

**5.5 V INPUT, 200 mA SYNCHRONOUS STEP-DOWN  
SWITCHING REGULATOR WITH 10  $\mu$ A QUIESCENT CURRENT**
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Rev.1.1\_01

The S-85V1A Series is a step-down switching regulator which features high efficiency and fast transient response. Since PWM / PFM switching control automatically switches to PFM control when under light load, high efficiency is realized. This IC is suitable for mobile devices and battery powered devices due to introduction of own distinctive COT (Constant On-Time) control.

Also, high-density mounting is realized by adopting super-small, thin SNT-6A package. Therefore, the occupancy area including a coil, an input capacitor and an output capacitor can be reduced to  $2.0 \text{ mm} \times 4.5 \text{ mm} = 9.0 \text{ mm}^2$ , and it contributes to miniaturization of electronic equipment.

**■ Features**

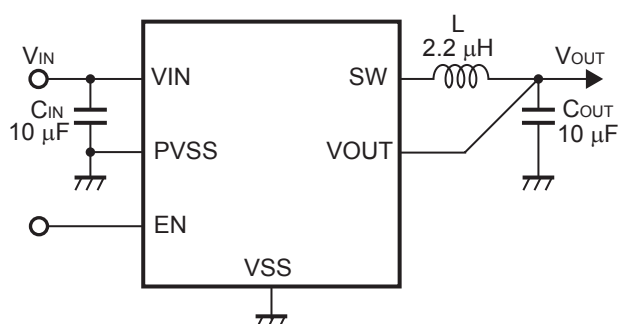
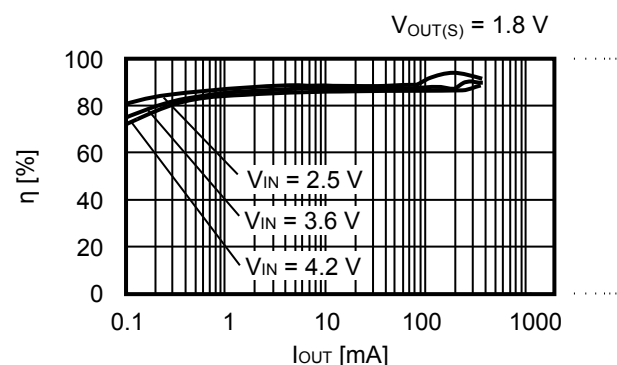
- Current consumption: 10  $\mu$ A quiescent current
- Efficiency: 93%
- Fast transient response: COT control
- Input voltage: 2.2 V to 5.5 V
- Output voltage: 0.7 V to 2.5 V, in 0.05 V step  
2.6 V to 3.9 V, in 0.1 V step
- Output voltage accuracy:  $\pm 1.5\%$  ( $1.0 \text{ V} \leq V_{\text{OUT}} \leq 3.9 \text{ V}$ )  
 $\pm 15 \text{ mV}$  ( $0.7 \text{ V} \leq V_{\text{OUT}} < 1.0 \text{ V}$ )
- Switching frequency: 1.0 MHz (at PWM operation)
- High side power MOS FET on-resistance: 450 m $\Omega$
- Low side power MOS FET on-resistance: 350 m $\Omega$
- Soft-start function: 1 ms typ.
- Under voltage lockout function (UVLO): 1.8 V typ. (detection voltage)
- Thermal shutdown function: 135°C typ. (detection temperature)
- Overcurrent limit function: 450 mA (at L = 2.2  $\mu$ H)
- Automatic recovery type short-circuit protection function: Hiccup control
- Input and output capacitors: Ceramic capacitor compatible
- Operation temperature range:  $T_a = -40^\circ\text{C}$  to  $+85^\circ\text{C}$
- Lead-free (Sn 100%), halogen-free

**■ Applications**

- Bluetooth device
- Wireless sensor network device
- Healthcare equipment
- Smart meter
- Portable game device
- Remote control

**■ Package**

- SNT-6A  
(1.80 mm  $\times$  1.57 mm  $\times$  t0.5 mm max.)

**■ Typical Application Circuit**

**■ Efficiency**


■ **Block Diagram**

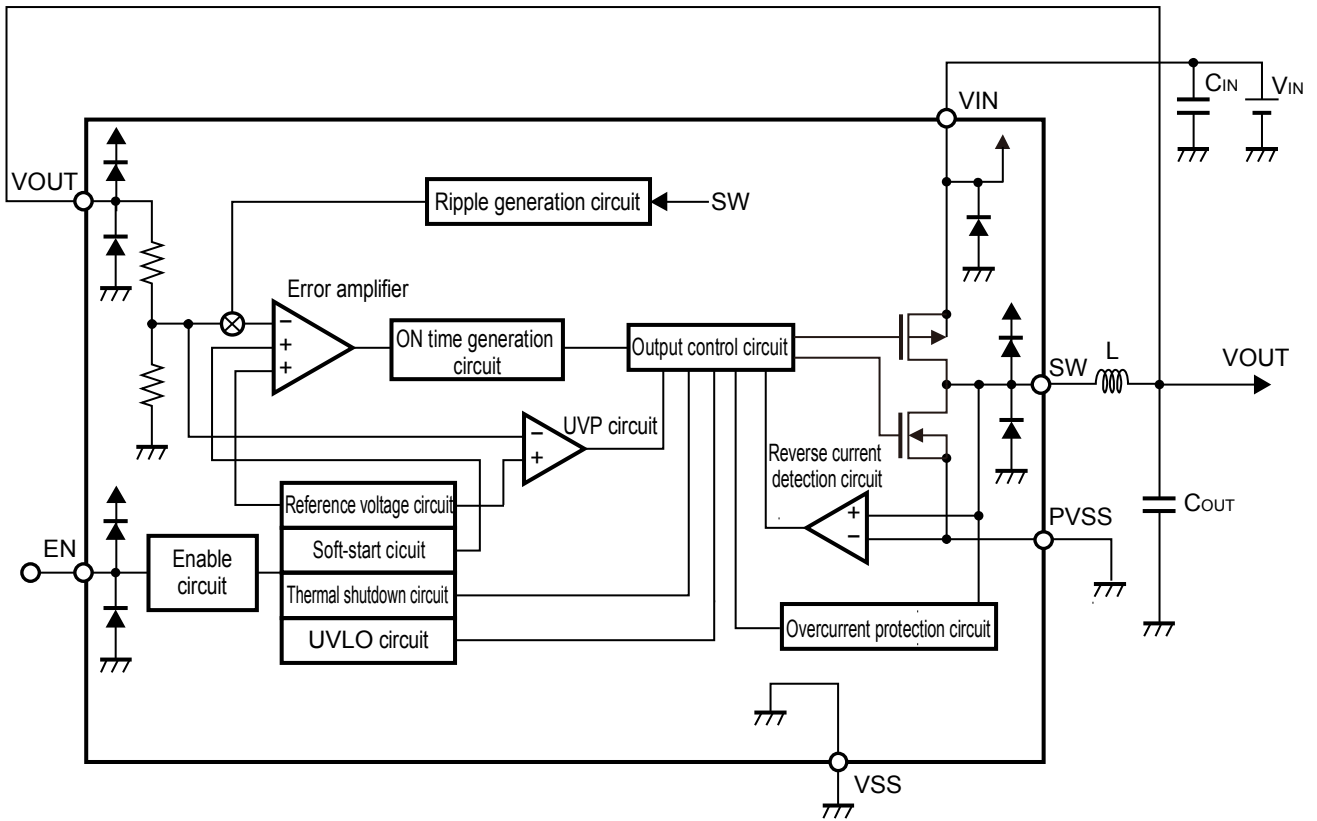
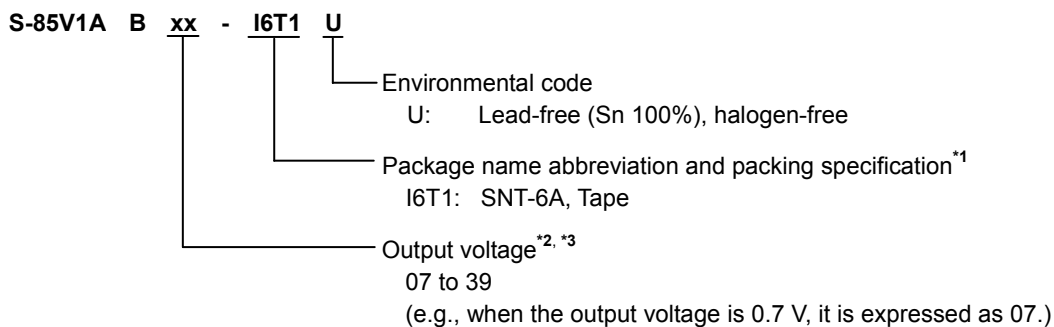


Figure 1

## ■ Product Name Structure

Users can select output voltage for the S-85V1A Series. Refer to "1. **Product name**" regarding the contents of product name, "2. **Package**" regarding the package, "3. **Product name list**" regarding details of the product name.

### 1. Product name



- \*1. Refer to the tape drawing.
- \*2. Refer to "3. **Product name list**".
- \*3. In the range from 0.7 V to 2.5 V, the products which have 0.05 V step are also available.  
Contact our sales office when the product is necessary.

### 2. Package

Table 1 Package Drawing Codes

Package Name	Dimension	Tape	Reel	Land
SNT-6A	PG006-A-P-SD	PG006-A-C-SD	PG006-A-R-SD	PG006-A-L-SD

3. Product name list

Table 2

Output Voltage ( $V_{OUT}$ )	S-85V1A Series
0.7 V $\pm$ 15 mV	S-85V1AB07-I6T1U
0.8 V $\pm$ 15 mV	S-85V1AB08-I6T1U
0.9 V $\pm$ 15 mV	S-85V1AB09-I6T1U
1.0 V $\pm$ 1.5%	S-85V1AB10-I6T1U
1.1 V $\pm$ 1.5%	S-85V1AB11-I6T1U
1.2 V $\pm$ 1.5%	S-85V1AB12-I6T1U
1.3 V $\pm$ 1.5%	S-85V1AB13-I6T1U
1.4 V $\pm$ 1.5%	S-85V1AB14-I6T1U
1.5 V $\pm$ 1.5%	S-85V1AB15-I6T1U
1.6 V $\pm$ 1.5%	S-85V1AB16-I6T1U
1.7 V $\pm$ 1.5%	S-85V1AB17-I6T1U
1.8 V $\pm$ 1.5%	S-85V1AB18-I6T1U
1.9 V $\pm$ 1.5%	S-85V1AB19-I6T1U
2.0 V $\pm$ 1.5%	S-85V1AB20-I6T1U
2.1 V $\pm$ 1.5%	S-85V1AB21-I6T1U
2.2 V $\pm$ 1.5%	S-85V1AB22-I6T1U
2.3 V $\pm$ 1.5%	S-85V1AB23-I6T1U
2.4 V $\pm$ 1.5%	S-85V1AB24-I6T1U
2.5 V $\pm$ 1.5%	S-85V1AB25-I6T1U
2.6 V $\pm$ 1.5%	S-85V1AB26-I6T1U
2.7 V $\pm$ 1.5%	S-85V1AB27-I6T1U
2.8 V $\pm$ 1.5%	S-85V1AB28-I6T1U
2.9 V $\pm$ 1.5%	S-85V1AB29-I6T1U
3.0 V $\pm$ 1.5%	S-85V1AB30-I6T1U
3.1 V $\pm$ 1.5%	S-85V1AB31-I6T1U
3.2 V $\pm$ 1.5%	S-85V1AB32-I6T1U
3.3 V $\pm$ 1.5%	S-85V1AB33-I6T1U
3.4 V $\pm$ 1.5%	S-85V1AB34-I6T1U
3.5 V $\pm$ 1.5%	S-85V1AB35-I6T1U
3.6 V $\pm$ 1.5%	S-85V1AB36-I6T1U
3.7 V $\pm$ 1.5%	S-85V1AB37-I6T1U
3.8 V $\pm$ 1.5%	S-85V1AB38-I6T1U
3.9 V $\pm$ 1.5%	S-85V1AB39-I6T1U

**Remark** Please contact our sales office for products with specifications other than the above.

■ Pin Configuration

1. SNT-6A

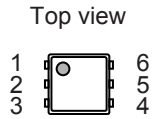


Figure 2

Table 3

Pin No.	Symbol	Description
1	VOUT	Voltage output pin
2	VSS	GND pin
3	SW	External inductor connection pin
4	PVSS	Power GND pin
5	VIN	Power supply pin
6	EN	Enable pin "H" : Enable (normal operation) "L" : Disable (standby)

■ **Absolute Maximum Ratings**

**Table 4**

(Unless otherwise specified: Ta = +25°C, V<sub>SS</sub> = 0 V)

Item	Symbol	Absolute Maximum Rating	Unit
VIN pin voltage	V <sub>IN</sub>	V <sub>SS</sub> - 0.3 to V <sub>SS</sub> + 6.0	V
EN pin voltage	V <sub>EN</sub>	V <sub>SS</sub> - 0.3 to V <sub>IN</sub> + 0.3 $\leq$ V <sub>SS</sub> + 6.0	V
VOOUT pin voltage	V <sub>OUT</sub>	V <sub>SS</sub> - 0.3 to V <sub>IN</sub> + 0.3 $\leq$ V <sub>SS</sub> + 6.0	V
SW pin voltage	V <sub>SW</sub>	V <sub>SS</sub> - 0.3 to V <sub>IN</sub> + 0.3 $\leq$ V <sub>SS</sub> + 6.0	V
PVSS pin voltage	V <sub>PVSS</sub>	V <sub>SS</sub> - 0.3 to V <sub>SS</sub> + 0.3 $\leq$ V <sub>SS</sub> + 6.0	V
Operation temperature	T <sub>opr</sub>	-40 to +85	°C
Storage temperature	T <sub>stg</sub>	-40 to +125	°C

**Caution** The absolute maximum ratings are rated values exceeding which the product could suffer physical damage. These values must therefore not be exceeded under any conditions.

■ **Thermal Resistance Value**

**Table 5**

Item	Symbol	Condition	Min.	Typ.	Max.	Unit	
Junction-to-ambient thermal resistance*1	$\theta_{ja}$	SNT-6A	Board A	-	224	-	°C/W
			Board B	-	176	-	°C/W
			Board C	-	-	-	°C/W
			Board D	-	-	-	°C/W
			Board E	-	-	-	°C/W

\*1. Test environment: compliance with JEDEC STANDARD JESD51-2A

**Remark** Refer to "■ Power Dissipation" and "Test Board" for details.

■ Electrical Characteristics

Table 6

( $V_{IN} = 3.6 \text{ V}^{*1}$ ,  $T_a = +25^\circ\text{C}$  unless otherwise specified)

Item	Symbol	Condition	Min.	Typ.	Max.	Unit
Operating input voltage	$V_{IN}$	–	2.2	3.6	5.5	V
Output voltage <sup>*2</sup>	$V_{OUT}$	$1.0 \text{ V} \leq V_{OUT} \leq 3.9 \text{ V}$ , no external parts	$V_{OUT(S)} \times 0.985$	$V_{OUT(S)}$	$V_{OUT(S)} \times 1.015$	V
		$0.7 \text{ V} \leq V_{OUT} < 1.0 \text{ V}$ , no external parts	$V_{OUT(S)} - 0.015$	$V_{OUT(S)}$	$V_{OUT(S)} + 0.015$	V
Current consumption during shutdown	$I_{SSS}$	$V_{EN} = 0 \text{ V}$	–	–	1	$\mu\text{A}$
Current consumption duringt switching off	$I_{SS1}$	$V_{OUT} = V_{OUT(S)} + 0.1 \text{ V}$ , $V_{EN} = V_{IN}$ , no external parts, no switching operation	–	10	20	$\mu\text{A}$
High level input voltage	$V_{SH}$	$V_{IN} = 2.2 \text{ V}$ to $5.5 \text{ V}$ , EN pin	1.1	–	–	V
Low level input voltage	$V_{SL}$	$V_{IN} = 2.2 \text{ V}$ to $5.5 \text{ V}$ , EN pin	–	–	0.3	V
High level input current	$I_{SH}$	$V_{IN} = 2.2 \text{ V}$ to $5.5 \text{ V}$ , EN pin, $V_{EN} = V_{IN}$	–0.1	–	0.1	$\mu\text{A}$
Low level input current	$I_{SL}$	$V_{IN} = 2.2 \text{ V}$ to $5.5 \text{ V}$ , EN pin, $V_{EN} = 0 \text{ V}$	–0.1	–	0.1	$\mu\text{A}$
High side power MOS FET on-resistance	$R_{HFET}$	$I_{SW} = 100 \text{ mA}$	–	450	–	$\text{m}\Omega$
Low side power MOS FET on-resistance	$R_{LFET}$	$I_{SW} = -100 \text{ mA}$	–	350	–	$\text{m}\Omega$
High side power MOS FET leakage current	$I_{HSW}$	$V_{IN} = 2.2 \text{ V}$ to $5.5 \text{ V}$ , $V_{EN} = 0 \text{ V}$ , $V_{SW} = 0 \text{ V}$	–	–	0.5	$\mu\text{A}$
Low side power MOS FET leakage current	$I_{LSW}$	$V_{IN} = 2.2 \text{ V}$ to $5.5 \text{ V}$ , $V_{EN} = 0 \text{ V}$ , $V_{SW} = V_{IN}$	–0.5	–	–	$\mu\text{A}$
Current limit <sup>*3</sup>	$I_{LIM}$	$L = 2.2 \mu\text{H}$	–	450	–	mA
ON time <sup>*4</sup>	$t_{ON}$	$t_{ON(S)} = 1/f_{SW}^{*5} \times V_{OUT}/V_{IN}$ , $V_{OUT} = V_{OUT(S)} \times 0.9$	$t_{ON(S)}/1.3$	$t_{ON(S)}$	$t_{ON(S)}/0.7$	ns
Minimum OFF time	$t_{OFF(MIN)}$	–	–	100	–	ns
UVLO detection voltage	$V_{UVLO-}$	When $V_{IN}$ falls	1.7	1.8	1.9	V
UVLO release voltage	$V_{UVLO+}$	When $V_{IN}$ rises	1.9	2.0	2.1	V
UVP detection voltage	$V_{UVP}$	–	–	$V_{OUT(S)} \times 0.7$	–	V
Soft-start wait time	$t_{SSW}$	Time until $V_{OUT}$ starts rising	–	1.5	–	ms
Soft-start time	$t_{SS}$	Time until $V_{OUT}$ reaches 90% after it starts rising	–	1.0	–	ms
Thermal shutdown detection temperature	$T_{SD}$	Junction temperature	–	135	–	$^\circ\text{C}$
Thermal shutdown release temperature	$T_{SR}$	Junction temperature	–	115	–	$^\circ\text{C}$

\*1.  $V_{IN} = V_{OUT(S)} + 1.0 \text{ V}$  ( $V_{OUT(S)} \geq 2.6 \text{ V}$ )

\*2.  $V_{OUT}$ : Actual output voltage  
 $V_{OUT(S)}$ : Set output voltage

\*3. The current limit changes according to the L value for the inductor to be used, input voltage, and output voltage.  
 Refer to "■ Operation" for details.

\*4.  $t_{ON}$ : Actual ON time  
 $t_{ON(S)}$ : Set ON time

\*5.  $f_{SW}$ : Switching frequency (1 MHz)

## ■ Operation

### 1. Fast transient response

Distinctive COT (Constant On-Time) control is used for DC-DC converter control.

The S-85V1A Series monitors the output voltage ( $V_{OUT}$ ) using a comparator and if  $V_{OUT}$  falls below the targeted value, the high side power MOS FET will turn on for a certain amount of time. Since the high side power MOS FET turns on and  $V_{OUT}$  rises immediately after the load current fluctuates rapidly and  $V_{OUT}$  falls, the fast transient response is realized.

The S-85V1A Series outputs ON time in proportion to  $V_{OUT}$  and in inverse proportion to power supply voltage. Therefore, when in continuous mode, even if the power supply voltage or  $V_{OUT}$  settings would change, it always operates at a quasi-fixed frequency of 1 MHz.

### 2. PWM / PFM switching control

The S-85V1A Series automatically switches between the pulse width modulation method (PWM) and pulse frequency modulation method (PFM) according to the load current. If the output current ( $I_{OUT}$ ) is large, the IC will operate using PWM control. If  $I_{OUT}$  is small, the IC using PFM control, the pulse will skip according to the load current. This reduces switching loss and improves efficiency when under light load.

The S-85V1A Series has a built-in reverse current detection circuit. The reverse current detection circuit monitors the current flowing through the inductor. If the bottom of ripple current in the inductor falls to 0 mA, the high side power MOS FET and low side power MOS FET will turn off and switching operation will stop. Switching frequency will fall from 1.0 MHz by skipping a pulse. This means that the smaller  $I_{OUT}$  is, the more the switching frequency ( $f_{SW}$ ) will drop, and it reduces switching loss.

### 3. EN pin

This pin starts and stops switching operation. When the EN pin is set to "L", the operation of all internal circuits, including the high side power MOS FET, is stopped, reducing current consumption. Current consumption increases when a voltage of 0.3 V to  $V_{IN} - 0.3$  V is applied to the EN pin. When not using the EN pin, connect it to the VIN pin. Since the EN pin is neither pulled down nor pulled up internally, do not use it in the floating status. The structure of the EN pin is shown in **Figure 3**.

Table 7

EN Pin	Internal Circuit	V <sub>OUT</sub> Pin Voltage
"H"	Enable (normal operation)	$V_{OUT}^{*2}$
"L"	Disable (standby)	"High-Z"

\*1. Refer to \*2 in Table 6 in "■ Electrical Characteristics".

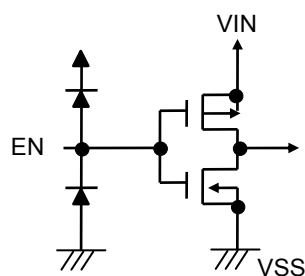


Figure 3

### 4. Under voltage lockout function (UVLO)

The S-85V1A Series has a built-in UVLO circuit to prevent the IC from malfunctioning due to a transient status at power-on or a momentary drop in the supply voltage. When UVLO status is detected, the high side power MOS FET and low side power MOS FET will turn off, and the SW pin will change to "High-Z". For this reason, switching operation will stop. The soft-start function is reset if UVLO status is detected once, and is restarted by releasing the UVLO status.

Note that the other internal circuits operate normally and the status is different from the disabled status.

Also, there is a hysteresis width for avoiding malfunctions due to generation of noise etc. in the input voltage.



### 5. Thermal shutdown function

The S-85V1A Series has a built-in thermal shutdown circuit to limit overheating. When the junction temperature increases to 135°C typ., the thermal shutdown circuit becomes the detection status, and the switching operation is stopped. When the junction temperature decreases to 115°C typ., the thermal shutdown circuit becomes the release status, and the switching operation is restarted.

If the thermal shutdown circuit becomes the detection status due to self-heating, the switching operation is stopped and output voltage ( $V_{OUT}$ ) decreases. For this reason, the self-heating is limited and the temperature of the IC decreases. The thermal shutdown circuit becomes release status when the temperature of the IC decreases, and the switching operation is restarted, thus the self-heating is generated again. Repeating this procedure makes the waveform of  $V_{OUT}$  into a pulse-like form. Switching operation stopping and starting can be stopped by either setting the EN pin to "L", lowering the output current ( $I_{OUT}$ ) to reduce internal power consumption, or decreasing the ambient temperature.

**Table 8**

Thermal Shutdown Circuit	V <sub>OUT</sub> Pin Voltage
Release: 115°C typ.*1	V <sub>OUT</sub>
Detection: 135°C typ.*1	"High-Z"

\*1. Junction temperature

### 6. Overcurrent protection function

The S-85V1A Series has a built-in current limit circuit.

The overcurrent protection circuit monitors the current that flows through the low side power MOS FET and limits current to prevent thermal destruction of the IC due to an overload, magnetic saturation in the inductor, etc.

When a current exceeding the current limit ( $I_{LIM}$ ) flows through the low side power MOS FET, the current limit circuit operates and prohibits turning on the high side power MOS FET until the current falls below the low side current limit ( $I_{LIMDET}$ ). If the value of the current that flows through the low side power MOS FET falls to the  $I_{LIMDET}$  or lower, the S-85V1A Series returns to normal operation.  $I_{LIMDET}$  is fixed at 270 mA typ. in the IC, and  $I_{LIM}$  will vary depending on the external parts to be used.

The relation between  $I_{LIM}$ , the inductor value (L), the input voltage ( $V_{IN}$ ), and the output voltage ( $V_{OUT}$ ) are shown in the following expression.

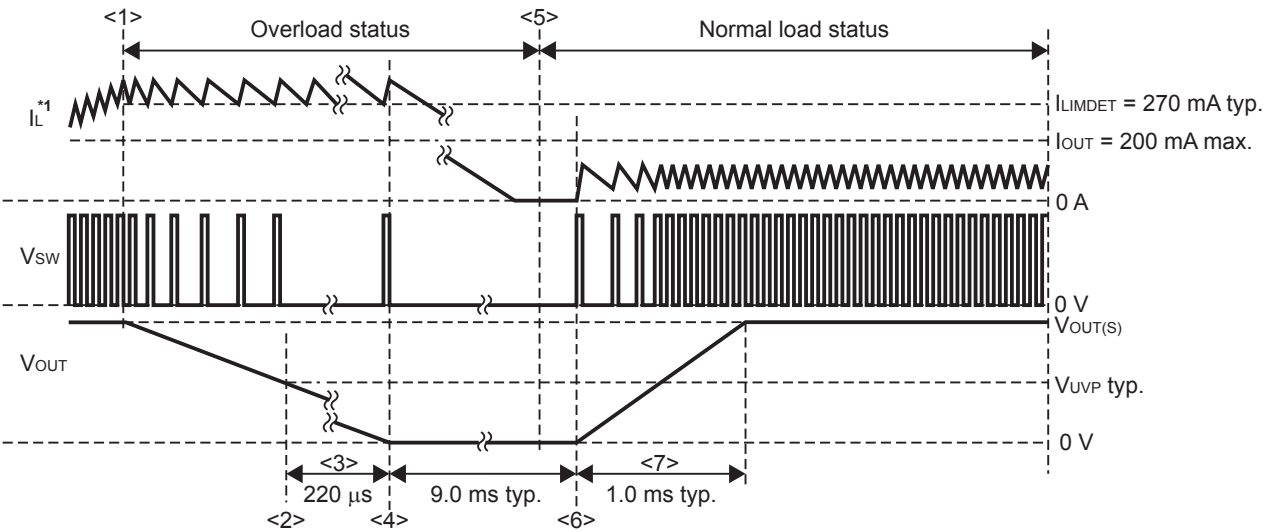
$$I_{LIM} = I_{LIMDET} + \frac{1}{2 \times L \times f_{SW}} \times \frac{(V_{IN} - V_{OUT}) \times V_{OUT}}{V_{IN}}$$

**7. Automatic recovery type short-circuit protection function (Hiccup control)**

The S-85V1A Series has a built-in automatic recovery type short-circuit protection function for Hiccup control. Hiccup control is a method for periodically carrying out automatic recovery when the IC detects overcurrent and stops the switching operation.

**7.1 When over load status is released**

- <1> Overcurrent detection
- <2> Under voltage protection circuit (UVP circuit) detects a drop in the output voltage ( $V_{OUT}$ ).
- <3> 220  $\mu$ s elapse
- <4> Switching operation stop (for 9 ms typ.)
- <5> Overload status release
- <6> The IC restarts, soft-start function starts.  
 In this case, it is unnecessary to input an external reset signal for restart.
- <7>  $V_{OUT}$  reaches  $V_{OUT(S)}$  after 1.0 ms typ. elapses.

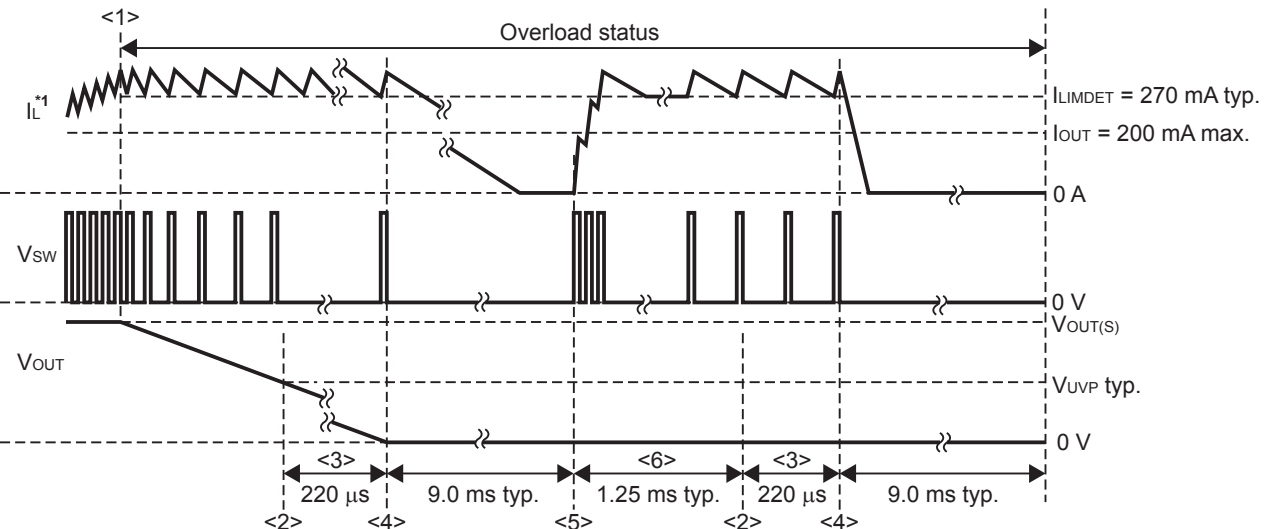


\*1. Inductor current

**Figure 4**

**7.2 When over load status continues**

- <1> Overcurrent detection
- <2> The UVP circuit detects a drop in  $V_{OUT}$ .
- <3> 220  $\mu$ s elapse
- <4> Switching operation stop (for 9 ms typ.)
- <5> The IC restarts, soft-start function starts.
- <6> The status returns to <2> when over load status continues after 1.25 ms typ. elapses.



\*1. Inductor current

**Figure 5**

### 8. Pre-bias compatible soft-start function

The S-85V1A Series has a built-in pre-bias compatible soft-start circuit.

If the pre-bias compatible soft-start circuit starts when electrical charge remains in the output voltage ( $V_{OUT}$ ) as a result of power supply restart, etc., or when  $V_{OUT}$  is biased beforehand (pre-bias status), switching operation is stopped until the soft-start voltage exceeds the internal feedback voltage, and then  $V_{OUT}$  is maintained. If the soft-start voltage exceeds the internal feedback voltage, switching operation will restart and  $V_{OUT}$  will rise to the output voltage setting value ( $V_{OUT(S)}$ ). This allows  $V_{OUT(S)}$  to be reached without lowering the pre-biased  $V_{OUT}$ .

In soft-start circuits which are not pre-bias compatible, a large current flows as a result of the discharge of the residual electric charge through the low side power MOS FET when switching operation starts, which could cause damage, however in a pre-bias compatible soft-start circuit, the IC is protected from the large current when switching operation starts, and it makes power supply design for the application circuit simpler.

In the S-85V1A Series,  $V_{OUT}$  reaches  $V_{OUT(S)}$  gradually due to the soft-start circuit.

In the following cases, rush current and  $V_{OUT}$  overshoot are reduced.

- At power-on
- When the EN pin changes from "L" to "H".
- When UVLO operation is released.
- When thermal shutdown is released.
- At short-circuit recovery

In addition, the soft-start circuit operates under the following conditions.

The soft-start circuit starts operating after "H" is input to the EN pin and the soft-start wait time ( $t_{SSW}$ ) = 1.5 ms typ. elapses. The soft-start time ( $t_{SS}$ ) is set to 1.0 ms typ.

- At power supply restart (the IC restart)
- At UVLO detection (after UVLO release)
- At thermal shutdown detection (after thermal shutdown release)
- After Hiccup control

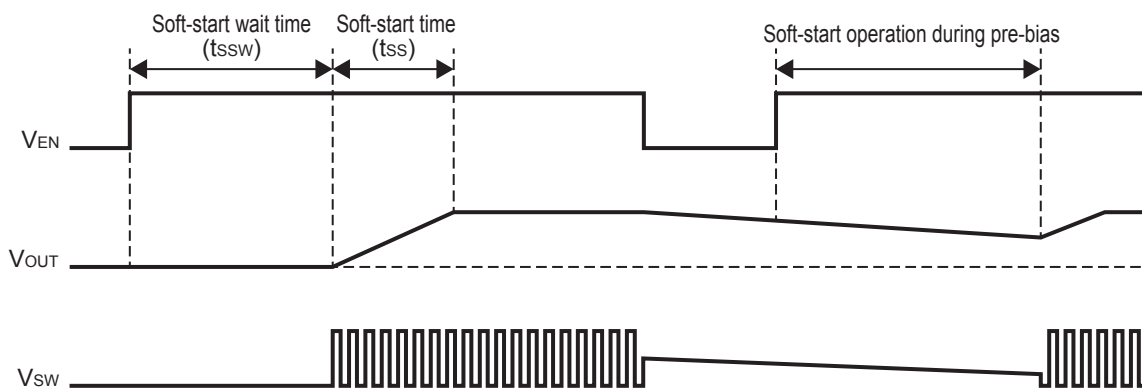


Figure 6

■ **Typical Circuit**

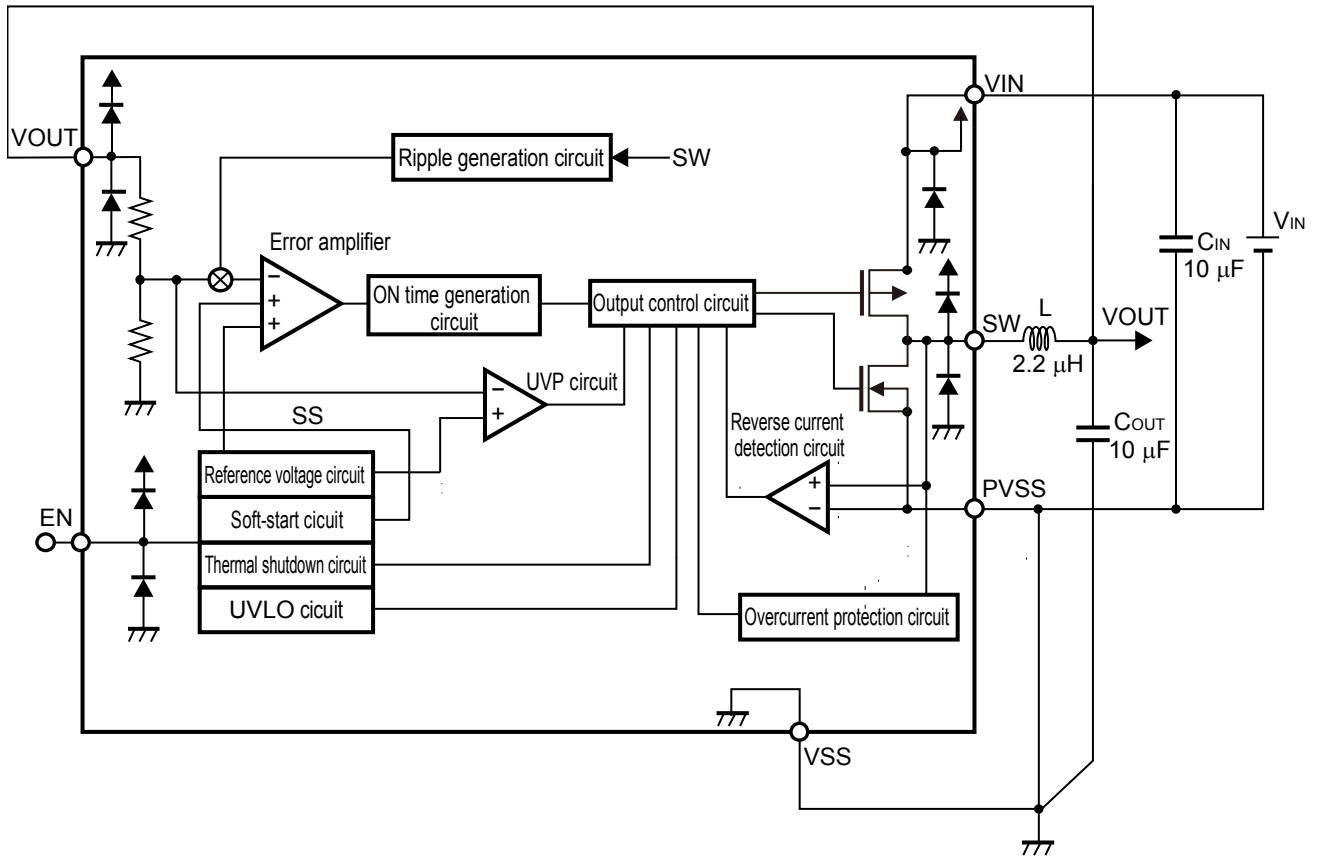


Figure 7

**Caution** The above connection diagram and constants will not guarantee successful operation. Perform thorough evaluation using an actual application to set the constants.

## External Parts Selection

Selectable values and recommended values for external parts are shown in **Table 9**.  
 Use ceramic capacitors for  $C_{IN}$  and  $C_{OUT}$ .

**Table 9**

Item	Input Capacitor ( $C_{IN}$ )	Output Capacitor ( $C_{OUT}$ )	Inductor (L)
Selectable value	2.2 $\mu$ F or larger	4.7 $\mu$ F to 100 $\mu$ F	1.5 $\mu$ H to 10 $\mu$ H
Recommended value	10 $\mu$ F	10 $\mu$ F	2.2 $\mu$ H

### 1. Input capacitor ( $C_{IN}$ )

$C_{IN}$  can lower the power supply impedance, average the input current, improve the efficiency and noise tolerance. Select a capacitor according to the impedance of the power supply to be used. Also take into consideration the DC bias characteristics of the capacitor to be used.

### 2. Output capacitor ( $C_{OUT}$ )

$C_{OUT}$  is used to smooth output voltage. If the capacitance is large, the overshoot and undershoot during load transient and output ripple voltage can be improved even more. Select a proper capacitor after the sufficient evaluation under actual conditions.

**Table 10 Recommended Capacitors ( $C_{IN}$ ,  $C_{OUT}$ ) List (at  $V_{OUT(S)} \leq 3.3$  V)**

Manufacturer	Part Number	Capacitance	Withstanding Voltage	Dimensions (L × W × H)
TDK Corporation	C1005X5R0J106M050BC	10 $\mu$ F	6.3 V	1.0 mm × 0.5 mm × 0.5 mm
TDK Corporation	C1608X5R0J106K080AB	10 $\mu$ F	6.3 V	1.6 mm × 0.8 mm × 0.8 mm
Murata Manufacturing Co., Ltd.	GRM185R60J106ME15	10 $\mu$ F	6.3 V	1.6 mm × 0.8 mm × 0.5 mm

**Table 11 Recommended Capacitors ( $C_{IN}$ ,  $C_{OUT}$ ) List (at  $V_{OUT(S)} > 3.3$  V)**

Manufacturer	Part Number	Capacitance	Withstanding Voltage	Dimensions (L × W × H)
TDK Corporation	C1608X5R0J106K080AB	10 $\mu$ F	6.3 V	1.6 mm × 0.8 mm × 0.8 mm
Murata Manufacturing Co., Ltd.	GRM185R60J106ME15	10 $\mu$ F	6.3 V	1.6 mm × 0.8 mm × 0.5 mm

### 3. Inductor (L)

When selecting L, note the allowable current. If a current exceeding this allowable current flows through the inductor, magnetic saturation may occur, and there may be risks which substantially lower efficiency and damage the IC as a result of large current.

Therefore, select an inductor so that peak current value ( $I_{PK}$ ), even during overcurrent detection, does not exceed the allowable current.

When prioritizing the load response, select an inductor with a small L value such as 2.2  $\mu$ H. When prioritizing the efficiency, select an inductor with a large L value such as 10  $\mu$ H.  $I_{PK}$  is calculated using the following expression.

$$I_{PK} = I_{OUT} + \frac{1}{2 \times L \times f_{SW}} \times \frac{(V_{IN} - V_{OUT}) \times V_{OUT}}{V_{IN}}$$

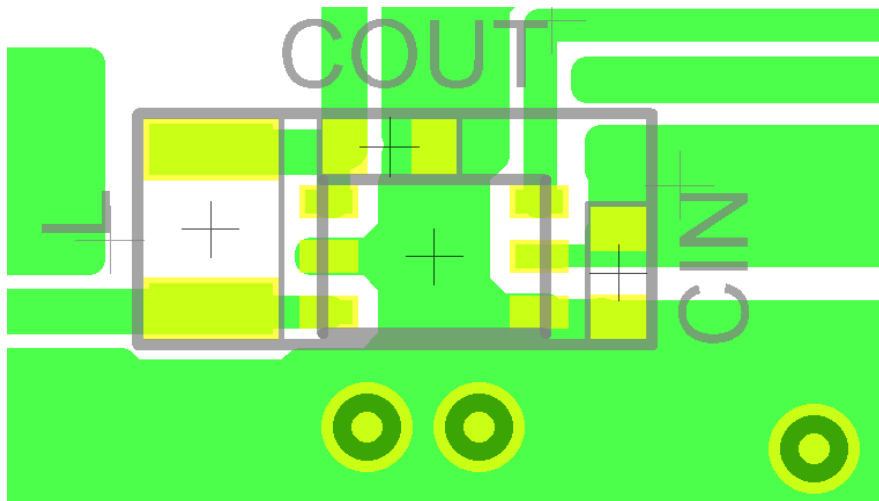
**Table 12 Recommended Inductors (L) List**

Manufacturer	Part Number	Inductance	Rated Current	Dimensions (L × W × H)
ALPS ELECTRIC CO., LTD.	GLUHK2R201A	2.2 $\mu$ H	1700 mA	2.0 mm × 1.6 mm × 1.0 mm
Murata Manufacturing Co., Ltd.	DFE201210S-2R2M=P2	2.2 $\mu$ H	2000 mA	2.0 mm × 1.2 mm × 1.0 mm
Würth Elektronik GmbH & Co. KG	74438343022	2.2 $\mu$ H	1100 mA	2.0 mm × 1.6 mm × 1.0 mm
Murata Manufacturing Co., Ltd.	LQM2MPN2R2MGH	2.2 $\mu$ H	1300 mA	2.0 mm × 1.6 mm × 0.9 mm
TDK Corporation	MLP2016G2R2M	2.2 $\mu$ H	850 mA	2.0 mm × 1.6 mm × 1.0 mm
Coilcraft, Inc.	PFL2015-222ME	2.2 $\mu$ H	1050 mA	2.2 mm × 1.45 mm × 1.5 mm

## ■ Board Layout Guidelines

Note the following cautions when determining the board layout for the S-85V1A Series.

- Place  $C_{IN}$  as close to the VIN pin and the PVSS pin as possible.
- Make the VIN pattern and GND pattern as wide as possible.
- Place thermal vias in the GND pattern to ensure sufficient heat dissipation.
- Keep thermal vias near  $C_{IN}$  and  $C_{OUT}$  approximately 3 mm to 4 mm away from capacitor pins.
- Large current flows through the SW pin. Make the wiring area of the pattern to be connected to the SW pin small to minimize parasitic capacitance and emission noise.
- Do not wire the SW pin pattern under the IC.



Total size 2.0 mm  $\times$  4.5 mm = 9.0 mm<sup>2</sup>

Figure 8 Reference Board Pattern

**Caution** The above pattern diagram does not guarantee successful operation. Perform thorough evaluation using the actual application to determine the pattern.

**Remark** Refer to the land drawing of SNT-6A and "SNT Package User's Guide".

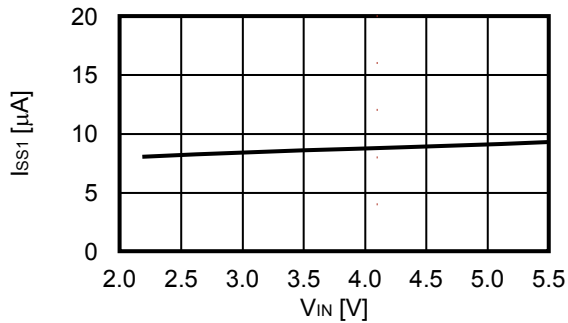
## ■ Precautions

- Mount external capacitors and inductors as close as possible to the IC, and make single GND.
- Characteristic ripple voltage and spike noise occur in the IC containing switching regulators. Moreover rush current flows at the time of a power supply injection. Because these largely depend on the inductor, the capacitor and impedance of power supply to be used, fully check them using an actually mounted model.
- The 10  $\mu$ F capacitor connected between the VIN pin and the VSS pin is a bypass capacitor. It stabilizes the power supply in the IC when application is used with a heavy load, and thus effectively works for stable switching regulator operation. Allocate the bypass capacitor as close to the IC as possible, prioritized over other parts.
- Although the IC contains a static electricity protection circuit, static electricity or voltage that exceeds the limit of the protection circuit should not be applied.
- The power dissipation of the IC greatly varies depending on the size and material of the board to be connected. Perform sufficient evaluation using an actual application before designing.
- ABLIC Inc. assumes no responsibility for the way in which this IC is used on products created using this IC or for the specifications of that product, nor does ABLIC Inc. assume any responsibility for any infringement of patents or copyrights by products that include this IC either in Japan or in other countries.

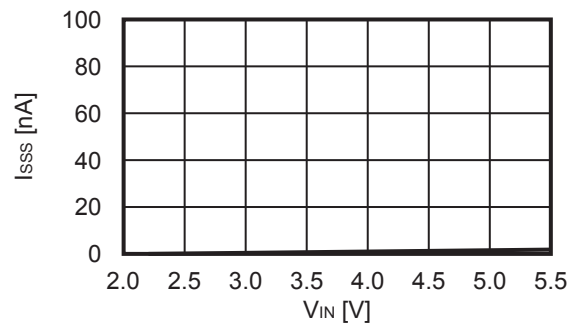
■ **Characteristics (Typical Data)**

1. Example of major power supply dependence characteristics (Ta = +25°C)

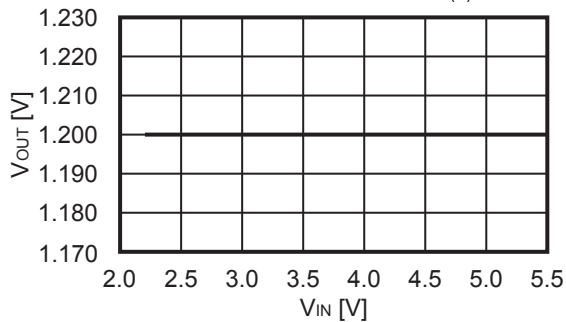
1.1 Current consumption during switching off (I<sub>SS1</sub>) vs. Input voltage (V<sub>IN</sub>)



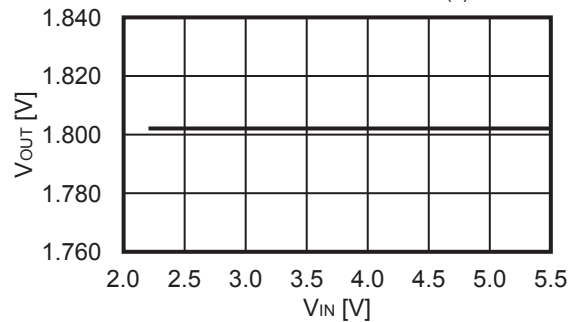
1.2 Current consumption during shutdown (I<sub>SSS</sub>) vs. Input voltage (V<sub>IN</sub>)



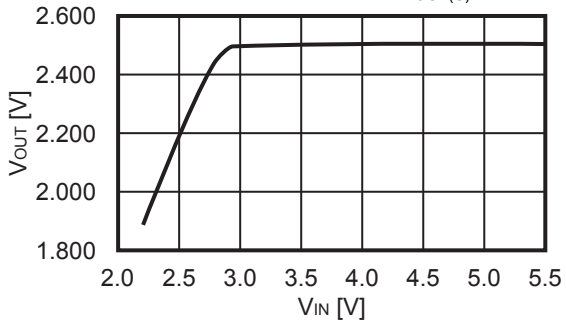
1.3 Output voltage (V<sub>OUT</sub>) vs. Input voltage (V<sub>IN</sub>)  
 V<sub>OUT(S)</sub> = 1.2 V



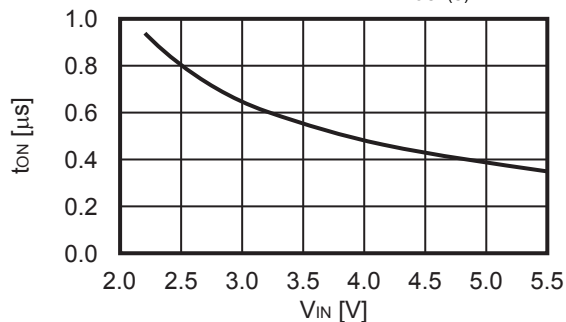
1.4 Output voltage (V<sub>OUT</sub>) vs. Input voltage (V<sub>IN</sub>)  
 V<sub>OUT(S)</sub> = 1.8 V



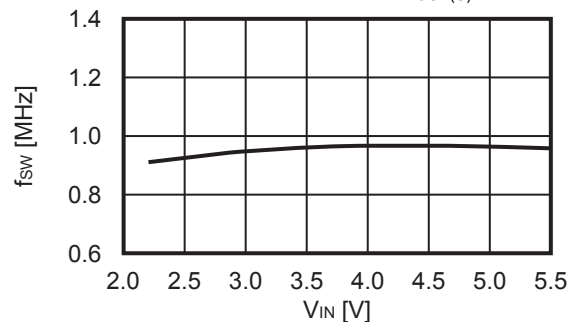
1.5 Output voltage (V<sub>OUT</sub>) vs. Input voltage (V<sub>IN</sub>)  
 V<sub>OUT(S)</sub> = 2.5 V



1.6 ON time (t<sub>ON</sub>) vs. Input voltage (V<sub>IN</sub>)  
 V<sub>OUT(S)</sub> = 1.8 V

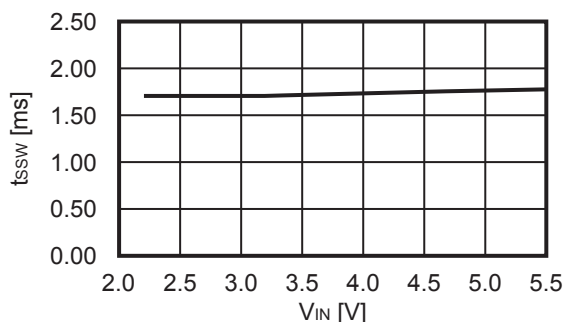


1.7 Switching frequency (f<sub>SW</sub>) vs. Input voltage (V<sub>IN</sub>)  
 V<sub>OUT(S)</sub> = 1.8 V

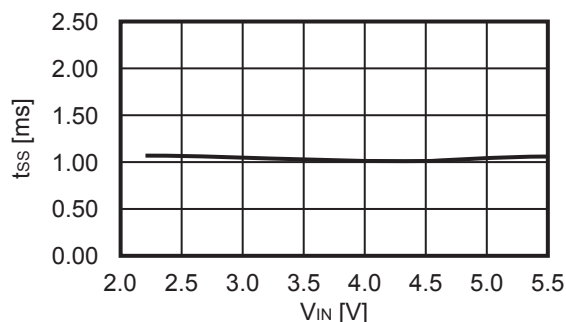




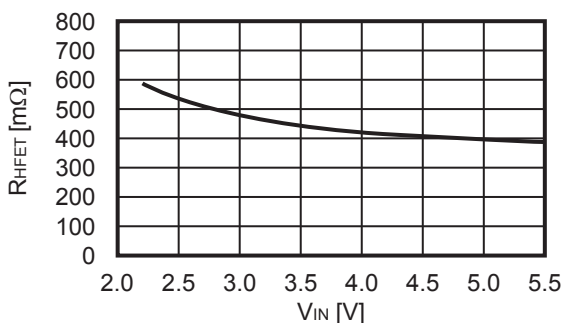
1. 8 Soft-start wait time ( $t_{SSW}$ ) vs. Input voltage ( $V_{IN}$ )



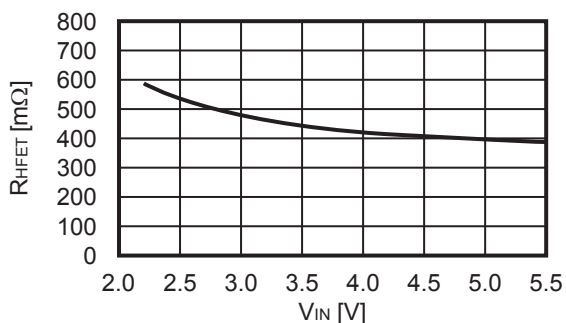
1. 9 Soft-start time ( $t_{SS}$ ) vs. Input voltage ( $V_{IN}$ )



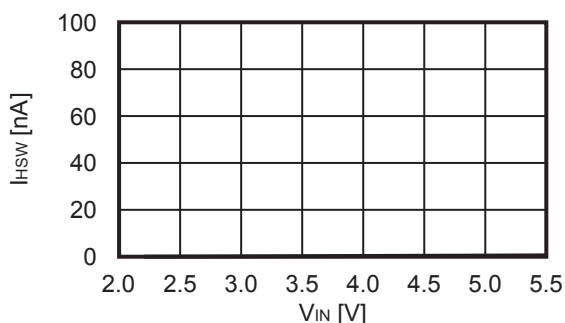
1. 10 High side power MOS FET on-resistance ( $R_{HFET}$ ) vs. Input voltage ( $V_{IN}$ )



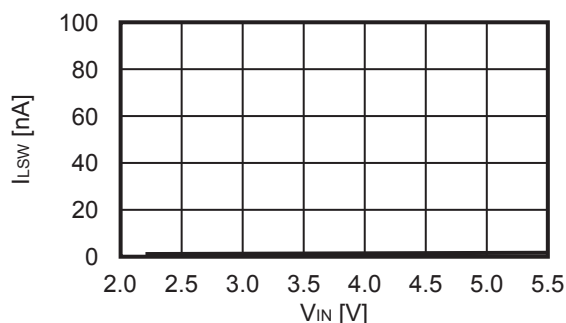
1. 11 Low side power MOS FET on-resistance ( $R_{LFET}$ ) vs. Input voltage ( $V_{IN}$ )



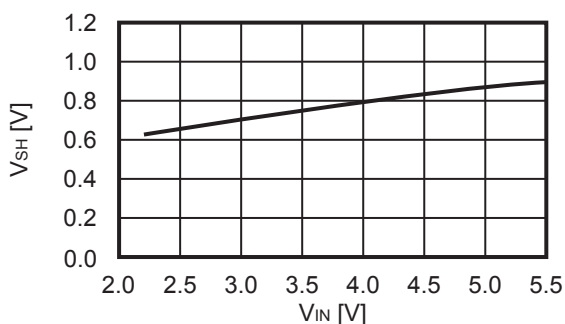
1. 12 High side power MOS FET leakage current ( $I_{HSW}$ ) vs. Input voltage ( $V_{IN}$ )



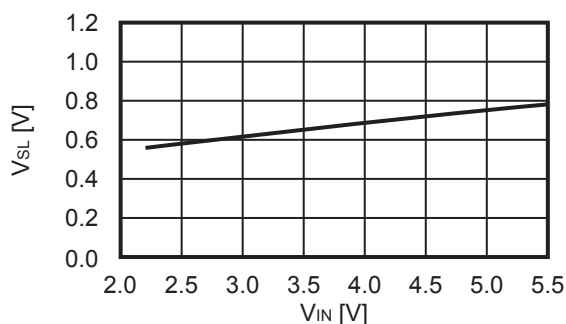
1. 13 Low side power MOS FET leakage current ( $I_{LSW}$ ) vs. Input voltage ( $V_{IN}$ )



1. 14 High level input voltage ( $V_{SH}$ ) vs. Input voltage ( $V_{IN}$ )

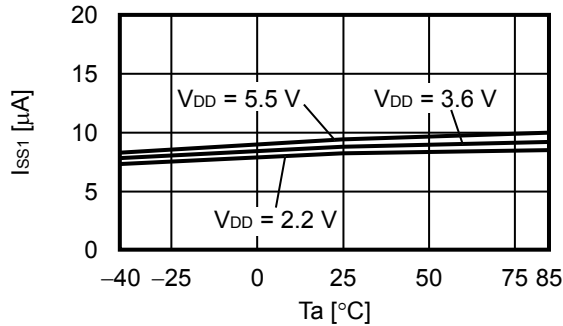


1. 15 Low level input voltage ( $V_{SL}$ ) vs. Input voltage ( $V_{IN}$ )

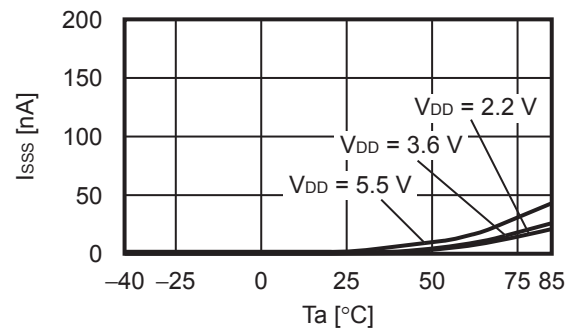


**2. Example of major temperature characteristics (Ta = -40°C to +85°C)**

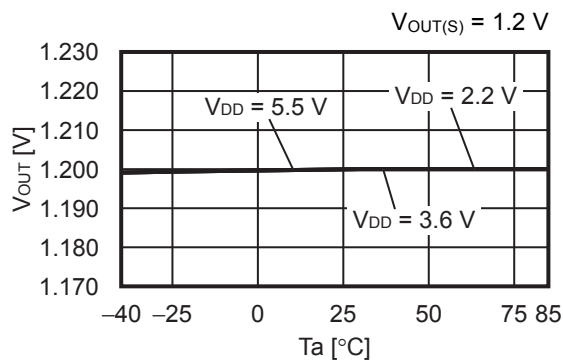
**2.1 Current consumption during switching off (I<sub>SS1</sub>) vs. Temperature (Ta)**



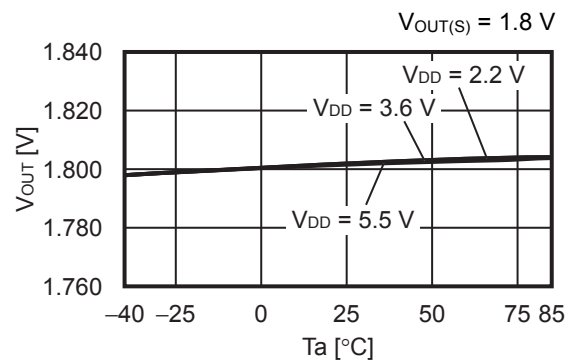
**2.2 Current consumption during shutdown (I<sub>SSS</sub>) vs. Temperature (Ta)**



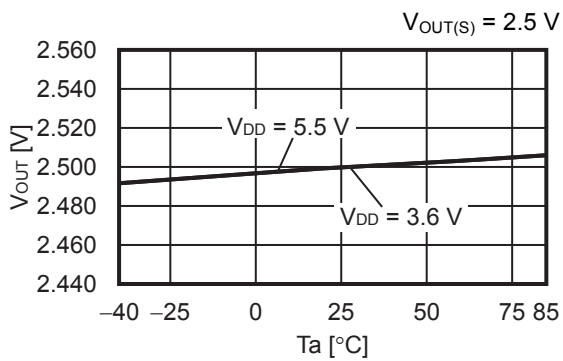
**2.3 Output voltage (V<sub>OUT</sub>) vs. Temperature (Ta)**



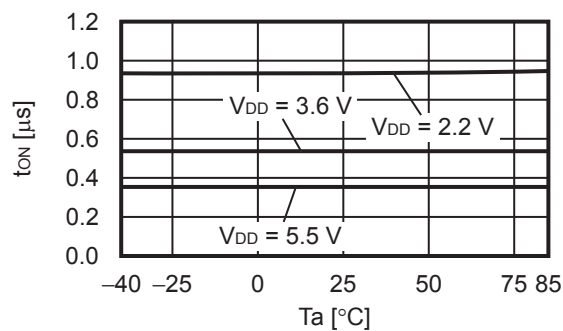
**2.4 Output voltage (V<sub>OUT</sub>) vs. Temperature (Ta)**



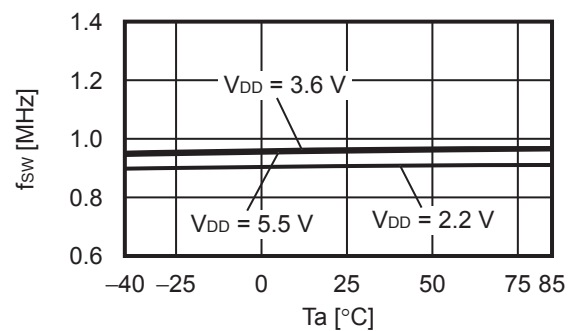
**2.5 Output voltage (V<sub>OUT</sub>) vs. Temperature (Ta)**



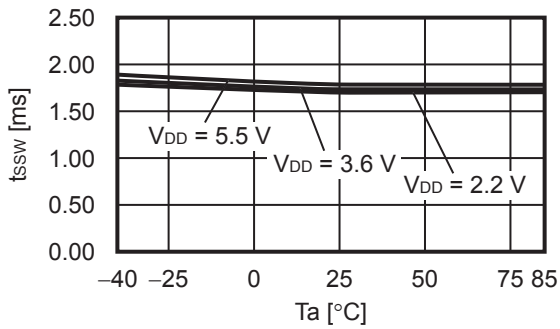
**2.6 ON time (t<sub>ON</sub>) vs. Temperature (Ta)**



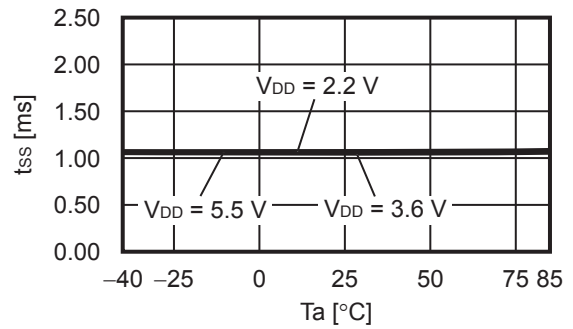
**2.7 Switching frequency (f<sub>sw</sub>) vs. Temperature (Ta)**



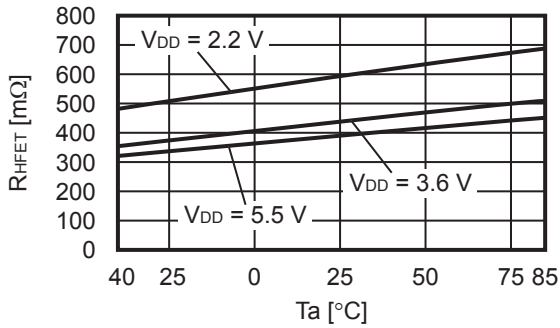
2. 8 Soft-start wait time ( $t_{SSW}$ ) vs. Temperature ( $T_a$ )



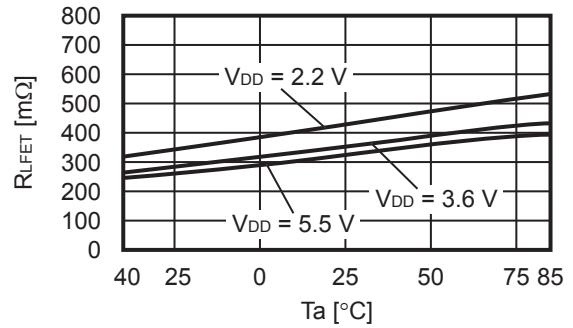
2. 9 Soft-start time ( $t_{SS}$ ) vs. Temperature ( $T_a$ )



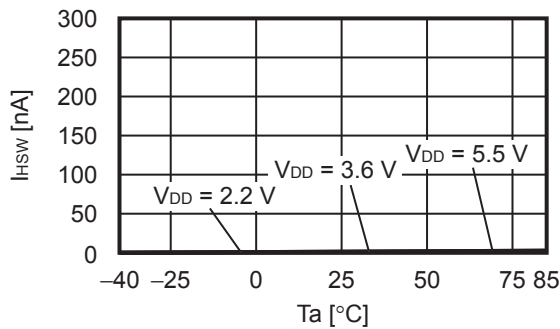
2. 10 High side power MOS FET on-resistance ( $R_{HFET}$ ) vs. Temperature ( $T_a$ )



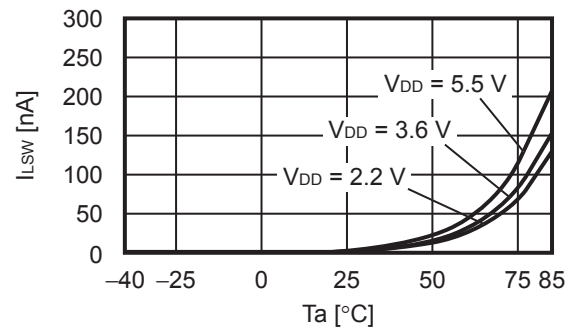
2. 11 Low side power MOS FET on-resistance ( $R_{LFET}$ ) vs. Temperature ( $T_a$ )



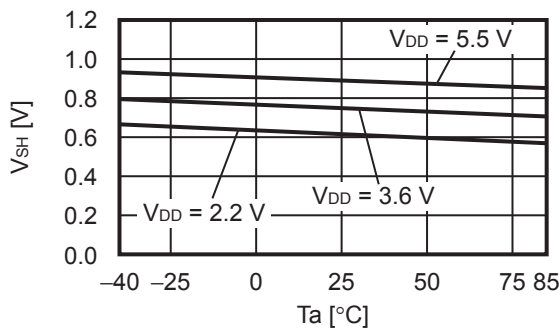
2. 12 High side power MOS FET leakage current ( $I_{HSW}$ ) vs. Temperature ( $T_a$ )



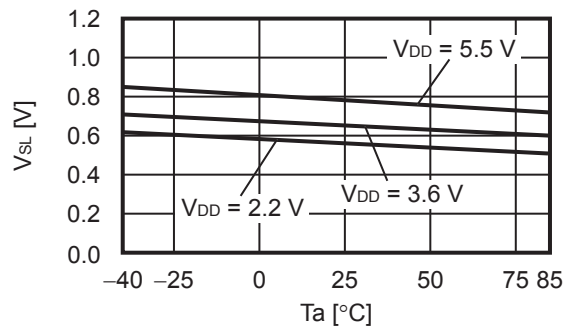
2. 13 Low side power MOS FET leakage current ( $I_{LSW}$ ) vs. Temperature ( $T_a$ )



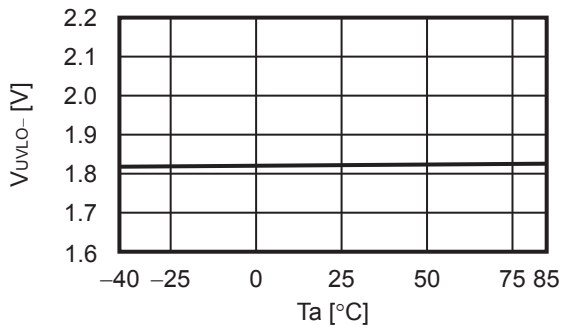
2. 14 High level input voltage ( $V_{SH}$ ) vs. Temperature ( $T_a$ )



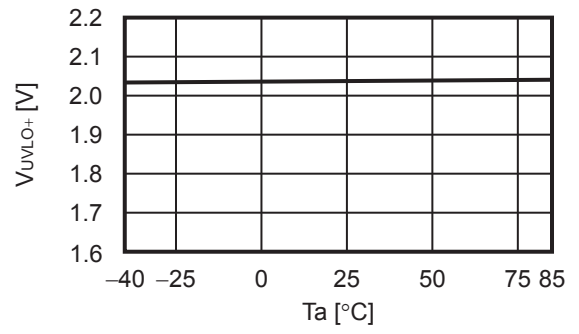
2. 15 Low level input voltage ( $V_{SL}$ ) vs. Temperature ( $T_a$ )



2. 16 UVLO detection voltage ( $V_{UVLO-}$ ) vs. Temperature ( $T_a$ )



2. 17 UVLO release voltage ( $V_{UVLO+}$ ) vs. Temperature ( $T_a$ )



### 3. Transient response characteristics

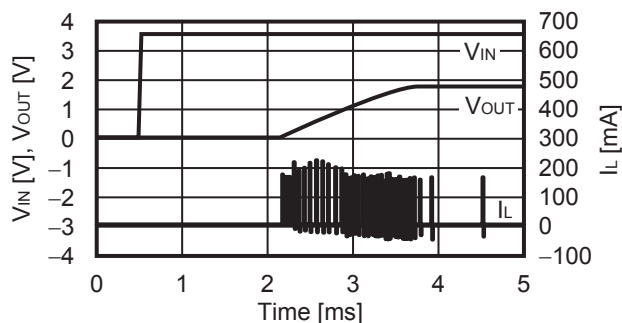
The external parts shown in Table 13 are used in "3. Transient response characteristics".

Table 13

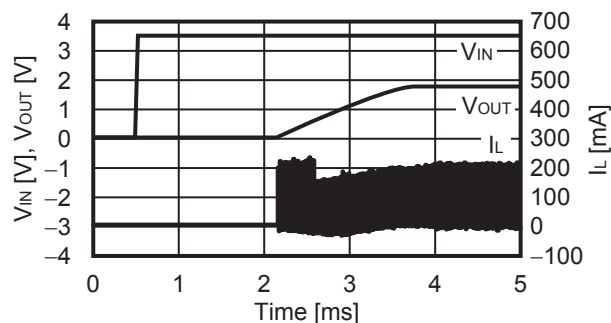
Element Name	Constant	Manufacturer	Part Number
Inductor	2.2 $\mu$ H	ALPS ELECTRIC CO., LTD.	GLUHK2R201A
Input capacitor	10 $\mu$ F	TDK Corporation	C1608X5R0J106K080AB
Output capacitor	10 $\mu$ F	TDK Corporation	C1608X5R0J106K080AB

#### 3.1 Power-on ( $V_{OUT} = 1.8$ V, $V_{IN} = 0$ V $\rightarrow$ 3.6 V, $T_a = +25^\circ$ C)

##### 3.1.1 $I_{OUT} = 0.1$ mA



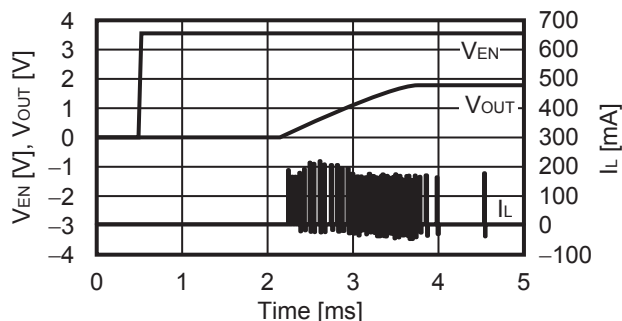
##### 3.1.2 $I_{OUT} = 200$ mA



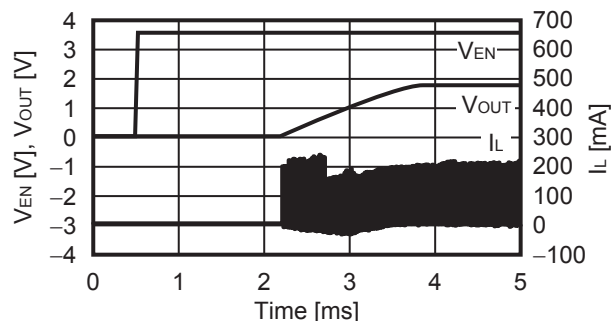
#### 3.2 Transient response characteristics of EN pin

( $V_{OUT} = 1.8$  V,  $V_{IN} = 3.6$  V,  $V_{EN} = 0$  V  $\rightarrow$  3.6 V,  $T_a = +25^\circ$ C)

##### 3.2.1 $I_{OUT} = 0.1$ mA

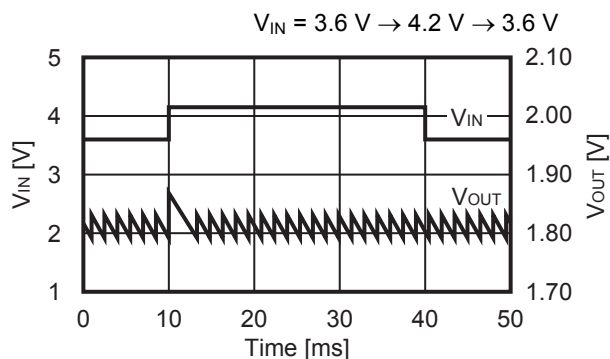


##### 3.2.2 $I_{OUT} = 200$ mA

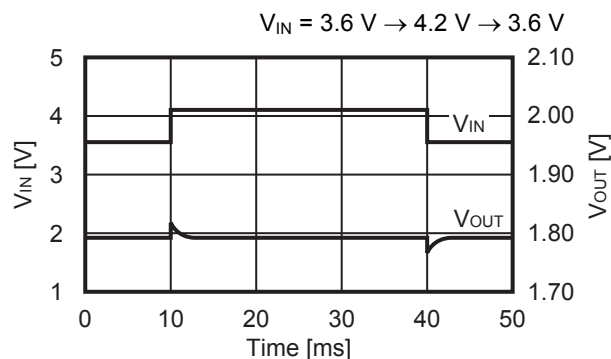


#### 3.3 Power supply fluctuation ( $V_{OUT} = 1.8$ V, $T_a = +25^\circ$ C)

##### 3.3.1 $I_{OUT} = 0.1$ mA

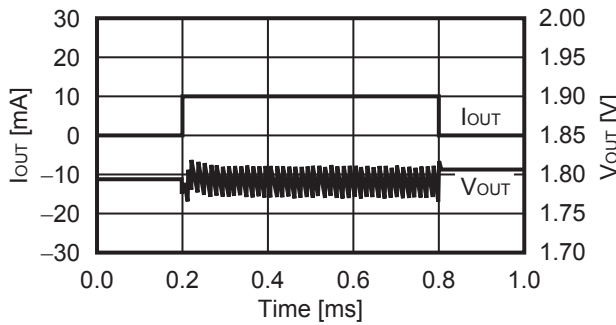


##### 3.3.2 $I_{OUT} = 200$ mA

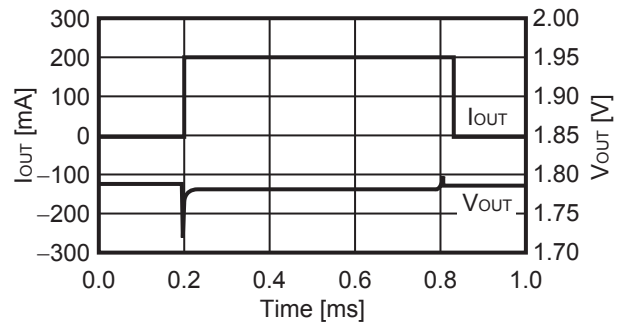


**3.4 Load fluctuation ( $V_{OUT} = 1.8$  V,  $V_{IN} = 3.6$  V,  $T_a = +25^\circ\text{C}$ )**

**3.4.1  $I_{OUT} = 0.1$  mA  $\rightarrow$  10 mA  $\rightarrow$  0.1 mA**



**3.4.2  $I_{OUT} = 0.1$  mA  $\rightarrow$  200 mA  $\rightarrow$  0.1 mA**



**Reference Data**

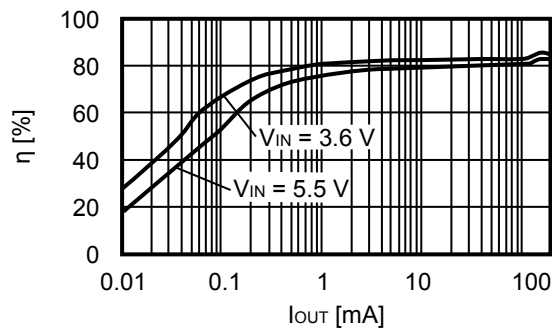
The external parts shown in **Table 14** are used in "Reference Data".

**Table 14**

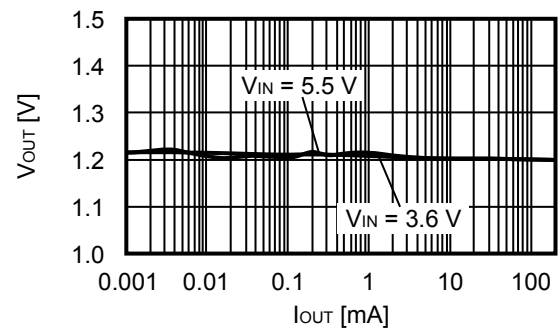
Condition	Inductor (L)	Input Capacitor ( $C_{IN}$ )	Output Capacitor ( $C_{OUT}$ )
<1>	GLUHK2R201A (2.2 $\mu$ H) ALPS ELECTRIC CO., LTD	C1005X5R0J106M050BC (10 $\mu$ F) TDK Corporation	C1005X5R0J106M050BC (10 $\mu$ F) TDK Corporation
<2>	DFE201210S (2.2 $\mu$ H) Toko Ink.	C1005X5R0J106M050BC (10 $\mu$ F) TDK Corporation	C1005X5R0J106M050BC (10 $\mu$ F) TDK Corporation

**1.  $V_{OUT} = 1.2$  V (External parts: Condition<1>)**

**1.1 Efficiency ( $\eta$ ) vs. Output current ( $I_{OUT}$ )**

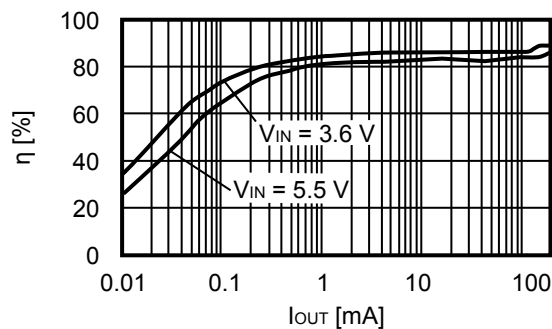


**1.2 Output voltage ( $V_{OUT}$ ) vs. Output current ( $I_{OUT}$ )**

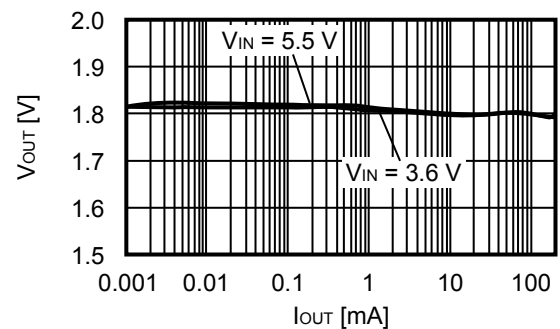


**2.  $V_{OUT} = 1.8$  V (External parts: Condition<1>)**

**2.1 Efficiency ( $\eta$ ) vs. Output current ( $I_{OUT}$ )**

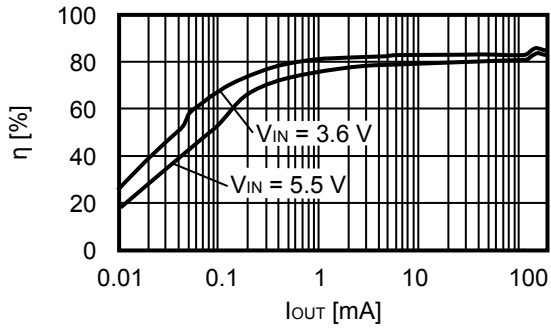


**2.2 Output voltage ( $V_{OUT}$ ) vs. Output current ( $I_{OUT}$ )**

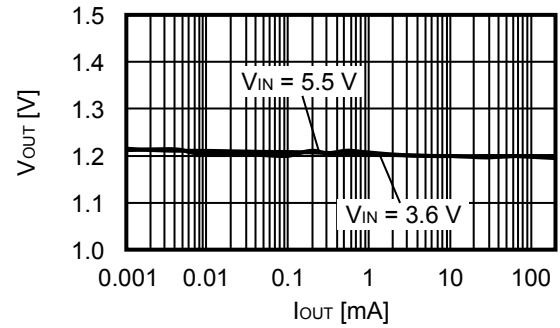


**3.  $V_{OUT} = 1.2$  V (External parts: Condition<2>)**

**3.1 Efficiency ( $\eta$ ) vs. Output current ( $I_{OUT}$ )**

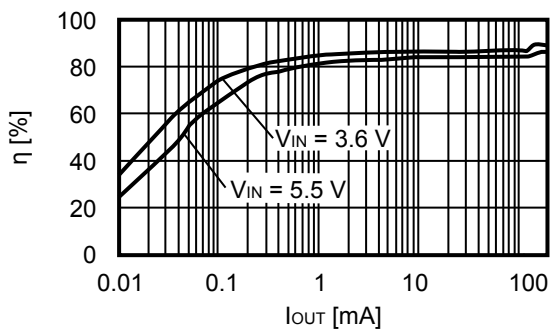


**3.2 Output voltage ( $V_{OUT}$ ) vs. Output current ( $I_{OUT}$ )**

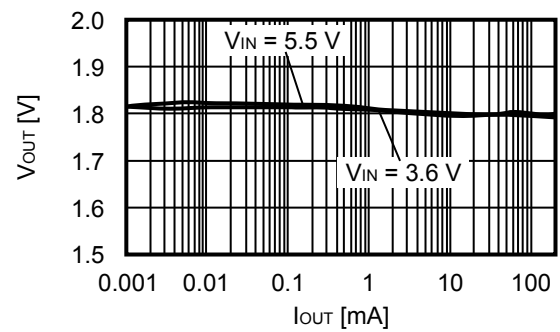


**4.  $V_{OUT} = 1.8$  V (External parts: Condition<2>)**

**4.1 Efficiency ( $\eta$ ) vs. Output current ( $I_{OUT}$ )**

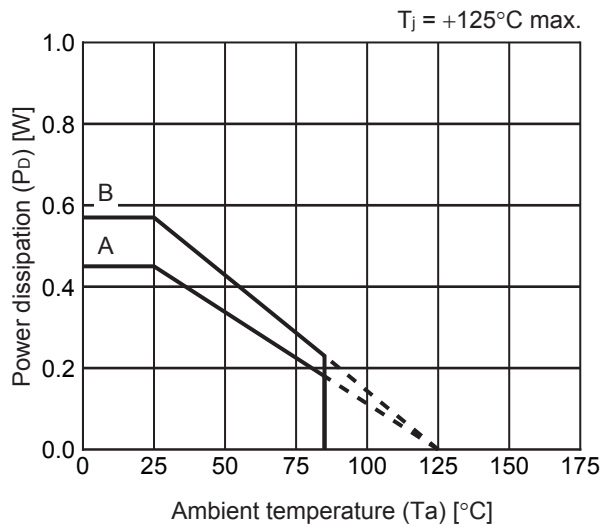


**4.2 Output voltage ( $V_{OUT}$ ) vs. Output current ( $I_{OUT}$ )**



■ **Power Dissipation**

SNT-6A



Board	Power Dissipation ( $P_D$ )
A	0.45 W
B	0.57 W
C	-
D	-
E	-



# SNT-6A Test Board

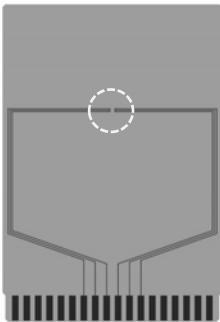
(1) Board A

 IC Mount Area



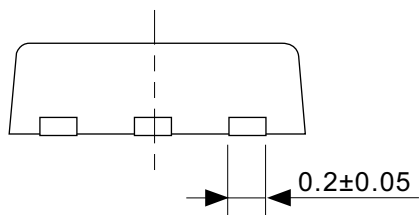
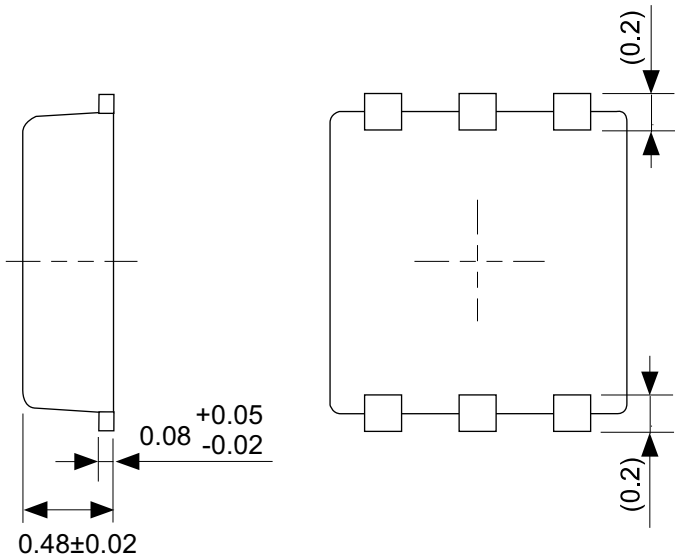
Item		Specification
Size [mm]		114.3 x 76.2 x t1.6
Material		FR-4
Number of copper foil layer		2
Copper foil layer [mm]	1	Land pattern and wiring for testing: t0.070
	2	-
	3	-
	4	74.2 x 74.2 x t0.070
Thermal via		-

(2) Board B



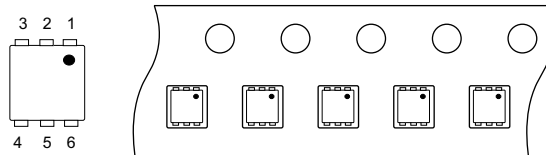
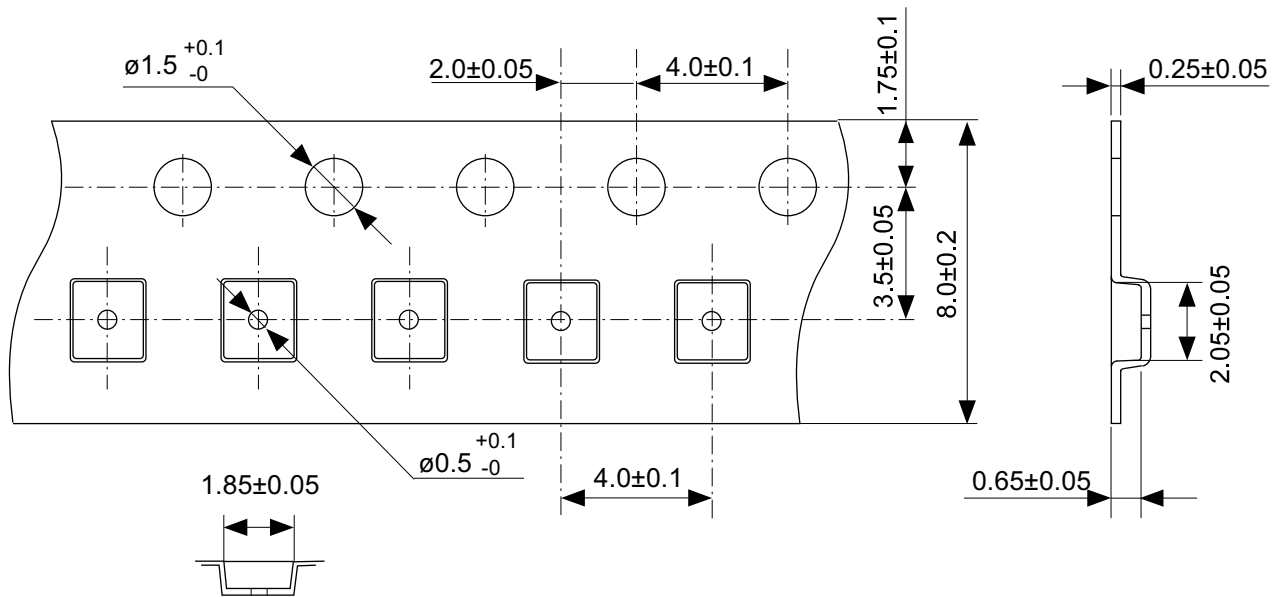
Item		Specification
Size [mm]		114.3 x 76.2 x t1.6
Material		FR-4
Number of copper foil layer		4
Copper foil layer [mm]	1	Land pattern and wiring for testing: t0.070
	2	74.2 x 74.2 x t0.035
	3	74.2 x 74.2 x t0.035
	4	74.2 x 74.2 x t0.070
Thermal via		-

No. SNT6A-A-Board-SD-1.0



No. PG006-A-P-SD-2.1

TITLE	SNT-6A-A-PKG Dimensions
No.	PG006-A-P-SD-2.1
ANGLE	
UNIT	mm
<b>ABLIC Inc.</b>	

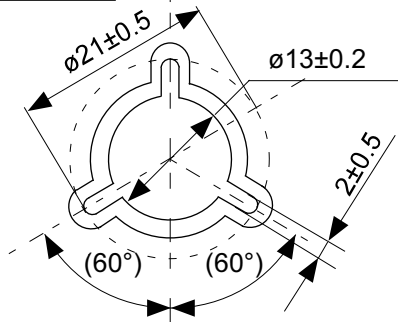


No. PG006-A-C-SD-2.0

TITLE	SNT-6A-A-Carrier Tape
No.	PG006-A-C-SD-2.0
ANGLE	
UNIT	mm
<b>ABLIC Inc.</b>	



Enlarged drawing in the central part



No. PG006-A-R-SD-1.0

TITLE	SNT-6A-A-Reel		
No.	PG006-A-R-SD-1.0		
ANGLE		QTY.	5,000
UNIT	mm		
<b>ABLIC Inc.</b>			



※1. ランドパターンの幅に注意してください (0.25 mm min. / 0.30 mm typ.).  
 ※2. パッケージ中央にランドパターンを広げないでください (1.30 mm ~ 1.40 mm)。

- 注意
1. パッケージのモールド樹脂下にシルク印刷やハンダ印刷などしないでください。
  2. パッケージ下の配線上のソルダーレジストなどの厚みをランドパターン表面から0.03 mm 以下にしてください。
  3. マスク開口サイズと開口位置はランドパターンと合わせてください。
  4. 詳細は "SNTパッケージ活用の手引き" を参照してください。

※1. Pay attention to the land pattern width (0.25 mm min. / 0.30 mm typ.).  
 ※2. Do not widen the land pattern to the center of the package ( 1.30 mm ~ 1.40 mm ).

- Caution**
1. Do not do silkscreen printing and solder printing under the mold resin of the package.
  2. The thickness of the solder resist on the wire pattern under the package should be 0.03 mm or less from the land pattern surface.
  3. Match the mask aperture size and aperture position with the land pattern.
  4. Refer to "SNT Package User's Guide" for details.

※1. 请注意焊盘模式的宽度 (0.25 mm min. / 0.30 mm typ.).  
 ※2. 请勿向封装中间扩展焊盘模式 (1.30 mm ~ 1.40 mm)。

- 注意
1. 请勿在树脂型封装的下面印刷丝网、焊锡。
  2. 在封装下、布线上的阻焊膜厚度 (从焊盘模式表面起) 请控制在 0.03 mm 以下。
  3. 钢网的开口尺寸和开口位置请与焊盘模式对齐。
  4. 详细内容请参阅 "SNT 封装的应用指南"。

No. PG006-A-L-SD-4.1

TITLE	SNT-6A-A -Land Recommendation
No.	PG006-A-L-SD-4.1
ANGLE	
UNIT	mm
<b>ABLIC Inc.</b>	

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