

POWER MANAGEMENT

Description

The SC1405CTS is a Dual-MOSFET Driver with an internal Overlap Protection Circuit to prevent shoot-through. Each driver is capable of driving a 3000pF load in 15ns rise/fall time and has ULTRA-LOW propagation delay from input transition to the gate of the power FET's. Adaptive Overlap Protection circuit ensures that the synchronous FET does not turn on until the top FET source has reached a voltage low enough to prevent shoot-through. The delay between the bottom gate going low to the top gate transitioning high is externally programmable via a capacitor to minimize dead time. The bottom FET may be disabled at light loads by keeping S_MOD low to trigger asynchronous operation, thus saving the bottom FET's gate drive current and inductor ripple current. An internal voltage reference allows threshold adjustment for an Output Over-Voltage protection circuitry, independent of the PWM controller. The device provides overvoltage protection independent of the PWM feedback loop with a unique "adaptive OVP" comparator which rejects noise but responds quickly to a true OVP situation. Under-Voltage-Lock-Out circuit is included to guarantee that both driver outputs are off when Vcc is less than or equal to 4.4V (typ) at supply ramp up (4.35V at supply ramp down). A CMOS output provides status indication of the 5V supply. A low enable input places the IC in stand-by mode, reducing supply current to less than 10µA. SC1405CTS is offered in a high pitch (.025" lead spacing) TSSOP package.

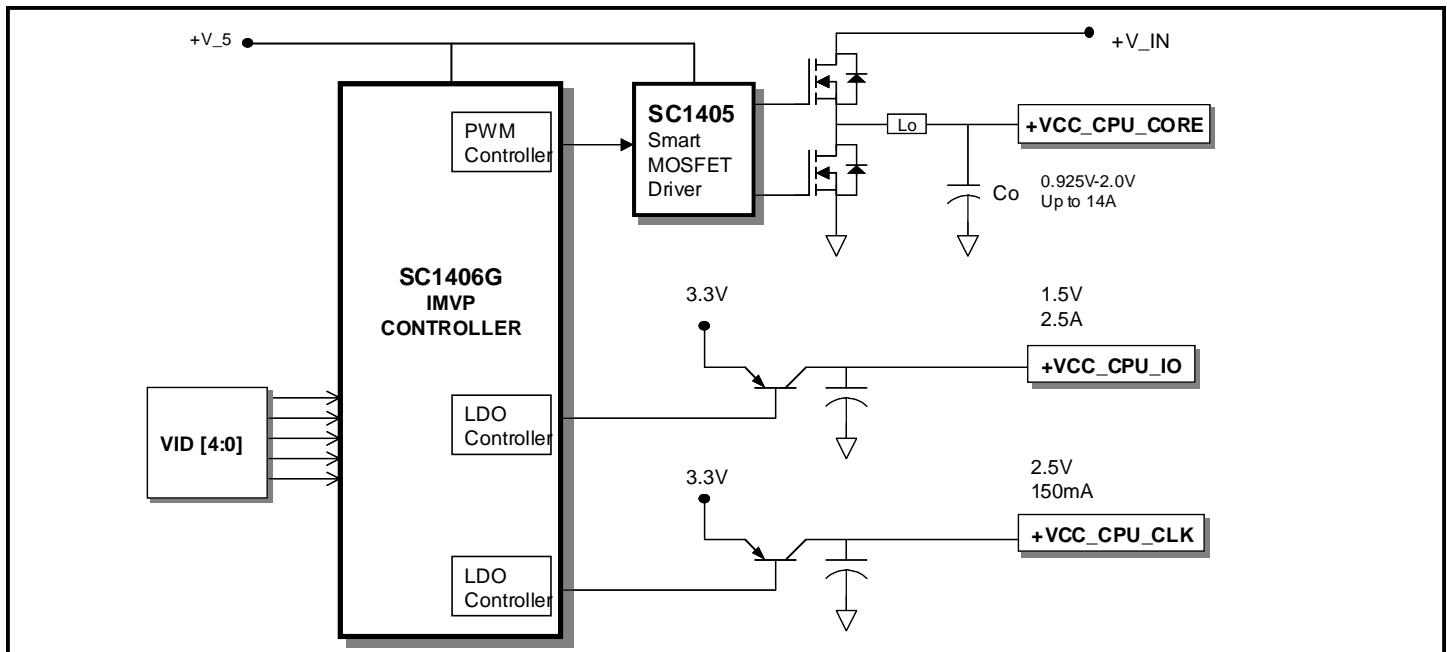
Features

- ◆ Fast rise and fall times (15ns with 3000pf load)
- ◆ **14ns max. Propagation delay (BG going with low)**
- ◆ Adaptive and programmable shoot-through protection
- ◆ Adaptive overvoltage protection
- ◆ Wide input voltage range (4.5-25V)
- ◆ Programmable delay between MOSFET's
- ◆ Power saving asynchronous mode control
- ◆ Output overvoltage protection/overtemp shutdown
- ◆ Under-Voltage lock-out and power ready signal
- ◆ Less than 10µA stand-by current (EN=low)
- ◆ Power ready output signal
- ◆ High frequency (to 1.2MHz) operation allows use of small inductors and low cost caps in place of electrolytics

Applications

- ◆ High Density/Fast transient microprocessor power supplies
- ◆ Motor Drives/Class-D amps
- ◆ High efficiency portable computers

Typical Application Circuit



POWER MANAGEMENT
Absolute Maximum Ratings

Exceeding the specifications below may result in permanent damage to the device, or device malfunction. Operation outside of the parameters specified in the Electrical Characteristics section is not implied.

Parameter	Symbol	Conditions	Minimum	Maximum	Units
V _{CC} Supply Voltage	V _{CCMAX}		-0.3	7	V
BST to PGND	VMAX _{BST-PGND}		-0.3	30	V
BST to DRN	VMAX _{BST-DRN}		-0.3	7	V
DRN to PGND	VMAX _{DRN-PGN}	DC	-2	25	V
DRN to PGND	VMAX _{DRN-PGN}	Transient, 100nS	-4	25	V
OVP_S to PGND	VMAX _{OVP S-PGND}		-0.3	10	V
EN, CO, DSPS, MODE, PRDY, DELAY to AGND			-0.3	V _{CC} + 0.3	V
AGND to PGND			-1	+1	V
Continuous Power Dissipation	Pd	Tamb = 25°C, T _J = 125°C Tcase = 25°C, T _J = 125°C		0.66 2.56	Ω
Thermal Impedance Junction to Case	θ _{JC}			40	°C/W
Thermal Impedance Junction to Ambient	θ _{JA}			150	°C/W
Operating Temperature Range	T _J		0	+125	°C
Storage Temperature Range	T _{STG}		-65	+150	°C
Lead Temperature (Soldering) 10 Sec.	T _{LEAD}			300	°C

NOTE:

(1) Specification refers to application circuit.

Electrical Characteristics - DC Operating Specifications

Unless specified: -0 < θ_J < 125°C; V_{CC} = 5V; 4V ≤ V_{BST} ≤ 26V

Parameter	Symbol	Conditions	Min	Typ	Max	Units
Power Supply						
Supply Voltage	V _{CC}		4.5	5	6.0	V
Quiescent Current	I _{q_stby}	EN = 0V			10	μA
Quiescent Current, operating	I _{q_op}	V _{CC} = 5V, CO = 0V		1		mA
PRDY						
High Level Output Voltage	V _{OH}	V _{CC} = 4.6V, I _{load} = 10mA	4.5	4.55		V
Low Level Output Voltage	V _{OL}	V _{CC} < UVLO threshold, I _{load} = 10μA		0.1	0.2	V

POWER MANAGEMENT
Electrical Characteristics - DC Operating Specifications

Parameter	Symbol	Conditions	Min	Typ	Max	Units
DSPS_DR						
High Level Output Voltage	V_{OH}	$V_{CC} = 4.6V$, $C_{load} = 100pF$	4.15			V
Low Level Output Voltage	V_{OL}	$V_{CC} = 4.6V$, $C_{load} = 100pF$			0.05	V
Under Voltage Lockout						
Start Threshold			4.2	4.4	4.6	V
Hysteresis	V_{hys}			0.05		V
Logic Active Threshold		EN is low			1.5	V
Overvoltage Protection						
Trip Threshold	V_{TRIP}		1.145	1.2	1.255	V
Hysteresis	$V_{hys_{OVP}}$			0.8		V
Trip Delay, 50mV Overdrive		$t = 0$ to $125^{\circ}C$	300	470	800	ns
Trip Delay, 100mV Overdrive		$t = 0$ to $125^{\circ}C$	125	225	400	ns
S_MOD						
High Level Input Voltage	V_{IH}		2.0			V
Low Level Input Voltage	V_{IL}				0.8	V
Enable						
High Level Input Voltage	V_{IH}		2.0			V
Low Level Input Voltage	V_{IL}				0.8	V
CO						
High Level Input Voltage	V_{IH}		2.0			V
Low Level Input Voltage	V_{IL}				0.8	V
Thermal Shutdown						
Over Temperature Trip Point	T_{OTP}			165		$^{\circ}C$
Hysteresis	T_{HYST}			10		$^{\circ}C$
High-Side Driver						
Peak Output Current	I_{PKH}			3		A
Output Resistance	$R_{src_{TG}}$ $R_{sink_{TG}}$	duty cycle < 2%, $tpw < 100\mu s$, $T_J = 125^{\circ}C$, $V_{BST} - V_{DRN} = 4.5V$, $V_{TG} = 4.0V (src) + V_{DRN}$ or $V_{TG} = 0.5V (sink) + V_{DRN}$		1 .7		Ω
Low-Side Drive						
Peak Output Current	I_{PKL}			3		A
Output Resistance	$R_{src_{BG}}$ $R_{sink_{BG}}$	duty cycle < 2%, $tpw < 100\mu s$, $T_J = 125^{\circ}C$, $V_{VS} = 4.6V$, $V_{BG} = 4V (src)$ or $V_{LOWDR} = 0.5V (sink)$		1.2 1.0		Ω

POWER MANAGEMENT
Electrical Characteristics - AC Operating Specifications

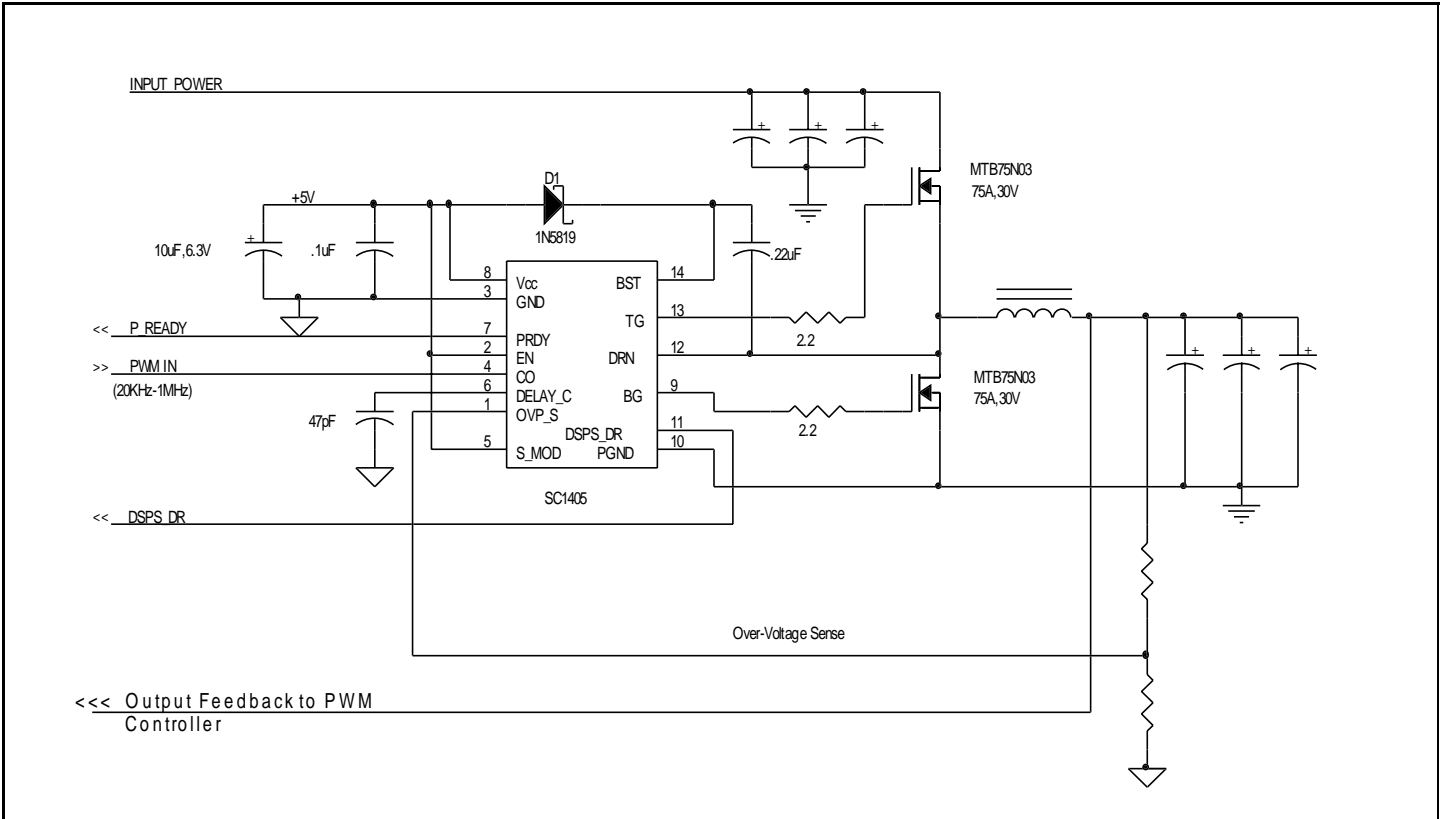
Parameter	Symbol	Conditions	Min	Typ	Max	Units
High Side Driver						
rise time	$t_{r_{TG1}}$	$CI = 3nF, V_{BST} - V_{DRN} = 4.6V,$		14	23	ns
fall time	$t_{f_{TG}}$	$CI = 3nF, V_{BST} - V_{DRN} = 4.6V,$		12	19	ns
propagation delay time, TG going high	$tpdh_{TG}$	$CI = 3nF, V_{BST} - V_{DRN} = 4.6V,$ C-delay=0		20	32	ns
propagation delay time, TG going low	$tpdl_{TG}$	$CI = 3nF, V_{BST} - V_{DRN} = 4.6V,$		15	24	ns
Low-Side Driver						
rise time	$t_{r_{BG}}$	$CI = 3nF, V_{V_S} = 4.6V,$		15	24	ns
fall time	$t_{f_{BG}}$	$CI = 3nF, V_{V_S} = 4.6V,$		13	21	ns
propagation delay time, BG going high	$tpdh_{BGHI}$	$CI = 3nF, V_{BST} - V_{DRN} = 4.6V,$ C-delay=0		12	19	ns
propagation delay time, TG going low	$tpdl_{BG}$	$CI = 3nF, V_{V_S} = 4.6V,$ $DRN < 1V$		7	12	ns
Under-Voltage Lockout						
V_5 ramping up	$tpdh_{UVLO}$	EN is High			10	us
V_5 ramping down	$tpdh_{UVLO}$	EN is High			10	us
PRDY						
EN is transitioning from low to high	$tpdhPRDY$	$V_5 > UVLO$ threshold, Delay measured from $EN > 2.0V$ to $PRDY > 3.5V$			10	μs
EN is transitioning fro high to low	$tpdhUVLO$	$V_5 > UVLO$ threshold, Delay measured from $EN < 0.8V$ to $PRDY < 10\%$ of V_5V			500	μs
DSPS_DR						
rise/fall time	$t_{r_{DSPS\ DR}}$	$CI = 100\ pf, V_5 = 4.6V$			20	ns
propagation delay, DSPS_DR going high	$tpdh_{DSPS\ DR}$	S_MOD goes high and BG goes high or S_MOD goes low			10	ns
propagation delay, DSPS_DR goes low	$tpdl_{DSPS\ DR}$	S_MOD goes high and BG goes low			10	ns

NOTE:

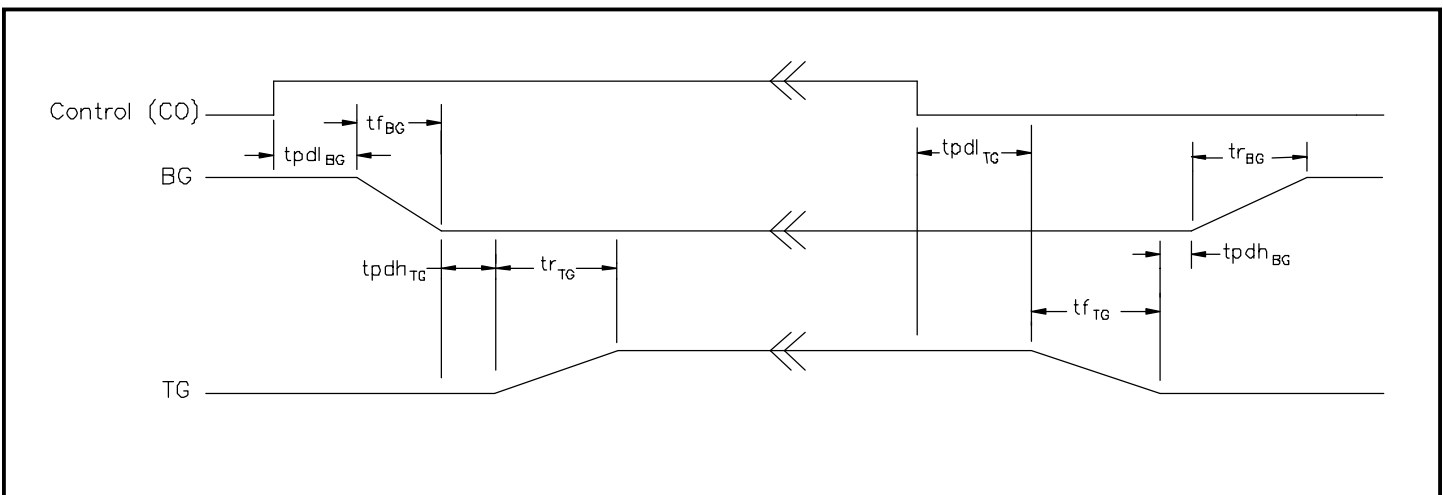
(1) This device is ESD sensitive. Use of standard ESD handling precautions is required.

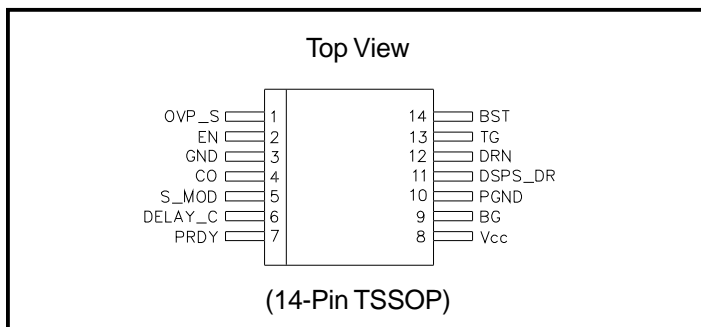
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Application Circuit



Timing Diagram



POWER MANAGEMENT
Pin Configuration

Ordering Information

Device ⁽¹⁾	Package	Temp Range (T _j)
SC1405CTS.TR	TSSOP-14	0 to 125°C

Note:

(1) Only available in tape and reel packaging. A reel contains 2500 devices.

Pin Descriptions

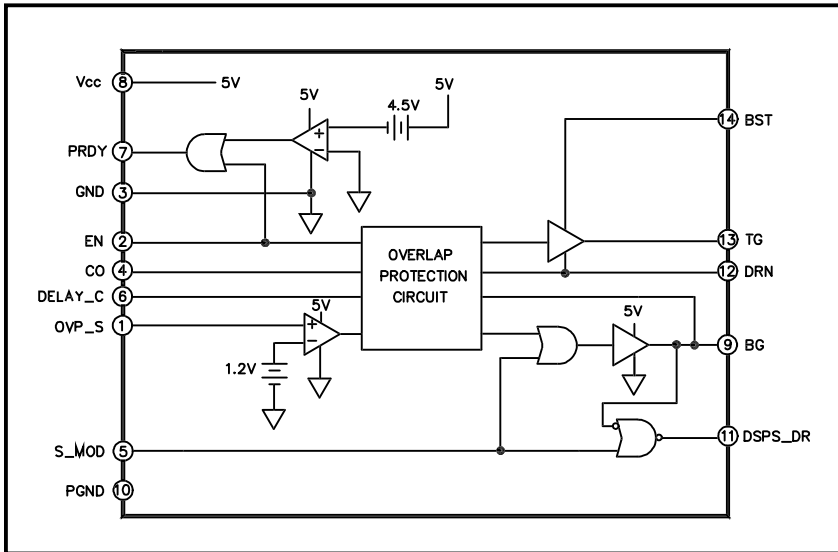
Pin #	Pin Name	Pin Function
1	OVP_S	Overvoltage protection sense. External scaling resistors required to set protection threshold.
2	EN	When high, this pin enables the internal circuitry of the device. When low, TG, BG, and PRDY are forced low and the supply current (5V) is less than 10µA.
3	GND	Logic GND.
4	CO	TTL-level input signal to the MOSFET drivers.
5	S_MOD	When low, this signal forces BG to be low, triggering asynchronous operation. When high, BG is not a function of this signal.
6	DELAY_C	The capacitance connected between this pin and GND sets the additional propagation delay for BG going low to TG going high. Total propagation delay = 20ns + 1ns/pF. If no capacitor is connected, the propagation delay = 20ns.
7	PRDY	This pin indicates the status of VCC. When VCC is less than the UVLO threshold, this output is driven low. When VCC is greater than or equals to the UVLO threshold this output goes high. This output has a 10mA source drive and 10µA sink capability.
8	VCC	Input supply of 5 - 8V. A .22-1µF ceramic capacitor should be connected from VCC to PGND very close to the chip.
9	BG	Output drive for the synchronous (bottom) MOSFET.
10	PGND	Power ground. Connect to the synchronous FET source pin (power ground).
11	DSPS_DR	Dynamic Set Point Switch Drive. TTL level output signal. When S-MOD is high, this pin follows the BG driver pin voltage.
12	DRN	This pin connects to the junction of the switching and synchronous MOSFET's. This pin can be subjected to a -2V minimum relative to PGND without affecting operation.
13	TG	Output gate drive for the switching (high-side) MOSFET.
14	BST	Bootstrap pin. A capacitor is connected between BST and DRN pins to develop the floating bootstrap voltage for the high-side MOSFET. The capacitor value is typically between 0.1µF and 1µF (ceramic).

NOTE:

(1) All logic level inputs and outputs are open collector TTL compatible.

POWER MANAGEMENT

Block Diagram



Applications Information

SC1405CTS is designed to drive Low Rds_{On} power MOSFET's with ultra-low rise/fall times and propagation delays. As the switching frequency of PWM controllers is increased to reduce power supply volume and cost, fast rise and fall times are necessary to minimize switching losses (TOP MOSFET) and reduce Dead-time (BOTTOM MOSFET) losses. While Low Rds_{On} MOSFET's present a power saving in I²R losses, the MOSFET's die area is larger and thus the effective input capacitance of the MOSFET is increased. Often a 50% decrease in Rds_{On} more than doubles the effective input gate charge, which must be supplied by the driver. The Rds_{On} power savings can be offset by the switching and dead-time losses with a suboptimum driver. While discrete solution can achieve reasonable drive capability, implementing shoot-through, programmable delay and other housekeeping functions necessary for safe operation can become cumbersome and costly. The SC1405 family of parts presents a total solution for the high-speed, high power density applications. Wide input supply range of 4.5V-25V allows use in battery powered applications, new high voltage, distributed power servers as well as Class-D amplifiers.

Theory of Operation

The control input (CO) to the SC1405CTS is typically supplied by a PWM controller that regulates the power supply output. (See Application Evaluation Schematic, Figure 3). The timing diagram demonstrates the sequence of events by which the top and bottom drive signals are

applied. The shoot-through protection is implemented by holding the bottom FET off until the voltage at the phase node (intersection of top FET source, the output inductor and the bottom FET drain) has dropped below 1V. This assures that the top FET has turned off and that a direct current path does not exist between the input supply and ground, a condition which both the top and bottom FET's are on momentarily. The top FET is also prevented from turning on until the bottom FET is off. This time is internally set to 20ns (typical) and may be increased by adding a capacitor from the C-Delay pin to GND. The delay is approximately 1ns/pf in addition to the internal 20ns delay. The external capacitor may be needed if multiple High input capacitance MOSFET's are used in parallel and the fall time is substantially greater than 20ns.

It must be noted that increasing the dead-time by high values of C-Delay capacitor will reduce efficiency since the parallel Schottky or the bottom FET body diode will have to conduct during dead-time.

Layout Guidelines

As with any high speed, high current circuit, proper layout is critical in achieving optimum performance of the SC1405CTS. The Evaluation board schematic (Refer to figure 3) shows a dual phase synchronous design with all surface mountable components.

While components connecting to C-Delay, OVP_S, EN, S-

POWER MANAGEMENT
Applications Information

MOD, DSPTS_DR and PRDY are relatively noncritical, tight placement and short, wide traces must be used in layout of The Drives, DRN, and especially PGND pin. The top gate driver supply voltage is provided by bootstrapping the +5V supply and adding it the phase node voltage (DRN). Since the bootstrap capacitor supplies the charge to the TOP gate, it must be less than .5" away from the SC1405. Ceramic X7R capacitors are a good choice for supply bypassing near the chip. The Vcc pin capacitor must also be less than .5" away from the SC1405. The ground node of this capacitor, the SC1405 PGND pin and the Source of the bottom FET must be very close to each other, preferably with common PCB copper land and multiple vias to the ground plane (if used). The parallel Schottky (if used) must be physically next to the Bottom FETS Drain and source. Any trace or lead inductance in these connections will drive current way from the Schottky and allow it to flow through the FET's Body diode, thus reducing efficiency.

Preventing Inadvertent Bottom FET Turn-on

At high input voltages, (12V and greater) a fast turn-on of the top FET creates a positive going spike on the Bottom FET's gate through the Miller capacitance, Crss of the bottom FET. The voltage appearing on the gate due to this spike is:

$$V_{spike} = \frac{V_{in} * cr_{ss}}{(Cr_{ss} + c_{iss})}$$

Where Ciss is the input gate capacitance of the bottom FET. This is assuming that the impedance of the drive path is too high compared to the instantaneous impedance of the capacitors. (since dV/dT and thus the effective frequency is very high). If the BG pin of the SC1405CTS is very close to the bottom FET, Vspike will be reduced depending on trace inductance, rate of rise of current, etc.

While not shown in Figure 3, a capacitor may be added from the gate of the Bottom FET to its source, preferably less than .5" away. This capacitor will be added to Ciss in the above equation to reduce the effective spike voltage, Vspike.

The selection of the bottom MOSFET must be done with attention paid to the Crss/Ciss ratio. A low ratio reduces

the Miller feedback and thus reduces Vspike. Also MOSFETs with higher Turn-on threshold voltages will conduct at a higher voltage and will not turn on during the spike. The MOSFET shown in the schematic has a 2 volt threshold and will require approximately 5 volts Vgs to be conducting, thus reducing the possibility of shoot-through. A zero ohm bottom FET gate resistor will obviously help keeping the gate voltage low.

Ultimately, slowing down the top FET by adding gate resistance will reduce di/dt which will in turn make the effective impedance of the capacitors higher, thus allowing the BG driver to hold the bottom gate voltage low.

Ringings on the Phase Node

The top MOSFET source must be close to the bottom MOSFET drain to prevent ringing and the possibility of the phase node going negative. This frequency is determined by:

$$F_{ring} = \frac{1}{(2 \pi * \text{Sqrt}(L_{st} * C_{oss}))}$$

Where:

L_{st} = The effective stray inductance of the top FET added to trace inductance of the connection between top FET's source and the bottom FET's drain added to the trace resistance of the bottom FET's ground connection.

Coss=Drain to source capacitance of bottom FET. If there is a Schottky used, the capacitance of the Schottky is added to the value.

Although this ringing does not pose any power losses due to a fairly high Q, it could cause the phase node to go too far negative, thus causing improper operation, double pulsing or at worst driver damage. This ringing is also an EMI nuisance due to its high resonant frequency. Adding a capacitor, typically 1000-2000pf, in parallel with Coss can often eliminate the EMI issue. If double pulsing is caused due to excessive ringing, placing 4.7-10 ohm resistor between the phase node and the DRN pin of the SC1405 should eliminate the double pulsing.

The negative voltage spikes on the phase node adds to the bootstrap capacitor voltage, thus increasing the voltage between VBST - VDRN. If the phase node negative

POWER MANAGEMENT**Applications Information (Cont.)**

spikes are too large, the voltage on the boost capacitor could exceed device's absolute maximum rating of 8V. To eliminate the effect of the ringing on the boost capacitor voltage, place a 4.7 - 10 Ohm resistor between boost Schottky diode and Vcc to filter the negative spikes on DRN Pin. Alternately, a Silicon diode, such as the commonly available 1N4148 can substitute for the Schottky diode and eliminate the need for the series resistor.

Proper layout will guarantee minimum ringing and eliminate the need for external components. Use of SO-8 or other surface mount MOSFETs will reduce lead inductance and their parasitic effects.

ASYNCHRONOUS OPERATION

The SC1405CTS can be configured to operate in Asynchronous mode by pulling S-MOD to logic LOW, thus disabling the bottom FET drive. This has the effect of saving power at light loads since the bottom FET's gate capacitance does not have to be charged at the switching frequency. There can be a significant savings since the bottom driver can supply up to 2A pulses to the FET at the switching frequency. There is an additional efficiency benefit to operating in asynchronous mode. When operating in synchronous mode, the inductor current can go

negative and flow in reverse direction when the bottom FET is on and the DC load is less than 1/2 inductor ripple current. At that point, the inductor core and wire losses, depending on the magnitude of the ripple current, can be quite significant. Operating in asynchronous mode at light loads effectively only charges the inductor by as much as needed to supply the load current, since the inductor never completely discharges at light loads. DC regulation can be an issue when operating in asynchronous mode, depending on the type of controller used and minimum load required to maintain regulation. If there are no Schottky diodes used in parallel with bottom FET, the FET's body diode will need to conduct in asynchronous mode. The high voltage drop of this diode must be considered when determining the criteria for this mode of operation.

DSPS DR

This pin produces an output which is a logical duplicate of the bottom FET's gate drive, if S-MOD is held LOW.

OVP_S/OVER TEMP SHUTDOWN

Output over-voltage protection (OVP) may be implemented on the SC1405CTS independent of the PWM controller. A voltage divider from the output is compared with the internal bandgap voltage of 1.2V (typical). Upon exceeding this voltage, the overvoltage comparator disables the top FET, while turning on the bottom FET to allow discharge of the output capacitors excessive voltage through the output inductor.

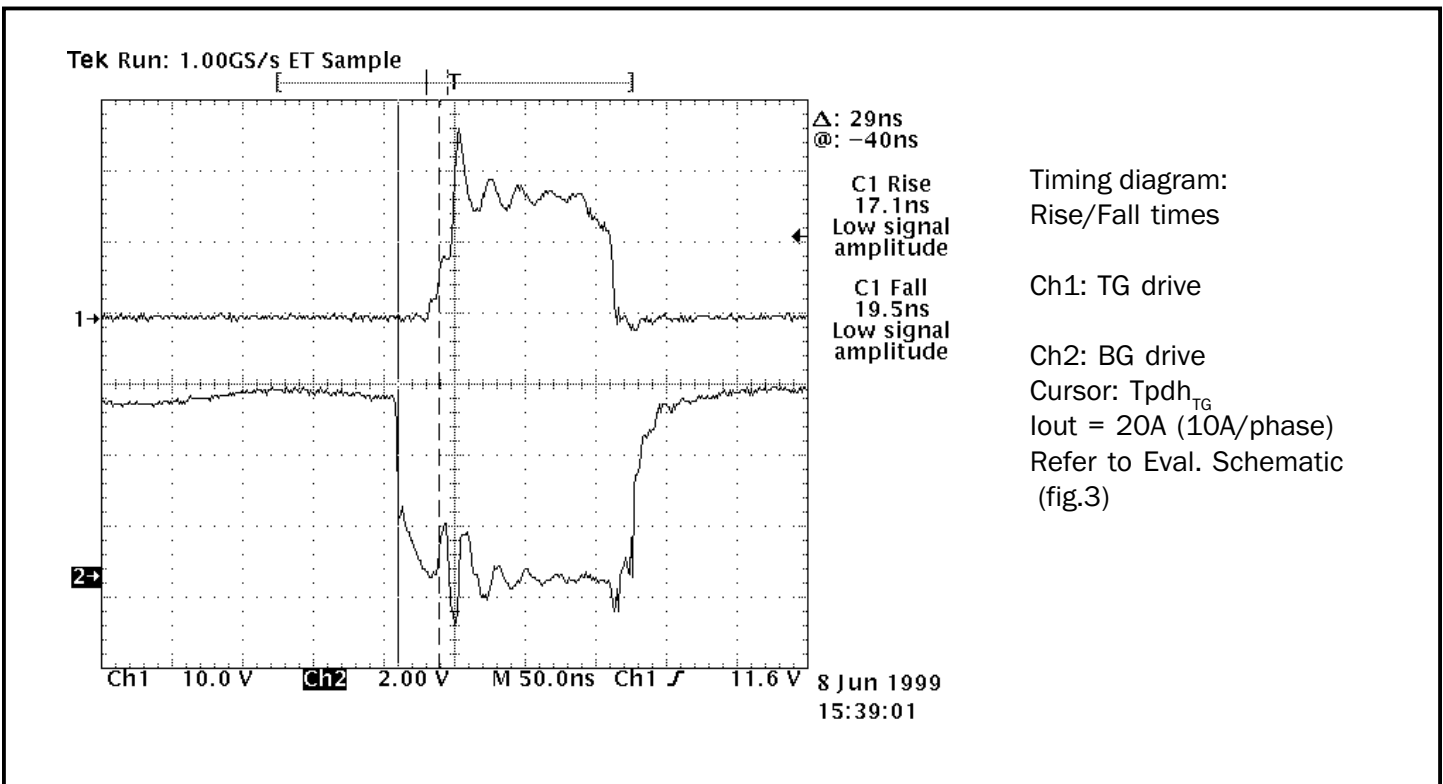
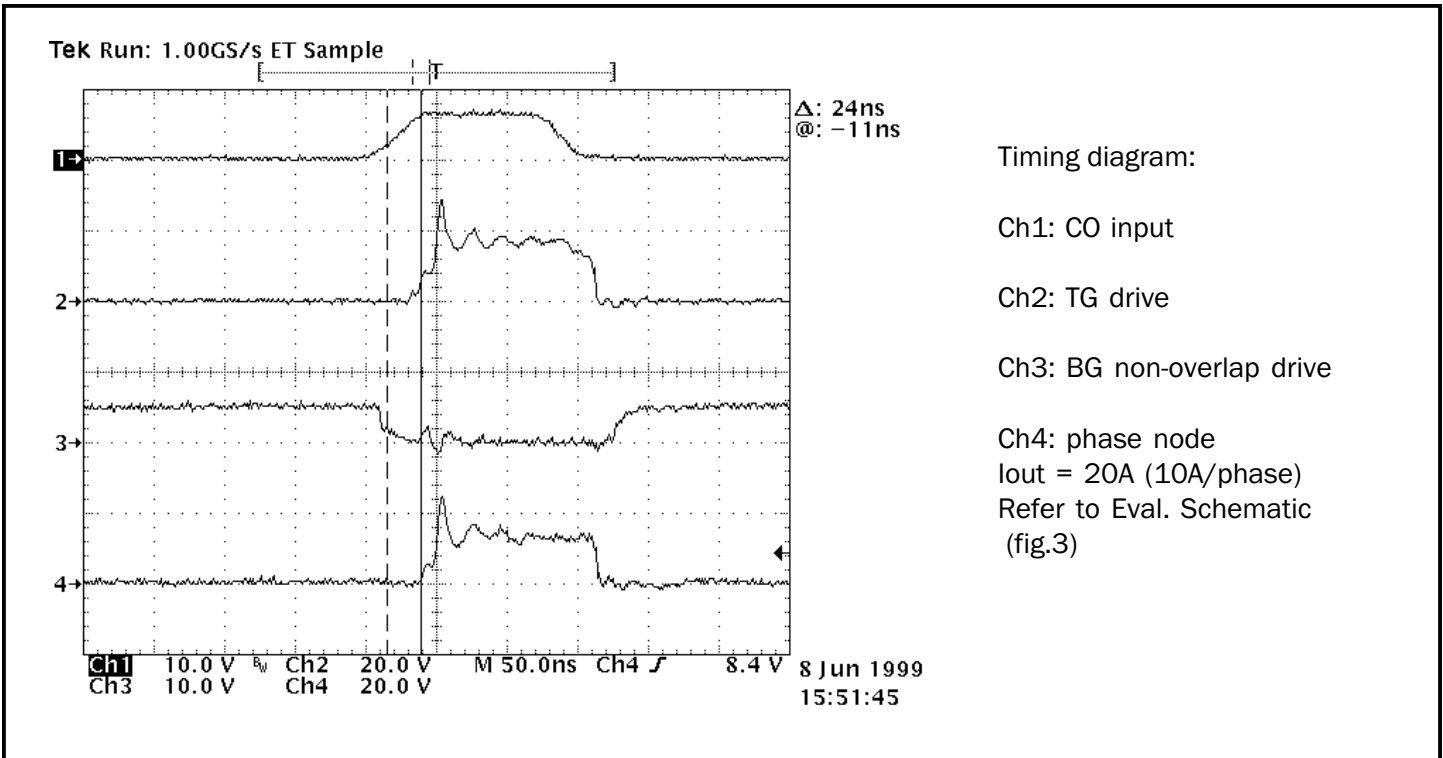
The SC1405CTS has a unique adaptive OVP circuit. Short noise pulses, less than ~100ns are rejected completely; longer pulses will trigger OVP if only of sufficient magnitude. A long term transient will trigger OVP with a smaller magnitude. To assure proper tripping, bypass the resistor from OVP_S pin to GND with a capacitor. The value of this capacitor must be selected to achieve a time constant equal to one switching period. Leave at least 250mV headroom on the OVP pin to prevent false OVP events.

The SC1405CTS will shutdown if its T_j exceeds 165 °C.

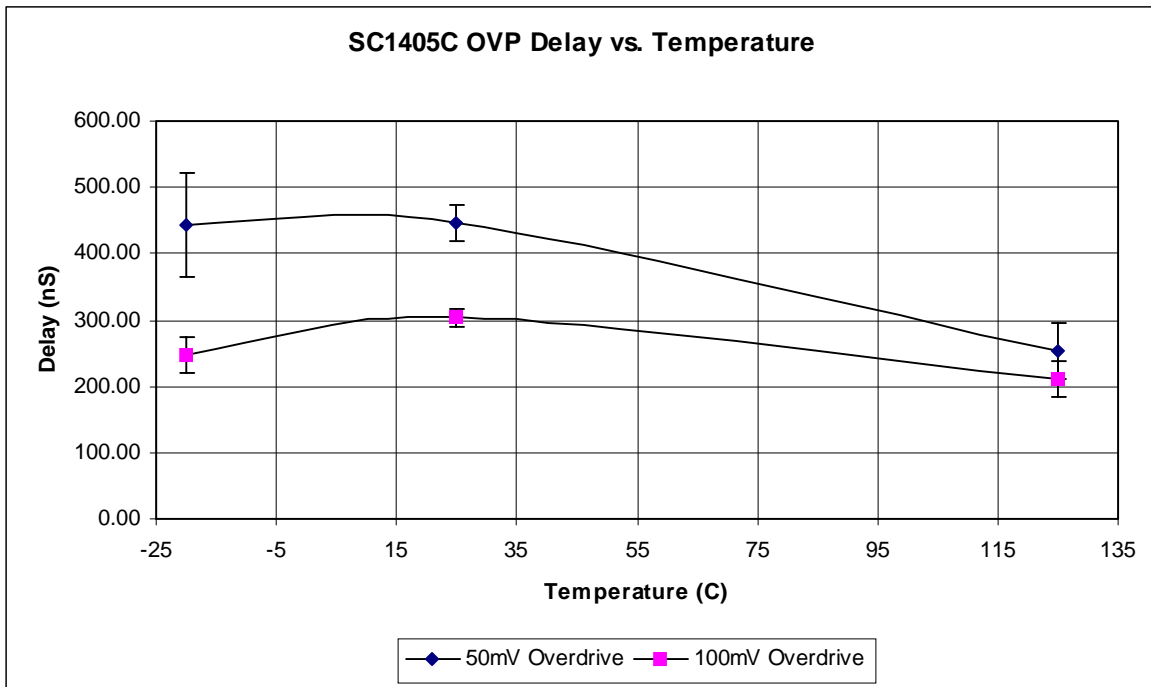
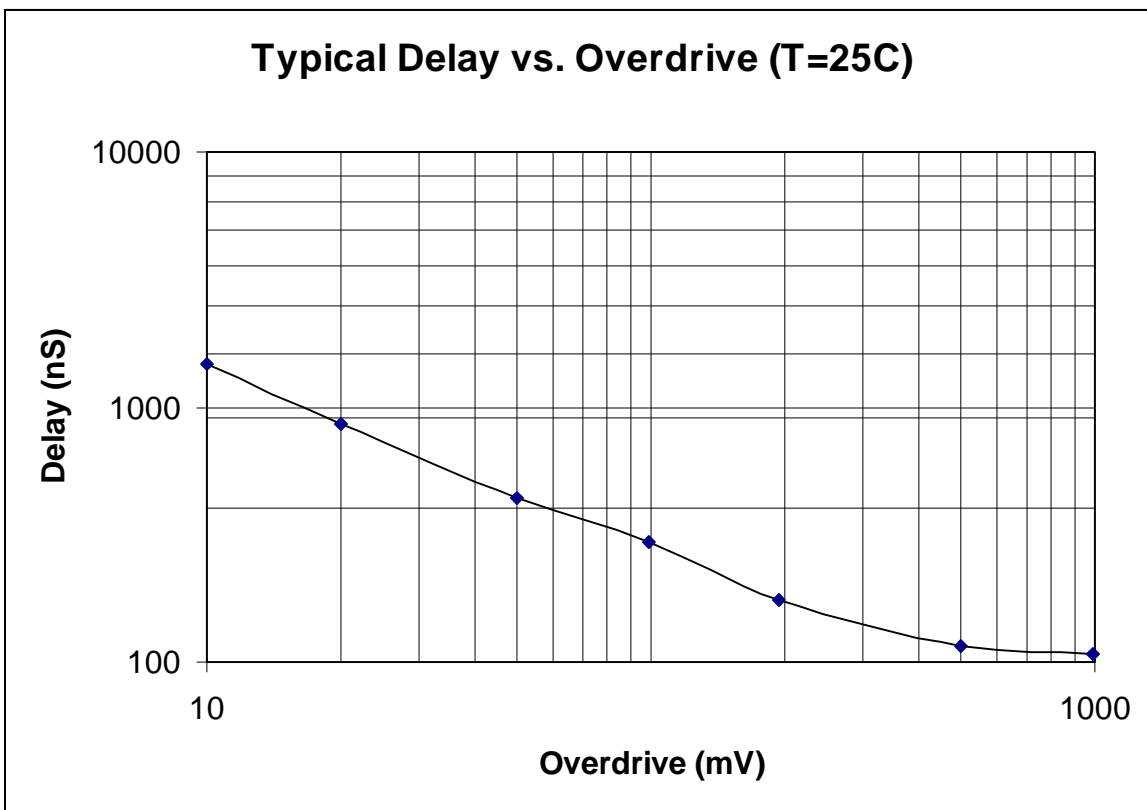
POWER MANAGEMENT

Typical Characteristics

Performance diagrams, Application Evaluation Board.

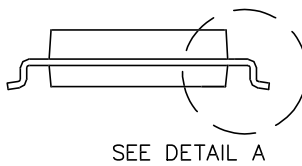
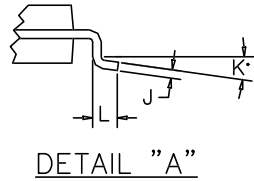
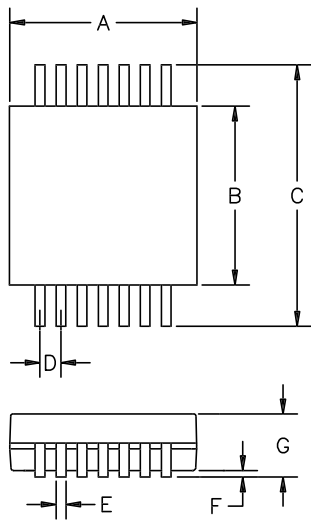


V_{IN} = 12V, V_{OUT} = 1.6V. Top FET = IR7811 FDB7030(BL) Q_{gd} = 23nc

POWER MANAGEMENT
Typical Characteristics (Cont.)
Delay vs. Temp.

Delay vs. Overdrive


POWER MANAGEMENT

Outline Drawing -TSSOP-14



DIM ^N	DIMENSIONS ①				NOTE
	INCHES		MM		
	MIN	MAX	MIN	MAX	
A	.193	.201	4.90	5.10	②
B	.169	.177	4.30	4.50	②
C	.252 BSC		6.40 BSC		—
D	.026 BSC		.65 BSC		—
E	.007	.012	.19	.30	—
F	.002	.006	.05	.15	—
G	.047		1.20		—
J	.004	.008	.09	.20	—
K	0° 8°		0° 8°		—
L	.018	.030	.45	.75	—

JEDEC MO-153AB1

- ② DIMENSIONS A AND B DO NOT INCLUDE MOLD PROTRUSIONS.
- ① CONTROLLING DIMENSIONS: MILLIMETERS.

Contact Information

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