

The mini-analog series is a group of ICs that incorporate a general purpose analog circuit in a small package. The S-19610A is a CMOS type operational amplifier that has a phase compensation circuit, and operates at a low voltage with low current consumption.

The S-19610A is a dual operational amplifier (with 2 circuits).

**Caution** This product can be used in vehicle equipment and in-vehicle equipment. Before using the product in the purpose, contact to ABLIC Inc. is indispensable.

## ■ Features

- Low input offset voltage:  $V_{IO} = 6.0 \text{ mV max. (Ta = -40°C to +125°C)}$
- Operation power supply voltage range:  $V_{DD} = 2.70 \text{ V to } 5.50 \text{ V}$
- Low current consumption (Per circuit):  $I_{DD} = 1.00 \text{ } \mu\text{A typ.}$
- No external parts required for internal phase compensation
- Operation temperature range:  $Ta = -40°C \text{ to } +125°C$
- Lead-free (Sn 100%), halogen-free
- AEC-Q100 qualified<sup>\*1</sup>

\*1. Contact our sales office for details.

## ■ Applications

- Current sensing
- Signal amplification
- Buffer
- Active filter

## ■ Package

- TMSOP-8

■ Block Diagram

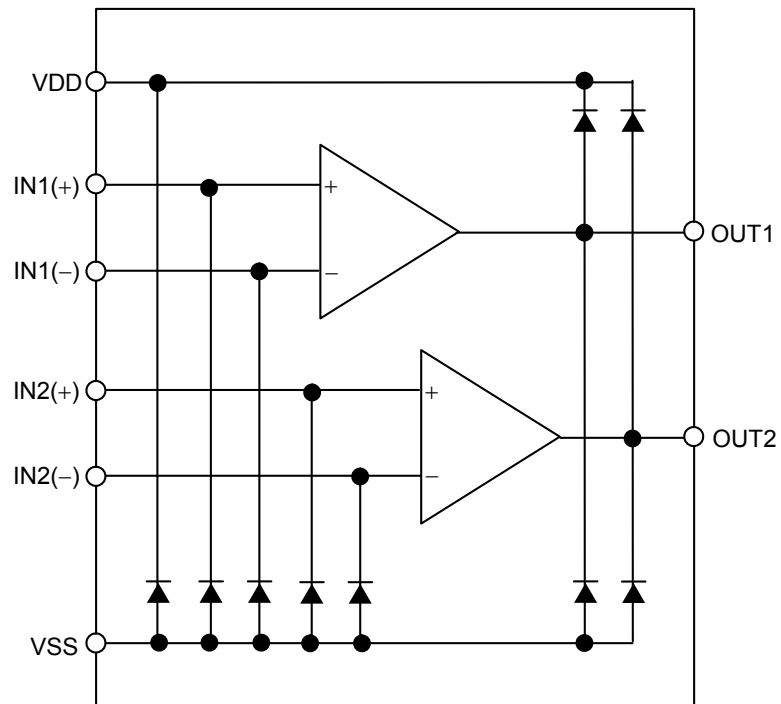


Figure 1

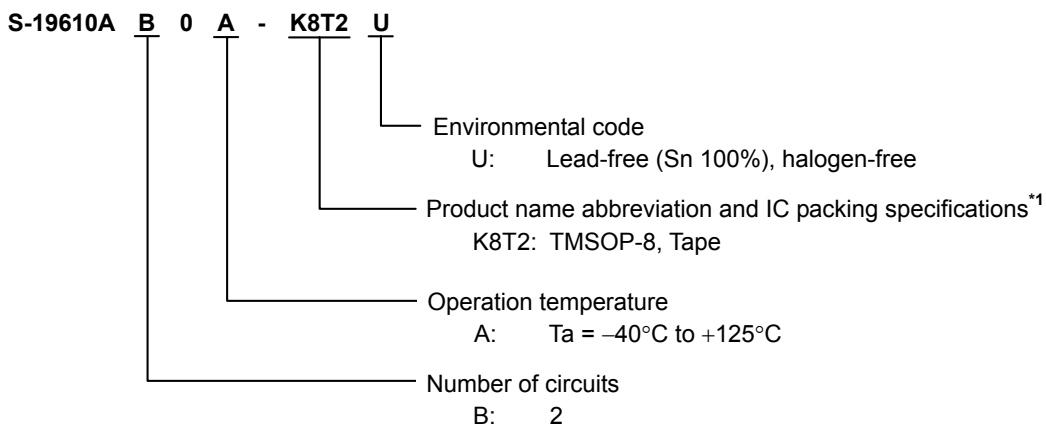
■ **AEC-Q100 Qualified**

This IC supports AEC-Q100 for the operation temperature grade 1.  
 Contact our sales office for details of AEC-Q100 reliability specification.

■ **Product Name Structure**

Refer to "1. Product name" regarding the contents of product name, "2. Package" regarding the package drawings and "3. Product name list" regarding the product type.

1. **Product name**



\*1. Refer to the tape drawing.

2. **Package**

Table 1 Package Drawing Codes

Package Name	Dimension	Tape	Reel
TMSOP-8	FM008-A-P-SD	FM008-A-C-SD	FM008-A-R-SD

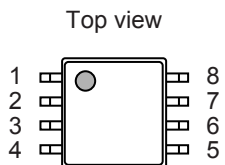
3. **Product name list**

Table 2

Product Name	Package
S-19610AB0A-K8T2U	TMSOP-8

■ Pin Configuration

1. TMSOP-8



**Figure 2**

**Table 3**

Pin No.	Symbol	Description
1	OUT1	Output pin 1
2	IN1(-)	Inverted input pin 1
3	IN1(+)	Non-inverted input pin 1
4	VSS	GND pin
5	IN2(+)	Non-inverted input pin 2
6	IN2(-)	Inverted input pin 2
7	OUT2	Output pin 2
8	VDD	Positive power supply pin

## ■ Absolute Maximum Ratings

Table 4

(Ta = +25°C unless otherwise specified)

Parameter	Symbol	Absolute Maximum Rating	Unit
Power supply voltage	V <sub>DD</sub>	V <sub>SS</sub> – 0.3 to V <sub>SS</sub> + 7.0	V
Input voltage	V <sub>IN(+)</sub> , V <sub>IN(-)</sub>	V <sub>SS</sub> – 0.3 to V <sub>SS</sub> + 7.0	V
Output voltage	V <sub>OUT</sub>	V <sub>SS</sub> – 0.3 to V <sub>DD</sub> + 0.3	V
Differential input voltage	V <sub>IND</sub>	±7.0	V
Output pin current	I <sub>SOURCE</sub>	20.0	mA
	I <sub>SINK</sub>	20.0	mA
Operating ambient temperature	T <sub>opr</sub>	–40 to +125	°C
Storage temperature	T <sub>stg</sub>	–55 to +150	°C

**Caution** The absolute maximum ratings are rated values exceeding which the product could suffer physical damage. These values must therefore not be exceeded under any conditions.

## ■ Thermal Resistance Value

Table 5

Item	Symbol	Condition	Min.	Typ.	Max.	Unit	
Junction-to-ambient thermal resistance*1	θ <sub>JA</sub>	TMSOP-8	Board 1	–	160	–	°C/W
			Board 2	–	133	–	°C/W

\*1. Test environment: compliance with JEDEC STANDARD JESD51-2A

**Remark** Refer to "■ Thermal Characteristics" for details of power dissipation and test board.

■ Electrical Characteristics

**Table 6**

DC Electrical Characteristics (V<sub>DD</sub> = 5.0 V, Ta = +25°C unless otherwise specified)

Item	Symbol	Condition	Min.	Typ.	Max.	Unit	Test Circuit
Operation power supply voltage range	V <sub>DD</sub>	–	2.70	5.00	5.50	V	–
Current consumption (2 circuits)	I <sub>DD</sub>	V <sub>CMR</sub> = V <sub>OUT</sub> = V <sub>DD</sub> / 2	–	2.00	2.50	mA	5
Input offset voltage	V <sub>IO</sub>	V <sub>CMR</sub> = V <sub>DD</sub> / 2, Ta = –40°C to +125°C	–6.0	±3.0	+6.0	mV	1
Input offset voltage drift	$\frac{\Delta V_{IO}}{\Delta T_a}$	V <sub>CMR</sub> = V <sub>DD</sub> / 2, Ta = –40°C to +125°C	–	±3	–	μV/°C	1
Input offset current	I <sub>IO</sub>	–	–	1	–	pA	–
Input bias current	I <sub>BIAS</sub>	–	–	1	–	pA	–
Common-mode input voltage range	V <sub>CMR</sub>	–	–0.1	–	3.8	V	2
Voltage gain (open loop)	A <sub>VOL</sub>	V <sub>OUT</sub> = V <sub>SS</sub> + 0.5 V to V <sub>DD</sub> – 0.5 V V <sub>CMR</sub> = V <sub>DD</sub> / 2, R <sub>L</sub> = 1.0 MΩ	88	110	–	dB	8
Maximum output swing voltage	V <sub>OH</sub>	R <sub>L</sub> = 1.0 MΩ, Ta = –40°C to +125°C	4.9	–	–	V	3
	V <sub>OL</sub>	R <sub>L</sub> = 1.0 MΩ, Ta = –40°C to +125°C	–	–	0.1	V	4
Common-mode input signal rejection ratio	CMRR	V <sub>CMR</sub> = V <sub>SS</sub> – 0.1 V to V <sub>DD</sub> – 1.2 V	70	85	–	dB	2
Power supply voltage rejection ratio	PSRR	V <sub>DD</sub> = 2.70 V to 5.50 V	70	90	–	dB	1
Source current	I <sub>SOURCE</sub>	V <sub>OUT</sub> = V <sub>DD</sub> – 0.12 V	5.0	–	–	mA	6
Sink current	I <sub>SINK</sub>	V <sub>OUT</sub> = 0.12 V	5.0	–	–	mA	7

**Table 7**

AC Electrical Characteristics (V<sub>DD</sub> = 5.0 V, Ta = +25°C unless otherwise specified)

Item	Symbol	Condition	Min.	Typ.	Max.	Unit
Slew rate	SR	R <sub>L</sub> = 1.0 MΩ, C <sub>L</sub> = 15 pF (Refer to <b>Figure 11</b> )	–	2.00	–	V/μs
Gain-bandwidth product	GBP	C <sub>L</sub> = 0 pF	–	3.00	–	MHz

■ Test Circuits (Per Circuit)

1. Power supply voltage rejection ratio, input offset voltage

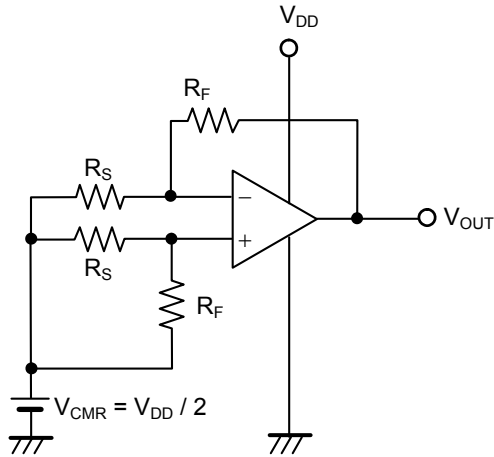


Figure 3 Test Circuit 1

• Power supply voltage rejection ratio (PSRR)

The power supply voltage rejection ratio (PSRR) can be calculated by the following expression, with  $V_{OUT}$  measured at each  $V_{DD}$ .

Test conditions:

$$V_{DD} = 2.70 \text{ V: } V_{DD} = V_{DD1}, V_{OUT} = V_{OUT1},$$

$$V_{DD} = 5.50 \text{ V: } V_{DD} = V_{DD2}, V_{OUT} = V_{OUT2}$$

$$PSRR = 20 \log \left( \left| \frac{V_{DD1} - V_{DD2}}{\left( V_{OUT1} - \frac{V_{DD1}}{2} \right) - \left( V_{OUT2} - \frac{V_{DD2}}{2} \right)} \right| \times \frac{R_F + R_S}{R_S} \right)$$

• Input offset voltage ( $V_{IO}$ )

$$V_{IO} = \left( V_{OUT} - \frac{V_{DD}}{2} \right) \times \frac{R_S}{R_F + R_S}$$

2. Common-mode input signal rejection ratio, common-mode input voltage range

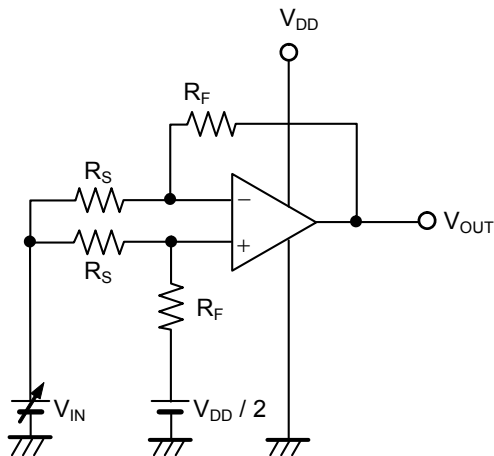


Figure 4 Test Circuit 2

• Common-mode input signal rejection ratio (CMRR)

The common-mode input signal rejection ratio (CMRR) can be calculated by the following expression, with  $V_{OUT}$  measured at each  $V_{IN}$ .

Test conditions:

$$V_{IN} = V_{CMR \text{ Max.}}: V_{IN} = V_{IN1}, V_{OUT} = V_{OUT1},$$

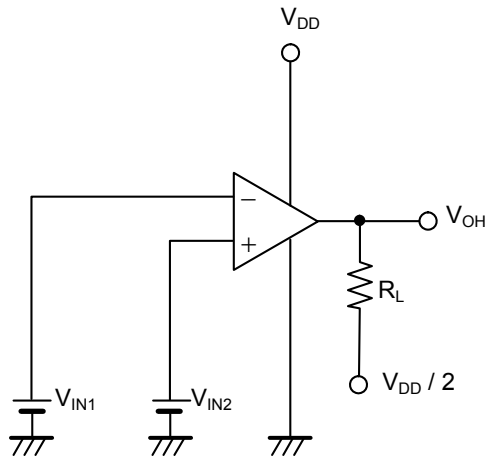
$$V_{IN} = V_{CMR \text{ Min.}}: V_{IN} = V_{IN2}, V_{OUT} = V_{OUT2}$$

$$CMRR = 20 \log \left( \left| \frac{V_{IN1} - V_{IN2}}{V_{OUT1} - V_{OUT2}} \right| \times \frac{R_F + R_S}{R_S} \right)$$

• Common-mode input voltage range ( $V_{CMR}$ )

The common mode input voltage range ( $V_{CMR}$ ) is the range of  $V_{IN}$  in which the common mode input signal rejection ratio (CMRR) is satisfied when  $V_{IN}$  is varied.

**3. Maximum output swing voltage ( $V_{OH}$ )**



**Figure 5 Test Circuit 3**

• **Maximum output swing voltage ( $V_{OH}$ )**

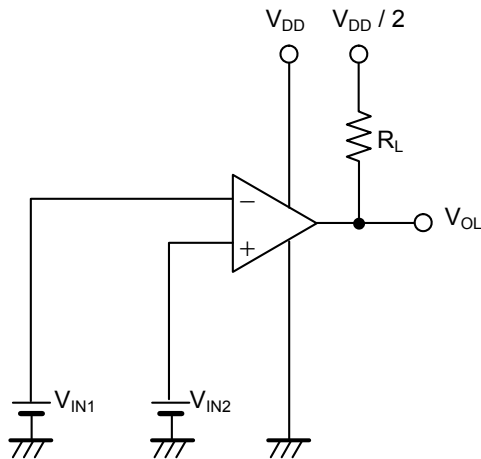
Test conditions

$$V_{IN1} = \frac{V_{DD}}{2} - 0.1 \text{ V}$$

$$V_{IN2} = \frac{V_{DD}}{2} + 0.1 \text{ V}$$

$$R_L = 1 \text{ M}\Omega$$

**4. Maximum output swing voltage ( $V_{OL}$ )**



**Figure 6 Test Circuit 4**

• **Maximum output swing voltage ( $V_{OL}$ )**

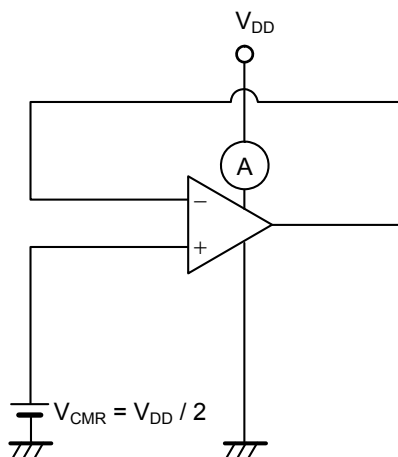
Test conditions:

$$V_{IN1} = \frac{V_{DD}}{2} + 0.1 \text{ V}$$

$$V_{IN2} = \frac{V_{DD}}{2} - 0.1 \text{ V}$$

$$R_L = 1 \text{ M}\Omega$$

**5. Current consumption**

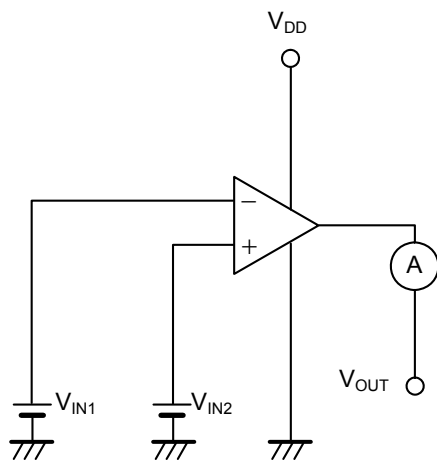


**Figure 7 Test Circuit 5**

• **Current consumption ( $I_{DD}$ )**



**6. Source current**



**Figure 8 Test Circuit 6**

• **Source current ( $I_{SOURCE}$ )**

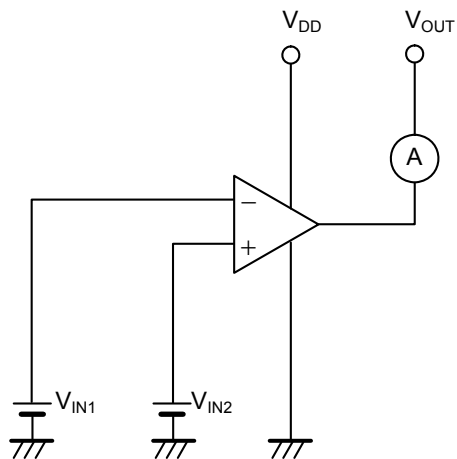
Test conditions:

$$V_{OUT} = V_{DD} - 0.12 \text{ V}$$

$$V_{IN1} = \frac{V_{DD}}{2} - 0.1 \text{ V}$$

$$V_{IN2} = \frac{V_{DD}}{2} + 0.1 \text{ V}$$

**7. Sink current**



**Figure 9 Test Circuit 7**

• **Sink current ( $I_{SINK}$ )**

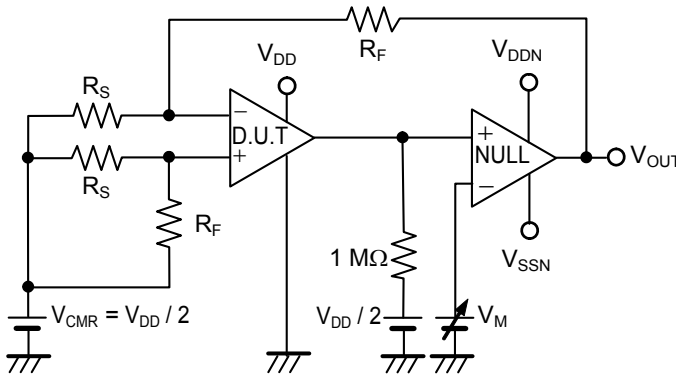
Test conditions:

$$V_{OUT} = V_{SS} + 0.12 \text{ V}$$

$$V_{IN1} = \frac{V_{DD}}{2} + 0.1 \text{ V}$$

$$V_{IN2} = \frac{V_{DD}}{2} - 0.1 \text{ V}$$

**8. Voltage gain**



**Figure 10 Test Circuit 8**

• **Voltage gain (open loop) ( $A_{VOL}$ )**

The voltage gain ( $A_{VOL}$ ) can be calculated by the following expression, with  $V_{OUT}$  measured at each  $V_M$ .

Test conditions:

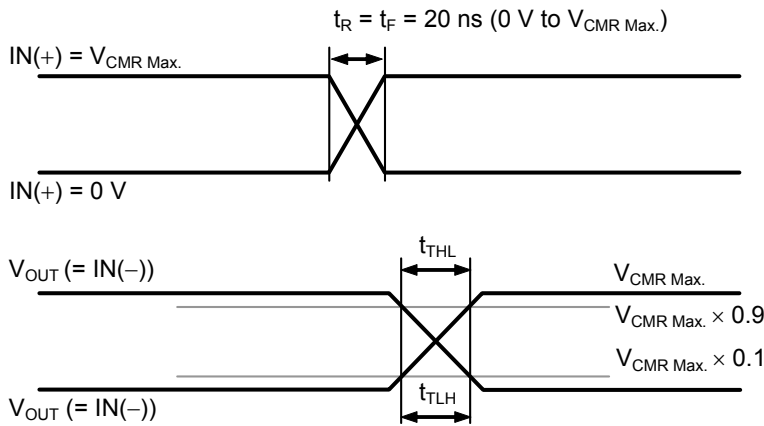
$$V_M = V_{DD} - 0.5 \text{ V}: V_M = V_{M1}, V_{OUT} = V_{OUT1},$$

$$V_M = 0.5 \text{ V}: V_M = V_{M2}, V_{OUT} = V_{OUT2}$$

$$A_{VOL} = 20 \log \left( \left| \frac{V_{M1} - V_{M2}}{V_{OUT1} - V_{OUT2}} \right| \times \frac{R_F + R_S}{R_S} \right)$$

**9. Slew rate (SR)**

Measured by the voltage follower circuit.



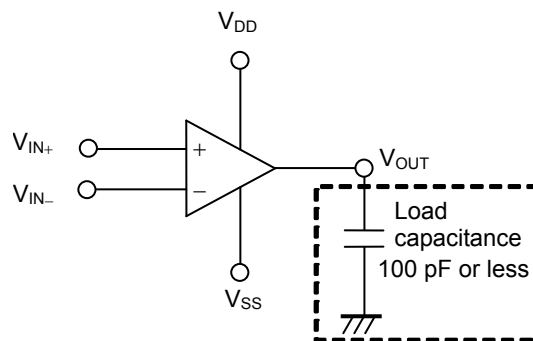
$$SR = \frac{V_{CMR \text{ Max.}} \times 0.8}{t_{TLH}}$$

$$SR = \frac{V_{CMR \text{ Max.}} \times 0.8}{t_{TLH}}$$

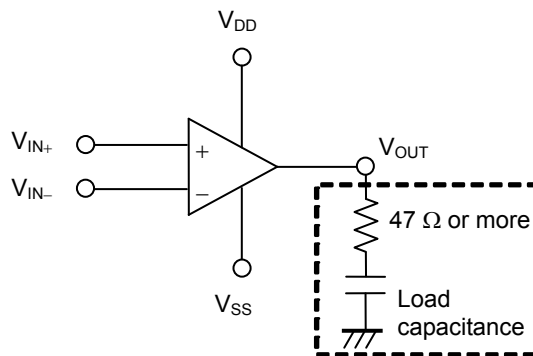
**Figure 11**

■ **Precautions**

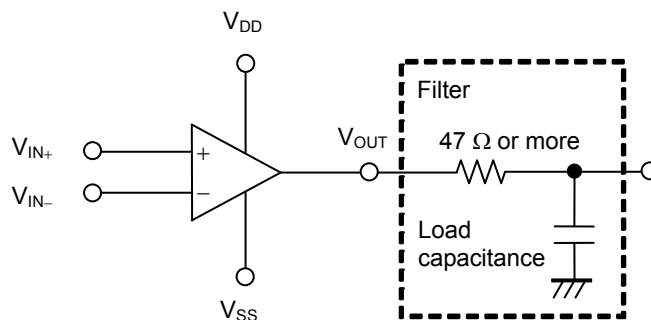
- Do not apply an electrostatic discharge to this IC that exceeds performance ratings of the built-in electrostatic protection circuit.
- ABLIC Inc. claims no responsibility for any disputes arising out of or in connection with any infringement by products including this IC of patents owned by a third party.
- Use this IC with the output current of 20 mA or less.
- This IC operates stably even directly connecting a load capacitance of 100 pF or less to the output pin, as seen in **Figure 12**. When using a load capacitance of 100 pF or larger, set a resistor of 47 Ω or more as seen in **Figure 13**. In case of connecting a filter for noise prevention, and using a load capacitance of 100 pF or more, also set a resistor of 47 Ω or more as seen in **Figure 14**.



**Figure 12**



**Figure 13**

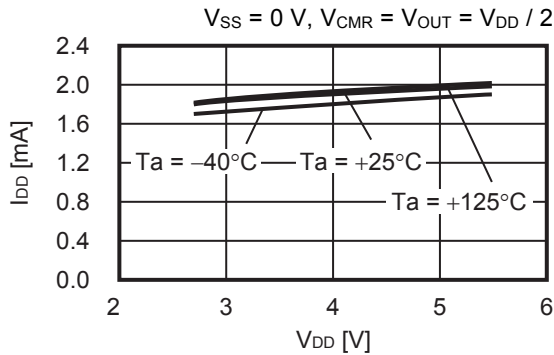


**Figure 14**

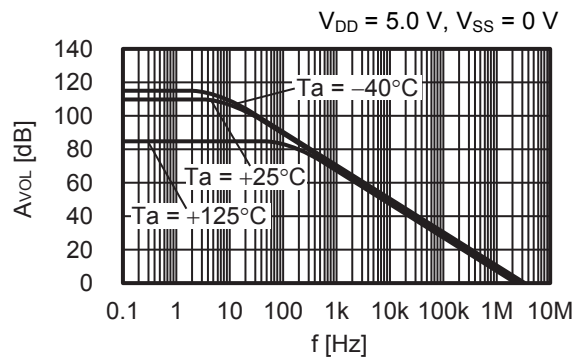
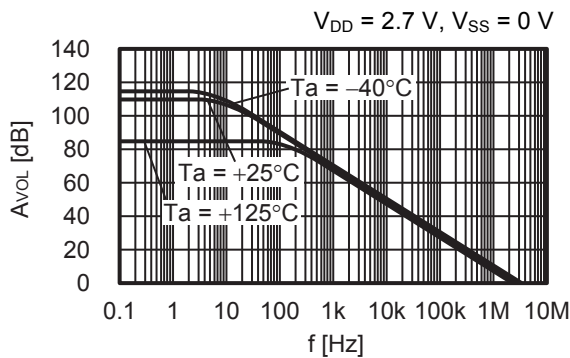
**Caution**      **The above connection diagram and constant will not guarantee successful operation. Perform through evaluation using the actual application to set the constant.**

■ **Characteristics (Typical Data)**

**1. Current consumption ( $I_{DD}$ ) (2 circuits) vs. Power supply voltage ( $V_{DD}$ )**

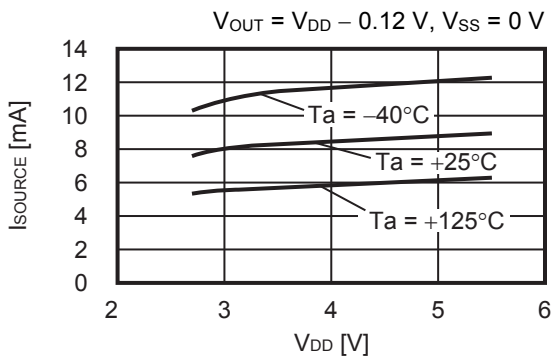


**2. Voltage gain ( $A_{VOL}$ ) vs. Frequency ( $f$ )**

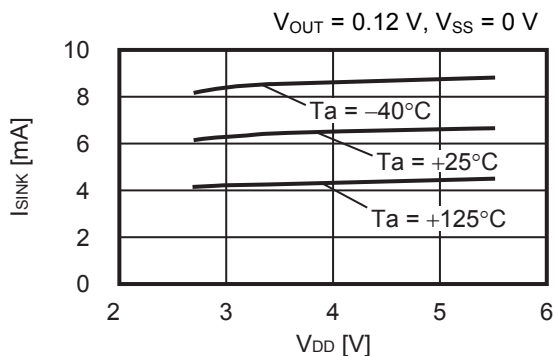


**3. Output current**

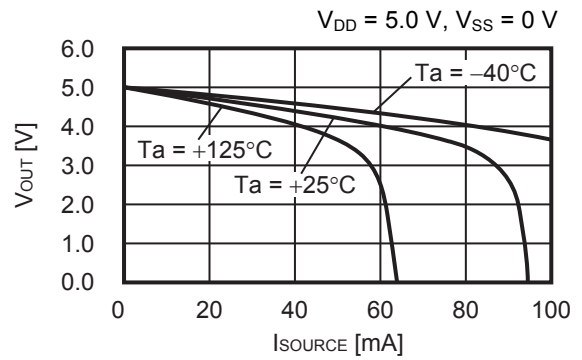
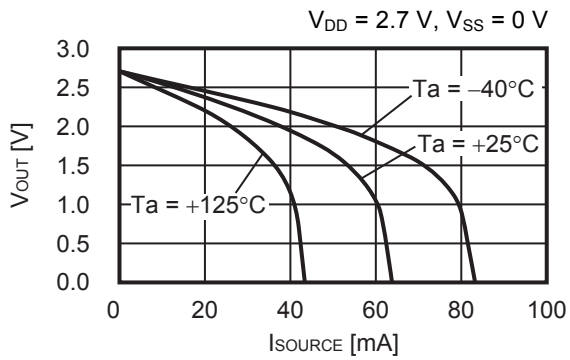
**3.1 Source current ( $I_{SOURCE}$ ) vs. Power supply voltage ( $V_{DD}$ )**



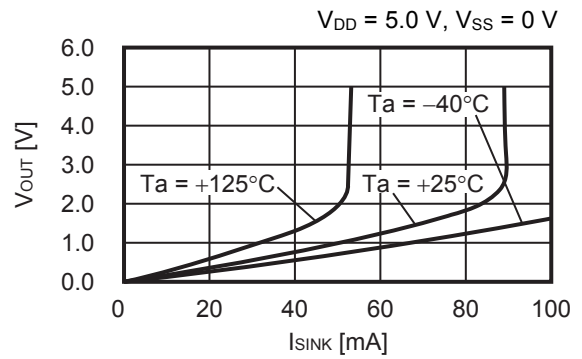
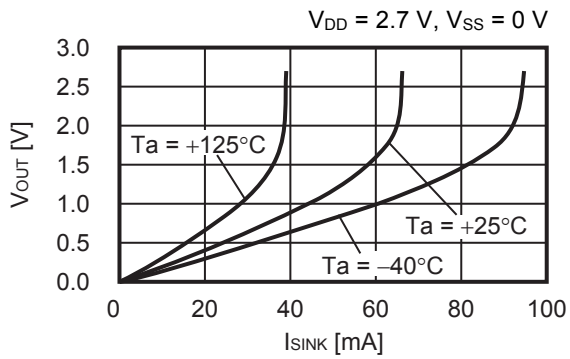
**3.2 Sink current ( $I_{SINK}$ ) vs. Power supply voltage ( $V_{DD}$ )**



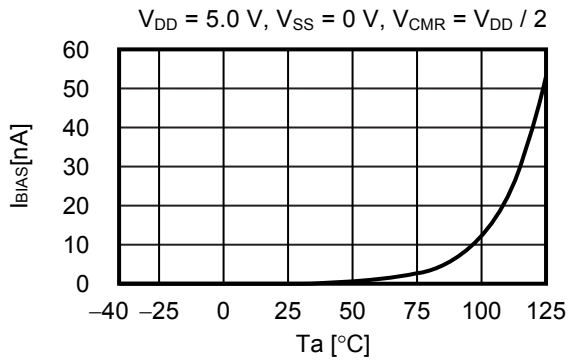
**3.3 Output voltage ( $V_{OUT}$ ) vs. Source current ( $I_{SOURCE}$ )**



**3.4 Output voltage ( $V_{OUT}$ ) vs. Sink current ( $I_{SINK}$ )**



**4. Input bias current ( $I_{BIAS}$ ) vs. Temperature ( $T_a$ )**



■ Thermal Characteristics

1. TMSOP-8

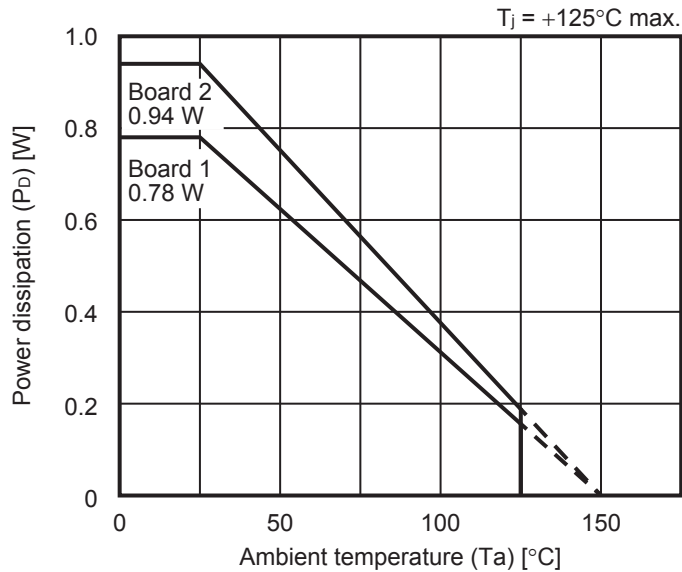


Figure 15 Power Dissipation of Package (When Mounted on Board)

1.1 Board 1

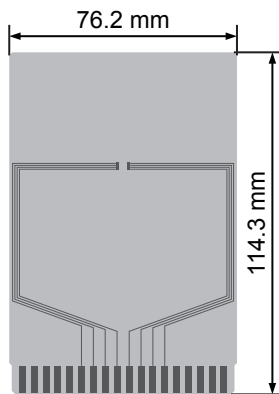


Figure 16

Table 8

Item	Specification
Thermal resistance value ( $\theta_{ja}$ )	160°C/W
Size	114.3 mm × 76.2 mm × t1.6 mm
Material	FR-4
Number of copper foil layer	2
Copper foil layer	1 Land pattern and wiring for testing: t0.070 mm
	2 -
	3 -
	4 74.2 mm × 74.2 mm × t0.070 mm
Thermal via	-

1.2 Board 2

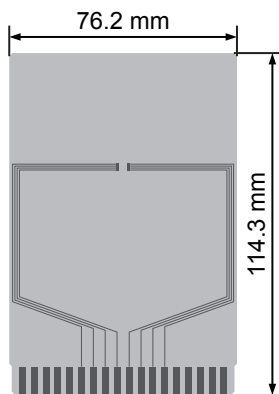


Figure 17

Table 9

Item	Specification
Thermal resistance value ( $\theta_{ja}$ )	133°C/W
Size	114.3 mm × 76.2 mm × t1.6 mm
Material	FR-4
Number of copper foil layer	4
Copper foil layer	1 Land pattern and wiring for testing: t0.070 mm
	2 74.2 mm × 74.2 mm × t0.035 mm
	3 74.2 mm × 74.2 mm × t0.035 mm
	4 74.2 mm × 74.2 mm × t0.070 mm
Thermal via	-



No. FM008-A-P-SD-1.2

TITLE	TMSOP8-A-PKG Dimensions
No.	FM008-A-P-SD-1.2
ANGLE	
UNIT	mm
<b>ABLIC Inc.</b>	



No. FM008-A-C-SD-2.0

TITLE	TMSOP8-A-Carrier Tape
No.	FM008-A-C-SD-2.0
ANGLE	
UNIT	mm
<b>ABLIC Inc.</b>	





Enlarged drawing in the central part



No. FM008-A-R-SD-1.0

TITLE	TMSOP8-A-Reel		
No.	FM008-A-R-SD-1.0		
ANGLE		QTY.	4,000
UNIT	mm		
<b>ABLIC Inc.</b>			

## Disclaimers (Handling Precautions)

1. All the information described herein (product data, specifications, figures, tables, programs, algorithms and application circuit examples, etc.) is current as of publishing date of this document and is subject to change without notice.
2. The circuit examples and the usages described herein are for reference only, and do not guarantee the success of any specific mass-production design.  
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Prior consultation with our sales office is required when considering the above uses.  
ABLIC Inc. is not responsible for damages caused by unauthorized or unspecified use of our products.
9. Semiconductor products may fail or malfunction with some probability.  
The user of the products should therefore take responsibility to give thorough consideration to safety design including redundancy, fire spread prevention measures, and malfunction prevention to prevent accidents causing injury or death, fires and social damage, etc. that may ensue from the products' failure or malfunction.  
The entire system must be sufficiently evaluated and applied on customer's own responsibility.
10. The products are not designed to be radiation-proof. The necessary radiation measures should be taken in the product design by the customer depending on the intended use.
11. The products do not affect human health under normal use. However, they contain chemical substances and heavy metals and should therefore not be put in the mouth. The fracture surfaces of wafers and chips may be sharp. Be careful when handling these with the bare hands to prevent injuries, etc.
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