 \sim 85 \text{ °C}, unless otherwise noticed.} \text{ Typical values are at Vcc} = 3.3 \text{V}, \text{ f carrier} = 88 \text{ MHz and } T_A = 25 \text{°C}).$ 

SYMBOL	PARAMETERS	CONDITIONS	MIN	TYP	MAX	UNIT
R <sub>audio_in</sub>	Audio input impedance	At pin ALI and ARI	10		80	kΩ
$C_{audio\_in}$	Audio input capacitance <sup>1</sup>	At pin ALI and ARI		2	5	pF
$G_{audio\_In}$	Audio input gain	RIN[1:0] = 01	-1.5		15	dB
$\Delta G_{audio\_In}$	Audio gain step	For any gain setting	0.5	1	1.5	dB
-	Pre-emphasis time	PETC = 1	71.3	75	78.7	110
$ au_{ m emph}$	constant <sup>1</sup>	PETC = 0	47.5	50	52.5	μs
		MONO, $\Delta f = 22.5 \text{ kHz}$	>	65 _		
SNR <sub>audio_tx</sub>	Tx audio SNR <sup>3</sup>	STEREO, $\Delta f = 67.5 \text{ kHz}$ , $\Delta f_{\text{pilot}} = 6.75 \text{ kHz}$		66		dB
THD <sub>audio_tx</sub>	Tx audio THD <sup>3</sup>	STEREO, $\Delta f = 67.5 \text{ kHz}$ , $\Delta f_{\text{pilot}} = 6.75 \text{ kHz}$		0.03	0.1	%
$\alpha_{LR\_tx}$	L/R separation <sup>2, 3</sup>		40	45		dB
$\mathrm{B}_{\mathrm{LR\_tx}}$	L/R channel imbalance 1,2	L and R channel gain imbalance at 1 kHz offset from DC			1	dB
$M_{pilot}$	19 kHz pilot modulation	Relative to 75 kHz deviation	7	9.0	15	%
$\mathrm{SUP}_{\mathrm{sub}}$	38 kHz sub-carrier <sup>2, 3</sup> suppression		70			dB
$C_{\text{tune}}$	Output capacitance tuning range		5		30	pF
P <sub>out</sub>	RF output voltage swing <sup>4</sup>	RF Channel frequency = 88 MHz	82		121	dBμV
$\Delta G_{RF\_Out}$	Power gain step	Over process, temperature		1.5		dB
$\Delta P_{out}$	Power gain flatness	Over 76 MHz ~ 108 MHz	-2		2	dB
	7.7	120 kHz to 240 kHz offset		-50	-45	
$P_{\text{mask}}$	RF output spectrum mask <sup>5</sup>	240 kHz to 600 kHz offset		-55	-45	dBc
		>600 kHz offset			-45	
$F_{rf}$	RF channel frequency		76		108	MHz
$F_{ch}$	Channel frequency step		50	100	200	kHz
$F_{err}$	Channel center frequency accuracy		-2		2	kHz
$F_{perr}$	Pilot Tone frequency accuracy <sup>1</sup>		-2		2	Hz
$F_{pk}$	Modulation peak frequency deviation			75		kHz



|--|

#### Notes:

- 1. Guaranteed by design.
- 2. Stereo ( $ST_MO_TX = 0$ ).
- 3. 1000mVp-p, 1 kHz tone at ALI pin, no input signal at ARI pin.
- 4. Into matched antenna (see application note for details).
- 5. Within operating band 76 MHz to 108 MHz.
- 6. Value set with GAIN\_TXPLT [5:0] (reg. 0Fh, bits 5:0). The user must conform to local regulatory requirements for low-power unlicensed FM broadcast operation when setting this value.

### **Table 7: Timing Characteristics**

 $(\text{Vcc} = 2.7 \sim 5.0 \text{ V}, \text{ T}_{\text{A}} = -25 \sim 85 \,^{\circ}\text{C}, \text{ unless otherwise noticed.}$  Typical values are at Vcc = 3.3V and T<sub>A</sub> = 25°C).

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
$ au_{ ext{pup}}$	Chip power-up time <sup>1</sup>	From rising edge of CEN to PLL settled and transmitter ready for transmission.			0.6	Sec
$ au_{ m astby}$	Auto Standby time <sup>2</sup>	TMOUT $[1:0] = 00$ TMOUT $[1:0] = 01$ TMOUT $[1:0] = 10$ TMOUT $[1:0] = 11$		1 3 5 Never		Min
$ au_{chsw}$	Channel switching time <sup>1</sup>	From any channel to any channel.			0.1	Sec
Transmitter Timing						
$ au_{ m wkup}$	Wake-up time from standby to transmit			25	200	msec
$ au_{ ext{CCS}}$	Clear channel scan time	Per channel.		5		msec

#### Notes:

- 1. Guaranteed by design.
- 2. Chip automatically goes from idle to standby mode; TMOUT = 11 equivalent to auto standby disabled.

### **Table 8: 2-Wire Interface Timing Characteristics**

 $(\text{Vcc} = 2.7 \sim 5.0 \text{ V}, \text{ T}_{\text{A}} = -25 \sim 85 \, ^{\circ}\text{C}, \text{ unless otherwise noticed}. \text{ Typical values are at Vcc} = 3.3 \text{V} \text{ and } \text{T}_{\text{A}} = 25 \, ^{\circ}\text{C}).$ 

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
$f_{SCL}$	2-wire clock frequency				400	kHz
$t_{LOW}$	Clock Low time		1.3			μs
t <sub>HI</sub>	Clock High time		0.8			μs
$t_{\mathrm{ST}}$	SCL input to SDA falling edge start <sup>1,3</sup>		0.6			μs
$t_{STHD}$	SDA falling edge to SCL falling edge start <sup>3</sup>	$\langle$	0.6			μs
$t_{rc}$	SCL rising edge <sup>3</sup>	Level from 30% to 70%			300	ns
$t_{fc}$	SCL falling edge <sup>3</sup>	Level from 70% to 30%			300	ns
$t_{ m dtHD}$	SCL falling edge to next SDA rising edge <sup>3</sup>		20			ns
t <sub>dtc</sub>	SDA rising edge to next SCL rising edge <sup>3</sup>				900	ns
$t_{stp}$	SCL rising edge to SDA rising edge <sup>2,3</sup>		0.6			μs
$t_{\rm w}$	Duration before restart <sup>3</sup>		1.3			μs
$C_b$	SCL, SDA capacitive loading <sup>3</sup>			10		pF

### Notes:

- 1. Start signaling of 2-wire interface.
- 2. Stop signaling of 2-wire interface.
- 3. Guaranteed by design.

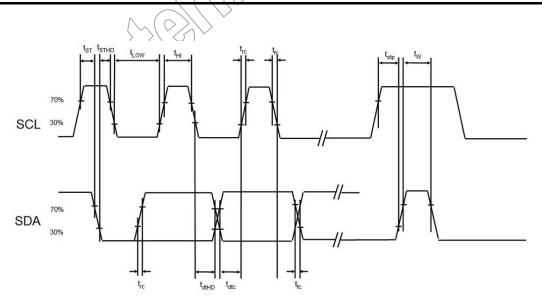


Figure 3: 2-wire Serial Control Interface Timing Diagram



## **Table 9: 3-Wire Interface Timing Characteristics**

(  $Vcc = 2.7 \sim 5.0 \text{ V}$ ,  $T_A = -25 \sim 85 \,^{\circ}\text{C}$ , unless otherwise noticed. Typical values are at Vcc = 3.3 V and  $T_A = 25 \,^{\circ}\text{C}$ ).

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
$f_{CLK}$	Bus clock frequency				2.5	MHz
$t_{ m HI}$	SCL high time		50			ns
$t_{LOW}$	SCL low time		25			ns
ts	SEB and SDA falling edge to clock rising edge <sup>1</sup>		20			ns
$t_h$	Data holding time <sup>1</sup>		10			ns
t <sub>tr</sub>	SCL rising edge to SDA output valid <sup>1</sup>	Only in read mode.	2	<	50	ns
t <sub>ed</sub>	SCL rising edge to SDA output high Z <sup>1</sup>		2		25	ns
Notes: 1. Guara	anteed by design.					



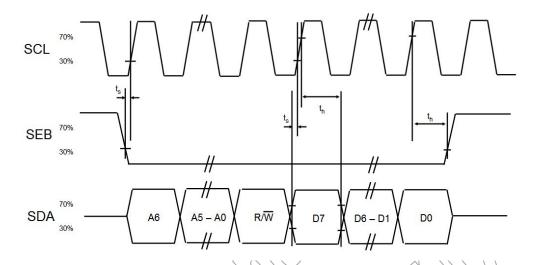


Figure 4: 3-Wire Serial Control Interface Write Timing Diagram

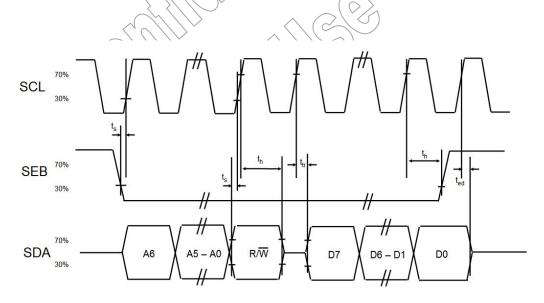


Figure 5: 3-Wire Serial Control Interface Read Timing Diagram



### **Table 10: Digital Audio Interface Timing Characteristics**

### Master Clock:

(  $Vcc = 2.7 \sim 5.0 \text{ V}$ ,  $T_A = -25 \sim 85$  °C, unless otherwise noticed. Typical values are at Vcc = 3.3 V and  $T_A = 25$  °C).

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
$f_{CLK\_M}$	Bus clock frequency				3.072	MHz
$ au_{rise\_M}$	Rise-time				0.5	ns
$ au_{\mathrm{fall\_M}}$	Fall-time				0.5	ns
F <sub>err_M</sub>	Frequency accuracy	According to PLL.			100	ppm

### Slave Clock:

(  $Vcc = 2.7 \sim 5.0 \text{ V}$ ,  $T_A = -25 \sim 85 \,^{\circ}\text{C}$ , unless otherwise noticed. Typical values are at  $Vcc = 3.3 \,^{\circ}\text{V}$  and  $T_A = 25 \,^{\circ}\text{C}$ )

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
$f_{CLK\_S}$	Bus clock frequency				V 10	MHz
$ au_{rise\_S}$	Rise-time			$\bigcirc)$	1.5	ns
$ au_{\mathrm{fall\_S}}$	Fall-time				1.5	ns
F <sub>err_S</sub>	Frequency accuracy				100	ppm



## 3.1 I<sup>2</sup>S Interface Timing

**Note:** The term 'receiver' as described below is from the QN8007's point of view.

Either the QN8007 or the external device can act as the system master by providing the necessary clock signals. The slave will usually derive its internal clock signal from an external clock input. This means, taking into account the propagation delay between the master clock and the data and/or word-select signals, that the total delay is simply the sum of:

- The delay between the external (master) clock and slave's internal clock;
- The delay between the internal clock and the data and/or word-select signals.

For data and word-select inputs, the external to internal clock delay is of no consequence because it only lengthens the effective set-up time (see Figure 6:). The major part of the time margin is to accommodate the difference between the propagation delay of the transmitter, and the time required to set up the receiver. All timing requirements are specified relative to the clock period of a device. This means that higher data rates can be used in the future.

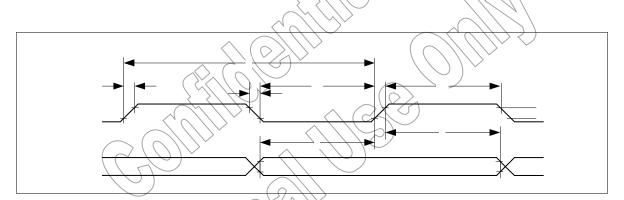


Figure 6: Timing for QN8007 as I<sup>2</sup>S Slave and Receiver

Table 11: Timing for QN8007 as I2S Slave and Receiver

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
T	I <sup>2</sup> S clock frequency		100			ns
$t_{ m LC}$	Clock low time		10			ns
$t_{HC}$	Clock high time		10			ns
$t_{\rm s}$	WS and SD setup time		10			ns
$t_{\rm h}$	WS and SD hold time		5			ns
$t_{RC}$	Clock rise-time				5	ns
$t_{FC}$	Clock fall-time				5	ns

Т

 $t_{\rm FC}$ 

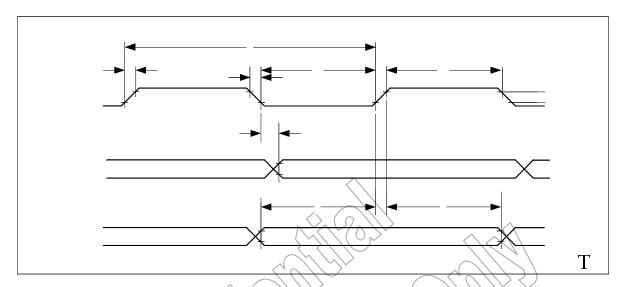


Figure 7: Timing for QN8007 as  $I^2S$  Master and Receiver  $t_{FC}$ 

Table 12: Timing for QN8007 as 12 S Master and Receiver

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
T	I <sup>2</sup> S clock period		330			ns
$t_{ m LC}$	Clock low time		120			ns
$t_{HC}$	Clock high time		120			ns
$t_{\rm s}$	SD setup time		10			ns
$t_h$	SD hold time	MIC	5			ns
$t_{ m dtr}$	WS delay time	W S			10	ns
$t_{RC}$	Clock rise-time				5	ns
$t_{FC}$	Clock fall-time				5	ns

DIN



### 4 FUNCTIONAL DESCRIPTION

The QN8007 is a high performance, low power, single chip FM transmitter IC that supports worldwide FM broadcast band operation. It has idle and standby modes for saving power. RDS/RBDS data service is also supported.

### 4.1 Transmit Mode

The QN8007 transmitter uses a highly digitized architecture. The input left and right analog audio signals are first adjusted by two automatic gain controlled (AGC) amplifiers, and then digitized by two high resolution ADCs into the digital domain. If a digital audio interface is used, the analog input circuits and ADCs will be bypassed. Pre-emphasis, soft clipping and MPX encoding are then performed. If RDS mode is enabled, the RDS signal will also be mixed with the MPX signal and the combined output will be fed into a high performance digital FM modulator which generates FM signal at RF carrier frequency. The FM signal is then filtered and amplified by the PA.

The QN8007 can deliver up to  $121 \ dB\mu V$  output signal to an external antenna and/or matching network. An RF VGA provides a  $42 \ dB$  of output power control range in 1.5 dB steps and can be programmed through the serial control bus. Output power control and in-band power flatness can be easily achieved by a calibration circuit. This wide range of control allows for various antenna configurations such as loop, monopole, or meandering traces on PCB. An integrated RF bandpass filter ensures optimal output spectral purity.

## 4.2 Idle and Standby Mode

The QN8007 features low power IDLE and STANDBY modes for fast turn around and power saving. After power up, the QN8007 will enter IDLE mode automatically. If there is no transmitting requirement in a pre-determined time period, the QN8007 will enter STANDBY mode automatically. The auto-standby function can be enabled or disabled through the serial control interface.

### 4.3 Audio Interface

The QN8007 supports both analog and digital audio interfaces providing maximum flexibility in real applications.

### **Analog Audio Interface**

The QN8007 has a highly flexible analog audio interface. For audio input, the signal is AC coupled with a 3dB corner frequency less than 50Hz. It has 4 different input impedances and 18dB adjustable gain range (in 1.5dB step) to optimize the SNR and linearity. The gain setting can be controlled automatically by integrated AGC or manually set through serial interface.

The single ended audio output level is 1V peak to peak and will be AC coupled to external audio driver.

#### **Digital Audio Interface**

With digital audio, the interface operates in slave mode and supports MSB-Justified, LSB-Justified, I<sup>2</sup>S, DSP1, and DSP2. The three interface lines are MCK, DIN, and WS. MCK and WS can be tri-stated to allow for multiplexing.



### 4.4 Audio Processing

The QN8007 supports audio AGC, programmable pre-emphasis, and soft clipping. The AGC state machine will detect the signal level and control the VGA gain to optimize both SNR and THD. A saturation indicator is also integrated which will be asserted when the input signal is out of the range of AGC. A soft clipping feature provides graceful performance degradation when the signal level is higher than a pre-determined level.

Stereo signal is generated by the MPX circuit. It combines the left and right channel signals in the following way:

$$m(t) = [L(t) + R(t)] + [L(t) - R(t)] \sin(4\pi ft + 2\theta_0) + \alpha \sin(2\pi ft + \theta_0) + d(t) \sin(6\pi ft + 3\theta_0)$$

Here, L(t) and R(t) correspond to the audio signals on left and right channels respectively, f = 19 kHz,  $\theta$  is the initial phase of pilot tone and  $\alpha$  is the magnitude of pilot tone, and d(t) is RDS signal. In mono mode, only the L+R portion of audio signal is transmitted. The 19 kHz pilot tone is generated by the MPX circuit which contributes 9% of peak modulation, and RDS signal will contribute 2.1% of peak modulation.

A pre-emphasis function is also integrated with both  $75\mu s$  and  $50\mu s$  time constants. The time constant can be programmed through the serial control interface.

## 4.5 Channel Setting

The QN8007 supports both clear channel scan and manual channel setting.

### **Manual Channel Setting**

By programming channel index CH[9:0], the RF channel can be set to any frequency between 76 MHz ~ 108 MHz in 50 kHz steps. The channel index and RF frequency have the following relationship:

 $F_{RF} = (76 + 0.05 \text{ x Channel Index})$ , where  $F_{RF}$  is the RF frequency in MHz.

The QN8007 has an integrated crystal oscillator and supports various crystal frequencies. Alternatively, the QN8007 can be driven externally by various clock frequencies.

#### **Clear Channel Scan**

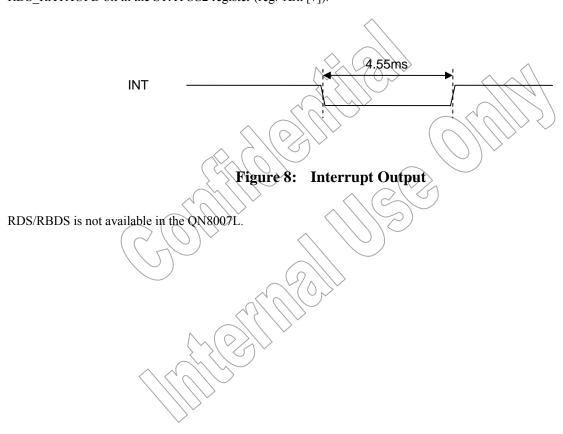
The QN8007 can automatically find the clearest channel and return the channel information for FM transmission. The start, stop and frequency step of searching as well as upward or downward searching can be programmed through serial interface.



### 4.6 RDS/RBDS

The QN8007 supports RDS/RBBS data transmitting, including station ID, Meta data, TMC information, etc. The integrated RDS processor performs all symbol encoding/decoding, block synchronization, error detection and correction functions. RDS/RBDS data communicates with an external MCU through the serial control interface.

The INT pin is used for the interrupt signal as shown in Figure 8. Ping-pong buffers are used so that the user can write into one buffer while the RDS data in the other buffer is being transmitted. When the internal RDS buffer (8 bytes) is full, an Interrupt signal is generated. The user should wait for the Interrupt signal (INT) before toggling the RDSTXRDY bit in the SYSTEM2 register (reg. 01h [2]). Alternatively, the user can also check the RDS buffer space by reading the RDS RXTXUPD bit in the STATUS2 register (reg. 1Bh [7]).





### 5 CONTROL INTERFACE PROTOCOL

The QN8007 supports 2- wire and 3-wire serial interfaces. The interface selection is controlled by the MOD pin which determines whether a 2- wire or a 3-wire serial interface will be used. MOD = HIGH selects a 3-wire bus and LOW selects a 2-wire bus.

### 5.1 2-Wire Serial Control Interface

The 2-wire bus is a simple bi-directional bus interface. The bus requires only serial data (SDA) and serial clock (SCL) signals. The bus is 8-bit oriented. Each device is recognized with a unique address. Each register is also recognized with a unique address. A third line (SEB) is used to choose the device address configuration. SEB = LOW selects the default address (0101011), SEB = HIGH selects register defined addressing. The L2 bus operates with a maximum frequency of 400 kHz. Each data put on the SDA must be 8 bits long (Byte) from MSB to LSB and each byte sent should be acknowledged by an "ACK" bit. In case a byte is not acknowledged, the transmitter should generate a stop condition or restart the transmission. If a stop condition is created before the whole transmission is completed, the remaining bytes will keep their old setting. In case a byte is not completely transferred, it will be discarded.

Data transfer to and from the QN8007 can begin when a start condition is created. This is the case if a transition from HIGH to LOW on the SDA line occurs while the SCL is HIGH. The first byte transferred represents the address of the IC plus the data direction. The default IC address is 0101011. A LOW LSB of this byte indicates data transmission (WRITE) while a HIGH LSB indicates data request (READ). This means that the first byte to be transmitted to the QN8007 should be "56" for a WRITE operation or "57" for a READ operation.

The second byte is the starting register address (N) for write/read operation. The following bytes are register data for address N, N+1, N+2, etc. There is no limit on the number of bytes in each transmission. A transmission can be terminated by generating a stop condition, which is SDA transition from LOW to HIGH while SCL is HIGH. For write operation, master stops transmission after the last byte. For read operation, master doesn't send ACK after receiving the last read back byte; then stops the transmission.

The following timing diagram is for both write and read.

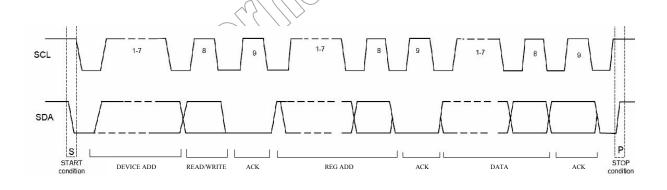


Figure 9: 2-Wire Serial Control Interface Protocol



### 5.2 3-Wire Serial Control Interface

For 3-wire serial operation, a transfer begins when the SEB pin is set LOW on a rising SCL edge. The control word is latched internally on rising SCL edges and is 8 bits in length, comprised of a 7-bit register address A6:A0, and a read/write bit (read = 1, write = 0). The ordering of the control word is A6:A0, R/W as shown in Figure 11.

For write operations, the serial control word is followed by an 8-bit data word and is latched internally on rising SCL edges. For read operations, a bus turn-around of half a cycle is followed by an 8-bit data word shifted out on rising SCL edges. The transfer ends on the rising SCL edge after SEB is set HIGH. After the 16th data bit, a full clock with both rising and falling edges is needed to shift in the control word.

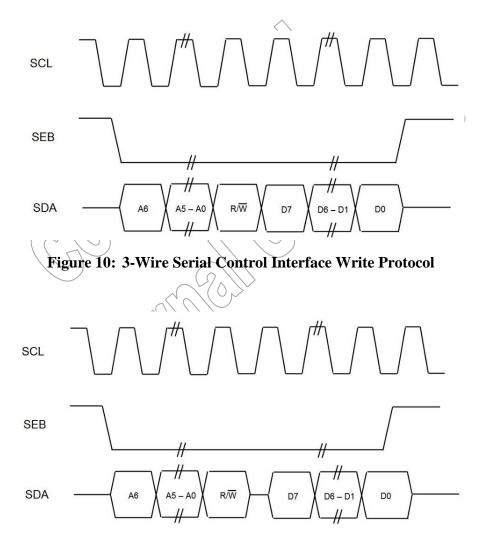


Figure 11: 3-Wire Serial Control Interface Read Protocol



## 6 DIGITAL AUDIO INTERFACE PROTOCOL

### 6.1 Introduction

The QN8007 uses an I<sup>2</sup>S interface to transfer audio data from the external source.

- Master or Slave modes are supported.
- Multiple data widths are supported: 8, 16, 24, and 32 bits.
- Multiple data alignments are supported: I<sup>2</sup>S, DSP1, DSP2, MSB-Justified, and LSB-Justified modes.
- The term 'transmitter' as described below is from the QN8007's point of view.

## 6.2 I<sup>2</sup>S BUS Signal Description

A 3-line serial bus is used consisting of a line for two time-multiplexed data channels, a word select line, and a clock line. The following figure shows how to use these three signals in the QN8007.

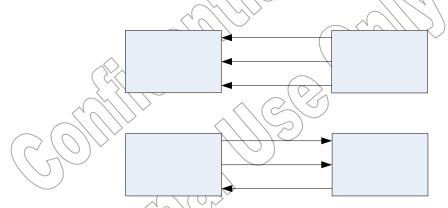


Figure 12: Top Level Block Diagram of I<sup>2</sup>S

In Master mode, the serial clock signal SCK and the word selection signal WS are generated by the QN8007 and are output to the external device. In Slave mode, those two signals are input signals from the external device.

The QN8007 receives the serial data from the external device by the SD signal.

QN8007 as Slave and Receiver

QN8007



## 6.3 I<sup>2</sup>S Interface Timing Description

The word select line indicates the channel being transmitted:

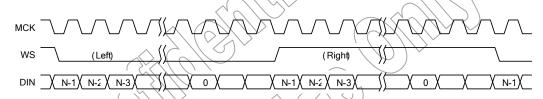
- WS = 0; channel 1 (left);
- WS = 1; channel 2 (right).

WS sent by the transmitter is synchronized by the trailing (HIGH-to-LOW) edge of the serial clock signal and is latched into the receiver on the leading (LOW-to-HIGH) edge of the serial clock signal.

The serial data signal SD has the same timing requirement as the WS signal. It is sent on the trailing edge of the clock signal by the transmitter, and received on the leading edge of the clock signal by the receiver. The serial data is transmitted in two's complement with the MSB first. The MSB is transmitted first because the transmitter and the receiver may have different word lengths. There are four data word lengths supported by the QN8007; 8, 16, 24 and 32.

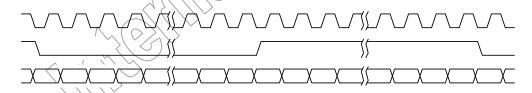
There are five data alignment modes supported by QN8007. The detailed timing descriptions are shown as below.

### 6.3.1 MSB-Justified (Format 0)



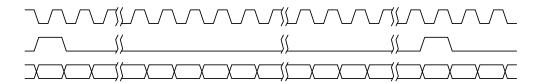
The transmitter sends the MSB (N-1) bit of the next word when the WS changes and sends the second MSB bit in the next clock period. Each bit is sent by the transmitter in one clock period until the LSB (0) bit is sent. The N indicates the word length that can be 8, 16, 24, and 32.

### **6.3.2** $I^2S$ (Format 1)



The transmitter sends the MSB(N-1) bit of the next word one clock period after the WS changes and sends the second MSB bit in the next clock period. Each bit is sent by the transmitter in one clock period until the LSB (0) bit is sent. In the case that the number of cycles equal the number of bits to be sent, the LSB (0) bit could be sent after the next WS change. The N indicates the word length that can be 8, 16, 24, and 32.

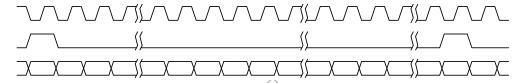
### **6.3.3 DSP1** (Format 2)





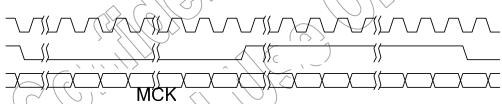
The transmitter sends the MSB (N-1) bit of the next word when the WS pulse occurs and sends the second MSB bit in the next clock period. Each bit is sent by the transmitter in one clock period until the LSB (0) bit is sent. Following the first LSB, transmitter sends the MSB (N-1) bit of a new word and then keeps sending data until the data transmission is finished. The N indicates the word length that can be 8, 16, 24, and 32.

### **6.3.4** DSP2 (Format 3)



The transmitter sends the MSB(N-1) bit of the next word one clock period after the WS pulse occurs and sends the second MSB bit in the next clock period. Each bit is sent by the transmitter in one clock period until the LSB (0) bit is sent. Following the first LSB, transmitter sends the MSB(N-1) bit of a new word and then keeps sending data until the data transmission is finished. The N indicates the word length that can be 8, 16, 24, and 32.

### 6.3.5 LSB-Justified (Format 4)



The transmitter sends the MSB (N-1) bit of the next word in the (N-1)th clock period back-counting from the next WS changes. The second MSB bit is sent in the next clock period. Finally, the LSB(0) bit is sent in the clock period that is just before the next WS changes in one clock period indicates the word length that the period be 8, 16, 24, and 32.

ŊłΩ

N-1 N-2

0

MCK

DIN

Rev 2.09 (11/09) Confidential A Copyright ©2009 by Quintic Corporation

Confidential Information contained herein is covered under Non-Displosure Agreement (NDA).

Page 24

N-1 N-2

1



## 7 USER CONTROL REGISTERS

----- THIS IS A PREVIEW LIST. Number and content of registers subject to change without notice ------

There are 28 user accessible control registers. All registers not listed below are for manufacturing use only.

**Table 13: Summary of User Control Registers** 

REGISTER	NAME	USER CONTROL FUNCTIONS
00h	SYSTEM1	Sets device modes.
01h	SYSTEM2	Sets device modes, resets.
02h	DEV_ADD	Sets device address:
03h	ANACTL1	Analog control functions.
04h	REG_VGA	TX mode input impedance, crystal cap load setting.
05h	CID1	Device ID numbers
06h	CID2	Device ID numbers.
07h	I <sup>2</sup> S	Sets I <sup>2</sup> S parameters.
08h	СН	Lower 8 bits of 10-bit channel index.
09h	CH_START	Lower 8 bits of 10-bit channel scan start channel index.
0Ah	CH_STOP	Lower 8 bits of 10-bit channel scan stop channel index.
0Bh	CH_STER	Channel scan frequency step. Highest 2 bits of channel indexes.
0Ch	PAC_TARGET	Output power calibration control.
0Dh \	TXAGC GAIN	Sets TX parameters.
0Eh	TX_FDEV	Specify total TX frequency deviation.
0Fh	GAIN_TXPLT	Gain of TX pilot frequency deviation, I <sup>2</sup> S buffer clear.
10h	RDSD0	RDS data byte 0.
11h	RDSD1	RDS data byte 1.
12h	RDSD2	RDS data byte 2.
13h	RDSD3	RDS data byte 3.
14h	RDSD4	RDS data byte 4.
15h	RDSD5	RDS data byte 5.
16h	RDSD6	RDS data byte 6.
17h	RDSD7	RDS data byte 7.
18h	RDSFDEV	Specify RDS frequency deviation, RDS mode selection.
19h	CCA	Sets CCA parameters.
1Ah	STATUS1	Device status indicators.
1Bh	STATUS2	Device status indicators.
49h	REG_XLT3	XCLK pin control.
59h	PAC_CAL	PA tuning cap calibration.
5Ah	PAG_CAL	PA gain calibration.



### Register Bit R/W Status:

RO - Read Only: You can not program these bits.

WO - Write Only: You can write and read these bits; the value you read back will be the same as written.

R/W - Read/Write: You can write and read these bits; the value you read back can be different from the value w

Typically, the value is set by the chip itself. This could be a calibration result, AGC FSM result, etc.

written.

Word: SYSTEM1 Address: 00h

Bit 7 (MSB)	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0 (LSB)
rsvd	txreq	chsc	stnby	rsvd	txi2s	rdsen	cca_ch_dis
wo	wo	wo	wo	wo	wo	wo	wo

Bit	Symbol	Default	Description
7	rsvd	0	Reserved
6	TXREQ	0	Transmission-request:
			0 Non TX mode. Either idle or standby mode.
		(	Enter Transmit mode.
5	CHSC	0	Channel Scan mode enable: Combined with TXREQ, the chip scans for an empty channel for transmission. After completing channel scanning, this bit will be cleared automatically.
		5	For CCS (TX Scan), the clearest channel (channel with weakest RSSI) will be selected (if TXCCAA is not equal zero, another prior condition should be met, see description of CCA register at 19h). To use the scanned channel, set CCA_CH_DIS to 0. (CCA_CH_DIS can be set to 0 at the same time CHSC=1).
		_	0 Normal operation.
		\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\	Channel Scan mode operation.
4	STNBY <		Request immediately to enter Standby mode if the chip is in IDLE and no TXREQ is received.
			0 Non standby mode. Either idle, or TX mode.
			1 Enter standby mode.
3	rsvd	0	Reserved
2	TXI2S	0	I <sup>2</sup> S enable:
			0 Use analog input for TX audio.
			1 Use I <sup>2</sup> S digital signal for TX audio.
1	RDSEN	0	RDS enable:
			0 RDS disable.
			1 RDS enable.
0	CCA_CH_DIS	1	CH (channel index) selection method: See description for CH register at 08h and 0Bh for more information.
			0 CH is determined by internal CCA (channel scan).



		1	CH is determined by the content in CH[9:0].
--	--	---	---

Note: TXREQ has highest priority, and STNBY has the lowest priority.

Word: SYSTEM2 Address: 01h

Bit 7 (MSB)	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0 (LSB)
swrst	recal	rsvd	st_mo_tx	tc	rdstxrdy	tmout[1]	tmout[0]
wo	wo	wo	wo	wo	wo	wo	wo

Bit	Symbol	Default	Description
7	SWRST	0	Reset all registers to default values:
			0 Keep the current value.
			1 Reset to default values.
6	RECAL	0 (	Reset the state to initial states and recalibrate all blocks:
			0 FSM runs normally.
			Reset the FSM. After this bit is de-asserted, FSM will go through
		>(0)	all the power up and calibration sequence.
5	rsvd	0	Reserved
4	ST_MO_TX	0	TX stereo and mono mode selection:
			0 Stereo
			1 Mono
3	TC	1 🚫	Pre-emphasis time constant: (μs)
			50
			1 75
2	RDSTXRDY	0	RDS transmit ready: Toggle this bit to transmit all 8 bytes in RDS0~RDS7.
		v	The chip will internally fetch these bytes after completing transmit of the
1.0	TD ( 0 X TTE ( 0 )	0.1	current group.
1:0	TMOUT[1:0]	01	Time out setting for IDLE to standby state transition: (min)
			0 0 1
			01 3
			10 5
			1 1 infinity (never)



Word: DEV\_ADD Address: 02h

Bit 7 (MSB)	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0 (LSB)
rsvd	dadd[6]	dadd [5]	dadd [4]	dadd [3]	dadd [2]	dadd [1]	dadd [0]
wo	wo	wo	wo	wo	wo	wo	wo

Bit	Symbol	Default	Description			
7	rsvd	0	Reserved			
6:0	DADD[6:0]	010 1010	Programmed device address when SEB=1: If SEB=0, the default device address (0101011) is used. After power up, if SEB=1, the device address is decided by this register (default 010 1010).			
			SEB Device address:			
			0 0101011			
			1 (DADDR[6:0]			

Word: ANACTL1 Address: 03h

Bit 7 (MSB)	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0 (LSB)
mute_en	i2s_sckinv	rstb_bb	ant_sel	xsel[3]	xsel[2]	xsel[1]	xsel[0]
wo	wo	wo	wo	wo	wo	wo	wo

Bit	Symbol	Default		Description		
7	MUTE_EN	\ Q\\\	TX audio	mute enable:		
			> 0	Un-mute		
			1	Mute		
6	I2S_SCKINV	0	I <sup>2</sup> S MCK	invert:		
			0	Non inverted		
			1 Inverted			
5	RSTB_BB	1	Reset signal of baseband data-path: (Low active)			
			0 Reset			
			1	No action		
4	ANT_SEL	0	Select the	antenna for TX channel scan mode:		
			0	Use the receiver antenna from RFI.		
			1	Use the transmitter antenna on RFO.		
3:0	XSEL[3:0]	1011	Crystal Frequency Selection (MHz):			
			0000	11.2896		
			0001	12		



		<del>-</del>
	0010	12.288
	0011	13
	0100	16.367
	0101	18.414
	0110	19.2
	0111	Reserved
	1000	22.5792
	1001	24
	1010	24.576
	1011	26 (default)
	1100	32.734
	1101	36.828
	1110	38.4
	1111	7.6
	1110	38.4

Word: REG\_VGA Address: 04h

Bit 7 (MSB)	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0 (LSB)
rin[1]	rin[0]	xcsel[5]	xcsel[4]	xcsel[3]	xcsel[2]	xcsel[1]	xcsel[0]
wo	wo	) wo	wo 🕜	wo	wo	wo	wo

Bit	Symbol	Default		Description				
7:6	RIN[1:0]	01	TX mode input impedance for both L/R channels: (kΩ)					
			00	10				
			0 1	20				
			10	40				
			1 1	80				
5:0	XCSEL[5:0]	10 0000	Crystal cap load setting: The loading cap on each side is:					
			10+XC	SEL*0.32 pF, i.e. it ranges from 10pF to 30pF. Default is 20 pF.				



Word: CID1 Address: 05h (RO)

Bit 7 (MSB)	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0 (LSB)
rsvd	rsvd	rsvd	cid1[2]	cid1[1]	cid1[0]	cid2[1]	cid2[0]
ro	ro	ro	ro	ro	ro	ro	ro

Bit	Symbol	Default	Description
7:5	Rsvd	rrr	reserved
4:2	CID1[2:0]	rrr	Chip ID for product family:
		000	000 FM
			001-111 Reserved
1:0	CID2[1:0]	rr	Chip ID for minor revision.
		01	00 1
			01 2
		(	3
			4

Word: CID2

Address: 06h (RO)

Bit 7 (MSB)	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0 (LSB)
cid3[3]	cid3[2]	cid3[1]	cid3[0]	cid4[3]	cid4[2]	cid4[1]	cid4[0]
ro	ro	40	ro	ro	ro	ro	ro

Bit	Symbol	Default		Description	
7:4	CID3[3:0]	rrrr	Chip ID for product ID:		
		0001	0000	Reserved	
			0001	Transmitter – QN8007	
			0010	Reserved	
			0011	Reserved	
			0100	Reserved	
			0101	Transmitter – QN8007L	
			0110-1111	Reserved	
3:0	CID4[3:0]	rrrr	Chip ID for major revision is 1+CID4:		
		0000	0000	1	
			0001	2	



0010	3
0011	4
0100-1111	Reserved

**Word:** I<sup>2</sup>S **Address:** 07h

Bit 7 (MSB)	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0 (LSB)
i2sbw[1]	i2sbw[0]	i2sdrate[1]	i2sdrate[0]	i2smode	i2sfmt[2]	i2sfmt[1]	i2sfmt[0]
wo	wo	wo	wo	wo	wo	wo	wo

Bit	Symbol	Default		Description
7:6	I2SBW[1:0]	01	I <sup>2</sup> S bit width:	
			200	8-bit
			$\langle \langle \langle g \rangle \rangle \rangle$	16-bit
			10	24-bit
			11	32-bit
5:4	I2SDRATE[1:0]	$\rangle \left( \exists I \right) \rangle$	I <sup>2</sup> S data rate:	
		5	00	32kbps
			01	40kbps
			10	44.1kbps
		/	11	48kbps
3	I2SMODE	0	l <sup>2</sup> S mode:	
			0	Slave
			1	Master
2:0	I2SFMT[2:0]	001	I <sup>2</sup> S format:	
			000	MSB justified mode
			001	I <sup>2</sup> S mode
			010	DSP1 mode
			011	DSP2 mode
			100	LSB justified mode
			101-111	Reserved



Word: CH Address: 08h (RW)

Bit 7 (MSB)	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0 (LSB)
ch[7]	ch[6]	ch[5]	ch[4]	ch[3]	ch[2]	ch[1]	ch[0]
rw	rw	rw	rw	rw	rw	rw	rw

Bit	Symbol	Default	Description
7:0	CH[7:0]	1010 0000	Lower 8 bits of 10-bit Channel index: Channel used for TX has two origins, one is from this register and CH[9:8] at 0Bh which can be written by the user, another is from CCA/CCS. CCA/CCS selected channels are stored in an internal register, which is physically different from the CH register, but it can be read out through register CH and be used for TX when CCA_CH_DIS (reg. 00h bit [0]) = 0.

Word: CH\_START Address: 09h

Bit 7 (MSB)	Bit 6	Bit 5	Bit 4	Bít 3	Bit 2	Bit 1	Bit 0 (LSB)
ch_sta[7]	ch_sta[6]	ch_sta[5]	ch_sta[4]	ch_sta[3]	ch_sta[2]	ch_sta[1]	ch_sta[0]
wo	wo	wo	wo	wo	wo	wo	wo
			(2)				

Bit	Symbol	Default	Description
7:0	CH_STA[7:0]	0000 0000	Lower 8 bits of 10-bit CCA (channel scan) start channel index.

Word: CH\_STOP Address: 0Ah

Bit 7 (MSB)	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0 (LSB)
ch_stp[7]	ch_stp[6]	ch_stp[5]	ch_stp[4]	ch_stp[3]	ch_stp[2]	ch_stp[1]	ch_stp[0]
wo	wo	wo	wo	wo	wo	wo	wo

Bit	Symbol	Default	Description
7:0	CH_STP[7:0]	1000 0000	Lower 8 bits of 10-bit CCA (channel scan) stop channel index.



Word: CH\_STEP Address: 0Bh

Bit 7 (MSB)	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0 (LSB)
fstep[1]	fstep[0]	ch_stp[9]	ch_stp[8]	ch_sta[9]	ch_sta[8]	ch[9]	ch[8]
wo	wo	wo	wo	wo	wo	rw	rw

Bit	Symbol	Default	Description
7:6	FSTEP[1:0]	01	CCA (channel scan) frequency step:
			00 50 kHz
			01 100 kHz
			10 200 kHz
			11 Reserved
5:4	CH_STP[9:8]	10	Highest 2 bits of 10-bit CCA(channel scan) stop channel index: Stop freq is (76+CH_STP*0.05) MHz.
3:2	CH_STA[9:8]	00	Highest 2 bits of 10-bit CCA(channel scan) start channel index: Start freq is (76+CH_STA*0.05) MHz.
1:0	CH[9:8]	00	Highest 2 bits of 10-bit channel index: Channel freq is (76+CH*0.05) MHz.

Word: PAC\_TARGET Address: 0Ch

Bit 7 (MSB)	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0 (LSB)
pac_target [7]	pac_target [6]	pac_target [5]	pac_target [4]	pac_target [3]	pac_target [2]	pac_target [1]	pac_target [0]
wo							

Bit	Symbol	Default	Description
7:0	PAC_TARGET [7:0]	1111 1111	PA calibration target value. PA output target is (0.37*PAC_TARGET+68) dBuV. Valid values are 31-131 dBuV.



Word: TXAGC GAIN Address: 0Dh (RW)

Bit 7 (MSB)	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0 (LSB)
tx_sftclpen	tagc_gain_ sel	rsvd	txagc _gdb	txagc_gvga [3]	txagc_gvga [2]	txagc_gvga [1]	txagc_gvga [0]
wo	wo	rw	rw	rw	rw	rw	rw

Bit	Symbol	Default			Description		
7	TX_SFTCLPEN	0	TX soft clip	ping enable:			
			0	Disable			
			1	Enabled		^	
6	TAGC_GAIN_	0	TX AGC G	ain selection m	ethod:		
	SEL		0		digital gain TX		
				buffer gain Tage AGC FSM or	XAGC_GVGA n chip.	[3:0] are deter	mined by TX
				determined by	digital gain and the contents of GA [3:0] respe	of TXAGC_GE	
5	rsvd	Ø///	Reserved		$(\mathcal{P})$		
4	TXAGC_GDB	(0)	TX digital g	ain:			
			0	0 dB			
			1	1 dB			
3:0	TXAGC_GVGA	0001	TX input bu	ffer gain: (dB)			
	[3:0]	N. (	VGAG		RIN	[1:0]	
			[3:0]	00	01	10	11
		140//	0000	4.5	-1.5	-7.5	-13.5
			0001	6	0	-6	-12
		$\searrow$	0010	7.5	1.5	-4.5	-10.5
			0011	9	3	-3	-9
			0100	10.5	4.5	-1.5	-7.5
			0101	12	6	0	-6
			0110	13.5	7.5	1.5	-4.5
			0111	15	9	3	-3
			1000	16.5	10.5	4.5	-1.5
			1001	18	12	6	0
			1010	19.5	13.5	7.5	1.5
			1011	21	15	9	3
			11XX		Rese	erved	



Word: TX\_FDEV Address: 0Eh

Bit 7 (MSB)	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0 (LSB)
tx_fdev[7]	tx_fdev[6]	tx_fdev[5]	tx_fdev[4]	tx_fdev[3]	tx_fdev[2]	tx_fdev[1]	tx_fdev[0]
wo	wo	wo	wo	wo	wo	wo	wo

Bit	Symbol	Default	Description
7:0	TX_FDEV[7:0]	0110 1100	Specify total TX frequency deviation: TX frequency deviation = 0.69 kHz*TX_FEDV.
			TX_FDEV[7:0] Value
			0000 0000 - 11 11 11 0 ~ 255

Word: GAIN\_TXPLT Address: 0Fh

Bit 7 (MSB)	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0 (LSB)
i2sundfl_ clr	i2sovfl_ clr	gain_txpl	gain_txplt [2]	gain_txplt [1]	gain_txplt [0]	rds_int_en	cca_int_en
wo	wo	wo	wo	wo	wo	wo	wo

Bit	Symbol	Default		Description
Dit	Symbol	Delauit		Description
7	I2SUNDFL_	0	I <sup>2</sup> S buffer unde	erflow clear: User has to de-assert this bit after clearing.
	CLR	\\( \cdot \)	0	No action
			1	Clear
6	I2SOVFL_		I <sup>2</sup> S buffer over	flow clear: User has to de-assert this bit after clearing.
	CLR		0	No action
		*	1	Clear
5:2	GAIN_TXPLT [3:0]	10 01	_	ot to adjust pilot frequency deviation: Refer to peak frequency PX signal when audio input is full scale.
			0111	7% * 75kHz
			1000	8% * 75kHz
			1001	9% * 75kHz
			1010	10% * 75kHz
1	RDS_INT_EN	0	output from DO	upt Enable: When RDS_INT_EN=1, a 4.5ms low pulse will be DUT/INT when a new group of data in RDSD0~RDSD7 is internal transmitting buffer after user toggles RDSTXRDY.
			0	Disable
			1	Enable



0	CCA_INT_EN	0		rupt Enable: When CCA_INT_EN=1, a 4.5ms low pulse will DOUT/INT (TX mode) when TXCCA is finished.
			0	Disable
			1	Enable

Word: RDSD0 Address: 10h (RW)

Bit 7 (MSB)	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0 (LSB)
rdsd0[7]	rdsd0[6]	rdsd0[5]	rdsd0[4]	rdsd0[3]	rdsd0[2]	rdsd0[1]	rdsd0[0]
rw	rw	rw	rw	rw	rw	rw	rw

Bit	Symbol	Default	Description
7:0	RDSD0[7:0]	0000 0000	RDS data byte 0: Data written into RDSD0~RDSD7 cannot be read out unless RDSTXRDY is toggled to allow the data to be loaded into the internal transmitting buffer.

Word: RDSD1 Address: 11h (RW)

Bit 7 (MSB)	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0 (LSB)
rdsd1[7]	rdsd1[6]	rdsd1[5]	rdsd1[4]	rdsd1[3]	rdsd1[2]	rdsd1[1]	rdsd1[0]
rw	rw	rw	\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\	rw	rw	rw	rw

Bit	Symbol	Default	Description
7:0	RDSD1[7:0]	0000,0000	RDS data byte 1



Word: RDSD2 Address: 12h (RW)

Bit 7 (MSB)	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0 (LSB)
rdsd2[7]	rdsd2[6]	rdsd2[5]	rdsd2[4]	rdsd2[3]	rdsd2[2]	rdsd2[1]	rdsd2[0]
rw	rw	rw	rw	rw	rw	rw	rw

Bit	Symbol	Default	Description
7:0	RDSD2[7:0]	0000 0000	RDS data byte 2

Word: RDSD3 Address: 13h (RW)

Bit 7 (MSB)	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0 (LSB)
rdsd3[7]	rdsd3[6]	rdsd3[5]	rdsd3[4]	rdsd3[3]	rdsd3[2]	rdsd3[1]	rdsd3[0]
rw	rw	rw	rw	rw	rw	rw	rw

Bit	Symbol	Default	Description
7:0	RDSD3[7:0]	0000 0000	RDS data byte 3

Address: 14h (RW) Word: RDSD4

Bit 7 (MSB)	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0 (LSB)
rdsd4[7]	rdsd4[6]	rdsd4[5]	rdsd4[4]	rdsd4[3]	rdsd4[2]	rdsd4[1]	rdsd4[0]
rw	rw	rw	rw	rw	rw	rw	rw

В	it	Symbol	Default	Description
7:	:0	RDSD4[7:0]	0000 0000	RDS data byte 4



Word: RDSD5 Address: 15h (RW)

Bit 7 (MSB)	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0 (LSB)
rdsd5[7]	rdsd5[6]	rdsd5[5]	rdsd5[4]	rdsd5[3]	rdsd5[2]	rdsd5[1]	rdsd5[0]
rw	rw	rw	rw	rw	rw	rw	rw

Bit	Symbol	Default	Description
7:0	RDSD5[7:0]	0000 0000	RDS data byte 5

Word: RDSD6 Address: 16h (RW)

Bit 7 (MSB)	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2 Bi	Bit 0 (LSB)
rdsd6[7]	rdsd6[6]	rdsd6[5]	rdsd6[4]	rdsd6[3]	rdsd6[2] rdsd	6[1] rdsd6[0]
rw	rw	rw	\\\\rw\	rw	rw r	w rw

Bit	Symbol	Default	Description
7:0	RDSD6[7:0]	0000 0000	RDS data byte 6

Word: RDSD7 Address: 17h (RW)

Bit 7 (MSB)	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0 (LSB)
rdsd7[7]	rdsd7[6]	rdsd7[5]	rdsd7[4]	rdsd7[3]	rdsd7[2]	rdsd7[1]	rdsd7[0]
rw	rw	(IM)	rw	rw	rw	rw	rw

Bit	Symbol	Default	Description	
7:0	RDSD7[7:0]	0000 0000	RDS data byte 7: Writing to this byte will cause the entire group (8 bytes) to be updated into the internal transmitting buffer.	



Word: RDSFDEV Address: 18h

Bit 7 (MSB)	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0 (LSB)
rsvd	rdsfdev[6]	rdsfdev[5]	rdsfdev[4]	rdsfdev[3]	rdsfdev[2]	rdsfdev[1]	rdsfdev[0]
wo	wo	wo	wo	wo	wo	wo	wo

Bit	Symbol	Default	Description
7	rsvd	1	Reserved
6:0	RDSFDEV[6:0]	000 0110	Specify RDS frequency deviation:
			RDS frequency deviation = 0.35 kHz*RDSFDEV.
			RDSFDEV[6:0] Value
			000 0000 - 111 1111 0 - 127

Word: CCA Address: 19h

Bit 7 (MSB)	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0 (LSB)
txccaa[2]	txccaa[1]	txccaa[0]	rsvd	rsvd	rsvd	rsvd	rsvd
wo	wo C	wo	wo	wo	wo	wo	wo

Bit	Symbol	Default	Description
7:5	TXCCAA[2:0]	010	Scaling factor to determine in-band noise power to out of band noise power ratio. Value 0 ~ 7 directly written in, default is 2. When TXCCAA is not zero, valid channels must satisfy the condition "in-band power > TXCCAA * out-of-band power", which usually is set to select channels without adjacent channel interference. When TXCCAA=0, this condition is omitted.
4:0	rsvd	0 1001	Reserved



Word: STATUS1 Address: 1Ah (RO)

Bit 7 (MSB)	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0 (LSB)
rsvd	rsvd	i2sovfl	i2sundfl	insat	rsvd	rsvd	rsvd
ro	ro	ro	ro	ro	ro	ro	ro

Bit	Symbol	Default	Description
7:6	rsvd	rr	Reserved
5	I2SOVFL	r	I <sup>2</sup> S overflow indicator:
			0 No overflow
			1 Overflow
4	I2SUNDFL	r	I <sup>2</sup> S underflow indicator:
			0 No underflow
			1 Underflow
3	INSAT	r	Input level saturation flag:
			No saturation
			1 Input level too high, Channel saturates.
2:0	rsvd		Reserved

Word: STATUS2 Address: 1Bh (RO)

Bit 7 (MSB)	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0 (LSB)
rds_rxtxupd	rsvd						
ro	ro	///10	ro	ro	ro	ro	ro

Bit	value	Symbol	Description	
7	r	RDS_RXTXUPD	RDS RX: RDS received bit will be toggled.	group updated. Each time a new group is received, this
			should toggle the register these bytes after complet	e chip transmitting all the 8 bytes in RDS0~RDS7, user r bit RDSTXRDY. Then the chip internally will fetch ing transmitting of current group. Once the chip bytes, it will toggle this bit.
			If RDS_INT_EN=1, then at the same time this bit is toggled, interrupt output out put a 4.5 ms low pulse.	
			RDS_RXTXUPD	Status



			0->1 or 1->0	A new set (8 Byte) of data is received
			0->0 or 1->1	New data is in receiving
6:0	rsvd	rr	Reserved	

### Word: XLT3 Address: 49h

Bit 7 (MSB)	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0 (LSB)
rsvd	rsvd	rsvd	xtlbyp	rsvd	rsvd	rsvd	rsvd
wo	wo	wo	wo	⟨ wo ∫	wo	wo	wo

Bit	Symbol	Default	Description
7:5	rsvd	rrr	Reserved
4	XTLBYP	0	Direct inject crystal oscillation from external XCLK pin.
			Use internal crystal oscillator.
			Inject external clock from pin XCLK.
3:0	rsvd	> trrt	Reserved

## Word: PAC\_CAL Address: 59h

Bit 7 (MSB)	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0 (LSB)
pac_req	pac_dis	pacap[5]	pacap[4]	pacap[3]	pacap[2]	pacap[1]	pacap[0]
wo	wo	rw	rw	rw	rw	rw	rw

Bit	Symbol	Default	Description		
7	PAC_REQ	0	Manually request PA tuning cap and gain calibration		
			PAC_REQ	Calibration request	
			1	Reset the calibration	
			0	At the 1->0 transition, calibration starts	
6	PAC_DIS	0	Disable PA tuning	g cap calibration and use PACAP as circuit setting	
			PAC_DIS	Status of calibration	
			0	Use calibrated value	
			1	No calibration and use user-set value	



5:0	PACAP[5:0]	000000	User-set PA Tuning cap. Each LSB is 0.3pF. The read back value is the calibration result.
-----	------------	--------	---

## Word: PAG\_CAL Address: 5Ah

Bit 7 (MSB)	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0 (LSB)
rsvd	pag_dis	paipow[1]	paipow[ 0]	pagain[3]	pagain[2]	pagain[1]	pagain[0]
wo	wo	rw	rw	rw	rw	rw	rw

Bit	Symbol	Default	Description		
7	rsvd	r	Reserved		
6	PAG_DIS	0	Disable PA output power calibration and use IPOW PAGAIN as circuit setting		
			PAG_DIS	Status of calibration	
			0 Use calibra	ted value	
			1 No calibrat	ion and use user-set value	
5:4	IPOW[1:0]	00	Set PA current. The read back value is the calibration result.		
			IPOW[1:0]	Current consumption in PA	
			00 4mA		
			01 2mA		
		MO	10 1mA		
			11 0.5mA		
3:0	PAGAIN[3:0]	(m)	Set PAGAIn setting. The read back value is the calibration result.  Transmitter output voltage on the RFO pin (dBuV) is 124dBuV-1.5dB*PAGAIN [3:0].		
			POUT[3:0]	Power w matching	
			0	124	
			1	122.5	
			10	121	
			11	119.5	
			100	118	
			101	116.5	
			110	115	



	111	113.5
	1000	112
	1001	110.5
	1010	109
	1011	107.5
	1100	106
	1101	104.5
	1110	103
	NA O	101.5



## **8 PACKAGE DESCRIPTION**

### 24-Lead plastic Quad Flat, No Lead Package (ML) – 4x4 mm Body [QFN]

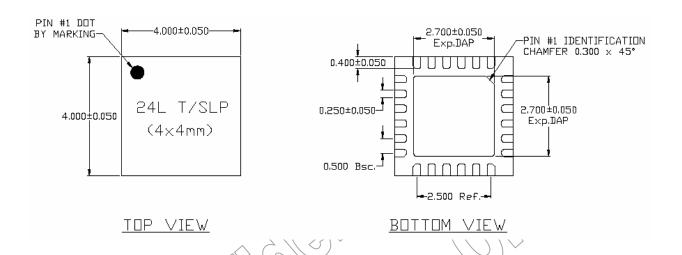


Figure 13: QN8007B/8007BL Device

Units	Millimeters			
<b>Dimension Limits</b>	MIN	NOM	MAX	
Number of pins		24		
Pitch		0.50 BSC		
Overall Height (SLP)	0.80	0.85	0.90	
Standoff	0.00		0.05	
Contact Thickness		0.203 REF		
Overall Width		4.00 BSC		
Exposed Pad Width	2.65	2.70	2.75	
Overall Length		4.00 BSC		
Exposed Pad Length	2.65	2.70	2.75	
Contact Width	0.20	0.25	0.30	
Contact Length	0.35	0.40	0.45	
Contact-to-Exposed Pad	-	0.25	-	

### Notes:

- 1. Pin 1 visual index feature may vary, but must be located within the hatched area.
- 2. Package is saw singulated.
- 3. Dimensioning and tolerance per ASME Y 14.5M.

BSC: Basic Dimension. The theoretically exact value is shown without tolerance.

REF: Reference Dimension, usually without tolerance, for information purpose only.

## **Carrier Tape Dimensions**

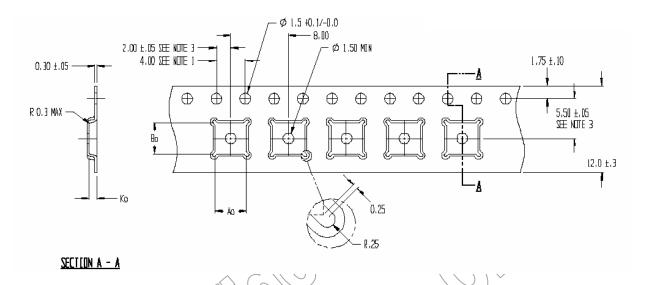


Figure 14: QN8007B/8007BL Carrier Tape Dimensions

### **NOTES:**

- 1. 10 sprocket hole pitch cumulative tolerance  $\pm 0.2$ .
- 2. Camber in compliance with EIA-481.
- 3. Pocket position relative to sprocket hole measured as true position of pocket, not pocket hole.
- 4.  $A_0 = 4.35$  $B_0 = 4.35$   $K_0 = 1.10$
- 5. Reels are shipped in one of two sizes: 2,500 or 5,000 pieces per reel.

## 9 SOLDER REFLOW PROFILE

## 9.1 Package Peak Reflow Temperature

QN800X is assembled in a lead-free QFN24 package. Since the geometrical size of QN800X is 4 mm  $\times$  4 mm  $\times$  0.85 mm, the volume and thickness is in the category of volume<350 mm<sup>3</sup> and thickness<1.6 mm in Table 4-2 of IPC/JEDEC J-STD-020C. The peak reflow temperature is:

$$T_p = 260^{\circ} C$$

The temperature tolerance is  $+0^{\circ}$ C and  $-5^{\circ}$ C. Temperature is measured at the top of the package.

### 9.2 Classification Reflow Profiles

Profile Fe	ature	Specification*	
Average Rar	mp-Up Rate (tsmax to tP)	3°C/second max.	
	Temperature Min (Tsmin)	150°C	
Pre-heat:	Temperature Max (Tsmax)	200°C	
	Time (ts)	60-180 seconds	
Time maintained	Temperature (T <sub>L</sub> )	217°C	
above:	Time (t <sub>L</sub> )	60-150 seconds	
Peak/Classif	ication Temperature (Tp)	260°C	
Time within to	5°C of Actual Reak	20-40 seconds	
Ramp-Down	Rate	6°C/second max.	
Time 25°C to	Peak Temperature	8 minutes max.	

\*Note: All temperatures are measured at the top of the package.



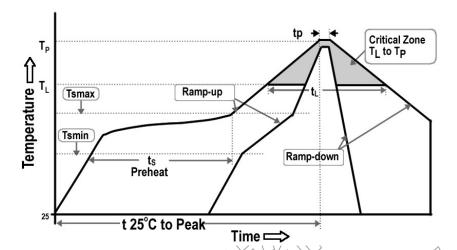


Figure 15: Reflow Temperature Profile

### 9.3 Maximum Reflow Times

All package reliability tests were performed and passed with a pre-condition procedure that repeat a reflow profile, which conforms to the requirements in Section 9.2, three (3) times.

#### **CONTACT INFORMATION**

#### **Quintic Corporation (USA)**

3211 Scott Blvd., Suite 203 Santa Clara, CA 95054 Tel: +1.408.970.8808 Fax: +1.408.970.8829

Email: <a href="mailto:support@quinticcorp.com">support@quinticcorp.com</a>
Web: <a href="mailto:www.quinticcorp.com">www.quinticcorp.com</a>

### **Quintic Microelectronics (China)**

Building 8 B-301A Tsinghua Science Park 1st East Zhongguancun Rd, Haidian Beijing, China 100084

Tel: +86 (10) 8215-1997 Fax: +86 (10) 8215-1570 Web: www.quinticcorp.com

Quintic Microelectronics and Quintic are trademarks of Quintic Corporation. All Rights Reserved.