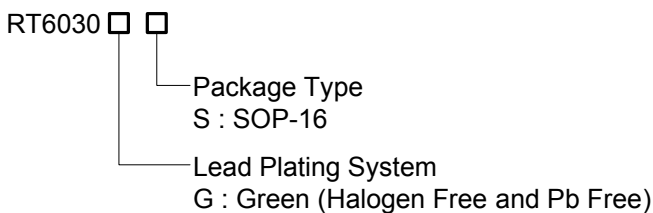


## 4-CH LED Current Source Controller

### General Description

The RT6030 is a current source controller, capable of driving up to 4-CH of LEDs. The part can also be used to drive an external BJT or N-MOSFET for various applications. With a wide operating voltage range from 3.8V to 13.5V, the RT6030 has the advantage of being flexible and cost-effective. The RT6030 is available in an SOP-16 package.

### Ordering Information

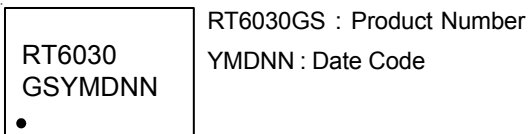


Note :

Richtek products are :

- ▶ RoHS compliant and compatible with the current requirements of IPC/JEDEC J-STD-020.
- ▶ Suitable for use in SnPb or Pb-free soldering processes.

### Marking Information



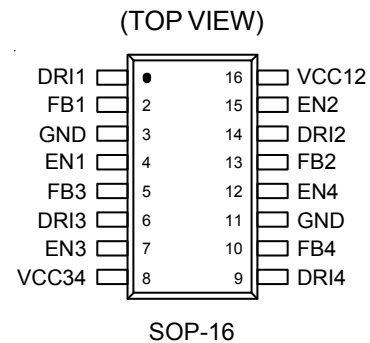
### Features

- 3.8V to 13.5V Operating Voltage
- 0.8V Voltage Reference with  $\pm 2\%$  High Accuracy
- Independent Enable Control for Each Channel
- Quick Transient Response
- Over Temperature Protection
- RoHS Compliant and Halogen Free

### Applications

- LED TV Backlight
- Lighting
- Intelligent Instruments
- Industrial Display Backlight

### Pin Configuration



### Part Status

Part No	Status	Package	Lead Plating System
RT6030	Lifebuy	SOP-16	G : Green (Halogen Free and Pb Free)

The part status values are defined as below :

**Active** : Device is in production and is recommended for new designs.

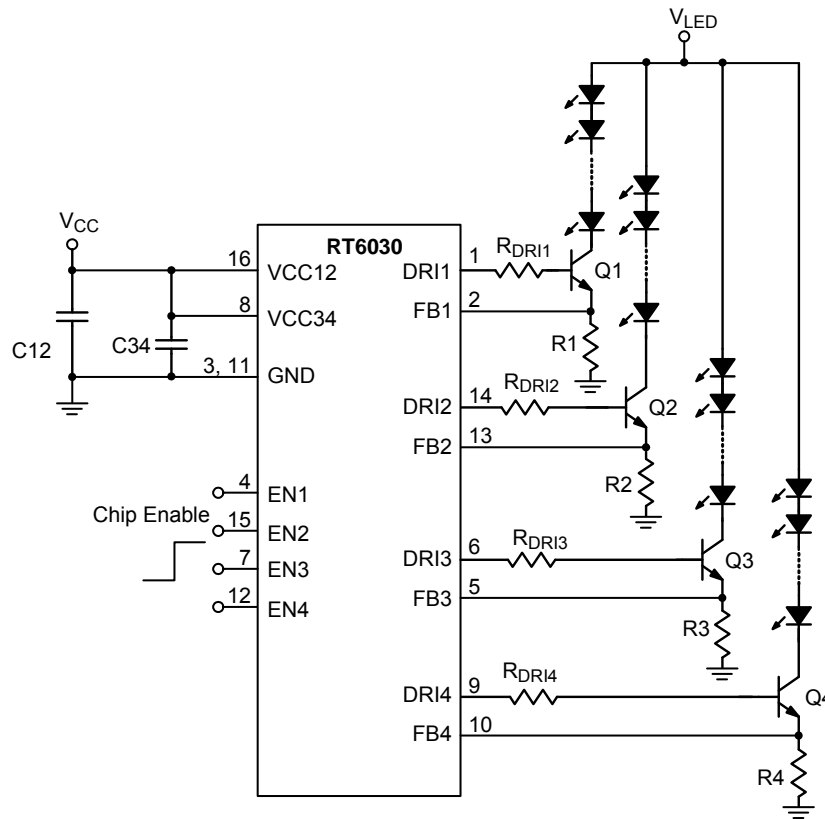
**Lifebuy** : The device will be discontinued, and a lifetime-buy period is in effect.

**NRND** : Not recommended for new designs.

**Preview** : Device has been announced but is not in production.

**Obsolete**: Richtek has discontinued the production of the device.

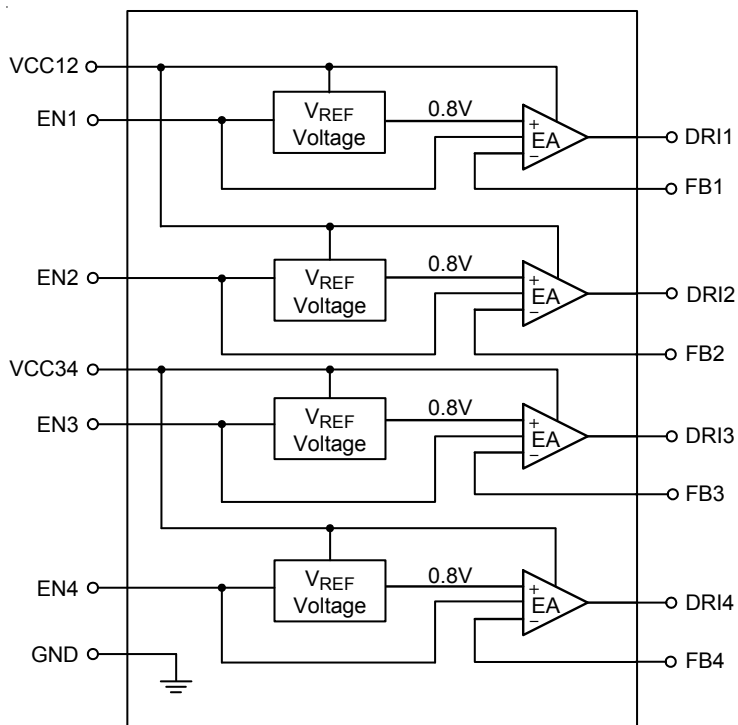
## Typical Application Circuit



## Functional Pin Description

Pin No.	Pin Name	Pin Function
1	DRI1	1 <sup>st</sup> CH driver output.
2	FB1	1 <sup>st</sup> CH current sense voltage feedback.
3, 11	GND	Ground.
4	EN1	1 <sup>st</sup> CH chip enable (Active High).
5	FB3	3 <sup>rd</sup> CH current sense voltage feedback.
6	DRI3	3 <sup>rd</sup> CH driver output.
7	EN3	3 <sup>rd</sup> CH chip enable (Active High).
8	VCC34	CH3 and CH4 power supply input.
9	DRI4	4 <sup>th</sup> CH driver output.
10	FB4	4 <sup>th</sup> CH current sense voltage feedback.
12	EN4	4 <sup>th</sup> CH chip enable (Active High).
13	FB2	2 <sup>nd</sup> CH current sense voltage feedback.
14	DRI2	2 <sup>nd</sup> CH driver output.
15	EN2	2 <sup>nd</sup> CH chip enable (Active High).
16	VCC12	CH1 and CH2 power supply input.

**Functional Block Diagram**



## Absolute Maximum Ratings (Note 1)

- VCC12, VCC34 ----- 15V
- All Other Inputs ----- 7V
- Power Dissipation, P<sub>D</sub> @ T<sub>A</sub> = 25°C
- SOP-16 ----- 1.053W
- Package Thermal Resistance (Note 2)
- SOP-16, θ<sub>JA</sub> ----- 95°C/W
- Lead Temperature (Soldering, 10 sec.) ----- 260°C
- Junction Temperature ----- 150°C
- Storage Temperature Range ----- -65°C to 150°C
- ESD Susceptibility (Note 3)
- HBM ----- 1.5kV
- MM ----- 150V

## Recommended Operating Conditions (Note 4)

- Supply Input Voltage, V<sub>CC12</sub>, V<sub>CC34</sub> ----- 3.8V to 13.5V
- Chip Enable Voltage, EN1, EN2, EN3, EN4 ----- 0V to 5.5V
- Junction Temperature Range ----- -40°C to 125°C
- Ambient Temperature Range ----- -40°C to 85°C

## Electrical Characteristics

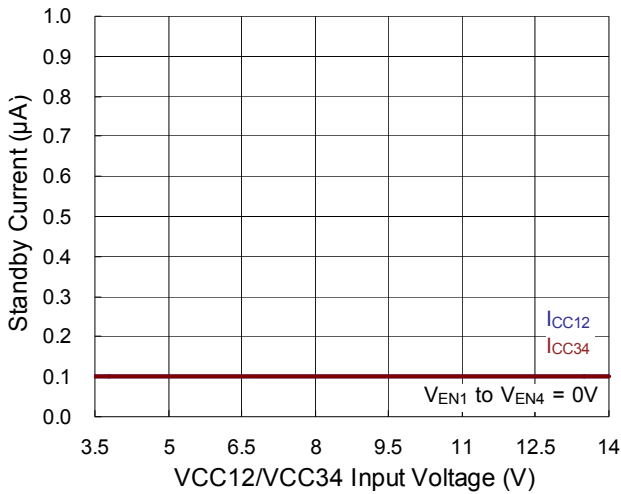
(V<sub>CC12</sub> = 5V/12V, V<sub>CC34</sub> = 5V/12V, T<sub>A</sub> = 25°C, unless otherwise specified)

Parameter	Symbol	Test Conditions	Min	Typ	Max	Unit
Under Voltage Lockout Threshold	V <sub>UVLO</sub>	V <sub>CC12</sub> and V <sub>CC34</sub> rising	3.15	3.4	3.65	V
Under Voltage Lockout Hysteresis	ΔV <sub>UVLO</sub>	V <sub>CC12</sub> and V <sub>CC34</sub> falling	0.1	0.2	0.3	V
V <sub>CC12</sub> and V <sub>CC34</sub> Supply Current		V <sub>CC12</sub> and V <sub>CC34</sub> = 12V	--	0.6	1.6	mA
Driver Source Current	I <sub>SR</sub>	V <sub>CC12</sub> and V <sub>CC34</sub> = 12V V <sub>DRI1</sub> to V <sub>DRI4</sub> = 6V	5	--	--	mA
Driver Sink Current	I <sub>SK</sub>	V <sub>CC12</sub> and V <sub>CC34</sub> = 12V V <sub>DRI1</sub> to V <sub>DRI4</sub> = 6V	5	--	--	mA
Reference Voltage (V <sub>FB1</sub> to V <sub>FB4</sub> )		V <sub>CC12</sub> and V <sub>CC34</sub> = 12V V <sub>DRI1</sub> to V <sub>DRI4</sub> = 5V	0.784	0.8	0.816	V
Reference Line Regulation (V <sub>FB1</sub> to V <sub>FB4</sub> )		V <sub>CC12</sub> and V <sub>CC34</sub> = 4.5V to 13.5V	--	3	6	mV
Amplifier Voltage Gain		V <sub>CC12</sub> and V <sub>CC34</sub> = 12V, no load	--	70	--	dB
<b>Chip Enable</b>						
EN Rising Threshold	V <sub>EN</sub>	V <sub>CC12</sub> and V <sub>CC34</sub> = 12V	--	0.7	--	V
EN Hysteresis	ΔV <sub>EN</sub>	V <sub>CC12</sub> and V <sub>CC34</sub> = 12V	--	30	--	mV
Standby Current		V <sub>CC12</sub> and V <sub>CC34</sub> = 12V V <sub>EN1</sub> to V <sub>EN4</sub> = 0V	--	--	10	μA

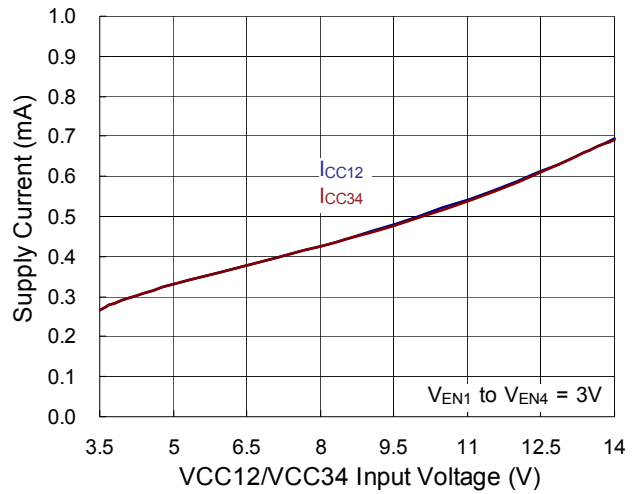
- Note 1.** Stresses beyond those listed under “Absolute Maximum Ratings” may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions may affect device reliability.
- Note 2.**  $\theta_{JA}$  is measured in natural convection at  $T_A = 25^\circ\text{C}$  on a low-effective thermal conductivity test board of JEDEC 51-3 thermal measurement standard.
- Note 3.** Devices are ESD sensitive. Handling precaution is recommended.
- Note 4.** The device is not guaranteed to function outside its operating conditions.

Typical Operating Characteristics

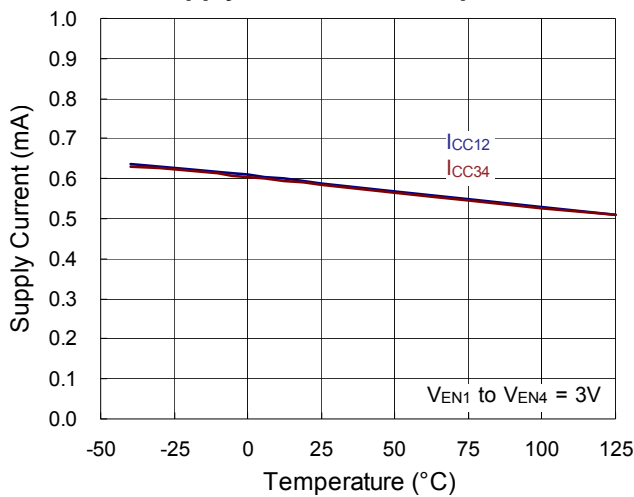
Standby Current vs. VCC12/VCC34 Input Voltage



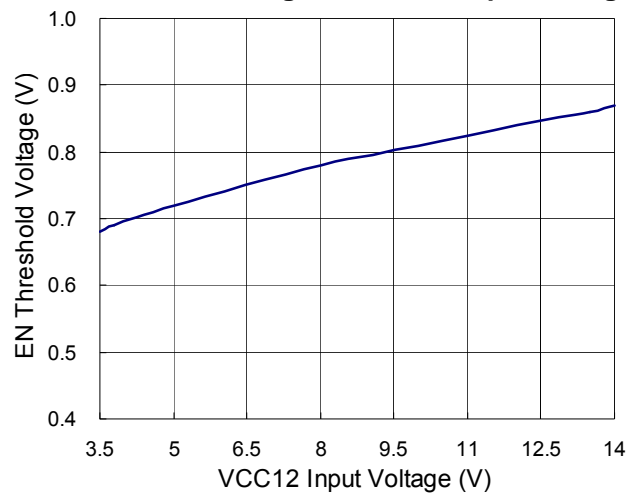
Supply Current vs. VCC12/VCC34 Input Voltage



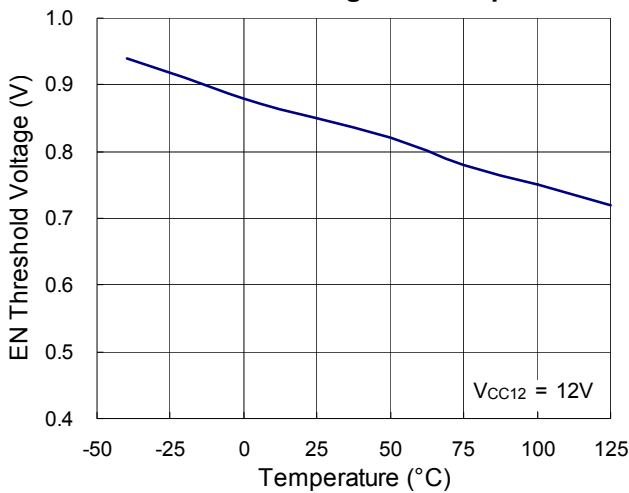
Supply Current vs. Temperature



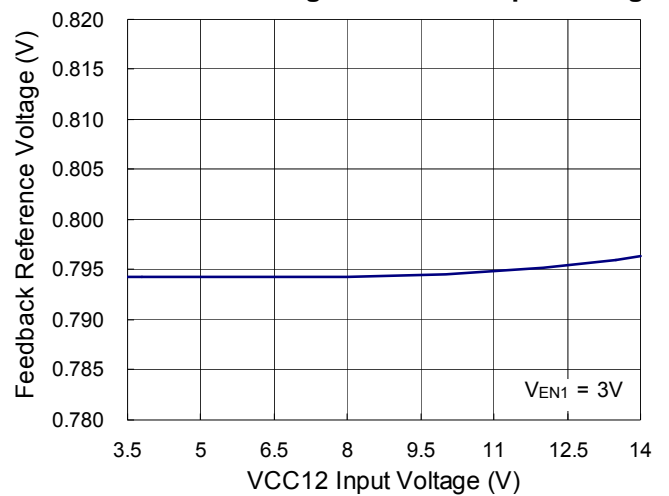
EN Threshold Voltage vs. VCC12 Input Voltage



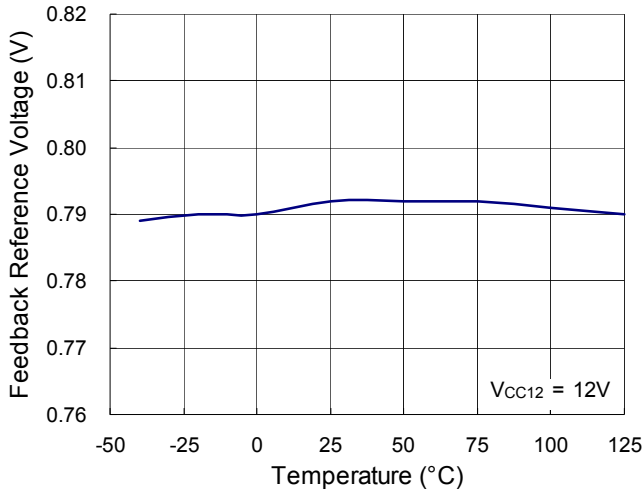
EN Threshold Voltage vs. Temperature



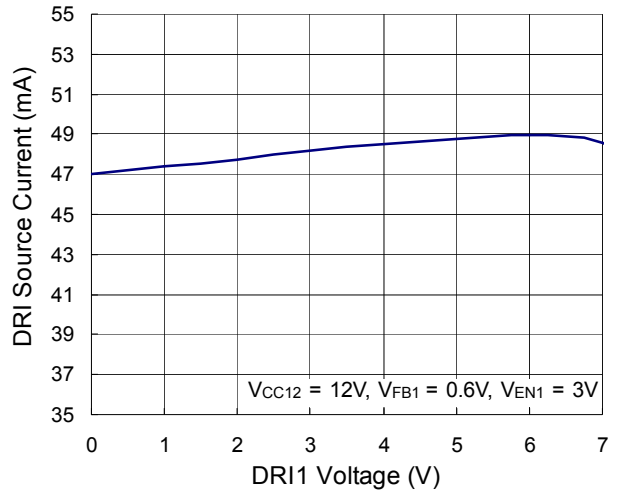
Feedback Reference Voltage vs. VCC12 Input Voltage



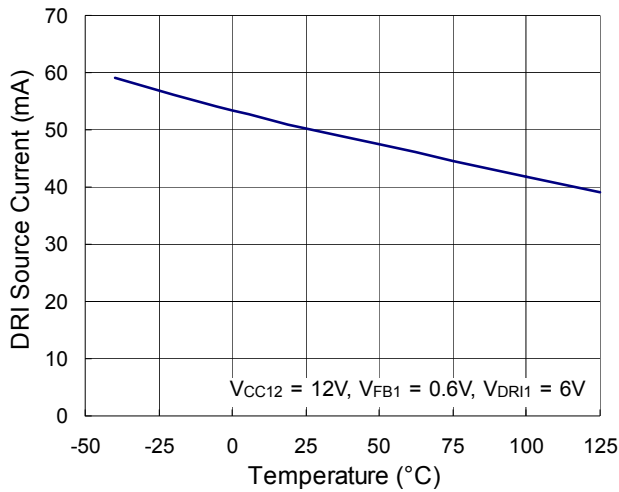
**Feedback Reference Voltage vs. Temperature**



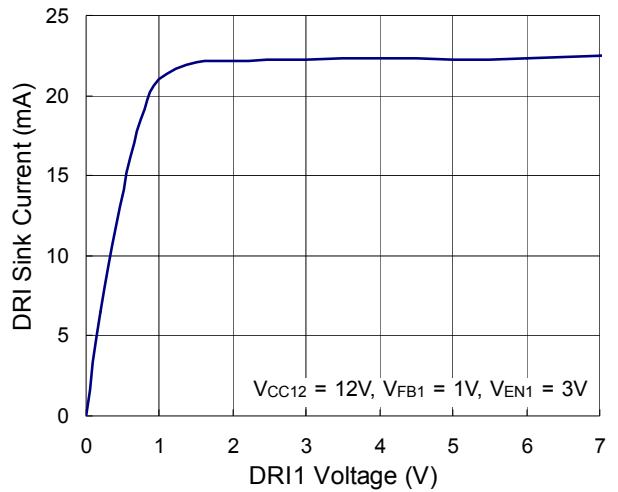
**DRI Source Current vs. DRI1 Voltage**



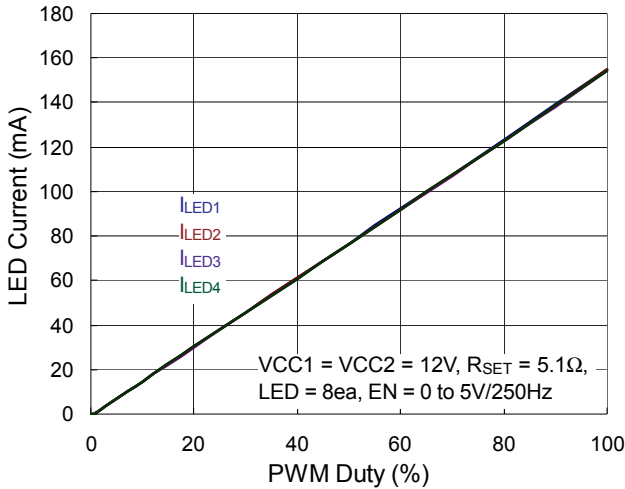
**DRI Source Current vs. Temperature**



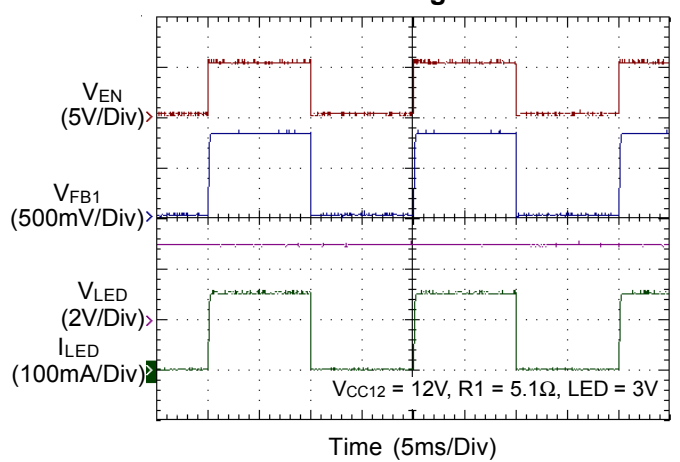
**DRI Sink Current vs. DRI1 Voltage**



**LED Current vs. PWM Duty**



**PWM Dimming From EN**



## Applications Information

The RT6030 is a 4-CH LED current source controller. This device can also drive an external BJT or N-MOSFET for various applications. Refer to topology in Typical Application Circuit for more details.

### Capacitors Selection

Careful selection of the external capacitors for the RT6030 is highly recommended in order to maintain high stability and performance. An input capacitor with minimum 1μF must be connected between VCC and ground. The capacitor improves the supply voltage stability for proper operation.

### Chip Enable Operation

Pull the EN pin low to drive the device into shutdown mode. During shutdown mode, the standby current drops to 10mA(MAX). Drive the EN pin high to turn on the device again. To control LED brightness, the RT6030 can perform dimming function by applying a PWM signal to the EN pin. The average LED current is proportional to the PWM signal duty cycle. To obtain correct dimming, the magnitude of the PWM signal should be higher than the threshold voltage of the EN pin.

### MOSFET Selection

The RT6030 is designed to drive external N-MOSFET pass element. MOSFET selection criteria include threshold voltage,  $V_{GS}$  ( $V_{TH}$ ), maximum continuous drain current,  $I_D$ , on resistance,  $R_{DS(ON)}$ , maximum drain-to-source voltage,  $V_{DS}$ , and package thermal resistance,  $\theta_{JA}$ . The most critical specification is the MOSFET  $R_{DS(ON)}$ .  $R_{DS(ON)}$  can be calculated from the following formula :

$$R_{DS(ON)} = \frac{(V_{IN} - V_{OUT})}{I_O}$$

For example, the MOSFET operates up to 2A when the input voltage is 1.5V and set the output voltage as 1.2V. Then,  $R_{DS(ON)} = (1.5V - 1.2V) / 2A = 150m\Omega$ . The MOSFET's  $R_{DS(ON)}$  must be lower than 150mΩ. Philip PHD3055E MOSFET with an  $R_{DS(ON)}$  of 120mΩ (typ.) is a suitable solution.

The power dissipation is calculated as :

$$P_D = (V_{IN} - V_{OUT}) \times I_{LOAD}$$

The thermal resistance from junction to ambient is :

$$\theta_{JA} = \frac{(T_J - T_A)}{P_D}$$

In this example,  $P_D = (1.5V - 1.2V) \times 2A = 0.6W$ . The PHD3055E's  $\theta_{JA}$  is 75°C/W for its D-PAK package, which translates to a 45°C temperature rise above ambient. The package provides exposed backsides that directly transfer heat to the PCB board.

### LED Current Setting

The RT6030 maintains an internal reference voltage of 0.8V. As shown in Typical Application Circuit, the LED current can be set accordingly via the Rx (x = 1, 2, 3, 4) resistor.

$$I_{LEDx} = \frac{0.8}{R_x} \text{ (A)}$$

### NPN Transistor Selection

The RT6030 drives the external NPN transistor via the DR1x pin (source Base current  $I_B$ ). NPN transistor selection criteria include DC current gain,  $h_{FE}$ , threshold voltage,  $V_{BE}$ , collector emitter voltage,  $V_{CE}$ , maximum continuous collector current,  $I_C$ , and package thermal resistance,  $\theta_{JA}$ .

### Thermal Considerations

For continuous operation, do not exceed absolute maximum junction temperature. The maximum power dissipation depends on the thermal resistance of the IC package, PCB layout, rate of surrounding airflow, and difference between junction and ambient temperature. The maximum power dissipation can be calculated by the following formula :

$$P_{D(MAX)} = (T_{J(MAX)} - T_A) / \theta_{JA}$$

where  $T_{J(MAX)}$  is the maximum junction temperature,  $T_A$  is the ambient temperature, and  $\theta_{JA}$  is the junction to ambient thermal resistance.

For recommended operating condition specifications of the RT6030, the maximum junction temperature is 125°C and  $T_A$  is the ambient temperature. The junction to ambient thermal resistance,  $\theta_{JA}$ , is layout dependent. For SOP-16 packages, the thermal resistance,  $\theta_{JA}$ , is 95°C/W on a standard JEDEC 51-3 single-layer thermal test board. The maximum power dissipation at  $T_A = 25^\circ\text{C}$  can be calculated by the following formula :



$$P_{D(MAX)} = (125^{\circ}\text{C} - 25^{\circ}\text{C}) / (95^{\circ}\text{C}/\text{W}) = 1.053\text{W for SOP-16 package}$$

The maximum power dissipation depends on the operating ambient temperature for fixed  $T_{J(MAX)}$  and thermal resistance,  $\theta_{JA}$ . For the RT6030 package, the derating curve in Figure 1 allows the designer to see the effect of rising ambient temperature on the maximum power dissipation.

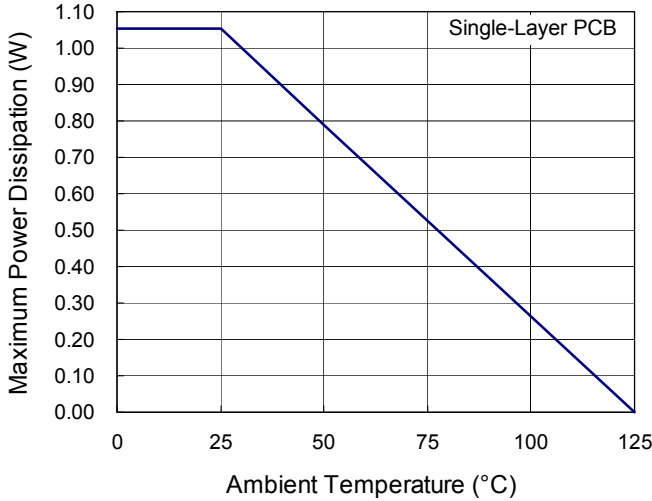


Figure 1. Derating Curve for RT6030 Package

**Layout Considerations**

There are three critical layout considerations.

- ▶ First the current setting resistor should be placed as close as possible to the RT6030 to prevent any noise coupling.
- ▶ Second of all  $C_{IN}$  and  $C_{OUT}$  should be placed near the RT6030 for good performance.
- ▶ Last of all, proper copper area for the pass element should be acknowledged. Pass elements operating under high power situations can result in abnormally junction temperature. In addition to the package thermal resistance limit, the copper area should be increased accordingly to improve the power dissipation.

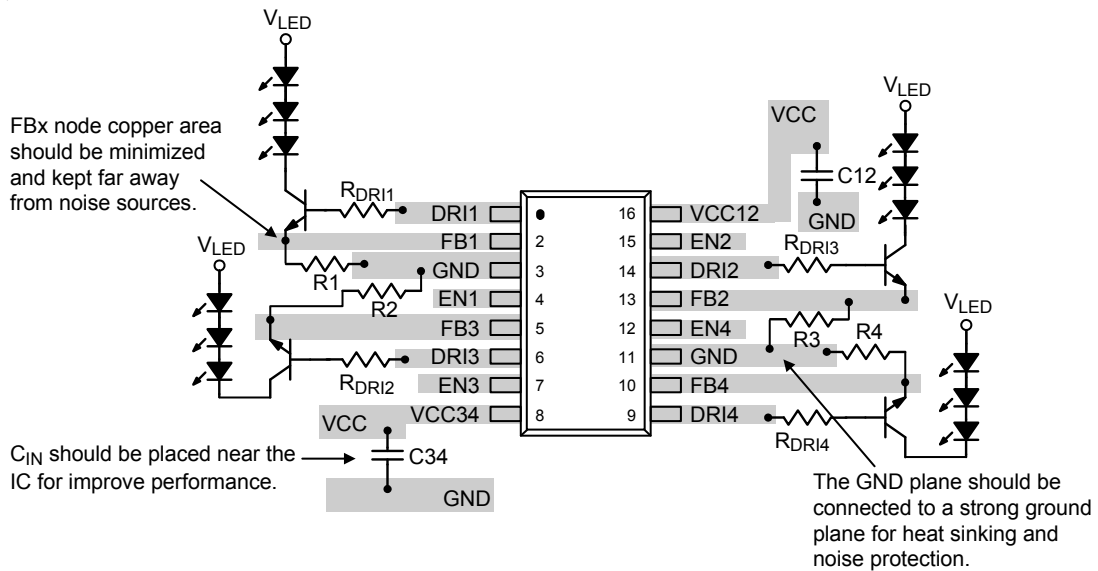
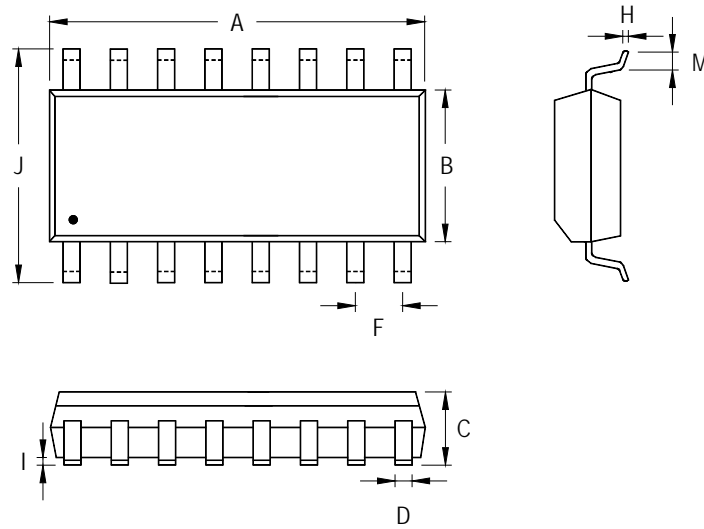


Figure 2. PCB Layout Guide

Outline Dimension



Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min	Max	Min	Max
A	9.804	10.008	0.386	0.394
B	3.810	3.988	0.150	0.157
C	1.346	1.753	0.053	0.069
D	0.330	0.508	0.013	0.020
F	1.194	1.346	0.047	0.053
H	0.178	0.254	0.007	0.010
I	0.102	0.254	0.004	0.010
J	5.791	6.198	0.228	0.244
M	0.406	1.270	0.016	0.050

16-Lead SOP Plastic Package

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