

Three-Phase PMSM/BLDC Motor Controller with Pre-Driver

1 General Description

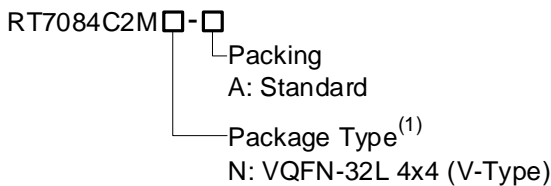
The RT7084C2M is an application-specific IC designed for PMSM/BLDC motor applications. This two-in-one ASIC integrates several functional circuits, including a 3-phase motor controller, a 3-phase gate driver, three bootstrap diodes, a 5V LDO, and a buck converter.

The RT7084C2M embeds the ARM 32-bit Cortex-M0 core with peripheral circuits to perform sensorless Field Oriented Control (FOC). In addition, this ASIC provides several system-level peripheral functions, including filters at the ADC input, a communication interface, a thermal sensor, short circuit protection (SCP) and locked-rotor protection, to reduce component count, PCB size and system cost.

Moreover, the RT7084C2M drives external N-Channel MOSFETs in a three half-bridge configuration with built-in bootstrap circuitry.

The RT7084C2M is available in a VQFN-32L 4x4 package. The recommended junction temperature range is -40°C to 125°C, and the ambient temperature range is -40°C to 105°C.

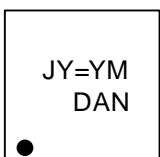
2 Ordering Information



Note 1.

Richtek products are Richtek Green Policy compliant and marked with ⁽¹⁾ indicates compatible with the current requirements of IPC/JEDEC J-STD-020.

3 Marking Information



JY=: Product Code
YMDAN: Date Code

4 Features

- Integrates 3-Phase PMSM/BLDC Controller, Gate Driver, Bootstrap Diodes, 5V LDO, and Buck Converter
- Input Voltage Range: 16V to 80V
- Sensorless, Sine-Wave Field Oriented Control (FOC)
- Integrates Filters at ADC Input
- Protections: SCP, UVLO, Locked-Rotor, and Thermal Detection
- PMSM/BLDC Motor Controller
 - ARM 32-bit Cortex-M0 CPU, Frequency up to 60MHz
 - Memories Size: 16kB MTP, Internal ROM with Embedded Motor Control Library and 4kB SRAM
 - Power Management: Deep Sleep
 - Communication Interface: I²C and UART
 - Configurable ADC Gain: x1, x4, and x8
 - 5-Channel 12-Bit ADC
 - AD0 to AD3 for Differential Mode Current Sense
 - AD6 to AD8 for System Application
 - 1-Channel Voltage Type 8-bit DAC
- Gate Driver
 - Floating Channel Designed for Bootstrap Operation
 - Sourcing/Sinking Current: 100mA/200mA
 - Built-In UVLO Functions for All Channels
 - Matched Propagation Delays for All Channels
- VQFN-32L 4x4 Package

5 Applications

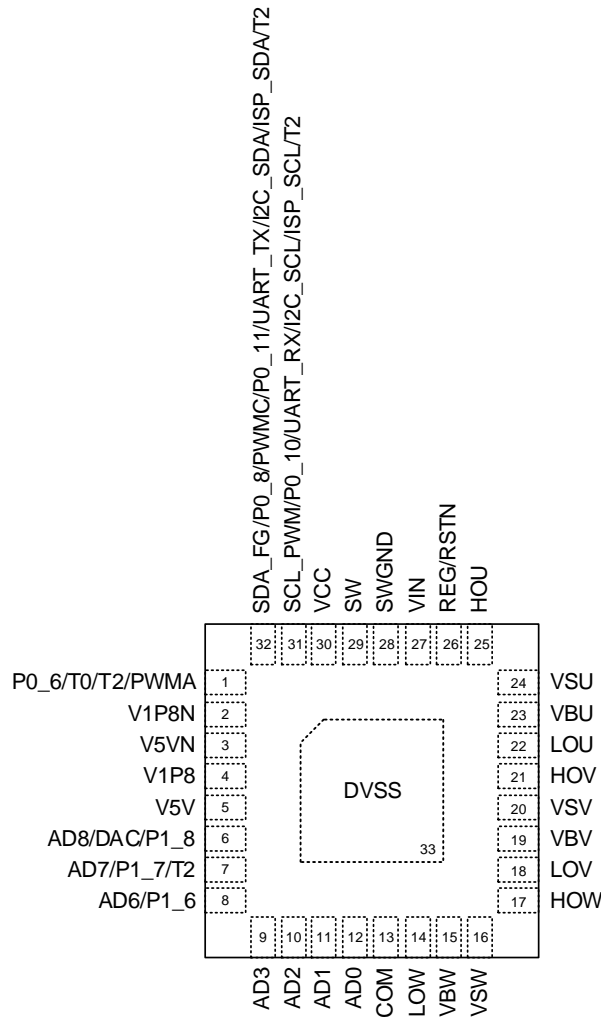
- Server Fan
- Telecom Fan
- Liquid Pump

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6 Pin Configuration

(TOP VIEW)



VQFN-32L 4x4

7 Functional Pin Description

Pin No.	Pin Name	Type	Pin Function
1	P0_6	DI/DO	Pin 6 of GPIO port 0.
	T0	DI	T0 external enable or external clock input pin.
	T2	DI	T2 external enable or external clock input pin.
	PWMA	DO	Programmable PWMA output pin.
2	V1P8N	GND	Digital ground.
3	V5VN	GND	Analog ground.
4	V1P8	P	1.8V power pin.
5	V5V	P	5V power pin.

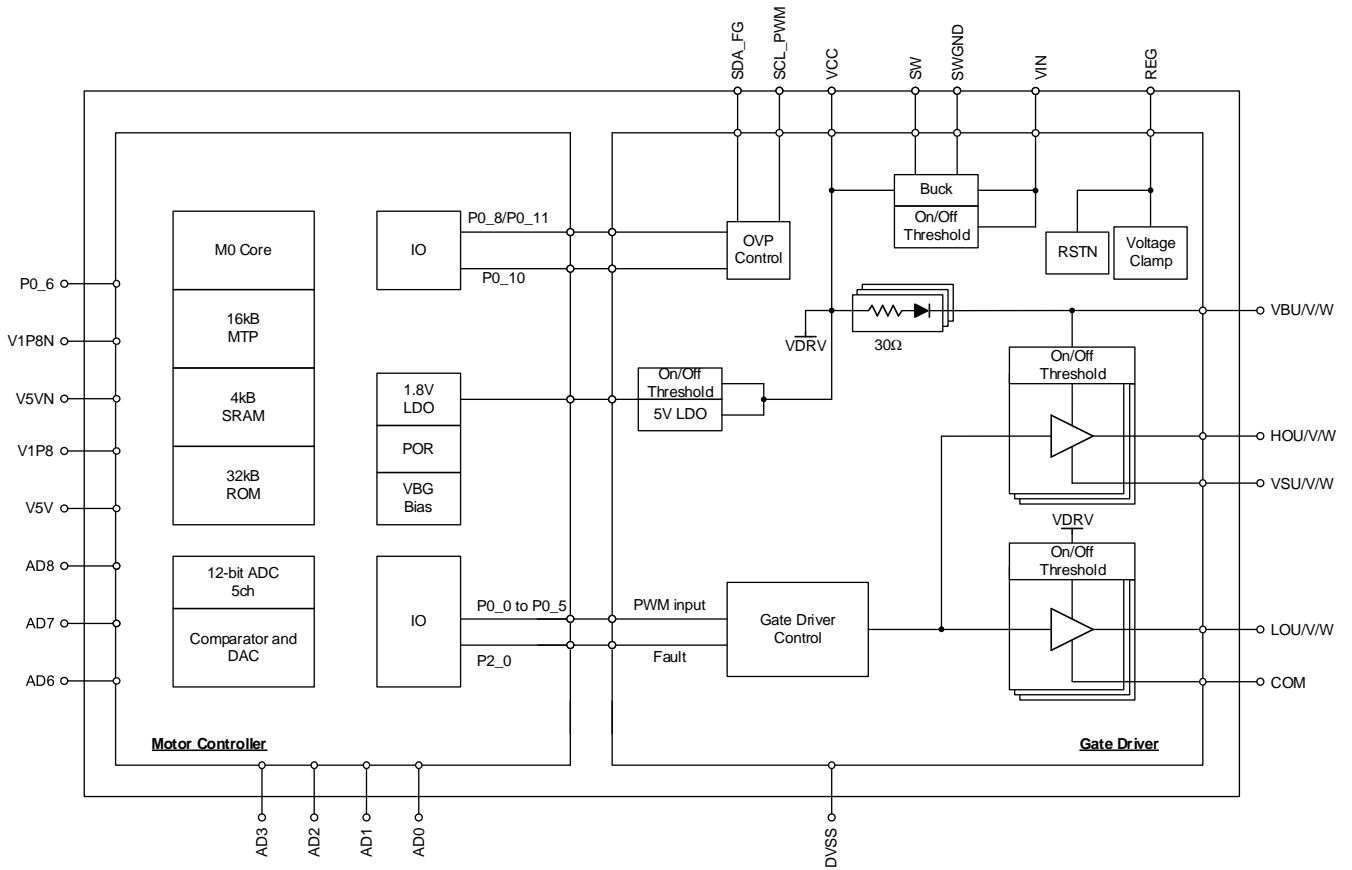
Pin No.	Pin Name	Type	Pin Function
6	AD8	AI	ADC channel 8 input pin.
	DAC	AO	Voltage type DAC output pin.
	P1_8	DI/DO	Pin 8 of GPIO port 1.
7	AD7	AI	ADC channel 7 input pin.
	P1_7	DI/DO	Pin 7 of GPIO port 1.
	T2	DI	T2 external enable or external clock input pin.
8	AD6	AI	ADC channel 6 input pin.
	P1_6	DI/DO	Pin 6 of GPIO port 1.
9	AD3	AI	ADC channel 3 for differential negative input pin only.
10	AD2	AI	ADC channel 2 for differential positive input pin only.
11	AD1	AI	ADC channel 1 for differential negative input pin only.
12	AD0	AI	ADC channel 0 for differential positive input pin only.
13	COM	GND	Gate driver ground.
14	LOW	VO	Low-side gate output of Phase W.
15	VBW	VI	High-side floating supply voltage of Phase W.
16	VSW	VI	High-side floating supply offset voltage of Phase W.
17	HOW	VO	High-side gate output of Phase W.
18	LOV	VO	Low-side gate output of Phase V.
19	VBV	VI	High-side floating supply voltage of Phase V.
20	VSV	VI	High-side floating supply offset voltage of Phase V.
21	HOV	VO	High-side gate output of Phase V.
22	LOU	VO	Low-side gate output of Phase U.
23	VBU	VI	High-side floating supply voltage of Phase U.
24	VSU	VI	High-side floating supply offset voltage of Phase U.
25	HOU	VO	High-side gate output of Phase U.
26	REG	VO	Voltage clamber for regulator with NMOS.
	RSTN	DI	Low active reset pin.
27	VIN	VI	Buck input voltage.
28	SWGND	GND	Buck ground.
29	SW	VO	Switch node of buck converter.
30	VCC	VI	Controller and gate driver supply voltage.
31	SCL_PWM	DI/DO	Clock pin with overvoltage protection.
	P0_10	DI/DO	Pin 10 of GPIO port 0.
	UART_RX	DI	UART receiving pin.
	I2C_SCL	DI/DO	I ² C clock pin.
	ISP_SCL	DI	In system programming clock pin.
	T2	DI	T2 external enable or external clock input pin.

Pin No.	Pin Name	Type	Pin Function
32	SDA_FG	DI/DO	Data pin with overvoltage protection.
	P0_8	DI/DO	Pin 8 of GPIO port 0.
	PWMC	DO	Programmable PWMC output pin.
	P0_11	DI/DO	Pin 11 of GPIO port 0.
	UART_TX	DO	UART transmitting pin.
	I2C_SDA	DI/DO	I ² C data pin.
	ISP_SDA	DI/DO	In system programming data pin.
	T2	DI	T2 external enable or external clock input pin.
33 (Exposed Pad)	DVSS	GND	Digital ground.

7.1 IO Type Definition

- DI: Digital Input Pin
- DO: Digital Output Pin
- AI: Analog Input Pin
- AO: Analog Output Pin
- P: Power Pin
- VI: Voltage Input Pin
- VO: Voltage Output Pin

8 Functional Block Diagram



9 Absolute Maximum Ratings

(Note 2)

• Controller and Driver Supply Voltage, VCC -----	-0.3V to 15V
• Controller Supply Voltage, V5V -----	-0.3V to 6.5V
• Controller Supply Voltage, V1P8-----	-0.3V to 2.5V
• Buck Supply Voltage, VIN -----	-0.3V to 66V
• Buck Switch Voltage, SW-----	-0.3V to 66V
• Clamping Voltage, REG-----	-0.3V to 50V
• VSU, VSV, VSW to COM -----	-2V to 90V
• VSU, VSV, VSW to COM (Transient, 2 μ s) -----	-5V to 90V
• VBU, VBV, VBW to COM -----	-0.3V to 100V
• HOU, HOV, HOW to COM -----	-0.3V to 100V
• LOU, LOV, LOW to COM -----	-0.3V to 15V
• VSU, VSV, VSW dv/dt-----	\pm 5V/ns
• SDA_FG, SCL_PWM to DVSS-----	-0.3V to 90V
• Voltage of I/O Pin, P0_6 -----	-0.3V to 6.5V
• Analog Input Voltage (AD0 to AD3) -----	-5V to 11V
• Analog Input Voltage (AD6 to AD8) -----	-0.3V to 6.5V
• Power Dissipation, P _D @ T _A = 25°C	
VQFN-32L 4x4 -----	3.59W
• Package Thermal Resistance (Note 3)	
VQFN-32L 4x4, θ_{JA} -----	27.8°C/W
VQFN-32L 4x4, θ_{JC} -----	4.6°C/W
• Lead Temperature (Soldering, 10 sec.) -----	260°C
• Junction Temperature -----	150°C
• Storage Temperature Range -----	-65°C to 150°C
• ESD Susceptibility (Note 4)	
HBM (Human Body Model)-----	2kV

Note 2. Stresses beyond those listed under “Absolute Maximum Ratings” may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions may affect device reliability.

Note 3. θ_{JA} is measured under natural convection (still air) at T_A = 25°C with the component mounted on a high effective-thermal-conductivity four-layer test board on a JEDEC 51-7 thermal measurement standard. θ_{JC} is measured at the bottom of the package.

Note 4. Devices are ESD sensitive. Handling precautions are recommended.

10 Recommended Operating Conditions

(Note 5)

• Controller and Driver Supply Voltage, VCC -----	9.5V to 10.5V
• Controller Supply Voltage, V5V -----	4.5V to 5.5V
• Buck Supply Voltage, VIN -----	16V to 60V
• Buck Switch Voltage, SW-----	-0.3V to 60V
• Clamping Voltage, REG-----	0V to 45V

- VSU, VSV, VSW to COM ----- -2V to 80V
- VSU, VSV, VSW to COM (Transient, 2 μ s) ----- -5V to 80V
- VBU, VBV, VBW to COM ----- 0V to 90V
- SDA_FG, SCL_PWM to DVSS ----- 0V to 20V
- Voltage of I/O Pin, P0_6 ----- 0V to 5V
- Analog Input Voltage (AD0 to AD3 and AD6 to AD8) ----- 0V to 5V
- Buck Input Capacitor on VIN ----- 0.1 μ F
- Controller and Driver Supply Voltage Capacitor on VCC ----- 4.7 μ F
- LDO Capacitor on V5V ----- 1 μ F
- LDO Capacitor on V1P8 ----- 1 μ F
- Minimum Time Period of RSTN, tRSTN ----- 100 μ s
- Ambient Temperature Range ----- -40°C to 105°C
- Junction Temperature Range ----- -40°C to 125°C

Note 5. The device is not guaranteed to function outside its operating conditions.

11 Electrical Characteristics

(V_{V5V} = 5V, V_{VIN} = 48V, V_{VCC} = 10V, T_A = 25°C, unless otherwise specified.)

Parameter	Symbol	Test Conditions	Min	Typ	Max	Unit
Clock Section						
System Frequency	fSCLK		58.8	60	61.2	MHz
Slow Clock for Sleep Mode	fLCLK		77.6	80	82.4	kHz
LDO Section						
V5V LDO Turn-On Threshold	V _{V5V_LDO_THON}		6.9	7.5	8.1	V
V5V LDO Turn-Off Threshold	V _{V5V_LDO_THOFF}		5.9	6.5	7.1	V
V5V LDO Threshold Hysteresis	V _{V5V_HYS}		--	1	--	V
V5V LDO Output Voltage	V _{V5V}		4.85	5	5.15	V
V5V Current at Operation Mode	I _{V5V_OPER}		--	20	--	mA
V5V Current at Deep Sleep Mode	I _{V5V_DSPLP}		--	1.3	--	mA
V5V LDO Current Limit	I _{V5V_OC}	V _{V5V} < 3.3V	--	15	--	mA
		V _{V5V} > 3.3V	70	--	130	
V1P8 LDO Turn-On Threshold	V _{V1P8_LDO_THON}		3.9	4.1	4.3	V
V1P8 LDO Turn-Off Threshold	V _{V1P8_LDO_THOFF}		3.4	3.6	3.8	V
V1P8 LDO Threshold Hysteresis	V _{V1P8_HYS}		--	0.5	--	V
V1P8 LDO Output Voltage	V _{V1P8}	C _{V1P8} = 1 μ F, I _{LOAD} = 40mA	1.62	1.8	1.98	V

Parameter	Symbol	Test Conditions	Min	Typ	Max	Unit
V5V Undervoltage for Interrupt Flag	VV5V_UV		4.1	4.3	4.5	V
V1P8 Undervoltage for Interrupt Flag	VV1P8_UV		--	1.5	--	V
ADC Section (0V to 3V, 12-bit, Single End Mode, Gain = 1) (Note 6)						
Minimum Conversion Voltage	VI_MIN	Code 000h	--	0	--	V
Maximum Conversion Voltage	VI_MAX	Code FFFh	--	3	--	V
ADC Offset	VOFFSET		-20	--	20	LSB
SCDAC Section (0V to 1.2V, 8-bit for Short Current) (Note 6)						
Minimum Conversion Voltage	VO_MIN	Code 00h	--	0	--	V
Maximum Conversion Voltage	VO_MIN	Code FFh	--	1.2	--	V
DAC Offset	VOFFSET		-4	--	4	LSB
VDAC Section (0V to 3V, 8-bit for General Purposed Comparator) (Note 6)						
Minimum Conversion Voltage	VO_MIN	Code 00h	--	0	--	V
Maximum Conversion Voltage	VO_MIN	Code FFh	--	3	--	V
DAC Offset	VOFFSET		-4	--	4	LSB
Output Resistance of DAC	Ro		--	5	--	kΩ
Current Limit Comparator (Short Circuit)						
Input Voltage Range of Comparator	VI_COMP		0	--	1.2	V
Comparator Offset	VOFFSET		-15	--	15	mV
General Purposed Comparator (Level Comparator)						
Input Voltage Range of Comparator	VI_COMP		0.5	--	3	V
Comparator Offset	VOFFSET		-20	--	20	mV
IO of P0_6 Section						
Input High Voltage	VIH		2.2	2.6	3	V
Input Low Voltage	VIL		1.1	1.6	2	V
Hysteresis	VH		--	1	--	V
Pull-Up Resistor	RUP		--	80	--	kΩ
Pull-Down Resistor	RDOWN		--	40	--	kΩ
High Level Output Current	IOH	@ 0.8 x V5V	--	5	--	mA
Low Level Output Current	IOl	@ 0.2 x V5V	--	10	--	mA

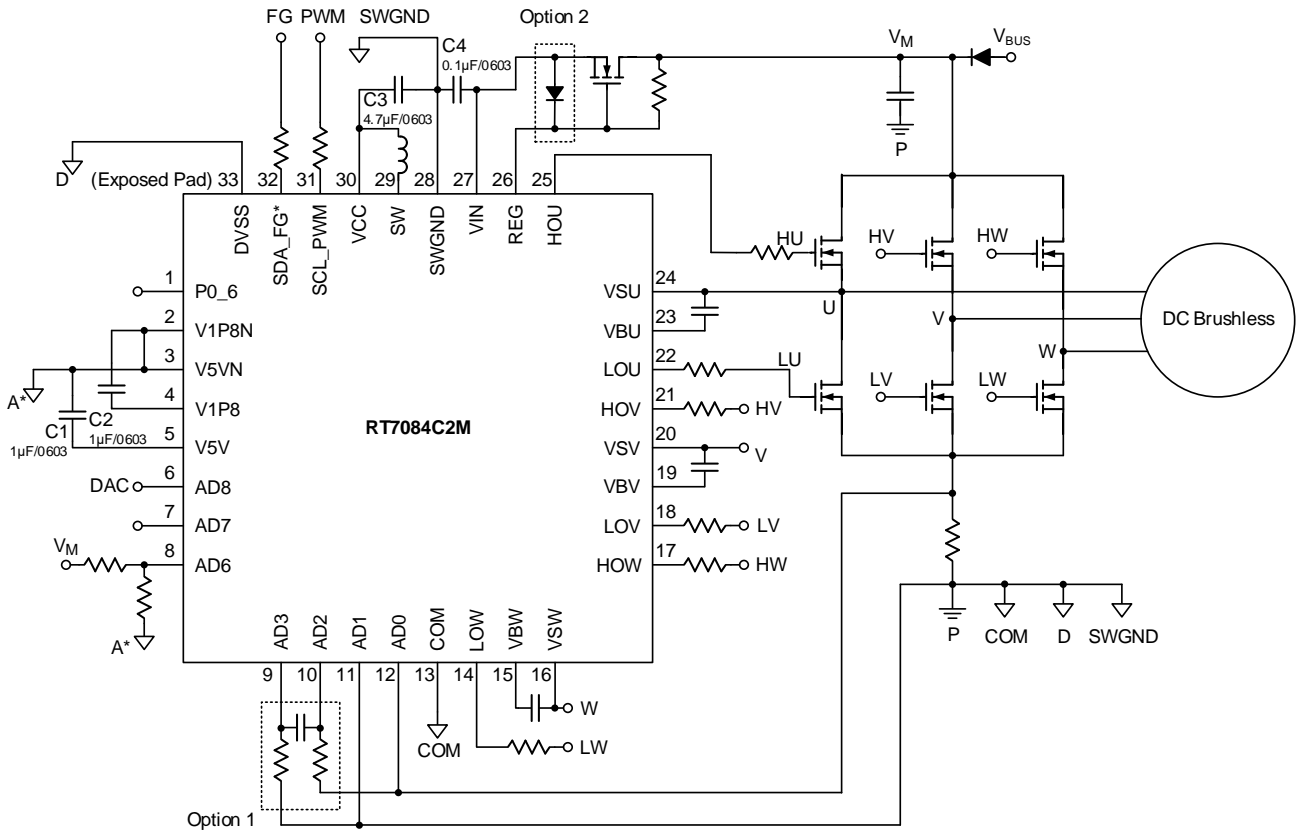
Parameter	Symbol	Test Conditions	Min	Typ	Max	Unit
IO of P1_6 to P1_8 Section						
Input High Voltage	V _{IH}		2.2	2.6	3	V
Input Low Voltage	V _{IL}		1.1	1.6	2	V
Hysteresis	V _H		--	1	--	V
Pull-Up Resistor	R _{UP}	Only P1_6/P1_7	--	80	--	kΩ
High Level Output Current	I _{OH}	@ 0.8 x V _{5V}	--	5	--	mA
Low Level Output Current	I _{OL}	@ 0.2 x V _{5V}	--	10	--	mA
IO of AD6 to AD8						
Time Constant of Input RC Filter at AD_FLT = 1	t _{AD6-8FLT=1}	Set register AD_FLT = 1, AD_FLT_ORD = 0	--	6	--	μs
Time Constant of Input RC Filter at AD_FLT = 0	t _{AD6-8FLT=0}	Set register AD_FLT = 0, AD_FLT_ORD = 0	--	1	--	μs
Bias Current at AD8	I _{AD8_BIAS}		95	100	105	μA
I²C Interface (t_{sys} = 1/ f_{SCLK})						
I ² C Clock Cycle Time	t _{SCL}		t _{sys} x 80	--	--	ns
I ² C Start Bit Setup Time	t _{START}		--	t _{SCL} / 2	--	ns
I ² C Stop Bit Setup Time	t _{STOP}		--	t _{SCL} / 2	--	ns
I ² C Data Setup Time	t _{SETUP}		--	t _{sys}	--	ns
I ² C Data Hold Time	t _{HOLD}		--	t _{sys}	--	ns
SDA/FG and SCL/PWM Section						
OVP Level	V _{OVP}		24	--	34	V
Input High Voltage	V _{IH}		2.2	2.6	3	V
Input Low Voltage	V _{IL}		1.1	1.45	2	V
Hysteresis	V _H		--	1.15	--	V
REG Section						
Zener Clamp Voltage	V _{CLAMP}	I _{CLAMP} = 100μA	33.4	39.2	45	V
Operation Current	I _Z		--	--	180	μA
Buck Converter Section						
V _{IN} Turn-On Threshold	V _{VIN_THON}		13	14	15	V
V _{IN} Turn-Off Threshold	V _{VIN_THOFF}		12	13	14	V
V _{IN} Threshold Hysteresis	V _{VIN_HYS}		--	1	--	V
V _{IN} Operation Current	I _{VIN_OPER}		--	1.2	--	mA
Buck High-Side R _{DS(ON)}	R _{DS(ON)_H}		--	4.35	--	Ω
Buck Low-Side R _{DS(ON)}	R _{DS(ON)_L}		--	4.6	--	Ω
Buck Minimum On-Time	t _{ON_MIN}	V _{IN} - V _{CC} = 45V	62.3	65.2	68.1	ns

Parameter	Symbol	Test Conditions	Min	Typ	Max	Unit
Buck High-Side Turn-On Threshold	VZCD		--	-20	--	mV
Blanking Time for VZCD	tzCD		--	130	--	ns
Buck Output Voltage	VVCC		9.45	--	10.53	V
High-Side and Low-Side Driver Section						
High-Side Driver Turn-On Threshold	VHS_THON		4.5	5	5.5	V
High-Side Driver Turn-Off Threshold	VHS_THOFF		4	4.5	5	V
High-Side Threshold Hysteresis	VHS_HYS		--	0.5	--	V
Low-Side Driver Turn-On Threshold	VLS_THON		6.9	7.5	8.1	V
Low-Side Driver Turn-Off Threshold	VLS_THOFF		6.4	7	7.6	V
Low-Side Threshold Hysteresis	VLS_HYS		--	0.5	--	V
Driver Operating Current	IvCC_OP	PWM = 20kHz (COUT = 1nF), VV5V = 5.5V	--	--	1.7	mA
Driver Quiescent Current	IvCC_Q	LOU/V/W and HOU/V/W output low, VV5V = 5.5V	--	--	1500	μA
VBS Quiescent Current for One Channel	IbSX_Q	HOU/V/W output low	--	--	450	μA
Bootstrap Diode Forward Voltage	VD_BOOT	Id = 5mA	--	0.8	--	V
		Id = 0.1A	--	3.3	--	
Bootstrap Diode Resistance for Current Limitation	RLMT	(Note 6)	--	30	--	Ω
The Difference between Input Voltage and Output Voltage	VOH	IO = 5mA, VVBU/V/W - VHOU/V/W, VVCC - VLOU/V/W	--	100	--	mV
	VOL	IO = -5mA, VHOU/V/W - VVSU/V/W, VLOU/V/W - VCOM	--	40	--	
HOU/V/W and LOU/V/W Sourcing Current	IO+	PWM input is high, VVBSU/V/W = VVCC = 10V, VHOU/V/W = VLOU/V/W = 7V (Note 6)	--	100	--	mA
HOU/V/W and LOU/V/W Sinking Current	IO-	PWM input is low, VHOU/V/W = VLOU/V/W = 3V (Note 6)	--	200	--	mA
Turn-On Rise Time	tR	VVCC = 10V, CLOAD = 1nF (Note 6)	--	100	--	ns
Turn-Off Fall Time	tF	VVCC = 10V, CLOAD = 1nF (Note 6)	--	50	--	ns
Turn-On Propagation Delay	tON	VVCC = 10V	30	--	110	ns
Turn-Off Propagation Delay	tOFF	VVCC = 10V	30	--	110	ns
Delay Matching	MT	VVCC = 10V, VVSU/V/W = 16V	--	--	60	ns
		VVCC = 10V, VVSU/V/W = 80V	--	--	60	

Note 6. This parameter is guaranteed by design.

12 Typical Application Circuit

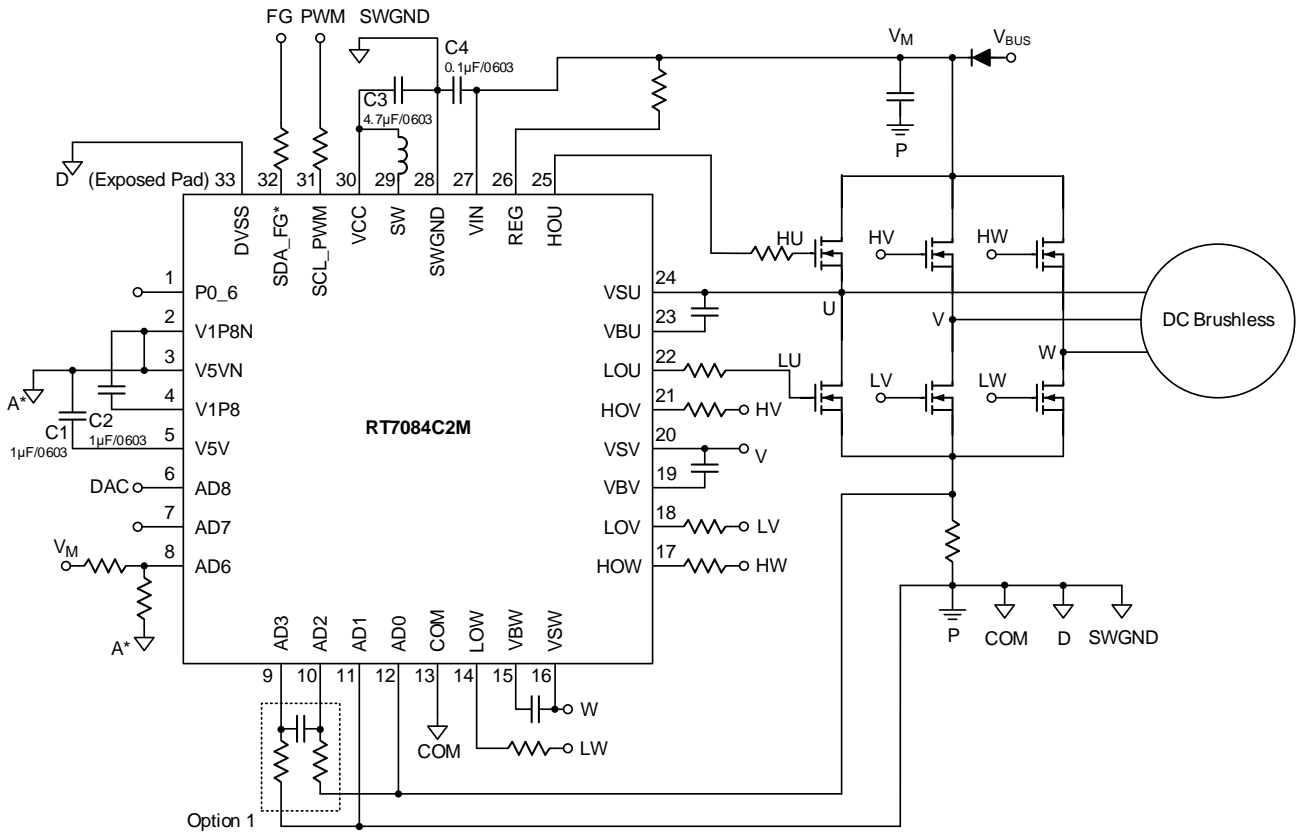
12.1 Typical Application Circuit 1 ($V_M > 60V$)



Note:

1. If C1 and C2 cannot be placed close to the IC, an additional 0.1 μ F capacitor should be placed close to the IC.
2. C3 and C4 should be placed as close as possible to the IC and SWGND.
3. A* is a separate loop, do not connect it to DVSS (D) and PGND (P).
4. SDA_FG* is a multi-function pin including P0_8 and P0_10.
5. Option 1 is for power control demand.
6. Option 2 is used to avoid damaging the MOSFET of voltage clamp during programming period.

12.2 Typical Application Circuit 2 ($V_M < 60V$)

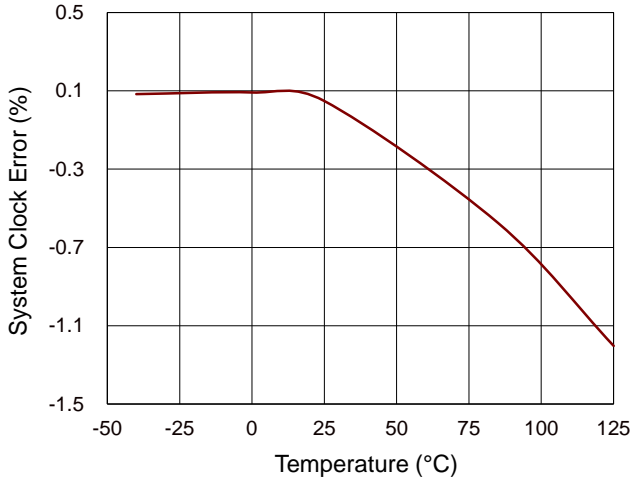


Note:

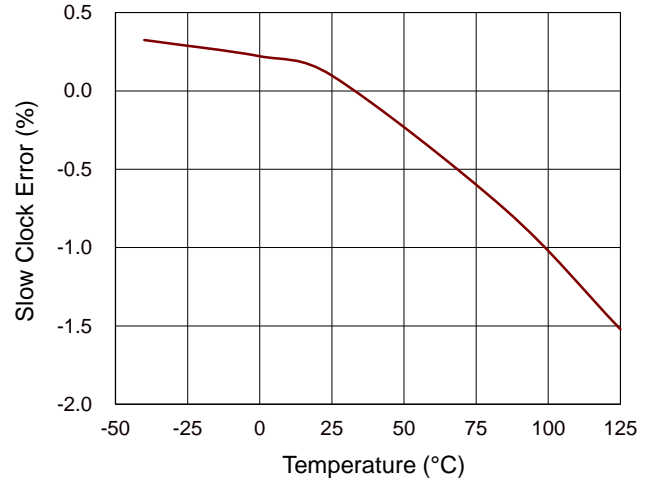
1. If C1 and C2 cannot be placed close to the IC, an additional 0.1 μ F capacitor should be placed close to the IC.
2. C3 and C4 should be placed as close as possible to the IC and SWGND.
3. A* is a separate loop, do not connect it to DVSS (D) and PGND (P).
4. SDA_FG* is a multi-function pin including P0_8 and P0_10.
5. Option 1 is for power control demand.

13 Typical Operating Characteristics

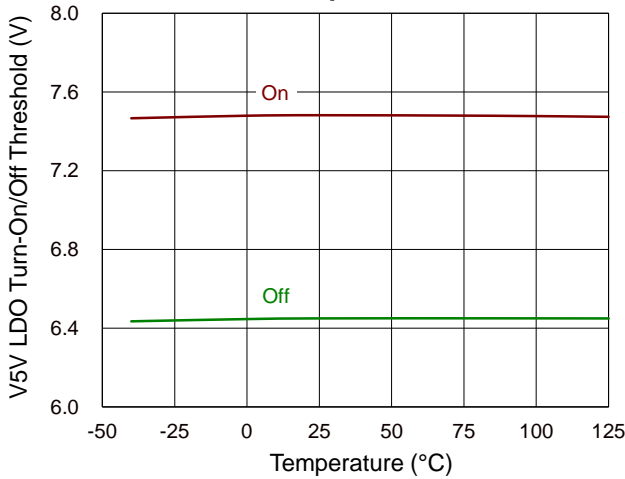
System Clock Error vs. Temperature



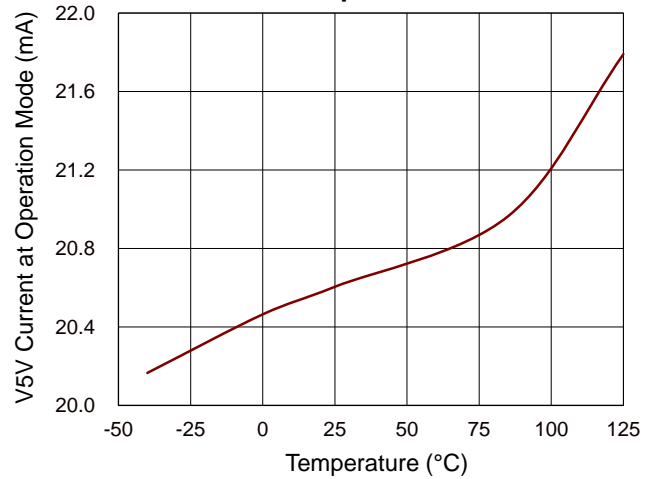
Slow Clock Error vs. Temperature



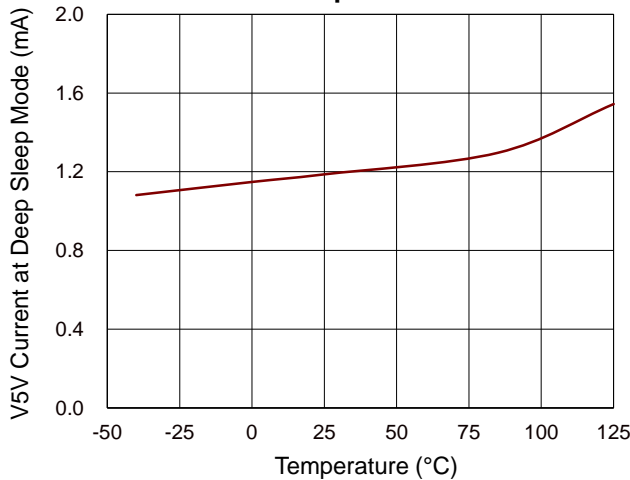
V5V LDO Turn-On/Off Threshold vs. Temperature



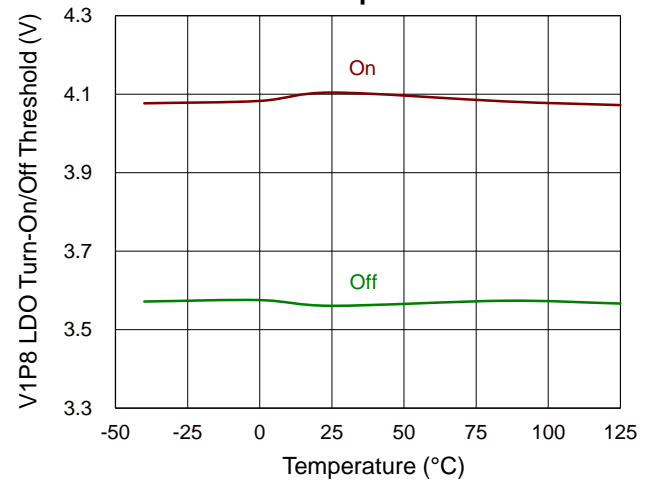
V5V Current at Operation Mode vs. Temperature



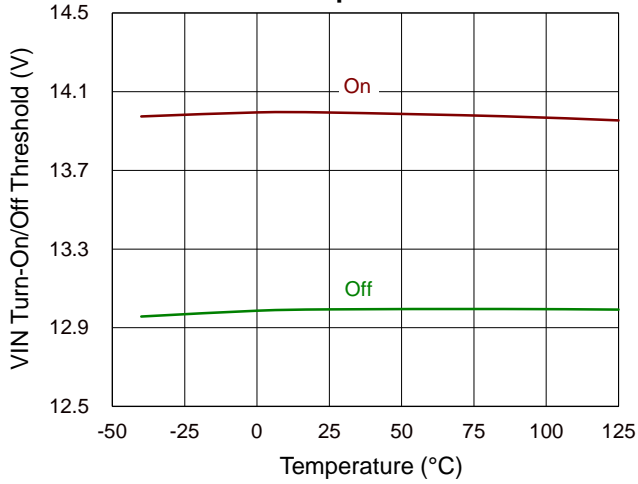
V5V Current at Deep Sleep Mode vs. Temperature



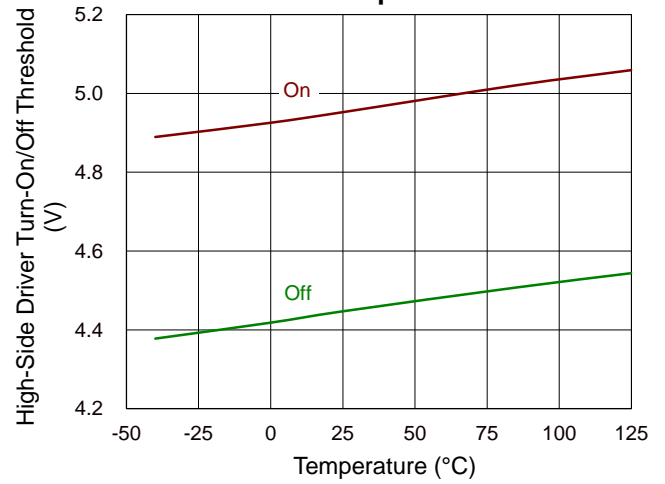
V1P8 LDO Turn-On/Off Threshold vs. Temperature



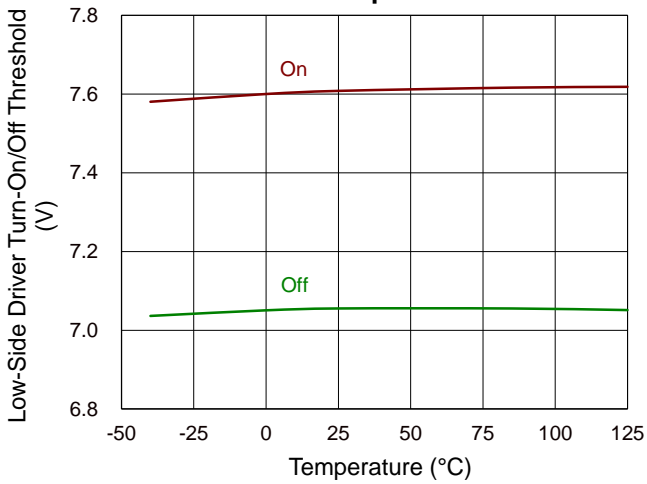
VIN Turn-On/Off Threshold vs. Temperature



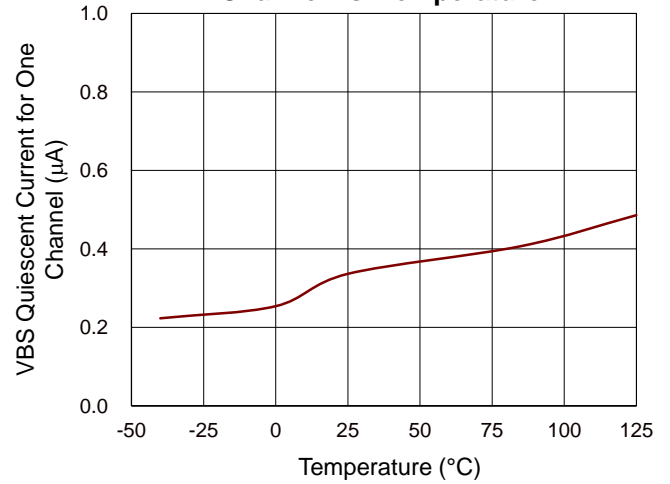
High-Side Driver Turn-On/Off Threshold vs. Temperature



Low-Side Driver Turn-On/Off Threshold vs. Temperature



VBS Quiescent Current for One Channel vs. Temperature



14 Application Information

(Note 7)

14.1 Negative Voltage of V_{phase}

When the high-side MOSFET turns off, a high di/dt current commutation is generated from the low-side MOSFET to the load during the switching transient. The fast transition slew rate with parasitic inductance induces a negative voltage spike that can be estimated with the equation $V_L = L \times di/dt$. The inductance $L = L_1 + L_2$ represents the parasitic inductance of the PCB trace and the wire bonding of the MOSFET.

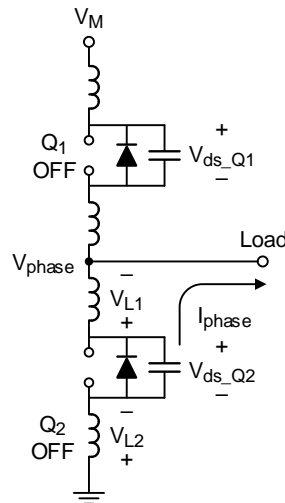


Figure 1. Negative V_{phase} Voltage Spike during Low-Side MOSFET Conducts Transition

According to the equation, higher parasitic inductance can contribute to a larger negative voltage at the V_{phase} point. To reduce the negative V_{phase} voltage, it is recommended to minimize the trace of the power loop and the distance between components. However, if the negative V_{phase} spike remains excessive, a further step can be taken as shown in [Figure 2](#). A resistance between 3.3Ω and 10Ω is recommended for R_{ext} which is placed between the VS pin and the switching node V_{phase} .

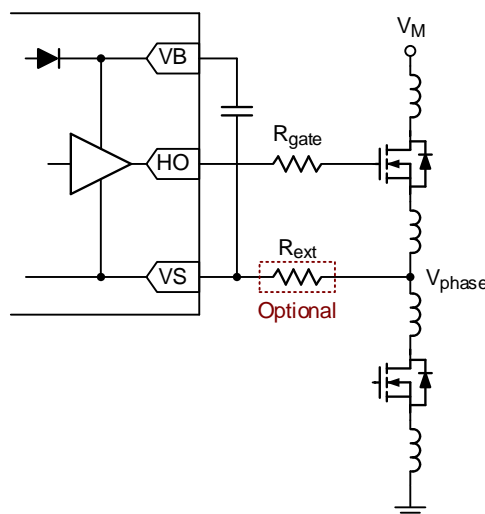


Figure 2. External Resistor between VS and V_{phase}

14.2 Buck Converter

The RT7084C2M contains a high-efficiency synchronous step-down (Buck) DC-DC converter with an input range from 16V to 60V. This Buck converter implements Constant On-Time (COT) control, adopting Boundary Conduction Mode (BCM). The values of peripheral circuit components L1, CIN, and COUT are listed in [Table 1](#) and [Table 2](#).

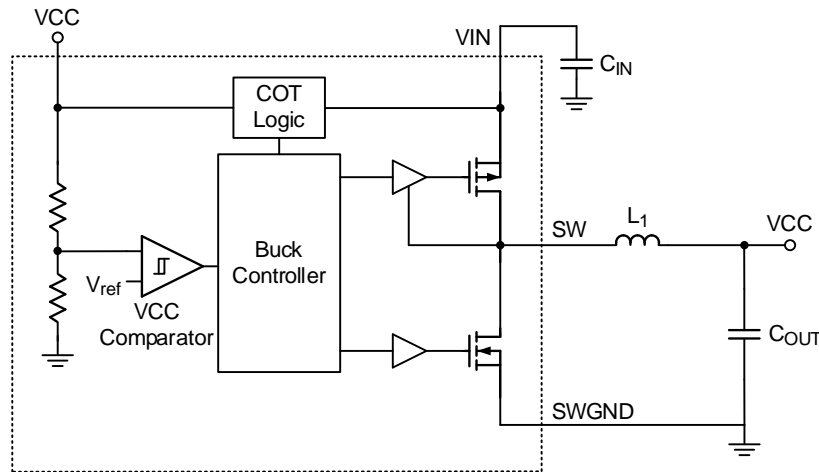


Figure 3. Buck Converter

Table 1. Recommended Value for Buck Converter with Regulator MOSFET or $V_{IN} < 45V$

L1 (μH)	10	15	22
L1 Saturation Current (mA)	>450	>300	>200
COUT (μF)	4.7		
CIN (μF)	0.1		

Table 2. Recommended Value for Buck Converter without Regulator MOSFET and $45V < V_{IN} < 60V$

L1 (μH)	10	15	22
L1 Saturation Current (mA)	>550	>400	>300
COUT (μF)	4.7		
CIN (μF)	0.1		

14.3 Regulator with Voltage Clamp Zener

When the input voltage V_M is higher than 45V, the regulator circuit (as shown in [Figure 4](#)) clamps the input voltage V_{IN} at 45V. Due to the high voltage drop across the regulator, it is important to pay attention to the power loss of the N-MOSFET and ensure proper heat dissipation. The values of peripheral circuit components C1 and R1 are listed in [Table 3](#).

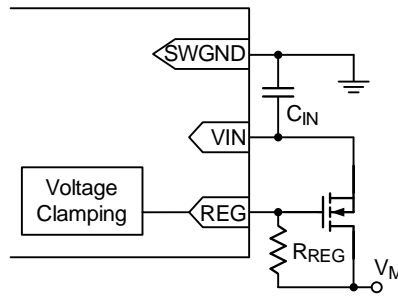


Figure 4. Regulator Circuit

Table 3. The Recommended Value for Component RREG and CIN

RREG (Ω)	270k
CIN (μF)	0.1

14.4 Thermal Considerations

The junction temperature should never exceed the absolute maximum junction temperature $T_{J(MAX)}$, listed under Absolute Maximum Ratings, to avoid permanent damage to the device. The maximum allowable power dissipation depends on the thermal resistance of the IC package, the PCB layout, the rate of surrounding airflow, and the difference between the junction and ambient temperatures. The maximum power dissipation can be calculated using the following formula:

$$P_{D(MAX)} = (T_{J(MAX)} - T_A) / \theta_{JA}$$

where $T_{J(MAX)}$ is the maximum junction temperature; T_A is the ambient temperature; and θ_{JA} is the junction-to-ambient thermal resistance.

For continuous operation, the maximum operating junction temperature indicated under Recommended Operating Conditions is 125°C. The junction-to-ambient thermal resistance, θ_{JA} , is highly package dependent. For a VQFN-32L 4x4 package, the thermal resistance, θ_{JA} , is 27.8°C/W on a standard JEDEC 51-7 high effective-thermal-conductivity four-layer test board. The maximum power dissipation at $T_A = 25^\circ\text{C}$ can be calculated as below:

$$P_{D(MAX)} = (125^\circ\text{C} - 25^\circ\text{C}) / (27.8^\circ\text{C/W}) = 3.59\text{W for a VQFN-32L 4x4 package.}$$

The maximum power dissipation depends on the operating ambient temperature for the fixed $T_{J(MAX)}$ and the thermal resistance, θ_{JA} . The derating curve in [Figure 5](#) allows the designer to estimate the effect of rising ambient temperature on the maximum power dissipation.

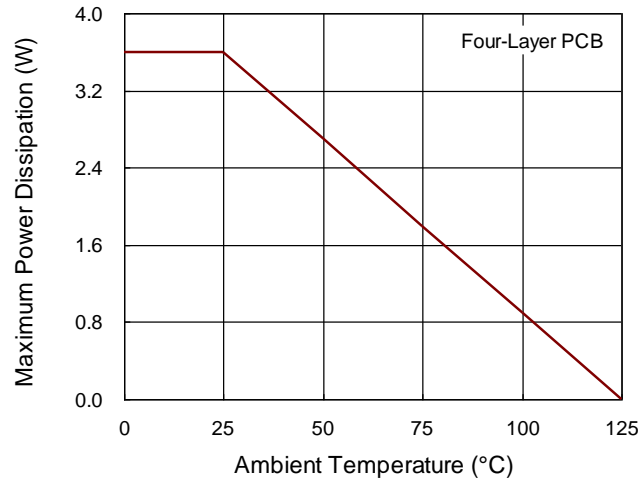
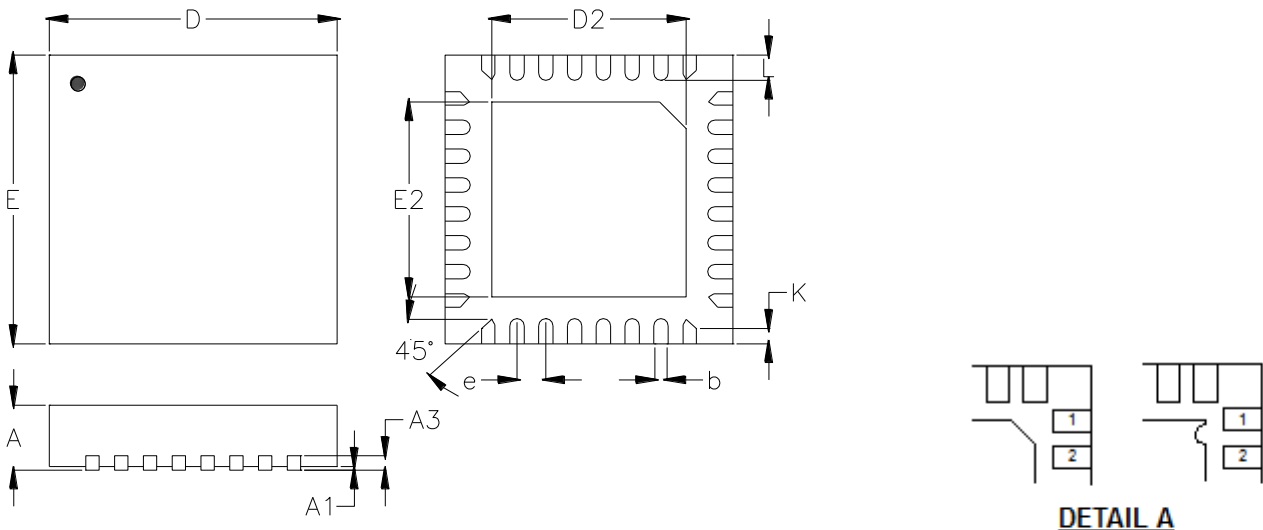


Figure 5. Derating Curve of Maximum Power Dissipation

Note 7. The information provided in this section is for reference only. The customer is solely responsible for designing, validating, and testing any applications incorporating Richtek’s product(s). The customer is also responsible for applicable standards and any safety, security, or other requirements.

15 Outline Dimension



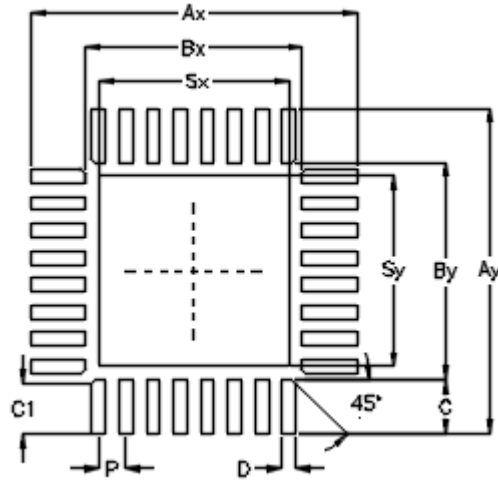
DETAIL A
Pin #1 ID and Tie Bar Mark Options

Note : The configuration of the Pin #1 identifier is optional, but must be located within the zone indicated.

Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min	Max	Min	Max
A	0.800	1.000	0.031	0.039
A1	0.000	0.050	0.000	0.002
A3	0.175	0.250	0.007	0.010
b	0.150	0.250	0.006	0.010
D	3.950	4.050	0.156	0.159
D2	2.650	2.750	0.104	0.108
E	3.950	4.050	0.156	0.159
E2	2.650	2.750	0.104	0.108
e	0.400		0.016	
L	0.300	0.400	0.012	0.016
K	0.200		0.008	

V-Type 32L QFN 4x4 Package

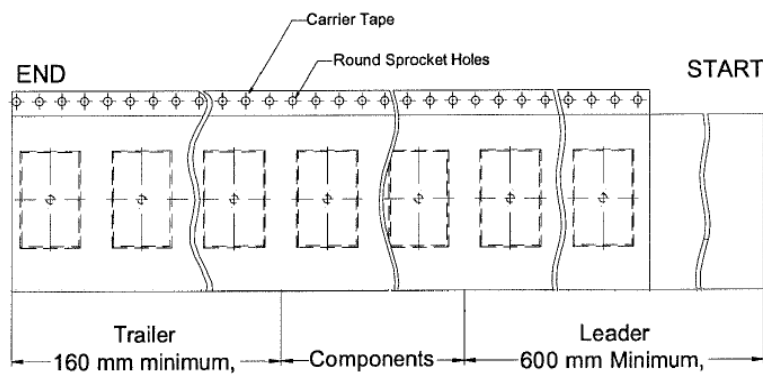
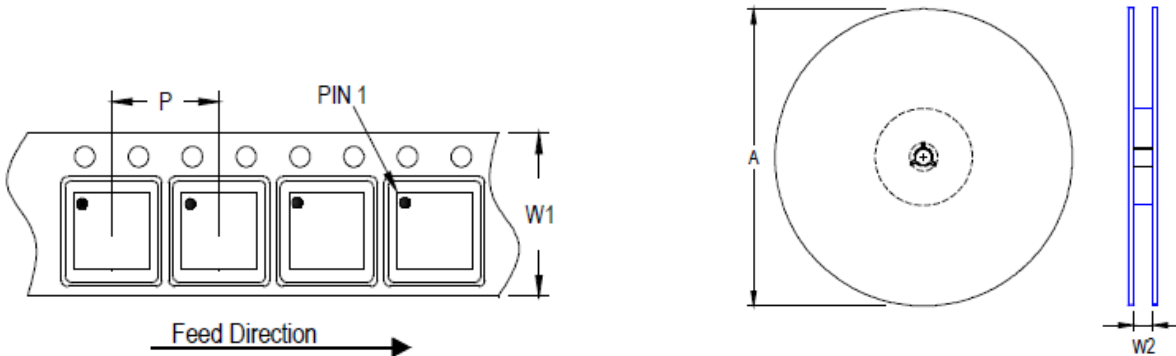
16 Footprint Information



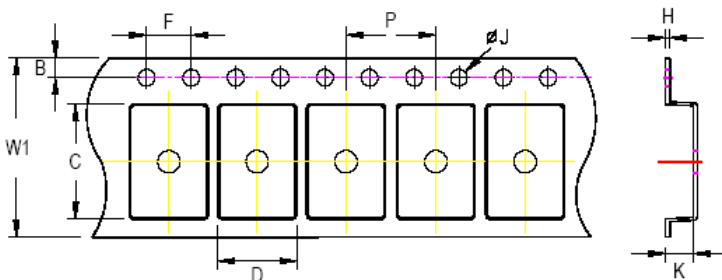
Package	Number of Pin	Footprint Dimension (mm)										Tolerance
		P	Ax	Ay	Bx	By	C*32	C1*8	D	Sx	Sy	
V/W/U/XQFN4*4-32	32	0.40	4.80	4.80	3.20	3.20	0.80	0.75	0.20	2.80	2.80	±0.05

17 Packing Information

17.1 Tape and Reel Data









Package Type	Tape Size (W1) (mm)	Pocket Pitch (P) (mm)	Reel Size (A)		Units per Reel	Trailer (mm)	Leader (mm)	Reel Width (W2) Min./Max. (mm)
			(mm)	(in)				
QFN/DFN 4x4	12	8	180	7	1,500	160	600	12.4/14.4



C, D, and K are determined by component size. The clearance between the components and the cavity is as follows:
- For 12mm carrier tape: 0.5mm max.

Tape Size	W1		P		B		F		ØJ		H
	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Max.	
12mm	12.3mm	7.9mm	8.1mm	1.65mm	1.85mm	3.9mm	4.1mm	1.5mm	1.6mm	0.6mm	

17.2 Tape and Reel Packing

Step	Photo/Description	Step	Photo/Description
1	 <p>Reel 7"</p>	4	 <p>3 reels per inner box Box A</p>
2	 <p>HIC & Desiccant (1 Unit) inside</p>	5	 <p>12 inner boxes per outer box</p>
3	 <p>Caution label is on backside of Al bag</p>	6	 <p>Outer box Carton A</p>

Container Package	Reel		Box			Carton		
	Size	Units	Item	Reels	Units	Item	Boxes	Unit
QFN/DFN 4x4	7"	1,500	Box A	3	4,500	Carton A	12	54,000
			Box E	1	1,500			

17.3 Packing Material Anti-ESD Property

Surface Resistance	Aluminum Bag	Reel	Cover tape	Carrier tape	Tube	Protection Band
Ω/cm^2	10^4 to 10^{11}	10^4 to 10^{11}	10^4 to 10^{11}	10^4 to 10^{11}	10^4 to 10^{11}	10^4 to 10^{11}

Richtek Technology Corporation14F, No. 8, Tai Yuen 1st Street, Chupei City

Hsinchu, Taiwan, R.O.C.

Tel: (8863)5526789

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18 Datasheet Revision History

Version	Date	Description	Item
00	2024/6/7	Final	