

SPEC No. | EL136017 ISSUE: June 12 2001 To; SPECIFICATIONS Product Type VIDEO PROCESSING IC FOR THE TFT LCD Model No. RB5P006AM *This specifications contains 52 pages including the cover and appendix. If you have any objections, please contact us before issuing purchasing order. CUSTOMERS ACCEPTANCE DATE: PRESENTED BY: T. Ohno Dept. General Manager REVIEWED BY: PREPARED BY:

Product Development Dept. I

H. Fujita S. Okuda

Analog IC Division

Integrated Circuits Group

SHARP CORPORATION

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1. General Description

The Sharp RB5P006AM is a video processing IC for the TFT color LCD panel using the source driver which power supply voltage is 5V. It converts composite video signals, Y/C separate video signals and RGB signals to the signals suited for the specific requirements of the TFT color LCD panel.

This IC has the serial data control function and it can simplify the external circuit drastically.

Features

- · Low power consumption: 125mW (TYP.)
- · Conforms to the source driver which power supply voltage is 5V.
- Supports composit video signals. Y/C separate video signals and analog RGB signals.
- · Built-in common driver.
- Built-in automatic gain control circuit of Y signal (AGC level adjustment function is available).
- · Built-in image enhancement circuit .
- · Built-in gamma correction circuit.
- · Built-in polarity inverting circuit.
- · Blanking function is available.
- · Contrast and brightness adjustment function is available.
- · White balance adjustment function is available.
- · Power save function is available.
- · Serial data control interface.
- · Built-in 15ch DAC.
- *Not designed or rated as radiation hardened.
- *Package material:
- *Chip material and wafer substrate type:
- *Number of pins and package type:
- *Process (Structure):

Plastic

P type silicon

48-pin quad flat package

(pin pitch 0.75mm)

Bipolar

Applications

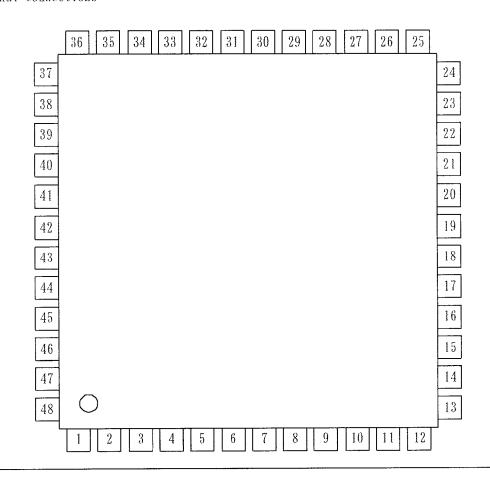
TFT LCD color monitors

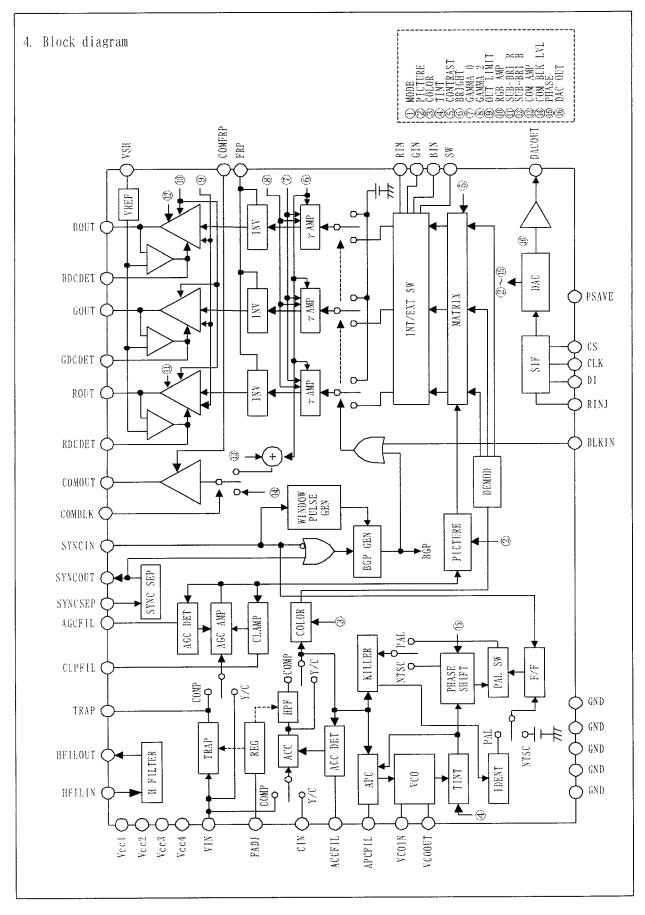
2. Terminal Name

Pin No.	Terminal Name	Pin No.	Terminal Name	Pin No.	Terminal Name
1	CLK	17	FADJ	33	GOUT
2	DI	18	AGCFIL	34	RDCDET
3	RINJ	19	RIN	35	ROUT
4	ACCFIL	20	GIN	36	N. C.
5	CIN	21	BIN	37	Vcc2
6	Vcc1	22	SW	38	Vce3
7	GND	23	BLKIN	39	COMOUT
8	Vcc4	24	VSH	40	GND
9	SYNCSEP	25	VCOIN	41	COMFRP
10	SYNCOUT	26	APCFIL	42	FRP
11	SYNCIN	27	VCOOUT	43	COMBLK
12	GND	28	BDCDET	44	DACOUT
13	VIN	29	BOUT	45	N. C.
14	PSAVE	30	GDCDET	46	HFILOUT
15	TRAP	31	GND	47	HFILIN
16	CLPFIL	32	GND	48	CS

≫N.C.:Non Connection Terminal

3. Terminal Connections





erm. No	Term. name	Voltage	Equivalent Circuit	Description
1	CLK	Threshold L→H: 1.6V H→L: 1.0V	Vec 1 P	This is the input terminal the clock for the serial interface. The serial data accepted at the rising edge of this clock.
2	D1	hystere- sis input	200	This is the input terminal the data for the serial interface.
48	CS		GND	This is the chip select terminal of the serial interface. When this is lot the serial data is acceptable.
3	RINJ	0. 78	Vcc 1 P P P P P P P P P P P P P P P P P P	The value of the resistor connected to this terminal decides the injection curr of the serial interface lo circuit. Connect to GND via 15kΩ.
4	ACCFIL	1.8V (at typical color bar input)	Vcc4 200 54k	This terminal is connected the ACC detection filter.
5	CIN	2. 55V	Vcc4 200 3. 6k	When using the Y/C video signal, input the chromina signal.

ferm. No	Term. name	Voltage	Equivalent Circuit	Description
6	Vcc1			This is the power supply terminal for the circuits except the RGB output circuits, the COM output circuit and the CHROMA signs processing circuits. Connect to the 3V power
7	GND			This is the GND terminal.
8	Vcc4			This is the power supply terminal for the CHROMA signal processing circuits. Connect to the 5V power supply.
9	SYNCSEP	1. 1V	Vec 1	This is the input terminal for the sync separator circuit. Input the signal of the HF1LOUT output. (Refer to the application circuit example)
1 0	SYNCOUT	Vcc t	Vcc 1	This is the output terminal of the composite synchronizing pulse separate by the sync separator circuit. (Active High) Because of the output is provided by an open circuit connect to Vec via IkΩ.
1 1	SYNCIN	Threshold 1. 5V MAX 5. 5V 0V	Vec 1	This is the input terminal the horizontal synchronizin pulse. The rising edge of SYNCIN input pulse must be before the falling edge of SYNCOUT output pulse.
12	GND		- III	This is the GND terminal.

Term. No	Term. name	Voltage	Equivalent Circuit	Description
1 3	VIN	1. 8V	VCC 1 200 GND	This is the input terminal of the composite video signal. (When using the Y/C video signal, input the luminance signal.)
1 4	PSAVE	Threshold 1.5V	Vcc 1 200 100k M	This is the power down control terminal for the signal processing circuits. When this is low, this circuits turn off. Even at this time the memorized data is not initialized. This terminal is internally pulled down to GND at 100kΩ.
15	TRAP	1. 0V	VCC I B B C C C C C C C C C C C C C C C C	This terminal is connected to the TRAP filter. Output impedance:1kΩ
16	CLPFIL	2. 0V	Vec 1 200 GND	This is the connection terminal of the capacitors for the pedestal clamp of luminance signal. Because of the high impedance, use low leakage capacitors.
1 7	FADJ	1. 2V	Vec 1 200 GND	The resistor between this terminal and the ground adjusts the frequency characteristics of the inner filters. Usual resistance:100kΩ Resistance accuracy:±2%, Temp. stability:±200ppm/℃

ferm. No	lerm. name	Voltage	Equivalent Circuit	Description
1. 8	AGCFIL	1. 04V at the input P. 30 SG3 (APL 50%)	25k Vcc 3	This terminal is connected to the AGC detection filter for luminance signal.
19	RIN	1. 8V	Vcc 200	These are the input terminal for analog RGB signals.
2 0	GIN			
2 1	BIN		GND	
2 2	SW	Threshold 1.5V	Vcc 1 200 100k ₹	This terminal selects the input mode as follows: Hi: RGB input mode Lo or Open: composite or Y/C input mode This terminal is internally pulled down to GND at 100kΩ
2 3	BLKIN	Threshold 1.5V	200 100k 100k	This is the input terminal of the blanking pulse. The RGB output signals turn to pedestal level when this terminal is high level. This terminal is internally pulled down to GND at 100kΩ
2 4	VSH		100k = 10c	The voltage of this terminal decides the central voltage of the RGB output. The central voltage of the RGB output is regulated to become half of the voltage of this terminal. Connect to the power supply of the source driver.

Term. No	Term. name	Voltage	Equivalent Circuit	Description
2 5	VCOIN	4. 2V	VCC 4 3 10k 200 200 200 10k 8. 4k 7. 6p 7 8 1. 3k 7 7 6p 7 8 8	This is the input terminal of the VCO circuit.
2 6	APCFIL	2. 3V	Vcc4 200 200 250	This terminal is connected the APC detection filter.
2 7	VCOOUT	2. 7V	Vec 4 200 3. 6k Z GND	This is the output terminal of the VCO circuit.
2 8	BDCDET	1. 8V	Vcc2 Vcc1	These are the connection terminals of the smoothing capacitors of the feedback
3 0	GDCDET		200 \$4k	circuits which control the I levels of the RGB output signals. Connect to GND via
3 4	RDCDET		GND	the capacitor. Use low leakage capacitors.
2 9	BOUT	VSH/2 (central voltage)	Vcc 2	These are the output terminals of the LCD compliant RGB signals.
3 3	GOUT		20 31k 90k	
3 5	ROUT		GND T	

Теги.	Term.	77.74	D : 1 (C: ::	Duranting
No	name	Voltage	Equivalent Circuit	Description
3 1	GND			This is the GND terminal.
3 2	GND			This is the GND terminal.
3 7	Vcc2			This is the power supply terminal for the RGB output circuits. Connect to the 5V power supply.
3 8	Vcc3			This is the power supply terminal for the COM output eircuit. Connect to the 13V power supply.
3 9	COMOUT	Vcc3/2 (central voltage)	Vcc3 5 16k 150k 16k	This is the output terminal of the COM signal. This output can drive a capacitive load up to 12000pF directly.
4 0	GND			This is the GND terminal.
41	COMFRP	Threshold 1.5V	Vcc 1 200 GND 777	This is the input terminal of the polarity control pulse for the COM output. The COM output voltage becomes high when this is low and becomes low when this is high.
4 2	FRP	Threshold 1.5V	WCC1 8	This is the input terminal of the polarity control pulse for the RGB outputs. The RGB outputs is inverted when this is low and not inverted when this is high.

Term. No	lerm. name	Voltage	Equivalent Circuit	Description
4 3	COMBLK	Threshold 1.5V	VCC 1 200 CND 100k	This is the input terminal of the blanking pulse for the COM output. When this is high, the COM output is blanked. This terminal is internally pulled down to GND at 100kΩ.
4 4	DACOUT		Vcc 1 120k	This is the output terminal of the voltage adjusted by the internal DAC (DAC OUT).
4 6	HFILOUT		Vcc1 200 2. 2p	The synchronizing area of the signal applied to Pin 47 for sync separator circuit, is amplified, then pass through the LPF, and then output frothis terminal.
4 7	HFILIN		Vcc1	This is the input terminal for the HF1LTER. Input the same signal applie to Pin 13. The tip of the synchronizing signal is clamped by an external capacitor.

7. Functional Operation

· Video AGC circuit

This circuit makes a gain change corresponding to the APL level of the luminance signal.

This circuit prevents the image from being whitish or being blackish.

• TRAP

In the case of the composite video input, the TRAP rejects the chroma signal from the luminance signal.

The frequency of the TRAP is switched between 3.58MHz (NTSC) and 4.43MHz (PAL). For the Y/C input, the signal does not pass through the TRAP.

· ACC detector, ACC amplifier

The ACC circuit detects the peak of the amplitude of the burst signal to form the ACC loop.

· VCO. APC detector

The local oscillator circuit (VCO) is a pierce type X' tal oscillator circuit. The APC and the VCO form the PLL loop to eliminate the adjustment work.

The APC detector compares the phase of the burst signal with that of the VCO oscillator output, and regulates the oscillation frequency of the VCO.

· RGB inputs

Because the each terminal of the RGB inputs is clamped, the signal is required to be AC coupled.

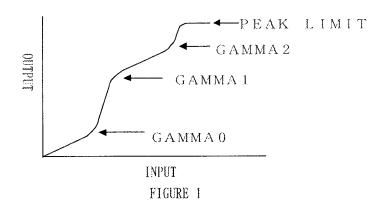
When giving the "Hi" level to the SW input, the RGB inputs are valid, and when giving the "Lo" level to it or opening it, the composite or Y/C input is valid. The RGB inputs accept the analog signals.

· Gamma correction

The relations while the inputted signal and the outputted signal are made nonlinear as the figure 1 to make it cope with the character of the LCD panel. Relative relations with GAMMA1 and GAMMA0 can be adjusted by built—in DAC [GAMMA0]. And relative relations with GAMMA1 and GAMMA2 can be adjusted by built—in DAC [GAMMA2].

The position of PEAKLIMIT follows the position of GAMMA2.

When built-in DAC [BRIGHT] is made to change, the position of all the tune points (GAMMAO, GAMMAI, GAMMA2, PEAKLIMIT) changes at the same time.



· Bright adjustment

When built-in DAC [BRIGHT] is made to change, the gamma compensation point of the RGB output and the voltage of the common output change as the figure 2. The relations between the voltage of the common output and the gamma compensation point can be adjusted by built-in DAC [COM AMPLITUDE].

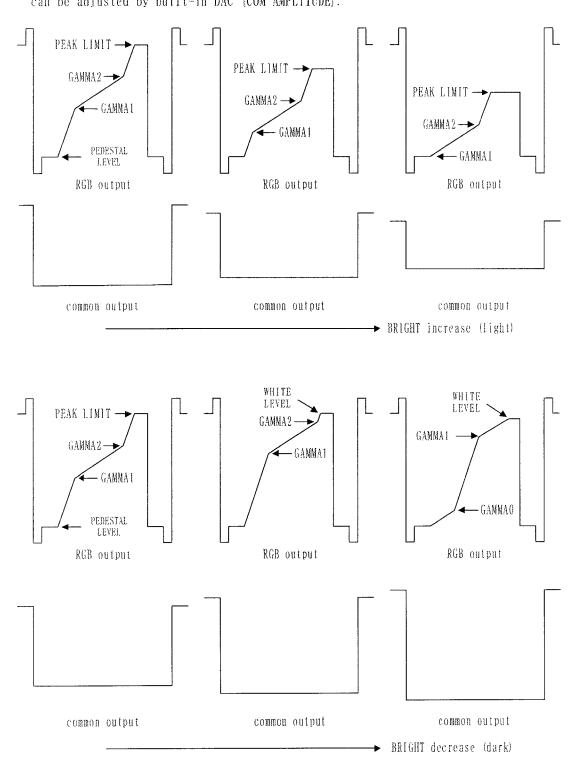
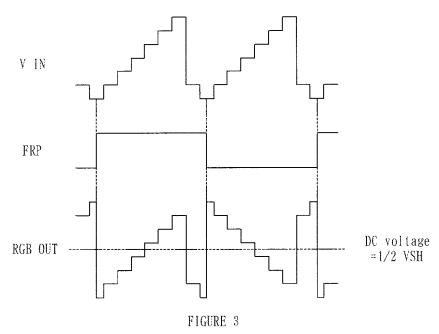


FIGURE 2

RGB outputs

The pulse turned over by every horizontal period is inputted to the FRP terminal, and the polarity of the RGB output is turned over as the figure 3 by every horizontal period. The feedback loop makes the DC voltage of the output signal to 1/2 of the voltage of the VSH terminal.



· Wide screen mode

A black mask territory can be inserted over and under the LCD screen. Make the BLKIN terminal high in the period which wants to make a black mask.

A black mask territory without an influence of built-in DAC[BRIGHT] is made by applying High level to COMBLK terminal at the suitable timing.

About common output

COMOUT output can drive a capacitive load up to 12,000pF directly. In the case of driving a capacitive load more than 12,000pF, connect to the external buffer amplifier.

· Color burst mask function

This IC clamps the pedestal level of the non-reversal period of the RGB outputs at the RGB outputs stage to stabilize the pedestal level of the RGB outputs. But when a color burst pulse is remaining in the clamp period of the image signal, the pedestal level changes by every frame, and there is a possibility that a screen flickers. To avoid this, a blanking pulse is internally generated in the clamp period, and the image signal of the clamp period is masked in the black level.

When it is usually used, a BSTMASK bit is made '1' (the image signal is masked in the black level.

· Power save function

The electric power of the signal processing circuits are cut down by making the PSAVE terminal low.

Even at this time the memorized data on built-in DAC do not change.

· About the internal clamp pulse

Internal clamp pulse, which width is $3\mu s*$ (typically), rises when the SYNC OUT pulse falls. The clamp pulse is able to generate during the WINDOW pulse is "Hi", that is $6\mu s*$ (typically) after SYNC IN pulse rises (Figure 4).

· The timing of the SYNC IN pulse

The SYNC IN pulse must be synchronized with HD.

Make sure the rising edge of SYNC IN is not behind the falling edge of SYNC OUT, because the black level might not be provided precisely if the picture area of the video signal is clamped (Figure 4).

Make sure the falling edge of SYNC IN does not cover the end of the picture area of the video signal, to avoid vertical lines on the LCD panels (Figure 4).

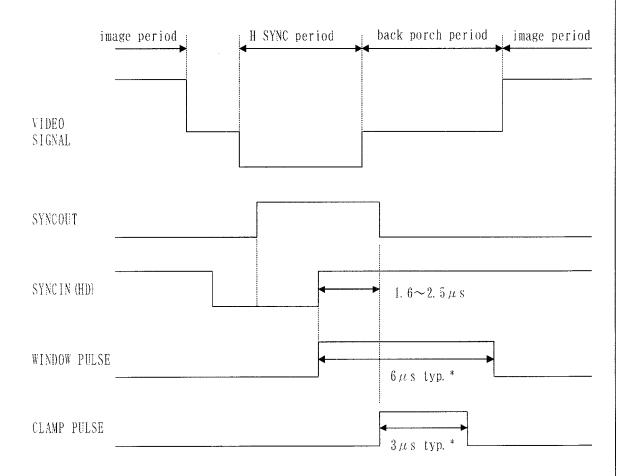


FIGURE 4

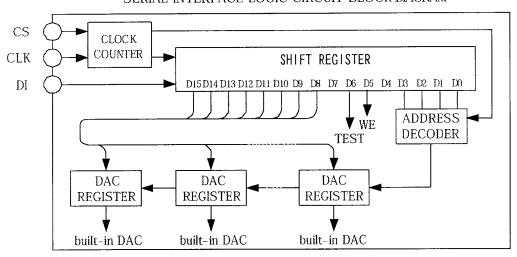
* It is the value when the external resistance of $100k\,\Omega$ is connected with the FADJ terminal.

· Serial interface logic circuit

The serial interface logic circuit is composed by the I²L logic circuit that decodes and memorizes the serial data, and the D/A converter circuit which gives each department an adjustment voltage.

When the power is turn on, the value of the register is initialized (refer P. 17).

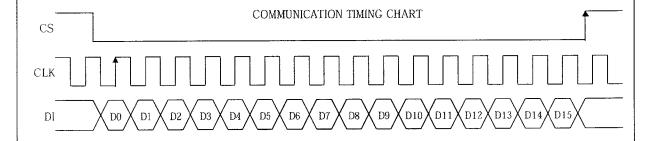
SERIAL INTERFACE LOGIC CIRCUIT BLOCK DIAGRAM



(1) Serial communication timing

A DI signal is taken in the shift register at the timing when a CLK signal stands up. The data taken in the shift register are transferred to the DAC register at the timing when a CS signal stands up.

The data are abandoned when the data taken in the shift register are less than 16 bits. When the data exceeds 16 bits, the last 16 bits taken in will be effective.



2 Construction of the serial data

The construction of the serial data which input DI terminal, is shown in the following.

First															Last
LSB															MSB
D.0	D 1	D2	D3	D4	D5	D6	D7	D8	D9	D10	D11	D12	D13	D14	D15
REG	ISTER	ADDR	ESS	*]	*2	*3	*1				DA	TA			

	۱ ۵	n n	nne	C			THE ACTION BUILDIN WALTE	X/ A	T TI	D 1	N	DEG	e na	ייף ד	NG
REGISTER NAME	\vdash		RES	-	RE	ESOLUTION	THE ACTION WHEN VALUE		LU	E J	[]N	RE.		_	
	1)()	1) [D2	1)3			INCREASED	U.	8	• • • •	• • • •			·])]	<u> </u>
INPUT MODE/SW	0	0	0	0_	6	$(D8 \sim D14)$	※ 1	0	0	1	0	1	0		
PICTURE	1	0	0	0	6	(D8~D13)	PICTURE → SHARP	0	0	0	0	0	1	_	_
COLOR	0	1	0	0	8	$(D8 \sim D15)$	$COLOR \rightarrow DEEP$	0	0	1	0	1	1	1	0
PHASE	1	1	0	0	6	(D8~D13)	R-Y/B-Y PHASE → LARGE	0	0	0	0	0	1	-	_
CONTRAST	0	0	1	0	8	(D8~D15)	$CONTRAST \rightarrow STRONG$	0	0	0	1	1	0	0	1
BRIGHT	1	0	1	0	8	(D8~D15)	BRIGHTNESS → LIGHT	0	0	0	0	0	0	0	1
GAMMA 0	0	1	1	0	8	(D8~D15)	SEPARATE FROM $\gamma 1 ightarrow LARGE$	0	0	0	()	1	0	1	0
GAMMA 2	1	1	1	0	8	(D8~D15)	SEPARATE FROM $\gamma 1 \rightarrow LARGE$	0	0	0	0	0	0	1	0
OUT LIMIT	0	0	0	1	5	(D8~D12)	AMPLITUDE → SMALL	0	1	0	0	1	_	-	-
RGB AMPLITUDE	1	0	0	1	8	(D8~D15)	AMPLITUDE → LARGE	0	0	0	0	0	1	0	1
SUB-BRIGHT R	0	1	0	1	8	(D8~D15)	RED AMPLITUDE→LARGE	0	0	0	0	0	0	0	1
SUB-BRIGHT B	1	1	0	1	8	(D8~D15)	BLUE AMPLITUDE→LARGE	0	0	0	0	0	0	0	1
COM AMPLITUDE	0	0	1	1	8	(D8~D15)	AMPLITUDE → LARGE	0	0	0	0	1	0	0	1
COM BLACK LEV	1	0	1	1	5	(D8~D12)	AMPLITUDE → SMALL	0	0	0	0	1	_	_	-
DAC OUT	0	1	1	1	8	(D8~D15)	VOLTAGE → HIGH	0	0	0	0	0	0	0	1
TINT	1	1	1		8	(D8~D15)	TINT → GREEN	0	0	0	0	0	0	0	1

<u> </u>	CONSTRUCTION	OF THE INPUT MODE/SW REGISTER
bit	FUNCTION NAME	EXPLANATION
D8	XCOP	Input mode choice. 0:COMPOSIT VIDEO INPUT 1:Y/C VIDEO INPUT
D9	PNS	PAL / NTSC mode choice. 0:NTSC 1:PAL
D10	AGCLO	AGC level set
D11	AGCL1	AGC level set
D12	BSTMASK	COLOR BURST MASKING function enable. 0:OFF 1:ON
D13	GAMTEST	TEST function for gamma adjustment enable. When this is turned on, an image signal is restricted in a gamma 1 point. 0:0FF 1:0N
014	(Don't care)	
D15	(Don't care)	

· About the DAC test bit (D6) and the writing permission bit (D5).

When a D6 bit is '1', the voltage set up in the register chosen with D0-D3 is outputted in the DACOUT terminal.

When a D5 bit is '0', data are not written in the register.

The voltage set up in the register can be confirmed without changing the value of the register when a D5 bit is made $^{\prime}$ 0 $^{\prime}$ and a D6 bit is made $^{\prime}$ 1 $^{\prime}$.

When the value of the register is 0, the voltage outputted in the DACOUT terminal rises most. But, as for the DACOUT register, the lowest voltage is outputted when the value is 0.

When a D6 bit is '0', the voltage set up in the register is outputted in the DACOUT terminal.

When it is usually used, a D5 bit is made '1', and a D6 bit is made '0'.

- 7. Precautions
 - Power supply pins
 All GND terminals (Pin 7, 12, 31, 32, and 40) must be at the same potential.
 Every wiring patterns from a bypass capacitor to Vcc or GND must be thick and short as much as possible.
 - *RINJ terminal A long, noise sensitive wiring of RINJ terminal may cause a malfunction at maximum clock rate of the serial interface. Connect 4.7k Ω resistor and 0.1 μ F capacitor in series between RINJ and GND (parallel connected to 15k Ω resistor of RINJ) to avoid this situation.
- ·White balance adjustment
 Adjusting built-in DAC [SUB-BRIGHT R] [SUB-BRIGHT B] or other white balance for each every product is recommended. Otherwise, the white balance of each every product may be uneven because of the deviation of the electrical parts.
- COM amplitude adjustment
 Adjusting built—in DAC [COM AMPLITUDE] for every product is recommended.
 Otherwise, the brightness of each product may be uneven because of the deviation of the amplitude of the COMOUT output.
- RGB output amplitude
 Do not set the black limit level of the non-inverted RGB output under 0.2V by adjusting built-in DAC[RGB AMPLITUDE] [OUT LIMIT]. If the minimum voltage level becomes under 0.2V, the waveform on inversion and non-inversion become asymmetric by transistors' saturation, which results central voltage stabilize circuit or other circuit operates wrong.
- Bright adjustable range (GAMMAO variable range)
 In the case of varying the built-in DAC [BRIGHT], make the GAMMAO point of
 the RGB output be below IV above the pedestal level, to avoid clamp failed.
- Input of the signal Synchronize all the input signals connected to the IC.
- · Input signals that are unusual

 Be careful to input signals which luminance changes each horizontal period (for instance, signals that repeats black level invert output and white level non-invert output), because irregular output might occur. This irregular output may be eased by increasing the capacitance connected to the DCDET terminals. Although it will take more time to stabilize the screen after the power is on.
- Sync separator input
 Make the impedance of the signal source connected to the sync-separator input as low as possible. Otherwise, the ability of the sync-separator may decrease.
- FADI terminal (pin 17)
 Because the impedance of this terminal is high, the crosstalk with neighboring signal lines may occur. Therefore, place the resistor connected to this terminal by the IC and design the PWB patterns in order to minimize the crosstalk.
- SYNCIN (pin 11)
 The horizontal synchronizing pulse synchronized with the output pulse of SYNCOUT (pin 10) must be applied to pin 11.
- POWER SAVE When PSAVE is on, make sure a current doesn't flow into IC from the RGB outputs.
- Circuits surrounding VCO
 Performance of the X'tal oscillator required is as follows.
 a. Load capacitance 16pF
 b. Frequency tolarance ±30ppm
 c. Temperature stability ±30ppm
 d. Frequency characteristics comparable to AT-49 of DAISHINKU CORPORATION Wiring between VCOIN (terminal 25) and X'tal oscillator should be made to keep the capacitive effect to minimum.
- Crosstalk between inputs A high frequency signal should not input into the input terminal of the side which is not chosen with SW terminal because the cross talk waveform of RGB outputs may appear by leak lump of the signal from RGB inputs side at the time of composite video input mode selection, and from composite video input side at the time of RGB inputs mode selection.

8. Absolute Maximum Ratings

Parameter	Symbol	Condition	Rating	Unit
	Veci		-0.3 ~ 4	V
Supply voltage	Vcc2		-0.3 ~ 7	V
Subbil Antrage	Vecs		-0. 3 ∼ 16. 5	V
	Vcc4		-0.3 ~ 7	V
Power dissipation	PD	Ta≦25℃	725	mW
Derating ratio	$\Delta P_{\rm D}$	Ta>25℃	5. 8	mW/℃
Operating temperature range	Topr		-30 ∼ 85	r
Storage temperature range	Tstg		-55 ∼ 150	C
SYNC IN input signal voltage		7	GND-0.3 ∼ 6	V
SYNC OUT output withstand voltage			7	V
COM FRP input signal voltage			GND-0.3 ∼ 6	V
FRP input signal voltage			GND-0.3 ∼ 6	V
COM BLK input signal voltage			GNI)-0. 3 ∼ 6	V
BLACK IN input signal voltage			GND-0.3 ∼ 6	V
CLK, D1. CS input signal voltage			GND-0.3 ∼ 6	V
VSH input signal voltage			GND-0.3 ∼ Vec2±0.3	V

9. Recommended Operating Conditions

Parameter	Crembol	Condition		Tim: +		
rarameter	Symbol Condition		MIN.	TYP.	MAX.	Unit
	Vcci		2. 7	3	3. 6	V
Operating supply	Vcc2		4. 5	5	5. 5	V
power voltage	Vссз		11	13	15. 5	V
	Vcc4		4. 5	5	5. 5	V
Composite video signal input voltage	$V_{\mathtt{CMP}}$	B/W amplitude at 100% color bar input	0	0. 35	0. 45	VP-P
Y signal input voltage	V_{γ}	B/W amplitude at 100% color bar input	0	0. 35	0. 45	VP-P
C signal input voltage	Vc	Burst amplitude at 100% color bar input	0	0. 15	0. 3	VP-P
RGB signal input voltage	VRGB	B/W amplitude at 100% color bar input	0	0. 7	1	VP-P
VSH terminal voltage	VSH		2. 7		Vcc2	V

10. Electrical Characteristics

DC Characteristics

Unless otherwise specified: Ta=25°C, Vcc1=3V, Vcc2=Vcc4=5V, Vcc3=13V, VSH=5V

Parameter	Symbol	Condition	MIN.	TYP.	MAX.	Unit
Current supply 1	1001	3V power supply current, Vcc1.		21	30	mA
Current supply 2	ICC 2	5V power supply current, Vcc2.		4. 5	6. 6	mΛ
Current supply 3	1003	13V power supply current, Vcc3.		0. 3	0. 5	mΑ
Current supply 4	1004	5V power supply current, Vcc4.		7. 0	10. 2	mA
Current supply 1 (At the signal processor is powered down)	IPSP1	P SAVE=0, 3V power supply current.		4. 9	7. 2	mΛ
Current supply 2 (At the signal processor is powered down)	IPSP2	P SAVE=0. 5V power supply current.		39	65	μΛ
Current supply 3 (At the signal processor is powered down)	IPSP3	P SAVE=0, 13V power supply current.		26	43	μΑ
Current supply 4 (At the signal processor is powered down)	IPSP4	P SAVE=0, 5V power supply current.		93	155	μΑ
RINJ terminal voltage	V3		0. 4	0.7	1. 0	V
CIN terminal voltage	V5		2. 3	2. 6	2. 9	V
VIN terminal voltage	V13		1. 5	1.8	2. 1	V
FADJ terminal voltage	V17		0. 9	1. 2	1. 5	V
RIN terminal voltage	V19		1, 5	1. 8	2. 1	V
GIN terminal voltage	V20		1. 5	1. 8	2. 1	V
BIN terminal voltage	V21		1. 5	1.8	2. 1	V
BDCDET terminal voltage	V28		1. 5	1.8	2. 1	V
GDCDET terminal voltage	V30		1. 5	1. 8	2. 1	V
RDCDET terminal voltage	V34		1. 5	1. 8	2. 1	V
HFILOUT terminal voltage	V46		1. 88	2. 18	2. 48	V
CLK terminal current	IL1	CLK = 0V.		-0.8	-4. ()	μ A
DI terminal current	IL2	DI = 0V.		-0. 8	-4. 0	μA
SYNCIN terminal current	IL11	SYNCIN = 0V.		-0.4	-1. 0	μΑ
APCFIL terminal current	IL26	APCFIL = 0V.		-0. 1	-1.0	μ A
COMFRP terminal current	IL41	COMFRP = OV.		-0. 1	-1.0	μΑ
FRP terminal current	IL42	FRP = OV.		-0. 1	-1. 0	μ A
CS terminal current	IL48	CS = 0V.		-0. 8	-4. 0	μ A
PSAVE input impedance	Z14		70	100	130	kΩ
TRAP input impedance	Z15		0. 7	1	1. 3	kΩ
SW input impedance	Z22		70	100	130	kΩ
BLKIN input impedance	Z23		70	100	130	kΩ
COMBLK input impedance	Z43		70	100	130	kΩ

AC characteristics

Unless otherwise specified: Ta=25%, Vcc1=3V, Vcc2=Vcc4=5V, Vcc3=13V. TP11=SG1. TP22=0V. TP41=TP42=SG2. $SW4\to 0N$, $SW5\to a$. $SW9\to 0N$, $SW10\to 0N$, $SW10\to 0N$, $SW20\to 0N$, the other is OFF. DAC setting is CONTRAST=70h, BRIGHT=00h, GAMMA0=GAMMA2=FFh, OUT LIMIT=00h, COLOR=58h, and RGB AMPLITUDE=FFh, the other is preset value. ('h' means hexadecimal). The sensitivity of each DAC is measured at the center of setting range.

Parameter		Symbol	Conditions	MIN.	TYP.	MAX.	Unit
	AGC LEV0	VA10 VA20	INPUT MODE/SW=10h, RGBAMP=FFh, TP13=SG3. Measure the amplitude of TP33 from the pedestal level to the fourth step.	1. 5	2. 0 1. 6	2. 6 2. 1	
AGC AGC LEVI	VA30 VA11 VA21 VA31	VA10:APL10%, VA20:APL50%, VA30:APL90% INPUT MODE/SW=14h. Measuring condition is same as AGC LEVO.	1. 0 2. 0 1. 6	1. 4 2. 7 2. 2 1. 5	1. 9 3. 6 2. 9 2. 0		
characteristics	AGC LEV2	VA12 VA22 VA32	INPUT MODE/SW=18h. Measuring condition is same as AGC LEVO.	2. 5 2. 1 1. 3	3. 4 2. 9 1. 8	4. 4 3. 8 2. 4	VP-P
	AGC LEV3	VA13 VA23 VA33	INPUT MODE/SW=1Ch. Measuring condition is same as AGC LEVO.	2. 9 2. 7 1. 6	3. 9 3. 6 2. 2	4. 7 4. 6 2. 9	
	NTSC	GpCN1	PICTURE=00h, TP13=SG4, SW33→0N. Sign wave amplitude ratio of TP33 (loaded 100pF) when SG4 is 100kHz to 2.0MHz.	-11	-7	-3	
Image quality control variable PAI	PAL	GpCP1	PICTURE=3Fh INPUT_MODE/SW=16h, PICTURE=00h	2 -9	6 -5	10 -1	₫₿
range	Y/C	GpCP2 GpYC1	INPUT MODE/SW=16h, PICTURE=3Fh INPUT MODE/SW=15h, PICTURE=00h, TP13=SG4, SW33→0N. Sign wave amplitude ratio of TP33 (loaded 100pF) when SG4 is 100kHz to 2.5MHz.	-4	0	12 4	
		GpYC2	INPUT MODE/SW=15h, PICTURE=3Fh	1()	14	18	
Color control var	iable	Gc11	TP13=SG9 (3.58MHz, OdB, burst/chroma phase=180°). The amplitude ratio of COLOR=00h to COLOR=58h at TP29.		-30	-20	dB
		Gc12	CONTRAST=3Ch.The amplitude ratio of COLOR=FFh to COLOR=58h at TP29.	10	13		
Y maximum gain		GMY	CONTRAST=FFh, TP13=SG5. Calculate gain from the amplitude of TP33.	34	37	40	dB
RGB maximum gain		GMRGB	TP22=3V, TP21, 20, 19=SG5. Calculate gain from the amplitude of TP29, 33, 35.	21	24	27	dB
Contrast aftenuat ratio	.е	Gcon	TP13=SG5. The amplitude ratio of TP33 at CONTRAST=00h to CONTRAST=FFh.		-30	-20	dB
DC reproduction r	atio	К	TP13=SG3 (APL10%, 90%). Measure the amplitude (black-black) of TP33 at 10% and 90%. Define the each value as V1 and V2. K=(V1-(V1-V2))/V1*100	95	100		%

Parameter	Symbol	Conditions	MIN.	TYP.	MAX.	Unit
Trap attenuation ratio (NTSC)	GefN	TP13=SG4(100kHz, 3.58MHz). Define the amplitude of TP15 at 100kHz as VI. Define the amplitude of TP15 at 3.58MHz as V2. GcfN=20log(V2/V1)	AU LIV.	-45	-30	dB
Trap attenuation ratio (PAL)	GefP	INPUT MODE/SW=16h, TP13=SG4 (100kHz, 4.43MHz). Measure the amplitude of TP15 at 100kHz and 4.43M. Define each value as V3 and V4. GcfP=20log(V4/V3)		-45	-30	dВ
ACC characteristic(1) (Composite NTSC in)	GA1NV	TP13=SG9 (0dB, +6dB, -25dB). Measure the amplitude of TP29 at 0dB, +6dB, -25dB.		()	2. 0	dB
ACC characteristic(2) (Composite NTSC in)	GA2NV	Define the each value as V0, V1 and V2. GA1NV=20log (V1/V0) GA2NV=20log (V2/V0)	-15. 0	-10. 0		dB
ACC characteristic (3) (Y/C NTSC in)	GAINY	INPUT MODE/SW=15h, TP5=SG9 (OdB, +6dB, -25dB). Measure the		0	2. 0	dB
ACC characteristic (4) (Y/C NTSC in)	GA2NY	amplitude of TP29 at OdB, +6dB, -25dB. Define the each value as V0.V1 and V2. GA1NY=20log(V1/V0) GA2NY=20log(V2/V0)	-9. 0	-4. 0		dB
ACC characteristic (5) (composite PAL in)	GAIPV	INPUT MODE/SW=16h. TP13=SG9 (OdB, +6dB, -25dB). Measure the		0	2. 0	dB
ACC characteristic(6) (composite PAL in)	GA2PV	amplitude of TP29 at OdB, +6dB, -25dB. Define the each value as V0, V1 and V2. GA1PV=20log(V1/V0) GA2PV=20log(V2/V0)	-17. 0	-12. 0		dB
ACC characteristic(7) (Y/C PAL in)	GA1PY	INPUT MODE/SW=17h, TP5=SG9 (OdB, +6dB, -25dB). Measure the		0	2. 0	dB
ACC characteristic(8) (Y/C PAL in)	GA2PY	amplitude of TP29 at OdB, +6dB, -25dB. Define the each value as V0, V1 and V2. GA1PY=20log(V1/V0) GA2PY=20log(V2/V0)	-11. ()	-6. 0		dB
APC capture range (NTSC, upper side)	fN+	INPUT MODE/SW=15h, TP5=SG9. Decrease the frequency from 3.584545MHz until output of TP29 is appeared. Work out the difference between the frequency at this time and 3.579545MHz.		+1600		Hz
APC capture range (NTSC, lower side)	fN-	INPUT MODE/SW=15h, TP5=SG9. Increase the frequency from 3.574545MHz until output of TP29 is appeared. Work out the difference between the frequency at this time and 3.579545MHz.		-1600	-500	Нz
APC capture range (PAL, upper side)	ſP+	INPUT MODE/SW=17h, TP5=SG9. Decrease the frequency from 4.438619MHz until output of TP29 is appeared. Work out the difference between the frequency at this time and 4.433619MHz.		+1800		Hz
APC capture range (PAL. lower side)	fP-	INPUT MODE/SW=17h, TP5=SG9. Increase the frequency from 4.428619MHz until output of TP29 is appeared. Work out the difference between the frequency at this time and 4.433619MHz.		-1800	-500	Hz

Parameter	Symbol	Conditions	MIN.	TYP.	MAX.	Unit
Killer operating	Vbk1A	INPUT MODE/SW=15h, TP5=SG9. Observe the waveform of TP29. Decrease the burst amplitude from 150mVp-p until the		-43	-38	dB
(NTSC)	Vbk1B	killer is turned on. Define the burst amplitude as VKIL1(mVp-p). Then, increase the burst amplitude from that			-35	ub
Killer operating input level	Vbk2A	situation until the killer is turned off. Define the burst amplitude as VKIL2(mVp-p). Vbk1A=20log(VKIL1/150). Vbk1B=20log(VKIL2/150).		-47	-42	dB
(PAL)	Vbk2B	PAL mode: INPUT MODE/SW=17h. Measuring condition is same as NTSC.			-39	άĐ
Killer color ghost (NTSC)	Vbs1	Measuring condition is same as Vbk1A and Vbk2A (the killer is turned on).		50	100	mV
Killer color ghost (PAL)	Vbs2	Measure the amplitude of the color difference output of TP29.		50	100	mV
Demodulation output amplitude ratio (1) (NTSC)	<u>R-Y</u> В-Y	INPUT MODE/SW=15h, TP5=SG9. Vary the chroma phase and observe the each amplitude of TP29, TP33 and TP35. Define the each maximum amplitude of TP29, TP33 and TP35 as VB, VG and VR. (R-Y) / (B-Y) = VR/VB, (G-Y) / (B-Y) = VG/VB	0. 56	0. 66	0. 76	
Demodulation output amplitude ratio (2) (NTSC)	<u>G-Y</u> B-Y		0. 29	0. 36	0. 44	
Demodulation output amplitude ratio (3) (PAL)	<u>R-Y</u> B-Y	INPUT MODE/SW=17h, TP5=SG9. Vary the chroma phase and observe the each amplitude of TP29, TP33 and TP35.	0. 56	0. 66	0. 76	
Demodulation output amplitude ratio (4) (PAL)	<u>G-Y</u> B-Y	Define the each maximum amplitude of TP29, TP33 and TP35 as VB, VG and VR. $(R-Y) / (B-Y) = VR/VB$, $(G-Y) / (B-Y) = VG/VB$	0. 29	0. 36	0. 44	
Demodulation relative phase (1) (NTSC)	QRB	INPUT MODE/SW=15h, TP5=SG9. Vary the chroma phase and observe the each amplitude and phase of TP29. TP33 and	80	90	100	deg
Demodulation relative phase (2) (NTSC)	QGB	TP35. Define the phase causing the maximum amplitude of TP29 as θ B. of TP33 as θ G, of TP35 as θ R. QRB= θ R- θ B, QGB= θ G- θ B	230	240	250	deg
Demodulation relative phase (3) (PAL)	QRB	INPUT MODE/SW=17h, TP5=SG9. Vary the chroma phase and observe the each amplitude and phase of TP29, TP33 and	80	90	100	deg
Demodulation relative phase (4) (PAL)	QGB	TP35. Define the phase causing the maximum amplitude of TP29 as θ B. of TP33 as θ G, of TP35 as θ R. QRB= θ R- θ B, QGB= θ G- θ B		240	250	deg

Parameter	Symbol	Conditions	MIN.	TYP.	MAX.	Unit
Demodulation output residual carrier (NTSC)	VCRN	TP42=3V, INPUT MODE/SW=15h, TP5=SG9. Adjust the chroma phase for maximizing the amplitude of TP29. Measure the ratio of the 7.15909MHz component to the 15.737kHz component with SPECTRUM ANALYZER.		-40	-30	dB
Demodulation output residual carrier (PAL)	VCRP	TP42=3V, INPUT MODE/SW=17h, TP5=SG9. Adjust the chroma phase for maximizing the amplitude of TP29. Measure the ratio of the 8.867238MHz component to the 15.625kHz component with SPECTRUM ANALYZER.		-50	-40	dB
TINT variable range	Q+	INPUT MODE/SW=15h, TP5=SG9 (NTSC). With T1NT=80h (preset value), define the chroma phase causing the maximum amplitude of TP29 as 00. Define the	40	55	70	deg
	Q	each chroma phase causing the maximum amplitude of TP29 at TINT=00h and TINT=FFh as θ 1 and θ 2. Q+= θ 1- θ 0, Q-= θ 2- θ 0	-60	-45	-30	deg
	PQ∔	INPUT MODE/SW=15h, TP5=SG9 (NTSC). With PHASE=20h (preset value), define the chroma phase causing the maximum amplitude of TP29 as \$\theta\$0. Define the	13	19	25	deg
PHASE variable range	PQ-	each chroma phase causing the maximum amplitude of TP29 at PHASE=00h and PHASE=3Fh as θ 1 and θ 2. PQ+= θ 1- θ 0, PQ-= θ 2- θ 0		-16	-10	deg
Frequency characteristics (Y/C input)	fycl	SW29, 33, 35→0N, INPUT MODE/SW=15h, PICTURE=00h, TP13=SG4 (amplitude is 20mV). Measure the frequency when the amplitude of TP29, 33, 35 (loaded 100pF) decreases 3dB compared to the amplitude of it at the frequency 100kHz.	2. 0	4. 0		MH2
	fyc2	PICTURE=3Fh. Measuring condition is same as fyc1.	6. 0	8. 0		
Frequency characteristics (RGB input)	frgb	SW29, 33, 35→0N. TP22=3V TP21, 20, 19=SG4 (amplitude is 50mV). Measure the frequency when the amplitude of TP29, 33, 35 (loaded 100pF) decreases 3dB compared to the amplitude of it at the frequency 100kHz.	4. 5	6. 0		MHz

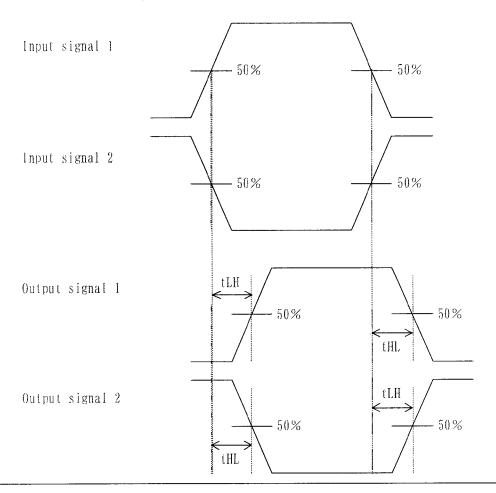
Parameter	Symbol	Conditions	MIN.	TYP.	MAX.	Unit
	Δγ1	TP22=3V, TP21, 20, 19=SG7 (0, 5VP-P). Compute the gain at TP29, 33, 35. BRIGHT=FFh, GAMMA2=FFh when measure Δ γ 2.	21	24	27	
	Δγ2	BRIGHT=AFh, GAMMA2=00h when measure $\Delta \gamma 3$, VW $\gamma 2$. GAMMA0=00h when measure $\Delta \gamma 0$.	0. 5	3. 5	6. 5	.tn
Gamma correction characteristics	Δγ3	VW γ 2 is voltage between γ 2 and PEAK LIMIT.	18. 5	21. 5	24. 5	d₿
	Δγθ	$\Delta r3$ $\uparrow vwr2$ $\Delta r2$	-2	0. 5	3	
	VW γ 2	Δγ1	0. 6	0. 9	1. 4	V
RGB AMP variable range	VRGB1	TP22=3V, RGB AMPLITUDE=00h. Measure the amplitude between black level of TP33 output.	1. 9	2. 2	2. 5	VP-P
	VRGB2	TP22=3V, RGB AMPLITUDE=FFh	4. 3	4. 6	4. 9	
RGB AMP sensitivity	△ VRGB	The amplitude varying per 1LSB.	7	10	13	mVP-P
SUB-BRIGHT variable	ΔVSB1	TP22=3V, TP21, 20, 19=SG5 ("Hi"level=100mV"). Measure the amplitude varying between white levels of TP29, 35 when SUB-BRIGHT R, B=00h based on 80h.	1. 2	1. 5	1.8	V
range	ΔVSB2	Measure the amplitude varying between white levels of TP29,35 when SUB-BRIGHT R,B=FFh based on 80h.	-1. 8	-1. 5	-1. 2	
SUB-BRIGHT sensitivity	Δ VSB	The amplitude varying per 1LSB.	9	12	16	mV
OUT LIMIT variable range	VBL1	OUT LIMIT=00h, RGB AMPLITUDE=00h, TP22=3V, TP20=SG6 (amplitude is 100mV). Measure non-inverted output voltage of TP33 between pedestal level and black limit level.	0. 45	0. 65	0. 85	V
	VBL2	OUT LIMIT=1Fh		0.0	0. 2	
OUT LIMIT sensitivity Gain difference among RGB	Δ VBL Δ GRGB	The amplitude varying per ILSB. TP22=3V, TP21, 20, 19=SG7 (50mVp-p). Measure the ratio of maximum to minimum amplitude (white-black) among TP29, 33, 35 non-inverted output.	-0. 8	20 0	0.8	mV dB
Gain difference between inverted and non-inverted	ΔGINV	Measure the ratio of the amplitude (white-black) of inverted to non- inverted output of TP29, 33, 35. Measuring condition is same as ΔGRGB.	-0. 7	0	0. 7	dB
Difference among RGB output black level	ΔVBKL	TP22=3V. Voltage difference of pedestal level among TP29, 33, 35.	-300	0	300	mV
DC voltage of RGB output	Vc	DC voltage of TP29, 33, 35 output.	2. 3	2. 5	2. 7	V
DC voltage difference among RGB output	ΔVc	DC voltage difference among TP29, 33, 35 output.	-100	0	100	mV

Parameter	Symbol	Conditions	MIN.	TYP.	MAX.	Unit
COM AMP adjustable	VCOM1	BRIGHT=80h, COM AMPLITUDE=00h. Measure the amplitude of TP39 output.	1. 5	2. 2	2. 8	VP-P
range	VCOM2	COM AMPLITUDE=FFh	7. 6	8. 3	9. 0	
COM AMP sensitivity	△ VCOM	The amplitude varying per 1LSB.	18	24	30	mV
COM-Gamma fracking ratio	Rγ	INPUT MODE/SW=34h, TP22=3V, TP21, 20, 19=SG7 (700mVp-p). Measure varying ratio of amplitude between inverted and non-inverted white level of TP33 to amplitude of TP39, when BRIGHT=60h and A0h.	0. 93	1. 00	1. 07	
COM BLACK LEV adjustable range	VCB1	TP43=3V, COM BLACK LEV=00h. Measure the amplitude of TP39 output.	8. 4	9. 1	-	VP-P
	VCB2	COM BLACK LEV=1Fh		3. 9	4. 6	
COM BLACK LEV sensitivity	ΔVCB	The amplitude varying per 1LSB.	140	190	240	шV
DAC OUT variable	VDAC1	DAC OUT=00h. TP44 voltage at 0.5mA sink.			0.3	3 V
range	VDAC2	DAC OUT=00h. TP44 voltage at 0.5mA source.	2. 7			V
DAC OUT sensitivity	ΔVDAC	The amplitude varying per 1LSB.	9	11	13	mV
Carrier leak	CR	INPUT MODE/SW=15h, PICTURE=3Fh, COLOR=FFh, TP42=3V, TP5=SG9 (NTSC). Measure the ratio of the 3.579545MHz component to the 15.734kHz component in the output of TP33 with SPECTRUM ANALYZER.		-35	-20	dB
Crosstalk among RGB	CTRGB	TP22=TP42=3V, TP21=SG4 (IMHz, 300mV). Measure the amplitude of IMHz component of TP29, 33 and 35 with SPECTRUM ANALYZER. Calculate the amplitude ratio of TP33 and TP35 to TP29. Similarly, TP20=SG4 (IMHz, 300mV), calculate the amplitude ratio of TP29 and TP35 to TP33. Similarly, TP19=SG4 (IMHz, 300mV), calculate the amplitude ratio of TP29 and TP33 to TP35.		-50	-40	dB
Crosstalk between inputs (EXT→1NT)	CTEI	TP42=3V. With TP21=SG4 (1MHz, 300mV), TP22=3V and SPECTRUM ANALYZER. Measure the amplitude of 1MHz component of TP29. Then with TP22=0V, measure the attenuation of 1MHz component on TP29. Similarly, with TP20=SG4 (1MHz, 300mV), measure the one of TP33. Similarly, with TP19=SG4 (1MHz, 300mV), measure the one of TP35.		-45	-35	₫B
Crosstalk between inputs (INT→EXT)	CTIE	TP42=3V, TP13=SG4 (1MHz, 300mV). With TP22=0V and SPECTRUM ANALYZER. Measure the amplitude of 1MHz component of TP29, TP33 and TP35. Then with TP22=3V, measure the attenuation of 1MHz component of TP29, TP33 and TP35.		-45	-35	dB

Parameter	Symbol	Conditions	MIN.	TYP.	MAX.	Unit
Sync separator input current sensitivity	ESYNC	SW9→OFF. Flow out the current from TP9. Measure the current when TP10 turns "Hi" from "Lo".		38	55	μΑ
Sync separator output ON-state voltage	VON	SW9→OFF. Measure the voltage of TP10		in the control	0.3	V
Sync separator OFF-state current Leak	10FF	SW9, $10 \rightarrow 0$ FF, TP10=10V. Flow out the $100~\mu$ A current from TP9. Measure the current from TP10 into pin 10.			1	μΛ
SYNC IN threshold voltage	VTHsyn	TP11	1. 2	1. 5	1. 8	V
FRP threshold voltage	VTHERP	TP42	1. 2	1. 5	1. 8	V
COM FRP threshold voltage	VTHCFRP	TP41	1. 2	1. 5	1. 8	V
COM BLACK threshold voltage	VTHeblk	TP43	1. 2	1. 5	1. 8	V
BLACK IN threshold voltage	VTHBKIN	TP23	1. 0	1. 3	1. 6	V
Serial input L→H threshold voltage	VTHLH	TP1, TP2, TP48		1. 6	1. 9	V
Serial input H→L threshold voltage	VTHHL	TP1, TP2, TP48	0. 7	1. 0		V
Input-Output	tLH1	SW29, 33, 35→0N, TP21, 20, 19=SG5, TP22=3V. Measure the propagation delay time of	30	70	110	
propagation delay time (RGB input)	t AL 1	TP29, 33, 35 output when the output rises (tLH1) and falls (tHL1).	30	70	110	ns
Input-Output propagation delay	t LH2	SW29, 33, 35→ON, TP13=SG5. Measure the propagation delay time of TP29, 33, 35	240	570	800	
time (composite input)	tHL2	output when the output rises(tLH2) and falls(tHL2)	240	570	800	ns
Input-Output	tLH3	SW29, 33. 35→ON, INPUT MODE/SW=15h. TP13=SG5. Measure the propagation	160	260	360	
propagation delay time (Y/C input)	tHL3	delay time of TP29,33.35 output when the output rises(tLH3) and falls(tHL3)	160	260	360	ns
BLACK IN	t LH4	SW29, 33, 35→ON. TP13=SG8, TP23=SG5 ("Hi" level=3V). Measure the propagation	100	170	240	
propagation delay time	tHL4	delay time of TP29,33.35 output when the output rises(tLH4) and falls(tHL4)	70	130	190	ns
FRP	tLH5	SW29, 33, 35→ON. Measure the propagation delay time from TP42 input	80	140	200	
propagation delay time	tHL5	to TP29, 33, 35 output when the output rises (tLH5) and falls (tHL5).	80	140	200	n s
COM FRP	tLH6	SW39→ON, BRIGHT=80h. Measure the propagation delay time from TP41 input	400	650	900	
propagation delay time	tHL6	to TP39 output when the output rises(tLH6) and falls(tHL6).	330	550	770	ns
COM BLACK propagation delay	tLH7	SW39→ON, BRIGHT=80h, COM BLACK LEV=1Fh, TP43=SG5 ("Hi" level=3V). Measure the propagation delay time of TP39 output	360	730	1100	ns
time	tHL7	when the output rises (tLH7) and falls (tHL7).	360	730	1100	11.9

Paramter	Symbol	Cnditions	MIN.	TYP.	MAX.	Unit
SW	tLH8	SW29.33,35→0N.TP21.20,19=SG10. TP22=SG5("Hi" level=3V). Measure the propagation delay time of TP29.33,35	70	130	190	n e
propagation delay time		output when the output rises(tLH8) and	70	130	190	ns
HFILTER propagation	tLH9	SW9→OFF, TP47=SG1("Lo" level=2.8V). Measure the propagation delay time of	190	520	850	ns
delay time	tHL9	TP46 output when the output rises(tLH9) and falls(tHL9).	180	330	480	11.0
Sync separator	tLH10	MD47 CO1/"I - " 1 1 0 0 C\	500	1000	1500	ns
propagation delay time	tHL10	TP10 output when the output rises(tLH10) and falls(tHL10).	130	190	260	по
Differential non linear error 1	DNL1	Apply to a DAC of lower 7 bit resolution.	-0. 8		0. 8	LSB
Differential non linear error 2	DNL2	Apply to a DAC of 8 bit resolution.	-1. 5		1. 5	LSB

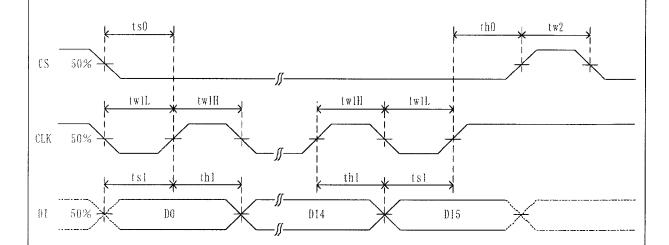
 \star The propagation delay time is measured by referring the 50% point of the input and the output. (As shown in the figure below.)

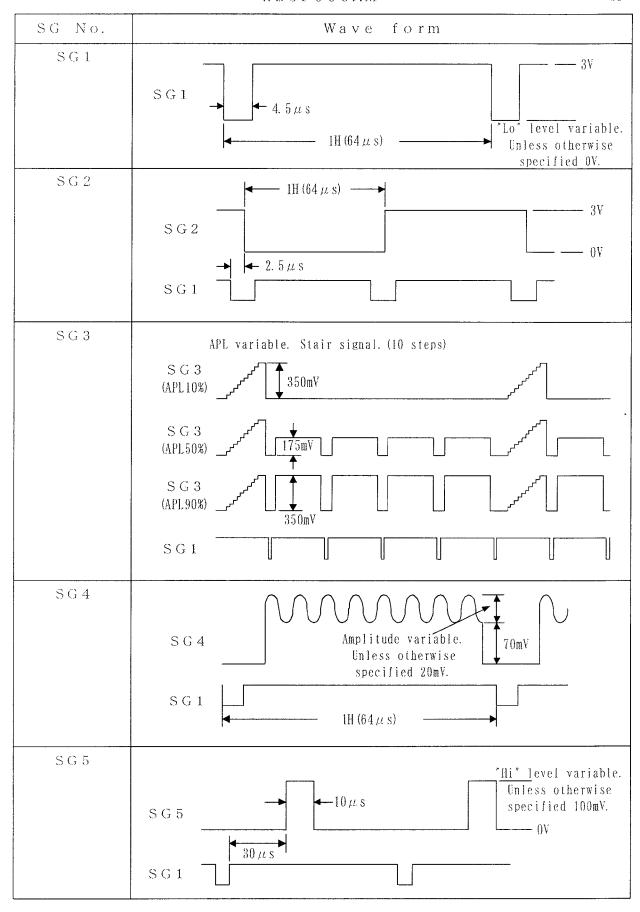


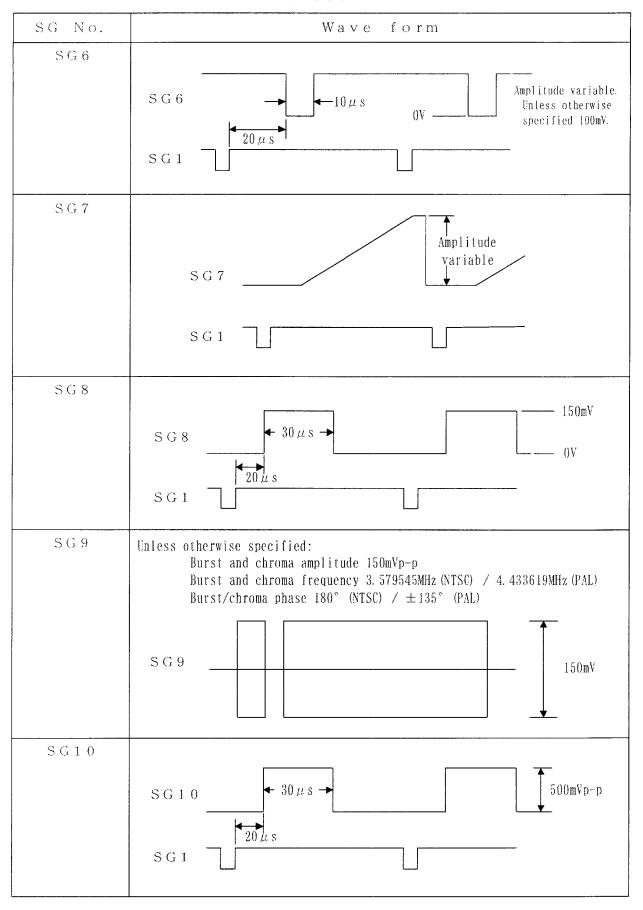
Serial I/O AC characteristics

 $Ta=-30\sim85$ °C, $Vec 1=2.7\sim3.6$ V, $Vec 2=Vec 4=4.5\sim5.5$ V, $Vec 3=11\sim15.5$ V

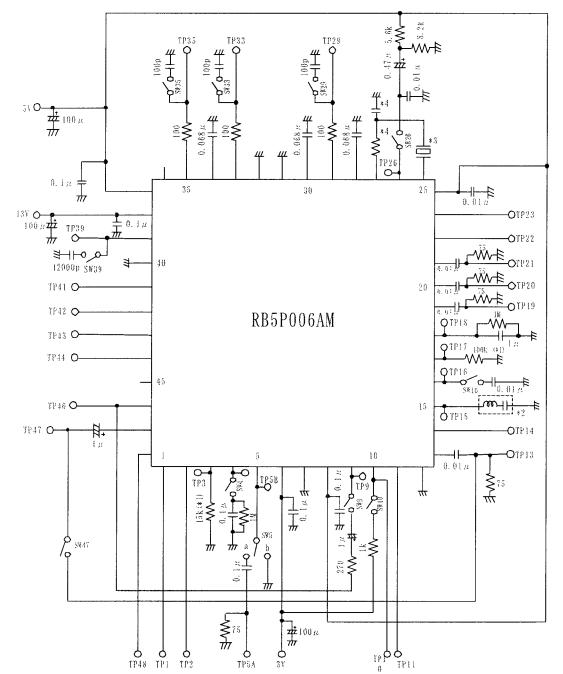
Parameter	Symbol	MIN.	TYP.	MAX.	Unit
CS setup time based on CLK rise	ts0	330			ns
DI setup time based on CLK rise	ts1	220			ns
CS hold time based on CLK rise	th0	330			ns
DI hold time based on CLK rise	thl	220			ns
CLK pulse width ("Lo" level period)	twlL	330			ns
CLK pulse width ("Hi" level period)	twlH	330			ns
CS pulse width ("Hi" level period)	t w2	330			ns







11. Test Circuit



Notes

- *1 Resistance accuracy $\pm 2\%$. Temperature stability $\pm 200 \text{pnm/} \mathfrak{V}$
- +2 TDK NLT4532-83R6B (NTSC mode) TDK NLT4532-84R4 (PAL mode)

1 s 1 Z 6*

DAISHINKU CORPORATION AT-49

Frequency: 3, 579545MHz (NTSC mode) Name-of-article Code: IAF7294CLA 4.433619MHz (PAL mode) Name-of-article Code: 1AF0218Cf.

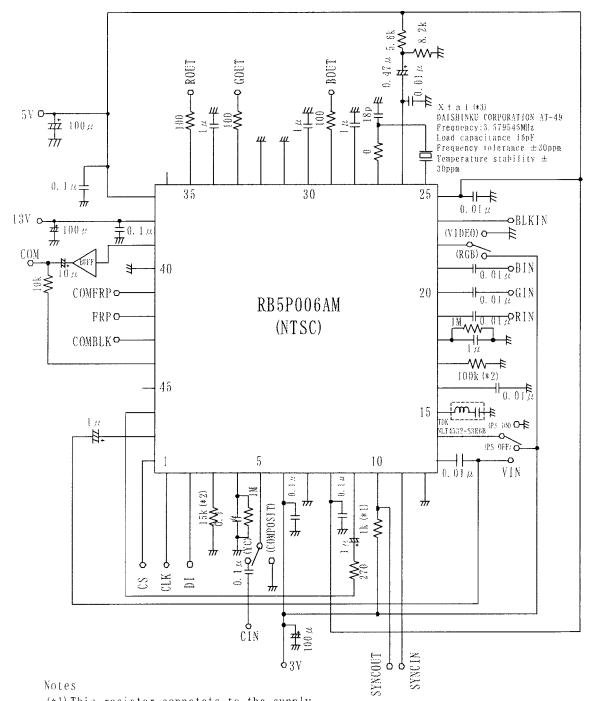
Load capacitance 16pF

Frequency tolerance ±30ppm Temperature stability ±30ppm

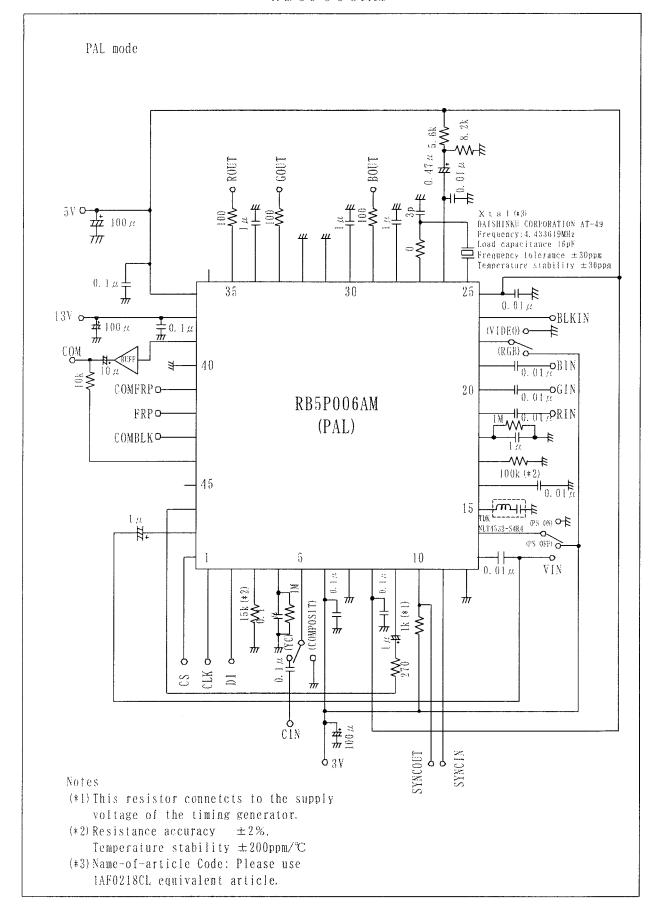
 $*4~0\,\Omega,~18 pF~(NTSC~mode)$

09, 3pF (PAL mode)

1 2. Application circuit example (NTSC mode)

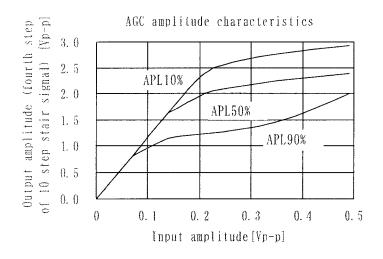


- (*1) This resistor connectts to the supply voltage of the timing generator.
- (*2) Resistance accuracy $\pm 2\%$, Temperature stability $\pm 200 \mathrm{ppm/C}$
- (*3) Name-of-article Code: Please use 1AF7294CLA equivalent article.



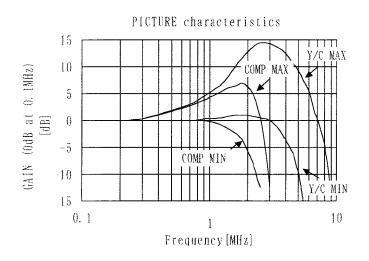
13. Typical Characteristics

The conditions are the same as AC Characteristics.

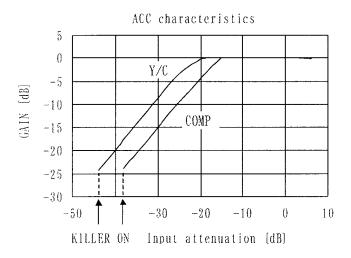


AGC LEVEL 1 Constants of AGC filter 1MΩ 1μF

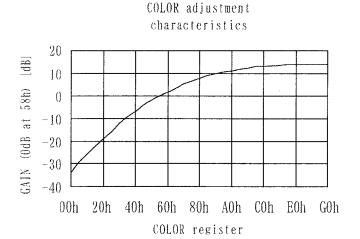
Measure the fourth step of 10 step stair signal.



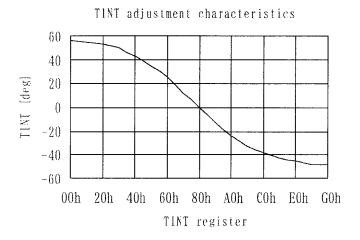
The gain at 0.1MHz frequency is 0dB.
NTSC mode.



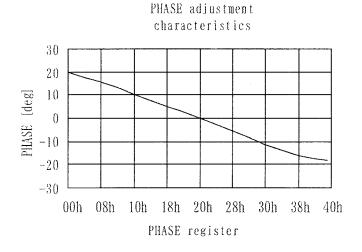
NTSC mode.



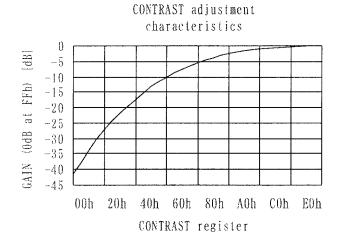
The gain at COLOR register = 58h is OdB.



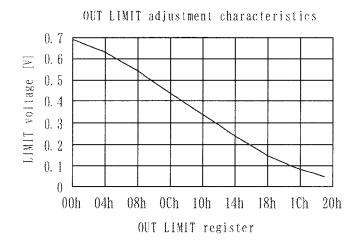
The degree at TINT register=80h (preset value) is Odeg.



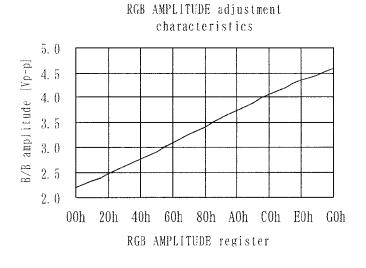
The degree at PHASE register=20h (preset value) is Odeg.



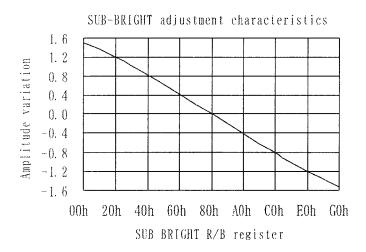
The gain at CONTRAST register = FFh is OdB.



Limit voltage:
The potential between pedestal level (non-inverted output) and black limit level.

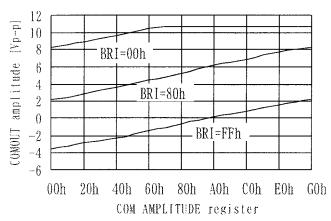


B/B amplitude : The amplitude of the black level between inverted and non-inverted output.



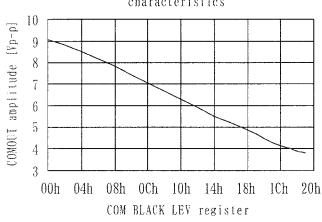
The amplitude at SUB-BRIGHT register =80h (preset value) is 0.



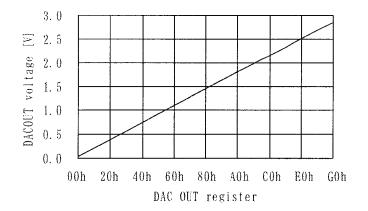


A COMOUT amplitude depends on both BRIGHT register and COM AMPLITUDE register.

COM BLACK LEV adjustment characteristics







DACOUT terminal is open.



14 Package and packing specification

- 1.Storage Conditions.
 - 1-1. Storage conditions required before opening the dry packing.
 - · Normal temperature: 5~40℃
 - · Normal humidity: 80% R.H. max.
 - 1-2. Storage conditions required after opening the dry packing.

In order to prevent moisture absorption after opening, ensure the following storage conditions apply:

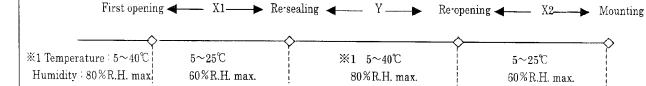
- (1) Storage conditions for one-time soldering. (Convection reflow*1, IR/Convection reflow.*1, V.P.S., or Manual soldering.)
 - · Temperature: 5~25℃
 - · Humidity: 60% R.H. max.
 - · Period: 96 hours max. after opening.
- (2) Storage conditions for one-time soldering. (Solder dipping.)
 - · Temperature:5~25℃
 - · Humidity: 60% R.H. Max.
 - · Period: 96 hours max. after opening.
- (3) Storage conditions for two-time soldering. (Convection reflow*1, IR/Convection reflow.*1)
 - a. Storage conditions following opening and prior to performing the 1st reflow.
 - Temperature : $5\sim25$ °C.
 - · Humidity: 60% R.H. max.
 - · Period: 96 hours max. after opening.
 - b. Storage conditions following completion of the 1st reflow and prior to performing the 2nd reflow.
 - Temperature: $5\sim25$ °C.
 - · Humidity: 60% R.H. max.
 - · Period: 96 hours max. after completion of the 1st reflow.
- *1:Air or nitrogen environment.
- 1-3. Temporary storage after opening.

To re-store the devices before soldering, do so only once and use a dry box or place desiccant (with a blue humidity indicator) with the devices and perform dry packing again using heat-sealing.

The storage period, temperature and humidity must be as follows:

(1) Storage temperature and humidity.

%1: External atmosphere temperature and humidity of the dry packing.



- (2) Storage period.
 - \cdot X1+X2: Refer to Section 1·2(1),(2), and (3)a, depending on the mounting method.
 - · Y : Two weeks max.



2. Baking Condition.

- (1) Situations requiring baking before mounting.
 - · Storage conditions exceed the limits specified in Section 1-2 or 1-3.
 - · Humidity indicator in the desiccant was already red (pink) when opened.
 - (Also for re-opening.)
- (2) Recommended baking conditions.
 - · Baking temperature and period:

120°C for 16∼24 hours.

- · The above baking conditions apply since the trays are heat-resistant.
- (3) Storage after baking.
 - After baking, store the devices in the environment specified in Section 1-2 and mount immediately.

3. Surface mount conditions.

The following soldering condition are recommended to ensure device quality.

3-1. Soldering.

- (1) Convection reflow or IR/Convection. (one-time soldering only in air or nitrogen environment)
 - · Temperature and period:

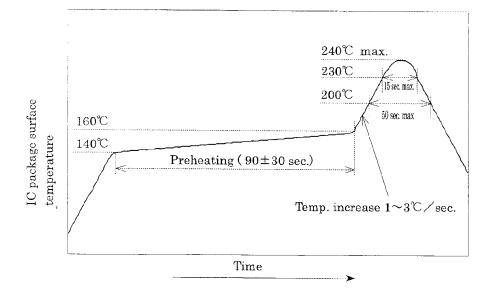
Peak temperature of 240℃ max., above 230℃ for 15 sec. max.

Above 200℃ for 50 sec. max.

Preheat temperature of $140 \sim 160\%$ for 90 ± 30 sec.

Temperature increase rate of $1\sim 3\%/\sec$.

- · Measuring point: IC package surface.
- · Temperature profile:





- (2) Convection reflow or IR/Convection. (two-time soldering only in air or nitrogen environment)
 - · Temperature and period :

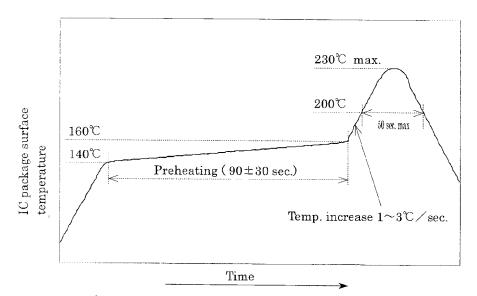
Peak temperature of 230°C max.

Above 200°C for 50 sec. max.

Preheat temperature of $140 \sim 160 \, \text{C}$ for $90 \pm 30 \, \text{sec}$.

Temperature increase rate of $1\sim3\%/\text{sec}$.

- · Measuring point: IC package surface.
- · Temperature profile:



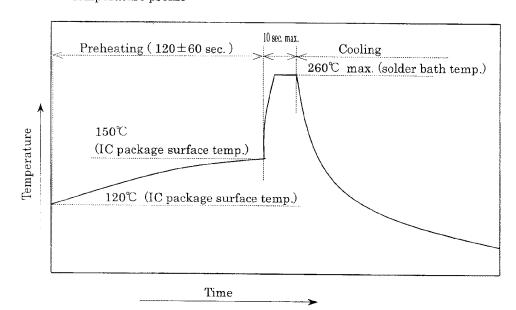
- (3) Solder dipping. (one-time dipping only)
 - Temperature and period :

260℃ max. for 10 sec. max.

Preheat temperature of $120\sim150$ °C for 120 ± 60 sec.

Measuring point: IC package surface and solder bath.

· Temperature profile:



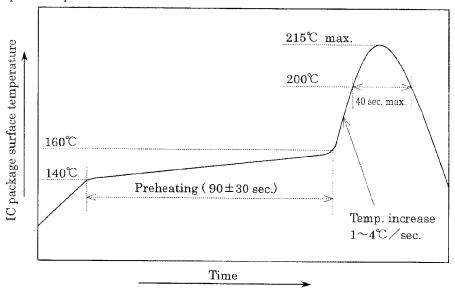
- (3) V.P.S.(one-time soldering only)
 - · Temperature and period:

Peak temperature of 215°C max., above 200°C for 40 sec. max.

Preheat temperature of $140 \sim 160\%$ for 90 ± 30 sec.

Temperature increase rate of $1 \sim 4\%/\text{sec}$.

- · Measuring point: IC package surface.
- · Temperature profile



(4) Manual soldering (soldering iron) (one-time soldering only)

Soldering iron should only touch the IC's outer leads.

• Temperature and period :

 $350\%\,$ max. for 3 sec. / pin max., or $260\%\,$ max. for 10 sec. / pin max.

(Soldering iron should only touch the IC's outer leads.)

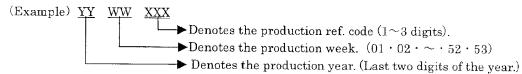
- · Measuring point: Soldering iron tip.
- 4. Condition for removal of residual flax.
- (1) Ultrasonic washing power: 25 watts / liter max.
- (2) Washing time: Total 1 minute max.
- (3) Solvent temperature: 15~40°C



5. Package outline specification.

Refer to the attached drawing.

- 6. Markings.
 - 6-1.Marking details. (The information on the package should be given as follows.)
 - (1) Product name : RB5P006AM
 - (2) Company name : SHARP
 - (3) Date code



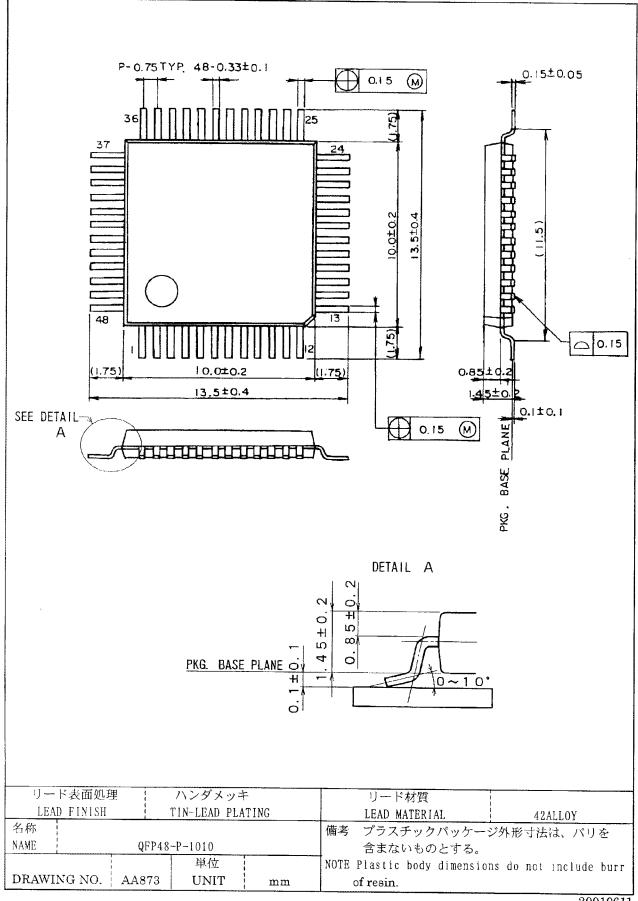
(4) "JAPAN" indicates the country of origin.

6.2. Marking layout.

The layout is shown in the attached drawing.

(However, this layout does not specify the size of the marking character and marking position.)









7.Packing Specifications (Dry packing for surface mount packages.)

7-1. Packing materials.

Material name	Material specifications	Purpose
Inner carton	Cardboard (800 devices / inner carton	Packing the devices.
	max.)	(10 trays / inner carton)
Tray	Conductive plastic (80 devices / tray)	Securing the devices.
Upper cover tray	Conductive plastic (1 tray / inner carton)	Securing the devices.
Laminated aluminum	Aluminum polyethylene	Keeping the devices dry.
bag		
Desiccant	Silica gel	Keeping the devices dry.
Label	Paper	Indicates part number,
		quantity, and packed date.
PP band	Polypropylene (3 pcs. / inner carton)	Securing the devices.
Outer carton	Cardboard (3200 devices / outer carton	Outer packing.
	max.)	

(Devices must be placed on the tray in the same direction.)

7-2. Outline dimension of tray.

Refer to the attached drawing.

7-3. Outline dimension of carton.

Refer to the attached drawing.

8. Precautions for use.

- Opening must be done on an anti-ESD treated workbench.
 All workers must also have undergone anti-ESD treatment.
- (2) The trays have undergone either conductive or anti-ESD treatment.

 If another tray is used, make sure it has also undergone conductive or anti-ESD treatment.
- (3) The devices should be mounted the devices within one year of the date of delivery.



